## PMOLED display power supply

## Features

■ Inductor switches boost controller

- PFM mode control

■ High efficiency over wide range of load (1 mA to 40 mA )
■ Integrated load disconnect switch

- Over voltage protection with automatic restart
- Soft start with adjustable peak current limit
- Enable pin
- Low shutdown current
- Small external inductor
- Supply voltage from 3.0 V to 5.5 V


## Application

■ PMOLED display driver

## Description

STOD2540 is dedicated to passive matrix OLED (PMOLED) displays for portable handsets and provides the pre-charge and biasing voltage of the column matrix driver as shown in figure 3. The current capability of STOD2540 allows feeding a $1^{\prime \prime}, 1.3^{\prime \prime}$ or $1.5^{\prime \prime}$ color PMOLED. STOD2540 is a boost converter that operates from 3.0 V to 5.5 V and can provide an output voltage as high as 25 V . The output current capability is maximum 40 mA up to 25 V output voltage. The regulation is performed by a resistor divider network (figure 3) that detects the output voltage.


In this state-of-the-art boost converter, a DC current path exists between the battery source and the load. In order to reduce the consumption in shutdown mode, a high-side load isolation switch is necessary to cut this DC current path in standby mode. The load disconnect switch (LDS) act as an isolation switch in shutdown mode.

Table 1. Device summary

| Order code | Package | Packaging |
| :---: | :---: | :---: |
| STOD2540PUR | QFN8 $(3 \times 3 \mathrm{~mm})$ | 3000 parts per reel |

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## 1 Diagram

Figure 1. Block diagram


## 2 Pin configuration

Figure 2. Pin connections (top through view)


Table 2. Pin description

| Pin $\mathbf{n}^{\circ}$ | Symbol |  |
| :---: | :---: | :--- |
| 1 | $\mathrm{~V}_{\mathrm{I}}$ | Supply voltage |
| 2 | $\mathrm{R}_{\text {SET }}$ | Peak inductor current adjust |
| 3 | AGND | Analog ground |
| 4 | FB | Feedback for the LED current regulation |
| 5 | $\mathrm{~V}_{\mathrm{O}}$ | Output voltage for LED supply |
| 6 | $\mathrm{~V}_{\text {CAP }}$ | Load disconnect switch input |
| 7 | EN | IC enable signal |
| 8 | $\mathrm{~V}_{\text {SW }}$ | Boost switch drain |
|  | PGND | Power ground |

## 3 Maximum ratings

Table 3. Absolute maximum ratings

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{VB}_{\text {SW, }}, \mathrm{VB}_{\mathrm{O}}$ | Breakdown voltage at OUT and SW pin | 40 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | Supply voltage range | 6 | V |
| $\mathrm{R}_{\mathrm{SET}}$ | $\mathrm{R}_{\text {SET }}$ pin | $\mathrm{V}_{1}+0.3$ | V |
| EN | Enable pin | $\mathrm{V}_{1}+0.3$ | V |
| $\mathrm{~V}_{\text {ESD }}$ | ESD ratings, HBM MIL STD 883C | 2 | kV |
| $\mathrm{T}_{\mathrm{STG}}$ | Storage temperature range | -65 to 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{OP}}$ | Operating junction temperature range | -40 to 85 | ${ }^{\circ} \mathrm{C}$ |

Table 4. Thermal data

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{R}_{\text {thJA }}$ | Thermal resistance junction-ambient | 52 | ${ }^{\circ} \mathrm{C} / \mathrm{W}$ |

## 4 Electrical characteristics

$\mathrm{T}_{\mathrm{J}}=40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{I}}=3.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{EN}}=3 \mathrm{~V}, \mathrm{C}_{\mathrm{I}}=\mathrm{C}_{\mathrm{O}}=4.7 \mu \mathrm{~F}, \mathrm{~L}=4.7 \mu \mathrm{H}, \mathrm{R}_{1}=180 \mathrm{k} \Omega$, $R_{2}=10 \mathrm{k} \Omega, \mathrm{V}_{\mathrm{O}}=24 \mathrm{~V}$, Typ. values @ $25^{\circ} \mathrm{C}$, unless otherwise specified.

