



TS486 TS487

100mW STEREO HEADPHONE AMPLIFIER WITH STANDBY MODE

- OPERATING FROM $V_{CC}=2V$ to 5.5V
- **STANDBY MODE** ACTIVE LOW (TS486) or HIGH (TS487)
- OUTPUT POWER: 102mW @5V, 38mW @3.3V into 16 Ω with 0.1% THD+N max (1kHz)
- **LOW CURRENT CONSUMPTION:** 2.5mA max
- High Signal-to-Noise ratio: 103dB(A) at 5V
- High Crosstalk immunity: 83dB (F=1kHz)
- PSRR: 58 dB (F=1kHz), inputs grounded
- ON/OFF click reduction circuitry
- Unity-Gain Stable
- **SHORT CIRCUIT LIMITATION**
- Available in SO8, MiniSO8 & DFN 3x3mm

DESCRIPTION

The TS486/7 is a dual audio power amplifier capable of driving, in single-ended mode, either a 16 or a 32 Ω stereo headset.

Capable of descending to low voltages, it delivers up to 90mW per channel (into 16 Ω loads) of continuous average power with 0.3% THD+N in the audio bandwidth from a 5V power supply.

An externally-controlled standby mode reduces the supply current to 10nA (typ.). The unity gain stable TS486/7 can be configured by external gain-setting resistors or used in a fixed gain version.

APPLICATIONS

- Headphone Amplifier
- Mobile phone, PDA, computer motherboard
- High end TV, portable audio player

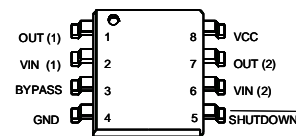
ORDER CODE

Part Number	Temperature Range: I	Package			Gain	Marking
		D	S	Q		
TS486	-40, +85°C	•			external	TS486I
TS487		•			external	TS487I
TS486			•	•	external	K86A
TS486-1			tba	tba	x1/0dB	K86B
TS486-2			tba	tba	x2/6dB	K86C
TS486-4			tba	tba	x4/12dB	K86D
TS487			•	•	external	K87A
TS487-1			tba	tba	x1/0dB	K87B
TS487-2			tba	tba	x2/6dB	K87C
TS487-4			tba	tba	x4/12dB	K87D

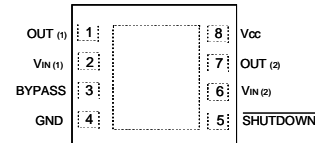
MiniSO & DFN only available in Tape & Reel with T suffix,
SO is available in Tube (D) and in Tape & Reel (DT)

PIN CONNECTIONS (top view)

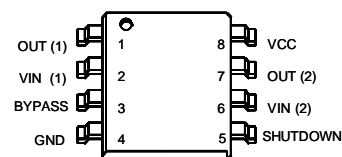
TS486IDT: SO8, TS486IST, TS486-1IST, TS486-2IST, TS486-4IST: MiniSO8



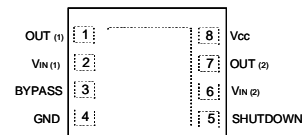
TS486-IQT, TS486-1IQT, TS486-2IQT, TS486-4IQT: DFN8



TS487IDT: SO8, TS487IST, TS487-1IST, TS487-2IST, TS487-4IST: MiniSO8



TS487-IQT, TS487-1IQT, TS487-2IQT, TS487-4IQT: DFN8



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_{CC}	Supply voltage ¹⁾	6	V
V_i	Input Voltage	-0.3v to $V_{CC}+0.3v$	V
T_{stg}	Storage Temperature	-65 to +150	°C
T_j	Maximum Junction Temperature	150	°C
R_{thja}	Thermal Resistance Junction to Ambient SO8 MiniSO8 DFN8	175 215 70	°C/W
P_d	Power Dissipation ²⁾ SO8 MiniSO8 DFN8	0.71 0.58 1.79	W
ESD	Human Body Model (pin to pin): TS486, TS487 ³⁾	1.5	kV
ESD	Machine Model - 220pF - 240pF (pin to pin)	100	V
Latch-up	Latch-up Immunity (All pins)	200	mA
	Lead Temperature (soldering, 10sec)	250	°C
	Output Short-Circuit to V_{CC} or GND	continuous ⁴⁾	

1. All voltage values are measured with respect to the ground pin.

2. P_d has been calculated with $T_{amb} = 25^{\circ}\text{C}$, $T_{junction} = 150^{\circ}\text{C}$.

3. TS487 stands 1.5KV on all pins except standby pin which stands 1KV.

4. Attention must be paid to continuous power dissipation ($V_{DD} \times 300\text{mA}$). Exposure of the IC to a short circuit for an extended time period is dramatically reducing product life expectancy.

OPERATING CONDITIONS

Symbol	Parameter	Value	Unit
V_{CC}	Supply Voltage	2 to 5.5	V
R_L	Load Resistor	≥ 16	Ω
T_{oper}	Operating Free Air Temperature Range	-40 to + 85	°C
C_L	Load Capacitor $R_L = 16$ to 100Ω $R_L > 100\Omega$	400 100	pF
V_{STB}	Standby Voltage Input TS486 ACTIVE / TS487 in STANDBY TS486 in STANDBY / TS487 ACTIVE	$1.5 \leq V_{STB} \leq V_{CC}$ $GND \leq V_{STB} \leq 0.4$ ¹⁾	V
R_{THJA}	Thermal Resistance Junction to Ambient SO8 MiniSO8 DFN8 ²⁾	150 190 41	°C/W

1. The minimum current consumption ($I_{STANDBY}$) is guaranteed at GND (TS486) or V_{CC} (TS487) for the whole temperature range.

2. When mounted on a 4-layer PCB.

FIXED GAIN VERSION SPECIFIC ELECTRICAL CHARACTERISTICS

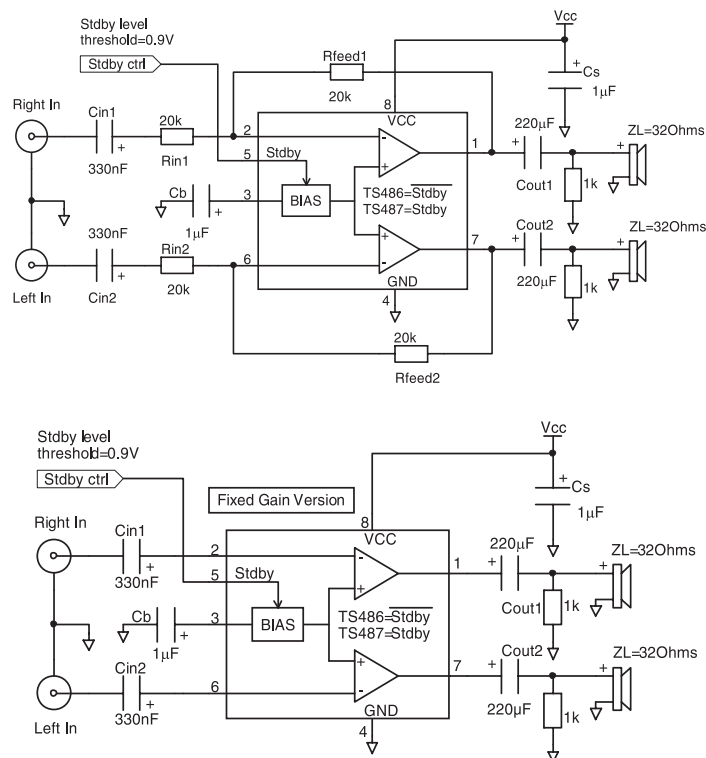
V_{CC} from +5V to +2V, GND = 0V, $T_{amb} = 25^{\circ}\text{C}$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
$R_{IN\ 1,2}$	Input Resistance ¹⁾		20		k Ω
G	Gain value for Gain TS486/TS487-1		0dB		dB
	Gain value for Gain TS486/TS487-2		6dB		
	Gain value for Gain TS486/TS487-4		12dB		

1. See figure 30 to establish the value of C_{in} vs. -3dB cut off frequency.

APPLICATION COMPONENTS INFORMATION

Components	Functional Description
$R_{IN1,2}$	Inverting input resistor which sets the closed loop gain in conjunction with R_{FEED} . This resistor also forms a high pass filter with C_{IN} ($f_c = 1 / (2 \times \pi \times R_{IN} \times C_{IN})$). Not needed in fixed gain versions.
$C_{IN1,2}$	Input coupling capacitor which blocks the DC voltage at the amplifier's input terminal.
$R_{FEED1,2}$	Feedback resistor which sets the closed loop gain in conjunction with R_{IN} . $A_V = \text{Closed Loop Gain} = -R_{FEED}/R_{IN}$. Not needed in fixed gain versions.
C_S	Supply Bypass capacitor which provides power supply filtering.
C_B	Bypass capacitor which provides half supply filtering.
$C_{OUT1,2}$	Output coupling capacitor which blocks the DC voltage at the load input terminal. This capacitor also forms a high pass filter with R_L ($f_c = 1 / (2 \times \pi \times R_L \times C_{OUT})$).

TYPICAL APPLICATION SCHEMATICS

ELECTRICAL CHARACTERISTICS $V_{CC} = +5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.8	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ¹⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 0.1% Max, $F = 1kHz$, $R_L = 16\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 16\Omega$	60 95	64 65 102 108		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 60mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 90mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ²⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, $F = 1kHz$, $V_{ripple} = 200mV_{pp}$	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	106	115		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	4.45 4.2	0.45 4.52 0.6 4.35	0.5 0.7	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ²⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	103		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$ Channel Separation, $R_L = 16\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$		83 79 80 72		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. Only for external gain version.

2. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS

$V_{CC} = +3.3V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified) ¹⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.8	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ²⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 0.1% Max, $F = 1kHz$, $R_L = 16\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 16\Omega$	23 36	26 28 38 42		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 35mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ³⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, $F = 1kHz$, $V_{ripple} = 200mV_{pp}$	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	64	75		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	2.85 2.68	0.3 3 0.45 2.85	0.38 0.52	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ³⁾ ($R_L = 32\Omega$, THD + N < 0.4%, $20Hz \leq F \leq 20kHz$)	80	98		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$ Channel Separation, $R_L = 16\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. All electrical values are guaranteed with correlation measurements at 2V and 5V.

2. Only for external gain version.

3. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS
 $V_{CC} = +2.5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)¹⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.7	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ²⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 0.1% Max, $F = 1kHz$, $R_L = 16\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 16\Omega$	12.5 17.5	13 14 21 22		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 10mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ³⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, $F = 1kHz$, $V_{ripple} = 200mV_{pp}$	53	58		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	45	56		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	2.14 1.97	0.25 2.25 0.35 2.15	0.32 0.45	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ³⁾ ($R_L = 32\Omega$, THD + N $< 0.4\%$, $20Hz \leq F \leq 20kHz$)	80	95		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$ Channel Separation, $R_L = 16\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. All electrical values are guaranteed with correlation measurements at 2V and 5V.

2. Only for external gain version.

3. Guaranteed by design and evaluation.

ELECTRICAL CHARACTERISTICS $V_{CC} = +2V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		1.7	2.5	mA
$I_{STANDBY}$	Standby Current No input signal, $V_{STANDBY}=GND$ for TS486, $R_L=32\Omega$ No input signal, $V_{STANDBY}=V_{CC}$ for TS487, $R_L=32\Omega$		10	1000	nA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1		mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$) ¹⁾		90	200	nA
P_O	Output Power THD+N = 0.1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 32\Omega$ THD+N = 0.3% Max, $F = 1kHz$, $R_L = 16\Omega$ THD+N = 1% Max, $F = 1kHz$, $R_L = 16\Omega$	7 9.5	8 9 12 13		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) $R_L = 32\Omega$, $P_{out} = 6.5mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 8mW$, $20Hz \leq F \leq 20kHz$		0.3 0.3		%
PSRR	Power Supply Rejection Ratio, inputs grounded ²⁾ ($A_v=-1$), $R_L \geq 16\Omega$, $C_B=1\mu F$, $F = 1kHz$, $V_{ripple} = 200mV_{pp}$	52	57		dB
I_O	Max Output Current THD + N $\leq 1\%$, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	33	41		mA
V_O	Output Swing $V_{OL} : R_L = 32\Omega$ $V_{OH} : R_L = 32\Omega$ $V_{OL} : R_L = 16\Omega$ $V_{OH} : R_L = 16\Omega$	1.67 1.53	0.24 1.73 0.33 1.63	0.29 0.41	V
SNR	Signal-to-Noise Ratio (A weighted, $A_v=-1$) ²⁾ ($R_L = 32\Omega$, THD + N $< 0.4\%$, $20Hz \leq F \leq 20kHz$)	80	93		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$ Channel Separation, $R_L = 16\Omega$, $A_v=-1$ $F = 1kHz$ $F = 20Hz$ to $20kHz$		80 76 77 69		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)		1.1		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)		0.4		V/ μs

1. Only for external gain version.

2. Guaranteed by design and evaluation.

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Crosstalk vs Frequency	92 to 95	24
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Fig. 1: Open Loop Gain and Phase vs Frequency

