

LP2950/LP2951

Series of Adjustable Micropower Voltage Regulators

General Description

The LP2950 and LP2951 are micropower voltage regulators with very low quiescent current (75μA typ.) and very low dropout voltage (typ. 40mV at light loads and 380mV at 100mA). They are ideally suited for use in battery-powered systems. Furthermore, the quiescent current of the LP2950/LP2951 increases only slightly in dropout, prolonging battery life

The LP2950-5.0 is available in the surface-mount D-Pak package, and in the popular 3-pin TO-92 package for pin-compatibility with older 5V regulators. The 8-lead LP2951 is available in plastic, ceramic dual-in-line, LLP, or metal can packages and offers additional system functions.

One such feature is an error flag output which warns of a low output voltage, often due to falling batteries on the input. It may be used for a power-on reset. A second feature is the logic-compatible shutdown input which enables the regulator to be switched on and off. Also, the part may be pin-strapped for a 5V, 3V, or 3.3V output (depending on the version), or programmed from 1.24V to 29V with an external pair of resistors.

Careful design of the LP2950/LP2951 has minimized all contributions to the error budget. This includes a tight initial tolerance (.5% typ.), extremely good load and line regulation

(.05% typ.) and a very low output voltage temperature coefficient, making the part useful as a low-power voltage reference.

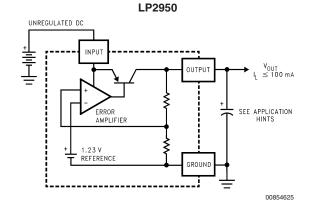
Features

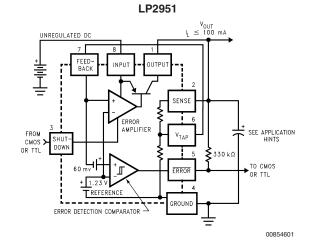
- 5V, 3V, and 3.3V versions available
- High accuracy output voltage
- Guaranteed 100mA output current
- Extremely low quiescent current
- Low dropout voltage
- Extremely tight load and line regulation
- Very low temperature coefficient
- Use as Regulator or Reference
- Needs minimum capacitance for stability
- Current and Thermal Limiting
- Stable with low-ESR output capacitors ($10m\Omega$ to 6Ω)

LP2951 versions only

- Error flag warns of output dropout
- Logic-controlled electronic shutdown
- Output programmable from 1.24 to 29V

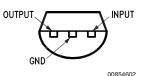
Block Diagram and Typical Applications





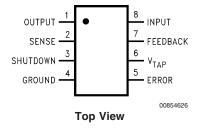
Connection Diagrams

TO-92 Plastic Package (Z)

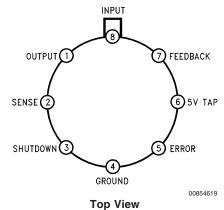


Bottom View

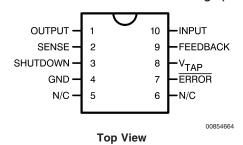
Dual-In-Line Packages (N, J) Surface-Mount Package (M, MM)

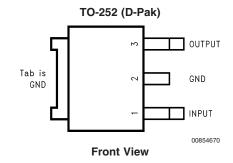


Metal Can Package (H)



10-Lead Ceramic Surface-Mount Package (WG)





SENSE 2 SHUTDOWN 3 SENSE 5 ERROR

Pin 4 is fused to center DAP

Top View

www.national.com

2

Ordering Information Package Temperature Part Number Transport Media NSC Drawing Package Marking Range TO-92 (Z) $-40 < T_{.1} < 125$ LP2950ACZ-3.0 2950A CZ3.0 Bag Z03A 2950 CZ3.0 LP2950CZ-3.0 Bag LP2950ACZ-3.3 2950A CZ3.3 Bag LP2950CZ-3.3 2950 CZ3.3 Bag LP2950ACZ-5.0 2950A CZ5.0 Bag LP2950CZ-5.0 2950 CZ5.0 Bag TO-252 $-40 < T_J < 125$ TD03B LP2950CDT-3.0 LP2950CDT-3.0 75 Units/Rail (D-Pak) LP2950CDTX-3.0 2.5k Units Tape and Reel LP2950CDT-3.3 75 Units/Rail LP2950CDT-3.3 LP2950CDTX-3.3 2.5k Units Tape and Reel LP2950CDT-5.0 LP2950CDT-5.0 75 Units/Rail LP2950CDTX-5.0 2.5k Units Tape and Reel N (N-08E) $-40 < T_J < 125$ 40 Units/Rail LP2951ACN-3.0 LP2951ACN-3.0 N08E LP2951CN-3.0 LP2951CN-3.0 40 Units/Rail LP2951ACN-3.3 40 Units/Rail LP2951ACN-3.3 40 Units/Rail LP2951CN-3.3 LP2951CN-3.3 LP2951ACN LP2951ACN 40 Units/Rail LP2951CN LP2951CN 40 Units/Rail $-40 < T_J < 125$ M (M08A) LP2951ACM-3.0 2951ACM30* 95 Units/Rail A80M (where * is die rev letter) LP2951ACMX-3.0 2.5k Units Tape and Reel 2951CM30* 95 Units/Rail LP2951CM-3.0 (where * is die rev letter) LP2951CMX-3.0 2.5k Units Tape and Reel LP2951ACM-3.3 2951ACM33* 95 Units/Rail (where * is die rev letter) LP2951ACMX-3.3 2.5k Units Tape and Reel 2951CM33* 95 Units/Rail LP2951CM-3.3 (where * is die rev letter) LP2951CMX-3.3 2.5k Units Tape and Reel LP2951ACM 95 Units/Rail 2951ACM* (where * is die rev letter) 2.5k Units Tape and Reel LP2951ACMX 95 Units/Rail LP2951CM 2951CM* (where * is die rev letter) LP2951CMX 2.5k Units Tape and Reel $-40 < T_J < 125$ MM L0BA 1k Units Tape and Reel MUA08A LP2951ACMM-3.0 (MUA08A) LP2951ACMMX-3.0 3.5k Units Tape and Reel L0BB LP2951CMM-3.0 1k Units Tape and Reel LP2951CMMX-3.0 3.5k Units Tape and Reel LP2951ACMM-3.3 L0CA 1k Units Tape and Reel LP2951ACMMX-3.3 3.5k Units Tape and Reel LP2951CMM-3.3 L0CB 1k Units Tape and Reel LP2951CMMX-3.3 3.5k Units Tape and Reel 1k Units Tape and Reel LP2951ACMM L0DA LP2951ACMMX 3.5k Units Tape and Reel 1k Units Tape and Reel LP2951CMM L0DB LP2951CMMX 3.5k Units Tape and Reel J (J08A) $-55 < T_J < 150$ LP2951J/883 See MIL/AERO Datasheet 40 Units/Rail J08A

