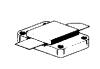
The RF Line **NPN Silicon RF Power Transistor**

... designed primarily for high-voltage applications as a high-power linear amplifier from 2 to 30 MHz. Ideal for marine and base station equipment.

- Specified 50 Volt, 30 MHz Characteristics Output Power = 600 W Minimum Gain = 10 dB Efficiency = 40%
- Intermodulation Distortion @ 600 W(PEP) -- IMD = -30 dB
- Diffused Emitter Resistors for Superior Ruggedness
- Low Thermal Resistance

MRF430

600 WATTS (LINEAR) 30 MHz RF POWER TRANSISTOR NPN SILICON



CASE 368-01, STYLE 1

MAXIMUM RATINGS

Rating	Symbol	Value	Unit Vdc	
Collector-Emitter Voltage	VCEO	50		
Collector-Base Voltage	Усво	110	Vdc	
Emitter-Base Voltage	VEBO	4	Vdc	
Collector Current — Continuous	¹c	60	Adc	
Operating Junction Temperature	TJ	200	°C	
Total Device Dissipation (a. T _C = 25 °C Derate Above 25 °C	PD	875 5	Watts W/°C	
Storage Temperature Range	T _{stg}	-65 to +150	°C	

THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	R _O JC	0.20	°C/W

ELECTRICAL CHARACTERISTICS (TC = 25°C unless otherwise noted)

Characteristic	Symbol	Min	Тур	Max	Unit
OFF CHARACTERISTICS					
Collector-Emitter Breakdown Voltage (IC = 500 mAdc, IB = 0)	V(BR)CEO	50	_	_	Vdc
Collector-Emitter Breakdown Voltage (I _C = 200 mAdc, V _{BE} = 0)	V(BR)CES	110			Vdc
Emitter-Base Breakdown Voltage (IE = 20 mAdc, IC = 0)	V(BR)EBO	4	-	_	Vdc
ON CHARACTERISTICS					
DC Current Gain (IC = 20 Adc, VCE = 10 Vdc)	pŁE	10	30	80	_

(continued)

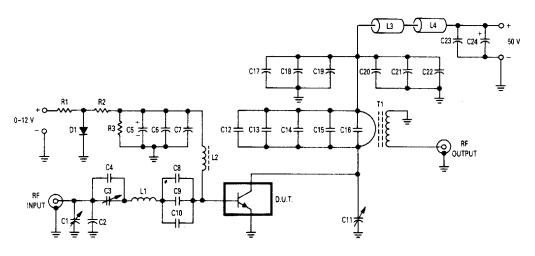
MOTOROLA RF DEVICE DATA

2-636

ELECTRICAL CHARACTERISTICS — continued (T_C = 25°C unless otherwise noted)

Characteristic		Symbol	Min	Тур	Max	Unit
YNAMIC CHARACTERISTICS				L	L1	
Output Capacitance (V _{CB} = 50 Vdc, I _E = 0, f = 1 MHz)		Cop	_	900	1200	pF
UNCTIONAL TEST						
Common-Emitter Amplifier Power Gain (V _{CC} = 50 Vdc, P _{out} = 600 W (CW), f = 30 MHz, I _{CQ} = 600 mA)		GPE	10	13	-	₫B
Collector Efficiency (V _{CC} = 50 Vdc, P _{out} = 600 W, f = 30 MHz, I _{CQ} = 600 mA)	(PEP) (CW)	η	_	40 60	_	% %
Intermodulation Distortion (1) (VCE = 50 Vdc, P_{Out} = 600 W (PEP), I_{CQ} = 600 mA, f = 30 MHz)		IMD	_	- 30	_	dB

⁽¹⁾ To Mil Std 1311 Version A, Test Method 2204B, Two Tone, Reference Each Tone.



- C1, C3, C11 Arco 469 or equivalent

- C1, C3, C11 Arco 469 or equivalent C2 820 pF C4 330 pF C5 1000 μ F/3 V Electrolytic C6, C8, C9, C10, C17, C18, C19 0.1 μ F C7, C22, C23 0.47 μ F, RMC Type 2225C or equivalent C12, C13, C14 470 pF

- C15 1000 pF C16 Two Unencapsulated 1000 pF Mica in Series
- C20, C21 0.039 μ F C24 10 μ F/100 V Electrolytic
- D1 1N4997 or equivalent R1 — 10 Ohms/10 W
- R2 0.1 Ohm/½ W R3 2.7 Ohms/2 W

- L1 -- 2 Turns #14 AWG, 1/2" ID, 1/2" Long
- L2 Ohmite Z-235 or equivalent
- L3, L4 Ferrite Beads, Fair-Rite Products Corp. #2673000801 or equivalent
- T1 RF Transformer, 1:25 Impedance Ratio. See Motorola Application Note AN749, Figure 4 for details. Ferrite Material: 2 Each, Fair-Rite Product Corp. #2667540001 or equivalent

All capacitors ATC type 100/200 chips or equivalent, unless otherwise noted.

