

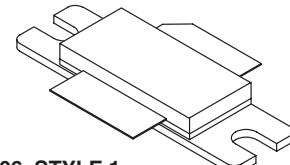
## The RF Sub-Micron MOSFET Line **RF Power Field Effect Transistors** N-Channel Enhancement-Mode Lateral MOSFETs

Designed for broadband commercial and industrial applications with frequencies from 865 to 895 MHz. The high gain and broadband performance of these devices make them ideal for large-signal, common-source amplifier applications in 26 volt base station equipment.

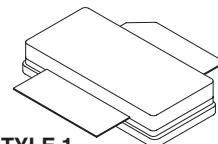
- Typical N-CDMA Performance @ 880 MHz, 26 Volts,  $I_{DQ} = 1100$  mA  
IS-95 CDMA Pilot, Sync, Paging, Traffic Codes 8 Through 13  
Output Power — 25 Watts Avg.  
Power Gain — 17.8 dB  
Efficiency — 25%  
Adjacent Channel Power —  
    750 kHz: -47 dBc @ 30 kHz BW
- Internally Matched, for Ease of Use
- High Gain, High Efficiency and High Linearity
- Integrated ESD Protection
- Designed for Maximum Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 26 Vdc, 880 MHz, 135 Watts CW Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters
- Available in Tape and Reel. R3 Suffix = 250 Units per 32 mm, 13 inch Reel.
- Available with Low Gold Plating Thickness on Leads. L Suffix Indicates 40 $\mu$ " Nominal.

**MRF9135L**  
**MRF9135LR3**  
**MRF9135LSR3**

880 MHz, 135 W, 26 V  
LATERAL N-CHANNEL  
RF POWER MOSFETs



CASE 465-06, STYLE 1  
NI-780  
MRF9135L



CASE 465A-06, STYLE 1  
NI-780S  
MRF9135LSR3

### MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Drain-Source Voltage	$V_{DSS}$	65	Vdc
Gate-Source Voltage	$V_{GS}$	+15, -0.5	Vdc
Total Device Dissipation @ $T_C >= 25^\circ\text{C}$ Derate above $25^\circ\text{C}$	$P_D$	298 1.7	Watts W/ $^\circ\text{C}$
Storage Temperature Range	$T_{stg}$	-65 to +200	$^\circ\text{C}$
Operating Junction Temperature	$T_J$	200	$^\circ\text{C}$

### ESD PROTECTION CHARACTERISTICS

Test Conditions	Class
Human Body Model	1 (Minimum)
Machine Model	M2 (Minimum)
Charge Device Model	C7 (Minimum)

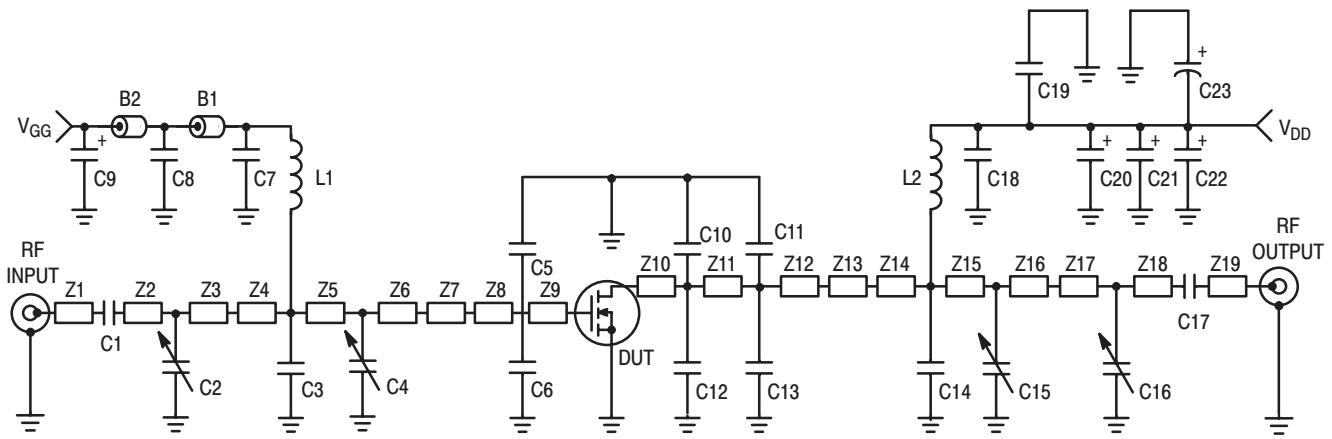
### THERMAL CHARACTERISTICS

Characteristic	Symbol	Max	Unit
Thermal Resistance, Junction to Case	$R_{\theta JC}$	0.6	$^\circ\text{C/W}$