See MIL/AERO Datasheet

See MIL/AERO Datasheet

Tray

Tray

H08C

WG10A

 $-55 < T_J < 150$

 $-55 < T_J < 150$

LP2951H/883

LP2951WG/883

H (H08C)

(WG10A)

WG

Ordering Information (Continued)

Package	Temperature Range	Part Number	Package Marking	Transport Media	NSC Drawing
8-lead	-40 < T _J < 125	LP2951ACSD-3.0	51AC30	1k Units Tape and Reel	SDC08A
LLP		LP2951ACSDX-3.0		4.5k Units Tape and Reel	
		LP2951CSD-3.0	51AC30B	1k Units Tape and Reel	
		LP2951CSDX-3.0		4.5k Units Tape and Reel	
		LP2951ACSD-3.3	51AC33	1k Units Tape and Reel	
		LP2951ACSDX-3.3		4.5k Units Tape and Reel	
		LP2951CSD-3.3	51AC33B	1k Units Tape and Reel	
		LP2951CSDX-3.3		4.5k Units Tape and Reel	
		LP2951ACSD	2951AC	1k Units Tape and Reel	
		LP2951ACSDX		4.5k Units Tape and Reel	
		LP2951CSD	2951ACB	1k Units Tape and Reel	
		LP2951CSDX		4.5k Units Tape and Reel	

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Input Supply Voltage -0.3 to +30V

SHUTDOWN Input Voltage, Error Comparator Output

Voltage, (Note 9)

FEEDBACK Input Voltage -1.5 to +30V

(Note 9) (Note 10)

Power Dissipation Internally Limited

Junction Temperature (T_J) +150 $^{\circ}$ C

Ambient Storage Temperature -65 $^{\circ}$ to +150 $^{\circ}$ C

Soldering Dwell Time, Temperature

Wave 4 seconds, 260°C
Infrared 10 seconds, 240°C
Vapor Phase 75 seconds, 219°C

ESD Rating

Human Body Model(Note 18) 2500V

Operating Ratings (Note 1)

Maximum Input Supply Voltage 30V

Junction Temperature Range

(T_J) (Note 8)

LP2950AC-XX, LP2950C-XX,

Electrical Characteristics (Note 2)

	0	LP2951		LP2950AC-XX LP2951AC-XX			LP2950C-XX LP2951C-XX			
Parameter	Conditions		Tested		Tested	Design		Tested	Design	Units
	(Note 2)	Тур	Limit	Тур	Limit	Limit	Тур	Limit	Limit	
			(Notes 3, 16)		(Note 3)	(Note 4)		(Note 3)	(Note 4)	
3V Versions (Note	17)							ı		
Output Voltage	$T_J = 25^{\circ}C$	3.0	3.015	3.0	3.015		3.0	3.030		V max
			2.985		2.985			2.970		V min
	$-25^{\circ}\text{C} \le \text{T}_{\text{J}} \le 85^{\circ}\text{C}$	3.0		3.0		3.030	3.0		3.045	V max
						2.970			2.955	V min
	Full Operating	3.0	3.036	3.0		3.036	3.0		3.060	V max
	Temperature		2.964			2.964			2.940	V min
	Range									
Output Voltage	100μA ≤ I _L ≤	3.0	3.045	3.0		3.042	3.0		3.072	V max
	100mA	3.0		0.0			3.0			
	$T_{J} \leq T_{JMAX}$		2.955			2.958			2.928	V min
3.3V Versions (Note										
Output Voltage	$T_J = 25^{\circ}C$	3.3	3.317	3.3	3.317		3.3	3.333		V max
			3.284		3.284			3.267		V min
	$-25^{\circ}\text{C} \le \text{T}_{\text{J}} \le 85^{\circ}\text{C}$	3.3		3.3		3.333	3.3		3.350	V max
						3.267			3.251	V min
	Full Operating	3.3	3.340	3.3		3.340	3.3		3.366	V max
	Temperature		3.260			3.260			3.234	V min
	Range									
Output Voltage	$100\mu A \leq I_L \leq$	3.3	3.350	3.3		3.346	3.3		3.379	V max
	100mA	0.0		0.0			0.0			
	$T_{J} \leq T_{JMAX}$		3.251			3.254			3.221	V min
5V Versions (Note										
Output Voltage	$T_J = 25^{\circ}C$	5.0	5.025	5.0	5.025		5.0	5.05		V max
			4.975		4.975			4.95		V min
	$-25^{\circ}\text{C} \le \text{T}_{\text{J}} \le 85^{\circ}\text{C}$	5.0		5.0		5.05	5.0		5.075	V max
						4.95			4.925	V min
	Full Operating	5.0	5.06	5.0		5.06	5.0		5.1	V max

	Conditions	LP2951		LP2950AC-XX LP2951AC-XX			LP2950C-XX LP2951C-XX			
Parameter	(Note 2)	Тур	Tested Limit (Notes 3, 16)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Тур	Tested Limit (Note 3)	Design Limit (Note 4)	Units
	Temperature Range		4.94			4.94			4.9	V min
Output Voltage	100μA ≤ I _L ≤ 100mA	5.0	5.075	5.0		5.075	5.0		5.12	V max
	$T_{J} \leq T_{JMAX}$		4.925			4.925			4.88	V min
All Voltage Options	T		ı		1					
Output Voltage Temperature Coefficient	(Note 12)	20	120	20		100	50		150	ppm/°C
Line Regulation	$(V_ONOM + 1)V \le$	0.03	0.1	0.03	0.1		0.04	0.2		% max
(Note 14)	V _{in} ≤ 30V (Note 15)		0.5			0.2			0.4	% max
Load Regulation	100μA ≤ I _L ≤	0.04	0.1	0.04	0.1		0.1	0.2		% max
(Note 14)	100mA		0.3			0.2			0.3	% max
Dropout Voltage (Note 5)	$I_L = 100 \mu A$		80		80			80		mV max
		50	150	50		150	50		150	mV max
	I _L = 100mA	200	450	200	450	600	200	450	600	mV max
Overvied	1 1004	380	600	380	100	600	380	100	600	mV max
Ground Current	I _L = 100μA	75	120 140	75	120	140	75	120	140	μA max μA max
	I _L = 100mA	8	12	8	12		8	12	- 110	mA max
			14			14			14	mA max
Dropout	$V_{in} = (V_O NOM - 0.5)V$	110	170	110	170		110	170		μA max
Ground Current	I _L = 100μA		200			200			200	μΑ max
Current Limit	V _{out} = 0	160	200	160	200		160	200		mA max
			220			220			220	mA max
Thermal Regulation	(Note 13)	0.05	0.2	0.05	0.2		0.05	0.2		%/W max
Output Noise,	$C_L = 1\mu F$ (5V Only)	430		430			430			μV rms
10 Hz to 100 kHz	C _L = 200µF	160		160			160			μV rms
	$C_L = 3.3 \mu F$ (Bypass = 0.01 μF Pins 7 to 1 (LP2951)	100		100			100			μV rms
8-pin Versions Only	•		LP2951		LP2951AC	-XX		LP2951C	-XX	
Reference		1.235	1.25	1.235	1.25		1.235	1.26		V max
Voltage			1.26			1.26			1.27	V max
			1.22		1.22			1.21		V min
D ((1)		1.2			1.2			1.2	V min
Reference	(Note 7)		1.27			1.27			1.285	V max
Voltage			1.19	00	40	1.19	00	40	1.185	V min
Feedback Pin		20	40	20	40	60	20	40	60	nA max
Bias Current			60			60			60	nA max

