

### LME49860

# 44V Dual High Performance, High Fidelity Audio Operational Amplifier

### **General Description**

The LME49860 is part of the ultra-low distortion, low noise, high slew rate operational amplifier series optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LME49860 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LME49860 combines extremely low voltage noise density  $(2.7\text{nV}/\sqrt{\text{Hz}})$  with vanishingly low THD+N (0.00003%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LME49860 has a high slew rate of  $\pm 20\text{V}/\mu\text{s}$  and an output current capability of  $\pm 26\text{mA}$ . Further, dynamic range is maximized by an output stage that drives  $2k\Omega$  loads to within 1V of either power supply voltage and to within 1.4V when driving  $600\Omega$  loads.

The LME49860's outstanding CMRR (120dB), PSRR (120dB), and  $\rm V_{OS}$  (0.1mV) give the amplifier excellent operational amplifier DC performance.

The LME49860 has a wide supply range of ±2.5V to ±22V. Over this supply range the LME49860 maintains excellent common-mode rejection, power supply rejection, and low input bias current. The LME49860 is unity gain stable. This Audio Operational Amplifier achieves outstanding AC performance while driving complex loads with values as high as 100pF.

The LME49860 is available in 8-lead narrow body SOIC and 8-lead Plastic DIP packages. Demonstration boards are available for each package.

### **Key Specifications**

<ul><li>Power Supply Voltage Range</li></ul>	±2.5V to ±22V
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■ THD+N  

$$(A_V = 1, V_{OUT} = 3V_{RMS}, f_{IN} = 1kHz)$$

$R_L = 2k\Omega$	0.00003% (typ)
$R_L = 600\Omega$	0.00003% (typ)
■ Input Noise Density	$2.7 \text{nV}/\sqrt{\text{Hz}}$ (typ)
■ Slew Rate	±20V/µs (typ)
■ Gain Bandwidth Product	55MHz (typ)
■ Open Loop Gain (R <sub>L</sub> = 600Ω)	140dB (typ)
■ Input Bias Current	10nA (typ)
■ Input Offset Voltage	0.1mV (typ)
■ DC Gain Linearity Error	0.000009%

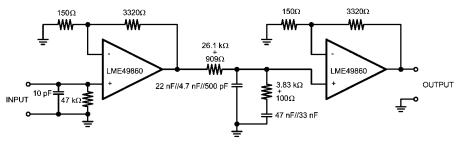
### **Features**

- Easily drives 600Ω loads
- Optimized for superior audio signal fidelity
- Output short circuit protection
- PSRR and CMRR exceed 120dB (typ)
- SOIC, DIP packages

### **Applications**

- Ultra high quality audio amplification
- High fidelity preamplifiers
- High fidelity multimedia
- State of the art phono pre amps
- High performance professional audio
- High fidelity equalization and crossover networks
- High performance line drivers
- High performance line receivers
- High fidelity active filters

### **Typical Application**

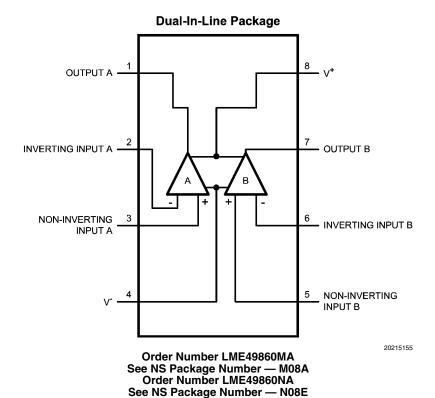


Note: 1% metal film resistors, 5% polypropylene capacitors

**Passively Equalized RIAA Phono Preamplifier** 

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### **Connection Diagrams**



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#### LME49860MA Top Mark

NZXTT L49860 MA

N — National Logo
Z — Assembly Plant code
X — 1 Digit Date code
TT — Die Traceability
L49860 — LME49860
MA — Package code

LME49860NA Top Mark

NUZXYTT LME 49860NA

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N — National Logo
U — Fabrication code
Z — Assembly Plant code
XY — 2 Digit Date code
TT — Die Traceability
NA — Package code

## Absolute Maximum Ratings (Notes 1, 2)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Power Supply Voltage

 $(V_S = V^+ - V^-)$  46V Storage Temperature -65°C to 150°C

Input Voltage (V-) - 0.7V to (V+) + 0.7V

Output Short Circuit (Note 3) Continuous ESD Susceptibility (Note 4) 2000V

ESD Susceptibility (Note 4)

Pins 1, 4, 7 and 8 200V

Pins 2, 3, 5 and 6 100V Junction Temperature 150°C Thermal Resistance

 $\theta_{JA}$  (SO) 145°C/W  $\theta_{JA}$  (NA) 102°C/W

### **Operating Ratings**

Temperature Range

 $T_{MIN} \le T_A \le T_{MAX}$   $-40^{\circ}C \le T_A \le 85^{\circ}C$  Supply Voltage Range  $\pm 2.5V \le V_S \le \pm 22V$ 

	Parameter	Conditions	LME49860		
Symbol			Typical	Limit	Units (Limits)
			(Note 6)	(Note 7)	
		$A_V = 1$ , $V_{OUT} = 3V_{rms}$			
THD+N	Total Harmonic Distortion + Noise	$R_1 = 2k\Omega$	0.00003		% (max)
		$R_L = 600\Omega$	0.00003	0.00009	
IMD	Intermodulation Distortion	A <sub>V</sub> = 1, V <sub>OUT</sub> = 3V <sub>RMS</sub> Two-tone, 60Hz & 7kHz 4:1	0.00005		%
GBWP	Gain Bandwidth Product		55	45	MHz (min)
SR	Slew Rate		±20	±15	V/µs (min)
FPBW	Full Power Bandwidth	V <sub>OUT</sub> = 1V <sub>P-P</sub> , -3dB referenced to output magnitude at f = 1kHz	10		MHz
t <sub>s</sub>	Settling time	$A_V = -1$ , 10V step, $C_L = 100pF$ 0.1% error range	1.2		μs
_	Equivalent Input Noise Voltage	f <sub>BW</sub> = 20Hz to 20kHz	0.34	0.65	μV <sub>RMS</sub> (max)
e <sub>n</sub>	Equivalent Input Noise Density	f = 1kHz f = 10Hz	2.7 6.4	4.7	nV/√ <del>Hz</del> (max)
i <sub>n</sub>	Current Noise Density	f = 1kHz f = 10Hz	1.6 3.1		pAJ√Hz
,,	<b></b>	V <sub>S</sub> = ±18V	±0.12	±0.7	mV (max)
V <sub>OS</sub>	Offset Voltage	V <sub>S</sub> = ±22V	±0.14	±0.7	mV (max)
ΔV <sub>OS</sub> /ΔTemp	Average Input Offset Voltage Drift vs Temperature	-40°C ≤ T <sub>A</sub> ≤ 85°C	0.2		μV/°C
PSRR	Average Input Offset Voltage Shift vs Power Supply Voltage	(Note 8) $V_S = \pm 18V$ , $\Delta V_S = 24V$ $V_S = \pm 22V$ , $\Delta V_S = 30V$	120 120	110	dB dB (min)
ISO <sub>CH-CH</sub>	Channel-to-Channel Isolation	$f_{IN} = 1kHz$ $f_{IN} = 20kHz$	118 112		dB
I <sub>B</sub>	Input Bias Current	V <sub>CM</sub> = 0V	10	72	nA (max)
ΔI <sub>OS</sub> /ΔTemp	Input Bias Current Drift vs Temperature	-40°C ≤ T <sub>A</sub> ≤ 85°C	0.1		nA/°C
I <sub>os</sub>	Input Offset Current	V <sub>CM</sub> = 0V	11	65	nA (max)
	Common-Mode Input Voltage Range	V <sub>S</sub> = ±18V	+17.1 -16.9	(V+) - 2.0 (V-) + 2.0	V (min) V (min)
V <sub>IN-CM</sub>		V <sub>S</sub> = ±22V	+21.0 -20.8	(V+) - 2.0 (V-) + 2.0	V (min) V (min)

