## User's Manual

NEC

# $\mu$ PD789167, 789177, 789167Y, 789177Y Subseries 

8-Bit Single-Chip Microcontrollers

| $\mu$ PD789166 | $\mu$ PD789166Y | $\mu$ PD789166(A1) |
| :--- | :--- | :--- |
| $\mu$ PD789167 | $\mu$ PD789167Y | $\mu$ PD789167(A1) |
| $\mu$ PD789176 | $\mu$ PD789176Y | $\mu$ PD789176(A1) |
| $\mu$ PD789177 | $\mu$ PD789177Y | $\mu$ PD789177(A1) |
| $\mu$ PD78F9177 | $\mu$ PD78F9177Y | $\mu$ PD78F9177A(A1) |
| $\mu$ PD78F9177A | $\mu$ PD78F9177AY | $\mu$ PD789166(A2) |
| $\mu$ PD789166(A) | $\mu$ PD789166Y(A) | $\mu$ PD789167(A2) |
| $\mu$ PD789167(A) | $\mu$ PD789167Y(A) | $\mu$ PD789176(A2) |
| $\mu$ PD789176(A) | $\mu$ PD789176Y(A) | $\mu$ PD789177(A2) |
| $\mu$ PD789177(A) | $\mu$ PD789177Y(A) |  |
| $\mu$ PD78F9177A(A) $\mu$ PD78F9177AY(A) |  |  |

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[MEMO]

## (1) VOLTAGE APPLICATION WAVEFORM AT INPUT PIN

Waveform distortion due to input noise or a reflected wave may cause malfunction. If the input of the CMOS device stays in the area between $\mathrm{V}_{\mathrm{IL}}$ (MAX) and $\mathrm{V}_{\mathrm{IH}}$ (MIN) due to noise, etc., the device may malfunction. Take care to prevent chattering noise from entering the device when the input level is fixed, and also in the transition period when the input level passes through the area between VIL (MAX) and VIH (MIN).

## (2) HANDLING OF UNUSED INPUT PINS

Unconnected CMOS device inputs can be cause of malfunction. If an input pin is unconnected, it is possible that an internal input level may be generated due to noise, etc., causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using pull-up or pull-down circuitry. Each unused pin should be connected to Vod or GND via a resistor if there is a possibility that it will be an output pin. All handling related to unused pins must be judged separately for each device and according to related specifications governing the device.

## (3) PRECAUTION AGAINST ESD

A strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it when it has occurred. Environmental control must be adequate. When it is dry, a humidifier should be used. It is recommended to avoid using insulators that easily build up static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work benches and floors should be grounded. The operator should be grounded using a wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with mounted semiconductor devices.
(4) STATUS BEFORE INITIALIZATION

Power-on does not necessarily define the initial status of a MOS device. Immediately after the power source is turned ON, devices with reset functions have not yet been initialized. Hence, power-on does not guarantee output pin levels, I/O settings or contents of registers. A device is not initialized until the reset signal is received. A reset operation must be executed immediately after power-on for devices with reset functions.
(5) POWER ON/OFF SEQUENCE

In the case of a device that uses different power supplies for the internal operation and external interface, as a rule, switch on the external power supply after switching on the internal power supply. When switching the power supply off, as a rule, switch off the external power supply and then the internal power supply. Use of the reverse power on/off sequences may result in the application of an overvoltage to the internal elements of the device, causing malfunction and degradation of internal elements due to the passage of an abnormal current.
The correct power on/off sequence must be judged separately for each device and according to related specifications governing the device.

## (6) INPUT OF SIGNAL DURING POWER OFF STATE

Do not input signals or an I/O pull-up power supply while the device is not powered. The current injection that results from input of such a signal or I/O pull-up power supply may cause malfunction and the abnormal current that passes in the device at this time may cause degradation of internal elements. Input of signals during the power off state must be judged separately for each device and according to related specifications governing the device.

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- Availability of related technical literature
- Development environment specifications (for example, specifications for third-party tools and components, host computers, power plugs, AC supply voltages, and so forth)
- Network requirements

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## INTRODUCTION

Readers

Purpose

This manual is intended for user engineers who wish to understand the functions of the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries in order to design and develop its application systems and programs.
Target products:

- $\mu$ PD789167 Subseries: $\quad \mu$ PD789166, 789167, 789166(A), 789167(A), 789166(A1), 789167(A1), 789166(A2), 789167(A2)
- $\mu$ PD789177 Subseries: $\quad \mu$ PD789176, 789177, 78F9177, 78F9177A, 789176(A), 789177(A), 78F9177A(A), 789176(A1), 789177(A1), 78F9177A(A1), 789176(A2), 789177(A2)
- $\mu$ PD789167Y Subseries: $\mu$ PD789166Y, 789167Y, 789166Y(A), 789167Y(A)
$\bullet \mu$ PD789177Y Subseries: $\mu$ PD789176Y, 789177Y, 78F9177Y, 78F9177AY, 789176Y(A), 789177Y(A), 78F9177AY(A)
The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries is a generic term for all the target devices in this manual.
The generic terms used in this manual indicate the following products.
"Standard quality grade products"... $\mu$ PD789166, 789167, 789176, 789177, 78F9177, 78F9177A, 789166Y, 789167Y, 789176Y, 789177Y, 78F9177Y, 78F9177AY
"(A) products"... $\quad \mu$ PD789166(A), 789167(A), 789176(A), 789177(A), 78F9177A(A), $789166 \mathrm{Y}(\mathrm{A}), 789167 \mathrm{Y}(\mathrm{A}), 789176 \mathrm{Y}(\mathrm{A})$, $789177 \mathrm{Y}(\mathrm{A}), 78 \mathrm{F9177AY}(\mathrm{~A})$
"(A1) products"... $\quad \mu$ PD789166(A1), 789167(A1), 789176(A1), 789177(A1), 78F9177A(A1)
"(A2) products"... $\quad \mu$ PD789166(A2), 789167(A2), 789176(A2), 789177(A2)
"Mask ROM versions"... $\mu$ PD789166, 789167, 789176, 789177, 789166Y, 789167Y, 789176Y, 789177Y, 789166(A), 789167(A), $789176(\mathrm{~A}), 789177(\mathrm{~A}), 789166 \mathrm{Y}(\mathrm{A}), 789167 \mathrm{Y}(\mathrm{A})$, 789176Y(A), 789177Y(A), 789166(A1), 789167(A1), 789176(A1), 789177(A1), 789166(A2), 789167(A2), 789176(A2), 789177(A2)
"Flash memory versions"... $\mu$ PD78F9177, 78F9177A, 78F9177A(A), 78F9177A(A1), 78F9177Y, 78F9177AY, 78F9177AY(A)

This manual is intended to give users an understanding of the functions described in the Organization below.

## Organization

The $\mu$ PD789167, 789177, 789167Y, 789177Y Subseries manual is divided into two parts: this manual and the instruction manual (common to the $78 \mathrm{~K} / 0$ S Series).


- Pin functions
- Internal block functions
- Interrupts
- Other internal peripheral functions
- Electrical specifications

- CPU function
- Instruction set
- Instruction description

It is assumed that the readers of this manual have general knowledge of electric engineering, logic circuits, and microcontrollers.
$\diamond$ For users who use this document as the manual for the $\mu$ PD789166(A), 789167(A), 789176(A), 789177(A), $789166 Y(A), 789167 Y(A), 789176 Y(A), 789177 Y(A)$, 789166(A1), 789167(A1), 789176(A1), 789177(A1), 789166(A2), 789167(A2), 789176(A2), 789177(A2), 78F9177A(A), 78F9177AY(A), and 78F9177A(A1)
$\rightarrow$ The only differences between standard products and (A) products, (A1) products, and (A2) products are quality grades, power supply voltage, operating ambient temperature, minimum instruction execution time, and electrical specifications. (Refer to 1.10 Differences Between Standard Quality Grade Products and (A) Products, (A1) Products, and (A2) Products, and 2.10 Differences Between Standard Quality Grade Products and (A) Products.) For (A) products, (A1) products, and (A2) products, read the part numbers indicated in Chapters 3 to 22 in the following manner.

```
\muPD789166 }\quad->\quad\mu\mathrm{ PD789166(A), 789166(A1), 789166(A2)
\muPD789167 }->\mu\mathrm{ PD789167(A), 789167(A1), 789167(A2)
\muPD789176 }->\mu\mathrm{ PD789176(A), 789176(A1), 789176(A2)
\muPD789177 }->\mu\mathrm{ PD789177(A), 789177(A1), 789177(A2)
\muPD789166Y }->\mu\mathrm{ PD789166Y(A)
\muPD789167Y }->\mu\textrm{PD789167Y(A)
\muPD789176Y }->\mu\textrm{PD}789176\textrm{Y}(\textrm{A}
\muPD789177Y }->\mu\textrm{PD}789177\textrm{Y}(\textrm{A}
\muPD78F9177A }->\mu\mathrm{ PD789177A(A), 78F9177A(A1)
\muPD78F9177AY }->\mu\mathrm{ PD78F9177AY(A)
```

$\diamond$ To understand the overall functions of the $\mu \mathrm{PD} 789167,789177,789167 \mathrm{Y}$, and 789177Y Subseries
$\rightarrow$ Read this manual in the order of the CONTENTS.
$\diamond$ How to read register formats
$\rightarrow$ The name of a bit whose number is enclosed with $<>$ is reserved in the assembler and is defined as an sfr variable by the \#pragma sfr directive in the C compiler.
$\diamond$ To learn the detailed functions of a register whose register name is known
$\rightarrow$ See APPENDIX C REGISTER INDEX.
$\diamond$ To learn the details of the instruction functions of the 78K/0S Series
$\rightarrow$ Refer to 78K/0S Series Instructions User's Manual (U11047E) separately available.
$\diamond$ To know the electrical specifications of the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries
$\rightarrow$ Refer to CHAPTER 23 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78916x, 17x, 16xY, 17xY, 16x(A), 17x(A), 16xY(A), 17xY(A)), CHAPTER 25 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78916x(A1), 78917x(A1), 78916x(A2), 78917x(A2)), CHAPTER 27 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177A, 78F9177AY, 78F9177A(A), 78F9177AY(A)), CHAPTER 28 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177, 78F9177AY), and CHAPTER 30 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177A(A1)).

Caution The application examples in this manual are created for "Standard" quality grade products for general electric equipment. When using the application examples in this manual for purposes which require "Special" quality grades, thoroughly examine the quality grade of each part and circuit actually used.

Differences between $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries
The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries differ in their package type, A/D converter resolution, and serial interface configuration.

| Item | Suk | $\mu \mathrm{PD} 789167$ | $\mu \mathrm{PD} 789177$ | $\mu \mathrm{PD} 789167 \mathrm{Y}$ | $\mu \mathrm{PD} 789177 \mathrm{Y}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Package |  | - 44-pin plastic LQFP |  | - 44-pin plastic LQFP <br> - 48-pin plastic TQFP |  |
| IC2 pin |  | Not provided |  | Provided |  |
| A/D converter resolution |  | 8 bits | 10 bits | 8 bits | 10 bits |
| Serial interface configuration | 3-wire serial I/O mode | 1 channel |  |  |  |
|  | SMB0 | Not provided |  | 1 channel |  |

Configuration of This Manual This manual uses separate chapters to describe the functions that vary between the subseries. The chapters related to each subseries are listed below.
For information about a certain subseries, see only the chapters indicated by checkmarks in that subseries' column.

| Chapter | $\mu$ PD789167 <br> Subseries | $\mu$ PD789177 <br> Subseries | $\begin{gathered} \mu \mathrm{PD} 789167 \mathrm{Y} \\ \text { Subseries } \end{gathered}$ | $\mu$ PD789177Y <br> Subseries |
| :---: | :---: | :---: | :---: | :---: |
| CHAPTER 1 GENERAL ( $\mu$ PD789167 AND 789177 SUBSERIES) | $\checkmark$ | $\checkmark$ | - | - |
| CHAPTER 2 GENERAL ( $\mu$ PD789167Y AND 789177Y SUBSERIES) | - | - | $\checkmark$ | $\checkmark$ |
| CHAPTER 3 PIN FUNCTIONS ( $\mu$ PD789167 AND 789177 SUBSERIES | $\checkmark$ | $\checkmark$ | - | - |
| CHAPTER 4 PIN FUNCTIONS ( $\mu$ PD789167Y AND 789177Y SUBSERIES) | - | - | $\checkmark$ | $\checkmark$ |
| CHAPTER 5 CPU ARCHITECTURE | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 6 PORT FUNCTIONS | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 7 CLOCK GENERATOR | $\checkmark$ | $\checkmark$ | $\sqrt{ }$ | $\checkmark$ |
| CHAPTER 8 16-BIT TIMER 90 | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 9 8-BIT TIMER/EVENT COUNTERS 80 TO 82 | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\sqrt{ }$ |
| CHAPTER 10 WATCH TIMER | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 11 WATCHDOG TIMER | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 12 8-BIT A/D CONVERTER ( $\mu$ PD789167 AND 789167Y SUBSERIES) | $\checkmark$ | - | $\checkmark$ | - |
| CHAPTER 13 10-BIT A/D CONVERTER ( $\mu$ PD789177 AND 789177Y SUBSERIES) | - | $\checkmark$ | - | $\checkmark$ |
| CHAPTER 14 SERIAL INTERFACE 20 | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 15 SMB0 ( $\mu$ PD789167Y AND 789177Y SUBSERIES) | - | - | $\checkmark$ | $\checkmark$ |
| CHAPTER 16 MULTIPLIER | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 17 INTERRUPT FUNCTIONS | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 18 STANDBY FUNCTION | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 19 RESET FUNCTION | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 20 FLASH MEMORY VERSION | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 21 MASK OPTION | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
| CHAPTER 22 INSTRUCTION SET | $\checkmark$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |


| Conventions | Data significance: | Higher digits on the left and lower digits on the right |
| :---: | :---: | :---: |
|  | Active low representation: | $\overline{x \times \times}$ (overscore over pin or signal name) |
|  | Note: | Footnote for item marked with Note in the text |
|  | Caution: | Information requiring particular attention |
|  | Remark: | Supplementary information |
|  | Numerical representation: | Binary ... $\times \times \times \times$ or $\times x \times \times$ B |
|  |  | Decimal ... $\times \times \times \times$ |
|  |  | Hexadecimal ... $\times \times \times \times \mathrm{H}$ |

Related Documents
The related documents indicated in this publication may include preliminary versions. However, preliminary versions are not marked as such.

## Documents Related to Devices

| Document Name | Document No. |
| :--- | :--- |
| $\mu$ PD789167, 789177, 789167Y, 789177Y Subseries User's Manual | This manual |
| 78K/0S Series Instructions User's Manual | U11047E |

$\star$ Documents Related to Development Tools (Software) (User's Manuals)

| Document Name |  | Document No. |
| :--- | :--- | :--- |
| RA78KOS Assembler Package | Operation | U16656E |
|  | Language | U14877E |
|  | Structured Assembly Language | U11623E |
| CC78KOS C Compiler | Operation | U16654E |
|  | Language | U14872E |
|  | Operation | U16768E |
|  | External Parts User Open Interface Specifications | U15802E |
| ID78K0S-NS Ver. 2.52 Integrated Debugger | Operation | U16584E |
| PM plus Ver. 5.10 |  | U16569E |

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Documents Related to Development Tools (Hardware) (User's Manuals)

| Document Name | Document No. |
| :--- | :---: |
| IE-78K0S-NS In-Circuit Emulator | U13549E |
| IE-78K0S-NS-A In-Circuit Emulator | U15207E |
| IE-789177-NS-EM1 Emulation Board | U14621E |

## Documents Related to Flash Memory Writing

| Document Name | Document No. |
| :--- | :---: |
| PG-FP3 Flash Memory Programmer User's Manual | U13502E |
| PG-FP4 Flash Memory Programmer User's Manual | U15260E |

## Other Related Documents

| Document Name | Document No. |
| :--- | :---: |
| SEMICONDUCTORS SELECTION GUIDE Product \& Packages | X13769X |
| Semiconductor Device Mount Manual | Note |
| Quality Grades on NEC Semiconductor Device | C11531E |
| NEC Semiconductor Device Reliability/Quality Control System | C10983E |
| Guide to Prevent Damage for Semiconductor Devices by Electrostatic Discharge (ESD) | C11892E |

Note See the "Semiconductor Device Mount Manual" website (http://www.necel.com/pkg/en/mount/index.html)

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## CHAPTER 1 GENERAL ( $\mu$ PD789167 AND 789177 SUBSERIES)

### 1.1 Expanded-Specification Products and Conventional Products

The expanded-specification products and conventional products refer to the following products.

Expanded-specification product... Products with a rank ${ }^{\text {Note } 1}$ other than K

- Mask ROM versions for which orders were received after December 1, 2001 (except (A1) products and (A2) products ${ }^{\text {Note } 2}$ ).
- $\mu$ PD78F9177A, 78F9177A(A)

Conventional product... Products with rank ${ }^{\text {Note } 1} \mathrm{~K}$

- Products other than the above expanded-specification products.

Notes 1. The rank is indicated by the 5th digit from the left in the lot number marked on the package.

2. For the (A1) products and (A2) products, refer to 1.10 Differences Between Standard Quality Grade Products and (A) Products, (A1) Products, and (A2) Products.

Expanded-specification products and conventional products differ in operating frequency ratings. The differences are shown in Table 1-1.

Table 1-1. Differences Between Expanded-Specification Products and Conventional Products

| Power Supply Voltage (VDD) | Guaranteed Operating Speed (Operating Frequency) |  |
| :--- | :--- | :--- |
|  | Conventional Products | Expanded-Specification Products |
| 4.5 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $10 \mathrm{MHz}(0.2 \mu \mathrm{~s})$ |
| 3.0 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $6 \mathrm{MHz}(0.33 \mu \mathrm{~s})$ |
| 2.7 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ |
| 1.8 to 5.5 V | $1.25 \mathrm{MHz}(1.6 \mu \mathrm{~s})$ | $1.25 \mathrm{MHz}(1.6 \mu \mathrm{~s})$ |

Remark The values in parentheses indicate the minimum instruction execution time.

### 1.2 Features

- ROM and RAM capacity

| Product Name | Program Memory (ROM) |  | Data Memory (Internal High-Speed RAM) |
| :---: | :---: | :---: | :---: |
| $\mu$ PD789166, 789176, 789166(A), 789176(A), 789166(A1), 789176(A1), 789166(A2), 789176(A2) | Mask ROM | 16 KB | 512 bytes |
| $\begin{aligned} & \mu \mathrm{PD} 789167,789177,789167(\mathrm{~A}), 789177(\mathrm{~A}), 789167(\mathrm{~A} 1), \\ & 789177(\mathrm{~A} 1), 789167(\mathrm{~A} 2), 789177(\mathrm{~A} 2) \end{aligned}$ |  | 24 KB |  |
| $\mu \mathrm{PD} 78 \mathrm{F9177}$, 78F9177A, 78F9177A(A), 78F9177A(A1) | Flash memory | 24 KB |  |

- Minimum instruction execution time changeable from high-speed ( $0.2 \mu \mathrm{~s}$ : Main system clock 10.0 MHz operation ${ }^{\text {Note }}$ ) to ultra-low speed ( $122 \mu \mathrm{~s}$ : Subsystem clock 32.768 kHz operation)
- I/O port: 31
- Serial interface: 1 channel
- 3-wire serial I/O mode/UART mode: 1 channel
- 8-bit resolution A/D converter: 8 channels ( $\mu$ PD789167 Subseries)
- 10-bit resolution A/D converter: 8 channels ( $\mu$ PD789177 Subseries)
- Timer: 6 channels
- 16-bit timer: 1 channel
- 8-bit timer/event counter: 2 channels
- 8-bit timer: 1 channel
- Watch timer: 1 channel
- Watchdog timer: 1 channel
- Vectored interrupt sources: 15
- Power supply voltage
- $V_{D D}=1.8$ to 5.5 V ( $\mu$ PD78916x, 78917x, 78916x(A), 78917x(A), 78F9177A, 78F9177A(A))
- $V_{D D}=4.5$ to 5.5 V ( $\left.\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1), 78916 x(\mathrm{~A} 2), 78917 x(\mathrm{~A} 2)\right)$
- Operating ambient temperature
- $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}, 78917 \mathrm{x}, 78916 \mathrm{x}(\mathrm{A}), 78917 \mathrm{x}(\mathrm{A}), 78 \mathrm{~F} 9177 \mathrm{~A}, 78 \mathrm{~F} 9177 \mathrm{~A}(\mathrm{~A}))$
- $\mathrm{T}_{\mathrm{A}}=-40$ to $110^{\circ} \mathrm{C}$ ( $\left.\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 \times(\mathrm{A} 1), 789177 \mathrm{~A}(\mathrm{~A} 1)\right)$
- $\mathrm{T}_{\mathrm{A}}=-40$ to $125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 2), 78917 \mathrm{x}(\mathrm{A} 2))$

Note When VDD $=4.5$ to 5.5 V and the product is an expanded-specification product

### 1.3 Applications

Power windows, keyless entry, battery management units, side air bags, etc.

### 1.4 Ordering Information

| Part Number | Package | Internal ROM |
| :---: | :---: | :---: |
| $\mu$ PD789166GB-×xx-8ES | 44-pin plastic LQFP (10 $\times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{GA}-\times \times \times-9 \mathrm{E}$ | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789167 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789167 \mathrm{GA}-\times \times \times-9 \mathrm{EU}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{GA}-\times \times \times-9 \mathrm{EU}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{GA}-\times \times \times-9 \mathrm{E}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789166GB-×x×-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789166GA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789167GB-×x×-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789167 \mathrm{GA}-\times \times \times-9 \mathrm{EU}-\mathrm{A}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789176GB-×xx-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789176GA-×××-9EU-A | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789177GB-×xx-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789177GA-×××-9EU-A | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A})-\times \times \times-8 \mathrm{SS}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789167GB(A)-×xx-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{~GB}(\mathrm{~A})-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789177GB(A)-×xx-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789167 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times$-8ES | 44 -pin plastic LQFP $(10 \times 10)$ | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789167 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD78F9177GB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AGB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AGA-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu$ PD78F9177GB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu \mathrm{PD} 78 \mathrm{F9177AGB}(\mathrm{~A})$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu \mathrm{PD} 78 \mathrm{F9177AGB}(\mathrm{~A} 1)-8 \mathrm{ES}$ | 44 -pin plastic LQFP (10 $\times 10$ ) | Flash memory |

Remarks 1. $\times x \times$ indicates ROM code suffix.
2. Products with additional order code "-A" are lead-free products.

### 1.5 Quality Grades

|  | Part Number | Package | Quality Grade |
| :---: | :---: | :---: | :---: |
|  | $\mu \mathrm{PD} 789166 \mathrm{~GB}-\times \times \times$-8ES | 44-pin plastic LQFP (10 $\times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789166 \mathrm{GA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789167 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789167 \mathrm{GA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789176 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789176 \mathrm{GA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789177 \mathrm{~GB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789177 \mathrm{GA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu \mathrm{PD} 789166 \mathrm{~GB}-\times \times \times-8 \mathrm{~S}-\mathrm{A}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789166GA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789167GB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789167GA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789176GB-×xx-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789176GA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| * | $\mu$ PD789177GB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789177GA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A})-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789167 \mathrm{~GB}(\mathrm{~A})-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789176 \mathrm{~GB}(\mathrm{~A})-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789177 \mathrm{~GB}(\mathrm{~A})-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP (10 $\times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789167 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789176 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP (10 $\times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789177 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{ES}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789167 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times-8 \mathrm{~S}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu$ PD789176GB(A2)-×xx-8ES | 44 -pin plastic LQFP (10 $\times 10$ ) | Special |
|  | $\mu \mathrm{PD} 789177 \mathrm{~GB}(\mathrm{~A} 2)-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu$ PD78F9177GB-8ES | 44 -pin plastic LQFP (10 $\times 10$ ) | Standard |
|  | $\mu$ PD78F9177AGB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu$ PD78F9177AGA-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD78F9177GB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD78F9177AGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD78F9177AGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu$ PD78F9177AGB(A)-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu \mathrm{PD} 78 \mathrm{F9177AGB}(\mathrm{~A} 1)-8 \mathrm{ES}$ | 44 -pin plastic LQFP (10 $\times 10$ ) | Special |

Remarks 1. $X X X$ indicates ROM code suffix.
$\star$
2. Products with additional order code "-A" are lead-free products.

Please refer to Quality Grades on NEC Semiconductor Devices (C11531E) published by NEC Electronics Corporation to know the specification of the quality grade on the devices and its recommended applications.

### 1.6 Pin Configuration (Top View)

- 44-pin plastic LQFP ( $10 \times 10$ )
$\mu$ PD789166GB-×xx-8ES $\quad \mu$ PD789166GB(A)-×××-8ES $\quad \mu$ PD789166GB(A2)-×××-8ES
$\mu$ PD789167GB-×××-8ES
$\mu$ PD789176GB-×××-8ES
$\mu$ PD789177GB-×xx-8ES
$\mu$ PD789166GB-×××-8ES-A
$\mu$ PD789167GB-×××-8ES-A
$\mu$ PD789176GB- $\times x \times-8 E S-A$
$\mu$ PD789177GB-×××-8ES-A
$\mu$ PD789167GB(A)-×Xx-8ES
$\mu$ PD789176GB(A)-XXX-8ES
$\mu$ PD789177GB(A)-XXX-8ES
$\mu$ PD789166GB(A1)-×××-8ES
$\mu$ PD789167GB(A1)-×××-8ES
$\mu$ PD789176GB(A1)-×××-8ES
$\mu$ PD789177GB(A1)-×××-8ES
$\mu$ PD789167GB(A2)-×××-8ES $\mu$ PD789176GB(A2)-××x-8ES $\mu$ PD789177GB(A2)-×××-8ES $\mu$ PD78F9177GB-8ES $\mu$ PD78F9177AGB-8ES $\mu$ PD78F9177GB-8ES-A $\mu$ PD78F9177AGB-8ES-A $\mu$ PD78F9177AGB(A)-8ES
$\mu$ PD78F9177AGB(A1)-8ES


Cautions 1. Connect the ICO (internally connected) pin directly to the Vsso or Vss1 pin.
2. Connect the $A V d d$ pin to the Vddo pin.
3. Connect the AVss pin to the Vsso pin.

Remark Pin connections in parentheses are intended for the $\mu$ PD78F9177, 78F9177A, 78F9177A(A), and 78F9177A(A1).

- 48-pin plastic TQFP (fine pitch) $(7 \times 7)$
$\mu$ PD789166GA-×XX-9EU $\quad \mu$ PD789166GA-×X×-9EU-A $\mu$ PD78F9177AGA-9EU
$\mu$ PD789167GA-×××-9EU $\quad \mu$ PD789167GA-×x×-9EU-A $\mu$ PD78F9177AGA-9EU-A
$\mu$ PD789176GA-×××-9EU $\quad \mu$ PD789176GA-×××-9EU-A
$\mu$ PD789177GA-××x-9EU $\quad \mu$ PD789177GA- $\times x \times-9 E U-A$


Cautions 1. Connect the IC0 (internally connected) pin directly to the Vsso or Vss1 pin.
2. Leave the IC3 pin open.
3. Connect the AVdd pin to the Vddo pin.
4. Connect the AVss pin to the Vsso pin.
5. The pin configuration of the 48-pin package for (A), (A1), and (A2) products is undefined.

Remark Pin connections in parentheses are intended for the $\mu$ PD78F9177A.

| ANI0 to ANI7: | Analog input | RESET: | Reset |
| :--- | :--- | :--- | :--- |
| ASCK20: | Asynchronous serial input | RxD20: | Receive data |
| AVDD: | Analog power supply | $\overline{S C K 20:}$ | Serial clock |
| AV | AneF: | Analog reference voltage | SI20: |

## * 1.7 78K/0S Series Lineup

The 78K/0S Series products are shown below. The subseries names are indicated in frames.


Remark VFD (Vacuum Fluorescent Display) is referred to as FIP ${ }^{\text {TM }}$ (Fluorescent Indicator Panel) in some documents, but the functions of the two are the same.

The major functional differences between the subseries are listed below.

