

# **Open-Drain Output Sub-Microamp Comparators**

#### Features

- Low Quiescent Current: 600 nA/comparator (typ.)
- Rail-to-Rail Input:  $V_{SS}$  0.3V to  $V_{DD}$  + 0.3V
- Open-Drain Output: V<sub>OUT</sub> ≤ 10V
- Propagation Delay: 4 µs (typ., 100 mV Overdrive)
- Wide Supply Voltage Range: 1.6V to 5.5V
- · Single available in SOT-23-5, SC-70-5 \* packages
- · Available in Single, Dual and Quad
- Chip Select (CS) with MCP6548
- Low Switching Current
- Internal Hysteresis: 3.3 mV (typ.)
- Temperature Range:
- Industrial: -40°C to +85°C
- Extended: -40°C to +125°C

#### **Typical Applications**

- · Laptop Computers
- Mobile Phones
- Metering Systems
- Hand-held Electronics
- RC Timers
- · Alarm and Monitoring Circuits
- Windowed Comparators
- Multi-vibrators

#### **Related Devices**

· CMOS/TTL-Compatible Output: MCP6541/2/3/4

### Description

The Microchip Technology Inc. MCP6546/7/8/9 family of comparators is offered in single (MCP6546, MCP6546R, MCP6546U), single with chip select (MCP6548), dual (MCP6547) and quad (MCP6549) configurations. The outputs are open-drain and are capable of driving heavy DC or capacitive loads.

These comparators are optimized for low power, single-supply application with greater than rail-to-rail input operation. The output limits supply current surges and dynamic power consumption while switching. The open-drain output of the MCP6546/7/8/9 family can be used as a level-shifter for up to 10V using a pull-up resistor. It can also be used as a wired-OR logic. The internal Input hysteresis eliminates output switching due to internal noise voltage, reducing current draw. These comparators operate with a single-supply voltage as low as 1.6V and draw a quiescent current of less than 1 µA/comparator.

The related MCP6541/2/3/4 family of comparators from Microchip has a push-pull output that supports rail-to-rail output swing and interfaces with CMOS/TTL logic.

\* SC-70-5 E-Temp parts not available at this release of the data sheet.

MCP6546U SOT-23-5 is E-Temp only.



# 1.0 ELECTRICAL CHARACTERISTICS

## Absolute Maximum Ratings †

V <sub>DD</sub> - V <sub>SS</sub>
Open-Drain outputV <sub>SS</sub> + 10.5V
Analog Input (V <sub>IN</sub> +, V <sub>IN</sub> -)††V <sub>SS</sub> - 1.0V to V <sub>DD</sub> + 1.0V
All other inputs and outputs $V_{SS}$ – 0.3V to $V_{DD}$ + 0.3V
Difference Input voltage $ V_{DD} - V_{SS} $
Output Short-Circuit Currentcontinuous
Current at Input Pins±2 mA
Current at Output and Supply Pins±30 mA
Storage temperature65°C to +150°C
Maximum Junction Temperature (T <sub>J</sub> )+150°C
ESD protection on all pins:
(HBM;MM)2 kV;200V (MCP6546U)
(HBM;MM) 4 kV; 200V (all other parts)

**† Notice:** Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

**††** See Section 4.1.2 "Input Voltage and Current Limits"

# DC CHARACTERISTICS

Electrical Specifications: Unless otherwise indicated, V <sub>DD</sub> = +1.6V to +5.5V, V <sub>SS</sub> = GND, T <sub>A</sub> = 25°C, V <sub>IN</sub> + = V <sub>DD</sub> /2, V	/ <sub>IN</sub> - = V <sub>SS</sub> ,
$R_{PLI} = 2.74 \text{ k}\Omega$ to $V_{PLI} = V_{DD}$ (Refer to Figure 1-3).	

Parameters	Sym	Min	Тур	Max	Units	Conditions
Power Supply						
Supply Voltage	V <sub>DD</sub>	1.6	_	5.5	V	$V_{PU} \ge V_{DD}$
Quiescent Current	Ι <sub>Q</sub>	0.3	0.6	1	μA	I <sub>OUT</sub> = 0
						[
Input Voltage Range	V <sub>CMR</sub>	V <sub>SS</sub> – 0.3	—	V <sub>DD</sub> + 0.3	V	
Common Mode Rejection Ratio	CMRR	55	70	_	dB	$V_{DD}$ = 5V, $V_{CM}$ = -0.3V to 5.3V
Common Mode Rejection Ratio	CMRR	50	65		dB	$V_{DD}$ = 5V, $V_{CM}$ = 2.5V to 5.3V
Common Mode Rejection Ratio	CMRR	55	70	_	dB	$V_{DD}$ = 5V, $V_{CM}$ = -0.3V to 2.5V
Power Supply Rejection Ratio	PSRR	63	80	_	dB	$V_{CM} = V_{SS}$
Input Offset Voltage	V <sub>OS</sub>	-7.0	±1.5	+7.0	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Drift with Temperature	$\Delta V_{OS} / \Delta T_A$	_	±3		µV/°C	$T_A = -40^{\circ}C$ to +125°C, $V_{CM} = V_{SS}$
Input Hysteresis Voltage	V <sub>HYST</sub>	1.5	3.3	6.5	mV	V <sub>CM</sub> = V <sub>SS</sub> (Note 1)
Linear Temp. Co.	TC <sub>1</sub>	_	6.7		µV/°C	$T_A = -40^{\circ}C \text{ to } +125^{\circ}C, V_{CM} = V_{SS}$ (Note 2)
Quadratic Temp. Co.	TC <sub>2</sub>	_	-0.035		µV/°C <sup>2</sup>	$T_A = -40^{\circ}C \text{ to } +125^{\circ}C, V_{CM} = V_{SS} \text{ (Note 2)}$
Input Bias Current	I <sub>B</sub>	_	1	_	pА	$V_{CM} = V_{SS}$
At Temperature (I-Temp parts)	I <sub>B</sub>	_	25	100	pА	T <sub>A</sub> = +85°C, V <sub>CM</sub> = V <sub>SS</sub> (Note 3)
At Temperature (E-Temp parts)	Ι <sub>Β</sub>	_	1200	5000	pА	T <sub>A</sub> = +125°C, V <sub>CM</sub> = V <sub>SS</sub> (Note 3)
Input Offset Current	I <sub>OS</sub>	_	±1	_	pА	$V_{CM} = V_{SS}$
Common Mode Input Impedance	Z <sub>CM</sub>	_	10 <sup>13</sup>   4	—	Ω∥pF	
Differential Input Impedance	Z <sub>DIFF</sub>	_	10 <sup>13</sup>   2	_	Ω∥pF	

**Note 1:** The input offset voltage is the center of the input-referred trip points. The input hysteresis is the difference between the input-referred trip points.

