LM158/LM258/LM358/LM2904 Low Power Dual Operational Amplifiers

General Description

The LM158 series consists of two independent, high gain, internally frequency compensated operational amplifiers which were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage.

Application areas include transducer amplifiers, dc gain blocks and all the conventional op amp circuits which now can be more easily implemented in single power supply systems. For example, the LM158 series can be directly operated off of the standard +5V power supply voltage which is used in digital systems and will easily provide the required interface electronics without requiring the additional ±15V power supplies.

The LM358 is also available in a chip sized package (8-Bump micro SMD) using National's micro SMD package technology.

Unique Characteristics

- In the linear mode the input common-mode voltage range includes ground and the output voltage can also swing to ground, even though operated from only a single power supply voltage.
- The unity gain cross frequency is temperature compensated.
- The input bias current is also temperature compensated.

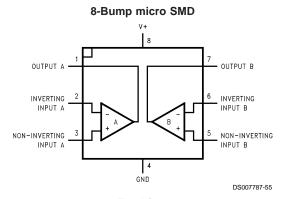
Advantages

- Two internally compensated op amps
- Eliminates need for dual supplies
- Allows direct sensing near GND and V_{OUT} also goes to GND
- Compatible with all forms of logic
- Power drain suitable for battery operation
- Pin-out same as LM1558/LM1458 dual op amp

Features

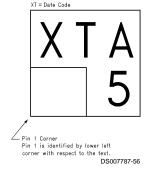
- Available in 8-Bump micro SMD chip sized package, (See AN-1112)
- Internally frequency compensated for unity gain
- Large dc voltage gain: 100 dB
- Wide bandwidth (unity gain): 1 MHz (temperature compensated)
- Wide power supply range:
 - Single supply: 3V to 32V
 - or dual supplies: ±1.5V to ±16V
- Very low supply current drain (500 µA) essentially independent of supply voltage
- Low input offset voltage: 2 mV
- Input common-mode voltage range includes ground
- Differential input voltage range equal to the power supply voltage
- Large output voltage swing: 0V to V+- 1.5V

Connection Diagrams



Top View (Bump Side Down)

micro SMD Marking Orientation

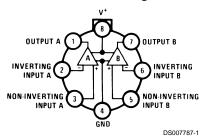


Bumps are numbered counter-clockwise

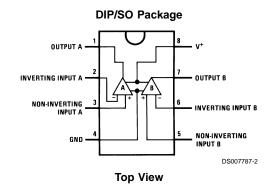
Top View

Connection Diagrams (Continued)

Metal Can Package



Top View



Ordering Information

Dookogo		NCC Descripe				
Package	−55°C to 125°C	−25°C to 85°C	0°C to 70°C	-40°C to 85°C	NSC Drawing	
SO-8			LM358AM LM358M	LM2904M	M08A	
8-Pin Molded DIP			LM358AN LM358N	LM2904N	N08E	
8-Pin Ceramic DIP	LM158AJ/883(Note 1) LM158J/883(Note 1) LM158J LM158AJLQML(Note 2) LM158AJQMLV(Note 2)				J08A	
TO-5, 8-Pin Metal Can	LM158AH/883(Note 1) LM158H/883(Note 1) LM158AH LM158H LM158AHLQML(Note 2) LM158AHLQMLV(Note 2)	LM258H	LM358H		H08C	
8-Bump micro SMD			LM358BP LM358BPX		BPA08AAA	

Note 1: LM158 is available per SMD #5962-8771001

LM158A is available per SMD #5962-8771002

Note 2: See STD Mil DWG 5962L87710 for Radiation Tolerant Devices

Absolute Maximum Ratings (Note 11)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

	LM158/LM258/LM358	LM2904
	LM158A/LM258A/LM358A	
Supply Voltage, V ⁺	32V	26V
Differential Input Voltage	32V	26V
Input Voltage	-0.3V to +32V	-0.3V to $+26V$
Power Dissipation (Note 3)		
Molded DIP	830 mW	830 mW
Metal Can	550 mW	
Small Outline Package (M)	530 mW	530 mW
micro SMD	435mW	
Output Short-Circuit to GND		
(One Amplifier) (Note 4)		
$V^+ \le 15V$ and $T_A = 25^{\circ}C$	Continuous	Continuous
Input Current ($V_{IN} < -0.3V$) (Note 5)	50 mA	50 mA
Operating Temperature Range		
LM358	0°C to +70°C	-40°C to +85°C
LM258	−25°C to +85°C	
LM158	−55°C to +125°C	
Storage Temperature Range	−65°C to +150°C	-65°C to +150°C
Lead Temperature, DIP		
(Soldering, 10 seconds)	260°C	260°C
Lead Temperature, Metal Can		
(Soldering, 10 seconds)	300°C	300°C
Soldering Information		
Dual-In-Line Package		
Soldering (10 seconds)	260°C	260°C
Small Outline Package		
Vapor Phase (60 seconds)	215°C	215°C
Infrared (15 seconds)	220°C	220°C
See AN-450 "Surface Mounting Methods and T surface mount devices.	heir Effect on Product Reliability" for other methods	of soldering
ESD Tolerance (Note 12)	250V	250V

