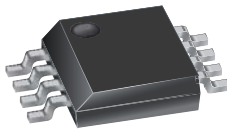


## Hall Effect Linear Current Sensor with Overcurrent Fault Output for <100 V Isolation Applications

### Features and Benefits

- No external sense resistor required; single package solution
- 1.2 mΩ internal conductor resistance; reduced power loss
- Economical low- and high-side current sensing
- Output voltage proportional to AC or DC currents
- ±12.5 A and ±25 A full scale sensing ranges
- Overcurrent FAULT trips and latches at 100% of full-scale current
- Low-noise analog signal path
- 100 kHz bandwidth
- Small footprint, low-profile SOIC8 package
- 3.0 to 5.5 V, single supply operation
- Integrated electrostatic shield for output stability
- Factory-trimmed for accuracy
- Extremely stable output offset voltage
- Zero magnetic hysteresis
- Ratiometric output from supply voltage

### Package: 8 Lead SOIC (suffix LC)



Approximate Scale 1:1



### Description

The Allegro® ACS711 provides economical and precise solutions for AC or DC current sensing in <100 V audio, communications systems, white goods, and automotive applications. The device package allows for easy implementation by the customer. Typical applications include circuit protection, current monitoring, and motor and inverter control.

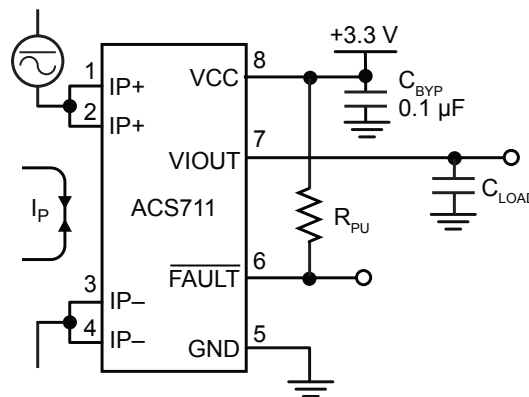
The device consists of a linear Hall sensor circuit with a copper conduction path located near the surface of the die. Applied current flowing through this copper conduction path generates a magnetic field which is sensed by the integrated Hall IC and converted into a proportional voltage. Device accuracy is optimized through the close proximity of the magnetic signal to the Hall transducer.

The output of the device has a positive slope proportional to the current flow from IP+ to IP- (pins 1 and 2, to pins 3 and 4). The internal resistance of this conductive path is 1.2 mΩ typical, providing a non-intrusive measurement interface that saves power in applications that require energy efficiency.

The ACS711 is optimized for low-side current sensing applications, although the terminals of the conductive path are electrically isolated from the sensor leads (pins 5 through 8), providing sufficient internal creepage and clearance dimensions for a low AC or DC working voltage applications. The thickness

*Continued on the next page...*

### Typical Application



Application 1. The ACS711 outputs an analog signal,  $V_{IOUT}$ , that varies linearly with the bi-directional AC or DC primary current,  $I_P$ , within the range specified. The FAULT pin trips when  $I_P$  reaches ±100% of its full-scale current.

# ACS711

## Hall Effect Linear Current Sensor with Overcurrent Fault Output for < 100 V Isolation Applications

### Description (continued)

of the copper conductor allows survival of the device at up to 5× overcurrent conditions.

The ACS711 is provided in a small, surface mount SOIC8 package. The leadframe is plated with 100% matte tin, which is compatible with standard lead (Pb) free printed circuit board assembly processes.

Internally, the device is Pb-free, except for flip-chip high-temperature Pb-based solder balls, currently exempt from RoHS. The device is fully calibrated prior to shipment from the factory.

### Selection Guide

Part Number	Packing <sup>1</sup>	T <sub>A</sub> (°C)	Optimized Accuracy Range, I <sub>P</sub> (A)	Sensitivity, Sens (Typ) (mV/A)
ACS711ELCTR-12AB-T <sup>2</sup>	Tape and reel, 3000 pieces/reel	-40 to 85	±12.5	110
ACS711ELCTR-25AB-T <sup>2</sup>	Tape and reel, 3000 pieces/reel	-40 to 85	±25	55
ACS711KLCTR-12AB-T <sup>2</sup>	Tape and reel, 3000 pieces/reel	-40 to 125	±12.5	110
ACS711KLCTR-25AB-T <sup>2</sup>	Tape and reel, 3000 pieces/reel	-40 to 125	±25	55

<sup>1</sup>Contact Allegro for additional packing options.

<sup>2</sup>Sensitivity measured with V<sub>CC</sub> = 3.3 V.

