**TENTATIVE** 

TOSHIBA BIPOLAR DIGITAL INTEGRATED CIRCUIT SILICON MONOLITHIC

# TD7623AFN

# 3-WIRE AND I $^2$ C BUS SYSTEM, 2.3 GHz DIRECT TWO MODULUS-TYPE FREQUENCY SYNTHESIZER FOR CATV

The TD7623AFN can be combined with a micro CPU to create a highly functional frequency synthesizer. The control data conforms to 3-wire bus and standard I<sup>2</sup>C bus formats. BUS-SW can be used to easily switch for easy tuner system set-up.

#### **FEATURES**

- Direct two modulus-type frequency synthesizer
- Standard I<sup>2</sup>C bus format control with built-in read mode
- 3-wire bus 27-bit format control
- 4-bit bandswitch drive transistor
- 5-level A / D converter (when I<sup>2</sup>C bus selected)
- Frequency step : 50 kHz, 62.5 kHz, 250 kHz, and 333.3 kHz (at 4 MHz X'tal used)
- Phase lock detector
- Various function settings via program data
- Four address settings via address selector (when I<sup>2</sup>C bus selected)
- Power on reset circuit
- Flat, compact package : SSOP16 (0.65 mm pitch)
- Power on reset operation condition

Bandswitch register 1 to 4 : OFF Tuning amplifier : ON Tuning Voltage output (Vt) : High Level Charge-pump output current : ±200 μA

Phase comparator reference frequency divider ratio: 1/80

SSOP16-P-225-0.65B Weight: 0.07 g (Typ.)

These devices are easy to be damaged by high static voltage or electric fields. Note: In regards to this, please handle with care. To input sub features items.

- damage to property.

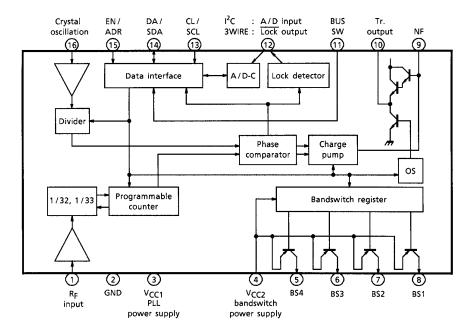
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#### **BLOCK DIAGRAM**



# **MAXIMUM RATINGS (Ta = 25°C)**

CHARACTERISTIC	SYMBOL	RATING	UNIT
Supply Voltage 1	V <sub>CC1</sub>	6.0	٧
Supply Voltage 2	V <sub>CC2</sub>	12	V
Power Consumption	$P_{D}$	560	mW
Operating Temperature	T <sub>opr</sub>	-20~85	°C
Storage Temperature	T <sub>stg</sub>	-55~150	°C

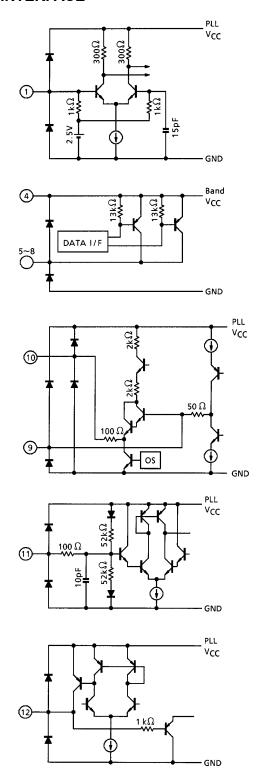
Note 1: When using the device at above Ta =  $25^{\circ}$ C, decrease the power dissipation by 4.5 mW for each increase of  $1^{\circ}$ C.