2:  $V_{HYST}$  at differential temperatures is estimated using:  $V_{HYST}$  (T<sub>A</sub>) =  $V_{HYST}$  + (T<sub>A</sub> - 25°C) TC<sub>1</sub> + (T<sub>A</sub> - 25°C)<sup>2</sup>TC<sub>2</sub>.

3: Input bias current at temperature is not tested for the SC-70-5 package

4: Do not short the output above V<sub>SS</sub> + 10V. Limit the output current to Absolute Maximum Rating of 30 mA. The minimum V<sub>PU</sub> test limit was V<sub>DD</sub> before Dec. 2004 (week code 52).

# **DC CHARACTERISTICS (CONTINUED)**

Electrical Specifications: Unless otherwise indicated, $V_{DD}$ = +1.6V to +5.5V, $V_{SS}$ = GND, $T_A$ = 25°C, $V_{IN}$ + = $V_{DD}/2$ , $V_{IN}$ - = $V_{SS}$ , $R_{PU}$ = 2.74 k $\Omega$ to $V_{PU}$ = $V_{DD}$ (Refer to Figure 1-3).									
Parameters	Sym	Min	Тур	Мах	Units	Conditions			
Open-Drain Output									
Output Pull-Up Voltage	V <sub>PU</sub>	1.6	_	10	V	(Note 4)			
High-Level Output Current	I <sub>OH</sub>	-100	_	—	nA	V <sub>DD</sub> = 1.6V to 5.5V, V <sub>PU</sub> = 10V (Note 4)			
Low-Level Output Voltage	V <sub>OL</sub>	V <sub>SS</sub>	_	V <sub>SS</sub> + 0.2	V	$I_{OUT}$ = 2 mA, $V_{PU}$ = $V_{DD}$ = 5V			
Short-Circuit Current	I <sub>SC</sub>	_	±1.5	—	mA	V <sub>PU</sub> = V <sub>DD</sub> = 1.6V (Note 4)			
	I <sub>SC</sub>	-	30	—	mA	V <sub>PU</sub> = V <sub>DD</sub> = 5.5V (Note 4)			
Output Pin Capacitance	C <sub>OUT</sub>	_	8	_	pF				

**Note 1:** The input offset voltage is the center of the input-referred trip points. The input hysteresis is the difference between the input-referred trip points.

2:  $V_{HYST}$  at differential temperatures is estimated using:  $V_{HYST}$  (T<sub>A</sub>) =  $V_{HYST}$  + (T<sub>A</sub> - 25°C) TC<sub>1</sub> + (T<sub>A</sub> - 25°C)<sup>2</sup>TC<sub>2</sub>.

3: Input bias current at temperature is not tested for the SC-70-5 package

4: Do not short the output above V<sub>SS</sub> + 10V. Limit the output current to Absolute Maximum Rating of 30 mA. The minimum V<sub>PU</sub> test limit was V<sub>DD</sub> before Dec. 2004 (week code 52).

# **AC CHARACTERISTICS**

**Electrical Specifications:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = 25°C,  $V_{IN}$ + =  $V_{DD}/2$ , Step = 200 mV, Overdrive = 100 mV,  $R_{PU}$  = 2.74 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 36 pF (Refer to Figure 1-2 and Figure 1-3).

Parameters	Sym	Min	Тур	Max	Units	Conditions
Fall Time	t <sub>F</sub>	_	0.7	—	μs	(Note 1)
Propagation Delay (High-to-Low)	t <sub>PHL</sub>		4.0	8.0	μs	
Propagation Delay (Low-to-High)	t <sub>PLH</sub>		3.0	8.0	μs	(Note 1)
Propagation Delay Skew	t <sub>PDS</sub>		-1.0	_	μs	(Notes 1 and 2)
Maximum Toggle Frequency	f <sub>MAX</sub>		225	_	kHz	V <sub>DD</sub> = 1.6V
	f <sub>MAX</sub>		165		kHz	V <sub>DD</sub> = 5.5V
Input Noise Voltage	E <sub>ni</sub>		200	_	μV <sub>P-P</sub>	10 Hz to 100 kHz

Note 1: t<sub>R</sub> and t<sub>PLH</sub> depend on the load (R<sub>L</sub> and C<sub>L</sub>); these specifications are valid for the indicated load only.

