TRANSFER-MOLD TYPE INSULATED TYPE

PS21267



INTEGRATED POWER FUNCTIONS

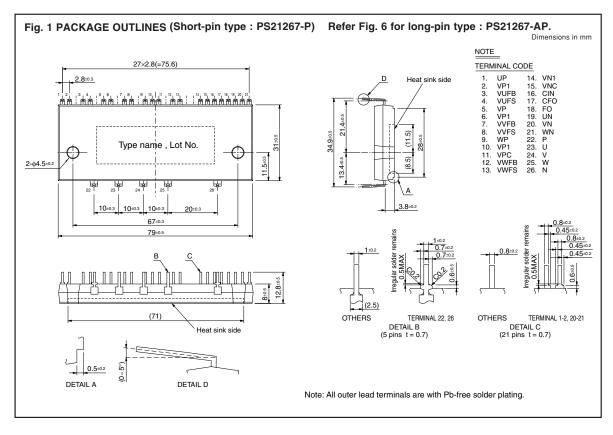
600V/30A low-loss CSTBTTM inverter bridge for three phase DC-to-AC power conversion

INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For upper-leg IGBTs :Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection.
- For lower-leg IGBTs: Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC).
- Fault signaling: Corresponding to an SC fault (Lower-side IGBT) or a UV fault (Lower-side supply).
- Input interface: 3, 5V line compatible. (High Active)
- UL Approved : Yellow Card No. E80276

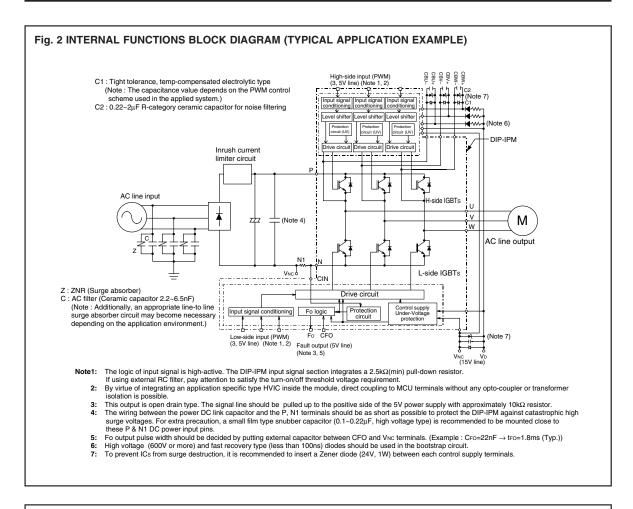
APPLICATION

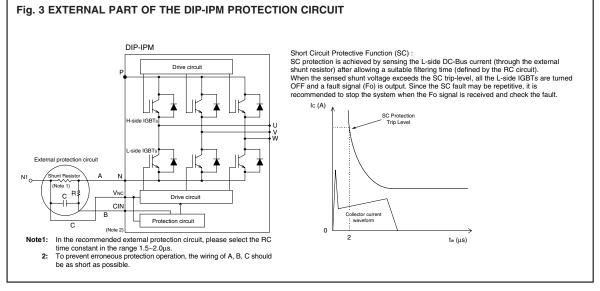
AC100V~200V three-phase inverter drive for small power motor control.





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$\textbf{MAXIMUM RATINGS} \ (T_{j} = 25^{\circ}\text{C}, \ unless \ otherwise \ noted)$

INVERTER PART

Symbol	Parameter	Condition	Ratings	Unit
Vcc	Supply voltage	Applied between P-N	450	V
VCC(surge)	ge) Supply voltage (surge) Applied between P-N		500	V
VCES	Collector-emitter voltage		600	V
±IC	Each IGBT collector current	Tc = 25°C	30	Α
±ICP	Each IGBT collector current (peak)	Tc = 25°C, less than 1ms	60	Α
Pc	Collector dissipation	Tc = 25°C, per 1 chip	55.5	W
Tj	Junction temperature	(Note 1)	-20~+125	°C

Note 1 : The maximum junction temperature rating of the power chips integrated within the DIP-IPM is 150° C (@ Tc $\leq 100^{\circ}$ C) however, to insure safe operation of the DIP-IPM, the average junction temperature should be limited to $T_{j(ave)} \leq 125^{\circ}$ C (@ Tc $\leq 100^{\circ}$ C).