Series for General-purpose applications and LCD drive

|  |  | ROM Capacity | Timer |  |  |  | $\begin{gathered} \text { 8-Bit } \\ \text { A/D } \end{gathered}$ | $\begin{gathered} 10-\mathrm{Bit} \\ \text { A/D } \end{gathered}$ | Serial <br> Interface | I/O | VDD | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 8-Bit | 16-Bit | Watch | WDT | MIN. <br> Value |  |  |  |  |  |
| Small-scale package, generalpurpose applications | $\mu$ PD789046 |  | 16 KB | 1 ch | 1 ch | 1 ch | 1 ch | - | - | 1 ch <br> (UART: 1 ch ) | 34 | 1.8 V | - |
|  | $\mu$ PD789026 | 4 KB to 16 KB | - |  |  |  |  |  |  |  |  |  |
|  | $\mu$ PD789088 | $\begin{aligned} & 16 \mathrm{~KB} \text { to } \\ & 32 \mathrm{~KB} \end{aligned}$ |  | 3 ch |  | 24 |  |  |  |  |  |  |  |
|  | $\mu$ PD789074 | 2 KB to 8 KB |  | 1 ch |  |  |  |  |  |  |  |  |  |
|  | $\mu$ PD789062 | 4 KB |  | 2 ch | - | - |  |  |  | 14 | RC oscillation version |  |  |
|  | $\mu$ PD789052 |  |  |  |  |  |  |  |  |  | - |  |  |
| Small-scale package, generalpurpose applications and $A / D$ converter | $\mu$ PD789177 | 16 KB to 24 KB | 3 ch | 1 ch | 1 ch | 1 ch | - | 8 ch | 1 ch <br> (UART: 1 ch ) | 31 | 1.8 V | - |  |
|  | $\mu$ PD789167 |  |  |  |  |  | 8 ch | - |  |  |  |  |  |
|  | $\mu$ PD789134A | 2 KB to 8 KB | 1 ch |  | - |  | - | 4 ch |  | 20 |  | RC oscillation version |  |
|  | $\mu \mathrm{PD} 789124 \mathrm{~A}$ |  |  |  |  |  | 4 ch | - |  |  |  |  |  |
|  | $\mu \mathrm{PD} 789114 \mathrm{~A}$ |  |  |  |  |  | - | 4 ch |  |  |  | - |  |
|  | $\mu$ PD789104A |  |  |  |  |  | 4 ch | - |  |  |  |  |  |
| LCD drive | $\mu$ PD789835B | 24 KB to 60 KB | 6 ch | - | 1 ch | 1 ch | 3 ch | - | 1 ch <br> (UART: 1 ch ) | 37 | $1.8 \mathrm{~V}^{\text {Note }}$ | Dot LCD supported |  |
|  | $\mu$ PD789830 | 24 KB | 1 ch | 1 ch |  |  | - |  |  | 30 | 2.7 V |  |  |
|  | $\mu$ PD789489 | $\begin{aligned} & 32 \mathrm{~KB} \text { to } \\ & 48 \mathrm{~KB} \end{aligned}$ | 3 ch |  |  |  |  | 8 ch | 2 ch <br> (UART: 1 ch ) | 45 | 1.8 V | - |  |
|  | $\mu$ PD789479 | 24 KB to 48 KB 48 KB |  |  |  |  | 8 ch | - |  |  |  |  |  |
|  | $\mu$ PD789417A | 12 KB to |  |  |  |  | - | 7 ch | $1 \text { ch }$ | 43 |  |  |  |
|  | $\mu \mathrm{PD} 789407 \mathrm{~A}$ | 24 KB |  |  |  |  | 7 ch | - | (UART: 1 ch ) |  |  |  |  |
|  | $\mu$ PD789456 | $\begin{aligned} & 12 \mathrm{~KB} \text { to } \\ & 16 \mathrm{~KB} \end{aligned}$ | 2 ch |  |  |  | - | 6 ch |  | 30 |  |  |  |
|  | $\mu \mathrm{PD} 789446$ |  |  |  |  |  | 6 ch | - |  |  |  |  |  |
|  | $\mu$ PD789436 |  |  |  |  |  | - | 6 ch |  | 40 |  |  |  |
|  | $\mu$ PD789426 |  |  |  |  |  | 6 ch | - |  |  |  |  |  |
|  | $\mu$ PD789316 | 8 KB to 16 KB |  |  |  |  | - |  | 2 ch <br> (UART: 1 ch$)$ | 23 |  | RC oscillation version |  |
|  | $\mu$ PD789306 |  |  |  |  |  |  |  |  |  |  | - |  |
|  | $\mu$ PD789467 | 4 KB to 24 KB |  | - |  |  | 1 ch |  | - | 18 |  |  |  |
|  | $\mu$ PD789327 |  |  |  |  |  | - |  | 1 ch | 21 |  |  |  |

Note Flash memory version: 3.0 V

Series for ASSP

|  |  | ROM Capacity | Timer |  |  |  | $\begin{gathered} \text { 8-Bit } \\ \text { A/D } \end{gathered}$ | $\left\|\begin{array}{c} 10-\mathrm{Bit} \\ \text { A/D } \end{array}\right\|$ | Serial Interface | I/O | VDD | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | 8-Bit | 16-Bit | Watch | WDT | MIN. <br> Value |  |  |  |  |  |
| USB | $\mu$ PD789800 |  | 8 KB | 2 ch | - | - | 1 ch | - | - | 2 ch <br> (USB: 1 ch ) | 31 | 4.0 V | - |
| Inverter control | $\mu$ PD789842 | 8 KB to 16 KB | 3 ch | Note 1 | 1 ch | 1 ch | 8 ch | - | 1 ch <br> (UART: 1 ch$)$ | 30 | 4.0 V | - |
| On-chip bus controller | $\mu$ PD789852 | 24 KB to 32 KB | 3 ch | 1 ch | - | 1 ch | - | 8 ch | 3 ch <br> (UART: 2 ch ) | 31 | 4.0 V | - |
|  | $\mu$ PD789850A | 16 KB | 1 ch |  |  |  | 4 ch | - | 2 ch <br> (UART: 1 ch) | 18 |  |  |
| Keyless entry | $\mu$ PD789861 | 4 KB | 2 ch | - | - | 1 ch | - | - | - | 14 | 1.8 V | RC oscillation version, onchip EEPROM |
|  | $\mu$ PD789860 |  |  |  |  |  |  |  |  |  |  | On-chip EEPROM |
|  | $\mu$ PD789862 | 16 KB | 1 ch | 2 ch |  |  |  |  | 1 ch <br> (UART: 1 ch$)$ | 22 |  |  |
| Sensor | $\mu$ PD789864 | 4 KB | 1 ch | Note 2 | - | 1 ch | - | 4 ch | - | 5 | 1.9 V | On-chip EEPROM |
|  | $\mu \mathrm{PD} 789863$ |  |  |  |  |  |  |  |  |  |  | RC oscillation version, onchip EEPROM |
| VFD drive | $\mu$ PD789871 | 4 KB to 8 KB | 3 ch | - | 1 ch | 1 ch | - | - | 1 ch | 33 | 2.7 V | - |
| Meter control | $\mu$ PD789881 | 16 KB | 2 ch | 1 ch | - | 1 ch | - | - | 1 ch <br> (UART: 1 ch$)$ | 28 | $2.7 \mathrm{~V}^{\text {Note } 3}$ | - |

Notes 1. 10-bit timer: 1 channel
2. 12-bit timer: 1 channel
3. Flash memory version: 3.0 V

### 1.8 Block Diagram



Remarks 1. The size of the internal ROM varies depending on the product.
2. Pin connections in parentheses are intended for the $\mu$ PD78F9177, 78F9177A, 78F9177A(A), and 78F9177A(A1).

### 1.9 Outline of Functions

| Part Number <br> Item |  | $\begin{gathered} \mu \text { PD789166, 789176, } \\ 789166(\mathrm{~A}), 789176(\mathrm{~A}), \\ 789166(\mathrm{~A} 1), 789176(\mathrm{~A} 1), \\ 789166(\mathrm{~A} 2), 789176(\mathrm{~A} 2) \end{gathered}$ | $\begin{gathered} \mu \mathrm{PD} 789167,789177, \\ 789167(\mathrm{~A}), 789177(\mathrm{~A}), \\ 789167(\mathrm{~A} 1), 789177(\mathrm{~A} 1), \\ 789167(\mathrm{~A} 2), 789177(\mathrm{~A} 2) \end{gathered}$ | $\mu$ PD78F9177, <br> 78F9177A, <br> 78F9177A(A), <br> 78F9177A(A1) |
| :---: | :---: | :---: | :---: | :---: |
| Internal memory | ROM | Mask ROM |  | Flash memory |
|  |  | 16 KB | 24 KB | 24 KB |
|  | High-speed RAM | 512 bytes |  |  |
| Minimum instruction execution time |  | Expanded-specification products of $\mu \mathrm{PD} 78916 \mathrm{x}, 78917 \mathrm{x}, 78916 \mathrm{x}(\mathrm{A})$, 78917x(A), 78F9177A, 78F9177A(A) <br> $0.2 / 0.8 \mu \mathrm{~s}$ (operation with main system clock operating at $10.0 \mathrm{MHz}, \mathrm{VDD}=4.5$ to 5.5 <br> V) <br> - $122 \mu$ (operation with subsystem clock operating at 32.768 kHz ) <br> Other than above products <br> - $0.4 / 1.6 \mu \mathrm{~s}$ (operation with main system clock operating at 5.0 MHz ) <br> - $122 \mu$ s (operation with subsystem clock operating at 32.768 kHz ) |  |  |
| General-purpose registers |  | 8 bits $\times 8$ registers |  |  |
| Instruction set |  | - 16-bit operations <br> - Bit manipulations (such as set, reset, and test) |  |  |
| Multiplier |  | 8 bits $\times 8$ bits $=16$ bits |  |  |
| I/O ports |  | Total: 31 <br> - CMOS input: 8 <br> - CMOS I/O: 17 <br> - N-ch open-drain: 6 |  |  |
| A/D converter |  | - 8 -bit resolution $\times 8$ channels ( $\mu$ PD789167 Subseries) <br> - 10 -bit resolution $\times 8$ channels ( $\mu$ PD789177 Subseries) |  |  |
| Serial interface |  | - Switchable between 3-wire serial I/O and UART modes: 1 channel |  |  |
| Timers |  | - 16 -bit timer: 1 channel <br> - 8 -bit timer/event counter: 2 channels <br> - 8 -bit timer: 1 channel <br> - Watch timer: 1 channel <br> - Watchdog timer: 1 channel |  |  |
| Timer output |  | Four outputs |  |  |
| Buzzer output |  | One output |  |  |
| Vectored interrupt sources | Maskable | Internal: 10, external: 4 |  |  |
|  | Non-maskable | Internal: 1 |  |  |
| Power supply voltage |  | $\begin{aligned} & V_{\mathrm{DD}}=1.8 \text { to } 5.5 \mathrm{~V}(\mu \mathrm{PD} 78916 \mathrm{x}, 78917 \mathrm{x}, 78916 \mathrm{x}(\mathrm{~A}), 78917 \mathrm{x}(\mathrm{~A}), 78 \mathrm{F9177}, \\ & 78 \mathrm{~F} 9177 \mathrm{~A}, 78 \mathrm{F9177A}(\mathrm{~A}) \\ & \mathrm{VDD}=4.5 \text { to } 5.5 \mathrm{~V} \begin{array}{l} (\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A} 1), 78917 \times(\mathrm{A} 1), 78916 \times(\mathrm{A} 2), 78917 \times(\mathrm{A} 2), \\ \\ \\ 78 \mathrm{~F} 9177 \mathrm{~A}(\mathrm{~A} 1) \end{array}, \end{aligned}$ |  |  |
| Operating ambient temperature |  | $\begin{array}{ll} \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to }+85^{\circ} \mathrm{C} & (\mu \mathrm{PD} 78916 \mathrm{x}, 78917 \mathrm{x}, 78916 \mathrm{x}(\mathrm{~A}), 78917 \times(\mathrm{A}), 78 \mathrm{F9177}, \\ & 78 \mathrm{~F} 9177 \mathrm{~A}, 78 \mathrm{~F} 9177 \mathrm{~A}(\mathrm{~A}) \\ \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to }+110^{\circ} \mathrm{C} & (\mu \mathrm{PD} 78916 \times(\mathrm{A} 1), 78917 \times(\mathrm{A} 1), 78 \mathrm{F9177A}(\mathrm{~A} 1) \\ \mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C} \text { to }+125^{\circ} \mathrm{C} & (\mu \mathrm{PD} 78916 \times(\mathrm{A} 2), 78917 \times(\mathrm{A} 2) \end{array}$ |  |  |
| Package |  | - 44-pin plastic LQFP $(10 \times 10)$ <br> - 48 -pin plastic TQFP (fine pitch) $(7 \times 7)^{\text {Note }}$ |  |  |

Note $\mu$ PD789166, 789167, 789176, 789177, and 78F9177A only

The timers are outlined below.

|  |  | $\begin{gathered} 16-\mathrm{Bit} \\ \text { Timer } 90 \end{gathered}$ | $\begin{gathered} \text { 8-Bit } \\ \text { Timer/Event } \\ \text { Counter } 80 \end{gathered}$ | $\begin{gathered} \text { 8-Bit } \\ \text { Timer/Event } \\ \text { Counter } 81 \end{gathered}$ | 8-Bit Timer 82 | Watch Timer | Watchdog Timer |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating <br> mode | Interval timer | - | 1 channel | 1 channel | 1 channel | 1 channel $^{\text {Note } 1}$ | 1 channel $^{\text {Note } 2}$ |
|  | External event counter | - | 1 channel | 1 channel | - | - | - |
| Function | Timer output | 1 output | 1 output | 1 output | 1 output | - | - |
|  | PWM output | - | 1 output | 1 output | 1 output | - | - |
|  | Square-wave output | - | 1 output | 1 output | 1 output | - | - |
|  | Buzzer output | 1 output | - | - | - | - | - |
|  | Capture | 1 input | - | - | - | - | - |
|  | Interrupt sources | 1 | 1 | 1 | 1 | 2 | 2 |

Notes 1. The watch timer can perform both watch timer and interval timer functions at the same time.
2. The watchdog timer provides a watchdog timer function and an interval timer function. Use either of the functions.

### 1.10 Differences Between Standard Quality Grade Products and (A) Products, (A1) Products, and (A2) Products

Standard quality grade products, (A) products, (A1) products, and (A2) products indicate the following products respectively.

Standard quality grade products... $\mu$ PD789166, 789167, 789176, 789177, 78F9177, 78F9177A
(A) products... $\quad \mu \mathrm{PD} 789166(\mathrm{~A}), 789167(\mathrm{~A}), 789176(\mathrm{~A}), 789177(\mathrm{~A}), 78 F 9177 \mathrm{~A}(\mathrm{~A})$
(A1) products... $\quad \mu$ PD789166(A1), 789167(A1), 789176(A1), 789177(A1), 78F9177A(A1)
(A2) products... $\mu$ PD789166(A2), 789167(A2), 789176(A2), 789177(A2)

Table 1-2 shows the differences between the standard quality grade products and (A) products, (A1) products, and (A2) products.

Table 1-2. Differences Between Standard Quality Grade Products and (A) Products, (A1) Products, and (A2) Products

| Part Number <br> Item | Standard Quality Grade Products | (A) Products | (A1) Products | (A2) Products |
| :---: | :---: | :---: | :---: | :---: |
| Quality grade | Standard | Special |  |  |
| Power supply voltage | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | $V_{\text {DD }}=4.5$ to 5.5 V |  |
| Operating ambient temperature | $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}$ |  | $\mathrm{T}_{\mathrm{A}}=-40$ to $110^{\circ} \mathrm{C}$ | $\mathrm{T}_{\mathrm{A}}=-40$ to $125^{\circ} \mathrm{C}$ |
| Minimum instruction execution time | Expanded-specification product ${ }^{\text {Note }}$ : <br> $0.2 \mu \mathrm{~s}$ (at 10.0 MHz operation) <br> Conventional product ${ }^{\text {Note }}$ : <br> $0.4 \mu \mathrm{~s}$ (at 5.0 MHz operation) |  | $0.4 \mu \mathrm{~s}$ (at 5.0 MHz operation) |  |
| Electrical specifications | Refer to the ELECTRICAL SPECIFICATIONS chapters. |  |  |  |

Note Refer to 1.1 Expanded-Specification Products and Conventional Products

## CHAPTER 2 GENERAL ( $\mu$ PD789167Y AND 789177Y SUBSERIES)

### 2.1 Expanded-Specification Products and Conventional Products

The expanded-specification products and conventional products refer to the following products.

Expanded-specification product... Products with a rank ${ }^{\text {Note }}$ other than K

- Mask ROM versions for which orders were received after December 1, 2001.
- $\mu$ PD78F9177AY, 78F9177AY(A)

Conventional product... Products with rank ${ }^{\text {Note }} \mathrm{K}$

- Products other than the above expanded-specification products.

Note The rank is indicated by the 5th digit from the left in the lot number marked on the package.


Expanded-specification products and conventional products differ in operating frequency ratings. The differences are shown in Table 2-1.

Table 2-1. Differences Between Expanded-Specification Products and Conventional Products

| Power Supply Voltage (VDD) | Guaranteed Operating Speed (Operating Frequency) |  |
| :--- | :--- | :--- |
|  | Conventional Products | Expanded-Specification Products |
| 4.5 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $10 \mathrm{MHz}(0.2 \mu \mathrm{~s})$ |
| 3.0 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $6 \mathrm{MHz}(0.33 \mu \mathrm{~s})$ |
| 2.7 to 5.5 V | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ | $5 \mathrm{MHz}(0.4 \mu \mathrm{~s})$ |
| 1.8 to 5.5 V | $1.25 \mathrm{MHz}(1.6 \mu \mathrm{~s})$ | $1.25 \mathrm{MHz}(1.6 \mu \mathrm{~s})$ |

Remark The values in parentheses indicate the minimum instruction execution time.

### 2.2 Features

- ROM and RAM capacity

| Product Name | Program Memory <br> (ROM) |  | Data Memory (Internal High-Speed RAM) |
| :---: | :---: | :---: | :---: |
| $\mu \mathrm{PD} 789166 \mathrm{Y}, 789176 \mathrm{Y}, 789166 \mathrm{Y}(\mathrm{A}), 789176 \mathrm{Y}(\mathrm{A})$ | Mask ROM | 16 KB | 512 bytes |
| $\mu \mathrm{PD} 789167 \mathrm{Y}, 789177 \mathrm{Y}, 789167 \mathrm{Y}(\mathrm{A}), 789177 \mathrm{Y}(\mathrm{A})$ |  | 24 KB |  |
| $\mu \mathrm{PD} 78 \mathrm{F9177Y}$, 78F9177AY, 78F9177AY(A) | Flash memory | 24 KB |  |

- Minimum instruction execution time changeable from high-speed ( $0.2 \mu \mathrm{~s}$ : Main system clock 10.0 MHz operation ${ }^{\text {Note }}$ ) to ultra-low speed ( $122 \mu \mathrm{~s}$ : Subsystem clock 32.768 kHz operation)
- I/O port: 31
- Serial interface: 2 channels
- 3-wire serial I/O mode/UART mode: 1 channel
- SMB: 1 channel
- 8-bit resolution A/D converter: 8 channels ( $\mu$ PD789167Y Subseries)
- 10-bit resolution A/D converter: 8 channels ( $\mu$ PD789177Y Subseries)
- Timer: 6 channels
- 16-bit timer: 1 channel
- 8-bit timer/event counter: 2 channels
- 8-bit timer: 1 channel
- Watch timer: 1 channel
- Watchdog timer: 1 channel
- Vectored interrupt sources: 17
- Supply voltage: VDD $=1.8$ to 5.5 V
- Operating ambient temperature: $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$

Note $\quad$ When $V_{D D}=4.5$ to 5.5 V and the product is an expanded-specification product.

### 2.3 Applications

Power windows, keyless entry, battery management units, side air bags, etc.

### 2.4 Ordering Information

| Part Number | Package | Internal ROM |
| :---: | :---: | :---: |
| $\mu \mathrm{PD} 789166 \mathrm{YGB}-\times x \times-8 \mathrm{ES}$ | 44-pin plastic LQFP (10 $\times 10$ ) | Mask ROM |
| $\mu$ PD789166YGA-×Xx-9EU | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789167YGB-×××-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789167YGA-×X×-9EU | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{YGB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{YGA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789177YGB-×xx-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{YGA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{YGB}-\times \times \times-8 \mathrm{ES}-\mathrm{A}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789166YGA-××x-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789167YGB- $\times \times \times$-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789167YGA- $\times \times \times$-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{YGB}-\times \times \times-8 \mathrm{ES}-\mathrm{A}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{YGA}-\times \times \times-9 E U-A$ | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789177 \mathrm{YGB}-\times \times \times-8 \mathrm{ES}-\mathrm{A}$ | 44 -pin plastic LQFP ( $10 \times 10$ ) | Mask ROM |
| $\mu$ PD789177YGA-×Xx-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789166 \mathrm{YGA}(\mathrm{A})-\times \times \times-9 \mathrm{U}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789167YGA(A)-×× - 9 EU | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu \mathrm{PD} 789176 \mathrm{YGA}(\mathrm{A})-\times \times \times-9 \mathrm{U}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD789177YGA(A)-×× - 9 -U | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Mask ROM |
| $\mu$ PD78F9177YGB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177YGA-9EU | 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu$ PD78F9177AYGB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AYGA-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu$ PD78F9177YGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177YGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu$ PD78F9177AYGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu$ PD78F9177AYGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |
| $\mu \mathrm{PD} 78 \mathrm{F9177AYGB}(\mathrm{~A})$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Flash memory |
| $\mu \mathrm{PD} 78 \mathrm{F9177AYGA}(\mathrm{~A})-9 \mathrm{EU}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Flash memory |

Remark $x \times x$ indicates ROM code suffix.

### 2.5 Quality Grades

|  | Part Number | Package | Quality Grade |
| :---: | :---: | :---: | :---: |
|  | $\mu \mathrm{PD} 789166 \mathrm{YGB}-\times \times \times$-8ES | 44-pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789166 \mathrm{YGA}-\times \times \times-9 \mathrm{U}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789167 \mathrm{YGB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789167 \mathrm{YGA}-\times \times \times-9 \mathrm{U}$ | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789176 \mathrm{YGB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789176 \mathrm{YGA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 789177 \mathrm{YGB}-\times \times \times$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu \mathrm{PD} 789177 \mathrm{YGA}-\times \times \times$-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789166YGB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789166YGA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789167YGB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789167YGA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789176YGB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD789176YGA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD789177YGB-×××-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu \mathrm{PD} 789177 \mathrm{YGA}-\times \times \times$-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu$ PD789166YGA(A)-×xx-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Special |
|  | $\mu$ PD789167YGA(A)-×××-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Special |
|  | $\mu$ PD789176YGA(A)-×××-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Special |
|  | $\mu$ PD789177YGA(A)-×××-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Special |
|  | $\mu$ PD78F9177YGB-8ES | 44 -pin plastic LQFP (10 $\times 10$ ) | Standard |
|  | $\mu$ PD78F9177YGA-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu$ PD78F9177AYGB-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
|  | $\mu$ PD78F9177AYGA-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD78F9177YGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD78F9177YGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
| $\star$ | $\mu$ PD78F9177AYGB-8ES-A | 44 -pin plastic LQFP ( $10 \times 10$ ) | Standard |
| $\star$ | $\mu$ PD78F9177AYGA-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Standard |
|  | $\mu \mathrm{PD} 78 \mathrm{F9177AYGB}(\mathrm{~A})$-8ES | 44 -pin plastic LQFP ( $10 \times 10$ ) | Special |
|  | $\mu$ PD78F9177AYGA(A)-9EU | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ | Special |

Remark $\times X \times$ indicates ROM code suffix.

Please refer to Quality Grades on NEC Semiconductor Devices (C11531E) published by NEC Electronics Corporation to know the specification of the quality grade on the device and its recommended applications.

### 2.6 Pin Configuration (Top View)

$\star \quad$ - 44-pin plastic LQFP ( $10 \times 10$ )
$\mu$ PD789166YGB- $x \times x-8 E S \quad \mu$ PD789166YGB- $-\times x$ - 8 ES $-A \quad \mu$ PD78F9177YGB-8ES $\mu$ PD789167YGB-×××-8ES
$\mu$ PD789176YGB $-\times x \times-8 E S$
$\mu$ PD789167YGB- $\times \times \times-8 E S-A \quad \mu$ PD78F9177AYGB-8ES
$\mu$ PD789176YGB $-\times \times \times-8 E S-A \quad \mu$ PD78F9177YGB-8ES-A
$\mu$ PD789177YGB $-x \times x-8 E S \quad \mu$ PD789177YGB $-x \times x-8 E S-A \quad \mu$ PD78F9177AYGB-8ES-A
$\mu$ PD78F9177AYGB(A)-8ES


Cautions 1. Connect the IC0 (internally connected) pin directly to the Vsso or Vss1 pin.
2. Connect the AVdd pin to the Vddo pin.
3. Connect the AVss pin to the Vsso pin.

Remark Pin connections in parentheses are intended for the $\mu$ PD78F9177Y, 78F9177AY, and 78F9177AY(A).
$\star \quad$ - 48-pin plastic TQFP (fine pitch) $(7 \times 7)$
$\mu$ PD789166YGA-×××-9EU $\quad \mu$ PD789166YGA(A)-×××-9EU $\quad \mu$ PD78F9177YGA-9EU
$\mu$ PD789167YGA-×××-9EU $\quad \mu$ PD789167YGA(A)-×X×-9EU $\quad \mu$ PD78F9177AYGA-9EU
$\mu$ PD789176YGA-×××-9EU $\quad \mu$ PD789176YGA(A)-×XX-9EU $\quad \mu$ PD78F9177YGA-9EU-A
$\mu$ PD789177YGA-xxx-9EU $\quad \mu$ PD789177YGA(A)-xxx-9EU $\quad \mu$ PD78F9177AYGA-9EU-A
$\mu$ PD789166YGA-×××-9EU-A
$\mu$ PD78F9177AYGA(A)-9EU
$\mu$ PD789167YGA-×××-9EU-A
$\mu$ PD789176YGA-×××-9EU-A
$\mu$ PD789177YGA-××x-9EU-A


Cautions 1. Connect the ICO (internally connected) pin directly to the Vsso or Vss1 pin.
2. Leave the IC2 pin open.
3. Connect the AVdd pin to the Vddo pin.
4. Connect the AVss pin to the Vsso pin.

Remark Pin connections in parentheses are intended for the $\mu$ PD78F9177Y, 78F9177AY, and 78F9177AY(A).

| ANI0 to ANI7: | Analog input |
| :--- | :--- |
| ASCK20: | Asynchronous serial input |
| AVDD: | Analog power supply |
| AV REF: | Analog reference voltage |
| AVss: | Analog ground |
| BZO90: | Buzzer output |
| CPT90: | Capture trigger input |
| IC0, IC2: | Internally connected |
| INTP0 to INTP3: | Interrupt from peripherals |
| P00 to P05: | Port 0 |
| P10, P11: | Port 1 |
| P20 to P26: | Port 2 |
| P30 to P33: | Port 3 |
| P50 to P53: | Port 5 |
| P60 to P67: | Port 6 |


| RESET: | Reset |
| :--- | :--- |
| RxD20: | Receive data |
| SCK20: | Serial clock (for SIO20) |
| SCL0: | Serial clock (for SMB0) |
| SDA0: | Serial data |
| SI20: | Serial input |
| SO20: | Serial output |
| SS20: | Chip select input |
| TI80, TI81: | Timer input |
| TO80 to TO82, TO90: | Timer output |
| TxD20: | Transmit data |
| VDD0, VDD1: | Power supply |
| VPP: | Programming power supply |
| Vsso, Vss1: | Ground |
| X1, X2: | Crystal (main system clock) |
| XT1, XT2: | Crystal (subsystem clock) |

## * 2.7 78K/OS Series Lineup

The $78 \mathrm{~K} / 0 \mathrm{~S}$ Series products are shown below. The subseries names are indicated in frames.


Remark VFD (Vacuum Fluorescent Display) is referred to as FIP ${ }^{\text {TM }}$ (Fluorescent Indicator Panel) in some documents, but the functions of the two are the same.

The functions of the Y Subseries are listed below.
$\left.\begin{array}{|l|l|l|l|l|l|c|}\hline & \text { Function } & \text { ROM Capacity } & \text { Serial Interface Configuration } & \begin{array}{c}\text { I/O } \\ \text { (pins) }\end{array} & \begin{array}{c}\text { Vod } \\ \text { Subseries Name }\end{array} & \text { Ralue }\end{array}\right]$

### 2.8 Block Diagram



Remarks 1. The size of the internal ROM varies depending on the model.
2. Pin connections in parentheses are intended for the $\mu$ PD78F9177Y, 78F9177AY, and 78F9177AY(A).

### 2.9 Outline of Function

| Part Number <br> Item |  | $\begin{gathered} \mu \mathrm{PD} 789166 \mathrm{Y}, 789176 \mathrm{Y} \\ 789166 \mathrm{Y}(\mathrm{~A}), 789176 \mathrm{Y}(\mathrm{~A}) \end{gathered}$ | $\begin{gathered} \mu \mathrm{PD} 789167 \mathrm{Y}, 789177 \mathrm{Y} \\ 789167 \mathrm{Y}(\mathrm{~A}), 789177 \mathrm{Y}(\mathrm{~A}) \end{gathered}$ | $\mu$ PD78F9177Y, 78F9177AY 78F9177AY(A) |
| :---: | :---: | :---: | :---: | :---: |
| Internal memory | ROM | Mask ROM |  | Flash Memory |
|  |  | 16 KB | 24 KB | 24 KB |
|  | High-speed RAM | 512 bytes |  |  |
| Minimum instruction execution time |  | Expanded-specification products of $\mu \mathrm{PD} 78916 \mathrm{xY}, 78917 \mathrm{xY}, 78916 \mathrm{xY}(\mathrm{A}), 78917 \mathrm{xY}(\mathrm{A})$, 78F9177AY, 78F9177AY(A) <br> - $0.2 / 0.8 \mu \mathrm{~s}$ (operation with main system clock operating at 10.0 MHz , $\left.V_{D D}=4.5 \text { to } 5.5 \mathrm{~V}\right)$ <br> - $122 \mu \mathrm{~s}$ (operation with subsystem clock operating at 32.768 kHz ) Other than above products <br> - $0.4 / 1.6 \mu \mathrm{~s}$ (operation with main system clock operating at 5.0 MHz ) <br> - $122 \mu \mathrm{~s}$ (operation with subsystem clock operating at 32.768 kHz ) |  |  |
| General-purpose registers |  | 8 bits $\times 8$ registers |  |  |
| Instruction set |  | - 16-bit operations <br> - Bit manipulations (such as set, reset, and test) |  |  |
| Multiplier |  | 8 bits $\times 8$ bits $=16$ bits |  |  |
| I/O ports |  | Total: 31 <br> - CMOS input: 8 <br> - CMOS I/O: 17 <br> - N-ch open-drain: 6 |  |  |
| A/D converter |  | - 8 -bit resolution $\times 8$ channels ( $\mu$ PD789167Y Subseries) <br> - 10 -bit resolution $\times 8$ channels ( $\mu$ PD789177Y Subseries) |  |  |
| Serial interface |  | - Switchable between 3 -wire serial I/O and UART modes: 1 channel <br> - SMB (System Management Bus): 1 channel |  |  |
| Timers |  | - 16 -bit timer: 1 channel <br> - 8 -bit timer/event counter: 2 channels <br> - 8 -bit timer: 1 channel <br> - Watch timer: 1 channel <br> - Watchdog timer: 1 channel |  |  |
| Timer output |  | Four outputs |  |  |
| Buzzer output |  | One output |  |  |
| Vectored interrupt sources | Maskable | Internal: 12, external: 4 |  |  |
|  | Nonmaskable | Internal: 1 |  |  |
| Power supply voltage |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |  |
| Operating ambient temperature |  | $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$ |  |  |
| Package |  | - 44-pin plastic LQFP $(10 \times 10)^{\text {Note }}$ <br> - 48 -pin plastic TQFP (fine pitch) $(7 \times 7)$ |  |  |

Note $\mu$ PD789166Y, 789167Y, 789176Y, 789177Y, 78F9177Y, 78F9177AY, and 78F9177AY(A) only

The timers are outlined below.

|  |  | $\begin{gathered} \text { 16-Bit } \\ \text { Timer } 90 \end{gathered}$ | 8-Bit <br> Timer/Event Counter 80 | 8-Bit <br> Timer/Event Counter 81 | 8-Bit Timer 82 | Watch Timer | Watchdog Timer |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating mode | Interval timer | - | 1 channel | 1 channel | 1 channel | 1 channel $^{\text {Note } 1}$ | 1 channel $^{\text {Note 2 }}$ |
|  | External event counter | - | 1 channel | 1 channel | - | - | - |
| Function | Timer output | 1 output | 1 output | 1 output | 1 output | - | - |
|  | PWM output | - | 1 output | 1 output | 1 output | - | - |
|  | Square-wave output | - | 1 output | 1 output | 1 output | - | - |
|  | Buzzer output | 1 output | - | - | - | - | - |
|  | Capture | 1 input | - | - | - | - | - |
|  | Interrupt sources | 1 | 1 | 1 | 1 | 2 | 2 |

Notes 1. The watch timer can perform both watch timer and interval timer functions at the same time.
2. The watchdog timer provides a watchdog timer function and an interval timer function. Use either of the functions.