**2:** Propagation Delay Skew is defined as:  $t_{PDS} = t_{PLH} - t_{PHL}$ .

# MCP6548 CHIP SELECT (CS) CHARACTERISTICS

<b>Electrical Specifications:</b> Unless otherwise indicated, $V_{DD}$ = +1.6V to +5.5V, $V_{SS}$ = GND, $T_A$ = 25°C, $V_{IN}$ + = $V_{DD}/2$ , $V_{IN}$ - = $V_{SS}$ , $R_{PU}$ = 2.74 k $\Omega$ to $V_{PU}$ = $V_{DD}$ , and $C_L$ = 36 pF (Refer to Figures 1-1 and 1-3).									
Parameters	Sym	Min	Тур	Max	Units	Conditions			
CS Low Specifications									
CS Logic Threshold, Low	V <sub>IL</sub>	$V_{SS}$	—	$0.2 V_{DD}$	V				
CS Input Current, Low	I <sub>CSL</sub>	_	5	—	pА	CS = V <sub>SS</sub>			
CS High Specifications									
CS Logic Threshold, High	V <sub>IH</sub>	$0.8  V_{DD}$	_	V <sub>DD</sub>	V				
CS Input Current, High	I <sub>CSH</sub>	—	1	—	pА	CS = V <sub>DD</sub>			
CS Input High, V <sub>DD</sub> Current	I <sub>DD</sub>	—	18	—	pА	CS = V <sub>DD</sub>			
CS Input High, GND Current	I <sub>SS</sub>	_	-20	—	pА	CS = V <sub>DD</sub>			
Comparator Output Leakage	I <sub>O(LEAK)</sub>	_	1	—	pА	$V_{OUT} = V_{SS}$ +10V, CS = $V_{DD}$			
CS Dynamic Specifications									
CS Low to Comparator Output Low Turn-on Time	t <sub>ON</sub>		2	50	ms	$\overline{\text{CS}}$ = 0.2V <sub>DD</sub> to V <sub>OUT</sub> = V <sub>DD</sub> /2, V <sub>IN</sub> - = V <sub>DD</sub>			
CS High to Comparator Output High Z Turn-off Time	t <sub>OFF</sub>	—	10	—	μs	$\overline{\text{CS}}$ = 0.8V <sub>DD</sub> to V <sub>OUT</sub> = V <sub>DD</sub> /2, V <sub>IN</sub> - = V <sub>DD</sub>			
CS Hysteresis	V <sub>CS_HYST</sub>	—	0.6	_	V	V <sub>DD</sub> = 5V			





FIGURE 1-2: Propagation Delay Timing Diagram.

**FIGURE 1-1:** Timing Diagram for the  $\overline{CS}$  pin on the MCP6548.

# **TEMPERATURE CHARACTERISTICS**

<b>Electrical Specifications:</b> Unless otherwise indicated, $V_{DD}$ = +1.6V to +5.5V and $V_{SS}$ = GND.									
Parameters	Sym	Min	Тур	Мах	Units	Conditions			
Temperature Ranges									
Specified Temperature Range	Τ <sub>Α</sub>	-40	_	+85	°C				
Operating Temperature Range	Τ <sub>Α</sub>	-40	_	+125	°C	Note			
Storage Temperature Range	Τ <sub>Α</sub>	-65	_	+150	°C				
Thermal Package Resistances									
Thermal Resistance, 5L-SC-70	$\theta_{JA}$	—	331	—	°C/W				
Thermal Resistance, 5L-SOT-23	$\theta_{JA}$	_	256	_	°C/W				
Thermal Resistance, 8L-PDIP	$\theta_{JA}$	_	85	_	°C/W				
Thermal Resistance, 8L-SOIC	$\theta_{JA}$	—	163	_	°C/W				
Thermal Resistance, 8L-MSOP	$\theta_{JA}$	_	206	_	°C/W				
Thermal Resistance, 14L-PDIP	$\theta_{JA}$	_	70	_	°C/W				
Thermal Resistance, 14L-SOIC	$\theta_{JA}$	—	120	—	°C/W				
Thermal Resistance, 14L-TSSOP	$\theta_{JA}$	_	100	_	°C/W				

Note: The MCP6546/7/8/9 I-temp family operates over this extended temperature range, but with reduced performance. In any case, the Junction Temperature  $(T_J)$  must not exceed the absolute maximum specification of +150°C.

# 1.1 Test Circuit Configuration

This test circuit configuration is used to determine the AC and DC specifications.



**FIGURE 1-3:** AC and DC Test Circuit for the Open-Drain Output Comparators.

# 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore outside the warranted range.



**FIGURE 2-1:** Input Offset Voltage at  $V_{CM} = V_{SS}$ .



**FIGURE 2-2:** Input Offset Voltage Drift at  $V_{CM} = V_{SS}$ .



FIGURE 2-3: The MCP6546/6R/6U/7/8/9 comparators show no phase reversal.



FIGURE 2-4: V<sub>CM</sub> = V<sub>SS</sub>.

Input Hysteresis Voltage at



**FIGURE 2-5:** Input Hysteresis Voltage Linear Temp. Co.  $(TC_1)$  at  $V_{CM} = V_{SS}$ .



**FIGURE 2-6:** Input Hysteresis Voltage Quadratic Temp. Co.  $(TC_2)$  at  $V_{CM} = V_{SS}$ .



**FIGURE 2-7:** Input Offset Voltage vs. Ambient Temperature at  $V_{CM} = V_{SS}$ .



**FIGURE 2-8:** Input Offset Voltage vs. Common Mode Input Voltage at  $V_{DD} = 1.6V$ .



**FIGURE 2-9:** Input Offset Voltage vs. Common Mode Input Voltage at  $V_{DD}$  = 5.5V.



**FIGURE 2-10:** Input Hysteresis Voltage vs. Ambient Temperature at  $V_{CM} = V_{SS}$ .



**FIGURE 2-11:** Input Hysteresis Voltage vs. Common Mode Input Voltage at  $V_{DD} = 1.6V$ .



**FIGURE 2-12:** Input Hysteresis Voltage vs. Common Mode Input Voltage at  $V_{DD} = 5.5V$ .

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}/2$ ,  $V_{IN}$ - = GND,  $R_{PU}$  = 2.74 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 36 pF.



FIGURE 2-13: CMRR,PSRR vs. Ambient Temperature.



FIGURE 2-14: Input Bias Current, Input Offset Current vs. Ambient Temperature.



**FIGURE 2-15:** Quiescent Current vs. Common Mode Input Voltage at  $V_{DD} = 1.6V$ .



FIGURE 2-16: Input Bias Current, Input Offset Current vs. Common Mode Input Voltage.



FIGURE 2-17: Quiescent Current vs. Power Supply Voltage.



**FIGURE 2-18:** Quiescent Current vs. Common Mode Input Voltage at  $V_{DD}$  = 5.5V.



FIGURE 2-19: Supply Current vs. Pull-Up Voltage.



FIGURE 2-20: Supply Current vs. Toggle Frequency.



**FIGURE 2-21:** Output Voltage Headroom vs. Output Current at  $V_{DD} = 1.6V$ .



FIGURE 2-22: Supply Current vs. Pull-Up to Supply Voltage Difference.



