

# Hi-performance Regulator IC Series for PCs

## Switching Regulators

### for DDR-SDRAM Cores


**BD9575FV**

No.09030EBT11

#### ●Description

BD9575FV is a switching regulator controller with high output current which can achieve low output voltage (0.7V to 2.0V) from a wide input voltage range (4.5V to 28V). High efficiency for the switching regulator can be realized by utilizing an external N-MOSFET power transistor. A new technology called H3Reg™ is a Rohm proprietary control method to realize ultra high transient response against load change. SLLM (Simple Light Load Mode) technology is also integrated to improve efficiency in light load mode, providing high efficiency over a wide load range. For protection and ease of use, the soft start function, variable frequency function, short circuit protection function with timer latch, over voltage protection function and REF synchronous function are all built in. This switching regulator is specially designed for the DDR-SDRAM core.

#### ●Features

- 1) H<sup>3</sup>Reg™ Switching Regulator Controller
- 2) Light Load Mode and Continuous Mode Changeable
- 3) Thermal Shut Down (TSD), Under Voltage LockOut (UVLO), Over Current Protection (OCP), Over Voltage Protection (OVP), Short Circuit Protection (SCP) built-in
- 4) Soft start function to minimize rush current during startup
- 5) Switching Frequency Variable (f=200KHz~600KHz)
- 6) SSOP-B20 Package

#### ●Applications

Laptop PC, Desktop PC, LCD-TV, Digital Components

#### ●Maximum Absolute Ratings (Ta=25°C)

Parameter	Symbol	Limit	Unit
Input Voltage 1	VCC	7* <sup>1</sup>	V
Input Voltage 2	VDD	7* <sup>1</sup>	V
Input Voltage 3	VIN	30* <sup>1</sup>	V
BOOT Voltage	BOOT	35* <sup>1</sup>	V
BOOT-SW Voltage	BOOT-SW	7* <sup>1</sup>	V
HG-SW Voltage	HG-SW	7* <sup>1</sup>	V
LG Voltage	LG	VDD	V
REF Voltage	REF	VCC	V
Output Voltage	VOOUT/Is+/Is-	VCC	V
ILIM/SCP/SS/FS/SLLM Voltage	ILIM/SCP/SS/FS/SLLM	VCC	V
VREG Voltage	VREG	VCC	V
EN Input Voltage	EN	7* <sup>1</sup>	V
Power Dissipation 1	Pd1	0.5* <sup>2</sup>	W
Power Dissipation 2	Pd2	0.81* <sup>3</sup>	W
Operating Temperature Range	Topr	-20~+100	°C
Storage Temperature Range	Tstg	-55~+150	°C
Maximum Junction Temperature	Tjmax	+150	°C

\*1 Not to exceed Pd.

\*2 Reduced by 4mW for each increase in Ta of 1°C over 25°C. IC only.

\*3 Reduced by 6.5mW for increase in Ta of 1°C over 25°C. (when mounted on a board 70.0mm × 70mm × 1.6mm Glass-epoxy PCB, 1 layer)

## ● Operating Conditions (Ta=25°C)

Parameter	Symbol	MIN	MAX	Unit
Input Voltage 1	VCC	4.5	5.5	V
Input Voltage 2	VDD	4.5	5.5	V
Input Voltage 3	VIN	4.5	28	V
BOOT Voltage	BOOT	4.5	33	V
SW Voltage	SW	-0.7	28	V
BOOT-SW Voltage	BOOT-SW	4.5	5.5	V
SLLM Input Voltage	SLLM	0	5.5	V
EN Input Voltage	EN	0	5.5	V
Output setting voltage	REF	0.7	2.0	V
Is Input Voltage	Is+/Is-	0.7	2.7	V
MIN ON Time	Tonmin	-	200	nsec

\*This product should not be used in a radioactive environment.

## ● Electrical characteristics

(unless otherwise noted, Ta=25°C VCC=5V, VDD=5V, EN/SLLM=5V, VIN=12V, REF=1.8V, RfS=68kΩ)

Parameter	Symbol	Standard Value			Unit	Condition
		MIN	TYP	MAX		
[Whole Device]						
VCC Bias Current	Icc	-	700	900	μA	
VIN Bias Current	Iin	-	100	200	μA	
VCC Standby Current	Iccstb	-	0	10	μA	EN=0V
VIN Standby Current	Iinstb	-	100	200	μA	EN=0V
EN Low Voltage	Enlow	GND	-	0.8	V	
EN High Voltage	Enhhigh	2.3	-	5.5	V	
EN Bias Current	Ien	-	7	10	μA	
VREG Voltage	Vreg	2.475	2.500	2.525	V	Ireg=100μA Ta=-10°C to 100°C*
[Under Voltage Locked Out]						
VCC threshold voltage	Vcc_UVLO	4.1	4.3	4.5	V	VCC:Sweep up
VCC hysteresis voltage	dVcc_UVLO	100	160	220	mV	VCC:Sweep down
VIN threshold voltage	Vin_UVLO	4.1	4.3	4.5	V	VIN:Sweep up
VIN hysteresis voltage	dVin_UVLO	100	160	220	mV	VIN:Sweep down
VREG threshold voltage	Vreg_UVLO	2.0	2.2	2.4	V	VREG:Sweep up
VREG hysteresis voltage	dVreg_UVLO	100	160	220	mV	VREG:Sweep down
[H <sup>3</sup> Reg™]						
Frequency	Fosc	-	300	-	kHz	
ON Time	Ton	400	500	600	nsec	
MAX ON Time	Tonmax	-	3	-	μsec	
MIN OFF Time	Toffmin	-	450	550	nsec	
[FET Driver]						
HG Higher side ON resistor	HGhon	-	3.0	6.0	Ω	
HG Lower side ON resistor	HGlon	-	2.0	4.0	Ω	
LG Higher side ON resistor	LGhon	-	2.0	4.0	Ω	
LG Lower side ON resistor	LGlon	-	0.5	1.0	Ω	
[Dead Time]						
HG rising	HGdead	-	50	-	nsec	
LG rising	LGdead	-	50	-	nsec	

## ●Electrical characteristics – Continued

(unless otherwise noted, Ta=25°C VCC=5V, VDD=5V, EN/SLLM=5V, VIN=12V, REF=1.8V, RFS=68kΩ)

Parameter	Symbol	Standard Value			Unit	Condition
		MIN	TYP	MAX		
[SCP]						
SCP Detect Voltage	Vscp	REF × 0.65	REF × 0.7	REF × 0.75	V	
SCP threshold voltage	Vscpth	1.2	1.25	1.3	V	
Charge current	Iscp	1.5	2	2.5	μA	
Standby voltage	Vscp_stb	-	-	50	mV	
[OVP]						
OVP Detect Voltage	VOVP	REF × 1.16	REF × 1.2	REF × 1.24	V	
[Soft start]						
Charge current	Iss	1.5	2	2.5	μA	
Standby voltage	Vss_stb	-	-	50	mV	
[Over Current Protection Block]						
Current limit threshold1	Ilim1	43	50	57	mV	ILIM=0.5V Ta=-20°C to 100°C*
Current limit threshold2	Ilim2	160	200	240	mV	ILIM=2.0V
Reverse current limit threshold1	Rellim1	-	-50	-	mV	ILIM=0.5V
Reverse current limit threshold2	Rellim2	-	-200	-	mV	ILIM=2.0V
[VOOUT setting ]						
VOOUT offset voltage1	Voutoff1	REF-7m	REF	REF+7m	V	Ta=-20°C to 100°C*
VOOUT bias current	Ivout	-100	0	100	nA	
REF bias current	Iref	-100	0	100	nA	
Is+ Input current	IIs+	-1	0	1	μA	Is+=1.8V
Is- Input current	IIs-	-1	0	1	μA	Is-=1.8V
[SLLM ]						
Continuous mode threshold	Vthcon	VCC-0.5	-	VCC	V	
SLLM threshold	VthSLLM	GND	-	0.5	V	

\* Design Guarantee

●Reference Data

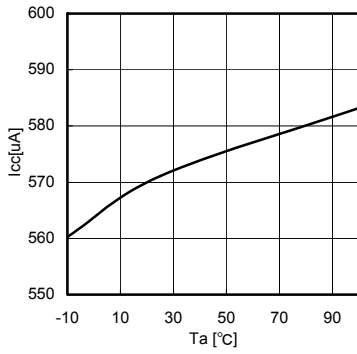


Fig.1 Ta vs Icc

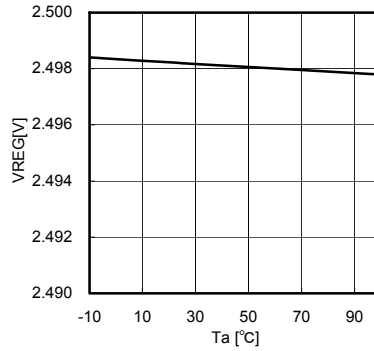


Fig.2 Ta vs VREG

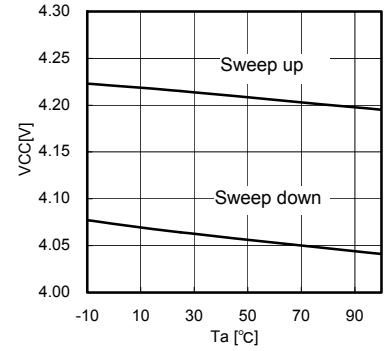


Fig.3 Ta vs UVLO (Vcc)

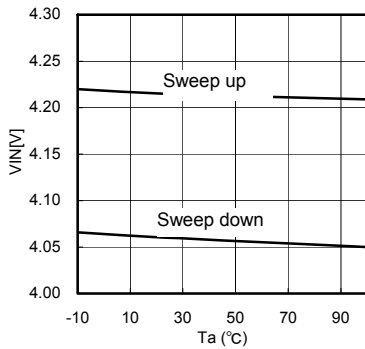


Fig.4 Ta vs UVLO (VIN)

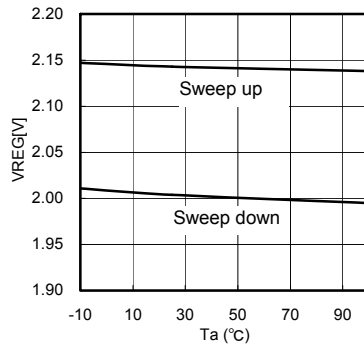


Fig.5 Ta vs UVLO (VREG)

