19-3676; Rev 0; 5/05

DirectDrive Video Amplifier with Reconstruction Filter

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General Description

The MAX9503 filters and amplifies standard-definition video signals. Maxim's DirectDrive™ technology eliminates large output-coupling capacitors and sets the video black level to ground. The input of the MAX9503 can be directly connected to the output of a video digital-to-analog converter (DAC). The MAX9503 provides a compact, integrated, and low-power solution.

An internal reconstruction filter smoothes the steps and reduces the spikes on the video signal from the DAC. The reconstruction filter typically has 3dB attenuation at 9MHz and 50dB attenuation at 27MHz, and ±1dB passband flatness to 5.5MHz.

Maxim's DirectDrive uses an integrated charge pump and a linear regulator to create a clean negative power supply to drive the sync below ground. The charge pump injects so little noise into the video output that the picture is visibly flawless.

The MAX9503 is available with +6dB (MAX9503G) and +12dB (MAX9503M) gains. The device operates from a 2.7V to 3.6V single supply and features a 10nA low-power shutdown mode.

The MAX9503 is offered in space-saving 16-pin QSOP and 16-pin TQFN packages and is specified over the -40°C to +85°C extended temperature range.

Applications

Digital Still Cameras

Mobile Phones/Smartphones

Security Cameras

Portable Media Players

Space-Constrained, Low-Power Portable Devices

Features

- ♦ **DC-Coupled Output**
- ♦ **Direct Connection to Video DAC**
- ♦ **Video Output Black Level Set to Ground**
- ♦ **Video Reconstruction Filter with 50dB Attenuation at 27MHz**
- ♦ **Preset Gain**
  6dB (MAX9503G)
  12dB (MAX9503M)
- ♦ **10nA Shutdown Supply Current**
- ♦ **2.7V to 3.6V Single-Supply Operation**

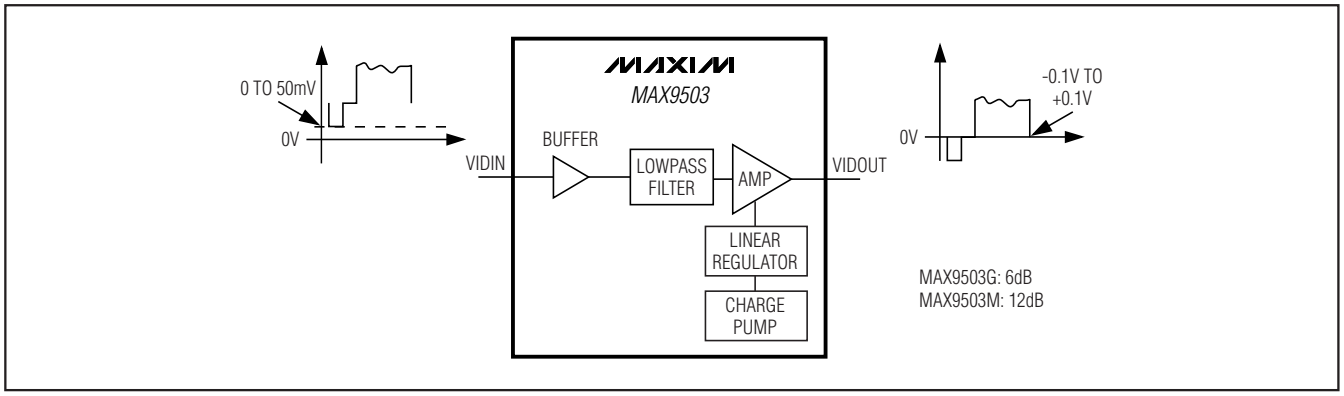
Ordering Information

PART*	PIN-PACKAGE	PKG CODE	TOP MARK
MAX9503GEEE	16 QSOP	E16-4	—
MAX9503GETE	16 TQFN	T1633-4	ACU
MAX9503MEEE	16 QSOP	E16-4	—
MAX9503METE	16 TQFN	T1633-4	ACV

**All devices are specified over the -40°C to +85°C operating temperature range.*

Functional Diagram/Typical Operating Circuit and Pin Configurations appear at end of data sheet.

Block Diagram



For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

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ABSOLUTE MAXIMUM RATINGS

V_{DD} to SGND .....-0.3V to +4V
VIDIN to SGND .....-0.3V to +4V
BIAS to SGND .....-0.3V to (V_{DD} + .3V)
$\overline{SHDN}$ to SGND .....-0.3V to +4V
VIDOUT to SGND.....(The greater of V_{SS} and -2V) to (V_{DD} + 0.3V)
CPV_{DD} to CPGND .....-0.3V to +4V
C1P, C1N, CPV_{SS} .....Capacitor Connection Only
CPGND, SGND, GND.....-0.1V to +0.1V
CPV_{SS} to V_{SS} .....-0.1V to +0.1V
VIDOUT Short Circuit to V_{DD}, SGND and the Greater of (V_{SS} and -2V).....Continuous
Continuous Current
VIDIN, BIAS, $\overline{SHDN}$ .....±20mA
Continuous Power Dissipation (T_A = +70°C)
16-Pin QSOP (derate 8.3mW/°C above +70°C).....667mW
16-Pin TQFN (derate 15.6mW/°C above +70°C) 1349mW
Operating Temperature Range .....-40°C to +85°C
Junction Temperature.....+150°C
Storage Temperature Range .....-65°C to +150°C
Lead Temperature (soldering, 10s) .....+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_L = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, T_A = +25°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Supply Voltage Range	V_{DD}, CPV_{DD}	Guaranteed by DC voltage gain and quiescent current		2.7		3.6	V
Quiescent Supply Current	I_{DD}	V_{DD} = 3.6V (I_{DD} = I_{VDD} + I_{CPVDD}, R_L = ∞)			12	15	mA
Shutdown Supply Current	$I_{\overline{SHDN}}$	V_{DD} = 3.6V ($I_{\overline{SHDN}}$ = I_{VDD} + I_{CPVDD}), $\overline{SHDN}$ = SGND			0.01	1	µA
Bias Voltage	V_{BIAS}				1		V
VIDEO AMPLIFIER							
Input Voltage Range	V_{RANGE}	Guaranteed by DC voltage gain, V_{DD} = 2.7V	MAX9503G	-0.10		+1.05	V
			MAX9503M	-0.050		+0.525	
		Guaranteed by DC voltage gain, V_{DD} = 3V	MAX9503G	-0.10		+1.28	
			MAX9503M	-0.05		+0.64	
Input Current	I_{IN}	V_{DD} = 2.7V		-2.5		+2.5	µA
Input Resistance	R_{IN}				1		MΩ
DC Voltage Gain (Note 2)	A_V	V_{DD} = 2.7V to 3.6V	MAX9503G	5.5	6	6.5	dB
			MAX9503M	11.5	12	12.5	
Output Black Level (Note 3)		V_{DD} = 2.7V	MAX9503G	-0.1	0	+0.1	V
			MAX9503M	-0.15	0	+0.15	
Output Short-Circuit Current	I_{SC}	Sinking or sourcing			50		mA
Output Resistance	R_{OUT}	MAX9503G			0.01		Ω
		MAX9503M			0.02		
Shutdown Output Impedance	$R_{OUT\overline{SHDN}}$	$\overline{SHDN}$ = SGND	MAX9503G		4.2		kΩ
			MAX9503M		8.2		