Electrical Characteristics

 $V^+ = +5.0V$, unless otherwise stated

Parameter	Conditions	LM158A			LM358A			LM158/LM258			Units
		Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	
Input Offset Voltage	(Note 7), T _A = 25°C		1	2		2	3		2	5	mV
Input Bias Current	$I_{IN(+)}$ or $I_{IN(-)}$, $T_A = 25^{\circ}C$,		20	50		45	100		45	150	nA
	V _{CM} = 0V, (Note 8)										
Input Offset Current	$I_{IN(+)} - I_{IN(-)}, V_{CM} = 0V, T_A = 25^{\circ}C$		2	10		5	30		3	30	nA
Input Common-Mode	V ⁺ = 30V, (Note 9)	0		V ⁺ –1.5	0		V ⁺ –1.5	0		V ⁺ -1.5	V
Voltage Range	(LM2904, $V^+ = 26V$), $T_A = 25^{\circ}C$										
Supply Current	Over Full Temperature Range										
	R _L = ∞ on All Op Amps										
	V ⁺ = 30V (LM2904 V ⁺ = 26V)		1	2		1	2		1	2	mA
	V ⁺ = 5V		0.5	1.2		0.5	1.2		0.5	1.2	mA

Electrical Characteristics

V⁺ = +5.0V, unless otherwise stated

Parameter	Conditions	LM358				Units		
		Min	Тур	Max	Min	Тур	Max	
Input Offset Voltage	(Note 7) , T _A = 25°C		2	7		2	7	mV
Input Bias Current	$I_{IN(+)}$ or $I_{IN(-)}$, $T_A = 25^{\circ}C$,		45	250		45	250	nA
	V _{CM} = 0V, (Note 8)							
Input Offset Current	$I_{IN(+)} - I_{IN(-)}, V_{CM} = 0V, T_A = 25^{\circ}C$		5	50		5	50	nA
Input Common-Mode	V ⁺ = 30V, (Note 9)	0		V ⁺ –1.5	0		V ⁺ –1.5	V
Voltage Range	(LM2904, $V^+ = 26V$), $T_A = 25^{\circ}C$							
Supply Current	Over Full Temperature Range							
	R _L = ∞ on All Op Amps							
	V ⁺ = 30V (LM2904 V ⁺ = 26V)		1	2		1	2	mA
	V ⁺ = 5V		0.5	1.2		0.5	1.2	mA

Electrical Characteristics

 $V^+ = +5.0V$, (Note 6), unless otherwise stated

Paramete	or.	Conditions		LM158	BA	LM358A			LM158/LM258			Units
Faranieu	# 1			Тур	Max	Min	Тур	Max	Min	Тур	Max	
Large Signal Vo	ltage	V ⁺ = 15V, T _A = 25°C,										
Gain		$R_L \ge 2 \text{ k}\Omega$, (For $V_O = 1V$	50	100		25	100		50	100		V/mV
		to 11V)										
Common-Mode		$T_A = 25^{\circ}C$,	70	85		65	85		70	85		٩D
Rejection Ratio		$V_{CM} = 0V$ to $V^+-1.5V$	/0	85		05	85		/0	85		dB
Power Supply		V ⁺ = 5V to 30V										
Rejection Ratio		(LM2904, V ⁺ = 5V	65	100		65	100		65	100		dB
		to 26V), T _A = 25°C										
Amplifier-to-Amp	olifier	f = 1 kHz to 20 kHz, T _A = 25°C		-120			400			400		٦D
Coupling		(Input Referred), (Note 10)		-120	-120		-120			–120		dB
Output Current	Source	$V_{IN}^{+} = 1V,$										
		$V_{IN}^- = 0V,$	20	40		20	40		20	40		m Λ
		V ⁺ = 15V,	20	40		20	40		20	40		mA
		$V_{O} = 2V, T_{A} = 25^{\circ}C$										
	Sink	$V_{IN}^{-} = 1V, V_{IN}^{+} = 0V$										
		V ⁺ = 15V, T _A = 25°C,	10	20		10	20		10	20		mA
		$V_O = 2V$										
		$V_{IN}^- = 1V$,										
		$V_{IN}^{+} = 0V$	12	50		12	50		12	50		
		$T_A = 25^{\circ}C, V_O = 200 \text{ mV},$	12	50		12	50		12	50		μA
		V ⁺ = 15V										
Short Circuit to	Ground	T _A = 25°C, (Note 4),		40	60		40	60		40	60	mΛ
		V+ = 15V		40	60		40	60		40	60	mA
Input Offset Volt	age	(Note 7)			4			5			7	mV
Input Offset Volt	age	$R_S = 0\Omega$		7	15		7	20		7		μV/°C
Drift				,	13		,	20		,		μν/ Ο
Input Offset Cur	rent	$I_{IN(+)} - I_{IN(-)}$			30			75			100	nA
Input Offset Cur	rent	$R_S = 0\Omega$		10	200		10	300		10		pA/°C
Drift				10	200		10	300		10		μ.ν. C
Input Bias Curre	nt	I _{IN(+)} or I _{IN(-)}		40	100		40	200		40	300	nA
Input Common-I	Mode	V ⁺ = 30 V, (Note 9)	0		V+-2	0		V+-2	0		V+-2	V
Voltage Range		(LM2904, V ⁺ = 26V)	0		v –2	"		v –2	"		v –2	V

