

## High-speed low-power quad operational amplifier with dual standby position

### Features

- Low supply current: 4.5 mA
- High speed: 150 MHz - 110 V/ $\mu$ s
- Unity gain stability
- Low offset voltage: 4 mV
- Low noise 4.2 nV/ $\sqrt{\text{Hz}}$
- Specified for 600  $\Omega$  and 150  $\Omega$  loads
- High video performances:
  - differential gain: 0.03%
  - differential phase: 0.07°
  - gain flatness: 6 MHz, 0.1 dB max. at 10 db gain

### Applications

- Video buffers
- A/D converter drivers

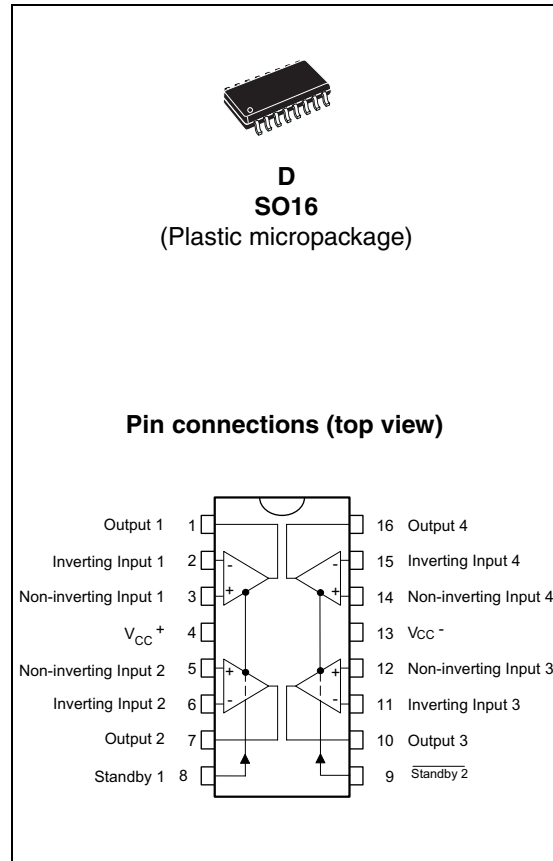
### Description

The TSH95 is a low-power, high-frequency quad operational amplifier designated for high-quality video processing. The device offers an excellent speed consumption ratio with 4.5 mA per amplifier for a 150 MHz bandwidth.

A high slew rate and low noise also make it suitable for high-quality audio applications.

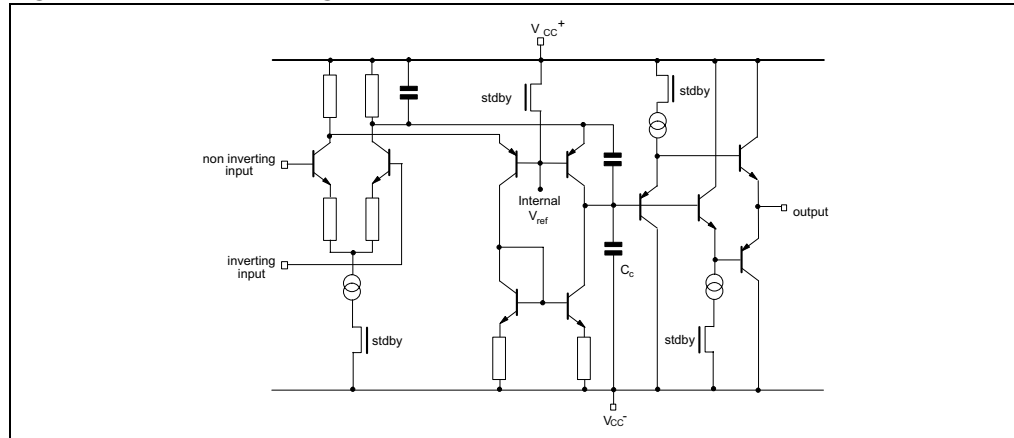
The TSH95 offers two separate complementary STANDBY pins: STANDBY 1 acting on operators 1 and 2, and STANDBY 2 acting on operators 3 and 4.

These pins reduce the consumption of the corresponding operators and put the output in a high impedance state.