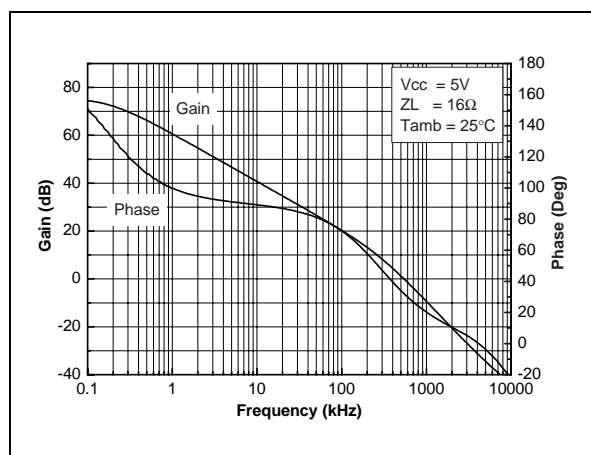


Fig. 2: Open Loop Gain and Phase vs Frequency

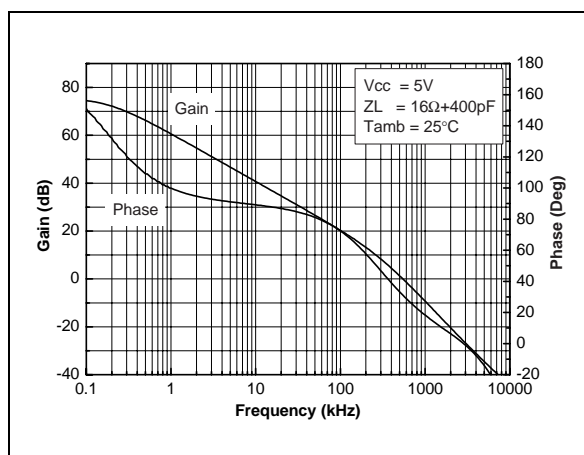


Fig. 3: Open Loop Gain and Phase vs Frequency

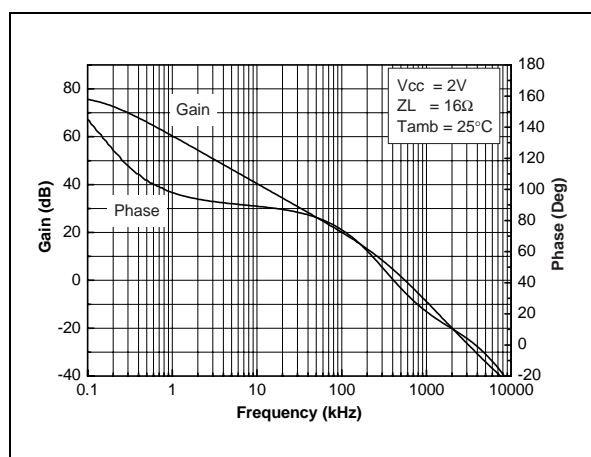


Fig. 4: Open Loop Gain and Phase vs Frequency

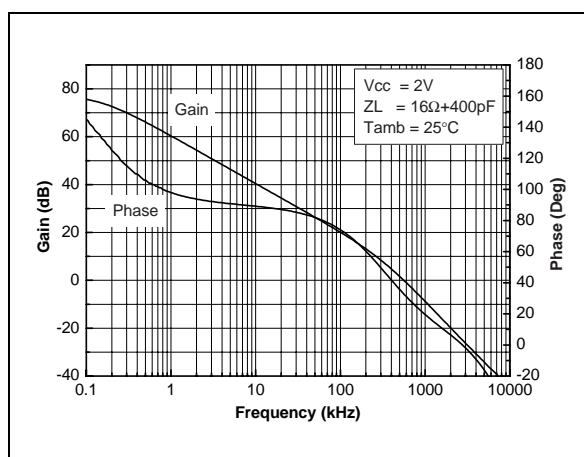


Fig. 5: Open Loop Gain and Phase vs Frequency

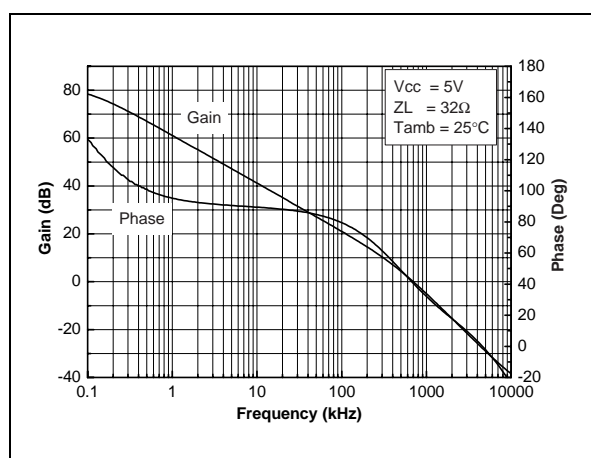


Fig. 6: Open Loop Gain and Phase vs Frequency

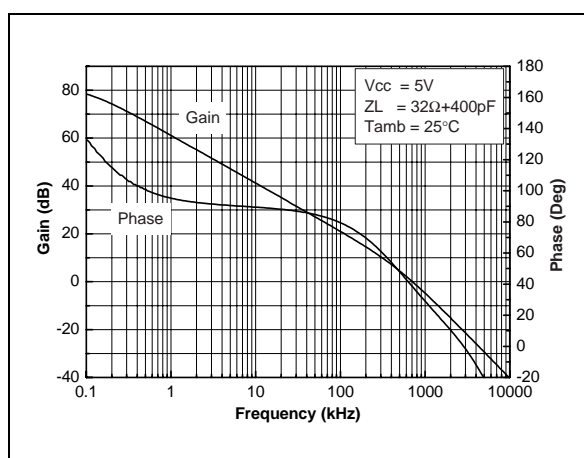


Fig. 7: Open Loop Gain and Phase vs Frequency

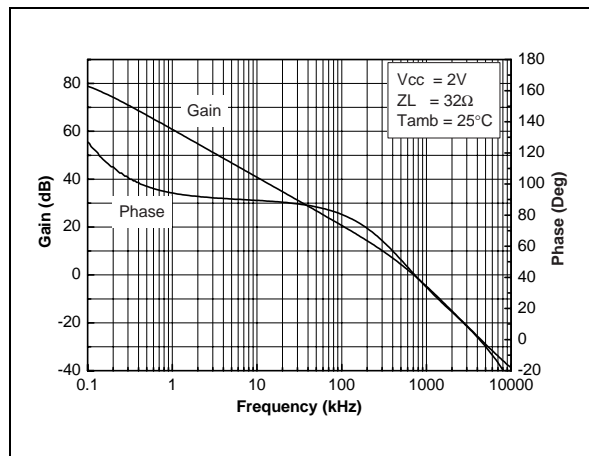


Fig. 8: Open Loop Gain and Phase vs Frequency

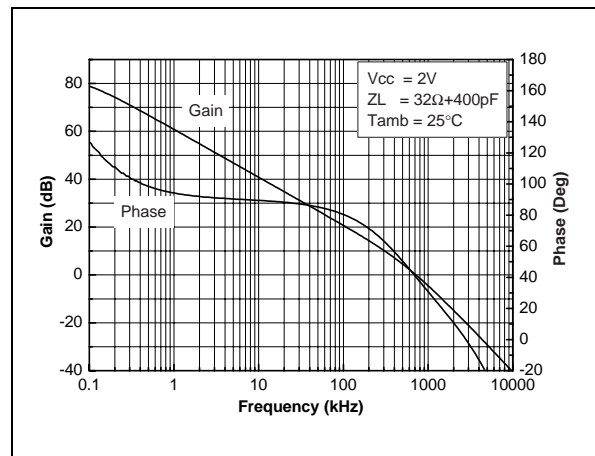


Fig. 9: Open Loop Gain and Phase vs Frequency

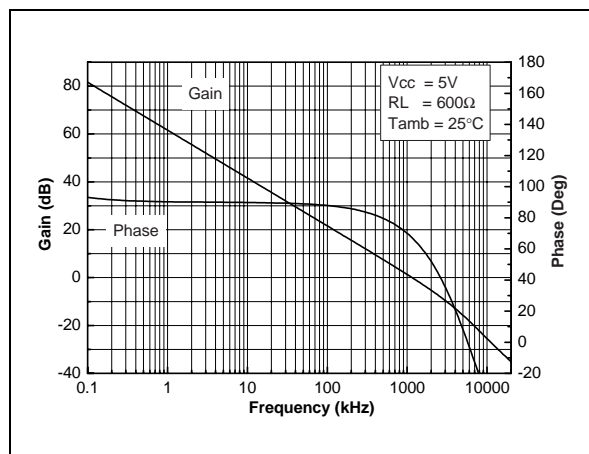


Fig. 10: Open Loop Gain and Phase vs Frequency

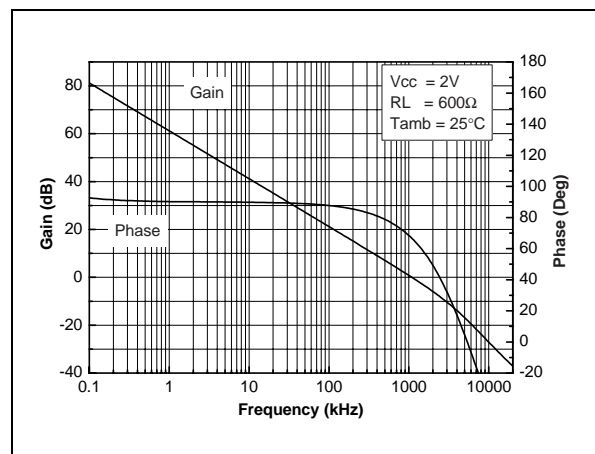


Fig. 11: Current Consumption vs Power Supply Voltage

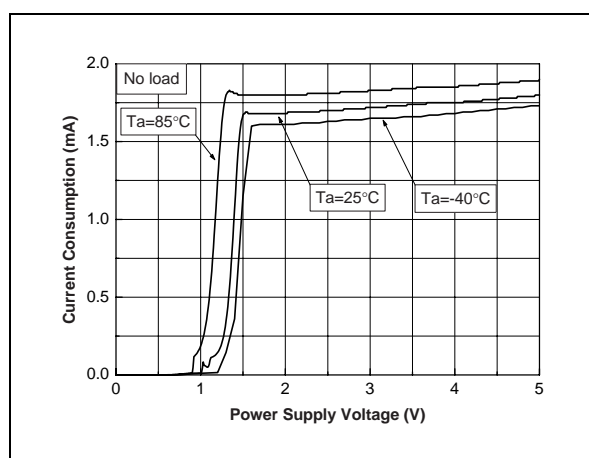


Fig. 12: Current Consumption vs Standby Voltage

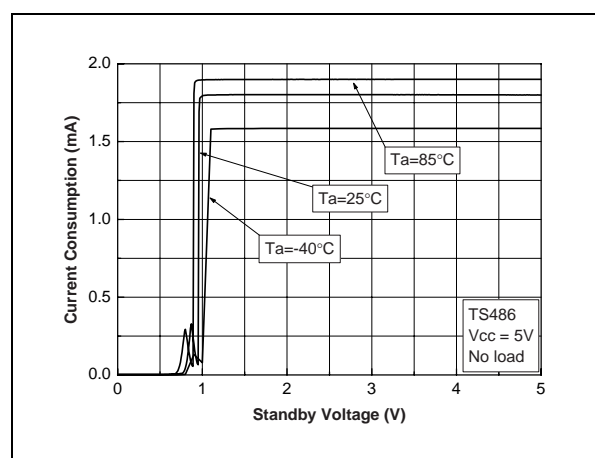


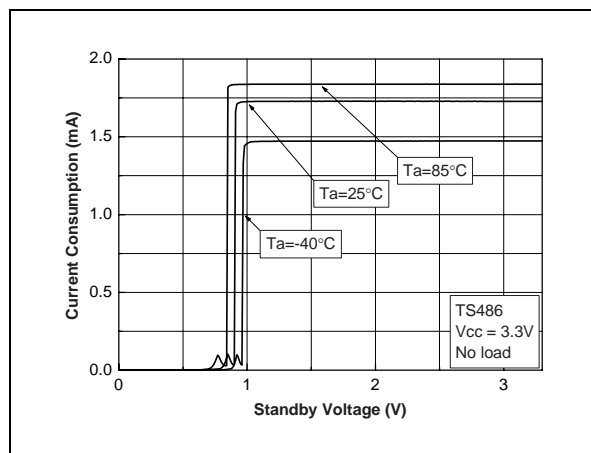
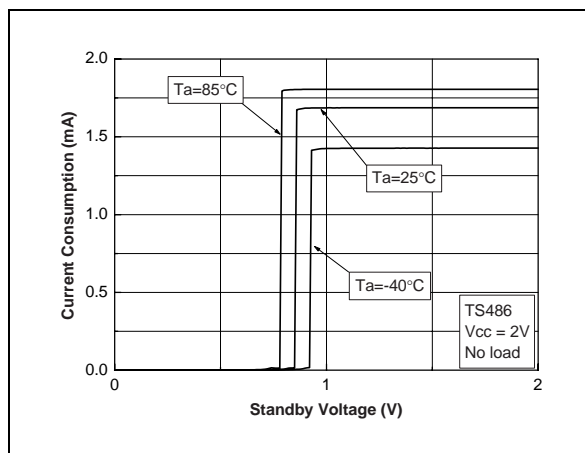
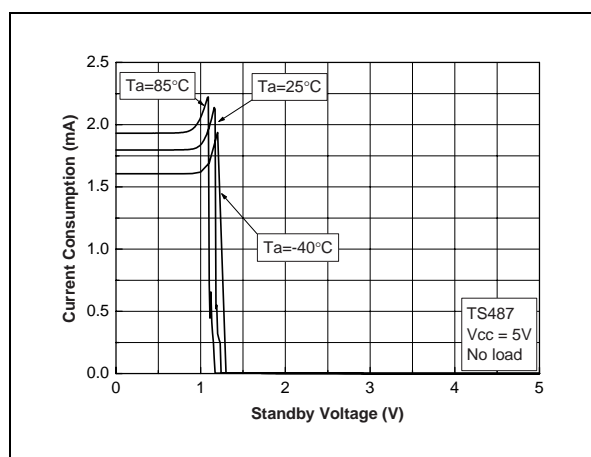
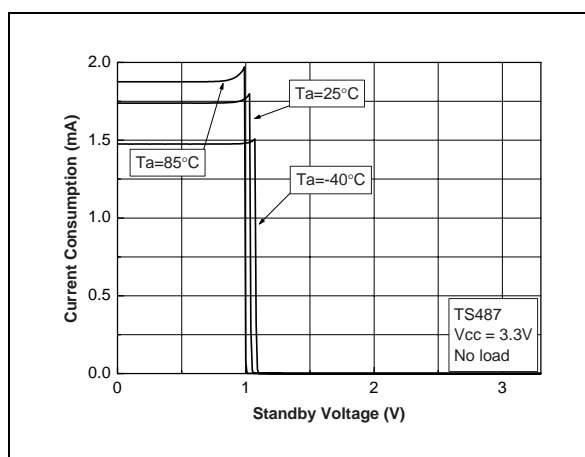
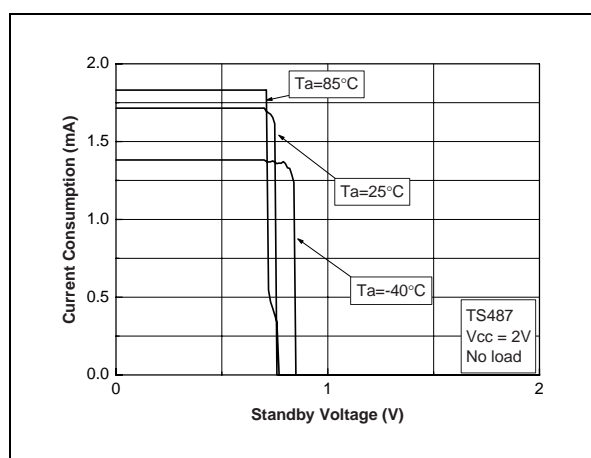
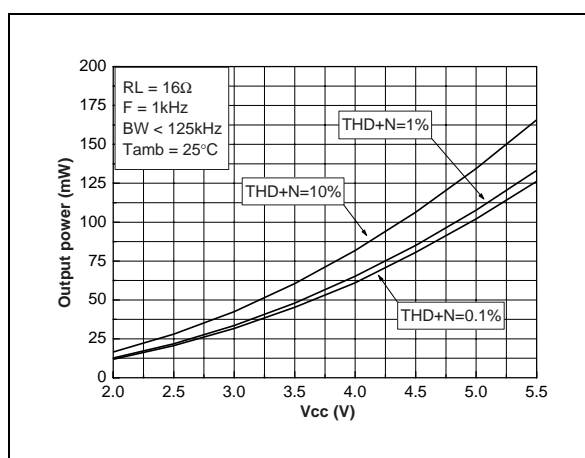
Fig. 13: Current Consumption vs Standby Voltage**Fig. 14: Current Consumption vs Standby Voltage****Fig. 15: Current Consumption vs Standby Voltage****Fig. 16: Current Consumption vs Standby Voltage****Fig. 17: Current Consumption vs Standby Voltage****Fig. 18: Output Power vs Power Supply Voltage**