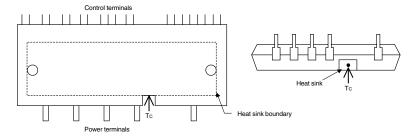
CONTROL (PROTECTION) PART

Symbol	Parameter	Condition	Ratings	Unit
VD	Control supply voltage	Applied between VP1-VPC, VN1-VNC	20	V
VDB	Control supply voltage	Applied between VUFB-VUFS, VVFB-VVFS, VWFB-VWFS	20	V
VIN	Input voltage	Applied between UP, VP, WP-VPC, UN, VN, WN-VNC	-0.5~VD+0.5	V
VFO	Fault output supply voltage	Applied between Fo-VNC	-0.5~VD+0.5	V
IFO	Fault output current	Sink current at Fo terminal	1	mA
Vsc	Current sensing input voltage	Applied between CIN-VNC	-0.5~VD+0.5	V

TOTAL SYSTEM

Symbol	Parameter	Condition	Ratings	Unit
VCC(PROT)	Self protection supply voltage limit (short circuit protection capability)	$VD = 13.5 \sim 16.5 V$, Inverter part $T_j = 125 °C$, non-repetitive, less than 2 μs	400	٧
Tc	Module case operation temperature	(Note 2)	− 20~+100	°C
Tstg	Storage temperature		− 40~+125	°C
Viso	Isolation voltage	60Hz, Sinusoidal, AC 1 minute, connecting pins to heat-sink plate	2500	Vrms

Note 2 : To measurement point





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THERMAL RESISTANCE

Symbol Parameter	Davasastav	O a madiki a m	Limits			Unit
	Condition		Тур.	Max.		
Rth(j-c)Q	Junction to case thermal	Inverter IGBT part (per 1/6 module)	_	_	1.80	°C/W
Rth(j-c)F	resistance (Note 3)	Inverter FWDi part (per 1/6 module)	_	_	3.00	°C/W
Rth(c-f)F	Contact thermal resistance	Case to fin (per 1 module) thermal grease applied	_	_	0.067	°C/W

Note 3 : Grease with good thermal conductivity should be applied evenly with a thickness of about +100μm~+200μm on the contact surface of DIP-IPM and heat-sink.

ELECTRICAL CHARACTERISTICS (Tj = 25°C, unless otherwise noted) **INVERTER PART**

Cymbol		O a madistica re		Limits			11.2	
Symbol	Parameter		Condition		Тур.	Max.	Unit	
V05(4)	Collector-emitter saturation	VD = VDB = 15V	Ic = 30A, Tj = 25°C	_	1.50	2.00		
VCE(sat) voltage	voltage	VIN = 5V	Ic = 30A, Tj = 125°C	_	1.50	2.00	V	
VEC	FWDi forward voltage	Tj = 25°C, -IC = 30A, VIN = 0V		_	1.50	2.00	V	
ton		hing times $ \begin{aligned} &\text{Vcc} = 300\text{V}, \text{Vd} = \text{VdB} = 15\text{V} \\ &\text{Ic} = 30\text{A}, \text{Tj} = 125^{\circ}\text{C}, \text{Vin} = 0 \Leftrightarrow 5\text{V} \\ &\text{Inductive load (upper-lower arm)} \end{aligned} $		0.65	1.25	1.85	μS	
trr				_	0.30	_	μS	
tc(on)	Switching times			_	0.30	0.50	μS	
toff				_	1.70	2.40	μS	
tc(off)				_	0.40	0.70	μS	
ICES	Collector-emitter cut-off		Tj = 25°C	_	_	1	mA	
current		VCE = VCES	Tj = 125°C	_		10	IIIA	