Electrical Characteristics (Continued)

 $V^+ = +5.0V$, (Note 6), unless otherwise stated

Daramete	- W	Condition		LM158A			LM358A			LM158/LM258			Units
Parameter		Condition	15	Min	Тур	Max	Min	Тур	Max	Min	Тур	Max	
Large Signal Vo	Itage	V ⁺ = +15V											
Gain		(V _O = 1V to 11V)		25			15			25			V/mV
		$R_L \ge 2 k\Omega$											
Output	V _{OH}	V+ = +30V	$R_L = 2 k\Omega$	26			26			26			V
Voltage		(LM2904, V ⁺ = 26V)	$R_L = 10 \text{ k}\Omega$	27	28		27	28		27	28		V
Swing	V _{OL}	$V^{+} = 5V, R_{L} = 10 \text{ k}\Omega$			5	20		5	20		5	20	mV
Output Current	Source	$V_{IN}^{+} = +1V, V_{IN}^{-} = 0V$	/,	10	20		10	20		10	20		mA
		V ⁺ = 15V, V _O = 2V		10	20		10	20		10	20		111/4
	Sink			10	15		5	8		5	8		mA
		$V^{+} = 15V, V_{O} = 2V$		10	13		J	0			0		ША

Electrical Characteristics

 $V^+ = +5.0V$, (Note 6), unless otherwise stated

Parameter		Conditions		LM358		LM2904			Units	
		Conditions	Min	Тур	Max	Min	Тур	Max		
Large Signal Voltage		V ⁺ = 15V, T _A = 25°C,								
Gain		$R_L \ge 2 \text{ k}\Omega$, (For $V_O = 1V$	25	100		25	100		V/mV	
		to 11V)								
Common-Mode		$T_A = 25^{\circ}C$	65	85		50	70		dB	
Rejection Ratio		$V_{CM} = 0V \text{ to } V^{+}-1.5V$	0.5	00		30	70		uБ	
Power Supply		V ⁺ = 5V to 30V								
Rejection Ratio		(LM2904, V ⁺ = 5V	65	100		50	100		dB	
		to 26V), T _A = 25°C								
Amplifier-to-Amplifier		f = 1 kHz to 20 kHz, T _A = 25°C		-120			-120		dB	
Coupling		(Input Referred), (Note 10)		-120			-120		uБ	
Output Current	Source	$V_{IN}^{+} = 1V,$								
		$V_{IN}^- = 0V,$	20	40		20	40		mA	
		V ⁺ = 15V,	20	40		20	40		ША	
		$V_{O} = 2V, T_{A} = 25^{\circ}C$								
	Sink	$V_{IN}^{-} = 1V, V_{IN}^{+} = 0V$								
		$V^{+} = 15V, T_{A} = 25^{\circ}C,$	10	20		10	20		mA	
		$V_O = 2V$								
		$V_{IN}^- = 1V,$								
		$V_{IN}^{+} = 0V$	12	50		12	50		μA	
		$T_A = 25^{\circ}C, V_O = 200 \text{ mV},$	12	30		12	30		μΑ	
		V ⁺ = 15V								
Short Circuit to Grou	nd	$T_A = 25^{\circ}C$, (Note 4),		40	60		40	60	mA	
		V ⁺ = 15V		40	60		40	00	IIIA	
Input Offset Voltage		(Note 7)			9			10	mV	
Input Offset Voltage		$R_S = 0\Omega$		7			7		μV/°C	
Drift				,			,		μν/ Ο	
Input Offset Current		$I_{IN(+)} - I_{IN(-)}$			150		45	200	nA	
Input Offset Current		$R_S = 0\Omega$		10			10		pA/°C	
Drift				10			10		pA/ C	
Input Bias Current		I _{IN(+)} or I _{IN(-)}		40	500		40	500	nA	
Input Common-Mode		V ⁺ = 30 V, (Note 9)	0		V+-2	0		V+ -2	V	
Voltage Range		(LM2904, V ⁺ = 26V)			v –2	U		v –2	V	

Electrical Characteristics (Continued)

 $V^+ = +5.0V$, (Note 6), unless otherwise stated

Parameter		Conditions			LM358		LM2904			Units
				Min	Тур	Max	Min	Тур	Max	
Large Signal Voltage		V ⁺ = +15V								
Gain		$(V_O = 1V \text{ to } 11V)$		15			15			V/mV
		$R_L \ge 2 \text{ k}\Omega$								
Output	V _{OH}	V+ = +30V	$R_L = 2 k\Omega$	26			22			V
Voltage		(LM2904, V ⁺ = 26V)	$R_L = 10 \text{ k}\Omega$	27	28		23	24		V
Swing	V _{OL}	$V^{+} = 5V, R_{L} = 10 \text{ k}\Omega$			5	20		5	100	mV
Output Current	Source	$V_{IN}^{+} = +1V, V_{IN}^{-} = 0V$,	10	20		10	20		mA
		$V^{+} = 15V, V_{O} = 2V$		10	20		10	20		l IIIA
	Sink	$V_{IN}^- = +1V, V_{IN}^+ = 0V,$		5	8		5	8		mA
		$V^{+} = 15V, V_{O} = 2V$		3	0		3	0		IIIA