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ELECTRICAL CHARACTERISTICS (continued)

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_L = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, T_A = +25°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
CHARGE PUMP						
Switching Frequency			150	250	300	kHz
LOGIC SIGNALS						
Logic-Low Threshold	V_{IL}	V_{DD} = 2.7V to 3.6V			0.5	V
Logic-High Threshold	V_{IH}	V_{DD} = 2.7V to 3.6V	1.5			V
Logic Input Current	I_{IL}				1	µA

AC ELECTRICAL CHARACTERISTICS

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_L = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, T_A = +25°C, unless otherwise noted.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Passband Flatness		V_{DD} = 2.7V, f = 100kHz to 5.5MHz		-1	0	+1	dB
Attenuation		V_{DD} = 2.7V, VIDOUT= $2V_{P-P}$, attenuation is referred to 100kHz	f = 9.1MHz		3		dB
			f = 27MHz	35	50		
Power-Supply Rejection Ratio	PSRR	f = 100kHz	MAX9503G		62		dB
			MAX9503M		56		
Output Impedance	Z_{OUT}	f = 5MHz	MAX9503G		0.5		Ω
			MAX9503M		0.65		
Differential Gain Error	DG	NTSC, VIDOUT = $2V_{P-P}$	MAX9503G		0.1		%
			MAX9503M		0.1		
Differential Phase Error	DP	NTSC, VIDOUT = $2V_{P-P}$	MAX9503G		0.2		degrees
			MAX9503M		0.2		
2T Pulse-to-Bar K Rating		2T = 250ns, bar time is 18µs, the beginning 2.5% and the ending 2.5% of the bar time are ignored			-0.3		K%
2T Pulse Response		2T = 250ns			0.3		K%
2T Bar Response		2T = 250ns, bar time is 18µs, the beginning 2.5% and the ending 2.5% of the bar time are ignored			0.7		K%
Nonlinearity		5-step staircase			0.2		%
Group-Delay Distortion	D_{Dt}	100kHz to 5.5MHz			10		ns
VIDOUT Capacitive-Load Stability	C_L	V_{OUT} = $2V_{P-P}$, no sustained oscillations			20		pF

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AC ELECTRICAL CHARACTERISTICS (continued)

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_L = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, T_A = +25°C, unless otherwise noted.) (Note 1)

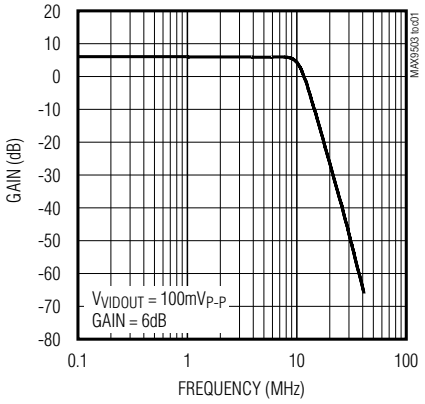
PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Peak Signal-to-RMS Noise	SNR	100kHz to 5.5MHz	MAX9503G		64		dB
			MAX9503M		58		
Enable Time	t_{ON}	VIDIN = 0.5V, VIDOUT settled to within 1% of the final voltage			0.2		ms
Disable Time	t_{OFF}	VIDIN = 0.5V, VIDOUT settled to below 1% of the output voltage			0.1		ms

Note 1: All devices are 100% production tested at T_A = +25°C. Specifications over temperature are guaranteed by design.
Note 2: Voltage gain (A_V) is a two-point measurement in which the output voltage swing is divided by the input voltage swing.
Note 3: With an output load attached, this offset will directly contribute to quiescent current.

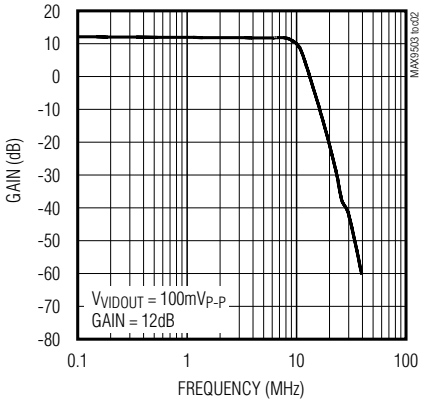
Typical Operating Characteristics

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, no load, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_{IN} = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at T_A = +25°C, unless otherwise noted.)

