ICs for Communications

PPP and HDLC Synchronous Serial Controller with 2 Channels PASSAT

PEB 20525 Version 1.1
PEF 20525 Version 1.1

Preliminary Data Sheet 09.99

DS 2

PEB 20525, PEF 20525 **Revision History:**		**Current Version: 09.99**
Previous Version:		07.99
Page (in previous Version)	Page (in current Version)	Subjects (major changes since last revision)
2-24, 2-25, 2-30	2-23, 2-24, 2-29	Corrected $\overline{\text{RESET}}$ polarity to active low
3-42, 5-188	3-41,5-193	revised description of transmit data underrun (XDU) handling
5-108	5-109	bit field coding of "GMODE:IPC(1:0)" changed (equal to ESCC2)
5-108	5-109	bit field "GMODE:DMODE(1:0)" reduced to "GMODE:EDMA"
5-108	5-110	bits "SHAPERPD" and "BYPASS" reduced to bit "DSHP"
-	-	HDLC "Non-Automode" renamed to "Address Mode 2"
5-123	5-128	Corrected reset value of register STARH to 10_H
-	6-215	Added description of external DMA support

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Preface

The PASSAT is a Protocol Controller for a wide range of data communication and telecommunication applications. This document provides complete reference information on hardware and software related issues as well as on general operation.

Organization of this Document

This Preliminary Data Sheet is divided into 9 chapters. It is organized as follows:

- Chapter 1, Introduction
  Gives a general description of the product, lists the key features, and presents some typical applications.

- Chapter 2, Pin Description
  Lists pin locations with associated signals, categorizes signals according to function, and describes signals.

- Chapters 3 Functional Description
  These chapters provide detailed descriptions of all PASSAT internal function blocks.

- Chapter 4, Detailed Protocol Descriptions
  Gives a detailed description of all protocols supported by the serial communication controllers SCCs.

- Chapter 5, Detailed Register Description
  Gives a detailed description of all PASSAT on chip registers.

- Chapter 6, Programming
  Provides programming help for PASSAT initialization procedure and operation.

- Chapter 7, Electrical Characteristics
  Gives a detailed description of all electrical DC and AC characteristics and provides timing diagrams and values for all interfaces.

- Chapter 8, Test Modes
  Gives a detailed description of the JTAG boundary scan unit.

- Chapter 9, Package Outline

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1 Introduction

The PASSAT is a Serial Communication Controller with two independent serial channels[1]. The serial channels are derived from updated protocol logic of the ESCC and DSCC4 device family providing a large set of protocol support and variety in serial interface configuration. This allows easy integration to different environments and applications.

A generic 8- or 16-bit multiplexed/demultiplexed slave interface provides fast device access with low bus utilization and easy software handshaking (in the P-LFBGA-80-2 package only an 8-bit data bus is provided). DMA handshake control signals allow connection to an external DMA controller.

Large on-chip FIFOs of 64 byte capacity per port and direction in combination with enhanced threshold control mechanisms allow decoupling of traffic requirements on host bus and serial interfaces with little exception probabilities such as data underuns or overflows.

Each of the two Serial Communication Controllers (SCC) contains an independent Baud Rate Generator, DPLL and programmable protocol processing (HDLC, PPP, ASYNC and BISYNC). Data rates of up to 2 Mbit/s (DPLL assisted modes) and 12.5 Mbit/s (HDLC, PPP, bit transparent) are supported. The channels can also handle a large set of layer-2 protocol functions (LAPD, SS7) reducing bus and host CPU load. Two channel specific timers are provided to support protocol functions.

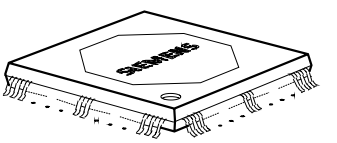
[1] The serial channels are also called ’ports’ or ’cores’ depending on the context.

PPP and HDLC Synchronous Serial Controller with 2 Channels
PASSAT

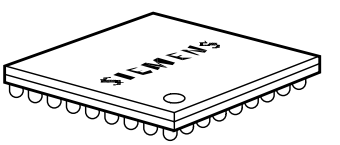
PEB 20525
PEF 20525

Version 1.1 **CMOS**

1.1 Features



P-TQFP-100-3



P-LFBGA-80-2

Serial communication controllers (SCCs)

- Two independent channels
- Full duplex data rates on each channel of up to 12.5 Mbit/s sync - 2 Mbit/s with DPLL
- 64 Bytes deep receive FIFO per SCC
- 64 Bytes deep transmit FIFO per SCC

Serial Interface

- On-chip clock generation or external clock sources
- On-chip DPLLs for clock recovery
- Baud rate generator
- Clock gating signals
- Clock gapping capability
- Programmable time-slot capability for connection to TDM interfaces (e.g. T1, E1)
- NRZ, NRZI, FM and Manchester data encoding
- Optional data flow control using modem control lines ($\overline{RTS}$, $\overline{CTS}$, CD)
- Support of bus configuration by collision detection and resolution

Bit Processor Functions

- HDLC/SDLC Protocol Modes
  - Automatic flag detection and transmission
  - Shared opening and closing flag
  - Generation of interframe-time fill ’1’s or flags
  - Detection of receive line status
  - Zero bit insertion and deletion

Type	Package
PEB 20525, PEF 20525	P-TQFP-100-3
	P-LFBGA-80-2

  - CRC generation and checking (CRC-CCITT or CRC-32)
  - Transparent CRC option per channel and/or per frame
  - Programmable Preamble (8 bit) with selectable repetition rate
  - Error detection (abort, long frame, CRC error, short frames)
- Bit Synchronous PPP Mode
  - Bit oriented transmission of HDLC frame (flag, data, CRC, flag)
  - Zero bit insertion/deletion
  - 15 consecutive ’1’ bits abort sequence
- Octet Synchronous PPP Mode
  - Octet oriented transmission of HDLC frame (flag, data, CRC, flag)
  - Programmable character map of 32 hard-wired characters (00_H-$1F_H$)
  - Four programmable characters for additional mapping
  - Insertion/deletion of control-escape character ($7D_H$) for mapped characters
- Extended Transparent Mode
  - Fully bit transparent (no framing, no bit manipulation)
  - Octet-aligned transmission and reception
- Protocol and Mode Independent
  - Data bit inversion
  - Data overflow and underrun detection
  - Timer

Protocol Support

- Address Recognition Modes
  - No address recognition (Address Mode 0)
  - 8-bit (high byte) address recognition (Address Mode 1)
  - 8-bit (low byte) or 16-bit (high and low byte) address recognition (Address Mode 2)
- HDLC Automode
  - 8-bit or 16-bit address generation/recognition
  - Support of LAPB/LAPD
  - Automatic handling of S- and I-frames
  - Automatic processing of control byte(s)
  - Modulo-8 or modulo-128 operation
  - Programmable time-out and retry conditions
  - SDLC Normal Response Mode (NRM) operation for slave
- Signaling System #7 (SS7) support
  - Detection of FISUs, MSUs and LSSUs
  - Unchanged Fill-In Signaling Units (FISUs) optionally not forwarded
  - Automatic generation of FISUs in transmit direction (incl. sequence number)
  - Counting of errored signaling units
- Optional $\overline{\text{DTACK}}$/$\overline{\text{READY}}$ controlled cycles

Microprocessor Interface

- 8-bit bus interface (P-LFBGA-80-2 package)
- 8/16-bit bus interface (P-TQFP-100-3 package)
- Multiplexed and De-multiplexed address/data bus
- Intel/Motorola style
- Asynchronous interface
- Maskable interrupts for each channel

General Purpose Port (GPP) Pins (up to 3 in P-LFBGA-80-2, up to 7 in P-TQFP-100-3 package)

General

- 3.3V power supply with 5V tolerant inputs
- Low power consumption
- Power safe features
- P-TQFP-100-3 Package (Thermal Resistance: R_{JA} = 42 K/W)
- Small P-LFBGA-80-2 Package (Thermal Resistance: R_{JA} = 51 K/W)

1.2 Logic Symbol

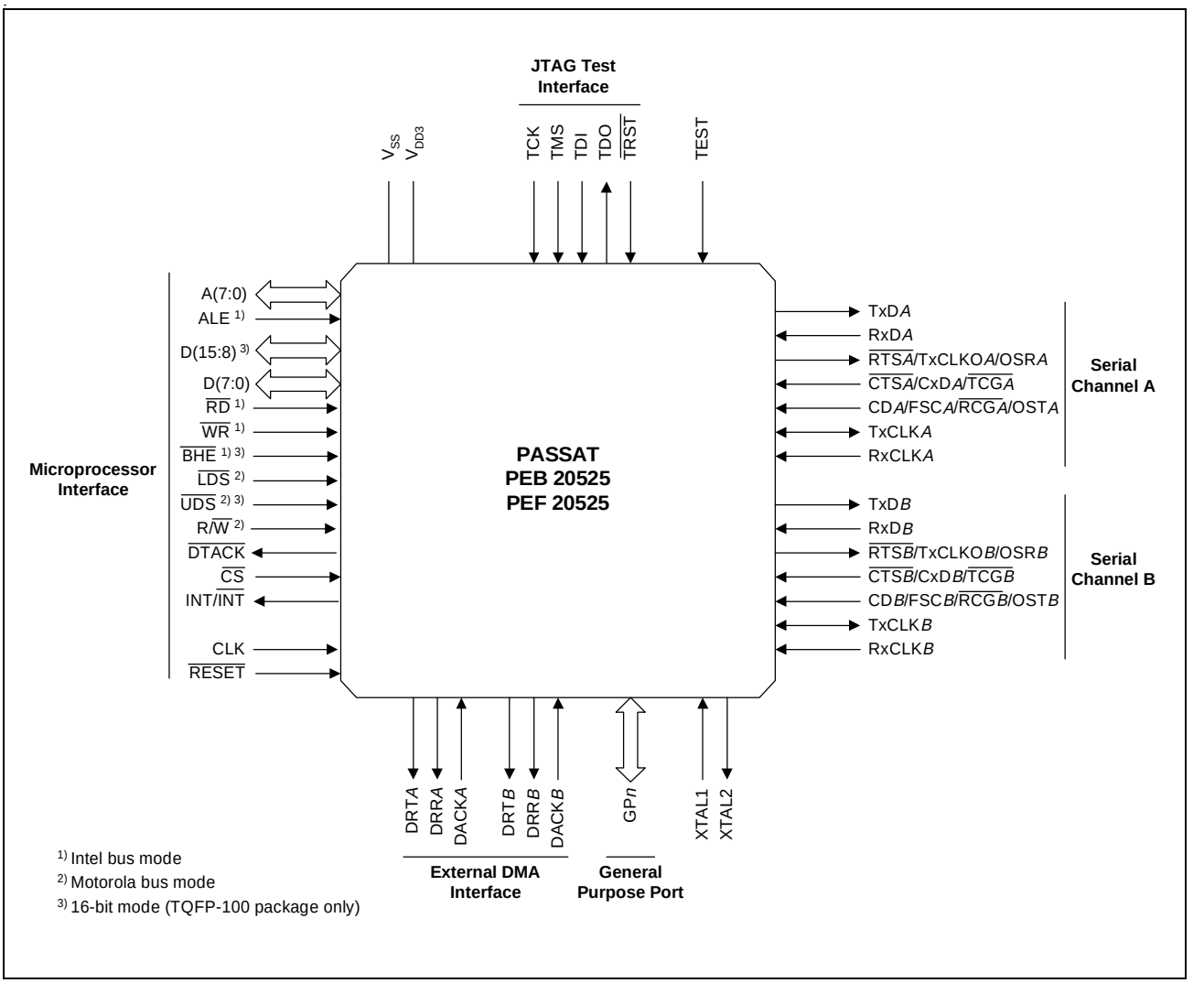


Figure 1-1 Logic Symbol

1.3 Typical Applications

PASSAT devices can be used in LAN-WAN inter-networking applications such as Routers, Switches and Trunk cards and support the common V.35, ISDN BRI (S/T) and RFC1662 standards. Its new features provide powerful hardware and software interfaces to develop high performance systems.

1.3.1 System Integration Example

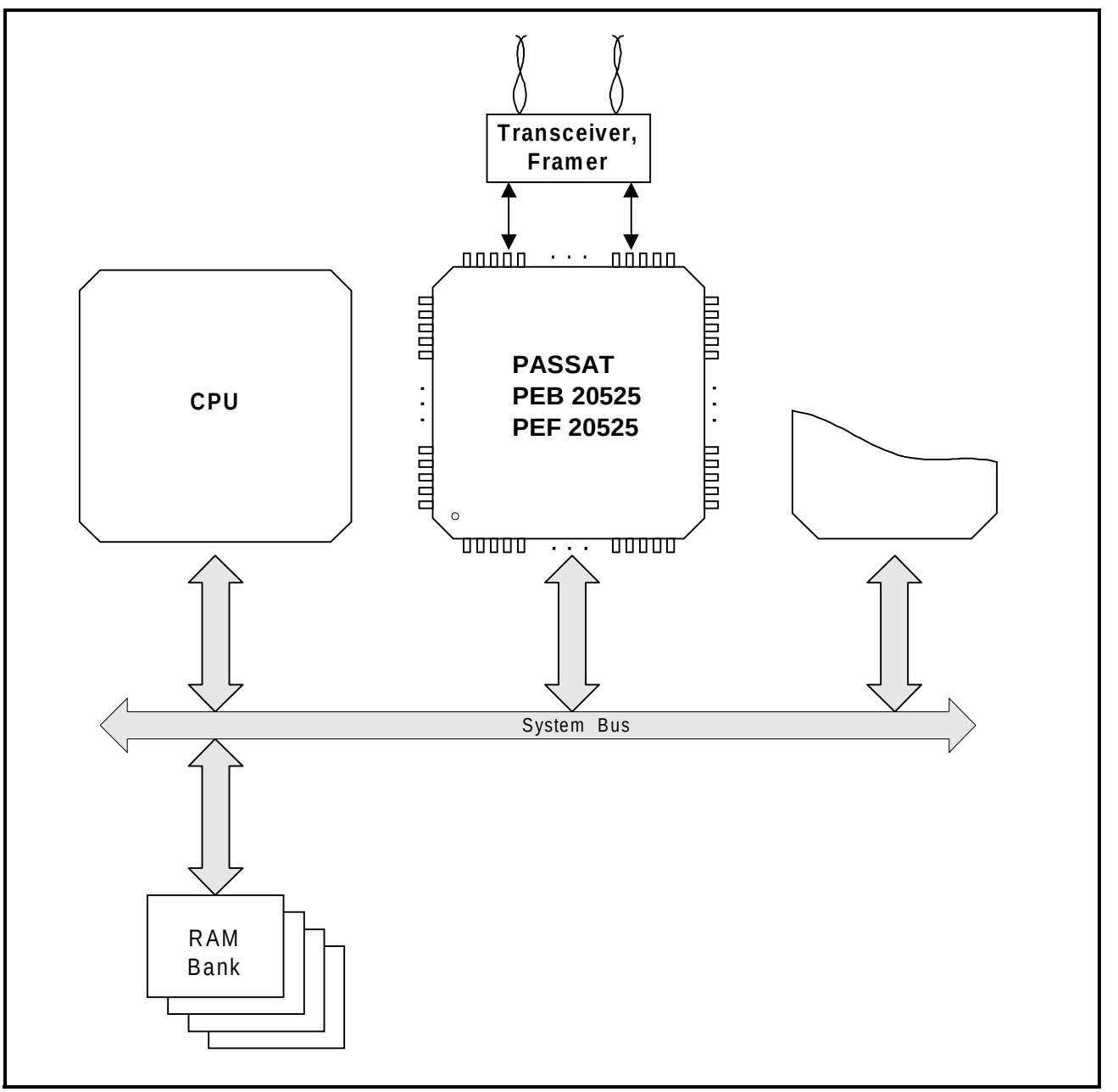


Figure 1-2 System Integration

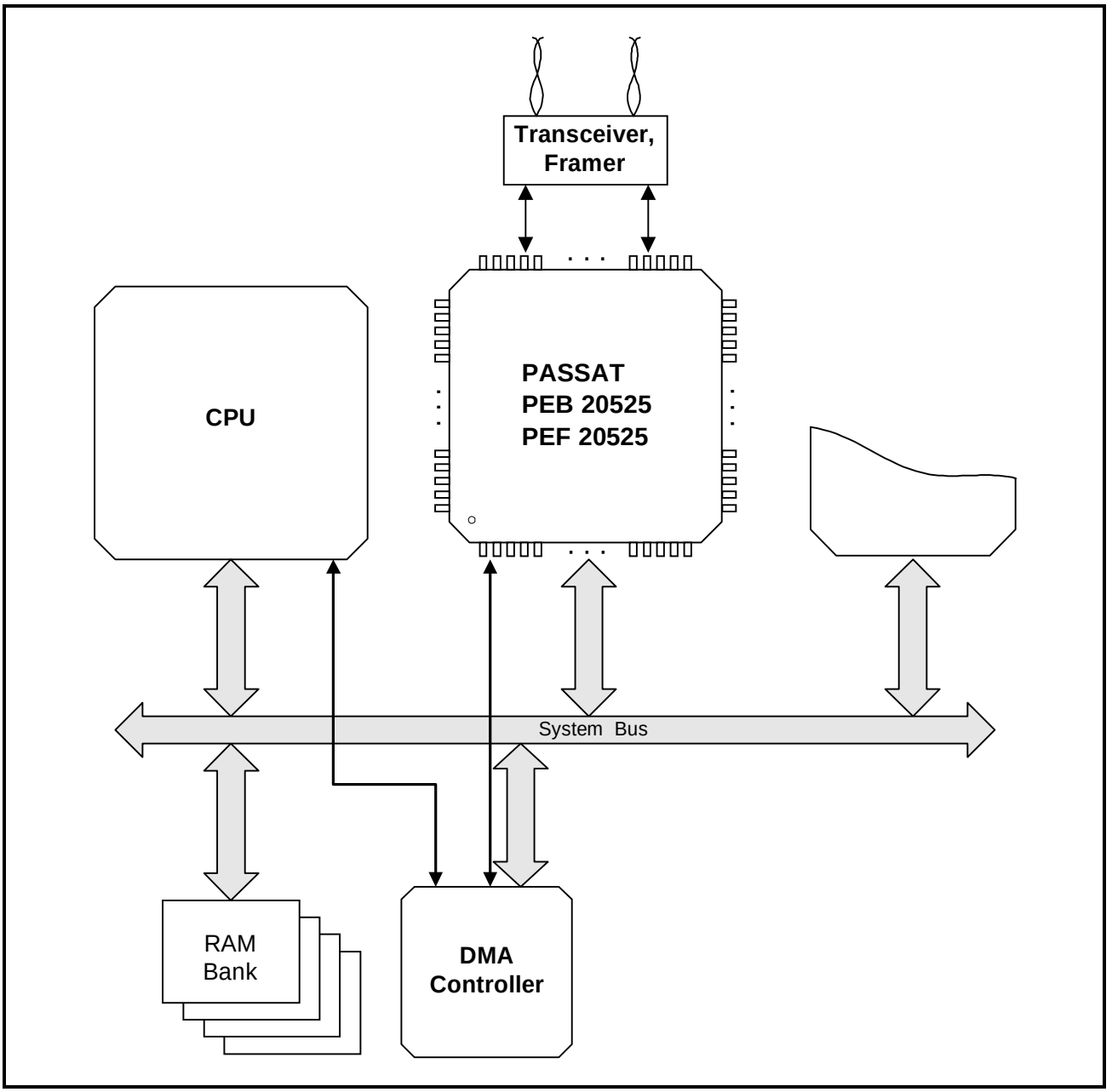


Figure 1-3 System Integration With External DMA Controller

1.3.2 Serial Configuration Examples

PASSAT supports a variety of serial configurations at Layer-1 and Layer-2 level. The outstanding variety of clock modes supporting a large number of combinations of external and internal clock sources allows easy integration in application environments.

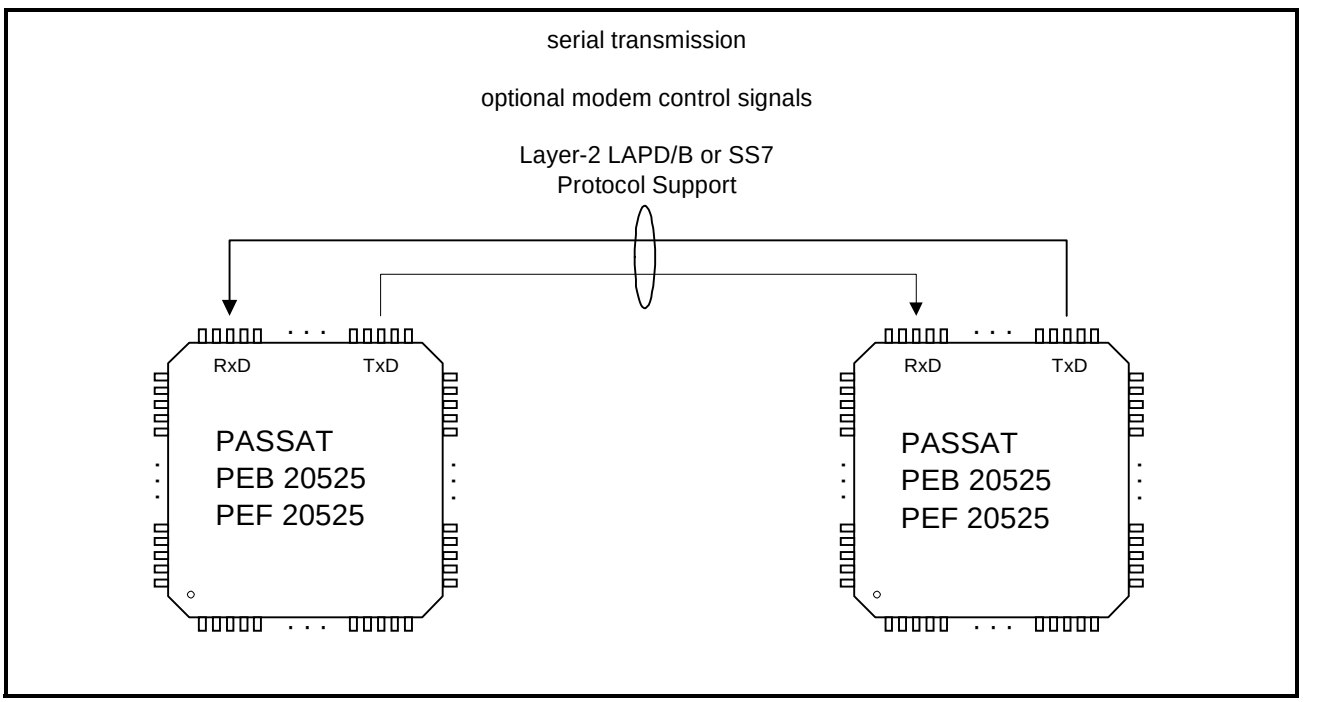


Figure 1-4 Point-to-Point Configuration

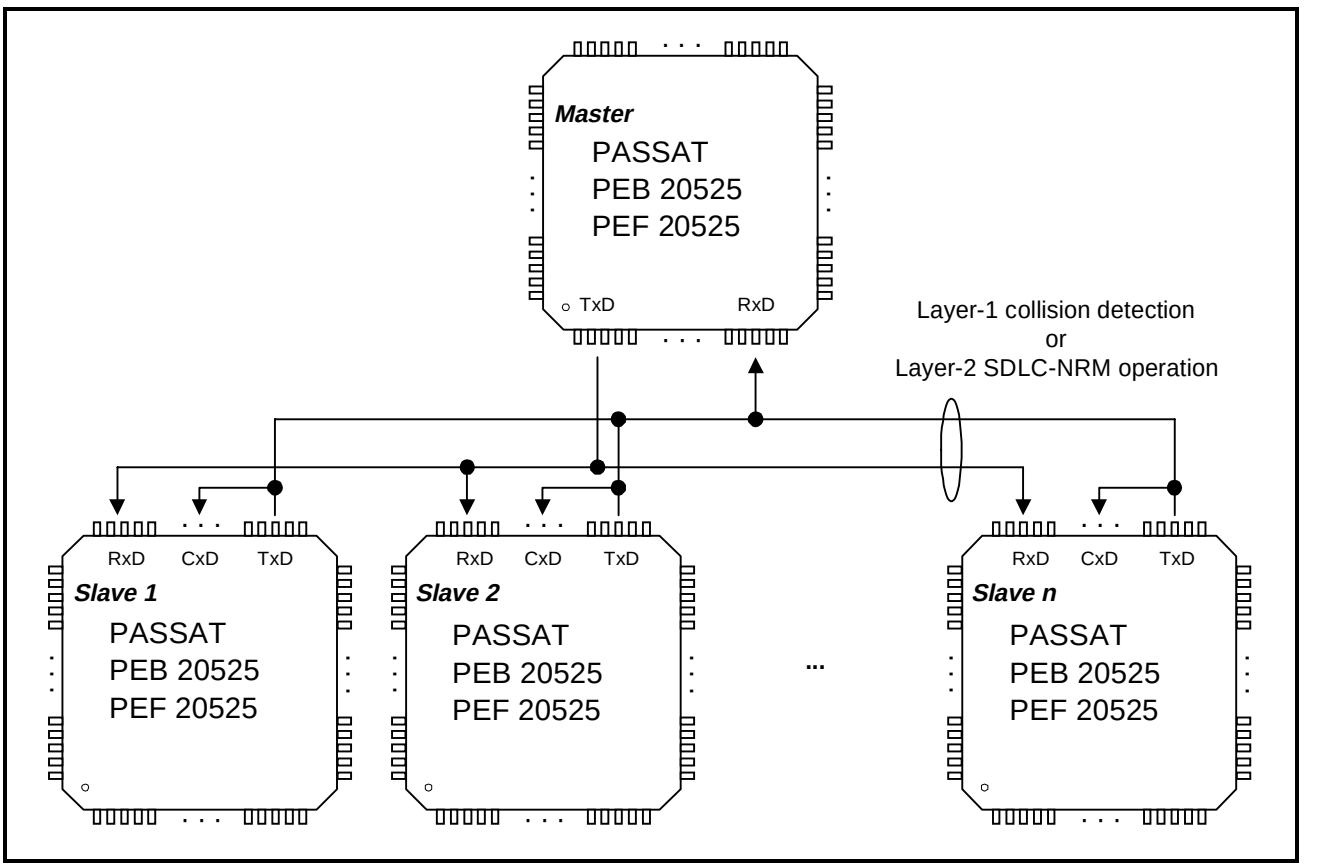


Figure 1-5 Point-to-Multipoint Bus Configuration

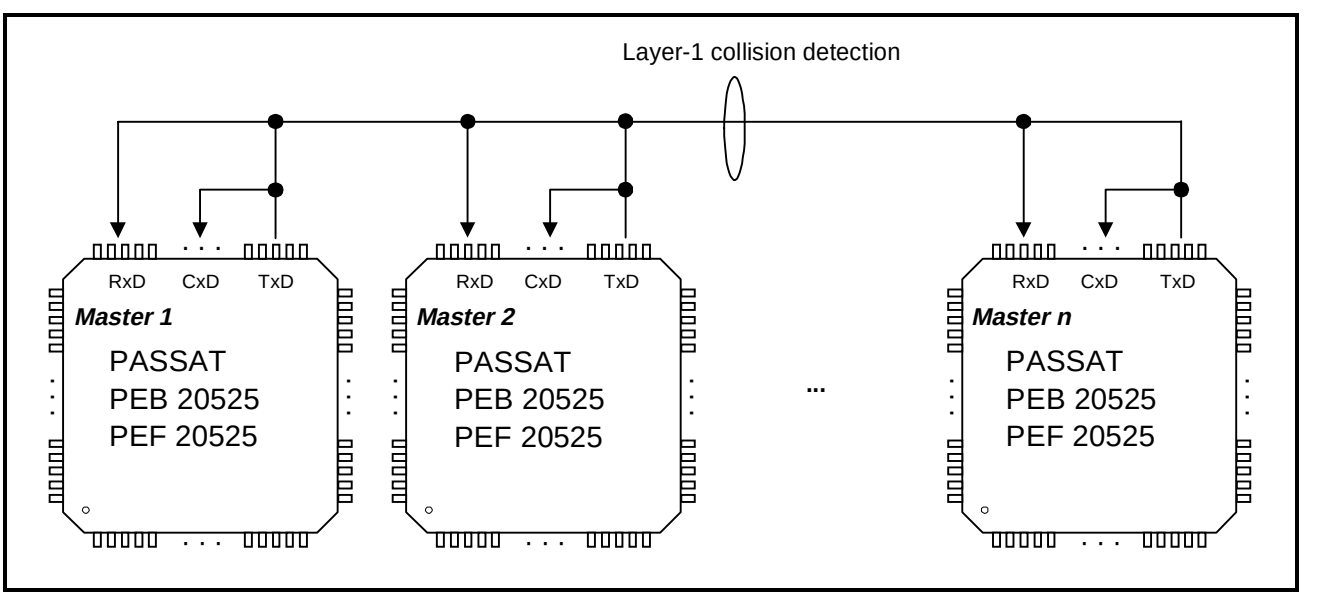


Figure 1-6 Multimaster Bus Configuration

1.4 Differences between PASSAT and the HSCX/ESCC Family

This chapter is useful for all being familiar with the HSCX/ESCC family.

1.4.1 Enhancements to the HSCX Serial Core

The PASSAT SCC cores contain the core logic of the HSCX as the heart of the device. Some enhancements are incorporated in the SCCs. These are:

- Octet-and Bit Synchronous PPP protocol support as in RFC-1662
- Signaling System #7 (SS7) support
- 4-kByte packet length byte counter
- Enhanced address filtering (16-bit maskable)
- Enhanced time slot assigner
- Support of high data rates (12.5 Mbit/s)

1.4.2 Simplifications to the HSCX Serial Core

The following features of the HSCX core have been removed:

- Extended transparent mode 0
  (this mode provided octet buffered data reception without usage of FIFOs; PASSAT supports octet buffered reception via appropriate threshold configurations for the SCC receive FIFOs)

2 Pin Descriptions

2.1 Pin Diagram P-LFBGA-80-2

(top view)

P-LFBGA-80-2

RD#	VSS	R/W#	BM/ ALE	A0/ BLE#/ UDS#	A3	A5	A7	VDD	J
READY# DTACK#	TMS	CS#	VDD	A2	A4	VSS	RESET#	INT/ INT#	H
VDD	WR#	CLK	DS#/ BHE#/ LDS#	A1	VDD	A6	VSS	RTSA#	G
D2	VSS	D1	VSS	VSS	TxDA	TxCLKA	RxCLKA	VDD	F
VSS	D3	VDD	D0		XTAL2	CDA/ FSCA/ RCGA#/ OSTA	CTSA#/ CxDA/ TCGA#/ OSRA	RxDA	E
D6	D4	TEST1	D5	DRTB/ GP0	TxDB	XTAL1	VSSA	VDDA	D
TEST2	D7	DRRB/ GP1	DACKB# GP2	VDD	VDD	VSS	TCK	CTSB#/ CxDB/ TCGB#/ OSRB	C
VSS	VDD	DACKA#	VSS	RxDB	RxCLKB	CDB/ FSCB/ RCGB#/ OSTB	TRST#	VDD	B
DRTA	DRRA	VDD	RTSB#	VSS	TxCLKB	VSS	TDI	TDO	A
9	8	7	6	5	4	3	2	1	

Figure 2-1 Pin Configuration P-LFBGA-80-2 Package

2.2 Pin Diagram P-TQFP-100-3

(top view)

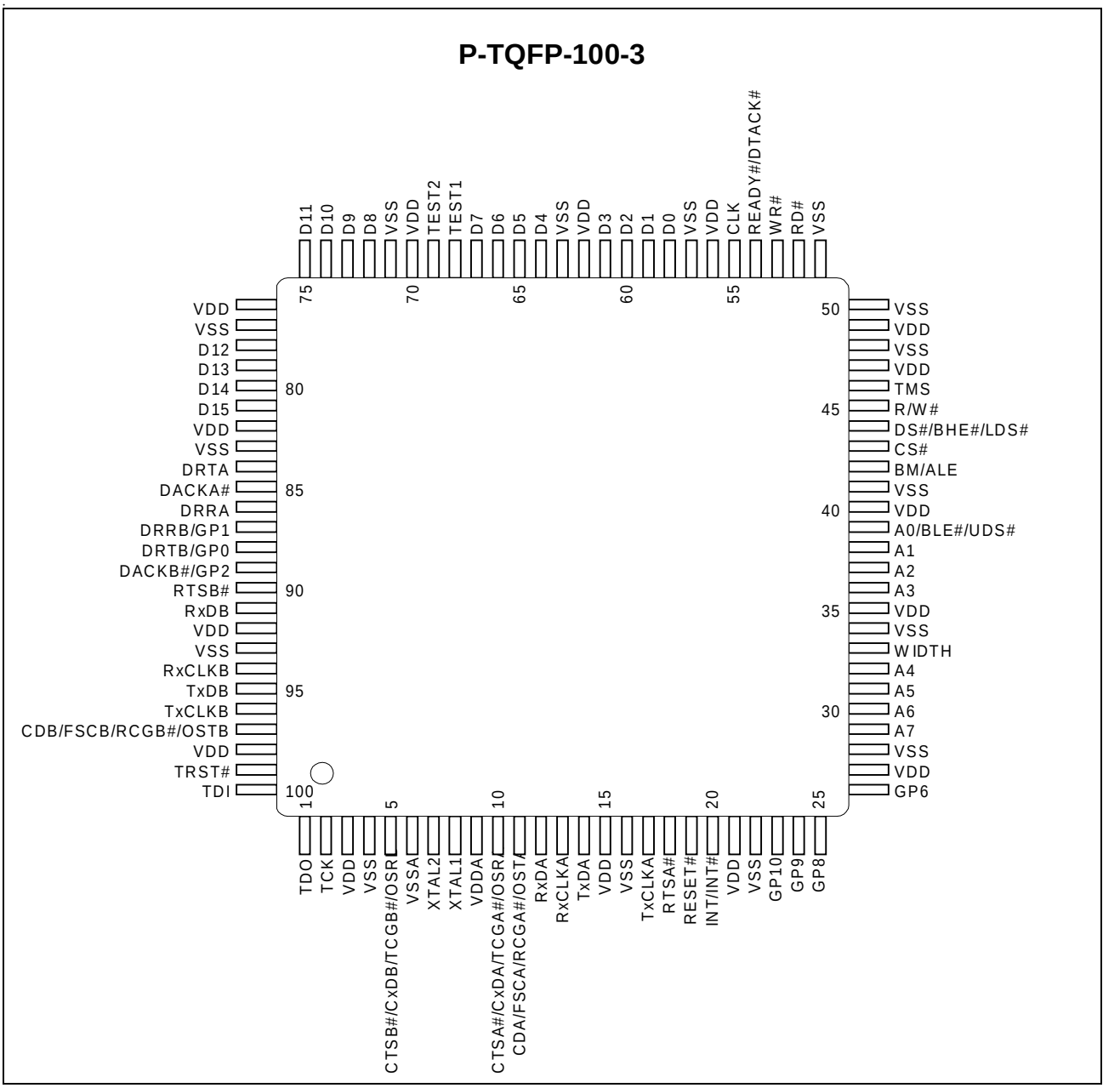


Figure 2-2 Pin Configuration P-TQFP-100-3 Package

2.3 Pin Definitions and Functions

Table 2-1 Microprocessor Bus Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
-	81	D15	I/O	**Data Bus** The data bus lines are bi-directional tri-state lines which interface with the system's data bus. *The PASSAT in the P-LFBGA-80-2 package does not support 16-bit bus modes.*
-	80	D14		
-	79	D13		
-	78	D12		
-	75	D11		
-	74	D10		
-	73	D9		
-	72	D8		
C8	67	D7		
D9	66	D6		
D6	65	D5		
D8	64	D4		
E8	61	D3		
F9	60	D2		
F7	59	D1		
E6	58	D0		
J2	29	A7	I	**Address Bus** These pins connect to the system's address bus to select one of the internal registers for read or write.
G3	30	A6	I	
J3	31	A5	I	
H4	32	A4	I	
J4	36	A3	I	
H5	37	A2	I	
G5	38	A1	I	

Table 2-1 Microprocessor Bus Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
J5	39	A0	I	**Address Line A0 (8-bit modes)** In Motorola and in Intel 8-bit mode this signal represents the least significant address line.
		$\overline{\text{BLE}}$	I	**Byte Low Enable (16-bit Intel bus mode)** This signal indicates a data transfer on the lower byte of the data bus (D7..D0). Together with signal $\overline{\text{BHE}}$ the type of bus access is determined (byte or word access at even or odd address).
		$\overline{\text{UDS}}$	I	**Upper Data Strobe (16-bit Motorola bus mode)** This active low strobe signal serves to control read/write operations. Together with signal $\overline{\text{LDS}}$ the type of bus access is determined.
J6	42	BM	I	**Bus Mode** – BM = static ’1’ for operation in Motorola bus mode (de-multiplexed). – BM = static ’0’ for operation in Intel bus mode with de-multiplexed address and data buses. – Pin BM/ALE has the function of an Address Latch Enable (ALE) for operation in Intel bus mode with a multiplexed address/data bus. A falling edge on this pin selects Intel multiplexed bus mode.
		ALE	I	**Address Latch Enable (mux’ed Intel bus)** The address is latched by the PASSAT with the falling edge of ALE. The address input pins A(7:0) must be externally connected to the data bus pins D(7:0). *For operation of the 8-bit PASSAT (P-LFBGA-80-2 package) in a 16-bit environment, A(7:0) should be connected to address/data lines AD(8:1) of the external bus. D(7:0) interface to AD(7:0) of the external bus.*

Table 2-1 Microprocessor Bus Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
G6	44	$\overline{DS}$	I	**Data Strobe (8-bit Motorola bus mode only)** This active low strobe signal serves to control read/write operations.
		$\overline{BHE}$	I	**Bus High Enable (16-bit Intel bus mode only)** This signal indicates a data transfer on the upper byte of the data bus (D15..D8). In 8-bit Intel bus mode this signal has no function.
		$\overline{LDS}$	I	**Lower Data Strobe (16-bit Motorola bus mode)** This active low strobe signal serves to control read/write operations. Together with signal $\overline{UDS}$ the type of bus access is determined (byte or word access at even or odd address). *In 8-bit Intel bus mode, a pull-up resistor to V_{DD3} is recommended on this pin.*
J9	52	$\overline{RD}$	I	**Read Strobe (Intel bus mode only)** This signal indicates a read operation. The bus is able to accept data on lines D(7:0) / D(15:0) during an active $\overline{RD}$ signal. *In Motorola bus mode, a pull-up resistor to V_{DD3} is recommended on this pin.*
J7	45	R/$\overline{W}$	I	**Read/Write Enable (Motorola bus mode)** This signal distinguishes between read and write operation. As an input it must be valid during data strobe ($\overline{DS}$). *In Intel bus mode, a pull-up resistor to V_{DD3} is recommended on this pin.*
H7	43	$\overline{CS}$	I	**Chip Select** A low signal selects PASSAT for read/write operations.

Table 2-1 Microprocessor Bus Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
G8	53	$\overline{WR}$	I	**Write Strobe (Intel bus mode only)** This signal indicates a write operation. Valid data is present data on lines D(7:0) / D(15:0) during an active $\overline{WR}$ signal. *In Motorola bus mode, a pull-up resistor to V_{DD3} is recommended on this pin.*
-	33	WIDTH	I	**Width Of Bus Interface** A low signal on this input selects the 8-bit bus interface mode. A high signal on this input selects the 16-bit bus interface mode. In this case word transfer to/from the internal registers is enabled. Byte transfers are implemented by using $\overline{BLE}$ and $\overline{BHE}$ (Intel bus mode) or $\overline{LDS}$ and $\overline{UDS}$ (Motorola bus mode) *In P-LFBGA-80-2 package this signal is not available, since only 8 bit bus width is supported.*
G7	55	CLK	I	**Clock** The system clock for PASSAT is provided through this pin.
H1	20	INT/$\overline{INT}$	O o/d	**Interrupt Request** The INT/$\overline{INT}$ goes active when one or more of the bits in registers ISR0..ISR2 are set to ’1’. A read to these registers clears the interrupt. The INT/$\overline{INT}$ line is inactive when all interrupt status bits are reset. Interrupt sources can be unmasked in registers IMR0..IMR2 by setting the corresponding bits to ’0’.

Table 2-1 Microprocessor Bus Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
H9	54	$\overline{\text{READY}}$ $\overline{\text{DTACK}}$	O O	**Ready (Intel bus mode)** **Data Transfer Acknowledge (Motorola mode)** During a slave access (register read/write) this signal (output) indicates, that the PASSAT is ready for data transfer. The signal remains active until the data strobe ($\overline{\text{DS}}$ in Motorola bus mode, $\overline{\text{RD}}$/$\overline{\text{WR}}$ in Intel bus mode) and/or the chip select ($\overline{\text{CS}}$) go inactive. *This line is tri-state when unused.* *A pull-up resistor to V_{DD3} is recommended if this function is not used.*
H2	19	$\overline{\text{RESET}}$	I	**Reset** With this active low signal the on-chip registers and state machines are forced to reset state. During Reset all pins are in a high impedance state.

Table 2-2 External DMA Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
A9	84	DRTA	O	**DMA Request Transmitter Channel A** The transmitter on a this channel requests a DMA transfer by activating the DRTA line. The request remains active as long as the Transmit FIFO requires data transfers. The amount of data bytes to be transferred from the system memory to the serial channel (= Byte Count) must be written first to the XBCL, XBCH registers. Always blocks of data (n x 32 bytes + rest ; n=0,1,…) are transferred till the Byte Count is reached. DRTA is deactivated with the beginning of the last write cycle.
A8	86	DRRA	O	**DMA Request Receiver Channel A** The receiver on this serial channel requests a DMA transfer by activating the DRRA line. The request remains active as long as the Receive FIFO requires data transfers, thus always blocks of data are transferred. DRRA is deactivated immediately following the falling edge of the last read cycle.
B7	85	$\overline{\text{DACKA}}$	I	**DMA Acknowledge Channel A** A low signal on this pin informs the PASSAT that the requested DMA cycle controlled via DRTA or DRRA of this channel is in progress, i.e. the DMA controller has achieved bus mastership from the CPU and will start data transfer cycles (either write or read). In conjunction with a read or write operation this input serves as Access Enable (similar to $\overline{\text{CS}}$) to the respective FIFOs. If $\overline{\text{DACKA}}$ is active, the input to pins A(7:0) and $\overline{\text{CS}}$ is ignored and the FIFOs are implicitly selected. *If not used, a pull-up resistor to V_{DD} is required for this pin.*

Table 2-2 External DMA Interface

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
D5	88	DRTB	O	**DMA Request Transmitter Channel B** (corresponding to channel A)
		GP0	I/O	**General Purpose Pin #0** If DMA support is not enabled, this pin serves as a general pupose input/output pin. *After reset this pin serves as a general purpose input. A pull-up resistor to V_{DD3} is recommended.*
C7	87	DRRB	O	**DMA Request Receiver Channel B** (corresponding to channel A)
		GP1	I/O	**General Purpose Pin #1** If DMA support is not enabled, this pin serves as a general pupose input/output pin. *After reset this pin serves as a general purpose input. A pull-up resistor to V_{DD3} is recommended.*
C6	89	$\overline{\text{DACKB}}$	I	**DMA Acknowledge Channel B** (corresponding to channel A)
		GP2	I/O	**General Purpose Pin #2** If DMA support is not enabled, this pin serves as a general pupose input/output pin. *A pull-up resistor to V_{DD3} is recommended if this pin is not used.*

Table 2-3 Serial Port Pins

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
F3	17	TxCLKA	I/O	**Transmit Clock Channel A** The function of this pin depends on the selected clock mode and the value of bit 'TOE' (CCR0L register, refer to Table 3-2 "Clock Modes of the SCCs" on page 3-45). **If programmed as Input (CCR0L.TOE='0')**, either – the transmit clock for the channel (clock mode 0a, 2a, 4, 5b, 6a), or – a transmit strobe signal for the channel (clock mode 1) can be provided to this pin. **If programmed as Output (CCR0L.TOE='1')**, this pin supplies either – the transmit clock from the baud rate generator (clock mode 0b, 2b, 3b, 6b, 7b), or – the transmit clock from the DPLL circuit (clock mode 3a, 7a), or – an active-low control signal marking the programmed transmit time-slot in clock mode 5a.
F2	13	RxCLKA	I	**Receive Clock Channel A** The function of this pin depends on the selected clock mode (refer to Table 3-2 "Clock Modes of the SCCs" on page 3-45). A signal provided on pin RxCLKA may supply – the receive clock (clock mode 0, 4, 5b), or – the receive and transmit clock (clock mode 1, 5a), or – the clock input for the baud rate generator (clock mode 2, 3).

Table 2-3 Serial Port Pins (cont'd)

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
E3	11	CDA	I	**Carrier Detect Channel A** The function of this pin depends on the selected clock mode. It can supply – either a modem control or a general purpose input (clock modes 0, 2, 3, 6, 7). If auto-start is programmed, it functions as a receiver enable signal. – or a receive strobe signal (clock mode 1). Polarity of CDA can be set to 'active low' with bit ICD in register CCR1H. Additionally, an interrupt may be issued if a state transition occurs at the CDA pin (programmable feature).
		FSCA	I	**Frame Sync Clock Channel A (cm 5a)** When the SCC is in the time-slot oriented clock mode 5a, this pin functions as the Frame Synchronization Clock input.
		$\overline{\text{RCGA}}$	I	**Receive Clock Gating Channel A (cm 4)** In clock mode 4 this pin is used as Receive Clock Gating signal. *If no clock gating function is required, a pull-up resistor to V_{DD3} is recommended.*
		OSTA	I	**Octet Sync Transmit Channel A (cm 5b)** When the SCC is in the time-slot oriented clock mode with octet-alignment (clock mode 5b), a synchronization pulse on this input pin aligns transmit octets.

Table 2-3 Serial Port Pins (cont'd)

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
G1	18	$\overline{\text{RTSA}}$	O	**Request to Send Channel A** The function of this pin depends on the settings of bits RTS, FRTS in register CCR1H . In bus configuration, $\overline{\text{RTS}}$ can be programmed to: – go low during the actual transmission of a frame shifted by one clock period, excluding collision bits. – go low during reception of a data frame. – stay always high ($\overline{\text{RTS}}$ disabled).
E2	10	$\overline{\text{CTSA}}$	I	**Clear to Send Channel A** A low on the $\overline{\text{CTSA}}$ input enables the transmitter. Additionally, an interrupt may be issued if a state transition occurs at the $\overline{\text{CTSA}}$ pin (programmable feature). *If no 'Clear To Send' function is required, a pull-down resistor to V_{SS} is recommended.*
		CxDA	I	**Collision Data Channel A** In a bus configuration, the external serial bus must be connected to the corresponding CxDA pin for collision detection. A collision is detected whenever a logical '1' is driven on the open drain TxDA output but a logical '0' is detected via CxDA input.
		$\overline{\text{TCGA}}$	I	**Transmit Clock Gating Channel A (cm 4)** In clock mode 4 these pins are used as Transmit Clock Gating signals. *If no clock gating function is required, a pull-up resistor to V_{DD3} is recommended.*
		OSRA	I	**Octet Sync Receive Channel A (cm 5b)** (clock mode 5b) When the SCC is in the time-slot oriented clock mode with octet-alignment (clock mode 5b), received octets are aligned to this synchronization pulse input.

Table 2-3 Serial Port Pins (cont'd)

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
F4	14	TxDA	O o/d	**Transmit Data Channel A** Transmit data is shifted out via this pin. It can be configured as push/pull or open drain output characteristic via bit 'ODS' in register CCR1L.
E1	12	RxDA	I	**Receive Data Channel A** Serial data is received on this pin.
A4	96	TxCLKB	I/O	**Transmit Clock Channel B** (corresponding to channel A)
B4	94	RxCLKB	I	**Receive Clock Channel B** (corresponding to channel A)
B3	97	CDB FSCB $\overline{\text{RCGB}}$ OSTB	I I I I	**Carrier Detect Channel B** **Frame Sync Clock Channel B (cm 5a)** **Receive Clock Gating Channel B (cm 4)** **Octet Sync Transmit Channel B (cm 5b)** (corresponding to channel A)
A6	90	$\overline{\text{RTSB}}$	O	**Request to Send Channel B** (corresponding to channel A)
C1	5	$\overline{\text{CTSB}}$ CxDB $\overline{\text{TCGB}}$ OSRB	I I I I	**Clear to Send Channel B** **Collision Data Channel B** **Transmit Clock Gating Channel B (cm 4)** **Octet Sync Receive Channel B (cm 5b)** (corresponding to channel A)
D4	95	TxDB	O o/d	**Transmit Data Channel B** (corresponding to channel A)

Table 2-3 Serial Port Pins (cont'd)

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
B5	91	RxDB	I	**Receive Data Channel B** (corresponding to channel A)
D3 E4	8 7	XTAL1 XTAL2	I O	**Crystal Connection** If the internal oscillator is used for clock generation (clock modes 0b, 6, 7) the external crystal has to be connected to these pins. The internal oscillator should be powered up (GMODE:OSCPD = '0') and the signal shaper may be activated (GMODE:DSHP = '0'). Moreover, XTAL1 may be used as input for a common clock source to both SCCs, provided by an external clock generator (oscillator). In this case the oscillator unit may be powered down and it is recommended to bypass the shaper of the internal oscillator unit by setting bit 'DSHP' to '1'. *A pull-down resistor to V_{SS} is recommended for pin XTAL1 if not used.*

Table 2-4 General Purpose Pins

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
-	23 24 25 26	GP10 GP9 GP8 GP6	I/O	**General Purpose Pins** These pins serve as general purpose input/output pins.

Table 2-5 Test Interface Pins

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
B2	99	$\overline{\text{TRST}}$	I	**JTAG Reset Pin (internal pull-up)** For proper device operation, a reset for the boundary scan controller must be supplied to this active low pin. *If the boundary scan of the PASSAT is not used, this pin can be connected to V_{SS} to keep it in reset state.*
C2	2	TCK	I	**JTAG Test Clock (internal pull-up)** *If the boundary scan of the PASSAT is not used, this pin may remain unconnected.*
A2	100	TDI	I	**JTAG Test Data Input (internal pull-up)** *If the boundary scan of the PASSAT is not used, this pin may remain unconnected.*
A1	1	TDO	O	**JTAG Test Data Output**
H8	46	TMS	I	**JTAG Test Mode Select (internal pull-up)** *If the boundary scan of the PASSAT is not used, this pin may remain unconnected.*
D7	68	TEST1	I	**Test Input 1** When connected to V_{DD3} the PASSAT works in a vendor specific test mode. *This pin must be connected to V_{SS}.*
C9	69	TEST2	I	**Test Input 2** When connected to V_{DD3} the PASSAT works in a vendor specific test mode. *This pin must be connected to V_{SS}.*

Table 2-6 Power Pins

Pin No.		Symbol	In (I) Out (O)	Function
P-LFBGA-80-2	P-TQFP-100-3			
A7, B1, B8, C4, C5, E7, F1, G4, G9, H6, J1	3, 15, 21, 27, 35, 40, 47, 49, 56, 62, 70, 76, 82, 92, 98	V_{DD3}	-	**Digital Supply Voltage** 3.3 V ± 0.3 V All pins must be connected to the same voltage potential.
A3, A5, B6, B9, C3, E9, F5, F6, F8, G2, H3, J8	4, 16, 22, 28, 34, 41, 48, 50, 51, 57, 63, 71, 77, 83, 93	V_{SS}	-	**Digital Ground** (0 V) All pins must be connected to the same voltage potential.
D1	9	V_{DDA}	-	**Analog Supply Voltage** 3.3 V ± 0.3 V This pin supplies the on-chip oscillator of the PASSAT. It can be directly connected to V_{DD3}.
D2	6	V_{SSA}	-	**Analog Ground** (0 V) This pin supplies the ground level to the on-chip oscillator of the PASSAT. It can be directly connected to V_{SS}.
-	--	N.C.	-	**Not Connected**

3 Functional Overview

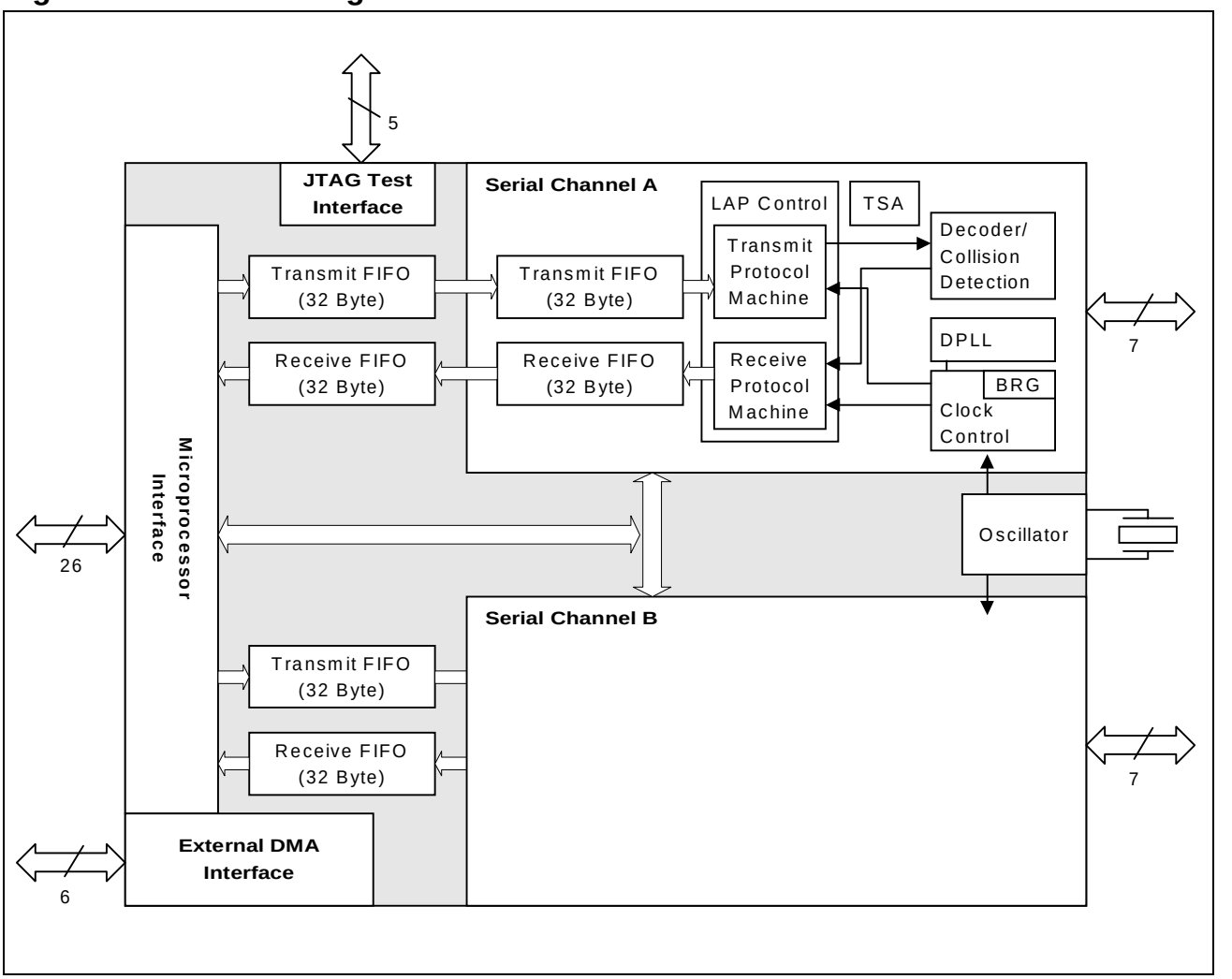
The functional blocks of PASSAT can be divided into two major domains:

– the microprocessor interface of PASSAT provides access to internal on-chip and to the system portion of the receive and transmit FIFOs (RFIFO/XFIFO). Optionally these FIFOs can be accessed by an external 4-channel DMA controller.
– the Serial Communication Controller (SCC) is capable of processing bit-synchronous (HDLC/SDLC/bitsync PPP) and octet-synchronous (octet-sync PPP) as well as fully transparent data traffic.

