## **TS2508A**

#### 3A / 330KHz PWM Buck Converter



SOP-8

#### Pin Definition:



1. FB 5. Vss 2. Enable 6. Vss 3. Ocset 7. SW

4. Vcc 5. SW

#### **General Description**

TS2508A consists of step-down switching regulator with PWM control. These devise include a reference voltage source, oscillation circuit, error amplifier, internal PMOS and etc. TS2508A provides low-ripple power, high efficiency, and excellent transient characteristics. The PWM control circuit is able to very the duty ratio linearly form 0 up to 100%. This converter also contains an error amplifier circuit as well as a soft-start circuit that prevents overshoot at startup. An enable function, an over current protect function and short circuit protect function are built inside, and when OCP or SCP happens, the operation frequency will be reduced. Also, an internal compensation block is built in to minimum external component count. With the addition of an internal P-channel Power MOS, a coil, capacitors, and a diode connected externally, these ICs can function as step-down switching regulators. They serve as ideal power supply units for portable devices when coupled with the SOP-8 package, providing such outstanding features as low current consumption. Since this converter can accommodate an input voltage up to 20V, it is also suitable for the operation via an AC adapter.

#### **Features**

Input Voltage: 3.6V~23VOutput Voltage: 0.8V~Vcc

Duty Ratio: 0%~100% PWM Control
 Oscillation Fragrupa and 230kl In turn

- Oscillation Frequency: 330kHz typ.
- Soft-Start (SS), Current Limit (CL), Enable Function
- Thermal Shutdown Function
- Short Circuit Protect (SCP)
- Built-in Internal SW P-Channel MOSFET

#### **Ordering Information**

Part No.	Package	Packing	
TS2508ACS RL	SOP-8	2.5Kpcs / 13" Reel	

#### **Application**

- Simple High-efficiency Step down Regulator
- On-Card Switching Regulators

#### **Absolute Maximum Rating**

Parameter	Symbol	Rating	Unit
VCC Pin Voltage	V <sub>CC</sub>	Gnd -0.3 to Gnd +25	V
Feedback Pin Voltage	$V_{FB}$	Gnd -0.3 to VCC	V
ON/OFF Pin Voltage	$V_{EN}$	Gnd -0.3 to VCC +0.3	V
Switch Pin Voltage	V <sub>SW</sub>	Gnd -0.3 to VCC +0.3	V
Power Dissipation	$P_{D}$	Internally limited	mW
Storage Temperature Range	T <sub>ST</sub>	-40 to +150	°C
Operating Temperature Range	T <sub>OP</sub>	-20 to +125	°C
Operating Supply Voltage	V <sub>OP</sub>	+3.6 to +23	V
Output Current	Io	3	Α
Thermal Resistance from Junction to case	$\theta_{JC}$	25	°C/W
Thermal Resistance from Junction to ambient	$\theta_{JA}$	70	°C/W