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

**ELECTRICAL CHARACTERISTICS** ( $T_C = 25^\circ\text{C}$ , 50 ohm system unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
<b>OFF CHARACTERISTICS</b>					
Zero Gate Voltage Drain Leakage Current ( $V_{DS} = 65 \text{ Vdc}$ , $V_{GS} = 0 \text{ Vdc}$ )	$I_{DSS}$	—	—	10	$\mu\text{A dc}$
Zero Gate Voltage Drain Leakage Current ( $V_{DS} = 26 \text{ Vdc}$ , $V_{GS} = 0 \text{ Vdc}$ )	$I_{DSS}$	—	—	1	$\mu\text{A dc}$
Gate-Source Leakage Current ( $V_{GS} = 5 \text{ Vdc}$ , $V_{DS} = 0 \text{ Vdc}$ )	$I_{GSS}$	—	—	1	$\mu\text{A dc}$
<b>ON CHARACTERISTICS</b>					
Gate Threshold Voltage ( $V_{DS} = 10 \text{ Vdc}$ , $I_D = 450 \mu\text{A}$ )	$V_{GS(\text{th})}$	2	2.8	4	Vdc
Gate Quiescent Voltage ( $V_{DS} = 26 \text{ Vdc}$ , $I_D = 1100 \text{ mA dc}$ )	$V_{GS(Q)}$	3	3.7	5	Vdc
Drain-Source On-Voltage ( $V_{GS} = 10 \text{ Vdc}$ , $I_D = 3 \text{ Adc}$ )	$V_{DS(\text{on})}$	—	0.19	0.4	Vdc
Forward Transconductance ( $V_{DS} = 10 \text{ Vdc}$ , $I_D = 9 \text{ Adc}$ )	$g_{fs}$	—	12	—	S
<b>DYNAMIC CHARACTERISTICS</b>					
Output Capacitance ( $V_{DS} = 26 \text{ Vdc} \pm 30 \text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$ )	$C_{oss}$	—	109	—	pF
Reverse Transfer Capacitance ( $V_{DS} = 26 \text{ Vdc} \pm 30 \text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0 \text{ Vdc}$ )	$C_{rss}$	—	4.4	—	pF
<b>FUNCTIONAL TESTS</b> (In Motorola Test Fixture) Single-Carrier N-CDMA, 1.2288 MHz Channel Bandwidth Carrier, Peak/Avg. Ratio = 9.8 dB @ 0.01% Probability on CCDF					
Common-Source Amplifier Power Gain ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 880.0 \text{ MHz}$ )	$G_{ps}$	16	17.8	—	dB
Drain Efficiency ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 880.0 \text{ MHz}$ )	$\eta$	22	25	—	%
Adjacent Channel Power Ratio ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 880.0 \text{ MHz}$ ; ACPR @ 25 W, 1.23 MHz Bandwidth, 750 kHz Channel Spacing)	ACPR	—	-47	-45	dBc
Input Return Loss ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 880.0 \text{ MHz}$ )	IRL	—	-13.5	-9	dB
Common-Source Amplifier Power Gain ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 865 \text{ MHz}$ and $895 \text{ MHz}$ )	$G_{ps}$	—	17	—	dB
Drain Efficiency ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 865 \text{ MHz}$ and $895 \text{ MHz}$ )	$\eta$	—	24	—	%
Adjacent Channel Power Ratio ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 865 \text{ MHz}$ and $895 \text{ MHz}$ ; ACPR @ 25 W, 1.23 MHz Bandwidth, 750 kHz Channel Spacing)	ACPR	—	-46	—	dBc
Input Return Loss ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 25 \text{ W Avg. N-CDMA}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 865 \text{ MHz}$ and $895 \text{ MHz}$ )	IRL	—	-12.5	—	dB
Output Mismatch Stress ( $V_{DD} = 26 \text{ Vdc}$ , $P_{out} = 135 \text{ W CW}$ , $I_{DQ} = 1100 \text{ mA}$ , $f = 880.0 \text{ MHz}$ , VSWR = 10:1, All Phase Angles at Frequency of Tests)	$\Psi$	No Degradation In Output Power			