Data exchange between the serial communication controller and the microprocessor interface is performed using FIFOs, decoupling these two clocking domains.

3.1 Block Diagram

Figure 3-1 Block Diagram



3.2 Serial Communication Controller (SCC)

3.2.1 Protocol Modes Overview

The SCC is a multi-protocol communication controller. The core logic provides different protocol modes which are listed below:

- HDLC Modes
  - HDLC Transparent Operation (Address Mode 0)
  - HDLC Address Recognition (Address Mode 1, Address Mode 2 8/16-bit)
  - Full-Duplex LAPB/LAPD Operation (Automode 8/16-bit)
  - Half-Duplex SDLC-NRM Operation (Automode 8-bit)
  - Signaling System #7 (SS7) Operation

- Point-to-Point Protocol (PPP) Modes
  - Bit Synchronous PPP
  - Octet Synchronous PPP

- Extended Transparent Mode

A detailed description of these protocol modes is given in **Chapter 4**, starting on page 4-80.

3.2.2 SCC FIFOs

Each SCC provides its own transmit and receive FIFOs to handle internal arbitration and microcontroller latencies.

3.2.2.1 SCC Transmit FIFO

The SCC transmit FIFO is divided into two parts of 32 bytes each (’transmit pools’). The interface between the two parts provides clock synchronization between the system clock domain and the protocol logic working with the serial transmit clock.

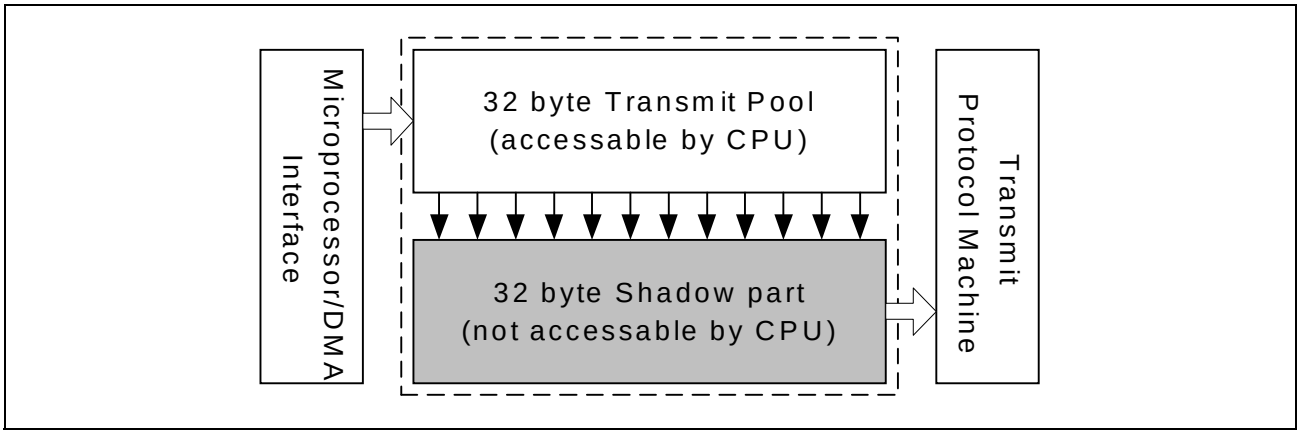


Figure 3-2 SCC Transmit FIFO

The 32 bytes system clocked FIFO part is accessable by the CPU/DMA controller; it accepts transmit data even if the SCC is in power-down condition (register CCR0H bit PU='0').

The only exception is a transmit data underrun (XDU) event. In case of an XDU event (e.g. after excessive bus latency), the FIFO will neither accept more data nor transfer another byte to the protocol logic. This XDU blocking mechanism prevents unexpected serial data. The blocking condition must be cleared by reading the interrupt status register ISR1 after the XDU interrupt was generated. Thus, the XDU interrupt indication should not be masked in register IMR1.

Transfer of data to the 32 byte shadow part only takes place if the SCC is in power-up condition and an appropriate transmit clock is provided depending on the selected clock mode.

Serial data transmission will start as soon as at least one byte is transferred into the shadow FIFO and transmission is enabled depending on the selected clock mode ($\overline{\text{CTS}}$ signal active, clock strobe signal active, timeslot valid or clock gapping signal inactive).

3.2.2.2 SCC Receive FIFO

The SCC receive FIFO is divided into two parts of 32 bytes each. The interface between the two parts provides clock synchronization between the system clock domain and the protocol logic working with the serial receive clock.

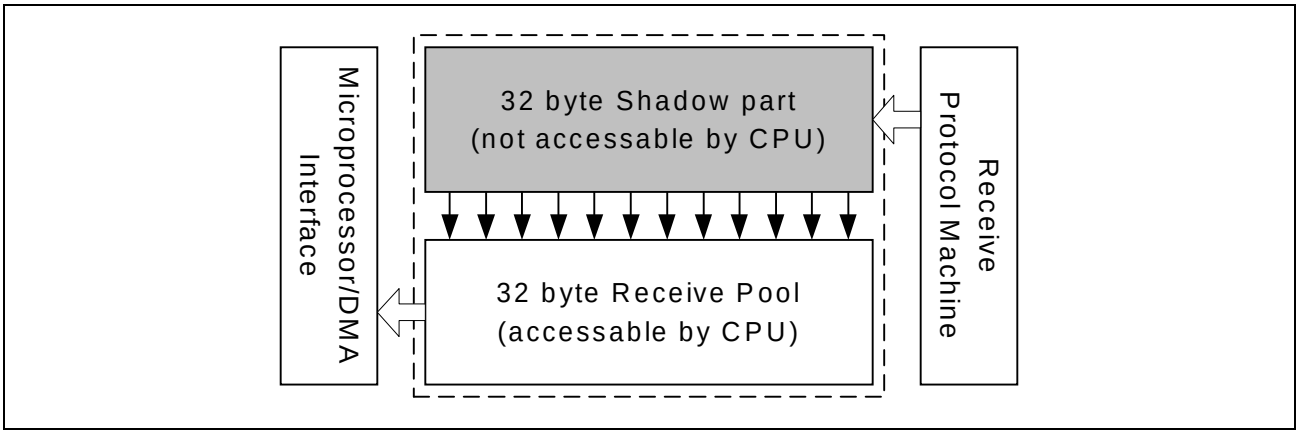


Figure 3-3 SCC Receive FIFO

New receive data is announced to the CPU with an interrupt latest when the FIFO fill level reaches a chosen threshold level (selected with bitfield 'RFTH(1..0)' in register "CCR3H" on page 5-149). Default value for this threshold level is 32 bytes.

If the SCC receive FIFO is completely filled, further incoming data is ignored and a receive data overflow condition ('RDO') is detected. As soon as the receive FIFO provides empty space, receive data is accepted again after a frame end or frame abort sequence. The automatically generated receive status byte (RSTA) will contain an 'RDO' indication in this case and the next incoming frame will be received in a normal way.

Therefore no further CPU intervention is necessary to recover the SCC from an 'RDO' condition.

A "frame" with 'RDO' status might be a mixture of a frame partly received before the 'RDO' event occured and the rest of this frame received after the receive FIFO again accepted data and the frame was still incoming. A quite arbitrary series of data or complete frames might get lost in case of an 'RDO' event. Every frame which is completely discarded because of an 'RDO' condition generates an 'RFO' interrupt.

The SCC receive FIFO can be cleared by command 'RRES' in register CMDRH. Note that clearing the receive FIFO during operation might delete a frame end / block end indication. A frame which was already partly transferred cannot be "closed" in this case. A new frame received after receiver reset command will be appended to this "open" frame.

3.2.2.3 SCC FIFO Access

Figure 3-4 and **Figure 3-5** illustrate byte interpretation for Intel and Motorola 16-bit accesses to the transmit and receive FIFOs.

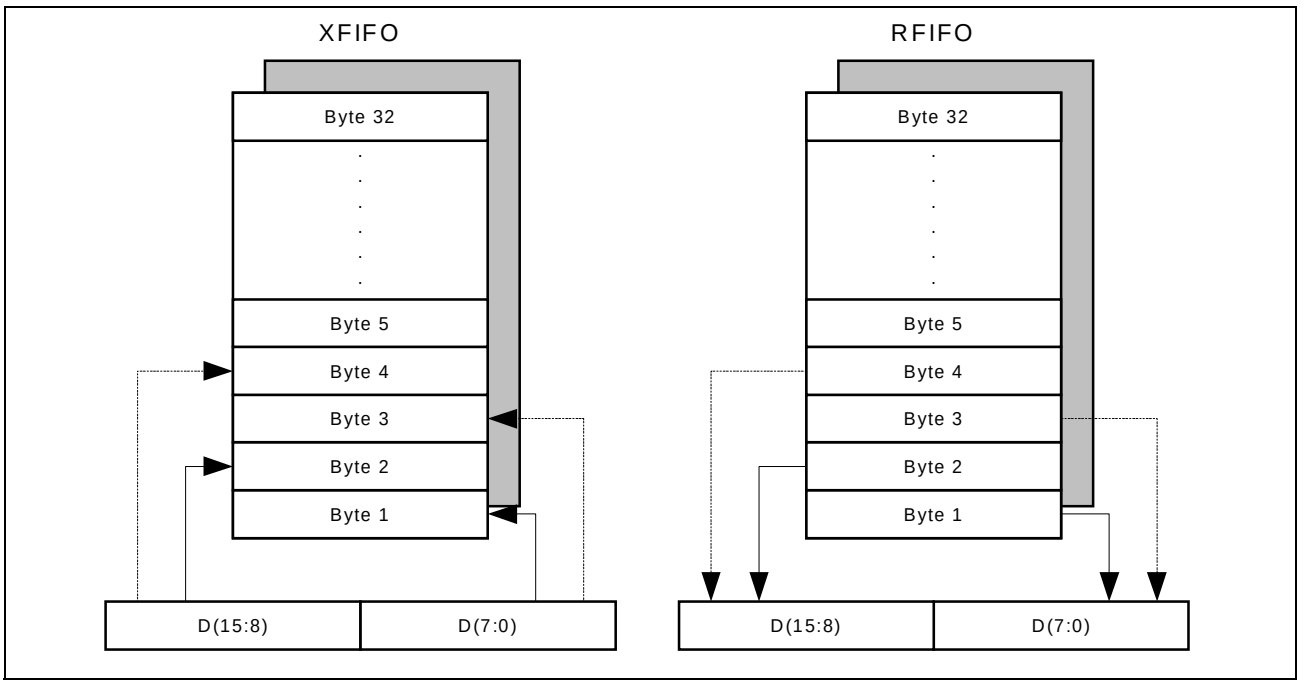


Figure 3-4 XFIFO/RFIFO Word Access (Intel Mode)

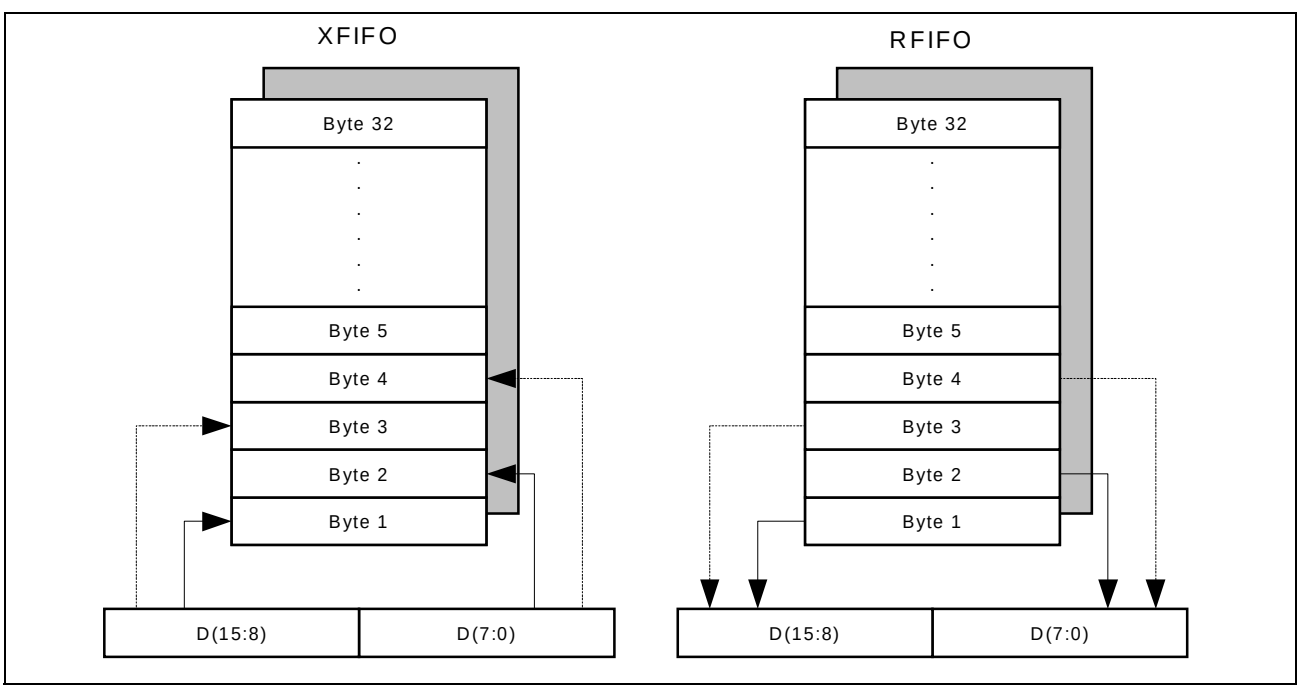


Figure 3-5 XFIFO/RFIFO Word Access (Motorola Mode)

3.2.3 Clocking System

The PASSAT includes an internal Oscillator (OSC) as well as two independent Baud Rate Generators (BRG) and two Digital Phase Locked Loop (DPLL) circuits.

The transmit and receive clock can be generated either

- externally, and supplied directly via the RxCLK and/or TxCLK pins (called external clock modes)
- internally, by selecting
  - the internal oscillator (OSC) and/or the channel specific baud rate generator (BRG)
  - the internal DPLL, recovering the receive (and optionally transmit) clock from the receive data stream.

  (called internal clock modes)

There are a total of 14 different clocking modes programmable via bit field 'CM' in register CCR0L, providing a wide variety of clock generation and clock pin functions, as shown in **Table 3-2.**

The transmit clock pins (TxCLK) may also be configured as output clock and control signals in certain clock modes if enabled via bit 'TOE' in register CCR0L.

The clocking source for the DPLL's is always the internal channel specific BRG; the scaling factor (divider) of the BRG can be programmed through BRRL and BRRH registers.

There are two channel specific internal operational clocks in the SCC:

One operational clock (= transmit clock) for the transmitter part and one operational clock (= receive clock) for the receiver part of the protocol logic.

Note: The internal timers always run using the internal transmit clock.

Table 3-1 Overview of Clock Modes

Clock			
Type	**Source**	**Generation**	**Clock Mode**
Receive Clock	RxCLK Pins	Externally	0, 1, 4, 5
	OSC, DPLL, BRG,	Internally	 2, 3a, 6, 7a 3b, 7b
Transmit Clock	TxCLK Pins, RxCLK Pins	Externally	0a, 2a, 4, 6a 1,5
	OSC, DPLL, BRG/BCR, BRG	Internally	 3a, 7a 2b, 6b 0b, 3b, 7b

The internal structure of each SCC channel consists of 3 clocking domains, transmit, receive, and system. These three function blocks are clocked with internal transmit frequency $\mathbf{f_{TRM}}$, internal receive frequency $\mathbf{f_{REC}}$ and system frequency $\mathbf{f_{SYS}}$, respectively (system frequency $\mathbf{f_{SYS}}$ only supplies the SCC receive and transmit FIFO part facing the microprocessor interface). The internal FIFO interfaces are used to transfer data between the different clock domains.

The clocks $\mathbf{f_{TRM}}$ and $\mathbf{f_{REC}}$ are internal clocks only and need not be identical to external clock inputs e.g. $\mathbf{f_{TRM}}$ and TxCLK input pin.

The features of the different clock modes are summarized in **Table 3-2**.

Table 3-2 Clock Modes of the SCCs

Channel Configuration		Clock Sources				Control Sources					
Clock Mode CCR0L: CM(2..0)	CCR0L: SSEL	to BRG	to DPLL	to REC	to TRM	CD	R- Strobe	X- Strobe	Frame-Sync Tx	Rx	Output via TxCLK (if CCR0L: TOE = '1')
0a	0	–	–	RxCLK	TxCLK	CD	–	–	–	–	–
0b	1	OSC	–	RxCLK	BRG	CD	–	–	–	–	BRG
1	X	–	–	RxCLK	RxCLK	–	CD	TxCLK	–	–	–
2a	0	RxCLK	BRG	DPLL	TxCLK	CD	–	–	–	–	–
2b	1	RxCLK	BRG	DPLL	BRG/16	CD	–	–	–	–	BRG/16
3a	0	RxCLK	BRG	DPLL	DPLL	CD	–	–	–	–	DPLL
3b	1	RxCLK	–	BRG	BRG	CD	–	–	–	–	BRG
4	X	–	–	RxCLK	TxCLK	–	$\overline{\text{RCG}}$	$\overline{\text{TCG}}$	–	–	-
5a	0	–	–	RxCLK	RxCLK	–	(TSAR/ PCMRX)	(TSAX/ PCMTX)	FSC	FSC	TS-Control
5b	1	–	–	RxCLK	TxCLK	–	(TSAR/ PCMRX)	(TSAX/ PCMTX)	OST	OSR	–
6a	0	OSC	BRG	DPLL	TxCLK	CD	–	–	–	–	–
6b	1	OSC	BRG	DPLL	BRG/16	CD	–	–	–	–	BRG/16
7a	0	OSC	BRG	DPLL	DPLL	CD	–	–	–	–	DPLL
7b	1	OSC	–	BRG	BRG	CD	–	–	–	–	BRG

Note: If one of the clock modes 0b, 6 or 7 is selected, the internal oscillator (OSC) is enabled which allows connection of an external crystal to pins XTAL1-XTAL2. The output signal of the OSC can be used for one serial channel, or for both serial channels (independent baud rate generators and DPLLs). Moreover, XTAL1 alone can be used as input for an externally generated clock.

The first two columns of **Table 3-2** list all possible clock modes configured via bit field 'CM' and bit 'SSEL' in register CCR0L.

For example, clock mode 6b is choosen by writing a '6' to register CCR0L.CM(2:0) and by setting bit CCR0L.SSEL equal to '1'. The following 4 columns (grouped as 'Clock Sources') specify the source of the internal clocks. Columns REC and TRM correspond to the domain clock frequencies $\mathbf{f_{REC}}$ and $\mathbf{f_{TRM}}$.

The columns grouped as 'Control Sources' cover additional clock mode dependent control signals like strobe signals (clock mode 1), clock gating signals (clock mode 4) or synchronization signals (clock mode 5). The last column describes the function of signal TxCLK which in some clock modes can be enabled as output signal monitoring the effective transmit clock or providing a time slot control signal (clock mode 5).

The following is an example of how to read **Table 3-2**:

For clock mode 6b (row '6b') the TRM clock (column 'TRM') is supplied by the baudrate generator (BRG) output divided by 16 (source BRG/16). The BRG (column 'BRG') is derived from the internal oscillator which is supplied by pin XTAL1 and XTAL2.

The REC clock (column 'REC') is supplied by the internal DPLL which itself is supplied by the baud rate generator (column 'DPLL') again.

Note: The REC clock is DPLL clock divided by 16.

If enabled by bit 'TOE' in register CCR0L the resulting transmit clock can be monitored via pin TxCLK (last column, row '6b').

The clocking concept is illustrated in a block diagram manner in the following figure:

Additional control signals are not illustrated (please refer to the detailed clock mode descriptions below).

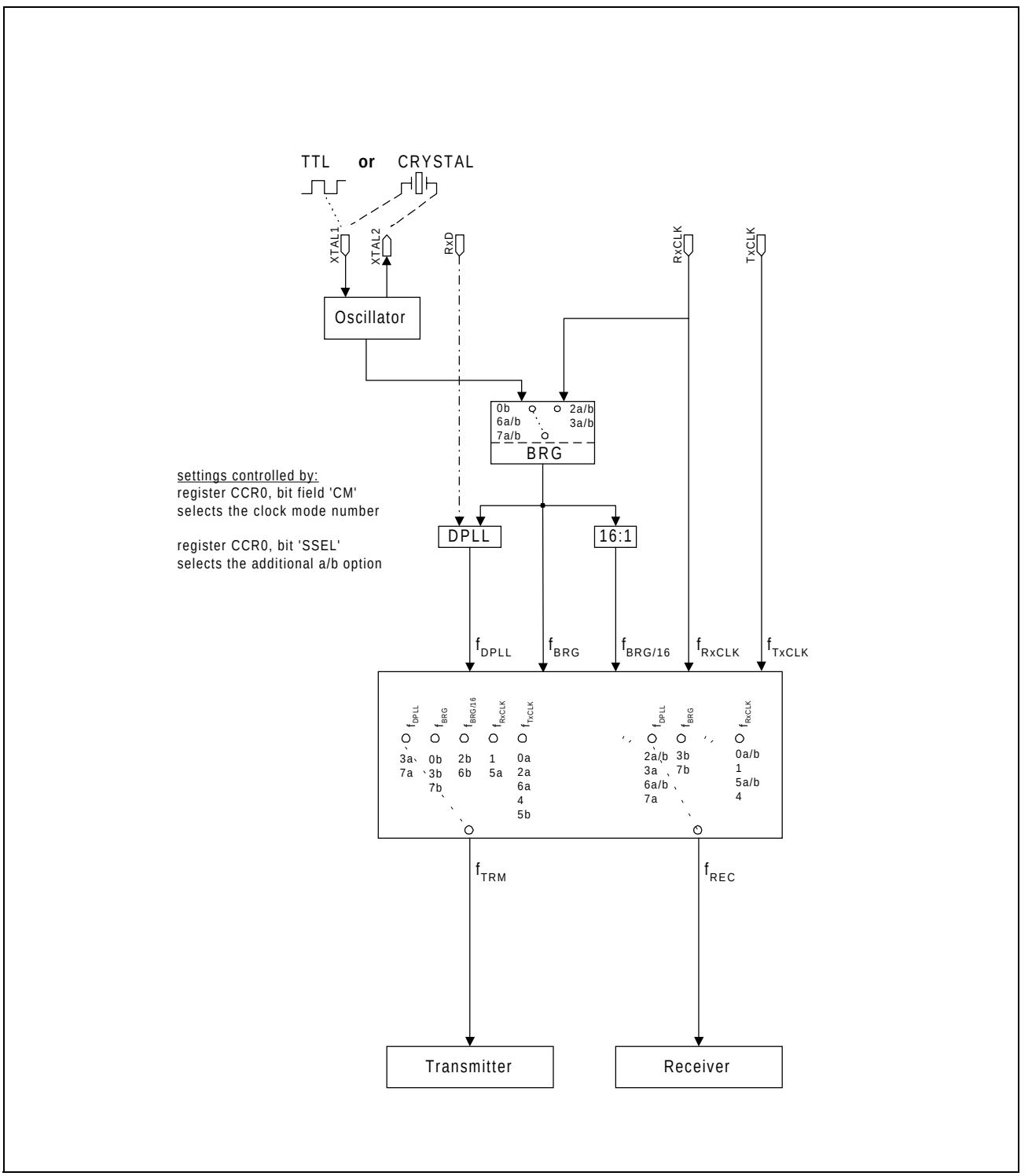


Figure 3-6 Clock Supply Overview

Clock Modes

3.2.3.1 Clock Mode 0 (0a/0b)

Separate, externally generated receive and transmit clocks are supplied to the SCC via their respective pins. The transmit clock may be directly supplied by pin TxCLK (clock mode 0a) or generated by the internal baud rate generator from the clock supplied at pin XTAL1 (clock mode 0b).

In clock mode 0b the resulting transmit clock can be driven out to pin TxCLK if enabled via bit 'TOE' in register CCR0L.

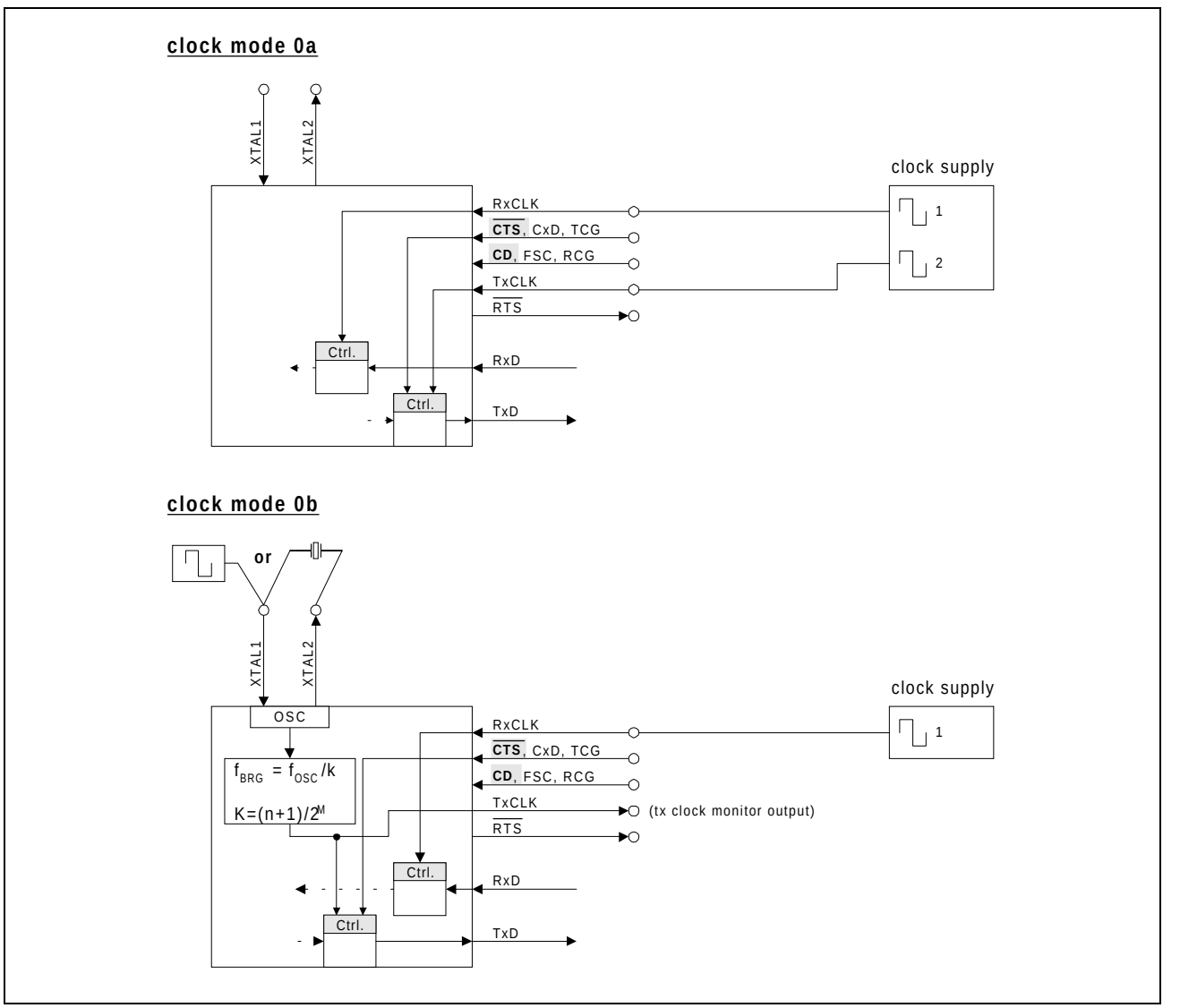


Figure 3-7 Clock Mode 0a/0b Configuration

3.2.3.2 Clock Mode 1

Externally generated RxCLK is supplied to both the receiver and transmitter. In addition, a receive strobe can be connected via CD and a transmit strobe via TxCLK pin. These strobe signals work on a per bit basis. This operating mode can be used in time division multiplex applications or for adjusting disparate transmit and receive data rates.

Note: In Extended Transparent Mode, the above mentioned strobe signals provide byte synchronization (byte alignment).
This means that the strobe signal needs to be detected once only to transmit or receive a complete byte.

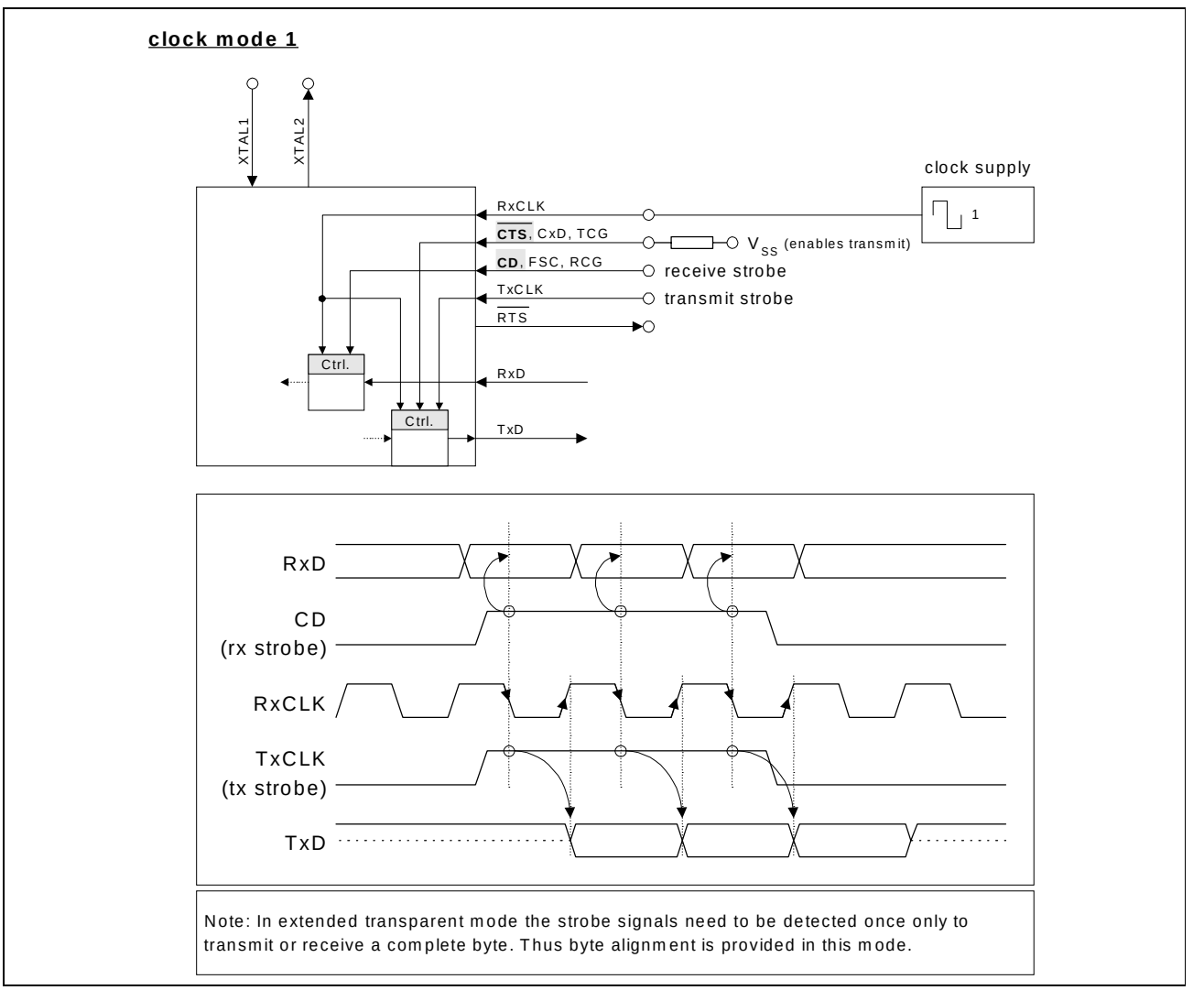


Figure 3-8 Clock Mode 1 Configuration

3.2.3.3 Clock Mode 2 (2a/2b)

The BRG is driven by an external clock (RxCLK pin) and delivers a reference clock for the DPLL which is 16 times of the resulting DPLL output frequency which in turn supplies the internal receive clock. Depending on the programming of register CCR0L bit 'SSEL', the transmit clock will be either an external input clock signal provided at pin TxCLK in clock mode 2a or the clock delivered by the BRG divided by 16 in clock mode 2b. In the latter case, the transmit clock can be driven out to pin TxCLK if enabled via bit 'TOE' in register CCR0L.

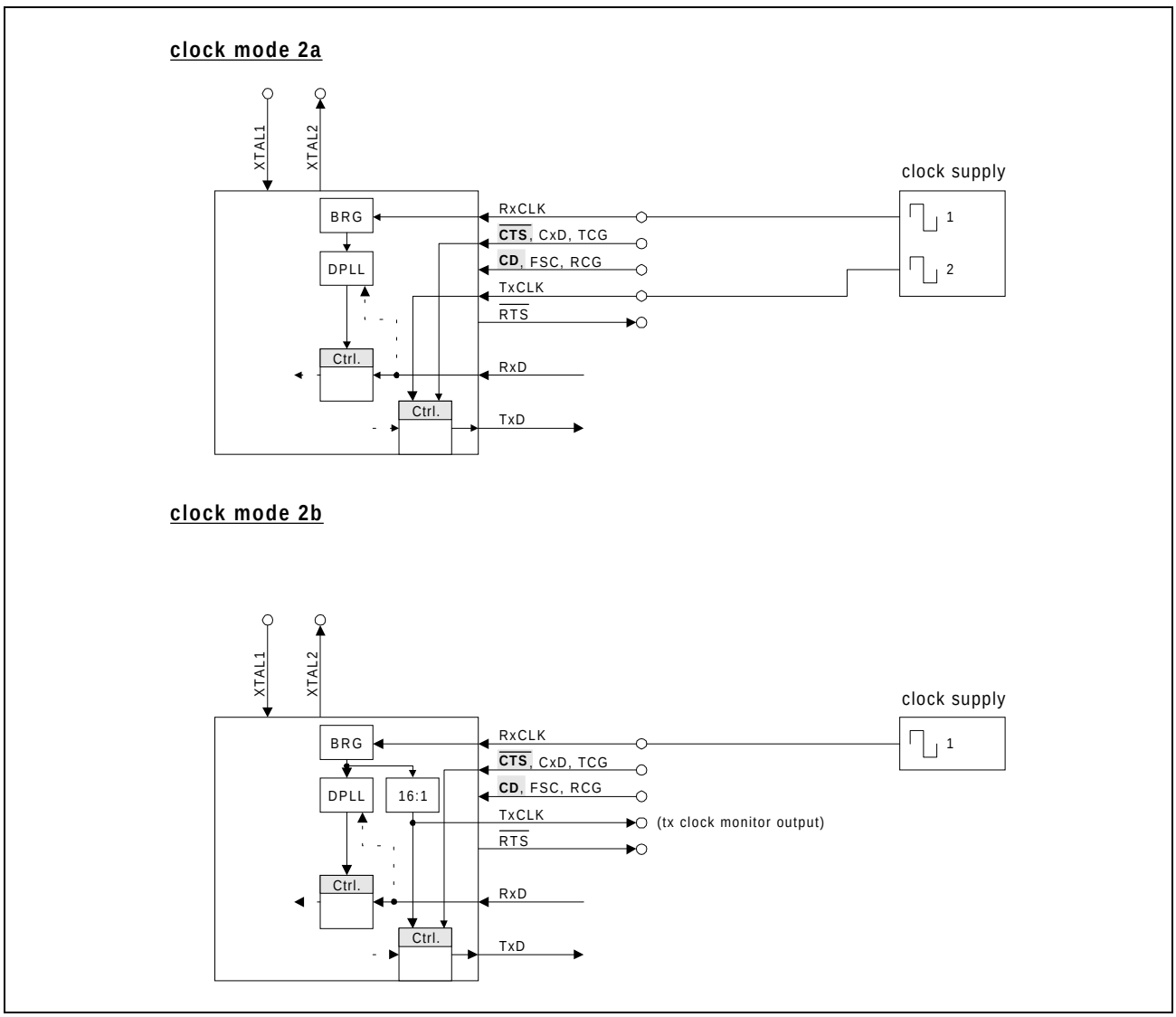


Figure 3-9 Clock Mode 2a/2b Configuration

3.2.3.4 Clock Mode 3 (3a/3b)

The BRG is fed with an externally generated clock via pin RxCLK. Depending on the value of bit 'SSEL' in register CCR0L the BRG delivers either a reference clock for the DPLL which is 16 times of the resulting DPLL output frequency (clock mode 3a) or delivers directly the receive and transmit clock (clock mode 3b). In the first case the DPLL output clock is used as receive and transmit clock.

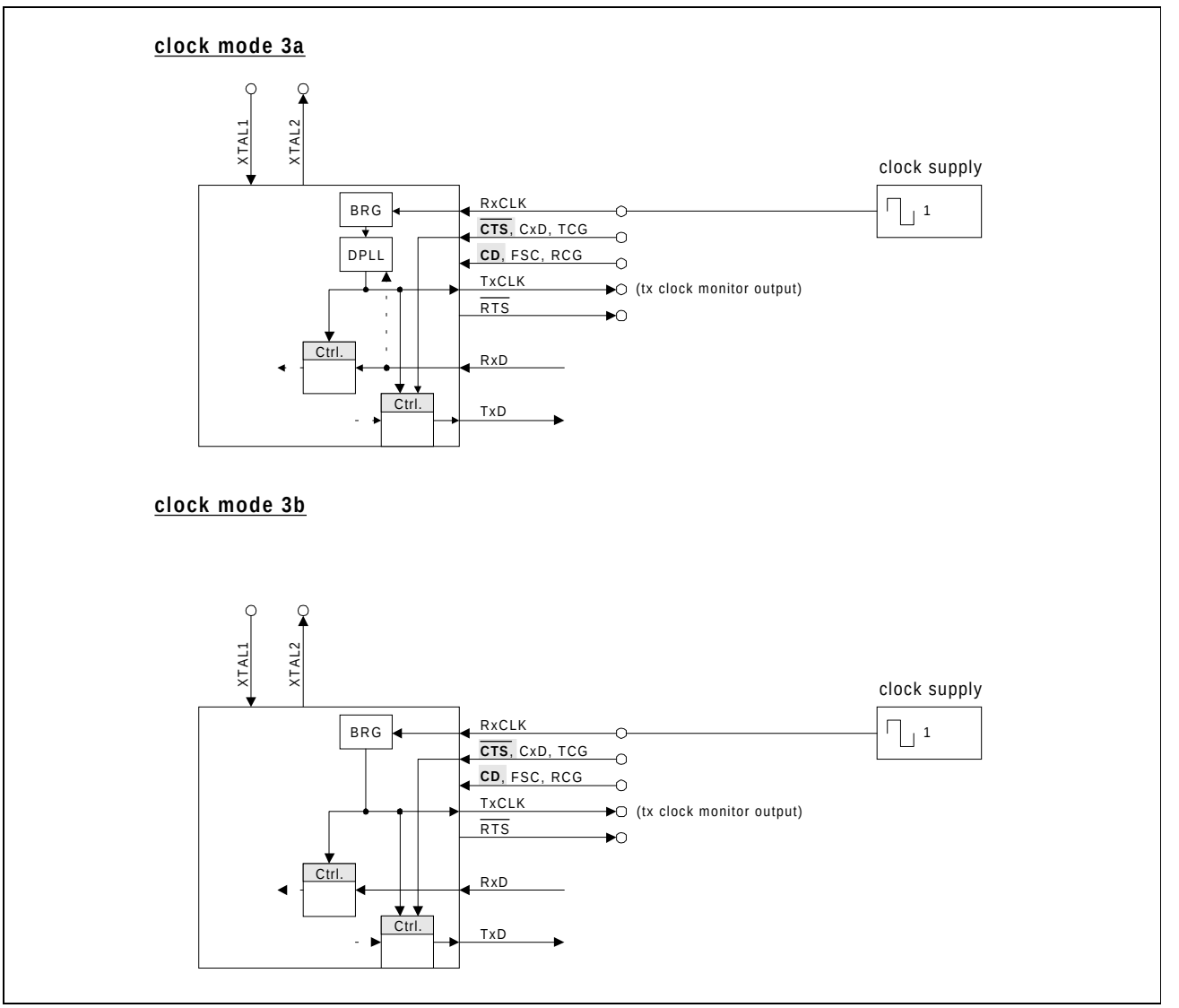


Figure 3-10 Clock Mode 3a/3b Configuration

3.2.3.5 Clock Mode 4

Separate, externally generated receive and transmit clocks are supplied via pins RxCLK and TxCLK. In addition separate receive and transmit clock gating signals are supplied via pins $\overline{\text{RCG}}$ and $\overline{\text{TCG}}$. These gating signals work on a per bit basis.

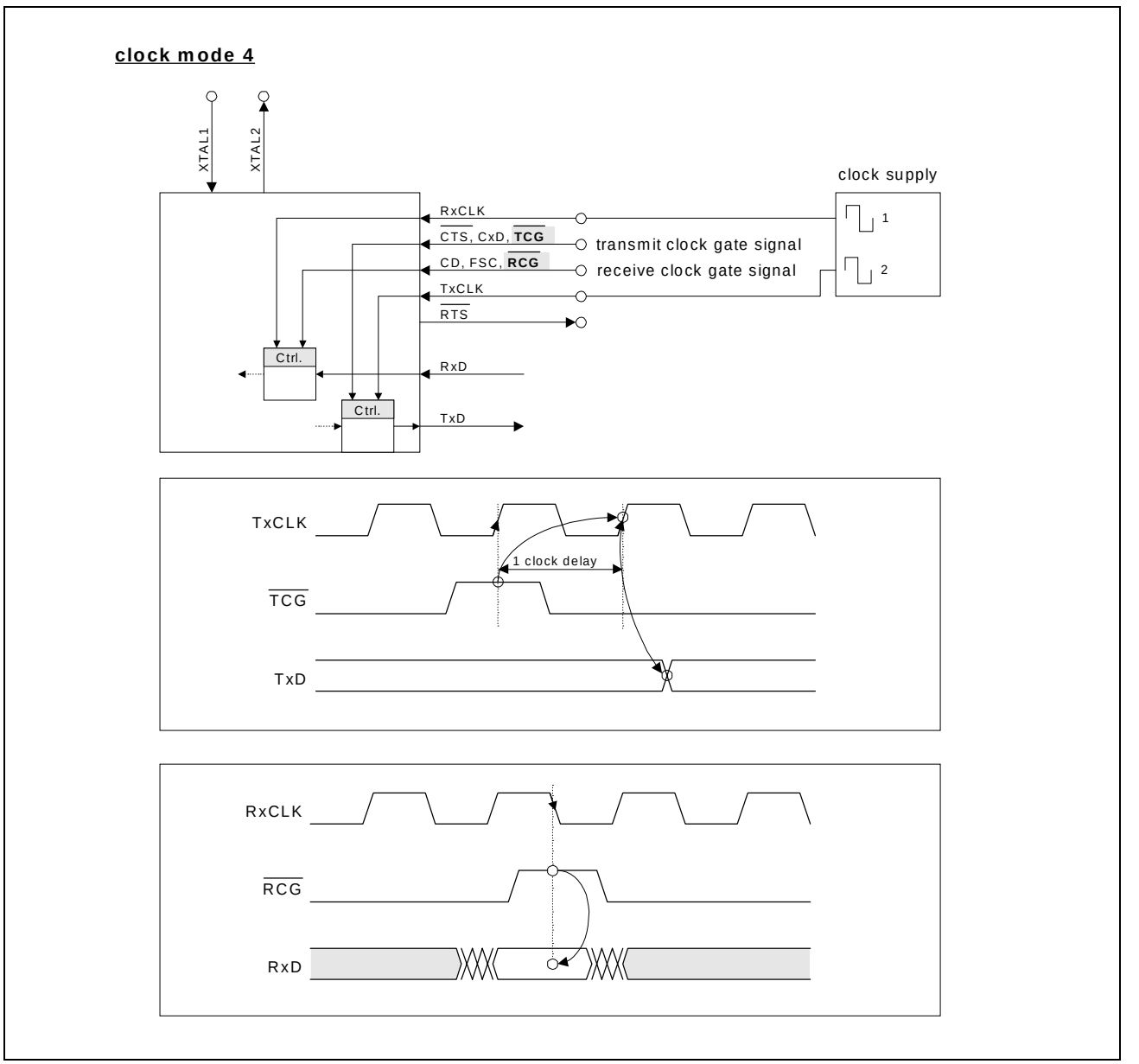


Figure 3-11 Clock Mode 4 Configuration

3.2.3.6 Clock Mode 5a (Time Slot Mode)

This operation mode has been designed for application in time-slot oriented PCM systems.

Note: For correct operation NRZ data coding/encoding should be used.

The receive and transmit clock are common for each channel and must be supplied externally via pin RxCLK. The SCC receives and transmits only during fixed time-slots. Either one time-slot

– of programmable width (1 … 512 bit, via TTSA and RTSA registers), and
– of programmable location with respect to the frame synchronization signal (via pin FSC)

or up to 32 time-slots

– of constant width (8 bits), and
– of programmable location with respect to the frame synchronization signal (via pin FSC)

can be selected.

The time-slot locations can be programmed independently for receive and transmit direction via TTSA/RTSA and PCMTX/PCMRX registers.

Depending on the value programmed via those registers, the receive/transmit time-slot starts with a delay of 1 (minimum delay) up to 1024 clock periods following the frame synchronization signal.

Figure 3-12 shows how to select a time-slot of programmable width and location and **Figure 3-13** shows how to select one or more time-slots of 8-bit width.

If bit ’TOE’ in register CCR0L is set, the selected transmit time-slot(s) is(are) indicated at an output status signal via pin TxCLK, which is driven to ‘low’ during the active transmit window.

Bit ’TSCM’ in register CCR1H determines whether the internal offset counters are continuously running even if no synchronization pulse is detected at FSC signal or stopping at their maximum value.

In the continuous case the repetition rate of offset counter operation is 1024 transmit or receive clocks respectively. An FSC pulse detected earlier resets the counters and starts operation again.

In the non-continuous case the time slot assigner offset counter is stopped after the counter reached its maximum value and is started again if an FSC pulse is detected.

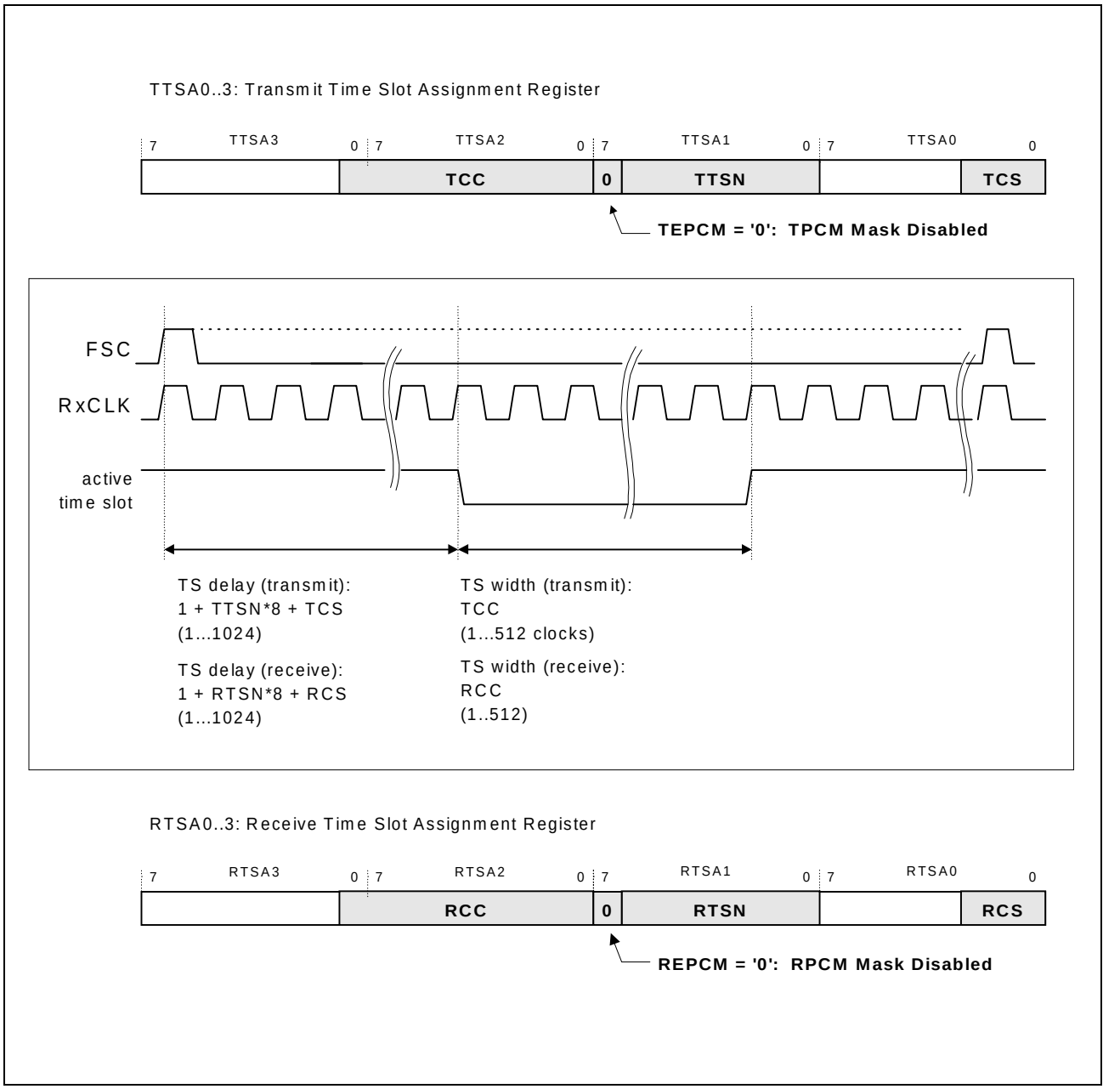


Figure 3-12 Selecting one time-slot of programmable delay and width

*Note: If time-slot 0 is to be selected, the DELAY has to be as long as the PCM frame itself to achieve synchronization (at least for the 2nd and subsequent PCM frames): DELAY = PCM frame length = 1 + xTSN*8 + xCS. xTSN and xCS have to be set appropriately.*

Example: Time-slot 0 in E1 (2.048 Mbit/s) system has to be selected.
*PCM frame length is 256 clocks. 256 = 1+ xTSN*8 + xCS. => xTSN = 31, xCS = 7.*

Note: In extended transparent mode the width xCC of the selected time-slot has to be n × 8 bit because of character synchronization (byte alignment). In all other modes the width can be used to define windows down to a minimum length of one bit.

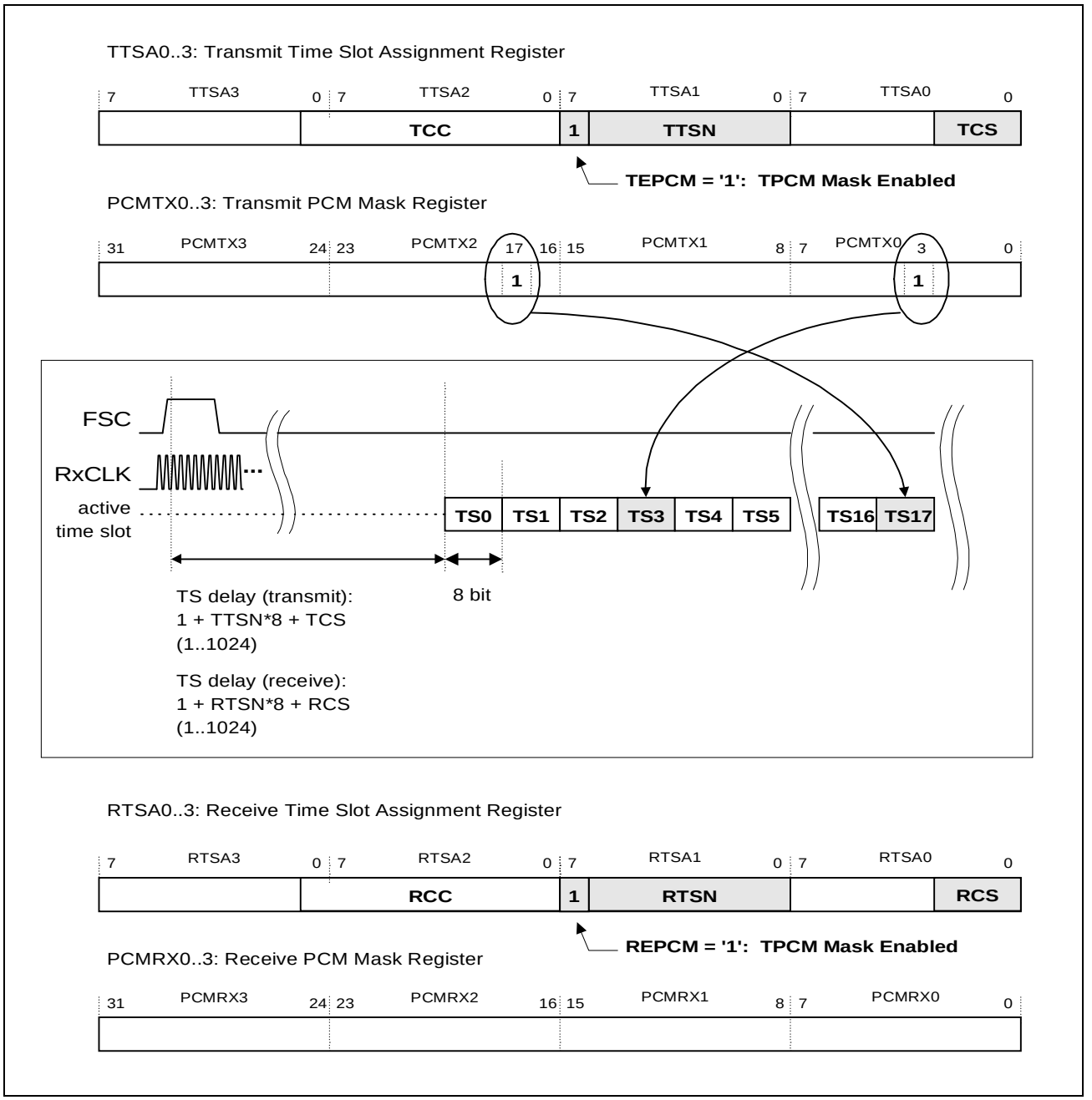


Figure 3-13 Selecting one or more time-slots of 8-bit width

The common transmit and receive clock is supplied at pin RxCLK and the common frame synchronisation signal at pin FSC. The "strobe signals" for active time slots are generated internally by the time slot assigner block (TSA) independent in transmit and receive direction.