Symbol	Parameter		LME4	LME49860	
		Conditions	Typical	Limit	Units
			(Note 6)	(Note 7)	(Limits)
		V <sub>S</sub> = ±18V			
		-12V ≤ V <sub>CM</sub> ≤ 12V	120		dB
CMRR	Common-Mode Rejection	$V_S = \pm 22V$			
		-15V ≤ V <sub>CM</sub> ≤ 15V	120	110	dB (min)
	Differential Input Impedance	CIVI	30		kΩ
$Z_{IN}$	Common Mode Input Impedance	-10V <vcm<10v< td=""><td>1000</td><td></td><td>MΩ</td></vcm<10v<>	1000		MΩ
		$V_S = \pm 18V$	1000		14132
		-12V≤Vout≤12V			
		$R_1 = 600\Omega$			
		$R_L = 2k\Omega$	140		dB
		$R_L = 10k\Omega$	140 140		dB dB
$A_{VOL}$	Open Loop Voltage Gain		140		ив
		V <sub>S</sub> = ±22V			
		–15V≤Vout≤15V			
		$R_L = 600\Omega$	140	125	dB (min)
		$R_L = 2k\Omega$	140		dB
		$R_L = 10k\Omega$	140		dB
		$R_L = 600\Omega$			
		$V_S = \pm 18V$	±16.7		V
		$V_S = \pm 22V$	±20.4	±19.0	V (min)
		$R_L = 2k\Omega$			
$V_{OUTMAX}$	Maximum Output Voltage Swing	V <sub>S</sub> = ±18V	±17.0		V
		$V_S = \pm 22V$	±21.0		V
		$R_I = 10k\Omega$			
		$V_{S} = \pm 18V$	±17.1		V
		$V_S = \pm 22V$	±21.2		V
I <sub>OUT</sub>	Output Current	$R_1 = 600\Omega$			
		$V_S = \pm 20V$	±31		mA
		$V_S = \pm 22V$	±37	±30	mA (min)
I <sub>OUT-CC</sub>			+53		
	Instantaneous Short Circuit Current		-42		mA
R <sub>OUT</sub>	Output Impedance	f <sub>IN</sub> = 10kHz			
		Closed-Loop	0.01		Ω
		Open-Loop	13		
C <sub>LOAD</sub>	Capacitive Load Drive Overshoot	100pF	16		%
		I <sub>OUT</sub> = 0mA			
I <sub>s</sub>	Total Quiescent Current	$V_S = \pm 18V$	10.2		mA
		$V_{S} = \pm 22V$	10.5	13	mA (max)

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.

**Note 2:** Operating Ratings indicate conditions for which the device is functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see the Electrical Characteristics. The guaranteed specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.

Note 3: Amplifier output connected to GND, any number of amplifiers within a package.

Note 4: Human body model, 100pF discharged through a  $1.5k\Omega$  resistor.

**Note 5:** Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under  $50\Omega$ ).

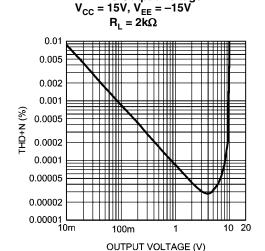
Note 6: Typical specifications are specified at +25°C and represent the most likely parametric norm.

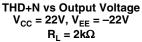
Note 7: Tested limits are guaranteed to National's AOQL (Average Outgoing Quality Level).

Note 8: PSRR is measured as follows: For  $V_S$  = ±22V,  $V_{OS}$  is measured at two supply voltages, ±7V and ±22V. PSRR = I  $20log(\Delta V_{OS}/\Delta V_S)$  I.

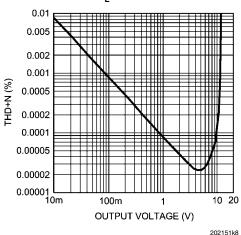
### **Typical Performance Characteristics**

THD+N vs Output Voltage

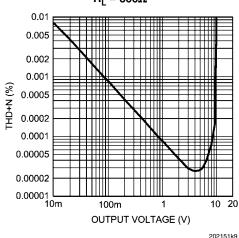




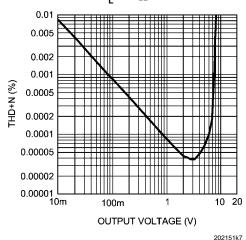
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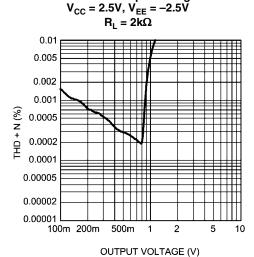
# THD+N vs Output Voltage $V_{\text{CC}} = 15 \text{V}, \, V_{\text{EE}} = -15 \text{V} \\ R_{\text{L}} = 600 \Omega$



# THD+N vs Output Voltage $V_{CC}$ = 12V, $V_{EE}$ = -12V $R_L$ = 2k $\Omega$

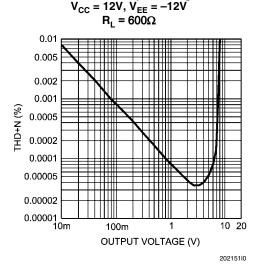


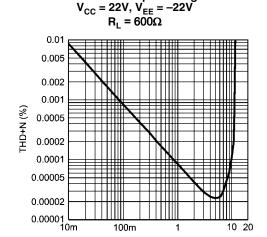
THD+N vs Output Voltage



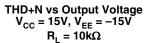
THD+N vs Output Voltage

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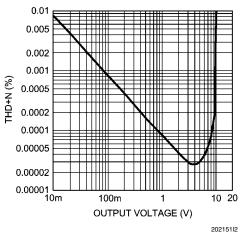


THD+N vs Output Voltage

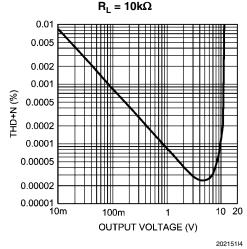


OUTPUT VOLTAGE (V)

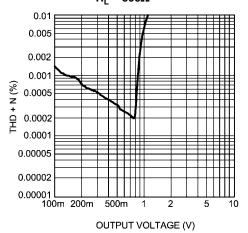
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THD+N vs Output Voltage V<sub>CC</sub> = 22V, V<sub>EE</sub> = -22V

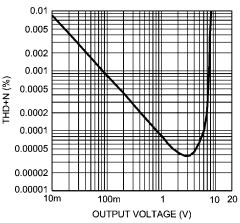


THD+N vs Output Voltage  $V_{CC}$  = 2.5V,  $V_{EE}$  = -2.5V  $R_L$  =  $600\Omega$ 



