

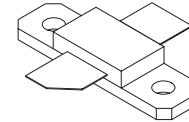
The RF Sub-Micron MOSFET Line **RF Power Field Effect Transistors** N-Channel Enhancement-Mode Lateral MOSFETs

Designed for broadband commercial and industrial applications with frequencies up to 1.0 GHz. The high gain and broadband performance of these devices make them ideal for large-signal, common-source amplifier applications in 26 volt base station equipment.

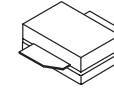
- Typical Two-Tone Performance at 945 MHz, 26 Volts
 - Output Power — 30 Watts PEP
 - Power Gain — 19 dB
 - Efficiency — 41.5%
 - IMD — -32.5 dBc
- Integrated ESD Protection
- Designed for Maximum Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 26 Vdc, 945 MHz, 30 Watts CW Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- In Tape and Reel. R1 Suffix = 500 Units per 32 mm, 13 inch Reel.

MRF9030R1 MRF9030SR1

945 MHz, 30 W, 26 V
LATERAL N-CHANNEL
BROADBAND
RF POWER MOSFETs



CASE 360B-05, STYLE 1
NI-360
MRF9030R1



CASE 360C-05, STYLE 1
NI-360S
MRF9030SR1

MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	V _{DSS}	68	Vdc
Gate-Source Voltage	V _{GS}	-0.5, +15	Vdc
Total Device Dissipation @ T _C = 25°C Derate above 25°C	MRF9030R1	P _D 92 0.53	Watts W/°C
Total Device Dissipation @ T _C = 25°C Derate above 25°C	MRF9030SR1	P _D 117 0.67	Watts W/°C
Storage Temperature Range	T _{stg}	-65 to +200	°C
Operating Junction Temperature	T _J	200	°C

ESD PROTECTION CHARACTERISTICS

Test Conditions	Class
Human Body Model	1 (Minimum)
Machine Model	M1 (Minimum)

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	MRF9030R1 MRF9030SR1	R _{θJC} 1.9 1.5	°C/W

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

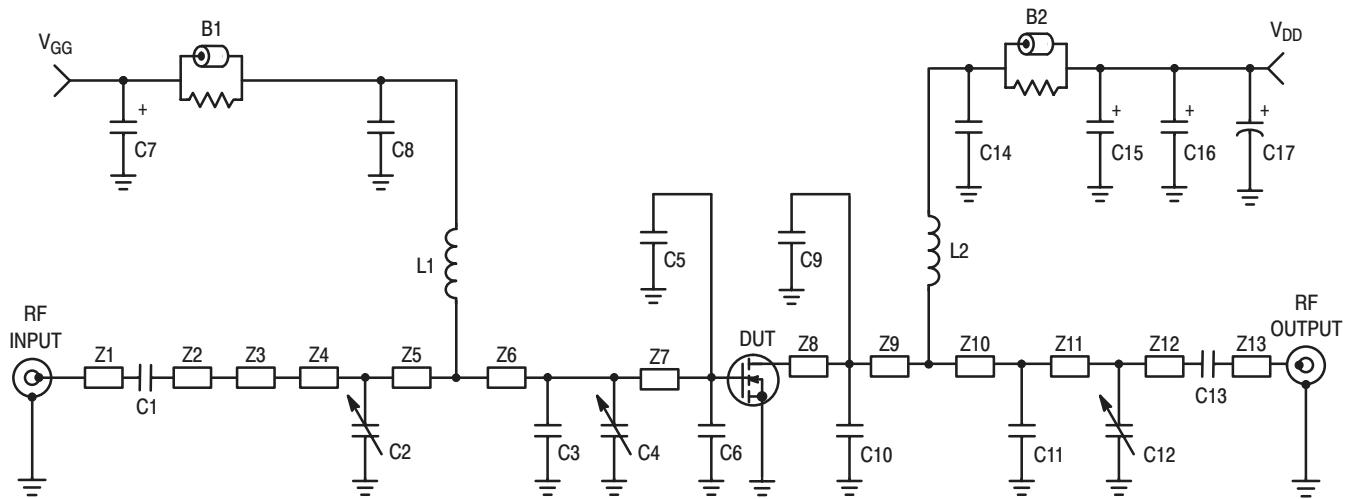
ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
OFF CHARACTERISTICS					
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 68 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	10	$\mu\text{A dc}$
Zero Gate Voltage Drain Leakage Current ($V_{DS} = 26 \text{ Vdc}$, $V_{GS} = 0 \text{ Vdc}$)	I_{DSS}	—	—	1	$\mu\text{A dc}$
Gate-Source Leakage Current ($V_{GS} = 5 \text{ Vdc}$, $V_{DS} = 0 \text{ Vdc}$)	I_{GSS}	—	—	1	$\mu\text{A dc}$
ON CHARACTERISTICS					
Gate Threshold Voltage ($V_{DS} = 10 \text{ Vdc}$, $I_D = 100 \mu\text{A dc}$)	$V_{GS(\text{th})}$	2	2.9	4	Vdc
Gate Quiescent Voltage ($V_{DS} = 26 \text{ Vdc}$, $I_D = 250 \text{ mA dc}$)	$V_{GS(Q)}$	—	3.8	—	Vdc
Drain-Source On-Voltage ($V_{GS} = 10 \text{ Vdc}$, $I_D = 0.7 \text{ Adc}$)	$V_{DS(\text{on})}$	—	0.19	0.4	Vdc
Forward Transconductance ($V_{DS} = 10 \text{ Vdc}$, $I_D = 2 \text{ Adc}$)	g_{fs}	—	3	—	S
DYNAMIC CHARACTERISTICS					
Input Capacitance ($V_{DS} = 26 \text{ Vdc} \pm 30 \text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{iss}	—	49.5	—	pF
Output Capacitance ($V_{DS} = 26 \text{ Vdc} \pm 30 \text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{oss}	—	26.5	—	pF
Reverse Transfer Capacitance ($V_{DS} = 26 \text{ Vdc} \pm 30 \text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$)	C_{rss}	—	1	—	pF