When the transmit and receive PCM masks are enabled, bit fields ’TCC’ and ’RCC’ are ignored because of the constant 8-bit time slot width.

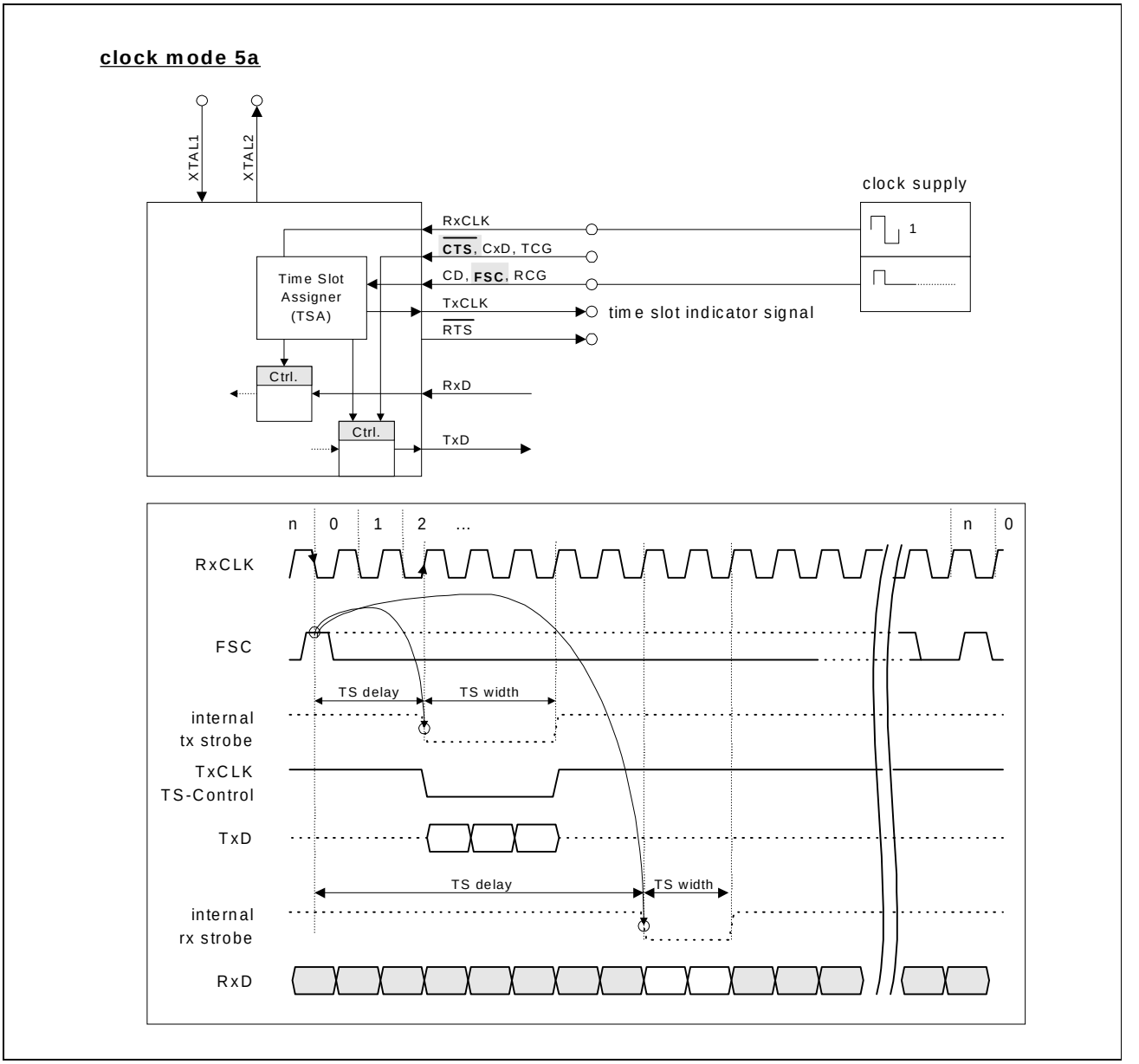


Figure 3-14 Clock Mode 5a Configuration

Note: The transmit time slot delay and width is programmable via bit fields 'TTSN', 'TCS' and 'TCC' in registers TTSA0..TTSA3.
The receive time slot delay and width is programmable via bit fields 'RTSN', 'RCS' and 'RCC' in registers RTSA0..RTSA3.

The following figures provide a more detailed description of the TSA internal counter operation and exceptional cases:

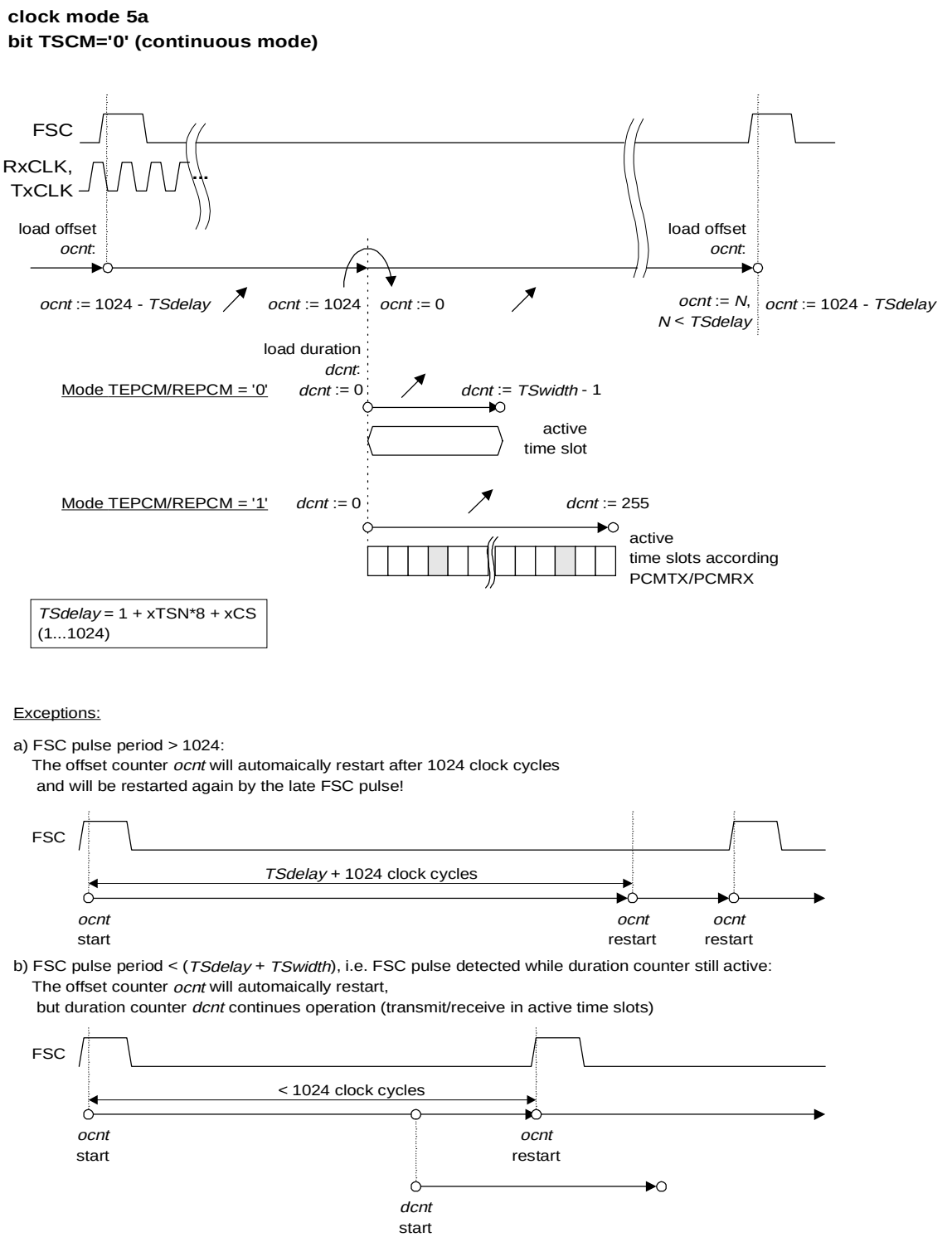


Figure 3-15 Clock Mode 5a "Continuous Mode"

Each frame sync pulse starts the internal offset counter with (1024 - TSdelay) whereas *TSdelay* is the configured value defining the start position. Whenever the offset counter reaches its maximum value 1024, it triggers the duration counter to start operation.

If continuous mode is selected (bit CCR1H.TSCM=’0’) the offset counter continues starting with value 0 until another frame sync puls is detected or again the maximum value 1024 is reached.

Once the duration counter is triggered it runs out independently from the offset counter, i.e. an active time slot period may overlap with the next frame beginning (frame sync event, refer to exception b) in **Figure 3-15**).

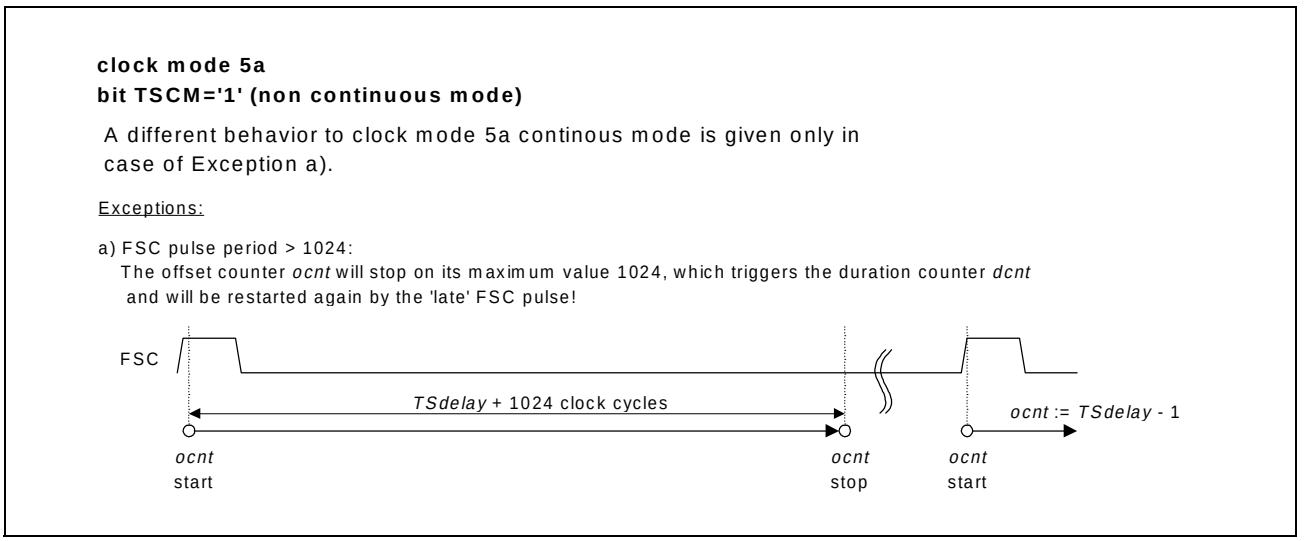


Figure 3-16 Clock Mode 5a "Non Continuous Mode"

If non-continuous mode is selected (bit CCR1H.TSCM=’1’) the offset counter is stopped on its maximum value 1024 until another frame sync puls is detected. This allows frame sync periods greater than 1024 clock cycles, but the accesible part is limited by the range of TSdelay value (1..1024) plus TSwidth (1..512) or plus 256 clock cycles if the PCM mask is selected.

3.2.3.7 Clock Mode 5b (Octet Sync Mode)

This operation mode has been designed for applications using Octet Synchronous PPP.

It is based on clock mode 5a, but only 8-bit (octet) wide time slot operation is supported, i.e. bits TTSA1.TEPCM and RTSA1.REPCM must be set to ’1’. Clock mode 5b provides octet alignment to time slots if Octet Synchronous PPP protocol mode or extended transparent mode is selected.

Note: For correct operation NRZ data coding/encoding should be used.

The receive and transmit clocks are separate and must be supplied at pins RxCLK and TxCLK. The SCC receives and transmits only during fixed octet wide time-slots of programmable location with respect to the octet synchronization signals (via pins OSR and OST)

The time-slot locations can be programmed independently for receive and transmit direction via registers TTSA0..TTSA3 / RTSA0..RTSA3 and PCMTX0..PCMTX3 / PCMRX0..PCMRX3.

Figure 3-17 shows how to select one or more octet wide time-slots.

Bit ’TSCM’ in register CCR1H determines whether the internal counters are continuously running even if no synchronization pulse is detected at OST/OSR signals or stopping at their maximum value.

In the continuous case the repetition rate of operation is 1024 transmit or receive clocks respectively. An OST/OSR pulse detected earlier resets the corresponding offset counter and starts operation again.

In the non-continuous case the transmit/receive time slot assigner offset counter is stopped after the counter reached its maximum value and is started again if an OST/ OSR pulse is detected.

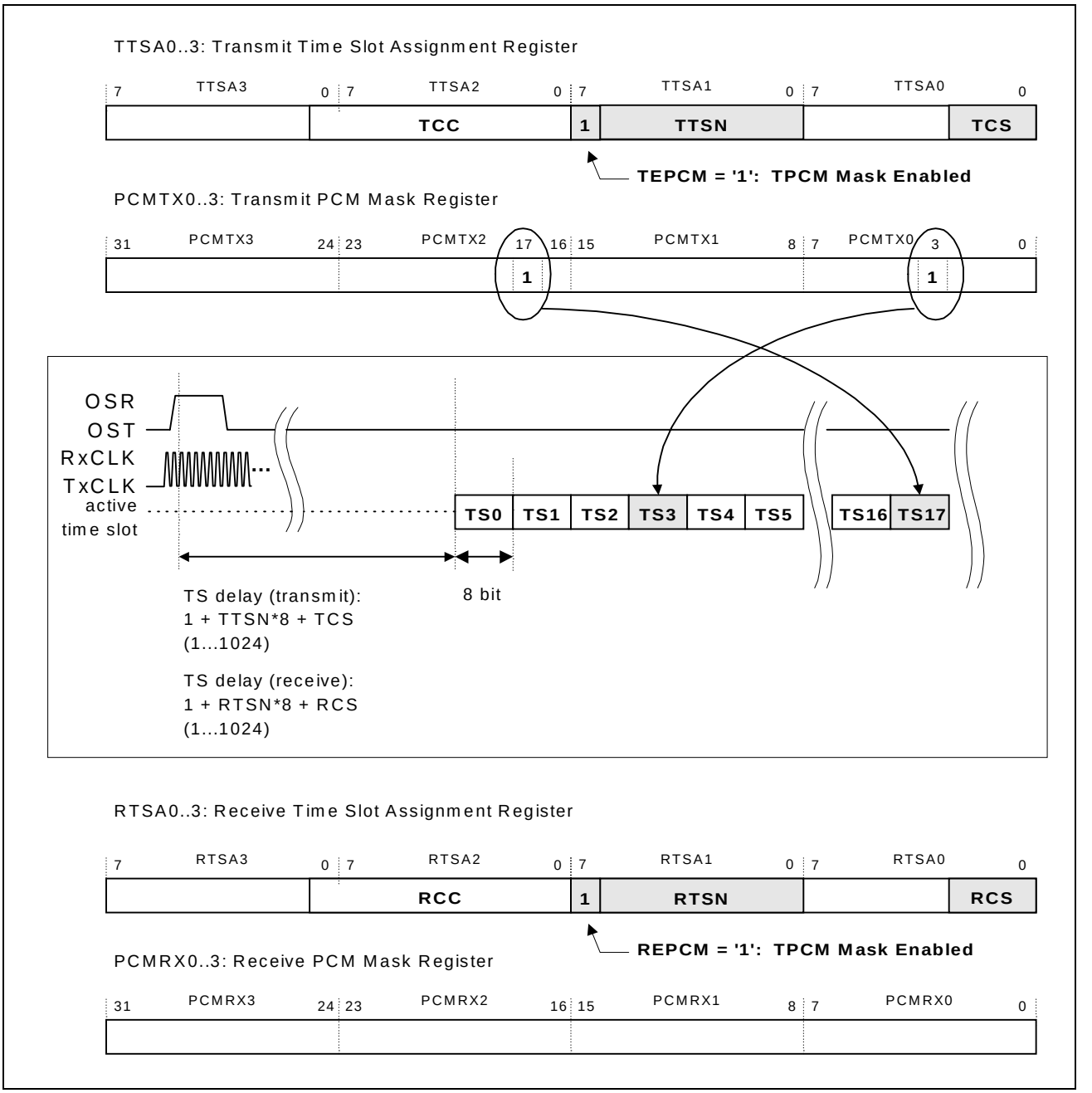


Figure 3-17 Selecting one or more octet wide time-slots

The transmit and receive clocks are supplied at pins RxCLK and TxCLK. The Octet synchronisation signals are supplied at pins OSR and OST. The "strobe signals" for active time slots are generated internally by the time slot assigner blocks (TSA) independent in transmit and receive direction.

Bit fields ’TCC’ and ’RCC’ are ignored because of the constant 8-bit time slot width.

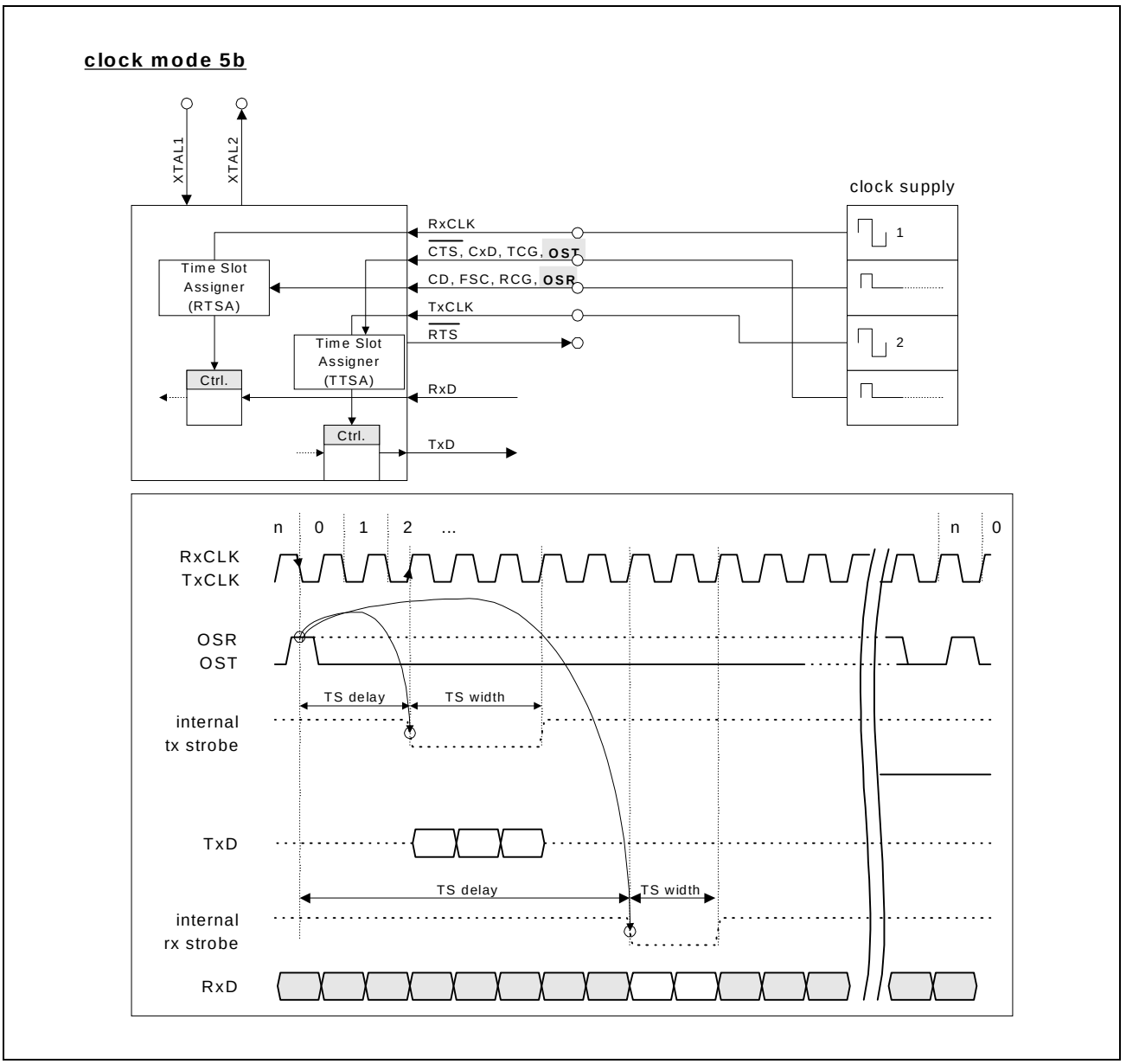


Figure 3-18 Clock Mode 5b Configuration

Note: The transmit time slot delay and width is programmable via bit fields 'TTSN', 'TCS' and 'TCC' in registers TTSA0..TTSA3.
The receive time slot delay and width is programmable via bit fields 'RTSN', 'RCS' and 'RCC' in registers RTSA0..RTSA3.

3.2.3.8 Clock Mode 6 (6a/6b)

This clock mode is identical to clock mode 2a/2b except that the clock source of the BRG is supplied at pin XTAL1.

The BRG is driven by the internal oscillator and delivers a reference clock for the DPLL which is 16 times the resulting DPLL output frequency which in turn supplies the internal receive clock. Depending on the programming of register CCR0L bit 'SSEL', the transmit clock will be either an external input clock signal provided at pin TxCLK in clock mode 6a or the clock delivered by the BRG divided by 16 in clock mode 6b. In the latter case, the transmit clock can be driven out to pin TxCLK if enabled via bit 'TOE' in register CCR0L.

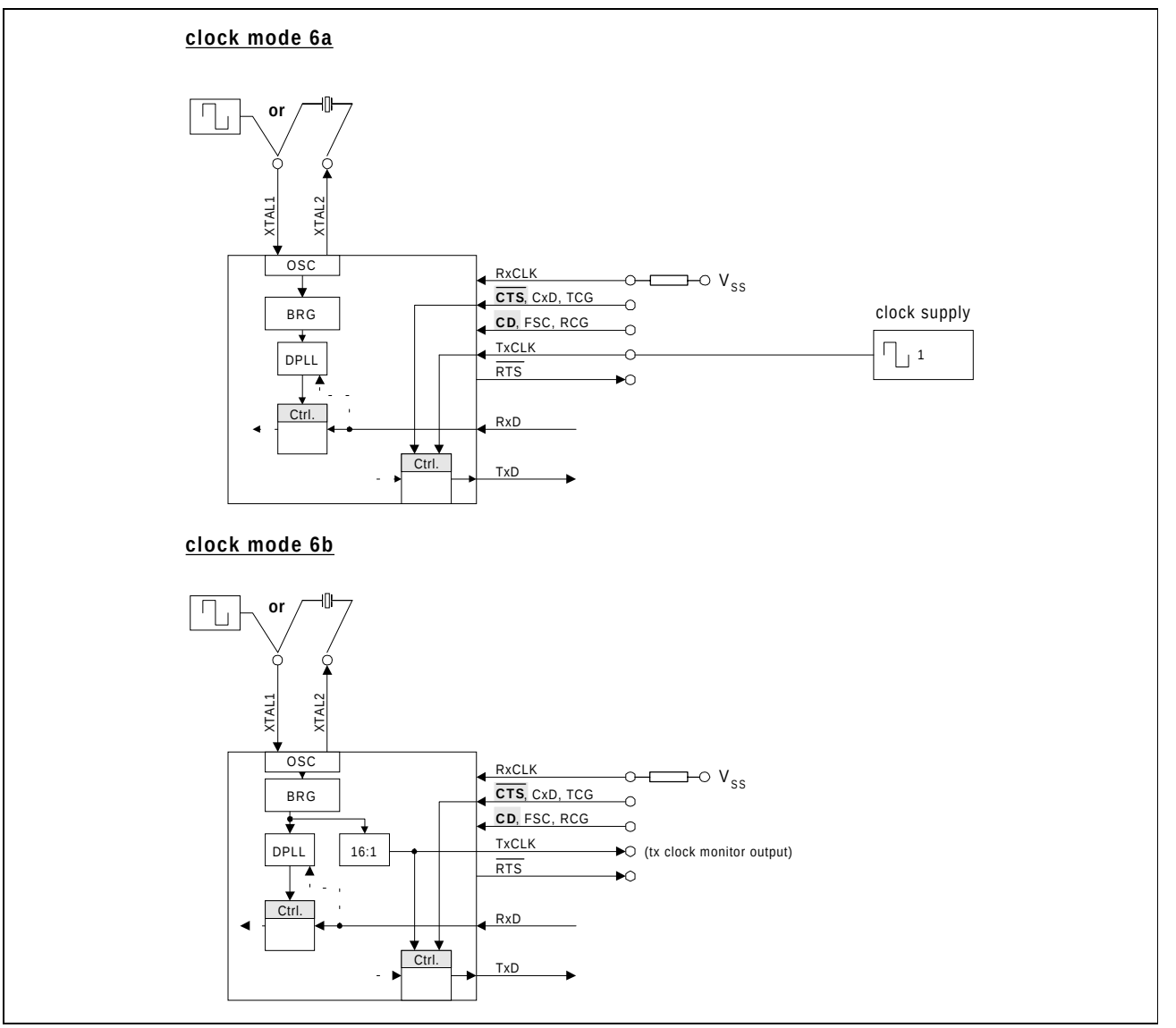


Figure 3-19 Clock Mode 6a/6b Configuration

3.2.3.9 Clock Mode 7 (7a/7b)

This clock mode is identical to clock mode 3a/3b except that the clock source of the BRG is supplied at pin XTAL1.

The BRG is driven by the internal oscillator. Depending on the value of bit 'SSEL' in register CCR0L the BRG delivers either a reference clock for the DPLL which is 16 times the resulting DPLL output frequency (clock mode 7a) or delivers directly the receive and transmit clock (clock mode 7b). In clock mode 7a the DPLL output clocks receive and transmit data.

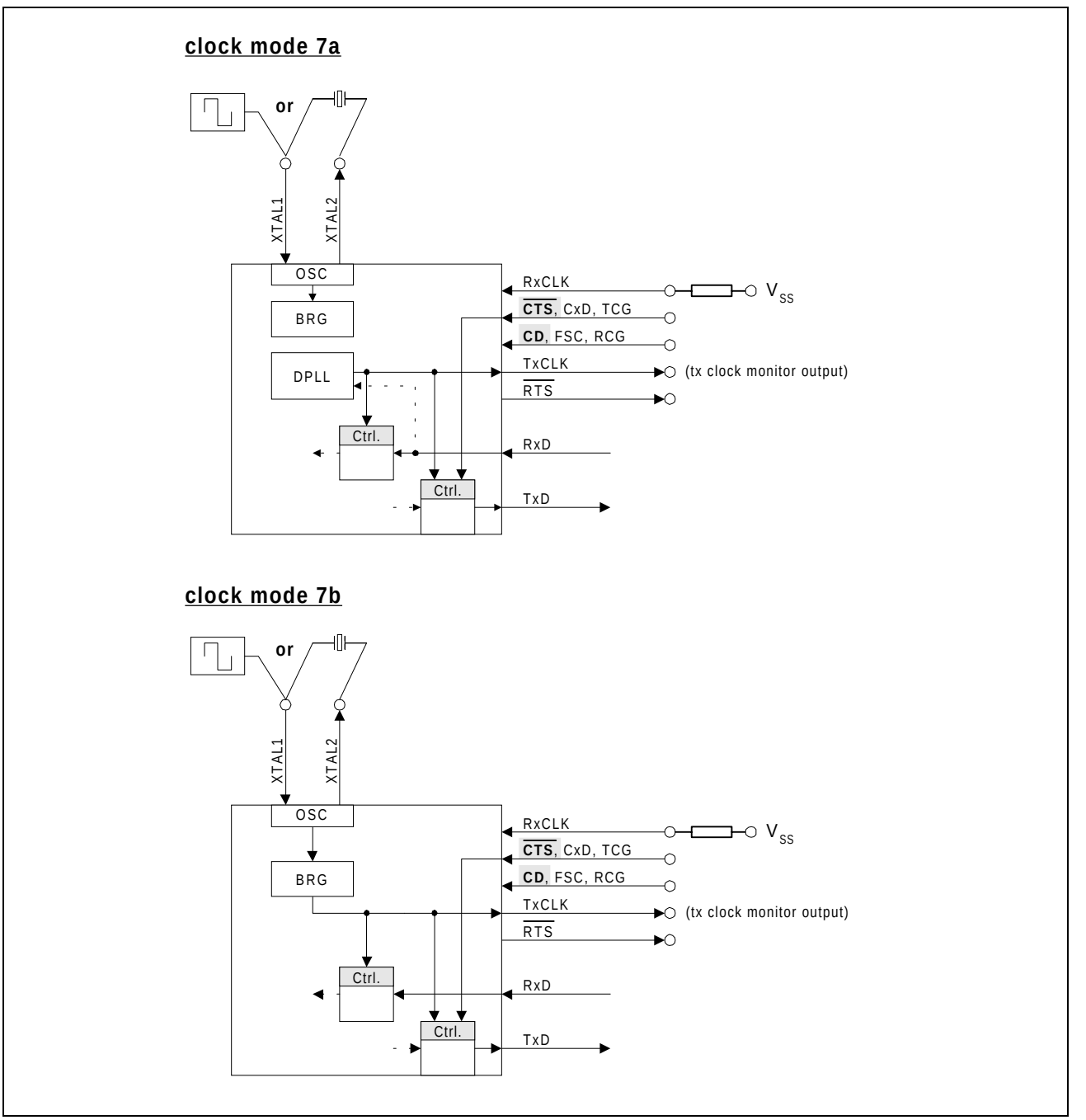


Figure 3-20 Clock Mode 7a/7b Configuration

3.2.4 Baud Rate Generator (BRG)

Each serial channel provides a baud rate generator (BRG) whose division factor is controlled by registers BRRL and BRRH. Whether the BRG is in the clocking path or not depends on the selected clock mode.

Table 3-3 BRRL/BRRH Register and Bit-Fields

Register	Bit-Fields			
Offset	Pos.	Name	Default	Description
BRRL $38_H/88_H$	5..0	BRN	0	Baud Rate Factor N range N = 0..63
BRRH $39_H/89_H$	11..8	BRM	0	Baud Rate Factor M, range M = 0..15

The clock division factor k is calculated by:

$$k = (N + 1) \times 2^M$$

$$f_{BRG} = f_{in} / k$$

3.2.5 Clock Recovery (DPLL)

The SCC offers the advantage of recovering the received clock from the received data by means of internal DPLL circuitry, thus eliminating the need to transfer additional clock information via a separate serial clock line. For this purpose, the DPLL is supplied with a ‘reference clock’ from the BRG which is 16 times the expected data clock rate (clock mode 2, 3a, 6, 7a). The transmit clock may be obtained by dividing the output of the BRG by a constant factor of 16 (clock mode 2b, 6b; bit ’SSEL’ in register CCR0L set) or also directly from the DPLL (clock mode 3a, 7a).

The main task of the DPLL is to derive a receive clock and to adjust its phase to the incoming data stream in order to enable optimal bit sampling.

The mechanism for clock recovery depends on the selected data encoding (see **"Data Encoding" on page 3-71**).

The following functions have been implemented to facilitate a fast and reliable synchronization:

Interference Rejection and Spike Filtering

Two or more edges in the same directional data stream within a time period of 16 reference clocks are considered to be interference and consequently no additional clock adjustment is performed.

Phase Adjustment (PA)

Referring to **Figure 3-21**, **Figure 3-22** and **Figure 3-23**, in the case where an edge appears in the data stream within the PA fields of the time window, the phase will be adjusted by 1/16 of the data.

Phase Shift (PS) (NRZ, NRZI only)

Referring to **Figure 3-21** in the case where an edge appears in the data stream within the PS field of the time window, a second sampling of the bit is forced and the phase is shifted by 180 degrees.

Note: Edges in all other parts of the time window will be ignored.

This operation facilitates a **fast** and reliable synchronization for most common applications. Above all, it implies a very fast synchronization because of the phase shift feature: one edge on the received data stream is enough for the DPLL to synchronize, thereby eliminating the need for synchronization patterns, sometimes called preambles. However, in case of **extremely** high jitter of the incoming data stream the reliability of the clock recovery cannot be guaranteed.

The SCC offers the option to disable the Phase Shift function for NRZ and NRZI encodings by setting bit 'PSD' in register CCR0L to '1'. In this case, the PA fields are extended as shown in **Figure 3-22**.

Now, the DPLL is more insensitive to high jitter amplitudes but needs **more time** to reach the optimal sampling position. To ensure correct data sampling, preambles should precede the data information.

Figure 3-21, **Figure 3-22** and **Figure 3-23** explain the DPLL algorithms used for the different data encodings.

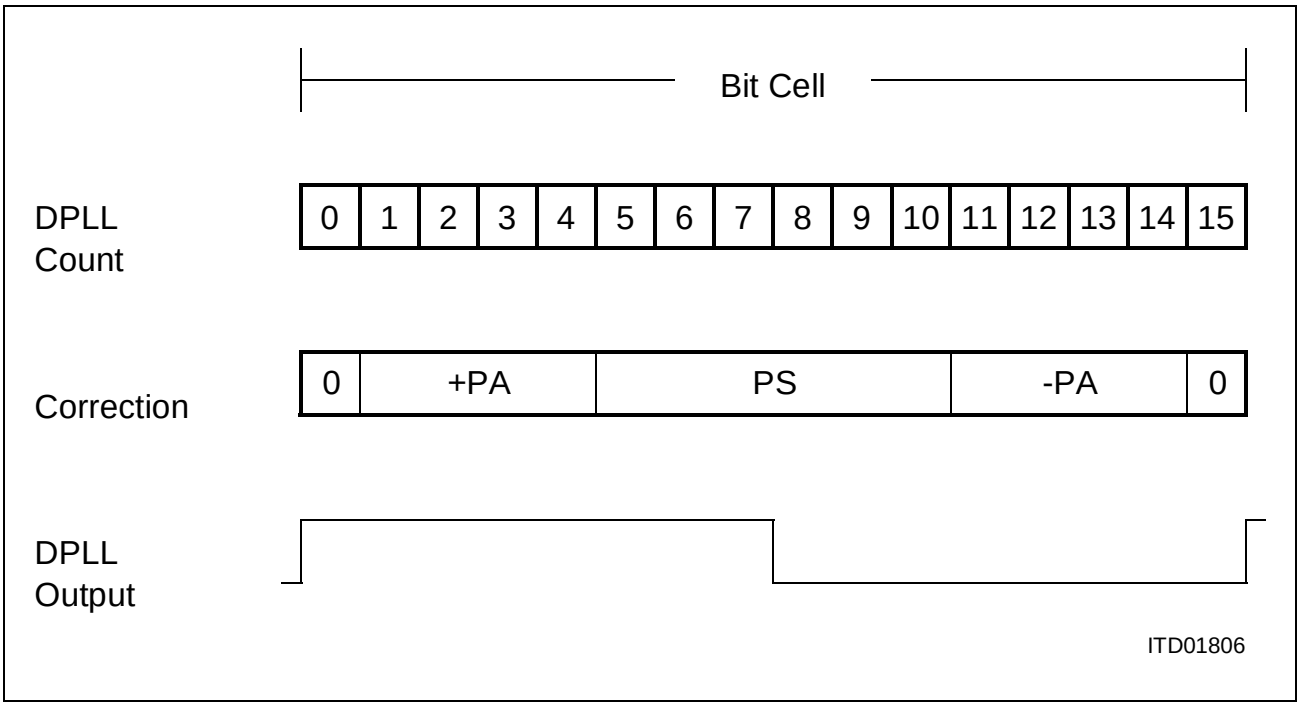


Figure 3-21 DPLL Algorithm (NRZ and NRZI Encoding, Phase Shift Enabled)

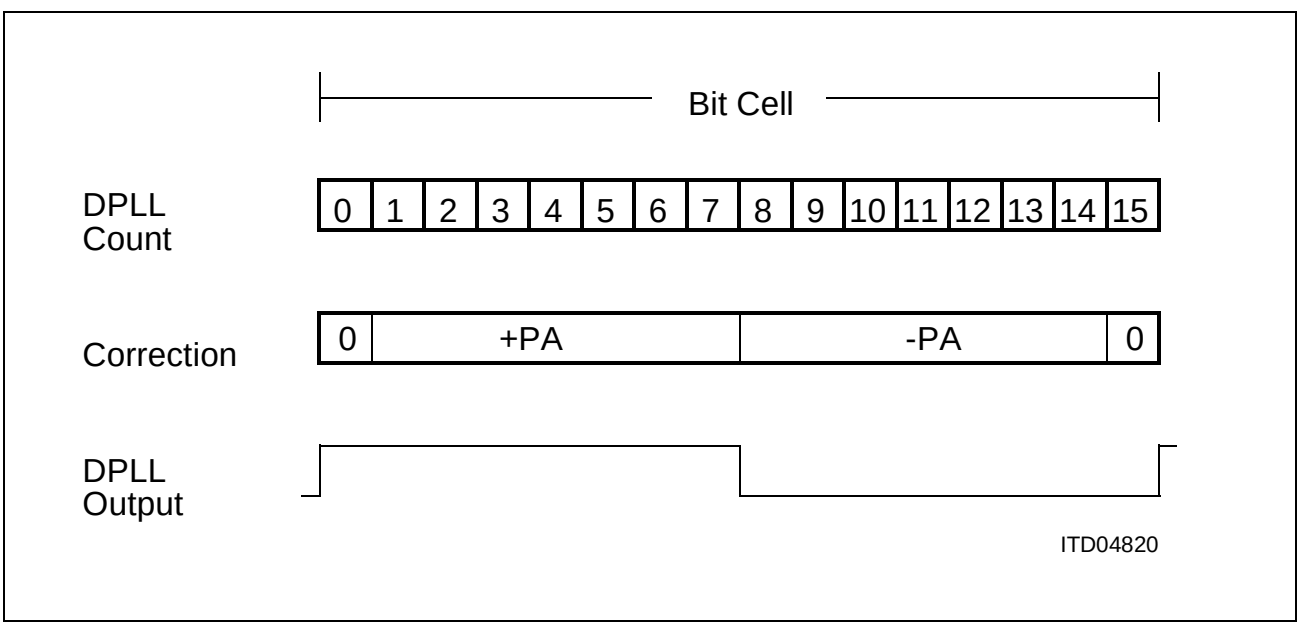


Figure 3-22 DPLL Algorithm (NRZ and NRZI Encoding, Phase Shift Disabled)

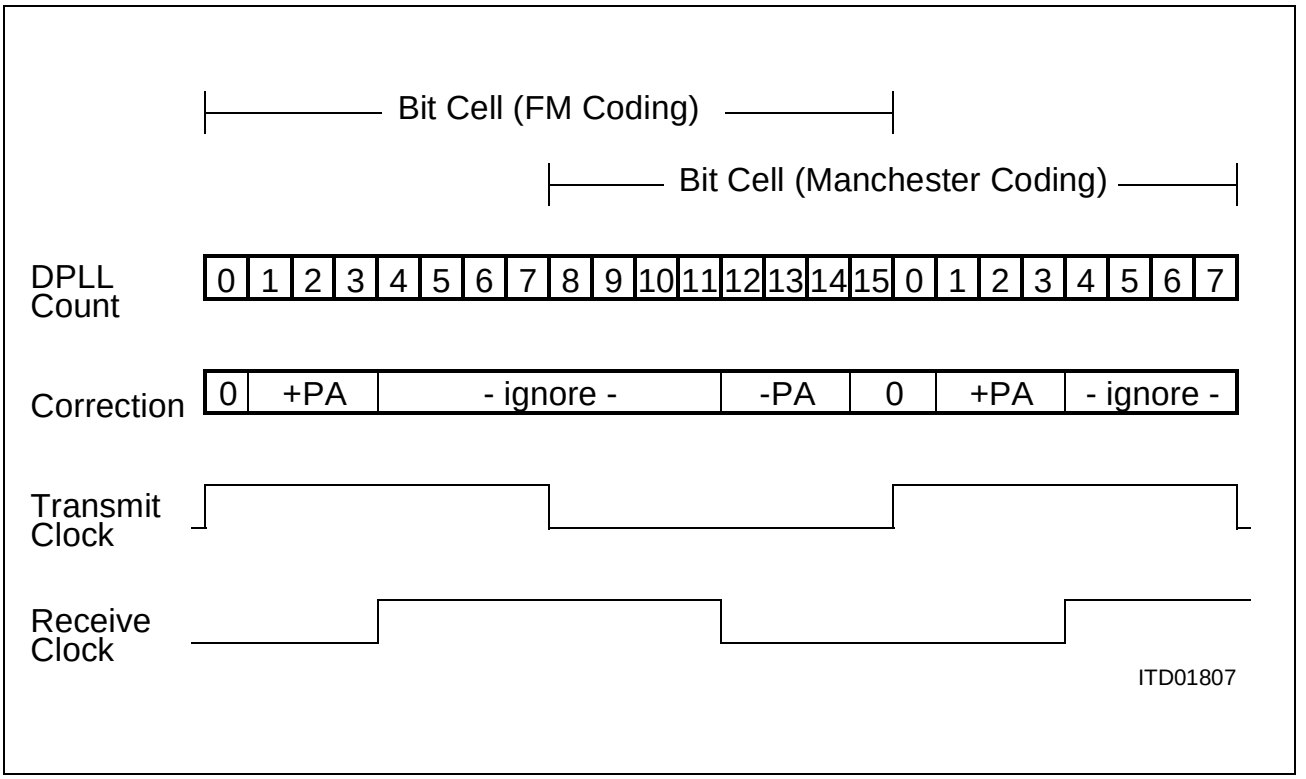


Figure 3-23 DPLL Algorithm for FM0, FM1 and Manchester Encoding

To supervise correct function when using bi-phase encoding, a status flag and a maskable interrupt inform about synchronous/asynchronous state of the DPLL.

3.2.6 SCC Timer Operation

Each SCC provides a general purpose timer e.g. to support protocol functions. In all operating modes the timer is clocked by the effective transmit clock. In clock mode 5 (time-slot oriented mode) the clock source for the timer can be optionally switched to the frame sync clock (input pin FSC) by setting bit 'SRC' in register TIMR3.

The timer is controlled by the CPU via access to registers CMDRL and TIMR0..TIMR3. The timer can be started any time by setting bit 'STI' in register CMDRL. After the timer has expired it generates a timer interrupt ('TIN').

With bit field 'CNT(2..0)' in register TIMR3 the number of automatic timer restarts can be programmed. If the maximum value '111' is entered, a timer interrupt is generated periodically, with the time period determined by bit field 'TVALUE' (registers TIMR0..TIMR3).

The timer can be stopped any time by setting bit 'TRES' in register CMDRL to '1'.

In HDLC Automode the timer is used internally for autonomous protocol functions (refer to the chapter "Automode" on page 4-81). If this operating mode is selected, bit 'TMD' in register TIMR3 must be set to '1'.

3.2.7 SCC Serial Bus Configuration Mode

Beside the point-to-point configuration, the SCC effectively supports point-to-multipoint (pt-mpt, or bus) configurations by means of internal idle and collision detection/collision resolution methods.

In a pt-mpt configuration, comprising a central station (master) and several peripheral stations (slaves), or in a multimaster configuration, data transmission can be initiated by each station over a common transmit line (bus). In case more than one station attempts to transmit data simultaneously (collision), the bus has to be assigned to only one station. A collision-resolution procedure is implemented in the SCC. Bus assignment is based on a priority mechanism with rotating priorities. This allows each station a bus access within a predetermined maximum time delay (deterministic CSMA/CD), no matter how many transmitters are connected to the serial bus.

Prerequisites for bus operation are:

- NRZ encoding
- ‘OR’ing of data from every transmitter on the bus (this can be realized as a wired-OR, using the TxD open drain capability)
- Feedback of bus information (CxD input).

The bus configuration is selected via bitfield SC(2:0) in register CCR0H.

Note: Central clock supply for each station is not necessary if both the receive and transmit clock is recovered by the DPLL (clock modes 3a, 7a). This minimizes the phase shift between the individual transmit clocks.

The bus configuration mode operates independently of the clock mode, e.g. also together with clock mode 1 (receive and transmit strobe operation).

3.2.8 Serial Bus Access Procedure

The idle state of the bus is identified by eight or more consecutive ‘1’s. When a device starts transmission of a frame, the bus is recognized to be busy by the other devices at the moment the first ‘zero’ is transmitted (e.g. first ‘zero’ of the opening flag in HDLC mode).

After the frame has been transmitted, the bus becomes available again (idle).

Note: If the bus is occupied by other transmitters and/or there is no transmit request in the SCC, logical ‘1’ will be continuously transmitted on TxD.

3.2.9 Serial Bus Collisions and Recovery

During the transmission, the data transmitted on TxD is compared with the data on CxD. In case of a mismatch (‘1’ sent and ‘0’ detected, or vice versa) data transmission is immediately aborted, and idle (logical ‘1’) is transmitted.

HDLC/SDLC: Transmission will be initiated again by the SCC as soon as possible if the first part of the frame is still present in the SCC transmit FIFO. If not, an XMR interrupt is generated.

Since a ‘zero’ (‘low’) on the bus prevails over a ‘1’ (high impedance) if a wired-OR connection is implemented, and since the address fields of the HDLC frames sent by different stations normally differ from one another, the fact that a collision has occurred will be detected prior to or at the latest within the address field. The frame of the transmitter with the highest temporary priority (determined by the address field) is not affected and is transmitted successfully. All other stations cease transmission immediately and return to bus monitoring state.

Note: If a wired-OR connection has been realized by an external pull-up resistor without decoupling, the data output (TxD) can be used as an open drain output and connected directly to the CxD input.
For correct identification as to which frame is aborted and thus has to be repeated after an XMR interrupt has occurred, the contents of SCC transmit FIFO have to be unique, i.e. SCC transmit FIFO should not contain data of more than one frame. For this purpose new data may be provided to the transmit FIFO only after ’ALLS’ interrupt status is detected.

3.2.10 Serial Bus Access Priority Scheme

To ensure that all competing stations are given a fair access to the transmission medium, a two-stage bus access priority scheme is supported by PASSAT:

Once a station has successfully completed the transmission of a frame, it is given a lower level of priority. This priority mechanism is based on the requirement that a station may attempt transmitting only when a determined number of consecutive ‘1’s are detected on the bus.

Normally, a transmission can start when eight consecutive ‘1’s on the bus are detected (through pin CxD). When an HDLC frame has been successfully transmitted, the internal priority class is decreased. Thus, in order for the same station to be able to transmit another frame, ten consecutive ‘1’s on the bus must be detected. This guarantees that the transmission requests of other stations are satisfied before the same station is allowed a second bus access. When ten consecutive ‘1’s have been detected, transmission is allowed again and the priority class (of all stations) is increased (to eight ‘1’s).

Inside a priority class, the order of transmission (individual priority) is based on the HDLC address, as explained in the preceding paragraph. Thus, when a collision occurs, it is always the station transmitting the only ‘zero’ (i.e. all other stations transmit a ‘one’) in a bit position of the address field that wins, all other stations cease transmission immediately.

3.2.11 Serial Bus Configuration Timing Modes

If a bus configuration has been selected, the SCC provides two timing modes, differing in the time interval between sending data and evaluation of the transmitted data for collision detection.

- Timing mode 1 (CCR0H:SC(2:0) = ‘001’)
  Data is output with the rising edge of the transmit clock via the TxD pin, and evaluated 1/2 a clock period later at the CxD pin with the falling clock edge.
- Timing mode 2 (CCR0H:SC(2:0) = ‘011’)
  Data is output with the falling clock edge and evaluated with the next falling clock edge. Thus one complete clock period is available between data output and collision detection.

3.2.12 Functions Of Signal $\overline{\text{RTS}}$ in HDLC Mode

In clock modes 0 and 1, the $\overline{\text{RTS}}$ output can be programmed via register CCR1 (SOC bits) to be active when data (frame or character) is being transmitted. This signal is delayed by one clock period with respect to the data output TxD, and marks all data bits that could be transmitted without collision (see **Figure 3-24**). In this way a configuration may be implemented in which the bus access is resolved on a local basis (collision bus) and where the data are sent one clock period later on a separate transmission line.

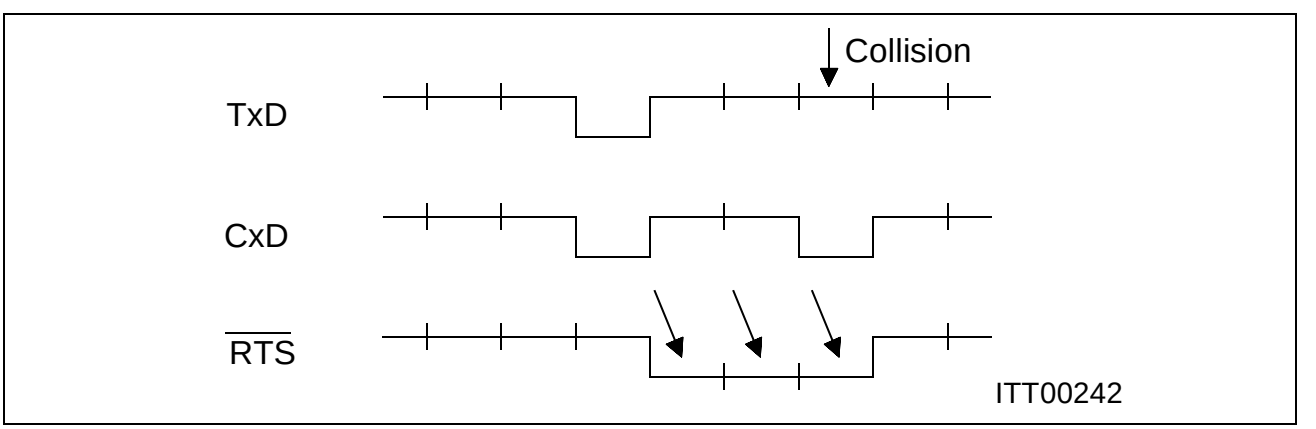


Figure 3-24 Request-to-Send in Bus Operation

Note: For details on the functions of the $\overline{\text{RTS}}$ pin refer to ***"Modem Control Signals (RTS, CTS, CD)" on page 3-74****.*

3.2.13 Data Encoding

The SCC supports the following coding schemes for serial data:

– Non-Return-To-Zero (NRZ)
– Non-Return-To-Zero-Inverted (NRZI)
– FM0 (also known as Bi-Phase Space)
– FM1 (also known as Bi-Phase Mark)

– Manchester (also known as Bi-Phase)

The desired line coding scheme can be selected via bit field ’SC(2:0)’ in register CCR0H.

3.2.13.1 NRZ and NRZI Encoding

NRZ: The signal level corresponds to the value of the data bit. By programming bit ’DIV’ (CCR1L register), the SCC may invert the transmission and reception of data.

NRZI: A logical ‘0’ is indicated by a transition and a logical ‘1’ by no transition at the beginning of the bit cell.

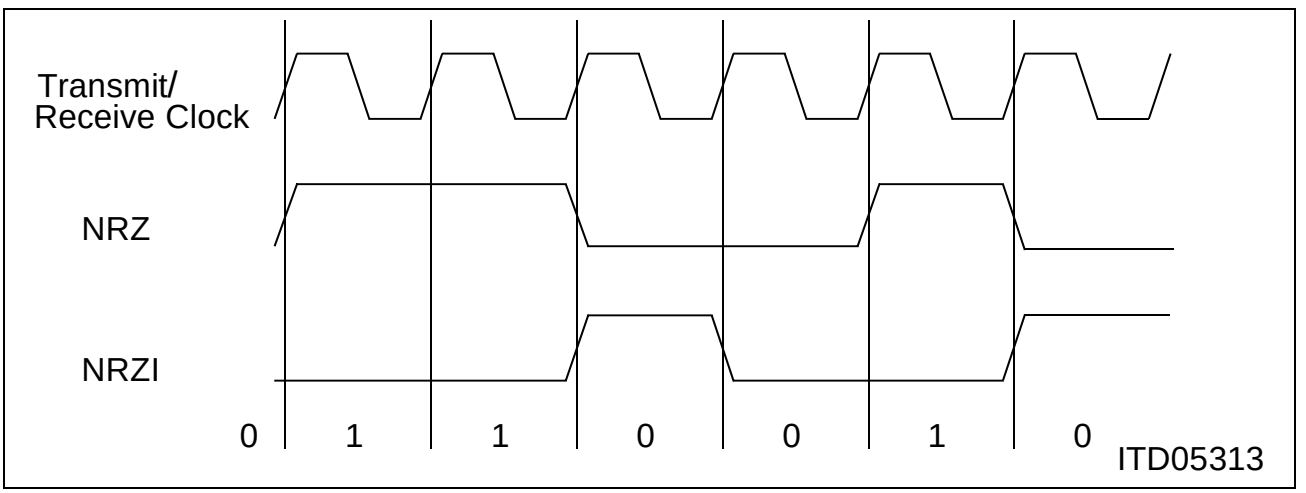


Figure 3-25 NRZ and NRZI Data Encoding

3.2.13.2 FM0 and FM1 Encoding

FM0: An edge occurs at the beginning of every bit cell. A logical ‘0’ has an additional edge in the center of the bit cell, whereas a logical ‘1’ has none. The transmit clock precedes the receive clock by 90°.

FM1: An edge occurs at the beginning of every bit cell. A logical ‘1’ has an additional edge in the center of the bit cell, a logical ‘0’ has none. The transmit clock precedes the receive clock by 90°.