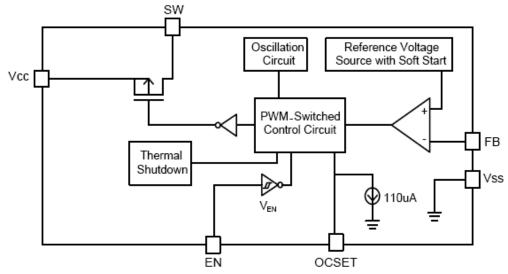




#### Electrical Specifications (V<sub>IN</sub>=12V, Ta=25°C unless otherwise noted)

Characteristics	Symbol	Conditions		Min	Тур	Max	Units
Feedback Voltage	$V_{FB}$	I <sub>OUT</sub> =0.1A		0.784	0.800	0.816	V
Quiescent Current	Iq	V <sub>FB</sub> =1.2V force	driver off		3	5	mA
Feedback Bias Current	I <sub>FB</sub>	I <sub>OUT</sub> =0.1A			0.1	0.5	uA
Shutdown Supply Current	I <sub>SD</sub>	V <sub>EN</sub> =0V			2	10	uA
OCSET pin bias current	I <sub>OCSET</sub>			110	130	150	uA
Switch Current	I <sub>SW</sub>			4			Α
Line Regulation	$\Delta V_{OUT}/V_{OUT}$	V <sub>CC</sub> = 5V~20V, I <sub>OUT</sub> =0.2A			1	2	%
Load Regulation	I <sub>OCSET</sub>	$I_{OUT} = 0.1 \text{ to } 3A$			0.3	0.5	%
Oscillation Frequency	Fosc	SW pin		260	330	400	KHz
EN Pin Logic input	$V_{SH}$	High (regulator ON)		2.0			V
threshold voltage	$V_{SL}$	Low (regulator OFF)				0.8	<b>'</b>
EN Pin Input Current	I <sub>SH</sub>	V <sub>EN</sub> =2.5V (ON)			20		uA
	I <sub>SL</sub>	V <sub>EN</sub> =0.3V (OFF)			-10		uA
Soft-Start Time	T <sub>SS</sub>			0.3	4	8	ms
Internal P-MOS R <sub>DSON</sub>	R <sub>DSON</sub>	$V_{CC} = 5V$ , $V_{FB} = 0V$			90	140	mΩ
Internal F-WOS INDSON	NDSON	V <sub>CC</sub> = 12V, V <sub>FB</sub> = 0V			55	90	mΩ
Efficiency	E <sub>FFI</sub>	V <sub>CC</sub> = 12V,	I <sub>OUT</sub> = 2A		92		%
	<b>∟</b> ⊦⊦1	V <sub>OUT</sub> = 5V	I <sub>OUT</sub> = 3A		91		/0

### **Block Diagram**



#### **Pin Assignment**

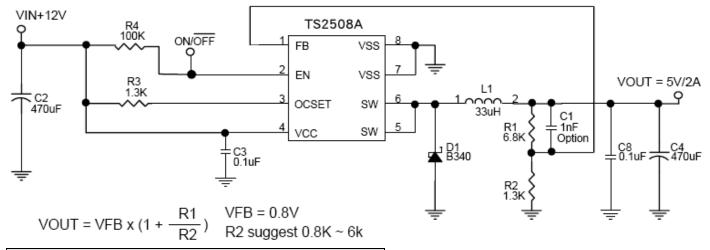
Name	Description		Description
FB	Feedback pin	Vss	GND pin
EN	H:Normal operation (Step-down) L:Step-down operation stopped (All circuit deactivated)	Vss	GND pin
OCSET	Add an external resistor to set max switch output current.	sw	Switch pin. Connect external inductor & diode here
Vcc	IC power supply pin	sw	Switch pin. Connect external inductor & diode here





#### **Application Circuit**

#### **Typical Circuit**



L1 recommend value (V <sub>IN</sub> =12V, I <sub>OUT</sub> =3A )					
V <sub>OUT</sub>	1.8V	2.5V	3.3V	5V	
L1 Value	18uH	22uH	27uH	33uH	

#### **Function Descriptions**

#### **PWM Control**

The TS2508A consists of DC/DC converters that employ a pulse-width modulation (PWM) system. In converters of the TS2508A, the pulse width varies in a range from 0 to 100%, according to the load current. The ripple voltage produced by the switching can easily be removed through a filter because the switching frequency remains constant. Therefore, these converters provide a low-ripple power over broad ranges of input voltage and load current.