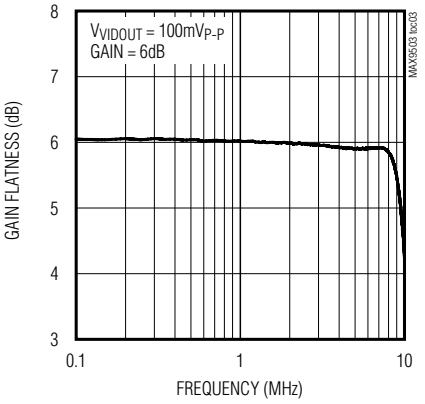
MAX9503G SMALL-SIGNAL GAIN vs. FREQUENCY



MAX9503M SMALL-SIGNAL GAIN vs. FREQUENCY



MAX9503G SMALL-SIGNAL GAIN FLATNESS vs. FREQUENCY



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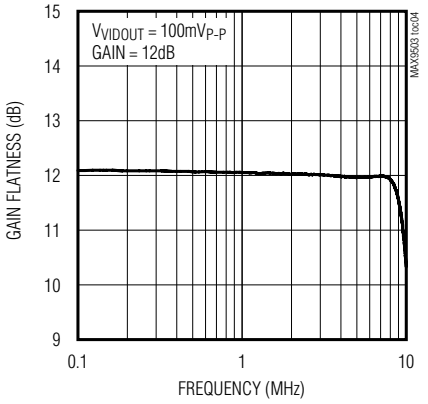
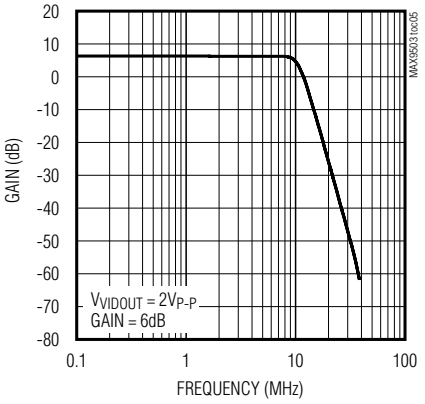
Typical Operating Characteristics (continued)

(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, no load, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_{IN} = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at T_A = +25°C, unless otherwise noted.)

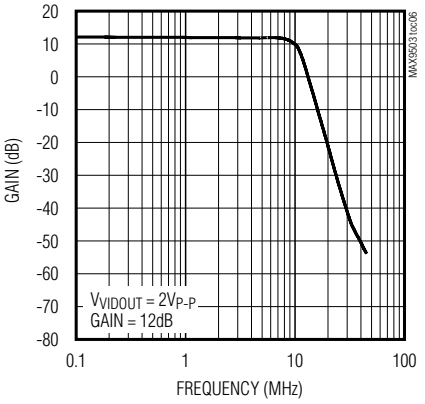
MAX9503M SMALL-SIGNAL GAIN FLATNESS vs. FREQUENCY



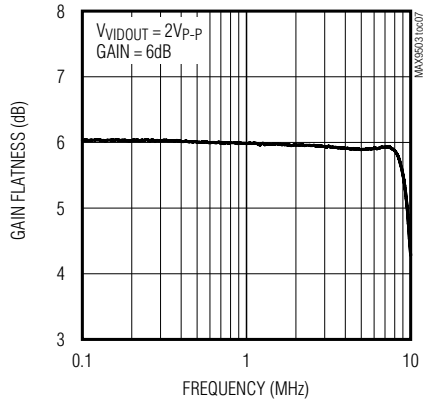
MAX9503G LARGE-SIGNAL GAIN vs. FREQUENCY



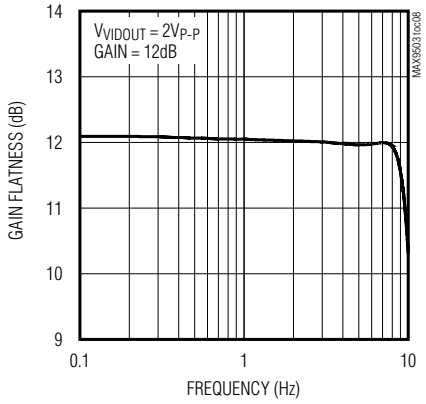
MAX9503M LARGE-SIGNAL GAIN vs. FREQUENCY



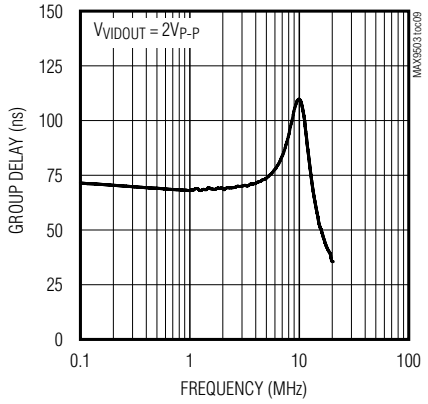
MAX9503G LARGE-SIGNAL GAIN FLATNESS vs. FREQUENCY



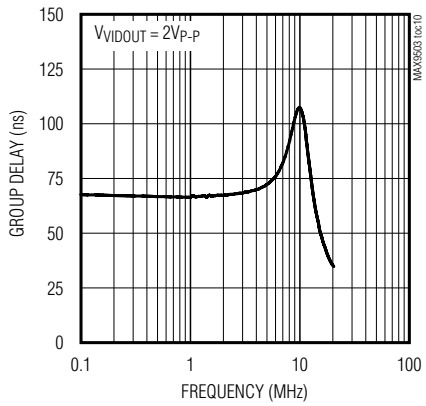
MAX9503M LARGE-SIGNAL GAIN FLATNESS vs. FREQUENCY



MAX9503G GROUP DELAY



MAX9503M GROUP DELAY



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Typical Operating Characteristics (continued)

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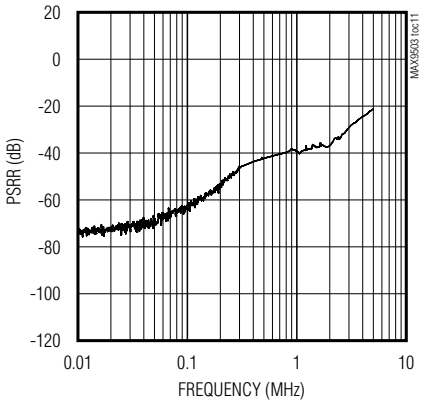
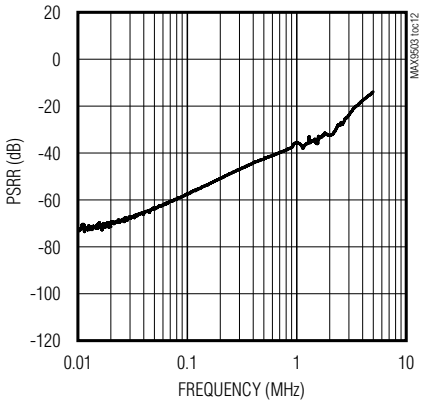
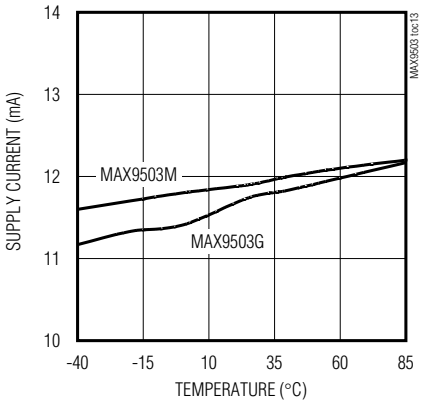
MAX9503G POWER-SUPPLY REJECTION RATIO vs. FREQUENCY