### Absolute Maximum Ratings

Characteristic	Symbol	Notes	Rating	Units
Supply Voltage	V <sub>CC</sub>		7	V
Reverse Supply Voltage	V <sub>RCC</sub>		-0.1	V
Output Voltage	V <sub>IOUT</sub>		7	V
Reverse Output Voltage	V <sub>RIOUT</sub>		-0.1	V
Working Voltage for Basic Isolation	V <sub>WORKING</sub>	Voltage applied between pins 1-4 and 5-8	100	VAC peak or VDC
FAULT Pin Voltage	V <sub>FAULT</sub>		7	V
Overcurrent Transient Tolerance	I <sub>POC</sub>	1 pulse, 100 ms	100	A
Nominal Operating Ambient Temperature	T <sub>A</sub>	Range E	-40 to 85	°C
		Range K	-40 to 125	°C
Maximum Junction Temperature	T <sub>J(max)</sub>		165	°C
Storage Temperature	T <sub>stg</sub>		-65 to 170	°C

### Thermal Characteristics

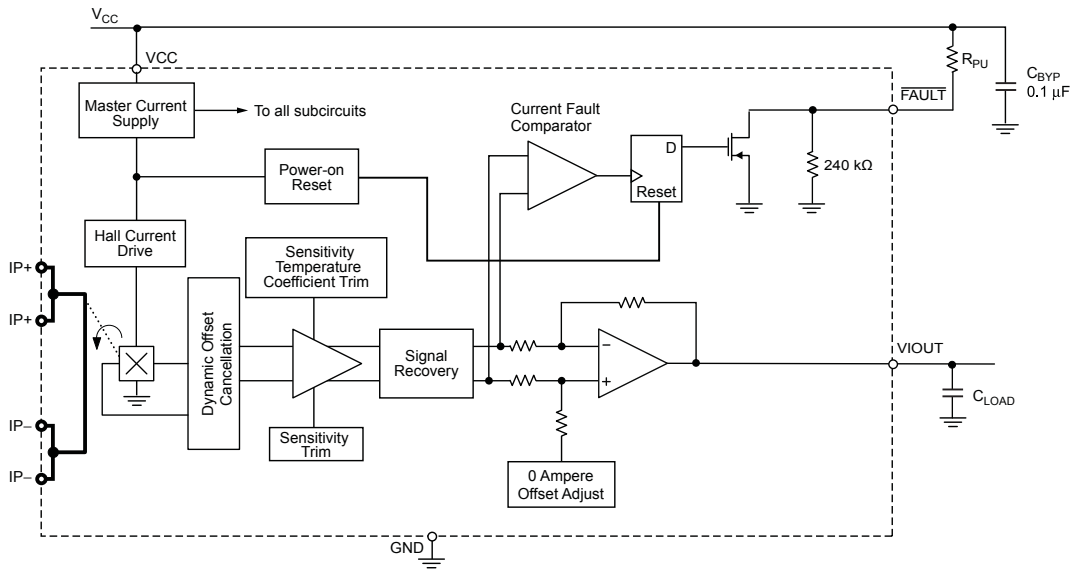
Characteristic	Symbol	Test Conditions <sup>1</sup>	Value	Units
Package Thermal Resistance, Junction to Lead	R <sub>θJL</sub>	Mounted on Allegro ASEK 711 evaluation board	5	°C/W
Package Thermal Resistance, Junction to Ambient <sup>2</sup>	R <sub>θJA</sub>	Mounted on Allegro 85-0404 evaluation board, includes the power consumed by the board	23	°C/W

<sup>1</sup>Additional thermal information available on the Allegro website

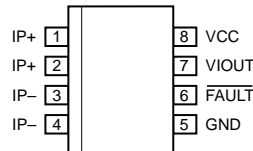
<sup>2</sup>The Allegro evaluation board has 1500 mm<sup>2</sup> of 2 oz. copper on each side, connected to pins 1 and 2, and to pins 3 and 4, with thermal vias connecting the layers. Performance values include the power consumed by the PCB. Further details on the board are available from the Frequently Asked Questions document on our website.



### Functional Block Diagram



### Pin-out Diagram



### Terminal List Table

Number	Name	Description
1 and 2	IP+	Terminals for current being sensed; fused internally
3 and 4	IP-	Terminals for current being sensed; fused internally
5	GND	Signal ground terminal
6	$\overline{\text{FAULT}}$	Overcurrent fault; active low
7	VIOUT	Analog output signal
8	VCC	Device power supply terminal

### COMMON OPERATING CHARACTERISTICS over full range of $T_A$ and $V_{CC} = 3.3$ V, unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>ELECTRICAL CHARACTERISTICS</b>						
Supply Voltage	$V_{CC}$		3	3.3	5.5	V
Supply Current	$I_{CC}$	$V_{CC} = 3.3$ V, output open	–	4	5.5	mA
Output Capacitance Load	$C_{LOAD}$	VIOUT to GND	–	–	1	nF
Output Resistive Load	$R_{LOAD}$	VIOUT to GND	15	–	–	k $\Omega$
Primary Conductor Resistance	$R_{IP}$	$T_A = 25^\circ\text{C}$	–	1.2	–	m $\Omega$
VIOUT Rise Time	$t_r$	$I_P = I_{P\text{MAX}}$ , $T_A = 25^\circ\text{C}$ , COU $\bar{T}$ = open	–	3.5	–	$\mu\text{s}$
Propagation Delay Time	$t_{PROP}$	$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , COU $\bar{T}$ = open	–	1.2	–	$\mu\text{s}$
Response Time	$t_{RESPONSE}$	$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , COU $\bar{T}$ = open	–	4.6	–	$\mu\text{s}$
Internal Bandwidth <sup>1</sup>	$BW_I$	–3 dB, $T_A = 25^\circ\text{C}$	–	100	–	kHz
Nonlinearity	$E_{LIN}$	Over full range of $I_P$	–	$\pm 1$	–	%
Symmetry	$E_{SYM}$	Apply full scale $I_P$	–	100	–	%
VIOUT Saturation Voltages	$V_{IOH}$		$V_{CC} - 0.3$	–	–	V
	$V_{IOL}$		–	–	0.3	V
Quiescent Output Voltage	$V_{IOUT(Q)}$	$I_P = 0$ A, $T_A = 25^\circ\text{C}$	–	$V_{CC} / 2$	–	V
Power-On Time	$t_{PO}$	Output reaches 90% of steady-state level, $T_A = 25^\circ\text{C}$ , 20 A present on primary conductor	–	35	–	$\mu\text{s}$
<b>FAULT Pin Characteristics</b>						
FAULT $\bar{O}$ Operating Point	$I_{FAULT}$		–	$\pm 1 \times I_P$	–	A
FAULT $\bar{O}$ Output Pullup Resistor	$R_{PU}$		1	–	–	k $\Omega$
FAULT $\bar{O}$ Output Voltage	$V_{OH}$		–	$V_{CC} - 0.3$	–	V
	$V_{OL}$	$R_{PU} = 1$ k $\Omega$	–	0.3	–	V
FAULT $\bar{O}$ Response Time	$t_{FAULT}$	Measured from $ I_P  >  I_{FAULT} $ to $V_{FAULT} \leq V_{OL}$	–	1.3	–	$\mu\text{s}$
$V_{CC}$ Off Voltage Level for Fault Reset <sup>2</sup>	$V_{CCFR}$		–	–	200	mV
$V_{CC}$ Off Duration for Fault Reset <sup>2</sup>	$t_{CCFR}$		100	–	–	$\mu\text{s}$