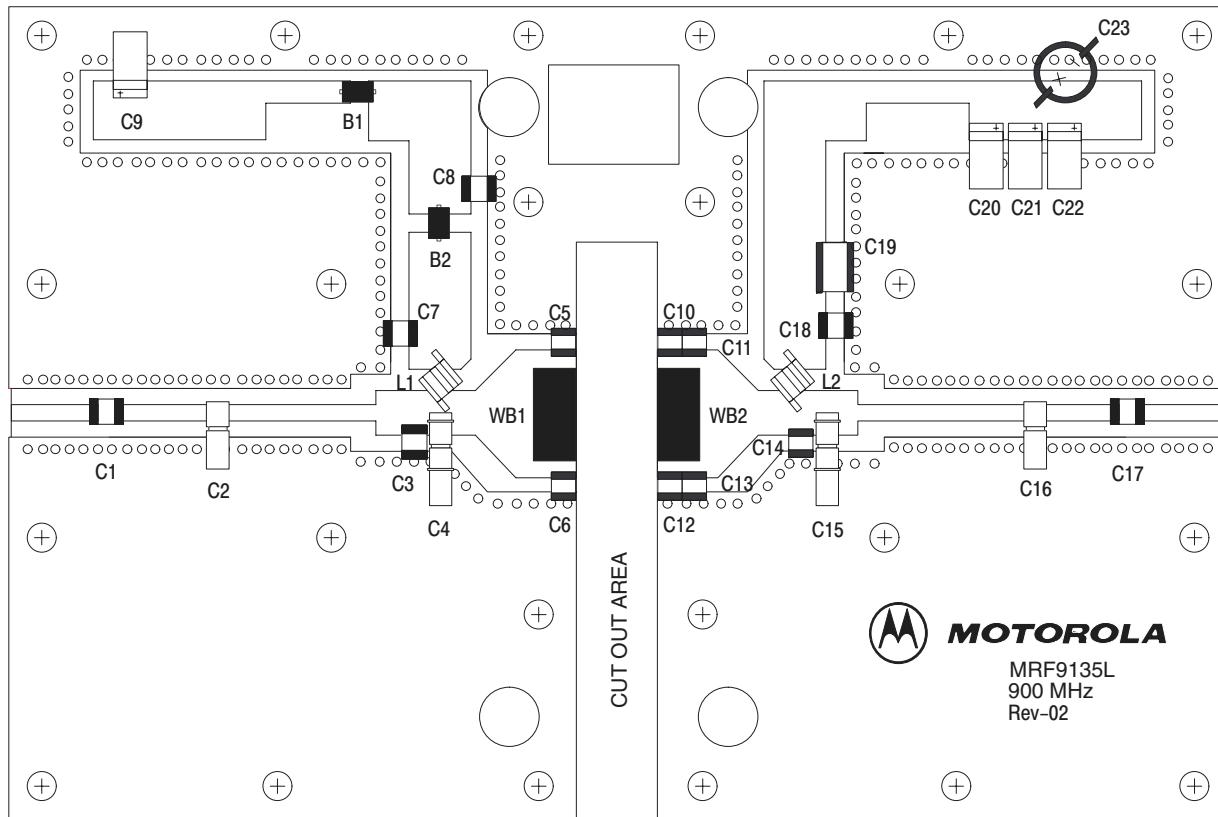


Z1	0.430" x 0.080" Microstrip	Z11	0.105" x 0.630" Microstrip
Z2	0.430" x 0.080" Microstrip	Z12	0.145" x 0.630" Microstrip
Z3	0.800" x 0.080" Microstrip	Z13	0.200" x 0.630" x 0.220" Taper
Z4	0.200" x 0.220" Microstrip	Z14	0.180" x 0.220" Microstrip
Z5	0.110" x 0.220" Microstrip	Z15	0.110" x 0.220" Microstrip
Z6	0.175" x 0.220" Microstrip	Z16	0.200" x 0.220" Microstrip
Z7	0.200" x 0.220" x 0.630" Taper	Z17	0.900" x 0.080" Microstrip
Z8	0.250" x 0.630" Microstrip	Z18	0.360" x 0.080" Microstrip
Z9	0.050" x 0.630" Microstrip	Z19	0.410" x 0.080" Microstrip
Z10	0.050" x 0.630" Microstrip		