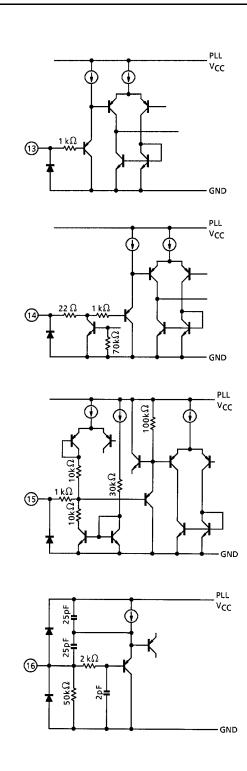
Note 2: These devices are easy to be damaged by high static voltage or electric fields. In regards to this, please handle with care.

# **RECOMMENDED SUPPLY VOLTAGE**

PIN No.	PIN NAME	MIN	TYP.	MAX	UNIT
3	V <sub>CC1</sub> : PLL Power Supply	4.5	5.0	5.5	V
4	V <sub>CC2</sub> : Band Switch Power Supply	V <sub>CC1</sub>		9.9	V

# **PIN INTERFACE**







# **ELECTRICAL CHARACTERISTICS** (Unless otherwise specified, $V_{CC1} = 5 \text{ V}$ , $V_{CC2} = 9 \text{ V}$ , $Ta = 25^{\circ}\text{C}$ )

CHARACTERISTIC	SYMBOL	TEST CIR- CUIT	TEST CONDITION	MIN	TYP.	MAX	UNIT
Supply Voltage 1	V <sub>CC1</sub>	_	_	4.5	5.0	5.5	V
Supply Current 1	I <sub>CC1</sub>	1	Bandswitch : OFF V <sub>t</sub> : OFF	24	32	40	mA
Supply Voltage 2	V <sub>CC2</sub>	_	_	V <sub>CC1</sub>	_	9.9	V
Supply Current 2	I <sub>CC2</sub> -1	1	Bandswitch : 1 Band ON I <sub>BD</sub> = 20 mA (LOAD)	ı	24	26	mA
Oupply Guilent 2	I <sub>CC2</sub> -2	'	Bandswitch : 2 Band ON I <sub>BD</sub> = 30 mA (TOTAL LOAD)	ı	38	42	IIIA
Bandswitch Drive Current	I <sub>BD</sub>	3	Maximum Drive Current / 1 port	l	_	20	mA
Bandswitch Drive Maximum LOAD	I <sub>BDMAX</sub>	3	Maximum Total Drive Current	1	_	40	mA
Bandswitch Drive Voltage Drop	V <sub>BD</sub> Sat	3	I <sub>BD</sub> = 20 mA	l	0.2	0.4	V
X'tal Operating Range	OSCfin	_	_	3.2	_	4.5	MHz
X'tal Negative Resistance	OSCR	1		1.0	1.5	_	kΩ
X'tal External Input Level	OSC <sub>in</sub>	_	3.2 MHz~4.5 MHz, Rx = 91 kΩ	250	_	1000	$mV_{p-p}$
Ratio Setting Range	N	_	15-bit counter		_	32767	Ratio
Prescaler Input Sensitivity	V <sub>inRF</sub>	2	f = 500~2300 MHz		_	+5	dBmW
Lock Output Low Voltage	$V_{LkL}$	1	(lock mode, 3-wire bus mode)	ı	_	0.4	V
Lock Output High Voltage	V <sub>LkH</sub>	1	(unlock mode, 3-wire bus mode)	4.6	_	_	V
Logic Input Low Voltage	V <sub>BsL</sub>	1	Pins 13 to 15	-0.3	_	1.5	V
Logic Input High Voltage	V <sub>BsH</sub>	1	Pins 13 to 15	2.5	_	V <sub>CC1</sub> +0.3	٧
Logic Input Current (low)	I <sub>BsL</sub>	1	Pins 13 to 15	-20	_	10	μA
Logic Input Current (high)	I <sub>BsH</sub>	1	Pins 13 to 15	-10	_	20	μΛ
BUS-SW Low Input Voltage	V <sub>BIL</sub>	1	_	0.0	_	0.8	V
BUS-SW High Input Voltage	V <sub>BIH</sub>	1		4.2	_	V <sub>CC1</sub>	v
BUS-SW Low Current (low)	I <sub>BIL</sub>	1		-200	_	_	μA
BUS-SW Low Current (high)	I <sub>BIH</sub>	1		_	_	200	μΛ
Charge Pump Output Current	lete	2	CP = [0]	±150	±200	±300	μA
Sharge Fump Output Ourrollt	I <sub>chg</sub> 2		CP = [1]	±600	±800	±1200	μ/ τ
ACK Output Voltage	V <sub>ACK</sub>	1	I <sub>SINK</sub> = 3 mA (I <sup>2</sup> C-bus mode)			0.4	V