(continued)

ELECTRICAL CHARACTERISTICS — continued ($T_C = 25^\circ\text{C}$ unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
FUNCTIONAL TESTS (In Motorola Test Fixture, 50 ohm system)					
Two-Tone Common-Source Amplifier Power Gain ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	G_{ps}	18	19	—	dB
Two-Tone Drain Efficiency ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	η	37	41.5	—	%
3rd Order Intermodulation Distortion ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	IMD	—	-32.5	-28	dBc
Input Return Loss ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$, $f_2 = 945.1 \text{ MHz}$)	IRL	—	-15.5	-9	dB
Two-Tone Common-Source Amplifier Power Gain ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	G_{ps}	—	19	—	dB
Two-Tone Drain Efficiency ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	η	—	41.5	—	%
3rd Order Intermodulation Distortion ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	IMD	—	-33	—	dBc
Input Return Loss ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W PEP}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 930.0 \text{ MHz}$, $f_2 = 930.1 \text{ MHz}$ and $f_1 = 960.0 \text{ MHz}$, $f_2 = 960.1 \text{ MHz}$)	IRL	—	-14	—	dB
Power Output, 1 dB Compression Point ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W CW}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$)	$P_{1\text{dB}}$	—	30	—	W
Common-Source Amplifier Power Gain ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W CW}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$)	G_{ps}	—	19	—	dB
Drain Efficiency ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W CW}$, $I_{DQ} = 250 \text{ mA}$, $f_1 = 945.0 \text{ MHz}$)	η	—	60	—	%
Output Mismatch Stress ($V_{DD} = 26 \text{ Vdc}$, $P_{out} = 30 \text{ W CW}$, $I_{DQ} = 250 \text{ mA}$, $f = 945.0 \text{ MHz}$, $VSWR = 10:1$, All Phase Angles at Frequency of Tests)	Ψ	No Degradation In Output Power			