#### **RDS(ON) Current Limiting**

The current limit threshold is setting by the external resistor (R3) connecting from  $V_{CC}$  supply to OCSET pin. The internal 110uA sink current crossing the resistor sets the voltage at pin of OCSET. When the PWM voltage is less than the voltage at OCSET, an over-current condition is triggered. Please refer to the formula for setting the minimum current limit value:

$$I_{SW(MIN)} = \frac{I_{OCSET} \times R3 + 0.08}{R_{DS(ON)}}$$
(Normally, The  $I_{SW(MAX)}$  setting more than  $I_{OUT}$  1.2A)

Example: 
$$I_{SW}$$
=(0.11mA \* 2k + 0.08) /  $70m\Omega$  = 4.3A - ( $V_{IN}$ =12V)  $I_{SW}$ =(0.11mA \* 3.3k + 0.08) /  $1000m\Omega$  = 4.4A - ( $V_{IN}$ =5V)





#### **Function Descriptions (Continue)**

#### **Setting the Output Voltage**

Application circuit item shows the basic application circuit with TS2508A adjustable output version. The external resistor sets the output voltage according to the following equation:

$$VOUT = 0.8V \times (1 + \frac{R1}{R2})$$

Table 1: Resistor select for output voltage setting

V <sub>OUT</sub>	R2	R1
5V	1.3K	6.8K
37	5.6K	30K
3.3V	1.5K	4.7K
3.5 V	5.6K	18K
2.5V	2.2K	4.7K
2.5 V	5.6K	12K
1.8V	1.2K	1.5K
1.5V	2.2K	2K
1.2V	2K	1K
1V	2K	0.5K

The R2 setting 5.6k that no load current can be reduce to under 4mA for EL CAP

#### **Input Capacitor Selection**

This capacitor should be located close to the IC using short leads and the voltage rating should be approximately 1.5 times the maximum input voltage. The RMS current rating requirement for the input capacitor of a buck regulator is approximately 1/2 the DC load current. A low ESR input capacitor sized for maximum RMS current must be used. A  $220\mu F$  low ESR capacitor for most applications is sufficient

#### **Output Capacitor Selection**

The output capacitor is required to filter the output and provide regulator loop stability. The important capacitor parameters are; the 100KHz Equivalent Series Resistance (ESR), the RMS ripples current rating, voltage rating, and capacitance value. For the output capacitor, the ESR value is the most important parameter. The ESR can be calculated from the following formula.

$$V_{RIPPLE} = \Delta I_L \times ESR = 0.33A \times 130m\Omega = 43mV$$

An aluminum electrolytic capacitor's ESR value is related to the capacitance and its voltage rating. In most case, higher voltage electrolytic capacitors have lower ESR values. Most of the time, capacitors with much higher voltage ratings may be needed to provide the low ESR values required for low output ripple voltage. It is recommended to replace this low ESR capacitor by using a  $330\mu$ F low ESR values <  $130m\Omega$ 

#### **Inductor Selection**

For most designs, the operate with inductors of  $12\mu H$  to  $33\mu H$ . The inductor value can be derived from the following equation:

$$L = \frac{V_{OUT} \times (V_{IN} - V_{OUT})}{V_{IN} \times \Delta IL \times f_{LX}}$$

Where is inductor Ripple Current. Large value inductors lower ripple current and small value inductors result in high ripple currents. Choose inductor ripple current approximately 15% of the maximum load current 3A,  $\Delta$ IL=0.4A. The DC current rating of the inductor should be at least equal to the maximum load current plus half the ripple current to prevent core saturation (3A+0.2A).



#### **Function Descriptions (Continue)**

#### **Layout Guidance**

If you need low Tc & Tj or large PD (Power Dissipation), The dual SW pins (5&6) and Gnd pins (7&8) on the SOP-8 package are internally connected to die pad, The PCB layout should allow for maximum possible copper area at the SW pins.

- 1. Connect C3 to V<sub>CC</sub> and Gnd pin as closely as possible to get good power filter effect.
- 2. Connect R3 to V<sub>CC</sub> and OCSET pin as closely as possible.
- 3. Connect ground side of the C2 & D1 as closely as possible.