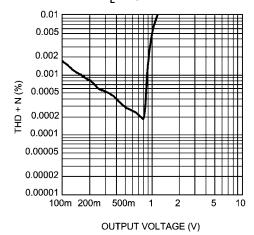
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THD+N vs Output Voltage  $V_{CC}$  = 12V,  $V_{EE}$  = -12V  $R_L$  = 10k $\Omega$ 



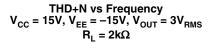
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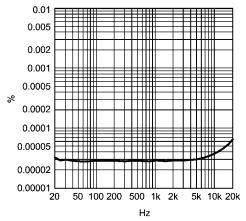
THD+N vs Output Voltage  $V_{CC}$  = 2.5V,  $V_{EE}$  = -2.5V  $R_{L}$  = 10k $\Omega$ 



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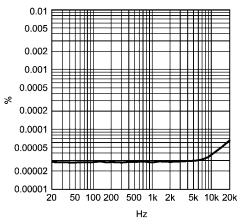
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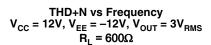


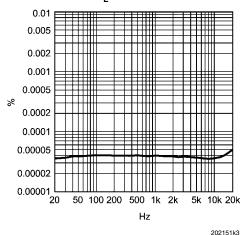
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# THD+N vs Frequency $\begin{aligned} \text{V}_{\text{CC}} &= \text{22V}, \, \text{V}_{\text{EE}} = -\text{22V}, \, \text{V}_{\text{OUT}} = \text{3V}_{\text{RMS}} \\ \text{R}_{\text{L}} &= 2\text{k}\Omega \end{aligned}$

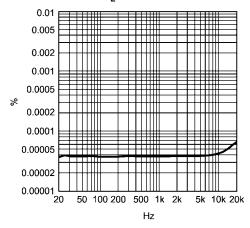


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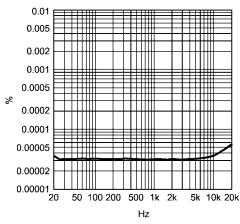


 $\begin{aligned} & \text{THD+N vs Frequency} \\ V_{\text{CC}} = 12V, \, V_{\text{EE}} = -12V, \, V_{\text{OUT}} = 3V_{\text{RMS}} \\ R_{\text{L}} = 2k\Omega \end{aligned}$ 



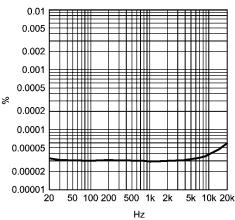
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# THD+N vs Frequency $\begin{aligned} \text{V}_{\text{CC}} = \text{15V}, \, \text{V}_{\text{EE}} = -\text{15V}, \, \text{V}_{\text{OUT}} = 3 \text{V}_{\text{RMS}} \\ \text{R}_{\text{L}} = 600 \Omega \end{aligned}$

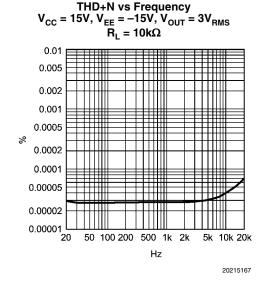


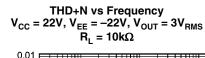
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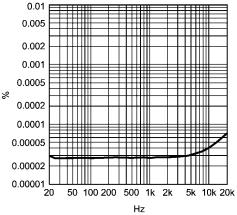
THD+N vs Frequency 
$$\begin{aligned} \text{V}_{\text{CC}} &= 22\text{V}, \, \text{V}_{\text{EE}} = -22\text{V}, \, \text{V}_{\text{OUT}} = 3\text{V}_{\text{RMS}} \\ \text{R}_{\text{L}} &= 600\Omega \end{aligned}$$



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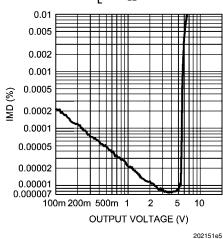


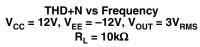


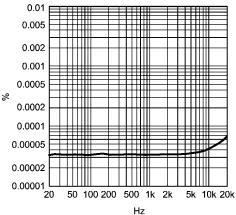
IMD vs Output Voltage  $V_{CC}$  = 12V,  $V_{EE}$  = -12V  $R_L$  = 2k $\Omega$ 

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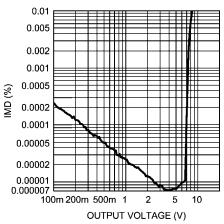
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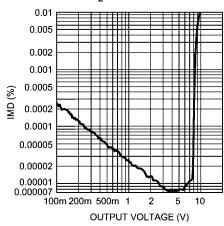


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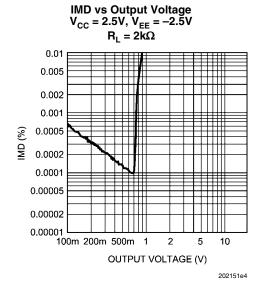


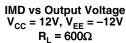
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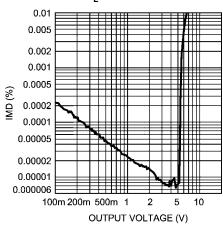
$$\begin{split} & \text{IMD vs Output Voltage} \\ & \text{V}_{\text{CC}} = 22\text{V}, \, \text{V}_{\text{EE}} = -22\text{V} \\ & \text{R}_{\text{L}} = 2\text{k}\Omega \end{split}$$



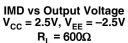
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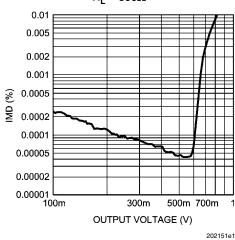




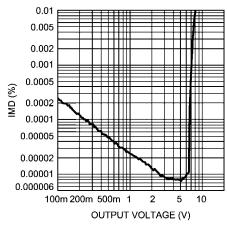


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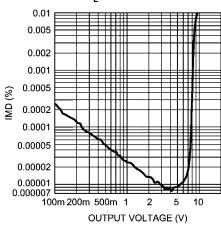


 $\begin{array}{c} \text{IMD vs Output Voltage} \\ V_{CC} = 15V, \, V_{EE} = -15V \\ R_L = 600\Omega \end{array}$ 

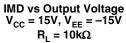


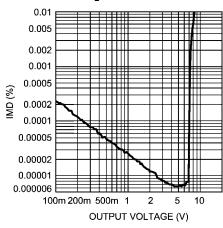
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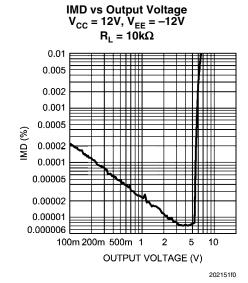


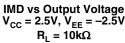
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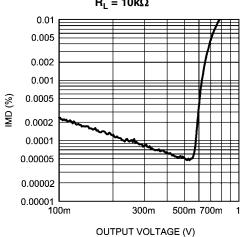




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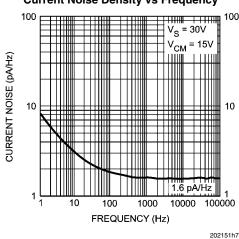