CHARACTERISTIC	SYMBOL	TEST CIR- CUIT	TEST CONDITION	MIN	TYP.	MAX	UNIT
Set-up Time	Ts	55.1		2	_	_	
Enable Hold Time	T <sub>sL</sub>			2	_	_	
Next Enable Stop Time	T <sub>NE</sub>			6	_	_	
Next Clock Stop Time	T <sub>NC</sub>		(3-wire bus mode) Refer to data timing chart	6	_	_	μs
Clock Width	T <sub>c</sub>		, and the second	2	_	_	
Enable Set-up Time	TL			10	_	_	
Data Hold Time	T <sub>H</sub>			2	_	_	
SCL Clock Frequency	f <sub>SCL</sub>			_	_	100	kHz
Bus Free Time Between a STOP and START Condition	t <sub>BUF</sub>			4.7	_	_	
Hold Time (Repeated) START Condition	t <sub>HD;STA</sub>	_		4.0	_	_	
Low Period of the SCL Clock	t <sub>LOW</sub>			4.7	_	_	μs
High Period of the SCL Clock	tHIGH			4.0	_	_	
Set-up Time for a Repeated START Condition	tsu;sta		(I <sup>2</sup> C bus mode) Refer to data timing chart	4.7	_	_	
Data Hold Time	t <sub>HD;DAT</sub>			0	_	_	
Data Set-up Time	t <sub>SU;DAT</sub>			250	_	_	
Rise Time of both SDA and SCL Signals	t <sub>R</sub>			_		1000	ns
Fall Time of both SDA and SCL Signals	t <sub>F</sub>			_	_	300	
Set-up Time for STOP Condition	tsu;sto			4.0	_	_	μs

Fig.1 3-wire bus data timing chart (Falling edge timing)

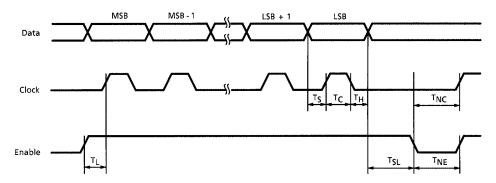
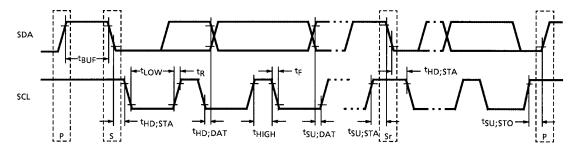


Fig.2 I<sup>2</sup>C bus data timing chart (Falling edge timing)



#### OPERATION INSTRUCTIONS

The TD7623AFN can be controlled with either the 3-wire bus or standard I<sup>2</sup>C bus.

The 3-wire bus mode, the device is controlled by 27-bit serial data.

The  $I^2C$  bus conforms to the standard  $I^2C$  bus format. The bus supports two-way bus communications control, consisting of WRITE mode where data are received and READ mode where data are transmitted. In READ mode, the voltage applied on the A/D converter input pin can be transmitted and output with 5-level resolution.

(This function is only valid when the  $I^2C$  bus is selected. When the 3-wire bus is selected, the A/D converter input pin function as the  $\overline{LOCK}$  output pin.)

Addresses can be set using the hardware bits. Three programmable addresses are supported. 3-wire bus and standard  $I^2C$  bus are switches by the voltage applied on the BUS-SW pin.