Fig. 19: Output Power vs Power Supply Voltage

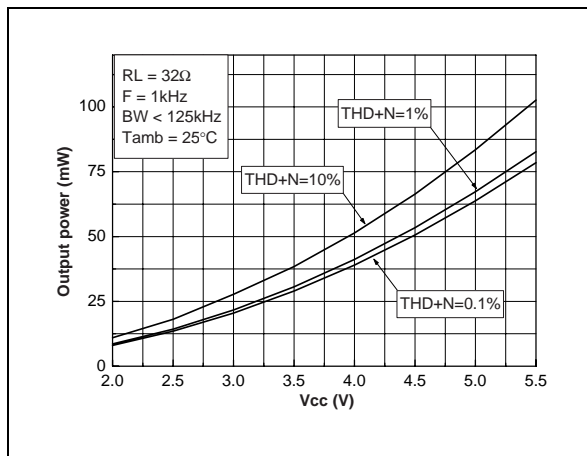


Fig. 20: Output Power vs Load Resistor

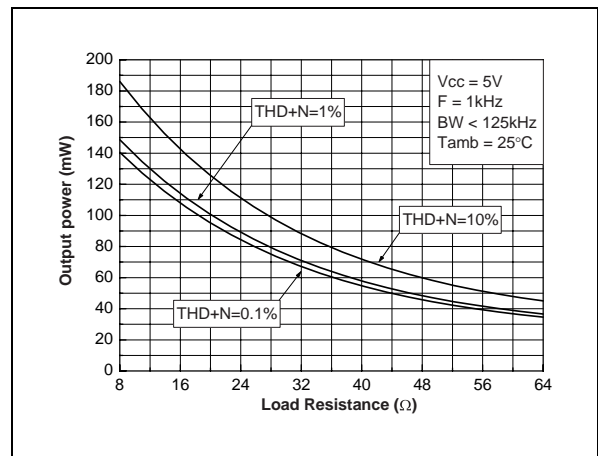


Fig. 21: Output Power vs Load Resistor

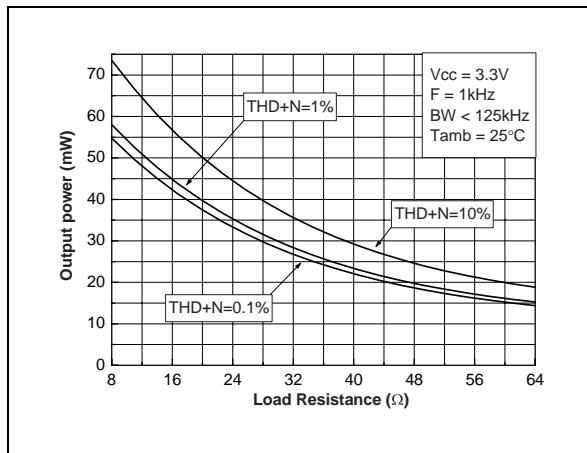


Fig. 22: Output Power vs Load Resistor

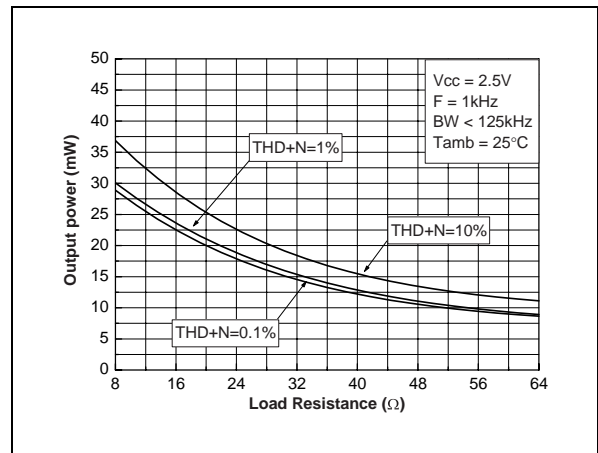


Fig. 23: Output Power vs Load Resistor

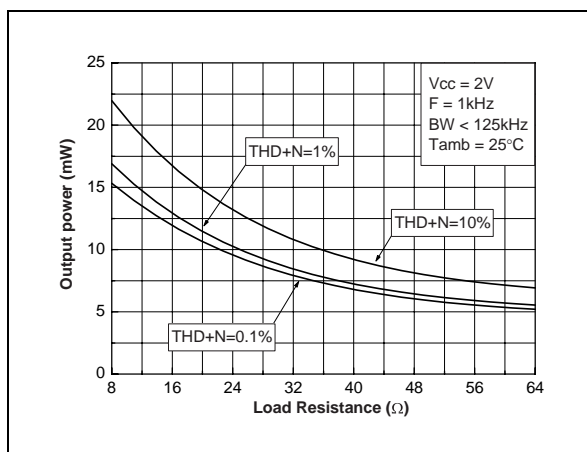


Fig. 24: Power Dissipation vs Output Power

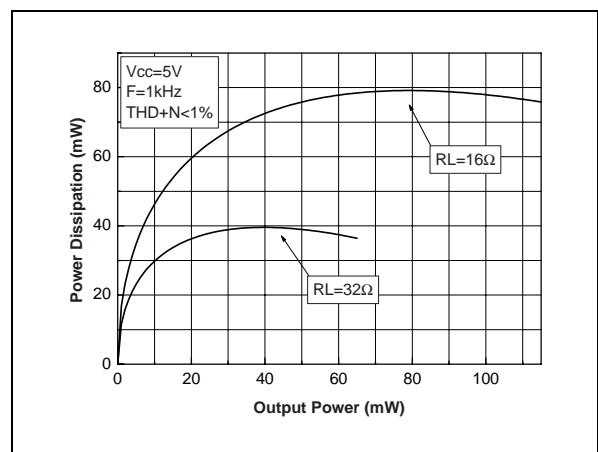


Fig. 25: Power Dissipation vs Output Power

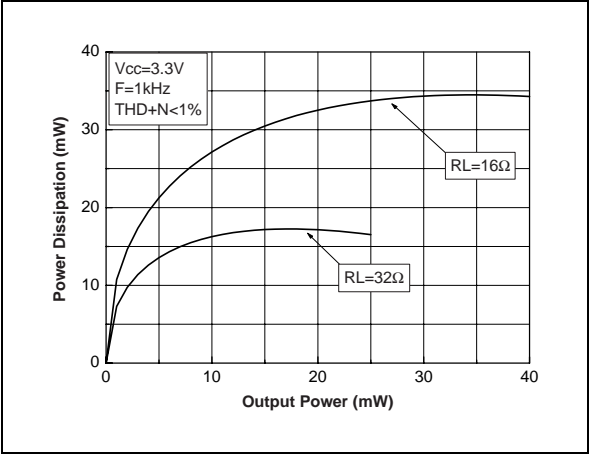


Fig. 26: Power Dissipation vs Output Power

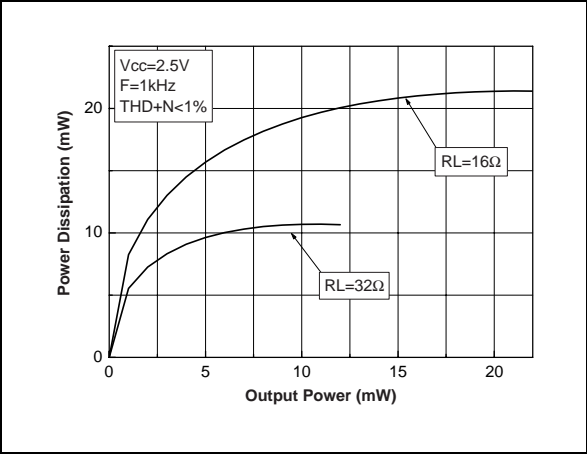


Fig. 27: Power Dissipation vs Output Power

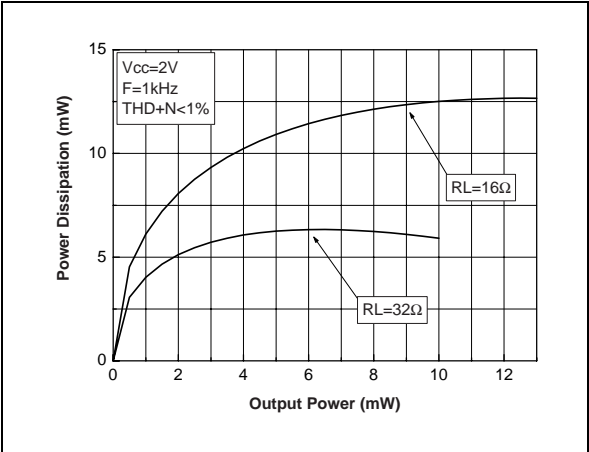


Fig. 28: Power Derating vs Ambient Temperature

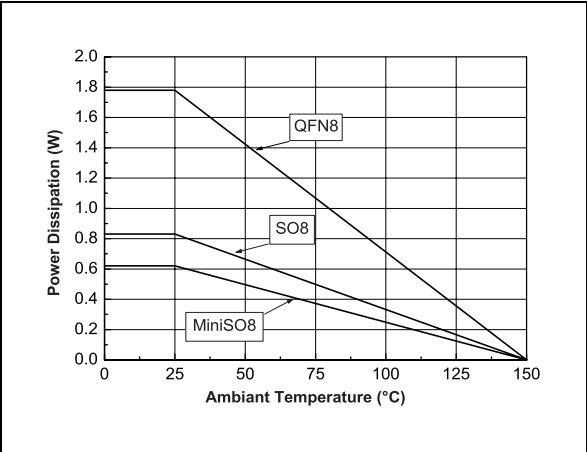


Fig. 29: Output Voltage Swing vs Power Supply Voltage

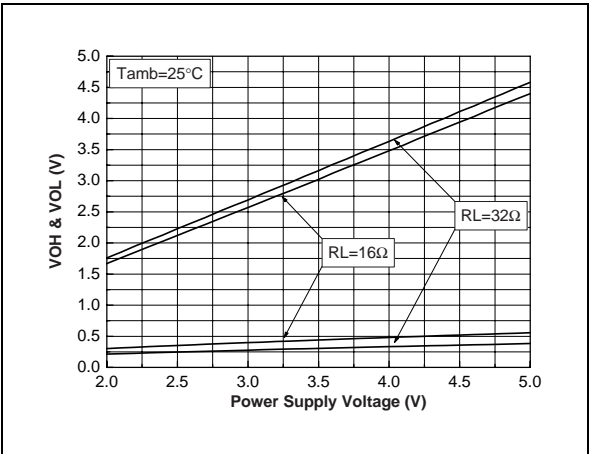


Fig. 30: Low Frequency Cut Off vs Input Capacitor for fixed gain versions.