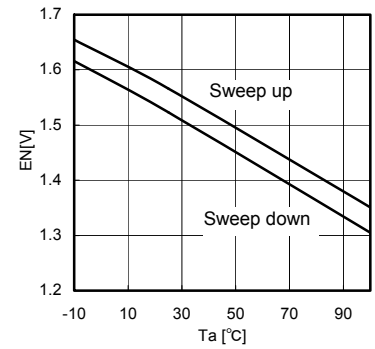


Fig.6 Ta vs EN Threshold

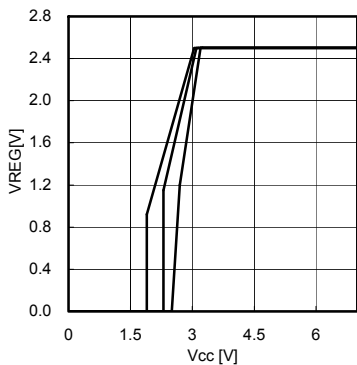


Fig.7 Vcc vs VREG (Start up)

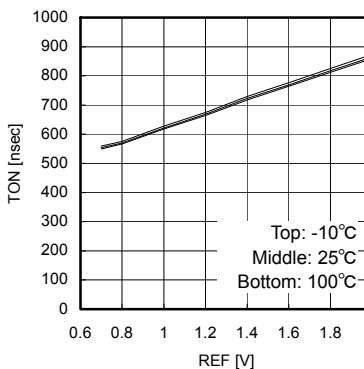


Fig.8 REF vs ON TIME (VIN=7V)

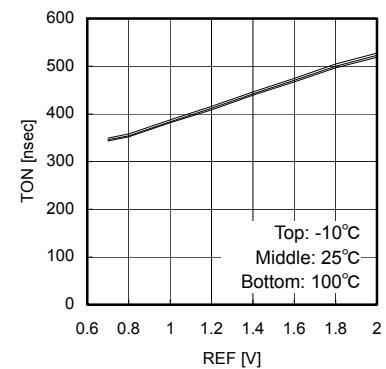


Fig.9 REF vs ON TIME (VIN=12V)

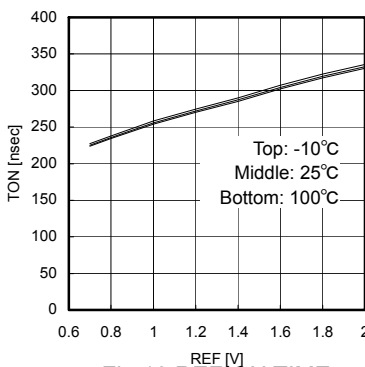


Fig.10 REF-ON TIME (VIN=25V)

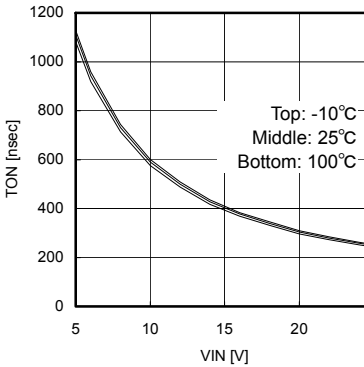


Fig.11 VIN-ON TIME (REF=1.8V)

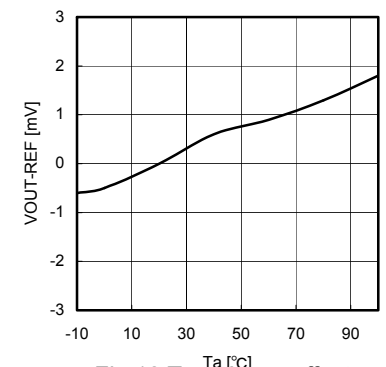


Fig.12 Ta vs Vout offset

●Reference Data – Continued

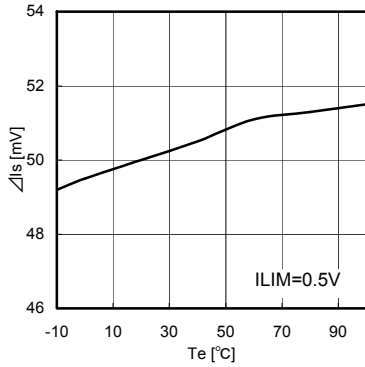


Fig.13  $T_a$  vs current limit threshold

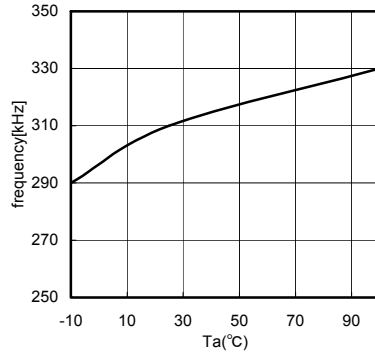


Fig.14  $T_a$  vs Frequency

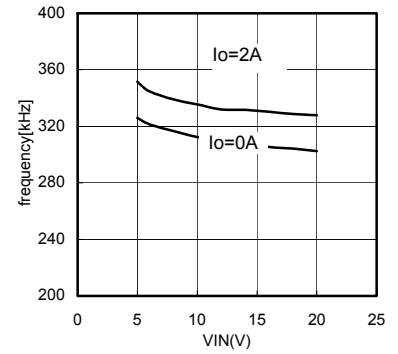


Fig.15  $V_{IN}$  vs Frequency

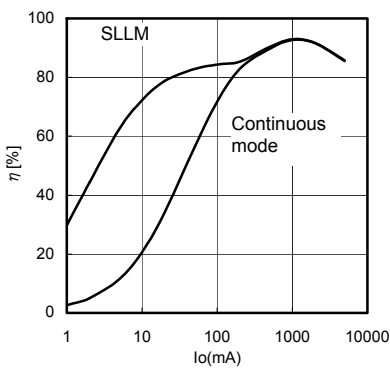


Fig.16  $I_o$  vs Efficiency ( $V_{IN}=7V$ )

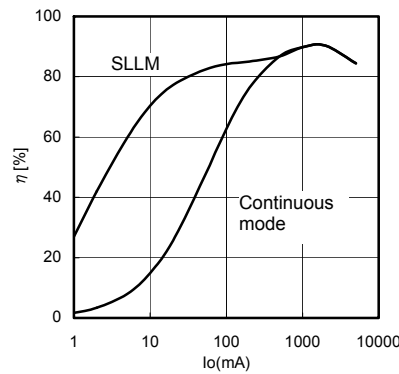


Fig.17  $I_o$  vs Efficiency ( $V_{IN}=12V$ )

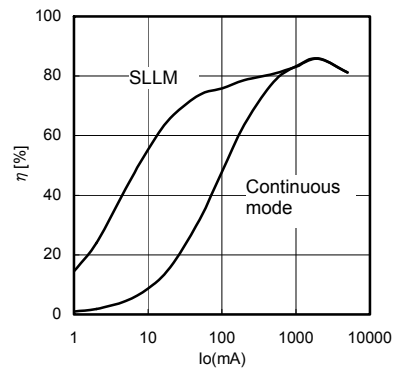


Fig.18  $I_o$  vs Efficiency ( $V_{IN}=20V$ )

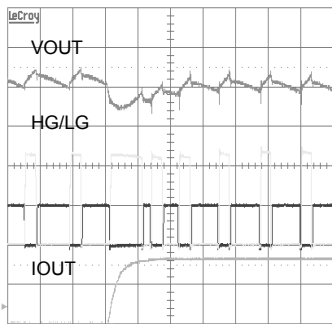


Fig.19 Load Transient Response ( $V_{IN}=7V$ )

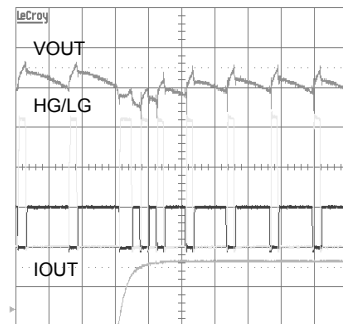


Fig.20 Load Transient Response ( $V_{IN}=12V$ )

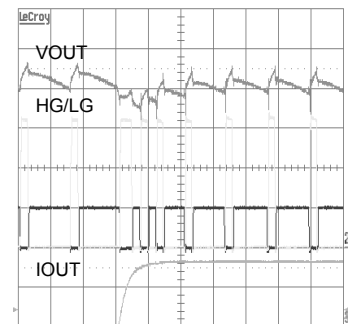


Fig.21 Load Transient Response ( $V_{IN}=19V$ )

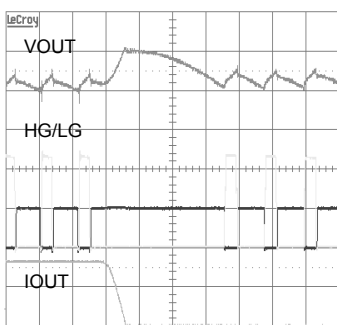


Fig.22 Load Transient Response ( $V_{IN}=7V$ )

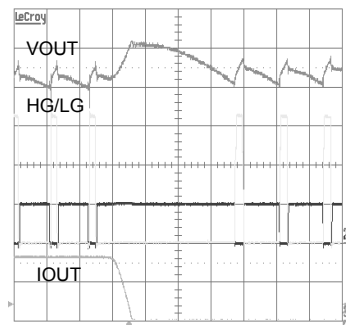


Fig.23 Load Transient Response ( $V_{IN}=12V$ )

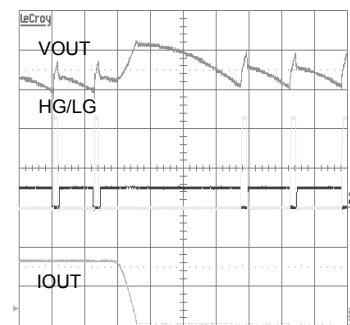


Fig.24 Load Transient Response ( $V_{IN}=19V$ )