Note 3: For operating at high temperatures, the LM358/LM358A, LM2904 must be derated based on a +125°C maximum junction temperature and a thermal resistance of 120°C/W for MDIP, 182°C/W for Metal Can, 189°C/W for Small Outline package, and 230°C/W for micro SMD, which applies for the device soldered in a printed circuit board, operating in a still air ambient. The LM258/LM258A and LM158/LM158A can be derated based on a +150°C maximum junction temperature. The dissipation is the total of both amplifiers — use external resistors, where possible, to allow the amplifier to saturate or to reduce the power which is dissipated in the integrated circuit.

Note 4: Short circuits from the output to V⁺ can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 40 mA independent of the magnitude of V⁺. At values of supply voltage in excess of +15V, continuous short-circuits can exceed the power dissipation ratings and cause eventual destruction. Destructive dissipation can result from simultaneous shorts on all amplifiers.

Note 5: This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the op amps to go to the V⁺voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which was negative, again returns to a value greater than -0.3V (at 25°C).

Note 6: These specifications are limited to $-55^{\circ}\text{C} \le \text{T}_{\text{A}} \le +125^{\circ}\text{C}$ for the LM158/LM158A. With the LM258/LM258A, all temperature specifications are limited to $-25^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$, the LM358/LM358A temperature specifications are limited to $0^{\circ}\text{C} \le \text{T}_{\text{A}} \le +70^{\circ}\text{C}$, and the LM2904 specifications are limited to $-40^{\circ}\text{C} \le \text{T}_{\text{A}} \le +85^{\circ}\text{C}$.

Note 7: $V_0 \cong 1.4V$, $R_S = 0\Omega$ with V⁺ from 5V to 30V; and over the full input common-mode range (0V to V⁺ -1.5V) at 25°C. For LM2904, V⁺ from 5V to 26V.

Note 8: The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the input lines.

Note 9: The input common-mode voltage of either input signal voltage should not be allowed to go negative by more than 0.3V (at 25°C). The upper end of the common-mode voltage range is V⁺ –1.5V (at 25°C), but either or both inputs can go to +32V without damage (+26V for LM2904), independent of the magnitude of V⁺.

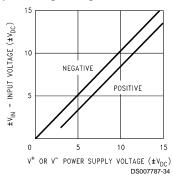
Note 10: Due to proximity of external components, insure that coupling is not originating via stray capacitance between these external parts. This typically can be detected as this type of capacitance increases at higher frequencies.

Note 11: Refer to RETS158AX for LM158A military specifications and to RETS158X for LM158 military specifications.

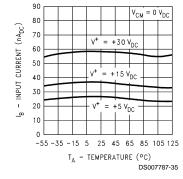
Note 12: Human body model, 1.5 k Ω in series with 100 pF.

Typical Performance Characteristics

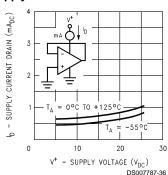
Input Voltage Range



Input Current

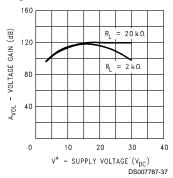


Supply Current

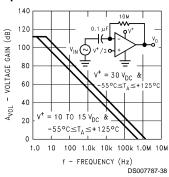


Typical Performance Characteristics (Continued)

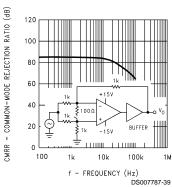
Voltage Gain



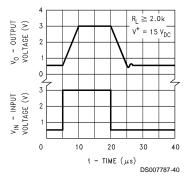
Open Loop Frequency Response



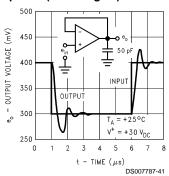
Common-Mode Rejection Ratio



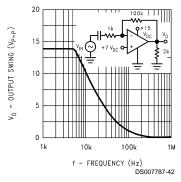
Voltage Follower Pulse Response



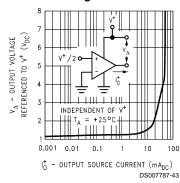
Voltage Follower Pulse Response (Small Signal)



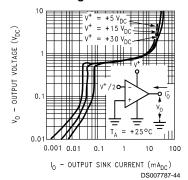
Large Signal Frequency Response



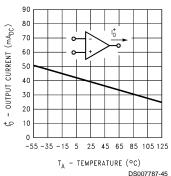
Output Characteristics Current Sourcing



Output Characteristics Current Sinking

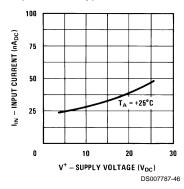


Current Limiting



Typical Performance Characteristics (Continued)