### 2.10 Differences Between Standard Quality Grade Products and (A) Products

Standard quality grade products and (A) products indicate the following products.

Standard quality grade products... $\mu$ PD789166Y, 789167Y, 789176Y, 789177Y, 78F9177Y, 78F9177AY
(A) products... $\mu \mathrm{PD} 789166 \mathrm{Y}(\mathrm{A}), 789167 \mathrm{Y}(\mathrm{A}), 789176 \mathrm{Y}(\mathrm{A}), 789177 \mathrm{Y}(\mathrm{A}), 78 \mathrm{F9177AY}(\mathrm{~A})$

Table 2-2 shows the differences between the standard quality grade products and (A) products

Table 2-2. Differences Between Standard Quality Grade Products and (A) Products

| Part Number <br> Item | Standard Quality Grade Products | (A) Products |
| :---: | :---: | :---: |
| Quality grade | Standard | Special |
| Power supply voltage | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |
| Operating ambient temperature | $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}$ |  |
| Minimum instruction execution time | Expanded-specification product ${ }^{\text {Note }}:$ $0.2 \mu \mathrm{~s}$ (at 10.0 MHz operation) <br> Conventional product ${ }^{\text {Note }}:$ $0.4 \mu \mathrm{~s}$ (at 5.0 MHz operation) |  |
| Electrical specifications | Refer to the ELECTRICAL SPECIFICATIONS chapters. |  |

Note Refer to 2.1 Expanded-Specification Products and Conventional Products.

## CHAPTER 3 PIN FUNCTIONS ( $\mu$ PD789167 AND 789177 SUBSERIES)

### 3.1 Pin Function List

(1) Port pins

| Pin Name | I/O | Function | After Reset | Alternate Function |
| :---: | :---: | :---: | :---: | :---: |
| P00 to P05 | I/O | Port 0 <br> 6-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P10, P11 | I/O | Port 1 <br> 2-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P20 | I/O | Port 2 <br> 7-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> For P20 to P22, P25, and P26, an on-chip pull-up resistor can be specified by means of pull-up resistor option register B2 (PUB2). <br> Only P23 and P24 can be used as N-ch open-drain I/O port pins. | Input | SCK20/ASCK20 |
| P21 |  |  |  | SO20/TxD20 |
| P22 |  |  |  | SI20/RxD20 |
| P23 |  |  |  | - |
| P24 |  |  |  | - |
| P25 |  |  |  | TI80/SS20 |
| P26 |  |  |  | TO80 |
| P30 | 1/O | Port 3 <br> 4-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> An on-chip pull-up resistor can be specified by means of pullup resistor option register B3 (PUB3). | Input | INTP0/T181/CPT90 |
| P31 |  |  |  | INTP1/TO81 |
| P32 |  |  |  | INTP2/TO90 |
| P33 |  |  |  | INTP3/TO82/BZO90 |
| P50 to P53 | I/O | Port 5 <br> 4-bit N-ch open-drain I/O port <br> I/O mode can be specified in 1-bit units. <br> For a mask ROM version, an on-chip pull-up resistor can be specified by the mask option. | Input | - |
| P60 to P67 | Input | Port 6 <br> 8-bit input-only port | Input | ANIO to ANI7 |

## (2) Non-port pins

| Pin Name | I/O | Function | After Reset | Alternate Function |
| :---: | :---: | :---: | :---: | :---: |
| INTPO | Input | External interrupt input for which the valid edge (rising edge, falling edge, or both rising and falling edges) can be specified | Input | P30/TI81/CPT90 |
| INTP1 |  |  |  | P31/TO81 |
| INTP2 |  |  |  | P32/TO90 |
| INTP3 |  |  |  | P33/TO82/BZO90 |
| SI20 | Input | Serial data input to serial interface | Input | P22/RxD20 |
| SO20 | Output | Serial data output from serial interface | Input | P21/TxD20 |
| $\overline{\text { SCK20 }}$ | I/O | Serial clock I/O for serial interface | Input | P20/ASCK20 |
| $\overline{\text { SS20 }}$ | Input | Chip select input to serial interface | Input | P25/T180 |
| ASCK20 | Input | Serial clock input for asynchronous serial interface | Input | P20/SCK20 |
| RxD20 | Input | Serial data input for asynchronous serial interface | Input | P22/SI20 |
| TxD20 | Output | Serial data output for asynchronous serial interface | Input | P21/SO20 |
| T180 | Input | External count clock input to 8-bit timer/event counter (TM80) | Input | P25/SS20 |
| T181 | Input | External count clock input to 8-bit timer/event counter (TM81) | Input | P30/INTP0/CPT90 |
| TO80 | Output | 8-bit timer/event counter (TM80) output | Input | P26 |
| TO81 | Output | 8-bit timer/event counter (TM81) output | Input | P31/INTP1 |
| TO82 | Output | 8-bit timer (TM82) output | Input | P33/INTP3/BZO90 |
| TO90 | Output | 16-bit timer (TM90) output | Input | P32/INTP2 |
| CPT90 | Input | Capture edge input | Input | P30/INTP0/TI81 |
| BZO90 | Output | Buzzer output | Input | P33/INTP3/TO82 |
| ANIO to ANI7 | Input | A/D converter analog input | Input | P60 to P67 |
| AVref | - | A/D converter reference voltage | - | - |
| AVss | - | A/D converter ground potential | - | - |
| AV ${ }_{\text {do }}$ | - | A/D converter analog power supply | - | - |
| X1 | Input | Connecting crystal resonator for main system clock oscillation | - | - |
| X2 | - |  | - | - |
| XT1 | Input | Connecting crystal resonator for subsystem clock oscillation | - | - |
| XT2 | - |  | - | - |
| RESET | Input | System reset input | Input | - |
| Vodo | - | Positive power supply | - | - |
| VDD1 | - | Positive power supply (other than ports) | - | - |
| Vsso | - | Ground potential | - | - |
| Vss1 | - | Ground potential (other than ports) | - | - |
| IC0 | - | Internally connected. Connect this pin directly to the Vsso or Vss1 pin. | - | - |
| IC3 | - | Internally connected. Leave open. | - | - |
| VPP | - | This pin is used to set flash memory programming mode and applies a high voltage when a program is written or verified. | - | - |

### 3.2 Description of Pin Functions

### 3.2.1 P00 to P05 (Port 0)

These pins constitute a 6-bit I/O port and can be set to input or output port mode in 1-bit units by using port mode register 0 ( PMO ). When these pins are used as an input port, an on-chip pull-up resistor can be used by setting pullup resistor option register 0 (PUO).

### 3.2.2 P10, P11 (Port 1)

These pins constitute a 2-bit l/O port and can be set to input or output port mode in 1-bit units by using port mode register 1 (PM1). When these pins are used as an input port, an on-chip pull-up resistor can be used by setting pullup resistor option register 0 (PU0).

### 3.2.3 P20 to P26 (Port 2)

These pins constitute a 7-bit I/O port. In addition, these pins provide a function to perform I/O to/from the timer and to I/O the data and clock of the serial interface.

Port 2 can be set to the following operation modes in 1-bit units.

## (1) Port mode

In port mode, P20 to P26 function as a 7-bit I/O port. Port 2 can be set to input or output mode in 1-bit units by using port mode register 2 (PM2). For P20 to P22, P25, and P26, whether to use on-chip pull-up resistors can be specified in 1-bit units by using pull-up resistor option register B2 (PUB2), regardless of the setting of port mode register 2 (PM2). P23 and P24 are N-ch open-drain I/O ports.

## (2) Control mode

In this mode, P20 to P26 function as the timer I/O, the data I/O and the clock I/O of the serial interface.
(a) T 180

This is the external clock input pin for 8-bit timer/event counter 80.
(b) TO 80

This is the timer output pin of 8-bit timer/event counter 80.
(c) $\mathrm{SI} 20, \mathrm{SO} 20$

These are the serial data I/O pins of the serial interface.

## (d) $\overline{\text { SCK20 }}$

This is the serial clock I/O pin of the serial interface.
(e) $\overline{\mathrm{SS} 20}$

This is the chip select input pin of the serial interface.

## (f) RxD20, TxD20

These are the serial data I/O pins of the asynchronous serial interface.

## (g) ASCK20

This is the serial clock input pin of the asynchronous serial interface.

Caution When using P20 to P26 as serial interface pins, the I/O mode and output latch must be set according to the function to be used. For details of the setting, see Table 14-2 Operating Mode Settings of Serial Interface 20.

### 3.2.4 P30 to P33 (Port 3)

These pins constitute a 4-bit I/O port. In addition, these pins function as the timer I/O and the external interrupt input.

Port 3 can be set to the following operation modes in 1-bit units.

## (1) Port mode

In port mode, P30 to P33 function as a 4-bit I/O port. Port 3 can be set to input or output mode in 1-bit units by using port mode register 3 (PM3). Whether to use the on-chip pull-up resistor can be specified in 1-bit units by using pull-up resistor option register B3 (PUB3), regardless of the setting of port mode register 3 (PM3).

## (2) Control mode

In this mode, P30 to P33 function as the timer I/O and the external interrupt input.
(a) TI 81

This is the external clock input pin for 8-bit timer/event counter 81.
(b) TO90, TO81, TO82

These are the output pins of 16 -bit timer 90, 8-bit timer/event counter 81, and 8-bit timer 82.
(c) CPT90

This is the capture edge input pin of 16 -bit timer 90 .
(d) BZO90

This is the buzzer output pin of 16-bit timer 90.
(e) INTPO to INTP3

These are external interrupt input pins for which the valid edge (rising edge, falling edge, and both the rising and falling edges) can be specified.

### 3.2.5 P50 to P53 (Port 5)

These pins constitute a 4-bit N-ch open-drain I/O port. Port 5 can be set to input or output mode in 1-bit units by using port mode register 5 (PM5). For a mask ROM version, whether a pull-up resistor is to be incorporated can be specified by a mask option.

### 3.2.6 P60 to P67 (Port 6)

These pins constitute an 8-bit input-only port. They can function as $A / D$ converter input pins as well as a generalpurpose input port.

## (1) Port mode

In port mode, P60 to P67 function as an 8-bit input-only port.

## (2) Control mode

In control mode, P60 to P67 function as A/D converter analog inputs (ANIO to ANI7).

### 3.2.7 RESET

A low-level active system reset signal is input to this pin.

### 3.2.8 X1, X2

These pins are used to connect a crystal resonator for main system clock oscillation.
To supply an external clock, input the clock to X1 and input the inverted signal to X2.

### 3.2.9 XT1, XT2

These pins are used to connect a crystal resonator for subsystem clock oscillation.
To supply an external clock, input the clock to XT1 and input the inverted signal to XT2.

### 3.2.10 AVdd

Analog power supply pin of the $A / D$ converter. Always use the same potential as that of the Vodo pin even when the $A / D$ converter is not used.

### 3.2.11 AVss

This is a ground potential pin of the A/D converter. Always use the same potential as that of the Vsso pin even when the $A / D$ converter is not used.

### 3.2.12 AVref

This is the A/D converter reference voltage input pin. When the $A / D$ converter is not used, connect this pin to Vddo or Vsso.

### 3.2.13 Vddo, Vdd1

VdDo is a positive power supply pin for ports.
$V_{D D 1}$ is a positive power supply pin for other than ports.

### 3.2.14 Vsso, Vss1

Vsso is a ground potential for ports pin.
Vss1 is a ground potential pin for other than ports.

### 3.2.15 Vpp (flash memory version only)

High voltage application pin for flash memory programming mode setting and program write/verify.
Connect this pin in either of the following ways.

- Independently connect to a $10 \mathrm{k} \Omega$ pull-down resistor.
- By using a jumper on the board, connect directly to the dedicated flash programmer in the programming mode or to Vss in the normal operation mode.
If the wiring between the Vpp pin and Vss pin is very long or external noise is superimposed on the Vpp pin, the user program may not run correctly.


### 3.2.16 ICO (mask ROM version only)

The IC0 (internally connected) pin is used to set the $\mu$ PD789167 and 789177 Subseries to test mode before shipment. In normal operation mode, directly connect this pin to the Vsso or Vss1 pin with as short a wiring length as possible.

If a potential difference is generated between the ICO pin and Vsso or Vss1 pin due to a long wiring length or external noise superimposed on the ICO pin, the user program may not run correctly.

- Directly connect the IC0 pin to the Vsso or Vssi pin.



### 3.2.17 IC3

The IC3 pin is internally connected. Leave this pin open.

### 3.3 Pin I/O Circuits and Recommended Connection of Unused Pins

The I/O circuit type of each pin and recommended connection of unused pins are shown in Table 3-1.
For the I/O circuit configuration of each type, refer to Figure 3-1.

Table 3-1. Types of I/O Circuits for Each Pin and Recommended Connection of Unused Pins

| Pin Name | I/O Circuit Type | I/O | Recommended Connection of Unused Pins |
| :---: | :---: | :---: | :---: |
| P00 to P05 | $5-\mathrm{H}$ | I/O | Input: Independently connect to $\mathrm{V}_{\mathrm{DD}}$, $\mathrm{V}_{\mathrm{DD} 1} \mathrm{~V}_{\mathrm{ss}}$, or $\mathrm{V}_{\mathrm{ss} 1}$ via a resistor. <br> Output: Leave open. |
| P10, P11 |  |  |  |
| P20/ $\overline{\text { SCK20/ASCK20 }}$ | 8-C |  |  |
| P21/SO20/TxD20 |  |  |  |
| P22/SI20/RxD20 |  |  |  |
| P23 | 13-X |  | Input: Independently connect to Vodo or VdD1 via a resistor. <br> Output: Leave open. |
| P24 |  |  |  |
| P25/TI80/ $\overline{\text { SS20 }}$ | 8-C |  | Input: Independently connect to VdDo, VdD1, Vsso, or Vss1 via a resistor. <br> Output: Leave open. |
| P26/TO80 |  |  |  |
| P30/INTP0/TI81/CPT90 |  |  | Input: Independently connect to Vsso or Vss1 via a resistor. Output: Leave open. |
| P31/INTP1/TO81 |  |  |  |
| P32/INTP2/TO90 |  |  |  |
| P33/INTP3/TO82/BZO90 |  |  |  |
| P50 to P53 (mask ROM version) | 13-U |  | Input: Connect to Vsso or Vss1. Output: Leave open. |
| P50 to P53 (flash memory version) | 13-T |  |  |
| P60/ANI0 to P67/ANI7 | 9-C | Input | Connect directly to Vddo, VdD1, Vsso, or Vss1. |
| XT1 | - | Input | Connect directly to V ${ }_{\text {sso }}$ or Vssi. |
| XT2 |  | - | Leave open. |
| RESET | 2 | Input | - |
| IC0 (mask ROM version) | - | - | Connect directly to Vsso or Vssi. |
| IC3 |  |  | Leave open. |
| VPP (flash memory version) |  |  | Independently connect via a $10 \mathrm{k} \Omega$ pull-down resistor, or connect directly to Vsso or Vssi. |

Figure 3-1. Pin I/O Circuits


## CHAPTER 4 PIN FUNCTIONS ( $\mu$ PD789167Y AND 789177Y SUBSERIES)

### 4.1 Pin Function List

## (1) Port pins

| Pin Name | I/O | Function | After Reset | Alternate Function |
| :---: | :---: | :---: | :---: | :---: |
| P00 to P05 | I/O | Port 0 <br> 6-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P10, P11 | I/O | Port 1 <br> 2-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P20 | I/O | Port 2 <br> 7-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> For P20 to P22, P25, and P26, an on-chip pull-up resistor can be specified by means of pull-up resistor option register B2 (PUB2). <br> Only P23 and P24 can be used as N-ch open-drain I/O port pins. | Input | $\overline{\text { SCK20/ASCK20 }}$ |
| P21 |  |  |  | SO20/TxD20 |
| P22 |  |  |  | SI20/RxD20 |
| P23 |  |  |  | SCL0 |
| P24 |  |  |  | SDA0 |
| P25 |  |  |  | T180/ $\overline{\text { SS20 }}$ |
| P26 |  |  |  | TO80 |
| P30 | I/O | Port 3 <br> 4-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> An on-chip pull-up resistor can be specified by means of pullup resistor option register B3 (PUB3). | Input | INTP0/TI81/CPT90 |
| P31 |  |  |  | INTP1/TO81 |
| P32 |  |  |  | INTP2/TO90 |
| P33 |  |  |  | INTP3/TO82/BZO90 |
| P50 to P53 | I/O | Port 5 <br> 4-bit N-ch open-drain I/O port <br> I/O mode can be specified in 1-bit units. <br> For a mask ROM version, an on-chip pull-up resistor can be specified by the mask option. | Input | - |
| P60 to P67 | Input | Port 6 <br> 8-bit input-only port | Input | ANIO to ANI7 |

## (2) Non-port pins

| Pin Name | I/O | Function | After Reset | Alternate Function |
| :---: | :---: | :---: | :---: | :---: |
| INTPO | Input | External interrupt input for which the valid edge (rising edge, falling edge, or both rising and falling edges) can be specified | Input | P30/TI81/CPT90 |
| INTP1 |  |  |  | P31/TO81 |
| INTP2 |  |  |  | P32/TO90 |
| INTP3 |  |  |  | P33/TO82/BZO90 |
| SI20 | Input | Serial data input to serial interface | Input | P22/RxD20 |
| SO20 | Output | Serial data output from serial interface | Input | P21/TxD20 |
| $\overline{\text { SCK20 }}$ | I/O | Serial clock I/O for serial interface | Input | P20/ASCK20 |
| SS20 | Input | Chip select input to serial interface | Input | P25/TI80 |
| ASCK20 | Input | Serial clock input for asynchronous serial interface | Input | P20/SCK20 |
| RxD20 | Input | Serial data input for asynchronous serial interface | Input | P22/SI20 |
| TxD20 | Output | Serial data output for asynchronous serial interface | Input | P21/SO20 |
| SCL0 | I/O | SMB0 clock I/O | Input | P23 |
| SDA0 | I/O | SMB0 data I/O | Input | P24 |
| TI80 | Input | External count clock input to 8-bit timer/event counter (TM80) | Input | P25/SS20 |
| TI81 | Input | External count clock input to 8-bit timer/event counter (TM81) | Input | P30/INTP0/CPT90 |
| TO80 | Output | 8-bit timer/event counter (TM80) output | Input | P26 |
| TO81 | Output | 8-bit timer/event counter (TM81) output | Input | P31/INTP1 |
| TO82 | Output | 8-bit timer (TM82) output | Input | P33/INTP3/BZO90 |
| TO90 | Output | 16-bit timer (TM90) output | Input | P32/INTP2 |
| CPT90 | Input | Capture edge input | Input | P30/INTP0/T181 |
| BZO90 | Output | Buzzer output | Input | P33/INTP3/TO82 |
| ANI0 to ANI7 | Input | A/D converter analog input | Input | P60 to P67 |
| AVref | - | A/D converter reference voltage | - | - |
| AVss | - | A/D converter ground potential | - | - |
| AVDD | - | A/D converter analog power supply | - | - |
| X1 | Input | Connecting crystal resonator for main system clock oscillation | - | - |
| X2 | - |  | - | - |
| XT1 | Input | Connecting crystal resonator for subsystem clock oscillation | - | - |
| XT2 | - |  | - | - |
| RESET | Input | System reset input | Input | - |
| VDDo | - | Positive power supply | - | - |
| VDD1 | - | Positive power supply (other than ports) | - | - |
| Vsso | - | Ground potential | - | - |
| Vss1 | - | Ground potential (other than ports) | - | - |
| ICO | - | Internally connected. Connect this pin directly to the Vsso or Vssi pin. | - | - |
| IC2 | - | Internally connected. Leave this pin open. | - | - |
| VPP | - | This pin is used to set flash memory programming mode and applies a high voltage when a program is written or verified. | - | - |

### 4.2 Description of Pin Functions

### 4.2.1 P00 to P05 (Port 0)

These pins constitute a 6-bit I/O port and can be set to input or output port mode in 1-bit units by using port mode register 0 ( PMO ). When these pins are used as an input port, an on-chip pull-up resistor can be used by setting pullup resistor option register 0 (PUO).

### 4.2.2 P10, P11 (Port 1)

These pins constitute a 2-bit l/O port and can be set to input or output port mode in 1-bit units by using port mode register 1 (PM1). When these pins are used as an input port, an on-chip pull-up resistor can be used by setting pullup resistor option register 0 (PU0).

### 4.2.3 P20 to P26 (Port 2)

These pins constitute a 7 -bit I/O port. In addition, these pins provide a function to perform I/O to/from the timer and to I/O the data and clock of the serial interface.

Port 2 can be set to the following operation modes in 1-bit units.

## (1) Port mode

In port mode, P20 to P26 function as a 7-bit I/O port. Port 2 can be set to input or output mode in 1-bit units by using port mode register 2 (PM2). For P20 to P22, P25, and P26, whether to use on-chip pull-up resistors can be specified in 1-bit units by using pull-up resistor option register B2 (PUB2), regardless of the setting of port mode register 2 (PM2). P23 and P24 are N-ch open-drain I/O ports.

## (2) Control mode

In this mode, P20 to P26 function as the timer I/O, the data I/O and the clock I/O of the serial interface.
(a) T 180

This is the external clock input pin for 8-bit timer/event counter 80.
(b) TO 80

This is the timer output pin of 8-bit timer/event counter 80.
(c) $\mathrm{SI} 20, \mathrm{SO} 20$

These are the serial data I/O pins of the serial interface.

## (d) $\overline{\text { SCK20 }}$

This is the serial clock I/O pin of the serial interface.
(e) $\overline{\mathrm{SS} 20}$

This is the chip select input pin of the serial interface.

## (f) RxD20, TxD20

These are the serial data I/O pins of the asynchronous serial interface.

## (g) ASCK20

This is the serial clock input pin of the asynchronous serial interface.
(h) SCLO

This is the clock I/O pin of SMBO.
(i) SDAO

This is the data I/O pin of SMBO.

## Caution When using P20 to P26 as serial interface pins, the I/O mode and output latch must be set according to the function to be used. For details of the setting, see Table 14-2 Operating Mode Setting of Serial Interface 20.

### 4.2.4 P30 to P33 (Port 3)

These pins constitute a 4-bit I/O port. In addition, these pins function as the timer I/O and the external interrupt input.

Port 3 can be set to the following operation modes in 1-bit units.

## (1) Port mode

In port mode, P30 to P33 function as a 4-bit I/O port. Port 3 can be set to input or output mode in 1-bit units by using port mode register 3 (PM3). Whether to use the on-chip pull-up resistor can be specified in 1-bit units by using pull-up resistor option register B3 (PUB3), regardless of the setting of port mode register 3 (PM3).
(2) Control mode

In this mode, P30 to P33 function as the timer I/O and the external interrupt input.
(a) TI 81

This is the external clock input pin for 8-bit timer/event counter 81.
(b) TO90, T081, TO82

These are the output pins of 16 -bit timer 90, 8-bit timer/event counter 81, and 8-bit timer 82.
(c) CPT90

This is the capture edge input pin of 16-bit timer 90.
(d) BZO90

This is the buzzer output pin of 16-bit timer 90.

## (e) INTPO to INTP3

These are external interrupt input pins for which the valid edge (rising edge, falling edge, and both the rising and falling edges) can be specified.

### 4.2.5 P50 to P53 (Port 5)

These pins constitute a 4-bit N-ch open-drain I/O port. Port 5 can be set to input or output mode in 1-bit units by using port mode register 5 (PM5). For a mask ROM version, whether a pull-up resistor is to be incorporated can be specified by a mask option.

### 4.2.6 P60 to P67 (Port 6)

These pins constitute an 8-bit input-only port. They can function as $A / D$ converter input pins as well as a generalpurpose input port.

## (1) Port mode

In port mode, P60 to P67 function as an 8-bit input-only port.

## (2) Control mode

In control mode, P60 to P67 function as A/D converter analog inputs (ANIO to ANI7).

### 4.2.7 RESET

A low-level active system reset signal is input to this pin.

### 4.2.8 X1, X2

These pins are used to connect a crystal resonator for main system clock oscillation.
To supply an external clock, input the clock to X1 and input the inverted signal to X2.

### 4.2.9 XT1, XT2

These pins are used to connect a crystal resonator for subsystem clock oscillation.
To supply an external clock, input the clock to XT1 and input the inverted signal to XT2.

### 4.2.10 AVdd

Analog power supply pin of the $A / D$ converter. Always use the same potential as that of the Vodo pin even when the $A / D$ converter is not used.

### 4.2.11 AVss

This is a ground potential pin of the A/D converter. Always use the same potential as that of the Vsso pin even when the $A / D$ converter is not used.

### 4.2.12 AVref

This is the A/D converter reference voltage input pin. When the A/D converter is not used, connect this pin to Vddo or Vsso.

### 4.2.13 Vddo, Vdd1

Vddo is a positive power supply pin for ports.
$V_{D D 1}$ is a positive power supply pin for other than ports.

### 4.2.14 Vsso, Vss1

Vsso is a ground potential pin for ports.
Vss1 is a ground potential pin for other than ports.

### 4.2.15 Vpp (flash memory version only)

High voltage apply pin for flash memory programming mode setting and program write/verify.
Connect this pin in either of the following ways.

- Independently connect to a $10 \mathrm{k} \Omega$ pull-down resistor.
- By using a jumper on the board, connect directly to the dedicated flash programmer in the programming mode or to Vss in the normal operation mode.
If the wiring between the Vpp pin and Vss pin is very long or external noise is superimposed on the Vpp pin, the user program may not run correctly.


### 4.2.16 ICO (mask ROM version only)

The IC0 (internally connected) pin is used to set the $\mu$ PD789167Y and 789177Y Subseries to test mode before shipment. In normal operation mode, directly connect this pin to the $\mathrm{V}_{\mathrm{ss}}$ or V ss1 pin with as short a wiring length as possible.

If a potential difference is generated between the ICO pin and Vsso or Vss1 pin due to a long wiring length or external noise superimposed on the ICO pin, the user program may not run correctly.

- Directly connect the IC0 pin to the Vsso or Vss1 pin.



### 4.2.17 IC2

The IC2 pin is internally connected. Leave this pin open.

### 4.3 Pin I/O Circuits and Recommended Connection of Unused Pins

The I/O circuit type of each pin and recommended connection of unused pins are shown in Table 4-1.
For the I/O circuit configuration of each type, refer to Figure 4-1.

Table 4-1. Types of I/O Circuits for Each Pin and Recommended Connection of Unused Pins

| Pin Name | I/O Circuit Type | I/O | Recommended Connection of Unused Pins |
| :---: | :---: | :---: | :---: |
| P00 to P05 | $5-\mathrm{H}$ | I/O | Input: Independently connect to $\mathrm{V}_{\mathrm{DD}}$, $\mathrm{V}_{\mathrm{DD} 1}$, $\mathrm{V}_{\mathrm{ss}}$, or $\mathrm{V}_{\mathrm{SS} 1}$ via a resistor. <br> Output: Leave open. |
| P10, P11 |  |  |  |
| P20/SCK20/ASCK20 | 8-C |  |  |
| P21/SO20/TxD20 |  |  |  |
| P22/SI20/RxD20 |  |  |  |
| P23/SCL0 | 13-X |  | Input: Independently connect to $\mathrm{V}_{\mathrm{DDO}}$ or $\mathrm{V}_{\mathrm{DD1}}$ via a resistor. |
| P24/SDA0 |  |  | Output: Leave open. |
| P25/TI80/ $\overline{\text { SS20 }}$ | 8-C |  | Input: Independently connect to Vddo, VDD1, Vsso, or Vss1 via a resistor. <br> Output: Leave open. |
| P26/TO80 |  |  |  |
| P30/INTP0/TI81/CPT90 |  |  | Input: Independently connect to Vsso or Vssı via a resistor. Output: Leave open. |
| P31/INTP1/TO81 |  |  |  |
| P32/INTP2/TO90 |  |  |  |
| P33/INTP3/TO82/BZO90 |  |  |  |
| P50 to P53 (mask ROM version) | 13-U |  | Input: Connect to Vsso or Vss1. <br> Output: Leave open. |
| P50 to P53 (flash memory version) | 13-T |  |  |
| P60/ANI0 to P67/ANI7 | 9-C | Input | Connect directly to Vddo, Vdd1, Vsso, or Vss1. |
| XT1 | - | Input | Connect directly to Vsso or Vssi. |
| XT2 |  | - | Leave open. |
| $\overline{\text { RESET }}$ | 2 | Input | - |
| IC0 (mask ROM version) | - | - | Connect directly to Vsso or Vsss. |
| IC2 |  |  | Leave open. |
| VPP (flash memory version) |  |  | Independently connect via a $10 \mathrm{k} \Omega$ pull-down resistor, or connect directly to Vsso or Vss1. |

Figure 4-1. Pin I/O Circuits


## CHAPTER 5 CPU ARCHITECTURE

### 5.1 Memory Space

Products in the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries can each access up to 64 KB of memory space. Figures 5-1 through 5-3 show the memory maps.

Figure 5-1. Memory Map ( $\mu$ PD789166, $\mu$ PD789176, $\mu$ PD789166Y, and $\mu$ PD789176Y)


Figure 5-2. Memory Map ( $\mu$ PD789167, $\mu$ PD789177, $\mu$ PD789167Y, and $\mu$ PD789177Y)


Figure 5-3. Memory Map ( $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY)


### 5.1.1 Internal program memory space

The internal program memory space stores programs and table data. This space is usually addressed by the program counter (PC).

The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries provide the following internal ROM (or flash memory) containing the following capacities.

Table 5-1. Internal ROM Capacity

| Part Number |  | Internal ROM |
| :--- | :--- | :--- |
|  |  | Structure |

The following areas are allocated to the internal program memory space.

## (1) Vector table area

A 36 -byte area of addresses 0000 H to 0023 H is reserved as a vector table area. This area stores program start addresses to be used when branching by RESET input or interrupt request generation. Of a 16-bit program address, the lower 8 bits are stored in an even address, and the higher 8 bits are stored in an odd address.