<sup>1</sup>Calculated using the formula  $BW_I = 0.35 / t_r$ .

<sup>2</sup>After the FAULT $\bar{O}$  pin is latched low, the only way to reset it is through a power-off and power-on cycle on the VCC pin. To get a guaranteed fault reset,  $V_{CC}$  must stay below  $V_{CCFR}$  for a period greater than  $t_{CCFR}$  before settling to the normal operation voltage (3 to 5.5 V).

### x12.5A PERFORMANCE CHARACTERISTICS for E TEMPERATURE RANGE $T_A = 25^\circ\text{C}$ and $V_{CC} = 3.3\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Optimized Accuracy Range	$I_P$		-12.5	-	12.5	A
Sensitivity	Sens	Over full range of $I_P$	-	110	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	110	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = 25^\circ\text{C}$ to $85^\circ\text{C}$	-	110	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^\circ\text{C}$ , no external low pass filter on VIOUT	-	11	-	mV
Electrical Offset Voltage	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $85^\circ\text{C}$	-	$\pm 40$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 50$	-	mV
Total Output Error <sup>3</sup>	$E_{TOT}$	$I_P = \pm 12.5\text{ A}$ , $T_A = -40^\circ\text{C}$ to $85^\circ\text{C}$	-	$\pm 5$	-	%

<sup>1</sup>See Characteristic Performance Data for parameter distributions over temperature.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup>Percentage of  $I_P$ , with  $I_P = \pm 12.5\text{ A}$ .

### x12.5A PERFORMANCE CHARACTERISTICS for K TEMPERATURE RANGE<sup>1</sup> $T_A = 25^\circ\text{C}$ and $V_{CC} = 3.3\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Optimized Accuracy Range	$I_P$		-12.5	-	12.5	A
Sensitivity	Sens	Over full range of $I_P$	-	110	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	110	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	110	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^\circ\text{C}$ , no external low pass filter on VIOUT	-	11	-	mV
Electrical Offset Voltage	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 40$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 50$	-	mV
Total Output Error <sup>3</sup>	$E_{TOT}$	$I_P = \pm 12.5\text{ A}$ , $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 5$	-	%

<sup>1</sup>See Characteristic Performance Data for parameter distributions over temperature.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup>Percentage of  $I_P$ , with  $I_P = \pm 12.5\text{ A}$ .

### x25A PERFORMANCE CHARACTERISTICS for E TEMPERATURE RANGE $T_A = 25^\circ\text{C}$ and $V_{CC} = 3.3\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Optimized Accuracy Range	$I_P$		-25	-	25	A
Sensitivity	Sens	Over full range of $I_P$	-	55	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	55	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = 25^\circ\text{C}$ to $85^\circ\text{C}$	-	55	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^\circ\text{C}$ , no external low pass filter on VIOUT	-	8	-	mV
Electrical Offset Voltage	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $85^\circ\text{C}$	-	$\pm 30$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 35$	-	mV
Total Output Error <sup>3</sup>	$E_{TOT}$	$I_P = \pm 25\text{ A}$ , $T_A = -40^\circ\text{C}$ to $85^\circ\text{C}$	-	$\pm 4$	-	%

<sup>1</sup>See Characteristic Performance Data for parameter distributions over temperature.

<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup>Percentage of  $I_P$ , with  $I_P = \pm 25\text{ A}$ .