Input Current (LM2902 only)



Application Hints

The LM158 series are op amps which operate with only a single power supply voltage, have true-differential inputs, and remain in the linear mode with an input common-mode voltage of 0 $V_{\rm DC}.$ These amplifiers operate over a wide range of power supply voltage with little change in performance characteristics. At $25\,^{\circ}\text{C}$ amplifier operation is possible down to a minimum supply voltage of $2.3~V_{\rm DC}.$

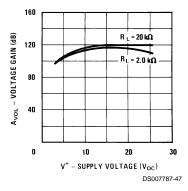
Precautions should be taken to insure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a test socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

Large differential input voltages can be easily accomodated and, as input differential voltage protection diodes are not needed, no large input currents result from large differential input voltages. The differential input voltage may be larger than V+ without damaging the device. Protection should be provided to prevent the input voltages from going negative more than $-0.3~V_{\rm DC}$ (at $25^{\circ}{\rm C}$). An input clamp diode with a resistor to the IC input terminal can be used.

To reduce the power supply current drain, the amplifiers have a class A output stage for small signal levels which converts to class B in a large signal mode. This allows the amplifiers to both source and sink large output currents. Therefore both NPN and PNP external current boost transistors can be used to extend the power capability of the basic amplifiers. The output voltage needs to raise approximately 1 diode drop above ground to bias the on-chip vertical PNP transistor for output current sinking applications.

For ac applications, where the load is capacitively coupled to the output of the amplifier, a resistor should be used, from the output of the amplifier to ground to increase the class A bias current and prevent crossover distortion. Where the load is directly coupled, as in dc applications, there is no crossover distortion.

Voltage Gain (LM2902 only)



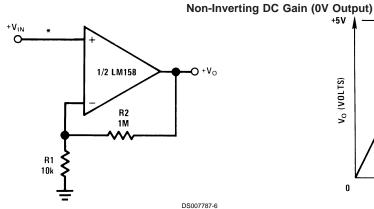
Capacitive loads which are applied directly to the output of the amplifier reduce the loop stability margin. Values of 50 pF can be accomodated using the worst-case non-inverting unity gain connection. Large closed loop gains or resistive isolation should be used if larger load capacitance must be driven by the amplifier.

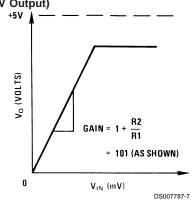
The bias network of the LM158 establishes a drain current which is independent of the magnitude of the power supply voltage over the range of 3 $V_{\rm DC}$ to 30 $V_{\rm DC}.$

Output short circuits either to ground or to the positive power supply should be of short time duration. Units can be destroyed, not as a result of the short circuit current causing metal fusing, but rather due to the large increase in IC chip dissipation which will cause eventual failure due to excessive function temperatures. Putting direct short-circuits on more than one amplifier at a time will increase the total IC power dissipation to destructive levels, if not properly protected with external dissipation limiting resistors in series with the output leads of the amplifiers. The larger value of output source current which is available at 25°C provides a larger output current capability at elevated temperatures (see typical performance characteristics) than a standard IC op amp.

The circuits presented in the section on typical applications emphasize operation on only a single power supply voltage. If complementary power supplies are available, all of the standard op amp circuits can be used. In general, introducing a pseudo-ground (a bias voltage reference of V $^+$ /2) will allow operation above and below this value in single power supply systems. Many application circuits are shown which take advantage of the wide input common-mode voltage range which includes ground. In most cases, input biasing is not required and input voltages which range to ground can easily be accommodated.

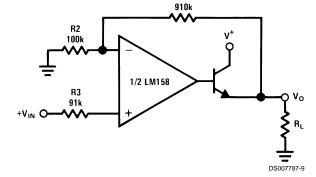
Typical Single-Supply Applications (V+ = $5.0 V_{DC}$)





*R not needed due to temperature independent $I_{\mbox{\scriptsize IN}}$

DC Summing Amplifier $(V_{\text{IN'S}} \geq 0 \ V_{\text{DC}} \ \text{and} \ V_{\text{O}} \geq 0 \ V_{\text{DC}})$ $+ V_{1} \bigcirc V_{\text{DC}}$ $+ V_{2} \bigcirc V_{\text{DC}}$ $+ V_{3} \bigcirc V_{\text{DC}}$ $+ V_{4} \bigcirc V_{\text{DC}}$ $+ V_{4} \bigcirc V_{\text{DC}}$



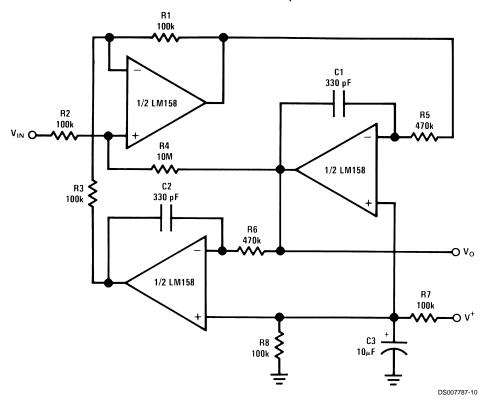
Power Amplifier

 $\begin{array}{c} V_{O}=0\ V_{DC}\ \text{for}\ V_{IN}=0\ V_{DC}\\ A_{V}=10 \end{array}$