The power-on reset circuit is built in this product, and the detection voltage is designed about 1.4 V.

If it raises to voltage of operation after making it stop for a while near the voltage of a power-on reset circuit of operation at the time of starting of a power supply, a power-on reset circuit may not operate normally.

#### **FUNCTION CHART**

NAME	3-WIRE BUS MODE	I <sup>2</sup> C BUS MODE
BUS-SW	[V <sub>CC</sub> ]	[GND]
CL / SCL	CLOCK INPUT	SCL INPUT
DA / SDA	DATA INPUT	SDA IN / OUTPUT
EN / ADR	ENABLE INPUT	ADDRESS
LOCK / ADC	LOCK	ADC

#### - 3-WIRE BUS COMMUNICATIONS CONTROL -

The 3-wire bus mode, the device is controlled by 27-bit serial data.

The 3-wire bus sets the following data: (bandswitch information and programmable counter information, charge-pump current setting, reference frequency divider ratio setting, and testing item functions.)

The program frequency can be calculated in the following formula:

 $fosc = f_r \times N$ 

fosc: Program frequency

fr : Phase comparator reference frequency (Step frequency)

N : Counter total ratio

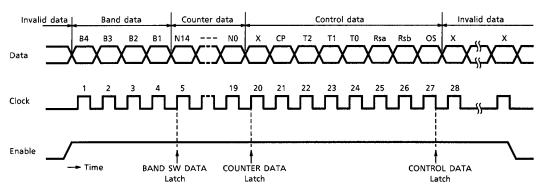


Figure 3. 3-wire bus data format

#### • 27-bit DATA TRANSMISSION

During a high level of the enable signal, the data is clocked into the register on the falling edge of the clock.

The clock number during a high level of the enable signal must be set to 27-bit or more of clock and data transmission.

The data are latched at the 27th falling edge of the clock signal, validating the previous 27-bit data. The 4-bit bandswitch data are latched at the 5th bit rising edge of the clock signal, and the data is updated.

The programmable counter data are latched at the 20th bit rising edge of the clock signal, and the data is updated.

The control data are latched at the 27th bit falling edge of the clock signal, and the data is updated. Details of the data timing, see the data timing chart. (Figure 1)

### **TEST DATA SPECIFICATIONS**

● B4~B1 : Band drive data

> [0]: OFF [1]: ON

N14~N0 : Programmable divider data CP: Charge pump output current

[0]:  $\pm 200 \,\mu\text{A}$  (Typ.)

[1]:  $\pm 800 \,\mu\mathrm{A}$  (Typ.)

• T2, T1, T0: Test mode setting bits

CHARACTERISTIC	T2	T1	T0	NOTE
Normal operation	0	0	1	_
Reference signal output	1	0	0	Reference signal output : B4, Counter output : B2
1 / 2 counter divider output	1	0	1	Reference signal output : B4, 1 / 2 counter output : B2
Phase comparator test	0	0	0	Comparative signal input : DA Reference signal input : CL (check output : NF)

Note: When testing the counter divider output, programmable counter data input is necessary.

• RSa, RSb : X'tal Reference frequency divider ratio select bits

RSa	RSb	DIVIDER RATIO	STEP FREQUENCY	TUNING FREQUENCY
0	0	1 / 12	333.3 k	500 MHz~~2300 MHz
0	1	1 / 16	250.0 k	500 MHz~2300 MHz
1	0	1 / 64	62.5 k	500 MHz~2000 MHz
1	1	1 / 80	50.0 k	500 MHz~1600 MHz

• os : Tuning amplifier control bit

[0]: Tuning amp ON (Normal operation)

[1]: Tuning amp OFF (Tr. Output is Low Level)

: Don't care



# — I<sup>2</sup>C BUS COMMUNICATIONS CONTROL—

The TD7623AFN conform to standard I<sup>2</sup>C bus format.

The  $I^2C$  bus mode enables two-way bus communications with the WRITE mode, which receives data, and READ mode, which status data.