Figure 1. 880 MHz Test Circuit Schematic

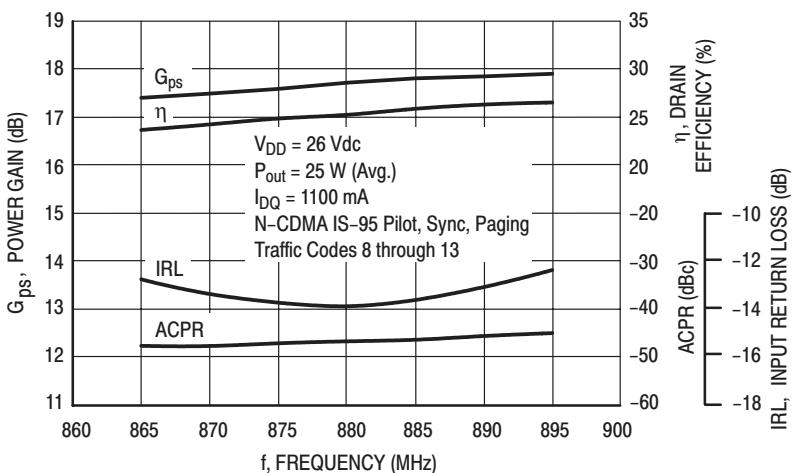
Table 1. 880 MHz Test Circuit Component Designations and Values

Part	Description	Value, P/N or DWG	Manufacturer
B1, B2	Short Ferrite Beads, Surface Mount	95F786	Newark
C1, C7, C17, C18	47 pF Chip Capacitors, B Case	100B470JP 500X	ATC
C2, C16	0.6–4.5 Gigatrim Variable Capacitors	44F3360	Newark
C3	8.2 pF Chip Capacitor, B Case	100B8R2BP 500X	ATC
C4, C15	0.8–8.0 Gigatrim Variable Capacitors	44F3360	Newark
C5, C6	12 pF Chip Capacitors, B Case	100B120JP 500X	ATC
C8	20K pF Chip Capacitor, B Case	200B203MP50X	ATC
C9, C20, C21, C22	10 $\mu$ F, 35 V Tantalum Capacitors	93F2975	Newark
C10, C11, C12, C13	7.5 pF Chip Capacitors, B Case	100B7R5JP 500X	ATC
C14	11 pF Chip Capacitor, B Case	100B110JP 500X	ATC
C19	0.56 $\mu$ F, 50 V Chip Capacitor	C1825C564K5RA7800	Kemet
C23	470 $\mu$ F Electrolytic Capacitor	14F185	Newark
L1, L2	12.5 nH Coilcraft inductors	A04T-5	Coilcraft
WB1, WB2	10 mil Brass Shim (0.205 x 0.530)	RF-Design Lab	RF-Design Lab
PCB	Etched Circuit Board	900 MHz 4X6 Cobra Rev 02	CMR
Bedstead	Circuit Bedstead	DWG #990528JAM2	RF-Design Lab
Board Material	30 mil Glass Teflon <sup>®</sup> , $\epsilon_r = 2.55$ , 2 oz Cu	GX-0300-55-22	Arlon