Table 5. Electrical characteristics

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $V_{1}$ | Input voltage range |  | 3.0 |  | 5.5 | V |
| $\mathrm{V}_{\mathrm{O}}$ | Regulated output voltage | $\mathrm{V}_{1}=3 \mathrm{~V}$ to 5.5 V | $V_{1}+0.5$ | 25 | 35 | V |
| $V_{\text {OVP }}$ | Over voltage protection on output |  | 35 |  |  | V |
| 10 | Continuous output current | $\mathrm{V}_{\mathrm{O}}=25 \mathrm{~V}$ | 1 |  | 40 | mA |
| $I_{\text {SD }}$ | Stand-by current | $\mathrm{V}_{\mathrm{EN}}=$ Low, $\mathrm{V}_{\mathrm{I}}=3.6 \mathrm{~V}$ |  |  | 3 | $\mu \mathrm{A}$ |
|  |  | $\mathrm{V}_{\text {EN }}=$ Low, $\mathrm{V}_{\mathrm{I}}=3 \mathrm{~V}$ to 4.2 V |  |  | 10 |  |
| $\mathrm{I}_{\mathrm{Q}}$ | Quiescent current consumption | $\mathrm{V}_{1}=3 \mathrm{~V}$ to $4.2 \mathrm{~V} @ 25^{\circ} \mathrm{C}$ |  | 0.4 | 0.8 | mA |
|  |  | $\mathrm{V}_{1}=5.5 \mathrm{~V}$ @ $25^{\circ} \mathrm{C}$ |  | 0.8 | 1.2 |  |
| $\mathrm{R}_{\text {DSON-SW }}$ | Boost switch $\mathrm{R}_{\text {DSON }}{ }^{(1)}$ | $\mathrm{V}_{\mathrm{I}}=4.2 \mathrm{~V}, \mathrm{I}_{\mathrm{SW}}=100 \mathrm{~mA}$ |  | 0.4 |  | $\Omega$ |
|  | BVDS Breakdown voltage |  | 40 |  |  | V |
| $\mathrm{R}_{\text {DSON-LDS }}$ | $\mathrm{R}_{\text {DSON }}{ }^{(1)}$ | $\mathrm{V}_{\mathrm{O}}=25 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=30 \mathrm{~mA}$ |  | 2 |  | $\Omega$ |
|  | BVDS Breakdown voltage |  | 40 |  |  | V |
| I lim-ADJ | Peak inductor limit range ${ }^{(1)}$ | $\mathrm{R}_{\text {SET }}=10 \mathrm{k} \Omega$ to $100 \mathrm{k} \Omega$ | 0.2 |  | 1.1 | A |
| ILIM-max | Maximum peak inductor current (1) | $\mathrm{V}_{\mathrm{I}}=3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{R}_{\mathrm{SET}}=\mathrm{V}_{\mathrm{I}}$ | 0.75 |  | 1.2 | A |
| FB | Feedback voltage | $5 \%$ @ $25^{\circ} \mathrm{C}$ | 1.18 | 1.24 | 1.30 | V |
| TON_MAX | Maximum ON time | $\mathrm{V}_{1}=4.2 \mathrm{~V}$ |  | 5.5 |  | $\mu \mathrm{s}$ |
| TOFF_MIN | Minimum OFF time | $\mathrm{V}_{1}=4.2 \mathrm{~V}$ |  | 300 |  | ns |
| Eff | Efficiency, $\mathrm{V}_{\mathrm{I}}=3.6 \mathrm{~V}^{(1)}$ | $\mathrm{I}_{\mathrm{O}}=1 \mathrm{~mA}$ to 5 mA | 65 |  |  | \% |
|  |  | $\mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}$ to 40 mA | 70 |  |  |  |
|  | Efficiency, $\mathrm{V}_{\mathrm{I}}=4.2 \mathrm{~V}{ }^{(1)}$ | $\mathrm{I}_{\mathrm{O}}=1 \mathrm{~mA}$ to 5 mA | 65 |  |  |  |
|  |  | $\mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}$ to 40 mA | 70 |  |  |  |
| Ripple | Output ripple and noise | $\mathrm{V}_{\mathrm{I}}=3.6 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}, \mathrm{~V}_{\mathrm{O}}=24 \mathrm{~V}$ |  | 1.3 |  | \% |
|  |  | $\mathrm{V}_{1}=3.6 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=30 \mathrm{~mA}, \mathrm{~V}_{\mathrm{O}}=24 \mathrm{~V}$ |  | 1.3 |  |  |
|  |  | $\mathrm{V}_{\mathrm{I}}=4.2 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}, \mathrm{~V}_{\mathrm{O}}=24 \mathrm{~V}$ |  | 1.3 |  |  |
|  |  | $\mathrm{V}_{\mathrm{I}}=4.2 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=30 \mathrm{~mA}, \mathrm{~V}_{\mathrm{O}}=24 \mathrm{~V}$ |  | 1.3 |  |  |
| $\mathrm{OV}_{\text {HYST }}$ | Over-voltage hysteresis |  |  | 2 |  | V |