Where: $V_O = V_1 + V_2 + V_3 + V_4$ $(V_1 + V_2) \ge (V_3 + V_4)$ to keep $V_O > 0$ V_{DC}

Typical Single-Supply Applications (V+ = 5.0 V_{DC}) (Continued)

"BI-QUAD" RC Active Bandpass Filter



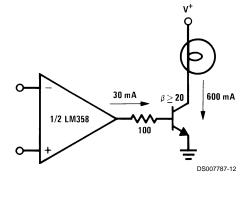
 $f_0 = 1 \text{ kHz}$ Q = 50 $A_v = 100 (40 \text{ dB})$

Fixed Current Sources

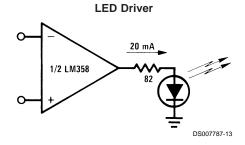
2V R3 2k 2V R1 R2 R2 R4 3k 1 mA

 $I_2 = \left(\frac{R1}{R2}\right)I_1$

Lamp Driver



Typical Single-Supply Applications $(V^+ = 5.0 V_{DC})$ (Continued)

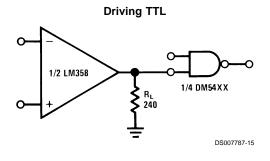


Current Monitor R1* 0.1 VL 0 R2 100 Vo R3 11 R3

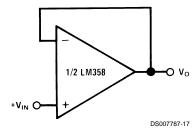
DS007787-14

$$V_O = \frac{1V (I_L)}{1A}$$

*(Increase R1 for I_L small) $V_L \le V^+$ –2V

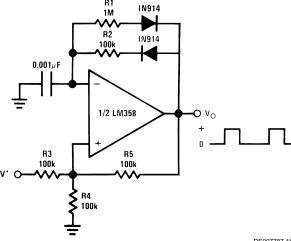


Voltage Follower



 $V_{O} = V_{IN}$

Pulse Generator

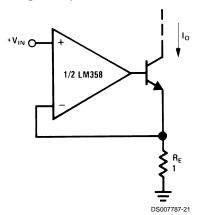


DS007787-16

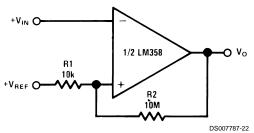
Typical Single-Supply Applications $(V^+ = 5.0 V_{DC})$ (Continued) **Squarewave Oscillator Pulse Generator** R1 30k R1 100k IN914 C 0.001μF 1/2 LM358 1/2 LM358 R4 100k R2 R4 100k 100k R3 100k R3 100k DS007787-18 DS007787-19 Low Drift Peak Detector 1/2 LM358 Z_{OUT} 1/2 LM358 **1**μ**F** (POLYCARBONATE OR Z_{IN} POLYETHYLENE) 2N929* $0.001 \mu F$ *hi β AT 100 nA 3R 3M 21_B 1/2 LM358 **AUX AMP** R IM INPUT CURRENT COMPENSATION DS007787-20 HIGH Z_{IN} LOW Z_{OUT}

Typical Single-Supply Applications $(V^+ = 5.0 V_{DC})$ (Continued)

High Compliance Current Sink

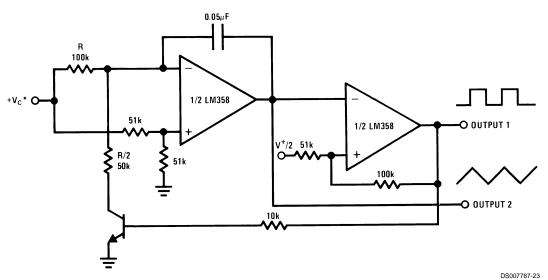


Comparator with Hysteresis



 $I_O = 1$ amp/volt V_{IN} (Increase R_E for I_O small)

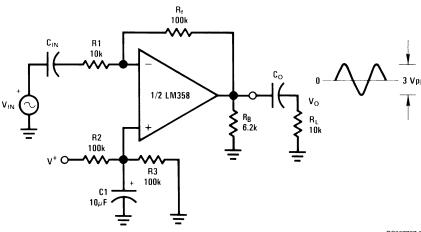
Voltage Controlled Oscillator (VCO)



WIDE CONTROL VOLTAGE RANGE: 0 $\rm V_{DC} \leq \rm V_{C} \leq 2$ (V –1.5V $_{DC})$

Typical Single-Supply Applications (V+ = 5.0 V_{DC}) (Continued)