B1	Short Ferrite Bead
B2	Long Ferrite Bead
C1, C8, C13, C14	47 pF Chip Capacitors, B Case
C2, C4	0.8 pF to 8.0 pF Trim Capacitors
C3	3.9 pF Chip Capacitor, B Case
C5, C6	7.5 pF Chip Capacitors, B Case
C7, C15, C16	10 µF, 35 V Tantalum Capacitors
C9, C10	10 pF Chip Capacitors, B Case
C11	9.1 pF Chip Capacitor, B Case
C12	0.6 pF to 4.5 pF Trim Capacitor
C17	220 µF, 50 V Electrolytic Capacitor
L1, L2	12.5 nH Surface Mount Inductors
Z1	0.260" x 0.060" Microstrip
Z2	0.240" x 0.060" Microstrip
Z3	0.500" x 0.100" Microstrip
Z4	0.215" x 0.270" Microstrip
Z5	0.315" x 0.270" Microstrip
Z6	0.160" x 0.270" x 0.520", Taper
Z7	0.285" x 0.520" Microstrip
Z8	0.140" x 0.270" Microstrip
Z9	0.450" x 0.270" Microstrip
Z10	0.250" x 0.060" Microstrip
Z11	0.720" x 0.060" Microstrip
Z12	0.490" x 0.060" Microstrip
Z13	0.290" x 0.060" Microstrip
Board	Taconic RF-35-0300, ($\epsilon_r = 3.5$) CAX1/CAX1