Table 5-2. Vector Table

| Vector Table Address | Interrupt Request | Vector Table Address | Interrupt Request |
| :--- | :--- | :--- | :--- |
| 0000 H | $\overline{\text { RESET }}$ input | 0014 H | INTWTI |
| 0004 H | INTWDT | 0016 H | INTTM80 |
| 0006 H | INTP0 | 0018 H | INTTM81 |
| 0008 H | INTP1 | 001 AH | INTTM82 |
| 000 AH | INTP2 | 001 CH | INTTM90 |
| 000 CH | INTP3 | 001 EH | INTSMB0 |
| 000 EH | INTSR20/INTCSI20 | 0020 H | INTSMBOV0 ${ }^{\text {Note }}$ |
| 0010 H | INTST20 | 0022 H | INTAD0 |
| 0012 H | INTWT |  |  |

## Note For the $\mu$ PD789167Y and 789177Y Subseries only

## (2) CALLT instruction table area

The subroutine entry address of a 1-byte call instruction (CALLT) can be stored in a 64-byte area of addresses 0040 H to 007 FH .

### 5.1.2 Internal data memory (internal high-speed RAM) space

The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries provide a 512-byte internal high-speed RAM.
The internal high-speed RAM can also be used as a stack memory.

### 5.1.3 Special-function register (SFR) area

Special-function registers (SFRs) of on-chip peripheral hardware are allocated to an area of FFOOH to FFFFH (see Table 5-3).

### 5.1.4 Data memory addressing

Each of the $\mu$ PD789167, 789177, 789167Y, 789177Y Subseries is provided with a wide range of addressing modes to make memory manipulation as efficient as possible. A data memory area (FDOOH to FFFFH) can be accessed using a unique addressing mode according to its use, such as a special-function register (SFR). Figures 54 through 5-6 illustrate the data memory addressing modes.

Figure 5-4. Data Memory Addressing Modes ( $\mu$ PD789166, $\mu$ PD789176, $\mu$ PD789166Y, and $\mu$ PD789176Y)


Figure 5-5. Data Memory Addressing Modes ( $\mu$ PD789167, $\mu$ PD789177, $\mu$ PD789167Y, and $\mu$ PD789177Y)


Figure 5-6. Data Memory Addressing Modes ( $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY)


### 5.2 Processor Registers

The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries provide the following on-chip processor registers.

### 5.2.1 Control registers

The control registers have special functions to control the program sequence statuses and stack memory. The control registers include a program counter, a program status word, and a stack pointer.

## (1) Program counter (PC)

The program counter is a 16 -bit register which holds the address information of the next program to be executed.
In normal operation, the PC is automatically incremented according to the number of bytes of the instruction to be fetched. When a branch instruction is executed, immediate data or register contents are set.
$\overline{\text { RESET }}$ input sets the reset vector table values at addresses 0000 H and 0001 H to the program counter.

Figure 5-7. Program Counter Configuration

| 15 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PC | PC15 | PC14 | PC13 | PC12 | PC11 | PC10 | PC9 | PC8 | PC7 | PC6 | PC5 | PC4 | PC3 | PC2 | PC1 | PC0 |

## (2) Program status word (PSW)

The program status word is an 8-bit register consisting of various flags to be set/reset by instruction execution.
Program status word contents are automatically stacked upon interrupt request generation or PUSH PSW instruction execution and are automatically restored upon execution of the RETI and POP PSW instructions. $\overline{\text { RESET }}$ input sets the PSW to 02 H .

Figure 5-8. Program Status Word Configuration


## (a) Interrupt enable flag (IE)

This flag controls interrupt request acknowledgment operations of the CPU.
When $\mathrm{IE}=0$, the interrupt disabled (DI) status is set. All interrupt requests except non-maskable interrupt are disabled.
When $I E=1$, the interrupt enabled ( EI ) status is set. Interrupt request acknowledgment is controlled with an interrupt mask flag for various interrupt sources.
This flag is reset to 0 upon DI instruction execution or interrupt acknowledgment and is set to 1 upon El instruction execution.
(b) Zero flag (Z)

When the operation result is zero, this flag is set to 1 . It is reset to 0 in all other cases.
(c) Auxiliary carry flag (AC)

If the operation result has a carry from bit 3 or a borrow at bit 3 , this flag is set to 1 . It is reset to 0 in all other cases.

## (d) Carry flag (CY)

This flag stores an overflow or underflow that occurs upon add/subtract instruction execution. It stores the shift-out value upon rotate instruction execution and functions as a bit accumulator during bit operation instruction execution.

## (3) Stack pointer (SP)

This is a 16-bit register used to hold the start address of the memory stack area. Only the internal highspeed RAM area can be set as the stack area.

Figure 5-9. Stack Pointer Configuration
15

| 15 | SP1 |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SP15 | SP14 | SP13 | SP12 | SP11 | SP10 | SP9 | SP8 | SP7 | SP6 | SP5 | SP4 | SP3 | SP2 | SP1 |
| SP0 |  |  |  |  |  |  |  |  |  |  |  |  |  |  |

The SP is decremented ahead of writing (saving) to the stack memory and is incremented after reading (restoring) from the stack memory.
Each stack operation saves/restores data as shown in Figures 5-10 and 5-11.

## Caution Since $\overline{\operatorname{RESET}}$ input makes SP contents undefined, be sure to initialize the SP before using the stack.

Figure 5-10. Data to Be Saved to Stack Memory


Figure 5-11. Data to Be Restored from Stack Memory


### 5.2.2 General-purpose registers

The general-purpose registers consist of eight 8-bit registers ( $X, A, C, B, E, D, L$, and $H$ ).
In addition that each register can be used as an 8-bit register, two 8-bit registers in pairs can be used as a 16-bit register ( $\mathrm{AX}, \mathrm{BC}, \mathrm{DE}$, and HL ).

They can be described in terms of functional names (X, A, C, B, E, D, L, H, AX, BC, DE, and HL) and absolute names (R0 to R7 and RP0 to RP3).

Figure 5-12. General-Purpose Register Configuration
(a) Absolute names

(b) Functional names


### 5.2.3 Special-function registers (SFR)

Unlike a general-purpose register, each special-function register has a special function.
They are allocated to the 256-byte area FFOOH to FFFFH.
The special-function registers can be manipulated, like the general-purpose registers, with operation, transfer, and bit manipulation instructions. Manipulatable bit units (1, 8, and 16) differ depending on the special-function register type.

Each manipulation bit unit can be specified as follows.

- 1-bit manipulation

Describes a symbol reserved by the assembler for the 1-bit manipulation instruction operand (sfr.bit). This manipulation can also be specified with an address.

- 8-bit manipulation

Describes a symbol reserved by the assembler for the 8 -bit manipulation instruction operand (sfr). This manipulation can also be specified with an address.

- 16-bit manipulation

Describes a symbol reserved by the assembler for the 16-bit manipulation instruction operand. When specifying an address, describe an even address.

Table 5-3 lists the special-function registers. The meanings of the symbols in this table are as follows.

- Symbol

Indicates the addresses of the implemented special-function registers. The symbols shown in this column are reserved words in the assembler, and have already been defined as sfr variables by the \#pragma sfr directive in the C compiler. Therefore, these symbols can be used as instruction operands if an assembler or integrated debugger is used.

- R/W

Indicates whether the special-function register can be read or written.
R/W: Read/write
R: Read only
W: Write only

- Bit units for manipulation

Indicates the bit units (1, 8, and 16) in which the special-function register can be manipulated.

- After reset

Indicates the status of the special-function register when the RESET signal is input.

Table 5-3. Special-Function Registers (1/2)

| Address | Special-Function Register (SFR) Name | Symbol |  | R/W | Bit Units for Manipulation |  |  | After Reset |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 1 Bit | 8 Bits | 16 Bits |  |
| FFOOH | Port 0 | PO |  |  | R/W | $\checkmark$ | $\checkmark$ | - | OOH |
| FF01H | Port 1 | P1 |  | $\checkmark$ |  | $\checkmark$ | - |  |  |
| FF02H | Port 2 | P2 |  | $\checkmark$ |  | $\checkmark$ | - |  |  |
| FF03H | Port 3 | P3 |  | $\checkmark$ |  | $\checkmark$ | - |  |  |
| FF05H | Port 5 | P5 |  | $\checkmark$ |  | $\checkmark$ | - |  |  |
| FF06H | Port 6 | P6 |  | R | $\sqrt{ }$ | $\checkmark$ | - |  |  |
| FF10H | 16-bit multiplication result storage register 0 | MULOL | MULO |  | - | - | $\sqrt{\text { Notes 2, }} 3$ | Undefined |  |
| FF11H |  | MULOH |  |  |  |  |  |  |  |
| FF14H | A/D conversion result register 0 | ADCR0 |  |  | - | $V^{\text {Note } 1}$ | $V^{\text {Note } 2}$ |  |  |
| FF15H |  |  |  |  |  |  |  |  |  |
| FF16H | 16-bit compare register 90 | CR90L | CR90 |  | W | - | - | $\sqrt{\text { Notes 2, } 3}$ | FFFFH |
| FF17H |  | $\mathrm{CR90H}$ |  |  |  |  |  |  |  |
| FF18H | 16-bit timer counter 90 | TM90L | TM90 | R | - | - | $\sqrt{\text { Notes 2, } 3}$ | 0000H |  |
| FF19H |  | TM90H |  |  |  |  |  |  |  |
| FF1AH | 16-bit capture register 90 | TCP90L | TCP90 |  | - | - | $\sqrt{\text { Notes 2, }} 3$ | Undefined |  |
| FF1BH |  | TCP90H |  |  |  |  |  |  |  |
| FF20H | Port mode register 0 | PMO |  | R/W | $\checkmark$ | $\checkmark$ | - | FFH |  |
| FF21H | Port mode register 1 | PM1 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF22H | Port mode register 2 | PM2 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF23H | Port mode register 3 | PM3 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF25H | Port mode register 5 | PM5 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF32H | Pull-up resistor option register B2 | PUB2 |  |  | $\checkmark$ | $\checkmark$ | - | OOH |  |
| FF33H | Pull-up resistor option register B3 | PUB3 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF42H | Timer clock selection register 2 | TCL2 |  |  | - | $\checkmark$ | - |  |  |
| FF48H | 16-bit timer mode control register 90 | TMC90 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF49H | Buzzer output control register 90 | BZC90 |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF4AH | Watch timer mode control register | WTM |  |  | $\checkmark$ | $\checkmark$ | - |  |  |
| FF50H | 8-bit compare register 80 | CR80 |  | W | - | $\checkmark$ | - | Undefined |  |
| FF51H | 8-bit timer counter 80 | TM80 |  | R | - | $\checkmark$ | - | OOH |  |
| FF53H | 8-bit timer mode control register 80 | TMC80 |  | R/W | $\checkmark$ | $\checkmark$ | - |  |  |

Notes 1. When using this register with an 8-bit A/D converter ( $\mu$ PD789167 or 789167 Y Subseries), the register can be accessed in 8-bit units. At this time, the address is FF15H.
When using this register with a 10-bit A/D converter ( $\mu$ PD789177 or 789177Y Subseries), the register can be accessed only in 16-bit units. When the $\mu$ PD78F9177 or $\mu$ PD78F9177A, the flash memory counterpart of the $\mu$ PD789166 or $\mu$ PD789167, is used, the register can be accessed in 8 -bit units. However, only an object file assembled with the $\mu$ PD789166 or $\mu$ PD789167 can be used. The same is also true for the $\mu$ PD78F9177Y or $\mu$ PD78F9177AY, the flash memory counterpart of the $\mu$ PD789166Y or $\mu$ PD789167Y. When the $\mu$ PD78F9177Y or $\mu$ PD78F9177AY is used, the register can be accessed in 8 -bit units. However, only an object file assembled with the $\mu \mathrm{PD} 789166 \mathrm{Y}$ and $\mu \mathrm{PD} 789167 \mathrm{Y}$ can be used.
2. 16-bit access is allowed only with short direct addressing.
3. MULO, CR90, TM90, and TCP90 are designed only for 16-bit access. With direct addressing, however, they can also be accessed in 8-bit mode.

Table 5-3. Special-Function Registers (2/2)

| Address | Special-Function Register (SFR) Name | Symbol |  | R/W | Bit Units for Manipulation |  |  | After Reset |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | 1 Bit | 8 Bits | 16 Bits |  |
| FF54H | 8-bit compare register 81 | CR81 |  |  | W | - | $\checkmark$ | - | Undefined |
| FF55H | 8-bit timer counter 81 | TM81 |  | R | - | $\checkmark$ | - | OOH |
| FF57H | 8-bit timer mode control register 81 | TMC81 |  | R/W | $\checkmark$ | $\checkmark$ | - |  |
| FF58H | 8-bit compare register 82 | CR82 |  | W | - | $\checkmark$ | - | Undefined |
| FF59H | 8-bit timer counter 82 | TM82 |  | R | - | $\checkmark$ | - | OOH |
| FF5BH | 8-bit timer mode control register 82 | TMC82 |  | R/W | $\checkmark$ | $\checkmark$ | - |  |
| FF70H | Asynchronous serial interface mode register 20 | ASIM20 |  |  | $\sqrt{ }$ | $\checkmark$ | - |  |
| FF71H | Asynchronous serial interface status register 20 | ASIS20 |  | R | $\checkmark$ | $\checkmark$ | - |  |
| FF72H | Serial operation mode register 20 | CSIM20 |  | R/W | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FF73H | Baud rate generator control register $20$ | BRGC20 |  |  | - | $\checkmark$ | - |  |
| FF74H | Transmission shift register 20 | TXS20 | SIO20 | W | - | $\sqrt{ }$ | - | FFH |
|  | Reception buffer register 20 | $\begin{aligned} & \text { RXB2 } \\ & 0 \end{aligned}$ |  | R | - | $\sqrt{ }$ | - | Undefined |
| FF78H | SMB control register $0^{\text {Note }}$ | SMBC0 |  | R/W | $\checkmark$ | $\sqrt{ }$ | - | OOH |
| FF79H | SMB status register $0^{\text {Note }}$ | SMBS0 |  | R | $\sqrt{ }$ | $\checkmark$ | - |  |
| FF7AH | SMB clock selection register $0^{\text {Note }}$ | SMBCL0 |  | R/W | $\checkmark$ | $\checkmark$ | - |  |
| FF7BH | SMB slave address register $0^{\text {Note }}$ | SMBSVA0 |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FF7CH | SMB mode register $0^{\text {Note }}$ | SMBM0 |  |  | $\checkmark$ | $\checkmark$ | - | 20 H |
| FF7DH | SMB input level setting register $0^{\text {Note }}$ | SMBVIO |  |  | $\checkmark$ | $\checkmark$ | - | OOH |
| FF7EH | SMB shift register $0^{\text {Note }}$ | SMB0 |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FF80H | A/D converter mode register 0 | ADM0 |  |  | $\checkmark$ | $\checkmark$ | - |  |
| FF84H | A/D input selection register 0 | ADS0 |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FFD0H | Multiplication data register A0 | MRA0 |  | W | $\checkmark$ | $\checkmark$ | - | Undefined |
| FFD1H | Multiplication data register B0 | MRB0 |  |  | $\checkmark$ | $\sqrt{ }$ | - |  |
| FFD2H | Multiplier control register 0 | MULC0 |  | R/W | $\sqrt{ }$ | $\sqrt{ }$ | - | OOH |
| FFEOH | Interrupt request flag register 0 | IFO |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FFE1H | Interrupt request flag register 1 | IF1 |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FFE4H | Interrupt mask flag register 0 | MKO |  |  | $\checkmark$ | $\checkmark$ | - | FFH |
| FFE5H | Interrupt mask flag register 1 | MK1 |  |  | $\checkmark$ | $\checkmark$ | - |  |
| FFECH | External interrupt mode register 0 | INTM0 |  |  | - | $\checkmark$ | - | OOH |
| FFEDH | External interrupt mode register 1 | INTM1 |  |  | - | $\sqrt{ }$ | - |  |
| FFFOH | Suboscillation mode register | SCKM |  |  | $\checkmark$ | $\checkmark$ | - |  |
| FFF2H | Subclock control register | CSS |  |  | $\sqrt{ }$ | $\checkmark$ | - |  |
| FFF7H | Pull-up resistor option register 0 | PU0 |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - |  |
| FFF9H | Watchdog timer mode register | WDTM |  |  | $\checkmark$ | $\checkmark$ | - |  |
| FFFAH | Oscillation stabilization time selection register | OSTS |  |  | - | $\checkmark$ | - | 04H |
| FFFBH | Processor clock control register | PCC |  |  | $\sqrt{ }$ | $\sqrt{ }$ | - | 02H |

Note For the $\mu$ PD789167Y and 789177Y Subseries only

### 5.3 Instruction Address Addressing

An instruction address is determined by the program counter (PC) contents. The PC contents are normally incremented ( +1 for each byte) automatically according to the number of bytes of an instruction to be fetched each time another instruction is executed. When a branch instruction is executed, the branch destination information is set to the PC and branched by the following addressing (for details of each instruction, refer to $\mathbf{7 8 K} / \mathbf{0 S}$ Series Instruction User's Manual (U11047E)).

### 5.3.1 Relative addressing

## [Function]

The value obtained by adding 8-bit immediate data (displacement value: jdisp8) of an instruction code to the start address of the following instruction is transferred to the program counter ( PC ) and branched. The displacement value is treated as signed two's complement data $(-128$ to +127$)$ and bit 7 becomes a sign bit. In other words, the range of branch in relative addressing is between -128 and +127 of the start address of the following instruction.
This function is carried out when the BR \$addr16 instruction or a conditional branch instruction is executed.

## [IIlustration]



When $S=0, \alpha$ indicates all bits " 0 ".
When $S=1, \alpha$ indicates all bits "1".

### 5.3.2 Immediate addressing

## [Function]

Immediate data in the instruction word is transferred to the program counter (PC) and branched. This function is carried out when the CALL !addr16 and BR !addr16 instructions are executed.
CALL !addr16 and BR !addr16 instructions can be used to branch to all the memory spaces.

## [IIlustration]

In case of CALL !addr16 and BR !addr16 instructions


### 5.3.3 Table indirect addressing

## [Function]

Table contents (branch destination address) of the particular location to be addressed by the immediate data of an instruction code from bit 1 to bit 5 are transferred to the program counter (PC) and branched.
Table indirect addressing is carried out when the CALLT [addr5] instruction is executed. This instruction can be used to branch to all the memory spaces according to the address stored in the memory table 40 H to 7 FH .

## [IIlustration]



### 5.3.4 Register addressing

## [Function]

Register pair (AX) contents to be specified with an instruction word are transferred to the program counter (PC) and branched.
This function is carried out when the BR AX instruction is executed.

## [IIlustration]



### 5.4 Operand Address Addressing

The following methods are available to specify the register and memory (addressing) which undergo manipulation during instruction execution.

### 5.4.1 Direct addressing

## [Function]

The memory indicated by immediate data in an instruction word is directly addressed.

## [Operand format]

| Identifier | Description |
| :---: | :---: |
| addr16 | Label or 16-bit immediate data |

## [Description example]

MOV A, !FEOOH; When setting !addr16 to FEOOH


| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $00 H$ |  |  |  |  |  |  |  |


| 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

[Illustration]


### 5.4.2 Short direct addressing

## [Function]

The memory to be manipulated in the fixed space is directly addressed with 8-bit data in an instruction word.
The fixed space where this addressing is applied to is the 256 -byte space FE20H to FF1FH. An internal highspeed RAM and special-function registers (SFR) are mapped at FE20H to FEFFH and FF00H to FF1FH, respectively.
The SFR area (FF00H to FF1FH) where short direct addressing is applied is a part of the overall SFR area. In this area, ports which are frequently accessed in a program and a compare register of the timer counter are mapped, and these SFRs can be manipulated with a small number of bytes and clocks.
When 8 -bit immediate data is at 20 H to FFH , bit 8 of an effective address is set to 0 . When it is at 00 H to 1 FH , bit 8 is set to 1 . See [Illustration] below.

## [Operand format]

| Identifier | Description |
| :---: | :--- |
| saddr | Label or FE20H to FF1FH immediate data |
| saddrp | Label or FE20H to FF1FH immediate data (even address only) |

## [Description example]

MOV FE90H, \#50H; When setting saddr to FE90H and the immediate data to 50 H

[IIlustration]


When 8-bit immediate data is 20 H to $\mathrm{FFH}, \alpha=0$. When 8-bit immediate data is 00 H to $1 \mathrm{FH}, \alpha=1$.

### 5.4.3 Special-function register (SFR) addressing

## [Function]

The memory-mapped special-function registers (SFR) are addressed with 8-bit immediate data in an instruction word.
This addressing is applied to the 256 -byte space FFOOH to FFFFH. However, the SFRs mapped at FFOOH to FF1FH can also be accessed with short direct addressing.

## [Operand format]

| Identifier | Description |
| :---: | :--- |
| sfr | Special-function register name |

## [Description example]

MOV PMO, A; When selecting PMO for sfr

Instruction code | 1 | 1 | 1 | 0 | 0 | 1 | 1 | 1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |

| 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

[IIlustration]


### 5.4.4 Register addressing

## [Function]

The general-purpose registers are accessed as operands. The general-purpose register to be accessed is specified by the register specification code and functional name in the instruction code.
Register addressing is carried out when an instruction with the following operand format is executed. When an 8 -bit register is specified, one of the eight registers is specified with 3 bits in the instruction code.

## [Operand format]

| Identifier |  | Description |
| :---: | :--- | :--- |
| $r$ | $\mathrm{X}, \mathrm{A}, \mathrm{C}, \mathrm{B}, \mathrm{E}, \mathrm{D}, \mathrm{L}, \mathrm{H}$ |  |
| rp | $\mathrm{AX}, \mathrm{BC}, \mathrm{DE}, \mathrm{HL}$ |  |

' $r$ ' and ' $r$ ' can be described with absolute names ( $R 0$ to R7 and RP0 to RP3) as well as function names ( $X, A$, $C, B, E, D, L, H, A X, B C, D E$, and $H L$ ).

## [Description example]

MOV A, C; When selecting the $C$ register for $r$

Instruction code | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |



INCW DE; When selecting the DE register pair for rp


### 5.4.5 Register indirect addressing

## [Function]

The memory is addressed with the contents of the register pair specified as an operand. The register pair to be accessed is specified with the register pair specify code in the instruction code. This addressing can be carried out for all the memory spaces.

## [Operand format]

| Identifier |  | Description |
| :---: | :--- | :--- |
| - | $[\mathrm{DE}],[\mathrm{HL}]$ |  |

## [Description example]

MOV A, [DE]; When selecting register pair [DE]

Instruction code | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

[Illustration]


### 5.4.6 Based addressing

## [Function]

8 -bit immediate data is added to the contents of the base register, that is, the HL register pair, and the sum is used to address the memory. Addition is performed by expanding the offset data as a positive number to 16 bits. A carry from the 16th bit is ignored. This addressing can be carried out for all the memory spaces.

## [Operand format]

| Identifier |  | Description |
| :---: | :--- | :--- |
| - | $[H L+$ byte $]$ |  |

## [Description example]

MOV A, [HL+10H]; When setting byte to 10 H

| Instruction code | 0 |
| ---: | :--- |
| 0 | 1 |
| 0 | 0 |
| 1 | 1 |
| 0 | 0 |
| 0 | 0 |
| 0 | 1 |
| 0 | 0 |

### 5.4.7 Stack addressing

## [Function]

The stack area is indirectly addressed with the stack pointer (SP) contents.
This addressing method is automatically employed when the PUSH, POP, subroutine call, and RETURN instructions are executed or the register is saved/reset upon generation of an interrupt request. Stack addressing can be used to access the internal high-speed RAM area only.

## [Description example]

In the case of PUSH DE


## CHAPTER 6 PORT FUNCTIONS

### 6.1 Port Functions

The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries are provided with the ports shown in Figure 6-1. These ports are used to enable several types of control. Table 6-1 lists the functions of each port.

These ports, while originally designed as digital I/O ports, have alternate functions, as summarized in 3.1 Pin Function List ( $\mu$ PD789167 and 789177 Subseries) and 4.1 Pin Function List ( $\mu$ PD789167Y and 789177Y Subseries).

Figure 6-1. Port Types


Table 6-1. Port Functions

| Pin Name | I/O | Function | After Reset | Alternate Function |
| :---: | :---: | :---: | :---: | :---: |
| P00 to P05 | I/O | Port 0 <br> 6-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P10, P11 | I/O | Port 1 <br> 2-bit I/O port <br> I/O mode can be specified in 1 -bit units. <br> When used as an input port, an on-chip pull-up resistor can be specified by means of pull-up resistor option register 0 (PUO). | Input | - |
| P20 | I/O | Port 2 <br> 7-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> For P20 to P22, P25, and P26, an on-chip pull-up resistor can be specified by means of pull-up resistor option register B2 (PUB2). <br> Only P23 and P24 can be used as N-ch open-drain I/O port pins. | Input | SCK20/ASCK20 |
| P21 |  |  |  | SO20/TxD20 |
| P22 |  |  |  | SI20/RxD20 |
| P23 |  |  |  | SCLO ${ }^{\text {Note }}$ |
| P24 |  |  |  | SDAO ${ }^{\text {Note }}$ |
| P25 |  |  |  | TI80/SS20 |
| P26 |  |  |  | TO80 |
| P30 | I/O | Port 3 <br> 4-bit I/O port <br> I/O mode can be specified in 1-bit units. <br> An on-chip pull-up resistor can be specified by means of pullup resistor option register B3 (PUB3). | Input | INTP0/T181/CPT90 |
| P31 |  |  |  | INTP1/TO81 |
| P32 |  |  |  | INTP2/TO90 |
| P33 |  |  |  | INTP3/TO82/BZO90 |
| P50 to P53 | I/O | Port 5 <br> 4-bit N -ch open-drain I/O port <br> I/O mode can be specified in 1-bit units. <br> For a mask ROM version, an on-chip pull-up resistor can be specified by a mask option. | Input | - |
| P60 to P67 | Input | Port 6 <br> 8-bit input-only port | Input | ANIO to ANI7 |

Note For the $\mu$ PD789167Y and 789177Y Subseries only

### 6.2 Port Configuration

Ports have the following hardware configuration.

Table 6-2. Configuration of Port

| Parameter |  |
| :--- | :--- |
| Control registers | Port mode registers (PMm: $\mathrm{m}=0$ to 3, 5) <br> Pull-up resistor option register 0 (PU0) <br> Pull-up resistor option registers B2, B3 (PUB2, PUB3) |
| Ports | Total: 31 (CMOS I/O: 17, CMOS input: 8, N-ch open-drain I/O: 6) |
| Pull-up resistors | - Mask ROM versions <br> Total: 21 (software control: 17, mask option control: 4) <br> - Flash memory versions <br> Total: 17 (software control only) |

### 6.2.1 Port 0

This is a 6-bit I/O port with output latches. Port 0 can be set to input or output mode in 1-bit units by using port mode register 0 (PMO). When the P00 to P05 pins are used as input port pins, on-chip pull-up resistors can be connected in 6-bit units by using pull-up resistor option register 0 (PUO).
$\overline{\text { RESET }}$ input sets port 0 to input mode.
Figure 6-2 shows a block diagram of port 0 .

Figure 6-2. Block Diagram of P00 to P05


PUO: Pull-up resistor option register 0
PM: Port mode register
RD: Port 0 read signal
WR: Port 0 write signal

### 6.2.2 Port 1

This is a 2-bit I/O port with output latches. Port 1 can be set to input or output mode in 1-bit units by using the port mode register 1 (PM1). When the P10 and P11 pins are used as input port pins, on-chip pull-up resistors can be connected in 2-bit units by using pull-up resistor option register 0 (PUO).

RESET input sets port 1 to input mode.
Figure $6-3$ shows a block diagram of port 1.

Figure 6-3. Block Diagram of P10 and P11


PUO: Pull-up resistor option register 0
PM: Port mode register
RD: Port 1 read signal
WR: Port 1 write signal

### 6.2.3 Port 2

This is a 7-bit I/O port with output latches. Port 2 can be set to input or output mode in 1-bit units by using port mode register 2 (PM2). For the P20 to P22, P25, and P26 pins, on-chip pull-up resistors can be connected in 1-bit units by using pull-up resistor option register B2 (PUB2).

The port is also used as a data I/O and clock I/O to and from the serial interface, and as the timer I/O.
$\overline{\text { RESET input sets port } 2 \text { to input mode. }}$
Figures 6-4 through 6-8 show block diagrams of port 2.

## Caution When using the pins of port 2 as the serial interface, the I/O and output latches must be set

 according to the function to be used. For details of the settings, see Table 14-2 Operating Mode Settings of Serial Interface 20.Figure 6-4. Block Diagram of P20


PUB2: Pull-up resistor option register B2
PM: Port mode register
RD: Port 2 read signal
WR: Port 2 write signal

Figure 6-5. Block Diagram of P21


PUB2: Pull-up resistor option register B2
PM: Port mode register
RD: Port 2 read signal
WR: Port 2 write signal

Figure 6-6. Block Diagram of P22 and P25


PUB2: Pull-up resistor option register B2
PM: Port mode register
RD: Port 2 read signal
WR: Port 2 write signal

Figure 6-7. Block Diagram of P23 and P24


PM: Port mode register
RD: Port 2 read signal
WR: Port 2 write signal

Note This function is provided for the $\mu$ PD789167Y and 789177Y Subseries only. For the $\mu$ PD789167 and 789177 Subseries, P23 and P24 cannot be used as alternate-function pins.

Figure 6-8. Block Diagram of P26


PUB2: Pull-up resistor option register B2
PM: Port mode register
RD: Port 2 read signal
WR: Port 2 write signal

### 6.2.4 Port 3

This is a 4-bit I/O port with output latches. Port 3 can be set to input or output mode in 1-bit units by using port mode register 3 (PM3). For the P30 to P33 pins, on-chip pull-up resistors can be connected in 1-bit units by using pull-up resistor option register B3 (PUB3).

The port is also used as an external interrupt input, capture input, timer output, and buzzer output.
$\overline{\text { RESET }}$ input sets port 3 to input mode.
Figures 6-9 through 6-11 show block diagrams of port 3 .

Figure 6-9. Block Diagram of P30


PUB3: Pull-up resistor option register B3
PM: Port mode register
RD: Port 3 read signal
WR: Port 3 write signal

Figure 6-10. Block Diagram of P31 and P32


PUB3: Pull-up resistor option register B3
PM: Port mode register
RD: Port 3 read signal
WR: Port 3 write signal

Figure 6-11. Block Diagram of P33


PUB3: Pull-up resistor option register B3
PM: Port mode register
RD: Port 3 read signal
WR: Port 3 write signal

### 6.2.5 Port 5

This is a 4-bit N-ch open-drain I/O port with output latches. Port 5 can be set to input or output mode in 1-bit units by using port mode register 5 (PM5). For a mask ROM version, whether a pull-up resistor is to be incorporated can be specified by the mask option.

RESET input sets port 5 to input mode.
Figure 6-12 shows a block diagram of port 5 .