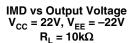


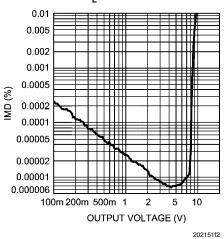


### **Current Noise Density vs Frequency**

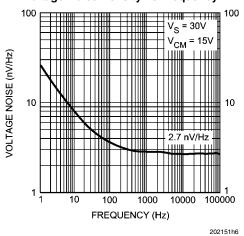
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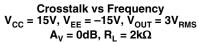


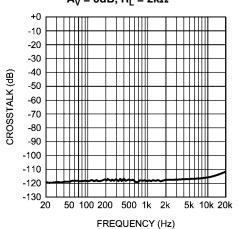




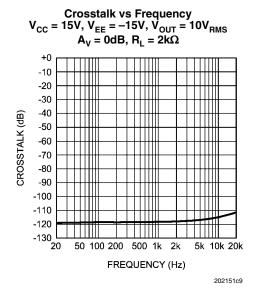
#### **Voltage Noise Density vs Frequency**

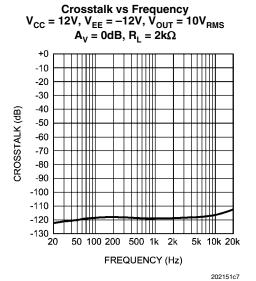


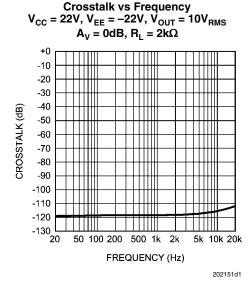


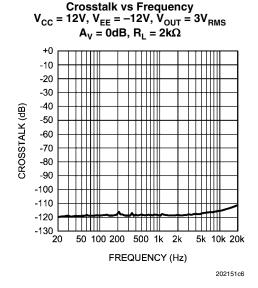


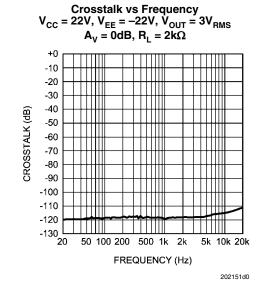
202151c8

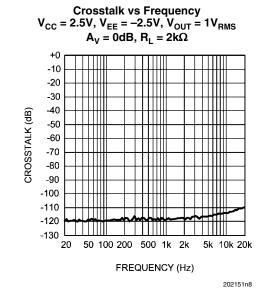


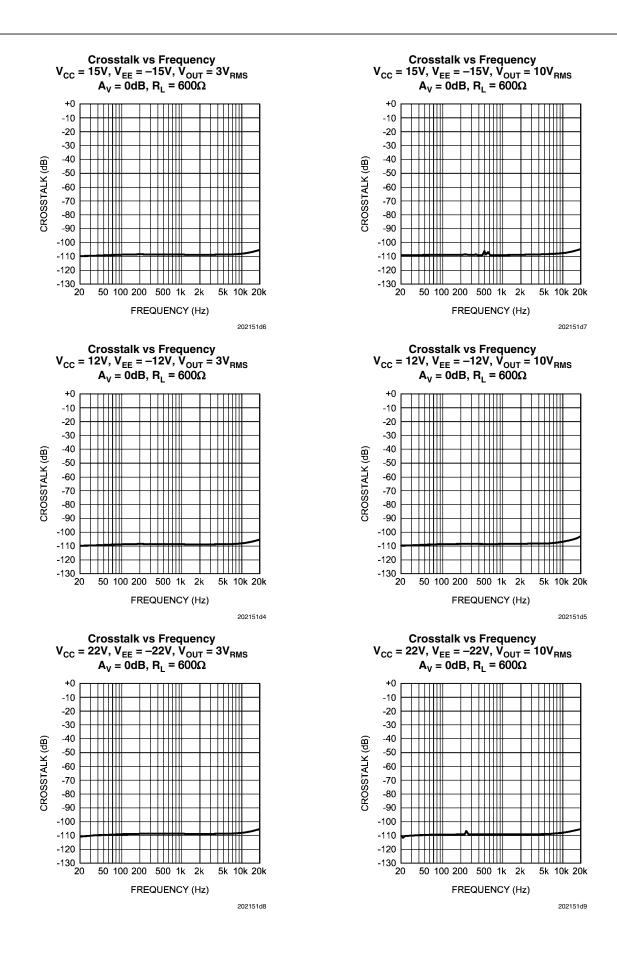


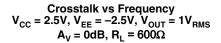


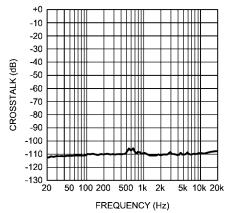






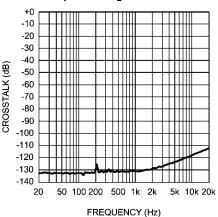






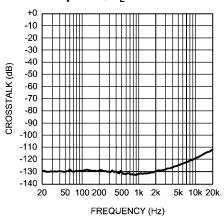
202151d2

# $\begin{array}{c} \text{Crosstalk vs Frequency} \\ \text{V}_{\text{CC}} = 15\text{V}, \, \text{V}_{\text{EE}} = -15\text{V}, \, \text{V}_{\text{OUT}} = 10\text{V}_{\text{RMS}} \\ \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{array}$



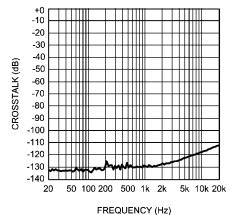
202151n7

# $\begin{aligned} & \text{Crosstalk vs Frequency} \\ & \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V}, \, \text{V}_{\text{OUT}} = 10\text{V}_{\text{RMS}} \\ & \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{aligned}$



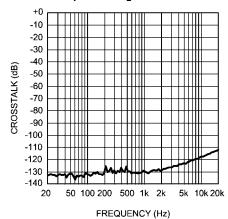
202151n6

$$\begin{aligned} & \text{Crosstalk vs Frequency} \\ V_{\text{CC}} &= 15\text{V}, V_{\text{EE}} = -15\text{V}, V_{\text{OUT}} = 3\text{V}_{\text{RMS}} \\ A_{\text{V}} &= 0\text{dB}, \, R_{\text{L}} = 10\text{k}\Omega \end{aligned}$$



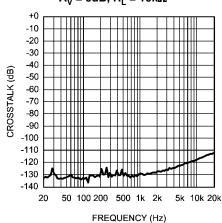
20215100

# $\begin{array}{c} \text{Crosstalk vs Frequency} \\ \text{V}_{\text{CC}} = 12\text{V}, \, \text{V}_{\text{EE}} = -12\text{V}, \, \text{V}_{\text{OUT}} = 3\text{V}_{\text{RMS}} \\ \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{array}$