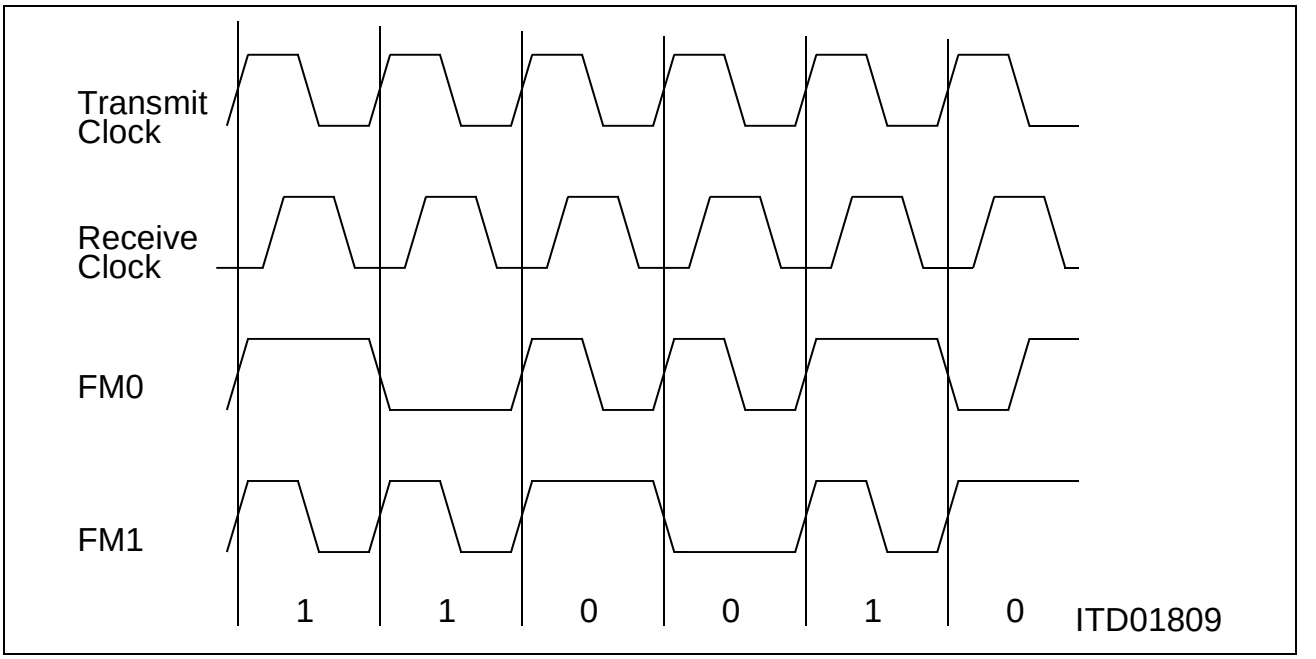


Figure 3-26 FM0 and FM1 Data Encoding

3.2.13.3 Manchester Encoding

Manchester: In the first half of the bit cell, the physical signal level corresponds to the logical value of the data bit. At the center of the bit cell this level is inverted. The transmit clock precedes the receive clock by 90°. The bit cell is shifted by 180° in comparison with FM coding.

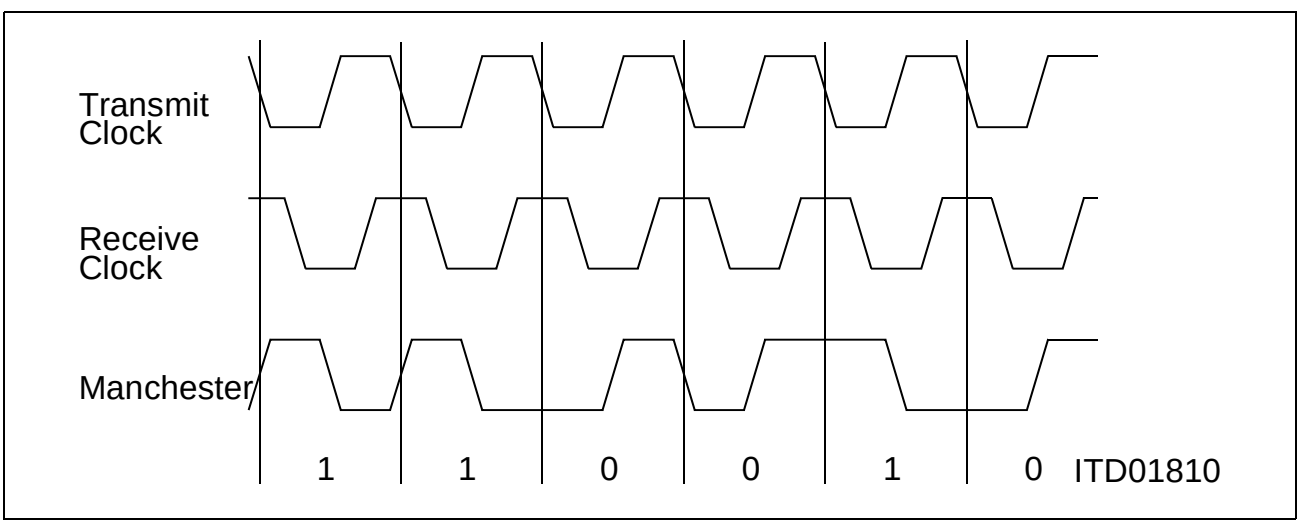


Figure 3-27 Manchester Data Encoding

3.2.14 Modem Control Signals ($\overline{RTS}$, $\overline{CTS}$, CD)

3.2.14.1 $\overline{RTS}$/$\overline{CTS}$ Handshaking

The SCC provides two pins ($\overline{RTS}$, $\overline{CTS}$) per serial channel supporting the standard request-to-send modem handshaking procedure for transmission control.

A transmit request will be indicated by outputting logical ‘0’ on the request-to-send output ($\overline{RTS}$). It is also possible to control the $\overline{RTS}$ output by software. After having received the permission to transmit ($\overline{CTS}$) the SCC starts data transmission.

In the case where permission to transmit is withdrawn in the course of transmission, the frame is aborted and IDLE is sent. After transmission is enabled again by re-activation of $\overline{CTS}$, and if the beginning of the frame is still available in the SCC, the frame will be re-transmitted (self-recovery). However, if the permission to transmit is withdrawn after the data available in the shadow part of the SCC transmit FIFO has been completely transmitted and the pool is released, the transmitter and the SCC transmit FIFO are reset, the $\overline{RTS}$ output is deactivated and an interrupt (XMR) is generated.

Note: For correct identification as to which frame is aborted and thus has to be repeated after an XMR interrupt has occurred, the contents of SCC transmit FIFO have to be unique, i.e. SCC transmit FIFO should not contain data of more than one frame, which could happen if transmission of a new frame is started by providing new data to the transmitter too early. For this purpose the ’All Sent’ interrupt (ISR1.ALLS) has to be waited for before providing new transmit data.

Note: In the case where permission to transmit is not required, the $\overline{CTS}$ input can be connected directly to V_{SS} and/or bit ’FCTS’ (register CCR1H) may be set to ’1’.

Additionally, any transition on the $\overline{CTS}$ input pin, sampled with the transmit clock, will generate an interrupt indicated via register ISR1, if this function is enabled by setting the ’CSC’ bit in register IMR1 to ’0’.

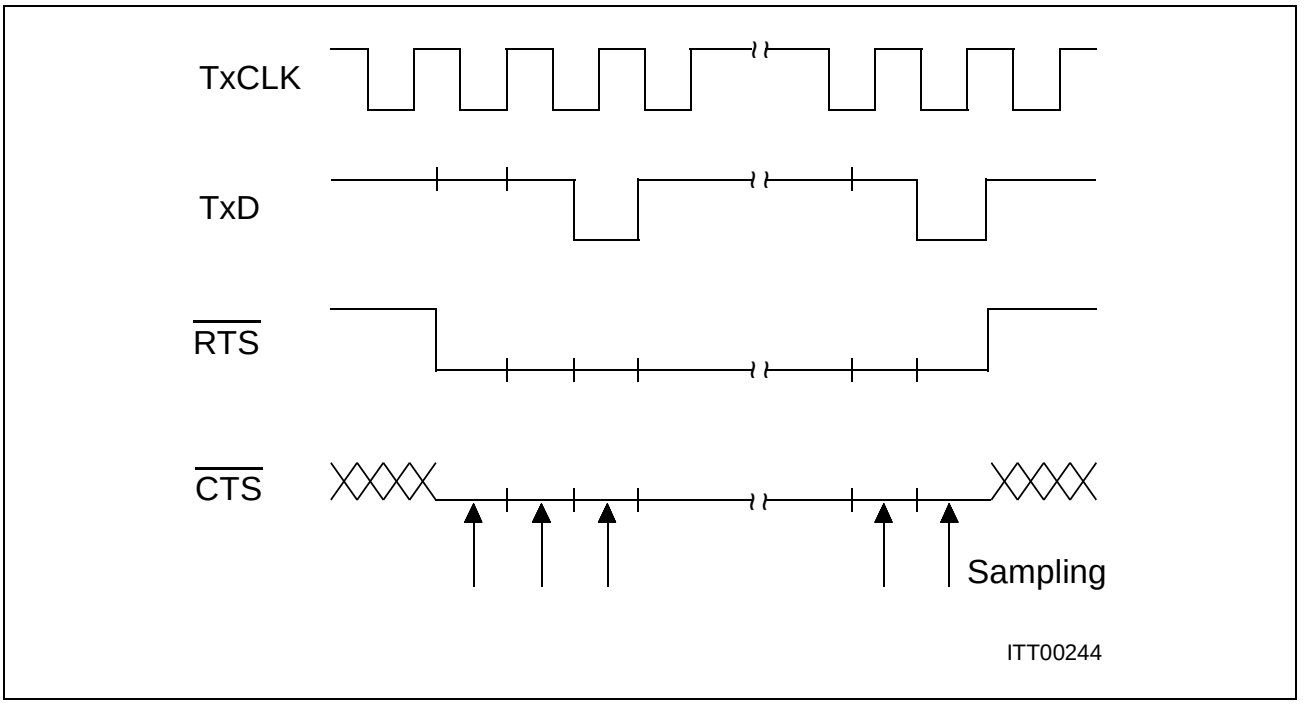


Figure 3-28 RTS/CTS Handshaking

Beyond this standard RTS function, signifying a transmission request of a frame (Request To Send), in HDLC mode the RTS output may be programmed for a special function via SOC1, SOC0 bits in the CCR1L register. This is only available if the serial channel is operating in a bus configuration mode in clock mode 0 or 1.

- If SOC1, SOC0 bits are set to '11', the RTS output is active (= low) during the reception of a frame.
- If SOC1, SOC0 bits are set to '10', the RTS output function is disabled and the RTS pin remains always high.

3.2.14.2 Carrier Detect (CD) Receiver Control

Similar to the RTS/CTS control for the transmitter, the SCC supports the carrier detect modem control function for the serial receiver if the Carrier Detect Auto Start (CAS) function is programmed by setting the 'CAS' bit in register CCR1H. This function is always available in clock modes 0, 2, 3, 6, 7 via the CD pin. In clock mode 1 the CD function is not supported. See **Table 3-2** for an overview.

If the CAS function is selected, the receiver is enabled and data reception is started when the CD input is detected to be high. If CD input is set to 'low', reception of the current character (byte) is still completed.

3.2.15 Local Loop Test Mode

To provide fast and efficient testing, the SCC can be operated in a test mode by setting the 'TLP' bit in register CCR2L. The on-chip serial data input and output signals (TxD,

RxD) are connected, generating a local loopback. As a result, the user can perform a self-test of the SCC.

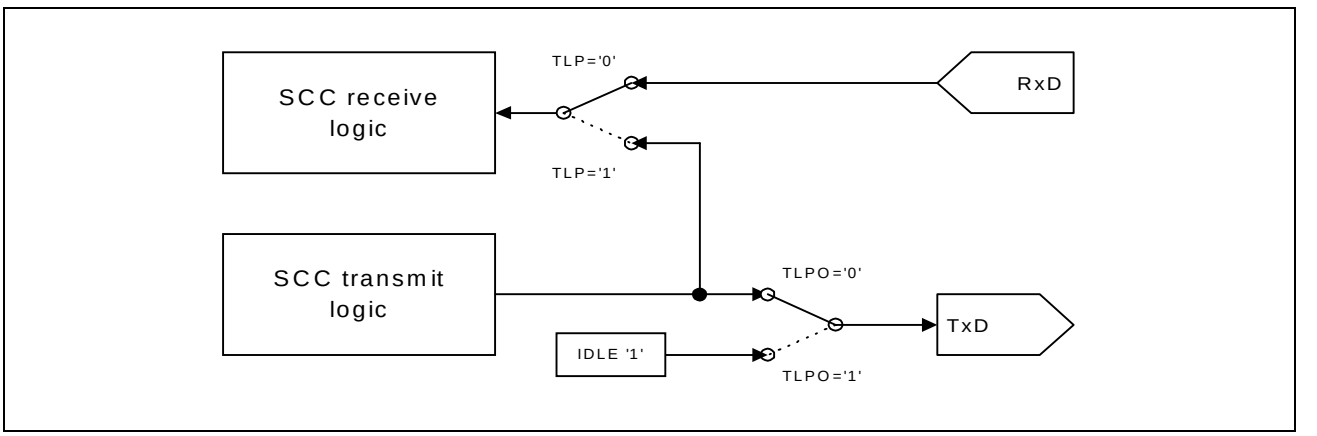


Figure 3-29 SCC Test Loop

Transmit data can be disconnected from pin TxD by setting bit TLPO in register CCR2L.

Note: A sufficient clock mode must be used for test loop operation such that receiver and transmitter operate with the same frequencies depending on the clock supply (e.g. clock mode 2b or 6b).

3.3 Microprocessor Interface

The communication between the CPU and PASSAT is done via a set of directly accessible registers. The interface may be configured as Intel or Motorola type (refer to description of pin 'BM') with a selectable data bus width of 8 or 16 bit (refer to description of pin 'WIDTH').

Note: For the PASSAT in P-LFBGA-80-2 package only an 8-bit wide bus interface is supported.

The CPU transfers data to/from PASSAT (via 64 byte deep FIFOs per direction and channel), sets the operating modes, controls function sequences, and gets status information by writing or reading control/status registers.

All accesses can be done as byte or word accesses if enabled. If 16-bit bus width is selected, access to the lower/upper part of the data bus is determined by signals $\overline{\text{BHE}}$/$\overline{\text{BLE}}$ as shown in **Table 3-4** (Intel mode) or by the upper and lower data strobe signals $\overline{\text{UDS}}$/$\overline{\text{LDS}}$ as shown in **Table 3-5** (Motorola mode).

Table 3-4 Data Bus Access 16-bit Intel Mode

$\overline{\text{BHE}}$	$\overline{\text{BLE}}$	Register Access	Data Pins Used
0	0	Word access (16 bit)	D(15:0)
0	1	Byte access (8 bit), odd address	D(15:8)
1	0	Byte access (8 bit), even address	D(7:0)
1	1	no data transfer	-

Table 3-5 Data Bus Access 16-bit Motorola Mode

$\overline{\text{UDS}}$	$\overline{\text{LDS}}$	Register Access	Data Pins Used
0	0	Word access (16 bit)	D(15:0)
0	1	Byte access (8 bit), even address	D(15:8)
1	0	Byte access (8 bit), odd address	D(7:0)
1	1	no data transfer	-

Each of the two serial channels of PASSAT is controlled via an identical, but completely independent register set (Channel A and B). Global functions that are common to or independent from the two serial channels are located in global registers.

3.4 External DMA Controller Support

The PASSAT comprises a 4-channel DMA interface for fast and effective data transfers using an external DMA controller. For both serial channels, a separate DMA Request output for Transmit (DRT) and Receive direction (DRR) as well as a DMA Acknowledgement input ($\overline{\text{DACK}}$) is provided.

The PASSAT activates the DRR/DRT line as long as data transfers are needed from/to the specific FIFO (level triggered demand transfer mode of DMA controller).

It is the responsibility of the DMA controller to perform the correct amount of bus cycles. Either read cycles will be performed if the DMA transfer has been requested from the receiver, or write cycles if DMA has been requested from the transmitter. If the DMA controller provides a DMA acknowledge signal ($\overline{\text{DACK}}$ pin, input to the PASSAT), each bus cycle implicitly selects the top of the specific FIFO and neither address (via A0..A7) nor chip select need to be supplied (I/O to Memory transfers). If no $\overline{\text{DACK}}$ signal is provided, normal read/write operations (providing addresses) must be performed (Memory to Memory transfers).

The PASSAT deactivates the DRR/DRT line immediately after the last read/write cycle of the data transfer has started.

3.5 Interrupt Architecture

For certain events in PASSAT an interrupt can be generated, requesting the CPU to read status information from PASSAT. The interrupt line INT/$\overline{\text{INT}}$ is asserted with the output characteristics programmed in bit field ’IPC(1..0)’ in register "GMODE" on page 5-109 (open drain/push pull, active low/high).

Since only one interrupt request output is provided, the cause of an interrupt must be determined by the CPU by reading the interrupt status registers (GSTAR, ISR0, ISR1, ISR2, DISR, GPISL/GPISH).

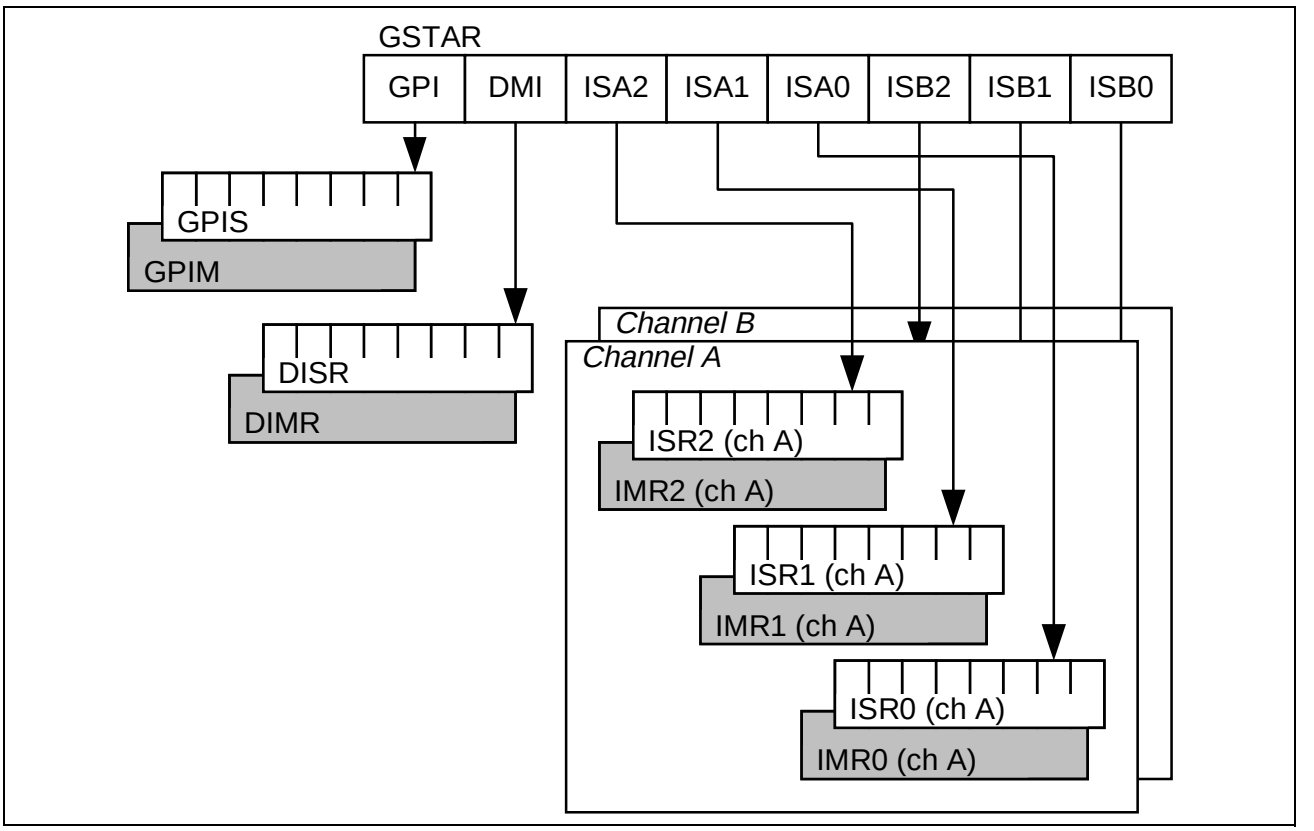


Figure 3-30 Interrupt Status Registers

Each interrupt indication of registers ISR0, ISR1, ISR2, DISR and GPISL/GPISH can be selectively unmasked by resetting the corresponding bit in the corresponding mask registers IMR0, IMR1, IMR2, DIMR and GPIML/GPIMH. Use of these registers depends on the selected serial mode.

If bit ’VIS’ in register CCR0L is set to ’1’, masked interrupt status bits are visible in the interrupt status registers ISR0..ISR2. Interrupts masked in registers IMR0..IMR2 will not generate an interrupt though. A read access to the interrupt status registers clears the bits.

A global interrupt mask bit (bit ’GIM’ in register GMODE) suppresses interrupt generation at all. To enable the interrupt system after reset, this bit must be set to ’0’.

The Global Interrupt Status Register (GSTAR) serves as pointer to pending channel related interrupts and general purpose port interrupts.

3.6 General Purpose Port Pins

3.6.1 GPP Functional Description

General purpose port pins are provided on pins GP6, GP8, GP9 and GP10 in P-TQFP-100-3 package (not provided in P-LFBGA-80-2 package). If external DMA support is not enabled, pins GP0...GP2 are available as general purpose pins (in both P-TQFP-100-3 and P-LFBGA-80-2 package).

Every pin is separately programmable via the General Purpose Port Direction registers GPDIRL/GPDIRH to operate as an output (bit GPnDIR=’0’) or as an input (bit GPnDIR=’1’, reset value).

If defined as output, the state of the pin is directly controlled via the General Purpose Port Data registers GPDATL/GPDATH. Read access to these registers delivers the current state of all GPP pins (input and output signals).

If defined as input, the state of the pin is monitored. The signal state of the corresponding GP pins is sampled with a rising edge of CLK and is readable via registers GPDATL/ GPDATH.

3.6.2 GPP Interrupt Indication

The GPP block generates interrupts for transitions on each input signal. All changes may be indicated via interrupt (optional). To enable interrupt generation, the corresponding interrupt mask bit in registers GPIML/GPIMH must be reset to ’0’.

Bit PI in the gloabl interrupt status register (GSTAR) is set to ’1’ if an interrupt was generated by any one or more of the the general purpose port pins. The GPP pin causing the interrupt can be located by reading the GPISL/GPISH registers.

4 Detailed Protocol Description

The following **Table 4-1** provides an overview of all supported protocol modes and . The desired protocol mode is selected via bit fields in the channel configuration registers CCR2L and CCR3L.

Table 4-1 Protocol Mode Overview

Protocol Mode		Register CCR2L - Bit Field:			CCR3L
		MDS	ADM	PPPM	ESS7
HDLC Automode (LAP D / LAP B / SDLC-NRM)	16 bit	’00’	’1’	’00’	’0’
	8 bit	’00’	’0’		
HDLC Address Mode 2	16 bit	’01’	’1’		
	8 bit	’01’	’0’		
HDLC Address Mode 1		’10’	’1’		
HDLC Address Mode 0		’10’	’0’		
Signaling System #7 (SS7) Operation		’10’	’0’	’00’	’1’
Bit Synchronous PPP Mode		’10’	’0’	’11’	’0’
Octet Synchronous PPP Mode				’01’	
Extended Transparent Mode[1)]		’11’	’1’	’00’	’0’

1) Extended transparent mode is a fully bit-transparent transmission/reception mode.

All modes are discussed in details in this chapter.

4.1 HDLC/SDLC Protocol Modes

The HDLC controller of each serial channel (SCC) can be programmed to operate in various modes, which are different in the treatment of the HDLC frame in receive direction. Thus, the receive data flow and the address recognition features can be performed in a very flexible way satisfying almost any application specific requirements.

There are 4 different HDLC operating modes which can be selected via register CCR2L.

The following table provides an overview of the different address comparison mechanisms in HDLC operating modes:

Table 4-2 Address Comparison Overview

Mode	Address Field	Recognized Address Bytes for a Match: High Address Byte		Low Address Byte
Address Mode 2 - Auto Mode	16 bit	FE_H / FC_H (1111 11 C/R 0_2)	***and***	RAL1
		FE_H / FC_H (1111 11 C/R 0_2)	***and***	RAL2
		RAH1	***and***	RAL1
		RAH2	***and***	RAL2
	8 bit	RAL1		*don't care*
		RAL2		*don't care*
Address Mode 1	8 bit	FE_H / FC_H (1111 11 C/R 0_2)		*don't care*
		RAH1		*don't care*
		RAH2		*don't care*
Address Mode 0	None	*don't care*		*don't care*

4.1.0.1 Automode

Characteristics: Window size 1, random message length, address recognition.

The SCC processes autonomously all numbered frames (S-, I-frames) of an HDLC protocol. The HDLC control field, I-field data of the frames and an additional status byte are temporarily stored in the SCC receive FIFO.

Depending on the selected address mode, the SCC can perform a 2-byte or 1-byte address recognition.

If a 2-byte address field is selected, the high address byte is compared with the fixed value FE_H or FC_H (group address) as well as with two individually programmable values in RAH1 and RAH2 registers. According to the ISDN LAPD protocol, bit 1 of the high byte address will be interpreted as COMMAND/RESPONSE bit (C/R), depending on the setting of the CRI bit in RAH1, and will be excluded from the address comparison.

Similarly, two comparison values can be programmed in special registers (RAL1, RAL2) for the low address byte. A valid address will be recognized in case the high and low byte of the address field correspond to one of the compare values. Thus, the SCC can be called (addressed) with 6 different address combinations, however, only the logical connection identified through the address combination RAH1/RAL1 will be processed in the auto-mode, all others in the non auto-mode. HDLC frames with address fields that do not match any of the address combinations, are ignored by the SCC.

In the case of a 1-byte address, only RAL1 and RAL2 will be used as comparison values. According to the X.25 LAPB protocol, the value in RAL1 will be interpreted as COMMAND and the value in RAL2 as RESPONSE.

The address bytes can be masked to allow selective broadcast frame recognition. For further information see **"Receive Address Handling" on page 4-85**.

4.1.0.2 Address Mode 2

Characteristics: address recognition, arbitrary window size.

All frames with valid addresses (address recognition identical to auto-mode) are forwarded directly to the RFIFO.

The HDLC control field, I-field data and an additional status byte are temporarily stored in the SCC receive FIFO.

In address mode 2, all frames with a valid address are treated similarly.

The address bytes can be masked to allow selective broadcast frame recognition.

4.1.0.3 Address Mode 1

Characteristics: address recognition high byte.

Only the high byte of a 2-byte address field will be compared. The address byte is compared with the fixed value FE_H or FC_H (group address) as well as with two individually programmable values RAH1 and RAH2. The whole frame excluding the first address byte will be stored in the SCC receive FIFO.

The address bytes can be masked to allow selective broadcast frame recognition.

4.1.0.4 Address Mode 0

Characteristics: no address recognitio

No address recognition is performed and each complete frame will be stored in the SCC receive FIFO.

4.1.1 HDLC Receive Data Processing

The following figures give an overview about the management of the received frames in the different HDLC operating modes. The graphics show the actual HDLC frame and how PASSAT interprets the incoming octets. Below that it is shown which octets are stored in the RFIFO and will thus be transferred into memory.

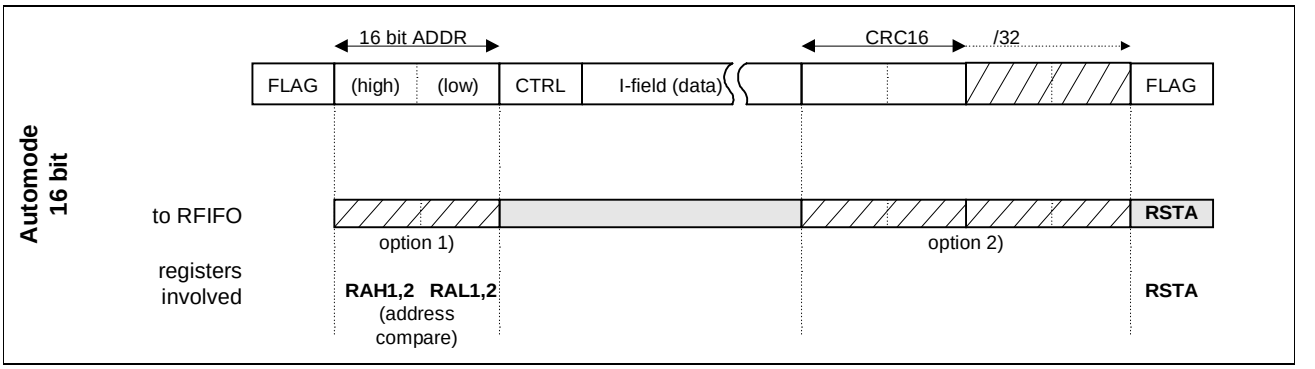


Figure 4-1 HDLC Receive Data Processing in 16 bit Automode

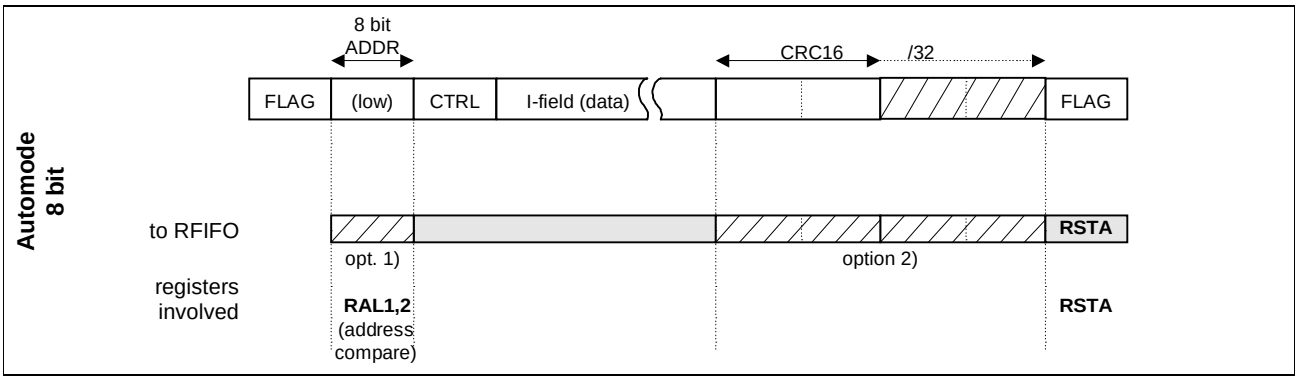


Figure 4-2 HDLC Receive Data Processing in 8 bit Automode

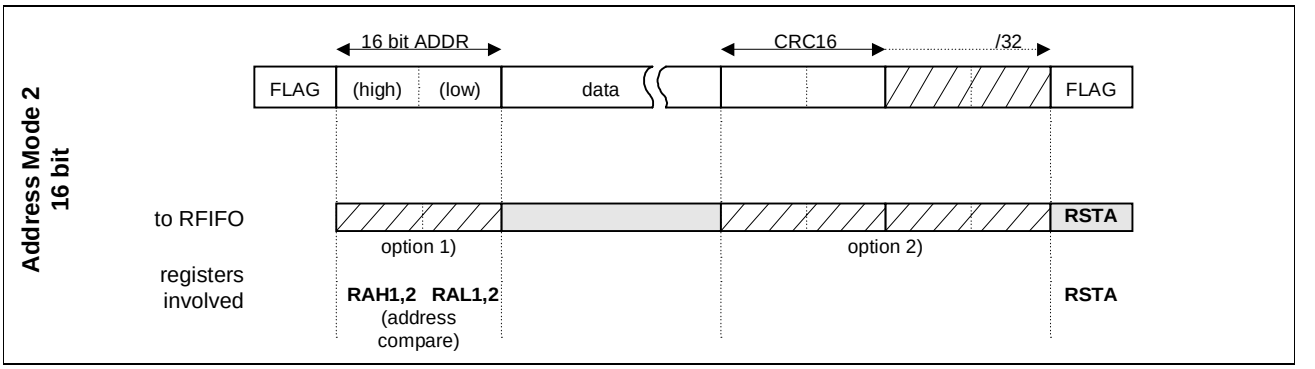


Figure 4-3 HDLC Receive Data Processing in Address Mode 2 (16 bit)

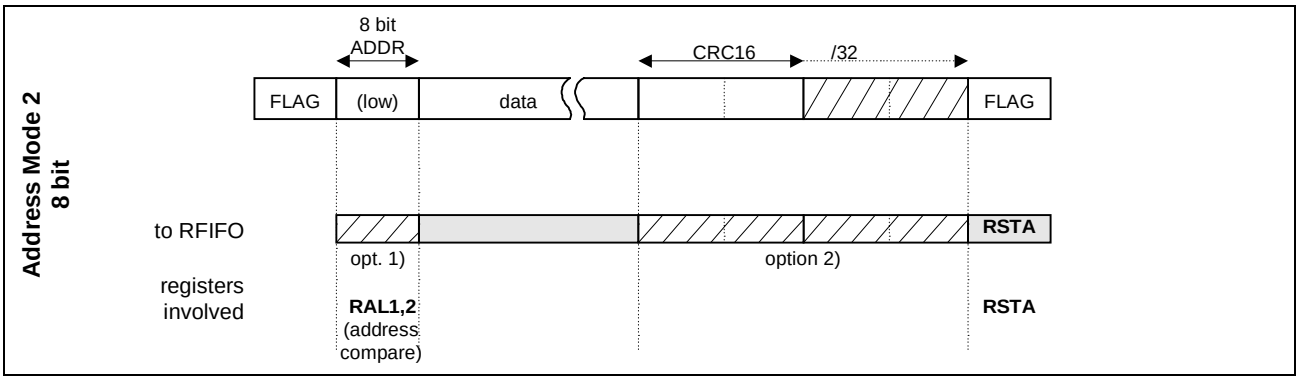


Figure 4-4 HDLC Receive Data Processing in Address Mode 2 (8 bit)

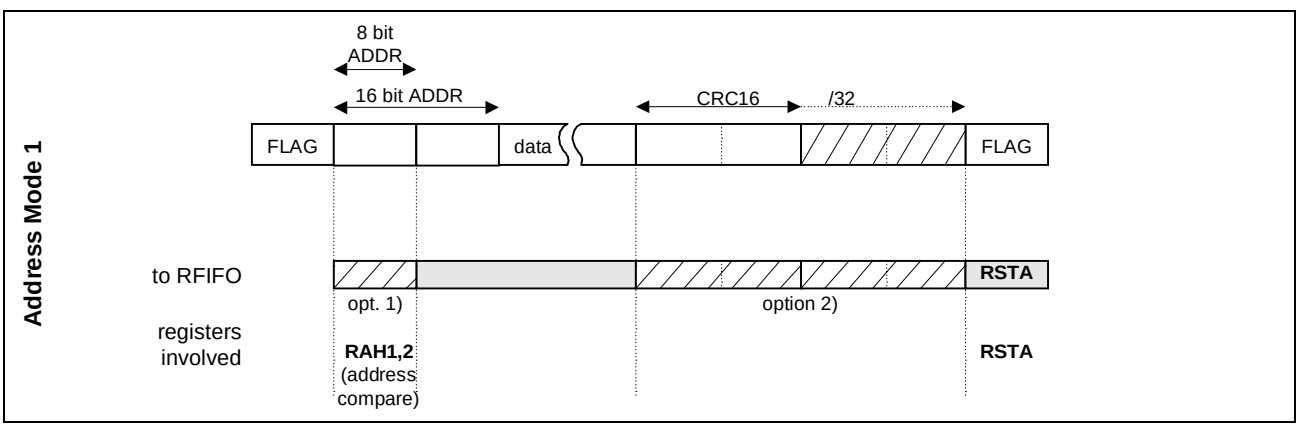


Figure 4-5 HDLC Receive Data Processing in Address Mode 1

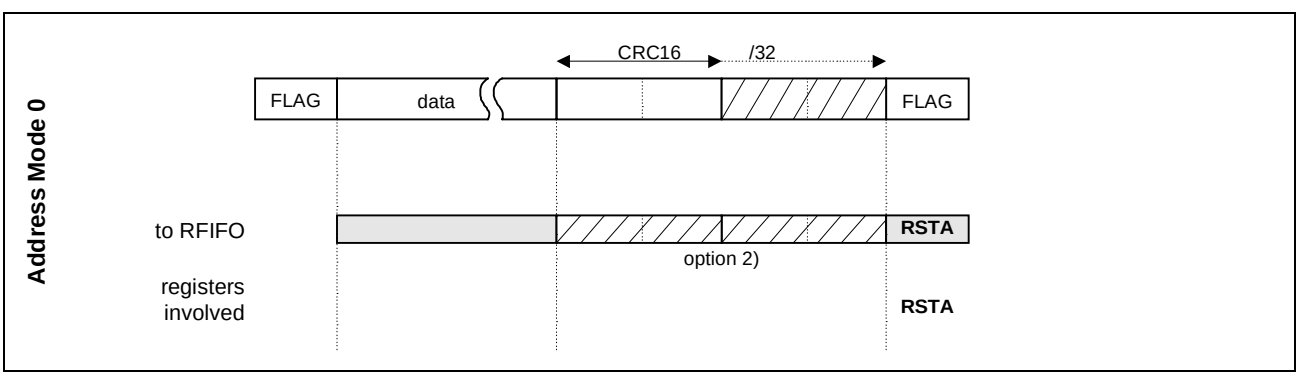


Figure 4-6 HDLC Receive Data Processing in Address Mode 0

option 1)

The address field (8 bit address, 16 bit address or the high byte of a 16 bit address) can optionally be forwarded to the RFIFO (bit 'RADD' in register CCR3H)

option 2)

The 16 bit or 32 bit CRC field can optionally be forwarded to the RFIFO (bit 'RCRC' in register CCR3H)

4.1.2 Receive Address Handling

The Receive Address Low/High Bytes (registers RAL1/RAH1 and RAL2/RAH2) can be masked on a per bit basis by setting the corresponding bits in the mask registers AMRAL1/AMRAH1 and AMRAL2/AMRAH2. This allows extended broadcast address recognition. Masked bit positions always match in comparison of the received frame address with the respective address fields in the Receive Address Low/High registers.

This feature is applicable to all HDLC protocol modes with address recognition (auto mode, address mode 2 and address mode 1). It is disabled if all bits of mask bit fields AMRAL1/AMRAH1 and AMRAL2/AMRAH2 are set to ‘zero’ (which is the reset value).

Detection of the fixed group address FE_H or FC_H, if applicable to the selected operating mode, remains unchanged.

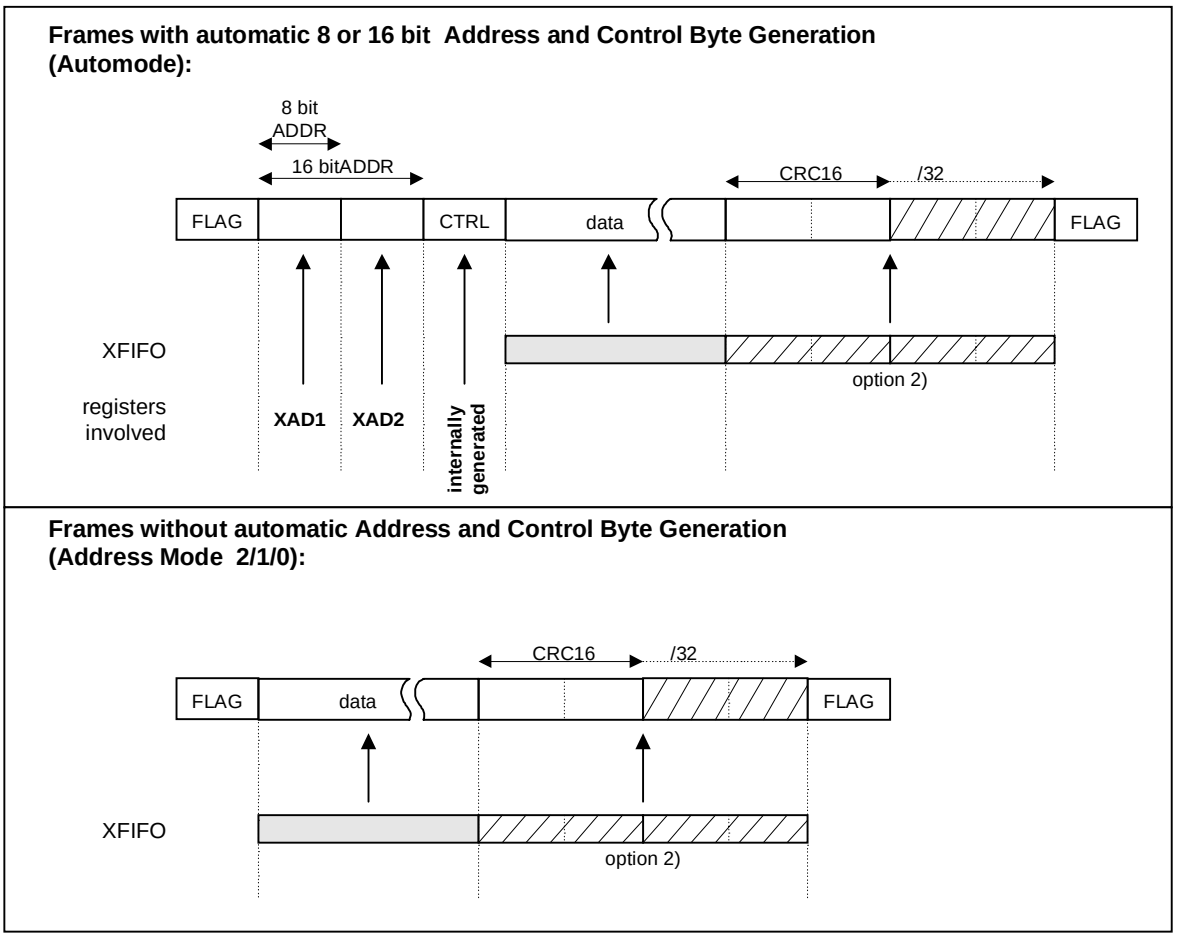
As an option in the auto mode, address mode 2 and address mode 1, the 8/16 bit address field of received frames can be pushed to the receive data buffer (first one/two bytes of the frame). This function is especially useful in conjunction with the extended broadcast address recognition. It is enabled by setting control bit ’RADD’ in register CCR3H.

4.1.3 HDLC Transmit Data Processing

Two different types of frames can be transmitted:

– I-frames and
– transparent frames

as shown below.



option 2)
Generation of the 16 or 32 bit CRC field can optionally be disabled by setting bit 'XCRC' in register CCR2H, in which case the CRC must be calculated and written into the last 2 or 4 bytes of the transmit FIFO, to immediately proceed closing flag.

Figure 4-7 SCC Transmit Data Flow (HDLC Modes)

For transmission of I-frames (selected via transmit command ’XIF’ in register CMDRL), the address and control fields are generated autonomously by the SCC and the data in the corresponding transmit data buffer is entered into the information field of the frame. This is possible only if the SCC is operated in Automode.

For (address-) transparent frames, the address and the control fields have to be entered in the transmit data buffer by software. This is possible in all operating modes and used also in auto-mode for sending U-frames.

If bit ’XCRC’ in register CCR2H is set, the CRC checksum will not be generated internally. The checksum has to be provided via the transmit data buffer as the last two or four bytes by software. The transmitted frame will be closed automatically only with a (closing) flag.

Note: The SCC does not check whether the length of the frame, i.e. the number of bytes, to be transmitted makes sense according the HDLC protocol or not.

4.1.4 Shared Flags

If the ‘Shared Flag’ feature is enabled by setting bit ’SFLG’ in register CCR1L the closing flag of a previously transmitted frame simultaneously becomes the opening flag of the following frame if there is one already available in the SCC transmit FIFO.

In receive direction the SCC always expects and handles ’Shared Flags’. ’Shared Zeroes’ of consecutive flags are also supported.

4.1.5 One Bit Insertion

Similar to the zero bit insertion (bit stuffing) mechanism, as defined by the HDLC protocol, the SCC offers a feature of inserting/deleting a ’one’ after seven consecutive ‘zeros’ into the transmit/receive data stream, if the serial channel is operating in bus configuration mode. This method is useful if clock recovery is performed by DPLL.

Since only NRZ data encoding is supported in a bus configuration, there are possibly long sequences without edges in the receive data stream in case of successive ‘0’s received, and the DPLL may lose synchronization.

Enabling the one bit insertion feature by setting bit ’OIN’ in register CCR2H, it is guaranteed that at least after

– 5 consecutive ‘1’s a ‘0’ will appear (bit stuffing), and after
– 7 consecutive ‘0’s a ‘1’ will appear (one insertion)

and thus a correct function of the DPLL is ensured.

Note: As with the bit stuffing, the ‘one insertion’ is fully transparent to the user, but it is not in accordance with the HDLC protocol, i.e. it can only be applied in proprietary systems using circuits that also implement this function, such as the PEB 20542 and PEB 20532.

4.1.6 Preamble Transmission

If enabled via bit ’EPT’ in register CCR2H, a programmable 8-bit pattern is transmitted with a selectable number of repetitions after Interframe Timefill transmission is stopped and a new frame is ready to be sent out. The 8 bit preamble pattern can be programmed in register PREAMB and the repetition time in bit field ’PRE’ of register CCR2H.

Note: Zero Bit Insertion is disabled during preamble transmission.

4.1.7 CRC Generation and Checking

In HDLC/SDLC mode, error protection is done by CRC generation and checking.

In standard applications, CRC-CCITT algorithm is used. The Frame Check Sequence at the end of each frame consists of two bytes of CRC checksum.

If required, the CRC-CCITT algorithm can be replaced by the CRC-32 algorithm, enabled via bit 'C32' in register CCR1L. In this case the Frame Check Sequence consists of four bytes.

Optionally the internal handling of received and transmitted CRC checksum can be influenced via control bits 'RCRC', 'DRCRC' in register CCR3H and 'XCRC' in register CCR2H.

Receive direction:
If not disabled by setting bit 'DRCRC' (register CCR3H), the received CRC checksum is always assumed to be in the 2 (CRC-CCITT) or 4 (CRC-32) last bytes of a frame, immediately preceding a closing flag. If bit 'RCRC' is set, the received CRC checksum is treated as data and will be forwarded to the RFIFO, where it precedes the frame status byte. Nevertheless the received CRC checksum is additionally checked for correctness. If CRC checking is disabled with bit CCR3H:DRCRC, the limits for 'Valid Frame' check are modified accordingly (refer to description of the Receive Status Byte, RSTA:VFR).

Transmit direction:
If bit 'XCRC' is set, the CRC checksum is not generated internally. The checksum has to be provided via the transmit data buffer by software. The transmitted frame will only be closed automatically with a (closing) flag.

Note: The SCC does not check whether the length of the frame, i.e. the number of bytes, to be transmitted makes sense or not according the HDLC protocol.

4.1.8 Receive Length Check Feature

The SCC offers the possibility to supervise the maximum length of received frames and to terminate data reception in the case that this length is exceeded.

This feature is controlled via the special Receive Length Check Registers RLCRL/ RLCRH.

The function is enabled by setting bit 'RCE' (Receive Length Check Enable) and the maximum frame length to be checked is programmed via bit field 'RL'. The maximum receive length can be determined as a multiple of 32-byte blocks as follows:

MAX_LENGTH = (RL + 1) $\times$ 32 ,

where RL is the value written to bit field 'RL'. Thus, the maximum length of receive frames can be programmed between 32 and 65536 bytes.

All frames exceeding this length are treated as if they had been aborted by the remote station, i.e. the CPU is informed via

– an 'RME' interrupt generated by the SCC, and
– the receive abort indication 'RAB' in the Receive Status Byte (RSTA).

Additionally an optional ’FLEX’ interrupt is generated prior to ’RME’, indicating that the maximum receive frame length was exceeded.

Receive operation continues with the beginning of the next receive frame.

4.2 Point-to-Point Protocol (PPP) Modes

PPP (as described in RFC1662) can work over 3 modes: asynchronous HDLC, synchronous HDLC, and octet synchronous. The PASSAT supports bit and octet synchronous HDLC PPP for use over dial-up connections. The octet synchronous mode of PPP protocol (RFC 1662) supports PPP over SONET applications.

The synchronous HDLC PPP modes are submodes of the HDLC mode. The appropriate PPP mode is selected via bit field ’PPPM’ in register CCR2L.

The PPP-support hardware allows software to perform segmentation and reassembly of PPP payloads, and allows PASSAT to perform the synchronous HDLC PPP protocol conversions as required for the network interface.

4.2.1 Bit Synchronous PPP

The PASSAT transmits a data block, inserts HDLC Header (Opening Flag), and appends the HDLC Trailer (CRC, Ending Flag). Zero-bit stuffing algorithm is also performed. No character mapping is performed. The bit-synchronous PPP mode differs from the HDLC mode (address mode 0) only in the abort sequence:

HDLC requires at least 7 consecutive ’1’ bits as abort sequence, whereas PPP requires at least 15 ’1’ bits.

For receive operation PASSAT monitors the incoming data stream for the Opening Flag (7E Hex) to identify the beginning of a HDLC packet. Subsequent bytes are part of data and are processed as normal HDLC packet including checking of CRC.

4.2.2 Octet Synchronous PPP

The PASSAT transmits a data block, inserts HDLC Header (Opening Flag), and appends the HDLC Trailer (CRC, Ending Flag). Beside this standard HDLC operation, zero-bit stuffing is not performed, but character mapping is performed.

For receive operation PASSAT monitors the incoming data stream for the Opening Flag (7E Hex) to identify the beginning of a HDLC packet. Subsequent bytes are part of data and are processed as normal HDLC packet including checking of CRC. Received mapped characters are unmapped.

4.2.3 Data Transparency in PPP Mode

When transporting bit-files (as opposed to text files), or compressed files, the characters could easily represent MODEM control characters (such as CTRL-Q, CTRL-S) which the

MODEM would not pass through. PASSAT maintains an Async Control Character Map (ACCM) for characters 00-1F Hex. Whenever there is a mapped character in the data stream, the transmitter precedes that character with a control-escape character of $7D_H$. After the control-escape, the character itself is transmitted with bit 5 inverted. character e.g. 13_H is mapped to $7D_H$, 33_H).

At the receive end, a $7D_H$ character is discarded and the following character is modified by inverting bit 5 (e.g. if $7D_H$, 33_H is received, the $7D_H$ is discarded and the 33_H is changed to 13_H the original character).

The 32 lookup octet values (00_H-$1F_H$) are stored within the on-chip registers ACCM0..3.

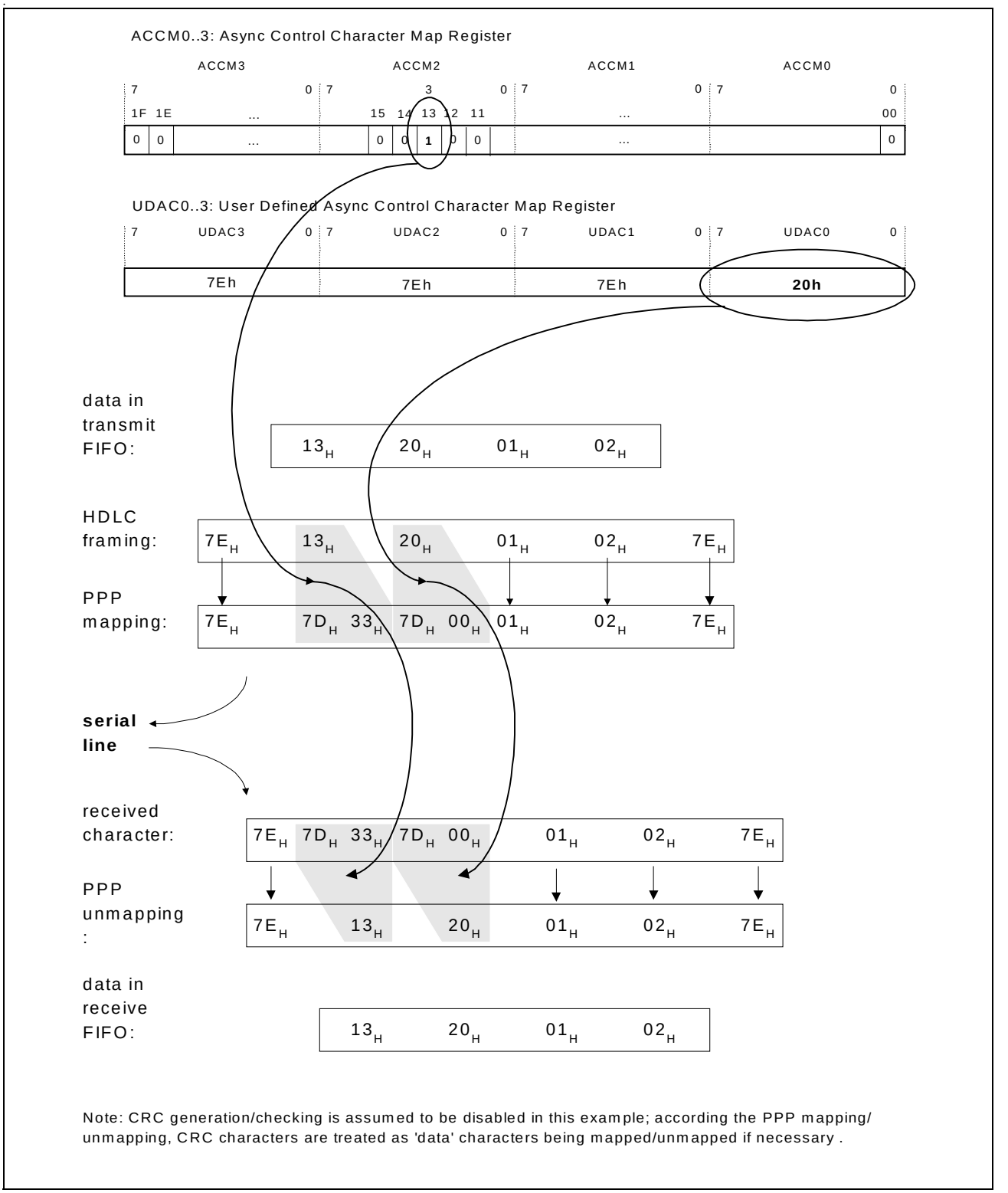
In addition to the ACCM, 4 user programmable characters (especially outside the range 00-1F Hex) can also be mapped using the control-escape sequence described above. These characters are specified in registers UDAC0..3.

The receiver discards all characters which are received unmapped, but expected to be mapped because of ACCM0..3 and UDAC0..3 register contents. If this occurs within an HDLC frame, the unexpected characters are discarded before forwarded to the receive CRC checking unit.

$7D_H$ (control-escape) and $7E_H$ (flag) octets in the data stream are mapped in general. The sequence of mapping control logic is:

1. $7D_H$ and $7E_H$ octets,
2. ACCM0..3,
3. UDAC0..3.

This mechanism is applied to octet synchronous HDLC PPP mode.



Note: CRC generation/checking is assumed to be disabled in this example; according the PPP mapping/ unmapping, CRC characters are treated as 'data' characters being mapped/unmapped if necessary .

Figure 4-8 PPP Mapping/Unmapping Example

4.3 Extended Transparent Mode

Characteristics: fully transparent

When programmed in the extended transparent mode via the CCR2L register (bits MDS1, MDS0, ADM = ‘111’), the SCC performs fully transparent data transmission and reception without HDLC framing, i.e. without

- FLAG insertion and deletion
- CRC generation and checking
- bit stuffing.

This feature can be profitably used e.g. for:

- user specific protocol variations
- line state monitoring, or
- test purposes, in particular for monitoring or intentionally generating HDLC protocol rule violations (e.g. wrong CRC)

Character or octet boundary synchronization can be achieved by using clock mode 5 or clock mode 1 with an external receive strobe input to pin CD.

Setting invokes this out-of-band flow control for the receiver. When the shadow part of the receive FIFO has reached a set threshold of 28 bytes, the signal is forced inactive (high). When the shadow part of the receive FIFO is empty, the is re-asserted (low). Note that the data is immediately transferred from the shadow receive FIFO to the user accessible RFIFO (as long as there is space available). So when the shadow receive FIFO reaches the 28 bytes threshold, there is 4 more byte storage available before overflow can occur. This allows sufficient time for the far end transmitter to react to the change in the signal and stop sending more data.