Figure 6-12. Block Diagram of P50 to P53


PM: Port mode register
RD: Port 5 read signal
WR: Port 5 write signal

Caution When using port 5 of the $\mu$ PD78F9177 and 78F9177Y as an input port, be sure to observe the restrictions listed below.

- When Vdd $=1.8$ to 5.5 V

Use within the range of $\mathrm{T}_{A}=25$ to $85^{\circ} \mathrm{C}$

- When $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}$

Use within the range of $V_{D D}=2.7$ to 5.5 V

- When $\mathrm{T}_{\mathrm{A}}=-40$ to $85^{\circ} \mathrm{C}$ and $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V

Issue three consecutive read instructions when reading port 5.

If the above restrictions are not observed, the input value may be read incorrectly.
Note, however, that these restrictions do not apply when port 5 pins are used as output pins, or when the product is other than the $\mu \mathrm{PD} 78 \mathrm{~F} 9177$ or 78F9177Y.

### 6.2.6 Port 6

This is an 8-bit input port.
The port is also used as an analog input to the A/D converter.
Figure 6-13 shows a block diagram of port 6 .

Figure 6-13. Block Diagram of P60 to P67


### 6.3 Port Function Control Registers

The following two types of registers are used to control the ports.

- Port mode registers (PM0 to PM3, and PM5)
- Pull-up resistor option registers (PU0, PUB2, and PUB3)
(1) Port mode registers (PM0 to PM3, and PM5)

The port mode registers separately set each port bit to either input or output.
Each port mode register is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input writes FFH into the port mode registers.
When port pins are used for alternate functions, the corresponding port mode register and output latch must
be set or reset as described in Table 6-3.

Caution When port 3 is acting as an output port and its output level is changed, an interrupt request flag is set, because this port is also used as the input for an external interrupt. To use port 3 in output mode, therefore, the interrupt mask flag must be set to 1 in advance.

Figure 6-14. Format of Port Mode Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | AddressFF2OH | After reset FFH | $\begin{aligned} & \text { R/W } \\ & \text { R/W } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PMO | 1 | 1 | PM05 | PM04 | PM03 | PM02 | PM01 | PM00 |  |  |  |
| PM1 | 1 | 1 | 1 | 1 | 1 | 1 | PM11 | PM10 | FF21H | FFH | R/W |
| PM2 | 1 | PM26 | PM25 | PM24 | PM23 | PM22 | PM21 | PM20 | FF22H | FFH | R/W |
| PM3 | 1 | 1 | 1 | 1 | PM33 | PM32 | PM31 | PM30 | FF23H | FFH | R/W |
| PM5 | 1 | 1 | 1 | 1 | PM53 | PM52 | PM51 | PM50 | FF25H | FFH | R/W |
|  | PMmn | Pmn pin I/O mode selection$\left(\begin{array}{l} m=0: n=0 \text { to } 5, m=1: n=0,1 \\ m=2: n=0 \text { to } 6, m=3: n=0 \text { to } 3 \\ m=5: n=0 \text { to } 3 \end{array}\right)$ |  |  |  |  |  |  |  |  |  |
|  | 0 | Output mode (output buffer on) |  |  |  |  |  |  |  |  |  |
|  | 1 | Input mode (output buffer off) |  |  |  |  |  |  |  |  |  |

Table 6-3. Port Mode Register and Output Latch Settings for Using Alternate Functions

| Pin Name | Alternate Function |  | PM $\times \times$ | Pxx |
| :---: | :---: | :---: | :---: | :---: |
|  | Name | I/O |  |  |
| P25 | TI80 | Input | 1 | $\times$ |
| P26 | TO80 | Output | 0 | 0 |
| P30 | INTP0 | Input | 1 | $\times$ |
|  | TI81 | Input | 1 | $\times$ |
|  | CPT90 | Input | 1 | $\times$ |
| P31 | INTP1 | Input | 1 | $\times$ |
|  | TO81 | Output | 0 | 0 |
| P32 | INTP2 | Input | 1 | $\times$ |
|  | TO90 | Output | 0 | 0 |
| P33 | INTP3 | Input | 1 | $\times$ |
|  | TO82 | Output | 0 | 0 |
|  | BZO90 | Output | 0 | 0 |

Caution When using the pins of port 2 as the serial interface, the I/O or output latch must be set according to the function to be used. For details of the settings, see Table 14-2 Operating Mode Settings of Serial Interface 20.

Remark $\times$ : don't care
PM $\times x$ : Port mode register
$P \times x$ : Port output latch
(2) Pull-up resistor option register 0 (PUO)

Pull-up resistor option register 0 (PUO) sets whether an on-chip pull-up resistor on each port is used. On the port which is specified to use the on-chip pull-up resistor in PUO, the pull-up resistor can be internally used only for the bits set to input mode. No on-chip pull-up resistors can be used for the bits set to output mode regardless of the setting of PUO. On-chip pull-up resistors cannot be used even when the pins are used as the alternate-function output pins.
PU0 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears PU0 to 00 H .

Figure 6-15. Format of Pull-up Resistor Option Register 0

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PU0 | 0 | 0 | 0 | 0 | 0 | 0 | PU01 | PU00 | FFF7H | OOH | R/W |
|  | PU0m | Pm on-chip pull-up resistor selection ( $m=0,1$ ) |  |  |  |  |  |  |  |  |  |
|  | 0 | On-chip pull-up resistor not used |  |  |  |  |  |  |  |  |  |
|  | 1 | On-chip pull-up resistor used |  |  |  |  |  |  |  |  |  |

## Caution Bits $\mathbf{2}$ to $\mathbf{7}$ must all be set to 0 .

## (3) Pull-up resistor option registers B2 and B3 (PUB2 and PUB3)

These registers specify whether an on-chip pull-up resistor is connected to each pin of ports 2 and 3 . The pin specified by PUB2 or PUB3 is connected to on-chip pull-up resistor regardless of the setting of the port mode register.
PUB2 and PUB3 are set with a 1-bit or 8-bit manipulation instruction.
$\overline{\text { RESET }}$ input clears this register to 00 H .

Figure 6-16. Format of Pull-up Resistor Option Register B2

| Symbol | 7 | <6> | <5> | 4 | 3 | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PUB2 | 0 | PUB26 | PUB25 | 0 | 0 | PUB22 | PUB21 | PUB20 | FF32H | OOH | R/W |


| PUB2n | P2n on-chip pull-up resistor selection ( $\mathrm{n}=0$ to 2, 5, 6) |
| :---: | :--- |
| 0 | On-chip pull-up resistor not used |
| 1 | On-chip pull-up resistor used |

## Caution Bits 3, 4, and 7 must all be set to 0 .

Figure 6-17. Format of Pull-up Resistor Option Register B3

| Symbol | 7 | 6 | 5 | 4 | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PUB3 | 0 | 0 | 0 | 0 | PUB33 | PUB32 | PUB31 | PUB30 | FF33H | 00H | R/W |
|  | PUB3n | P3n on-chip pull-up resistor selection ( $\mathrm{n}=0$ to 3) |  |  |  |  |  |  |  |  |  |
|  | 0 | On-chip pull-up resistor not used |  |  |  |  |  |  |  |  |  |
|  | 1 | On-chip pull-up resistor used |  |  |  |  |  |  |  |  |  |

## Caution Bits $\mathbf{4}$ to $\mathbf{7}$ must all be set to 0 .

### 6.4 Operation of Port Functions

The operation of a port differs depending on whether the port is set to input or output mode, as described below.

### 6.4.1 Writing to I/O port

(1) In output mode

A value can be written to the output latch of a port by using a transfer instruction. The contents of the output latch can be output from the pins of the port.
The data once written to the output latch is retained until new data is written to the output latch.
(2) In input mode

A value can be written to the output latch by using a transfer instruction. However, the status of the port pin is not changed because the output buffer is OFF.
The data once written to the output latch is retained until new data is written to the output latch.
Caution A 1-bit memory manipulation instruction is executed to manipulate 1 bit of a port. However, this instruction accesses the port in 8-bit units. When this instruction is executed to manipulate a bit of a port consisting both of inputs and outputs, therefore, the contents of the output latch of the pin that is set to input mode and not subject to manipulation become undefined.

### 6.4.2 Reading from I/O port

(1) In output mode

The contents of the output latch can be read by using a transfer instruction. The contents of the output latch are not changed.
(2) In input mode

The status of a pin can be read by using a transfer instruction. The contents of the output latch are not changed.

### 6.4.3 Arithmetic operation of I/O port

(1) In output mode

An arithmetic operation can be performed with the contents of the output latch. The result of the operation is written to the output latch. The contents of the output latch are output from the port pins.
The data once written to the output latch is retained until new data is written to the output latch.
(2) In input mode

The contents of the output latch become undefined. However, the status of the pin is not changed because the output buffer is OFF.

Caution A 1-bit memory manipulation instruction is executed to manipulate 1 bit of a port. However, this instruction accesses the port in 8-bit units. When this instruction is executed to manipulate a bit of a port consisting both of inputs and outputs, therefore, the contents of the output latch of the pin that is set to input mode and not subject to manipulation become undefined.

## CHAPTER 7 CLOCK GENERATOR

### 7.1 Clock Generator Functions

The clock generator generates the clock to be supplied to the CPU and peripheral hardware. The following two types of system clock oscillators are used.

## - Main system clock oscillator

<Expanded-specification products>
This circuit oscillates at 1.0 to 10.0 MHz . Oscillation can be stopped by executing the STOP instruction or setting the processor clock control register (PCC).
<Conventional products>
This circuit oscillates at 1.0 to 5.0 MHz . Oscillation can be stopped by executing the STOP instruction or setting the processor clock control register (PCC).

- Subsystem clock oscillator

This circuit oscillates at 32.768 kHz . Oscillation can be stopped by setting the suboscillation mode register (SCKM).

### 7.2 Clock Generator Configuration

The clock generator consists of the following items of hardware.

Table 7-1. Configuration of Clock Generator

| Item |  |
| :--- | :--- |
| Control registers | Processor clock control register (PCC) <br> Suboscillation mode register (SCKM) <br> Subclock control register (CSS) |
| Oscillator | Main system clock oscillator <br> Subsystem clock oscillator |

Figure 7-1. Block Diagram of Clock Generator


### 7.3 Registers Controlling Clock Generator

The clock generator is controlled by the following registers.

- Processor clock control register (PCC)
- Suboscillation mode register (SCKM)
- Subclock control register (CSS)


## (1) Processor clock control register (PCC)

PCC selects the CPU clock and the ratio of division.
PCC is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input sets PCC to 02 H .

Figure 7-2. Format of Processor Clock Control Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PCC | MCC | 0 | 0 | 0 | 0 | 0 | PCC1 | 0 | FFFBH | 02H | R/W |
|  | MCC | Control of main system clock oscillator operation |  |  |  |  |  |  |  |  |  |
|  | 0 | Operation enabled |  |  |  |  |  |  |  |  |  |
|  | 1 | Operation disabled |  |  |  |  |  |  |  |  |  |
|  | CSSO | PCC1 | CPU clock (fcpu) selectionNote ${ }^{1}$ |  |  | Minimum instruction execution time: 2/fcpu |  |  |  |  |  |
|  |  |  |  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note } 2}$ <br> or $\mathrm{f}_{\mathrm{xt}}=32.768 \mathrm{kHz}$ operation |  |  | At $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{f}_{\mathrm{XT}}=32.768 \mathrm{kHz}$ operation |  |  |
|  | 0 | 0 | $\mathrm{fx}^{\text {d }}$ |  |  | $0.2 \mu \mathrm{~s}$ |  |  | $0.4 \mu \mathrm{~s}$ |  |  |
|  | 0 | 1 | $\mathrm{fx}_{\mathrm{x}} 2^{2}$ |  |  | $0.8 \mu \mathrm{~s}$ |  |  | $1.6 \mu \mathrm{~s}$ |  |  |
|  | 1 | 0 | $\mathrm{fxT}^{\text {/2 }}$ |  |  | $122 \mu \mathrm{~s}$ |  |  | $122 \mu \mathrm{~s}$ |  |  |
|  | 1 | 1 |  |  |  |  |  |  |  |  |  |

Notes 1. The CPU clock is selected according to a combination of the PCC1 flag in the processor clock control register (PCC) and the CSSO flag in the subclock control register (CSS). See 7.3 (3) Subclock control register (CSS).
2. Expanded-specification products only.

## Cautions 1. Bits 0 and 2 to 6 must all be set to 0 .

2. MCC can be set only when the subsystem clock has been selected as the CPU clock.

Remarks 1. fx: Main system clock oscillation frequency
2. fx : Subsystem clock oscillation frequency
(2) Suboscillation mode register (SCKM)

SCKM specifies whether to use a feedback resistor for the subsystem clock, and controls the oscillation of the clock.
SCKM is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears SCKM to 00 H .

Figure 7-3. Format of Suboscillation Mode Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SCKM | 0 | 0 | 0 | 0 | 0 | 0 | FRC | SCC | FFFOH | 00H | R/W |


| FRC |  | Use of feedback resistorNote |
| :---: | :--- | :--- |
| 0 | On-chip feedback resistor used |  |
| 1 | On-chip feedback resistor not used |  |


| SCC |  | Control of subsystem clock oscillator operation |
| :---: | :--- | :--- |
| 0 | Operation enabled |  |
| 1 | Operation disabled |  |

Note The feedback resistor is necessary to adjust the bias point of the oscillation waveform to close to the mid point of the supply voltage. Only when the subclock is not used, the power consumption in STOP mode can be further reduced by setting $F R C=1$.

## Caution Bits $\mathbf{2}$ to $\mathbf{7}$ must all be set to 0 .

## (3) Subclock control register (CSS)

CSS specifies whether the main system or subsystem clock oscillator is to be used. It also specifies how the CPU clock operates.
CSS is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears CSS to 00 H .

Figure 7-4. Format of Subclock Control Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSS | 0 | 0 | CLS | CSSO | 0 | 0 | 0 | 0 | FFF2H | OOH | R/W ${ }^{\text {Note }}$ |


| CLS | CPU clock operation status |
| :---: | :--- |
| 0 | Operation based on the (divided) main system clock |
| 1 | Operation based on the subsystem clock |


| CSSO | Selection of main system or subsystem clock oscillator |
| :---: | :--- |
| 0 | (Divided) output from the main system clock oscillator |
| 1 | Output form the subsystem clock oscillator |

Note Bit 5 is read-only.

Caution Bits 0 to 3, 6, and 7 must all be set to 0 .

### 7.4 System Clock Oscillators

### 7.4.1 Main system clock oscillator

The main system clock oscillator is oscillated by the crystal or ceramic resonator (5.0 MHz TYP.) connected across the X 1 and X 2 pins.

An external clock can also be input to the circuit. In this case, input the clock signal to the X 1 pin, and input the reversed signal to the X2 pin.

Figure 7-5 shows the external circuit of the main system clock oscillator.

Figure 7-5. External Circuit of Main System Clock Oscillator
(a) Crystal or ceramic oscillation
(b) External clock

or
or
ceramic resonator
ceramic resonator

Caution When using the main system or subsystem clock oscillator, wire in the area enclosed by the broken lines in Figures 7-5 and 7-6 as follows to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines. Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso. Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.


### 7.4.2 Subsystem clock oscillator

The subsystem clock oscillator is oscillated by the crystal resonator ( 32.768 kHz TYP.) connected across the XT1 and XT2 pins.

An external clock can also be input to the circuit. In this case, input the clock signal to the XT1 pin, and input the inverted signal to the XT2 pin.

Figure 7-6 shows the external circuit of the subsystem clock oscillator.

Figure 7-6. External Circuit of Subsystem Clock Oscillator
(a) Crystal oscillation
(b) External clock



Caution When using the main system or subsystem clock oscillator, wire in the area enclosed by the broken lines in Figures 7-5 and 7-6 as follows to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines. Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso. Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

The subsystem clock oscillator is designed as low-amplitude circuit for reducing current consumption. Particular care is therefore required with the wiring method when the subsystem clock is used.

### 7.4.3 Examples of incorrect oscillator connection

Figure 7-7 shows examples of incorrect oscillator connections.
Figure 7-7. Examples of Incorrect Oscillator Connection (1/2)
(a) Wiring too long

(c) Wiring near high alternating current
(b) Crossed signal line

(d) Current flowing through ground line of oscillator (potential at points A, B, and C fluctuates)


Remark When using the subsystem clock, read X1 and X2 as XT1 and XT2, respectively, and connect a resistor to the XT2 pin in series.

Figure 7-7. Examples of Incorrect Oscillator Connection (2/2)
(e) Signals are fetched

(f) Signal conductors of the main and subsystem clocks are parallel and near to each other


Remark When using the subsystem clock, read X1 and X2 as XT1 and XT2, respectively, and connect a resistor to the XT2 pin in series.

Caution If the X1 wire is in parallel with the XT2 wire, crosstalk noise may occur between the X1 and XT2, resulting in a malfunction.
To avoid this, do not lay the X1 and XT2 wires in parallel.

### 7.4.4 Scaler

The scaler divides the main system clock oscillator output ( fx ) and generates clocks.

### 7.4.5 When no subsystem clocks are used

If it is not necessary to use subsystem clocks for low power consumption operations and watch operations, connect the XT1 and XT2 pins as follows.

XT1: Connect to Vsso or Vss1
XT2: Open

In this state, however, some current may leak via the internal feedback resistor of the subsystem clock oscillator when the main system clock stops. To minimize the leakage current, the internal feedback resistor can be removed by setting bit 1 (FRC) of the suboscillation mode register (SCKM). In this case, also connect the XT1 and XT2 pins as described above.

### 7.5 Clock Generator Operation

The clock generator generates the following clocks and controls operation modes of the CPU, such as standby mode.

- Main system clock fx
- Subsystem clock fxt
- CPU clock fcpu
- Clock to peripheral hardware

The operation of the clock generator is determined by the processor clock control register (PCC), suboscillation mode register (SCKM), and subclock control register (CSS), as follows.
(a) The slow mode ( $0.8 \mu \mathrm{~s}$ : at 10.0 MHz operation) of the main system clock is selected when the $\overline{\mathrm{RESET}}$ signal is generated $(P C C=02 H)$. While a low level is input to the $\overline{\text { RESET }}$ pin, oscillation of the main system clock is stopped.
(b) Three types of minimum instruction execution time ( $0.2 \mu \mathrm{~s}$ and $0.8 \mu \mathrm{~s}$ : main system clock (at 10.0 MHz operation), $122 \mu \mathrm{~s}$ : subsystem clock (at 32.768 kHz operation)) can be selected by the PCC, SCKM, and CSS settings.
(c) Two standby modes, STOP and HALT, can be used with the main system clock selected. In a system where no subsystem clock is used, setting bit 1 (FRC) of SCKM so that the built-in feedback resistor cannot be used reduces current drain during STOP mode. In a system where a subsystem clock is used, setting the SCKM bit 0 to 1 can cause the subsystem clock to stop oscillation.
(d) CSS bit 4 (CSSO) can be used to select the subsystem clock so that a low current operation operation is used ( $122 \mu \mathrm{~s}$ : at 32.768 kHz operation).
(e) With the subsystem clock selected, it is possible to cause the main system clock to stop oscillating by using bit 7 (MCC) of PCC. HALT mode can be used, but STOP mode cannot.
(f) The clock for the peripheral hardware is generated by dividing the frequency of the main system clock. The subsystem clock is supplied to 16 -bit timer 90,8 -bit timer 82 , and the watch timer only. So, even in standby mode, 16 -bit timer 90, 8 -bit timer 82, and the watch function can continue operating. The other hardware stops when the main system clock stops, because it operates based on the main system clock (except for an external clock).

### 7.6 Changing Setting of System Clock and CPU Clock

### 7.6.1 Time required for switching between system clock and CPU clock

The CPU clock can be selected by using bit 1 (PCC1) of the processor clock control register (PCC) and bit 4 (CSSO) of the subclock control register (CSS).

Actually, the specified clock is not selected immediately after the setting of PCC has been changed, and the old clock is used for the duration of several instructions after that (see Table 7-2).

Table 7-2. Maximum Time Required for Switching CPU Clock


Remarks 1. Two clocks are the minimum instruction execution time of the CPU clock before switching.
2. The values in paraentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$ or $\mathrm{fxt}=32.768 \mathrm{kHz}$. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{fxt}=32.768 \mathrm{kHz}$.
3. $x$ : don't care

### 7.6.2 Switching between system clock and CPU clock

The following figure illustrates how the CPU clock and system clock switch.

Figure 7-8. Switching Between System Clock and CPU Clock

$<1>$ The CPU is reset when the RESET pin is made low on power application. The effect of resetting is released when the $\overline{\text { RESET }}$ pin is later made high, and the main system clock starts oscillating. At this time, the time during which oscillation stabilizes $\left(2^{15} / \mathrm{fx}\right)$ is automatically secured.
After that, the CPU starts instruction execution at the slow speed of the main system clock ( $0.8 \mu \mathrm{~s}$ : at 10.0 MHz operation).
<2> After the time required for the VdD voltage to rise to the level at which the CPU can operate at the high speed has elapsed, bit 1 (PCC1) of the processor clock control register (PCC) and bit 4 (CSS0) of the subclock control register (CSSO) are rewritten so that the high speed operation can be selected.
$<3>$ When a drop of the Vdd voltage is detected with an interrupt request signal, the clock is switched to the subsystem clock. (At this moment, the subsystem clock must be in the oscillation stabilized status.)
<4> When a recover of the VDD voltage is detected with an interrupt request signal, bit 7 (MCC) of PCC is set to 0 to make the main system clock start oscillating. After the time required for the oscillation to stabilize has elapsed, PCC1 and CSS0 are rewritten so that high-speed operation can be selected again.

## Caution When the main system clock is stopped and the subsystem clock is operating, allow sufficient time for the oscillation to stabilize by coding the program before switching again from the subsystem clock to the main system clock.

## CHAPTER 8 16-BIT TIMER 90

### 8.1 16-Bit Timer 90 Functions

16-bit timer 90 has the following functions.

- Timer interrupt
- Timer output
- Buzzer output
- Count value capture


## (1) Timer interrupt

An interrupt is generated when a count value and compare value matches.
(2) Timer output

Timer output can be controlled when a count value and compare value matches.
(3) Buzzer output

Buzzer output can be controlled by software.
(4) Count value capture

The count value of 16 -bit timer counter 90 (TM90) is latched into the capture register in synchronization with the capture trigger and retained.

### 8.2 16-Bit Timer 90 Configuration

16-bit timer 90 consists of the following hardware.

Table 8-1. Configuration of 16-Bit Timer 90

| Item |  |
| :--- | :--- |
| Timer counter | 16 bits $\times 1$ (TM90) |
| Registers | Compare register: $\quad 16$ bits $\times 1$ (CR90) <br> Capture register: $\quad 16$ bits $\times 1$ (TCP90) |
| Timer outputs | 1 (TO90) |
| Control registers | 16 -bit timer mode control register 90 (TMC90) <br> Buzzer output control register 90 (BZC90) <br> Port mode register 3 (PM3) |

Figure 8-1. Block Diagram of 16-Bit Timer 90


Note See Figure 9-3 Block Diagram of 8-Bit Timer 82.
$\stackrel{\rightharpoonup}{*}$
(1) 16-bit compare register 90 (CR90)

The value specified in CR90 is compared with the count in 16-bit timer register 90 (TM90). If they match, an interrupt request (INTTM90) is issued by CR90.
CR90 is set with an 8-bit or 16-bit memory manipulation instruction. Any value from 0000 H to FFFFH can be set.
$\overline{\text { RESET }}$ input sets CR90 to FFFFH.

Cautions 1. CR90 is designed to be manipulated with a 16-bit memory manipulation instruction. It can also be manipulated with 8-bit memory manipulation instructions, however. When an 8-bit memory manipulation instruction is used to set CR90, it must be accessed using direct addressing.
2. To re-set CR90 during a count operation, it is necessary to disable interrupts in advance, using interrupt mask flag register 1 (MK1). It is also necessary to disable inversion of the timer output data, using 16-bit timer mode control register 90 (TMC90). If the value in CR90 is rewritten in the interrupt-enabled state, an interrupt request may occur at the moment of rewrite.
(2) 16-bit timer counter 90 (TM90)

TM90 is used to count the number of pulses.
The contents of TM90 are read with an 8-bit or 16-bit memory manipulation instruction.
$\overline{R E S E T}$ input clears TM90 to 0000H.

Cautions 1. The count becomes undefined when STOP mode is released, because the count operation is performed during the oscillation stabilization time.
2. TM90 is designed to be manipulated with a 16-bit memory manipulation instruction. It can also be manipulated with 8-bit memory manipulation instructions, however. When an 8-bit memory instruction is used to manipulate TM90, it must be accessed using direct addressing.
3. When an 8-bit memory manipulation instruction is used to manipulate TM90, the lower and higher bytes must be read as a pair, in this order.
(3) 16-bit capture register 90 (TCP90)

TCP90 captures the contents of TM90.
It is set with an 8-bit or 16-bit memory manipulation instruction.
RESET input makes TCP90 undefined.

Caution TCP90 is designed to be manipulated with a 16-bit memory manipulation instruction. It can also be manipulated with 8-bit memory manipulation instructions, however. When an 8-bit memory manipulation instruction is used to manipulate TCP90, it must be accessed using direct addressing.
(4) 16-bit counter read buffer 90

This buffer is used to latch and hold the count for TM90.

### 8.3 Registers Controlling 16-Bit Timer 90

The following three registers control 16-bit timer 90 .

- 16-bit timer mode control register 90 (TMC90)
- Buzzer output control register 90 (BZC90)
- Port mode register 3 (PM3)
(1) 16-bit timer mode control register 90 (TMC90)

16-bit timer mode control register 90 (TMC90) controls the setting of the count clock, capture edge, etc.
TMC90 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears TMC90 to 00H.

Figure 8-2. Format of 16-Bit Timer Mode Control Register 90

| Symbol | 7 | $<6>$ | 5 | 4 | 3 | 2 | 1 | $<0>$ | Address | After reset | R/W |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMC90 | TOD90 | TOF90 | CPT901 | CPT900 | TOC90 | TCL901 | TCL900 | TOE90 |  | FF48H | $00 H$ | R/W |


| TOD90 |  | Timer output data |
| :---: | :--- | :--- |
| 0 | Timer output of 0 |  |
| 1 | Timer output of 1 |  |


| TOF90 |  | Overflow flag control |
| :---: | :--- | :--- |
| 0 | Reset or cleared by software |  |
| 1 | Set when the 16-bit timer overflows |  |


| CPT901 | CPT900 |  |
| :---: | :---: | :--- |
| 0 | 0 | Capture operation disabled |
| 0 | 1 | Captured at the rising edge of the CPT90 pin |
| 1 | 0 | Captured at the falling edge of the CPT90 pin |
| 1 | 1 | Captured at both the rising and falling edges of the CPT90 pin |


| TOC90 | Timer output data inversion control |  |
| :---: | :--- | :--- |
| 0 | Inversion disabled |  |
| 1 | Inversion enabled |  |


| TCL901 | CL900 | 16-bit time counter 90 count clock (fcl) section |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | At $\mathrm{fx}_{\mathrm{x}}=10.0 \mathrm{MHz}^{\text {Note } 2}$ <br> or $\mathrm{fxT}^{\mathrm{A}}=32.768 \mathrm{kHz}$ operation | At $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{f}_{\mathrm{Xt}}=32.768 \mathrm{kHz}$ operation |
| 0 | 0 | $\mathrm{fx} / 2^{2}$ | 2.5 MHz | 1.25 MHz |
| 0 | 1 | $\mathrm{fx} / 2^{6}$ | 156 kHz | 78.1 kHz |
| 1 | 0 | $\mathrm{fx} / 2^{7}$ | 78.1 kHz | 39.1 kHz |
| 1 | 1 | $\mathrm{fxT}^{\text {t }}$ | 32.768 kHz |  |


| TOE90 |  | 16-bit timer counter 90 output control |
| :---: | :--- | :--- |
| 0 | Output disabled (port mode) |  |
| 1 | Output enabled |  |

Notes 1. Bit 7 is read-only.
2. Expanded-specification products only.

## Caution Disable interrupts in advance by using the interrupt mask flag register (MK1) to change the data of TCL901 and TCL900. Also, prevent the timer output data from being inverted by setting TOC90 to 1 .

Remarks 1. fx: Main system clock oscillation frequency
2. fxt: Subsystem clock oscillation frequency

## (2) Buzzer output control register 90 (BZC90)

This register selects the buzzer frequency based on fcl selected with the count clock select bits (TCL901 and TCL900), and controls the output of a square wave.
BZC90 is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears BZC90 to 00H.

Figure 8-3. Format of Buzzer Output Control Register 90

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BZC90 | 0 | 0 | 0 | 0 | BCS902 | BCS901 | BCS900 | BZOE90 | FF49H | OOH | R/W ${ }^{\text {Note }} 1$ |


| BCS902 | BCS901 | BCS900 | Buzzer frequency selection |
| :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | $\mathrm{fc} / 2^{4}$ |
| 0 | 0 | 1 | $\mathrm{fc} / 2^{5}$ |
| 0 | 1 | 0 | $\mathrm{fcl} / 2^{8}$ |
| 0 | 1 | 1 | $\mathrm{fc} / 2^{9}$ |
| 1 | 0 | 0 | $\mathrm{fcl} / 2^{10}$ |
| 1 | 0 | 1 | $\mathrm{fcl} / 2^{11}$ |
| 1 | 1 | 0 | $\mathrm{fc} / 2^{12}$ |
| 1 | 1 | $\mathrm{fcl} / 2^{13}$ |  |
| 1 |  |  |  |


| BZOE90 |  | Buzzer port output control |
| :---: | :--- | :--- |
| 0 | Disable buzzer port output. |  |
| 1 | Enable buzzer port output.Note 2 |  |

Notes 1. Bits 4 to 7 must all be set to 0 .
2. When setting BZOE90 to 1 , TOE82 must be set to 0 . (See Figure 9-6 Format of 8-Bit Timer Mode Control Register 82.)

Caution If the subclock is selected as the count clock (TCL901 = 1, TCL900 = 1: see Figure 8-2 Format of 16-Bit Timer Mode Control Register 90), the subclock is not synchronized when buzzer port output is enabled. In this case, the capture function and TM90 read function are disabled. In addition, the count value of TM90 is undefined.