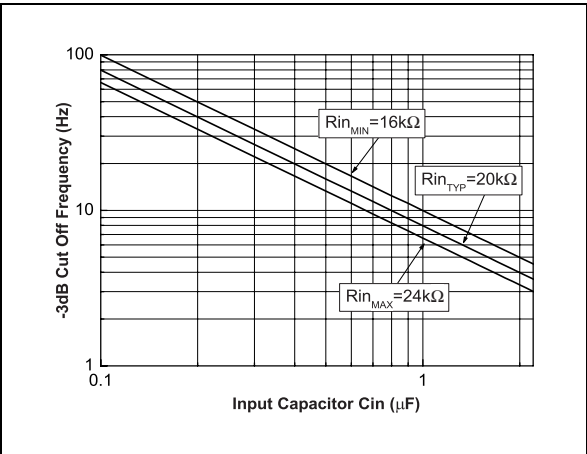


Fig. 31: THD + N vs Output Power

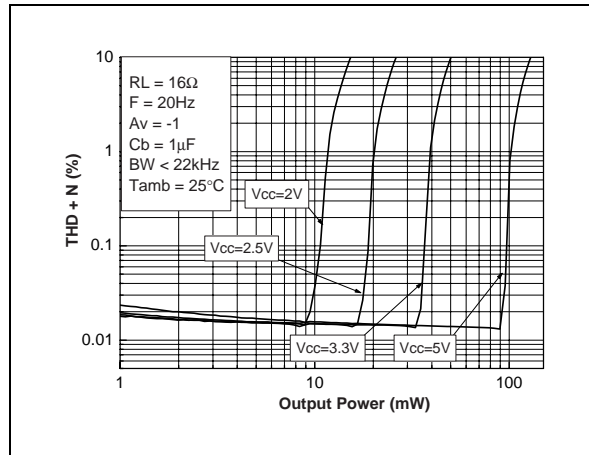


Fig. 32: THD + N vs Output Power

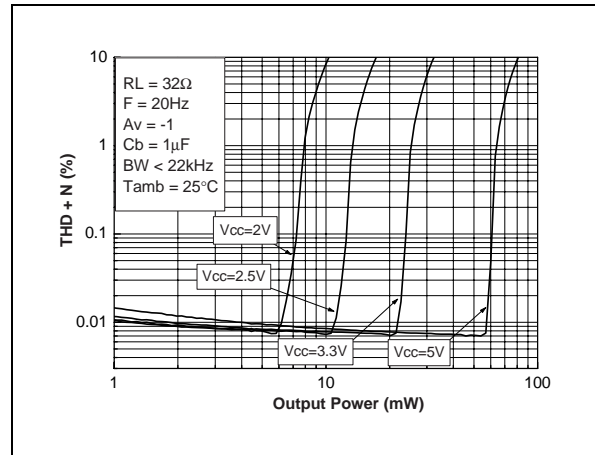


Fig. 33: THD + N vs Output Power

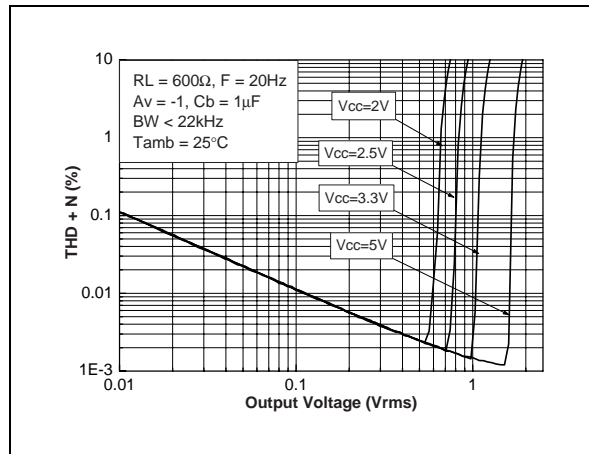


Fig. 34: THD + N vs Output Power

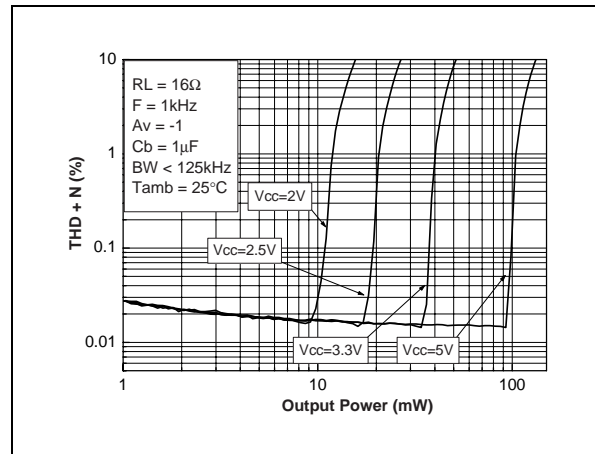


Fig. 35: THD + N vs Output Power

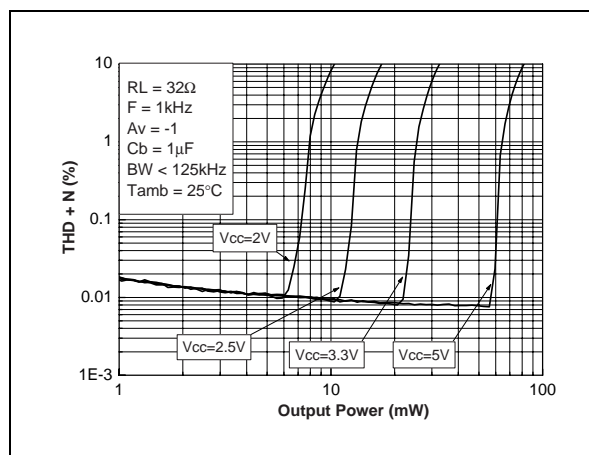


Fig. 36: THD + N vs Output Power

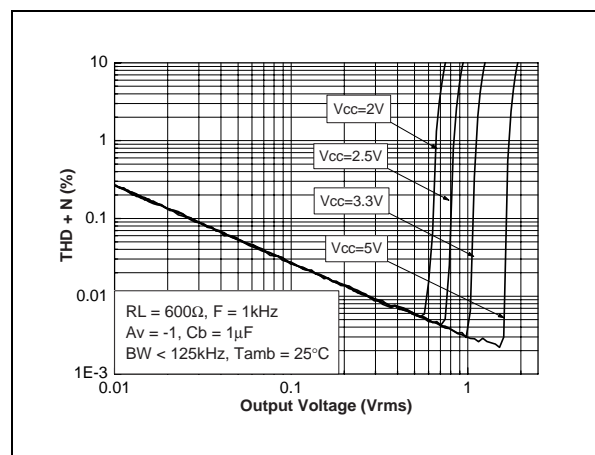


Fig. 37: THD + N vs Output Power

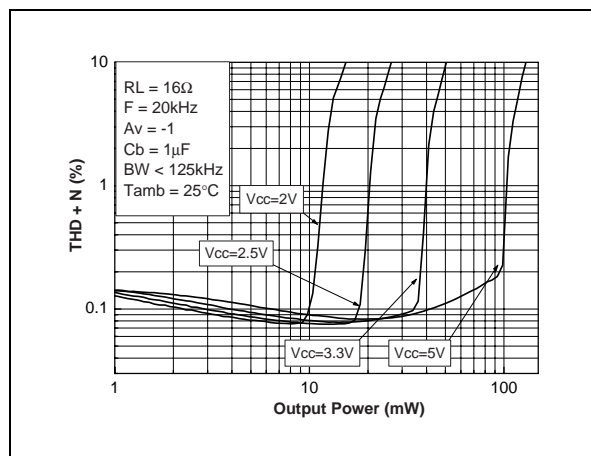


Fig. 38: THD + N vs Output Power

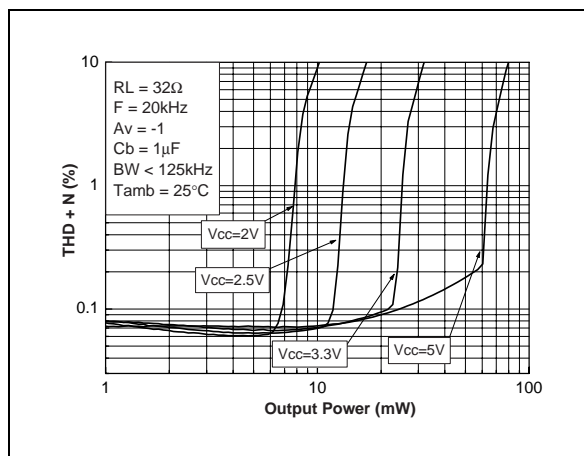


Fig. 39: THD + N vs Output Power

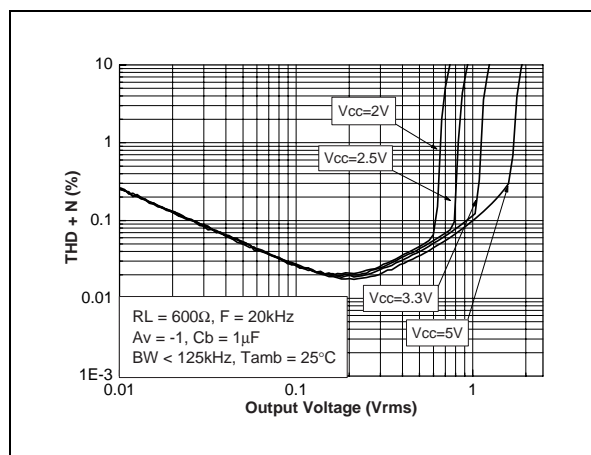


Fig. 40: THD + N vs Frequency

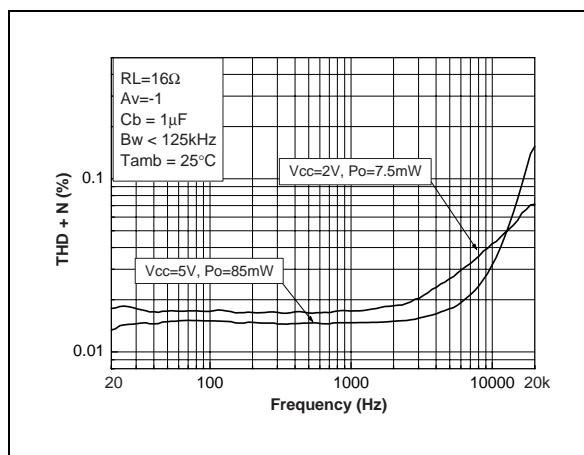


Fig. 41: THD + N vs Frequency

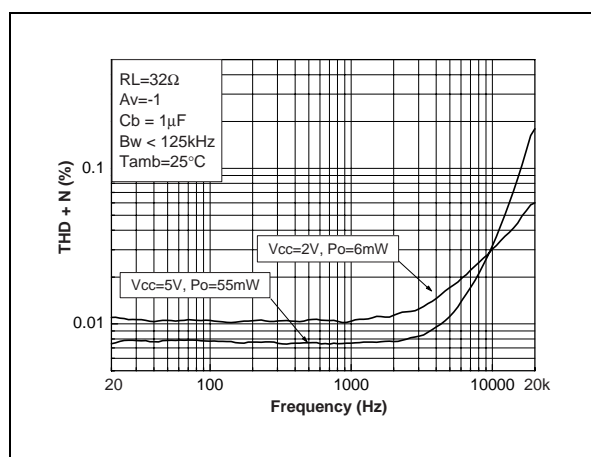


Fig. 42: THD + N vs Frequency

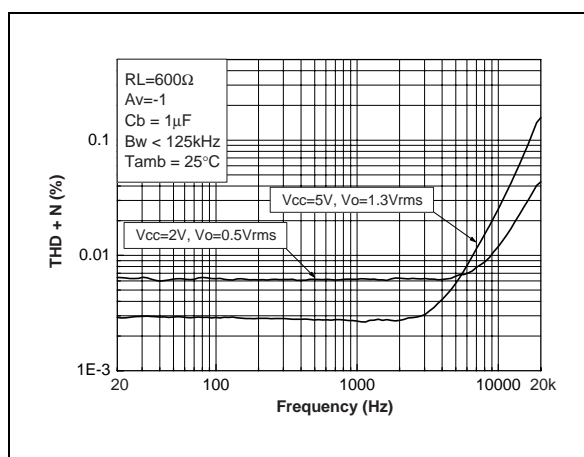


Fig. 43: Crosstalk vs Frequency

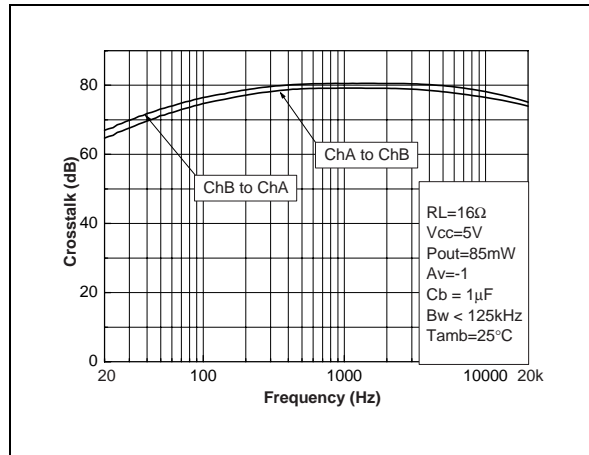


Fig. 44: Crosstalk vs Frequency

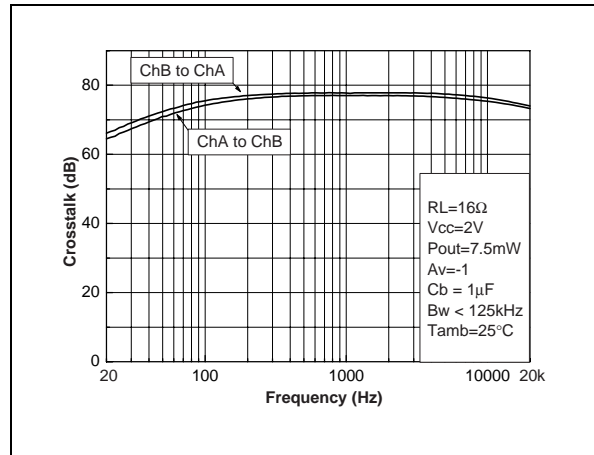


Fig. 45: Crosstalk vs Frequency

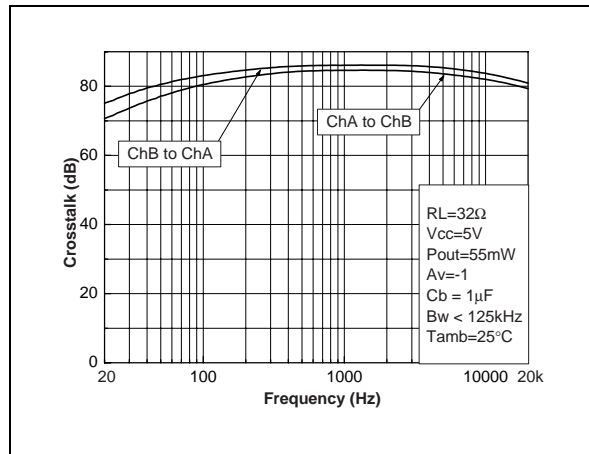


Fig. 46: Crosstalk vs Frequency

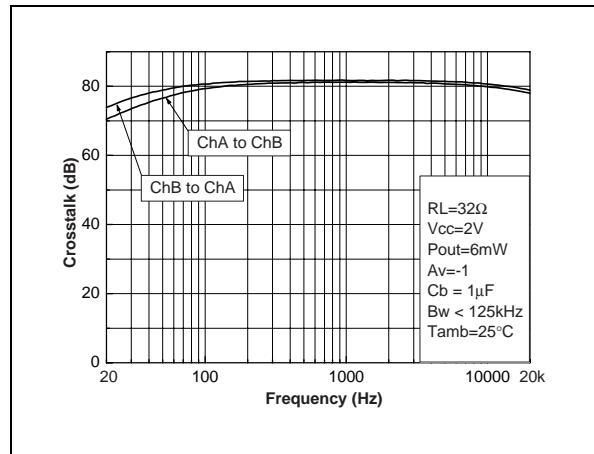


Fig. 47: Crosstalk vs Frequency

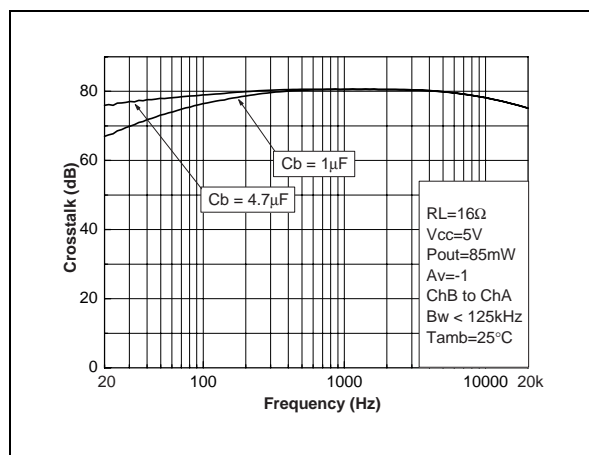