6

	Conditions (Note 2)	LP2951		LP2950AC-XX LP2951AC-XX			LP2950C-XX LP2951C-XX			
Parameter		Тур	Tested Limit	Тур	Tested Limit	Design Limit	Тур	Tested Limit	Design Limit	Units
All Voltage Options			(Notes 3, 16)		(Note 3)	(Note 4)		(Note 3)	(Note 4)	
Temperature Coefficient										
Feedback Pin Bias Current Temperature Coefficient		0.1		0.1			0.1			nA/°C
Error Comparator										<u> </u>
Output Leakage Current	V _{OH} = 30V	0.01	1 2	0.01	1	2	0.01	1	2	μΑ ma μΑ ma
Output Low	$V_{in} = (V_O NOM - 0.5)V$	150	250	150	250		150	250		mV ma
Voltage	$I_{OL} = 400 \mu A$		400			400			400	mV ma
Upper Threshold Voltage	(Note 6)	60	40 25	60	40	25	60	40	25	mV mii mV mii
Lower Threshold Voltage	(Note 6)	75	95 140	75	95	140	75	95	140	mV ma
Hysteresis	(Note 6)	15		15			15			mV
Shutdown Input		1								
Input		1.3		1.3			1.3			V
Logic	Low (Regulator ON)		0.6			0.7			0.7	V max
Voltage	High (Regulator OFF)		2.0			2.0			2.0	V min
Shutdown Pin Input	V _{shutdown} = 2.4V	30	50	30	50		30	50		μA ma

Note 1: Absolute Maximum Ratings are limits beyond which damage to the device may occur. Operating Ratings are conditions under which operation of the device is guaranteed. Operating Ratings do not imply guaranteed performance limits. For guaranteed performance limits and associated test conditions, see the Electrical Characteristics tables.

450

3

600

10

100

600

750

10

20

Note 2: Unless otherwise specified all limits guaranteed for $V_{IN} = (V_{ONOM} + 1)V$, $I_L = 100\mu A$ and $C_L = 1\mu F$ for 5V versions and 2.2 μ F for 3V and 3.3V versions. Limits appearing in **boldface** type apply over the entire junction temperature range for operation. Limits appearing in normal type apply for $T_A = T_J = 25^{\circ}C$. Additional conditions for the 8-pin versions are FEEDBACK tied to V_{TAP} , OUTPUT tied to SENSE, and $V_{SHUTDOWN} \le 0.8V$.

Note 3: Guaranteed and 100% production tested.

Current

Regulator Output

Current in Shutdown

Note 4: Guaranteed but not 100% production tested. These limits are not used to calculate outgoing AQL levels.

450

3

- Note 5: Dropout Voltage is defined as the input to output differential at which the output voltage drops 100 mV below its nominal value measured at 1V differential. At very low values of programmed output voltage, the minimum input supply voltage of 2V (2.3V over temperature) must be taken into account.
- Note 6: Comparator thresholds are expressed in terms of a voltage differential at the Feedback terminal below the nominal reference voltage measured at $V_{in} = (V_O NOM + 1)V$. To express these thresholds in terms of output voltage change, multiply by the error amplifier gain = $V_{out}V_{ref} = (R1 + R2)/R2$. For example, at a programmed output voltage of 5V, the Error output is guaranteed to go low when the output drops by 95mV x 5V/1.235V = 384 mV. Thresholds remain constant as a percent of V_{out} as V_{out} is varied, with the dropout warning occurring at typically 5% below nominal, 7.5% guaranteed.
- Note 7: $V_{ref} \le V_{out} \le (V_{in} 1V), \ 2.3V \le V_{in} \le 30V, \ 100 \mu A \le I_L \le 100 mA, \ T_J \le T_{JMAX}.$

 $V_{\text{shutdown}} = 30V$

(Note 11)

Note 8: The junction-to-ambient thermal resistances are as follows: 180°C/W and 160°C/W for the TO-92 package with 0.40 inch and 0.25 inch leads to the printed circuit board (PCB) respectively, 105°C/W for the molded plastic DIP (N), 130°C/W for the ceramic DIP (J), 160°C/W for the molded plastic SOP (M), 200°C/W for the molded plastic MSOP (MM), and 160°C/W for the metal can package (H). The above thermal resistances for the N, J, M, and MM packages apply when the package is soldered directly to the PCB. Junction-to-case thermal resistance for the H package is 20°C/W. Junction-to-case thermal resistance for the TO-252 package is 5.4°C/W. The value of 0_{JA} for the LLP package is typically 51°C/W but is dependent on the PCB trace area, trace material, and the number of layers and thermal vias. For details of thermal resistance and power dissipation for the LLP package, refer to Application Note AN-1187.

Note 9: May exceed input supply voltage.

100

750

20

450

3

600

10

100

750

20

μA max

µA max

µA max

µA max

uA max

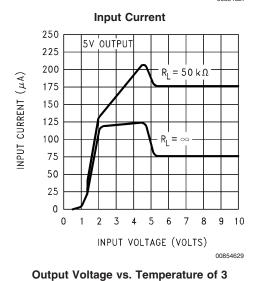
Electrical Characteristics (Note 2) (Continued)

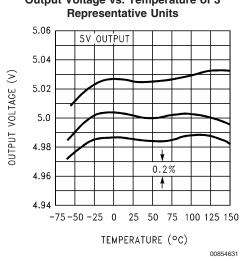
Note 10: When used in dual-supply systems where the output terminal sees loads returned to a negative supply, the output voltage should be diode-clamped to ground.