### x25A PERFORMANCE CHARACTERISTICS for K TEMPERATURE RANGE<sup>1</sup> $T_A = 25^\circ\text{C}$ and $V_{CC} = 3.3\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Optimized Accuracy Range	$I_P$		-25	-	25	A
Sensitivity	Sens	Over full range of $I_P$	-	55	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	55	-	mV/A
		Full scale of $I_P$ applied for 5 ms, $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	55	-	mV/A
Noise <sup>2</sup>	$V_{NOISE}$	$T_A = 25^\circ\text{C}$ , no external low pass filter on VIOUT	-	8	-	mV
Electrical Offset Voltage	$V_{OE(TA)}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$	-	$\pm 5$	-	mV
	$V_{OE(TOP)HT}$	$I_P = 0\text{ A}$ , $T_A = 25^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 30$	-	mV
	$V_{OE(TOP)LT}$	$I_P = 0\text{ A}$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 35$	-	mV
Total Output Error <sup>3</sup>	$E_{TOT}$	$I_P = \pm 25\text{ A}$ , $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$	-	$\pm 4$	-	%

<sup>1</sup>See Characteristic Performance Data for parameter distributions over temperature.

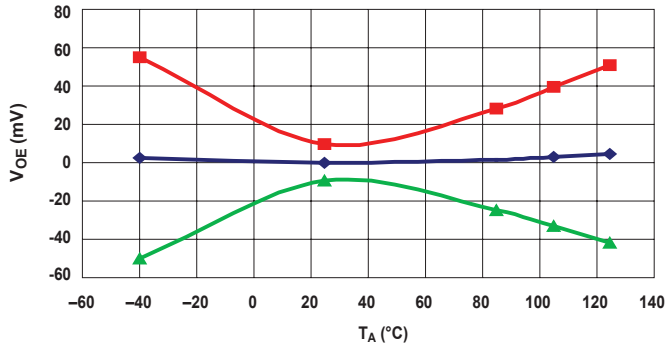
<sup>2</sup> $\pm 3$  sigma noise voltage.

<sup>3</sup>Percentage of  $I_P$ , with  $I_P = \pm 25\text{ A}$ .

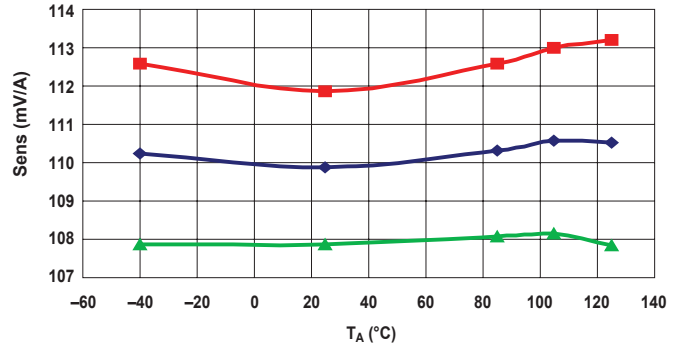
### Characteristic Performance Data Data taken using the ACS711KLC-12A, $V_{CC} = 3.3\text{ V}$

#### Accuracy Data

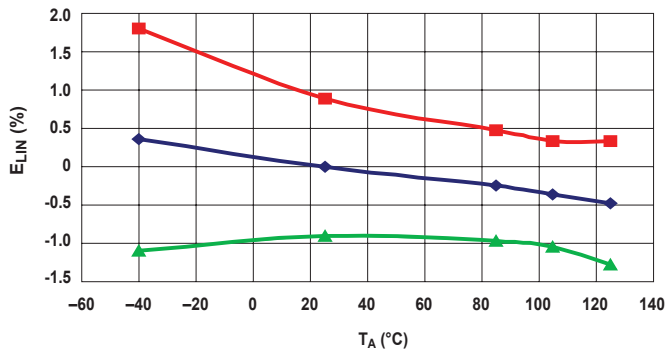
Electrical Offset Voltage versus Ambient Temperature