Fig. 48: Crosstalk vs Frequency

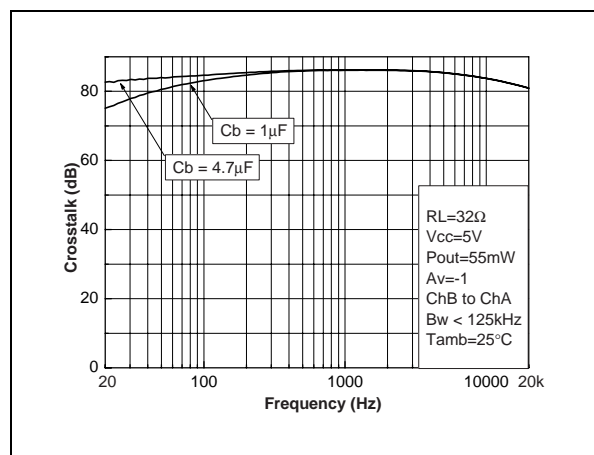


Fig. 49: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

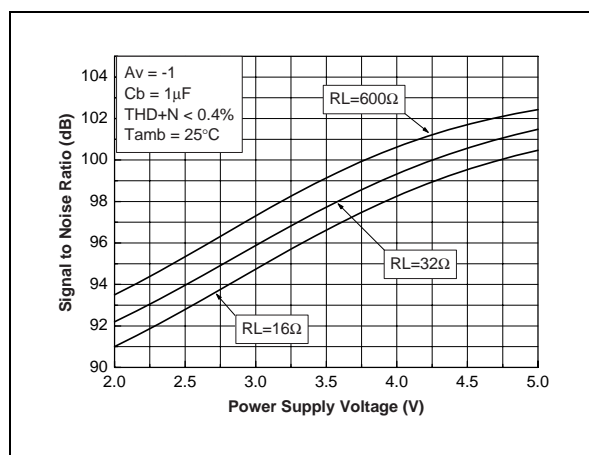


Fig. 50: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

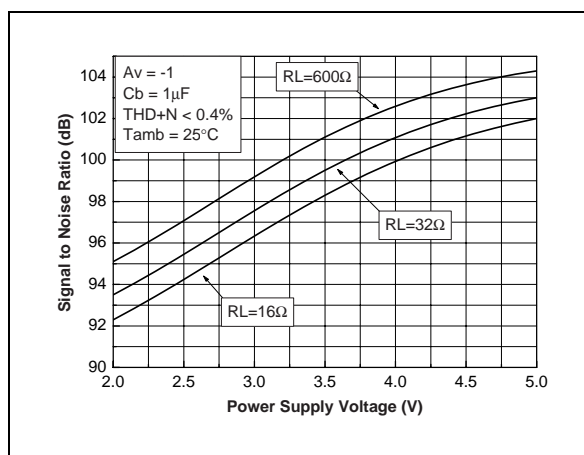


Fig. 51: PSRR vs Power Supply Voltage

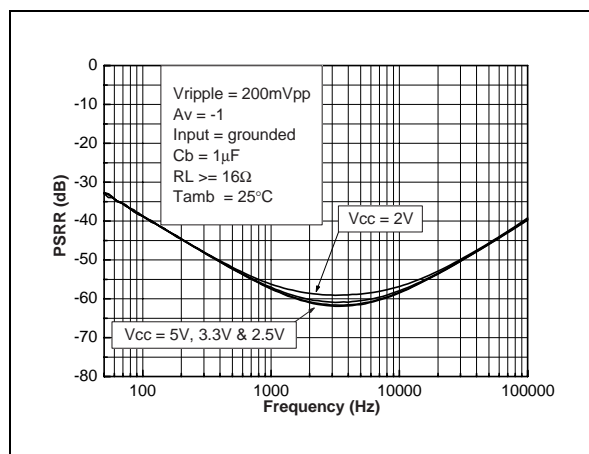


Fig. 52: PSRR vs Bypass Capacitor

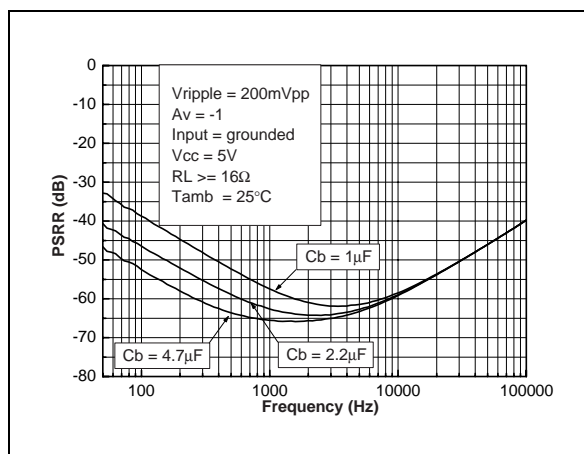


Fig. 53: PSRR vs Input Capacitor

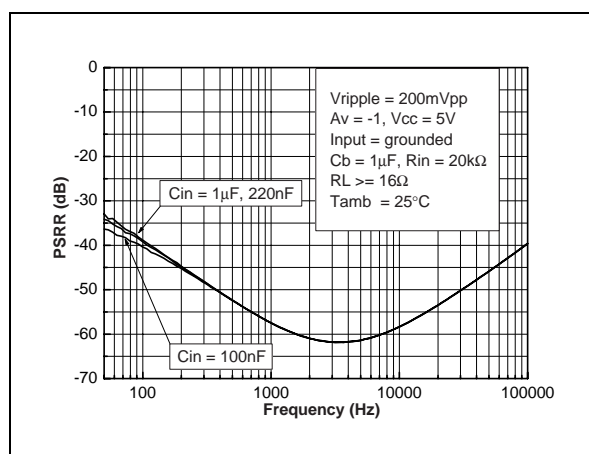


Fig. 54: PSRR vs Output Capacitor

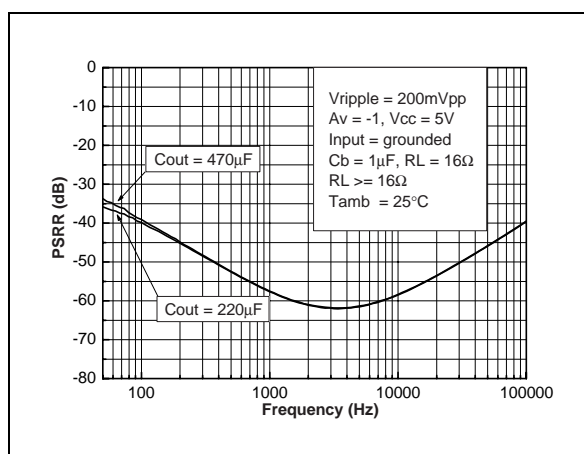


Fig. 55: PSRR vs Output Capacitor

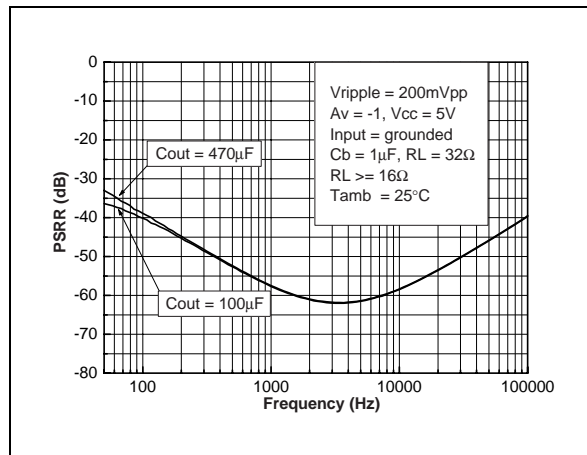


Fig. 56: PSRR vs Power Supply Voltage

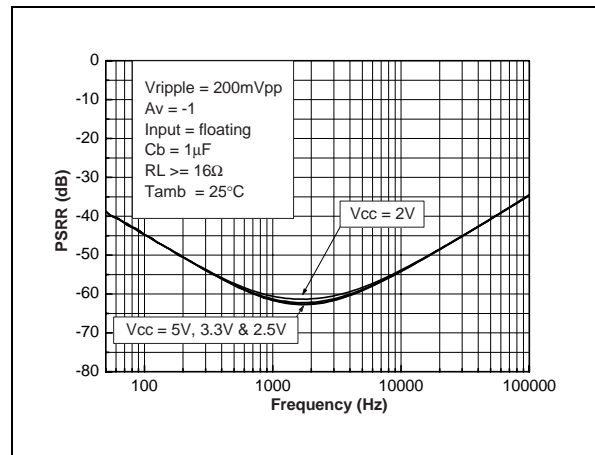


Fig. 57: THD + N vs Output Power

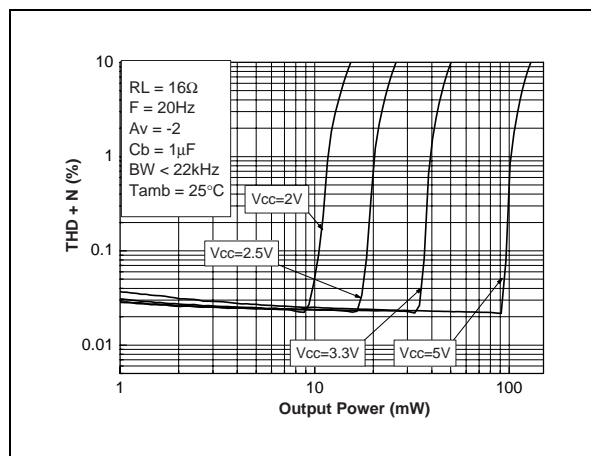


Fig. 58: THD + N vs Output Power

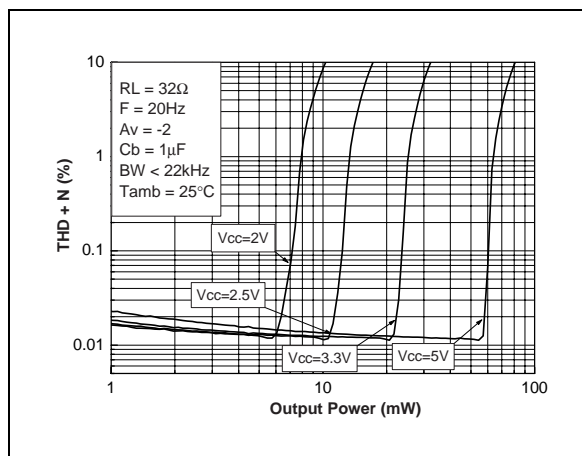


Fig. 59: THD + N vs Output Power

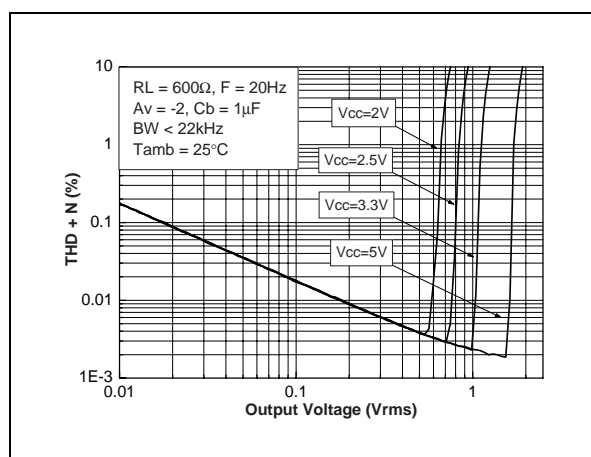


Fig. 60: THD + N vs Output Power

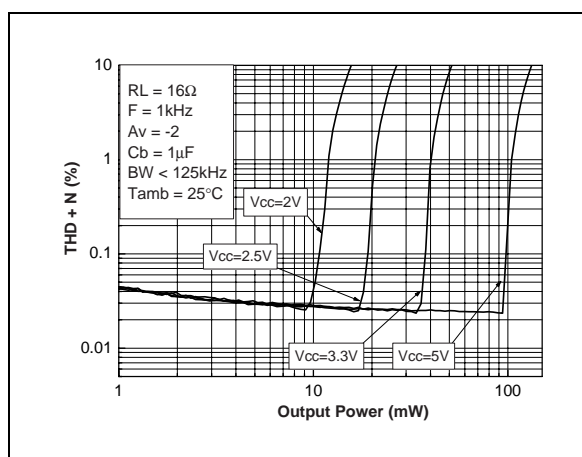


Fig. 61: THD + N vs Output Power

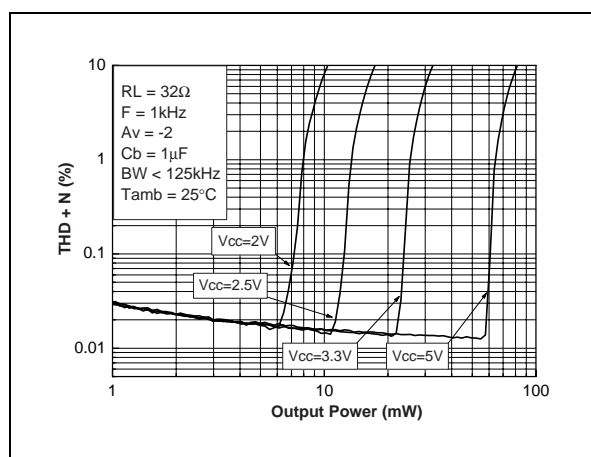


Fig. 62: THD + N vs Output Power

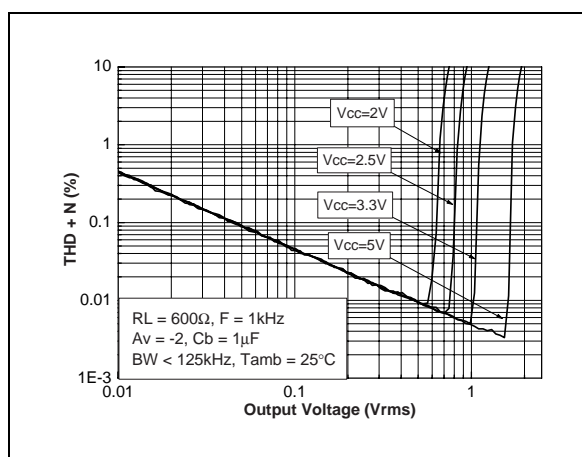


Fig. 63: THD + N vs Output Power

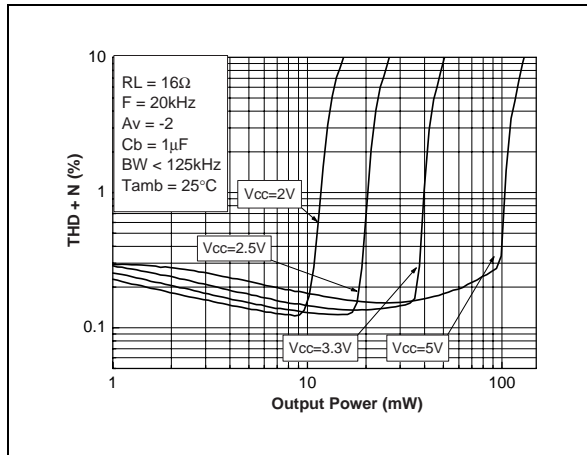


Fig. 64: THD + N vs Output Power