A transmit data underrun condition in the XFIFO is indicated with an ’XDU’ interrupt. Nevertheless, transmission continues inserting SYN characters into the data stream until new data is available in the transmit FIFO. Inserted SYN characters are not part of the frame and thus not used for CRC calculation.

4.4 Procedural Support (Layer-2 Functions)

When operating in the auto mode, the SCC offers a high degree of protocol support. In addition to address recognition, the SCC autonomously processes all (numbered) S- and I-frames (window size 1 only) with either normal or extended control field format (modulo-8 or modulo-128 sequence numbers – selectable via register CCR2H bit ’MCS’).

The following functions will be performed:

– updating of transmit and receive counter
– evaluation of transmit and receive counter
– processing of S commands

– flow control with RR/RNR
– generation of responses
– recognition of protocol errors
– transmission of S commands, if acknowledgement is not received
– continuous status query of remote station after RNR has been received
– programmable timer/repeater functions.

In addition, all unnumbered frames are forwarded directly to the processor. The logical link can be initialized by software at any time (Reset HDLC Receiver by RRES command in register CMDRH).
Additional logical connections can be operated in parallel by software.

4.4.1 Full-Duplex LAPB/LAPD Operation

Initially (i.e. after RESET), the LAP controllers of the two serial channels are configured to function as a combined (primary/secondary) station, where they autonomously perform a subset of the balanced X.25 LAPB/ISDN LAPD protocol.

Reception of Frames:

The logical processing of received S-frames is performed by the SCC without interrupting the host. The host is merely informed by interrupt of status changes in the remote station (receiver ready / receiver not ready) and protocol errors (unacceptable N(R), or S-frame with I-field).

I-frames are also processed autonomously and checked for protocol errors. The I-frame will not be accepted in the case of sequence errors (no interrupt is forwarded to the host), but is immediately confirmed by an S-response. If the host sets the SCC into a ‘receive not ready’ status, an I-frame will not be accepted (no interrupt) and an RNR response is transmitted. U-frames are always stored in the RFIFO and forwarded directly to the host. The logical sequence and the reception of a frame in auto mode is illustrated in **Figure 4-9**.

Note: The state variables N(S), N(R) are evaluated within the window size 1, i.e. the SCC checks only the least significant bit of the receive and transmit counter regardless of the selected modulo count.

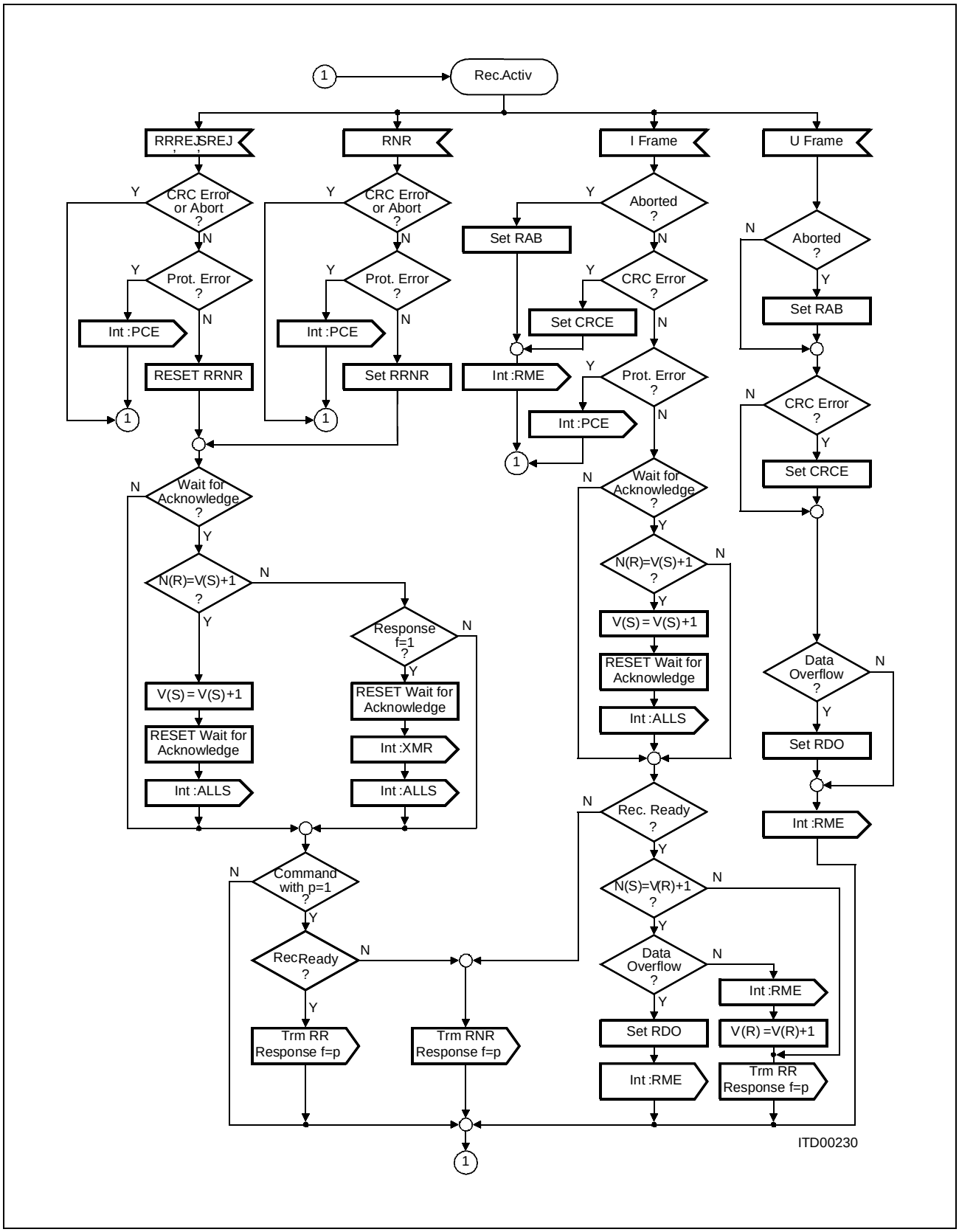


Figure 4-9 Processing of Received Frames in Auto Mode

Transmission of Frames:

The SCC autonomously transmits S commands and S responses in the auto mode. Either transparent or I-frames can be transmitted by the user. The software timer has to be operated in the internal timer mode to transmit I-frames. After the frame has been transmitted, the timer is self-started, the XFIFO is inhibited, and the SCC waits for the arrival of a positive acknowledgement. This acknowledgement can be provided by means of an S- or I-frame.

If no positive acknowledgement is received during time t_1, the SCC transmits an S-command (p = '1'), which must be answered by an S-response (f = '1'). If the S-response is not received, the process is performed n1 times (in HDLC known as N2, refer to register TIMR3).

Upon the arrival of an acknowledgement or after the completion of this poll procedure the XFIFO is enabled and an interrupt is generated. Interrupts may be triggered by the following:

- message has been positively acknowledged (ALLS interrupt)
- message must be repeated (XMR interrupt)
- response has not been received (TIN interrupt).

In automode, only when the ALLS interrupt has been issued data of a new frame may be provided to the XFIFO!

Upon arrival of an RNR frame, the software timer is started and the status of the remote station is polled periodically after expiration of t_1, until the status 'receive ready' has been detected. The user is informed via the appropriate interrupt. If no response is received after n1 times, a TIN interrupt, and t_1 clock periods thereafter an ALLS interrupt is generated and the process is terminated.

Note: The internal timer mode should only be used in the auto mode.

Transparent frames can be transmitted in all operating modes.

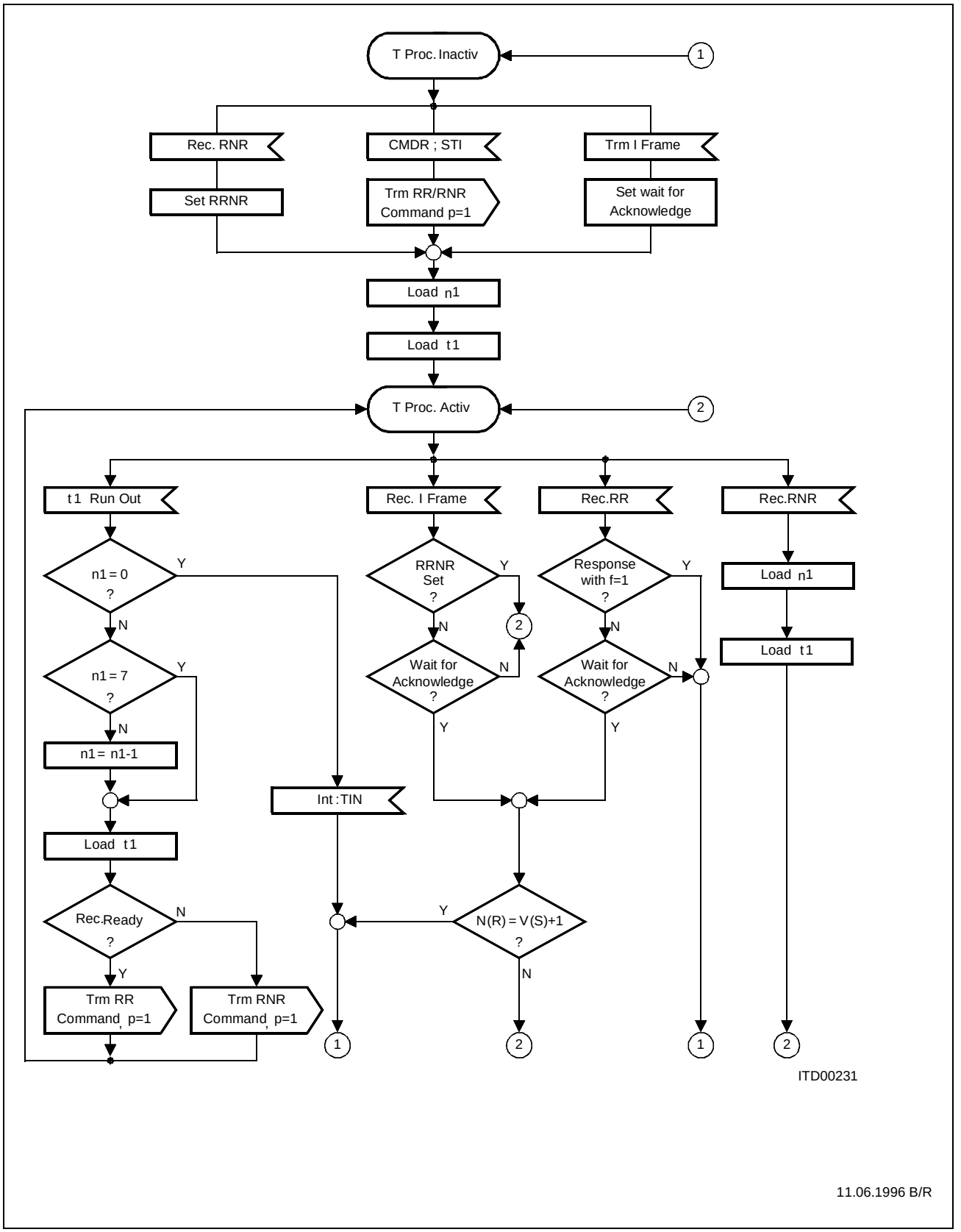


Figure 4-10 Timer Procedure/Poll Cycle

Examples

The interaction between SCC and the host during transmission and reception of I-frames is illustrated in the following two figures. The flow control with RR/RNR of I-frames during transmission/reception is illustrated in **Figure 4-11**. Both, the sequence of the poll cycle and protocol errors are shown in **Figure 4-12**.

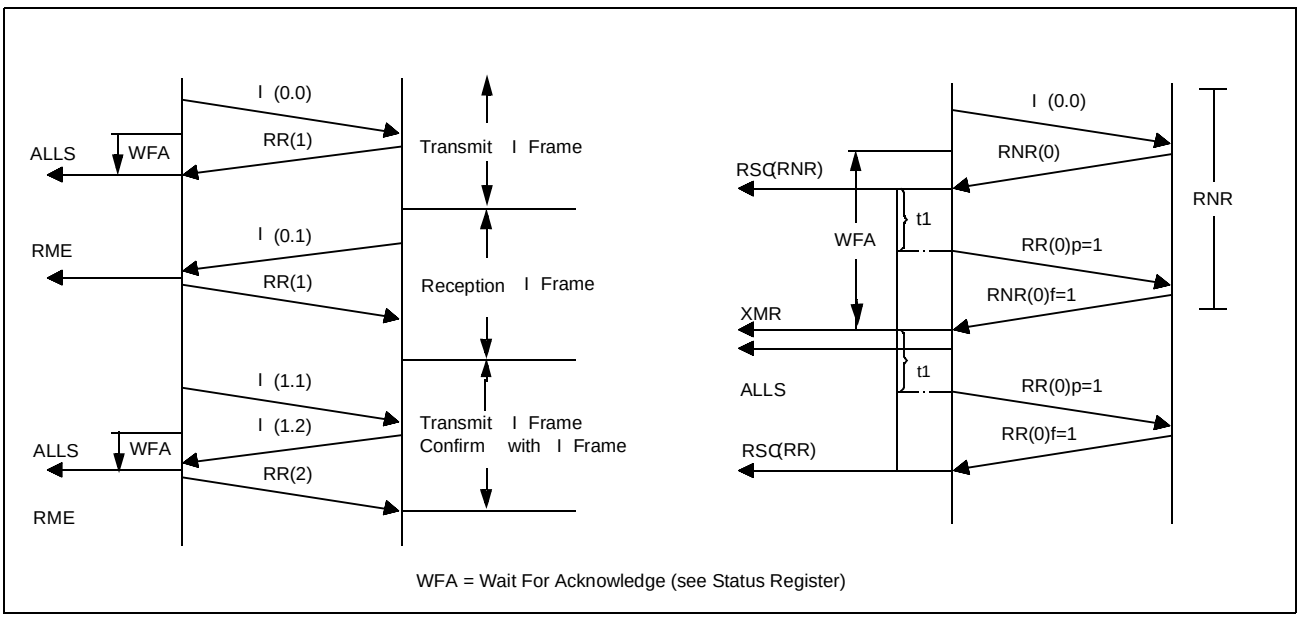


Figure 4-11 Transmission/Reception of I-Frames and Flow Control

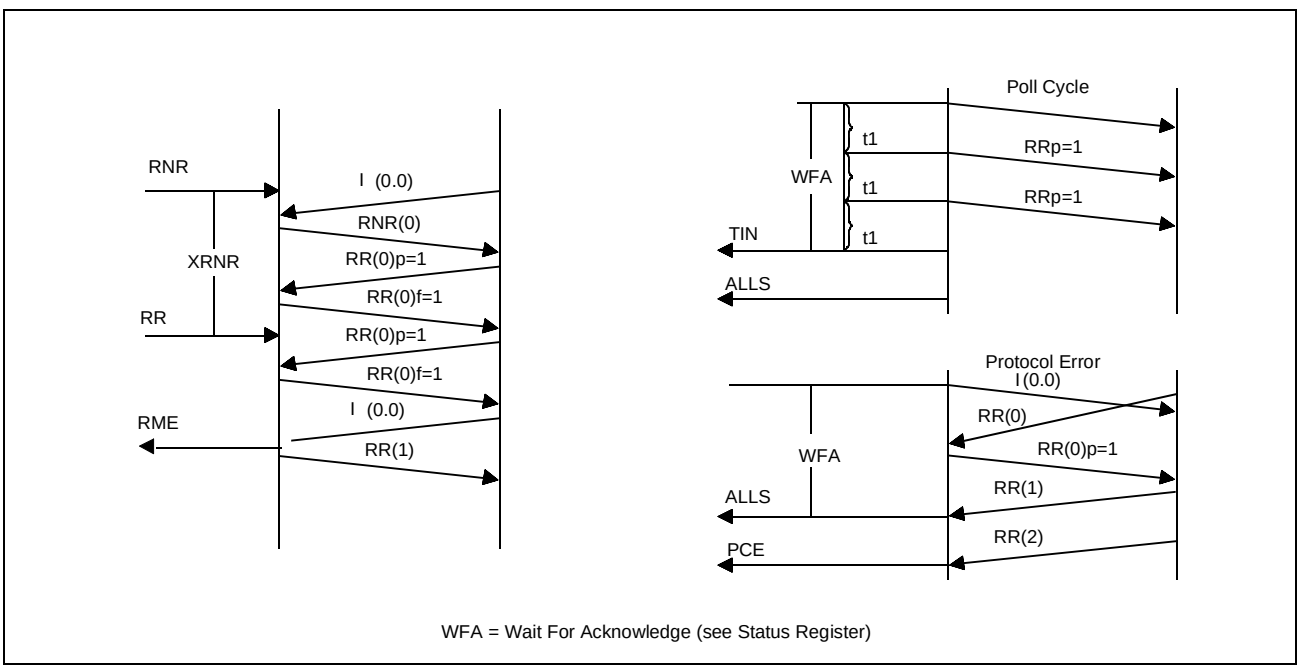


Figure 4-12 Flow Control: Reception of S-Commands and Protocol Errors

Protocol Error Handling:

Depending on the error type, erroneous frames are handled according to **Table 4-3**.

Table 4-3 Error Handling

Frame Type	Error Type	Generated Response	Generated Interrupt	Rec. Status
I	CRC error Aborted Unexpected N(S) Unexpected N(R)	– – S-frame –	RME RME – PCE	CRC error Abort – –
S	CRC error Aborted Unexpected N(R) With I-field	– – – –	– – PCE PCE	– – – –

Note: The station variables (V(S), V(R)) are not changed.

4.4.2 Half-Duplex SDLC-NRM Operation

The LAP controllers of the two serial channels can be configured to function in a half-duplex Normal Response Mode (NRM), where they operate as a slave (secondary) station, by setting the NRM bit in the CCR2L register of the corresponding channel.

In contrast to the full-duplex LAP B/LAP D operation, where the combined (primary + secondary) station transmits both commands and responses and may transmit data at any time, the NRM mode allows only responses to be transmitted **and** the secondary station may transmit only when instructed to do so by the master (primary) station. The SCC gets the permission to transmit from the primary station via an S-, or I-frame with the poll bit (p) **set**.

The NRM mode can be profitably used in a point-to-multipoint configuration with a fixed master-slave relationship, which guarantees the absence of collisions on the common transmit line. It is the responsibility of the master station to poll the slaves periodically and to handle error situations.

Prerequisite for NRM operation is:

– auto mode with 8-bit address field selected
  Register CCR2L bit fields ’MDS1’, ’MDS0’, ’ADM’ = ‘000’
– Register TIMR3 bit ’TMD’ = ‘0’
– same transmit and receive addresses, since only responses can be transmitted, i.e.
  Register XAD1 = XAD2 and register RAL1 = RAL2 (address of secondary).

Note: The broadcast address may be programmed in register RAL2 if broadcasting is required.
In this case registers RAL1 and RAL2 are not equal.

The primary station has to operate in transparent HDLC mode.

Reception of Frames:

The reception of frames functions similarly to the LAPB/LAPD operation (see **"Full-Duplex LAPB/LAPD Operation" on page 4-93**).

Transmission of Frames:

The SCC does **not** transmit S-, or I-frames if not instructed to do so by the primary station via an S-, or I-frame with the poll bit set.

The SCC can be told to send an I-frame issuing the transmit command 'XIF' in register CMDRL. The transmission of the frame, however, will not be initiated by the SCC until reception of either an

- RR, or
- I-frame

with poll bit set (p = '1').

After the frame has been transmitted (with the final bit set), the host has to wait for an ALLS or XMR interrupt.

A secondary does not poll the primary for acknowledgements, thus timer supervision must be done by the primary station.

Upon the arrival of an acknowledgement the SCC transmit FIFO is enabled and an interrupt is forwarded to the host, either the

– message has been positively acknowledged (ALLS interrupt), or the
– message must be repeated (XMR interrupt).

Additionally, the on-chip timer can be used **under host control** to provide timer recovery of the secondary if no acknowledgements are received at all.

Note: A secondary will transmit transparent frames only if the permission to send is given by receiving an S-frame or I-frame with poll bit set (p = '1').

Examples:

A few examples of SCC/host interaction in the case of normal response mode (NRM) mode are shown in **Figure 4-13** and **Figure 4-14**.

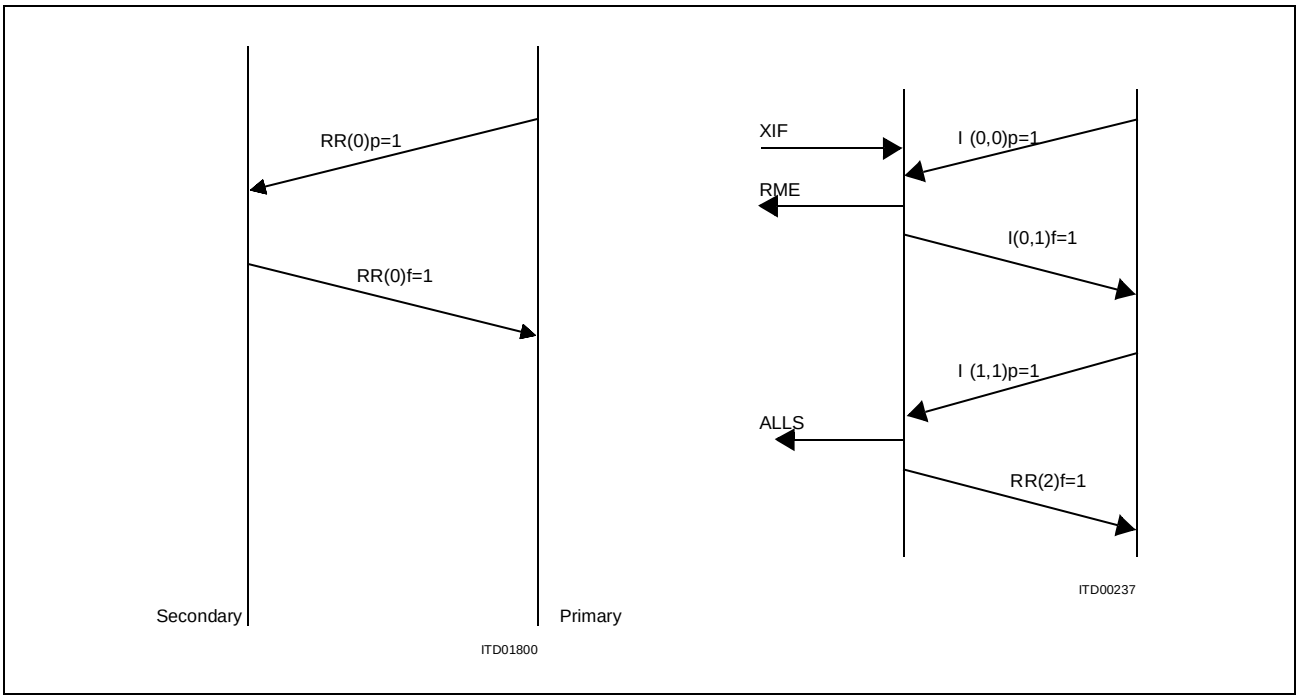


Figure 4-13 No Data to Send: Data Reception/Transmission

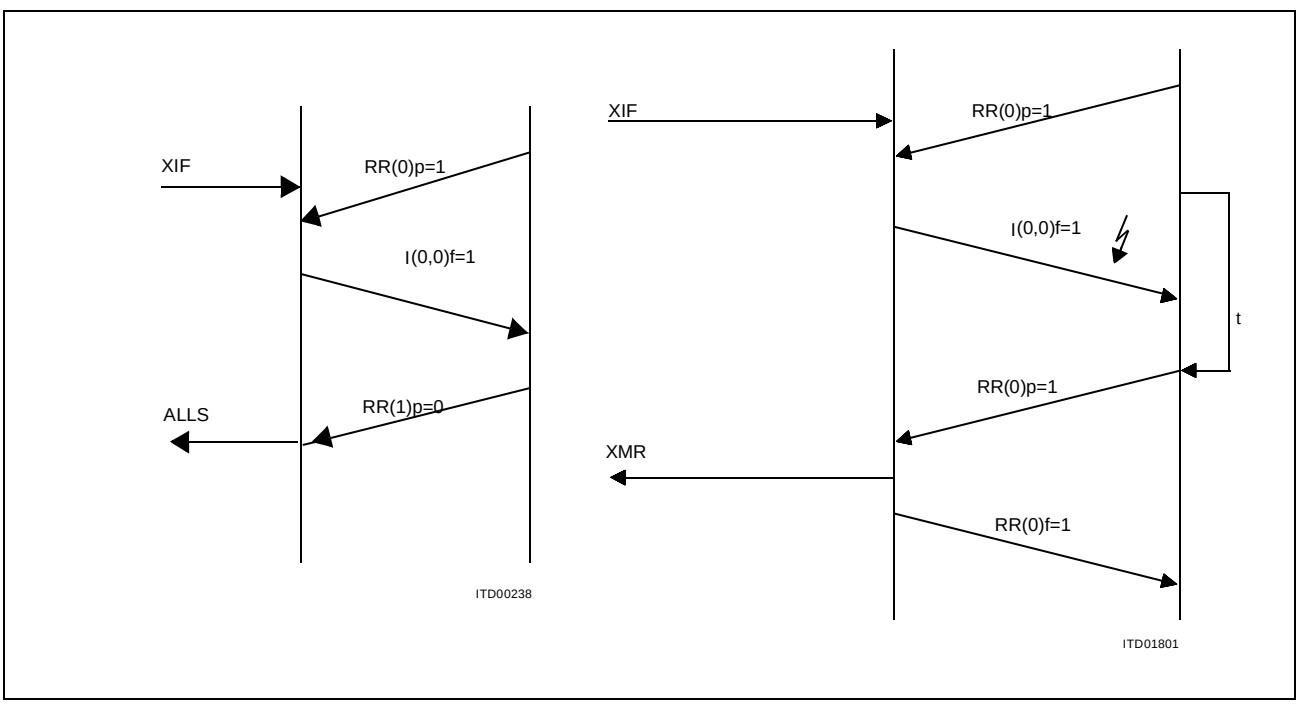


Figure 4-14 Data Transmission (without error), Data Transmission (with error)

4.4.3 Signaling System #7 (SS7) Operation

The PASSAT supports the signaling system #7 (SS7) which is described in ITU-Q.703. SS7 support must be activated by setting bit 'ESS7' in register CCR3L.

Receive

The SS7 protocol is supported by the following hardware features in receive direction:

- Recognition of Signaling Unit type
- Discard of repeatedly received FISUs and LSSUs if content is unchanged (optional)
- Check if the length of the received signaling unit is at least six octets (including the opening flag)
- Check if the signal information field of a received signaling unit consists of more than 272 octets (enabled with bit CCR3L.ELC). In this case, reception of the current signaling unit will be aborted.
- Counting and processing of errored signaling units

In order to reduce the microprocessor load, Fill In Signaling Units (FISUs) are processed automatically. By examining the length indicator of a received Signal Unit (SU) PASSAT decides whether a FISU has been received. Consecutively received FISUs will be compared and optionally not stored in the RFIFO, if the content is equal to the previous one. The same applies to Link Status Signaling Units (LSSUs), if enabled with bit CCR3L.CSF. The different types of Signaling Units as Message Signaling Unit (MSU), Link Status Signaling Unit (LSSU) and Fill-In Signaling Units (FISU) are indicated in the RSTA byte (bit field 'SU'), which is automatically added to the RFIFO with each received Signaling Unit. The complete Signaling Unit except start and end flags is stored in the receive FIFO. The functions of bits CCR3H.RCRC and CCR3H.RADD are also valid in SS7 mode, with bit 'RADD' related to BSN (backward sequence number) and FSN (forward sequence number).

Errored signaling units are counted and processed according to ITU-T Q.703. The SU counter and errored-SU counter are reset by setting CMDRH.RSUC to '1'. The error threshold can be selected to be 64 (default) or 32 by clearing/setting bit CCR3L.SUET. If the defined error limit is exceeded, an interrupt (ISR1.SUEX) is generated, if not masked by bit IMR1.SUEX.

Transmit

In transmit direction, following features are supported:

- single or repetitive transmission of signaling units
- automatic generation of Fill-In Signaling Units (FISU)

Each Signaling Unit (SU) written to the transmit FIFO (XFIFO) will be sent once or repeatedly including flags, CRC checksum and stuffed bits. After e.g. an MSU has been transmitted completely, PASSAT optionally starts sending of Fill In Signaling Units (FISUs) containing the forward sequence number (FSN) and the backward sequence number (BSN) of the previously transmitted signaling unit. Setting bit CCR3L.AFX to '1' causes FISUs to be sent continuously if no Signaling Unit is to be transmitted from XFIFO. After a new signaling unit has been written to the XFIFO and a transmission has been initiated, the current FISU is completed and the new SU is sent. After this,

transmission of FISUs continues. The internally generated FISUs contain FSN and BSN of the last transmitted signaling unit written to XFIFO.

Using CMDRL.XREP='1', the contents of XFIFO (1..32 bytes) can be sent continuously. This cyclic transmission can be stopped with the CMDRL.XRES command.

5 Register Description

5.1 Register Overview

The PASSAT global registers are used to configure and control the Serial Communication Controllers (SCCs), General Purpose Pins (GPP) and DMA operation.

All registers are 8-bit organized registers, but grouped and optimized for 16 bit access. 16 bit access (P-TQFP-100-3 package) is supported to even addresses only.

Table 5-1 provides an overview about all on-chip registers:

Table 5-1 Register Overview

Offset Ch A	Offset Ch B	Register read	Register write	Res Val	Meaning	Page
Global registers:						
00_H		GCMDR		00_H	Global Command Register	5-108
01_H		GMODE		$0F_H$	Global Mode Register	5-109
02_H		*Reserved*				
03_H		GSTAR		00_H	Global Status Register	5-112
04_H		GPDIRL		07_H	GPP Direction Register (Low Byte)	5-114
05_H		GPDIRH		FF_H	GPP Direction Register (High Byte)	5-114
06_H		GPDATL		-	GPP Data Register (Low Byte)	5-116
07_H		GPDATH		-	GPP Data Register (High Byte)	5-116
08_H		GPIML		07_H	GPP Interrupt Mask Register (Low Byte)	5-118
09_H		GPIMH		FF_H	GPP Interrupt Mask Register (High Byte)	5-118
$0A_H$		GPISL		00_H	GPP Interrupt Status Register (Low Byte)	5-120
$0B_H$		GPISH		00_H	GPP Interrupt Status Register (High Byte)	5-120
$0C_H$		DCMDR		00_H	DMA Command Register	5-122
$0D_H$		*Reserved*				
$0E_H$		DISR		00_H	DMA Interrupt Status Register	5-123
$0F_H$		DIMR		77_H	DMA Interrupt Mask Register	5-124
Channel specific registers:						
10_H	60_H	RFIFO	XFIFO	-	Receive/Transmit FIFO (Low Byte)	5-125
11_H	61_H			-	Receive/Transmit FIFO (High Byte)	5-125

Table 5-1 Register Overview (cont'd)

Offset Ch A	Offset Ch B	Register read	Register write	Res Val	Meaning	Page
12_H	62_H	STARL		00_H	Status Register (Low Byte)	5-128
13_H	63_H	STARH		10_H	Status Register (High Byte)	5-128
14_H	64_H	CMDRL		00_H	Command Register (Low Byte)	5-132
15_H	65_H	CMDRH		00_H	Command Register (High Byte)	5-132
16_H	66_H	CCR0L		00_H	Channel Configuration Register 0 (Low Byte)	5-136
17_H	67_H	CCR0H		00_H	Channel Configuration Register 0 (High Byte)	5-136
18_H	68_H	CCR1L		00_H	Channel Configuration Register 1 (Low Byte)	5-139
19_H	69_H	CCR1H		00_H	Channel Configuration Register 1 (High Byte)	5-139
$1A_H$	$6A_H$	CCR2L		00_H	Channel Configuration Register 2 (Low Byte)	5-144
$1B_H$	$6B_H$	CCR2H		00_H	Channel Configuration Register 2 (High Byte)	5-144
$1C_H$	$6C_H$	CCR3L		00_H	Channel Configuration Register 3 (Low Byte)	5-149
$1D_H$	$6D_H$	CCR3H		00_H	Channel Configuration Register 3 (High Byte)	5-149
$1E_H$	$6E_H$	PREAMB		00_H	Preamble Register	5-153
$1F_H$	$6F_H$	*Reserved*				
20_H	70_H	ACCM0		00_H	PPP ASYNC Control Character Map 0	5-154
21_H	71_H	ACCM1		00_H	PPP ASYNC Control Character Map 1	5-154
22_H	72_H	ACCM2		00_H	PPP ASYNC Control Character Map2	5-155
23_H	73_H	ACCM3		00_H	PPP ASYNC Control Character Map 3	5-155
24_H	74_H	UDAC0		$7E_H$	User Defined PPP ASYNC Control Character Map 0	5-157
25_H	75_H	UDAC1		$7E_H$	User Defined PPP ASYNC Control Character Map 1	5-157
26_H	76_H	UDAC2		$7E_H$	User Defined PPP ASYNC Control Character Map 2	5-158

Table 5-1 Register Overview (cont'd)

Offset Ch A	Offset Ch B	Register read write	Res Val	Meaning	Page
27_H	77_H	UDAC3	$7E_H$	User Defined PPP ASYNC Control Character Map 3	5-158
28_H	78_H	TTSA0	00_H	Transmit Time Slot Assignment Register 0	5-160
29_H	79_H	TTSA1	00_H	Transmit Time Slot Assignment Register 1	5-160
$2A_H$	$7A_H$	TTSA2	00_H	Transmit Time Slot Assignment Register 2	5-161
$2B_H$	$7B_H$	TTSA3	00_H	Transmit Time Slot Assignment Register 3	5-161
$2C_H$	$7C_H$	RTSA0	00_H	Receive Time Slot Assignment Register 0	5-163
$2D_H$	$7D_H$	RTSA1	00_H	Receive Time Slot Assignment Register 1	5-163
$2E_H$	$7E_H$	RTSA2	00_H	Receive Time Slot Assignment Register 2	5-164
$2F_H$	$7F_H$	RTSA3	00_H	Receive Time Slot Assignment Register 3	5-164
30_H	80_H	PCMTX0	00_H	PCM Mask Transmit Direction Register 0	5-166
31_H	81_H	PCMTX1	00_H	PCM Mask Transmit Direction Register 1	5-166
32_H	82_H	PCMTX2	00_H	PCM Mask Transmit Direction Register 2	5-167
33_H	83_H	PCMTX3	00_H	PCM Mask Transmit Direction Register 3	5-167
34_H	84_H	PCMRX0	00_H	PCM Mask Receive Direction Register 0	5-169
35_H	85_H	PCMRX1	00_H	PCM Mask Receive Direction Register 1	5-169
36_H	86_H	PCMRX2	00_H	PCM Mask Receive Direction Register 2	5-170
37_H	87_H	PCMRX3	00_H	PCM Mask Receive Direction Register 3	5-170
38_H	88_H	BRRL	00_H	Baud Rate Register (Low Byte)	5-172
39_H	89_H	BRRH	00_H	Baud Rate Register (High Byte)	5-172
$3A_H$	$8A_H$	TIMR0	00_H	Timer Register 0	5-174
$3B_H$	$8B_H$	TIMR1	00_H	Timer Register 1	5-174
$3C_H$	$8C_H$	TIMR2	00_H	Timer Register 2	5-175
$3D_H$	$8D_H$	TIMR3	00_H	Timer Register 3	5-175
$3E_H$	$8E_H$	XAD1	00_H	Transmit Address 1 Register	5-178
$3F_H$	$8F_H$	XAD2	00_H	Transmit Address 2 Register	5-178
40_H	90_H	RAL1	00_H	Receive Address 1 Low Register	5-180
41_H	91_H	RAH1	00_H	Receive Address 1 High Register	5-180
42_H	92_H	RAL2	00_H	Receive Address 2 Low Register	5-181
43_H	93_H	RAH2	00_H	Receive Address 2 High Register	5-181

Table 5-1 Register Overview (cont'd)

Offset Ch A	Offset Ch B	Register read	Register write	Res Val	Meaning	Page
44_H	94_H	AMRAL1		00_H	Mask Receive Address 1 Low Register	5-183
45_H	95_H	AMRAH1		00_H	Mask Receive Address 1 High Register	5-183
46_H	95_H	AMRAL2		00_H	Mask Receive Address 2 Low Register	5-184
47_H	96_H	AMRAH2		00_H	Mask Receive Address 2 High Register	5-184
48_H	98_H	RLCRL		00_H	Receive Length Check Register (Low Byte)	5-186
49_H	99_H	RLCRH		00_H	Receive Length Check Register (High Byte)	5-186
$4A_H$	$9A_H$	*Reserved*				
...						
$4F_H$	$9F_H$					
50_H	$A0_H$	ISR0		00_H	Interrupt Status Register 0	5-188
51_H	$A1_H$	ISR1		00_H	Interrupt Status Register 1	5-188
52_H	$A2_H$	ISR2		00_H	Interrupt Status Register 2	5-189
53_H	$A3_H$	Reserved				
54_H	$A4_H$	IMR0		FF_H	Interrupt Mask Register 0	5-194
55_H	$A5_H$	IMR1		FF_H	Interrupt Mask Register 1	5-194
56_H	$A6_H$	IMR2		03_H	Interrupt Mask Register 2	5-195
57_H	$A7_H$	Reserved				
58_H	$A8_H$	RSTA		00_H	Receive Status Byte	5-197
59_H	$A9_H$	Reserved				
...						
$5F_H$	AF_H					
Channel specific DMA registers:						
$B0_H$	CA_H	*Reserved*				
...						
$B7_H$	$D1_H$					
$B8_H$	$D2_H$	XBCL		00_H	Transmit Byte Count (Low Byte)	5-202
$B9_H$	$D3_H$	XBCH		00_H	Transmit Byte Count (High Byte)	5-202
BA_H	$D4_H$					

Table 5-1 Register Overview (cont'd)

Offset Ch A	Offset Ch B	Register read write	Res Val	Meaning	Page
...		*Reserved*			
$C3_H$	DD_H				
$C4_H$	DE_H	RMBSL	00_H	Receive Maximum Buffer Size (Low Byte)	5-204
$C5_H$	DF_H	RMBSH	00_H	Receive Maximum Buffer Size (High Byte)	5-204
$C6_H$	$E0_H$	RBCL	00_H	Receive Byte Count (Low Byte)	5-206
$C7_H$	$E1_H$	RBCH	00_H	Receive Byte Count (High Byte)	5-206
$C8_H$	$E2_H$	Reserved			
$C9_H$	$E3_H$	Reserved			
Miscellaneous:					
$E4_H$		Reserved			
...					
EB_H					
EC_H		VER0	03_H	Version Register 0	5-208
ED_H		VER1	$F0_H$	Version Register 1	5-208
EE_H		VER2	05_H	Version Register 2	5-209
EF_H		VER3	10_H	Version Register 3	5-209

5.2 Detailed Register Description

5.2.1 Global Registers

Each register description is organized in three parts:

- a head with general information about reset value, access type (read/write), offset address and usual handling;
- a table containing the bit information (name of bit positions);
- a section containing the detailed description of each bit.

Register 5-1 GCMDR
Global Command Register

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **00_H**
typical usage: written by CPU,
evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Global Command Bits							
	0	0	0	0	0	0	0	SWR

SWR **Software Reset Command**

Self clearing command bit:

bit='0' No software reset command is issued.

bit='1' Causes PASSAT to perform a complete reset identical to hardware reset.

Register 5-2 GMODE
Global Mode Register

CPU Accessibility: **read/write**
Reset Value: **$0F_H$**
Offset Address: **01_H**
typical usage: written by CPU
evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	DMA and Global Interrupt Control							
	0	EDMA	IPC(1:0)		OSCPD	*Reserved*	DSHP	GIM

EDMA **Enable External DMA Support**

This bit field controls the DMA operation mode:

EDMA=’0’ The external DMA controller support functions are disabled. PASSAT is operated in standard register access controlled mode.

EDMA=’1’ External DMA controller support functions are enabled.

IPC(1:0) **Interrupt-Port Configuration**

These bits control the function of interrupt output pin INT/$\overline{\text{INT}}$:

IPC(1:0)	Output Function:
’00’	Open Drain active low
’01’	Push/Pull active low
’10’	Reserved.
’11’	Push/Pull active high

OSCPD **Oscillator Power Down**

Setting this bit to ’0’ enables the internal oscillator. For power saving purposes (escpecially if clock modes are used which do not need the internal oscillator) this bit may remain set to ’1’.

OSCPD=’0’ The internal oscillator is active.

OSCPD=’1’ The internal oscillator is in power down mode.

Note: After reset this bit is set to ’1’, i.e. the oscillator is in power down mode!

Reserved ***Reserved Bit***

The reset value of this bit is ’1’.
It should be set to ’0’ during configuration in any case.

Note: This bit is a redundant control bit for the shaper in the oscillator unit. In later revisions of PASSAT the shaper will be controlled with bit ’DSHP’ only!

DSHP **Disable Shaper**

This bit has to be set to ’0’ if the shaping function in the oscillator unit is desired. The shaper amplifies the oscillator signal and improves the slope of the clock edges.

DSHP=’0’ Shaper is enabled. Recommended setting if a crystal is connected to pins XTAL1/XTAL2.

DSHP=’1’ Shaper is disabled (bypassed). Recommended setting if
- a TTL level clock signal is supplied to pin XTAL1
- the oscillator unit is unused

Note: (1) After reset this bit is set to ’1’, i.e. the shaper is disabled!
(2) For correct operation the reserved bit 2 must be set to ’0’ (in later revisions the shaper will be controlled with bit ’DSHP’ only)!

GIM **Global Interrupt Mask**

This bits disables all interrupt indications via pin INT/$\overline{\text{INT}}$. Internal operation (interrupt generation, interrupt status register update,...) is not affected.
If set, pin INT/$\overline{\text{INT}}$ immediately changes or remains in inactive state.

GIM='0' Global interrupt mask is cleared. Pin INT/$\overline{\text{INT}}$ is controlled by the internal interrupt control logic and activated as long as at least one unmasked interrupt indication is pending (not yet confirmed by read access to corresponding interrupt status register).

GIM='1' Global interrupt mask is set. Pin INT/$\overline{\text{INT}}$ remains inactive.

Note: After reset this bit is set to '1', i.e. all interrupts are disabled!

Register 5-3 GSTAR
Global Status Register

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **03_H**
typical usage: written by PASSAT evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Global Interrupt Status Information							
	GPI	DMI	ISA2	ISA1	ISA0	ISB2	ISB1	ISB0

GPI **General Purpose Port Indication** (-)

This bit indicates, that a GPP port interrupt indication is pending:

GPI=’0’ No general purpose port interrupt indication is pending.

GPI=’1’ General purpose port interrupt indication is pending. The source for this interrupt can be further determined by reading registers GPISL/GPISH (refer to page 5-120).

DMI **DMA Interrupt Indication** (-)

This bit indicates, that a DMA interrupt indication is pending:

DMI=’0’ No DMA interrupt indication is pending.

DMI=’1’ DMA interrupt indication is pending. The source for this interrupt (channel A/B, receive/transmit) can be further determined by reading register DISR (refer to page 5-123).

ISA2 **Channel A Interrupt Status Register 2**

ISA1 **Channel A Interrupt Status Register 1**

ISA0 **Channel A Interrupt Status Register 0**

ISB2 **Channel B Interrupt Status Register 2**

ISB1 **Channel B Interrupt Status Register 1**

ISB0 **Channel B Interrupt Status Register 0**

These bits indicate, that an interrupt indication is pending in the corresponding interrupt status register(s) ISR0/ISR1/ISR2 of the serial communication controller (SCC):

bit=’0’ No interrupt indication is pending.

bit=’1’ An interrupt indication is pending.

Register 5-4 GPDIRL
GPP Direction Register (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **07_H**
Offset Address: **04_H**
typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	GPP I/O Direction Control							
	0	0	0	0	0	GP10DIR	GP9DIR	GP8DIR

Register 5-5 GPDIRH
GPP Direction Register (High Byte)

CPU Accessibility: **read/write**
Reset Value: **FF_H**
Offset Address: **05_H**
typical usage: written by CPU evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	GPP I/O Direction Control							
	1	GP6DIR	1	1	1	GP2DIR	GP1DIR	GP0DIR

GPnDIR **GPP Pin n Direction Control** (-)

This bit selects between input and output function of the corresponding GPP pin:

bit = ’0’ output

bit = ’1’ input (reset value)

Register 5-6 GPDATL
GPP Data Register (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **-**
Offset Address: **06_H**
typical usage: written by CPU(outputs) and PASSAT(inputs), evaluated by PASSAT(outputs) and CPU(inputs)

Bit	7	6	5	4	3	2	1	0
	GPP Data I/O							
	-	-	-	-	-	GP10DAT	GP9DAT	GP8DAT

Register 5-7 GPDATH
GPP Data Register (High Byte)

CPU Accessibility: **read/write**
Reset Value: **-**
Offset Address: **07_H**
typical usage: written by CPU(outputs) and PASSAT(inputs), evaluated by PASSAT(outputs) and CPU(inputs)

Bit	7	6	5	4	3	2	1	0
	GPP Data I/O							
	-	GP6DAT	-	-	-	GP2DAT	GP1DAT	GP0DAT

GPnDAT **GPP Pin n Data I/O Value** (-)

This bit indicates the value of the corresponding GPP pin:

bit = ’0’ If direction is input: input level is ’low’;
if direction is output: output level is ’low’.

bit = ’1’ If direction is input: input level is ’high’;
if direction is output: output level is ’high’.

Register 5-8 GPIML
GPP Interrupt Mask Register (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **07_H**
Offset Address: **08_H**
typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	GPP Interrupt Mask Bits							
	0	0	0	0	0	GP10IM	GP9IM	GP8IM

Register 5-9 GPIMH
GPP Interrupt Mask Register (High Byte)

CPU Accessibility: **read/write**
Reset Value: **FF_H**
Offset Address: **09_H**
typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	GPP Interrupt Mask Bits							
	1	GP6IM	1	1	1	GP2IM	GP1IM	GP0IM

GPnIM **GPP Pin n Interrupt Mask** (-)

This bit controls the interrupt mask of the corresponding GPP pin:

bit = ’0’ Interrupt generation is enabled. An interrupt is generated on any state transition of the corresponding port pin (inputs).

bit = ’1’ Interrupt generation is disabled (reset value).

Register 5-10 GPISL
GPP Interrupt Status Register (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **$0A_H$**
typical usage: written by PASSAT, read and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	GPP Interrupt Status Bits							
	0	0	0	0	0	GP10I	GP9I	GP8I

Register 5-11 GPISH
GPP Interrupt Status Register (High Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **$0B_H$**
typical usage: written by PASSAT, read and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	GPP Interrupt Status Bits							
	0	GP6I	0	0	0	GP2I	GP1I	GP0I

GPnI **GPP Pin n Interrupt Indiction** (-)

This bit indicates if an interrupt event occured on the corresponding GPP pin:

bit = ’0’ No interrupt indication is pending at this pin (no state transition has occured).

bit = ’1’ An interrupt indication is pending (a state transition occured). The interrupt indication is cleared after read access.

Register 5-12 DCMDR
DMA Command Register

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **$0C_H$**
typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	DMA Controller Reset Command Bits							
	RDTB	0	RDRB	0	RDTA	0	RDRA	0

RDTB **Reset DMA Transmit Channel B**
RDRB **Reset DMA Receive Channel B**
RDTA **Reset DMA Transmit Channel A**
RDRA **Reset DMA Receive Channel A**

Self-clearing command bit.
These bits bring the external DMA support logic to the reset state:

bit=’0’ No reset is performed.
bit=’1’ Reset is performed.

Register 5-13 DISR
DMA Interrupt Status Register

CPU Accessibility: **read/write**
Reset Value: **00_H**
Offset Address: **$0E_H$**
typical usage: written by PASSAT, evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	DMA Interrupt Status Register							
	0	RBFB	RDTEB	TDTEB	0	RBFA	RDTEA	TDTEA

RBFB **Receive Buffer Full Channel B**

RBFA **Receive Buffer Full Channel A**

If a receive buffer size is defined in registers RMBSL/RMBSH and during reception the end of the receive buffer is reached this interrupt is generated indicating that the receive buffer is full. If the external DMA controller supports length protection for receive buffers itself this interrupt is obsolete. In that case, the receive buffer length check can be disabled by settin bit RMBSH:DRMBS to ’1’.

RDTEB **Receive DMA Transfer End Channel B**

RDTEA **Receive DMA Transfer End Channel A**

This bit set to ’1’ indicates that a DMA transfer of receive data is finished and the receive data is completely moved to the corresponding receive buffer in host memory.

TDTEB **Transmit DMA Transfer End Channel B**

TDTEA **Transmit DMA Transfer End Channel A**

This bit set to ’1’ indicates that the data is completely moved from the transmit buffer to the on-chip transmit FIFO, i.e. the transmit byte count programmed in registers XBCL/XBCH is reached.

Register 5-14 DIMR
DMA Interrupt Mask Register

CPU Accessibility: **read/write**
Reset Value: **77_H**
Offset Address: **$0F_H$**
typical usage:

Bit	7	6	5	4	3	2	1	0
	DMA Interrupt Mask Register							
	0	MRBFB	MRDTEB	MTDTEB	0	MRBFA	MRDTEA	MTDTEA

MRBFB **Mask Receive Buffer Full Interrupt Channel B**
MRBFA **Mask Receive Buffer Full Interrupt Channel A**
MRDTEB **Mask Receive DMA Transfer End Interrupt Channel B**
MRDTEA **Mask Receive DMA Transfer End Interrupt Channel A**
MTDTEB **Mask Transmit DMA Transfer End Interrupt Channel B**
MTDTEA **Mask Transmit DMA Transfer End Interrupt Channel A**

If a bit in this interrupt mask register is set to ’1’, the corresponding interrupt is not generated and not indicated in the corresponding bit position in the DISR register. After reset all interrupts are masked.

5.2.2 Channel Specific SCC Registers

Each register description is organized in three parts:

- a head with general information about reset value, access type (read/write), channel specific offset addresses and usual handling;
- a table containing the bit information (name of bit positions);
- a section containing the detailed description of each bit.

Register 5-15 FIFOL
Receive/Transmit FIFO (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **-**

	Channel A	Channel B
Offset Address:	**10_H**	**60_H**

typical usage: XFIFO: written by CPU, evaluated by PASSAT
RFIFO: written by PASSAT, evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	RFIFO/XFIFO Access Low Byte							
	FIFO(7:0)							

Register 5-16 FIFOH
Receive/Transmit FIFO (High Byte)

CPU Accessibility: **read/write**

Reset Value: **-**

	Channel A	Channel B
Offset Address:	**11_H**	**61_H**

typical usage: XFIFO: written by CPU, evaluated by PASSAT
RFIFO: written by PASSAT, evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	RFIFO/XFIFO Access High Byte							
	FIFO(15:8)							

Receive FIFO (RFIFO)

Reading data from the RFIFO can be done in 8-bit (byte) or 16-bit (word) accesses, depending on the selected microprocessor bus width using signal 'WIDTH'.

Note: The 'WIDTH' signal is available for the P-TQFP-100-3 package only. With the P-LFBGA-80-2 package only 8-bit accesses are supported.

The size of the accessible part of RFIFO is determined by programming the RFIFO threshold level in bit field CCR3H.RFTH(1:0). The threshold can be adjusted to 32 (reset value), 16, 4 or 2 bytes.

- Interrupt Controlled Data Transfer (GMODE.EDMA='0')

Up to 32 bytes/16 words of received data can be read from the RFIFO following an RPF or an RME interrupt (see ISR0 register). The address provided during an RFIFO read access is not incremental; it is always 10_H for channel A or 60_H for channel B.

RPF Interrupt: This interrupt indicates that the adjusted receive threshold level is reached. The message is not yet complete. A fix number of bytes, dependent from the threshold level, has to be read.

RME Interrupt: The message is completely received. The number of valid **bytes** is determined by reading the RBCL, RBCH registers.

The content of the RFIFO is released by issuing the "Receive Message Complete" command (CMDRH.RMC).

- DMA Controlled Data Transfer (GMODE.EDMA='1')

If DMA operation is enabled, the PASSAT autonomously requests data transfer by asserting the DRR line to the external DMA controller. The DRR line remains active until the beginning of the last receive data byte/word transfer. For a detailed decription of the external DMA interface operation refer to "External DMA Controller Support" on page 3-77.

Transmit FIFO (XFIFO)

Writing data to the XFIFO can be done in 8-bit (byte) or 16-bit (word) accesses, depending on the selected microprocessor bus width using signal 'WIDTH'.

Note: The 'WIDTH' signal is available for the P-TQFP-100-3 package only. With the P-LFBGA-80-2 package only 8-bit accesses are supported.

- Interrupt Controlled Data Transfer (GMODE.EDMA='0')

Following an XPR (or an ALLS) interrupt, up to 32 bytes/16 words of new transmit data can be written into the XFIFO. Transmit data can be released for transmission with an XTF command. The address provided during an XFIFO write access is not incremental; it is always 10_H for channel A or 60_H for channel B.

- DMA Controlled Data Transfer (GMODE.EDMA=’1’)

If DMA operation is enabled, the PASSAT autonomously requests data transfer to the XFIFO by asserting the DRT line to the external DMA controller. The DRT line remains active until the beginning of the last transmit data byte/word transfer. For a detailed decription of the external DMA interface operation refer to "External DMA Controller Support" on page 3-77.

Register 5-17 STARL
Status Register (Low Byte)

CPU Accessibility: **read only**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**12_H**	**62_H**
typical usage:	updated by PASSAT	
	read and evaluated by CPU	

Bit	7	6	5	4	3	2	1	0
	Command Status				Transmitter Status			
	XREPE	0	0	CEC	0	XDOV	XFW	CTS

Register 5-18 STARH
Status Register (High Byte)

CPU Accessibility: **read only**

Reset Value: **10_H**

	Channel A	Channel B
Offset Address:	**13_H**	**63_H**
typical usage:	updated by PASSAT	
	read and evaluated by CPU	

Bit	7	6	5	4	3	2	1	0
	Receiver Status					Automode Status		
	0	0	CD	RLI	DPLA	WFA	XRNR	RRNR

XREPE **Transmit Repetition Executing**

XREPE=’0’ No transmit repetition command is in execution.

XREPE=’1’ A XREP command (register CMDRL) is currently in execution.

CEC **Command Executing**

CEC=’0’ No command is currently in execution. The command registers CMDRL/CMDRH can be written by CPU.

CEC=’1’ A command (written previously to registers CMDRL/CMDRH) is currently in execution. No further command can be written to registers CMDRL/CMDRH by CPU.

Note: CEC will be active at most 2.5 receive or transmit clock cycles (depending on whether a receiver or transmitter related command is executed).
CEC will stay active if the SCC is in power-down mode or if no serial clock, needed for command execution, is available.

XDOV **Transmit FIFO Data Overflow**

XDOV=’0’ Less than or equal to 32 bytes have been written to the XFIFO.

XDOV=’1’ More than 32 bytes have been written to the XFIFO. This bit is reset by:
– a transmitter reset command ’XRES’
– or when all bytes in the accessible half of the XFIFO have been moved into the inaccessible half.