●Reference Data – Continued

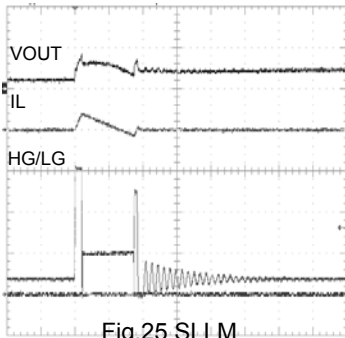


Fig.25 SLLM  
(Iout=0A)

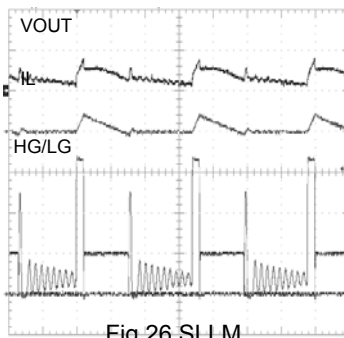


Fig.26 SLLM  
(Iout=0.4A)

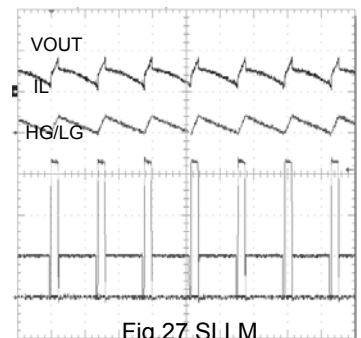


Fig.27 SLLM  
(Iout=1A)

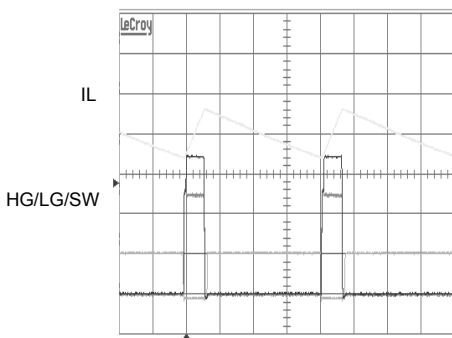


Fig.28 Continuous MODE  
(Io=0A)

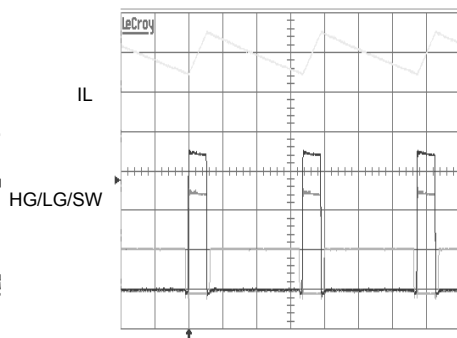


Fig.29 Continuous MODE  
(Io=4A)

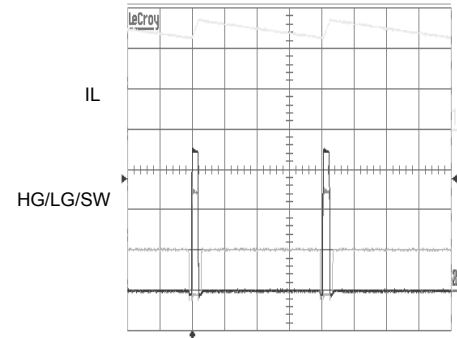
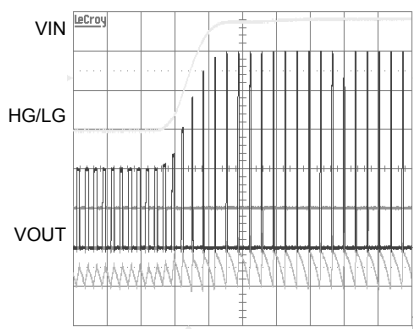


Fig.30 Continuous MODE  
(Io=5A)



(5→19V)

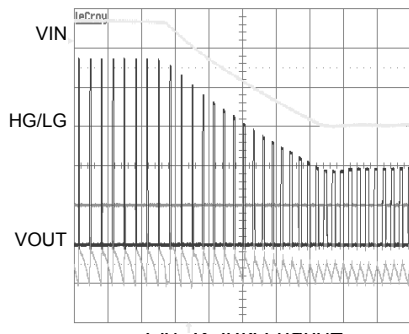


Fig.32 VIN change  
(19→5V)

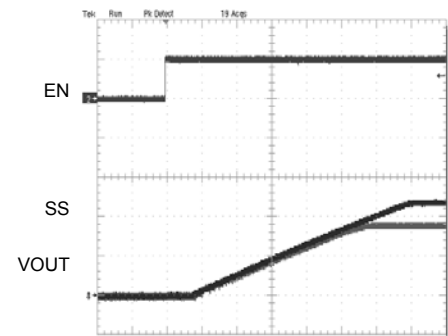


Fig.33 FS VIN wake up

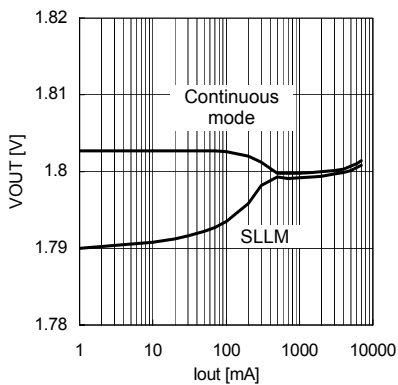


Fig.34 Iout-Vout

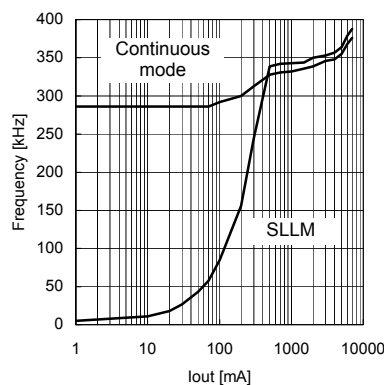


Fig.35 Iout-Frequency

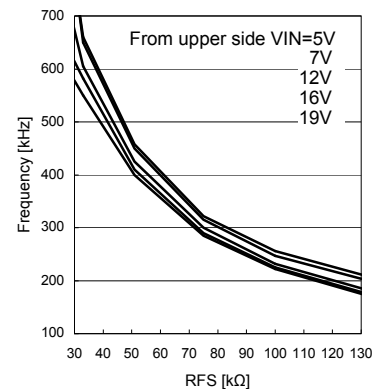
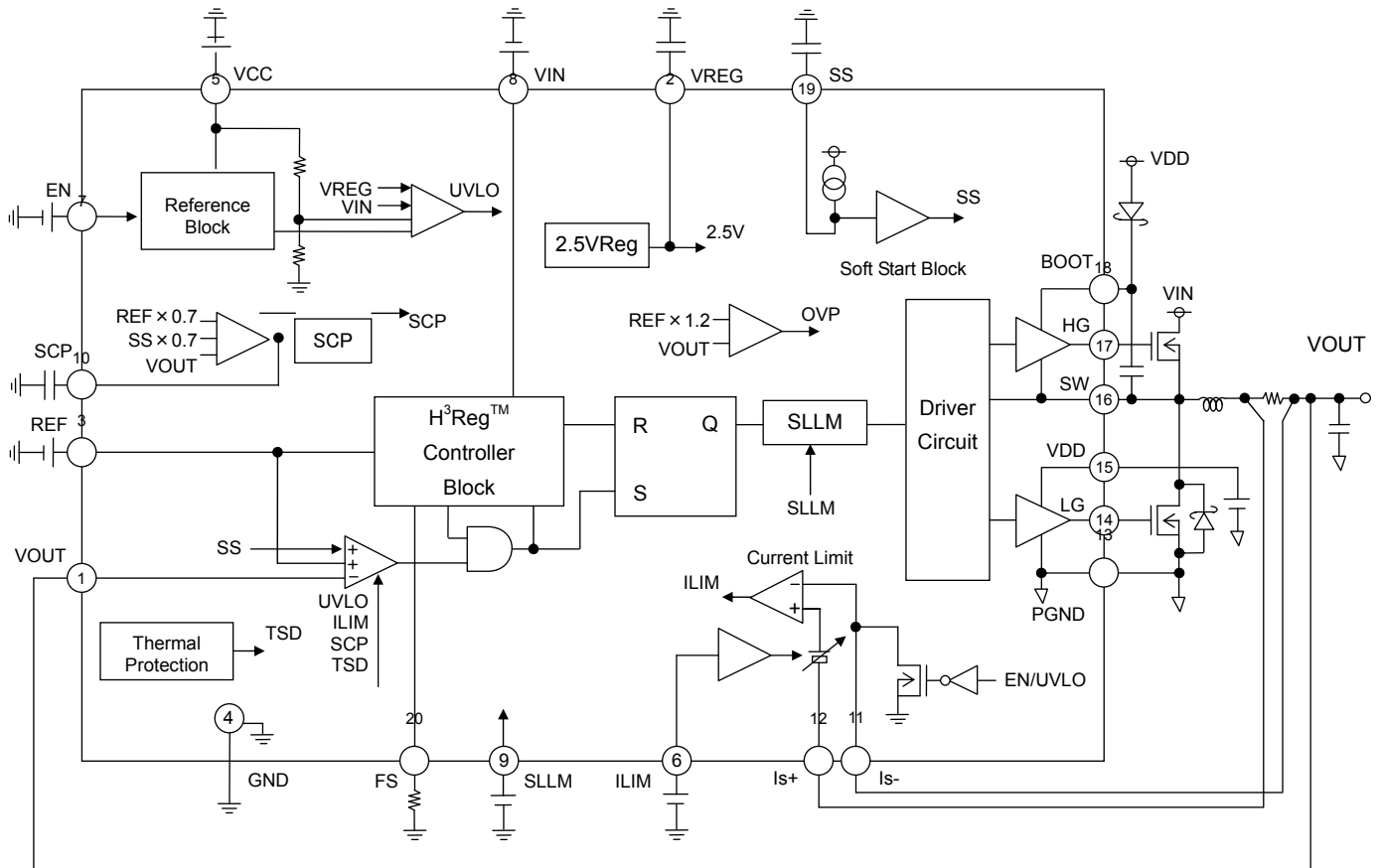
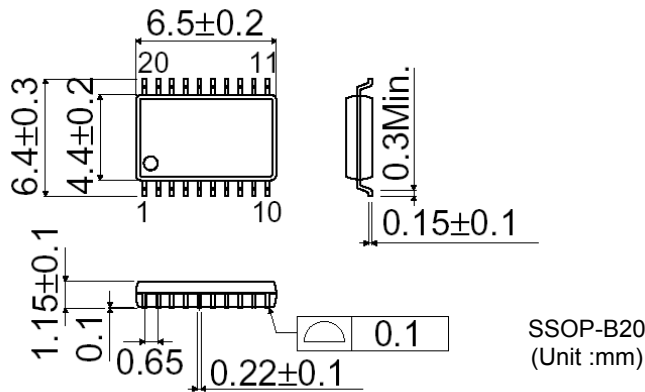