WRITE and READ mode are set using the last bit (R / W bit) of the address byte.

If the last address bit is set to [0], WRITE mode is set; if set to [1] READ mode is set.

Address can be set using the hardware bits. Three programmable address can be programmed.

With this setting, multiple frequency synthesizers can be used in the same I<sup>2</sup>C bus line.

The address for the hardware bit setting can be selected by applying voltage to the address setting pin

(ADR: Pin 15). An address is selected according to the set bits.

When the correct address byte is received, during acknowledgment, serial data (SDA) line is "Low".

If WRITE mode is set at this time, when the data byte is programmed, the serial data (SDA) line is "Low" during the next acknowledgment.

### a) WRITE mode (setting command)

When WRITE mode is set, Byte 1 segment the address data; Bytes 2 and 3 segment the frequency data; Byte 4 segment the divider ratio setting and function setting data; and Byte 5 segment the output port data.

Data are latched and transferred at the end of Byte 3, Byte 4 and Byte 5.

Byte 2 and Byte 3 are latched and transferred is done with a two byte set (Byte 2+Byte 3).

Once a correct address is received and acknowledged, the data type is determined according to [0] or [1]set in the first bit of the next byte. That is, if the first bit is [0], the data are frequency data; if [1], function setting or output port data.

Until the  $I^2C$  bus STOP CONDITION is detected, the additional data can be input without transmitting the address again. (Ex: Frequency sweep is possible with additional frequency data.)

If data transmission is aborted, data programmed before the abort are valid.

Byte 1 can set the hardware bit with address data.

The hardware bit is set with voltage applied to the address setting pin (ADR: Pin 15).

Bytes 2 and 3 are stored in the 15-bit shift register with counter data for the frequency setting, and control the 15-bit programmable counter ratio.

The program frequency can be calculated in the following formula:

 $fosc = f_r \times N$ 

fosc : Program frequency

fr : Phase comparator reference frequency (Step frequency)

N : Counter total ratio

 $f_r$  is calculated using the crystal oscillator frequency and the reference frequency divider ratio set in Byte 4 (control byte). ( $f_r = X$ 'tal oscillator frequency / reference frequency divider ratio)

The reference frequency divider ratio can be set to 1 / 12, 1 / 16, 1 / 64 and 1 / 80.

When using a 4 MHz crystal oscillator,  $f_r = 333.33 \text{ kHz}$ , 250 kHz, 62.5 kHz and 50 kHz.

The step frequency are 333.33 kHz, 250 kHz, 62.5 kHz and 50 kHz.

Byte 4 is a control byte used to set function. Bit 2 (CP) controls the output current of the charge-pump circuit. When bit 2 is set to [0]: the output current is set to  $\pm 200~\mu A$ ; when set to [1],  $\pm 800~\mu A$ .

Bit 3  $(T_2)$ , Bit 4  $(T_1)$  and Bit 5  $(T_0)$  are used to set test mode. They are used to set the phase comparator reference signal output, and counter divider output.

For details of test mode, see the test mode setting table.

Bit 6 (RSa) and Bit 7 (RSb) are used to set the X'tal reference frequency divider ratio.

For details of the X'tal reference frequency divider ratios, see the table for X'tal reference frequency divider ratios.

Bit 8 (OS) is used to set the charge-pump drive amplifier output setting. When bit 8 is set to [0] the output is ON (Normal Use); when set to [1] the output is OFF (Tr. Output is Low Level).

Byte 5 is used to set and control the output port (Bands 1~4).

When an output port set to [0] is OFF; when set to [1] is ON.

Two output ports can be operation turned on, but be sure to keep the total output current under 40 mA.

#### b) READ mode (status request)

When READ mode is set, power-on reset operation status, phase comparator lock detector output status, and 5-level A / D converter pin input voltage status are output to the master device.

