MIL-PRF-38534 QUALIFIED

MSK163

(315) 701-6751



4707 Dey Road Liverpool, N.Y. 13088

FEATURES:

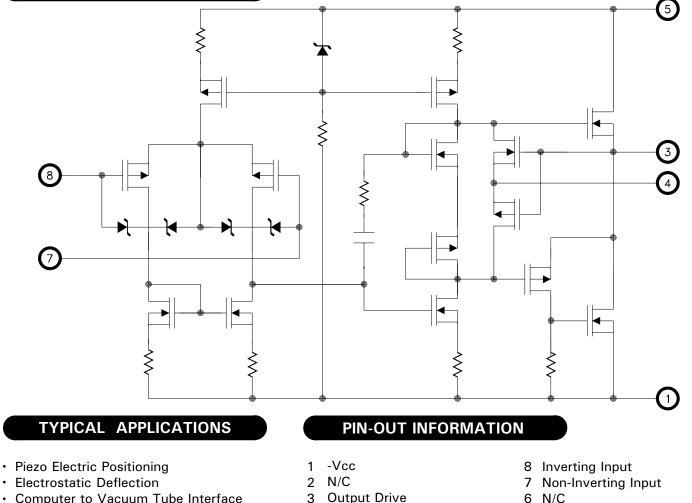
- Monolithic MOS Technology
- Low Cost
- High Voltage Operation : 350V
- Low Quiescent Current : 2mA Max.
- High Output Current : 60mA Min.
- No Second Breakdown
- High Speed : $20V/\mu S$ Typ.
- Internally Compensated For Gains >10 V/V

DESCRIPTION:

The MSK 162 is an ultra high voltage monolithic MOSFET operational amplifier ideally suited for electrostatic transducer and electrostatic deflection applications. With a total supply voltage rating of 350 volts and 60mA of available output current, the MSK 162 is also an excellent low cost choice for high voltage piezo drive circuits. The MOSFET output frees the MSK 162 from secondary breakdown limitations and power dissipation is kept to a minimum with a quiescent current rating of only 2mA. The MSK 162 is packaged in a hermetically sealed 8 pin power dip and is internally compensated for closed loop gains of 10 V/V or greater. For applications requiring heat sinking, the MSK 163 is available with bolt down tabs and is otherwise identical to the MSK 162 (see mechanical specifications). If the application calls for a gain of less than 10 V/V, refer to the MSK 158/159 data sheet.

MSK162

EQUIVALENT SCHEMATIC



Ultra High Voltage Op-Amp Applications

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Current Sense

- 6 N/C
- 5 + Vcc

ABSOLUTE MAXIMUM RATINGS

Vcc ②	Total Supply Voltage
± І оит	Output Current (within S.O.A.) 60mA
$\pm I$ outp	Output Current Peak
VIND	Input Voltage (Differential) <u>±16V</u>
Vin	Input Voltage (Common Mode) ± Vcc
TJ	Junction Temperature

Ts⊤	Storage Temperature	65°C to +150°C
Tld	Lead Temperature	300°C
Тc	Case Operating Temperature	
	(MSK162B/163B)	55°C to +125°C
	(MSK162/163)	40°C to +85°C
Rтн	Thermal Resistance (DC)	
	Junction to Case	10°C/W

ELECTRICAL SPECIFICATIONS

Parameter	Test Conditions ①	Group A	A MSK162B/163B			MSK162/163			
Parameter		Subgroup	Min.	Тур.	Max.	Min.	Тур.	Max.	Units
STATIC									
Supply Voltage Range 249		-	±50	±150	±175	±50	±150	±175	V
		1	-	±1.4	±2.0	-	±1.4	±2.0	mA
Quiescent Current	VIN=OV	2	-	±2.0	±3.0	-	-	-	mA
		3	-	±1.0	±2.1	-	-	-	mA
INPUT									
Offset Voltage	$V_{IN} = OV$	1	-	±15	± 30	-	±15	±30	mV
Offset Voltage Drift ④	VIN=OV	2,3	-	±40	±65	-	±40	-	µV/°C
Offset Voltage vs \pm Vcc $\textcircled{4}$	VIN=OV	1	-	±20	±32	-	±20	±32	$\mu V/V$
Input Bias Current ④	Vcm=0V	1,3	-	±5	±50	-	±5	±100	pА
		2	-	-	±50	-	-	-	nA
Input Impedance ④	(DC)	-	-	10 ¹¹	-	-	10 ¹¹	-	Ω
Input Capacitance ④		-	-	5	-	-	5	-	pF
Common Mode Rejection ④	$V_{CM} = \pm 90 VDC$	-	84	94	-	84	94	-	dB
Noise	1Hz≤f≤10Hz	-	-	50	-	-	50	-	μVrms
OUTPUT									
Output Voltage Swing	$IOUT = \pm 40 mA$ Peak	4	±138	±141	-	±138	±141	-	V
Output Current	Vout = MAX	4	±60	±120	-	±60	±120	-	mA
Power Bandwidth \oplus	Vout = 280Vpp	-	-	26	-	-	26	-	KHz
Resistance ④	No Load, RcL = 0Ω	-	-	150	-	-	150	-	Ω
Settling Time to 0.1% 34	10V Step	-	-	12	-	-	12	-	μS
Capacitive Load ④	Av = + 1V/V	-	10	-	-	10	-	-	nF
TRANSFER CHARACTERISTICS									
Slew Rate Cc=Open		4	10	20	-	10	20	-	V/µS
Open Loop Voltage Gain ④ F=15Hz RL=5KS		4	94	106	-	94	106	-	dB