Fig.36 FS resistance- Frequency

●Block Diagram



●PHYSICAL DIMENSIONS



●Pin Number · Pin Name

PinNo.	Pin Name
1	VOUT
2	VREG
3	REF
4	GND
5	VCC
6	ILIM
7	EN
8	VIN
9	SLLM
10	SCP
11	IS-

PinNo.	Pin Name
12	IS+
13	PGND
14	LG
15	VDD
16	SW
17	HG
18	BOOT
19	SS
20	FS
-	FIN

\*Please short FIN to the 1PIN

### ● Pin Descriptions

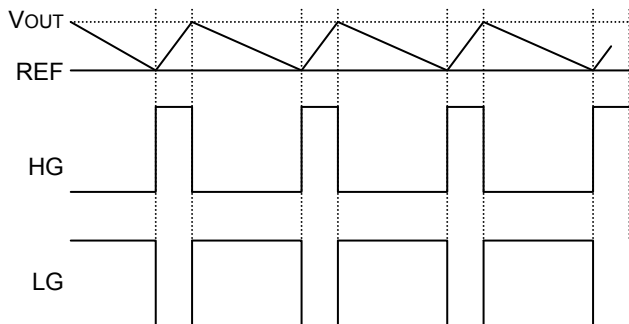
- VOUT (1pin)  
This is the output voltage sense pin. It is also possible to adjust the output voltage using external resistor divider based on the equation,  $REF \approx VOUT$ .
- VREG (2pin)  
This is the reference voltage output pin. The voltage is 2.5V, with 100 $\mu$ A current ability. It is recommended that a 1 $\mu$ F capacitor (X5R or X7R) be established between VREG and GND. When REF is not adjusted from the external voltage supply, the REF voltage can be adjusted using the external resistor divider of VREF.
- REF (3pin)  
This is the output voltage adjustment pin. It is very convenient for synchronizing external voltage supply. The IC controls the output voltage ( $REF \approx VOUT$ ).
- GND (4pin)  
This is the ground pin for IC internal circuits. It is equivalent to FIN voltage.
- VCC (5pin)  
This is the power supply pin for IC internal circuits, except the FET driver. The maximum circuit current is 900 $\mu$ A. The input supply voltage range is 4.5V to 5.5V. It is recommended that a 0.1 $\mu$ F bypass capacitor be put in this pin.
- ILIM (6pin)  
BD9575FV detects the voltage between Is+ pin and Is- pin and limits the output current (OCP). Voltage equivalent to 1/10 of the ILIM voltage is the voltage drop of external current sense resistor. A very low current sense resistor or inductor DCR can also be used for this platform.
- EN (7pin)  
When EN pin voltage is at least 2.3V, the status of this switching regulator becomes active. Conversely, the status switches off when EN pin voltage goes lower than 0.8V and circuit current becomes 10 $\mu$ A.
- VIN (8pin)  
The duty cycle is determined by input voltage and output voltage. In other words, the output voltage is affected by input voltage. Therefore, when VIN voltage fluctuates, the output voltage becomes also unstable. Since the VIN line is also the input voltage of the switching regulator, stability depends on the impedance of the voltage supply. It is recommended to establish a bypass capacitor or CR filter suitable for the actual application.
- SLLM (9pin)  
This is the switch shift pin for Simple Light Load Mode. The efficiency in SLLM is improved when SLLM pin voltage goes lower than 0.5V.
- SCP (10pin)  
This is the pin to adjust the timer latch time for short circuit protection. The timer circuit is active when the pin voltage becomes 70% of REF, and the output switches OFF and latched after the specified time. When the UVLO circuit is active or EN is low, this latch function is cancelled.
- Is- (11pin) ,Is+(12pin)  
These pins are connected to both sides of the current sense resistor to detect output current. The voltage drop between Is+ and Is- is compared with the voltage equivalent to 1/10 of ILIM voltage. When this voltage drop reaches the specified voltage level, the output voltage goes OFF.
- PGND (13pin)  
This is the power ground pin connected to the source of the low side FET.
- LG (14pin)  
This is the voltage supply to drive the Gate of the low side FET. This voltage swings between VDD and PGND. High-speed Gate driving for the low side FET is achieved due to the low on-resistance (2  $\Omega$  when LG is high, 0.5  $\Omega$  when LG is low) of the driver.
- VDD (15pin)  
This is the power supply pin to drive the LOW side FET. It is recommended that a 1 $\mu$ F bypass capacitor be established to compensate for rush current during the FET ON/OFF transition.
- SW (16pin)  
This is the source pin for the high side FET. The maximum absolute ratings are 30V (from GND). SW voltage swings between VIN and GND.
- HG (17pin)  
This is the voltage supply to drive the Gate of the high side FET. This voltage swings between BOOT and SW. High-speed Gate driving for the high side FET is achieved due to the low on-resistance (3  $\Omega$  when HG is high, 2  $\Omega$  when HG is low) of the driver.
- BOOT (18pin)  
This is the voltage supply to drive the high side FET. The maximum absolute ratings are 35V (from GND) and 7V (from SW). BOOT voltage swings between VIN+Vcc and Vcc during active operation.
- SS (19pin)  
This is the adjustment pin to set the soft start time. SS voltage is low during standby status. When EN is ON, the soft start time can be determined by the SS charge current and capacitor between SS-GND. Until SS reaches REF voltage, the output voltage is equivalent to SS voltage.
- FS (20pin)  
This is the pin to adjust the switching frequency based on the resistor value. The frequency range is f=200KHz - 600KHz.



●Explanation of Operation

The BD9575FV is a synchronous buck regulator controller incorporating ROHM's proprietary H<sup>3</sup>Reg™ CONTROLLA control system. When V<sub>OUT</sub> drops due to a rapid load change, the system quickly restores V<sub>OUT</sub> by extending the T<sub>ON</sub> time interval. Thus, it serves to improve the regulator's transient response. Activating the Light Load Mode will also exercise Simple Light Load Mode (SLLM) control when the load is light, to further increase efficiency.

H<sup>3</sup>Reg™ control  
(Normal operation)

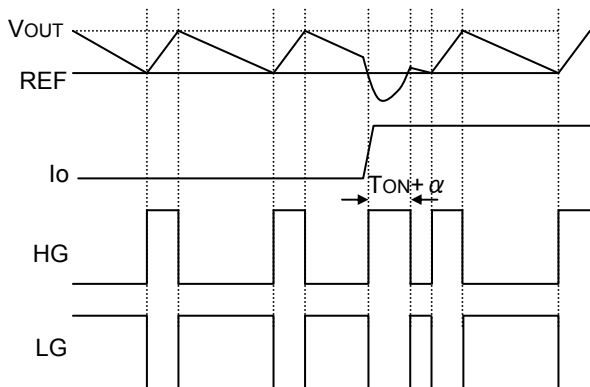


When V<sub>OUT</sub> falls to a threshold voltage (REF), the drop is detected, activating the H<sup>3</sup>Reg™ CONTROLLA system.

$$T_{ON} = \frac{REF}{V_{IN}} \times \frac{1}{f} \text{ [sec]} \cdot \cdot \cdot (1)$$

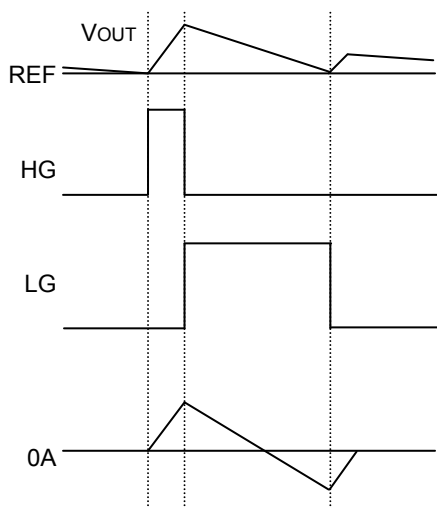
HG output is determined with the formula above. LG outputs until the status of V<sub>OUT</sub> is lower than REF after the status of HG is off.

(V<sub>OUT</sub> drops due to a rapid load change)



When V<sub>OUT</sub> drops due to a rapid load change, and the voltage remains below REF after the programmed T<sub>ON</sub> time interval has elapsed, the system quickly restores V<sub>OUT</sub> by extending the T<sub>ON</sub> time, improving the transient response.

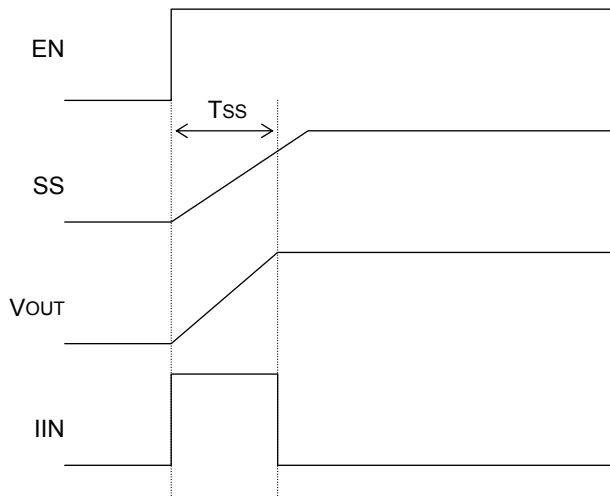
(SLLM )



In SLLM (SLLM=0V), SLLM function is operated when LG pin is OFF and the coil current is lower than 0A (the current goes from V<sub>OUT</sub> to SW). And it stops to output next HG.

When V<sub>OUT</sub> goes lower than REF voltage again, the status of HG is ON.

● Timing Chart  
 • Soft Start Function



Soft start is exercised with the EN pin set high. Current control takes effect at startup, enabling a moderate output voltage "ramping start." Soft start timing and incoming current are calculated with formulas (2) and (3) below.