Table 5. Electrical characteristics (continued)

| Symbol | Parameter | Test condition | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{EN}}$ | Enable input logic low | Disable Low $\mathrm{V}_{\text {IL }}$ |  |  | 0.3 | V |
|  | Enable input logic high | Enable High $\mathrm{V}_{\mathrm{IH}}$ | 1.2 |  |  |  |
| Line_V $\mathrm{V}_{\text {FB }}$ | Line regulation $\mathrm{V}_{\mathrm{FB}}$ | $\mathrm{V}_{\mathrm{I}}=3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}$ |  | 5 | 35 | mV |
| Load_V ${ }_{\text {FB }}$ | Line regulation $\mathrm{V}_{\mathrm{FB}}$ | $\mathrm{V}_{\mathrm{I}}=3 \mathrm{~V}$ to $5.5 \mathrm{~V}, \mathrm{I}_{\mathrm{O}}=5 \mathrm{~mA}$ |  | 5 | 35 | mV |

1. Guaranteed by design.

## 5 Functional description

### 5.1 Boost controller

STOD2540 is a boost converter operating in PFM (pulsed frequency modulation) mode. The converter monitors the output voltage through the bridge resistor divider $R_{1}$ and $R_{2}$ and when the feedback voltage falls below the reference voltage, REF2, the boost switch $t_{\text {sw }}$ turns ON and the current ramps up. The inductor current is measured by detect the temperature compensated drain voltage of the boost MOSFET. The boost turns off when its drain voltage reaches the reference REF1, the main switch remains off until the minimum off time ( 300 ns typical) has passed and the feedback voltage is below the reference again. A maximum ON time of $4 \mu$ s prevent the switch $t_{\text {Sw }}$ to stay ON during a too long period of time.

In order to well calculate the bridge resistors values with a fixed $\mathrm{V}_{\mathrm{O}}$, the following formula can be used:
$\left(\mathrm{V}_{\mathrm{O}} / 1.24\right)-1=\mathrm{R}_{1} / \mathrm{R}_{2}$

### 5.2 Adjustable peak inductor current limit

The peak inductor current is monitored by sensing the drain voltage of the switch $\mathrm{t}_{\mathrm{sw}}$.
Since it exceeds the temperature compensated and supply voltage compensated reference REF1, the RS flip flop is reset and $\mathrm{t}_{\mathrm{Sw}}$ is turned OFF.

By connecting a resistance between the pin $R_{\text {SET }}$ and GND, the peak current limit can be adjusted from 200 mA to 1.1 A ( $\mathrm{R}_{\mathrm{SET}}$ from $10 \mathrm{k} \Omega$ to $100 \mathrm{k} \Omega$ ). When the pin $R_{\mathrm{SET}}$ is directly connected to $V_{1}$, the default value is 1 A .

### 5.3 Enable

The ENABLE pin is a high logic input signal and allows turning on/off the controller without cutting the input voltage from the boost regulator circuit. With a high input voltage ( $1.2 \mathrm{~V}<$ $\mathrm{V}_{\mathrm{EN}}<\mathrm{V}_{\mathrm{I}}+0.3 \mathrm{~V}$ ) on this pin, the device is allowed to work normally. No pull-up or pull down is present on this pin.

### 5.4 OVP

If the regulation loop is cut, there is no signal at the feedback pin, the PFM controller will then continue to switch without control and generate an output voltage at the $\mathrm{SW}, \mathrm{V}_{\mathrm{CAP}}$ and $\mathrm{V}_{\mathrm{O}}$ pin exceeding the breakdown value $\mathrm{V}_{\mathrm{BSW}}, \mathrm{V}_{\mathrm{BCAP}}$ and $\mathrm{V}_{\mathrm{BO}}$.