Bit 1 (POR) indicates the power-on reset operation status. When the power supply of  $V_{CC1}$  stops, bit 1 is set to [1] . The condition for reset to [0] , voltage supplied to  $V_{CC1}$  is 3 V or higher, transmission is requested in READ mode, and the status is output. (when  $V_{CC1}$  is turned on, bit 1 is also set to [1] .)

Bit 2 (FL) indicates the phase comparator lock status. When locked, [1] is output; when unlocked, [0] is output.

Bit 6, 7 and 8 (A2, A1, A0) indicate the 5-level A / D converter status. The voltage applied to the A / D converter input pin (pin 12) is output through a 5-level resolution.

For the voltage applied on the A / D converter input pin, 5-level resolution, and the output bits, see the table.

(Ex: The AFT output voltage data can be given to the master device.)

### **DATA FORMAT**

# a) WRITE MODE

	BYTE	MSB							LSB	
1	Address Byte	1	1	0	0	0	MA1	MA0	R / W=0	ACK
2	Divider Byte (1)	0	N14	N13	N12	N11	N10	N9	N8	ACK
3	Divider Byte (2)	N7	N6	N5	N4	N3	N2	N1	N0	ACK (L)
4	Control Byte	1	СР	T2	T1	T0	RSa	RSb	os	ACK (L)
5	Band SW Byte	×	×	×	×	B4	В3	B2	B1	ACK (L)

: DON'T Care ACK : Acknowledged

Latch and transfer timing

# b) READ MODE

	BYTE	MSB							LSB	
1	Address Byte	1	1	0	0	0	MA1	MA0	R/W=1	ACK
2	Status Byte	POR	FL	1	1	1	A2	A1	A0	_

ACK : Acknowledged

# **DATA SPECIFICATIONS**

MA1, MA0 : Programmable hardware address bits

- its (i) its to it regionimasie manarer	G G G G	2.10
ADDRESS PIN APPLIED VOLTAGE	MA1	MA0
0~0.1 V <sub>CC1</sub>	0	0
0.4 V <sub>CC1</sub> ~0.6 V <sub>CC1</sub>	1	0
0~V <sub>CC1</sub>	0	1
0.9 V <sub>CC1</sub> ~V <sub>CC1</sub>	1	1

• CP : Charge-pump output current setting

[0]  $\pm 200 \,\mu\text{A}$  (Typ.) [1]  $\pm 800 \, \mu A \, (Typ.)$  • T2, T1, T0 : Test mode setting

CHARACTERISTIC	T2	T1	T0	NOTE
Normal operation	0	0	1	-
Reference signal output	1	0	0	Reference signal output : B4, Counter output : B2
1 / 2 counter divider output	1	0	1	Reference signal output : B4, 1 / 2 counter output : B2
Phase comparator test	0	0	0	Comparative signal input : SDA Reference signal input : SCL (check output : NF)

Note: When testing the counter divider output, programmable counter data input is necessary.

• RSa, RSb: X'tal reference frequency divider ratio select bits

RSa	RSb	DIVIDER RATIO	STEP FREQUENCY	TUNING FREQUENCY
0	0	1 / 12	333.3 k	500 MHz~2300 MHz
0	1	1 / 16	250.0 k	500 MHz~2300 MHz
1	0	1 / 64	62.5 k	500 MHz~2000 MHz
1	1	1 / 80	50.0 k	500 MHz~1600 MHz

OS : Tuning amplifier control setting

[0]: Tuning amp ON (Normal operation)

[1]: Tuning amp OFF (Tr. Output is Low Level)

• POR : Power-on reset flag

[0] : Normal operation[1] : Reset operation

• FL: Lock detect flag

[0] : Unlocked [1] : Locked

• A2, A1, A0 : 5-level A / D converter status.