Soft start time

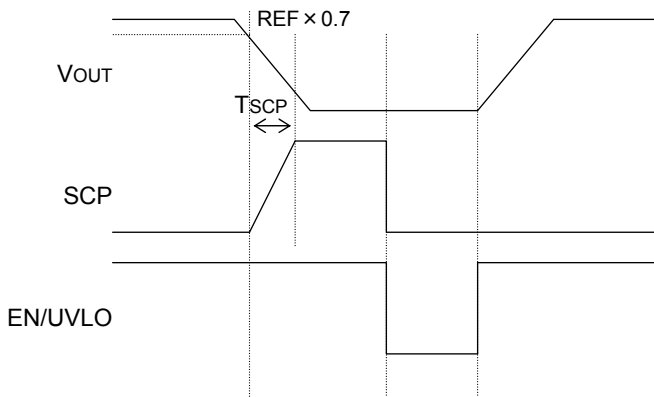
$$T_{ss} = \frac{REF \times C_{ss}}{2\mu A(\text{typ})} \quad [\text{sec}] \quad \dots (2)$$

Incoming current

$$I_{in} = \frac{C_o \times V_{OUT}}{T_{ss}} \quad [A] \quad \dots (3)$$

(C<sub>ss</sub>: Soft start capacitor; C<sub>o</sub>: Output capacitor)

• Timer Latch Type Short Circuit Protection

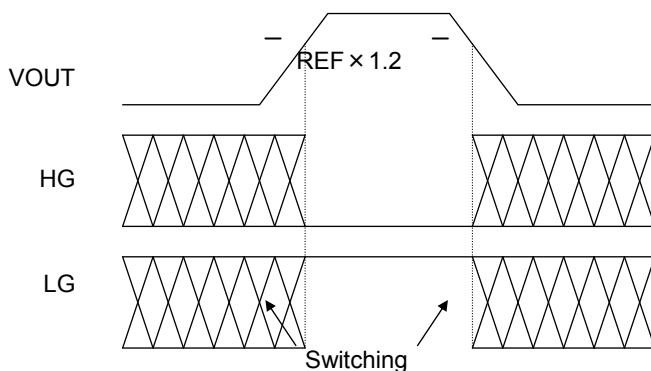


Short protection kicks in when output falls to or below REF X 0.7. When the programmed time period elapses, output is latched OFF to prevent destruction of the IC. Output voltage can be restored either by reconnecting the EN pin or disabling UVLO. Short circuit protection time is programmed using formula (4) below.

Short protection time setting

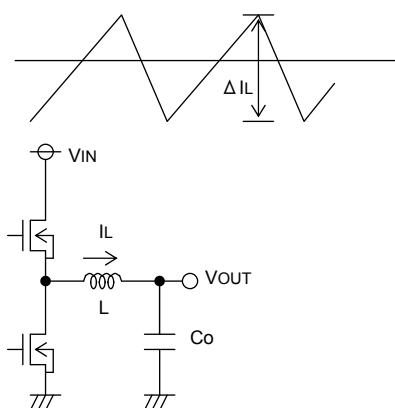
$$T_{scp} = \frac{1.25(V) \times C_{SCP}}{2\mu A(\text{typ})} \quad [\text{sec}] \quad \dots (4)$$

• Over Voltage Protection



● External Component Selection

1. Inductor (L) selection



The inductor value is a major influence on the output ripple current. As formula (5) below indicates, the greater the inductor or the switching frequency, the lower the ripple current.

$$\Delta I_L = \frac{(V_{IN} - V_{OUT}) \times V_{OUT}}{\Delta I_L \times V_{IN} \times f} \text{ [A]} \dots (5)$$

The proper output ripple current setting is about 30% of maximum output current.

$$\Delta I_L = 0.3 \times I_{OUTmax} \text{ [A]} \dots (6)$$

$$L = \frac{(V_{IN} - V_{OUT}) \times V_{OUT}}{\Delta I_L \times V_{IN} \times f} \text{ [H]} \dots (7)$$

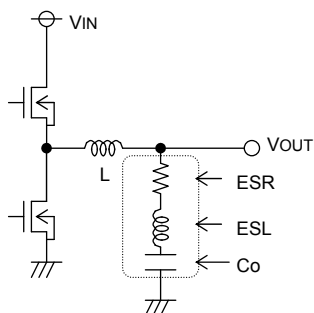
( $\Delta I_L$ : output ripple current; f: switch frequency)

Output ripple current

※ Passing a current larger than the inductor's rated current will cause magnetic saturation in the inductor and decrease system efficiency. In selecting the inductor, be sure to allow enough margin to assure that peak current does not exceed the inductor rated current value.

※ To minimize possible inductor damage and maximize efficiency, choose a inductor with a low (DCR, ACR) resistance.

2. Output Capacitor (Co) Selection



Output capacitor

When determining the proper output capacitor, be sure to factor in the equivalent series resistance required to smooth out ripple volume and maintain a stable output voltage range.

Output ripple voltage is determined as in formula (8) below.

$$\Delta V_{OUT} = \Delta I_L \times ESR + ESL \times \Delta I_L / T_{ON} \text{ [V]} \dots (8)$$

( $\Delta I_L$ : Output ripple current; ESR: Co equivalent series resistance, ESR: equivalent series inductance)

※ In selecting a capacitor, make sure the capacitor rating allows sufficient margin relative to output voltage. Note that a lower ESR can minimize output ripple voltage.

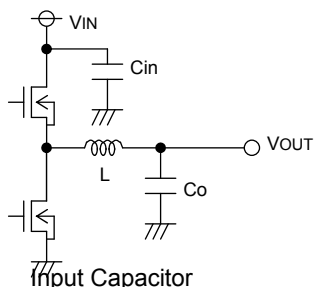
Please give due consideration to the conditions in formula (9) below for output capacity, bearing in mind that output rise time must be established within the soft start time frame.

$$C_o \leq \frac{T_{ss} \times (\text{Limit} - I_{OUT})}{V_{OUT}} \dots (9)$$

T<sub>ss</sub>: Soft start time  
Limit: Over current detection

Note: Improper capacitor may cause startup malfunctions.

3. Input Capacitor (Cin) Selection



Input Capacitor

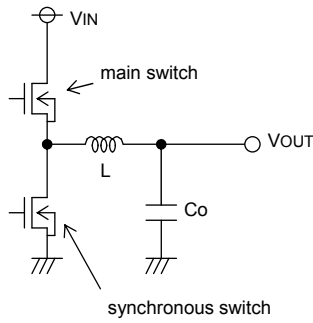
The input capacitor selected must have low enough ESR resistance to fully support large ripple output, in order to prevent extreme over current. The formula for ripple current IRMS is given in (10) below.

$$I_{RMS} = I_{OUT} \times \frac{\sqrt{V_{OUT}(V_{IN} - V_{OUT})}}{V_{IN}} \text{ [A]} \dots (10)$$

Where  $V_{CC} = 2 \times V_{OUT}$ ,  $I_{RMS} = \frac{I_{OUT}}{2}$

A low ESR capacitor is recommended to reduce ESR loss and maximize efficiency.

4. MOSFET Selection



Loss on the main MOSFET

$$P_{main} = P_{RON} + P_{GATE} + P_{TRAN}$$

$$= \frac{V_{OUT}}{V_{IN}} \times R_{ON} \times I_{OUT}^2 + C_{iss} \times f \times V_{DD} + \frac{V_{IN}^2 \times C_{rss} \times I_{OUT} \times f}{I_{DRIVE}} \dots (11)$$

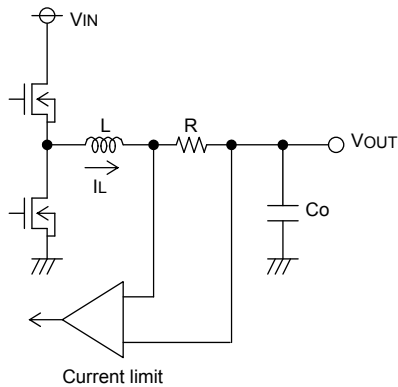
(Ron: On-resistance of FET; Ciss: FET gate capacity;  
f: Switching frequency Crss: FET inverse transfer function;  
IDRIVE: Gate peak current)

Loss on the synchronous MOSFET

$$P_{syn} = P_{RON} + P_{GATE}$$

$$= \frac{V_{IN} - V_{OUT}}{V_{IN}} \times R_{ON} \times I_{OUT}^2 + C_{iss} \times f \times V_{DD} \dots (12)$$

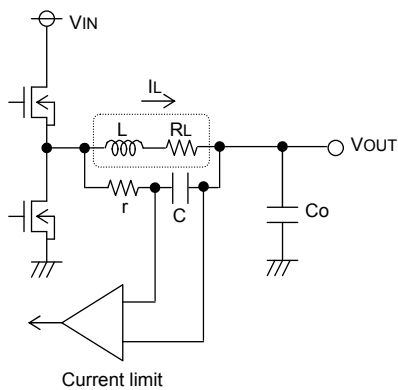
5. Setting Detection Resistance



The over current protection function detects the output ripple current peak value. This parameter (setting value) is determined as in formula (13) below.

$$I_{LIMIT} = \frac{V_{ILIM} \times 0.1 \text{ [A]}}{R} \dots (13)$$

(VILIM: ILIM voltage; R: Detection resistance)



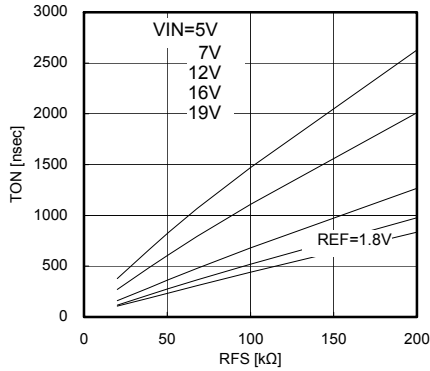
When the over current protection is detected by DCR of coil L, this parameter (setting value) is determined as in formula (14) below. (Application circuit:P20)

$$I_{LIMIT} = V_{ILIM} \times 0.1 \times \frac{r \times C}{L} \text{ [A]} \dots (14)$$