#### **Electrical Characteristics Curve**

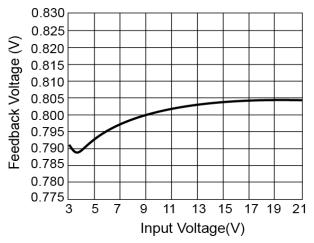


Figure 1. Feedback Voltage vs. Input Voltage

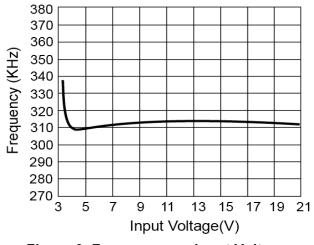


Figure 3. Frequency vs. Input Voltage

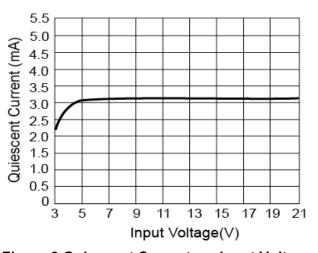


Figure 2. Quiescent Current vs. Input Voltage

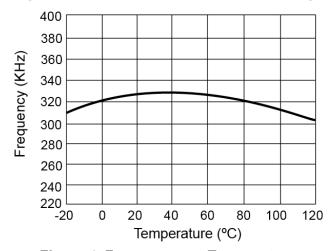


Figure 4. Frequency vs. Temperature





#### **Electrical Characteristics Curve**

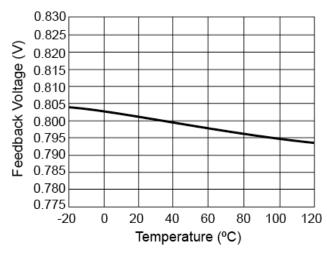


Figure 7. Feedback Voltage vs. Temperature

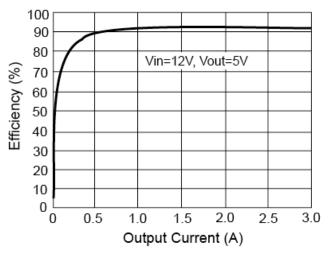


Figure 9. Efficiency vs. Output Current

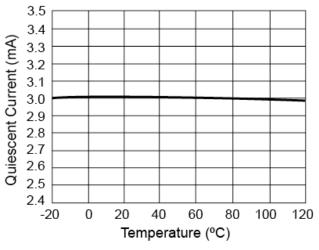


Figure 8. Quiescent Current vs. Temperature

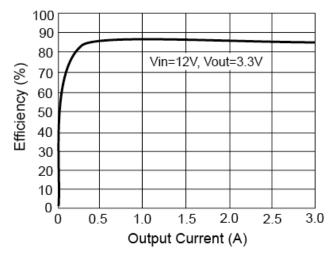
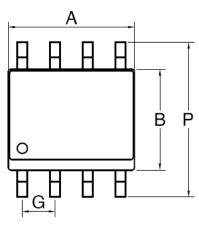


Figure 10. Efficiency vs. Output Current

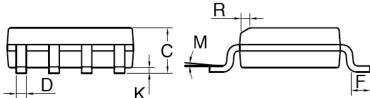




# **SOP-8 Mechanical Drawing**



SOP-8 DIMENSION					
DIM	MILLIMETERS		INCHES		
DIIVI	MIN	MAX	MIN	MAX.	
Α	4.80	5.00	0.189	0.196	
В	3.80	4.00	0.150	0.157	
С	1.35	1.75	0.054	0.068	
D	0.35	0.49	0.014	0.019	
F	0.40	1.25	0.016	0.049	
G	1.27	BSC	0.05	BSC	
K	0.10	0.25	0.004	0.009	
M	0°	7°	0°	7°	
Р	5.80	6.20	0.229	0.244	
R	0.25	0.50	0.010	0.019	



# **Marking Diagram**



Y = Year Code

M = Month Code

(A=Jan, B=Feb, C=Mar, D=Apl, E=May, F=Jun, G=Jul, H=Aug, I=Sep,

J=Oct, K=Nov, L=Dec)

L = Lot Code



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