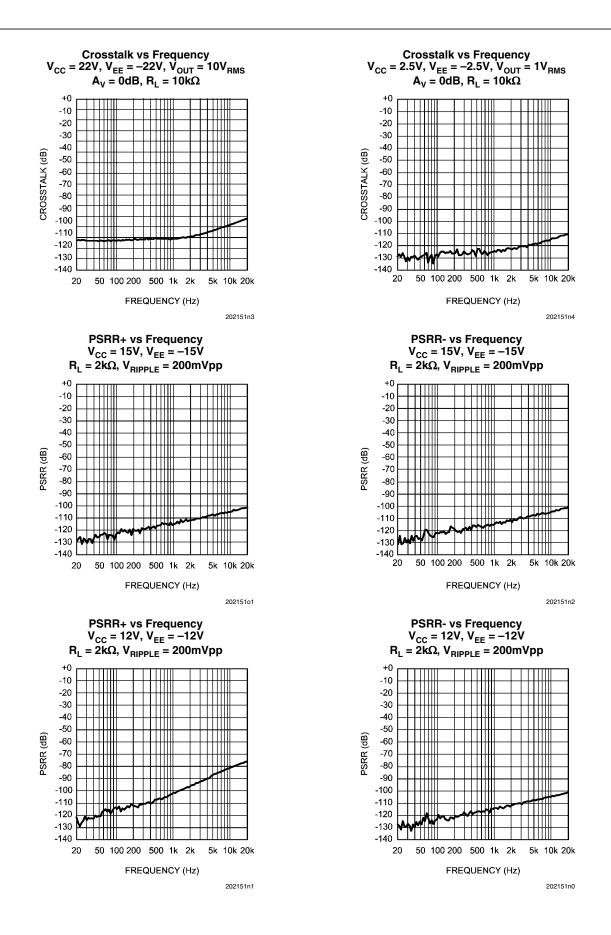


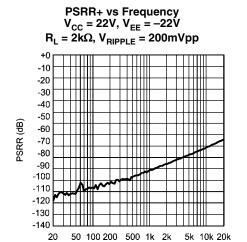
202151n9

# $\begin{aligned} & \text{Crosstalk vs Frequency} \\ & \text{V}_{\text{CC}} = 22\text{V}, \, \text{V}_{\text{EE}} = -22\text{V}, \, \text{V}_{\text{OUT}} = 3\text{V}_{\text{RMS}} \\ & \text{A}_{\text{V}} = 0\text{dB}, \, \text{R}_{\text{L}} = 10\text{k}\Omega \end{aligned}$



202151n5



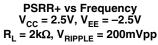


202151m9

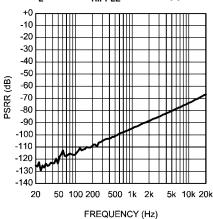
#### +0 -10 -20 -30 -40 -50 -60 -70 -80 -90 -100 -110 -120 -130 -140 20 50 100 200 500 1k 2k 5k 10k 20k FREQUENCY (Hz)

PSRR- vs Frequency  $V_{CC}$  = 22V,  $V_{EE}$  = -22V  $R_L$  = 2k $\Omega$ ,  $V_{RIPPLE}$  = 200mVpp

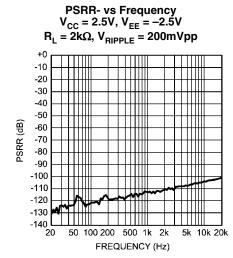
20215103



FREQUENCY (Hz)

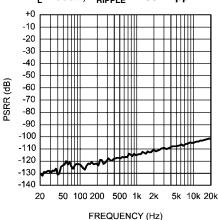


202151m8

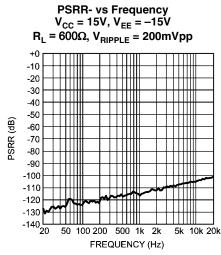


20215106

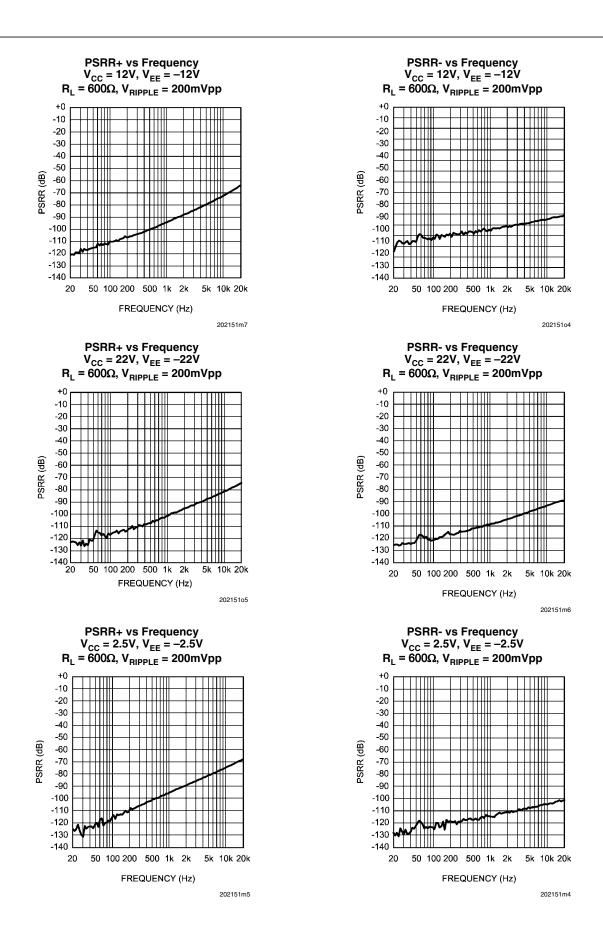
# PSRR+ vs Frequency $V_{CC}$ = 15V, $V_{EE}$ = -15V $R_L$ = 600 $\Omega$ , $V_{RIPPLE}$ = 200mVpp