# 1 Schematic diagram

Figure 1. Schematic diagram



## 2 Absolute maximum ratings and operating conditions

**Table 1. Absolute maximum ratings**

Symbol	Parameter	Value	Unit
$V_{CC}$	Supply voltage <sup>(1)</sup>	14	V
$V_{id}$	Differential input voltage <sup>(2)</sup>	±5	V
$V_i$	Input voltage <sup>(3)</sup>	-0.3 to 12	V
$T_{oper}$	Operating free-air temperature range	-40 to +125	°C
$T_{stg}$	Storage temperature range	-65 to +150	°C
ESD	CDM: charged device model <sup>(4)</sup>	1.5	kV
	HBM: human body model <sup>(5)</sup>	2	kV
	MM: machine model <sup>(6)</sup>	200	V

- All voltages values, except differential voltage, are with respect to network ground terminal.
- Differential voltages are the non-inverting input terminal with respect to the inverting input terminal.
- The magnitude of input and output voltages must never exceed  $V_{CC}^+ + 0.3$  V.
- Charged device model: all pins and the package are charged together to the specified voltage and then discharged directly to the ground through only one pin. This is done for all pins.
- Human body model: a 100 pF capacitor is charged to the specified voltage, then discharged through a 1.5 kΩ resistor between two pins of the device. This is done for all couples of connected pin combinations while the other pins are floating.
- Machine model: a 200 pF capacitor is charged to the specified voltage, then discharged directly between two pins of the device with no external series resistor (internal resistor < 5 Ω). This is done for all couples of connected pin combinations while the other pins are floating

**Table 2. Operating conditions**

Symbol	Parameter	Value	Unit
$V_{CC}$	Supply voltage	7 to 12	V
$V_{ic}$	Common mode input voltage range	$V_{CC}^- + 2$ to $V_{CC}^+ - 1$	V

### 3 Electrical characteristics

**Table 3.**  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = -5\text{ V}$ , pin 8 connected to 0 V, pin 9 connected to  $V_{CC+}$ ,  $T_{amb} = 25^\circ\text{ C}$  (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
$V_{io}$	Input offset voltage $V_{ic} = V_o = 0\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$			4 6	mV
$I_{io}$	Input offset current $T_{min.} \leq T_{amb} \leq T_{max.}$		1	2 5	$\mu\text{A}$
$I_{ib}$	Input bias current $T_{min.} \leq T_{amb} \leq T_{max.}$		5	15 20	$\mu\text{A}$
$I_{CC}$	Supply current (per amplifier, no load) $T_{min.} \leq T_{amb} \leq T_{max.}$		4.5	6 8	mA
CMR	Common-mode rejection ratio $V_{ic} = -3\text{ V to } +4\text{ V}$ , $V_o = 0\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$	80 70	100		dB
SVR	Supply voltage rejection ratio $V_{CC} = \pm 5\text{ V to } \pm 3\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$	60 50	75		dB
$A_{vd}$	Large signal voltage gain $R_L = 10\text{ k}\Omega$ , $V_o = \pm 2.5\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$	57 54	70		dB
$V_{OH}$	High level output voltage $V_{id} = 1\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$ $R_L = 600\ \Omega$ $R_L = 150\ \Omega$ $R_L = 150\ \Omega$	3 2.5 2.4	3.5 3		V
$V_{OL}$	Low level output voltage $V_{id} = 11\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$ $R_L = 600\ \Omega$ $R_L = 150\ \Omega$ $R_L = 150\ \Omega$		-3.5 -2.8	-3 -2.5 -2.4	V
$I_o$	Output short circuit current $V_{id} = \pm 1\text{ V}$ $T_{min.} \leq T_{amb} \leq T_{max.}$ Source Sink Source Sink	20 20 15 15	36 40		mA
GBP	Gain bandwidth product $A_{VCL} = 100$ , $R_L = 600\ \Omega$ , $C_L = 15\text{ pF}$ , $f = 7.5\text{ MHz}$	90	150		MHz
$f_T$	Transition frequency		90		MHz
SR	Slew rate $V_{in} = -2\text{ to } +2\text{ V}$ , $R_L = 600\ \Omega$ , $C_L = 15\text{ pF}$	62	110		V/ $\mu\text{s}$
$e_n$	Equivalent input voltage noise $R_s = 50\ \Omega$ , $f = 1\text{ kHz}$		4.2		nV/ $\sqrt{\text{Hz}}$
$\phi_m$	Phase margin $A_{VM} = +1$		35		Degrees
$V_{O1}/V_{O2}$	Channel separation $f = 1\text{ MHz to } 10\text{ MHz}$		65		dB
Gf	Gain flatness $f = \text{DC to } 6\text{ MHz}$ , $A_{VCL} = 10\text{ dB}$			0.1	dB
THD	Total harmonic distortion $f = 1\text{ kHz}$ , $V_o = \pm 2.5\text{ V}$ , $R_L = 600\ \Omega$		0.01		%

**Table 3.**  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = -5\text{ V}$ , pin 8 connected to 0 V, pin 9 connected to  $V_{CC+}$ ,  $T_{amb} = 25^\circ\text{ C}$  (unless otherwise specified) (continued)