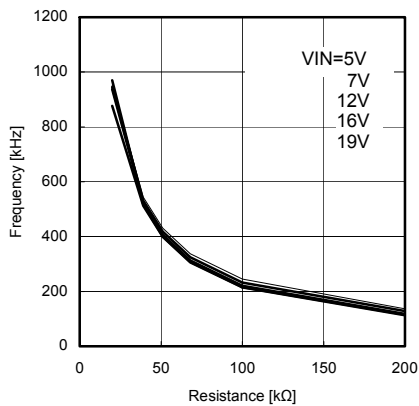
$$(R_L = \frac{L}{r \times C})$$

(VILIM:ILIM voltage RL: the DCR value of coil)

6. Setting frequency



↓  
Calculated frequency using TON above

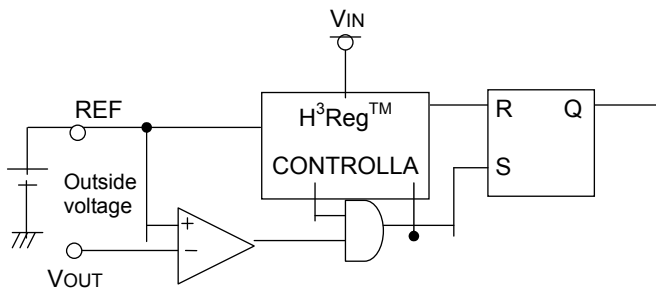


On Time (TON) at steady state is determined by resistance value connected to FS pin as shown in the figure. The frequency can be calculated with TON, input voltage VIN, the reference voltage REF as follows.

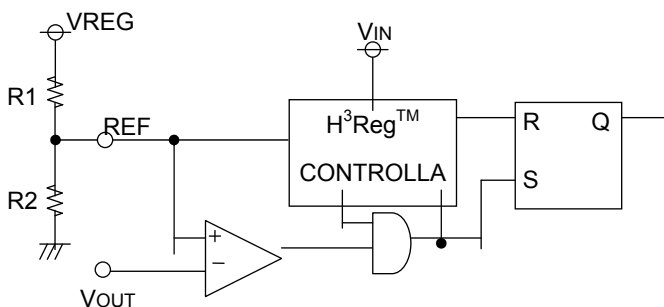
$$F = \frac{REF}{VIN \times TON} \dots (15)$$

However, the actual frequency becomes lower than the formula due to expanded TON time caused by the rise/fall time of the MOSFET (This rise/fall time is affected by the switching speed). TON is also influenced by Dead Time under light load in continuous mode. In this case the frequency becomes lower than setting frequency. The setting frequency noted above is under heavy load (when the inductor current does not go back to VIN through the body diodes of the external MOSFET's).

7. Setting standard voltage (REF)



It is available to synchronize setting the reference voltage (REF) with outside supply voltage [V] by using outside power supply voltage.



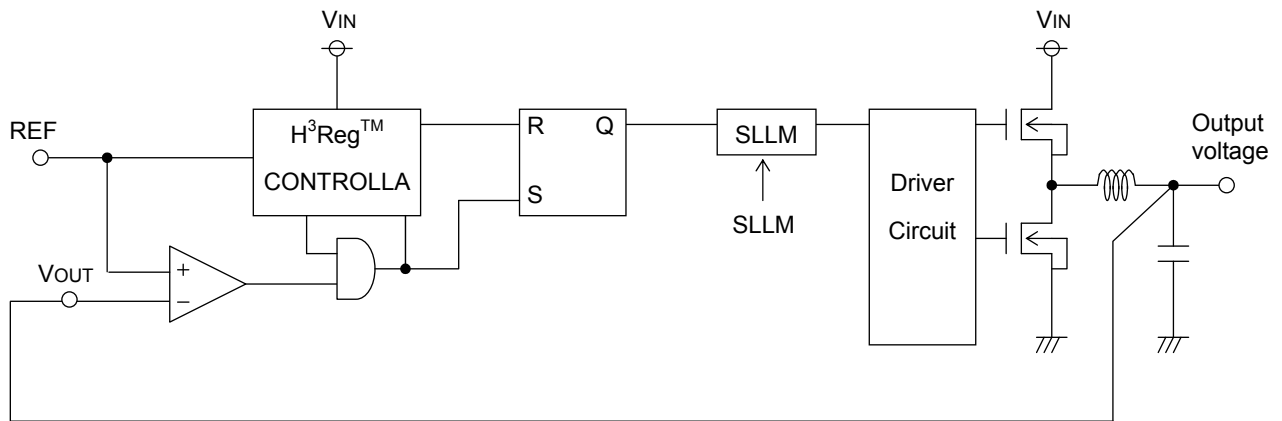
It is available to set the reference voltage (REF) by the resistance division value from VREG in case it is not set REF from an external power supply.

$$REF = \frac{R2}{R1+R2} \times VREG [V] \dots (16)$$

8. Setting output voltage

This IC is operated that output voltage is  $REF \doteq VOUT$ .

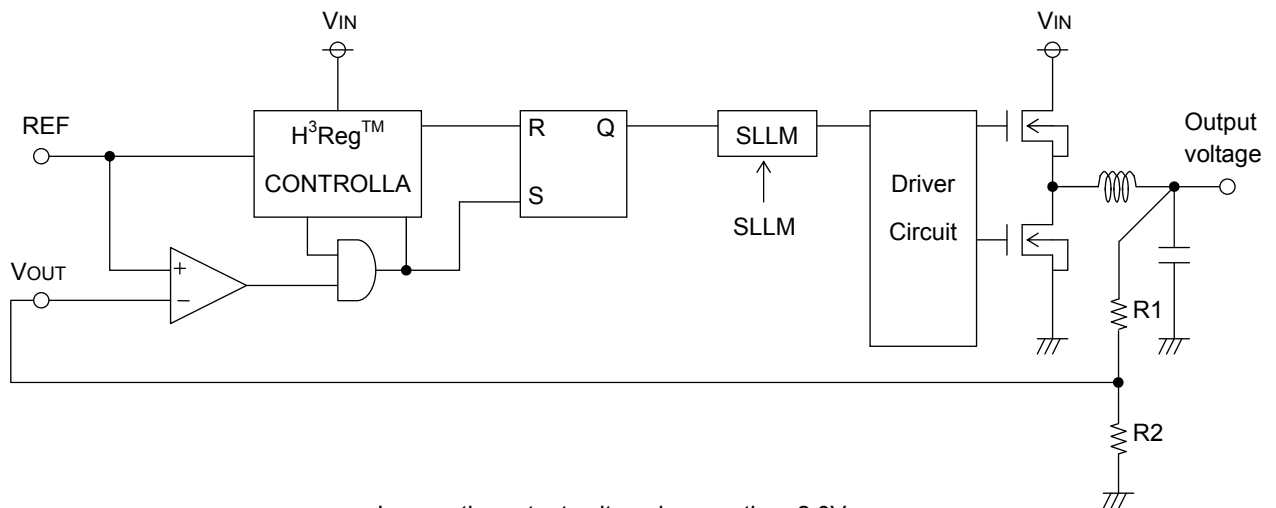
And it is operated that output voltage is feed back to FB pin in case the output voltage is 0.7V to 2.0V.



In case the output voltage range is 0.7V to 2.0V.

It is operated that the resistance division value of the output voltage is feed back to VOUT pin in case the output voltage is more than 2.0V.

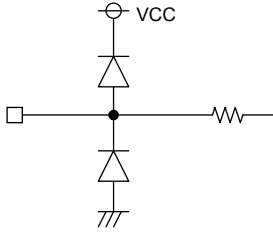
$$\text{output voltage} \doteq \frac{R1+R2}{R2} \times REF [V] \dots (17)$$



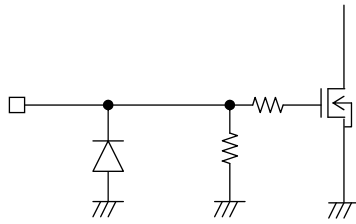
In case the output voltage is more than 2.0V.

● I/O Equivalent Circuit

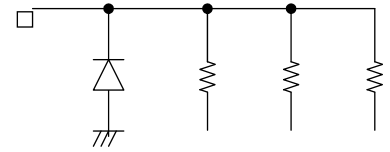
6pin (ILIM)



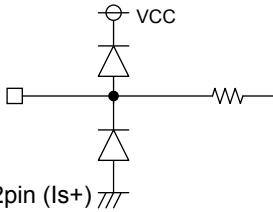
7pin (EN)



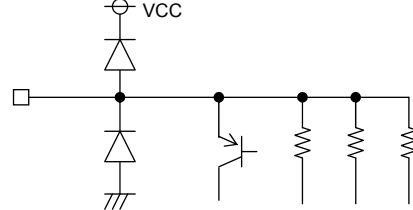
8pin (VIN)



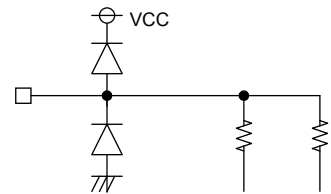
9pin (SLLM)



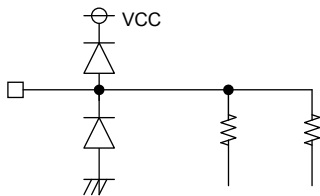
10pin (SCP)



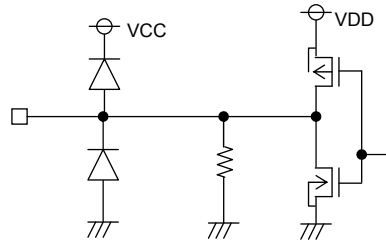
11pin (Is-)



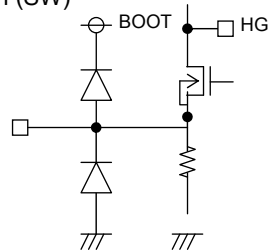
12pin (Is+)



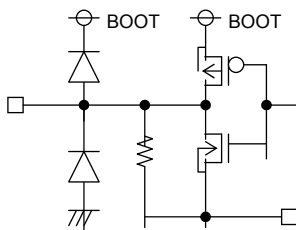
14pin (LG)