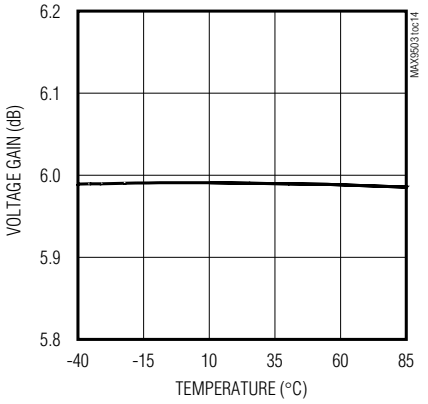
MAX9503M POWER-SUPPLY REJECTION RATIO vs. FREQUENCY



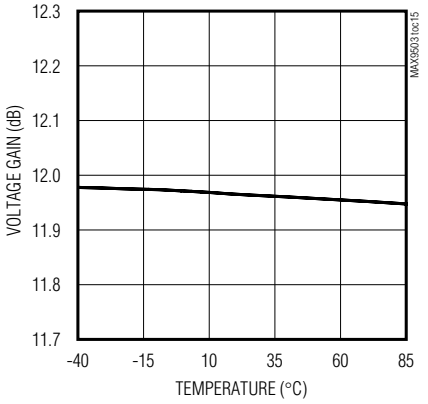
QUIESCENT SUPPLY CURRENT vs. TEMPERATURE



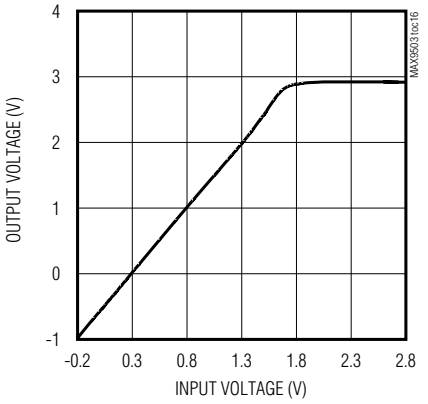
MAX9503G VOLTAGE GAIN vs. TEMPERATURE



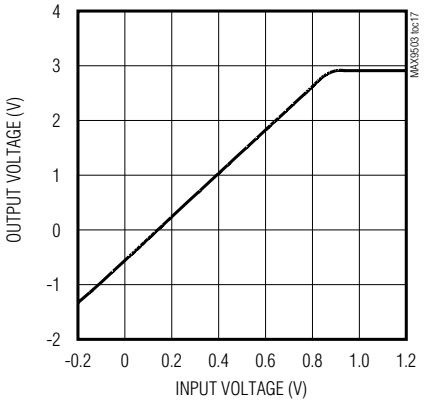
MAX9503M VOLTAGE GAIN vs. TEMPERATURE



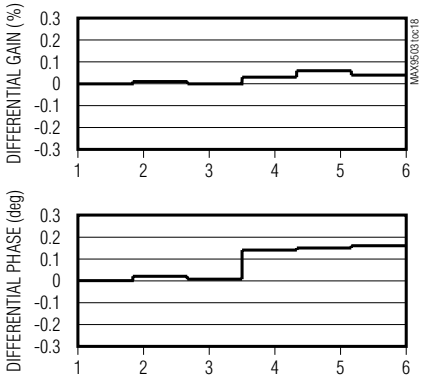
MAX9503G OUTPUT VOLTAGE vs. INPUT VOLTAGE



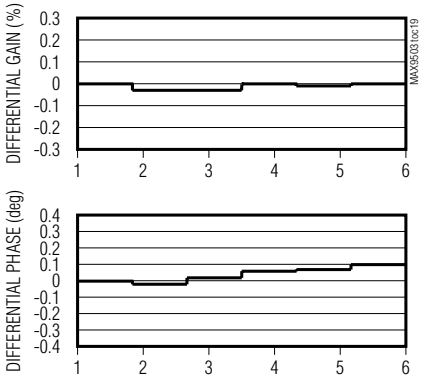
MAX9503M OUTPUT VOLTAGE vs. INPUT VOLTAGE



MAX9503G DIFFERENTIAL GAIN AND PHASE



MAX9503M DIFFERENTIAL GAIN AND PHASE



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Typical Operating Characteristics (continued)

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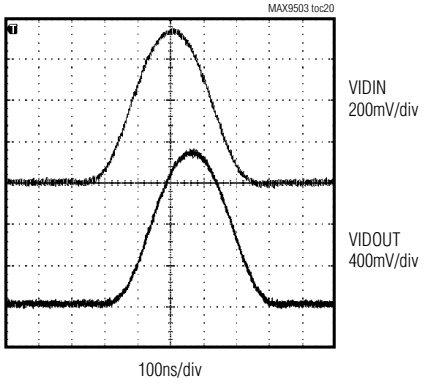
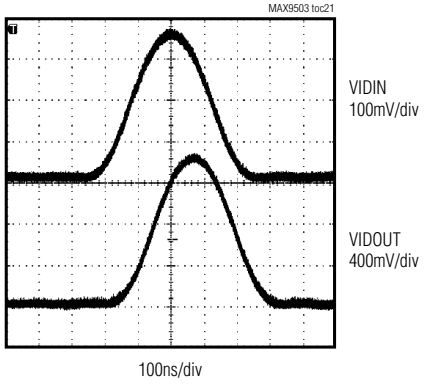
(V_{DD} = CPV_{DD} = $\overline{SHDN}$ = 3.0V, SGND = GND = CPGND = 0V, no load, C1 = C2 = C3 = C4 = 1µF, R_{BIAS} = 100kΩ, T_A = T_{MIN} to T_{MAX}. R_{IN} = 150Ω to SGND, unless otherwise noted. V_{VIDIN} = 286mV (MAX9503G), V_{VIDIN} = 143mV (MAX9503M). Typical values are at T_A = +25°C, unless otherwise noted.)