Figure 1. 30 MHz Test Circuit Schematic

MOTOROLA RF DEVICE DATA

2-637

46E D = 6367254 0094660 2 = MOT6 **MRF430**

TYPICAL CHARACTERISTICS

POWER GAIN

ę. G

1.5

MOTOROLA SC (XSTRS/R F) 1000 P_{OUL} OUTPUT POWER (WATTS CW) Vcc ÷ 50 \ ICQ = 600 mA VCC 40 \

T-33-15

V_{CC} = 50 V I_{CQ} = 600 mA P_{out} = 600 W

Figure 3. Power Gain versus Frequency

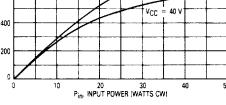


Figure 2. Output Power versus Input Power

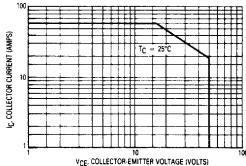
800 = 30,30,001 MH P_{OUI}. OUTPUT POWER (WATTS PEP) ICO = 600 mAMD - 30 dB – 35 dB 20 30 40 VCC, SUPPLY VOLTAGE (VOLTS)

IMD, INTERMODULATION DISTORTION (dB) V_{CC} = 50 V f = 30,30.001 MHz - 50 L 300 400 P_{OUT}, OUTPUT POWER (WATTS PEP) 600

6 8 10 f, FREQUENCY (MHz)

Figure 4. Output Power versus Supply Voltage

Figure 5. IMD versus Output Power





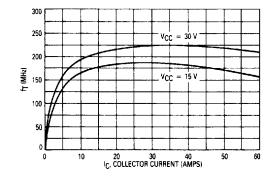


Figure 7. f_T versus Collector Current

MOTOROLA RF DEVICE DATA

2-638



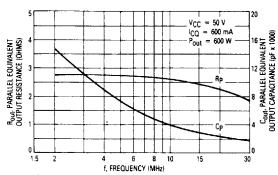


Figure 8. Output Resistance and Capacitance versus Frequency

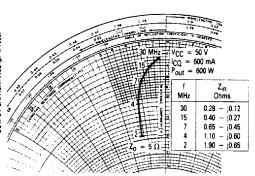


Figure 9. Series Equivalent Input Impedance

MOUNTING OF HIGH POWER RF **POWER TRANSISTORS**

The package of this device is designed for conduction cooling. It is extremely important to minimize the thermal resistance between the device flange and the heat dissipator.

Since the device mounting flange is made of soft copper, it may be deformed during various stages of handling or during transportation. It is recommended that the user makes a final inspection on this before the device installation. ±0.0005" is considered sufficient for the flange bottom.

The same applies to the heat dissipator in the device mounting area. If copper heatsink is not used, a copper head spreader is strongly recommended between the device mounting surfaces and the main heatsink, it should be at least 1/4" thick and extend at least one inch from the flange edges. A thin layer of thermal compound in all interfaces is, of course, essential. The recommended torque on the 4-40 mounting screws should be in the area of 4-5 lbs.-inch, and spring type lock washers along with flat washers are recommended.

For die temperature calculations, the Δ temperature from a corner mounting screw area to the bottom center of the flange is approximately 5°C and 10°C under normal operating conditions (dissipation 150 W and 300 W respectively).

The main heat dissipator must be sufficiently large and have low $R_{oldsymbol{ heta}}$ for moderate air velocity, unless liquid cooling is employed.

CIRCUIT CONSIDERATIONS

At high power levels (500 W and up), the circuit layout becomes critical due to the low impedance levels and high RF currents associated with the output matching. Some of the components, such as capacitors and inductors must also withstand these currents. The component losses are directly proportional to the operating frequency. The manufacturers specifications on capacitor ratings should be consulted on these aspects prior to design.

Push-pull circuits are less critical in general, since the ground referenced RF loops are practically eliminated, and the impedance levels are higher for a given power output. High power broadband transformers are also easier to design than comparable LC matching networks.