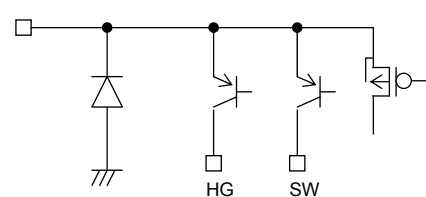
16pin (SW)



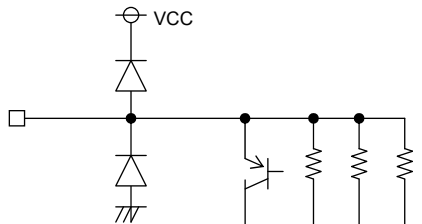
17pin (HG)



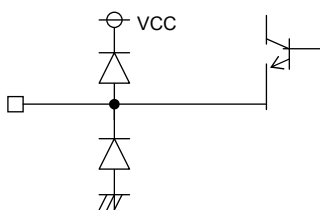
18pin (BOOT)



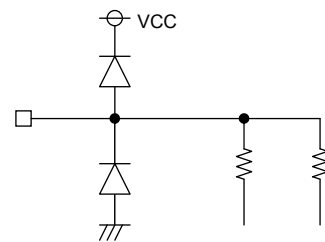
19pin (SS)



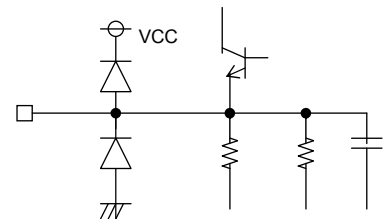
20pin (FS)



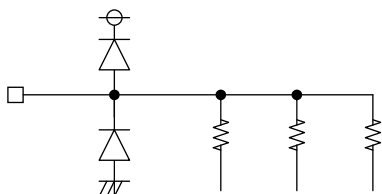
1pin (VOUT)



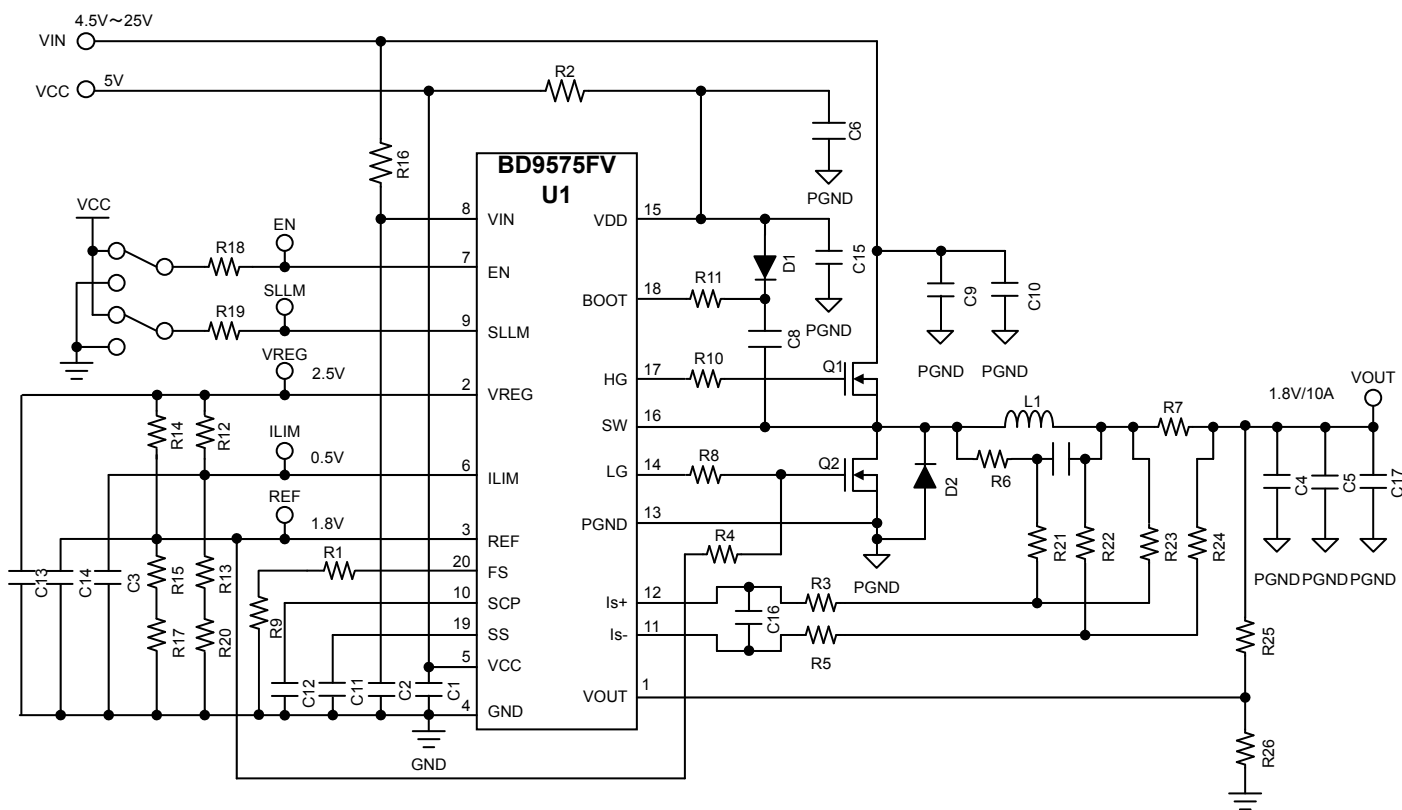
2pin (VREG)



3pin (REF)



●Evaluation Board Circuit (Frequency=300kHz application circuit in Continuous mode/SLLM)



●Evaluation Board Parts List

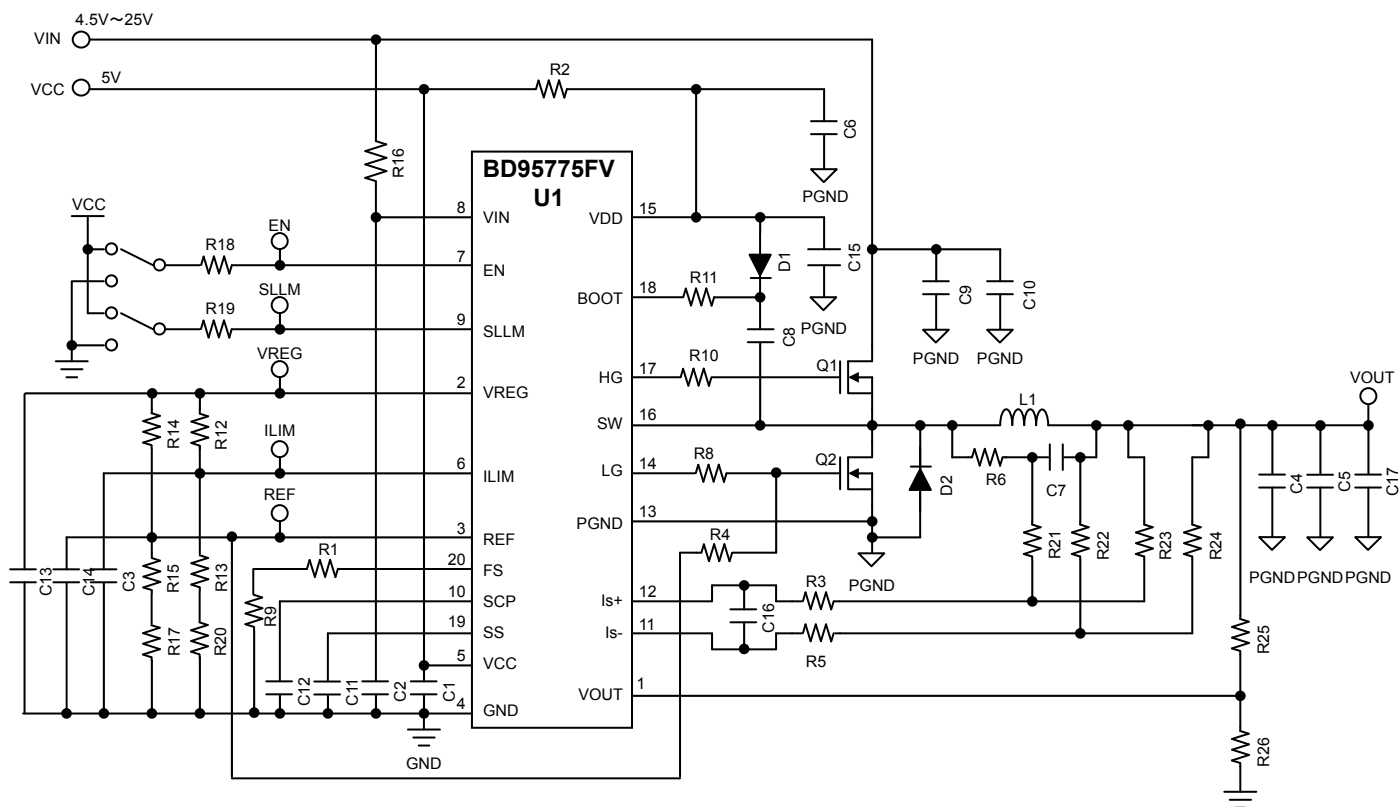
Part No	Value	Company	Part name
U1		ROHM	BD9575FV
Q1		ROHM	RSS100N03
Q2		ROHM	RSS100N03
D1		ROHM	RB521S-30
D2		ROHM	RB051L-40
C1	1μF	KYOCERA	CM105B105K06A
C2	1μF	KYOCERA	CM105B105K16A
C3	100pF	MURATA	GRM39C0G101J50
C4			
C5			
C6	10μF	KYOCERA	CM21B106K06
C7			
C8	0.1μF	KYOCERA	CM05B104K25A
C9	10μF	KYOCERA	CM316B106M16A
C10	10μF	KYOCERA	CM316B106M16A
C11	1000pF	MURATA	GRM39X7R102K50
C12	1500pF	MURATA	GRM39X7R152K50
C13	1μF	KYOCERA	CM105B105K06A
C14			
C15	10μF	KYOCERA	CM21B106M06A
C16			
C17	470μF	SANYO	2R5TPE470ML
R1	68KΩ	ROHM	MCR03
R2	0Ω	ROHM	MCR03
R3	0Ω	ROHM	MCR03