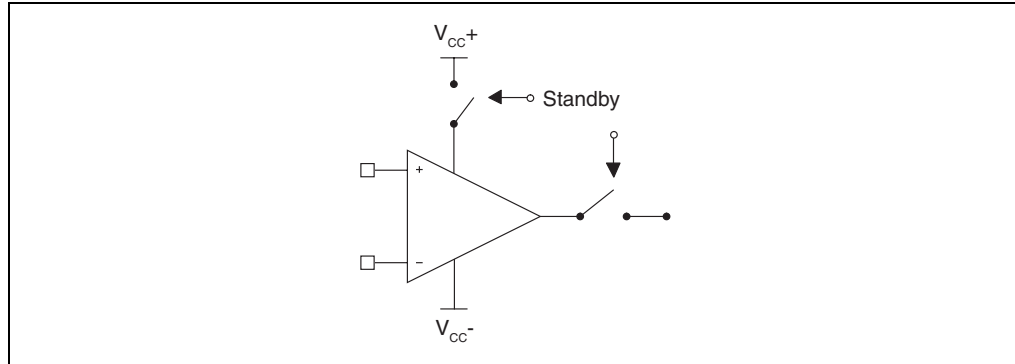
Symbol	Parameter	Min.	Typ.	Max.	Unit
$\Delta G$	Differential gain $f = 3.58\text{ MHz}$ , $A_{VCL} = +2$ , $R_L = 150\ \Omega$		0.03		%
$\Delta\phi$	Differential phase $f = 3.58\text{ MHz}$ , $A_{VCL} = +2$ , $R_L = 150\ \Omega$		0.07		Degree

**Table 4.** Standby mode:  $V_{CC+} = 5\text{ V}$ ,  $V_{CC-} = -5\text{ V}$ ,  $T_{amb} = 25^\circ\text{ C}$  (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
$V_{SBY}$	Pin 8/9 threshold voltage for STANDBY mode	$V_{CC+} - 2.2$	$V_{CC+} - 1.6$	$V_{CC+} - 1.0$	V
$I_{CC\ SBY}$	Total consumption: Pin 8 (Standby 1) = 0, Pin 9 ( <u>Standby 2</u> ) = 0 Pin 8 (Standby 1) = 0, Pin 9 ( <u>Standby 2</u> ) = 1 Pin 8 (Standby 1) = 0, Pin 9 ( <u>Standby 2</u> ) = 0		9.4 9.4 0.8		mA
$I_{sol}$	Input/output isolation ( $f = 1\text{ MHz to }10\text{ MHz}$ )		70		dB
$t_{ON}$	Time from Standby mode to Active mode		200		ns
$t_{OFF}$	Time from Active mode to Standby mode		200		ns
$I_D$	Standby driving current		2		pA
$I_{OL}$	Output leakage current		20		pA
$I_{IL}$	Input leakage current		20		pA

**Table 5.** Standby control pin status

Logic input		Status	
Standby 1	Standby 2	Op-amps 1 and 2	Op-amps 2 and 3
0	0	Enable	Standby
0	1	Enable	Enable
1	0	Standby	Standby
1	1	Standby	Enable

**Figure 2. Standby position**

To put the device in standby, a logic level must be applied on the standby MOS input. Since ground is a virtual level for the device, the threshold voltage has been referred to  $V_{CC+}$  at  $V_{CC+} - 1.6$  V typical.

In standby mode, the output goes into high impedance in 200 ns. Note that all maximum ratings must still be followed in this mode. This mode leads to a swing limitation while using the device in a signal multiplexing configuration with followers; the differential input voltage must not exceed  $\pm 5$  V, limiting the input swing to 2.5 Vpp.

## 4 Application information

Figure 3. Signal multiplexing

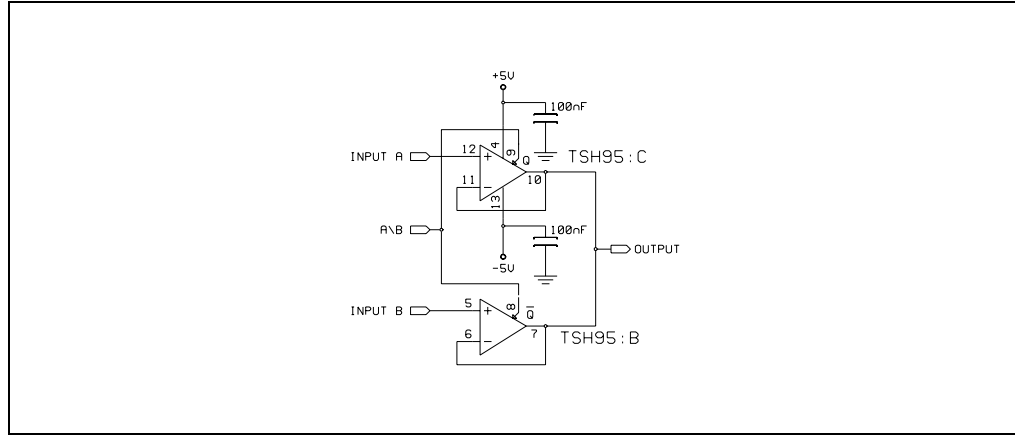
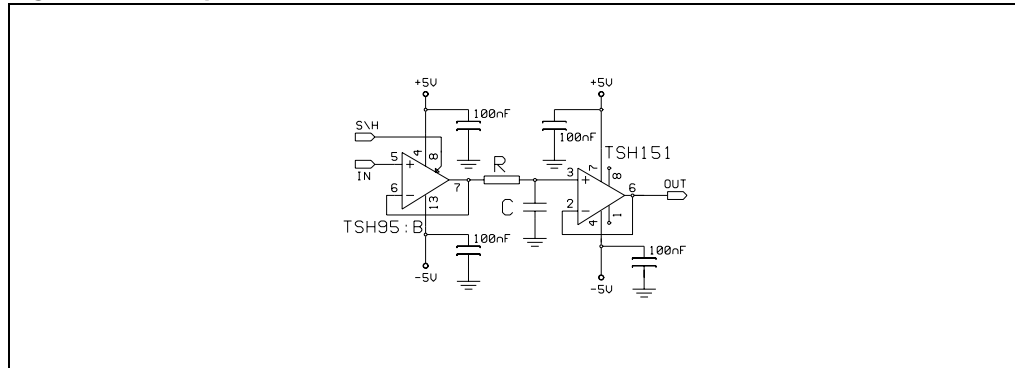