CONTROL (PROTECTION) PART

Symbol	Parameter	Condition			Limits			
Symbol	Farameter			Min.	Тур.	Max.	Unit	
		VD = VDB = 15V	VD = VDB = 15V Total of VP1-VPC, VN1-VNC		_	_	7.00	mA
l _D	Circuit current	VIN = 5V	VUFB-	VUFS, VVFB-VVFS, VWFB-VWFS	_	_	0.55	mA
ם ו	Circuit current	VD = VDB = 15V	Total o	of VP1-VPC, VN1-VNC	_	_	7.00	mA
		VIN = 0V	VUFB-	VUFS, VVFB-VVFS, VWFB-VWFS	_	_	0.55	mA
VFOH	Fault output voltage	Vsc = 0V, Fo circuit pull-up to 5V with $10k\Omega$			4.9	_	_	V
VFOL	Fault output voltage	VSC = 1V, IFO = 1mA		_	_	0.95	V	
VSC(ref)	Short circuit trip level	$TC = -20 \sim 100 ^{\circ}C$, $VD = 15V$ (Note 4)		0.45	_	0.52	V	
lin	Input current	VIN = 5V		1.0	1.5	2.0	mA	
UVDBt				Trip level	10.0	_	12.0	V
UVDBr	Control supply under-voltage	Ti≤ 125°C		Reset level	10.5	_	12.5	V
UVDt	protection	1] ≤ 125 C		Trip level	10.3	_	12.5	V
UVDr				Reset level	10.8	_	13.0	V
tFO	Fault output pulse width	CFO = 22nF (Note 5)		1.0	1.8	_	ms	
Vth(on)	ON threshold voltage	Applied between LIP, VP, WP, VPC, LIN, VN, WN, VNC		2.1	2.3	2.6	V	
Vth(off)	OFF threshold voltage	Applied between UP, VP, WP-VPC, UN, VN, WN-VNC			0.8	1.4	2.1	V

Note 4: Short circuit protection is functioning only at the low-arms. Please select the external shunt resistance such that the SC trip-level is



less than 2.0 times of the collector current rating.

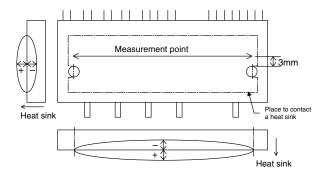
5: Fault signal is output when the low-arms short circuit or control supply under-voltage protective functions operate. The fault output pulse-width tFO depends on the capacitance value of CFO according to the following approximate equation: CFO = 12.2 × 10⁻⁶ × tFO [F].

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MECHANICAL CHARACTERISTICS AND RATINGS

Dave-market.	0	Condition			Limits		
Parameter					Max.	Unit	
Mounting torque	Mounting screw : M4	Recommended : 1.18 N·m	0.98	_	1.47	N·m	
Weight			_	54	_	g	
Heat-sink flatness (Note 6)		-50	_	100	μm		

Note 6 :



RECOMMENDED OPERATION CONDITIONS

Cymbol Doromotor		Condition		Recommended value			Unit
Symbol	Parameter	Condition	1	Min.	Тур.	Max.	Uriit
Vcc	Supply voltage	Applied between P-N		0	300	400	V
VD	Control supply voltage	Applied between VP1-VPC, VN1-	Vnc	13.5	15.0	16.5	V
VDB	Control supply voltage	Applied between VUFB-VUFS, VV	/FB-VVFS, VWFB-VWFS	13.0	15.0	18.5	V
ΔV D, ΔV DB	Control supply variation			-1	_	1	V/μs
tdead	Arm shoot-through blocking time	For each input signal, Tc ≤ 100°	C	2	_	_	μS
fPWM	PWM input frequency	Tc ≤ 100°C, Tj ≤ 125°C		_	_	20	kHz
	IO Allowable r.m.s. current	VCC = 300V, VD = VDB = 15V,	fPWM = 5kHz	_	_	19.0	
lo		P.F = 0.8, sinusoidal PWM TC \leq 100°C, T _i \leq 125°C (Note 7)	fPWM = 15kHz	_	_	11.6	Arms
PWIN(on)			(Note 8)	0.3	_	_	
		200 ≤ VCC ≤ 350V, 13.5 ≤ VD ≤ 16.5V,	Below rated current	1.5	_	_	
PWIN(off)	-2 N-	13.0 ≤ VDB ≤ 18.5V, -20°C ≤ TC ≤ 100°C,	Between rated current and 1.7 times of rated current	3.0	_	_	μS
		N-line wiring inductance less than 10nH (Note 9)	Between 1.7 times and 2.0 times of rated current	3.6	_	_	
VNC	VNC variation	between VNC-N (including surge)	-5.0	_	5.0	V



Note 7: The Allowable r.m.s. current value depends on the actual application conditions.

8: Input signal with ON pulse width less than PWIN(on) might make no response.

9: IPM might make no response or response delay to next turn-on pulse if off-pulse width is less than PWIN(off). (Please refer to Fig. 4) Please refer to Fig. 9 for recommended wiring method too.