NOTES:

- (1) Unless otherwise noted, \pm VCC = \pm 150VDC. (2) Derate maximum supply voltage 0.5V/°C below TC = + 25°C. No derating is needed above TC = 25°C. (3) AV = -10V/V measured in false summing junction circuit.
- Devices shall be capable of meeting the parameter, but need not be tested. Typical parameters are for reference only.
- (5) Industrial grade devices shall be tested to subgroups 1 and 4 unless otherwise requested.
- Military grade devices ('B' suffix) shall be 100% tested to subgroups 1,2,3 and 4.
 Subgroup 5 and 6 testing available upon request.

8 Subgroup 1,4 Tc = +25°C Subgroup 2,5 Tc = +125°C

Subgroup 3,6 TA = -55°C

(9) Electrical specifications are derated for power supply voltages less than \pm 50VDC.

CURRENT LIMIT

Current limit resistor value can be calculated as follows:

Rcl = 3/Ilim

It is recommended that the user set up the value of current limit as close as possible to the maximum expected output current to protect the amplifier. The minumum value of current limit resistance is 33 ohms. The maximum practical value is 500 ohms. Current limit will vary with case temperature. Refer to the typical performance graphs as a guide. Since load current passes through the current limit resistor, a loss in output voltage swing will occur. The following formula approximates output voltage swing reduction:

Vr = Io * Rcl

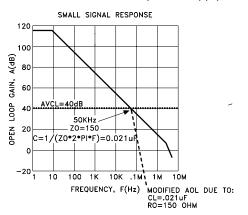
When the device is in current limit, there will be spurious oscillations present on the negative half cycle. The frequency of the oscillation is application dependent and can not be predicted. Oscillation will cease when the device comes out of current limit.

INPUT PROTECTION

Input protection curcuitry within the MSK 162/163 will clip differential input voltages greater than 16 volts. The inputs are also protected against common mode voltages up to the supply rails as well as static discharge. There are 300 ohm current limiting resistors in series with each input. These resistors may become damaged in the event the input overload is capable of driving currents above 1mA. If severe overload conditions are expected, external input current limiting resistors are recommended.

OUTPUT SNUBBER NETWORK

A 100 ohm resistor and a 330pF capacitor connected in series from the output of the amplifier to ground is recommended for applications where load capacitance is less than 330pF. For larger values of load capacitance, the output snubber network may be omitted. If loop stability becomes a problem due to excessively high load capacitance, a 100 ohm resistor may be added between the output of the amplifier and the load. A small tradeoff with bandwidth must be made in this configuration. The graph below illustrates the effect of capacitive load. Note that the compensation capacitor must have a voltage rating greater than the total rail to rail power supply voltage.

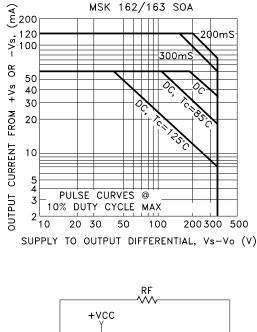


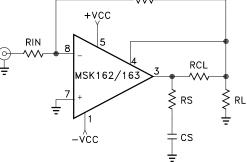
SAFE OPERATING AREA (SOA)

The MOSFET output stage of this power operational amplifier has two distinct limitations:

- 1. The current handling capability of the die metallization.
- 2. The temperature of the output MOSFET's.

NOTE: The output stage is protected against transient flyback. However, for protection against sustained, high energy flyback, external fast-recovery reverse biased diodes should be used.