Figure 4. Sample and hold



### 4.1 Printed circuit layout recommendations

As with any high-frequency device, a few rules must be observed when designing the PCB so as to maximize performance.

From the most to the least important points:

- Each power supply lead must be bypassed to ground with a 10 nF ceramic capacitor and a 10  $\mu$ F capacitor placed very close to the device.
- To provide low inductance and low resistance common return, use a ground plane or common point return for power and signal.
- All leads must be wide and as short as possible, especially for the inputs, in order to decrease parasitic capacitance and inductance.
- Use small resistor values to decrease the time constant with parasitic capacitance.
- Choose the smallest-possible component sizes (SMD).
- Decrease the capacitor load at the output to avoid degrading the circuit's stability and cause oscillation. You can also add a serial resistor to minimize its influence.

Figure 5. Large signal follower response

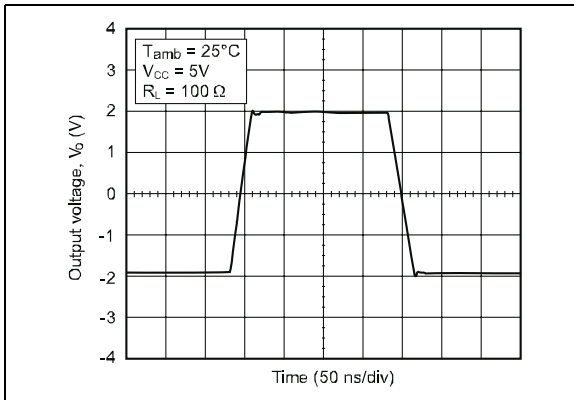


Figure 6. Static open-loop voltage gain

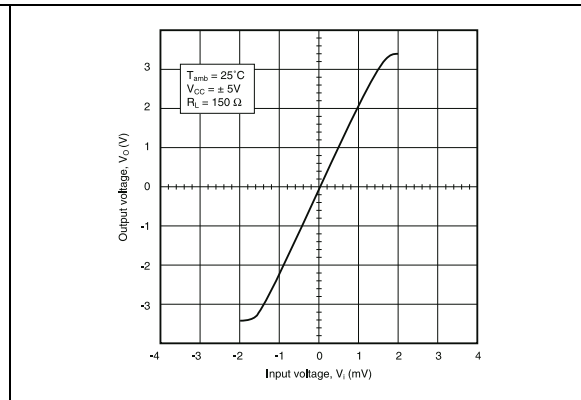


Figure 7. Input offset voltage drift versus temperature

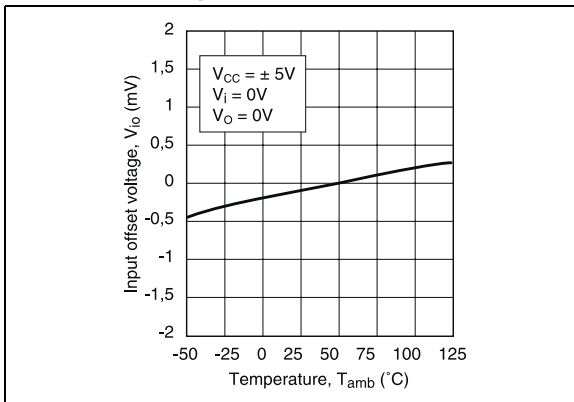


Figure 8. Small signal follower response

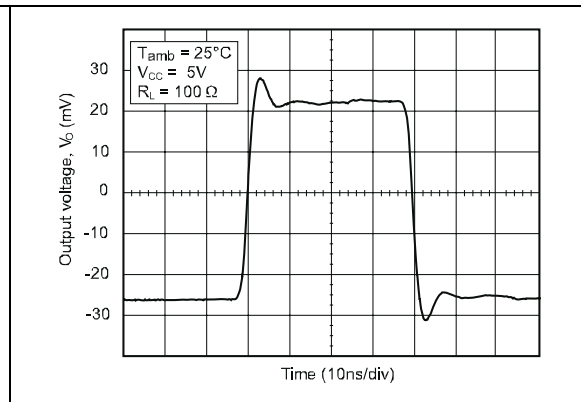


Figure 9. Closed-loop frequency response and phase shift

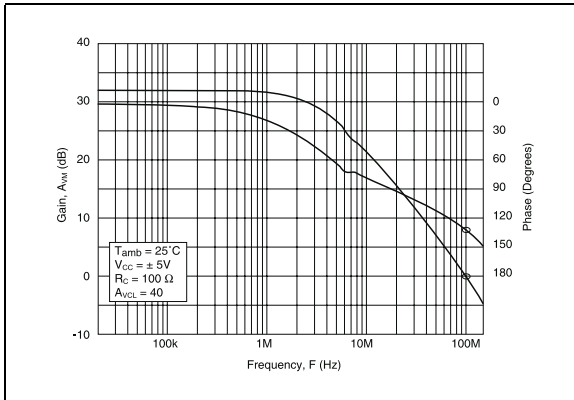
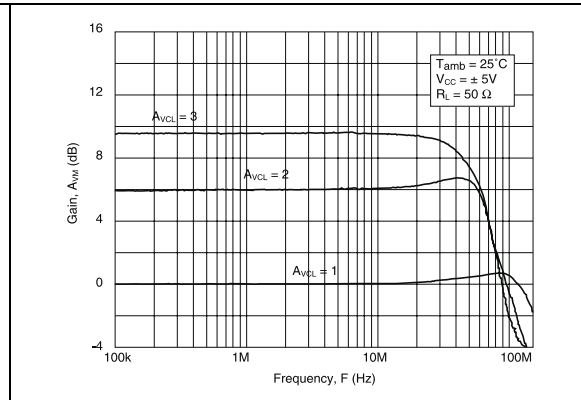
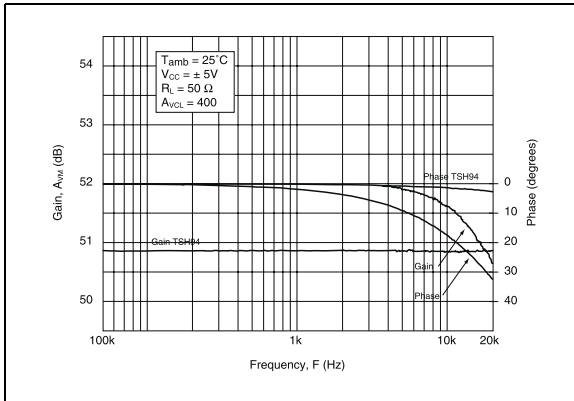


Figure 10. Closed-loop frequency response

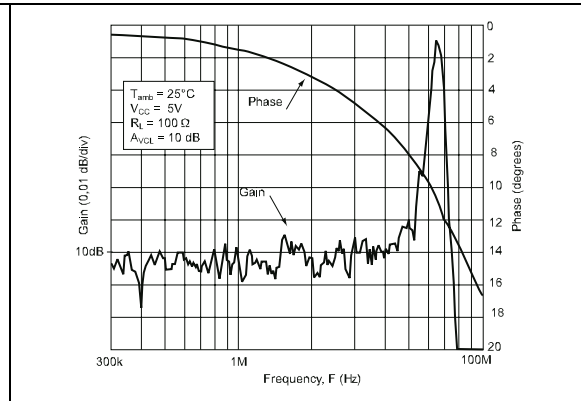




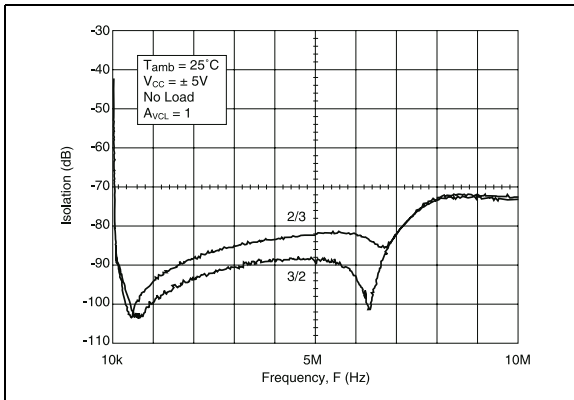
**Figure 11. Audio bandwidth frequency response & phase shift (TSH95 vs standard 15 MHz audio op-amp)**