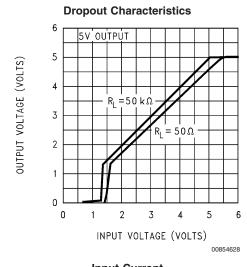
- Note 11: $V_{shutdown} \ge 2V$, $V_{in} \le 30V$, $V_{out} = 0$, Feedback pin tied to V_{TAP} .
- Note 12: Output or reference voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
- Note 13: Thermal regulation is defined as the change in output voltage at a time T after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 50mA load pulse at $V_{IN} = 30V$ (1.25W pulse) for T = 10ms.
- Note 14: Regulation is measured at constant junction temperature, using pulse testing with a low duty cycle. Changes in output voltage due to heating effects are covered under the specification for thermal regulation.
- Note 15: Line regulation for the LP2951 is tested at 150 $^{\circ}$ C for I_L = 1mA. For I_L = 100 μ A and T_J = 125 $^{\circ}$ C, line regulation is guaranteed by design to 0.2%. See Typical Performance Characteristics for line regulation versus temperature and load current.
- Note 16: A Military RETS specification is available on request. At time of printing, the LP2951 RETS specification complied with the boldface limits in this column. The LP2951H, WG, or J may also be procured as Standard Military Drawing Spec #5962-3870501MGA, MXA, or MPA.
- Note 17: All LP2950 devices have the nominal output voltage coded as the last two digits of the part number. In the LP2951 products, the 3.0V and 3.3V versions are designated by the last two digits, but the 5V version is denoted with no code at this location of the part number (refer to ordering information table).
- Note 18: Human Body Model $1.5k\Omega$ in series with 100pF.

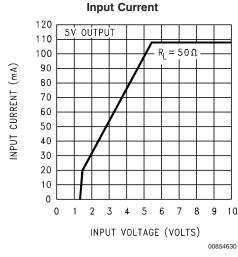
Typical Performance Characteristics

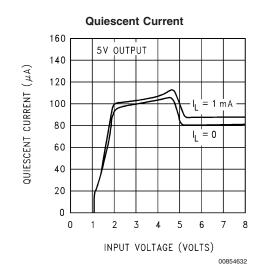
Quiescent Current 10 (Yeu) 10 0.1 10 10 10 10 10 100 LOAD CURRENT (mA)

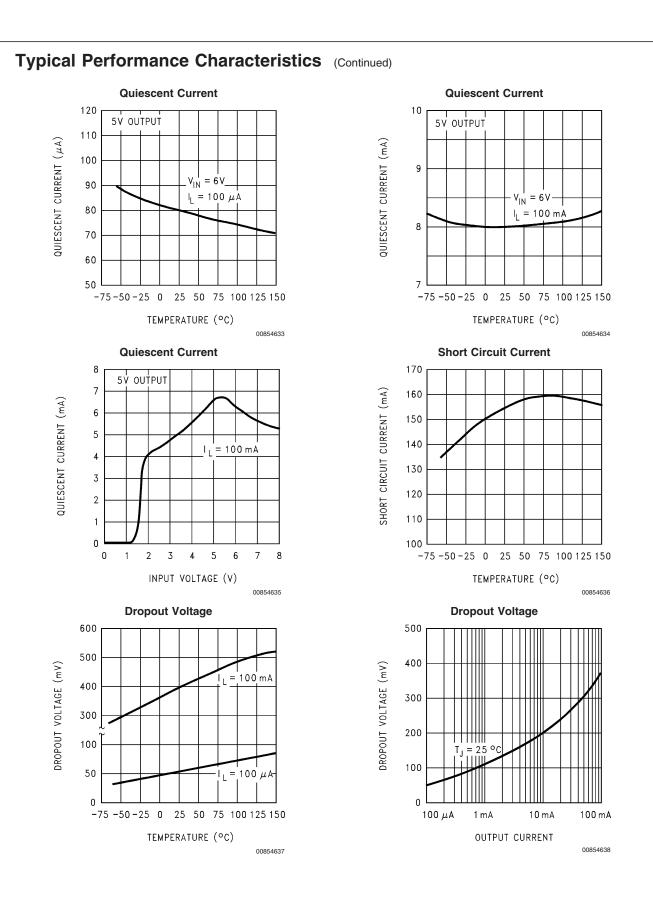






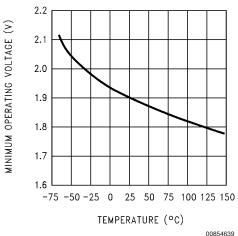




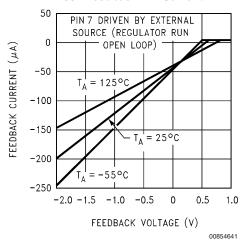


Typical Performance Characteristics (Continued)