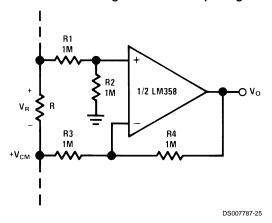
AC Coupled Inverting Amplifier



DS007787-24

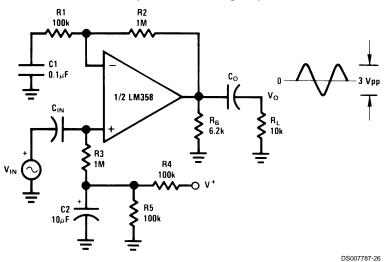
$$A_V = \frac{R_f}{R1} \text{ (As shown, } A_V = 10)$$

Ground Referencing a Differential Input Signal



Typical Single-Supply Applications $(V^+ = 5.0 V_{DC})$ (Continued)

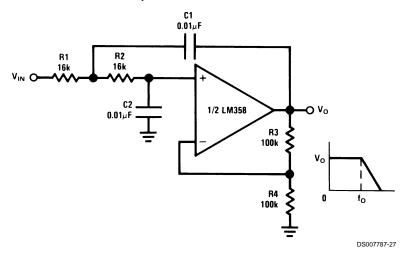
AC Coupled Non-Inverting Amplifier



$$A_V = 1 + \frac{R2}{R1}$$

 $A_v = 11$ (As Shown)

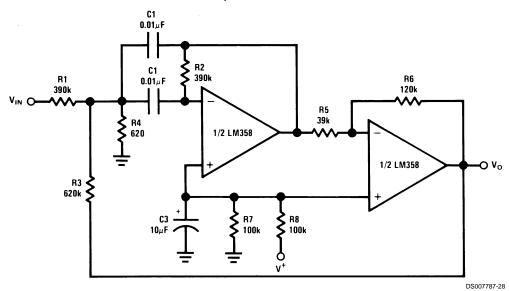
DC Coupled Low-Pass RC Active Filter



 $f_0 = 1 \text{ kHz}$ Q = 1 $A_V = 2$

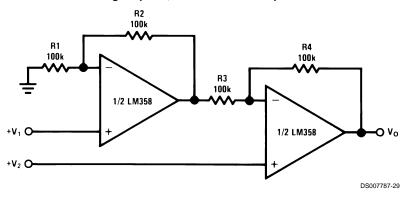
Typical Single-Supply Applications (V⁺ = 5.0 V_{DC}) (Continued)

Bandpass Active Filter



 $f_0 = 1 \text{ kHz}$ Q = 25

High Input Z, DC Differential Amplifier



For $\frac{R1}{R2} = \frac{R4}{R3}$ (CMRR depends on this resistor ratio match)

$$V_0 = 1 + \frac{R4}{R3} (V_2 - V_1)$$

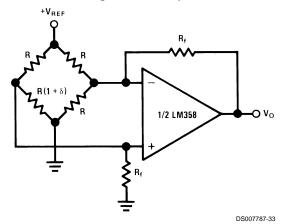
As Shown: $V_0 = 2 (V_2 - V_1)$

Typical Single-Supply Applications $(V^+ = 5.0 V_{DC})$ (Continued)

Photo Voltaic-Cell Amplifier

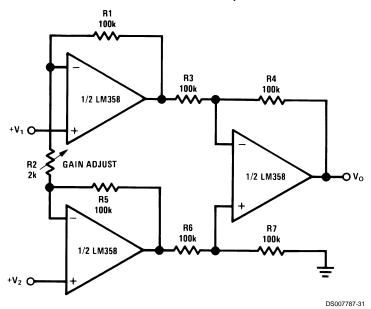
(CELL HAS OV ACROSS IT) DS007787-30

Bridge Current Amplifier



For $\delta <<$ 1 and $R_f >> R$ $V_O \cong V_{REF} \left(\frac{\delta}{2}\right) \frac{R_f}{R}$

High Input Z Adjustable-Gain DC Instrumentation Amplifier



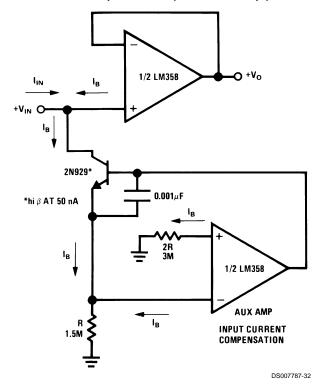
If R1 = R5 & R3 = R4 = R6 = R7 (CMRR depends on match)

$$V_0 = 1 + \frac{2R1}{R2} (V_2 - V_1)$$

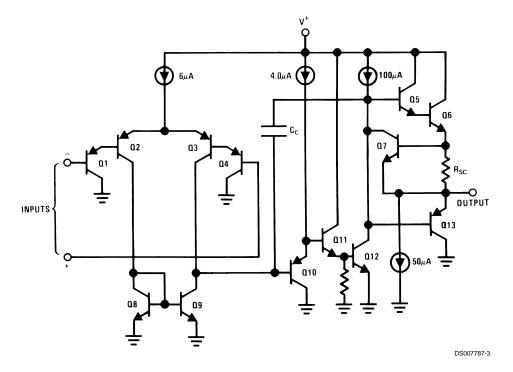
As shown $V_0 = 101 (V_2 - V_1)$

Typical Single-Supply Applications (V⁺ = 5.0 V_{DC}) (Continued)