XFW **Transmit FIFO Write Enable**

XFW=’0’ The XFIFO is not able to accept further transmit data.

XFW=’1’ Transmit data can be written to the XFIFO.

CTS **$\overline{\text{CTS}}$ (Clear To Send) Input Signal State**

CTS=’0’ $\overline{\text{CTS}}$ input signal is inactive (high level)

CTS=’1’ $\overline{\text{CTS}}$ input signal is active (low level)

Note: A transmit clock must be provided in order to detect the signal state of the $\overline{\text{CTS}}$ input pin.
Optionally this input can be programmed to generate an interrupt on signal level changes.

CD **CD (Carrier Detect) Input Signal State**

CD=’0’ CD input signal is low.

CD=’1’ CD input signal is high.

Note: A receive clock must be provided in order to detect the signal state of the CD input pin.
Optionally this input can be programmed to generate an interrupt on signal level changes.

RLI **Receive Line Inactive**

This bit indicates that neither flags as interframe time fill nor data are being received via the receive line.

RLI=’0’ Receive line is active, no constant high level is detected.

RLI=’1’ Receive line is inactive, i.e. more than 7 consecutive ’1’ are detected on the line.

Note: A receive clock must be provided in order to detect the receive line state.

DPLA **DPLL Asynchronous**

This bit is only valid if the receive clock is recovered by the DPLL and FM0, FM1 or Manchester data encoding is selected. It is set when the DPLL has lost synchronization. In this case reception is disabled (receive abort condition) until synchronization has been regained. In addition transmission is interrupted in all cases where transmit clock is derived from the DPLL (clock mode 3a, 7a). Interruption of transmission is performed the same way as on deactivation of the $\overline{\text{CTS}}$ signal.

DPLA=’0’ DPLL is synchronized.

DPLA=’1’ DPLL is asynchronous (re-synchronization process is started automatically).

WFA **Wait For Acknowledgement**

This status bit is significant in Automode only. It indicates whether the Automode state machine expects an acknowledging I- or S-Frame for a previously sent I-Frame.

WFA=’0’ No acknowledge I/S-Frame is expected.

WFA=’1’ The Automode state machine is waiting for an achnowledging S- or I-Frame.

XRNR **Transmit RNR Status**

This status bit is significant in Automode only. It indicates the receiver status of the local station (SCC).

XRNR=’0’ The receiver is ready and will automatically answer poll-frames with a S-Frame with ’receiver-ready’ indication.

XRNR=’1’ The receiver is NOT ready and will automatically answer poll-frames with a S-Frame with a ’receiver-not-ready’ indication.

RRNR **Received RNR (Receiver Not Ready) Status**

This status bit is significant in Automode only. It indicates the receiver status of the remote station.

RRNR=’0’ The remote station receiver is ready.

RRNR=’1’ The remote receiver is NOT ready.
(A ’receiver-not-ready’ indication was received from the remote station)

Register 5-19 CMDRL
Command Register (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**14_H**	**64_H**

typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Timer		Transmitter Commands					
	STI	TRES	XIF	XRES	XF	XME	XREP	0

Register 5-20 CMDRH
Command Register (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**15_H**	**65_H**

typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receiver Commands							
	RMC	RNR	0	0	RSUC	0	0	RRES

STI **Start Timer Command**

Self-clearing command bit:
HDLC Automode:
In HDLC Automode the timer is used internally for the autonomous protocol support functions. The timer is started automatically by the SCC when an I-Frame is sent out and needs to be acknowledged.
If the 'STI' command is issued by software:

STI='1' An S-Frame with poll bit set is sent out and the internal timer is started expecting an acknowledge from the remote station via an I- or S-Frame.
The timer is stopped after receiving an acknowledge otherwise the timer expires generating a timer interrupt.

Note: In HDLC Automode, bit 'TMD' in register TIMR3 must be set to '1'

All protocol modes except HDLC Automode:
In these modes the timer is operating as a general purpose timer.

STI='1' This commands starts timer operation.
The timer can be stopped by setting bit 'TRES'.

Note: Bit 'TMD' in register TIMR3 must be cleared for proper operation

TRES **Timer Reset**

Self-clearing command bit.
This bit deactivates timer operation:

TRES='0' Timer operation enabled.

TRES='1' Timer operation stopped.

XIF **Transmit I-Frame**

Self-clearing command bit.
This command bit is significant in HDLC Automode only.

XIF='1' Initiates the transmission of an I-frame in auto-mode. Additional to the opening flag, the address and control fields of the frame are added by PASSAT.

XRES **Transmitter Reset Command**

Self-clearing command bit:

XRES=’1’ The SCC transmit FIFO is cleared and the transmitter protocol engines are reset to their initial state.
A transmitter reset command is recommended after all changes in protocol mode configurations (e.g. switching between sub-modes of HDLC).

XF **Transmit Frame**

This self-clearing command bit is significant in interrupt driven operation only (GMODE.EDMA=’0’).

XF=’1’ After having written up to 32 bytes to the XFIFO, this command initiates transmission. In packet oriented protocols like HDLC/PPP the opening flag is automatically added by PASSAT. If the end of the packet is part of the transmit data, bit ’XME’ should be set in addition.

DMA Mode

After having written the length of the data block to be transmitted to registers XBCL and XBCH, this command initiates the data transfer from host memory to PASSAT by DMA. Transmission on the serial side starts as soon as 32 bytes are transferred to the XFIFO or the transmit byte counter value is reached.

XME **Transmit Message End**

Self-clearing command bit:

XME=’1’ Indicates that the data block written last to the XFIFO contains the end of the packet. This bit should always be set in conjunction with a transmit command (’XF’ or ’XIF’).

XREP **Transmission Repeat Command**

Self-clearing command bit:

XREP=’1’ If bit ’XREP’ is set together with bit ’XME’ and ’XF’, PASSAT repeatedly transmits the contents of the XFIFO (1..32 bytes).
The cyclic transmission can be stopped with the ’XRES’ command.

RMC **Receive Message Complete**

Self-clearing command bit:

RMC=’1’ With this bit the CPU indicates to PASSAT that the current receive data has been fetched out of the RFIFO. Thus the corresponding space in the RFIFO can be released and re-used by PASSAT for further incoming data.

RNR **Receiver Not Ready Command**

NON self-clearing command bit:
This command bit is significant in HDLC Automode only.

RNR=’0’ Forces the receiver to enter its ’receiver-ready’ state. The receiver acknowledges received poll or I-Frames with a ’receiver-ready’ indication.

RNR=’1’ Forces the receiver to enter its ’receiver-not-ready’ state. The receiver acknowledges received poll or I-Frames with a ’receiver-not-ready’ indication.

RSUC **Reset Signaling Unit Counter**

Self-clearing command bit:
This command bit is significant if HDLC SS7 mode is selected.

RSUC=’1’ The Signaling System #7 (SS7) unit counter is reset.

RRES **Receiver Reset Command**

Self-clearing command bit:

RRES=’1’ The SCC receive FIFO is cleared and the receiver protocol engines are reset to their initial state.
The SCC receive FIFO accepts new receive data from the protocol engine immediately after receiver reset procedure.
It is recommended to disable data reception before issuing a receiver reset command by setting bit CCR3L.RAC = ’0’ and enabling data reception afterwards.
A ’receiver reset’ command is recommended after all changes in protocol mode configurations.

Register 5-21 CCR0L
Channel Configuration Register 0 (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**16_H**	**66_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.				Clock Mode Selection			
	VIS	PSD	0	TOE	SSEL	CM(2:0)		

Register 5-22 CCR0H
Channel Configuration Register 0 (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**17_H**	**67_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Power	Line Coding						
	PU	SC(2:0)			0	0	0	0

PU **Power Up**

PU=’0’ The SCC is in ’power-down’ mode. The protocol engines are switched off (standby) and no operation is performed. This may be used to save power when SCC is not in use.

Note: The SCC transmit FIFO accepts transmit data even in ’power-down’ mode.

PU=’1’ The SCC is in ’power-up’ mode.

SC(2:0) **Serial Port Configuration**

This bit field selects the line coding of the serial port.
Note, that special operation modes and settings may require or exclude operation in special line coding modes. Refer to the ’prerequisites’ in the dedicated mode descriptions.

SC = ’000’ NRZ data encoding

SC = ’001’ Bus configuration, timing mode 1 (NRZ data encoding)

SC = ’010’ NRZI data encoding

SC = ’011’ Bus configuration, timing mode 2 (NRZ data encoding)

SC = ’100’ FM0 data encoding

SC = ’101’ FM1 data encoding

SC = ’110’ Manchester data encoding

SC = ’111’ Reserved

Note: If bus configuration mode is selected, only NRZ data encoding is supported.

VIS **Masked Interrupts Visible**

VIS=’0’ Masked interrupt status bits are not displayed in the interrupt status registers (ISR0..ISR2).

VIS=’1’ Masked interrupt status bits are visible and automatically cleared after interrupt status register (ISR0..ISR2) read access.

Note: Interrupts masked in registers IMR0..IMR2 will not generate an interrupt.

PSD **DPLL Phase Shift Disable**

This option is only applicable in the case of NRZ or NRZI line encoding is selected.

PSD=’0’ Normal DPLL operation.

PSD=’1’ The phase shift function of the DPLL is disabled. The windows for phase adjustment are extended.

TOE **Transmit Clock Out Enable**

For clock modes 0b, 2b, 3a, 3b, 6b, 7a and 7b, the internal transmit clock can be monitored on pin TxCLK as an output signal. In clock mode 5, a time slot control signal marking the active transmit time slot is output on pin TxCLK.
Bit ’TOE’ is invalid for all other clock modes.

TOE=’0’ TxCLK pin is input.

TOE=’1’ TxCLK pin is switched to output function if applicable for the selected clock mode.

SSEL **Clock Source Select**

Distinguishes between the ’a’ and ’b’ option of clock modes 0, 2, 3, 5, 6 and 7.

SSEL=’0’ Option ’a’ is selected.

SSEL=’1’ Option ’b’ is selected.

CM(2:0) **Clock Mode**

This bit field selects one of main clock modes 0..7. For a detailed description of the clock modes refer to **Chapter 3.2.3**

CM = ’000’ clock mode 0

CM = ’001’ clock mode 1

CM = ’010’ clock mode 2

CM = ’011’ clock mode 3

CM = ’100’ clock mode 4

CM = ’101’ clock mode 5 (time-slot oriented clocking modes)

CM = ’110’ clock mode 6

CM = ’111’ clock mode 7

Register 5-23 CCR1L
Channel Configuration Register 1 (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**18_H**	**68_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	CRL	C32	SOC(1:0)		SFLG	DIV	ODS	0

Register 5-24 CCR1H
Channel Configuration Register 1 (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**19_H**	**69_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	0	ICD	0	RTS	FRTS	FCTS	CAS	TSCM

CRL **CRC Reset Value**

This bit defines the initial value of the internal transmit/receive CRC generators:

CRL='0' Initial value is $0xFFFF_H$ (16 bit CRC), $0xFFFFFFFF_H$ (32 bit CRC).
This is the default value for most HDLC/PPP applications.

CRL='1' Initial value is $0x0000_H$ (16 bit CRC), $0x00000000_H$ (32 bit CRC).

C32 **CRC 32 Select**

This bit enables 32-bit CRC operation for transmit and receive.

C32='0' 16-bit CRC-CCITT generation/checking.

C32='1' 32-bit CRC generation/checking.

Note: The internal 'valid frame' criteria is updated depending on the selected number of CRC-bytes.

SOC(1:0) **Serial Output Control**

This bit field selects the $\overline{RTS}$ signal output function.
(This bit field is only valid in bus configuration modes selected via bit field SC(2:0) in register CCR0H).

SOC = '0X' $\overline{RTS}$ ouput signal is active during transmission of a frame (active low).

SOC = '10' $\overline{RTS}$ ouput signal is always inactive (high).

SOC = '11' $\overline{RTS}$ ouput signal is active during reception of a frame (active low).

SFLG **Shared Flags Transmission**

This bit enables 'shared flag transmission' in HDLC protocol mode. If another transmit frame begin is stored in the SCC transmit FIFO, the closing flag of the preceding frame becomes the opening flag of the next frame (shared flags):

SFLG = '0' Shared flag transmission disabled.

SFLG = '1' Shared flag transmission enabled.

Note: The receiver always supports shared flags and shared zeros of consecutive flags.

DIV **Data Inversion**

This bit is only valid if NRZ data encoding is selected via bit field SC(2:0) in register CCR0H.

DIV=’0’ No Data Inversion.

DIV=’1’ Data is transmitted/received inverted (on a per bit basis). In HDLC and HDLC Synchronous PPP modes the continuous ’1’ idle sequence is NOT inverted. Interframe time fill flag transmission is inverted.

ODS **Output Driver Select**

The transmit data output pin TxD can be configured as push/pull or open drain output chracteristic.

ODS=’0’ TxD pin is open drain output.

ODS=’1’ TxD pin is push/pull output.

ICD **Invert Carrier Detect Pin Polarity**

ICD=’0’ Carrier Detect (CD) input pin is active high.

ICD=’1’ Carrier Detect (CD) input pin is active low.

RTS **Request To Send Pin Control**

The request to send pin $\overline{\text{RTS}}$ can be controlled by PASSAT as an output autonomously or via setting/clearing bit ’RTS’.
This bit is not valid in clock mode 4.

RTS=’0’ Pin $\overline{\text{RTS}}$ (output) pin is controlled by PASSAT autonomously.
$\overline{\text{RTS}}$ is activated during transmission. In bus configuration mode the functionality depends on bit field ’SOC’ setting.

RTS=’1’ Pin $\overline{\text{RTS}}$ can be controlled by software. The output level of this pin depends on bit ’FRTS’.

Note: For $\overline{\text{RTS}}$ pin control a transmit clock is necessary.

FRTS **Flow Control (using signal $\overline{\text{RTS}}$)**

Bit 'FRTS' together with bit 'RTS' determine the function of signal $\overline{\text{RTS}}$:

RTS, FRTS		
0,	0	Pin $\overline{\text{RTS}}$ is controlled by PASSAT autonomously. $\overline{\text{RTS}}$ is activated (low) as soon as transmit data is available within the SCC transmit FIFO.
0,	1	Pin $\overline{\text{RTS}}$ is controlled by PASSAT autonomously supporting bi-directional data flow control. $\overline{\text{RTS}}$ is activated (low) if the shadow part of the SCC receive FIFO is empty and de-activated (high) when the SCC receive FIFO fill level reaches its receive FIFO threshold.
1,	0	Forces pin $\overline{\text{RTS}}$ to active state (low).
1,	1	Forces pin $\overline{\text{RTS}}$ to inactive state (high).

Note: For $\overline{\text{RTS}}$ pin control a transmit clock is necessary.

FCTS **Flow Control (using signal $\overline{\text{CTS}}$)**

This bit controls the function of pin $\overline{\text{CTS}}$.

FCTS = '0'	The transmitter is stopped if $\overline{\text{CTS}}$ input signal is inactive (high) and enabled if active (low).
FCTS = '1'	The transmitter is enabled, disregarding $\overline{\text{CTS}}$ input signal.

CAS **Carrier Detect Auto Start**

CAS = '0'	The CD pin is used as general input. In clock mode 1, 4 and 5, clock mode specific control signals must be provided at this pin (receive strobe, receive gating $\overline{\text{RCG}}$, frame sync clock $\overline{\text{FSC}}$). A pull-up/down resistor is recommended if unused.
CAS = '1'	The CD pin enables/disables the receiver for data reception. (Polarity of CD pin can be configured via bit 'ICD'.)

Note: (1) In clock mode 1, 4 and 5 this bit must be set to '0'.(3) A receive clock must be provided in order to detect the signal state of the CD input pin.

TSCM **Time Slot Control Mode**

This bit controls internal counter operation in time slot oriented clock mode 5:

TSCM=’0’ The internal counter keeps running, restarting with zero after being expired.

TSCM=’1’ The internal counter stops at its maximum value and restarts with the next frame sync pulse again.

Register 5-25 CCR2L
Channel Configuration Register 2 (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$1A_H$**	**$6A_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	MDS1	MDS0	ADM	NRM	PPPM(1:0)		TLPO	TLP

Register 5-26 CCR2H
Channel Configuration Register 2 (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$1B_H$**	**$6B_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	MCS	EPT	NPRE(1:0)		ITF	0	OIN	XCRC

MDS(1:0) **Mode Select**

This bit field selects the HDLC protocol sub-mode including the 'extended transparent mode'.

MDS = '00' Automode.

MDS = '01' Address Mode 2.

MDS = '10' Address Mode 0/1.
(Option '0' or '1' is selected via bit 'ADM'.)

MDS = '11' Extended transparent mode (bit transparent transmission/ reception).

Note: 'MDS(1:0)' must be set to '10' if any PPP mode is enabled via bit field 'PPPM' or if SS7 is enabled via bit 'ESS7' in register CCR3L.

ADM **Address Mode Select**

The meaning of this bit depends on the selected protocol sub-mode:

Automode, Address Mode 2:
Determines the address field length of an HDLC frame.

ADM = '0' 8-bit address field.

ADM = '1' 16-bit address field.

Address Mode 0/1:
Determines whether address mode 0 or 1 is selected.

ADM = '0' Address Mode 0 (no address recognition).

ADM = '1' Address Mode 1 (high byte address recognition).

Extended Transparent Mode:

ADM = '1' recommended setting

NRM **Normal Response Mode**

This bit is valid in HDLC Automode operation only and determines the function of the Automode LAP-Controller:

NRM = '0' Full-duplex LAP-B / LAP-D operation.

NRM = '1' Half-duplex normal response mode (NRM) operation.

PPPM(1:0) **PPP Mode Select**

This bit field enables and selects the HDLC PPP protocol modes:

PPPM = ’00’ No PPP protocol operation. The HDLC sub-mode is determined by bit field ’MDS’.

PPPM = ’01’ Octet synchronous PPP protocol operation.

PPPM = ’10’ *Reserved*

PPPM = ’11’ Bit synchronous PPP protocol operation.

Note: ’Address Mode 0’ must be selected by setting bit field ’MDS(1:0)’ to ’10’ and bit ’ADM’ to ’0’ if any PPP mode is enabled.

TLPO **Test Loop Out Function**

This bit is only valid if test loop is enabled and controls whether test loop transmit data is driven on pin TxD:

TLPO = ’0’ Test loop transmit data is driven to TxD pin.

TLPO = ’1’ Test loop transmit data is NOT driven to TxD pin. TxD pin is idle ’1’. Depending on the selected output characteristic the pin is high impedance (bit CCR1L.ODS =’0’) or driving high (CCR1L.ODS =’1’).

TLP **Test Loop**

This bit controls the internal test loop between transmit and receive data signals. The test loop is closed at the far end of serial transmit and receive line just before the respective TxD and RxD pins:

TLP = ’0’ Test loop disabled.

TLP = ’1’ Test loop enabled.
The software is responsible to select a clock mode which allows correct reception of transmit data depending on the external clock supply. Transmit data is sent out via pin TxD if not disabled with bit ’TLPO’. The receive input pin RxD is internally disconnected during test loop operation.

MCS **Modulo Count Select**

This bit is valid in HDLC Automode operation only and determines the control field format:

MCS = ’0’ Basic operation, one byte control field (modulo 8 counter operation).

MCS = ’1’ Extended operation, two bytes control field (modulo 128 counter operation).

EPT **Enable Preamble Transmission**

This bit enables preamble transmission. The preamble is started after interframe time fill (ITF) transmission is stopped because a new frame is ready to be transmitted. The preamble pattern consists of 8 bits defined in register PREAMB, which is sent repetitively. The number of repetitions is determined by bit field ’PRE(1:0)’:

EPT=’0’ Preamble transmission is disabled.

EPT=’1’ Preamble transmission is enabled.

Note: Preamble operation does NOT influence HDLC shared flag transmission if enabled.

NPRE(1:0) **Number of Preamble Repetitions**

This bit field determines the number of preambles transmitted:

NPRE = ’00’ 1 preamble.

NPRE = ’01’ 2 preambles.

NPRE = ’10’ 4 preambles.

NPRE = ’11’ 8 preambles.

ITF **Interframe Time Fill**

This bit selects the idle state of the transmit pin TxD:

ITF=’0’ Continuous logical ’1’ is sent during idle phase.

ITF=’1’ Continuous flag sequences are sent (’01111110’ flag pattern).

Note: It is recommended to clear bit ’ITF’ in bus configuration modes, i.e. continuous ’1’s are sent as idle sequence and data encoding is NRZ.

OIN **One Insertion**

In HDLC mode a one-insertion mechanism similar to the zero-insertion can be activated:

OIN=’0’ The ’1’ insertion mechanism is disabled.

OIN=’1’ In transmit direction a logical ’1’ is inserted to the serial data stream after 7 consecutive zeros.
In receive direction a ’1’ is deleted from the receive data stream after receiving 7 consecutive zeros.
This enables clock information to be recovered from the receive data stream by means of a DPLL, even in the case of NRZ data encoding, because a transition at bit cell boundary occurs at least every 7 bits.

XCRC **Transmit CRC Checking Mode**

XCRC=’0’ The transmit checksum (2 or 4 bytes) is generated and appended to the transmit data automatically.

XCRC=’1’ The transmit checksum is not generated automatically. The checksum is expected to be provided by software as the last 2 or 4 bytes in the transmit data buffer.

Register 5-27 CCR3L
Channel Configuration Register 3 (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$1C_H$**	**$6C_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	ELC	AFX	CSF	SUET	RAC	0	0	ESS7

Register 5-28 CCR3H
Channel Configuration Register 3 (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$1D_H$**	**$6D_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	misc.							
	0	DRCRC	RCRC	RADD	0	0	RFTH(1:0)	

ELC **Enable Length Check**

This bit is only valid in HDLC SS7 mode:
If the number of received octets exceeds 272 + 7 within one Signaling Unit, reception is aborted and bit RSTA.RAB is set.

ELC=’0’ Length Check disabled.

ELC=’1’ Length Check enabled.

AFX **Automatic FISU Transmission**

This bit is only valid in HDLC SS7 mode:
After the contents of the transmit FIFO (XFIFO) has been transmitted completely, FISUs are transmited automatically. These FISUs contain the FSN and BSN of the last transmitted Signaling Unit (provided in XFIFO).

AFX=’0’ Automatic FISU transmission disabled.

AFX=’1’ Automatic FISU transmission enabled.

CSF **Compare Status Field**

This bit is only valid in HDLC SS7 mode:
If the status fields of consecutive LSSUs are equal, only the first will be stored and every following is ignored

CSF=’0’ Compare is disabled, all received LSSUs are stored in the receive FIFO.

CSF=’1’ Compare is enabled, only the first one of consecutive equal LSSUs is stored in the receive FIFO.

SUET **Signalling Unit Counter Threshold**

This bit is only valid in HDLC SS7 mode:
Defines the number of signaling units received in error that will cause an error rate high indication (ISR1.SUEX).

SUET=’0’ threshold is 64 errored signaling units.

SUET=’1’ threshold is 32 errored signaling units.

RAC **Receiver active**

Switches the receiver between operational/inoperational states:

RAC=’0’ Receiver inactive, receive line is ignored.

RAC=’1’ Receiver active.

ESS7 **Enable SS7 Mode**

This bit is only valid in HDLC mode only.

ESS7=’0’ Disable signaling system #7 (SS7) support.

ESS7=’1’ Enable signaling system #7 (SS7) support.

Note: If SS7 mode is enabled, ’Address Mode 0’ must be selected by setting bit field CCR2L:MDS(1:0) to ’10’ and bit CCR2L:ADM to ’0’.

DRCRC **Disable Receive CRC Checking**

DRCRC=’0’ The receiver expects a 16 or 32 bit CRC within a HDLC frame. CRC processing depends on the setting of bit ’RCRC’.
Frames shorter than expected are marked ’invalid’ or are discarded (refer to RSTA description).

DRCRC=’1’ The receiver does not expect any CRC within a HDLC frame. The criteria for ’valid frame’ indication is updated accordingly (refer to RSTA description).
Bit ’RCRC’ is ignored.

RCRC **Receive CRC Checking Mode**

RCRC=’0’ The received checksum is evaluated, but NOT forwarded to the receive FIFO.

RCRC=’1’ The received checksum (2 or 4 bytes) is evaluated and forwarded to the receive FIFO as data.

RADD **Receive Address Forward to RFIFO**

This bit is only valid
– if an HDLC sub-mode with address field support is selected (Automode, Address Mode 2, Address Mode 1)
– in SS7 mode

RADD=’0’ The received HDLC address field (either 8 or 16 bit, depending on bit ’ADM’) is evaluated, but NOT forwarded to the receive FIFO.
In SS7 mode, the signaling unit fields ’FSN’ and ’BSN’ are NOT forwarded to the receive FIFO.

RADD=’1’ The received HDLC address field (either 8 or 16 bit, depending on bit ’ADM’) is evaluated and forwarded to the receive FIFO.
In SS7 mode, the signaling unit fields ’FSN’ and ’BSN’ are forwarded to the receive FIFO.

RFTH(1:0) **Receive FIFO Threshold**

This bit field defines the level up to which the SCC receive FIFO is filled with valid data before an ’RPF’ interrupt is generated.
(In case of a ’frame end’ condition the PASSAT notifies the CPU immediately, disregarding this threshold.)

RFTH(1:0)	Threshold level in number of data bytes.
’00’	32 byte
’01’	16 byte
’10’	4 byte
’11’	2 byte

Register 5-29 PREAMB
Preamble Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$1E_H$**	**$6E_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Preamble Pattern							
	PRE(7:0)							

PRE(7:0) Preamble

This bit field determines the preamble pattern which is send out during preamble transmission.

Note: In HDLC-mode, zero-bit insertion is disabled during preamble transmission.

Register 5-30 ACCM0
PPP ASYNC Control Character Map 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**20_H**	**70_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character Control Map 07..00							
	07	06	05	04	03	02	01	00

Register 5-31 ACCM1
PPP ASYNC Control Character Map 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**21_H**	**71_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character Control Map 0F..08							
	0F	0E	0D	0C	0B	0A	09	08

Register 5-32 ACCM2
PPP ASYNC Control Character Map2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**22_H**	**72_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character Control Map 17..10							
	17	16	15	14	13	12	11	10

Register 5-33 ACCM3
PPP ASYNC Control Character Map 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**23_H**	**73_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character Control Map 1F..18							
	1F	1E	1D	1C	1B	1A	19	18

ACCM **ASYNC Character Control Map**

This bit field is valid in HDLC octet-synchronous PPP mode only:

Each bit selects the corresponding character (indicated as hex value $1F_H..00_H$ in the register description table) as control character which has to be mapped into the transmit data stream.

Register 5-34 UDAC0
User Defined PPP ASYNC Control Character Map 0

CPU Accessibility: **read/write**

Reset Value: **$7E_H$**

	Channel A	Channel B
Offset Address:	**24_H**	**74_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character 0							
	AC0							

Register 5-35 UDAC1
User Defined PPP ASYNC Control Character Map 1

CPU Accessibility: **read/write**

Reset Value: **$7E_H$**

	Channel A	Channel B
Offset Address:	**25_H**	**75_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character 1							
	AC1							

Register 5-36 UDAC2
User Defined PPP ASYNC Control Character Map 2

CPU Accessibility: **read/write**

Reset Value: **$7E_H$**

	Channel A	Channel B
Offset Address:	**26_H**	**76_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character 2							
	AC2							

Register 5-37 UDAC3
User Defined PPP ASYNC Control Character Map 3

CPU Accessibility: **read/write**

Reset Value: **$7E_H$**

	Channel A	Channel B
Offset Address:	**27_H**	**77_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	ASYNC Character 3							
	AC3							

AC3..0 **User Defined ASYNC Character Control Map**

This bit field is valid in HDLC octet-synchronous PPP mode only:

These bit fields define user determined characters as control characters which have to be mapped into the transmit data stream.
In register ACCM only characters $00_H..1F_H$ can be selected as control characters. Register UDAC allows to specify any four characters in the range $00_H..FF_H$.
The default value is a $7E_H$ flag which must be always mapped. Thus no additional character is mapped if $7E_H$'s are programed to bit fields AC3...0 (reset value).
($7E_H$ is mapped automatically, even if not defined via a AC bit field.)

Register 5-38 TTSA0
Transmit Time Slot Assignment Register 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**28_H**	**78_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
					Tx Clock Shift			
	0	0	0	0	0	TCS(2:0)		

Register 5-39 TTSA1
Transmit Time Slot Assignment Register 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**29_H**	**79_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Tx Time Slot Number							
	TEPCM	TTSN(6:0)						

Register 5-40 TTSA2
Transmit Time Slot Assignment Register 2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2A_H$**	**$7A_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Transmit Channel Capacity							
	TCC(7:0)							

Register 5-41 TTSA3
Transmit Time Slot Assignment Register 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2B_H$**	**$7B_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Transmit Channel Capacity							
	0	0	0	0	0	0	0	TCC8

The following register bit fields allow flexible assignment of bit- or octet-aligned transmit time-slots to the serial channel. For more detailed information refer to chapters "Clock Mode 5a (Time Slot Mode)" on page 3-53 and "Clock Mode 5b (Octet Sync Mode)" on page 3-60.

TCS(2:0) **Transmit Clock Shift**

This bit field determines the transmit clock shift.

TEPCM **Enable PCM Mask Transmit**

This bit selects the additional Transmit PCM Mask (refer to register PCMTX0..PCMTX3):

TEPCM=’0’ Standard time-slot configuration.

TEPCM=’1’ The time-slot width is constant 8 bit, bit fields ’TTSN’ and ’TCS’ determine the offset of the PCM mask and ’TCC’ is ignored. Each time-slot selected via register PCMTX0..PCMTX3 is an active transmit timeslot.

TTSN(6:0) **Transmit Time Slot Number**

This bit field selects the start position of the timeslot in time-slot configuration mode (clock mode 5a/5b):

Offset = 1+TTSN*8 + TCS (1..1024 clocks)

TCC(8:0) **Transmit Channel Capacity**

This bit field determines the transmit time-slot width in standard time-slot configuration (bit TEPCM=’0’):
Number of bits = TCC + 1, (1..512 bits/time-slot)

Register 5-42 RTSA0
Receive Time Slot Assignment Register 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2C_H$**	**$7C_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
					Rx Clock Shift			
	0	0	0	0	0	RCS(2:0)		

Register 5-43 RTSA1
Receive Time Slot Assignment Register 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2D_H$**	**$7D_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Rx Time Slot Number							
	REPCM	RTSN(6:0)						

Register 5-44 RTSA2
Receive Time Slot Assignment Register 2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2E_H$**	**$7E_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Channel Capacity							
	RCC(7:0)							

Register 5-45 RTSA3
Receive Time Slot Assignment Register 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$2F_H$**	**$7F_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Channel Capacity							
	0	0	0	0	0	0	0	RCC8

The following register bit fields allow flexible assignment of bit- or octet-aligned receive time-slots to the serial channel. For more detailed information refer to chapters "Clock Mode 5a (Time Slot Mode)" on page 3-53 and "Clock Mode 5b (Octet Sync Mode)" on page 3-60.

RCS(2:0) **Receive Clock Shift**

This bit field determines the receive clock shift.

REPCM **Enable PCM Mask Receive**

This bit selects the additional Receive PCM Mask (refer to register PCMRX0..PCMRX3):

REPCM='0' Standard time-slot configuration.

REPCM='1' The time-slot width is constant 8 bit, bit fields 'RTSN' and 'RCS' determine the offset of the PCM mask and 'RCC' is ignored. Each time-slot selected via register PCMRX0..PCMRX3 is an active receive timeslot.

RTSN(6:0) **Receive Time Slot Number**

This bit field selects the start position of the timeslot in time-slot configuration mode (clock mode 5a/5b):

Offset = 1+RTSN*8 + RCS (1..1024 clocks)

RCC(8:0) **Receive Channel Capacity**

This bit field determines the receive time-slot width in standard time-slot configuration (bit REPCM='0'):
Number of bits = RCC + 1, (1..512 bits/time-slot)

Register 5-46 PCMTX0
PCM Mask Transmit Direction Register 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**30_H**	**80_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	PCM Mask for Transmit Direction							
	T07	T06	T05	T04	T03	T02	T01	T00

Register 5-47 PCMTX1
PCM Mask Transmit Direction Register 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**31_H**	**81_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	PCM Mask for Transmit Direction							
	T15	T14	T13	T12	T11	T10	T09	T08

Register 5-48 PCMTX2
PCM Mask Transmit Direction Register 2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**32_H**	**82_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	PCM Mask for Transmit Direction							
	T23	T22	T21	T20	T19	T18	T17	T16

Register 5-49 PCMTX3
PCM Mask Transmit Direction Register 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**33_H**	**83_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	15	14	13	12	11	10	9	8
	PCM Mask for Transmit Direction							
	T31	T30	T29	T28	T27	T26	T25	T24

PCMTX **PCM Mask for Transmit Direction**

This bit field is valid in HDLC clock mode 5 only and the PCM mask must be enabled via bit ’TEPCM’ in register TTSA1.

Each bit selects one of 32 (8-bit) transmit time-slots. The offset of time-slot zero to the frame sync pulse can be programmed via register TTSA1 bit field ’TTSN’.

Register 5-50 PCMRX0
PCM Mask Receive Direction Register 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**34_H**	**84_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	PCM Mask for Receive Direction							
	T07	T06	T05	T04	T03	T02	T01	T00

Register 5-51 PCMRX1
PCM Mask Receive Direction Register 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**35_H**	**85_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	15	14	13	12	11	10	9	8
	PCM Mask for Receive Direction							
	T15	T14	T13	T12	T11	T10	T09	T08

Register 5-52 PCMRX2
PCM Mask Receive Direction Register 2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**36_H**	**86_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	PCM Mask for Receive Direction							
	T23	T22	T21	T20	T19	T18	T17	T16

Register 5-53 PCMRX3
PCM Mask Receive Direction Register 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**37_H**	**87_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	15	14	13	12	11	10	9	8
	PCM Mask for Receive Direction							
	T31	T30	T29	T28	T27	T26	T25	T24

PCMRX **PCM Mask for Receive Direction**

This bit field is valid in HDLC clock mode 5 only and the PCM mask must be enabled via bit ’REPCM’ in register RTSA1.

Each bit selects one of 32 (8-bit) receive time-slots. The offset of time-slot zero to the frame sync pulse can be programmed via register RTSA1 bit field ’RTSN’.

Register 5-54 BRRL
Baud Rate Register (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**38_H**	**88_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Baud Rate Generator Factor N							
	0	0	BRN(5:0)					

Register 5-55 BRRH
Baud Rate Register (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**39_H**	**89_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Baud Rate Generator Factor M							
	0	0	0	0	BRM(3:0)			

BRM(3:0) **Baud Rate Factor 'M'**

BRN(5:0) **Baud Rate Factor 'N'**

These bit fields determine the division factor of the internal baud rate generator. The baud rate generator input clock and the usage of baud rate generator output depends on the selected clock mode.
The division factor k is calculated by:

$$k = (N + 1) \times 2^{M}$$

with M=0..15 and N=0..63.

$$f_{BRG} = f_{in} / k$$

Register 5-56 TIMR0
Timer Register 0

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3A_H$**	**$8A_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Timer Value							
	TVALUE(7:0)							

Register 5-57 TIMR1
Timer Register 1

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3B_H$**	**$8B_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Timer Value							
	TVALUE(15:0)							

Register 5-58 TIMR2
Timer Register 2

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3C_H$**	**$8C_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Timer Value							
	TVALUE(23:16)							

Register 5-59 TIMR3
Timer Register 3

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3D_H$**	**$8D_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Timer Configuration							
	SRC	0	0	TMD	0	CNT(2:0)		

SRC **Clock Source (valid in clock mode 5 only)**

This bit selects the clock source of the internal timer:

SRC = ’0’ The timer is clocked by the effective transmit clock.

SRC = ’1’ The timer is clocked by the frame-sync synchronization signal supplied via the FSC pin in clock mode 5.

TMD **Timer Mode**

This bit must be set to ’1’ if HDLC Automode operation is selected. In all other protocol modes it must remain ’0’:

TMD=’0’ The timer is controlled by the CPU via access to registers CMDRL and TIMR0..TIMR3.
The timer can be started any time by setting bit ’STI’ in register CMDRL. After the timer has expired it generates a timer interrupt. The timer can be stopped any time by setting bit ’TRES’ in register CMDRL to ’1’.

TMD=’1’ The timer is used by the PASSAT for protocol specific time-out and retry transactions in HDLC Automode.

CNT(2:0) **Counter**

The meaning of this bit field depends on the selected protocol mode.
In HDLC Automode, with bit TMD=’1’:

- Retry Counter (in HDLC protocol known as ’N2’):
  Bit field ’CNT’ indicates the number of S-Command frames (with poll bit set) which are transmitted autonomously by PASSAT after every expiration of the time out period ’t’ (determined by ’TVALUE’), in case an I-Frame gets not acknowledged by the opposite station. The maximum value is 6 S-command frames. If ’CNT’ is set to ’7’, the number of S-commands is unlimited in case of no acknowledgement.

In all other modes, with bit TMD=’0’:

- Restart Counter :
  Bit field ’CNT’ indicates the number of automatic restarts which are performed by PASSAT after every expiration of the time-out period ’t’, in case the timer is not stopped by setting bit ’TRES’ in register CMDRL to ’1’. The maximum value is 6 restarts. If ’CNT’ is set to ’7’, a timer interrupt is generated periodically with time period ’t’ determined by bit field ’TVALUE’.

TVALUE (23:0) **Timer Expiration Value**

This bit field determines the timer expiration period ’t’:

$$t = (TVALUE + 1) \cdot CP$$

(’CP’ is the clock period, depending on bit ’SRC’.)

Register 5-60 XAD1
Transmit Address 1 Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3E_H$**	**$8E_H$**

typical usage: written by CPU; read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Transmit Address (high)							
	XAD1 (high byte)						0	XAD1_0
	or XAD1 (COMMAND)							

Register 5-61 XAD2
Transmit Address 2 Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$3F_H$**	**$8F_H$**

typical usage: written by CPU; read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Transmit Address (low)							
	XAD2 (low byte)							
	or XAD2 (RESPONSE)							

XAD1 and XAD2 bit fields are valid in HDLC modes with automatic address field handling only (Automode, Address Mode 1, Non-Automode). They can be programmed with one individual address byte which is inserted automatically into the address field (8 or 16 bit) of a HDLC transmit frame. The function depends on the selected protocol mode and address field size (bit 'ADM' in register CCR2L).

XAD1 **Transmit Address 1**

– 2-byte address field:
Bit field XAD1 constitutes the high byte of the 2-byte address field. Bit 1 must be set to '0'. According to the ISDN LAP-D protocol, bit 1 is interpreted as the C/R (COMMAND/RESPONSE) bit. This bit is manipulated automatically by PASSAT according to the setting of bit 'CRI' in register RAH1. The following is the C/R value (on bit 1), when:
- transmitting COMMANDs: '1' (if 'CRI'='1') ; '0' (if 'CRI'='0')
- transmitting RESPONSEs: '0' (if 'CRI'='1') ; '1' (if 'CRI'='0')
(In ISDN LAP-D, the high byte is known as 'SAPI'.)
In accordance with the HDLC protocol, bit 'XAD1_0' should be set to '0', to indicate that the address field contains (at least) one more byte.

– 1-byte address field:
According to the X.25 LAP-B protocol, XAD1 is the address of a 'COMMAND' frame.

XAD2 **Transmit Address 2**

– 2-byte address field:
Bit field XAD2 constitutes the low byte of the 2-byte address field.
(In ISDN LAP-D, the low byte is known as 'TEI'.)

– 1-byte address field:
According to the X.25 LAP-B protocol, XAD2 is the address of a 'RESPONSE' frame.

Register 5-62 RAL1
Receive Address 1 Low Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**40_H**	**90_H**

typical usage: written by CPU; read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Address 1 (low)							
	RAL1							
	RAL1							

Register 5-63 RAH1
Receive Address 1 High Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**41_H**	**91_H**

typical usage: written by CPU; read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Address 1 (high)							
	RAH1						CRI	RAH1_0
	or RAH1							

Register 5-64 RAL2
Receive Address 2 Low Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**42_H**	**92_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Address 2 (low)							
	RAL2							

Register 5-65 RAH2
Receive Address 2 High Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**43_H**	**93_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Address 2 (high)							
	RAH2							

In operating modes that provide address recognition, the high/low byte of the received address is compared with the individually programmable values in register RAH2/ RAL2/RAH1/RAL1.
This addresses can be masked on a per bit basis by setting the corresponding bits in registers AMRAL1/AMRAH1/AMRAL2/AMRAH2 to allow extended broadcast address recognition. This feature is applicable to all HDLC sub-modes with address recognition.

RAH1 **Receive Address 1 Byte High**

In HDLC Automode bit '1' is reserved for 'CRI' (Command Response Interpretation). In all other modes RAH1 is an 8 bit address.

CRI **Command/Response Interpretation**

The setting of this bit effects the meaning of the 'C/R' bit in the receive status byte (RSTA). This status bit 'C/R' should be interpreted after reception as follows:

'0' (if 'CRI'='1') ; '1' (if 'CRI'='0') : COMMAND received
'1' (if 'CRI'='1') ; '0' (if 'CRI'='0') : RESPONSE received

Note: If 1-byte address field is selected in HDLC Automode, RAH1 must be set to $0x00_H$.

RAL1 **Receive Address 1 Byte Low**

The general function and its meaning depends on the selected HDLC operating mode:

- **Automode / Address Mode 2 (16-bit address)**
  RAL1 can be programmed with the value of the first individual low address byte.
- **Automode / Address Mode 2 (8-bit address)**
  According to X.25 LAP-B protocol, the address in RAL1 is considered as the address of a 'COMMAND' frame.

RAH2 **Receive Address 2 Byte High**

RAL2 **Receive Address 2 Byte Low**

Value of the second individually programmable high/low address byte. If a 1-byte address field is selected, RAL2 is considered as the address of a 'RESPONSE' frame according to X.25 LAP-B protocol.

Register 5-66 AMRAL1
Mask Receive Address 1 Low Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**44_H**	**94_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Mask Address 1 (low)							
	AMRAL1							

Register 5-67 AMRAH1
Mask Receive Address 1 High Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**45_H**	**95_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Mask Address 1 (high)							
	AMRAH1							

Register 5-68 AMRAL2
Mask Receive Address 2 Low Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**46_H**	**96_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Mask Address 2 (low)							
	AMRAL2							

Register 5-69 AMRAH2
Mask Receive Address 2 High Register

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**47_H**	**97_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Mask Address 2 (high)							
	AMRAH2							

AMRAH2 **Receive Mask Address 2 Byte High**

AMRAL2 **Receive Mask Address 2 Byte Low**

AMRAH1 **Receive Mask Address 1 Byte High**

AMRAL1 **Receive Mask Address 1 Byte Low**

Setting a bit in this registers to ’1’ masks the corresponding bit in registers RAH2/RAL2/RAH1/RAL1. A masked bit position always matches when comparing the received frame address with registers RAH2/RAL2/RAH1/RAL1, allowing extended broadcast mechanism.

bit = ’0’ The dedicated bit position is NOT masked. This bit position in the received address must match with the corresponding bit position in registers RAH2/RAL2/RAH1/RAL1 to accept the frame.

bit = ’1’ The dedicated bit position is masked. This bit position in the received address NEED NOT match with the corresponding bit position in registers RAH2/RAL2/RAH1/RAL1 to accept the frame.

Register 5-70 RLCRL
Receive Length Check Register (Low Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**48_H**	**98_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Length Limit							
	RL(7:0)							

Register 5-71 RLCRH
Receive Length Check Register (High Byte)

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**49_H**	**99_H**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Length Check Control				Receive Length Limit			
	RCE	0	0	0	0	RL(10:8)		

RCE **Receive Length Check Enable**

This bit is valid in HDLC mode only and enables/disables the receive length check function:

RCE = ’0’ No receive length check on received HDLC frames is performed.

RCE = ’1’ The receive length check is enabled. All bytes of a HDLC frame which are transferred to the receive FIFO (depending on the selected protocol sub-mode and receive CRC handling) are counted and checked against the maximum length check limit which is programmed in bit field ’RL’.
A frame exceeding the maximum length is treated as if it were aborted on the receive line (’RME’ interrupt and bit ’RAB’ (receive abort) set in the RSTA byte).
In addition a ’FLEX’ interrupt is generated prior to ’RME’, if enabled.

Note: The Receive Status Byte (RSTA) is part of the frame length checking.

RL(10:0) **Receive Length Check Limit**

This bit-field defines the receive length check limit (32..65536 bytes) if checking is enabled via bit ’RCE’:

RL(10:0) The receive length limit is calculated by:

$$Limit = (RL + 1) \cdot 32$$

Register 5-72 ISR0
Interrupt Status Register 0

CPU Accessibility: **read only**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**50_H**	**$A0_H$**

typical usage: updated by PASSAT
read and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	ISR0							
	RDO	RFO	PCE	RSC	RPF	RME	RFS	FLEX

Register 5-73 ISR1
Interrupt Status Register 1

CPU Accessibility: **read only**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**51_H**	**$A1_H$**

typical usage: updated by PASSAT
read and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	ISR1							
	TIN	CSC	XMR	XPR	ALLS	XDU	SUEX	0

Register 5-74 ISR2
Interrupt Status Register 2

CPU Accessibility: **read only**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**52_H**	**$A2_H$**

typical usage: updated by PASSAT
read and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	ISR2							
	0	0	0	0	0	0	PLLA	CDSC

RDO **Receive Data Overflow Interrupt**

This bit is set to ’1’, if receive data of the current frame got lost because of a SCC receive FIFO full condition. However the rest of the frame is received and discarded as long as the receive FIFO remains full and is stored as soon as FIFO space is available again. The receive status byte (RSTA) of such a frame contains an ’RDO’ indication. In DMA operation the ’RDO’ indication is also set in the receive byte count register RBCH.

RFO **Receive FIFO Overflow Interrupt**

This bit is set to ’1’, if the SCC receive FIFO is full and a complete frame must be discarded. This interrupt can be used for statistical purposes, indicating that the host was not able to service the SCC receive FIFO quickly enough, e.g. due to high bus latency.

PCE **Protocol Error Interrupt**

This bit is valid in HDLC Automode only.

It is set to ’1’, if the receiver has detected a protocol error, i.e. one of the following events occured:

- an S- or I-frame was received with wrong N(R) counter value;
- an S-frame containing an Information field was received.

RSC **Receive Status Change Interrupt**

This bit is valid in HDLC Automode only.
It is set to ’1’, if a status change of the remote station receiver has been detected by receiving a S-frame with receiver ready (RR) or receiver not ready (RNR) indication. Because only a status change is indicated via this interrupt, the current status can be evaluated by reading bit ’RRNR’ in status register STARH.

RPF **Receive Pool Full Interrupt**

This bit is set to ’1’ if the RFIFO threshold level, set with bit field ’RFTH(1:0)’ in register CCR3H, is reached. Default threshold level is 32 data bytes.

RME **Receive Message End Interrupt**

This bit set to ’1’ indicates that the reception of one message is completed, i.e. either
– one message which fits into RFIFO not exceeding the receive FIFO threshold, or
– the last part of a message, all in all exceeding the receive FIFO threshold

is stored in the RFIFO.
The complete message length can be determined by reading the RBCL/RBCH registers. The number of bytes stored in RFIFO is given by the 5, 4, 2 or 1 least significant bits of register RBCL, depending on the selected RFIFO threshold (bit field ’RFTH(1:0)’ in register CCR3H).
Additional frame status information is available in the RSTA byte, stored in the RFIFO as the last byte of each frame.

RFS **Receive Frame Start Interrupt**

This bit is set to ’1’, if the beginning of a valid frame is detected by the receiver. A valid frame start is detected either if a valid address field is recognized (in all operating modes with address recognition) or if a start flag is recognized (in all operating modes with no address recognition).

FLEX **Frame Length Exceeded Interrupt**

This bit is set to ’1’, if the frame length check feature is enabled and the current received frame is aborted because the programmed frame length limit was exceeded (refer to registers RLCRL/RLCRH for detailed description).

TIN **Timer Interrupt**

This bit is set to ’1’, if the internal timer was activated and has expired (refer also to description of timer registers TIMR0..TIMR3).

CSC **$\overline{\text{CTS}}$ Status Change**

This bit is set to ’1’, if a transition occurs on signal $\overline{\text{CTS}}$. The current state of signal $\overline{\text{CTS}}$ is monitored by status bit ’CTS’ in status register STARL.

XMR **Transmit Message Repeat**

This bit is set to ’1’, if transmission of the last frame has to be repeated (by software), because

- the SCC has received a negative acknowledge to an I-frame (in HDLC Automode operation);
- a collision occured after at least 31 bytes of data have been completely sent out, i.e. automatic re-transmission cannot be performed by the SCC;
- $\overline{\text{CTS}}$ signal was deasserted after at least 31 bytes of data have been completely sent out.

Note: For easy recovery from a collision event (in bus configuration only), the SCC transmit FIFO should not contain more than one complete frame. This can be achieved by using the ’ALLS’ interrupt to control the corresponding transmit channel forwarding a new frame on all sent (ALLS) event only.

XPR **Transmit Pool Ready Interrupt**

This bit is set to ’1’, if a transmitter reset command was executed successfully (command bit ’XRES’ in register CMDRL) and whenever the XFIFO is able to accept new transmit data again.

An ’XPR’ interrupt is not generated, if no sufficient transmit clock is available (depending on the selected clock mode).

ALLS **ALL Sent Interrupt**

This bit is set to ’1’:

- if the last bit of the current HDLC frame is sent out via pin TxD and no further frame is stored in the SCC transmit FIFO, i.e. the transmit FIFO is empty (Address Mode 2/1/0);
- if an I-frame is sent out completely via pin TxD and either a valid acknowledge S-frame has been received or a time-out condition occured because no valid acknowledge S-frame has been received in time (Automode).

XDU **Transmit Data Underrun Interrupt**

This bit is set to ’1’, if the current frame was terminated by the SCC with an abort sequence, because neither a ’frame end’ indication was detected in the FIFO (to complete the current frame) nor more data is available in the SCC transmit FIFO.

Note: The transmitter is stopped if this condition occurs. The XDU condition MUST be cleared by reading register ISR1, thus bit ’XDU’ should not be masked via register IMR1.

SUEX **Signalling Unit Counter Exceeded Interrupt**

This bit is set to ’1’, if 256 correct or incorrect SU’s have been received and the internal counter is reset to 0.

PLLA **DPLL Asynchronous Interrupt**

This bit is only valid, if the receive clock is derived from the internal DPLL and FM0, FM1 or Manchester data encoding is selected (depending on the selected clock mode and data encoding mode). It is set to ’1’ if the DPLL has lost synchronization. Reception is disabled until synchronization has been regained again. If the transmitter is supplied with a clock derived from the DPLL, transmission is also interrupted.

CDSC **Carrier Detect Status Change Interrupt**

This bit is set to ’1’, if a state transition has been detected at signal CD. Because only a state transition is indicated via this interrupt, the current status can be evaluated by reading bit ’CD’ in status register STARH.

Register 5-75 IMR0
Interrupt Mask Register 0

CPU Accessibility: **read/write**

Reset Value: **FF_H**

	Channel A	Channel B
Offset Address:	**54_H**	**$A4_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	IMR0							
	RDO	RFO	PCE	RSC	RPF	RME	RFS	FLEX

Register 5-76 IMR1
Interrupt Mask Register 1

CPU Accessibility: **read/write**

Reset Value: **FF_H**

	Channel A	Channel B
Offset Address:	**55_H**	**$A5_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	IMR1							
	TIN	CSC	XMR	XPR	ALLS	XDU	SUEX	1

Register 5-77 IMR2
Interrupt Mask Register 2

CPU Accessibility: **read/write**

Reset Value: **03_H**

	Channel A	Channel B
Offset Address:	**56_H**	**$A6_H$**

typical usage: written by CPU;
read and evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	IMR2							
	0	0	0	0	0	0	PLLA	CDSC

(IM) **Interrupt Mask Bits**

Each SCC interrupt event can generate an interrupt signal indication via pin INT/$\overline{\text{INT}}$. Each bit position of registers IMR0..IMR2 is a mask for the corresponding interrupt event in the interrupt status registers ISR0..ISR2. Masked interrupt events never generate an interrupt indication via pin INT/$\overline{\text{INT}}$.

bit = ’0’ The corresponding interrupt event is NOT masked and will generate an interrupt indication via pin INT/$\overline{\text{INT}}$.

bit = ’1’ The corresponding interrupt event is masked and will NEITHER generate an interrupt vector NOR an interrupt indication via pin INT/$\overline{\text{INT}}$.

Moreover, masked interrupt events are:

- not displayed in the interrupt status registers ISR0..ISR2 if bit ’VIS’ in register CCR0L is programmed to ’0’.
- displayed in interrupt status registers ISR0..ISR2 if bit ’VIS’ in register CCR0L is programmed to ’1’.

Note: After RESET, all interrupt events are masked.

For detailed interrupt event description refer to the corresponding bit position in registers ISR0..ISR2.

Register 5-78 RSTA
Receive Status Byte

CPU Accessibility: **read/write**

Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**58_H**	**$A8_H$**

typical usage: written by PASSAT to RFIFO;
read from RFIFO and evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Receive Status Byte							
	VFR	RDO	CRCOK	RAB	HA(1:0)/ SU(1:0)		C/R	LA

The Receive Status Byte ’RSTA’ contains comprehensive status information about the last received frame (HDLC/PPP).

The SCC attaches this status byte to the receive data and thus it should be read from the RFIFO.

In HDLC/PPP modes the RSTA value can optionally be read from this register address. In extended transparent mode this status field does not apply.

VFR **Valid Frame**

Determines whether a valid frame has been received.

VFR=’0’ The received frame is invalid.
An invalid frame is either a frame which is not an integer number of 8 bits (n * 8 bits) in length (e.g. 25 bits), or a frame which is too short, taking into account the operation mode selected via CCR2L (MDS1, MDS0, ADM) and the selected CRC algorithm (CCR1L:C32) as follows:

for CCR3H:DRCRC = ’0’ (CRC reception enabled):
- automode / address mode 2 (16-bit address)
  4 bytes (CRC-CCITT) or 6 (CRC-32)
- automode / address mode 2 (8-bit address)
  3 bytes (CRC-CCITT) or 5 (CRC-32)
- address mode 1:
  3 bytes (CRC-CCITT) or 5 (CRC-32)
- address mode 0:
  2 bytes (CRC-CCITT) or 4 (CRC-32)

for CCR3H:DRCRC = ’1’ (CRC reception disabled):
- automode / address mode 2 (16-bit address):
  2 bytes
- automode / address mode 2 (8-bit address):
  1 byte
- address mode 1:
  1 byte
- address mode 0:
  1 byte

Note: Shorter frames are not reported at all.

VFR=’1’ The received frame is valid.

RDO **Receive Data Overflow**

RDO=’0’ No receive data overflow has occurred.

RDO=’1’ A data overflow has occurred during reception of the frame. Additionally, an interrupt can be generated (refer to ISR0:RDO/IMR0:RDO).

CRCOK **CRC Compare/Check**

CRCOK=’0’ CRC check failed, received frame contains errors.

CRCOK=’1’ CRC check OK; the received frame does not contain CRC errors.