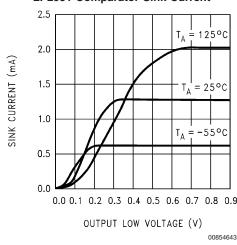
LP2951 Minimum Operating Voltage



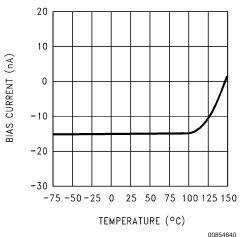
LP2951 Feedback Pin Current



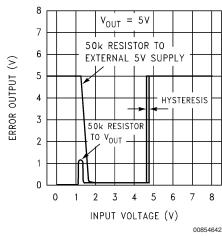
LP2951 Comparator Sink Current



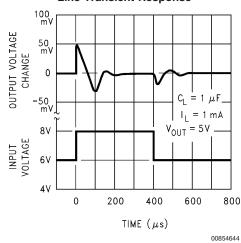
LP2951 Feedback Bias Current



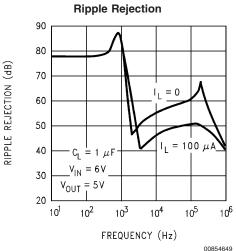
LP2951 Error Comparator Output

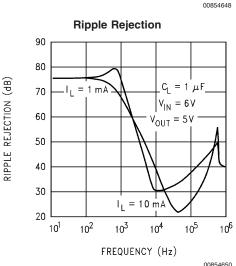


Line Transient Response

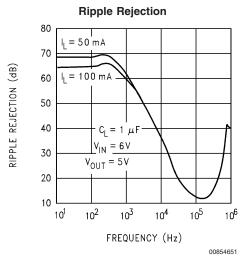


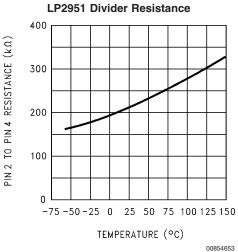
Typical Performance Characteristics (Continued) **Load Transient Response Load Transient Response** 250 80 200 60 OUTPUT VOLTAGE CHANGE (mV) CHANGE (mV) 100 50 -50 OUTPUT VOLTAGE 40 20 0 -20 = 10 μF -40 = 5V ÷ 1 μF ${\rm C}_{\rm L}$ -100 V_{OUT} = 5V 100 mA 100 μA 100 mA 100 μΑ 0 1 2 3 0 12 16 20 TIME (ms) TIME (ms) 00854645 00854646 LP2951 Enable Transient **Output Impedance** 10 $|_0 = 100 \,\mu$ 6 5 OUTPUT IMPEDANCE (OHMS) OUTPUT VOLTAGE (V) 5 2 4 $I_L = 10 \overline{\text{mA}}$ 3 $V_{IN} = 8V$ 0.5 2 $V_{OUT} = 5V$ = 10 µF 0.2 1 0.1 0 PIN VOLTAGE (V) 0.05 2 0 0.01 -100 0 100 200 300 400 500 600 700 10 100 100K TIME (μs) FREQUENCY (Hz) 00854647 00854648

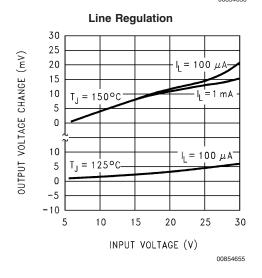


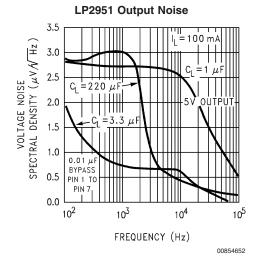


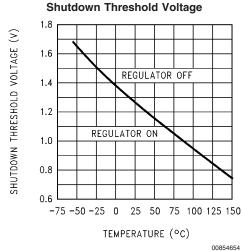
Typical Performance Characteristics (Continued)

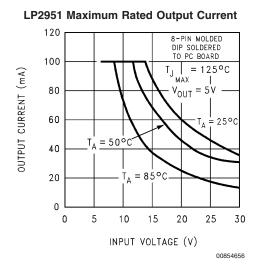






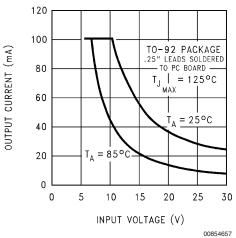






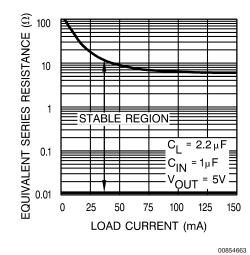
Typical Performance Characteristics (Continued)

LP2950 Maximum Rated Output Current



00854658

Output Capacitor ESR Range



Application Hints

EXTERNAL CAPACITORS

A 1.0µF (or greater) capacitor is required between the output and ground for stability at output voltages of 5V or more. At lower output voltages, more capacitance is required (2.2µF or more is recommended for 3V and 3.3V versions). Without this capacitor the part will oscillate. Most types of tantalum or aluminum electrolytics work fine here; even film types work but are not recommended for reasons of cost. Many aluminum electrolytics have electrolytes that freeze at about $-30\,^{\circ}\text{C}$, so solid tantalums are recommended for operation below $-25\,^{\circ}\text{C}$. The important parameters of the capacitor are an ESR of about 5Ω or less and a resonant frequency above 500kHz. The value of this capacitor may be increased without limit.

Ceramic capacitors whose value is greater than 1000pF should not be connected directly from the LP2951 output to ground. Ceramic capacitors typically have ESR values in the range of 5 to $10m\Omega$, a value below the lower limit for stable operation (see curve Output Capacitor ESR Range).

The reason for the lower ESR limit is that the loop compensation of the part relies on the ESR of the output capacitor to provide the zero that gives added phase lead. The ESR of ceramic capacitors is so low that this phase lead does not occur, significantly reducing phase margin. A ceramic output capacitor can be used if a series resistance is added (recommended value of resistance about 0.1Ω to 2Ω).

At lower values of output current, less output capacitance is required for stability. The capacitor can be reduced to $0.33\mu F$ for currents below 10mA or $0.1\mu F$ for currents below 1mA. Using the adjustable versions at voltages below 5V runs the error amplifier at lower gains so that *more* output capacitance is needed. For the worst-case situation of a 100mA load at 1.23V output (Output shorted to Feedback) a $3.3\mu F$ (or greater) capacitor should be used.

Unlike many other regulators, the LP2950 will remain stable and in regulation with no load in addition to the internal voltage divider. This is especially important in CMOS RAM keep-alive applications. When setting the output voltage of the LP2951 versions with external resistors, a minimum load of $1\mu A$ is recommended.

A $1\mu F$ tantalum, ceramic or aluminum electrolytic capacitor should be placed from the LP2950/LP2951 input to ground if there is more than 10 inches of wire between the input and the AC filter capacitor or if a battery is used as the input.

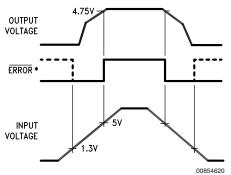
Stray capacitance to the LP2951 Feedback terminal can cause instability. This may especially be a problem when using high value external resistors to set the output voltage. Adding a 100pF capacitor between Output and Feedback and increasing the output capacitor to at least $3.3\mu F$ will fix this problem.

ERROR DETECTION COMPARATOR OUTPUT

The comparator produces a logic low output whenever the LP2951 output falls out of regulation by more than approximately 5%. This figure is the comparator's built-in offset of about 60mV divided by the 1.235 reference voltage. (Refer to the block diagram in the front of the datasheet.) This trip level remains "5% below normal" regardless of the programmed output voltage of the 2951. For example, the error flag trip level is typically 4.75V for a 5V output or 11.4V for a 12V output. The out of regulation condition may be due either to low input voltage, current limiting, or thermal limiting.