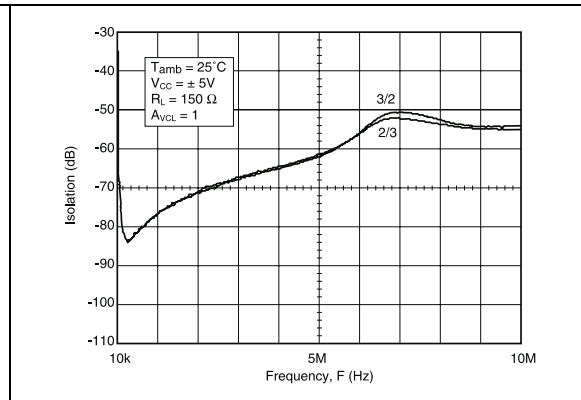
**Figure 12. Gain flatness and phase shift vs. frequency**



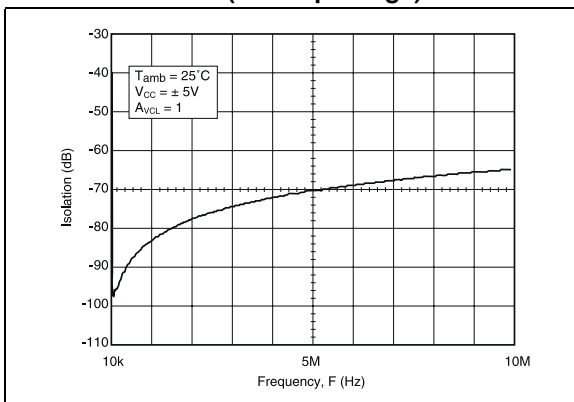
**Figure 13. Crosstalk isolation vs. frequency (SO-16 package)**



**Figure 14. Crosstalk isolation vs. frequency (SO-16 package)**



**Figure 15. Input/output isolation in standby mode (SO-16 package)**



**Figure 16. Standby switching**

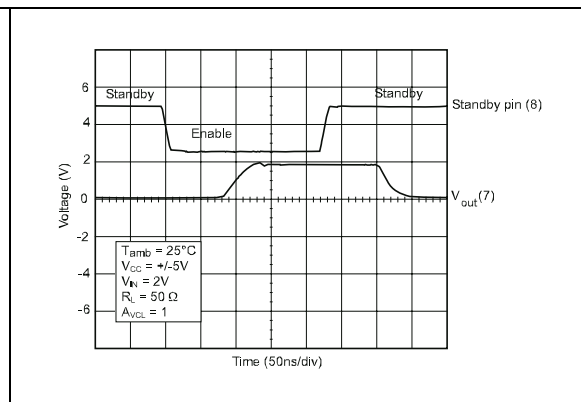


Figure 17. Signal multiplexing

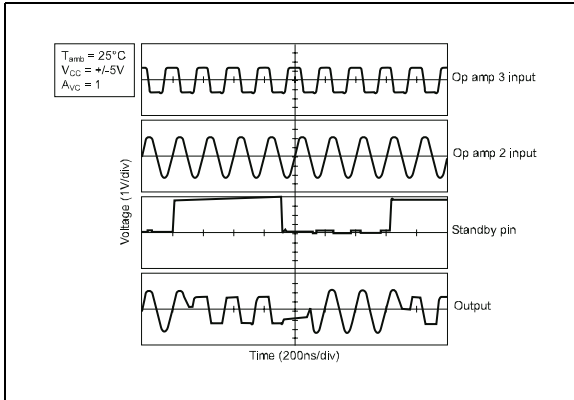


Figure 18. Differential input impedance versus frequency

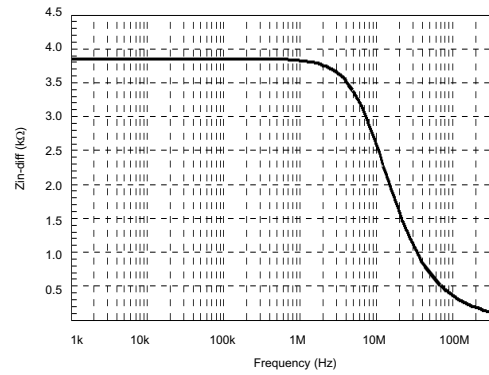
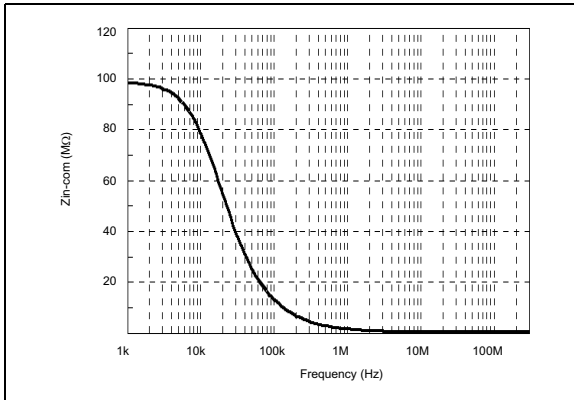


Figure 19. Common input impedance versus frequency



## 5 Macromodel information

The information below applies to the TSH95I.