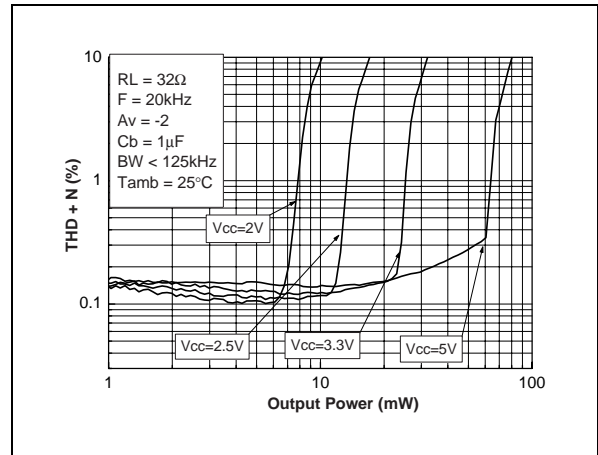


Fig. 65: THD + N vs Output Power

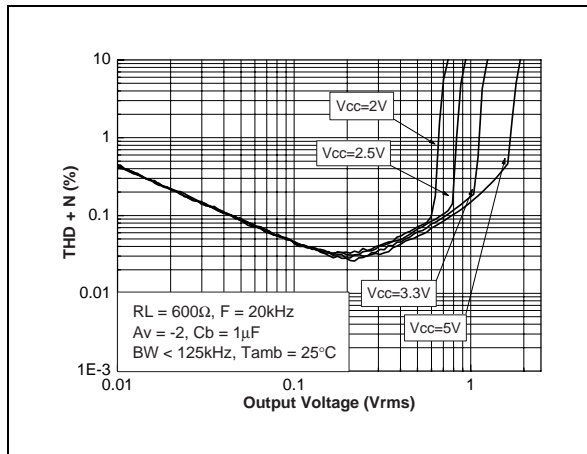


Fig. 66: THD + N vs Frequency

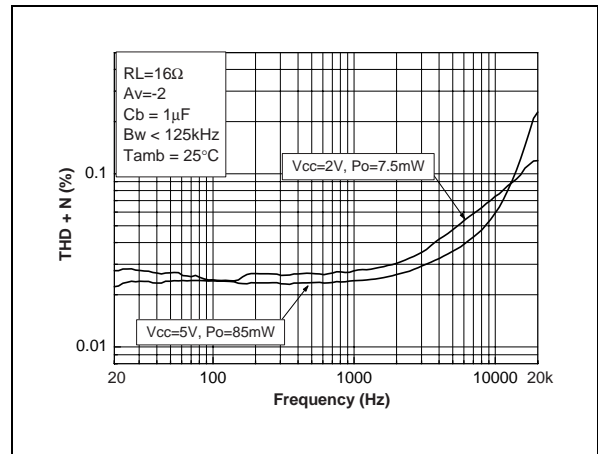


Fig. 67: THD + N vs Frequency

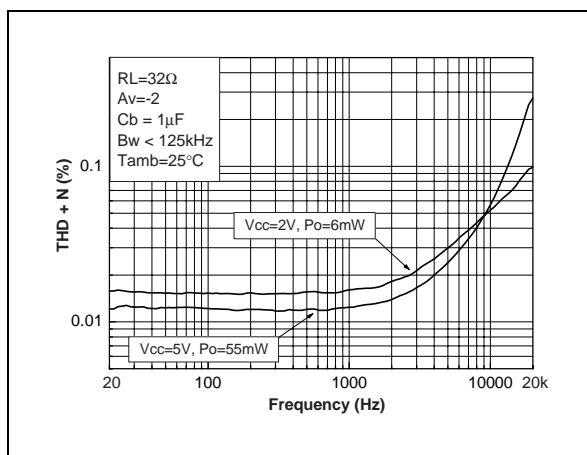


Fig. 68: THD + N vs Frequency

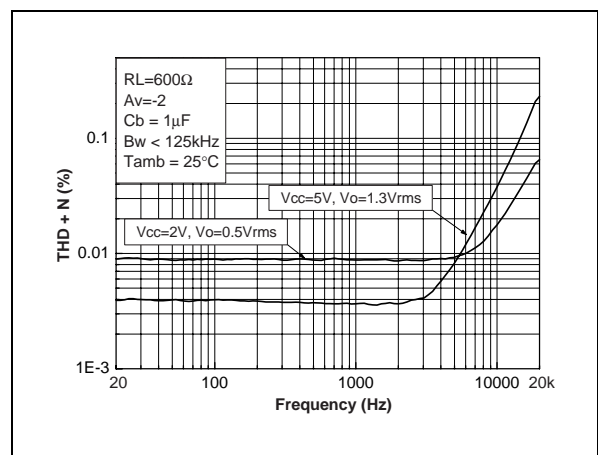


Fig. 69: Crosstalk vs Frequency

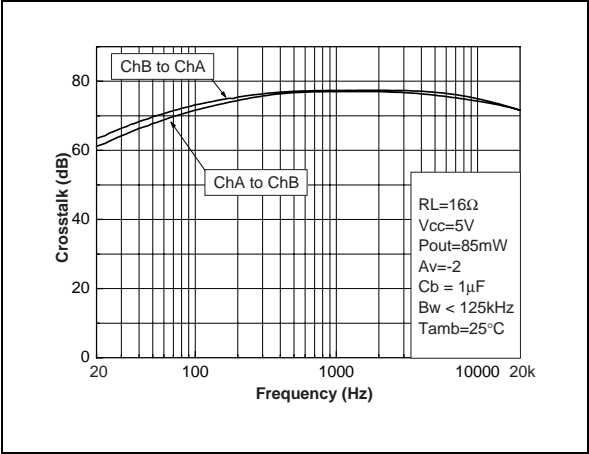


Fig. 70: Crosstalk vs Frequency

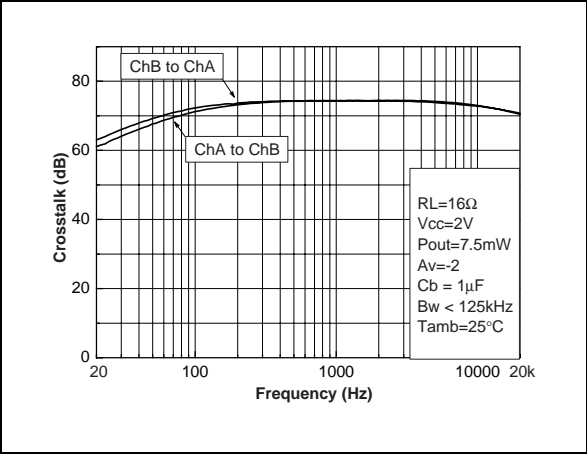


Fig. 71: Crosstalk vs Frequency

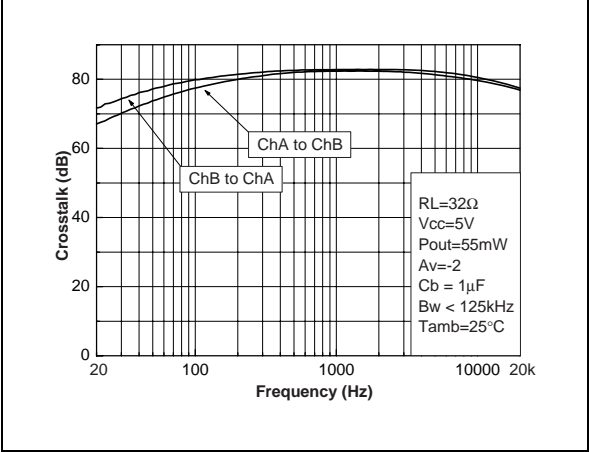


Fig. 72: Crosstalk vs Frequency

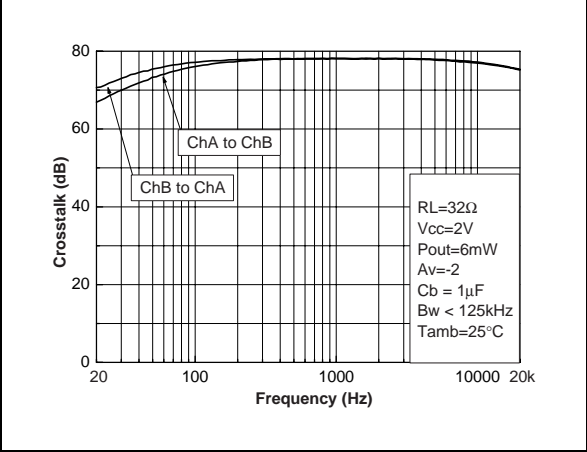


Fig. 73: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

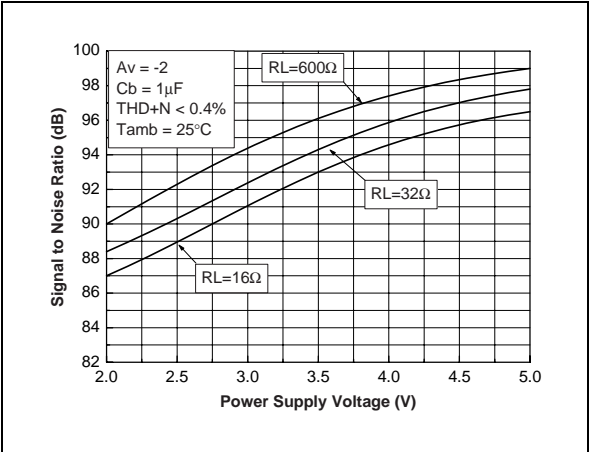


Fig. 74: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

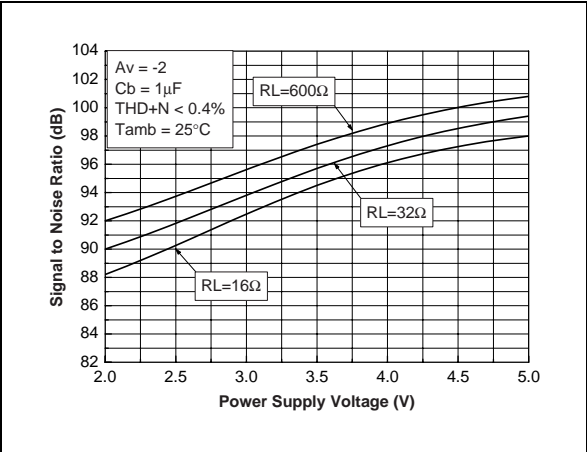


Fig. 75: PSRR vs Power Supply Voltage

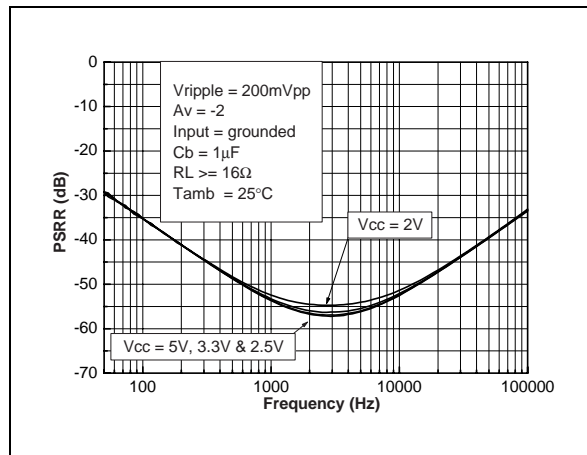


Fig. 76: PSRR vs Bypass Capacitor

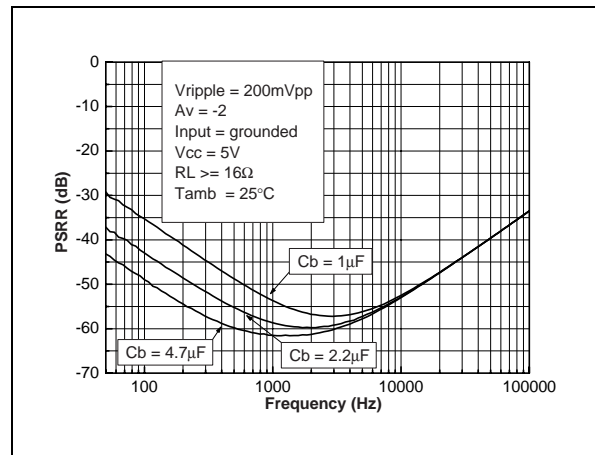


Fig. 77: PSRR vs Input Capacitor

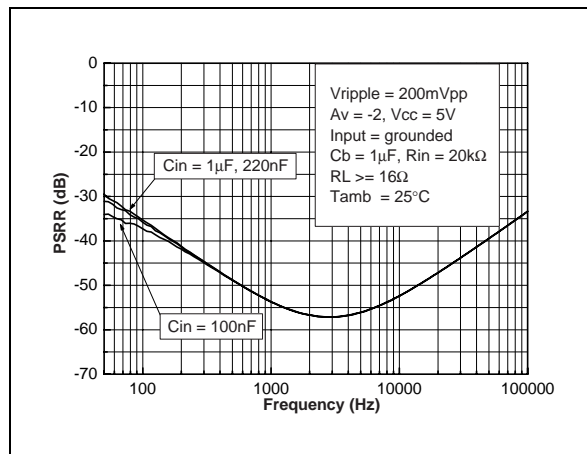


Fig. 78: PSRR vs Output Capacitor

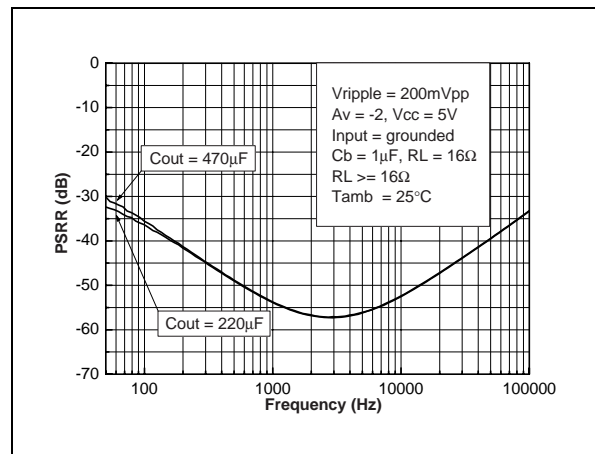


Fig. 79: PSRR vs Output Capacitor

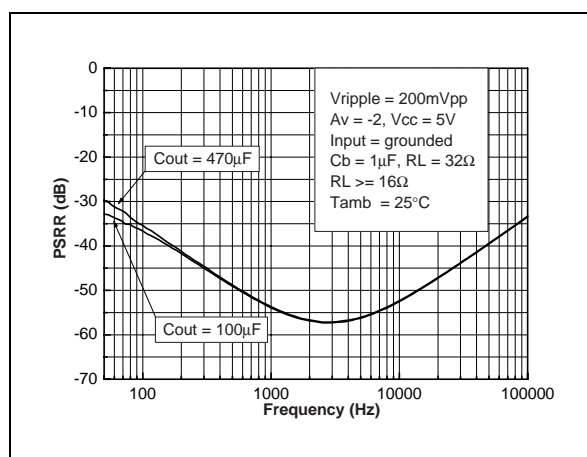


Fig. 80: THD + N vs Output Power

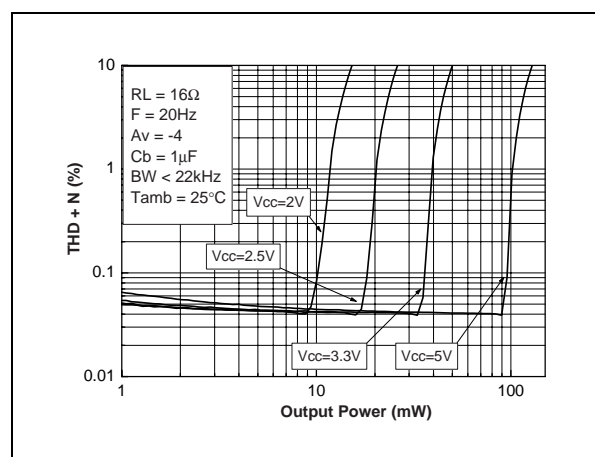


Fig. 81: THD + N vs Output Power