Table 8-2. Buzzer Frequency of 16-Bit Timer 90

| BCS902 | BCS901 | BCS900 | Buzzer Frequency |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ operation |  |  | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |  |  | At $\mathrm{fxt}^{\mathrm{x}}=$ 32.768 kHz operation$\mathrm{fcl}=\mathrm{fxt}_{\mathrm{x}}$ |
|  |  |  | $\mathrm{fcl}=\mathrm{fx} / 2^{2}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{6}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{7}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{2}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{6}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{7}$ |  |
| 0 | 0 | 0 | 156 kHz | 9.76 kHz | 4.88 kHz | 78.1 kHz | 4.88 kHz | 2.44 kHz | 2.05 kHz |
| 0 | 0 | 1 | 78.1 kHz | 4.88 kHz | 2.44 kHz | 39.1 kHz | 2.44 kHz | 1.22 kHz | 1.02 kHz |
| 0 | 1 | 0 | 9.76 kHz | 610 Hz | 305 Hz | 4.88 kHz | 305 Hz | 152 Hz | 128 Hz |
| 0 | 1 | 1 | 4.88 kHz | 305 Hz | 152 Hz | 2.44 kHz | 152 Hz | 76 Hz | 64 Hz |
| 1 | 0 | 0 | 2.44 kHz | 152 Hz | 76 Hz | 1.22 kHz | 76 Hz | 38 Hz | 32 Hz |
| 1 | 0 | 1 | 1.22 kHz | 76 Hz | 38 Hz | 610 Hz | 38 Hz | 19 Hz | 16 Hz |
| 1 | 1 | 0 | 610 Hz | 38 Hz | 19 Hz | 305 Hz | 19 Hz | 10 Hz | 8 Hz |
| 1 | 1 | 1 | 305 Hz | 19 Hz | 10 Hz | 153 Hz | 10 Hz | 5 Hz | 4 Hz |

Note Expanded-specification products only.

Remarks 1. fx: Main system clock oscillation frequency
2. fxT : Subsystem clock oscillation frequency
(3) Port mode register 3 (PM3)

PM3 is used to set each bit of port 3 to input or output.
When the P32/INTP2/TO90 pin is used for timer output, reset the output latch of P32 and PM32 to 0; when the P33/INTP3/TO82/BZO90 pin is used for buzzer output, ${ }^{\text {Note }}$ reset the output latch of P33 and PM33 to 0. PM3 is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input sets PM3 to FFH.

Note Never output the TO82 and BZO90 signals at the same time.

Figure 8-4. Format of Port Mode Register 3

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PM3 | 1 | 1 | 1 | 1 | PM33 | PM32 | PM31 | PM30 | FF23H | FFH | R/W |


| PM3n |  | P3n pin I/O mode ( $\mathrm{n}=2$ or 3) |
| :---: | :--- | :--- |
| 0 | Output mode (output buffer on) |  |
| 1 | Input mode (output buffer off) |  |

### 8.4 Operation of 16-Bit Timer 90

### 8.4.1 Operation as timer interrupt

16-bit timer 90 can generate interrupts repeatedly each time the free-running counter value reaches the value set to CR90. Since this counter is not cleared and holds the count even after an interrupt is generated, the interval time is equal to one cycle of the count clock set in TCL901 and TCL900.

To operate 16-bit timer 90 as a timer interrupt, the following settings are required.

- Set count values in CR90
- Set 16-bit timer mode control register 90 (TMC90) as shown in Figure 8-5.

Figure 8-5. Settings of 16-Bit Timer Mode Control Register 90 for Timer Interrupt Operation

TOD90 TOF90 CPT901 CPT900 TOC90 TCL901 TCL900 TOE90
TMC90


Setting of count clock (see Table 8-3)

Caution If both the CPT901 and CPT900 flags are set to 0 , the capture edge is disabled.

When the count value of 16 -bit timer counter 90 (TM90) matches the value set in CR90, counting of TM90 continues and an interrupt request signal (INTTM90) is generated.

Table 8-3 shows interval time, and Figure $8-6$ shows timing of the timer interrupt operation.

Caution When rewriting the value in CR90 during a count operation, be sure to execute the following processing.
<1> Set interrupts to disabled (set TMMK90 (bit 4 of interrupt mask flag register 1 (MK1)) to 1).
<2> Disable inversion control of timer output data (set TOC90 to 0)
If the value in CR90 is rewritten in the interrupt-enabled state, an interrupt request may occur at the moment of rewrite.

Table 8-3. Interval Time of 16-Bit Timer 90

| TCL901 | TCL900 | Count Clock |  |  | Interval Time |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{aligned} & \text { At } \mathrm{f}_{\mathrm{x}}=10.0 \mathrm{MHz}^{\text {Note }} \\ & \text { or } \mathrm{fxT}^{=} 32.768 \mathrm{kHz} \\ & \text { operation } \end{aligned}$ | $\begin{aligned} & \text { At } \mathrm{fx}=5.0 \mathrm{MHz} \text { or } \\ & \mathrm{fxT}=32.768 \mathrm{kHz} \\ & \text { operation } \\ & \hline \end{aligned}$ |  | At $f_{x}=10.0 \mathrm{MHz}^{\text {Note }}$ or $\mathrm{fxT}^{\text {a }}=32.768 \mathrm{kHz}$ operation | $\begin{aligned} & \text { At } \mathrm{fx}=5.0 \mathrm{MHz} \text { or } \\ & \mathrm{fxt}=32.768 \mathrm{kHz} \\ & \text { operation } \\ & \hline \end{aligned}$ |
| 0 | 0 | $2^{2} / f x$ | $0.4 \mu \mathrm{~s}$ | $0.8 \mu \mathrm{~s}$ | $2^{18} / \mathrm{fx}$ | 26.2 ms | 52.4 ms |
| 0 | 1 | $2^{6} / \mathrm{fx}$ | $6.4 \mu$ s | $12.8 \mu \mathrm{~s}$ | $2^{22} / \mathrm{fx}$ | 419 ms | 839 ms |
| 1 | 0 | $2^{7} / f x$ | $12.8 \mu \mathrm{~s}$ | $25.6 \mu \mathrm{~s}$ | $2^{23} / \mathrm{fx}$ | 839 ms | 1.68 s |
| 1 | 1 | 1/fxt | $30.5 \mu \mathrm{~s}$ |  | 216/fxt | 2.0 s |  |

Note Expanded-specification products only

Remarks 1. fx: Main system clock oscillation frequency
2. fxt: Subsystem clock oscillation frequency

Figure 8-6. Timing of Timer Interrupt Operation


Remark $\mathrm{N}=0000 \mathrm{H}$ to FFFFH

### 8.4.2 Operation as timer output

16-bit timer 90 can invert the timer output repeatedly each time the free-running counter value reaches the value set to CR90. Since this counter is not cleared and holds the count even after the timer output is inverted, the interval time is equal to one cycle of the count clock set in TCL901 and TCL900.

To operate the 16 -bit timer as a timer output, the following settings are required.

- Set P32 to output mode (PM32 = 0).
- Reset the output latch of P32 to 0.
- Set the count value in CR90.
- Set 16-bit timer mode control register 90 (TMC90) as shown in Figure 8-7.

Figure 8-7. Settings of 16-Bit Timer Mode Control Register 90 for Timer Output Operation


## Caution If both the CPT901 flag and CPT900 flag are set to 0 , the capture edge is disabled.

When the count value of 16-bit timer counter 90 (TM90) matches the value set in CR90, the output status of the TO90/P32/INTP2 pin is inverted. This enables timer output. At that time, the TM90 count is continued and an interrupt request signal (INTTM90) is generated.

Figure 8-8 shows the timing of timer output (see Table 8-3 for the interval time of the 16 -bit timer).

Figure 8-8. Timer Output Timing


Note The TO90 initial value becomes low level during output enable (TOE90 = 1).

Remark $\mathrm{N}=0000 \mathrm{H}$ to FFFFH

### 8.4.3 Capture operation

The capture operation consists of latching the count value of 16-bit timer register 90 (TM90) into a capture register in synchronization with a capture trigger, and retaining the count value.

Set TMC90 as shown in Figure 8-9 to allow the 16-bit timer to start the capture operation.

Figure 8-9. Settings of 16-Bit Timer Mode Control Register 90 for Capture Operation


16-bit capture register 90 (TCP90) starts a capture operation after a CPT90 capture trigger edge is detected, and latches and retains the count value of 16-bit timer register 90 . TCP90 fetches the count value within 2 clocks and retains the count value until the next capture edge detection.

Table 8-4 and Figure 8-10 shows the settings of the capture edge and capture operation timing, respectively.

Table 8-4. Settings of Capture Edge

| CPT901 | CPT900 |  |
| :---: | :---: | :--- |
| 0 | 0 | Capture operation prohibited |
| 0 | 1 | CPT90 pin rising edge |
| 1 | 0 | CPT90 pin falling edge |
| 1 | 1 | CPT90 pin both edges |

Caution Because TCP90 is rewritten when a capture trigger edge is detected during TCP90 read, disable capture trigger edge detection during TCP90 read.

Figure 8-10. Capture Operation Timing (Both Edges of CPT90 Pin Are Specified)


### 8.4.4 16-bit timer counter 90 readout

The count value of 16-bit timer counter 90 (TM90) is read out with a 16-bit manipulation instruction.
TM90 readout is performed via a counter read buffer. The counter read buffer latches the TM90 count value. The buffer operation is held pending at the CPU clock falling edge after the read signal of the TM90 lower byte rises and the count value is retained. The counter read buffer value in the retention state can be read out as the count value.

Cancellation of pending is performed at the CPU clock falling edge after the read signal of the TM90 higher byte falls.

RESET input clears TM90 to 0000H and TM90 starts freerunning.
Figure 8-11 shows the timing of 16-bit timer counter 90 readout.

Cautions 1. The count value after releasing the stop mode becomes undefined because the count operation is executed during the oscillation stabilization time.
2. Though TM90 is designed for a 16-bit transfer instruction, 8-bit transfer instruction can also be used.
When using the 8-bit transfer instruction, execute it using direct addressing.
3. When using the 8-bit transfer instruction, execute in the order from lower byte to higher byte in pairs. If only the lower byte is read, the pending state of the counter read buffer is not canceled, and if only the higher byte is read, an undefined count value is read.

Figure 8-11. 16-Bit Timer Counter 90 Readout Timing


Remark $\mathrm{N}=0000 \mathrm{H}$ to FFFFH

### 8.4.5 Buzzer output operation

The buzzer frequency is set using buzzer output control register 90 (BZC90) based on the count clock selected with TCL901 and TCL900 of TMC90 (source clock). A square wave of the set buzzer frequency is output.

Table 8-5 shows the buzzer frequency.
Set the 16 -bit timer as follows to use it for buzzer output.

- Set P33 to output mode (PM33 = 0).
- Reset output latch of P33 to 0.
- Set a count clock by using TCL901 and TCL900.
- Set BZC90 as shown in Figure 8-12.
- Clear TOE82 of 8-bit timer mode control register 82 (TMC82) to 0 to disable the output of 8 -bit timer 82.

Figure 8-12. Settings of Buzzer Output Control Register 90 for Buzzer Output Operation


Table 8-5. Buzzer Frequency of 16-Bit Timer 90

| BCS902 | BCS901 | BCS900 | Buzzer Frequency |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | At $\mathrm{fx}_{\mathrm{x}}=10.0 \mathrm{MHz}{ }^{\text {Note }}$ operation |  |  | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |  |  | At $\mathrm{f}_{\mathrm{X}}=$ <br> 32.768 kHz <br> operation $\mathrm{fcl}=\mathrm{f}_{\mathrm{XT}}$ |
|  |  |  | $\mathrm{fcl}=\mathrm{fx} / 2^{2}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{6}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{7}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{2}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{6}$ | $\mathrm{fcl}=\mathrm{fx} / 2^{7}$ |  |
| 0 | 0 | 0 | 156 kHz | 9.76 kHz | 4.88 kHz | 78.1 kHz | 4.88 kHz | 2.44 kHz | 2.05 kHz |
| 0 | 0 | 1 | 78.1 kHz | 4.88 kHz | 2.44 kHz | 39.1 kHz | 2.44 kHz | 1.22 kHz | 1.02 kHz |
| 0 | 1 | 0 | 9.76 kHz | 610 Hz | 305 Hz | 4.88 kHz | 305 Hz | 152 Hz | 128 Hz |
| 0 | 1 | 1 | 4.88 kHz | 305 Hz | 152 Hz | 2.44 kHz | 152 Hz | 76 Hz | 64 Hz |
| 1 | 0 | 0 | 2.44 kHz | 152 Hz | 76 Hz | 1.22 kHz | 76 Hz | 38 Hz | 32 Hz |
| 1 | 0 | 1 | 1.22 kHz | 76 Hz | 38 Hz | 610 Hz | 38 Hz | 19 Hz | 16 Hz |
| 1 | 1 | 0 | 610 Hz | 38 Hz | 19 Hz | 305 Hz | 19 Hz | 10 Hz | 8 Hz |
| 1 | 1 | 1 | 305 Hz | 19 Hz | 10 Hz | 153 Hz | 10 Hz | 5 Hz | 4 Hz |

Note Expanded-specification products only

Remarks 1. fx: Main system clock oscillation frequency
2. fxt: Subsystem clock oscillation frequency

### 8.5 Notes on 16-Bit Timer 90

### 8.5.1 Notes on using 16-bit timer 90

Usable functions differ according to the settings of the count clock selection, CPU clock operation, system clock oscillation status, and BZOE90 (bit 0 of buzzer output control register 90 (BZC90)).

Refer to the following table.

| Count <br> Clock | CPU Clock | System Clock |  | BZOE90 | Capture | TM90 Read | Buzzer <br> Output | Timer <br> Output | Timer Interrupt |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Main System Clock | Subsystem Clock |  |  |  |  |  |  |
| $\begin{aligned} & \mathrm{f}_{\mathrm{x} / 2^{2},} \\ & \mathrm{fx}_{2} 2^{6}, \\ & \mathrm{fx}_{2} / 2^{7} \end{aligned}$ | Main | Oscillating | Oscillating/Stopped | 1/0 | $\checkmark$ | $\checkmark^{\text {Note } 1}$ | Note 2 | $\checkmark$ | $\checkmark$ |
|  |  | Stopped |  |  | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
|  | Sub | Oscillating | Oscillating |  | $\checkmark$ | $\times$ | Note 2 | $\checkmark$ | $\checkmark$ |
|  |  | Stopped |  |  | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
| $\mathrm{fx}_{\mathrm{X}}$ | Main | Oscillating | Oscillating | 0 | $\checkmark$ | $\checkmark$ | $\times$ | $\checkmark$ | $\checkmark$ |
|  |  |  |  | 1 | $\times$ | $\times$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
|  |  |  | Stopped | 1/0 | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
|  |  | Stopped (STOP mode) | Oscillating | 0 | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
|  |  |  |  | 1 | $\times$ | $\times$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
|  |  |  | Stopped | 1/0 | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
|  | Sub | Oscillating | Oscillating | 0 | $\checkmark$ | $\checkmark$ | $\times$ | $\checkmark$ | $\checkmark$ |
|  |  |  |  | 1 | $\times$ | $\times$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |
|  |  | Stopped |  | 0 | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
|  |  |  |  | 1 | $\times$ | $\times$ | $\checkmark$ | $\checkmark$ | $\checkmark$ |

Notes 1. TM90 is enabled only when the CPU clock is in high-speed mode.
2. Output is enabled when $\mathrm{BZOE} 90=1$.

Cautions 1. The capture function uses $\mathrm{fx} / 2$ for control (refer to Figure 8-1 Block Diagram of 16-Bit Timer 90). Therefore, the capture function cannot be used when the main system clock is stopped.
2. The read function of TM90 uses the CPU clock for control (refer to Figure 8-1), and reads an undefined value when the CPU clock is slower than the count clock (values are not guaranteed). When reading TM90, set the count clock to the same speed as the CPU clock (when the CPU clock is the main system clock, high-speed mode is set), or select a clock slower than the CPU clock.
3. When the subsystem clock is selected as the count clock and BZOE90 is set to 0 , the subsystem clock selected as the TM90 count clock is one that has been synchronized with the main system clock (refer to Figure 8-1). Therefore, when the main system clock oscillation is stopped, the timer operation is stopped because the clock supplied to 16 -bit timer 90 is stopped (timer interrupt is not generated).
Moreover, when the subsystem clock is selected as the count clock and BZOE90 is set to 1 , the capture and TM90 read values are not guaranteed because the subsystem clock is not synchronized. Therefore, be sure to set BZOE90 to 0 when using the capture and TM90 read functions (when the subsystem clock is selected as the count clock, buzzer output, and the capture and TM90 read functions cannot be used at the same time).

Make the following settings to enable low-current consumption when stopping the main system clock oscillation and releasing the HALT mode.

| Count clock: | Subsystem clock |
| :--- | :--- |
| CPU clock: | Subsystem clock |
| Main system clock: | Oscillation stopped |
| BZOE90: | 1 (Buzzer output enable) |

At this time, when the setting of P33, the buzzer output alternate function pin, is "PM33 $=0, \mathrm{P} 33=0$ ", a square wave of the buzzer frequency is output from P33. When making the above settings, perform either of the following.

- Set P33 to input mode (PM33 = 1)
- If P33 cannot be set input mode, set the port latch value of P33 to $1(P 33=1)$ (in this case a high level is output from P33)


### 8.5.2 Restrictions on rewriting of 16-bit compare register 90

(1) When rewriting the compare register (CR90), be sure to disable interrupts (TMMK90 = 1), and disable inversion control of timer output $(\mathrm{TOC90}=0)$ first.
If CR90 is rewritten with interrupts enabled, an interrupt request may be generated at the moment of rewrite.
(2) The interval time may be double the intended time depending on to the timing at which the compare register (CR90) is rewritten. Likewise, the timer output waveform may be shorter or double the intended output. To avoid this, rewrite using one of the following procedures.
<Preventing method A> Rewriting by 8-bit access
$<1>$ Disable interrupts (TMMK90 = 1), and disable inversion control of timer output (TOC90 = 0)
<2> Rewrite the higher byte of CR90 (16 bits) first
$<3>$ Next, rewrite the lower byte of CR90 (16 bits)
$<4>$ Clear the interrupt request flag (TMIF90)
<5> After more than half the cycle of the count clock has passed from the start of the interrupt, enable timer interrupt and timer output inversion


Note This is because the INTTM90 signal is set to the high level for a period of half the cycle of the count clock after an interrupt is generated, so the output will be inverted if TOC90 is set to 1 during this period.
<Preventing method B> Rewriting by 16-bit access
<1> Disable interrupt (TMMK90 = 1), and disable inversion control of timer output (TOC90 = 0)
<2> Rewrite CR90 (16 bits)
<3> Wait for more than one cycle of the count clock
<4> Clear the interrupt request flag (TMIF90)
<5> Enable timer interrupt and timer output inversion
<Program example B> (When count clock $=64 / \mathrm{fx}$, CPU clock $=\mathrm{fx}$ )
TM90_VCT: SET1 TMMK90; Timer interrupt disable
CLR1 TMC90.3; Timer output inversion disable
MOVW AX, \#xxyyH; CR90 rewrite value setting
MOVW CR90, AX; CR90 rewriting
NOP
NOP
$: \quad\} \quad$ NOP $32(\text { Wait for 64/fx })^{\text {Note }}$
NOP
NOP
CLR1 TMIF90; Interrupt request flag clearing
CLR1 TMMK90; Timer interrupt enable
SET1 TMC90.3; Timer output inversion enable

Note Wait for more than one cycle of the count clock after the CR90 rewriting instruction (MOVW CR90, AX) before clearing the interrupt request flag (TMIF90).

## CHAPTER 9 8-BIT TIMER/EVENT COUNTERS 80 TO 82

### 9.1 Functions of 8-Bit Timer/Event Counters $\mathbf{8 0}$ to $\mathbf{8 2}$

8-bit timer/event counters 80 and 81 and 8-bit timer 82 have the following functions.

- Interval timer (TM80, TM81, TM82)
- External event counter (TM80, TM81 only)
- Square wave output (TM80, TM81, TM82)
- PWM output (TM80, TM81, TM82)


## (1) 8-bit interval timer

When an 8-bit timer/event counter is used as an interval timer, it generates an interrupt at any time interval set in advance.

Table 9-1. Interval Time of 8-Bit Timer/Event Counter 80

| Minimum Interval Time | Maximum Interval Time | Resolution |
| :--- | :--- | :--- |
| $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ |
| $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ | $2^{11} / \mathrm{fx}(204.8 \mu \mathrm{~s})[409.6 \mu \mathrm{~s}]$ | $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses ( ) apply to operation at $f x=10.0 \mathrm{MHz}$ (expandedspecification products only).
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-2. Interval Time of 8-Bit Timer/Event Counter 81

| Minimum Interval Time | Maximum Interval Time | Resolution |
| :--- | :--- | :--- |
| $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})(3.2 \mu \mathrm{~s})$ | $2^{12} / \mathrm{fx}(409.6 \mu \mathrm{~s})[819.2 \mu \mathrm{~s}]$ | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ |
| $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})(51.2 \mu \mathrm{~s})$ | $2^{16} / \mathrm{fx}(6.55 \mathrm{~ms})[13.1 \mathrm{~ms}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ |

Remarks 1. fx : Main system clock oscillation frequency
2. The values in parentheses ( ) apply to operation at $f x=10.0 \mathrm{MHz}$ (expandedspecification products only).
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-3. Interval Time of 8-Bit Timer 82

| Minimum Interval Time | Maximum Interval Time | Resolution |
| :--- | :--- | :--- |
| $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ | $2^{13} / \mathrm{fx}(0.82 \mathrm{~ms})[1.64 \mathrm{~ms}]$ | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ |
| $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ | $2^{15} / \mathrm{fx}(3.27 \mathrm{~ms})[6.55 \mathrm{~ms}]$ | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ |
| $1 / \mathrm{fxT}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ | $2^{8} / \mathrm{fxx}(7.81 \mathrm{~ms})[7.81 \mathrm{~ms}]$ | $1 / \mathrm{fxx}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. fxt: Subsystem clock oscillation frequency
3. The values in parentheses ( ) apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$ or $\mathrm{fxt}=32.768 \mathrm{kHz}$ (expanded-specification products only).
4. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{fx}=32.768 \mathrm{kHz}$.

## (2) External event counter

The number of pulses of an externally input signal can be counted.
(3) Square wave output

A square wave of arbitrary frequency can be output.

Table 9-4. Square Wave Output Range of 8-Bit Timer/Event Counter 80

| Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :--- | :--- | :--- |
| $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ |
| $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ | $2^{11} / \mathrm{fx}(204.8 \mu \mathrm{~s})[409.6 \mu \mathrm{~s}]$ | $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses ( ) apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$. (expandedspecification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-5. Square Wave Output Range of 8-Bit Timer/Event Counter 81

| Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :--- | :--- | :--- |
| $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ | $2^{12} / \mathrm{fx}(409.6 \mu \mathrm{~s})[819.2 \mu \mathrm{~s}]$ | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ |
| $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $2^{16} / \mathrm{fx}(6.55 \mathrm{~ms})[13.1 \mathrm{~ms}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses ( ) apply to operation at $f x=10.0 \mathrm{MHz}$. (expandedspecification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-6. Square Wave Output Range of 8-Bit Timer 82

| Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :--- | :--- | :--- |
| $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ | $2^{13} / \mathrm{fx}(819 \mu \mathrm{~s})[1.64 \mathrm{~ms}]$ | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ |
| $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ | $2^{15} / \mathrm{fx}(3.27 \mathrm{~ms})[6.55 \mathrm{~ms}]$ | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ |
| $1 / \mathrm{fxT}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ | $2^{8} / \mathrm{fxx}(7.81 \mathrm{~ms})[7.81 \mathrm{~ms}]$ | $1 / \mathrm{fxx}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. $f x$ : Subsystem clock oscillation frequency
3. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$ or $\mathrm{fxt}^{\mathrm{t}}=32.768 \mathrm{kHz}$. (expanded-specification products only)
4. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{fxt}=32.768 \mathrm{kHz}$.
(4) PWM output

8-bit resolution PWM output can be produced.

### 9.2 Configuration of 8-Bit Timer/Event Counters 80 to 82

8 -bit timer/event counters 80 to 82 consist of the following hardware.

Table 9-7. Configuration of 8-Bit Timer/Event Counters 80 to 82

| Item |  |
| :--- | :--- |
| Timer counter | 8 bits $\times 3$ (TM80 to TM82) |
| Register | Compare register: 8 bits $\times 3$ (CR80 to CR82) |
| Timer outputs | 3 (TO80 to TO82) |
| Control registers | 8-bit timer mode control register 80 to 82 (TMC80 to TMC82) <br> Port mode register 2, 3 (PM2, PM3) |

Figure 9-1. Block Diagram of 8-Bit Timer/Event Counter 80


Figure 9-2. Block Diagram of 8-Bit Timer/Event Counter 81


Figure 9-3. Block Diagram of 8-Bit Timer 82


## Note See Figure 8-1 Block Diagram of 16-Bit Time 90.

(1) 8-bit compare register 8n (CR8n)

A value specified in CR8n is compared with the count in 8 -bit timer counter 8 n (TM8n). If they match, an interrupt request (INTTM8n) is issued.
CR8n is set with an 8-bit memory manipulation instruction. Any value from 00 H to FFH can be set.
RESET input makes CR8n undefined.

Cautions 1. Before rewriting CR8n, stop the timer operation once. If CR8n is rewritten in the timer operation-enabled state, a match interrupt request signal may occur at the moment of rewrite.
2. Do not clear CR8n to 00 H in PWM output mode (when PWME8n = 1: bit 6 of 8 -bit timer mode control register 8 n (TMC8n)); otherwise, PWM output may not be produced normally.

Remark $\mathrm{n}=0$ to 2
(2) 8-bit timer counter 8n (TM8n)

TM8n is used to count the number of pulses.
Its contents are read with an 8-bit memory manipulation instruction.
RESET input clears TM8n to 00H.

Remark $\mathrm{n}=0$ to 2

### 9.3 8-Bit Timer/Event Counters 80 to 82 Control Registers

The following two types of registers are used to control the 8-bit timer/event counter.

- 8-bit timer mode control registers 80, 81, and 82 (TMC80, TMC81, and TMC82)
- Port mode registers 2 and 3 (PM2 and PM3)
(1) 8-bit timer mode control register 80 (TMC80)

TMC80 determines whether to enable or disable 8-bit timer counter 80 (TM80), specifies the count clock for TM80, and controls the operation of the output controller of 8-bit timer/event counter 80.
TMC80 is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears TMC80 to 00 H .

Figure 9-4. Format of 8-Bit Timer Mode Control Register 80

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMC80 | TCE80 | PWME80 | 0 | 0 | 0 | TCL801 | TCL800 | TOE80 | FF53H | OOH | R/W |


| TCE80 | TM80 operation control |
| :---: | :--- |
| 0 | Operation disabled (TM80 is cleared to 00 H ) |
| 1 | Operation enabled |


| PWME80 |  | PWM output selection |
| :---: | :--- | :--- |
| 0 | Timer counter operation mode |  |
| 1 | PWM output operation mode |  |


| TCL801 | TCL800 | 8 -bit timer counter 80 count clock selection |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}{ }^{\text {Note }}$ operation | At fx $=5.0 \mathrm{MHz}$ operation |
| 0 | 0 | fx | 10.0 MHz | 5.0 MHz |
| 0 | 1 | $\mathrm{fx} / 2^{3}$ | 1.25 MHz | 625 kHz |
| 1 | 0 | Rising edge of TI80 |  |  |
| 1 | 1 | Falling edge of TI80 |  |  |


| TOE80 | 8-bit timer/event counter 80 output control |
| :---: | :--- |
| 0 | Output disabled (port mode) |
| 1 | Output enabled |

Note Expanded-specification products only.

Cautions 1. Always stop the timer before setting TMC80.
2. For PWM mode operation, the interrupt mask flag (TMMK80) must be set.

Remarks fx: Main system clock oscillation frequency
(2) 8-bit timer mode control register 81 (TMC81)

TMC81 determines whether to enable or disable 8-bit timer counter 81 (TM81), specifies the count clock for TM81, and controls the operation of the output controller of 8 -bit timer/event counter 81.
TMC81 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears TMC81 to 00 H .

Figure 9-5. Format of 8-Bit Timer Mode Control Register 81

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMC81 | TCE81 | PWME81 | 0 | 0 | 0 | TCL811 | TCL810 | TOE81 | FF57H | OOH | R/W |


| TCE81 | TM81 operation control |
| :---: | :--- |
| 0 | Operation disabled (TM81 is cleared to 00H) |
| 1 | Operation enabled |


| PWME81 | PWM output selection |
| :---: | :---: |
| 0 |  |
| 1 |  |


| TCL811 | TCL810 |  | 8-bit | election |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | At fx | At fx $=5.0 \mathrm{MHz}$ operation |
| 0 | 0 | $\mathrm{fx} / 2^{4}$ | 625 kHz | 312 kHz |
| 0 | 1 | $\mathrm{fx} / 2^{8}$ | 39.1 kHz | 19.5 kHz |
| 1 | 0 | Rising edge of T181 |  |  |
| 1 | 1 | Falling edge of TI81 |  |  |


| TOE81 | 8-bit timer/event counter 81 output control |
| :---: | :--- |
| 0 | Output disabled (port mode) |
| 1 | Output enabled |

Note Expanded-specification products only.

Cautions 1. Always stop the timer before setting TMC81.
2. For PWM mode operation, the interrupt mask flag (TMMK81) must be set.

Remark fx: Main system clock oscillation frequency
(3) 8-bit timer mode control register 82 (TMC82)

TMC82 determines whether to enable or disable 8-bit timer counter 82 (TM82) and specifies the count clock for TM82. It also controls the operation of the output controller of 8-bit timer 82.
TMC82 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears TMC82 to 00H.

Figure 9-6. Format of 8-Bit Timer Mode Control Register 82

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMC82 | TCE82 | PWME82 | 0 | 0 | 0 | TCL821 | TCL820 | TOE82 | FF5BH | OOH | R/W |


| TCE82 | TM82 operation control |
| :---: | :--- |
| 0 | Operation disabled (TM82 is cleared to 00 H ) |
| 1 | Operation enabled |


| PWME82 |  | PWM output selection |
| :---: | :--- | :--- |
| 0 | Timer counter operation mode |  |
| 1 | PWM output operation mode |  |


| TCL821 | TCL820 | 8 -bit time counter 82 count clock section |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }} 1$ <br> or $f_{x t}=32.768 \mathrm{kHz}$ operation | $\begin{gathered} \text { At } \mathrm{fx}_{\mathrm{x}}=5.0 \mathrm{MHz} \text { or } \\ \mathrm{f}_{\mathrm{x} T}=32.768 \mathrm{kHz} \text { operation } \end{gathered}$ |
| 0 | 0 | $\mathrm{fx}_{\mathrm{x}} 2^{5}$ | 312 kHz | 156 kHz |
| 0 | 1 | $\mathrm{fx} / 2^{7}$ | 78.1 kHz | 39.1 kHz |
| 1 | 0 | $\mathrm{fxT}^{\text {d }}$ | 32.768 kHz |  |
| 1 | 1 | Setting prohibited |  |  |


| TOE82 |  | 8-bit timer 82 output control |
| :---: | :--- | :--- |
| 0 | Output disabled (port mode) |  |
| 1 | Output enabled ${ }^{\text {Note 2 }}$ |  |

Notes 1. Expanded-specification products only.
2. When TOE82 is set to 1 , BZOE90 must be set to 0 (see Figure 8-3 Format of Buzzer Output Control Register 90).