Figure 1. 945 MHz Broadband Test Circuit Schematic

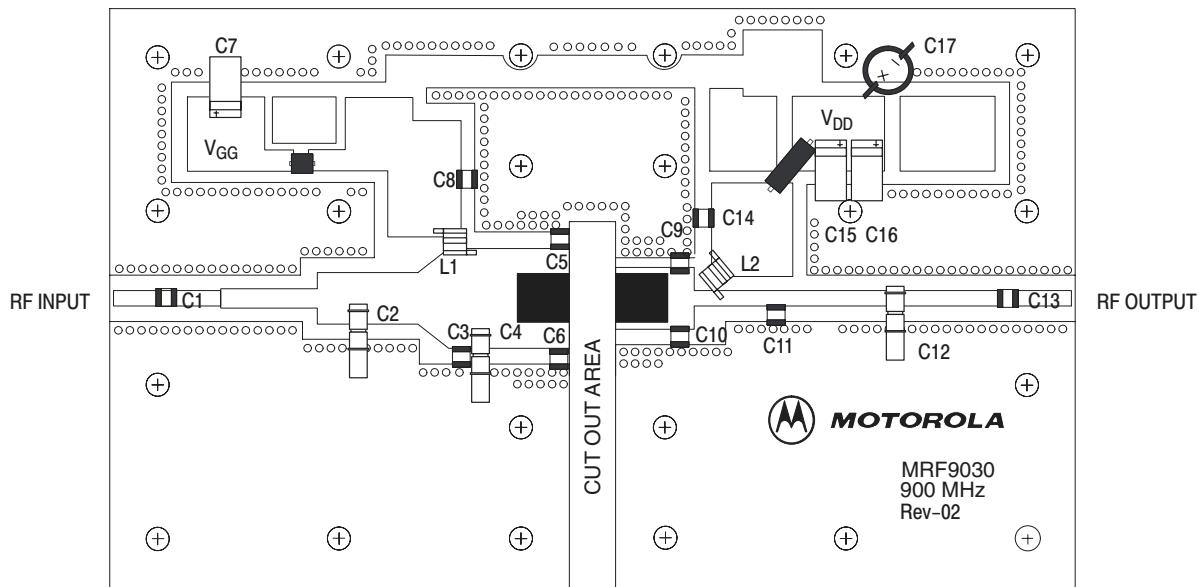


Figure 2. 945 MHz Broadband Test Circuit Component Layout

TYPICAL CHARACTERISTICS

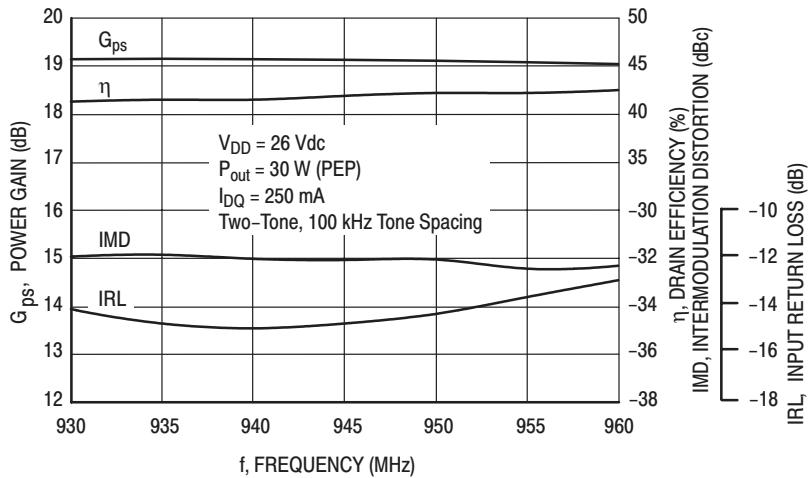


Figure 3. Class AB Broadband Circuit Performance

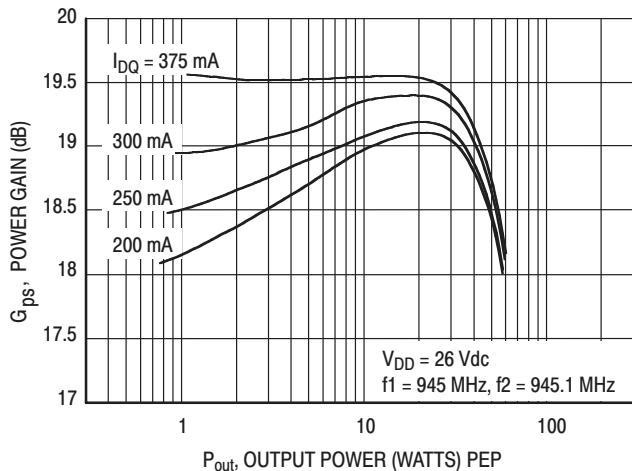


Figure 4. Power Gain versus Output Power