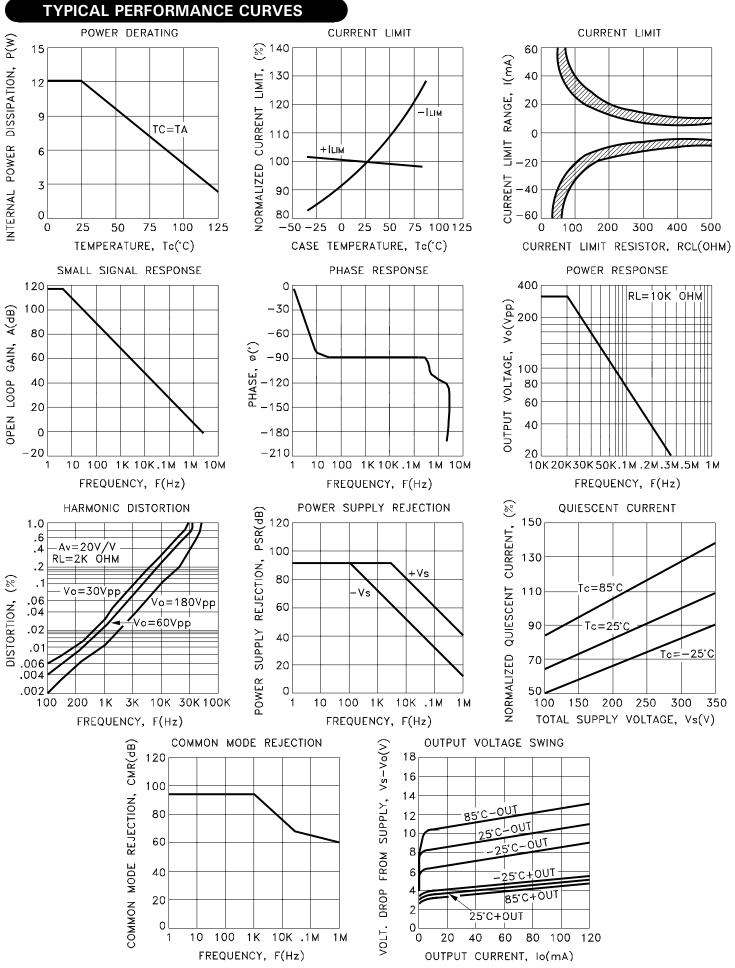


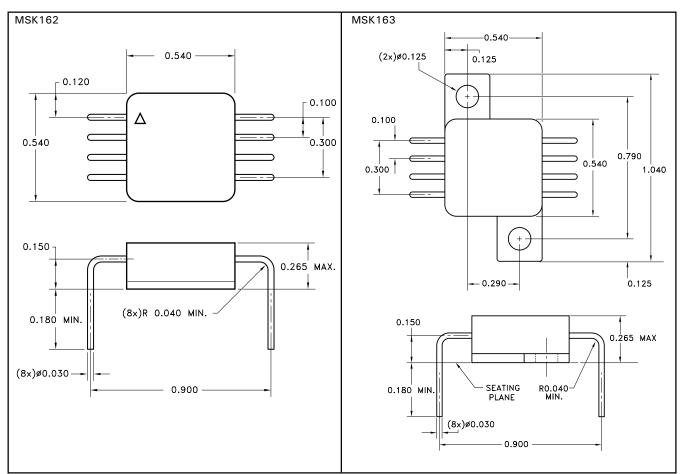


TYPICAL CONNECTION DIAGRAM

STABILITY

The MSK 162/163 are internally compensated for closed loop gains of 10 V/V or greater. The majority of applications the MSK 162/163 are used in involve gains greater than 10 V/V because the output is capable of swinging up to +/- 141V and the maximum differential input voltage is only +/- 16V. A large gain is necessary to make full use of the output voltage swing capability of the amplifier when input voltages are small. If closed loop gains of less than 10 V/V are required, refer to the MSK 158/159 data sheets. The MSK 158/159 operational amplifiers are identical to the MSK 162/163 except pins two and six are compensation pins. The user can tailor op-amp performance with the external connection of a series resistorcapacitor snubber network. An effective method of checking amplifier stability is to apply the worst case capacitive load to the output of the amplifier and drive a small signal square wave across it. If overshoot is less than 25%, the system will generally be stable.





ESD TRIANGLE INDICATES PIN 1. ALL DIMENSIONS ARE ± 0.010 INCHES UNLESS OTHERWISE LABELED.

ORDERING INFORMATION

Part Number	Screening Level
MSK162	Industrial
MSK162B	Military-Mil-PRF-38534
MSK163	Industrial
MSK163B	Military-Mil-PRF-38534

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