Part No	Value	Company	Part name
R4			
R5	0Ω	ROHM	MCR03
R6			
R7	5mΩ	ROHM	PMR100
R8	0Ω	ROHM	MCR03
R9	0Ω	ROHM	MCR03
R10	0Ω	ROHM	MCR03
R11	0Ω	ROHM	MCR03
R12	200kΩ	ROHM	MCR03
R13	51kΩ	ROHM	MCR03
R14	68kΩ	ROHM	MCR03
R15	180kΩ	ROHM	MCR03
R16	10Ω	ROHM	MCR03
R17	0Ω	ROHM	MCR03
R18	10kΩ	ROHM	MCR03
R19	10kΩ	ROHM	MCR03
R20	0Ω	ROHM	MCR03
R21			
R22			
R23	0Ω	ROHM	MCR03
R24	0Ω	ROHM	MCR03
R25	0Ω	ROHM	MCR03
R26			
L1	1.8μH	SUMIDA	CDEP104-1R8ML



●Evaluation Board Circuit

(Frequency=300kHz application circuit for detecting DCR current in Continuous mode/SLLM)



●Evaluation Board Parts List

Part No	Value	Company	Part name
U1		ROHM	BD9575FV
Q1		ROHM	RSS100N03
Q2		ROHM	RSS100N03
D1		ROHM	RB521S-30
D2		ROHM	RB051L-40
C1	1μF	KYOCERA	CM105B105K06A
C2	1μF	KYOCERA	CM105B105K16A
C3	100pF	MURATA	GRM39C0G101J50
C4			
C5			
C6	10μF	KYOCERA	CM21B106K06
C7	0.1μF	KYOCERA	CM05B104K25A
C8	0.1μF	KYOCERA	CM05B104K25A
C9	10μF	KYOCERA	CM316B106M16A
C10	10μF	KYOCERA	CM316B106M16A
C11	1000pF	MURATA	GRM39X7R102K50
C12	1500pF	MURATA	GRM39X7R152K50
C13	1μF	KYOCERA	CM105B105K06A
C14			
C15	10μF	KYOCERA	CM21B106M06A
C16			
C17	330μF	SANYO	6TPB330M
R1	68KΩ	ROHM	MCR03
R2	0Ω	ROHM	MCR03
R3	0Ω	ROHM	MCR03

Part No	Value	Company	Part name
R4			
R5	0Ω	ROHM	MCR03
R6	1kΩ	ROHM	MCR03
R7			
R8	0Ω	ROHM	MCR03
R9	0Ω	ROHM	MCR03
R10	0Ω	ROHM	MCR03
R11	0Ω	ROHM	MCR03
R12	51kΩ	ROHM	MCR03
R13	200kΩ	ROHM	MCR03
R14	68kΩ	ROHM	MCR03
R15	180kΩ	ROHM	MCR03
R16	10Ω	ROHM	MCR03
R17	0Ω	ROHM	MCR03
R18	10kΩ	ROHM	MCR03
R19	10kΩ	ROHM	MCR03
R20	0Ω	ROHM	MCR03
R21	0Ω	ROHM	MCR03
R22	0Ω	ROHM	MCR03
R23			
R24			
R25	0Ω	ROHM	MCR03
R26			
L1	3.3μH	NEC/TOKIN	MPLC0730L3R3

● Notes for use

1. Absolute maximum ratings

An excess in the absolute maximum ratings, such as supply voltage, temperature range of operating conditions, etc., can break down the devices, thus making impossible to identify breaking mode, such as a short circuit or an open circuit. If any over rated values will expect to exceed the absolute maximum ratings, consider adding circuit protection devices, such as fuses.

2. Connecting the power supply connector backward

Connecting of the power supply in reverse polarity can damage IC. Take precautions when connecting the power supply lines. An external direction diode can be added.

3. Power supply lines

Design PCB layout pattern to provide low impedance GND and supply lines. To obtain a low noise ground and supply line, separate the ground section and supply lines of the digital and analog blocks. Furthermore, for all power supply terminals to ICs, connect a capacitor between the power supply and the GND terminal. When applying electrolytic capacitors in the circuit, not that capacitance characteristic values are reduced at low temperatures.

4. GND voltage

The potential of GND pin must be minimum potential in all operating conditions.

5. Thermal design

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) in actual operating conditions.

6. Inter-pin shorts and mounting errors

Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error or if pins are shorted together.

7. Actions in strong electromagnetic field

Use caution when using the IC in the presence of a strong electromagnetic field as doing so may cause the IC to malfunction.

8. ASO

When using the IC, set the output transistor so that it does not exceed absolute maximum ratings or ASO.

9. Thermal shutdown circuit

The IC incorporates a built-in thermal shutdown circuit (TSD circuit). The thermal shutdown circuit (TSD circuit) is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation. Do not continue to use the IC after operating this circuit or use the IC in an environment where the operation of this circuit is assumed.

	TSD on temperature [°C] (typ.)	Hysteresis temperature [°C] (typ.)
BD9575FV	175	15

10. Testing on application boards

When testing the IC on an application board, connecting a capacitor to a pin with low impedance subjects the IC to stress. Always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from a jig or fixture during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

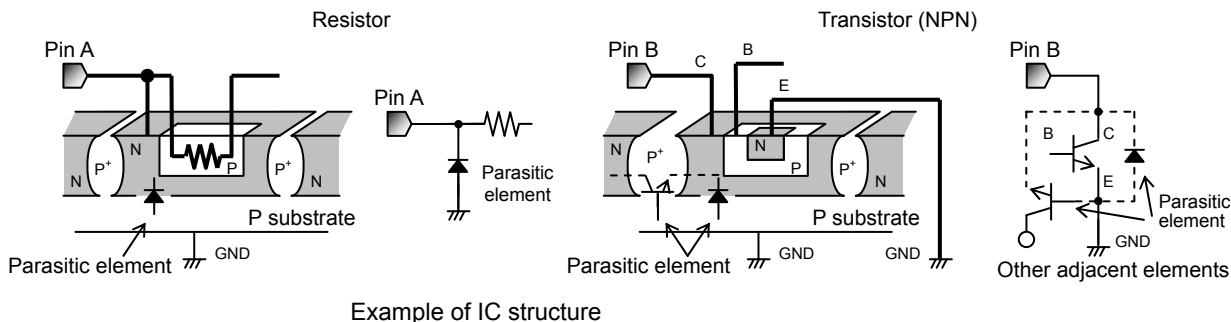
11. Regarding input pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

When  $GND > Pin A$  and  $GND > Pin B$ , the P-N junction operates as a parasitic diode.

When  $GND > Pin B$ , the P-N junction operates as a parasitic transistor.

Parasitic diodes can occur inevitable in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Accordingly, methods by which parasitic diodes operate, such as applying a voltage that is lower than the GND (P substrate) voltage to an input pin, should not be used.



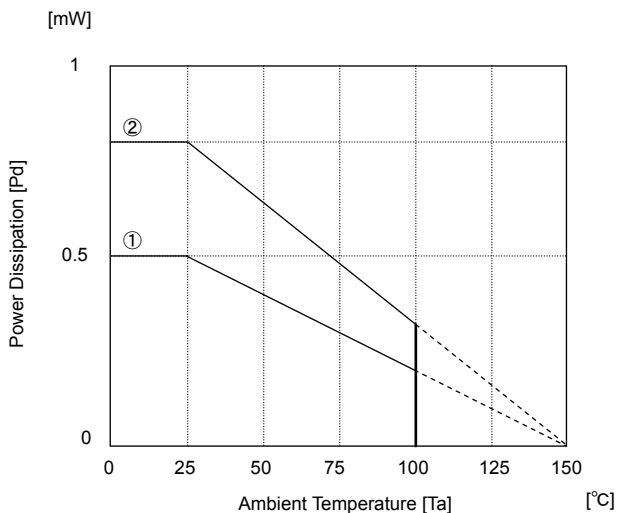
Example of IC structure

12. Ground Wiring Pattern

When using both small signal and large current GND patterns, it is recommended to isolate the two ground patterns, placing a single ground point at the ground potential of application so that the pattern wiring resistance and voltage variations caused by large currents do not cause variations in the small signal ground voltage. Be careful not to change the GND wiring pattern of any external components, either.

● Thermal Derating Curve

◎SSOP-B20



① IC only

$\theta_{ja}=250^{\circ}\text{C}/\text{W}$

② 1 layer board

(70mm × 70mm × 1.6mm 70.0mm × 70mm × 1.6mm Glass-epoxy PCB).

$\theta_{ja}=153.8^{\circ}\text{C}/\text{W}$

●Ordering part number

B	D
---	---

Part No.

9	5	7	5
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Part No.

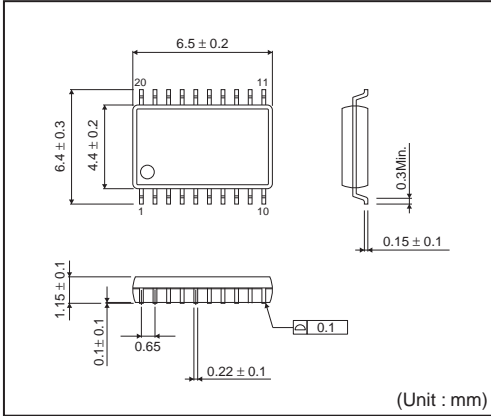
F	V
---	---

Package  
FV : SSOP-B20

E	2
---	---

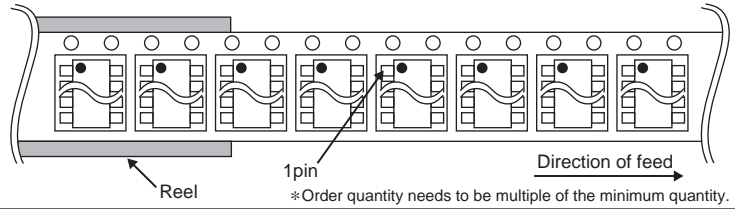
Packaging and forming specification  
E2: Embossed tape and reel

SSOP-B20



<Tape and Reel information>

Tape	Embossed carrier tape
Quantity	2500pcs
Direction of feed	E2 (The direction is the 1pin of product is at the upper left when you hold reel on the left hand and you pull out the tape on the right hand )



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