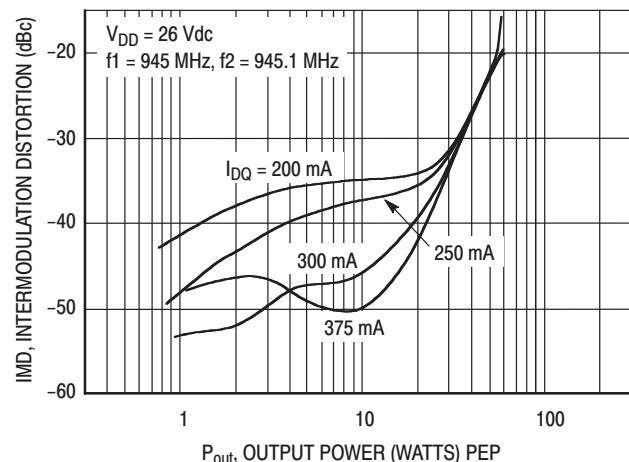


Figure 5. Intermodulation Distortion versus Output Power

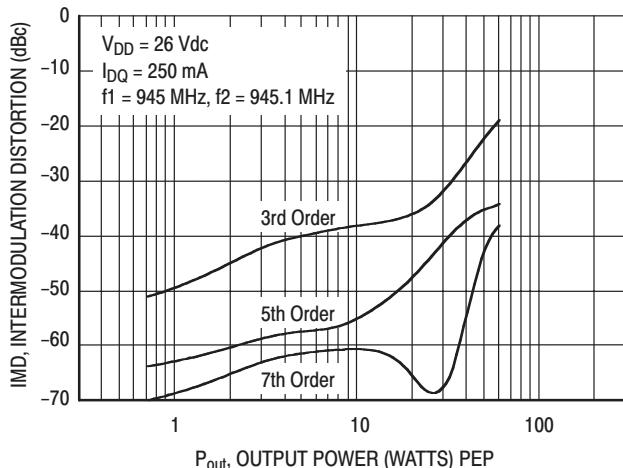


Figure 6. Intermodulation Distortion Products versus Output Power

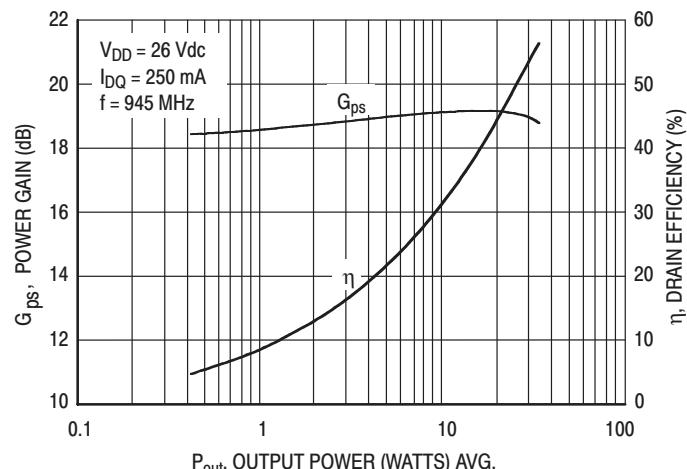


Figure 7. Power Gain and Efficiency versus Output Power

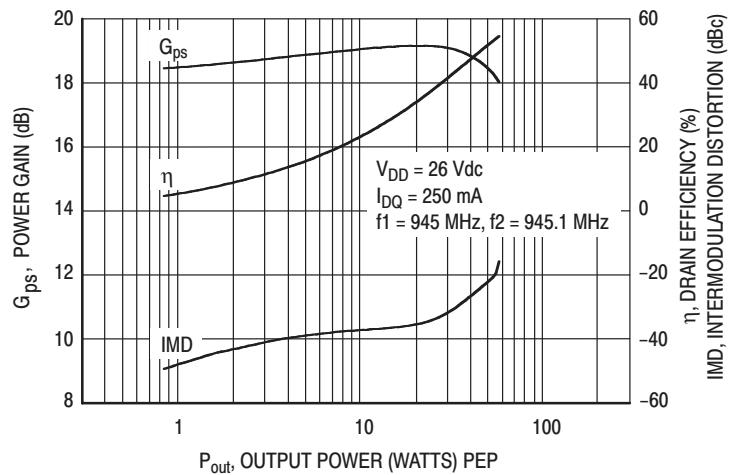
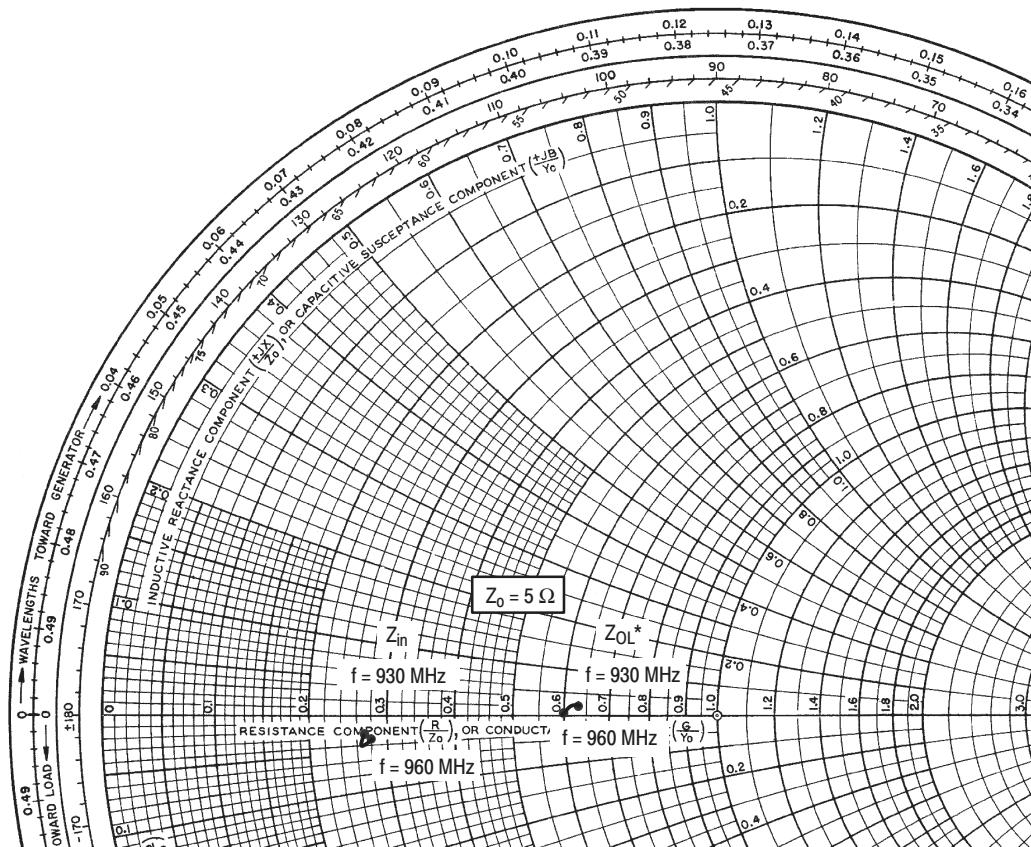