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Fig. 4 CURRENT OUTPUT WHEN INPUT SIGNAL IS LESS THAN ALLOWABLE MINIMUM INPUT PULSE WIDTH PWIN(off) (P-side only)

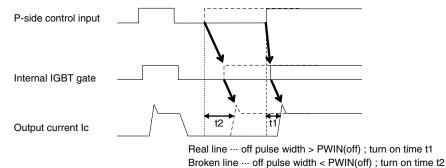
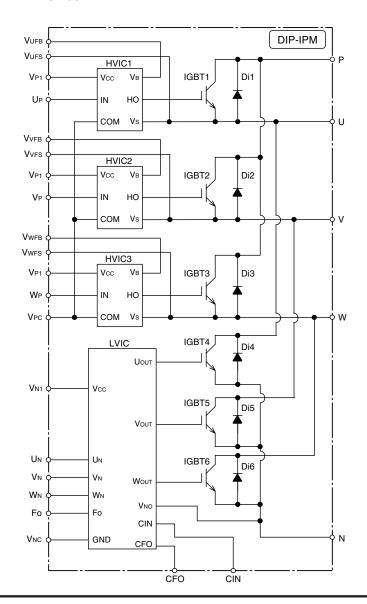


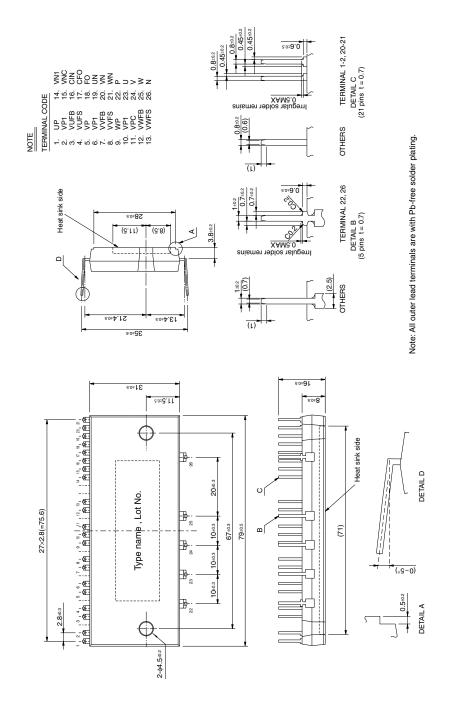
Fig. 5 THE DIP-IPM INTERNAL CIRCUIT





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Fig. 6 PACKAGE OUTLINES (Long-pin type: PS21267-AP)



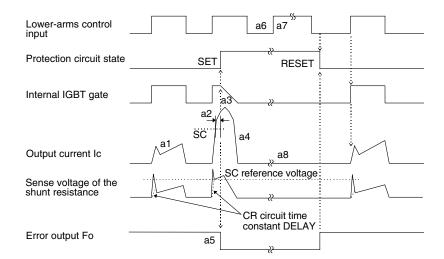


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Fig. 7 TIMING CHARTS OF THE DIP-IPM PROTECTIVE FUNCTIONS

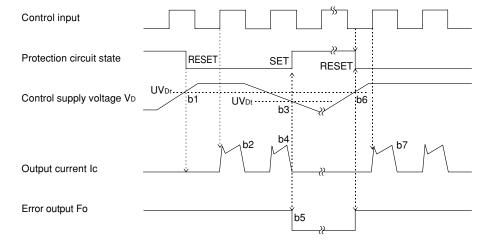
[A] Short-Circuit Protection (Lower-arms only) (with external shunt resistor and CR connection)

- a1. Normal operation: IGBT ON and carrying current.
- a2. Short circuit current detection (SC trigger).
- a3. Hard IGBT gate interrupt.
- a4. IGBT turns OFF.
- a5. Fo timer operation starts: The pulse width of the Fo signal is set by the external capacitor CFo.
- a6. Input "L": IGBT OFF state.
- a7. Input "H": IGBT ON state, but during the Fo signal active period the IGBT doesn't turn ON.
- a8. IGBT OFF in spite of "H" input.