\*\* Standard Linear Ics Macromodels, 1996.

\*\* CONNECTIONS :

- \* 1 INVERTING INPUT
- \* 2 NON-INVERTING INPUT
- \* 3 OUTPUT
- \* 4 POSITIVE POWER SUPPLY
- \* 5 NEGATIVE POWER SUPPLY
- \* 6 STANDBY

.SUBCKT TSH95 1 3 2 4 5 6 (analog)

\*\*\*\*\*

\*\*\*\*\* switch \*\*\*\*\*

.SUBCKT SWITCH 20 10 IN OUT COM

.MODEL DIDEAL D N=0.1 IS=1E-08

DP IN 1 DIDEAL 400E-12

DN OUT 2 DIDEAL 400E-12

EP 1 OUT COM 10 2

EN 2 IN COM 10 2

RFUIT1 IN 1 1E+09

RFUIT2 OUT 2 1E+09

RCOM COM 0 1E+12

.ENDS SWITCH

\*\*\*\*\* inverter \*\*\*\*\*

.SUBCKT INV 20 10 IN OUT

.MODEL DIDEAL D N=0.1 IS=1E-08

RP1 20 15 1E+09

RN1 15 10 1E+09

RIN IN 10 1E+12

RIP IN 20 1E+12

DPINV OUT 20 DIDEAL 400E-12

DNINV 10 OUT DIDEAL 400E-12

GINV 0 OUT IN 15 -6.7E-7

CINV 0 OUT 210f

```
.ENDS INV
***** AOP *****
.MODEL MDTH D IS=1E-8 KF=1.809064E-15
CJO=10F
* INPUT STAGE
CIP 2 5 1.000000E-12
CIN 1 5 1.000000E-12
EIP 10 5 2 5 1
EIN 16 5 1 5 1
RIP 10 11 2.600000E-01
RIN 15 16 2.600000E-01
RIS 11 15 3.645298E-01
DIP 11 12 MDTH 400E-12
DIN 15 14 MDTH 400E-12
VOFP 12 13 DC 0.000000E+00
VOFN 1314DC 0
FPOL 13 5 VSTB 1E+03
CPS 11 15 2.986990E-10
DINN 17 13 MDTH 400E-12
VIN 17 5 2.000000e+00
DINR 15 18 MDTH 400E-12
VIP 4 18 1.000000E+00
FCP 4 5 VOFP 3.500000E+00
FCN 5 4 VOFN 3.500000E+00
ISTB0 4 5 130UA
FIBP 2 5 VOFP 1.000000E-02
FIBN 5 1 VOFN 1.000000E-02
* AMPLIFYING STAGE
FIP 5 19 VOFP 2.530000E+02
FIN 5 19 VOFN 2.530000E+02
RG1 19 120 3.160721E+03
XCOM1 4 0 120 5 COM SWITCH
RG2 19 121 3.160721E+03
```

```
XCOM2 4 0 4 121 COM SWITCH
CC 19 5 2.000000E-09
DOPM 19 22 MDTH 400E-12
DONM 21 19 MDTH 400E-12
HOPM 22 28 VOUT 1.504000E+03
VIPM 28 4 5.000000E+01
HONM 21 27 VOUT 1.400000E+03
VINM 5 27 5.000000E+01
***** ZP *****
RZP1 5 80 1E+06
RZP2 4 80 1E+06
GZP 5 82 19 80 2.5E-05
RZP2H 83 4 10000
RZP1H 83 82 80000
RZP2B 84 5 10000
RZP1B 82 84 80000
LZPH 4 83 3.535e-02
LZPB 84 5 3.535e-02
*****
EOUT26 2382 51
VOUT 23 5 0
ROUT 26 103 35
COUT 103 5 30.000000E-12
XCOM 4 0 103 3 COM SWITCH
DOP 19 25 MDTH 400E-12
VOP 4 25 2.361965E+00
DON 24 19 MDTH 400E-12
VON 24 5 2.361965E+00
***** STAND BY *****
RMI1 4 111 1E+7
```

```

RMI2 0 111 2E+7
RONOFF 6 60 1K
CONOGG 60 0 10p
RSTBIN 60 0 1E+12
ESTBIN 106 0 6 0 1
ESTBREF 106 107 111 0 1
DSTB1 107 108 MDTH 400E-12
VSTB 108 109 0
ISTB 109 0 1U
RSTB 109 110 1
DSTB2 0 110 MDTH 400E-12
XINV 4 0 6 COM INV
.ENDS