FIGURE 2-23: Output Short Circuit Current Magnitude vs. Power Supply Voltage.



**FIGURE 2-24:** Output Voltage Headroom vs. Output Current at  $V_{DD} = 5.5V$ .

**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}/2$ ,  $V_{IN}$ - = GND,  $R_{PU}$  = 2.74 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 36 pF.



FIGURE 2-25: High-to-Low Propagation Delay.



FIGURE 2-26: Propag

Propagation Delay Skew.



**FIGURE 2-27:** Propagation Delay vs. Power Supply Voltage.



FIGURE 2-28: Low-to-High Propagation Delay.



**FIGURE 2-29:** Propagation Delay vs. Ambient Temperature.



FIGURE 2-30: Propagation Delay vs. Input Overdrive.



FIGURE 2-31:Propagation Delay vs.Common Mode Input Voltage at  $V_{DD} = 1.6V.$ 



*FIGURE 2-32:* Propagation Delay vs. Pull-up Resistor.



FIGURE 2-33: Pull-up Voltage.



FIGURE 2-34:Propagation Delay vs.Common Mode Input Voltage at  $V_{DD}$  = 5.5V.



*FIGURE 2-35:* Propagation Delay vs. Load Capacitance.



FIGURE 2-36:Output Leakage Current $\overline{(CS} = V_{DD})$  vs. Output Voltage (MCP6548 only).



**Note:** Unless otherwise indicated,  $V_{DD}$  = +1.6V to +5.5V,  $V_{SS}$  = GND,  $T_A$  = +25°C,  $V_{IN}$ + =  $V_{DD}/2$ ,  $V_{IN}$ - = GND,  $R_{PU}$  = 2.74 k $\Omega$  to  $V_{PU}$  =  $V_{DD}$ , and  $C_L$  = 36 pF.

**FIGURE 2-37:** Supply Current (shoot through current) vs. Chip Select ( $\overline{CS}$ ) Voltage at  $V_{DD} = 1.6V$  (MCP6548 only).



**FIGURE 2-38:** Supply Current (charging current) vs. Chip Select (CS) pulse at  $V_{DD}$  = 1.6V (MCP6548 only).



**FIGURE 2-39:** Chip Select  $\overline{(CS)}$  Step Response (MCP6548 only).



**FIGURE 2-40:** Supply Current (shoot through current) vs. Chip Select ( $\overline{CS}$ ) Voltage at  $V_{DD} = 5.5V$  (MCP6548 only).



**FIGURE 2-41:** Supply Current (charging current) vs. Chip Select (CS) pulse at  $V_{DD}$  = 5.5V (MCP6548 only).



FIGURE 2-42: Input Bias Current vs. Input Voltage.

# 3.0 PIN DESCRIPTIONS

Descriptions of the pins are listed in Table 3-1.

#### TABLE 3-1: PIN FUNCTION TABLE

MCP6546 (PDIP, SOIC, MSOP)	MCP6546 (SOT-23-5, SC-70-5)	MCP6546R	MCP6546U	MCP6547	MCP6548	MCP6549	Symbol	Description
6	1	1	4	1	6	1	OUT, OUTA	Digital Output (comparator A)
2	4	4	1	2	2	2	V <sub>IN</sub> –, V <sub>INA</sub> –	Inverting Input (comparator A)
3	3	3	3	3	3	3	V <sub>IN</sub> +, V <sub>INA</sub> +	Non-inverting Input (comparator A)
7	5	2	5	8	7	4	V <sub>DD</sub>	Positive Power Supply
—	—	_	—	5	—	5	V <sub>INB</sub> +	Non-inverting Input (comparator B)
—	—	_		6	—	6	V <sub>INB</sub> –	Inverting Input (comparator B)
—	—	-	_	7	—	7	OUTB	Digital Output (comparator B)
—	—	_	—		—	8	OUTC	Digital Output (comparator C)
—	—		—		—	9	V <sub>INC</sub> -	Inverting Input (comparator C)
_	—		—		—	10	V <sub>INC</sub> +	Non-inverting Input (comparator C)
4	2	5	2	4	4	11	V <sub>SS</sub>	Negative Power Supply
—	—	_			—	12	V <sub>IND</sub> +	Non-inverting Input (comparator D)
—	—	-	_		—	13	V <sub>IND</sub> -	Inverting Input (comparator D)
—	—	_	—	—	—	14	OUTD	Digital Output (comparator D)
—	_	_	—		8	_	CS	Chip Select
1, 5, 8	_	_			1, 5		NC	No Internal Connection

#### 3.1 Analog Inputs

The comparator non-inverting and inverting inputs are high-impedance CMOS inputs with low bias currents.

# 3.2 CS Digital Input

This is a CMOS, Schmitt-triggered input that places the part into a low power mode of operation.

#### 3.3 Digital Outputs

The comparator outputs are CMOS, open-drain digital outputs. They are designed to make level shifting and wired-OR easy to implement.

# 3.4 Power Supply (V<sub>SS</sub> and V<sub>DD</sub>)

The positive power supply pin (V<sub>DD</sub>) is 1.6V to 5.5V higher than the negative power supply pin (V<sub>SS</sub>). For normal operation, the other pins are at voltages between V<sub>SS</sub> and V<sub>DD</sub>, except the output pins which can be as high as 10V above V<sub>SS</sub>.

Typically, these parts are used in a single (positive) supply configuration. In this case,  $V_{SS}$  is connected to ground and  $V_{DD}$  is connected to the supply.  $V_{DD}$  will need a local bypass capacitor (typically 0.01 µF to 0.1 µF) within 2 mm of the  $V_{DD}$  pin. These can share a bulk capacitor with nearby analog parts (within 100 mm), but it is not required.

# 4.0 APPLICATIONS INFORMATION

The MCP6546/7/8/9 family of push-pull output comparators are fabricated on Microchip's state-of-theart CMOS process. They are suitable for a wide range of applications requiring very low power consumption.