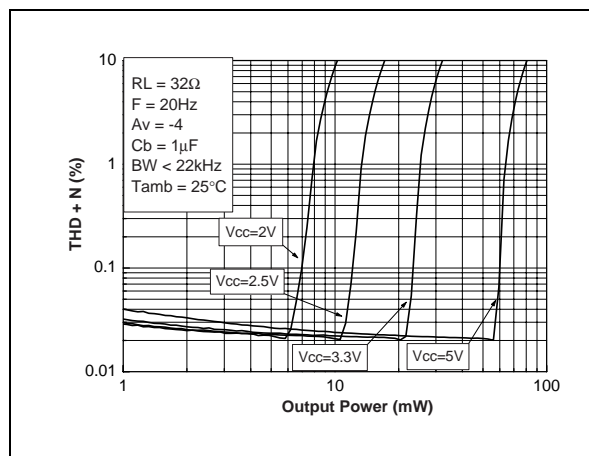


Fig. 82: THD + N vs Output Power

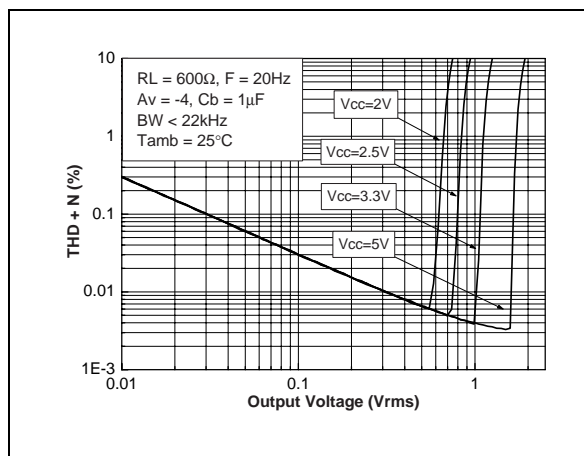


Fig. 83: THD + N vs Output Power

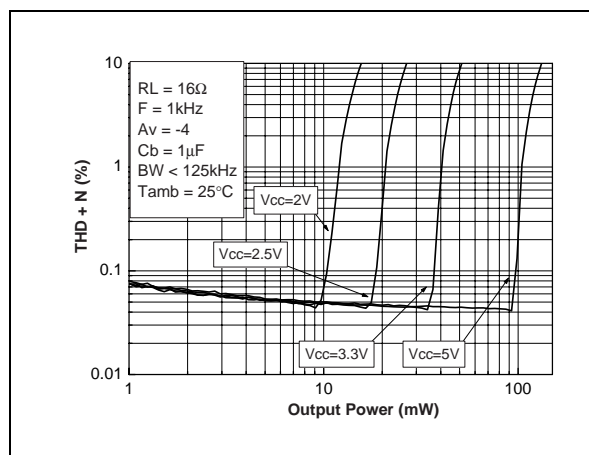


Fig. 84: THD + N vs Output Power

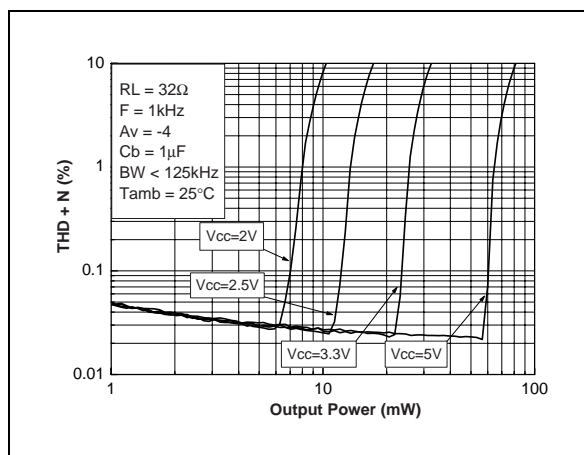


Fig. 85: THD + N vs Output Power

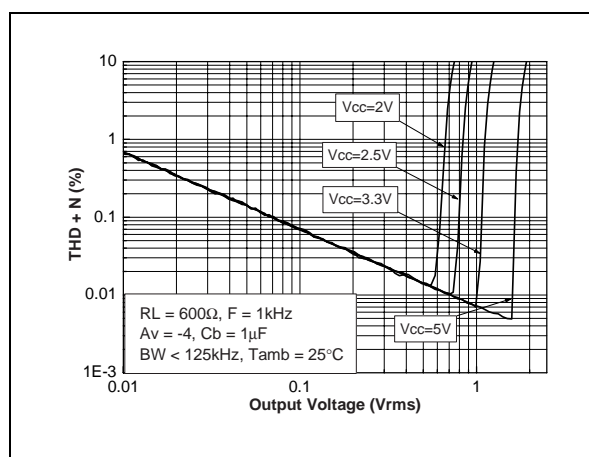


Fig. 86: THD + N vs Output Power

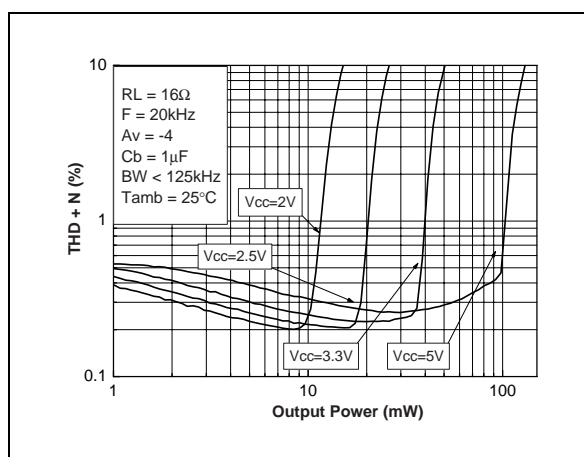


Fig. 87: THD + N vs Output Power

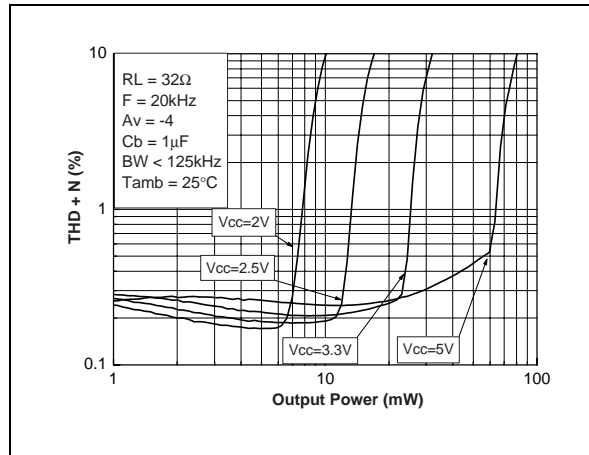


Fig. 88: THD + N vs Output Power

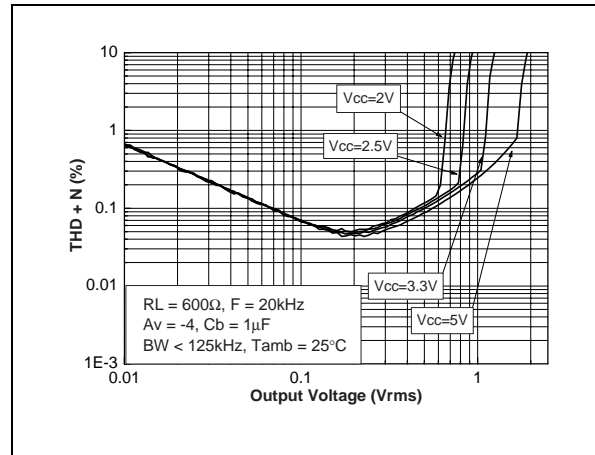


Fig. 89: THD + N vs Frequency

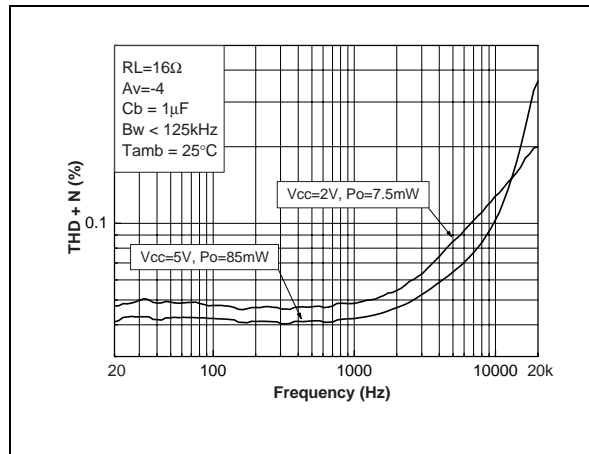


Fig. 90: THD + N vs Frequency

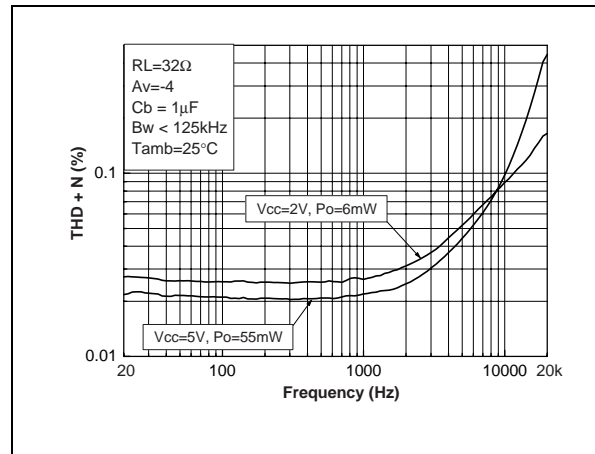


Fig. 91: THD + N vs Frequency

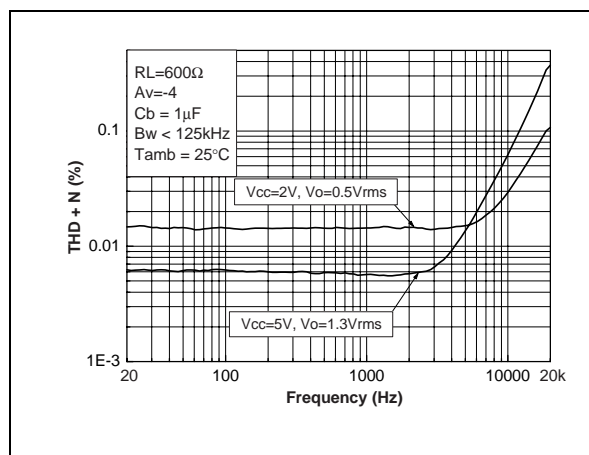


Fig. 92: Crosstalk vs Frequency

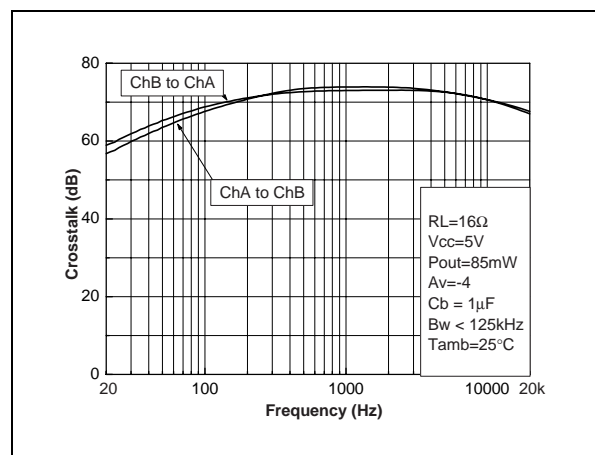


Fig. 93: Crosstalk vs Frequency

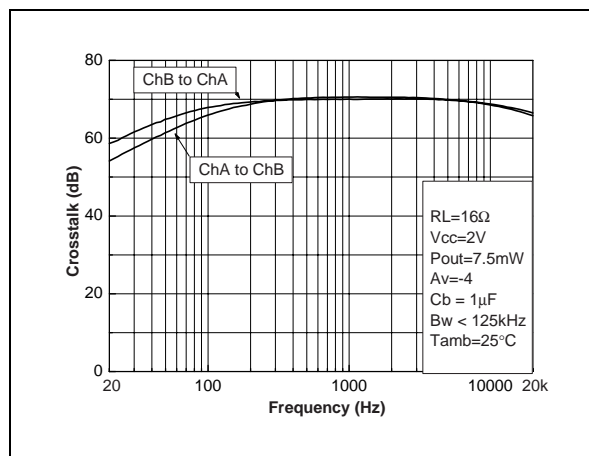


Fig. 94: Crosstalk vs Frequency

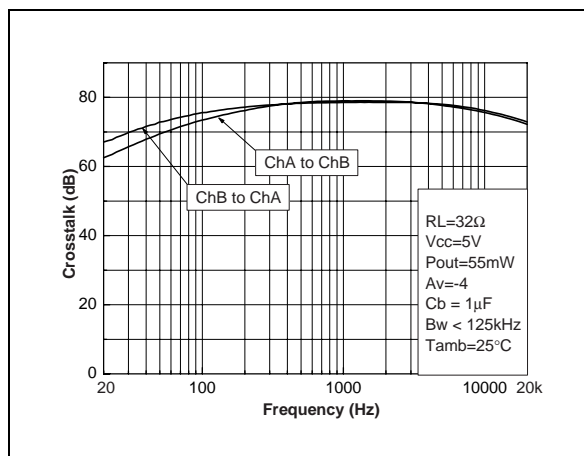


Fig. 95: Crosstalk vs Frequency

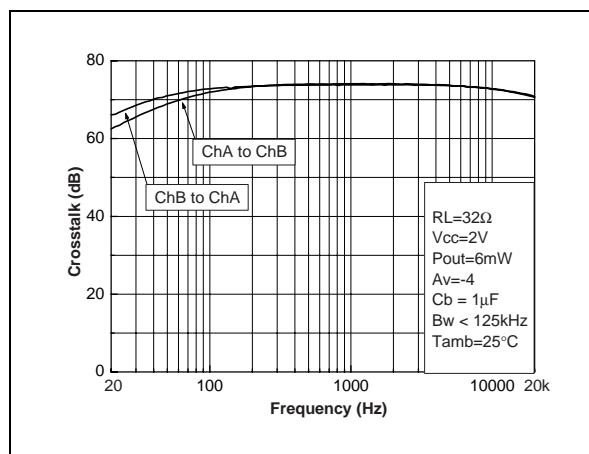


Fig. 96: Signal to Noise Ratio vs Power Supply Voltage with Unweighted Filter (20Hz to 20kHz)

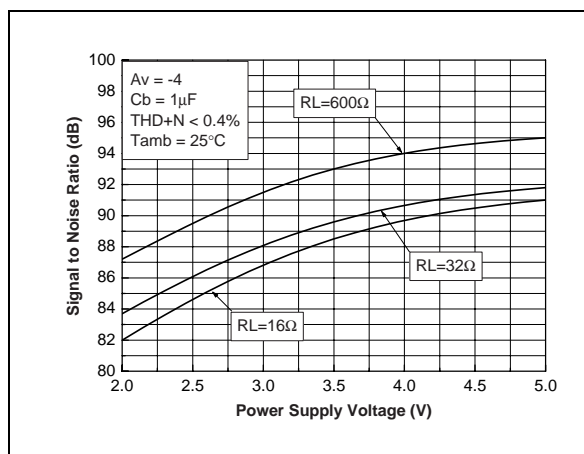


Fig. 97: Signal to Noise Ratio vs Power Supply Voltage with Weighted Filter Type A