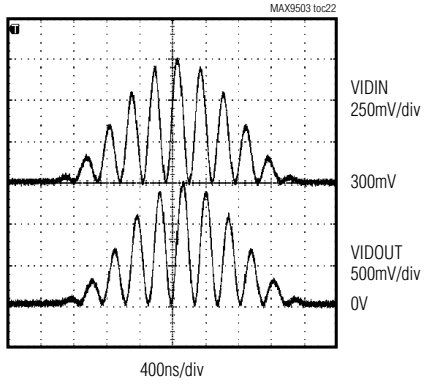
MAX9503G 2T RESPONSE



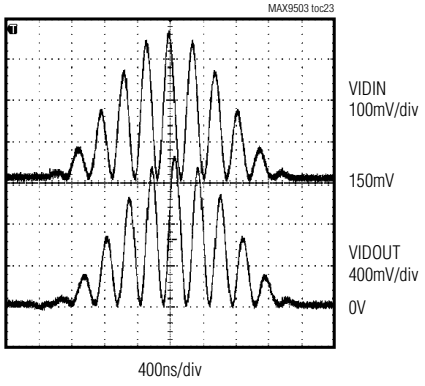
MAX9503M 2T RESPONSE



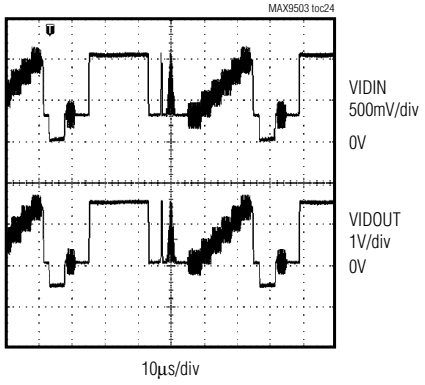
MAX9503G 12.5T RESPONSE



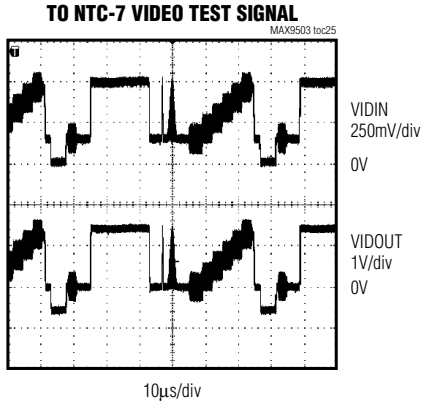
MAX9503M 12.5T RESPONSE



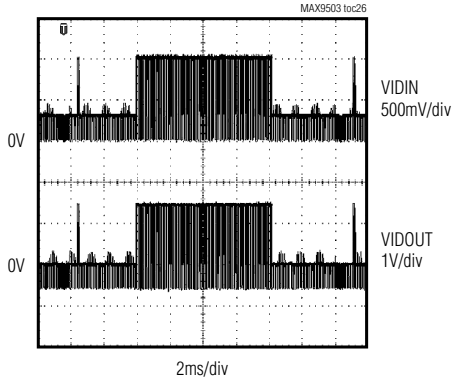
MAX9503G OUT RESPONSE TO NTC-7 VIDEO TEST SIGNAL



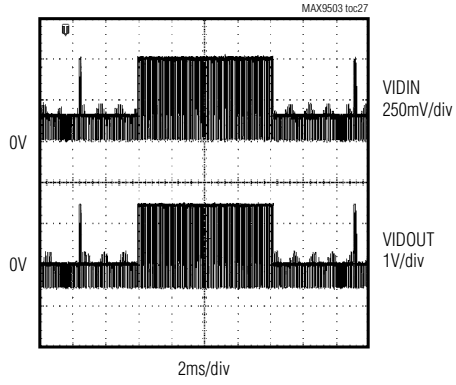
MAX9503M OUT RESPONSE TO NTC-7 VIDEO TEST SIGNAL



MAX9503G OUT RESPONSE TO BLACK-TO-WHITE VIDEO TEST SIGNAL



MAX9503M OUT RESPONSE TO BLACK-TO-WHITE VIDEO TEST SIGNAL



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DirectDrive Video Amplifier with Reconstruction Filter

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Pin Description

PIN		NAME	FUNCTION
QSOP	TQFN		
1	15	V_{SS}	Negative Power Supply. Connect to CPV_{SS}.
2	16	CPV_{SS}	Charge-Pump Negative Power Supply. Bypass with a 1µF capacitor to CPGND.
3	1	C1N	Charge-Pump Flying Capacitor Negative Terminal. Connect a 1µF capacitor from C1P to C1N.
4	2	CPGND	Charge-Pump Power Ground
5	3	C1P	Charge-Pump Flying Capacitor Positive Terminal. Connect a 1µF capacitor from C1P to C1N.
6	4	CPV_{DD}	Charge-Pump Positive Power Supply. Bypass with a 1µF capacitor to CPGND.
7	5	BIAS	Common-Mode Voltage. Connect a 100kΩ resistor from BIAS to SGND.
8	6	SGND	Signal Ground. Connect to GND.
9	7	VIDIN	Video Input
10, 14, 15	8, 12, 14	N.C.	No Connection. Not internally connected. Connect to SGND.
11	9	$\overline{SHDN}$	Active-Low Shutdown. Connect to V_{DD} for normal operation.
12	10	GND	Ground. Connect to SGND.
13	11	V_{DD}	Positive Power Supply. Bypass with a 1µF capacitor to SGND.
16	13	VIDOUT	Video Output
—	EP	EP	Exposed Paddle. Connect to GND.

Detailed Description

The MAX9503 completely eliminates the need for capacitors in the video output by using Maxim's DirectDrive technology that includes an inverting charge pump and linear regulator. The charge pump and linear regulator create a clean negative supply allowing the amplifier output to swing below ground. The amplifier output can swing both positive and negative so that the video signal black level can be placed at ground. The MAX9503 features a six-pole, Butterworth filter to perform reconstruction filtering on the video input signal from the DAC.