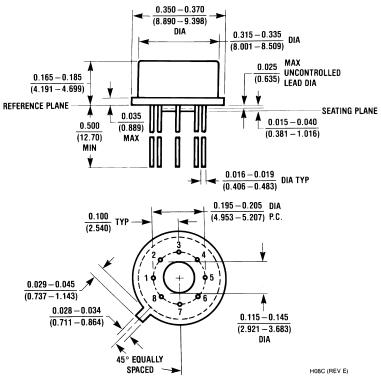
Using Symmetrical Amplifiers to Reduce Input Current (General Concept)



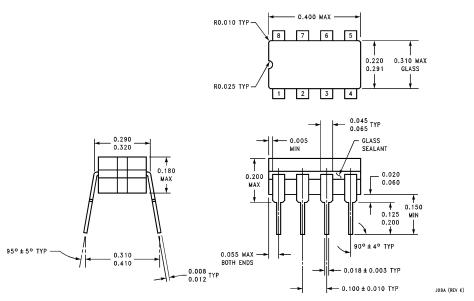
Schematic Diagram (Each Amplifier)



Physical Dimensions inches (millimeters) unless otherwise noted

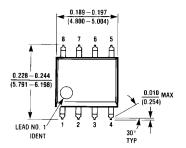


Metal Can Package (H)
Order Number LM158AH, LM158AH/883, LM158H,
LM158H/883, LM258H or LM358H
NS Package Number H08C



Cerdip Package (J)
Order Number LM158J, LM158J/883, LM158AJ or LM158AJ/883
NS Package Number J08A

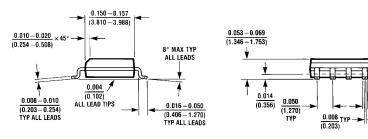
Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



0.004 - 0.010 (0.102 - 0.254)

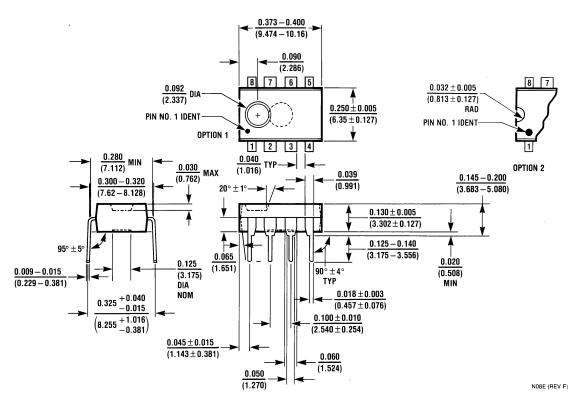
0.014 - 0.020 (0.356 - 0.508)

SEATING PLANE



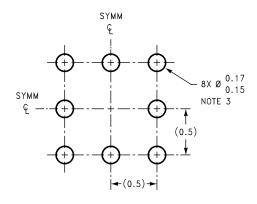
S.O. Package (M)
Order Number LM358M, LM358AM or LM2904M
NS Package Number M08A

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)

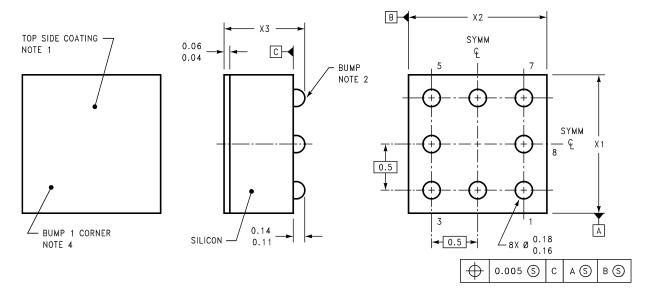


Molded Dip Package (N)
Order Number LM358AN, LM358N or LM2904N
NS Package Number N08E

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



LAND PATTERN RECOMMENDATION



DIMENSIONS ARE IN MILLIMETERS

BPA08XXX (REV A)

NOTES: UNLESS OTHERWISE SPECIFIED

- 1. EPOXY COATING
- 2. 63Sn/37Pb EUTECTIC BUMP
- 3. RECOMMEND NON-SOLDER MASK DEFINED LANDING PAD.
- 4. PIN 1 IS ESTABLISHED BY LOWER LEFT CORNER WITH RESPECT TO TEXT ORIENTATION REMAINING PINS ARE NUMBERED COUNTERCLOCKWISE.
- 5. XXX IN DRAWING NUMBER REPRESENTS PACKAGE SIZE VARIATION WHERE $\rm X_1$ IS PACKAGE WIDTH, $\rm X_2$ IS PACKAGE LENGTH AND $\rm X_3$ IS PACKAGE HEIGHT.
- 6. REFERENCE JEDEC REGISTRATION MO-211, VARIATION BC.

8-Bump micro SMD NS Package Number BPA08AAA X₁ = 1.285 X₂ = 1.285 X₃ = 0.700

Notes

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- A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.



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