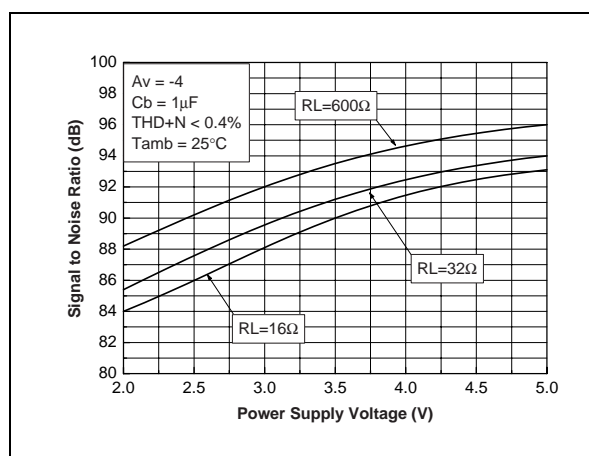


Fig. 98: PSRR vs Power Supply Voltage

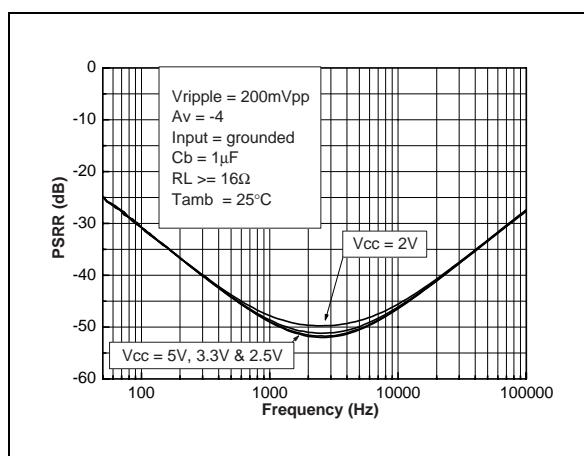


Fig. 99: PSRR vs Input Capacitor

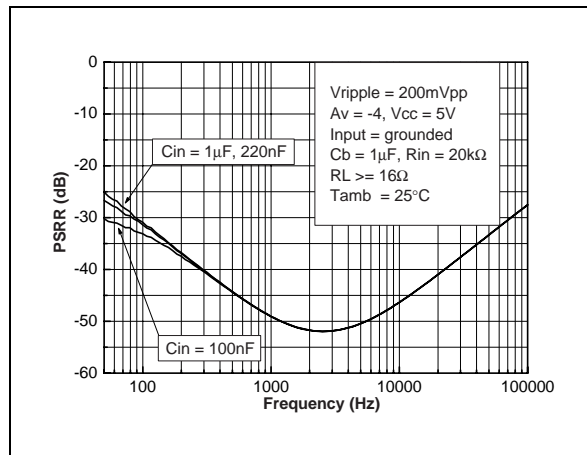


Fig. 100: PSRR vs Bypass Capacitor

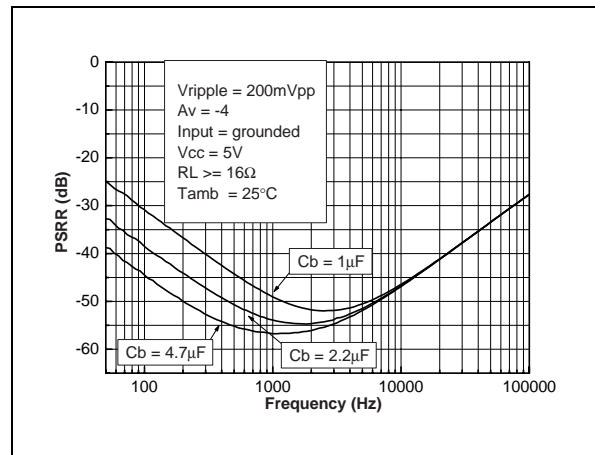


Fig. 101: PSRR vs Output Capacitor

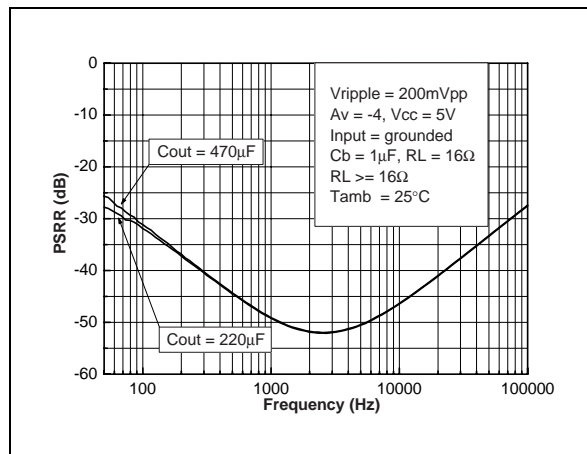
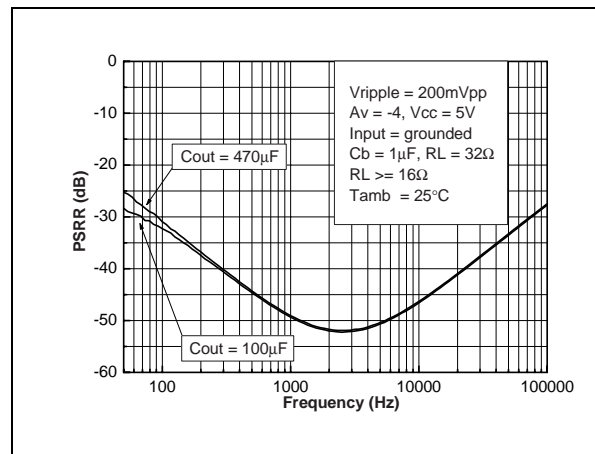
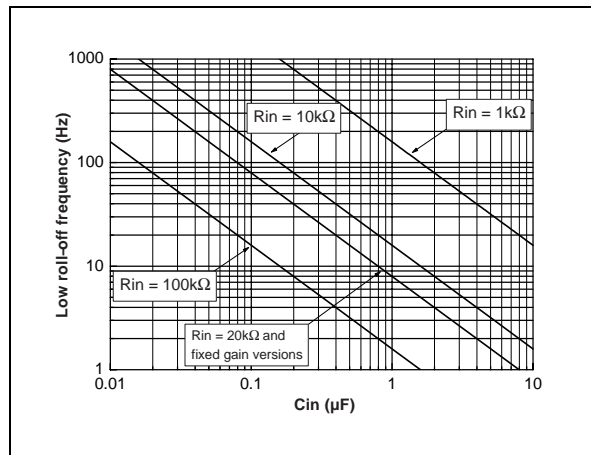


Fig. 102: PSRR vs Output Capacitor

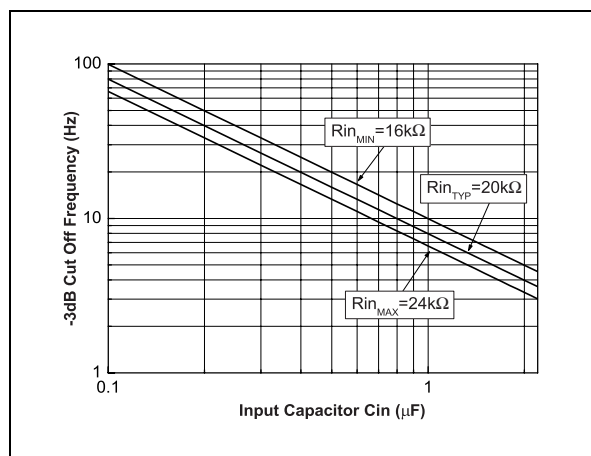


and for various values of the input resistor R_{IN} .



The input resistance of the fixed gain version is typically $20k\Omega$.

The following curve shows the limits of the roll off frequency depending on the min. and max. values of R_{in} :



LOW FREQUENCY ROLL OFF WITH OUTPUT CAPACITORS

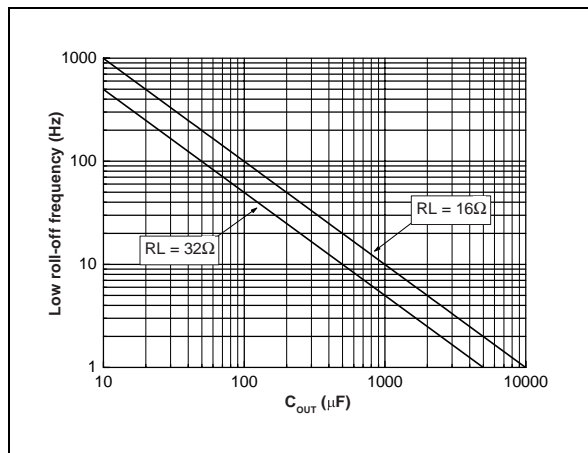
The DC voltage on the outputs of the TS486/487 is blocked by the output capacitors C_{OUT1} and C_{OUT2} . Each output capacitor C_{OUT} in series with the resistance of the load R_L is equivalent to a first order high pass filter.

Assuming that F_{min} is the lowest frequency to be amplified (with a 3dB attenuation), the minimum value of C_{OUT} is:

$$C_{OUT} > 1 / (2 * \pi * F_{min} * R_L)$$

The following curve gives directly the low roll-off

frequency versus the output capacitor C_{OUT} in μF and for the two typical 16Ω and 32Ω impedances:



DECOUPLING CAPACITOR C_B

The internal bias voltage at $V_{cc}/2$ is decoupled with the external capacitor C_B .

The TS486 and TS487 have a specified Power Supply Rejection Ratio parameter with $C_B = 1\mu F$. A higher value of C_B improves the PSRR, for example, a $4.7\mu F$ improves the PSRR by 15dB at 200Hz (please, refer to fig. 76 "PSRR vs Bypass Capacitor").

POP PRECAUTIONS

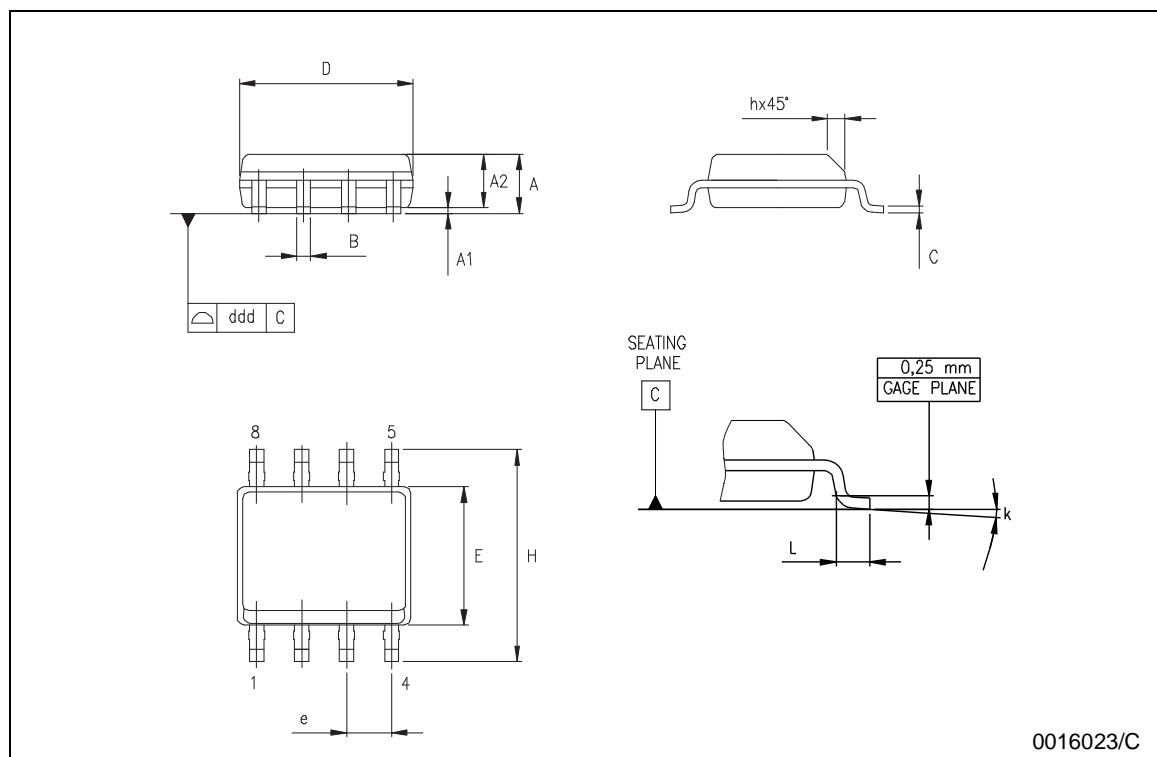
Generally headphones are connected using a connector as a jack. To prevent a pop in the headphones when plugged in the jack, a resistor should be connected in parallel with each headphone output. This allows the capacitors C_{out} to be charged even when no headphone is plugged.

A resistor of $1k\Omega$ is high enough to be a negligible load, and low enough to charge the capacitors C_{out} in less than one second.

PACKAGE MECHANICAL DATA

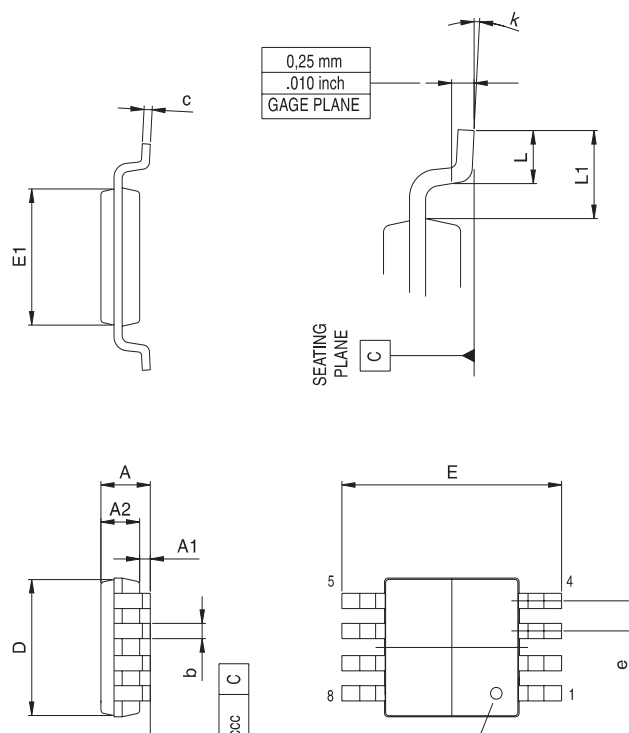
SO-8 MECHANICAL DATA

DIM.	mm.			inch		
	MIN.	TYP	MAX.	MIN.	TYP.	MAX.
A	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.04		0.010
A2	1.10		1.65	0.043		0.065
B	0.33		0.51	0.013		0.020
C	0.19		0.25	0.007		0.010
D	4.80		5.00	0.189		0.197
E	3.80		4.00	0.150		0.157
e		1.27			0.050	
H	5.80		6.20	0.228		0.244
h	0.25		0.50	0.010		0.020
L	0.40		1.27	0.016		0.050
k	8° (max.)					
ddd			0.1			0.04



PACKAGE MECHANICAL DATA

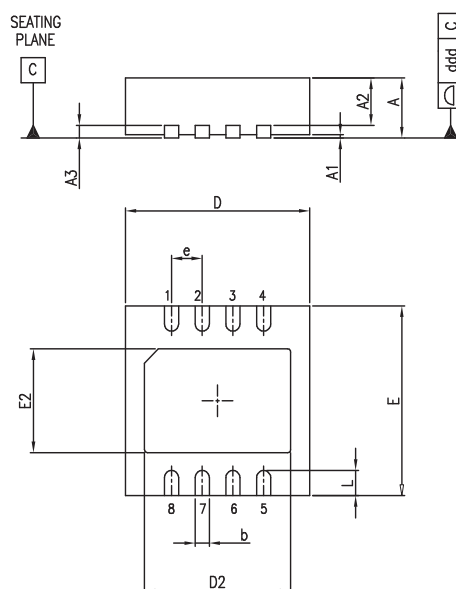
DIM.	mm.			inch		
	MIN.	TYP	MAX.	MIN.	TYP.	MAX.
A			1.1			0.043
A1	0.05	0.10	0.15	0.002	0.004	0.006
A2	0.78	0.86	0.94	0.031	0.031	0.037
b	0.25	0.33	0.40	0.010	0.13	0.013
c	0.13	0.18	0.23	0.005	0.007	0.009
D	2.90	3.00	3.10	0.114	0.118	0.122
E	4.75	4.90	5.05	0.187	0.193	0.199
E1	2.90	3.00	3.10	.0114	0.118	0.122
e		0.65			0.026	
K	0°		6°	0°		6°
L	0.40	0.55	0.70	0.016	0.022	0.028
L1			0.10			0.004



PACKAGE MECHANICAL DATA

DFN8 (3x3) MECHANICAL DATA

DIM.	mm.			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	0.80	0.90	1.00	31,5	35,4	39,4
A1		0.02	0.05		0.8	2.0
A2		0.70			27,6	
A3		0.20			7.9	
b	0.18	0.23	0.30	7.1	9.1	11.8
D		3.00			118,1	
D2	2.23	2.38	2.48	87.8	93.7	97.7
E		3.00			118,1	
E2	1.49	1.64	1.74	58.7	64.6	68.5
e		0.50			19.7	
L	0.30	0.40	0.50	11.8	15.7	19.7



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