Figure 1 below gives a timing diagram depicting the ERROR signal and the regulated output voltage as the LP2951 input is ramped up and down. For 5V versions, the ERROR signal becomes valid (low) at about 1.3V input. It goes high at about 5V input (the input voltage at which V_{OUT} = 4.75V). Since the LP2951's dropout voltage is load-dependent (see curve in typical performance characteristics), the **input** voltage trip point (about 5V) will vary with the load current. The **output** voltage trip point (approx. 4.75V) does not vary with load

The error comparator has an open-collector output which requires an external pullup resistor. This resistor may be returned to the output or some other supply voltage depending on system requirements. In determining a value for this resistor, note that while the output is rated to sink 400 μ A, this sink current adds to battery drain in a low battery condition. Suggested values range from 100k to 1 $M\Omega$. The resistor is not required if this output is unused.



*When $V_{IN} \le 1.3V$, the error flag pin becomes a high impedance, and the error flag voltage rises to its pull-up voltage. Using V_{OUT} as the pull-up voltage (see Figure 2), rather than an external 5V source, will keep the error flag voltage under 1.2V (typ.) in this condition. The user may wish to divide down the error flag voltage using equal-value resistors ($10k\Omega$ suggested), to ensure a low-level logic signal during any fault condition, while still allowing a valid high logic level during normal operation.

FIGURE 1. ERROR Output Timing

PROGRAMMING THE OUTPUT VOLTAGE (LP2951)

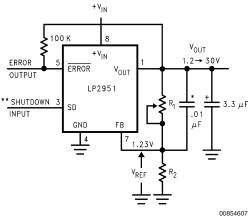
The LP2951 may be pin-strapped for the nominal fixed output voltage using its internal voltage divider by tying the output and sense pins together, and also tying the feedback and V_{TAP} pins together. Alternatively, it may be programmed for any output voltage between its 1.235V reference and its 30V maximum rating. As seen in *Figure 2*, an external pair of resistors is required.

The complete equation for the output voltage is

$$V_{OUT} = V_{REF} \bullet \left(1 + \frac{R_1}{R_2}\right) + I_{FB}R_1$$

where V_{REF} is the nominal 1.235 reference voltage and I_{FB} is the feedback pin bias current, nominally -20nA. The minimum recommended load current of 1µA forces an upper limit of 1.2 $M\Omega$ on the value of R_2 , if the regulator must work with no load (a condition often found in CMOS in standby). I_{FB} will produce a 2% typical error in V_{OUT} which may be eliminated at room temperature by trimming $R_1.$ For better accuracy, choosing R_2 = 100k reduces this error to 0.17% while increasing the resistor program current to 12µA. Since the LP2951 typically draws $60\mu A$ at no load with Pin 2 opencircuited, this is a small price to pay.

Application Hints (Continued)



*See Application Hints

$$V_{out} = V_{Ref} \left(1 + \frac{R_1}{R_2} \right)$$

**Drive with TTL-high to shut down. Ground or leave open if shutdown feature is not to be used.

Note: Pins 2 and 6 are left open.

REDUCING OUTPUT NOISE

In reference applications it may be advantageous to reduce the AC noise present at the output. One method is to reduce the regulator bandwidth by increasing the size of the output capacitor. This is the only way noise can be reduced on the 3 lead LP2950 but is relatively inefficient, as increasing the capacitor from $1\mu F$ to $220\mu F$ only decreases the noise from $430\mu V$ to $160\mu V$ rms for a 100kHz bandwidth at 5V output.

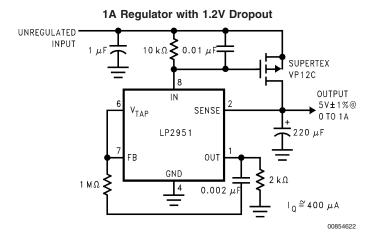
Noise can be reduced fourfold by a bypass capacitor across ${\rm R}_{\rm 1},$ since it reduces the high frequency gain from 4 to unity. Pick

$$C_{\text{BYPASS}} \cong \frac{1}{2\pi R_1 \cdot 200 \text{ Hz}}$$

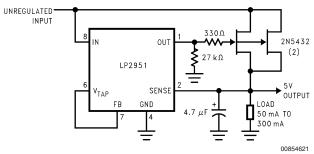
or about $0.01\mu F$. When doing this, the output capacitor must be increased to $3.3\mu F$ to maintain stability. These changes reduce the output noise from $430\mu V$ to $100\mu V$ rms for a 100kHz bandwidth at 5V output. With the bypass capacitor added, noise no longer scales with output voltage so that improvements are more dramatic at higher output voltages.

FIGURE 2. Adjustable Regulator

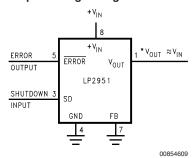
Typical Applications



300mA Regulator with 0.75V Dropout

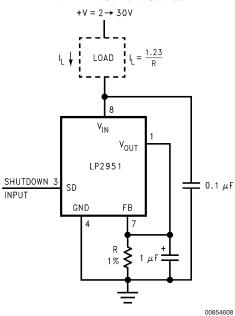


Wide Input Voltage Range Current Limiter

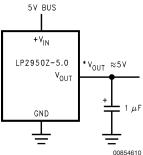


*Minimum input-output voltage ranges from 40mV to 400mV, depending on load current. Current limit is typically 160mA.

Low Drift Current Source

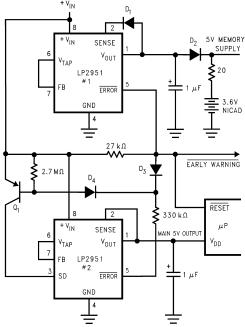


5 Volt Current Limiter



*Minimum input-output voltage ranges from 40mV to 400mV, depending on load current. Current limit is typically 160mA.

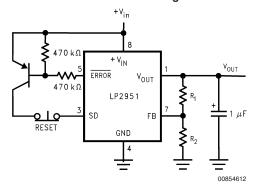
Regulator with Early Warning and Auxiliary Output



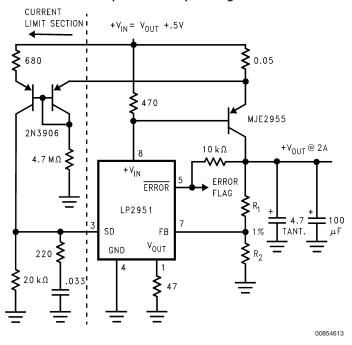
00854611

- Early warning flag on low input voltage
- Main output latches off at lower input voltages
- \blacksquare Battery backup on auxiliary output
- Operation: Reg. #1's V_{out} is programmed one diode drop above 5V. Its error flag becomes active when $V_{in} \le 5.7V$. When V_{in} drops below 5.3V, the error flag of Reg. #2 becomes active and via Q1 latches the main output off. When V_{in} again exceeds 5.7V Reg. #1 is back in regulation and the early warning signal rises, unlatching Reg. #2 via D3.