DirectDrive

Background

Integrated video filter/amplifier circuits operating from a single, positive supply usually create video output signals that are level-shifted above ground to keep the signal within the linear range of the output amplifier. For applications in which the positive DC level shift of the video signal is not acceptable, a series capacitor can be inserted in the output connection in an attempt to eliminate the positive DC level shift. The series capacitor cannot truly level shift a video signal because the average level of the video varies with picture content. The series capacitor biases the video output signal around ground, but the actual level of the video signal can vary significantly depending upon the RC time constant and the picture content.

The series capacitor creates a highpass filter. Since the lowest frequency in video is the frame rate, which can be between 24Hz and 30Hz, the pole of the highpass filter should ideally be an order of magnitude lower in frequency than the frame rate. Therefore, the series capacitor must be very large, typically from 220µF to 3000µF. For space-constrained equipment, the series capacitor is unacceptable. Changing from a single series capacitor to a SAG network that requires two smaller capacitors can only reduce space and cost slightly.

The series capacitor in the usual output connection also prevents damage to the output amplifier if the connector is shorted to a supply or to ground. While the output connection of the MAX9503 does not have a series capacitor, the MAX9503 will not be damaged if the connector is shorted to a supply or to ground (see the *Short-Circuit Protection* section).

Video Amplifier

Typically, the black level of the video signal created by the video DAC is around 300mV. The MAX9503 shifts the black level to ground at the output so that the active video is above ground, and sync is below ground. The amplifier needs a negative supply for its output stage to remain in its linear region when driving sync below ground.

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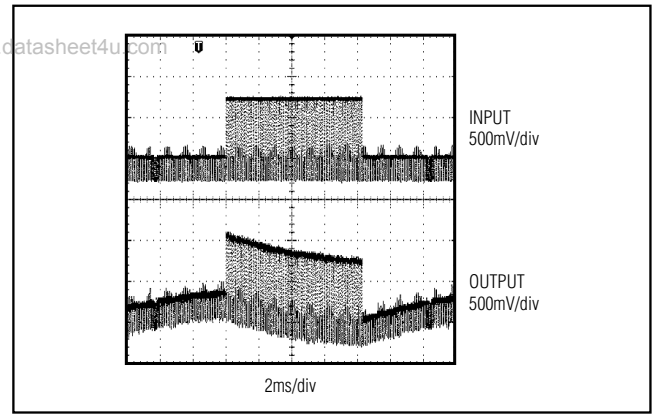

Figure 1. AC-Coupled Output

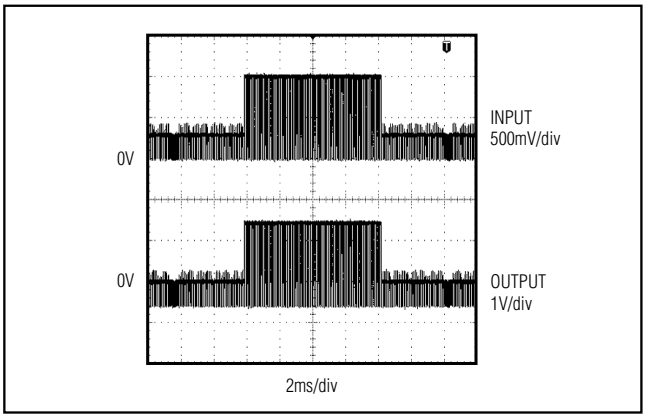

Figure 2. DirectDrive Output

The MAX9503 has an integrated charge pump and linear regulator to create a low-noise negative supply from the positive supply voltage. The charge pump inverts the positive supply to create a raw negative voltage that is then fed into the linear regulator, which outputs -2V. The linear regulator filters out the charge-pump noise.

Comparison Between DirectDrive Output vs. AC-Coupled Output

The actual level of the video signal varies less with a DirectDrive output than an AC-coupled output. The video signal average can change greatly depending upon the picture content. With an AC-coupled output, the average will change according to the time constant formed by the series capacitor and series resistance (usually 150Ω). For example, Figure 1 shows an AC-coupled video signal alternating between a completely black screen and a completely white screen. Notice the excursion of the video signal as the screen changes.

With the DirectDrive amplifier, the black level is held at ground. The video signal is constrained between -0.3V to +0.7V. Figure 2 shows the video signal from a DirectDrive amplifier with the same input signal as the AC-coupled system.

Video Reconstruction Filter

Before the video signal from the DAC can be amplified, it must be lowpass filtered to smooth the steps and to reduce the spikes created whenever the DAC output changes value. In the frequency domain, the steps and spikes cause images of the video signal to appear at multiples of the sampling clock. The MAX9503 contains a six-pole Butterworth lowpass filter. The passband extends to 5.5MHz, and the minimum attenuation is 35dB at 27MHz.

Short-Circuit Protection

The MAX9503 typical application circuit includes a 75Ω back-termination resistor that limits short-circuit current if an external short is applied to the video output. The MAX9503 features internal output, short-circuit protection to prevent device damage in prototyping and applications where the amplifier output can be directly shorted.

Shutdown

The MAX9503 features a low-power shutdown mode for battery-powered/portable applications. Shutdown reduces the quiescent current to less than 10nA. Connecting $\overline{SHDN}$ to ground (SGND) disables the outputs and places the MAX9503 into a low-power shutdown mode. In shutdown mode, the amplifier, charge pump, and linear regulator are turned off and the video output impedance is 4kΩ.