Cautions 1. Always stop the timer before setting TMC82.
2. For PWM mode operation, the interrupt mask flag (TMMK82) must be set.

Remarks 1. fx: Main system clock oscillation frequency
2. fxt: Subsystem clock oscillation frequency

## (4) Port mode registers 2 and 3 (PM2 and PM3)

PM2 and PM3 specify whether each bit of port 2 and port 3 is used for input or output.
To use the P26/TO80 pin for timer output, the PM26 and P26 output latch must be reset to 0 .
To use the P31/TO81/INTP1 pin for timer output, the PM31 and P31 output latch must be reset to 0 .
To use the P33/INTP3/TO82/BZO90 pin for timer output, the PM33 and P33 output latch must be reset to 0 .
PM2 and PM3 are set with a 1-bit or 8-bit memory manipulation instruction.
RESET input sets PM2 and PM3 to FFH.

Figure 9-7. Format of Port Mode Register 2

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PM2 | 1 | PM26 | PM25 | PM24 | PM23 | PM22 | PM21 | PM20 | FF22H | FFH | R/W |
|  | PM26 | P26 pin I/O mode selection |  |  |  |  |  |  |  |  |  |
|  | 0 | Output mode (output buffer on) |  |  |  |  |  |  |  |  |  |
|  | 1 | Input mode (output buffer off) |  |  |  |  |  |  |  |  |  |

Figure 9-8. Format of Port Mode Register 3


### 9.4 Operation of 8-Bit Timer/Event Counters 80 to 82

### 9.4.1 Operation as interval timer

The interval timer repeatedly generates an interrupt at time intervals specified by the count value set in 8 -bit compare register 8n (CR8n) in advance.

To operate 8 -bit timer/event counters 80 to 82 as an interval timer, the following settings are required.
<1> Set 8-bit timer counter 8n (TM8n) to operation disable (by setting TCE8n (bit 7 of 8 -bit timer mode control register 8n (TMC8n)) to 0).
$<2>$ Set the count clock of 8-bit timer/event counters 80 to 82 (see Tables 9-8 to 9-10).
$<3>$ Set a count value in CR8n.
$<4>$ Set TM8n to operation enabled (TCE8n $=1$ ).

When the count value of 8 -bit timer counter 8 n (TM8n) matches the value set in CR8n, TM8n is cleared to 00 H and continues counting. At the same time, an interrupt request signal (INTTM8n) is generated.

Tables 9-8 to $9-10$ show interval time, and Figure 9-9 shows the timing of interval timer operation.
Cautions 1. Before rewriting CR8n, stop the timer operation once. If CR8n is rewritten in the timer operation-enabled state, a match interrupt request signal may occur at the moment of rewrite.
2. If the count clock setting and TM8n operation-enabled are set in TCM8n simultaneously using an 8-bit memory manipulation instruction, an error of more than a clock in one cycle may occur after the timer start. Therefore, always follow the above procedure when operating the 8-bit timer/event counter as an interval timer.

Remark $\mathrm{n}=0$ to 2
Table 9-8. Interval Time of 8-Bit Timer/Event Counter 80

| TCL801 | TCL800 | Minimum Interval Time | Maximum Interval Time | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ |
| 0 | 1 | $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ | $2^{11} / \mathrm{fx} \times(204.8 \mu \mathrm{~s})[409.6 \mu \mathrm{~s}]$ | $2^{3} / \mathrm{fx} \times(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ |
| 1 | 0 | TI80 input cycle | $2^{8} \times$ TI80 input cycle | TI80 input edge cycle |
| 1 | 1 | TI80 input cycle | $2^{8} \times$ TI80 input cycle | TI80 input edge cycle |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$. (expanded-specification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-9. Interval Time of 8-Bit Timer/Event Counter 81

| TCL811 | TCL810 | Minimum Interval Time | Maximum Interval Time | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ | $2^{12} / \mathrm{fx}(409.6 \mu \mathrm{~s})[819.2 \mu \mathrm{~s}]$ | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ |
| 0 | 1 | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $2^{16} / \mathrm{fx}(6.55 \mathrm{~ms})[13.1 \mathrm{~ms}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ |
| 1 | 0 | TI81 input cycle | $2^{8} \times$ TI81 input cycle | TI81 input edge cycle |
| 1 | 1 | TI81 input cycle | $2^{8} \times$ TI81 input cycle | TI81 input edge cycle |

Remarks 1. fx : Main system clock oscillation frequency
2. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$. (expanded-specification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-10. Interval Time of 8-Bit Timer 82

| TCL821 | TCL820 | Minimum Interval Time | Maximum Interval Time | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ | $2^{13} / \mathrm{fx}(819 \mu \mathrm{~s})[1.64 \mathrm{~ms}]$ | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ |
| 0 | 1 | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ | $2^{15} / \mathrm{fx}(3.27 \mathrm{~ms})[6.55 \mathrm{~ms}]$ | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ |
| 1 | 0 | $1 / \mathrm{fxT}^{(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]}$ | $2^{8} / \mathrm{fxT}(7.81 \mathrm{~ms})[7.81 \mathrm{~ms}]$ | $1 / \mathrm{fxT}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ |
| 1 | 1 | Setting prohibited |  |  |

Remarks 1. fx: Main system clock oscillation frequency
2. $f \times \mathrm{x}$ : Subsystem clock oscillation frequency
3. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$ or $\mathrm{fxt}=32.768 \mathrm{kHz}$ (expandedspecification products only).
4. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{fx}=32.768 \mathrm{kHz}$.

Figure 9-9. Interval Timer Operation Timing


Remarks 1. Interval time $=(N+1) \times t: N=00 H$ to $F F H$
2. $\mathrm{n}=0$ to 2

### 9.4.2 Operation as external event counter ${ }^{\text {Note }}$

The external event counter counts the number of external clock pulses input to the TI80/P25/ $\overline{\mathrm{SS} 20}$ or TI81/P30/INTP0/CPT90 pin by using 8-bit timer counters 80 or 81 (TM80 or TM81).

To operate 8-bit timer/event counter 8 n as an external event counter, the following settings are required.
<1> Set P25 or P30 to input mode (PM25 = 1, PM30 = 1).
<2> Set 8-bit timer register 8n (TM8n) to operation disabled (by setting TCE8n (bit 7 of 8 -bit timer mode control register 8n (TMC8n)) to 0).
<3> Specify the rising/falling edges of TI8n (see Tables 9-4 and 9-5).
$<4>$ Set a count value in CR8n.
$<5>$ Set TM8n to operation enabled (TCE8n $=1$ ).

Note Only TM80 and TM81 have this function.

Each time the valid edge specified by bit 1 (TCL8n0) of TMC8n is input, the value TM8n is incremented.
When the count value of TM8n matches the value set in CR8n, TM8n is cleared to 0 and continues counting. At the same time, an interrupt request signal (INTTM8n) is generated.

Figure $9-10$ shows the timing of the external event counter operation (with rising edge specified).

Cautions 1. Before rewriting CR8n, stop the timer operation once. If CR8n is rewritten in the timer operation-enabled state, a match interrupt request signal may occur at the moment of rewrite.
2. If the count clock setting and TM8n operation-enabled are set in TCM8n simultaneously using an 8-bit memory manipulation instruction, an error of more than a clock in one cycle may occur after the timer start. Therefore, always follow the above procedure when operating the 8-bit timer/event counter as an external event counter.

Remark $\mathrm{n}=0,1$

Figure 9-10. External Event Counter Operation Timing (with Rising Edge Specified)


Remarks 1. $\mathrm{N}=00 \mathrm{H}$ to FFH
2. $\mathrm{n}=0,1$

### 9.4.3 Operation as square wave output

The 8-bit timer/event counter can generate output square waves of an arbitrary frequency at intervals specified by the count value set in 8 -bit compare registers 8 n (CR8n) in advance.

To operate 8 -bit timer/event counters 8 for square wave output, the following settings are required.
$<1>$ Set P26, P31, or P33 to output mode (PM26 = 0, PM31 = 0, PM33 = 0).
<2> Reset the output latches of P26, P31, or P33 to 0.
<3> Set 8-bit timer counter 8n (TM8n) to operation disable (by setting TCE8n (bit 7 of 8-bit timer mode control register 8 n (TMC8n)) to 1).
<4> Set the count clock of 8-bit timer/event counter 8n and set TO8n to output enable (TOE8n (bit 0 of TMC8n) = 1).
$<5>$ Set count value in CR8n.
$<6>$ Set TM8n to operation enable $($ TCE8 $n=1)$.

When the count value of TM8n matches the value set in CR8n, the TO8n pin output will be inverted. Through application of this mechanism, square waves of any frequency can be output. As soon as a match occurs, TM8n will be cleared to 00 H and resumes to count, generating an interrupt request signal (INTTM8n).

Setting 0 for bit 7 (TCE8n) of TMC8n clears the square-wave output to 0.
Tables 9-11 through 9-13 show square wave output range, and Figure $9-11$ shows timing of square wave output.

Cautions 1. Before rewriting CR8n, stop the timer operation once. If CR8n is rewritten in the timer operation-enabled state, a match interrupt request signal may occur at the moment of rewrite.
2. If the count clock setting and TM8n operation-enabled are set in TCM8n simultaneously using an 8-bit memory manipulation instruction, an error of more than a clock in one cycle may occur after the timer start. Therefore, always follow the above procedure when operating the 8 -bit timer/event counter for square wave output.

Remark $\mathrm{n}=0$ to 2

Table 9-11. Square Wave Output Range of 8-Bit Timer/Event Counter 80

| TCL801 | TCL800 | Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $1 / \mathrm{fx}(100 \mathrm{~ns})[200 \mathrm{~ns}]$ |
| 0 | 1 | $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ | $2^{11} / \mathrm{fx}(204.8 \mu \mathrm{~s})[409.6 \mu \mathrm{~s}]$ | $2^{3} / \mathrm{fx}(0.8 \mu \mathrm{~s})[1.6 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$. (expanded-specification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-12. Square Wave Output Range of 8-Bit Timer/Event Counter 81

| TCL811 | TCL810 | Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ | $2^{12} / \mathrm{fx}(409.6 \mu \mathrm{~s})[819.2 \mu \mathrm{~s}]$ | $2^{4} / \mathrm{fx}(1.6 \mu \mathrm{~s})[3.2 \mu \mathrm{~s}]$ |
| 0 | 1 | $2^{8} / \mathrm{fx}(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ | $2^{16} / \mathrm{fx}(6.55 \mathrm{~ms})[13.1 \mathrm{~ms}]$ | $2^{8} / \mathrm{fx} \times(25.6 \mu \mathrm{~s})[51.2 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. The values in parentheses () apply to operation at $\mathrm{fx}=10.0 \mathrm{MHz}$. (expanded-specification products only)
3. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$.

Table 9-13. Square Wave Output Range of 8-Bit Timer 82

| TCL821 | TCL820 | Minimum Pulse Width | Maximum Pulse Width | Resolution |
| :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ | $2^{13} / \mathrm{fx}(819 \mu \mathrm{~s})[1.64 \mathrm{~ms}]$ | $2^{5} / \mathrm{fx}(3.2 \mu \mathrm{~s})[6.4 \mu \mathrm{~s}]$ |
| 0 | 1 | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ | $2^{15} / \mathrm{fx}(3.27 \mathrm{~ms})[6.55 \mathrm{~ms}]$ | $2^{7} / \mathrm{fx}(12.8 \mu \mathrm{~s})[25.6 \mu \mathrm{~s}]$ |
| 1 | 0 | $1 / \mathrm{fxx}^{(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]}$ | $2^{8} / \mathrm{fxT}(7.81 \mathrm{~ms})[7.81 \mathrm{~ms}]$ | $1 / \mathrm{fxx}(30.5 \mu \mathrm{~s})[30.5 \mu \mathrm{~s}]$ |

Remarks 1. fx: Main system clock oscillation frequency
2. $f \times \mathrm{t}$ : Subsystem clock oscillation frequency
3. The values in parentheses () apply to operation at $f x=10.0 \mathrm{MHz}$ or $\mathrm{fxT}=32.768 \mathrm{kHz}$. (expandedspecification products only)
4. The values in brackets [ ] apply to operation at $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{fx}=32.768 \mathrm{kHz}$.

Figure 9-11. Square Wave Output Timing


Note The initial value of TO8 $n$ is low for output enable $($ TOE8 $n=1)$.

Remark $\mathrm{n}=0$ to 2

### 9.4.4 PWM output operation

PWM output enables generation of an interrupt repeatedly at intervals specified by the count value set in 8 -bit compare register 8n (CR8n) in advance.

To use 8-bit timer/event counter 8n for PWM output, the following settings are required.
$<1>$ Set P26, P31, or P33 to output mode ( $\mathrm{PM} 26=0, \mathrm{PM} 31=0, \mathrm{PM} 33=0$ ).
<2> Reset the output latches of P26, P31, or P33 to 0.
$<3>$ Set 8 -bit timer counter 8 n (TM8n) to operation disable (by setting TCE8n (bit 7 of 8 -bit timer mode control register 8n (TMC8n)) to 0).
<4> Set the count clock of 8-bit timer/event counter 8n, and set TO8n to output enable (TOE8n (bit 0 of TMC8n) $=1$ ), and to PWM output mode (PWME8n = 1).
<5> Set a count value in CR8n.
$<6>$ Set TM8n to operation enable $($ TCE8 $n=1)$.

When the count value of TM8n matches the value set in CR8n, TM8n continues counting, and an interrupt request signal (INTTM8n) is generated.

Cautions 1. Before rewriting CR8n, stop the timer. If CR8n is rewritten in the timer operation-enabled state, a high-level signal may be output for the next cycle (256 count pulses) (for details, see 9.5 (4) Timer operation after compare register is rewritten during PWM output).
2. If the count clock setting and TM8n operation-enabled are set in TCM8n simultaneously using an 8-bit memory manipulation instruction, an error of more than a clock in one cycle may occur after the timer start. Therefore, always follow the above procedure when operating the 8-bit timer/event counter for PWM output.

Remark $\mathrm{n}=0$ to 2

Figure 9-12. PWM Output Timing

$\mathrm{M}=01 \mathrm{H}$ to FFH

Note The initial value of TO8n is low for output enable (TOE8n $=1$ ).

Caution Do not set CR8n to 00H in PWM output mode; otherwise, PWM may not be output normally.

Remark $\mathrm{n}=0$ to 2

### 9.5 Notes on Using 8-Bit Timer/Event Counters 80 to 82

## (1) Error on starting timer

An error of up to 1.5 clocks is included in the time between when the timer is started and when a match signal is generated. This is because the rising edge is detected and the counter is incremented if the timer is started while the selected clock is high (see Figure 9-13).

Figure 9-13. Case of Error Occurrence of up to 1.5 Clocks


Remark $\mathrm{n}=0$ to 2

## (2) Count value if external clock input from TI8n pin is selected

When the rising edge of the external clock signal input from the TI8n pin is selected as the count clock, the count value may start from 01H if the timer is enabled (TCE8n $=0 \rightarrow 1$ ) while the TI8n pin is high. This is because the input signal of the TI8n pin is internally ANDed with the TCE8n signal. Consequently, the counter is incremented because the rising edge of the count clock is input to the timer immediately when the TCE8n pin is set. Depending on the delay timing, the count value is incremented by one if the rising edge is input after the counter is cleared. Counting is not affected if the rising edge is input before the counter is cleared (the counter operates normally).
By the same factor as the above, when the falling edge of the external clock signal input from the TI8n pin is selected as the count clock, the count value may start from 01 H if the timer is enabled (TCE8n =0 $0 \rightarrow 1$ ) while the TI8n pin is low.
Use the timer being aware that it has an error of one count, or take either of the following actions A or B.
<Action A> Always start the timer while the TI8n pin is low if the rising edge is selected.
Always start the timer while the TI8n pin is high if the falling edge is selected
<Action B> Save the count value to a control register when the timer is started, SUB the count value with the count value saved to the control register when reading the count value, and take the result of SUB as the true count value.

Figure 9-14. Counting Operation if Timer Is Started When TI8n Is High (When the rising edge is selected)


H
(3) Setting of 8-bit compare register 8n

8 -bit compare register 8 n (CR8n) can be set to 00 H .
Therefore, one pulse can be counted when an 8-bit timer/event counter operates as an event counter.

Figure 9-15. External Event Counter Operation Timing


Interrupt request flag


Cautions 1. When CR8n is rewritten to timer counter operation mode (PWME8n (8-bit timer mode control register $8 \mathrm{n}(\mathrm{TMC8n}))=0$ ), be sure to stop the timer operation beforehand. If CR8n is rewritten in the timer operation-enabled state, a match interrupt request signal may occur at the moment of rewrite.
2. If CR8n is rewritten during timer operation in the PWM operation mode (PWME8n =1), pulses may not be generated for one cycle after the rewrite.
3. Do not set CR8n to 00H in PWM operation mode; otherwise, PWM may not be output normally.

Remark $\mathrm{n}=0$ to 2
(4) Timer operation after compare register is rewritten during PWM output

When 8 -bit compare register 8 n (CR8n) is rewritten during PWM output, if the new value is smaller than that of 8 -bit timer/counter 8 n (TM8n), a high-level signal may be output for the next cycle ( 256 count pulses) after the CR8n value is rewritten. Figure $9-16$ shows the timing at which the high-level signal is output.

Figure 9-16. Operation Timing After Compare Register Is Rewritten During PWM Output


$$
\mathrm{M}=01 \mathrm{H} \text { to FFH }
$$

Remark $\mathrm{n}=0$ to 2
(5) Cautions when STOP mode is set

Be sure to stop timer operations $(T C E 8 n=0)$ before executing the STOP instruction.

## CHAPTER 10 WATCH TIMER

### 10.1 Watch Timer Functions

The watch timer has the following functions.

- Watch timer
- Interval timer

The watch and interval timers can be used at the same time.
Figure $10-1$ is a block diagram of the watch timer.

Figure 10-1. Block Diagram of Watch Timer

(1) Watch timer

The 4.19 MHz main system clock or 32.768 kHz subsystem clock is used to issue an interrupt request (INTWT) at 0.5 -second intervals.

Caution When the main system clock is operating at 5.0 MHz , it cannot be used to generate a 0.5second interval. In this case, the subsystem clock, which operates at $32.768 \mathbf{~ k H z}$, should be used instead.
(2) Interval timer

The interval timer is used to generate an interrupt request (INTWTI) at specified intervals.

Table 10-1. Interval Generated Using Interval Timer

| Interval | At $\mathrm{fx}=10.0 \mathrm{MHz}{ }^{\text {Note }}$ | At fx $=5.0 \mathrm{MHz}$ | At $\mathrm{fx}=4.19 \mathrm{MHz}$ | At $\mathrm{fxT}=32.768 \mathrm{kHz}$ |
| :--- | :--- | :--- | :--- | :--- |
| $2^{4} \times 1 / \mathrm{fw}$ | $204 \mu \mathrm{~s}$ | $409 \mu \mathrm{~s}$ | $489 \mu \mathrm{~s}$ | $488 \mu \mathrm{~s}$ |
| $2^{5} \times 1 / \mathrm{fw}$ | $409 \mu \mathrm{~s}$ | $819 \mu \mathrm{~s}$ | $978 \mu \mathrm{~s}$ | $977 \mu \mathrm{~s}$ |
| $2^{6} \times 1 / \mathrm{fw}$ | $819 \mu \mathrm{~s}$ | 1.64 ms | 1.96 ms | 1.95 ms |
| $2^{7} \times 1 / \mathrm{fw}$ | 1.64 ms | 3.28 ms | 3.91 ms | 3.91 ms |
| $2^{8} \times 1 / \mathrm{fw}$ | 3.28 ms | 6.55 ms | 7.82 ms | 7.81 ms |
| $2^{9} \times 1 / \mathrm{fw}$ | 6.55 ms | 13.1 ms | 15.6 ms | 15.6 ms |

Note Expanded-specification products only.

Remarks 1. fw: Watch timer clock frequency ( $\mathrm{f} \times / \mathrm{L}^{7}$ or fxT )
2. fx : Main system clock oscillation frequency
3. fxt: Subsystem clock oscillation frequency

### 10.2 Watch Timer Configuration

The watch timer consists of the following hardware.

Table 10-2. Watch Timer Configuration

| Item |  |
| :--- | :--- |
| Counter | 5 bits $\times 1$ |
| Prescaler | 9 bits $\times 1$ |
| Control register | Watch timer mode control register (WTM) |

### 10.3 Watch Timer Control Register

The watch timer mode control register (WTM) is used to control the watch timer.

- Watch timer mode control register (WTM) WTM selects a count clock for the watch timer and specifies whether to enable operation of the timer. It also specifies the prescaler interval and how the 5-bit counter is controlled.
WTM is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input sets WTM to 00 H .

Figure 10-2. Format of Watch Timer Mode Control Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WTM | WTM7 | WTM6 | WTM5 | WTM4 | 0 | 0 | WTM1 | WTM0 | FF4AH | 00H | R/W |


| WTW7 | Watch timer count clock (fw) section |  |  |
| :---: | :---: | :---: | :---: |
|  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ <br> or $\mathrm{f}_{\mathrm{xt}}=32.768 \mathrm{kHz}$ operation | At $\mathrm{fx}=5.0 \mathrm{MHz}$ or $\mathrm{f}_{\mathrm{Xt}}=32.768 \mathrm{kHz}$ operation |
| 0 | $\mathrm{fx} / 2^{7}$ | 78.2 kHz | 39.1 kHz |
| 1 | $\mathrm{fx}^{\text {t }}$ | 32.768 kHz |  |


| WTM6 | WTM5 | WTM4 | Prescaler interval selection |  |  |
| :---: | :---: | :---: | :--- | :---: | :---: |
| 0 | 0 | 0 | $2^{4} / \mathrm{fw}(488 \mu \mathrm{~s})$ |  |  |
| 0 | 0 | 1 | $2^{5 / f w}(977 \mu \mathrm{~s})$ |  |  |
| 0 | 1 | 0 | $2^{6 / f w}(1.95 \mathrm{~ms})$ |  |  |
| 0 | 1 | 1 | $2^{7 / f w}(3.91 \mathrm{~ms})$ |  |  |
| 1 | 0 | 0 | $2^{8 / f w}(7.81 \mathrm{~ms})$ |  |  |
| 1 | 0 | 1 | $2^{9 / f w}(15.6 \mathrm{~ms})$ |  |  |
| Other than above |  |  |  |  | Setting prohibited |


| WTM1 | Control of 5-bit counter operation |  |
| :---: | :--- | :--- |
| 0 | Cleared after stop |  |
| 1 | Started |  |


| WTM0 | Watch timer operation |
| :---: | :--- |
| 0 | Operation disabled (both prescaler and timer cleared) |
| 1 | Operation enabled |

Note Expanded-specification products only.
Remarks 1. fw: Watch timer clock frequency ( $\mathrm{fx} / \mathrm{L}^{7}$ or fxT )
2. fx: Main system clock oscillation frequency
3. fxt: Subsystem clock oscillation frequency
4. The values in parentheses apply to operation at $\mathrm{f} w=32.768 \mathrm{kHz}$.

### 10.4 Watch Timer Operation

### 10.4.1 Operation as watch timer

The main system clock ( 4.19 MHz ) or subsystem clock ( 32.768 kHz ) is used to enable the watch timer to operate at 0.5 -second intervals.

The watch timer is used to generate an interrupt request at specified intervals.
By setting bits 0 and 1 (WTM0 and WTM1) of the watch timer mode control register (WTM) to 1, the watch timer starts counting. By setting them to 0 , the 5 -bit counter is cleared and the watch timer stops counting.

Only the watch timer can be started form zero seconds by clearing WTM1 to 0 when the interval timer and watch timer operate at the same time. In this case, however, an error of up to $2^{9} \times 1 / \mathrm{fw}$ seconds may occur in the overflow (INTWT) after the zero-second start of the watch timer because the 9 -bit prescaler is not cleared to 0 .

### 10.4.2 Operation as interval timer

The interval timer is used to repeatedly generate an interrupt request at the interval specified by a count value set in advance.

The interval can be selected by bits 4 to 6 (WTM4 to WTM6) of the watch timer mode control register (WTM).

Table 10-3. Interval Generated Using Interval Timer

| WTM6 | WTM5 | WTM4 | Interval | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ | At $\mathrm{fx}=4.19 \mathrm{MHz}$ | At $\mathrm{fxT}=32.768$ <br> kHz |
| :---: | :---: | :---: | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | $2^{4} \times 1 / \mathrm{fw}$ | $204 \mu \mathrm{~s}$ | $409 \mu \mathrm{~s}$ | $489 \mu \mathrm{~s}$ | $488 \mu \mathrm{~s}$ |
| 0 | 0 | 1 | $2^{5} \times 1 / \mathrm{fw}$ | $409 \mu \mathrm{~s}$ | $819 \mu \mathrm{~s}$ | $978 \mu \mathrm{~s}$ | $977 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $2^{6} \times 1 / \mathrm{fw}$ | $819 \mu \mathrm{~s}$ | 1.64 ms | 1.96 ms | 1.95 ms |
| 0 | 1 | 1 | $2^{7} \times 1 / \mathrm{fw}$ | 1.64 ms | 3.28 ms | 3.91 ms | 3.91 ms |
| 1 | 0 | 0 | $2^{8} \times 1 / \mathrm{fw}$ | 3.27 ms | 6.55 ms | 7.82 ms | 7.81 ms |
| 1 | 0 | 1 | $2^{9} \times 1 / \mathrm{fw}$ | 6.55 ms | 13.1 ms | 15.6 ms | 15.6 ms |
| Other than above |  |  |  |  |  |  | Setting prohibited |

Note Expanded-specification products only.

Remarks 1. fx: Main system clock oscillation frequency
2. $\mathrm{fxt}_{\mathrm{T}}$ : Subsystem clock oscillation frequency
3. fw: Watch timer clock frequency

Figure 10-3. Watch Timer/Interval Timer Operation Timing


Caution When operation of the watch timer and 5-bit counter operation is enabled by setting bit 0 (WTMO) of the watch timer mode control register (WTM) to 1 , the interval until the first interrupt request (INTWT) is generated after the register is set does not exactly match the watch timer interrupt time ( 0.5 s ). This is because there is a delay of one 9 -bit prescaler output cycle until the 5 -bit counter starts counting. Subsequently, however, the INTWT signal is generated at the specified intervals.

Remarks 1. fw: Watch timer clock frequency
2. The values in parentheses apply to operation at $f w=32.768 \mathrm{kHz}$.

## CHAPTER 11 WATCHDOG TIMER

### 11.1 Watchdog Timer Functions

The watchdog timer has the following functions.

- Watchdog timer
- Interval timer


## Caution Select the watchdog timer mode or interval timer mode by using the watchdog timer mode

 register (WDTM).
## (1) Watchdog timer

The watchdog timer is used to detect inadvertent program loops. When an inadvertent loop is detected, a non-maskable interrupt or a $\overline{\text { RESET }}$ signal can be generated.

Table 11-1. Inadvertent Loop Detection Time of Watchdog Timer

| Inadvertent Loop Detection <br> Time | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ |
| :--- | :--- | :--- | At fx $=5.0 \mathrm{MHz}$

Note Expanded-specification products only.

Remark fx: Main system clock oscillation frequency
(2) Interval timer

The interval timer generates an interrupt at an arbitrary interval set in advance.

Table 11-2. Interval Time

| Interval | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ | At fx $=5.0 \mathrm{MHz}$ |
| :--- | :--- | :--- |
| $2^{11} \times 1 / \mathrm{fx}$ | $205 \mu \mathrm{~s}$ | $410 \mu \mathrm{~s}$ |
| $2^{13} \times 1 / \mathrm{fx}$ | $819 \mu \mathrm{~s}$ | 1.64 ms |
| $2^{15} \times 1 / \mathrm{fx}$ | 3.27 ms | 6.55 ms |
| $2^{17} \times 1 / \mathrm{fx}$ | 13.1 ms | 26.2 ms |

Note Expanded-specification products only.

Remark fx: Main system clock oscillation frequency

### 11.2 Watchdog Timer Configuration

The watchdog timer consists of the following hardware.

Table 11-3. Configuration of Watchdog Timer

| Item |  | Configuration |
| :---: | :--- | :--- |
| Control registers | Timer clock selection register 2 (TCL2) <br> Watchdog timer mode register (WDTM) |  |

Figure 11-1. Block Diagram of Watchdog Timer


### 11.3 Watchdog Timer Control Registers

The following two types of registers are used to control the watchdog timer.

- Timer clock selection register 2 (TCL2)
- Watchdog timer mode register (WDTM)


## (1) Timer clock selection register 2 (TCL2)

This register sets the watchdog timer count clock.
TCL2 is set with an 8-bit memory manipulation instruction.
RESET input clears TCL2 to 00 H .

Figure 11-2. Format of Timer Clock Selection Register 2

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | Address | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TCL2 | 0 | 0 | 0 | 0 | 0 | TCL22 | TCL21 | TCL20 | FF42H | OOH | R/W |


| TCL22 | TCL21 | TCL20 |  | chdog timer coun | clock selection |  | Interv |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | $\begin{gathered} \text { At } \mathrm{fx}=10.0 \\ \mathrm{MHz}^{\text {Note }} \end{gathered}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ |  | $\begin{gathered} \text { At } \mathrm{fx}=10.0 \\ \mathrm{MHz}^{\text {Note }} \end{gathered}$ | At fx $=5.0 \mathrm{MHz}$ |
| 0 | 0 | 0 | $\mathrm{fx} / 2^{4}$ | 625.0 kHz | 312.5 kHz | $2^{11} / \mathrm{fx}$ | $205 \mu \mathrm{~s}$ | $410 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $\mathrm{fx} / 2^{6}$ | 156.2 kHz | 78.1 kHz | $2^{13} / \mathrm{fx}$ | $819 \mu \mathrm{~s}$ | 1.64 ms |
| 1 | 0 | 0 | $\mathrm{fx} / 2^{8}$ | 39.0 kHz | 19.5 kHz | $2^{15} / \mathrm{fx}$ | 3.27 ms | 6.55 ms |
| 1 | 1 | 0 | $\mathrm{fx} / 2^{10}$ | 9.76 kHz | 4.88 kHz | $2^{17} / f x$ | 13.1 ms | 26.2 ms |
| Other than above |  |  | Setting prohibited |  |  |  |  |  |

Note Expanded-specification products only.