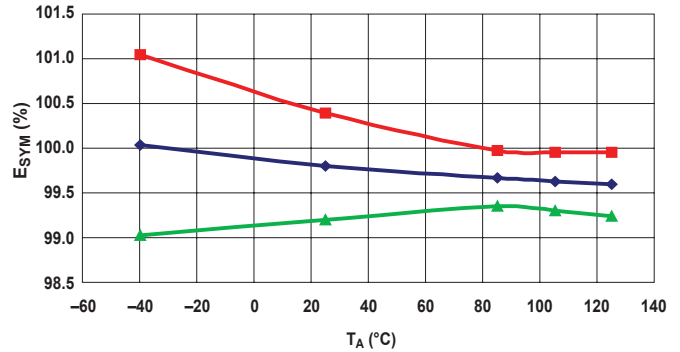
Sensitivity versus Ambient Temperature



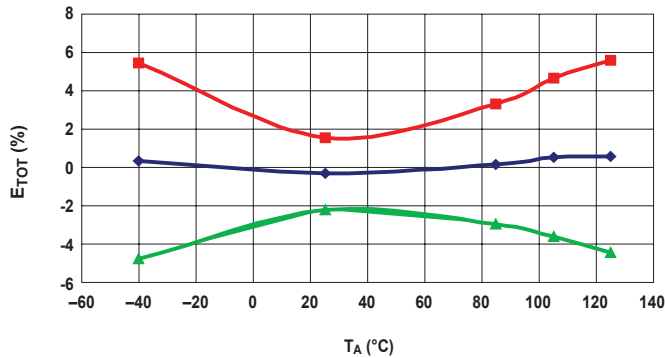
Nonlinearity versus Ambient Temperature



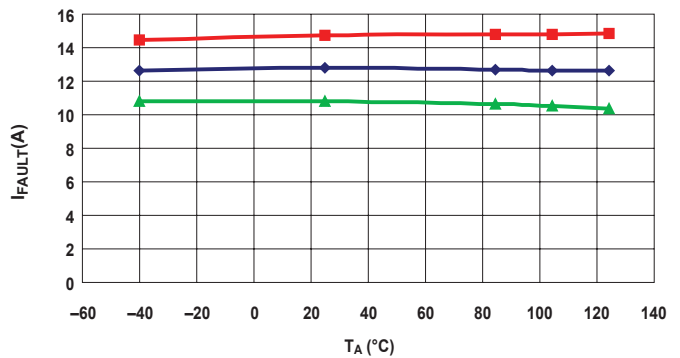
Symmetry versus Ambient Temperature



Total Output Error versus Ambient Temperature



Fault Operating Point versus Ambient Temperature

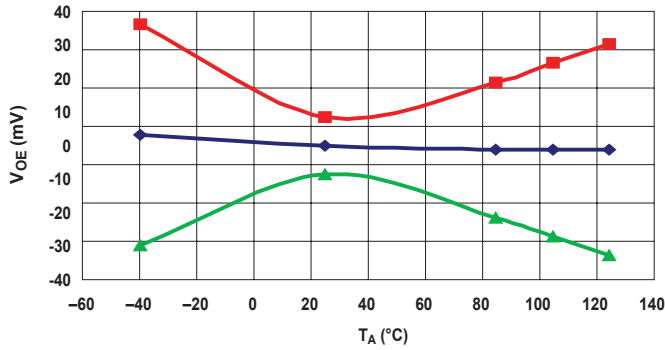


—■— Typical Maximum Limit —◆— Mean —▲— Typical Minimum Limit

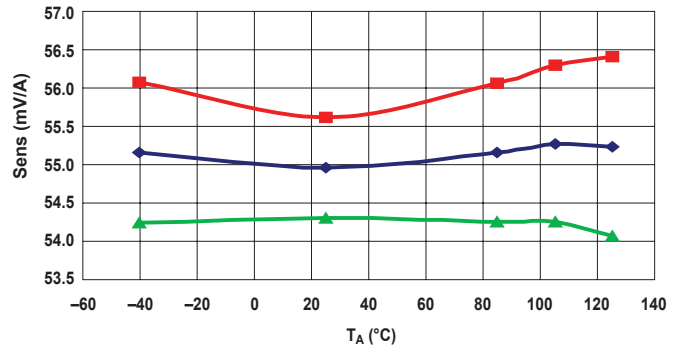
### Characteristic Performance Data Data taken using the ACS711KLC-25A, $V_{CC} = 3.3\text{ V}$

#### Accuracy Data

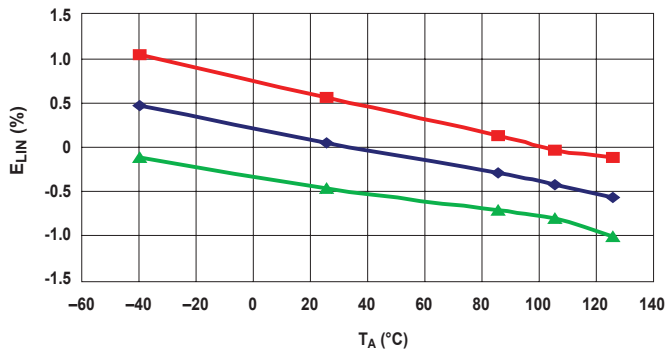
Electrical Offset Voltage versus Ambient Temperature