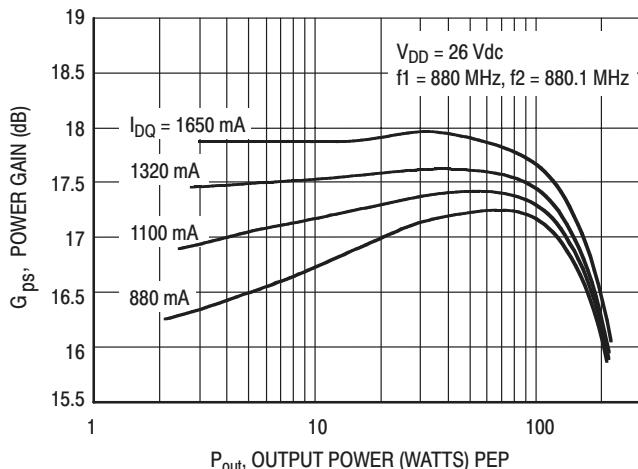


**Figure 2. 880 MHz Test Circuit Component Layout**

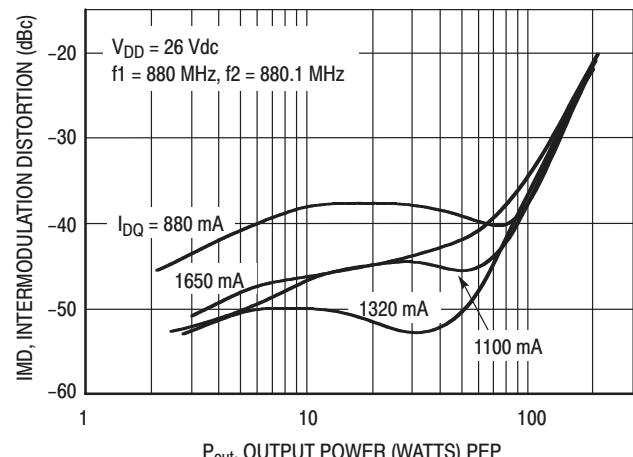
## TYPICAL CHARACTERISTICS



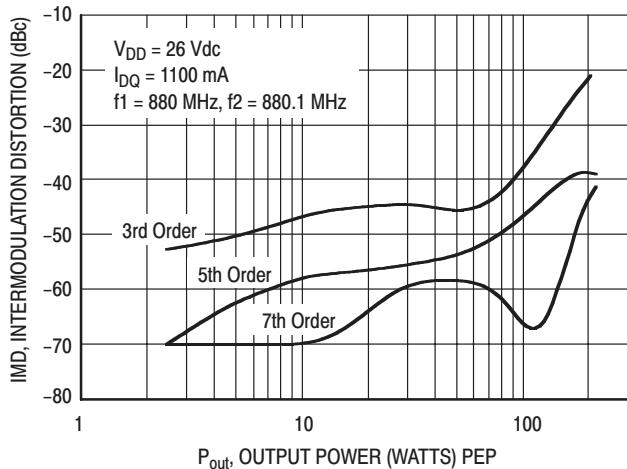
**Figure 3. Class AB Broadband Circuit Performance**



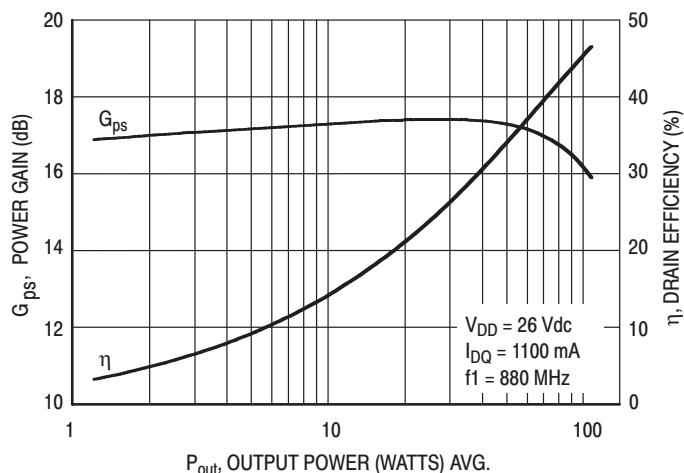
**Figure 4. Power Gain versus Output Power**