Latch Off When Error Flag Occurs



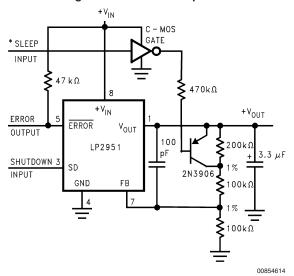
2 Ampere Low Dropout Regulator



$$V_{out} = 1.23V \left(1 + \frac{R_1}{R_2}\right)$$

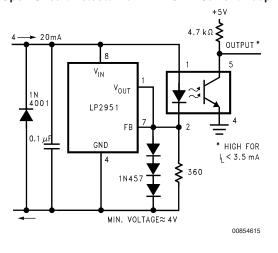
For $5V_{out}$, use internal resistors. Wire pin 6 to 7, & wire pin 2 to $+V_{out}$ Bus.

5V Regulator with 2.5V Sleep Function

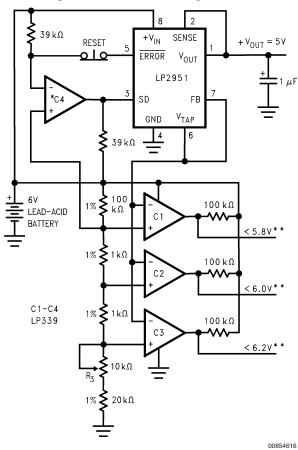


*High input lowers V_{out} to 2.5V

Open Circuit Detector for 4 ightarrow 20mA Current Loop

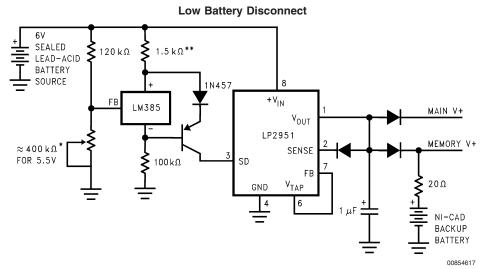


Regulator with State-of-Charge Indicator



*Optional Latch off when drop out occurs. Adjust R3 for C2 Switching when V_{in} is 6.0V.

 $[\]ensuremath{^{**}}\mbox{Outputs}$ go low when $\mbox{V}_{\mbox{\scriptsize in}}$ drops below designated thresholds.

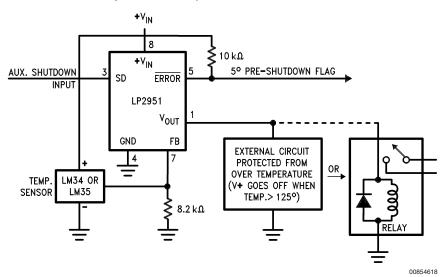


For values shown, Regulator shuts down when $V_{in} < 5.5V$ and turns on again at 6.0V. Current drain in disconnected mode is $\approx 150 \mu A$.

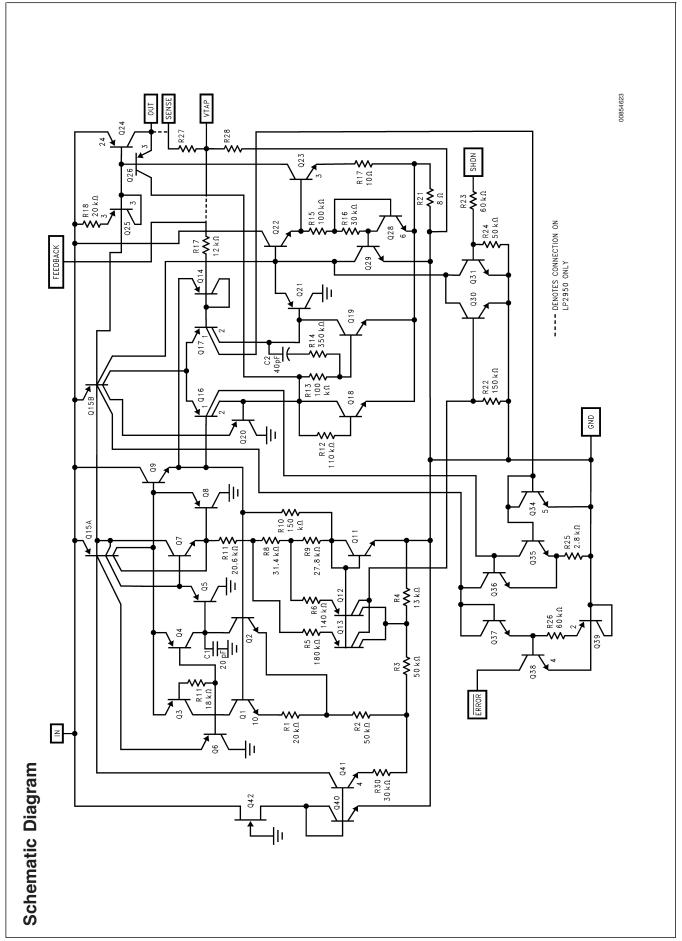
^{*}Sets disconnect Voltage

^{**}Sets disconnect Hysteresis

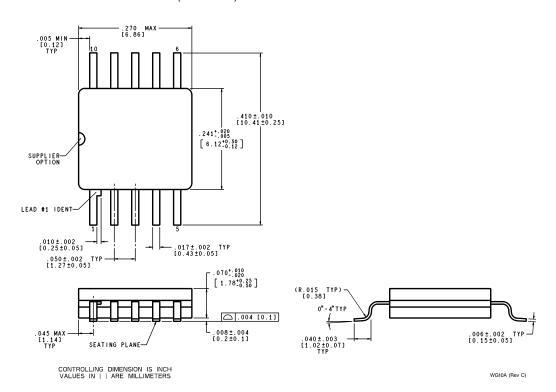
System Overtemperature Protection Circuit



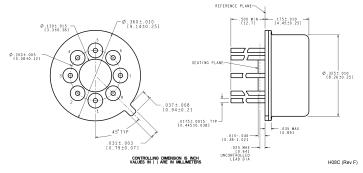
LM34 for 125°F Shutdown LM35 for 125°C Shutdown