The over voltage protection (OVP) senses the voltage at the $\mathrm{V}_{\mathrm{CAP}}$ pin. When the voltage exceed the breakdown voltage of the device the controller is automatically turned off.
A hysteresis control enables the device to automatically restart when the output voltage drops below a 2 V typical value.

### 5.5 Load isolation switch

When the device is in shutdown mode, a DC current path always exists between the power source and the load; increasing the standby consumption. A high side switch LDS isolates the load from the source when the STOD2540 is disabled.

### 5.6 Efficiency

The total consumption of some PMOLED display, can be as low as 1 mA . In order to increase the battery run time of the device, STOD2540 offers a high efficiency over a wide range of load and input voltage range.

### 5.7 Under voltage lockout (UVLO)

The minimum supply voltage is 3.0 V , under this value the under voltage lockout circuit operates with typical threshold 2.8 V .
When supply voltage is below 3.0 V , possible noise in the supply line could disturb the UVLO circuit causing loss of output regulation. This behavior is eliminated choosing $\mathrm{C}_{\mathrm{l}}=10 \mu \mathrm{~F}$ or higher.

## 6 Typical application

Figure 3. Basic connection


Table 6. External components (see Figure 3)

| Symbol | Parameter | Test | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| D | Boost schottky diode | VRRM | 30 |  |  | V |
|  |  | $V_{F}$ at $I_{F}=300 \mathrm{~mA}, \mathrm{~T}_{J}=25^{\circ} \mathrm{C}$ |  |  | 0.5 | V |
|  |  | $\mathrm{I}_{\mathrm{R}}$ at $\mathrm{V}_{\mathrm{R}}=10 \mathrm{~V}, \mathrm{~T}_{J}=25^{\circ} \mathrm{C}$ |  |  | 30 | $\mu \mathrm{A}$ |
| $\mathrm{R}_{1}$ | Feedback resistor |  |  | 180 |  | $\mathrm{k} \Omega$ |
| $\mathrm{R}_{2}$ | Feedback resistor |  |  | 10 |  |  |
| $\mathrm{R}_{\text {SET }}$ | Peak current limit adjust | $\mathrm{I}_{\mathrm{PK}}=200 \mathrm{~mA}$ to 1.1 A | 10 |  | 100 |  |
| $\mathrm{C}_{1}$ | Input ceramic type low ESR | Ceramic type |  | 4.7 |  | $\mu \mathrm{F}$ |
| $\mathrm{Co}_{0}$ | Output capacitance: ceramic low ESR | Capacitance | 4.7 |  |  | $\mu \mathrm{F}$ |
|  |  | Voltage | 42 |  |  | V |
|  |  | ESR |  |  | 1.6 | W |
| L | Boost inductor (height < 2mm) | Inductance |  |  | 4.7 | $\mu \mathrm{H}$ |
|  |  | $\mathrm{I}_{\text {SAT }}, \mathrm{R}_{\text {SET }}$ pin to $\mathrm{V}_{\text {I }}$ |  |  | 1 | A |

Note: $\quad$ The external components suggested in this document should be considered as a design reference guide. The performances mentioned in the electrical characteristics table are not guaranteed for all the possible electrical parameters of the components included in this list. On other hand, the operation of STOD2540 is not limited to the use of components included in this list.

### 6.1 Demonstration board

Figure 4. Suggested demonstration board schematic (top layer view)


## 7 Typical performance characteristics

$$
\begin{aligned}
& \mathrm{T}_{J}=40^{\circ} \mathrm{C} \text { to } 85^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{l}}=3.6 \mathrm{~V}, \mathrm{~V}_{\mathrm{EN}}=3 \mathrm{~V}, \mathrm{C}_{I}=\mathrm{C}_{\mathrm{O}}=4.7 \mu \mathrm{~F}, \mathrm{~L}=4.7 \mu \mathrm{H}, \mathrm{R}_{1}=180 \mathrm{k} \Omega, \\
& \mathrm{R}_{2}=10 \mathrm{k} \Omega, \mathrm{~V}_{\mathrm{O}}=24 \mathrm{~V} \text {, Typ. values } @ 25^{\circ} \mathrm{C} \text {, unless otherwise specified. }
\end{aligned}
$$