```

**Table 6. Electrical characteristics with  $V_{CC} = \pm 5\text{ V}$ ,  $T_{amb} = 25^\circ\text{ C}$  (unless otherwise specified)**

Symbol	Conditions	Value	Unit
$V_{io}$		0	mV
$A_{vd}$	$R_L = 600\ \Omega$	3.2	V/mV
$I_{CC}$	No load/amplifier	5.2	mA
$V_{icm}$		-3 to 4	V
$V_{OH}$	$R_L = 600\ \Omega$	+3.6	V
$V_{OL}$	$R_L = 600\ \Omega$	-3.6	V
$I_{sink}$	$V_o = 0\text{ V}$	40	mA
$I_{source}$	$V_o = 0\text{ V}$	40	mA
GBP	$R_L = 600\ \Omega$ , $C_L = 15\text{ pF}$	147	MHz
SR	$R_L = 600\ \Omega$ , $C_L = 15\text{ pF}$	110	V/ $\mu\text{s}$
$\phi_m$	$R_L = 600\ \Omega$ , $C_L = 15\text{ pF}$	42	Degrees

## 6 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK<sup>®</sup> is an ST trademark.

## 6.1 SO-16 package information

Figure 20. SO-16 package mechanical drawing

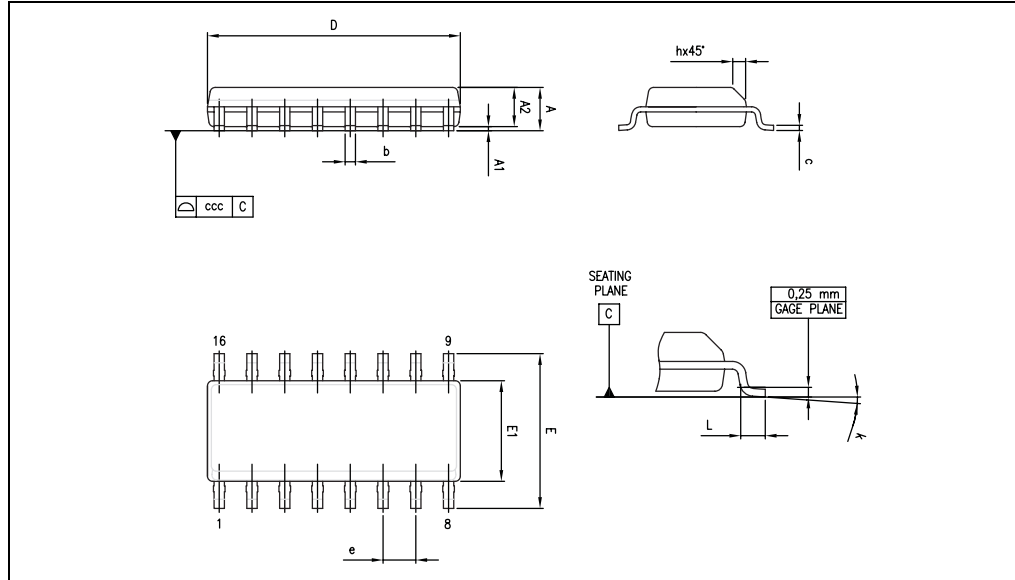


Table 7. SO-16 package mechanical data

Ref.	Dimensions					
	Millimeters			Inches		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A			1.75			0.069
A1	0.10		0.25	0.004		0.010
A2	1.25			0.049		
b	0.31		0.51	0.012		0.020
c	0.17		0.25	0.007		0.010
D <sup>(1)</sup>	9.80	9.90	10.00	0.386	0.390	0.394
E	5.80	6.00	6.20	0.228	0.236	0.244
E1 <sup>(2)</sup>	3.80	3.90	4.00	0.150	0.154	0.157
e		1.27			0.050	
h	0.25		0.50	0.010		0.020
L	0.40		1.27	0.016		0.050
k	0		8			
ccc			0.10			0.004

- Does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs not to exceed 0.15 mm in total.
- Does not include interlead flash or protrusions. Interlead flash or protrusions not to exceed 0.25 mm per side.



## 7 Ordering information

**Table 8. Order codes**

Part number	Temperature range	Package	Packing	Marking
TSH95ID	-40° C to +125° C	SO-16	Tube or Tape & reel	TSH95I
TSH95IDT				
TSH95IYD <sup>(1)</sup>		SO-16 (Automotive grade)		TSH95IY
TSH95IYDT <sup>(1)</sup>				

1. Qualification and characterization according to AEC Q100 and Q003 or equivalent, advanced screening according to AEC Q001 & Q 002 or equivalent are on-going.

## 8 Revision history

Table 9. Document revision history

Date	Revision	Changes
01-Nov-2000	1	Initial release.
27-Aug-2009	2	Document format updated. Updated SO-16 package information in <a href="#">Chapter 6</a> . Added automotive grade order codes in <a href="#">Table 8</a> .

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