Physical Dimensions inches (millimeters) unless otherwise noted



Order Number LP2951WG/883 NS Package Number WG10A



Metal Can Package (H) NS Package Number H08C

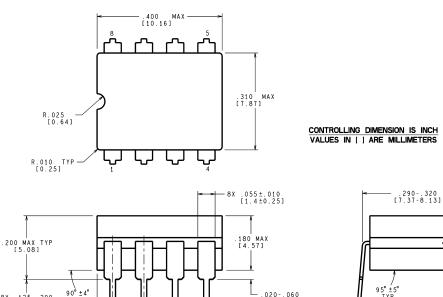
GLASS SEALANT

.010 ±.002 [0.25 ±0.05]

J08A (Rev M)

. 310 - . 410 [7.87-10.41]

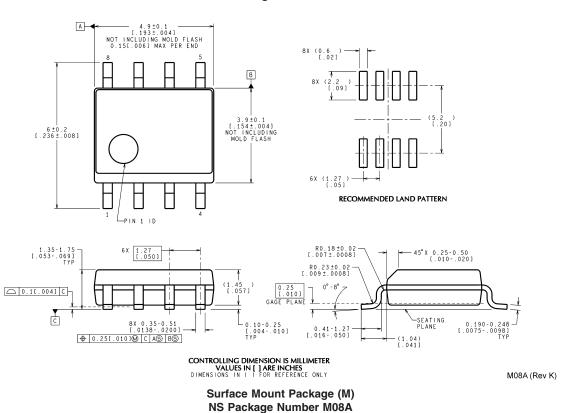
Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



.100±.010 [2.54±0.25]

Ceramic Dual-In-Line Package (J) NS Package Number J08A

- 8X .018 ± .003 [0.46 ±0.07]



Physical Dimensions inches (millimeters) unless otherwise noted (Continued) 0.373 - 0.400 $\overline{(9.474 - 10.16)}$ 0.090 (2.286) 7 6 5 8 7 0.092 $\boldsymbol{0.032 \pm 0.005}$ (2.337) (0.813 ± 0.127) $\underline{\textbf{0.250}\pm\textbf{0.005}}$ RAD PIN NO. 1 IDENT $\overline{(6.35 \pm 0.127)}$ PIN NO. 1 IDENT OPTION 1 1 2 3 0.040 TYP → 0.280 MIN $\frac{0.030}{(0.762)}$ MAX (7.112) OPTION 2 (1.016) 0.039 (0.991) 0.145 - 0.200 0.300 - 0.320

(3.683 - 5.080)

N08E (REV F)

0.020

(0.508)

 0.130 ± 0.005 (3.302 ± 0.127) 0.125 - 0.140

(3.175 - 3.556)

 $\frac{0.018 \pm 0.003}{(0.457 \pm 0.076)}$

 0.100 ± 0.010

 (2.540 ± 0.254)

0.060 (1.524)

20° ± 1°

0.065

(1.651)

<u>0.12</u>5

(3.175) DIA NOM

 $\underline{0.045 \pm 0.015}$ $\overline{(1.143\pm0.381)}$

(7.62 - 8.128)

 $0.325 \,{}^{+\, 0.040}_{-\, 0.015}$

 $\overline{\left(8.255 \, {}^{+\, 1.016}_{-\, 0.381}\right)}$

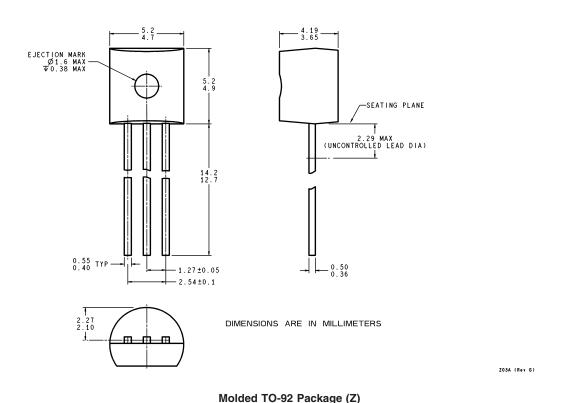
95° ± 5°

0.009 - 0.015

(0.229 - 0.381)

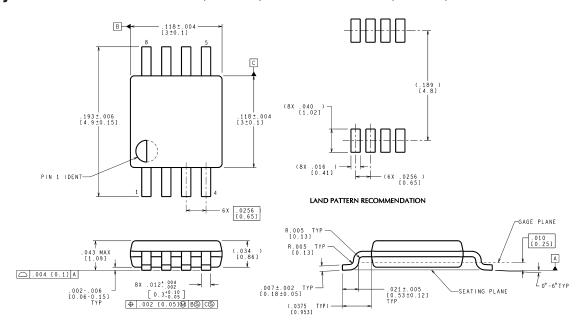
(1.270) Molded Dual-In-Line Package (N) NS Package Number N08E

0.050



NS Package Number Z03A

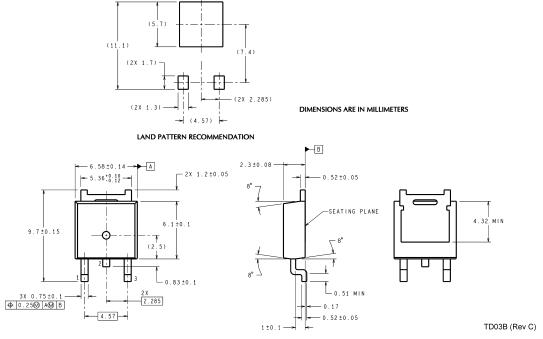
Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



CONTROLLING DIMENSION IS INCH VALUES IN [] ARE MILLIMETERS

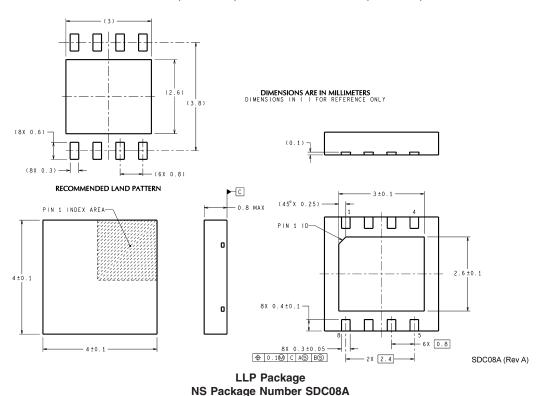
MUA08A (Rev E)

Surface Mount Package (MM) NS Package Number MUA08A



D-Pak Package NS Package Number TD03B

Physical Dimensions inches (millimeters) unless otherwise noted (Continued)



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