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DirectDrive Video Amplifier with Reconstruction Filter

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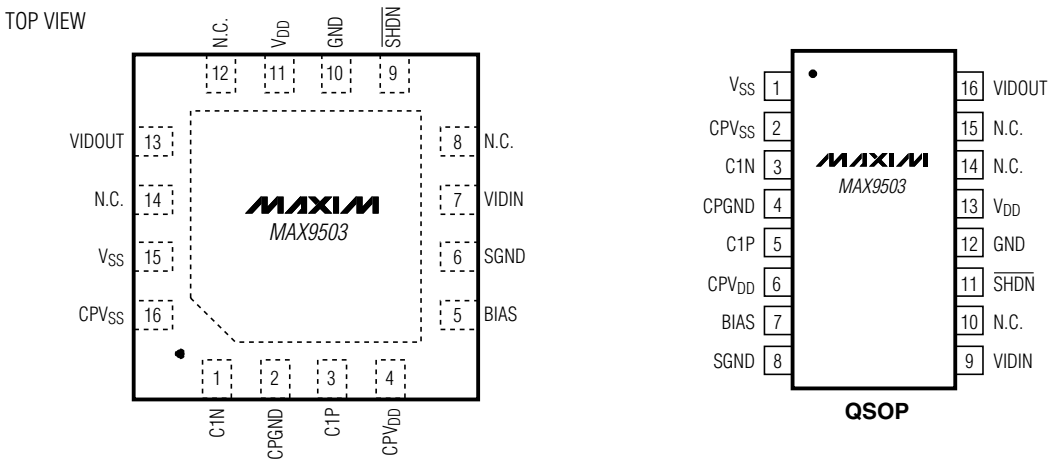
Applications Information

Power-Supply Bypassing and Ground Management

The MAX9503 operates from a 2.7V to 3.6V single supply and requires proper layout and bypassing. For the best performance, place the components as close to the device as possible.

Proper grounding improves performance and prevents any switching noise from coupling into the video signal. Connect GND and SGND together at a single point on the PC board. Route all traces that carry switching transients away from SGND. Return SGND to the lowest impedance ground available. Route CPGND and all traces carrying switching transients away from SGND, GND, and other traces and components in the video signal path. Bypass the analog supply (V_{DD}) with a 1µF capacitor to SGND, placed as close to the device as possible. Bypass the charge-pump supply (CPV_{DD}) with a 1µF capacitor to CPGND, placed as close to the device as possible. Connect CPV_{SS} to V_{SS} and bypass with a 1µF capacitor to CPGND as close to the device as possible.

Pin Configurations



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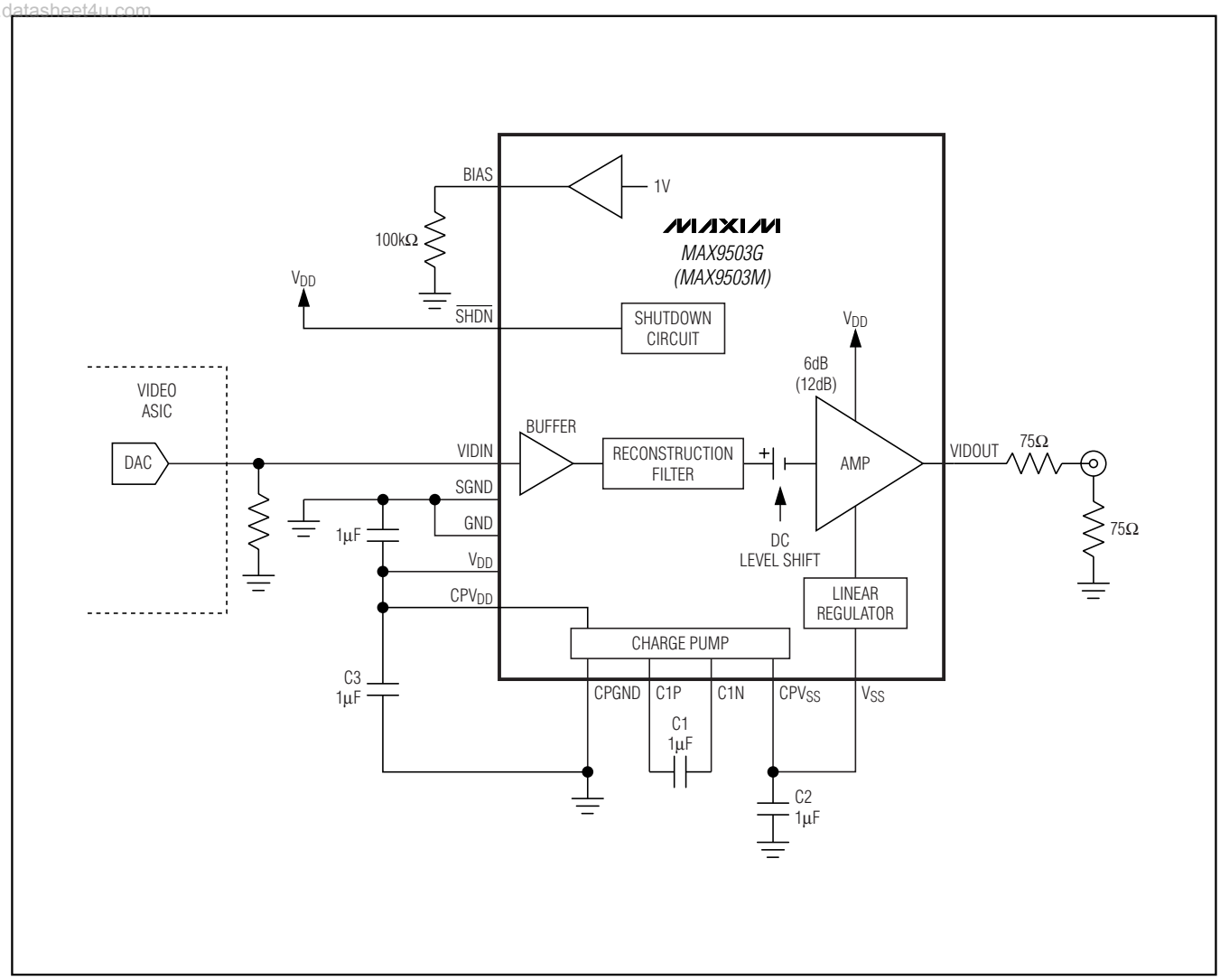
Functional Diagram/Typical Operating Circuit



Chip Information

PROCESS: BICMOS

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DirectDrive Video Amplifier with Reconstruction Filter