#### 4.1 Comparator Inputs

#### 4.1.1 PHASE REVERSAL

The MCP6546/6R/6U/7/8/9 comparator family uses CMOS transistors at the input. They are designed to prevent phase inversion when the input pins exceed the supply voltages. Figure 2-3 shows an input voltage exceeding both supplies with no resulting phase inversion.

#### 4.1.2 INPUT VOLTAGE AND CURRENT LIMITS

The ESD protection on the inputs can be depicted as shown in Figure 4-1. This structure was chosen to protect the input transistors, and to minimize input bias current (IB). The input ESD diodes clamp the inputs when they try to go more than one diode drop below  $V_{SS}$ . They also clamp any voltages that go too far above  $V_{DD}$ ; their breakdown voltage is high enough to allow normal operation, and low enough to bypass ESD events within the specified limits.



FIGURE 4-1: Simplified Analog Input ESD Structures.

In order to prevent damage and/or improper operation of these amplifiers, the circuits they are in must limit the currents (and voltages) at the V<sub>IN</sub>+ and V<sub>IN</sub>- pins (see **Absolute Maximum Ratings †** at the beginning of **Section 1.0 "Electrical Characteristics"**). Figure 4-3 shows the recommended approach to protecting these inputs. The internal ESD diodes prevent the input pins (V<sub>IN</sub>+ and V<sub>IN</sub>-) from going too far below ground, and the resistors R<sub>1</sub> and R<sub>2</sub> limit the possible current drawn out of the input pin. Diodes D<sub>1</sub> and D<sub>2</sub> prevent the input pin (V<sub>IN</sub>+ and V<sub>IN</sub>-) from going too far above V<sub>DD</sub>. When implemented as shown, resistors R<sub>1</sub> and R<sub>2</sub> also limit the current through D<sub>1</sub> and D<sub>2</sub>.



FIGURE 4-2: Protecting the Analog Inputs.

It is also possible to connect the diodes to the left of the resistors R<sub>1</sub> and R<sub>2</sub>. In this case, the currents through the diodes D<sub>1</sub> and D<sub>2</sub> need to be limited by some other mechanism. The resistor then serves as in-rush current limiter; the DC current into the input pins (V<sub>IN</sub>+ and V<sub>IN</sub>-) should be very small.

A significant amount of current can flow out of the inputs when the common mode voltage ( $V_{CM}$ ) is below ground ( $V_{SS}$ ); see Figure 2-42. Applications that are high impedance may need to limit the useable voltage range.

#### 4.1.3 NORMAL OPERATION

The input stage of this family of devices uses two differential input stages in parallel: one operates at low input voltages and the other at high input voltages. With this topology, the input voltage is 0.3V above V<sub>DD</sub> and 0.3V below V<sub>SS</sub>. The input offset voltage is measured at both V<sub>SS</sub> - 0.3V and V<sub>DD</sub> + 0.3V to ensure proper operation.

The MCP6546/6R/6U/7/8/9 family has internally-set hysteresis that is small enough to maintain input offset accuracy (<7 mV), and large enough to eliminate output chattering caused by the comparator's own input noise voltage (200  $\mu$ V<sub>P-P</sub>). Figure 4-3 illustrates this capability.



**FIGURE 4-3:** The MCP6546/7/8/9 comparators' internal hysteresis eliminates output chatter caused by input noise voltage.

# 4.2 Open-Drain Output

The open-drain output is designed to make levelshifting and wired-OR logic easy to implement. The output can go as high as 10V for 9V battery-powered applications. The output stage minimizes switching current (shoot-through current from supply-to-supply) when the output changes state. See Figures 2-15, 2-18 and 2-37 through 2-41, for more information.

# 4.3 MCP6548 Chip Select (CS)

The MCP6548 is a single comparator with a Chip Select (CS) pin. When CS is pulled high, the total current consumption drops to 20 pA (typ.). 1 pA (typ.) flows through the CS pin, 1 pA (typ.) flows through the V<sub>DD</sub> pin, as shown in Figure 1-1. When this happens, the comparator output is put into a high-impedance state. By pulling CS low, the comparator is enabled. If the CS pin is left floating, the comparator will not operate properly. Figure 1-1 shows the output voltage and supply current response to a CS pulse.

The internal  $\overline{CS}$  circuitry is designed to minimize glitches when cycling the  $\overline{CS}$  pin. This helps conserve power, which is especially important in battery-powered applications.

# 4.4 Externally Set Hysteresis

Greater flexibility in selecting hysteresis, or input trip points, is achieved by using external resistors.

Input offset voltage ( $V_{OS}$ ) is the center (average) of the (input-referred) low-high and high-low trip points. Input hysteresis voltage ( $V_{HYST}$ ) is the difference between the same trip points. Hysteresis reduces output chattering when one input is slowly moving past the other, thus reducing dynamic supply current. It also helps in systems where it is best not to cycle between states too frequently (e.g., air conditioner thermostatic control).

### 4.4.1 INVERTING CIRCUIT

Figure 4-4 shows an inverting circuit for a single-supply application using three resistors, besides the pull-up resistor. The resulting hysteresis diagram is shown in Figure 4-5.



*FIGURE 4-4:* Inverting circuit with hysteresis.



*FIGURE 4-5:* Hysteresis diagram for the inverting circuit.

In order to determine the trip voltages (V<sub>THL</sub> and V<sub>TLH</sub>) for the circuit shown in Figure 4-4, R<sub>2</sub> and R<sub>3</sub> can be simplified to the Thevenin equivalent circuit with respect to V<sub>DD</sub>, as shown in Figure 4-6.



FIGURE 4-6: Thevenin Equivalent Circuit.