C/R **Command/Response**

Significant only if 2-byte address mode has been selected.
Value of the C/R bit (bit 1 of high address byte) in the received frame. The interpretation depends on the setting of the 'CRI' bit in the RAH1 register (See "RAH1" on page 180.).

LA **Low Byte Address Compare**

Significant in automode and address mode 2 only.
The low byte address of a 2-byte address field, or the single address byte of a 1-byte address field is compared with two addresses (RAL1, RAL2).

LA='0' RAL2 has been recognized.

LA='1' RAL1 has been recognized.

According to the X.25 LAPB protocol, RAL1 is interpreted as the address of a COMMAND frame and RAL2 is interpreted as the address of a RESPONSE frame.

5.2.3 Channel Specific DMA Registers

Each register description is organized in three parts:

- a head with general information about reset value, access type (read/write), channel specific offset address and usual handling;
- a table containing the bit information (name of bit positions);
- a section containing the detailed description of each bit.

Register 5-79 XBCL
Transmit Byte Count (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$B8_H$**	**$D2_H$**

typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	XBC(7:0)							

Register 5-80 XBCH
Transmit Byte Count (High Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$B9_H$**	**$D3_H$**

typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	XME	XF	XIF	0	XBC(11:8)			

XBC (11:0) **Transmit Byte Count**

This register is used in DMA Mode only, to program the length (1…4096 bytes) of the next frame to be transmitted. The length of the block in number of bytes is:

$$\text{Length} = \text{XBC} + 1$$

This allows the PASSAT to request the correct amount of DMA cycles after an ’XF’ or’ XIF’ command.

XME **Transmit Message End Command**

Only valid in external DMA controller mode.
This bit is identical to ’XME’ command bit (refer to register "CMDRL" on page 5-132).

XF **Transmit Frame Command**

Only valid in external DMA controller mode.
This bit is identical to ’XF’ command bit (refer to register "CMDRL" on page 5-132).

XIF **Transmit I-Frame Command**

Only valid in external DMA controller mode.
This bit is identical to ’XIF’ command bit (refer to register "CMDRL" on page 5-132).

Register 5-81 RMBSL
Receive Maximum Buffer Size (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$C4_H$**	**DE_H**

typical usage: written by CPU, evaluated by PASSAT

Bit	7	6	5	4	3	2	1	0
	Receive Maximum Buffer Size							
	RMBS(7:0)							

Register 5-82 RMBSH
Receive Maximum Buffer Size (High Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$C5_H$**	**DF_H**

typical usage: written by CPU, evaluated by PASSAT

Bit	15	14	13	12	11	10	9	8
	Receive Maximum Buffer Size							
	RE	DRMBS	0	0	RMBS(11:8)			

RE **Receive DMA Enable**

Only valid if external DMA controller support is enabled.
Self-clearing command bit:

RE=’0’ The DMA controller is not set up to forward receive data into a buffer in memory.

RE=’1’ Setting this bit to ’1’ enables the DMA support logic to request the external DMA controller to transfer receive data when available in RFIFO.

DRMBS **Disable Receive Maximum Buffer Size (RMBS) Check**

Only valid if external DMA controller support is enabled.

DRMBS=’0’ Evaluation of bit field RMBS(11:0) is enabled.

DRMBS=’1’ Evaluation of bit field RMBS(11:0) is disabled.

RMBS(11:0) **Receive Maximum Buffer Size**

Only valid if external DMA controller support is enabled.
The size of the receive buffer in host memory can be set up in this bit field to ensure that request for DMA transfers are inhibited when the maximum buffer size is reached. An RBF interrupt is generated (if unmasked) to inform the CPU. If the external DMA controller supports this function, it can be disabled by setting bit ’DRMBS’ to ’1’.

Register 5-83 RBCL
Receive Byte Count (Low Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$C6_H$**	**$E0_H$**

typical usage: written by PASSAT, evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	RBC(7:0)							

Register 5-84 RBCH
Receive Byte Count (High Byte)

CPU Accessibility: **read/write**
Reset Value: **00_H**

	Channel A	Channel B
Offset Address:	**$C7_H$**	**$E1_H$**

typical usage: written by PASSAT, evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	RDO	0	0	0	RBC(11:8)			

RBC(11:0) **Receive Byte Count**

This bit field determines the receive byte count (1..4095) of the currently received frame/block.

RDO **RDO Indication**

Only valid in DMA controller mode.
This bit is identical to the ’RDO’ status bit belonging to this frame (see description of register "RSTA" on page 5-197).

5.2.4 Miscellaneous Registers

Register 5-85 VER0
Version Register 0

CPU Accessibility: **read/write**
Reset Value: **83_H**
Offset Address: **EC_H**
typical usage: evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Manufacturer Code							Fix '1'
	VER(7:0)							

Register 5-86 VER1
Version Register 1

CPU Accessibility: **read/write**
Reset Value: **$F0_H$**
Offset Address: **ED_H**
typical usage: evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Device Code (bits 3 .. 0)				Manufacturer Code			
	VER(15:8)							

Register 5-87 VER2
Version Register 2

CPU Accessibility: **read/write**
Reset Value: **05_H**
Offset Address: **EE_H**
typical usage: evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Device Code (bits 11 .. 4)							
	VER(23:16)							

Register 5-88 VER3
Version Register 3

CPU Accessibility: **read/write**
Reset Value: **10_H**
Offset Address: **EF_H**
typical usage: evaluated by CPU

Bit	7	6	5	4	3	2	1	0
	Version Number				Device Code (bits 15 .. 12)			
	VER(31:24)							

VER(31:0) Version Register

Identical to 32 bit boundary scan ID string.

The 32 bit string consists of the bit fields:

VER(31:28)	1_H	Version Number
VER(27:12)	$005F_H$	Device Code
VER(11:0)	083_H	Manufacturer Code (LSB fixed to '1')

6 Programming

6.1 Initialization

After Reset the CPU has to write a minimum set of registers and an optional set depending on the required features and operating modes.

First, the following initialization steps must be taken:

- Select serial protocol mode (refer to Table 4-1 "Protocol Mode Overview" on page 4-80),
- Select encoding of the serial data (refer to **Chapter 3.2.13** "Data Encoding" on page 3-71),
- Program the output characteristics of
  - pin TxD (selected with bit 'ODS' in "Channel Configuration Register 1 (Low Byte)" on page 5-139) and
  - interrupt pin INT/$\overline{\text{INT}}$ (selected with bit field 'IPC(1:0)' in "Global Mode Register" on page 5-109),
- Choose a clock mode (refer to Table 3-1 "Overview of Clock Modes" on page 3-44).
- Power-up the oscillator unit (with or without shaper) by re-setting bit GMODE:OSCPD to '0', if appropriate (GMODE:DSHP='0' enables the shaper).

The clock mode must be set before power-up (CCR0H.PU). The CPU may switch the PASSAT between power-up and power-down mode. This has no influence upon the contents of the registers, i.e. the internal state remains stored. In power-down mode however, all internal clocks are disabled, no interrupts from the corresponding channel are forwarded to the CPU. This state can be used as a standby mode, when the channel is (temporarily) not used, thus substantially reducing power consumption.

The PASSAT should usually be initialized in Power-Down mode.

The need for programming further registers depends on the selected features (serial mode, clock mode specific features, operating mode, address mode, user demands).

6.2 Interrupt Mode

6.2.1 Data Transmission (Interrupt Driven)

In transmit direction 2 × 32 byte FIFO buffers (transmit pools) are provided for each channel. After checking the XFIFO status by polling the Transmit FIFO Write Enable bit (bit 'XFW' in STARL register) or after a Transmit Pool Ready ('XPR') interrupt, up to 32 bytes may be entered by the CPU into the XFIFO.

The transmission of a packet can be started by issuing an 'XF' or 'XIF' command via the CMDRL register. If enabled, a specified number of preambles (refer to registers CCR2H and PREAMB) are sent out optionally before transmission of the current packet starts.

If the transmit command does not include an end of message indication (CMDRL.XME), PASSAT will repeatedly request for the next data block by means of an 'XPR' interrupt as soon as no more than 32 bytes are stored in the XFIFO, i.e. a 32-byte pool is accessible to the CPU.

This process will be repeated until the CPU indicates the end of message per 'XME' command, after which packet transmission is finished correctly by appending the CRC and closing flag sequence. Consecutive packets may be transmitted as back-to-back packets and may even share a flag (enabled via CCR1L.SFLG), if service of XFIFO is quick enough.

In case no more data is available in the XFIFO prior to the arrival of the end-of-message indiction ('XME'), the transmission of the packet is terminated with an abort sequence and the CPU is notified per interrupt (ISR1.XDU, transmit data underrun). The packet may also be aborted per software at any time (CMDRL.XRES).

The data transmission sequence, from the CPU's point of view, is outlined in **Figure 6-1**.

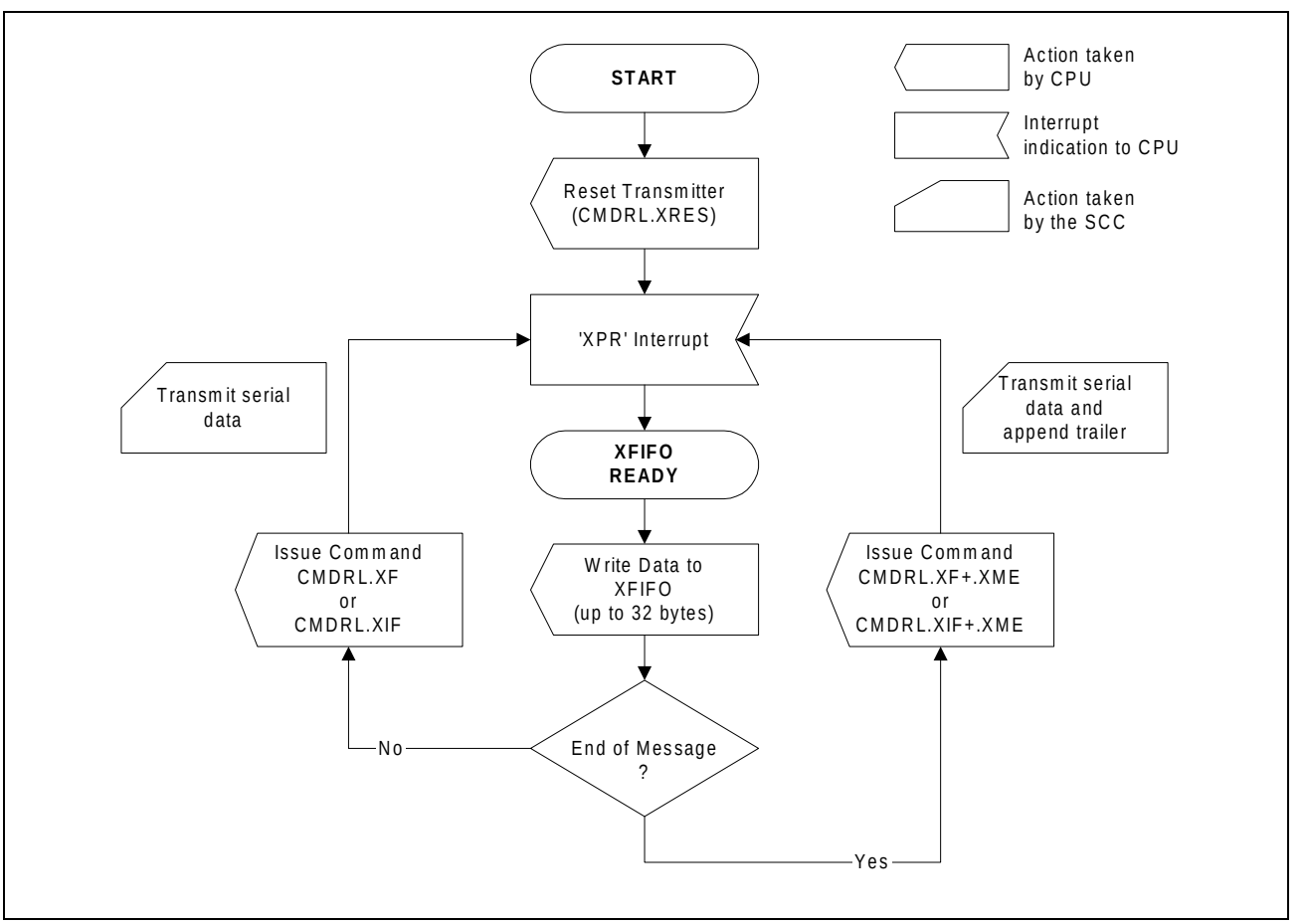


Figure 6-1 Interrupt Driven Data Transmission (Flow Diagram)

6.2.2 Data Reception (Interrupt Driven)

Also 2 × 32 byte FIFO buffers (receive pools) are provided for each channel in receive direction.

There are different interrupt indications concerned with the reception of data:

- 'RPF' (Receive Pool Full) interrupt, indicating that a specified number of bytes (limited with the receive FIFO threshold in register CCR3H, bit field 'RFTH(1..0)'; default is 32 bytes) can be read from RFIFO and the received message is not yet complete.
- 'RME' (Receive Message End) interrupt, indicating that the reception of one message is completed, i.e. either
  - one message which fits into RFIFO not exceeding the receive FIFO threshold, or
  - the last part of a message, all in all exceeding the receive FIFO threshold
  is stored in the RFIFO.

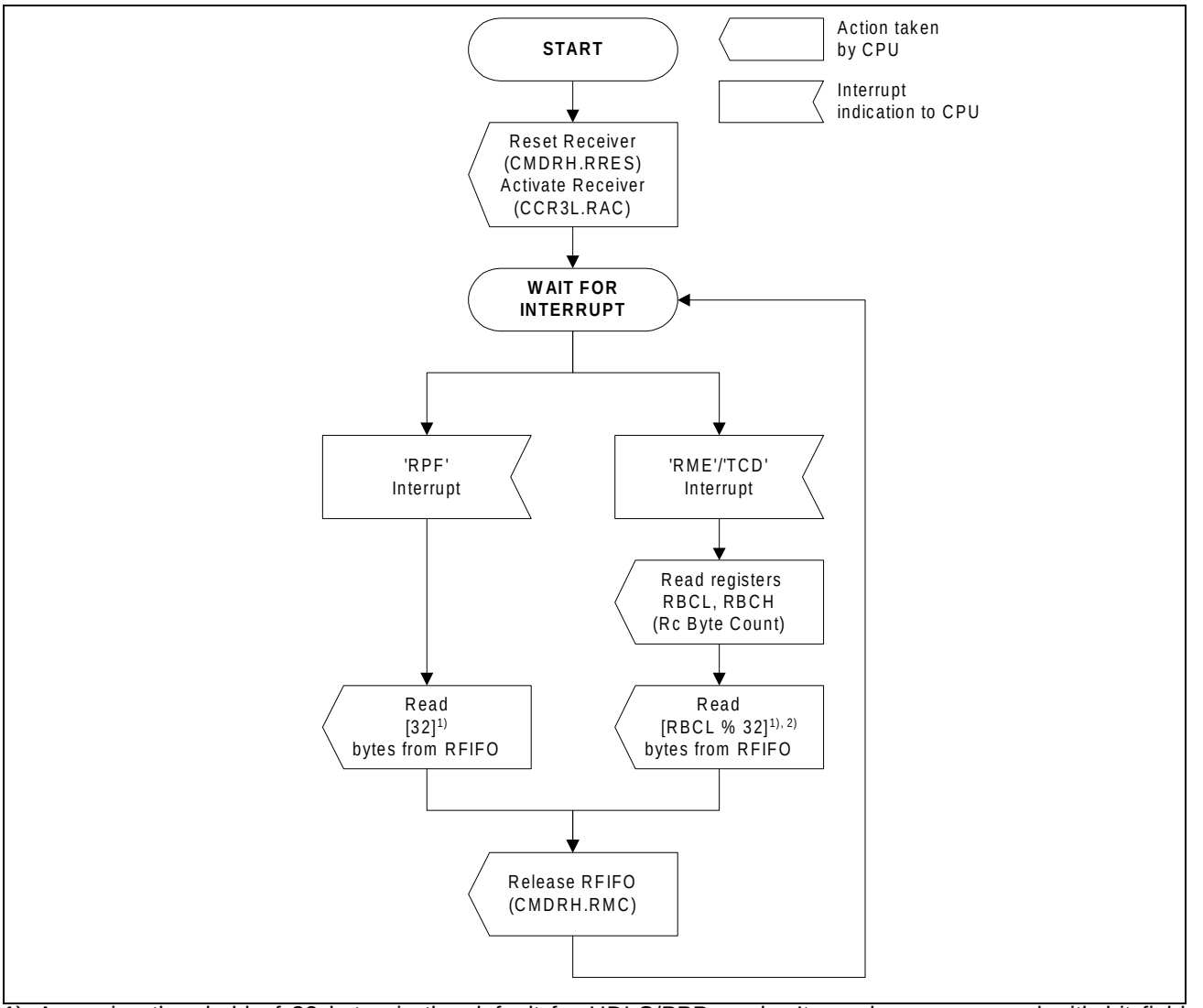
In addition to the message end ('RME') interrupt the following information about the received packet is stored by PASSAT in special registers and/or RFIFO:

Table 6-1 Status Information after RME interupt

Status Information	Location
Length of received message	registers RBCH, RBCL
CRC result (good/bad)	RSTA register (or last byte of received data)
Valid frame (yes/no)	RSTA register (or last byte of received data)
ABORT sequence recognized (yes/no)	RSTA register (or last byte of received data)
Data overflow (yes/no)	RSTA register (or last byte of received data)
Results from address comparison (with automatic address handling)	RSTA register (or last byte of received data)
Type of frame (COMMAND/RESPONSE) (with automatic address handling)	RSTA register (or last byte of received data)
Type of Signaling Unit (in SS7 mode)	RSTA register (or last byte of received data)

Note: After the received data has been read from the RFIFO, this must be explicitly acknowledged by the CPU issuing an 'RMC' (Receive Message Complete) command. The CPU has to handle the 'RPF' interrupt before the complete 2 x 32-byte FIFO is filled up with receive data which would cause a "Receive Data Overflow" condition.

The data reception sequence, from the CPU's point of view, is outlined in **Figure 6-2**.



1) A receive threshold of 32 bytes is the default for HDLC/PPP mode. It can be programmed with bit field RFTH(1:0) in register .
2) The number of bytes stored in RFIFO can be determined by evaluating the lower bits in register (depending on the selected receive threshold RFTH(1:0)).

Figure 6-2 Interrupt Driven Data Reception (Flow Diagram)

6.3 External DMA Supported Mode

The following table provides a definition of terms used in this chapter to describe the operation with external DMA controller support.

Table 6-2 DMA Terminology

Packet	A "Packet" is a connected block of data bytes. If a receive status byte (RSTA) is attached to data bytes, it is also considered as part of the packet.
Buffer	A "Buffer" is a limited space in memory that is reserved for DMA reception/transmission. PASSAT can optionally keep track of predefined (receive) buffer limits and notify the CPU with an appropriate interrupt if this functionality is not provided by the external DMA controller. A packet can go into one single buffer, or it can go fragmented into multiple buffers.
Block	A "Block" is the amount of data that is transfered from the memory to the XFIFO (transmit DMA transfer) or from the RFIFO to the memory. The block size is 32 bytes by default. It can be lowered with the receive FIFO threshold in register CCR3H, bit field 'RFTH(1..0)'.
Bus Cycle	A "Bus Cycle" corresponds to a single byte/word transfer. Multiple bus cycles make up a block transfer.
DMA Transfer	A "DMA Transfer" is the movement of complete buffers and/or packets between the XFIFO/RFIFO and the memory by the external DMA controller.

6.3.1 Data Transmission (With External DMA Support)

Any packet transmission is prepared by initializing the external DMA controller with the transmit buffer start address and writing the packet size in number of bytes to registers XBCL/XBCH.

Now there are two possible scenarios:

- If the prepared transmit buffer in memory contains a complete packet, the start command for DMA transmission is issued by setting bits 'XF' and 'XME' in register XBCH to '1'. The DMA support logic will request the external DMA controller to transfer data into the XFIFO . After the last byte has been transmitted, the protocol machine appends the trailer (e.g. CRC and Flag in HDLC), if applicable. The Transmit DMA Transfer End (TDTE) interrupt is generated (refer to **Figure 6-3**).
- If a transmit packet is distributed over more than one transmit buffer in memory, the 'XF' command (without setting the 'XME' bit) forces PASSAT to request data transfers from the external DMA controller from this buffer. A Transmit DMA Transfer End

(TDTE) interrupt is generated whenever a block of <XBC> bytes is completely transferred. For the last buffer, containing the end of the transmit packet, the 'XF' command is issued together with bit 'XME' set (refer to **Figure 6-4**).

After transmission is complete, the optional generation of the ALLS interrupt indicates that all transmit data has been sent on pin TxD.

Note: In HDLC Automode, the 'XF' command may be replaced by the 'XIF' command in the same register, when transmission of an I-frame is desired.

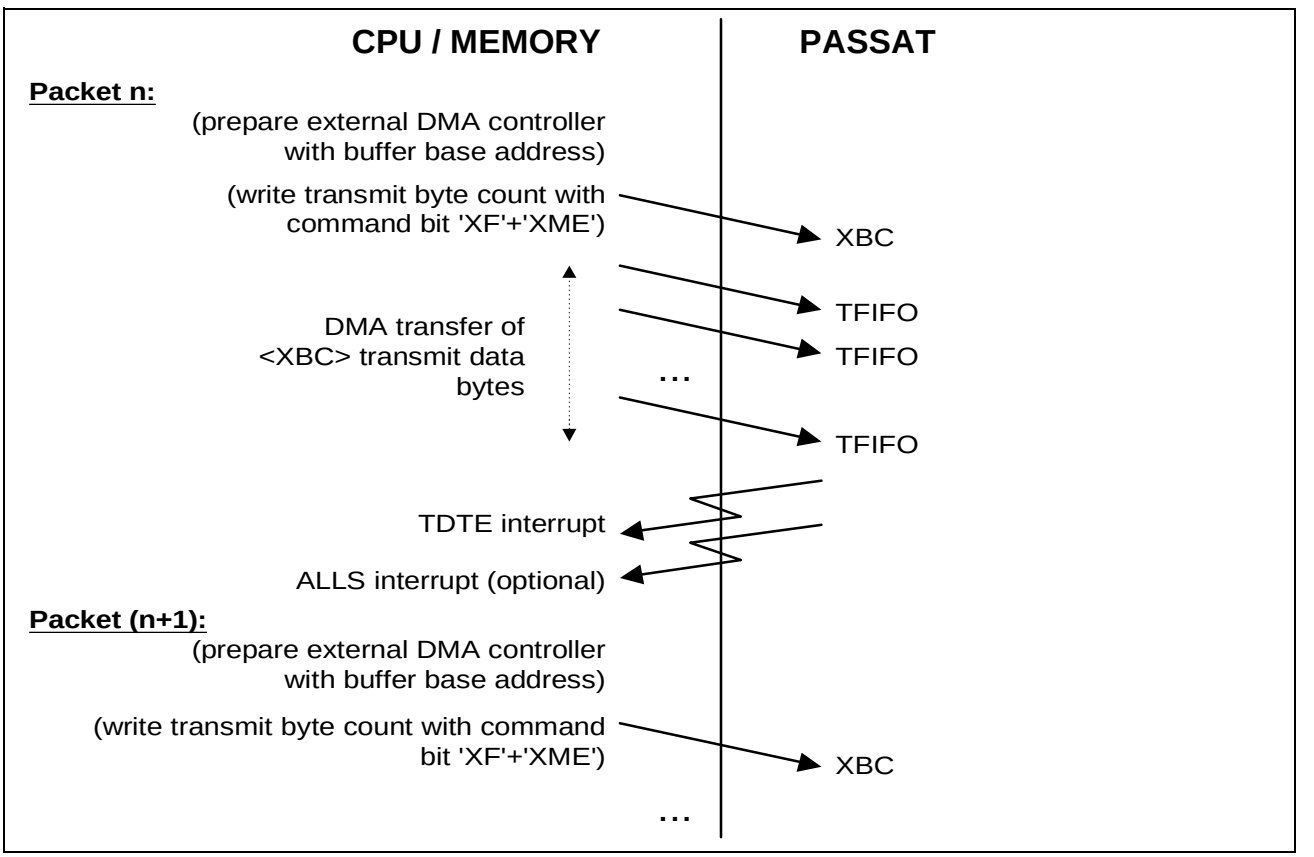


Figure 6-3 DMA Transmit (Single Buffer per Packet)

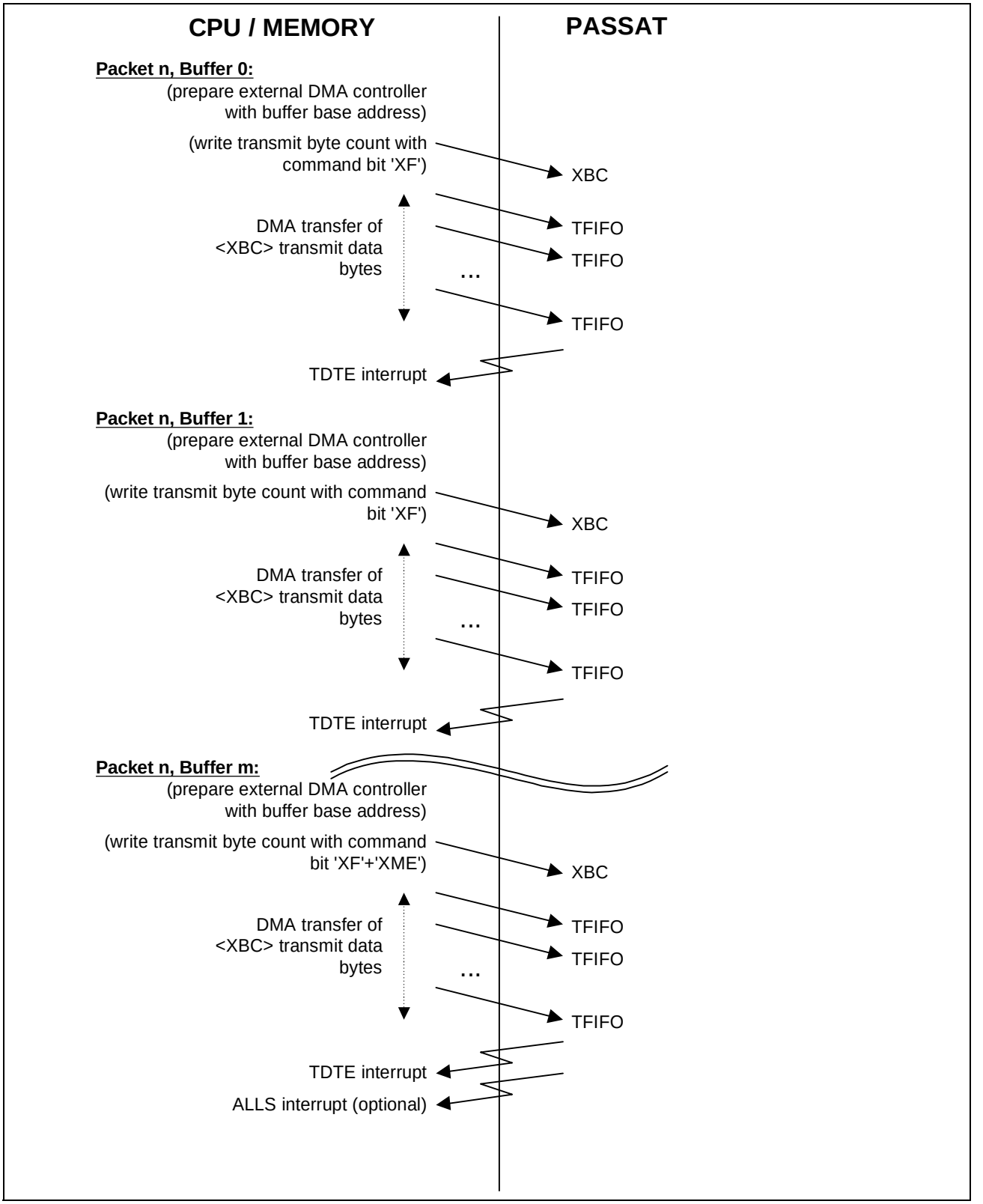


Figure 6-4 Fragmented DMA Transmission (Multiple Buffers per Packet)

6.3.2 Data Reception (With External DMA Support)

The receive DMA support logic is able to limit its requesting for data transfers to a byte count programmed in register RMBSL/RMBSH. If the external DMA controller is capable of handling maximum receive buffer sizes itself, this feature can be disabled by setting bit RMBSH:DRMBS to ’1’.

If a new packet is received by the SCC, the DMA support logic will request the external DMA controller to move receive data out of the RFIFO.

Now there are two possible scenarios:

- If the maximum buffer size programmed in register RMBSL/RMBSH has been transferred (only if RMBSH:DRMBS = ’0’), PASSAT stops requesting for data transfers and a Receive Buffer Full (RBF) interrupt is generated. The CPU now updates the receive buffer base address in the external DMA controller and releases the receive DMA control logic by setting the ’RE’ bit in register RMBSH. Optionally the maximum buffer size value can be updated with the same register write access.
- If the end of a received packet/block is part of the curent DMA transfer, PASSAT generates a Receive DMA Transfer End (RDTE) interrupt and stops operation. The CPU now reads the received byte count from registers RBCL/RBCH. The receive DMA support logic will not continue requesting for data transfer until it is set up again with the ’RE’ command in register RMBSH.

If in packet oriented protocol modes (HDLC, PPP) the maximum receive buffer size RMBS is chosen to be larger than the expected receive packets, each buffer will contain the whole packet (see **Figure 6-5**). In this case (or if RMBSH:DRMBS = ’1’) a Receive Buffer Full (RBF) interrupt will never occur, simplifying the software. To ensure that no packets exceeding the maximum buffer size are forwarded from the SCC to the RFIFO, the receive packet length should be limited with registers RLCRL/RLCRH.

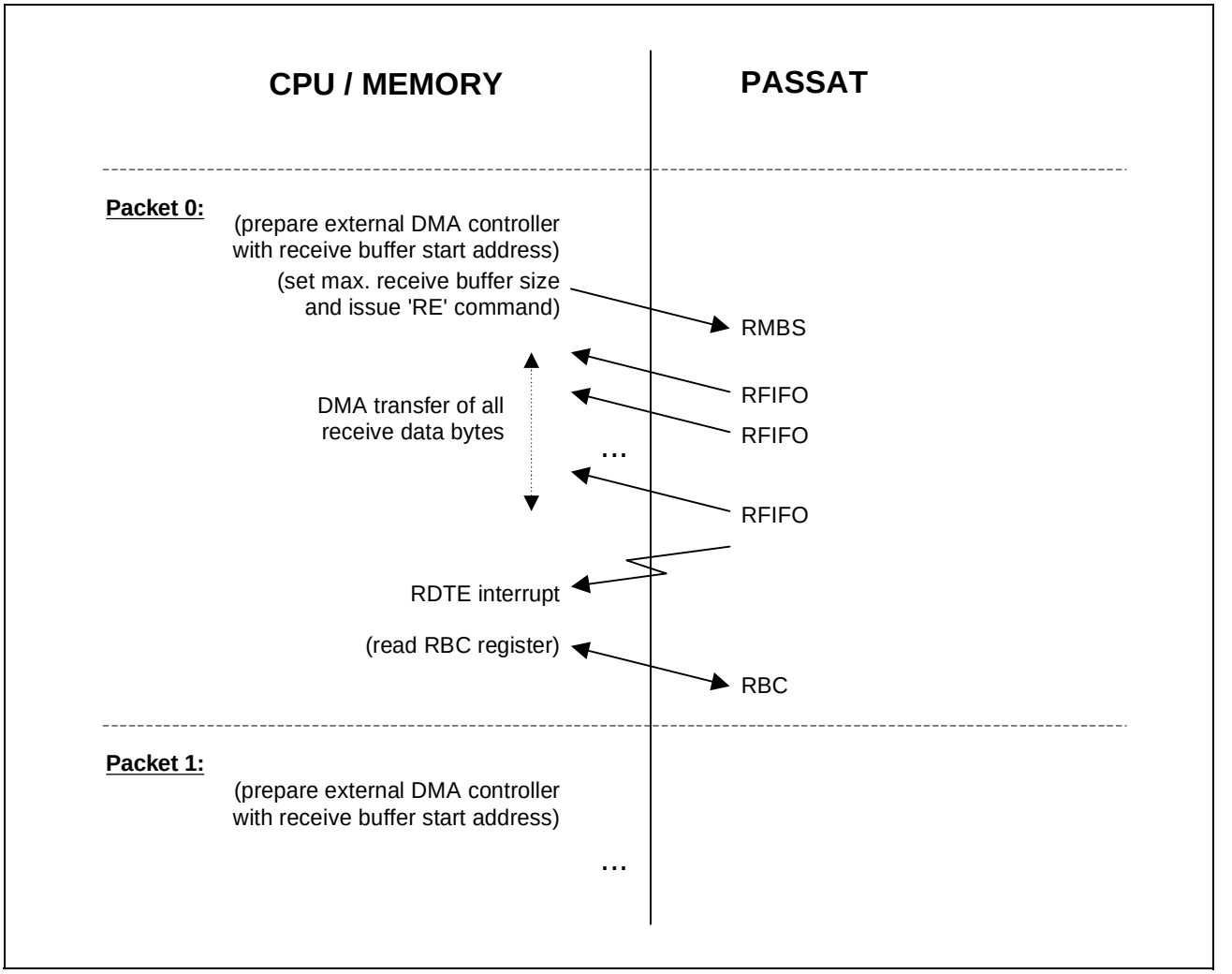


Figure 6-5 DMA Receive (Single Buffer per Packet)

Figure 6-6 shows an example for fragmented reception of a packet larger than the prepared receive buffers in memory. In this case the length of the received packet is 199 bytes, each of the buffers in host memory is 128 bytes deep:

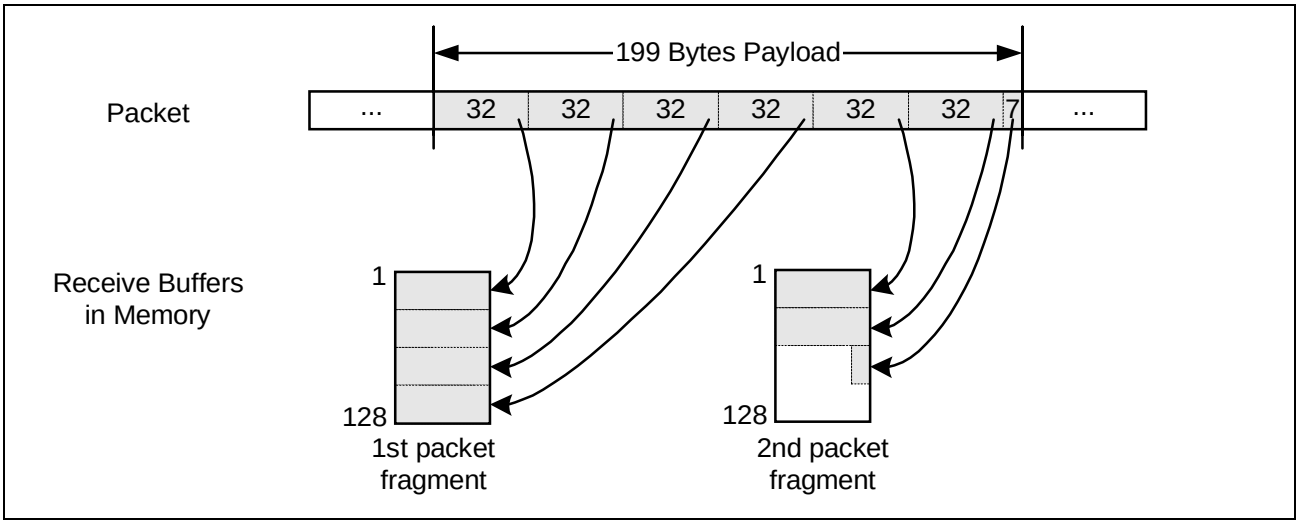


Figure 6-6 Fragmented Reception per DMA (Example)

After the external DMA controller is initialized with the base address of receive buffer #1 and the maximum buffer size RMBS is written to PASSAT, simultaneously activated with the 'RE' command, requesting of DMA transfer from the RFIFO to the receive buffer takes place in blocks of 32 bytes (unless changed with bit field 'RFTH' in register CCR3H).

After four 32-byte-blocks have been transferred, the first receive buffer is filled up completely with receive data. The PASSAT indicates this by generating the RBF interrupt.

Now the CPU has to provide the base address of the second receive buffer to the external DMA controller and issue the 'RE' command to PASSAT again. This allows the external DMA controller to continue data transfers into the second receive buffer. After another two 32-byte-blocks have been transferred, the DMA request for the remaining 7 bytes (including the RSTA byte) is generated to the external DMA controller, follwed by the generation of the RDTE interrupt. Now the DMA transfer is completed and software has to read the number of received bytes from the Receive Byte Count registers RBCL/ RBCH.

The following figure (**Figure 6-7**) gives the sequence of actions from both, the PASSAT and the CPU for this example (fragmented reception of 199 bytes into two receive buffers):

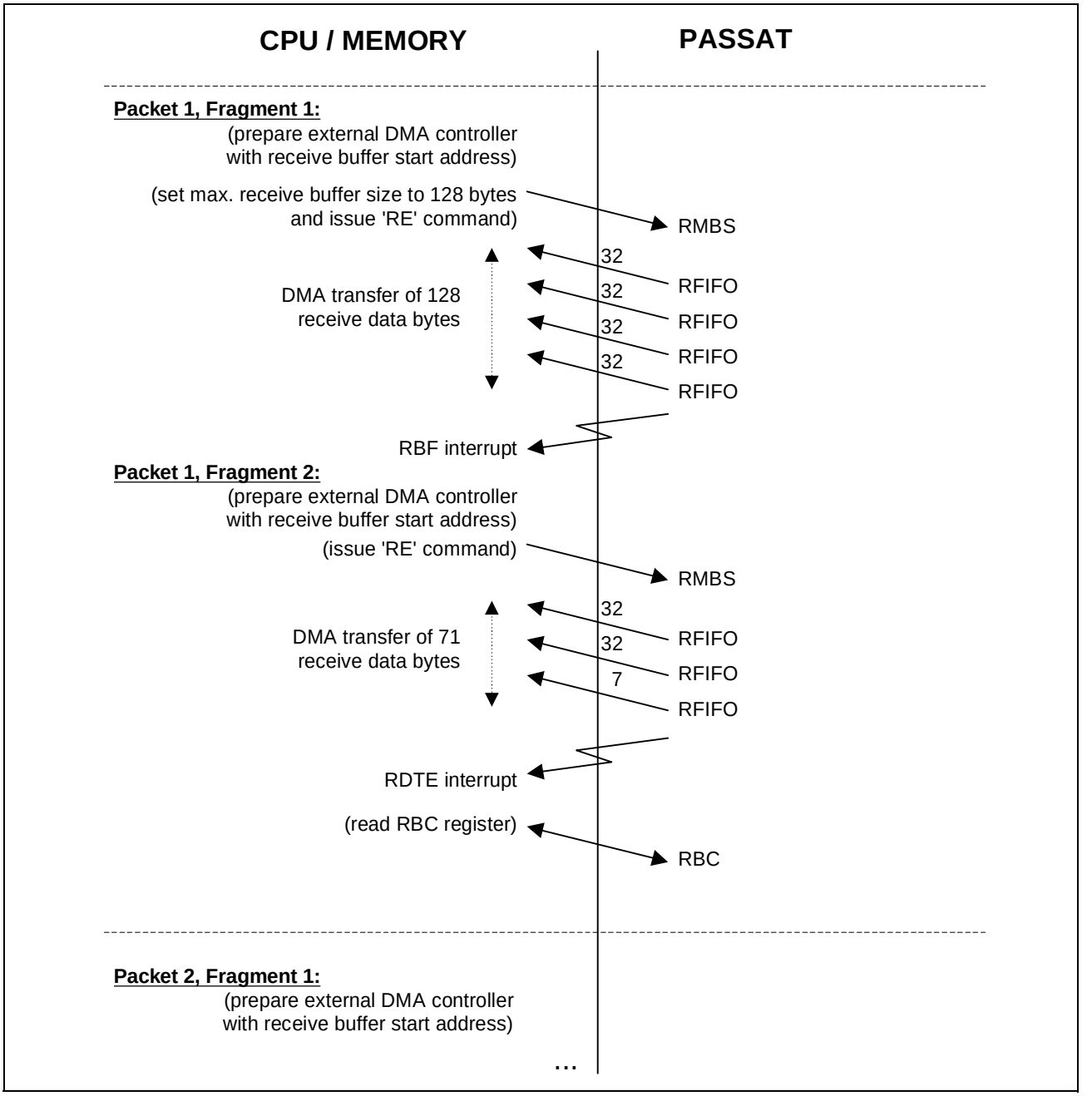


Figure 6-7 Fragmented Reception Sequence (Example)

7 Electrical Characteristics (Preliminary)

All electrical characteristics given in this chapter are preliminary and subject to change.

7.1 Absolute Maximum Ratings

Parameter	Symbol	Limit Values	Unit
Ambient temperature under bias PEB PEF	T_A T_A	0 to 70 – 40 to 85	°C °C
Storage temperature	T_{stg}	– 65 to 125	°C
IC supply voltage	V_{DD3}	– 0.3 to 3.6	V
Voltage on any signal pin with respect to ground	V_S	– 0.3 to 5.5	V
ESD robustness[1)] HBM: 1.5 kΩ, 100 pF	$V_{ESD,HBM}$	2500	V

1) According to MIL-Std 883D, method 3015.7 and ESD Ass. Standard EOS/ESD-5.1-1993.

Note: Stresses above those listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

7.2 Operating Range

Parameter	Symbol	Limit Values		Unit	Test Condition
		min.	max.		
Ambient temperature PEB PEF	T_A T_A	0 -40	70 85	°C °C	
Junction temperature	T_J	0	125	°C	
Supply voltage	V_{DD3}	3.0	3.6	V	
Ground	V_{SS}	0	0	V	

Note: In the operating range, the functions given in the circuit description are fulfilled.

7.3 Thermal Package Characteristics

Table 7-1 Thermal Package Characteristics P-TQFP-100-3

Parameter		Symbol	Value	Unit
Thermal Package Resistance Junction to Ambient				
Airflow:	Ambient Temperature:			
without airflow	T_A=-40°C	$\theta_{JA(0,-40)}$	45.7	K/W
without airflow	T_A=+25°C	$\theta_{JA(0,25)}$	41.5	K/W
airflow 1 m/s (~200 lfpm)	T_A=+25°C	$\theta_{JA(1,25)}$	39.6	K/W
airflow 2 m/s (~400 lfpm)	T_A=+25°C	$\theta_{JA(2,25)}$	38.8	K/W
airflow 3 m/s (~600 lfpm)	T_A=+25°C	$\theta_{JA(3,25)}$	38.4	K/W

Table 7-2 Thermal Package Characteristics P-LFBGA-80-2

Parameter		Symbol	Value	Unit
Thermal Package Resistance Junction to Ambient				
Airflow:	Ambient Temperature:			
without airflow	T_A=-40°C	$\theta_{JA(0,-40)}$	56.1	K/W
without airflow	T_A=+25°C	$\theta_{JA(0,25)}$	50.6	K/W
airflow 1 m/s (~200 lfpm)	T_A=+25°C	$\theta_{JA(1,25)}$	48.2	K/W
airflow 2 m/s (~400 lfpm)	T_A=+25°C	$\theta_{JA(2,25)}$	47.2	K/W
airflow 3 m/s (~600 lfpm)	T_A=+25°C	$\theta_{JA(3,25)}$	46.6	K/W

7.4 DC Characteristics

Parameter		Symbol	Limit Values		Unit	Notes
			min.	max.		
Input low voltage		V_{IL}	– 0.4	0.8	V	
Input high voltage		V_{IH}	2.0	5.5	V	
Output low voltage		V_{OL}		0.45	V	I_{OL} = 7 mA [1] I_{OL} = 2 mA [2]
Output high voltage		V_{OH}	2.4		V	I_{OH} = – 1.0 mA
Power supply current	operational (average)	I_{CC} (AV)		*tbd*	mA	V_{DD} = 3.3 V, T_A = 25 °C, CLK = 20 MHz, XTAL = 20 MHz, inputs at V_{SS}/V_{DD}, no output loads
	power down (no clocks)	I_{CC} (PD)		*tbd*	mA	V_{DD} = 3.3 V, T_A = 25 °C
Power dissipation		P		100	mW	V_{DD} = 3.3 V, T_A = 25 °C, CLK = 20 MHz, XTAL = 20 MHz, inputs at V_{SS}/V_{DD}, no output loads
Input leakage current		I_{IL}		*tbd*	µA	V_{DD} = 3.3 V, GND = 0 V; inputs at V_{SS}/V_{DD}, no output loads
Output leakage current		I_{OZ}		*tbd*	µA	V_{DD} = 3.3 V, GND = 0 V; V_{OUT} = 0 V, V_{DDP} + 0.4

[1] Apply to the next pins: *tbd*.

[2] Apply to all the I/O and O pins that do not appear in the list in note [1], except *tbd*.

The listed characteristics are ensured over the operating range of the integrated circuit. Typical characteristics specify mean values expected over the production spread. If not otherwise specified, typical characteristics apply at T_A = 25 °C and the given supply voltage.

7.5 AC Characteristics

Interface Pins

T_A = 0 to + 70 °C; V_{DD3} = 3.3 V ± 0.3 V

Inputs are driven to 2.4 V for a logical “1” and to 0.4 V for a logical “0”. Timing measurements are made at 2.0 V for a logical “1” and at 0.8 V for a logical “0”.

The AC testing input/output waveforms are shown below.

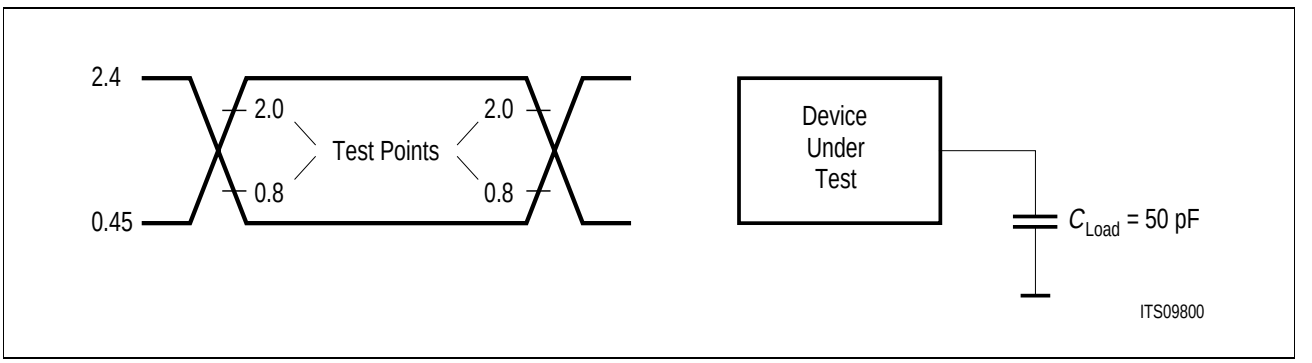


Figure 7-1 Input/Output Waveform for AC Tests

7.6 Capacitances

Interface Pins

Table 7-3 Capacitances

T_A = 25 °C; V_{DD3} = 3.3 V ± 0.3 V, V_{SS} = 0 V

Parameter	Symbol	Limit Values		Unit	Test Condition
		min.	max.		
Input capacitance	C_{IN}	*tbd*	*tbd*	pF	
Output capacitance	C_{OUT}	*tbd*	*tbd*	pF	
I/O-capacitance	C_{IO}	*tbd*	*tbd*	pF	

7.7 Timing Diagrams

7.7.1 Microprocessor Interface Timing

7.7.1.1 Microprocessor Interface Clock Timing

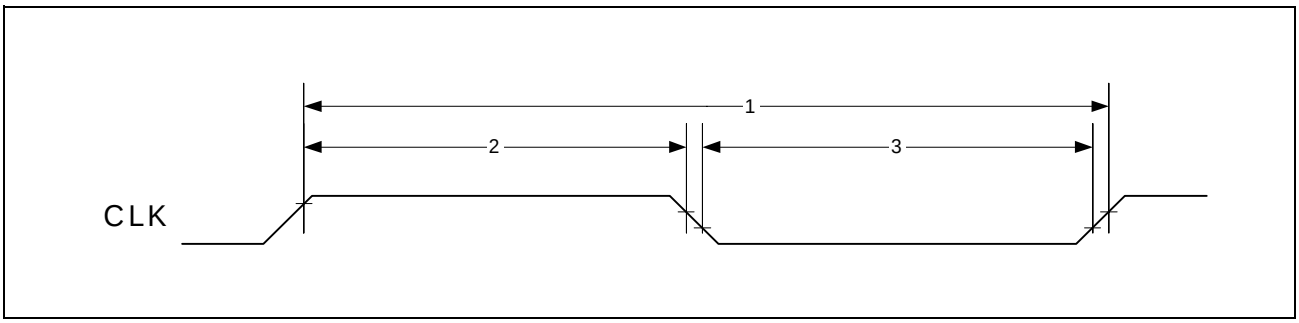


Figure 7-2 Microprocessor Interface Clock Timing

Table 7-4 Microprocessor Interface Clock Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
1	CLK clock period	30		ns
	CLK frequency		33	MHz
2	CLK high time	11		ns
3	CLK low time	11		ns

7.7.1.2 Siemens/Intel Bus Interface Timing

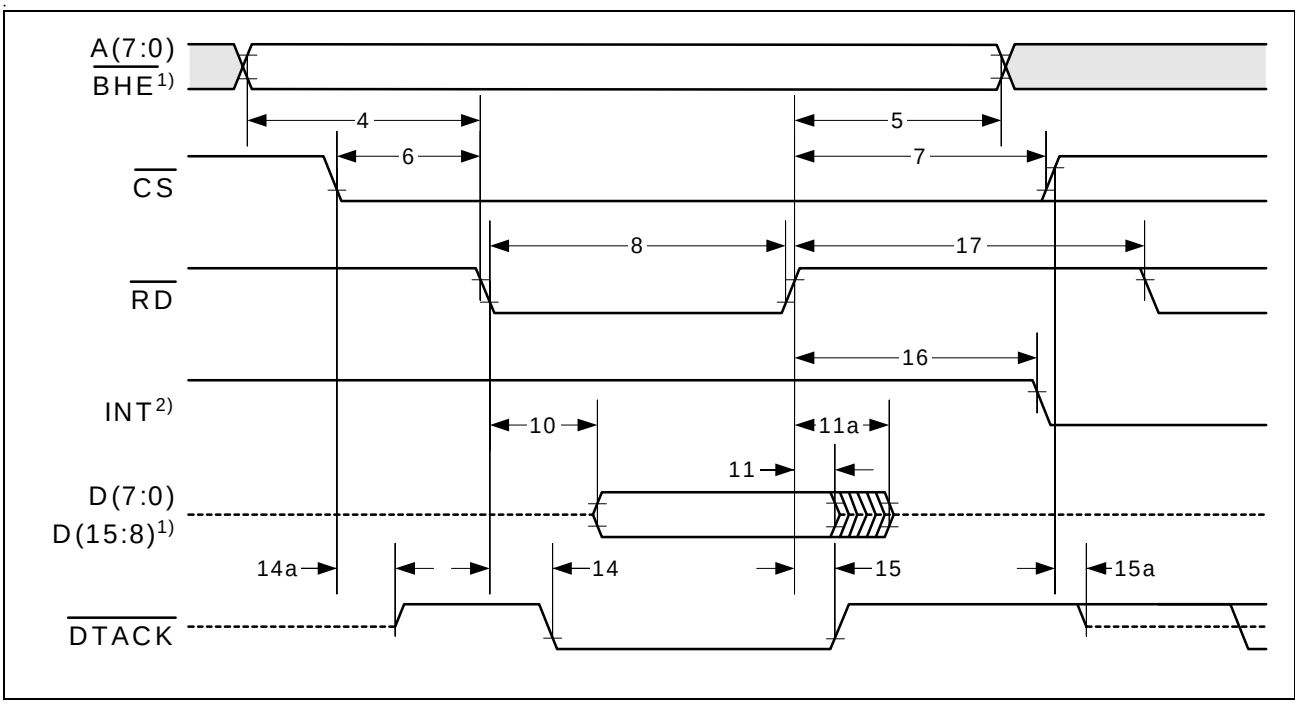


Figure 7-3 Siemens/Intel Read Cycle Timing

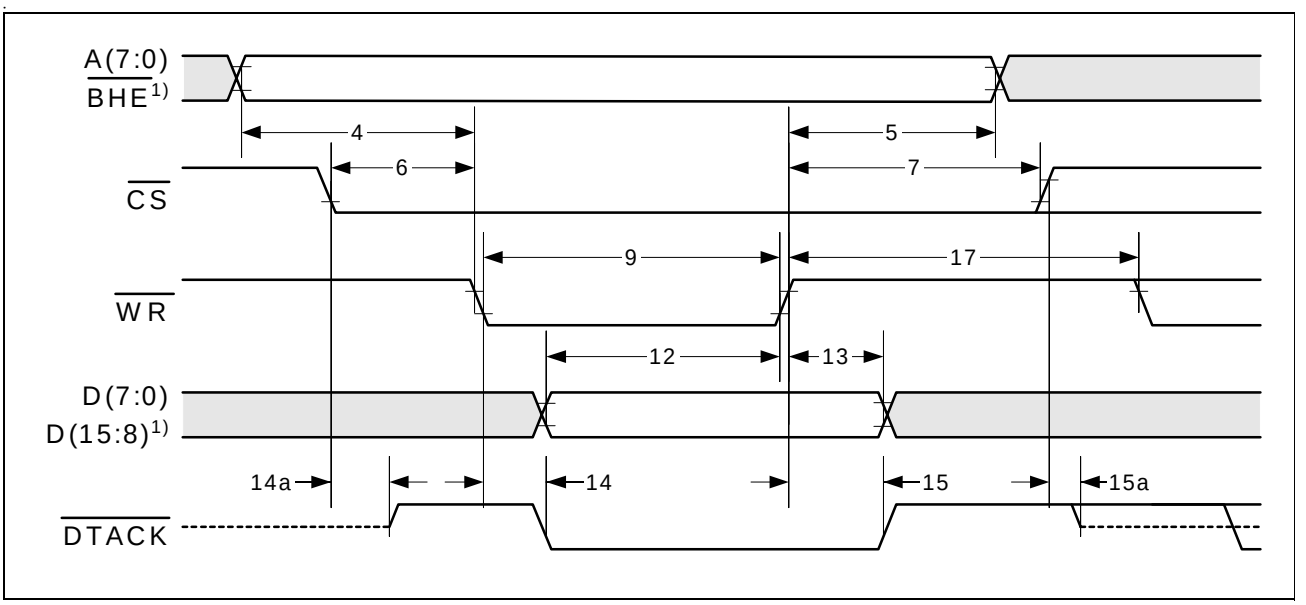


Figure 7-4 Siemens/Intel Write Cycle Timing

(1) Signals $\overline{\text{BHE}}$ and D(15:8) only available in 16-bit Intel bus mode
(2) Interrupt signal shown is push-pull, active high. Same timings apply to push-pull, active low interrupt signal. In case of open-drain output the timing depends on external components.