[B] Under-Voltage Protection (Lower-arm, UVD)

- b1. Control supply voltage rises: After the voltage reaches UVDr level, the circuits start to operate when the next input is applied.
- b2. Normal operation : IGBT ON and carrying current.
- b3. Under voltage trip (UVDt).
- b4. IGBT OFF in spite of control input condition.
- b5. Fo operation starts. The minimum pulse width of Fo is set by the external capacitor CFo, and Fo outputs continuously during UV period.
- b6. Under voltage reset (UVDr).
- b7. Normal operation: IGBT ON and carrying current.





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[C] Under-Voltage Protection (Upper-arm, UVDB)

- c1. Control supply voltage rises : Operation starts soon after UVDBr. c2. Normal operation : IGBT ON and carrying current.

- c3. Under voltage trip (UVDBt).
 c4. IGBT OFF in spite of control input condition, but there is no Fo signal output.
- c5. Under voltage reset (UVDBr).
- c6. Normal operation: IGBT ON and carrying current.

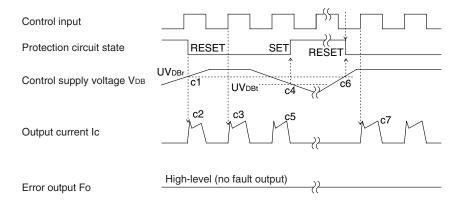
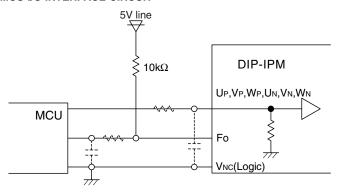
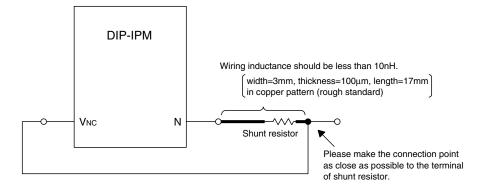


Fig. 8 RECOMMENDED MCU I/O INTERFACE CIRCUIT



Note: RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The DIP-IPM input signal section integrates a 2.5kΩ(min) pull-down resistor. Therefore, if using external RC filter, pay attention to satisfy the turn-on/off threshold voltage requirement.

Fig. 9 RECOMMENDED WIRING OF SHUNT RESISTOR





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Fig. 10 EXAMPLE OF TYPICAL DIP-IPM APPLICATION CIRCUIT

C1:Tight tolerance temp-compensated electrolytic type C2,C3: $0.22^{\sim}2\mu F$ R-category ceramic capacitor for noise filtering. (Note: The capacitance value depends on the PWM control used in the applied system.) C2 VufB DIP-IPM Vufs HVIC1 L. V_{P1} **V**cc ۷в __C3 UР HC Vs СОМ Vvfb V Vvrs HVIC2 لها ۷в ___C3 М сом ۷s VWFB C1 Vwes HVIC3 L V_{P1} Vcc VR ONTROLLER ___C3 Wp HC сом ۷s LVIC 垃 Uou V_{N1} Vcc ==C3 5V line Un Un ٧N ٧n If this wiring is too long, short circuit might be W_N W٨ caused Fo CIN GND CFO С CIN CFO G4(CFO) В T_{C5} Resisto N1 The long wiring of GND might generate If this wiring is too long, the SC level noise on input and cause IGBT to fluctuation might be larger and cause be malfunction. SC malfunction.

Note 1: To prevent the input signals oscillation, the wiring of each input should be as short as possible. (Less than 2-3cm)

- 2: By virtue of integrating an application specific type HVIC inside the module, direct coupling to MCU terminals without any opto-coupler or transformer isolation is possible.
- 3: Fo output is open drain type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 10kΩ resistor.
- 4: Fo output pulse width is determined by the external capacitor between CFO and VNc terminals (CFo). (Example : CFo = 22nF → tFO = 1.8ms (typ.))
- 5: The logic of input signal is high-active. The DIP-IPM input signal section integrates a 2.5kΩ (min) pull-down resistor. If using external RC filter, pay attention to satisfy the turn-on/off threshold voltage requirement.
- 6: To prevent malfunction of protection, the wiring of A, B, C should be as short as possible.
- 7: Please set the R1C5 time constant in the range 1.5~2µs.
- 8: Each capacitor should be located as nearby the pins of the DIP-IPM as possible.
- 9: To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 pins should be as short as possible. Approximately a 0.1~0.22μF snubber capacitor between the P-N1 pins is recommended.
- 10: To prevent ICs from surge destruction, it is recommended to insert a Zener diode (24V, 1W) between each control supply terminals.