#### **EQUATION 4-1:**

$$R_{23} = \frac{R_2 R_3}{R_2 + R_3}$$
$$V_{23} = \frac{R_3}{R_2 + R_3} \times V_{DD}$$

Using this simplified circuit, the trip voltage can be calculated using the following equation:

#### **EQUATION 4-2:**

$$V_{THL} = V_{PU} \left( \frac{R_{23}}{R_{23} + R_F + R_{PU}} \right) + V_{23} \left( \frac{R_F + R_{PU}}{R_{23} + R_F + R_{PU}} \right)$$
$$V_{TLH} = V_{OL} \left( \frac{R_{23}}{R_{23} + R_F} \right) + V_{23} \left( \frac{R_F}{R_{23} + R_F} \right)$$
$$V_{TLH} = \text{trip voltage from low to high}$$
$$V_{THL} = \text{trip voltage from high to low}$$

Figure 2-21 and Figure 2-24 can be used to determine typical values for  $V_{OL}$ . This voltage is dependent on the output current  $I_{OL}$  as shown in Figure 4-4. This current can be determined using the equation below:

#### **EQUATION 4-3:**

$$\begin{split} I_{OL} &= I_{PU} + I_{RF} \\ I_{OL} &= \left(\frac{V_{PU} - V_{OL}}{R_{PU}}\right) + \left(\frac{V_{23} - V_{OL}}{R_{23} + R_{F}}\right) \end{split}$$

V<sub>OH</sub> can be calculated using the equation below:

#### **EQUATION 4-4:**

$$V_{OH} = (V_{PU} - V_{23}) \times \left(\frac{R_{23} + R_F}{R_{23} + R_F + R_{PU}}\right)$$

As explained in **Section 4.1 "Comparator Inputs"**, it is important to keep the non-inverting input below  $V_{DD}$ +0.3V when  $V_{PU}$  >  $V_{DD}$ .

#### 4.5 Supply Bypass

With this family of comparators, the power supply pin (V<sub>DD</sub> for single supply) should have a local bypass capacitor (i.e., 0.01  $\mu$ F to 0.1  $\mu$ F) within 2 mm for good edge rate performance.

#### 4.6 Capacitive Loads

Reasonable capacitive loads (e.g., logic gates) have little impact on propagation delay (see Figure 2-27). The supply current increases with increasing toggle frequency (Figure 2-30), especially with higher capacitive loads.

#### 4.7 Battery Life

In order to maximize battery life in portable applications, use large resistors and small capacitive loads. Avoid toggling the <u>output</u> more than necessary. Do not use Chip Select (CS) too frequently in order to conserve power. Capacitive loads will draw additional power at start-up.

#### 4.8 PCB Surface Leakage

In applications where low input bias current is critical, PCB (Printed Circuit Board) surface leakage effects need to be considered. Surface leakage is caused by humidity, dust or other contamination on the board. Under low-humidity conditions, a typical resistance between nearby traces is  $10^{12}\Omega$ . A 5V difference would cause 5 pA of current to flow. This is greater than the MCP6546/6R/6U/7/8/9 family's bias current at 25°C (1 pA, typ.).

The easiest way to reduce surface leakage is to use a guard ring around sensitive pins (or traces). The guard ring is biased at the same voltage as the sensitive pin. An example of this type of layout is shown in Figure 4-7.



**FIGURE 4-7:** Example Guard Ring Layout for Inverting Circuit.

- 1. Inverting Configuration (Figures 4-4 and 4-7):
  - a. Connect the guard ring to the non-inverting input pin (V<sub>IN</sub>+). This biases the guard ring to the same reference voltage as the comparator (e.g.,  $V_{DD}/2$  or ground).
  - b. Connect the inverting pin (V\_{IN}-) to the input pad without touching the guard ring.

#### 4.9 Unused Comparators

An unused amplifier in a quad package (MCP6549) should be configured as shown in Figure 4-8. This circuit prevents the output from toggling and causing crosstalk. It uses the minimum number of components and draws minimal current (see Figure 2-15 and Figure 2-18).





Unused Comparators.

#### 4.10 Typical Applications

#### 4.10.1 PRECISE COMPARATOR

Some applications require higher DC precision. An easy way to solve this problem is to use an amplifier (such as the MCP6041) to gain-up the input signal before it reaches the comparator. Figure 4-9 shows an example of this approach.



FIGURE 4-9: Precise Inverting Comparator.

#### 4.10.2 WINDOWED COMPARATOR

Figure 4-10 shows one approach to designing a windowed comparator. The wired-OR connection produces a high output (logic 1) when the input voltage is between V<sub>RB</sub> and V<sub>RT</sub> (where V<sub>RT</sub> > V<sub>RB</sub>).





Example: (I-temp)

# 5.0 PACKAGING INFORMATION









# Package Marking Information (Continued)

14-Lead PDIP (300 mil) (MCP6549)





## 5-Lead Plastic Package (LT) (SC-70)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging







	Units		INCHES		MILLIMETERS*		
Dimension Limits	3	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		5			5	
Pitch	р	.0:	26 (BSC)		0.6	65 (BSC)	
Overall Height	А	.031		.043	0.80		1.10
Molded Package Thickness	A2	.031		.039	0.80		1.00
Standoff	A1	.000		.004	0.00		0.10
Overall Width	Е	.071		.094	1.80		2.40
Molded Package Width	E1	.045		.053	1.15		1.35
Overall Length	D	.071		.087	1.80		2.20
Foot Length	L	.004		.012	0.10		0.30
Top of Molded Pkg to Lead Shoulder	Q1	.004		.016	0.10		0.40
Lead Thickness	С	.004		.007	0.10		0.18
Lead Width	В	.006		.012	0.15		0.30

\* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

JEITA (EIAJ) Standard: SC-70 Drawing No. C04-061

Revised 07-19-05

## 5-Lead Plastic Small Outline Transistor (OT) (SOT23)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging









		INCHES*		MILLIMETERS			
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		6			6	
Pitch	р		038 BSC		(	).95 BSC	
Outside lead pitch	p1		075 BSC			1.90 BSC	
Overall Height	А	.035	.046	.057	0.90	1.18	1.45
Molded Package Thickness	A2	.035	.043	.051	0.90	1.10	1.30
Standoff	A1	.000	.003	.006	0.00	0.08	0.15
Overall Width	Е	.102	.110	.118	2.60	2.80	3.00
Molded Package Width	E1	.059	.064	.069	1.50	1.63	1.75
Overall Length	D	.110	.116	.122	2.80	2.95	3.10
Foot Length	L	.014	.018	.022	0.35	0.45	0.55
Foot Angle	¢	0	5	10	0	5	10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.014	.017	.020	0.35	0.43	0.50
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

\* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

JEITA (formerly EIAJ) equivalent: SC-74A Drawing No. C04-120

Revised 09-12-05

#### 8-Lead Plastic Dual In-line (P) – 300 mil Body (PDIP)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging





	Units		INCHES*		N	IILLIMETERS	3
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р		.100			2.54	
Top to Seating Plane	Α	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.360	.373	.385	9.14	9.46	9.78
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing §	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

\* Controlling Parameter § Significant Characteristic

Significant Char Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-001

Drawing No. C04-018

#### 8-Lead Plastic Small Outline (SN) – Narrow, 150 mil (SOIC)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging









	Units		INCHES*			MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n		8			8		
Pitch	р		.050			1.27		
Overall Height	Α	.053	.061	.069	1.35	1.55	1.75	
Molded Package Thickness	A2	.052	.056	.061	1.32	1.42	1.55	
Standoff §	A1	.004	.007	.010	0.10	0.18	0.25	
Overall Width	E	.228	.237	.244	5.79	6.02	6.20	
Molded Package Width	E1	.146	.154	.157	3.71	3.91	3.99	
Overall Length	D	.189	.193	.197	4.80	4.90	5.00	
Chamfer Distance	h	.010	.015	.020	0.25	0.38	0.51	
Foot Length	L	.019	.025	.030	0.48	0.62	0.76	
Foot Angle	φ	0	4	8	0	4	8	
Lead Thickness	С	.008	.009	.010	0.20	0.23	0.25	
Lead Width	В	.013	.017	.020	0.33	0.42	0.51	
Mold Draft Angle Top	α	0	12	15	0	12	15	
Mold Draft Angle Bottom	β	0	12	15	0	12	15	

\* Controlling Parameter

§ Significant Characteristic Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012

Drawing No. C04-057

#### 8-Lead Plastic Micro Small Outline Package (MS) (MSOP)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging









	Units	INCHES		MILLIMETERS*			
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		8			8	
Pitch	р	.026 BSC		0.65 BSC			
Overall Height	А	-	-	.043	-	-	1.10
Molded Package Thickness	A2	.030	.033	.037	0.75	0.85	0.95
Standoff	A1	.000	-	.006	0.00	-	0.15
Overall Width	Е	.193 BSC			4.90 BSC		
Molded Package Width	E1	.118 BSC		3.00 BSC			
Overall Length	D	.118 BSC		3.00 BSC			
Foot Length	L	.016	.024	.031	0.40	0.60	0.80
Footprint (Reference)	F	.037 REF		0.95 REF			
Foot Angle	¢	0°	-	8°	0°	-	8°
Lead Thickness	С	.003	.006	.009	0.08	-	0.23
Lead Width	В	.009	.012	.016	0.22	-	0.40
Mold Draft Angle Top	α	5°	-	15°	5°	-	15°
Mold Draft Angle Bottom	β	5°	-	15°	5°	-	15°

\* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

REF: Reference Dimension, usually without tolerance, for information purposes only.

See ASME Y14.5M

JEDEC Equivalent: MO-187

Drawing No. C04-111

Revised 07-21-05

#### 14-Lead Plastic Dual In-line (P) – 300 mil Body (PDIP)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



	Units	INCHES*		MILLIMETERS			
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		14			14	
Pitch	р		.100			2.54	
Top to Seating Plane	Α	.140	.155	.170	3.56	3.94	4.32
Molded Package Thickness	A2	.115	.130	.145	2.92	3.30	3.68
Base to Seating Plane	A1	.015			0.38		
Shoulder to Shoulder Width	E	.300	.313	.325	7.62	7.94	8.26
Molded Package Width	E1	.240	.250	.260	6.10	6.35	6.60
Overall Length	D	.740	.750	.760	18.80	19.05	19.30
Tip to Seating Plane	L	.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С	.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B1	.045	.058	.070	1.14	1.46	1.78
Lower Lead Width	В	.014	.018	.022	0.36	0.46	0.56
Overall Row Spacing §	eB	.310	.370	.430	7.87	9.40	10.92
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

\* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-001

Drawing No. C04-005

#### 14-Lead Plastic Small Outline (SL) – Narrow, 150 mil (SOIC)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging



Dimension Limits MIN NOM MAX MIN NOM   Number of Pins n 14 14 14	MAX
Number of Pins n 14 14	7
	7
Pitch P .050 1.2	5 175
Overall Height A .053 .061 .069 1.35 1.5	1.75
Molded Package Thickness A2 .052 .056 .061 1.32 1.4	2 1.55
Standoff § A1 .004 .007 .010 0.10 0.1	8 0.25
Overall Width E .228 .236 .244 5.79 5.9	9 6.20
Molded Package Width E1 .150 .154 .157 3.81 3.9	0 3.99
Overall Length D .337 .342 .347 8.56 8.6	9 8.81
Chamfer Distance h .010 .015 .020 0.25 0.3	8 0.51
Foot Length L .016 .033 .050 0.41 0.8	4 1.27
Foot Angle φ 0 4 8 0	4 8
Lead Thickness C .008 .009 .010 0.20 0.2	3 0.25
Lead Width B .014 .017 .020 0.36 0.4	2 0.51
Mold Draft Angle Top α 0 12 15 0	2 15
Mold Draft Angle Bottom β 0 12 15 0	2 15

\* Controlling Parameter

§ Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-012

Drawing No. C04-065

#### 14-Lead Plastic Thin Shrink Small Outline (ST) – 4.4 mm (TSSOP)

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at http://www.microchip.com/packaging









	Units	INCHES		MILLIMETERS*				
Dimension Limits		MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n	14			14			
Pitch	р	.026 BSC			0.65 BSC			
Overall Height	А	.039	.041	.043	1.00	1.05	1.10	
Molded Package Thickness	A2	.033	.035	.037	0.85	0.90	0.95	
Standoff	A1	.002	.004	.006	0.05	0.10	0.15	
Overall Width	Е	.246	.251	.256	6.25	6.38	6.50	
Molded Package Width	E1	.169	.173	.177	4.30	4.40	4.50	
Molded Package Length	D	.193	.197	.201	4.90	5.00	5.10	
Foot Length	L	.020	.024	.028	0.50	0.60	0.70	
Foot Angle	ф	0°	4°	8°	0°	4°	8°	
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20	
Lead Width	В	.007	.010	.012	0.19	0.25	0.30	
Mold Draft Angle Top	α	12° REF		12° REF				
Mold Draft Angle Bottom	β	12° REF		12° REF				

\* Controlling Parameter

Notes:

Dimensions D and E1 do not include mold fla sh or protrusions. Mold flash or protrusions shall not exceed .005" (0.127mm) per side. BSC: Basic Dimension. Theoretically exact value shown without tolerances.