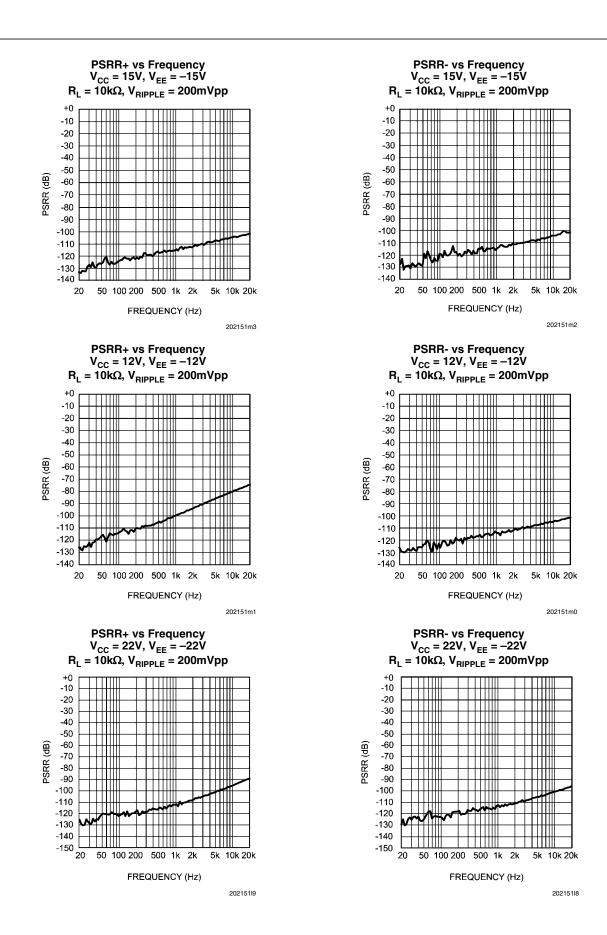


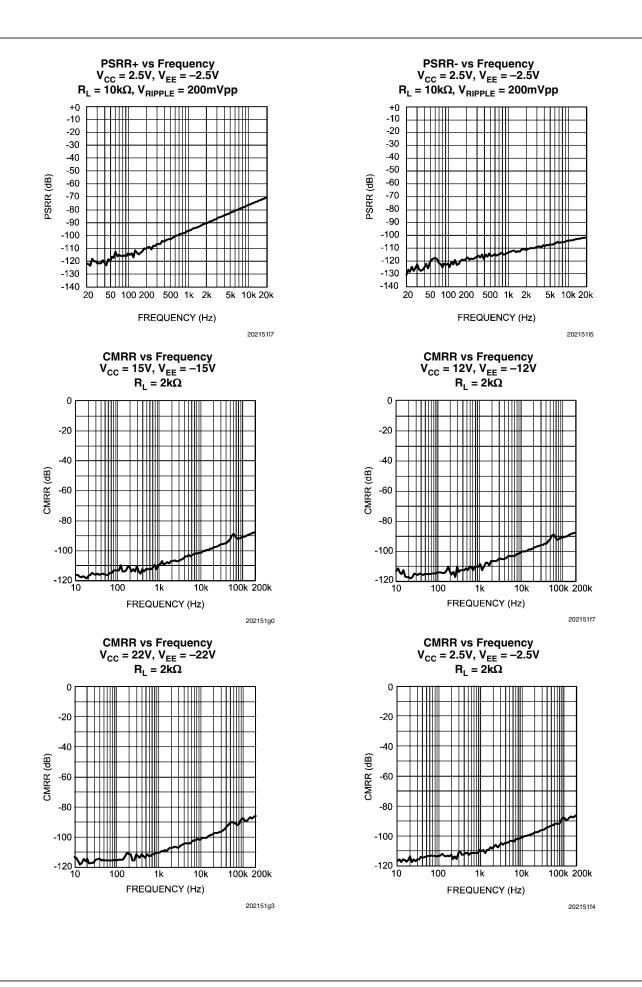
20215102

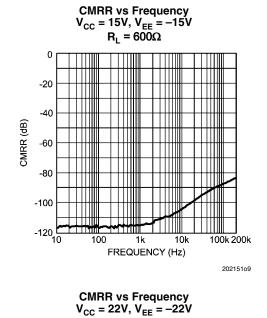


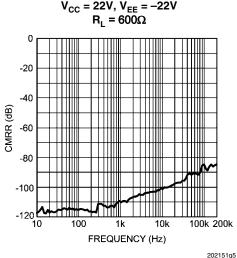
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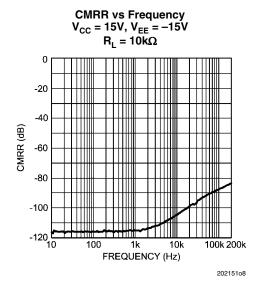


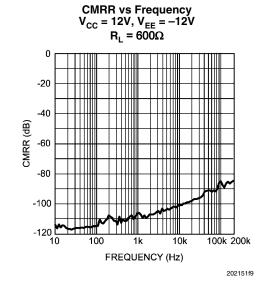


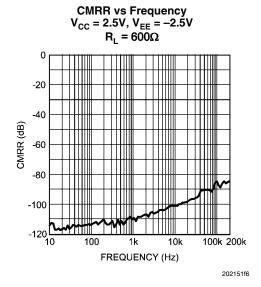


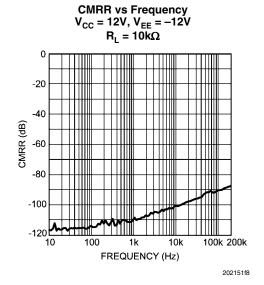


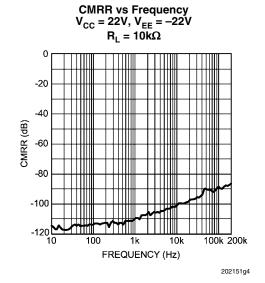


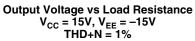


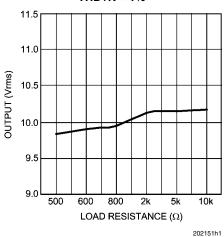




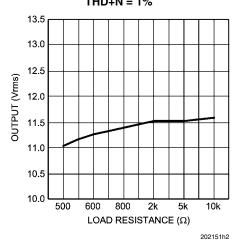




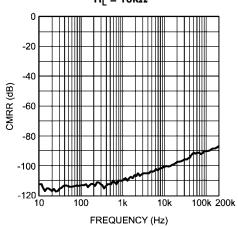




# Output Voltage vs Load Resistance $V_{CC}$ = 22V, $V_{EE}$ = -22V THD+N = 1%

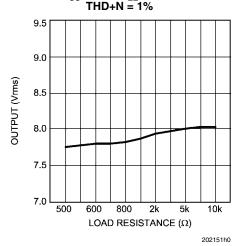


# CMRR vs Frequency $V_{CC}$ = 2.5V, $V_{EE}$ = -2.5V $R_L$ = $10k\Omega$

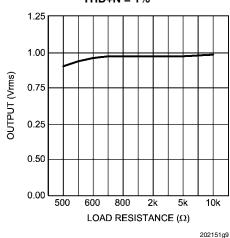


Output Voltage vs Load Resistance  $V_{CC}$  = 12V,  $V_{EE}$  = -12V THD+N = 1%

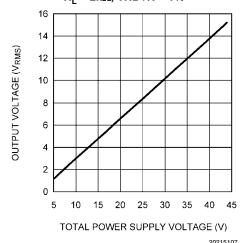
202151f5



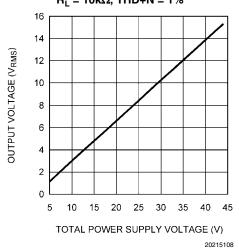
Output Voltage vs Load Resistance  $V_{CC}$  = 2.5V,  $V_{EE}$  = -2.5V THD+N = 1%



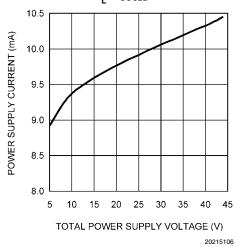
## Output Voltage vs Total Power Supply Voltage $R_1 = 2k\Omega$ , THD+N = 1%



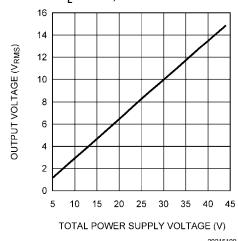
## Output Voltage vs Total Power Supply Voltage $R_L = 10k\Omega,\, THD + N = 1\%$



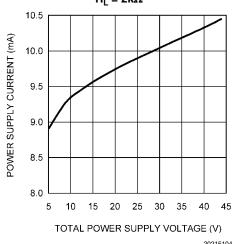
## Power Supply Current vs Total Power Supply Voltage $R_L = 600\Omega$



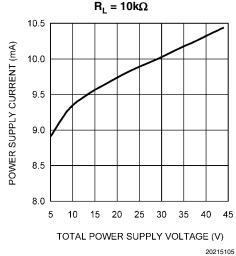
## Output Voltage vs Total Power Supply Voltage $R_1 = 600\Omega$ , THD+N = 1%



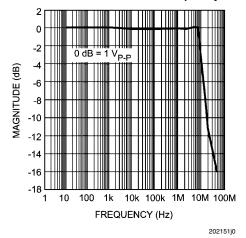
## Power Supply Current vs Total Power Supply Voltage $R_L = 2k\Omega \label{eq:RL}$