ADC PIN APPLIED VOLTAGE	A2	A1	A0
0.60 V <sub>CC1</sub> ~V <sub>CC1</sub>	1	0	0
0.45 V <sub>CC1</sub> ~0.60 V <sub>CC1</sub>	0	1	1
0.30 V <sub>CC1</sub> ~0.45 V <sub>CC1</sub>	0	1	0
0.15 V <sub>CC1</sub> ~0.30 V <sub>CC1</sub>	0	0	1
0~0.15 V <sub>CC1</sub>	0	0	0

\*: Accuracy is ±0.03 × V<sub>CC1</sub>

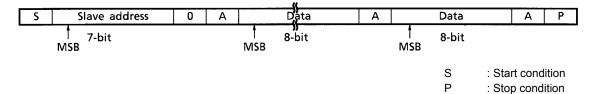
• X: DON'T Care

: Acknowledge

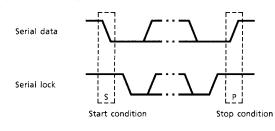
# 1<sup>2</sup>C BUS CONTROL SUMMARY

The bus control format of TD7623AFN conforms to the Philips I<sup>2</sup>C bus control format.

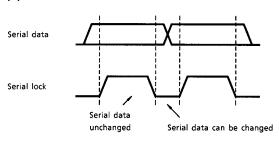
#### **Data transmission format**



(1) Start / Stop condition

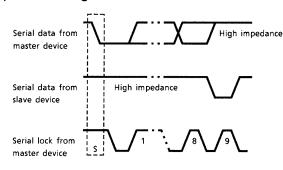


(2) Bit transfer



Α

(3) Acknowledge



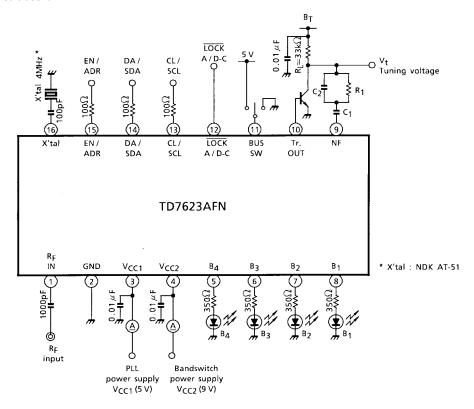
(4) Slave address

A <sub>6</sub>	A <sub>5</sub>	A <sub>4</sub>	A <sub>3</sub>	A <sub>2</sub>	A <sub>1</sub>	A <sub>0</sub>	R/W
1	1	0	0	0	*	*	0

Purchase of TOSHIBA  $I^2C$  components conveys a license under the Philips  $I^2C$  Patent Rights to use these components in an  $I^2C$  system, provided that the system conforms to the  $I^2C$  Standard Specification as defined by Philips.