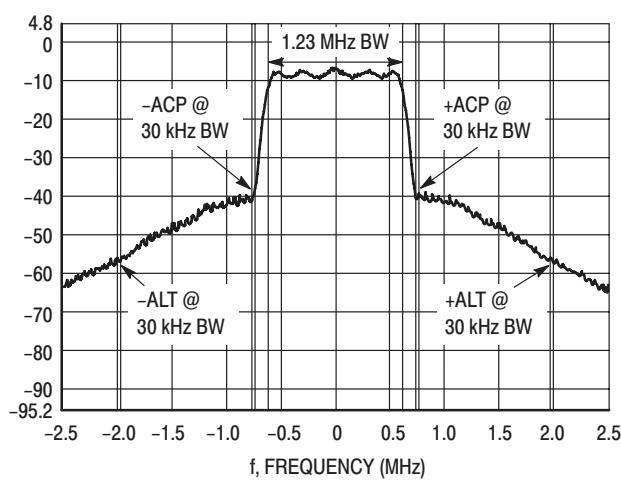
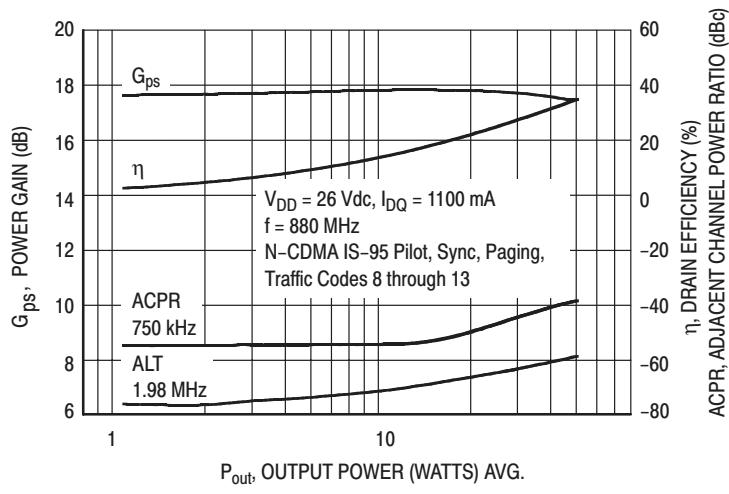
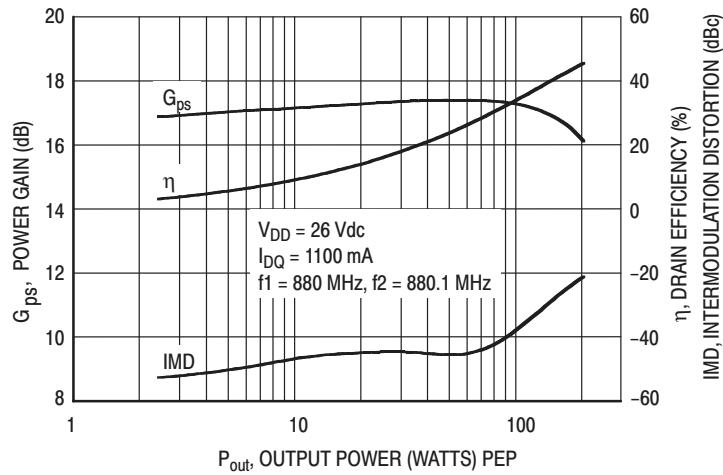
**Figure 5. Intermodulation Distortion versus Output Power**

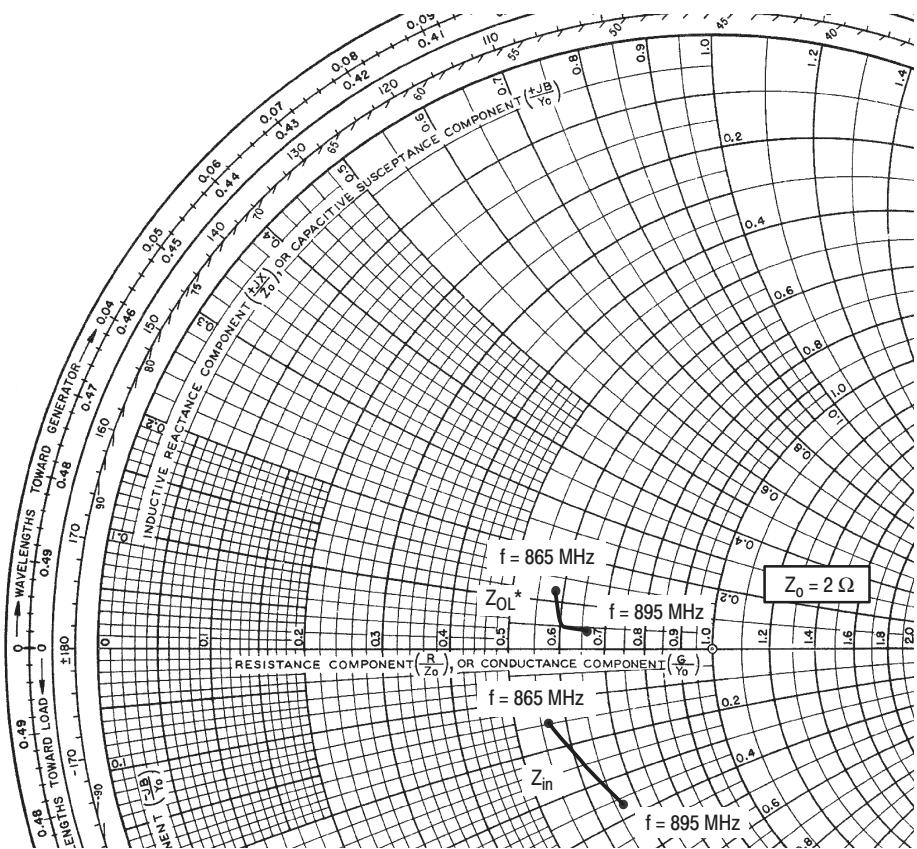


**Figure 6. Intermodulation Distortion Products versus Output Power**



**Figure 7. Power Gain and Efficiency versus Output Power**





$V_{DD} = 26 \text{ V}$ ,  $I_{DQ} = 1100 \text{ mA}$ ,  $P_{out} = 25 \text{ W Avg.}$

$f$ MHz	$Z_{in}$ $\Omega$	$Z_{OL^*}$ $\Omega$
865	$1.15 - j0.3$	$1.17 + j0.24$
880	$1.25 - j0.5$	$1.22 + j0.1$
895	$1.35 - j0.75$	$1.32 + j0.07$

$Z_{in}$  = Complex conjugate of source impedance.