Table 7-5 Siemens/Intel Bus Interface Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
4	active address to active $\overline{RD}/\overline{WR}$ setup time	10		ns
5	inactive $\overline{RD}/\overline{WR}$ to inactive address hold time	0		ns
6	active $\overline{CS}$ to active $\overline{RD}/\overline{WR}$ setup time	0		ns
7	inactive $\overline{RD}/\overline{WR}$ to inactive $\overline{CS}$ hold time	0		ns
8	$\overline{RD}$ active pulse width	tbd		ns
9	$\overline{WR}$ active pulse width	30		ns
10	active $\overline{RD}$ to valid data delay		15	ns
11	inactive $\overline{RD}$ to invalid data hold time	5		ns
11a	inactive $\overline{RD}$ to data high impedance delay		15	ns
12	valid data to inactive $\overline{WR}$ setup time	15		ns
13	inactive $\overline{WR}$ to invalid data hold time	5		ns
14	active $\overline{RD}/\overline{WR}$ to active $\overline{DTACK}$ delay		10	ns
14a	active $\overline{CS}$ to driven $\overline{DTACK}$ delay		tbd	ns
15	inactive $\overline{RD}/\overline{WR}$ to inactive $\overline{DTACK}$ delay		10	ns
15a	inactive $\overline{CS}$ to $\overline{DTACK}$ high impedance delay		tbd	ns
16	inactive $\overline{RD}$ to inactive INT/$\overline{INT}$ delay		1	T_{CLK}
17	$\overline{RD}/\overline{WR}$ inactive pulse width	tbd		ns

7.7.1.3 Motorola Bus Interface Timing

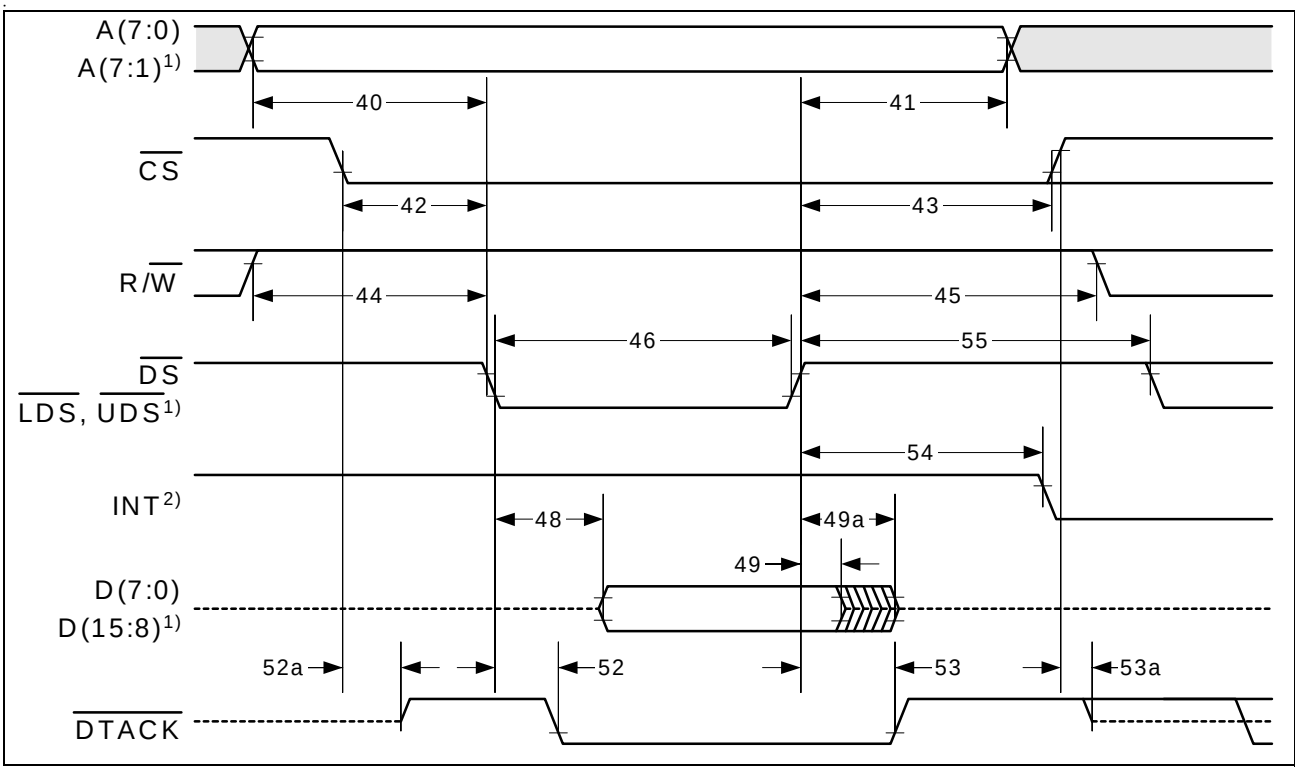


Figure 7-5 Motorola Read Cycle Timing

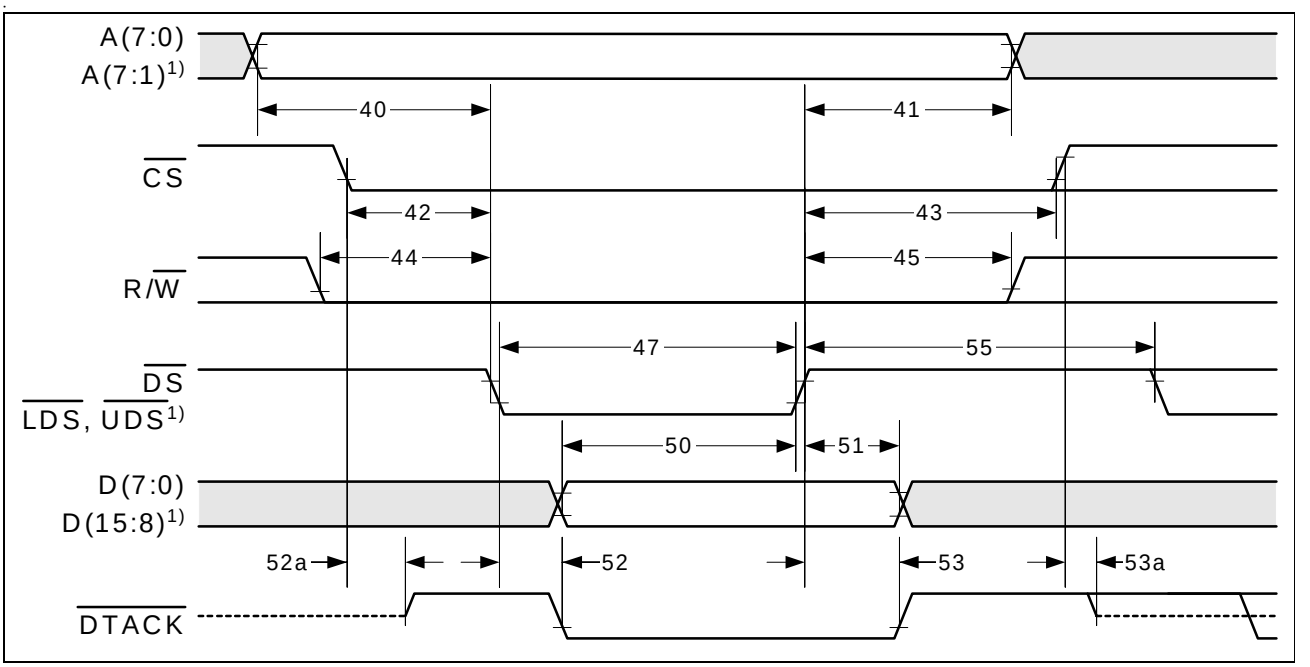


Figure 7-6 Motorola Write Cycle Timing

(1) Signals $\overline{LDS}$, $\overline{UDS}$ and D(15:8) only available in 16-bit Motorola bus mode
(2) Interrupt signal shown is push-pull, active high. Same timings apply to push-pull, active low interrupt signal. In case of open-drain output the timing depends on external components.

Table 7-6 Motorola Bus Interface Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
40	active address to active $\overline{DS}$ setup time	10		ns
41	inactive $\overline{DS}$ to inactive address hold time	0		ns
42	active $\overline{CS}$ to active $\overline{DS}$ setup time	0		ns
43	inactive $\overline{DS}$ to inactive $\overline{CS}$ hold time	0		ns
44	active R/$\overline{W}$ to active $\overline{DS}$ setup time	0		ns
45	inactive $\overline{DS}$ to inactive R/$\overline{W}$ hold time	0		ns
46	$\overline{DS}$ active pulse width (read access)	tbd		ns
47	$\overline{DS}$ active pulse width (write access)	30		ns
48	active $\overline{DS}$ (read) to valid data delay		15	ns
49	inactive $\overline{DS}$ (read) to invalid data hold time	5		ns
49a	inactive $\overline{DS}$ (read) to data high impedance delay		15	ns
50	valid data to inactive $\overline{DS}$ (write) setup time	15		ns
51	inactive $\overline{DS}$ (write) to invalid data hold time	5		ns
52	active $\overline{DS}$ to active $\overline{DTACK}$ delay		10	ns
52a	active $\overline{CS}$ to driving $\overline{DTACK}$ delay		tbd	ns
53	inactive $\overline{DS}$ to inactive $\overline{DTACK}$ delay		10	ns
53a	inactive $\overline{CS}$ to $\overline{DTACK}$ high impedance delay		tbd	ns
54	inactive $\overline{DS}$ (read) to inactive INT/$\overline{INT}$ delay		1	T_{CLK}
55	$\overline{DS}$ inactive pulse width	tbd		ns

7.7.2 PCM Serial Interface Timing

7.7.2.1 Clock Input Timing

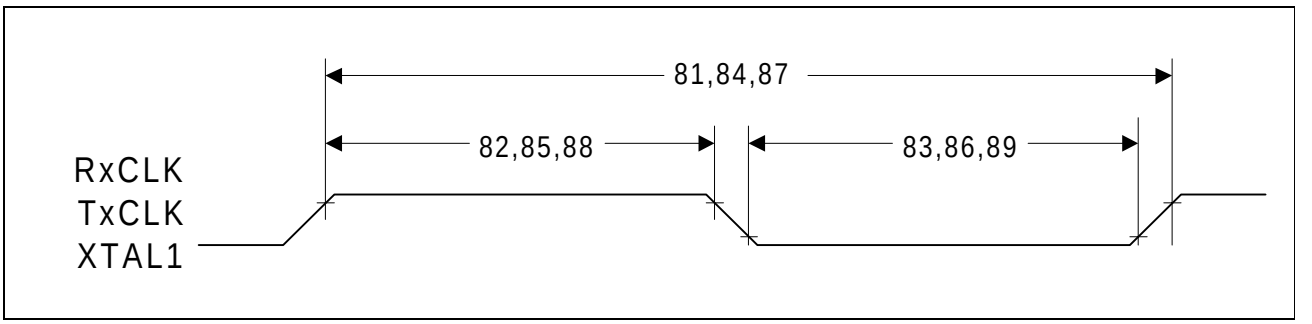


Figure 7-7 Clock Input Timing

Table 7-7 Clock Input Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
81	RxCLK clock period	tbd		ns
82	RxCLK high time	tbd		ns
83	RxCLK low time	tbd		ns
84	TxCLK clock period	tbd		ns
85	TxCLK high time	tbd		ns
86	TxCLK low time	tbd		ns
87	XTAL1 clock period (internal oscillator used)	tbd		ns
	XTAL1 clock period (TTL clock signal supplied)	tbd		ns
88	XTAL1 high time (internal oscillator used)	tbd		ns
	XTAL1 high time (TTL clock signal supplied)	tbd		ns
89	XTAL1 low time (internal oscillator used)	tbd		ns
	XTAL1 low time (TTL clock signal supplied)	tbd		ns

7.7.2.2 Receive Cycle Timing

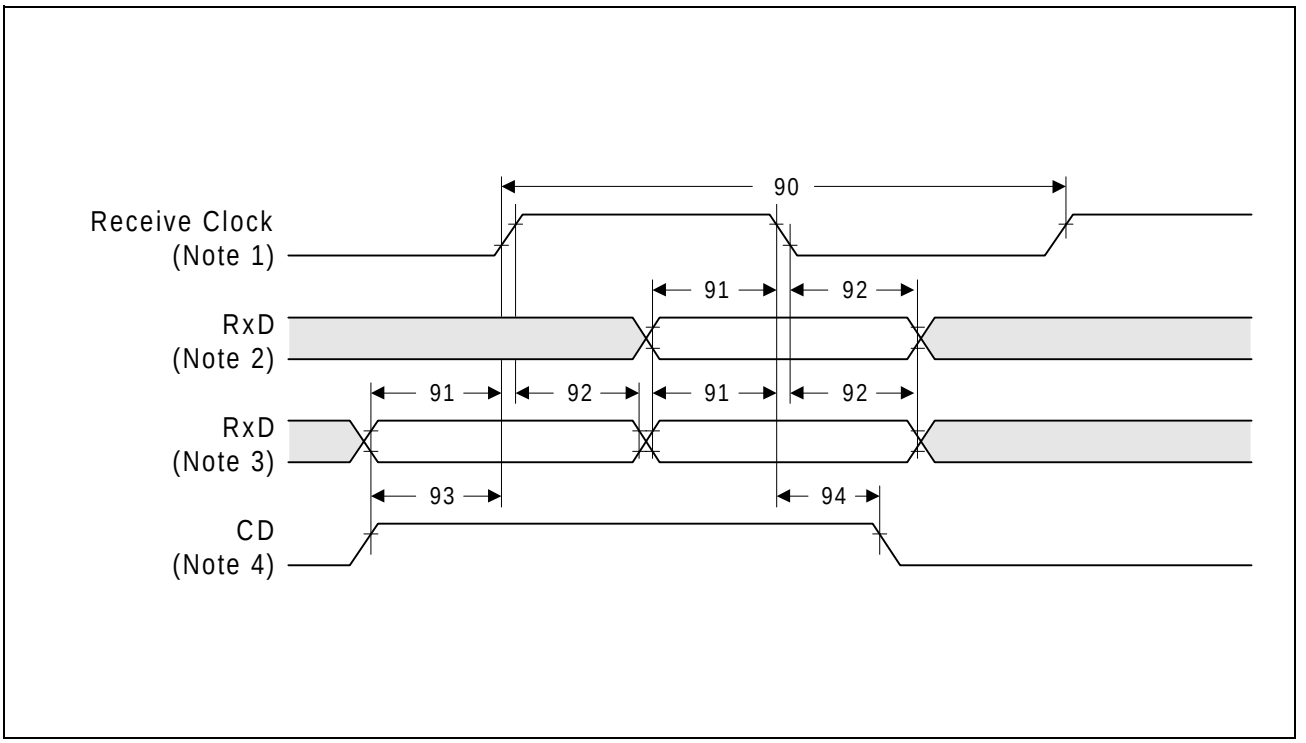


Figure 7-8 Receive Cycle Timing

Note:

1. *Whichever supplies the receive clock depending on the selected clock mode:*
   externally clocked via RxCLK or XTAL1 or
   internally clocked via DPLL or BRG.
   (No edge relation can be measured if the internal receive clock is derived from the external clock source by devision stages (BRG) or DPLL)
2. *NRZ, NRZI and Manchester data encoding*
3. *FM0 and FM1 data encoding*
4. *If Carrier Detect auto start feature enabled (not for clock modes 1 and 5)*

Table 7-8 Receive Cycle Timing

No.	Parameter		Limit Values		Unit
			min.	max.	
Receive data rates		externally clocked (HDLC)		16	Mbit/s
		internally clocked (DPLL modes)		2	Mbit/s
		internally clocked (non DPLL modes)		2	Mbit/s
90	Clock period	externally clocked	*tbd*		ns
		internally clocked (DPLL modes)	*tbd*		ns
		internally clocked (non DPLL modes)	*tbd*		ns
91	RxD to RxCLK setup time		*tbd*		ns
92	RxD to RxCLK hold time		*tbd*		ns
93	CD to RxCLK rising edge setup time		*tbd*		ns
94	CD to RxCLK falling edge hold time		*tbd*		ns

7.7.2.3 Transmit Cycle Timing

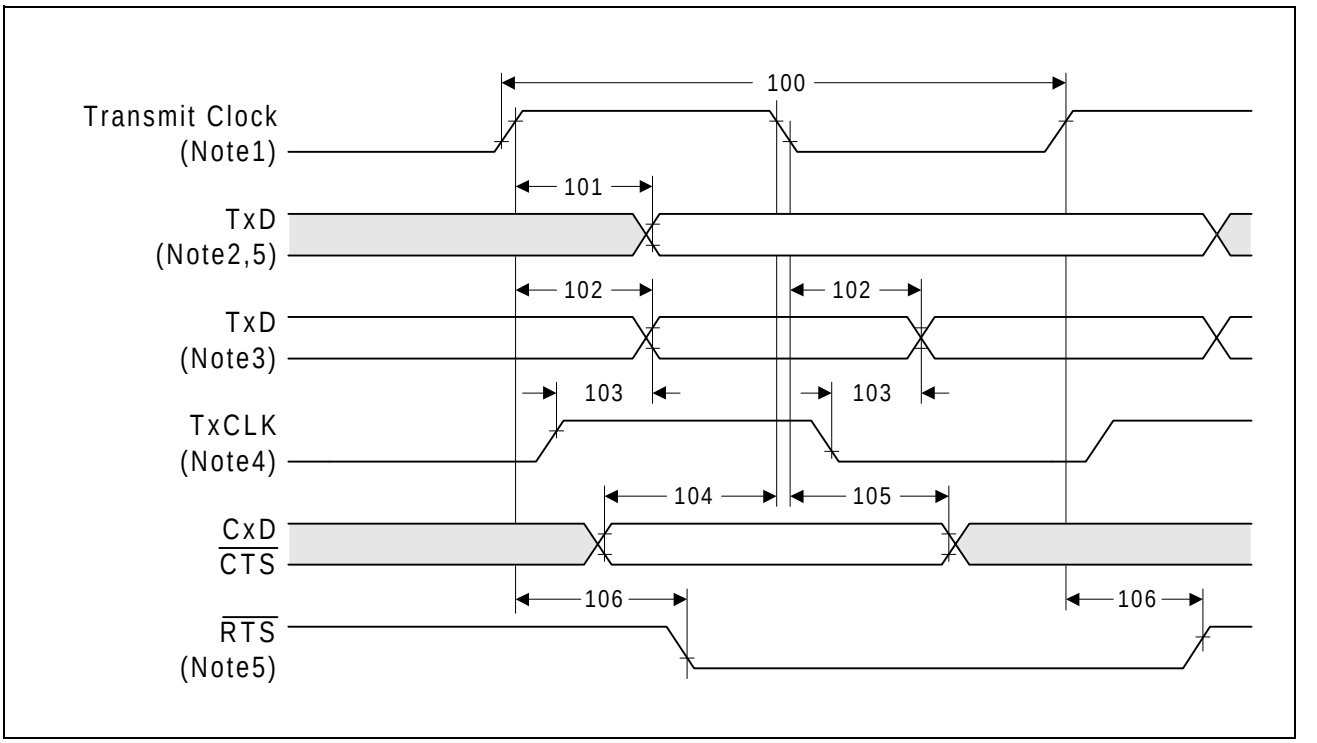


Figure 7-9 Transmit Cycle Timing

Note:

1. Whichever supplies the transmit clock depending on the selected clock mode:
   externally clocked via TxCLK, RxCLK or XTAL1 or
   internally clocked via DPLL or BRG.
   (No edge relation can be measured if the internal transmit clock is derived from the external clock source by devision stages (BRG) or DPLL)
2. NRZ, NRZI and Manchester data encoding
3. FM0 and FM1 data encoding
4. If TxCLK output feature is enabled (only in some clock modes)
5. The timing is valid for non bus configuration modes and bus configuration mode 1. In bus configuration mode 2, TxD and $\overline{\text{RTS}}$ are right shifted for 0.5 TxCLK periods i.e. driven by the falling TxCLK edge.

Table 7-9 Transmit Cycle Timing

No.	Parameter		Limit Values		Unit
			min.	max.	
Transmit data rates		externally clocked		16	Mbit/s
		internally clocked (DPLL modes)		2	Mbit/s
		internally clocked (non DPLL modes)		2	Mbit/s
100	Clock period	externally clocked	*tbd*		ns
		internally clocked (DPLL modes)	*tbd*		ns
		internally clocked (non DPLL modes)	*tbd*		ns
101	TxD to TxCLK delay (NRZ, NRZI encoding)			*tbd*	ns
102	TxD to TxCLK delay (FM0, FM1, Manchester encoding)			*tbd*	ns
103	TxD to TxCLK(out) delay (output function enabled)		*tbd*	*tbd*	ns
104	$\overline{\text{CxD}}$ to TxCLK setup time, $\overline{\text{CTS}}$ to TxCLK setup time		*tbd*		ns
105	$\overline{\text{CxD}}$ to TxCLK hold time, $\overline{\text{CTS}}$ to TxCLK hold time		*tbd*		ns
106	$\overline{\text{RTS}}$ to TxCLK delay (not bus configuration mode)			*tbd*	ns
	$\overline{\text{RTS}}$ to TxCLK delay (bus configuration mode)			*tbd*	ns

7.7.2.4 Clock Mode 1 Strobe Timing

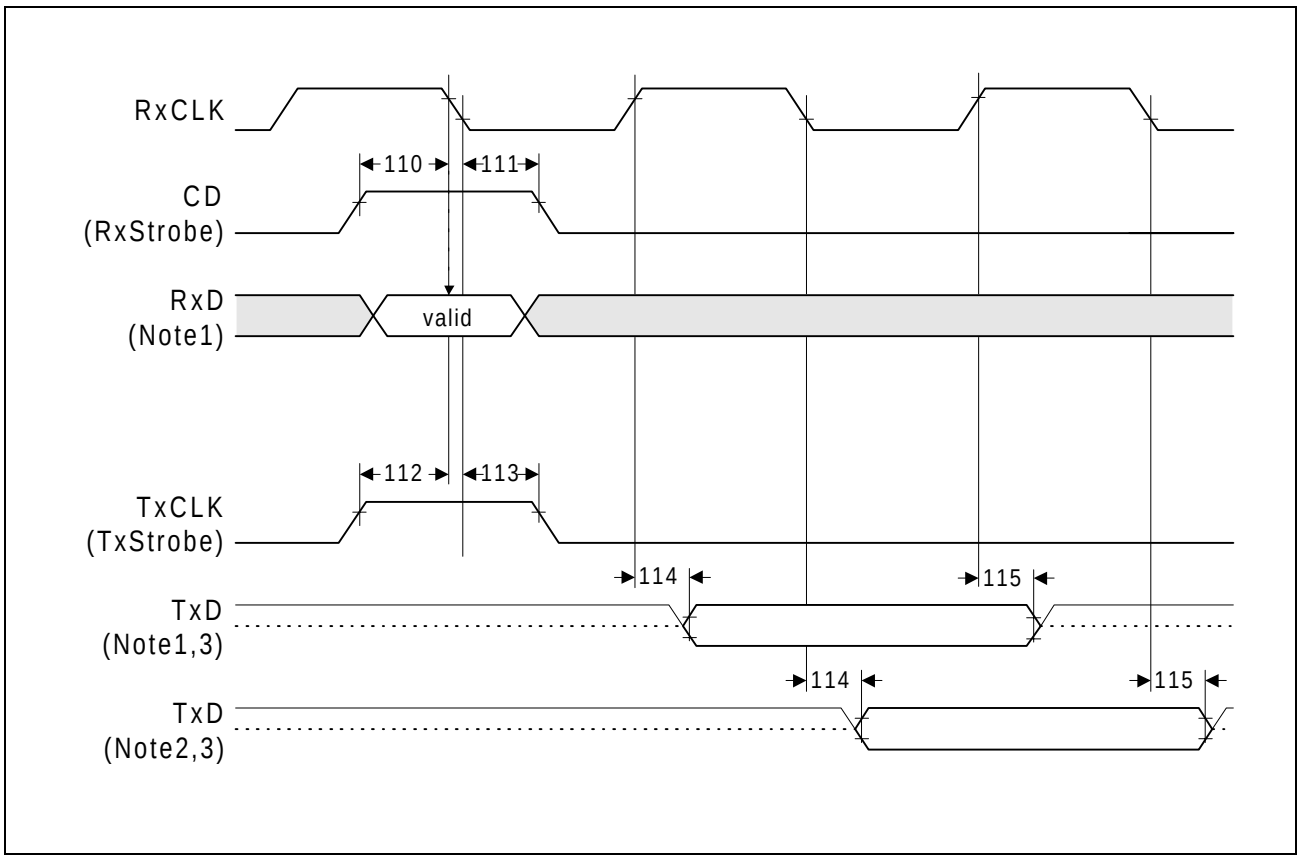


Figure 7-10 Clock Mode 1 Strobe Timing

Note:

1. No bus configuration mode and bus configuration mode 1
2. Bus configuration mode 2
3. TxD Idle is either *active high* or *high impedance* if 'open drain' output type is selected.

Table 7-10 Clock Mode 1 Strobe Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
110	Receive strobe to RxCLK setup	*tbd*		ns
111	Receive strobe to RxCLK hold	*tbd*		ns
112	Transmit strobe to RxCLK setup	*tbd*		ns
113	Transmit strobe to RxCLK hold	*tbd*		ns
114	TxD to RxCLK delay	*tbd*		ns
115	TxD to RxCLK high impedance delay	*tbd*		ns

7.7.2.5 Clock Mode 5 Frame Synchronisation Timing

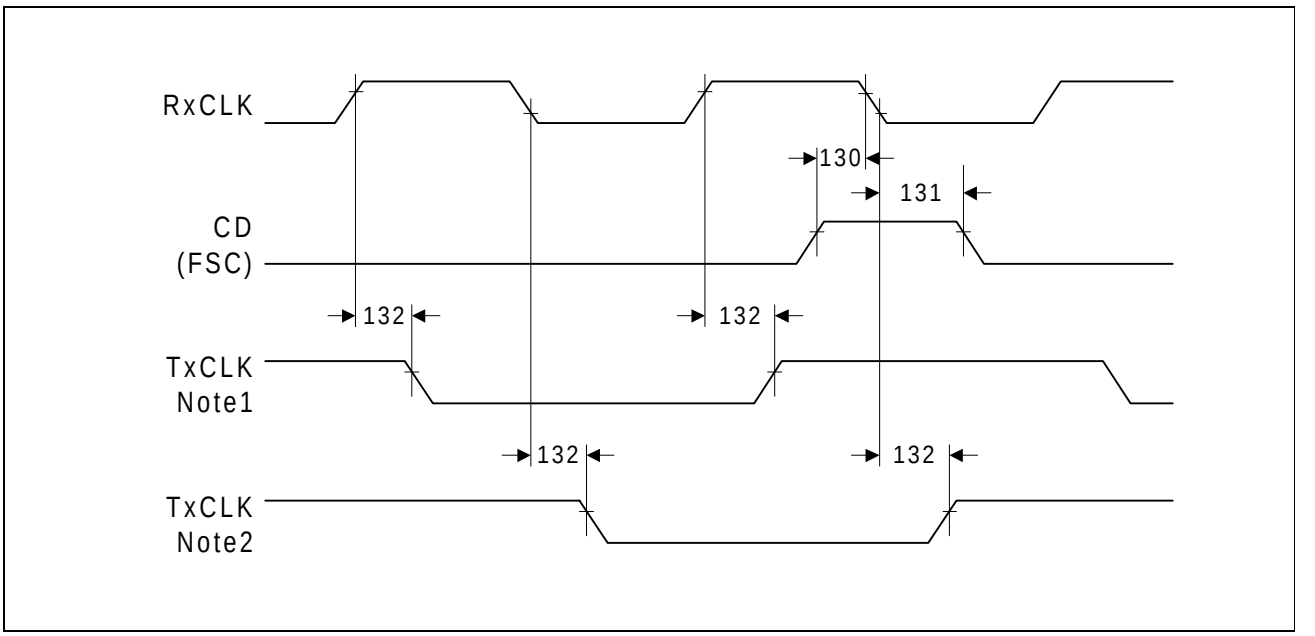


Figure 7-11 Clock Mode 5 Frame Synchronisation Timing

Note:

1. Normal operation and bus configuration mode 1
2. Bus configuration mode 2

Table 7-11 Clock Mode 5 Frame Synchronisation Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
130	Sync pulse to RxCLK setup time	*tbd*		ns
131	Sync pulse to RxCLK hold time	*tbd*		ns
132	TxCLKout to RxCLK delay (time slot monitor)	*tbd*	*tbd*	ns

7.7.2.6 Clock Mode 4 Receive Cycle Timing

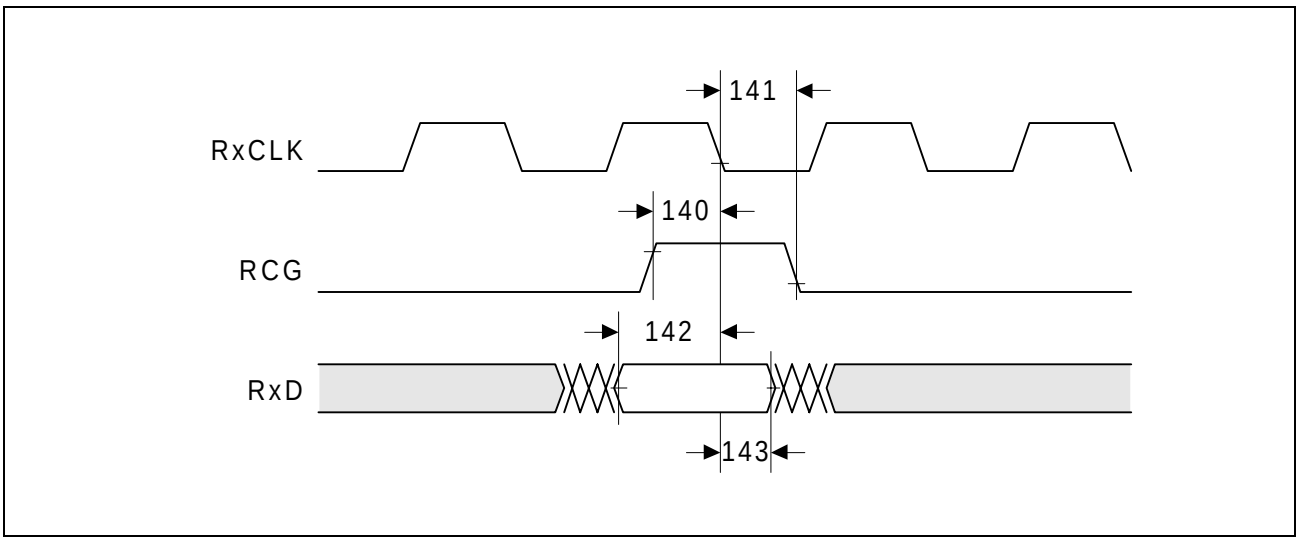


Figure 7-12 Clock Mode 4 Receive Timing

Table 7-12 Clock Mode 4 Receive Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
140	RCG setup time	*tbd*		ns
141	RCG hold time	*tbd*		ns
142	RxD setup time	*tbd*		ns
143	RxD hold time	*tbd*		ns

7.7.2.7 Clock Mode 4 Transmit Cycle Timing

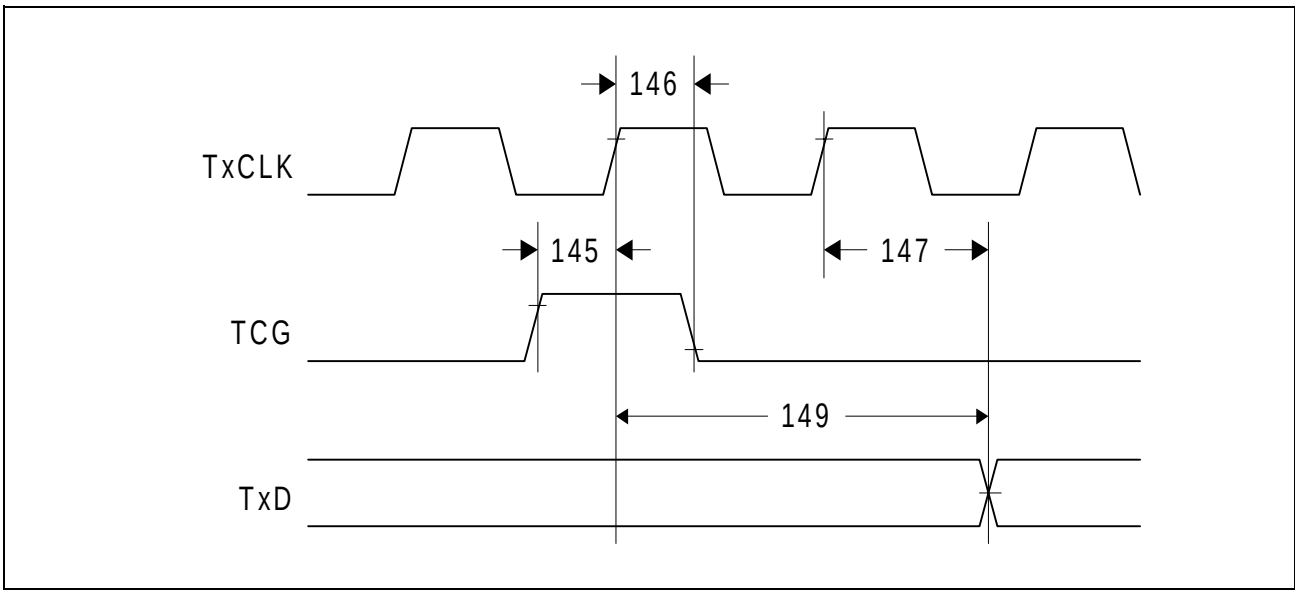


Figure 7-13 Clock Mode 4 Transmit Timing

Table 7-13 Clock Mode 4 Transmit Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
145	TCG setup time	*tbd*		ns
146	TCG hold time	*tbd*		ns
147	TxCLK to TxD delay		*tbd*	ns
149	TxD to TCG active delay		1 T_{TxCLK} + *tbd*	(ns)

Note: T_{TxCLK} is the TxCLK signal time period.
Timing 149 results from a constant functional one clock offset + signal delay.

7.7.3 Reset Timing

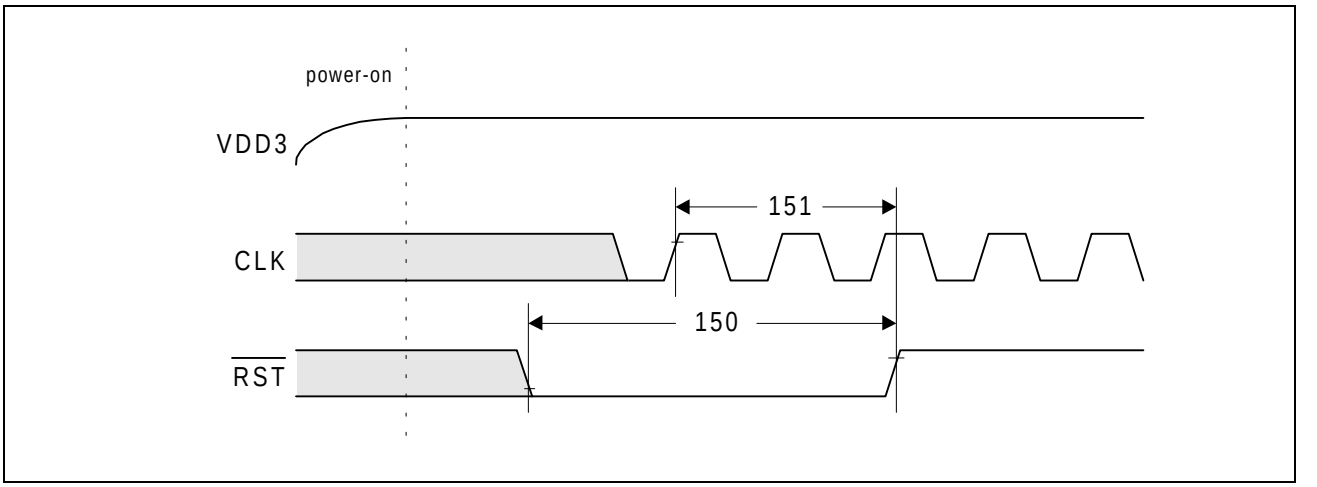


Figure 7-14 Reset Timing

Table 7-14 Reset Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
150	*RESET* pulse width	*tbd*		ns
150	Number of CLK cycles during RESET active	*tbd*		CLK cycles

Note: RESET may be asynchronous to CLK when asserted or deasserted. RESET may be asserted during power-up or asserted after power-up. Nevertheless deassertion must be clean.

7.7.4 JTAG-Boundary Scan Timing

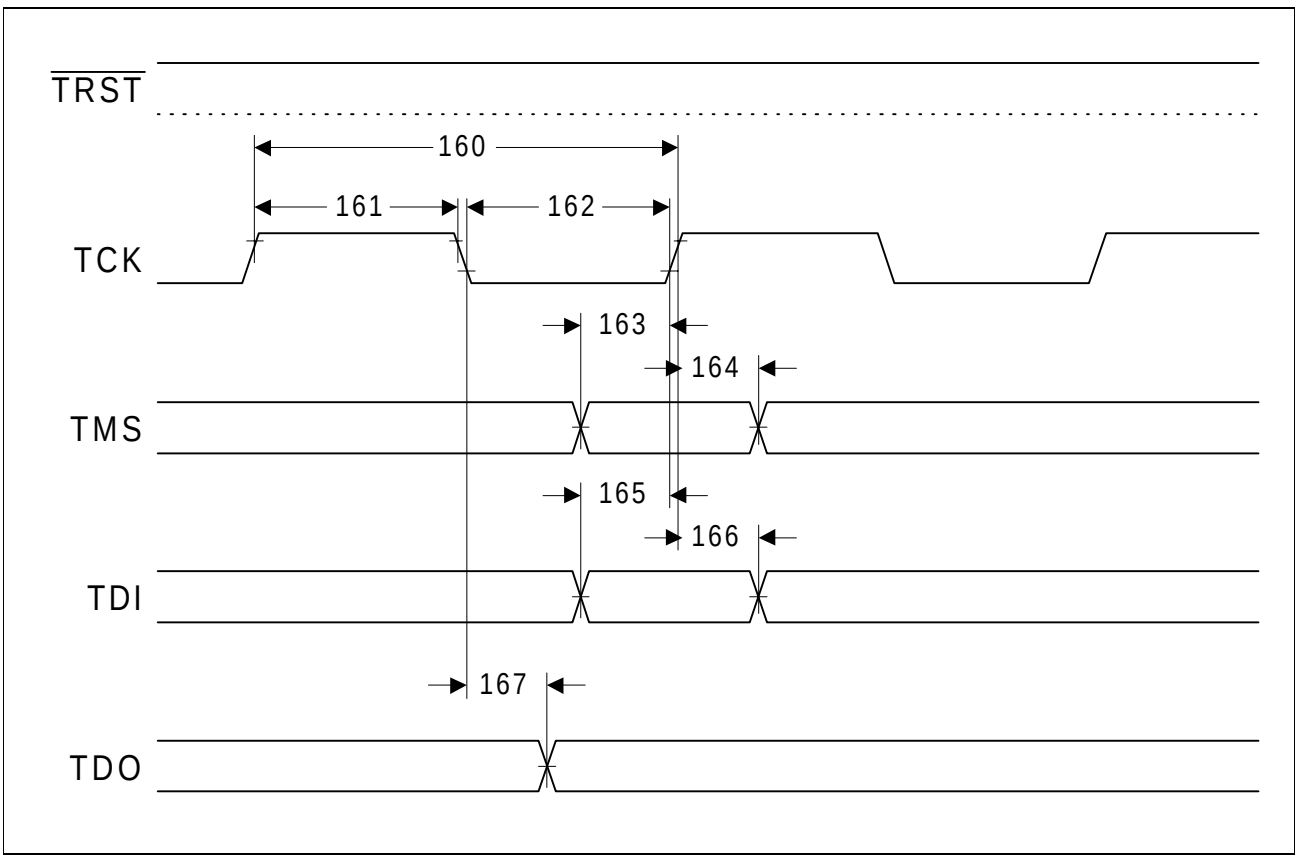


Figure 7-15 JTAG-Boundary Scan Timing

Table 7-15 JTAG-Boundary Scan Timing

No.	Parameter	Limit Values		Unit
		min.	max.	
160	TCK period	*tbd*	∞	ns
161	TCK high time	*tbd*		ns
162	TCK low time	*tbd*		ns
163	TMS setup time	*tbd*		ns
164	TMS hold time	*tbd*		ns
165	TDI setup time	*tbd*		ns
166	TDI hold time	*tbd*		ns
167	TDO valid delay	*tbd*		ns

8 Test Modes

8.1 JTAG Boundary Scan Interface

In the PASSAT a Test Access Port (TAP) controller is implemented. The essential part of the TAP is a finite state machine (16 states) controlling the different operational modes of the boundary scan. Both, TAP controller and boundary scan, meet the requirements given by the JTAG standard: IEEE 1149.1. **Figure 8-1** gives an overview about the TAP controller.

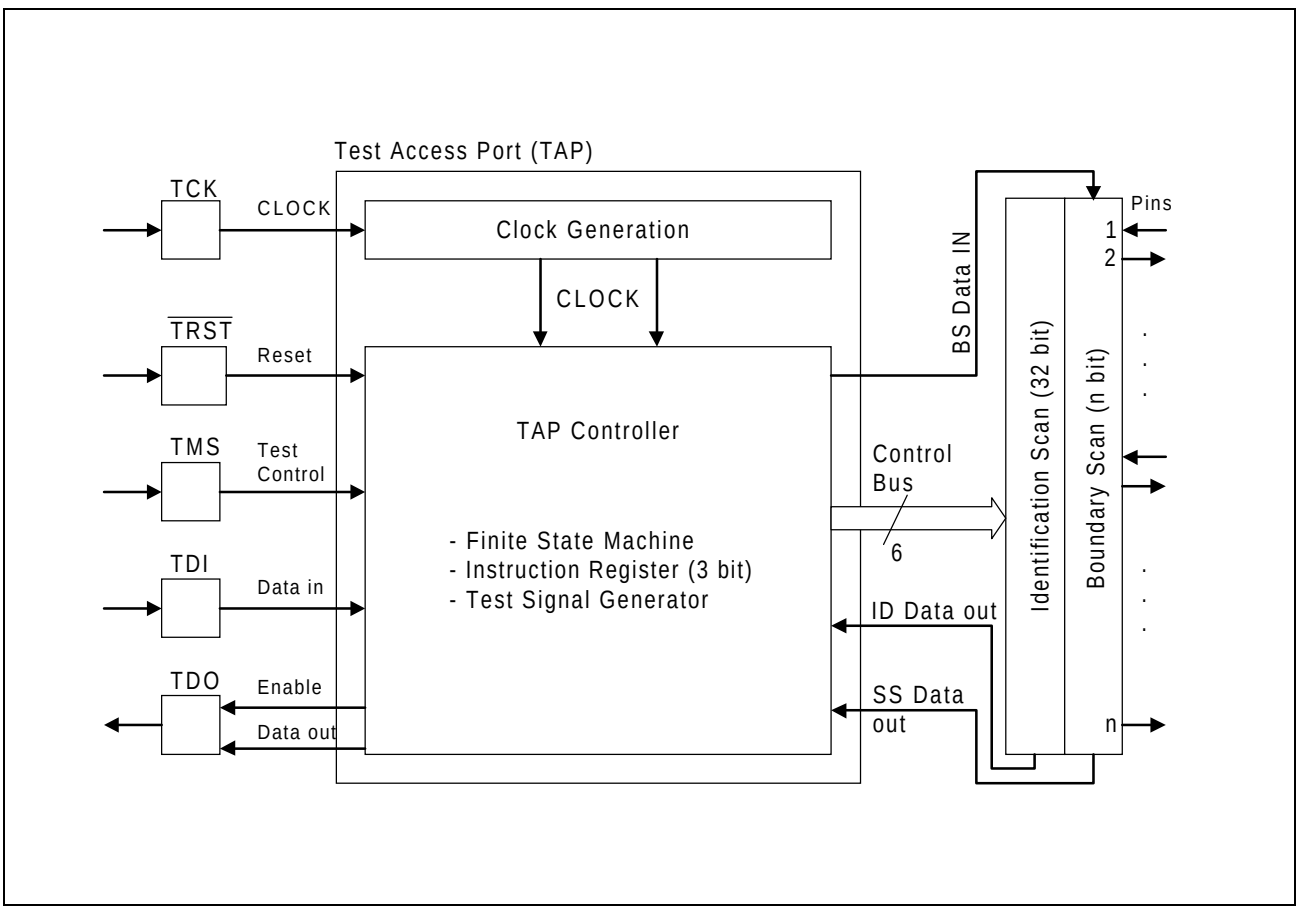


Figure 8-1 Block Diagram of Test Access Port and Boundary Scan Unit

If no boundary scan operation is planned $\overline{TRST}$ has to be connected with V_{SS}. TMS, TCK and TDI do not need to be connected since pull-up transistors ensure high input levels in this case. Nevertheless it would be a good practice to put these unused inputs to defined levels, using pull-up resistors.

Test handling (boundary scan operation) is performed via the pins TCK (Test Clock), TMS (Test Mode Select), TDI (Test Data Input) and TDO (Test Data Output) when the TAP controller is not in its reset state, i.e. $\overline{TRST}$ is connected to V_{DD} or it remains unconnected due to its internal pull-up. Test data at TDI are loaded with a 4-MHz clock

signal connected to TCK. ‘1’ or ‘0’ on TMS causes a transition from one controller state to another; constant ’1’ on TMS leads to normal operation of the chip.

Table 8-1 Boundary Scan Sequence of PASSAT

TDI ->

Seq. No.	Pin	I/O	Number of Boundary Scan Cells	Constant Value In, Out, Enable
1			2	**00**
2			1	**0**
3			3	**100**
4			2	**00**
5			1	**0**
6			1	**0**
7			1	**0**
8			3	**001**
9			3	**011**
10			3	**111**
11			3	**000**
12			1	**0**
13			3	**100**
14			1	**0**
15			2	**00**
16			2	**11**
17			2	00
18			2	00
19			2	00
20			2	00
21			2	00
22			2	00
23			1	0
24			2	00
25			2	00
26			2	00

Seq. No.	Pin	I/O	Number of Boundary Scan Cells	Constant Value In, Out, Enable
27			2	00
28			2	00
29			2	00
30			2	00
31			2	00
32			1	0
33			3	000
34			3	000
35			3	000
36			2	00
37			2	00
38			2	00
39			2	00
40			2	00
41			2	00
42			2	00
43			2	00
44			3	000
45			3	000
46			1	0
47			1	0
48			3	000
49			3	0
50			3	000
51			3	000
52			1	0
53			3	000
54			3	000
55			3	000
56			3	000

Seq. No.	Pin	I/O	Number of Boundary Scan Cells	Constant Value In, Out, Enable
57			3	000
58			3	000
59			3	000
60			3	000
61			3	000
62			3	000
63			3	000
64			3	000
65			2	00
66			1	0
67			1	0
68			3	000
69			2	00
70			1	0
71			1	0
72			1	0
73			1	0
74			1	0

-> TDO

An input pin (I) uses one boundary scan cell (data in), an output pin (O) uses two cells (data out, enable) and an I/O-pin (I/O) uses three cells (data in, data out, enable). Note that some functional output and input pins of PASSAT are tested as I/O pins in boundary scan, hence using three cells. The boundary scan unit of PASSAT contains a total of n = 158 scan cells.

The right column of **Table 8-1** gives the initialization values of the cells.

The desired test mode is selected by serially loading a 3-bit instruction code into the instruction register via TDI (LSB first); see **Table 8-2.**

EXTEST is used to examine the interconnection of the devices on the board. In this test mode at first all input pins **capture** the current level on the corresponding external

Table 8-2 Boundary Scan Test Modes

Instruction (Bit 2 … 0)	Test Mode
000	EXTEST (external testing)
001	INTEST (internal testing)
010	SAMPLE/PRELOAD (snap-shot testing)
011	IDCODE (reading ID code)
111	BYPASS (bypass operation)
others	handled like BYPASS

interconnection line, whereas all output pins are held at constant values (‘0’ or ‘1’, according to **Table 8-1**). Then the contents of the boundary scan is **shifted** to TDO. At the same time the next scan vector is loaded from TDI. Subsequently all output pins are **updated** according to the new boundary scan contents and all input pins again capture the current external level afterwards, and so on.

INTEST supports internal testing of the chip, i.e. the output pins **capture** the current level on the corresponding internal line whereas all input pins are held on constant values (‘0’ or ‘1’, according to **Table 8-1**). The resulting boundary scan vector is **shifted** to TDO. The next test vector is serially loaded via TDI. Then all input pins are **updated** for the following test cycle.

Note: In capture IR-state the code ‘001’ is automatically loaded into the instruction register, i.e. if INTEST is wanted the shift IR-state does not need to be passed.

SAMPLE/PRELOAD is a test mode which provides a snap-shot of pin levels during normal operation.

IDCODE: A 32-bit identification register is serially read out via TDO. It contains the version number (4 bits), the device code (16 bits) and the manufacturer code (11 bits). The LSB is fixed to ‘1’.

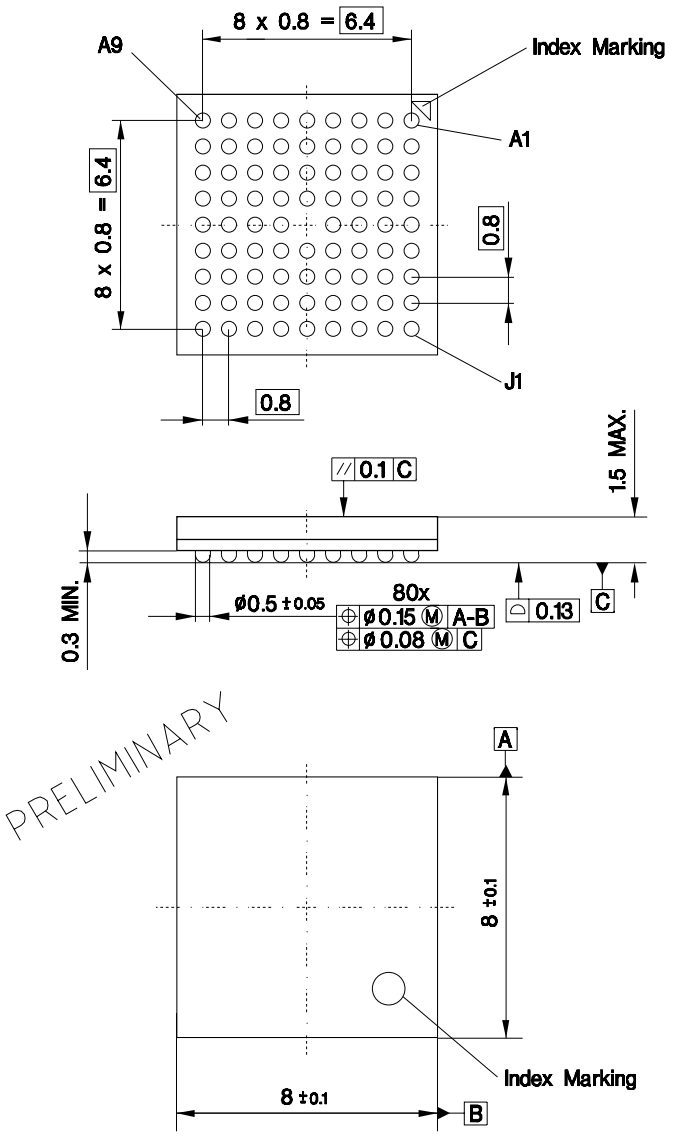
TDI ->	0001	0000 0000 0101 1111	0000 1000 001	1	-> TDO

Note: Since in test logic reset state the code ‘011’ is automatically loaded into the instruction register, the ID code can easily be read out in shift DR state which is reached by TMS = 0, 1, 0, 0.

BYPASS: A bit entering TDI is shifted to TDO after one TCK clock cycle.

9 Package Outlines

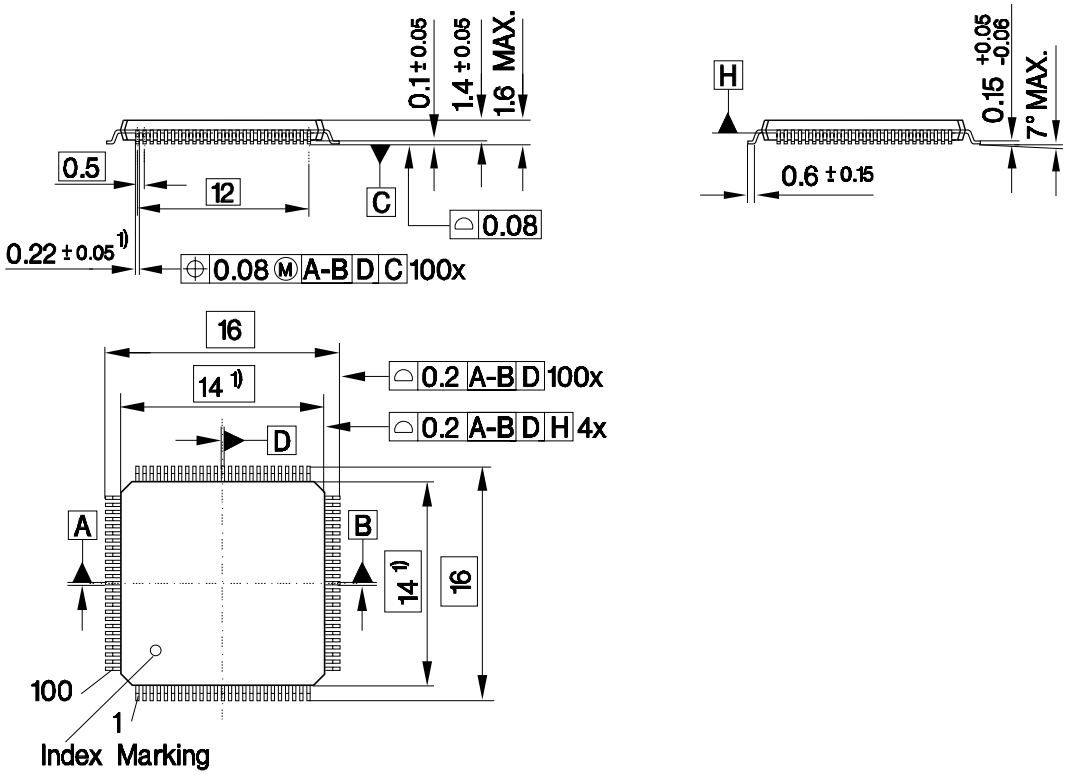
P-LFBGA-80-2
(Low-Profile Fine-Pitch Ball Grid Array)



Sorts of Packing
Package outlines for tubes, trays etc. are contained in our Data Book “Package Information”.

Dimensions in mm

P-TQFP-100-3
(Plastic Thin Quad Flat Package)



1) Does not include plastic or metal protrusion of 0.25 max. per side
2) Does not include dambar protrusion of 0.08 max. per side at max. material condition

GPP09189

Sorts of Packing
Package outlines for tubes, trays etc. are contained in our Data Book “Package Information”.

Dimensions in mm