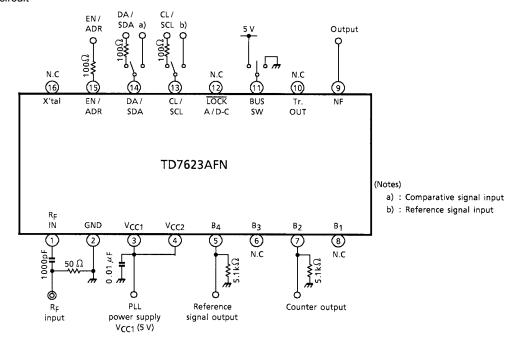
#### **TEST CIRCUIT 1**

Evaluation circuit board



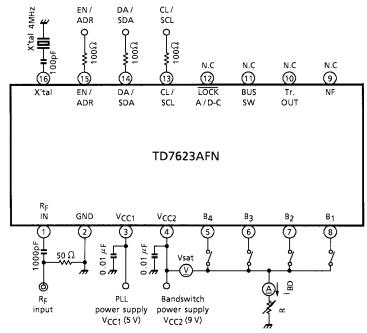
### **TEST CIRCUIT 2**

Input sensitivity test circuit
Test mode circuit

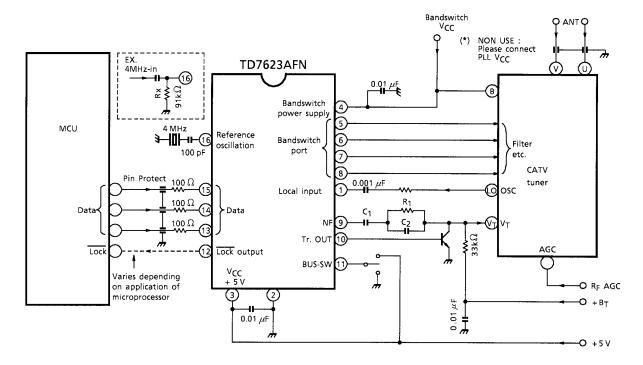


#### **TEST CIRCUIT 3**

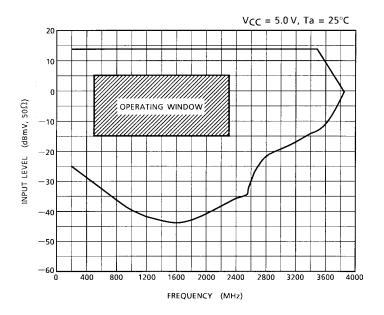
Bandswitch drive test circuit



### SYSTEM APPLICATION DIAGRAM



### TYPICAL INPUT SENSITIVITY CURVE



### FILTER COMPONENT EXPRESSION

=  $[Kv * Icomp/(2\pi)] / (\omega n^2 * N)$ C1

R1 $= [2 * \varepsilon] / (\omega n * C1)$ 

C2 $= 1 / (2\pi * fc * R1)$ 

with :

= Oscillator control sensitivity (radian / Second / Volts)

Icomp = Charge-pump current (A)

= Natural radian frequency (radian / Second) ωn

Ν = Total counter ratio

= Dumping-factor (generally: dumping-factor is about 0.5~1.0) 3

fc= filter cut-off frequency with combination resistor R1.

(generally: fc is about fr (reference frequency) / 20)

#### HANDLING PRECAUTIONS

1. The device should not be inserted into or removed from the test jig while the voltage is being applied: otherwise the device may be degraded or break down.

Do not abruptly increase or decrease the power supply to the device either. (See Figure 1.)

Overshoot or chattering of the power supply may cause the IC to be degraded.

To avoid this filters should be incorporated on the power supply line.

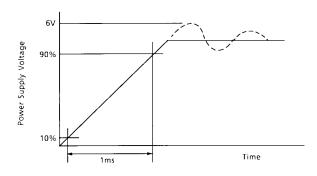


Figure 1

- 2. The peripheral circuits described in this datasheet are given only as system examples for evaluating the device's performance. Toshiba intend neither to recommend the configuration or related values of the peripheral circuits nor to manufacture such application systems in large quantities. Please note that high-frequency characteristics of the device may vary depending on the external components, mounting method and other factors relating to the application design. Therefore, the characteristics of application circuits must be evaluated at the responsibility of the users incorporating the device into their design.
  - To shiba only guarantee the quality and characteristics of the device as described in this datasheet and do not assume any responsibility for the customers application design.
- 3. In order to better understand the quality and reliability of Toshiba semiconductor products and to incorporate them into design in an appropriate manner, please refer to the latest Semiconductor Reliability Handbook (Integrated Circuits) published by Toshiba Semiconductor Company.

The handbook can also be viewed online at

http://doc.semicon.toshiba.co.jp/noseek/us/sinrai/sinraifm.htm.

# **PACKAGE DIMENSIONS**

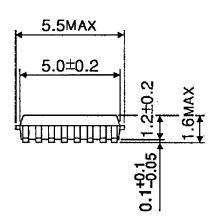
0.23TYP

SSOP16-P-225-0.65B

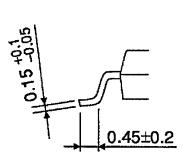
Unit: mm

**⊕** 0.13 **⋈** 

0.22±0.1



0.65



Weight: 0.07 g (Typ.)