Figure 5. Efficiency vs. output current


Figure 6. Efficiency vs. input voltage

Figure 7. $\quad V_{\text {EN }}$ vs. temperature


Figure 8. $\quad \mathrm{V}_{\mathrm{FB}}$ vs. input voltage


Figure 9. $\quad V_{F B}$ vs. output current


Figure 10. $\mathrm{V}_{\mathrm{FB}}$ vs. output current


Figure 11. $V_{\text {OVP }}$ vs. temperature


Figure 13. ILIM_mAX vs. input voltage


Figure 15. $\mathrm{l}_{\text {LIM_MAX }}$ vs. $\mathrm{R}_{\text {SET }}$


Figure 12. $V_{\text {RIPPLE }}$ vs. input voltage


Figure 14. ILIM_MAX vs. temperature


Figure 16. $\mathrm{I}_{\mathrm{Q}}$ Vs. temperature


Figure 17. $t_{\text {ON_MAX }}$ vs. temperature


Figure 19. Line $\mathrm{V}_{\mathrm{FB}}$ vs. temperature


Figure 18. $t_{\text {OFF_MIN }}$ vs. temperature


Figure 20. Load $\mathrm{V}_{\mathrm{FB}}$ vs. temperature


## 8 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK ${ }^{\circledR}$ packages, depending on their level of environmental compliance. ECOPACK ${ }^{\circledR}$ specifications, grade definitions and product status are available at: www.st.com. ECOPACK ${ }^{\circledR}$ is an ST trademark.

Table 7. QFN8 (3 x 3 mm ) mechanical data

| Dim. | mm. |  |  |
| :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |
| A | 0.80 | 0.90 | 1.00 |
| A1 | 0 | 0.02 | 0.05 |
| A3 |  | 0.20 |  |
| b | 0.25 | 0.30 | 0.35 |
| D | 2.85 | 3.00 | 3.15 |
| D2 | 2.49 | 2.64 | 2.74 |
| E | 2.85 | 3.00 | 3.15 |
| E2 | 1.75 | 1.90 | 2.00 |
| e |  | 0.65 |  |
| L | 0.30 | 0.40 | 0.50 |

Figure 21. Drawing dimension QFN8 (3x3 mm)


Tape \& reel QFNxx/DFNxx (3x3) mechanical data

| Dim. | mm. |  |  | inch. |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. | Min. | Typ. | Max. |
| A |  |  | 180 |  |  | 7.087 |
| C | 12.8 |  | 13.2 | 0.504 |  | 0.519 |
| D | 20.2 |  |  | 0.795 |  |  |
| N | 60 |  |  | 2.362 |  |  |
| T |  |  | 14.4 |  | 0.130 |  |
| Ao |  | 3.3 |  |  | 0.130 |  |
| Bo |  | 1.1 |  |  | 0.043 |  |
| Ko |  | 4 |  |  | 0.157 |  |
| Po |  | 8 |  |  | 0.315 |  |
| P |  |  |  |  |  |  |



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Figure 22. QFN8 (3x3) footprint recommended data
DIMENSIONS IN MILLIMETRES


## 9 Revision history

Table 8. Document revision history

| Date | Revision | Changes |
| :---: | :---: | :--- |
| 22-Mar-2006 | 1 | Initial release. |
| 03-Apr-2006 | 2 | Add fig. 2 demonstration board on page 3. |
| 08-Jun-2006 | 3 | Description in cover page updated. |
| 23-Jun-2006 | 4 | Change range of $R_{\text {SET }}$ value and add description paragraph 5.7. |
| 11-Sep-2006 | 5 | Mistake on table 4 Ripple test value 3.6 V ==> 4.2 V. |
| 27-Feb-2009 | 6 | Modified mechanical data. |
| 03-Mar-2009 | 7 | Modified packaging Table 1 on page 1. |
| 11-Mar-2009 | 8 | Modified Figure 2 on page 4 and added Figure 22 on page 18. |
| 25-Nov-2009 | 9 | Modified Table 1 on page 1. |
| 01-Jul-2010 | 10 | Modified: Table 7 on page 15, Figure 21 on page 16 and Figure 22 on page 18. |

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