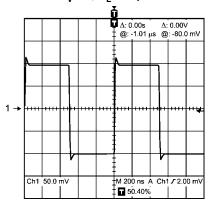
## Power Supply Current vs Total Power Supply Voltage R. = 10kO



### Full Power Bandwidth vs Frequency

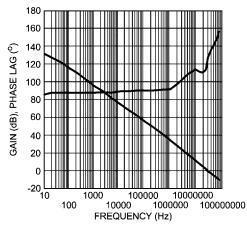


## Small-Signal Transient Response $A_V = 1$ , $C_L = 10pF$



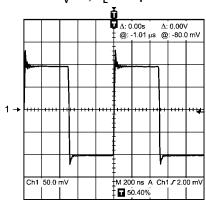
202151i7

### Gain Phase vs Frequency



202151j1

## Small-Signal Transient Response $A_V = 1$ , $C_L = 100pF$



202151i8

### **Application Information**

#### **DISTORTION MEASUREMENTS**

The vanishingly low residual distortion produced by LME49860 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49860's low residual distortion is an input referred internal error. As shown in Figure 1, adding the  $10\Omega$  resistor connected between the amplifier's inverting and non-inverting

inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 1.

This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.

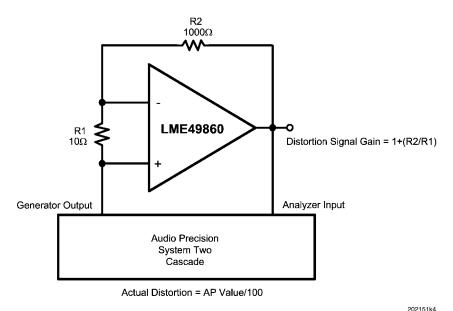
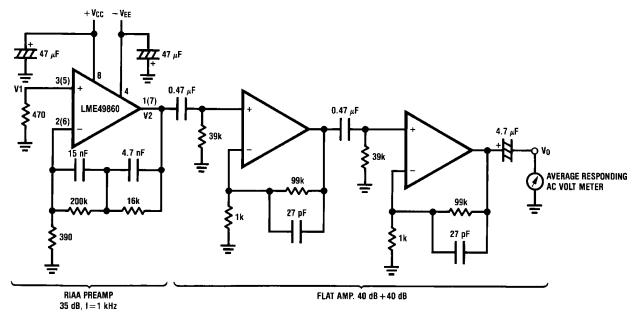


FIGURE 1. THD+N and IMD Distortion Test Circuit

The LME49860 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put

a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

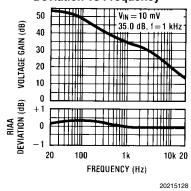


20215127

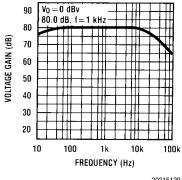
Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

**Noise Measurement Circuit** Total Gain: 115 dB @f = 1 kHz Input Referred Noise Voltage: e<sub>n</sub> = V0/560,000 (V)

#### **RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency**

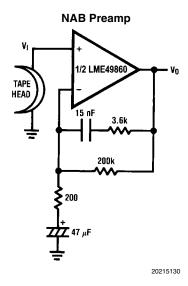


#### Flat Amp Voltage Gain vs Frequency

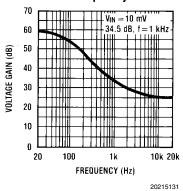


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### **TYPICAL APPLICATIONS**

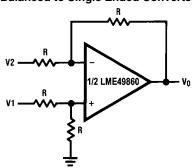


NAB Preamp Voltage Gain vs Frequency



 $A_V = 34.5$  F = 1 kHz  $E_n = 0.38 \mu\text{V}$ A Weighted

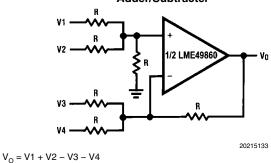
### **Balanced to Single Ended Converter**



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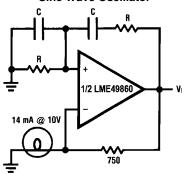
20215132

Adder/Subtracter



V<sub>O</sub> = V1–V2

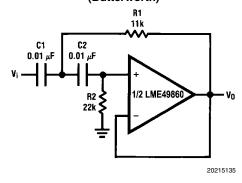
Sine Wave Oscillator



20215134

$$f_0 = \frac{1}{2\pi RC}$$

## Second Order High Pass Filter (Butterworth)

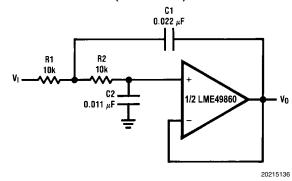


if 
$$C1 = C2 = C$$

$$R1 = \frac{\sqrt{2}}{2\omega_0 C}$$

Illustration is  $f_0 = 1 \text{ kHz}$ 

## Second Order Low Pass Filter (Butterworth)



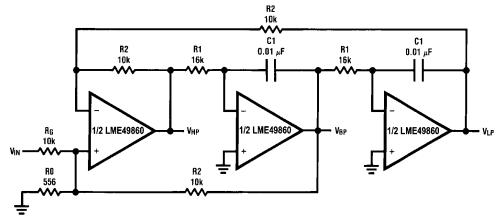
if R1 = R2 = R

$$C1 = \frac{\sqrt{2}}{m_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is  $f_0 = 1 \text{ kHz}$ 

### **State Variable Filter**

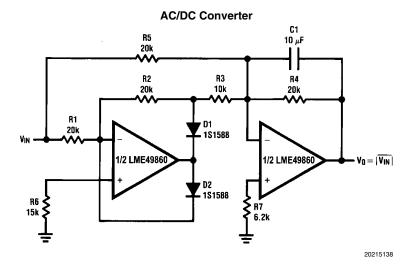


20215137

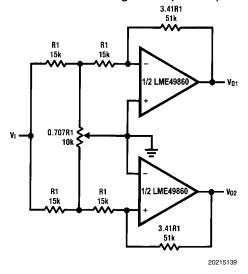
$${\rm f}_0 = \frac{1}{2\pi C 1 R 1}, {\rm Q} = \frac{1}{2} \left( 1 + \frac{R2}{R0} + \frac{R2}{RG} \right), {\rm A}_{BP} = {\rm QA}_{LP} = {\rm QA}_{LH} = \frac{R2}{RG}$$

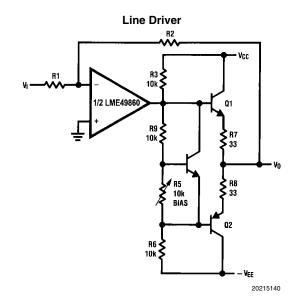
26

Illustration is  $f_0 = 1 \text{ kHz}$ , Q = 10,  $A_{BP} = 1$ 



### 2 Channel Panning Circuit (Pan Pot)





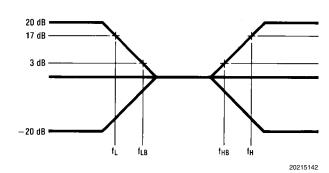
### 

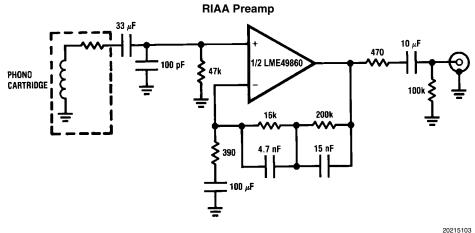
$$\begin{split} f_L &= \frac{1}{2\pi R2C1}, f_{LB} = \frac{1}{2\pi R1C1} \\ f_H &= \frac{1}{2\pi R5C2}, f_{HB} = \frac{1}{2\pi (R1 + R5 + 2R3)C2} \end{split}$$