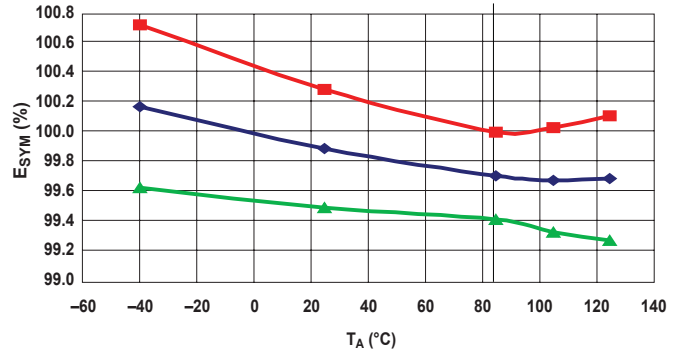
Sensitivity versus Ambient Temperature



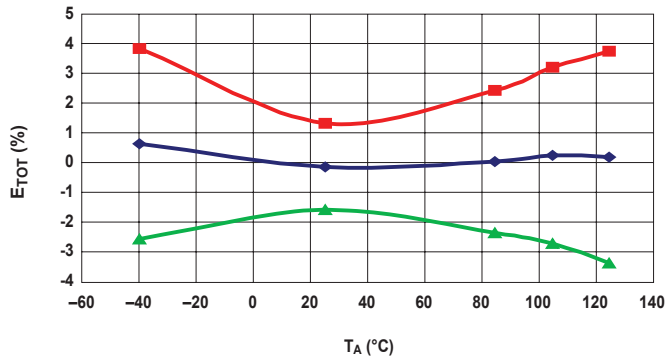
Nonlinearity versus Ambient Temperature



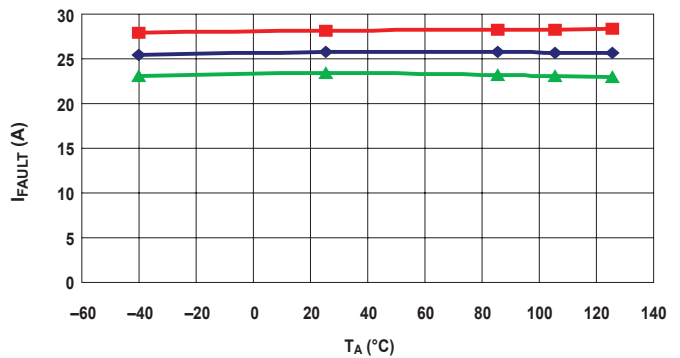
Symmetry versus Ambient Temperature



Total Output Error versus Ambient Temperature



Fault Operating Point versus Ambient Temperature



—■— Typical Maximum Limit    —◆— Mean    —▲— Typical Minimum Limit

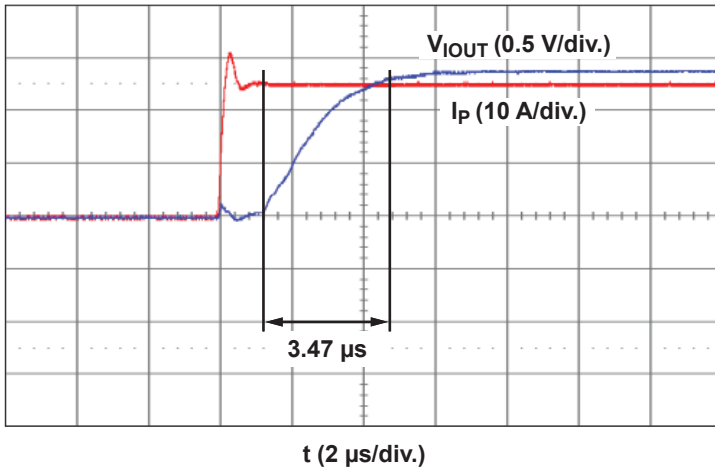


### Characteristic Performance Data

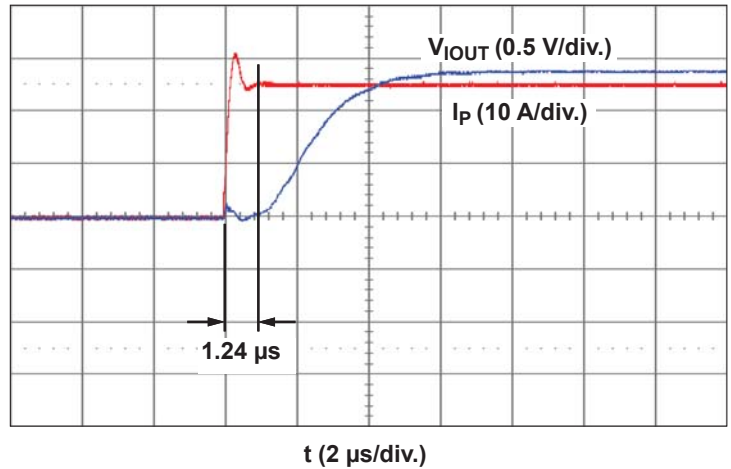
Data taken using the ACS711KLC-25A

#### Timing Data

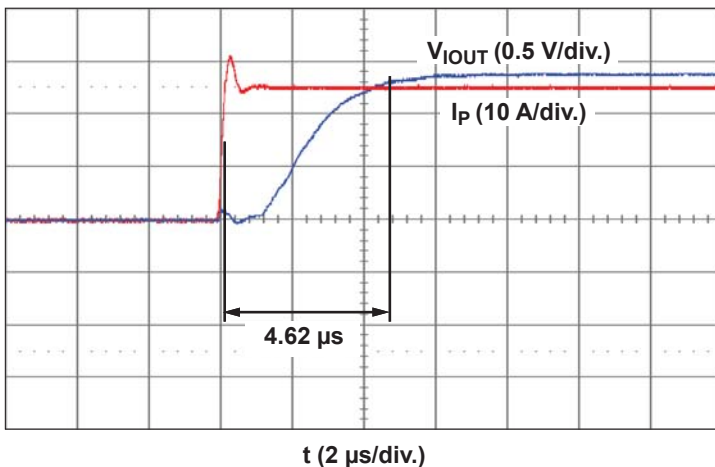
Rise Time



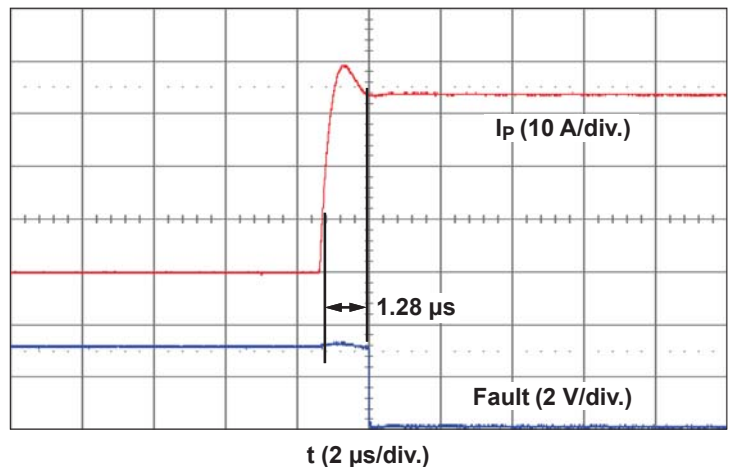
Propagation Delay Time



Response Time



Fault Response



### Definitions of Accuracy Characteristics

**Sensitivity (Sens).** The change in sensor output in response to a 1 A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the full-scale current of the device.