Figure 8. Power Gain, Efficiency and IMD versus Output Power



$V_{DD} = 26 \text{ V}, I_{DQ} = 250 \text{ mA}, P_{out} = 30 \text{ W PEP}$

f MHz	Z_{in} Ω	Z_{OL^*} Ω
930	$1.34 - j0.1$	$3.175 + j0.09$
945	$1.36 - j0.2$	$3.1 + j0.08$
960	$1.4 - j0.14$	$3.0 + j0.05$

Z_{in} = Complex conjugate of source impedance.

Z_{OL^*} = Complex conjugate of the optimum load impedance at a given output power, voltage, IMD, bias current and frequency.

Note: Z_{OL^*} was chosen based on tradeoffs between gain, output power, drain efficiency and intermodulation distortion.

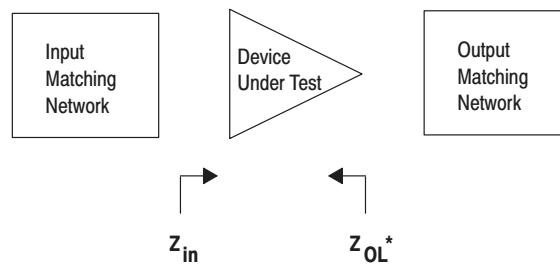


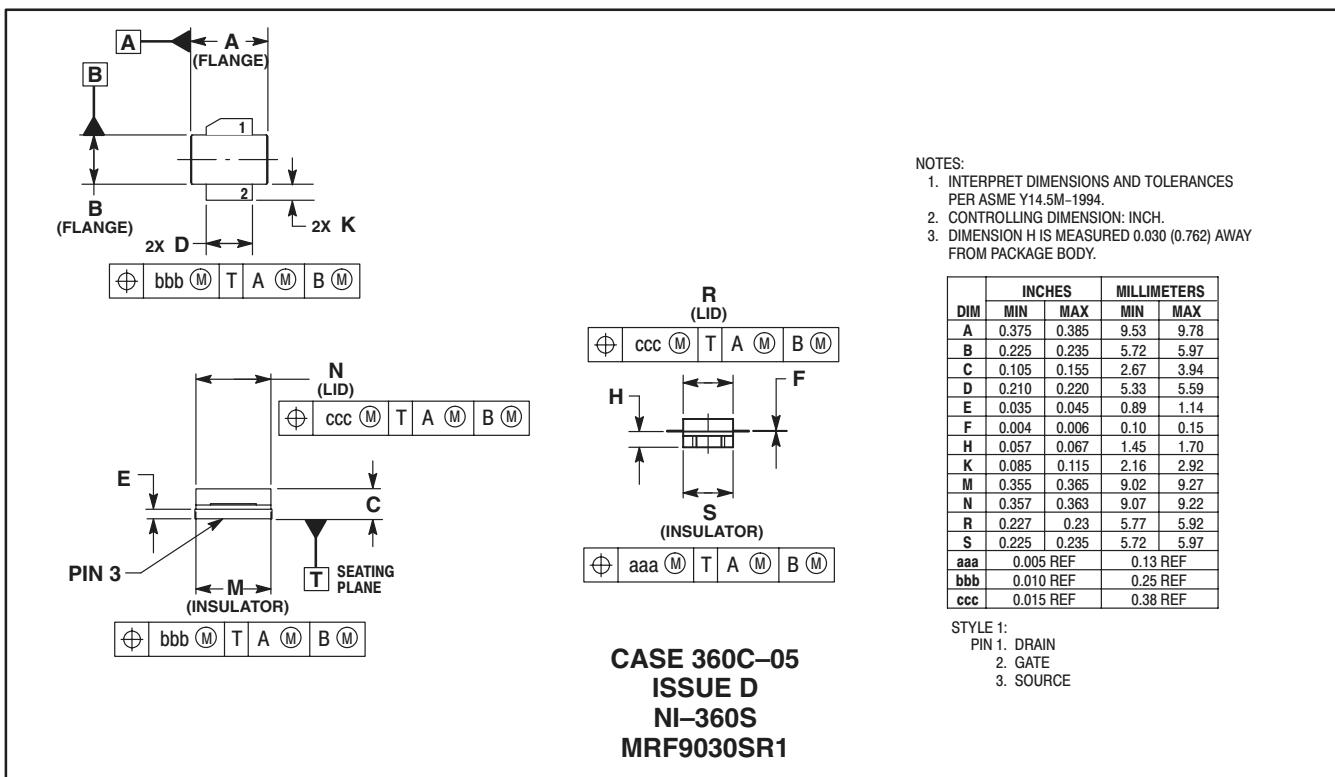
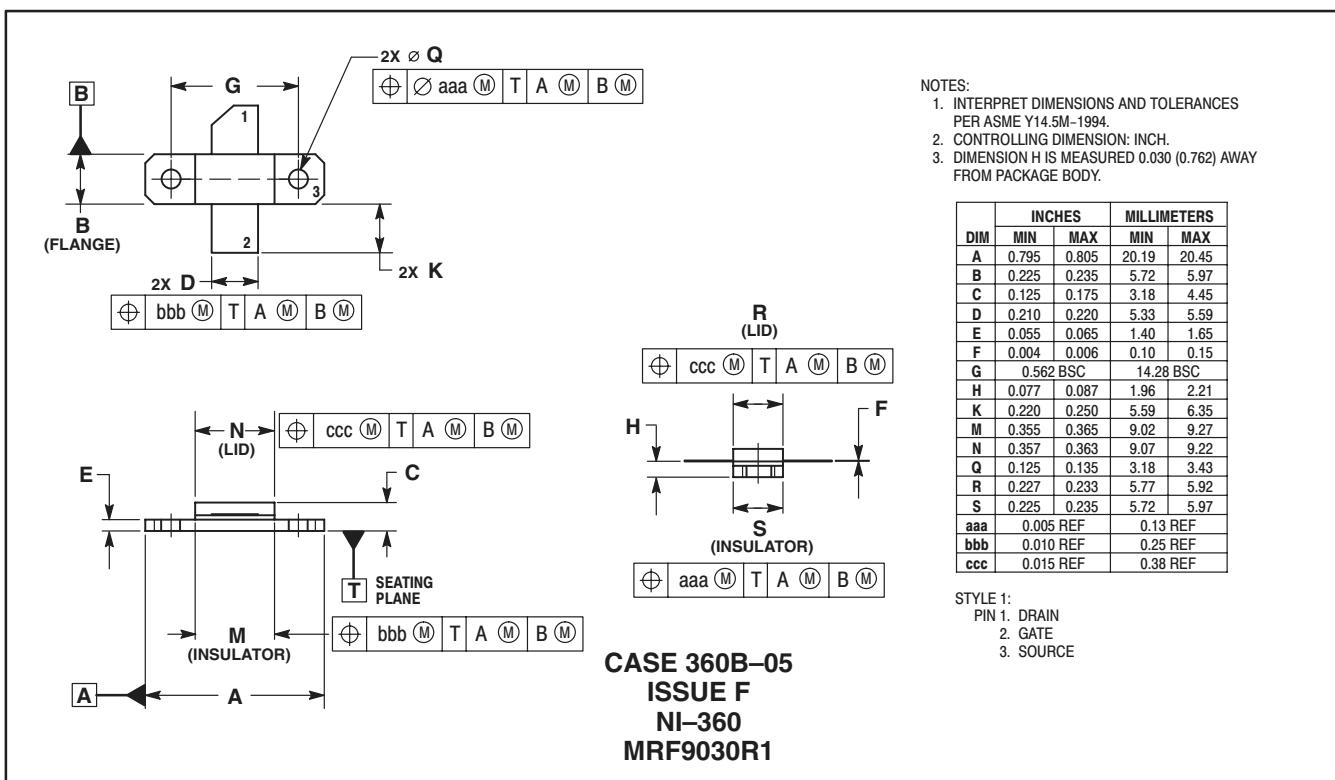
Figure 9. Series Equivalent Input and Output Impedance

NOTES

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PACKAGE DIMENSIONS



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