20215141

Illustration is:

$$f_L = 32 \text{ Hz}, f_{LB} = 320 \text{ Hz}$$
  
 $f_H = 11 \text{ kHz}, f_{HB} = 1.1 \text{ kHz}$ 





 $A_v = 35 \text{ dB}$ 

 $E_n = 0.33 \, \mu V$ 

S/N = 90 dB

f = 1 kHz

A Weighted

A Weighted,  $V_{IN} = 10 \text{ mV}$ 

@f = 1 kHz

# 

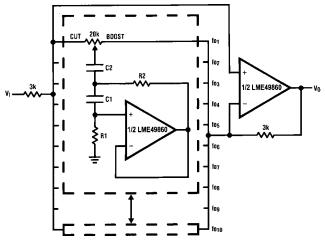
20215143

If R2 = R5, R3 = R6, R4 = R7  

$$V0 = \left(1 + \frac{2R2}{R1}\right) \frac{R4}{R3} (V2 - V1)$$

Illustration is: V0 = 101(V2 - V1)

### 10 Band Graphic Equalizer



20215144

fo (Hz)	C <sub>1</sub>	C <sub>2</sub>	R <sub>1</sub>	R <sub>2</sub>
32	0.12µF	4.7µF	75kΩ	500Ω
64	0.056µF	3.3µF	68kΩ	510Ω
125	0.033µF	1.5µF	62kΩ	510Ω
250	0.015µF	0.82µF	68kΩ	470Ω
500	8200pF	0.39µF	62kΩ	470Ω
1k	3900pF	0.22µF	68kΩ	470Ω
2k	2000pF	0.1µF	68kΩ	470Ω
4k	1100pF	0.056µF	62kΩ	470Ω
8k	510pF	0.022µF	68kΩ	510Ω
16k	330pF	0.012µF	51kΩ	510Ω

Note 9: At volume of change =  $\pm 12 \text{ dB}$ 

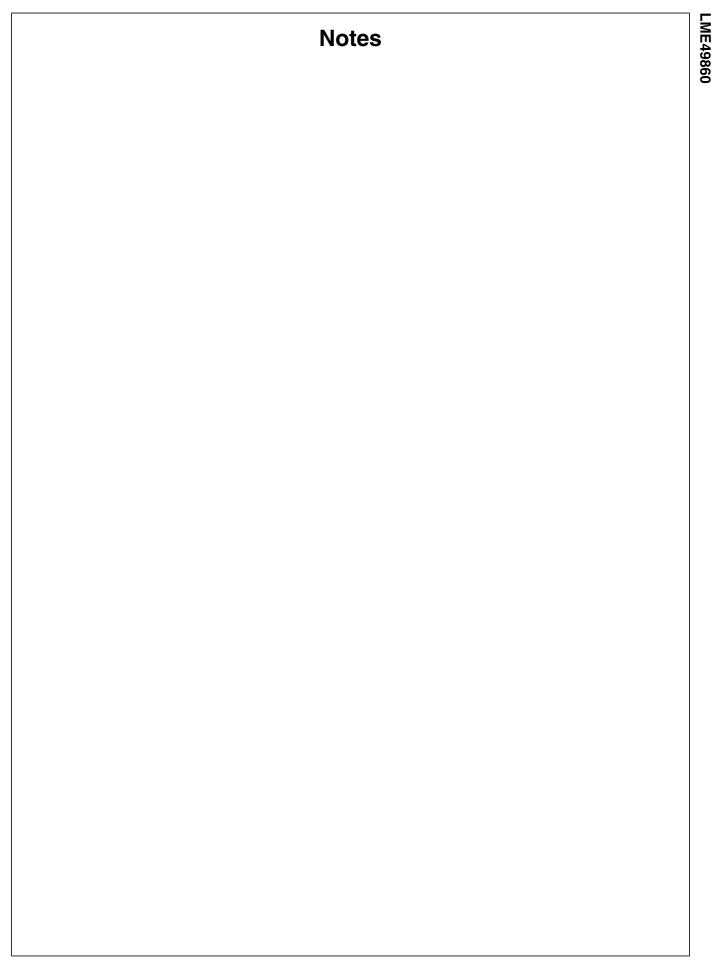
Q = 1.7

Reference: "AUDIO/RADIO HANDBOOK", National Semiconductor, 1980, Page 2-61

## **Revision History**

Rev	Date	Description
1.0	06/01/07	Initial release.
1.1	06/11/07	Added the LME49860MA and LME49860NA Top Mark Information.

#### Physical Dimensions inches (millimeters) unless otherwise noted 4.9±0.1 [.193±.004] NOT INCLUDING MOLD FLASH 0.15[.006] MAX PER END 3.9±0.1 [.154±.004] NOT INCLUDING MOLD FLASH 6±0.2 [.236±.008] 6X (1.27 RECOMMENDED LAND PATTERN 1.35-1.75 [.053-.069] 6X [1.27 [.050] R0.18±0.02 [.007±.0008] 0.1[.004] C C SEATING PLANE 8X 0.35-0.51 [.0138-.0200] .10-0.25 .004-.0101 ф 0.25Г.0101M C AS BS CONTROLLING DIMENSION IS MILLIMETER VALUES IN [ ] ARE INCHES DIMENSIONS IN ( ) FOR REFERENCE ONLY M08A (Rev L) Narrow SOIC Package Order Number LME49860MA **NS Package Number M08A** 0.373 - 0.400 $\overline{(9.474 - 10.16)}$ 0.090 (2.286) 7 6 5 $0.032 \pm 0.005$ 0.092 DIA (2.337) $(0.813 \pm 0.127)$ $0.250 \pm 0.005$ RAD + PIN NO. 1 IDENT $(6.35 \pm 0.127)$ PIN NO. 1 IDENT OPTION 1 1 2 3 4 0.280 (7.112) MIN 0.040 0.030 OPTION 2 (1.016)0.039 MAX 0.145 - 0.200(0.762)0.300 - 0.320(0.991)(3.683 - 5.080)(7.62 - 8.128) $0.130 \pm 0.005$ $\overline{(3.302 \pm 0.127)}$ 0.125 - 0.140 95° ± 5° (3.175 - 3.556)0.065 0.020 0.125 (1.651) 0.009 - 0.01590°±4° (0.508) MIN (3.175) (0.229 - 0.381) $0.018 \pm 0.003$ $0.325 \,{}^{+\, 0.040}_{-\, 0.015}$ $(0.457 \pm 0.076)$ $\underline{\textbf{0.100} \pm \textbf{0.010}}$ $(2.540 \pm 0.254)$ $0.045 \pm 0.015$ $(1.143 \pm 0.381)$ 0.060 (1.524)0.050 (1.270)N08E (REV F) **Dual-In-Line Package** Order Number LME49860NA **NS Package Number N08E**



### **Notes**

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