$Z_{OL^*}$  = Complex conjugate of the optimum load impedance at a given output power, voltage, IMD, bias current and frequency.

Note:  $Z_{OL^*}$  was chosen based on tradeoffs between gain, output power, drain efficiency and intermodulation distortion.

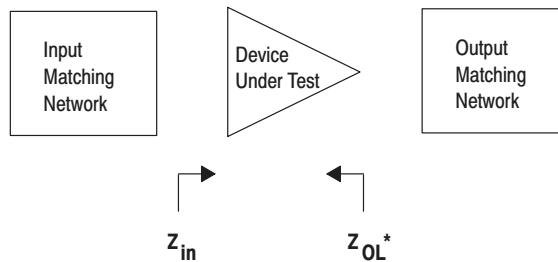


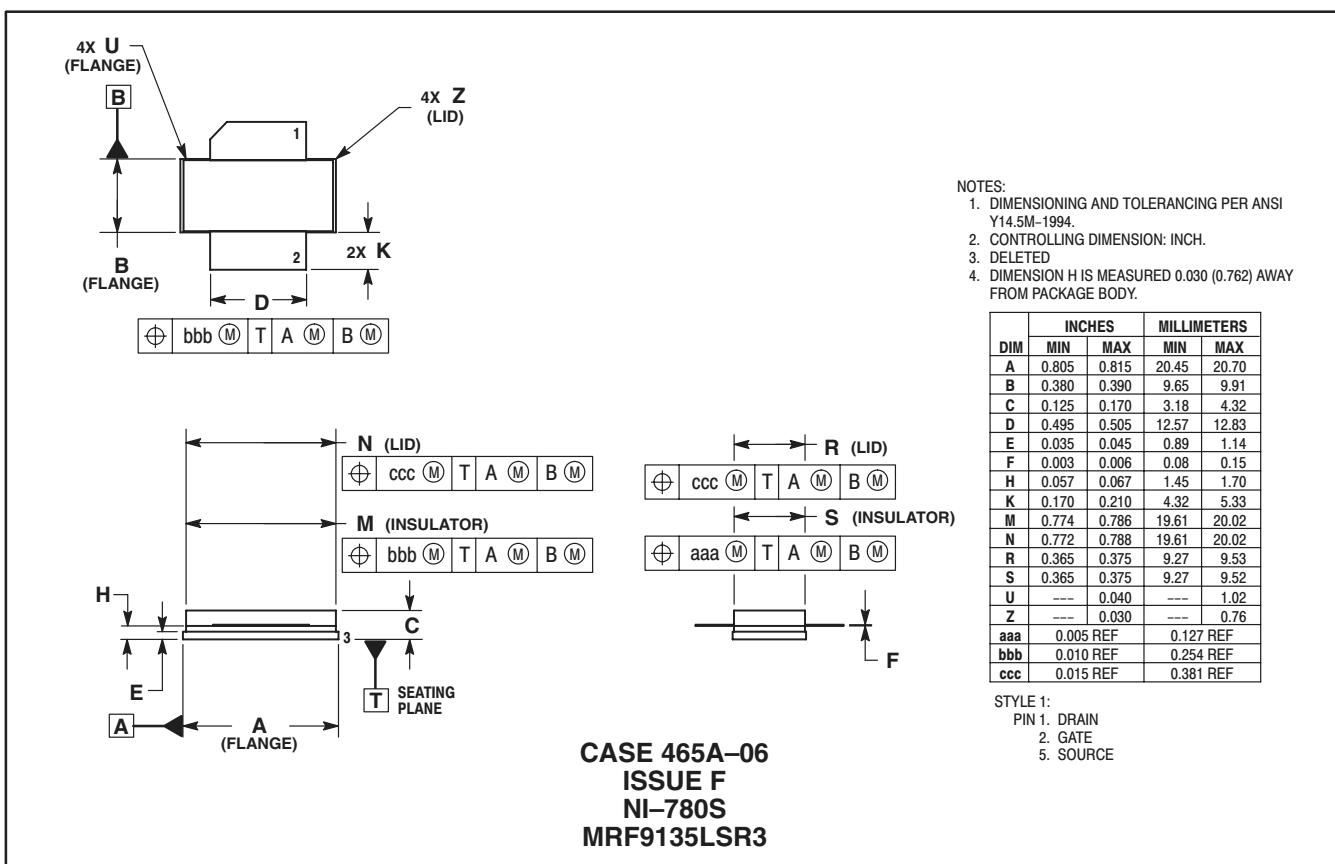
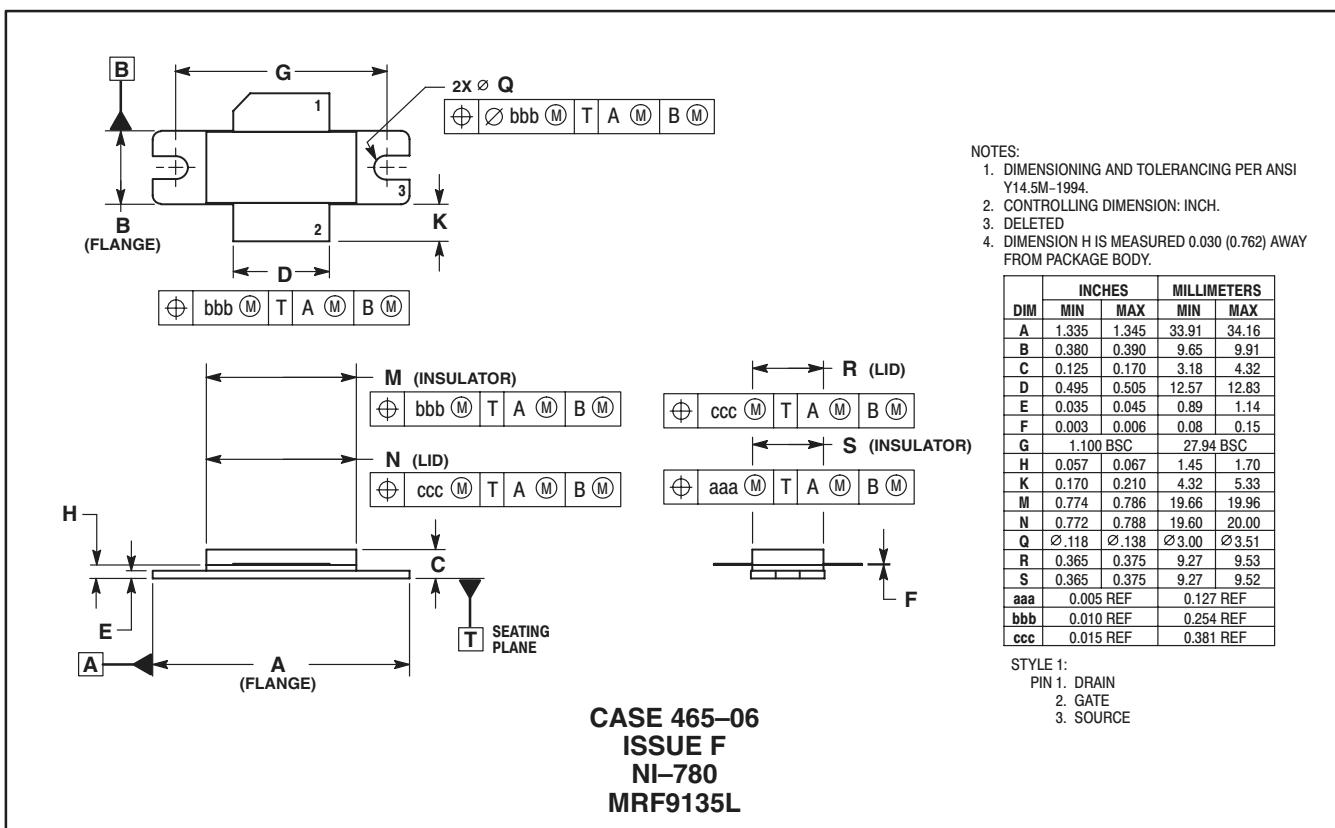
Figure 11. Series Equivalent Input and Output Impedance

# NOTES

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## PACKAGE DIMENSIONS



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