See ASME Y14.5M

REF: Reference Dimension, usually without tole rance, for information purposes only.

See ASME Y14.5M

JEDEC Equivalent: MO-153 AB-1

Drawing No. C04-087

Revised: 08-17-05

# APPENDIX A: REVISION HISTORY

#### **Revision E (September 2006)**

The following is the list of modifications:

- 1. Added MCP6546U pinout for the SOT-23-5 package.
- 2. Clarified Absolute Maximum Analog Input Voltage and Current Specifications.
- 3. Added applications writeups on unused comparators.
- 4. Added disclaimer to package outline drawings.

# Revision D (May 2006)

The following is the list of modifications:

- 1. Added E-temp parts.
- Changed minimum pull-up voltage specification (V<sub>PU</sub>) to 1.6V for parts starting Dec. 2004 (week code 52); previous parts are specified at a minimum of V<sub>DD</sub>.
- Changed V<sub>HYST</sub> temperature specifications to linear and quadratic temperature coefficients.
- 4. Changed specifications and plots to include E-Temp parts.
- 5. Added Section 3.0 "Pin Descriptions".
- 6. Corrected package markings (Section 5.1 "Package Marking Information").
- 7. Added Appendix A: "Revision History".

#### Revision C (May 2003)

#### **Revision B (December 2002)**

#### Revision A (February 2002)

• Original Release of this Document.

NOTES:

# **PRODUCT IDENTIFICATION SYSTEM**

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

PART NO.	<u>-X /XX</u>	Exa	Examples:				
 Device Temp Ra	│ │ │ erature Package nge	a)	MCP6546T-I/LT:	Tape and Reel, Industrial Temperature, 5LD SC-70.			
Device:	MCP6546: Single Comparator MCP6546T: Single Comparator (Tape and Reel)	b)	MCP6546T-I/OT:	Tape and Reel, Industrial Temperature, 5LD SOT-23.			
	(SC-70, SOT-23, SOIC, MSOP) MCP6546RT: Single Comparator (Rotated - Tape and	c)	MCP6546-E/P:	Extended Temperature, 8LD PDIP.			
	Reel) (SO 1-23 only) MCP6546UT: Single Comparator (Tape and Reel) (SOT-23-5 is E-Temp only) MCP6547: Dual Comparator	d)	MCP6546RT-I/OT	Tape and Reel, Industrial Temperature, 5LD SOT23.			
	MCP6547T: Dual Comparator (Tape and Reel for SOIC and MSOP)	e)	MCP6546-E/SN:	Extended Temperature, 8LD SOIC.			
	MCP6548: Single Comparator with CS (Tape and Reel for SOIC and MSOP) MCP6549: Quad Comparator MCP6549: Quad Comparator	f)	MCP6546UT-E/O	F:Tape and Reel, Extended Temperature, 5LD SOT23.			
	(Tape and Reel for SOIC and TSSOP)	a)	MCP6547-I/MS:	Industrial Temperature, 8LD MSOP.			
Temperature Range:	I = $-40^{\circ}$ C to $+85^{\circ}$ C E * = $-40^{\circ}$ C to $+125^{\circ}$ C * SC 70 5 E Tomp parts not available at this release of the	b)	MCP6547T-I/MS:	Tape and Reel, Industrial Temperature, 8LD MSOP.			
Dookogo:	data sheet.	C)	MCP6547-I/P:	Industrial Temperature, 8LD PDIP.			
Раскаде:	OT = Plastic Small Outline Transistor (SOT-23), 5-lead MS = Plastic MSOP, 8-lead P = Plastic DIP (300 mil Body), 8-lead, 14-lead	d)	MCP6547-E/SN:	Extended Temperature, 8LD SOIC.			
	SN = Plastic SOIC (150 mil Body), 8-lead SL = Plastic SOIC (150 mil Body), 14-lead (MCP6549)	a)	MCP6548-I/SN:	Industrial Temperature, 8LD SOIC.			
	ST = Plastic TSSOP (4.4mm Body), 14-lead (MCP'6549)	b)	MCP6548T-I/SN:	Tape and Reel, Industrial Temperature, 8LD SOIC.			
		c)	MCP6548-I/P:	Industrial Temperature, 8LD PDIP.			
		d)	MCP6548-E/SN:	Extended Temperature, 8LD SOIC.			
		a)	MCP6549T-I/SL:	Tape and Reel, Industrial Temperature, 14LD SOIC.			
		b)	MCP6549T-E/SL:	Tape and Reel, Extended Temperature, 14LD SOIC.			
		C)	MCP6549-I/P:	Industrial Temperature, 14LD PDIP.			
		d)	MCP6549-E/ST:	Extended Temperature, 14LD TSSOP.			

NOTES:

#### Note the following details of the code protection feature on Microchip devices:

- · Microchip products meet the specification contained in their particular Microchip Data Sheet.
- Microchip believes that its family of products is one of the most secure families of its kind on the market today, when used in the intended manner and under normal conditions.
- There are dishonest and possibly illegal methods used to breach the code protection feature. All of these methods, to our knowledge, require using the Microchip products in a manner outside the operating specifications contained in Microchip's Data Sheets. Most likely, the person doing so is engaged in theft of intellectual property.
- Microchip is willing to work with the customer who is concerned about the integrity of their code.
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