**Noise (V<sub>NOISE</sub>).** The product of the linear IC amplifier gain (mV) and the noise floor for the Allegro Hall effect linear IC. The noise floor is derived from the thermal and shot noise observed in Hall elements. Dividing the noise (mV) by the sensitivity (mV/A) provides the smallest current that the device is able to resolve.

**Linearity (E<sub>LIN</sub>).** The degree to which the voltage output from the sensor varies in direct proportion to the primary current through its full-scale amplitude. Nonlinearity in the output can be attributed to the saturation of the flux concentrator approaching the full-scale current. The following equation is used to derive the linearity:

where  $V_{IOUT\_full\text{-scale amperes}}$  = the output voltage (V) when the

$$100 \left\{ 1 - \left[ \frac{\Delta \text{ gain} \times \% \text{ sat} (V_{IOUT\_full\text{-scale amperes}} - V_{IOUT(Q)})}{2 (V_{IOUT\_half\text{-scale amperes}} - V_{IOUT(Q)})} \right] \right\}$$

sensed current approximates full-scale  $\pm I_p$ .

**Symmetry (E<sub>SYM</sub>).** The degree to which the absolute voltage output from the sensor varies in proportion to either a positive or negative full-scale primary current. The following formula is used to derive symmetry:

$$100 \left( \frac{V_{IOUT\_+ \text{ full-scale amperes}} - V_{IOUT(Q)}}{V_{IOUT(Q)} - V_{IOUT\_ - \text{ full-scale amperes}}} \right)$$

**Quiescent output voltage (V<sub>IOUT(Q)</sub>).** The output of the sensor when the primary current is zero. For a unipolar supply voltage, it nominally remains at  $V_{CC}/2$ . Thus,  $V_{CC} = 3.3 \text{ V}$  translates into  $V_{IOUT(Q)} = 1.65 \text{ V}$ . Variation in  $V_{IOUT(Q)}$  can be attributed to the resolution of the Allegro linear IC quiescent voltage trim and thermal drift.

**Electrical offset voltage (V<sub>OE</sub>).** The deviation of the device output from its ideal quiescent value of  $V_{CC}/2$  due to nonmagnetic causes. To convert this voltage to amperes, divide by the device sensitivity, Sens.

**Accuracy (E<sub>TOT</sub>).** The accuracy represents the maximum deviation of the actual output from its ideal value. This is also known as the total output error. The accuracy is illustrated graphically in the output voltage versus current chart below.

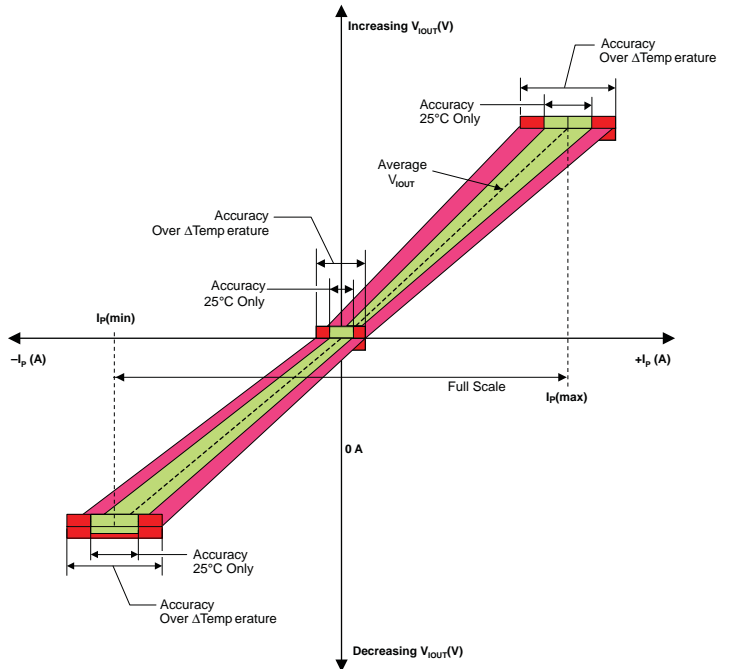
**Ratiometry.** The ratiometric feature means that its 0 A output,  $V_{IOUT(Q)}$ , (nominally equal to  $V_{CC}/2$ ) and sensitivity, Sens, are proportional to its supply voltage,  $V_{CC}$ . The following formula is used to derive the ratiometric change in 0 A output voltage,  $\Delta V_{IOUT(Q)RAT}$  (%).

$$100 \left( \frac{V_{IOUT(Q)VCC} / V_{IOUT(Q)3.3V}}{V_{CC} / 3.3 \text{ V}} \right)$$

The ratiometric change in sensitivity,  $\Delta \text{Sens}_{RAT}$  (%), is defined as:

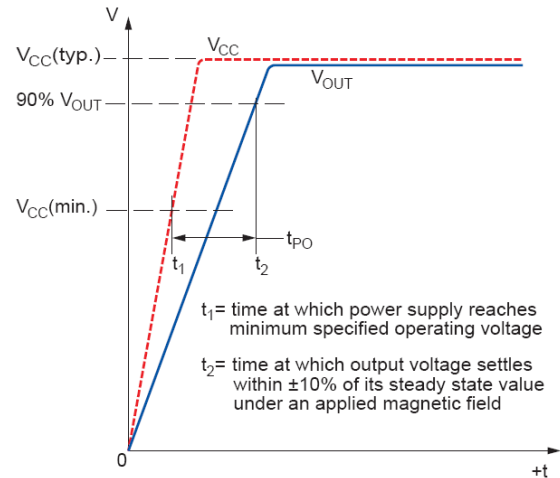
$$100 \left( \frac{\text{Sens}_{VCC} / \text{Sens}_{3.3V}}{V_{CC} / 3.3 \text{ V}} \right)$$

**Output Voltage versus Sensed Current**  
Accuracy at 0 A and at Full-Scale Current

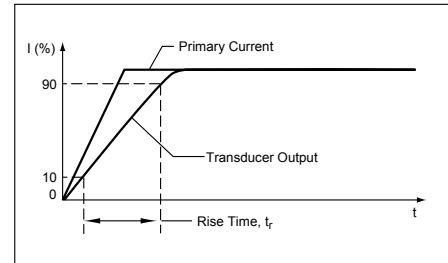


### Definitions of Dynamic Response Characteristics

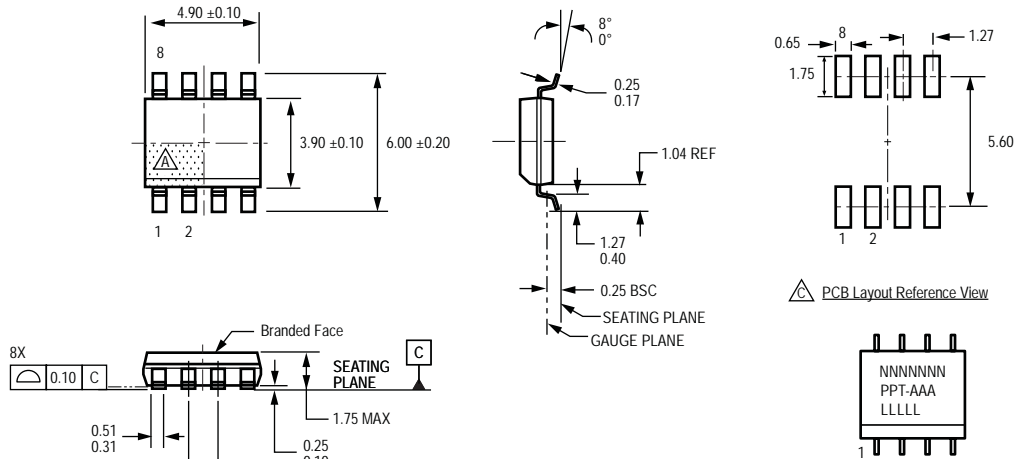
**Power-On Time ( $t_{PO}$ ).** When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field. Power-On Time,  $t_{PO}$ , is defined as the time it takes for the output voltage to settle within  $\pm 10\%$  of its steady state value under an applied magnetic field, after the power supply has reached its minimum specified operating voltage,  $V_{CC(min)}$ , as shown in the chart at right.



**Rise time ( $t_r$ ).** The time interval between a) when the sensor reaches 10% of its full scale value, and b) when it reaches 90% of its full scale value. The rise time to a step response is used to derive the bandwidth of the current sensor, in which  $f(-3 \text{ dB}) = 0.35/t_r$ . Both  $t_r$  and  $t_{RESPONSE}$  are detrimentally affected by eddy current losses observed in the conductive IC ground plane.



### Package LC, 8-pin SOIC



For Reference Only; not for tooling use (reference MS-012AA)  
 Dimensions in millimeters  
 Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown

- Terminal #1 mark area
- Branding scale and appearance at supplier discretion
- Reference land pattern layout (reference IPC7351)
- SOIC127P600X175-8M); all pads a minimum of 0.20 mm from all adjacent pads; adjust as necessary to meet application process requirements and PCB layout tolerances

#### PCB Layout Reference View

Standard Branding Reference View

N = Device part number  
 P = Package Designator  
 T = Device temperature range  
 A = Amperage  
 L = Lot number  
 Belly Brand = Country of Origin

Copyright ©2008-2010, Allegro MicroSystems, Inc.

Allegro MicroSystems, Inc. reserves the right to make, from time to time, such departures from the detail specifications as may be required to permit improvements in the performance, reliability, or manufacturability of its products. Before placing an order, the user is cautioned to verify that the information being relied upon is current.

Allegro's products are not to be used in life support devices or systems, if a failure of an Allegro product can reasonably be expected to cause the failure of that life support device or system, or to affect the safety or effectiveness of that device or system.

The information included herein is believed to be accurate and reliable. However, Allegro MicroSystems, Inc. assumes no responsibility for its use; nor for any infringement of patents or other rights of third parties which may result from its use.

For the latest version of this document, visit our website:

[www.allegromicro.com](http://www.allegromicro.com)



Allegro MicroSystems, Inc.  
 115 Northeast Cutoff  
 Worcester, Massachusetts 01615-0036 U.S.A.  
 1.508.853.5000; [www.allegromicro.com](http://www.allegromicro.com)