Package Information

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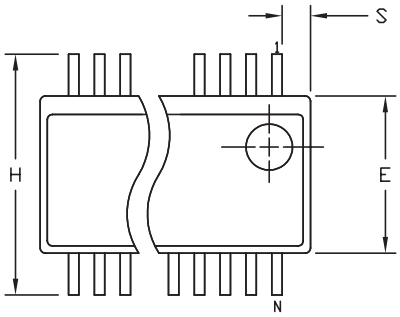
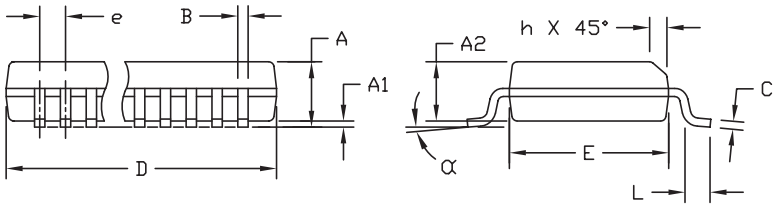
(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)

QSOP.EPS



DIM	INCHES MIN	INCHES MAX	MILLIMETERS MIN	MILLIMETERS MAX
A	.061	.068	1.55	1.73
A1	.004	.0098	0.102	0.249
A2	.055	.061	1.40	1.55
B	.008	.012	0.20	0.30
C	.0075	.0098	0.191	0.249
D	SEE VARIATIONS			
E	.150	.157	3.81	3.99
e	.025 BSC		0.635 BSC	
H	.230	.244	5.84	6.20
h	.010	.016	0.25	0.41
L	.016	.035	0.41	0.89
N	SEE VARIATIONS			
α	0°	8°	0°	8°



VARIATIONS:

	INCHES MIN.	INCHES MAX.	MILLIMETERS MIN.	MILLIMETERS MAX.	N	
D	.189	.196	4.80	4.98	16	AB
S	.0020	.0070	0.05	0.18		
D	.337	.344	8.56	8.74	20	AD
S	.0500	.0550	1.270	1.397		
D	.337	.344	8.56	8.74	24	AE
S	.0250	.0300	0.635	0.762		
D	.386	.393	9.80	9.98	28	AF
S	.0250	.0300	0.635	0.762		

NOTES:
1). D & E DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
2). MOLD FLASH OR PROTRUSIONS NOT TO EXCEED .006″ PER SIDE.
3). CONTROLLING DIMENSIONS: INCHES.
4). MEETS JEDEC MO137.

DALLAS SEMICONDUCTOR MAXIM
PROPRIETARY INFORMATION

TITLE:
PACKAGE OUTLINE, QSOP .150", .025" LEAD PITCH

APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0055	E	1/1

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Package Information (continued)

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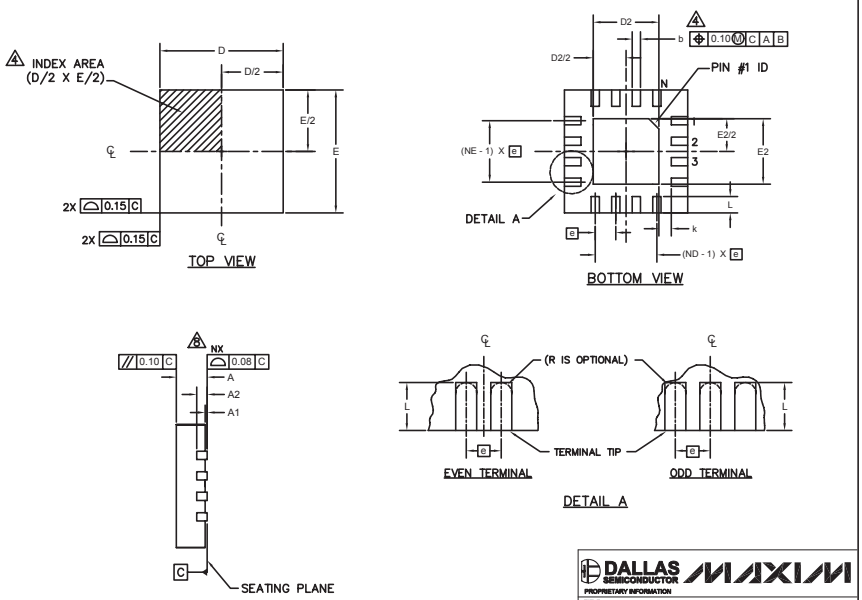
(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to **www.maxim-ic.com/packages**.)

12x16L QFN THIN.EPS



DALLAS SEMICONDUCTOR MAXIM
PROPRIETARY INFORMATION

TITLE: PACKAGE OUTLINE
12, 16L, THIN QFN, 3x3x0.8mm

APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0136	E	1/2

PKG	12L 3x3			16L 3x3		
REF.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.
A	0.70	0.75	0.80	0.70	0.75	0.80
b	0.20	0.25	0.30	0.20	0.25	0.30
D	2.90	3.00	3.10	2.90	3.00	3.10
E	2.90	3.00	3.10	2.90	3.00	3.10
e	0.50 BSC.			0.50 BSC.		
L	0.45	0.55	0.65	0.30	0.40	0.50
N	12			16		
ND	3			4		
NE	3			4		
A1	0	0.02	0.05	0	0.02	0.05
A2	0.20 REF			0.20 REF		
k	0.25	-	-	0.25	-	-

EXPOSED PAD VARIATIONS

PKG. CODES	D2			E2			PIN ID	JEDEC	DOWN BONDS ALLOWED
	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.			
T1233-1	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-1	NO
T1233-3	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-1	YES
T1633-1	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	NO
T1633-2	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	YES
T1633F-3	0.65	0.80	0.95	0.65	0.80	0.95	0.225 x 45°	WEED-2	N/A
T1633-4	0.95	1.10	1.25	0.95	1.10	1.25	0.35 x 45°	WEED-2	NO

NOTES:

1. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.
2. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES.
3. N IS THE TOTAL NUMBER OF TERMINALS.
4. THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JESD 95-1 SPP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE.
5. DIMENSION b APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.20 mm AND 0.25 mm FROM TERMINAL TIP.
6. ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY.
7. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.
8. COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.
9. DRAWING CONFORMS TO JEDEC MO220 REVISION C.

DALLAS SEMICONDUCTOR MAXIM
PROPRIETARY INFORMATION

TITLE: PACKAGE OUTLINE
12, 16L, THIN QFN, 3x3x0.8mm

APPROVAL	DOCUMENT CONTROL NO.	REV.	
	21-0136	E	2/2

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