Remark fx: Main system clock oscillation frequency
(2) Watchdog timer mode register (WDTM)

This register sets an operation mode of the watchdog timer, and enables/disables counting of the watchdog timer.
WDTM is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears WDTM to 00 H .

Figure 11-3. Format of Watchdog Timer Mode Register

| Symbol | <7> | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WDTM | RUN | 0 | 0 | WDTM4 | WDTM3 | 0 | 0 | 0 | FFF9H | 00 H | R/W |


| RUN |  |
| :---: | :--- |
| 0 | Stop counting |
| 1 | Watchdog timer operation selection ${ }^{\text {Note } 1}$ |


| WDTM4 | WDTM3 | Watchdog timer operation mode selection ${ }^{\text {Note } 2}$ |
| :---: | :---: | :--- |
| 0 | 0 | Operation stop |
| 0 | 1 | Interval timer mode (generates a maskable interrupt upon overflow occurrence) ${ }^{\text {Note } 3}$ |
| 1 | 0 | Watchdog timer mode 1 (generates a non-maskable interrupt upon overflow occurrence) |
| 1 | 1 | Watchdog timer mode 2 (starts reset operation upon overflow occurrence.) |

Notes 1. Once RUN has been set to 1, it cannot be cleared to 0 by software. Therefore, when counting is started, it cannot be stopped by any means other than RESET input.
2. Once WDTM3 and WDTM4 have been set to 1 , they cannot be cleared to 0 by software.
3. The watchdog timer starts operations as an interval timer when RUN is set to 1 .

Cautions 1. When the watchdog timer is cleared by setting RUN to 1, the actual overflow time is up to $0.8 \%$ shorter than the time set by timer clock selection register 2 (TCL2).
2. To set watchdog timer mode 1 or 2 , set WDTM4 to 1 after confirming that TMIF4 (bit 0 of interrupt request flag register 0 (IFO)) is set to 0 . When watchdog timer mode 1 or 2 is selected with TMIF4 set to 1, a non-maskable interrupt is generated upon the completion of rewriting WDTM4.

### 11.4 Watchdog Timer Operation

### 11.4.1 Operation as watchdog timer

The watchdog timer detects an inadvertent program loop when bit 4 (WDTM4) of the watchdog timer mode register (WDTM) is set to 1 .

The count clock (inadvertent loop detection time interval) of the watchdog timer can be selected by bits 0 to 2 (TCL20 to TCL22) of timer clock selection register 2 (TCL2). By setting bit 7 (RUN) of WDTM to 1, the watchdog timer is started. Set RUN to 1 within the set inadvertent loop detection time interval after the watchdog timer has been started. By setting RUN to 1 , the watchdog timer can be cleared and start counting. If RUN is not set to 1 , and the inadvertent loop detection time is exceeded, a system reset signal or a non-maskable interrupt is generated, depending on the value of bit 3 (WDTM3) of WDTM.

The watchdog timer continues operation in HALT mode, but stops in STOP mode. Therefore, first set RUN to 1 to clear the watchdog timer before executing the STOP instruction.

Cautions 1. The actual inadvertent loop detection time may be up to $0.8 \%$ shorter than the set time.
2. When the subsystem clock is selected as the CPU clock, watchdog timer count operation is stopped. Even when the main system clock continues oscillating in this case, the watchdog timer count operation is stopped.

Table 11-4. Inadvertent Loop Detection Time of Watchdog Timer

| TCL22 | TCL21 | TCL20 | Inadvertent Loop Detection Time | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ |
| :---: | :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | 0 | $2^{11} \times 1 / \mathrm{fx}$ | $205 \mu \mathrm{~s}$ | $410 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $2^{13} \times 1 / \mathrm{fx}$ | $819 \mu \mathrm{~s}$ | 1.64 ms |
| 1 | 0 | 0 | $2^{15} \times 1 / \mathrm{fx}$ | 3.27 ms | 6.55 ms |
| 1 | 1 | 0 | $2^{17} \times 1 / \mathrm{fx}$ | 13.1 ms | 26.2 ms |

Note Expanded-specification products only.

Remark fx: Main system clock oscillation frequency

### 11.4.2 Operation as interval timer

When bits 4 and 3 (WDTM4, WDTM3) of the watchdog timer mode register (WDTM) are set to 0 and 1, respectively, the watchdog timer operates as an interval timer that repeatedly generates an interrupt at intervals specified by a count value set in advance.

Select a count clock (or interval) by setting bits 0 to 2 (TCL20 to TCL22) of timer clock selection register 2 (TCL2). The watchdog timer starts operation as an interval timer when the RUN bit (bit 7 of WDTM) is set to 1 .

In interval timer mode, the interrupt mask flag (TMMK4) is valid, and a maskable interrupt (INTWDT) can be generated. The priority of INTWDT is set as the highest of all the maskable interrupts.

The interval timer continues operation in HALT mode, but stops in STOP mode. Therefore, first set RUN to 1 to clear the interval timer before executing the STOP instruction.

Cautions 1. Once bit 4 (WDTM4) of WDTM is set to 1 (when watchdog timer mode is selected), interval timer mode is not set unless a $\overline{\text { RESET }}$ signal is input.
2. The interval time may be up to $0.8 \%$ shorter than the set time when WDTM has just been set.

Table 11-5. Interval Time of Interval Timer

| TCL22 | TCL21 | TCL20 | Interval | At $f x=10.0 \mathrm{MHz}^{\text {Note }}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ |
| :---: | :---: | :---: | :--- | :--- | :--- |
| 0 | 0 | 0 | $2^{11} \times 1 / \mathrm{fx}$ | $205 \mu \mathrm{~s}$ | $410 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $2^{13} \times 1 / \mathrm{fx}$ | $819 \mu \mathrm{~s}$ | 1.64 ms |
| 1 | 0 | 0 | $2^{15} \times 1 / \mathrm{fx}$ | 3.27 ms | 6.55 ms |
| 1 | 1 | 0 | $2^{17} \times 1 / \mathrm{fx}$ | 13.1 ms | 26.2 ms |

Note Expanded-specification products only.

Remark fx: Main system clock oscillation frequency

## CHAPTER 12 8-BIT A/D CONVERTER ( $\mu$ PD789167 AND 789167Y SUBSERIES)

### 12.1 8-Bit A/D Converter Functions

The 8 -bit A/D converter is an 8 -bit resolution converter that converts an analog input to a digital signal. This converter can control eight channels (ANIO to ANI7) of analog inputs.

A/D conversion can be started only by software.
One of analog inputs ANIO to ANI7 is selected for A/D conversion. A/D conversion is performed repeatedly, with an interrupt request (INTADO) being issued each time A/D conversion is completed.

### 12.2 8-Bit A/D Converter Configuration

The 8-bit A/D converter consists of the following hardware.

Table 12-1. Configuration of 8-Bit A/D Converter

| Item |  |
| :--- | :--- |
| Analog input | 8 channels (ANIO to ANI7) |
| Registers | Successive approximation register (SAR) <br> A/D conversion result register 0 (ADCRO) |
| Control registers | A/D converter mode register 0 (ADM0) <br> A/D input selection register 0 (ADSO) |

Figure 12-1. Block Diagram of 8-Bit A/D Converter

(1) Successive approximation register (SAR)

SAR receives the result of comparing an analog input voltage and a voltage at the voltage tap (comparison voltage), received from the series resistor string, starting from the most significant bit (MSB).
Upon receiving all the bits, down to the least significant bit (LSB), that is, upon the completion of $A / D$ conversion, SAR sends its contents to A/D conversion result register 0 (ADCRO).

## (2) A/D conversion result register 0 (ADCRO)

ADCRO holds the result of $A / D$ conversion. Each time $A / D$ conversion ends, the conversion result in the successive approximation register is loaded into ADCRO, which is an 8-bit register.
ADCR0 can be read with an 8-bit memory manipulation instruction.
RESET input makes this register undefined.

## (3) Sample-and-hold circuit

The sample-and-hold circuit samples consecutive analog inputs from the input circuit, one by one, and sends them to the voltage comparator. The sampled analog input voltage is held during A/D conversion.

## (4) Voltage comparator

The voltage comparator compares an analog input with the voltage output by the series resistor string.
(5) Series resistor string

The series resistor string is configured between $A V_{\text {ref }}$ and $A V s s$. It generates the reference voltages against which analog inputs are compared.
(6) ANIO to ANI7

The ANIO to ANI7 pins are the 8-channel analog input pins for the A/D converter. They are used to receive the analog signals for $A / D$ conversion.

Caution Do not supply the ANIO to ANI7 pins with voltages that fall outside the rated range. If a voltage greater than or equal to $A V_{\text {ref }}$ or less than $A V$ ss (even if within the absolute maximum ratings) is supplied to any of these pins, the conversion value for the corresponding channel will be undefined. Furthermore, the conversion values for the other channels may also be affected.
(7) AVref

This pin inputs the A/D converter reference voltage.
It converts signals input to ANIO to ANI7 into digital signals according to the voltage applied between AVREF and $A V$ ss.
(8) AVss pin

The $A V$ ss pin is a ground potential pin for the $A / D$ converter. This pin must be held at the same potential as the Vsso pin, even while the A/D converter is not being used.
(9) AVdo pin

The $A V_{D D}$ pin is an analog power supply pin for the $A / D$ converter. This pin must be held at the same potential as the Vodo pin, even while the A/D converter is not being used.

### 12.3 8-Bit A/D Converter Control Registers

The following two registers are used to control the 8-bit A/D converter.

- A/D converter mode register 0 (ADMO)
- $\mathrm{A} / \mathrm{D}$ input selection register 0 (ADSO)


## (1) $A / D$ converter mode register 0 (ADMO)

ADM0 specifies the conversion time for analog inputs. It also specifies whether to enable conversion.
ADMO is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears ADM0 to 00 H .

Figure 12-2. Format of A/D Converter Mode Register 0

| Symbol | <7> | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADM0 | ADCSO | 0 | FR02 | FR01 | FR00 | 0 | 0 | 0 | FF80H | 00 H | R/W |


| ADCSO |  | A/D conversion control |
| :---: | :--- | :--- |
| 0 | Conversion disabled |  |
| 1 | Conversion enabled |  |


| FR02 | FR01 | FR00 |  | A/D conversion time selection |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}$ operation ${ }^{\text {Note } 2}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |
| 0 | 0 | 0 | 144/fx | $14.4 \mu \mathrm{~s}$ | 28.8 us |
| 0 | 0 | 1 | 120/fx | $12.0 \mu \mathrm{~s}$ | $24.0 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | 96/fx | Setting prohibited ${ }^{\text {Note } 3}$ | $19.2 \mu \mathrm{~s}$ |
| 1 | 0 | 0 | 72/fx | Setting prohibited ${ }^{\text {Note } 3}$ | $14.4 \mu \mathrm{~s}$ |
| 1 | 0 | 1 | 60/fx | Setting prohibited ${ }^{\text {Note } 3}$ | $12.0 \mu \mathrm{~s}^{\text {Note } 4}$ |
| 1 | 1 | 0 | 48/fx | Setting prohibited ${ }^{\text {Note } 3}$ |  |
| Other than above |  |  | Setting prohibited |  |  |

Notes 1. Set the A/D conversion time so that it satisfies the following ratings.
<Expanded-specification products>

$$
4.5 \leq A V_{R E F} \leq A V_{D D}=V_{D D} \leq 5.5 \vee \ldots . .12 \mu \mathrm{~s} \text { min. }
$$

$2.7 \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV}$ DD $=\mathrm{V}_{\mathrm{DD}}<4.5 \mathrm{~V} \ldots . .14 \mu \mathrm{~s} \mathrm{~min}$.
$1.8 \leq A V_{\text {REF }} \leq A V_{D D}=\operatorname{VdD}<2.7 \mathrm{~V} \ldots . . .28 \mu \mathrm{smin}$.
<Conventional products>
$2.7 \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}}=\mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V} \ldots . .14 \mu \mathrm{~s}$ min.
$1.8 \leq A V_{R E F} \leq A V_{D D}=V_{D D}<2.7 \mathrm{~V} \ldots . .28 \mu \mathrm{~s}$ min.
2. Expanded-specification products only.
3. Setting prohibited because the $A / D$ conversion time cannot satisfy the ratings in Note 1 during operation under these fx conditions.
4. Can only be set for expanded-specification products under the following conditions:
$4.5 \leq A V_{R E F} \leq A V_{D D}=V_{D D} \leq 5.5 \mathrm{~V}$. Other settings are prohibited.
Cautions 1. The result of conversion performed immediately after bit 7 (ADCSO) is set is undefined.
2. The conversion result may be undefined after ADCSO has been cleared to 0 (for details, see 12.5 (5) Timing of undefined A/D conversion result).

Remark fx: Main system clock oscillation frequency
(2) $A / D$ input selection register 0 (ADS0)

ADS0 specifies the port used to input the analog voltage to be converted to a digital signal.
ADSO is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears ADSO to 00 H .

Figure 12-3. Format of A/D Input Selection Register 0

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADSO | 0 | 0 | 0 | 0 | 0 | ADS02 | ADS01 | ADS00 | FF84H | OOH | R/W |


| ADS02 | ADS01 | ADS00 | Analog input channel specification |
| :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | ANI0 |
| 0 | 0 | 1 | ANI1 |
| 0 | 1 | 0 | ANI2 |
| 0 | 1 | 1 | ANI3 |
| 1 | 0 | 0 | ANI4 |
| 1 | 0 | 1 | ANI5 |
| 1 | 1 | 0 | ANI6 |
| 1 | 1 | 1 | ANI7 |

## Caution Bits $\mathbf{3}$ to $\mathbf{7}$ must all be set to $\mathbf{0}$.

### 12.4 8-Bit A/D Converter Operation

### 12.4.1 Basic operation of 8-bit A/D converter

<1> Select a channel for A/D conversion, using A/D input selection register 0 (ADS0).
$<2>$ The voltage supplied to the selected analog input channel is sampled using the sample and hold circuit.
<3> After sampling continues for a certain period of time, the sample and hold circuit is put on hold to keep the input analog voltage until $A / D$ conversion is completed.
<4> Bit 7 of the successive approximation register (SAR) is set. The series resistor string tap voltage at the tap selector is set to half of $A V_{\text {ref. }}$
<5> The series resistor string tap voltage is compared with the analog input voltage using the voltage comparator. If the analog input voltage is higher than half of $A V_{\text {ref, }}$ the MSB of SAR is left set. If it is lower than half of $A V_{\text {REF, }}$ the MSB is reset.
<6> Bit 6 of SAR is set automatically, and comparison shifts to the next stage. The next tap voltage of the series resistor string is selected according to bit 7, which reflects the previous comparison result, as follows:

- Bit $7=1$ : Three quarters of $A V_{\text {REF }}$
- Bit $7=0$ : One quarter of $A V_{\text {ref }}$

The tap voltage is compared with the analog input voltage. Bit 6 is set or reset according to the result of comparison.

- Analog input voltage $\geq$ tap voltage: Bit $6=1$
- Analog input voltage < tap voltage: Bit $6=0$
$<7>$ Comparison is repeated until bit 0 of SAR is reached.
<8> When comparison is completed for all of the 8 bits, a significant digital result is left in SAR. This value is sent to and latched in A/D conversion result register 0 (ADCRO). At the same time, it is possible to generate an A/D conversion end interrupt request (INTADO).

Cautions 1. The first A/D conversion value immediately after A/D conversion has been started may be undefined.
2. In standby mode, A/D converter operation is stopped.

Figure 12-4. Basic Operation of 8-Bit A/D Converter

$A / D$ conversion continues until bit 7 (ADCSO) of $A / D$ converter mode register 0 (ADMO) is reset to 0 by software.
If an attempt is made to write to $A D M 0$ or $A / D$ input selection register 0 (ADSO) during $A / D$ conversion, the ongoing $A / D$ conversion is canceled. In this case, $A / D$ conversion is restarted from the beginning, if $A D C S 0$ is set to 1.
$\overline{\text { RESET }}$ input makes A/D conversion result register 0 (ADCRO) undefined.

### 12.4.2 Input voltage and conversion result

The relationships between the analog input voltage at the analog input pins (ANIO to ANI7) and the A/D conversion result (A/D conversion result register 0 (ADCRO)) are represented by:
$\mathrm{ADCRO}=\operatorname{INT}\left(\frac{\mathrm{V}_{\mathrm{IN}}}{\mathrm{AV}_{\mathrm{REF}}} \times 256+0.5\right)$
or
$(A D C R 0-0.5) \times \frac{A V_{\text {REF }}}{256} \leq \mathrm{V}_{\mathrm{IN}}<(\mathrm{ADCRO}+0.5) \times \frac{\mathrm{A} \mathrm{V}_{\text {REF }}}{256}$
INT( ): Function that returns the integer part of a parenthesized value
VIN: Analog input voltage
AVref: Voltage of AVref pin
ADCRO: Value in A/D conversion result register 0 (ADCRO)

Figure $12-5$ shows the relationship between the analog input voltage and the $A / D$ conversion result.

Figure 12-5. Relationship Between Analog Input Voltage and A/D Conversion Result

A/D conversion result (ADCRO)


### 12.4.3 Operation mode of 8-bit A/D converter

The $A / D$ converter is initially in select mode. In this mode, $A / D$ input selection register 0 (ADSO) is used to select an analog input channel from ANIO to ANI7 for A/D conversion.
$A / D$ conversion can be started only by software; that is, by setting $A / D$ converter mode register 0 (ADM0).
The $A / D$ conversion result is saved to $A / D$ conversion result register 0 (ADCRO). At the same time, an interrupt request signal (INTADO) is generated.

## - Software-started A/D conversion

Setting bit 7 (ADCSO) of $A / D$ converter mode register 0 (ADMO) to 1 triggers $A / D$ conversion for the voltage applied to the analog input pin specified in A/D input selection register 0 (ADSO). Upon completion of A/D conversion, the conversion result is saved to A/D conversion result register 0 (ADCRO). At the same time, an interrupt request signal (INTADO) is generated. Once A/D conversion is activated, and completed, another session of $A / D$ conversion is started. A/D conversion is repeated until new data is written to ADMO. If data where ADCS0 is 1 is written to ADM0 again during A/D conversion, the ongoing session of $A / D$ conversion is discontinued, and a new session of $A / D$ conversion begins for the new data. If data where ADCSO is 0 is written to ADM0 again during A/D conversion, A/D conversion is stopped immediately.

Figure 12-6. Software-Started A/D Conversion


Remarks 1. $\mathrm{n}=0$ to 7
2. $\mathrm{m}=0$ to 7

### 12.5 Cautions Related to 8-Bit A/D Converter

## (1) Current consumption in standby mode

In standby mode, the A/D converter stops operating. Setting bit 7 (ADCSO) of $A / D$ converter mode register 0 (ADMO) to 0 can reduce the current consumption.
Figure 12-7 shows how to reduce the current consumption in standby mode.

Figure 12-7. How to Reduce Current Consumption in Standby Mode

(2) Input range for ANIO to ANI7 pins

Be sure to keep the input voltage at ANIO to ANI7 within its rating. If a voltage greater than or equal to $A V_{\text {ref }}$ or less than or equal to AVss (even within the absolute maximum ratings) is input into a conversion channel, the conversion output of the channel becomes undefined, and the conversion output of the other channels may also be affected.

## (3) Conflict

<1> Conflict between writing to A/D conversion result register 0 (ADCRO) at the end of conversion and reading from ADCR0 using instruction
Reading from ADCR0 takes precedence. After reading, the new conversion result is written to ADCRO.
<2> Conflict between writing to ADCR0 at the end of conversion and writing to A/D converter mode register 0 (ADMO) or A/D input selection register 0 (ADSO)
Writing to ADM0 or ADS0 takes precedence. ADCR0 is not written to. No A/D conversion end interrupt request signal (INTADO) is generated.
(4) Conversion result immediately after start of A/D conversion

The first $A / D$ conversion value immediately after $A / D$ conversion has been started is undefined. Poll the $A / D$ conversion end interrupt request (INTADO) and drop the first conversion result.

## (5) Timing of undefined $A / D$ conversion result

The $A / D$ conversion value may become undefined if the timing of the completion of $A / D$ conversion and that to stop the $A / D$ conversion operation conflict. Therefore, read the $A / D$ conversion result while the $A / D$ conversion operation is in progress. To read the A/D conversion result after the A/D conversion operation has been stopped, stop the A/D conversion operation before the next conversion operation is completed. Figures $12-8$ and $12-9$ show the timing at which the conversion result is read.

Figure 12-8. Conversion Result Read Timing (If Conversion Result Is Undefined)


Figure 12-9. Conversion Result Read Timing (If Conversion Result Is Normal)


## (6) Noise prevention

To maintain a resolution of 8 bits, watch out for noise on the $A V_{\text {ref }}$ and ANIO to ANI7 pins. The higher the output impedance of the analog input source, the larger the effect from noise. To reduce noise, attach an external capacitor to the relevant pins as shown in Figure 12-10.

Figure 12-10. Analog Input Pin Handling


## (7) ANIO to ANI7

The analog input pins (ANIO to ANI7) are alternate-function pins. They are used also as port pins (P60 to P67).
If any of ANIO to ANI7 has been selected for A/D conversion, do not execute input instructions for the ports. Otherwise, the conversion resolution may become lower.
If a digital pulse is applied to a pin adjacent to the analog input pins during A/D conversion, coupling noise may occur which prevents an A/D conversion result from being obtained as expected. Avoid applying a digital pulse to pins adjacent to the analog input pins during A/D conversion.

## (8) Input impedance of ANIO to ANI7 pins

This A/D converter charges the internal sampling capacitor for about $1 / 10$ of the conversion time, and performs sampling.
Therefore at times other than sampling, only the leak current is output. During sampling, the current for charging the capacitor is also output, so the input impedance fluctuates and has no meaning.
However, to ensure adequate sampling, it is recommended that the output impedance of the analog input source be set to below $10 \mathrm{k} \Omega$, or a 100 pF capacitor be connected to the ANIO to ANI7 pins (see Figure 1210).
(9) Input impedance of the AVref pin

A series resistor string of several $10 \mathrm{k} \Omega$ is connected across the $A V_{\text {REF }}$ and $A V s s$ pins.
If the output impedance of the reference voltage source is high, this high impedance is eventually connected in series with the series resistor string across the $A V_{\text {ref }}$ and $A V s s$ pins, leading to a higher reference voltage error.
(10) Interrupt request flag (ADIF0)

Changing the content of $A / D$ converter mode register 0 (ADMO) does not clear the interrupt request flag (ADIFO).
If the analog input pins are changed during A/D conversion, therefore, the $A / D$ conversion result and the conversion end interrupt request flag may reflect the previous analog input immediately before writing to ADM0 occurs. In this case, ADIF0 may already be set if it is read-accessed immediately after ADM0 is writeaccessed, even when A/D conversion has not been completed for the new analog input.
In addition, when A/D conversion is restarted, ADIF0 must be cleared beforehand.

Figure 12-11. A/D Conversion End Interrupt Request Generation Timing


Remarks 1. $\mathrm{n}=0$ to 7
2. $m=0$ to 7

## (11) AVdd pin

The AVdd pin is used to supply power to the analog circuit. It is also used to supply power to the ANIO to ANI7 input circuit.
If your application is designed to be changed to backup power, the AVDD pin must be supplied with the same voltage level as for the Vodo pin, as shown in Figure 12-12.

Figure 12-12. AVdd Pin Handling


## CHAPTER 13 10-BIT A/D CONVERTER ( $\mu$ PD789177 AND 789177Y SUBSERIES)

### 13.1 10-Bit A/D Converter Functions

The 10-bit A/D converter is a 10-bit resolution converter that converts an analog input to a digital signal. This converter can control eight channels (ANIO to ANI7) of analog inputs.

A/D conversion can be started only by software.
One of analog inputs ANIO to ANI7 is selected for A/D conversion. A/D conversion is performed repeatedly, with an interrupt request (INTADO) being issued each time A/D conversion is completed.

### 13.2 10-Bit A/D Converter Configuration

The 10-bit A/D converter consists of the following hardware.

Table 13-1. Configuration of 10-Bit A/D Converter

| Item |  |
| :--- | :--- |
| Analog input | 8 channels (ANIO to ANI7) |
| Registers | Successive approximation register (SAR) <br> A/D conversion result register 0 (ADCRO) |
| Control registers | A/D converter mode register 0 (ADM0) <br> A/D input selection register 0 (ADSO) |

Figure 13-1. Block Diagram of 10-Bit A/D Converter

(1) Successive approximation register (SAR)

SAR receives the result of comparing an analog input voltage and a voltage at a voltage tap (comparison voltage), received from the series resistor string, starting from the most significant bit (MSB).
Upon receiving all the bits, down to the least significant bit (LSB), that is, upon the completion of $A / D$ conversion, SAR sends its contents to A/D conversion result register 0 (ADCRO).
(2) $A / D$ conversion result register 0 (ADCRO)

ADCRO is a 16-bit register that holds the result of A/D conversion. The lower 6 bits are fixed to 0 . Each time A/D conversion ends, the conversion result in the successive approximation register is loaded into ADCRO. The results are stored in ADCR0 from the most significant bit (MSB).
ADCRO can be read with a 16-bit memory manipulation instruction.
RESET input sets ADCR0 to 0000 H .


Caution When the $\mu$ PD78F9177, the flash memory counterpart of the $\mu$ PD789166 or $\mu$ PD789167, is used, the register can be accessed in 8-bit units. However, only an object file assembled with the $\mu$ PD789166 or $\mu$ PD789167 can be used. The same is also true for the $\mu$ PD78F9177Y, the flash memory counterpart of the $\mu$ PD789166Y or $\mu$ PD789167Y. When the $\mu$ PD78F9177Y is used, the register can be accessed in 8-bit units. However, only an object file assembled with the $\mu$ PD789166Y or $\mu$ PD789167Y can be used.

## (3) Sample-and-hold circuit

The sample-and-hold circuit samples consecutive analog inputs from the input circuit, one by one, and sends them to the voltage comparator. The sampled analog input voltage is held during A/D conversion.
(4) Voltage comparator

The voltage comparator compares an analog input with the voltage output by the series resistor string.
(5) Series resistor string

The series resistor string is configured between $A V_{\text {ref }}$ and $A V s s$. It generates the reference voltages against which analog inputs are compared.

## (6) ANIO to ANI7

The ANIO to ANI7 pins are the 8-channel analog input pins for the A/D converter. They are used to receive the analog signals for $A / D$ conversion.

Caution Do not supply the ANIO to ANI7 pins with voltages that fall outside the rated range. If a voltage greater than or equal to AVref or less than or equal to AVss (even if within the absolute maximum ratings) is supplied to any of these pins, the conversion value for the corresponding channel will be undefined. Furthermore, the conversion values for the other channels may also be affected.
(7) AVref pin

This pin inputs the A/D converter reference voltage.
It converts signals input to ANIO to ANI7 into digital signals according to the voltage applied between AVref and $A V$ ss.
(8) AVss pin

The $A V$ ss pin is a ground potential pin for the $A / D$ converter. This pin must be held at the same potential as the $V_{s s o}$ pin, even while the A/D converter is not being used.
(9) AVdo pin

The AVDd pin is an analog power supply pin for the A/D converter. This pin must be held at the same potential as the Vodo pin, even while the A/D converter is not being used.

### 13.3 10-Bit A/D Converter Control Registers

The following two registers are used to control the 10 -bit $\mathrm{A} / \mathrm{D}$ converter.

- A/D converter mode register 0 (ADMO)
- $A / D$ input selection register 0 (ADSO)


## (1) $A / D$ converter mode register 0 (ADMO)

ADM0 specifies the conversion time for analog inputs. It also specifies whether to enable conversion.
ADMO is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ADMO to 00H.

Figure 13-2. Format of A/D Converter Mode Register 0

| Symbol | <7> | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADM0 | ADCS0 | 0 | FR02 | FR01 | FR00 | 0 | 0 | 0 | FF80H | 00H | R/W |


| ADCSO |  |
| :---: | :--- |
| 0 | Conversion disabled |
| 1 | Conversion enabled conversion control |


| FR02 | FR01 | FR00 |  | A/D conversion time selectio | $\mathrm{n}^{\text {Note } 1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | At $\mathrm{fx}_{\mathrm{x}}=10.0 \mathrm{MHz}$ operation ${ }^{\text {Nole } 2}$ | At $\mathrm{fx}_{\mathrm{x}}=5.0 \mathrm{MHz}$ operation |
| 0 | 0 | 0 | 144/fx | $14.4 \mu \mathrm{~s}$ | 28.8 ¢ |
| 0 | 0 | 1 | 120/fx | 12.0 /s | $24.0 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | 96/fx | Setting prohibited ${ }^{\text {Note }} 3$ | $19.2 \mu \mathrm{~s}$ |
| 1 | 0 | 0 | 72/fx | Setting prohibited ${ }^{\text {Note } 3}$ | $14.4 \mu \mathrm{~s}$ |
| 1 | 0 | 1 | 60/fx | Setting prohibited ${ }^{\text {Note } 3}$ | $12.0 \mu \mathrm{~s}^{\text {Note } 4}$ |
| 1 | 1 | 0 | 48/fx | Setting prohibited ${ }^{\text {Note } 3}$ |  |
| Other than above |  |  | Setting prohibited |  |  |

Notes 1. Set the A/D conversion time so that it satisfies the following ratings.
<Expanded-specification products>
$4.5 \leq A V_{\text {REF }} \leq A V_{D D}=V_{D D} \leq 5.5 \vee \ldots . .12 \mu \mathrm{~s}$ min.
$2.7 \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV}$ dD $=\mathrm{V}_{\mathrm{DD}}<4.5 \mathrm{~V} \ldots . .14 \mu \mathrm{~s} \mathrm{~min}$.
$1.8 \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV}$ DD $=\mathrm{V}_{\mathrm{DD}}<2.7 \mathrm{~V} \ldots . .28 \mu \mathrm{~s} \mathrm{~min}$.
<Conventional products>
$2.7 \leq A V_{\text {REF }} \leq A V_{d D}=V_{d D} \leq 5.5 \mathrm{~V} \ldots . .14 \mu \mathrm{~s}$ min.
$1.8 \leq A V_{\text {REF }} \leq A V_{D D}=V_{D D}<2.7 \mathrm{~V} \ldots . .28 \mu \mathrm{~s} \mathrm{~min}$.
2. Expanded-specification products only.
3. Setting prohibited because the A/D conversion time cannot satisfy the ratings in Note 1 during operation under these fx conditions.
4. Can only be set for expanded-specification products under the following conditions:
$4.5 \leq \mathrm{AV}$ REF $\leq \mathrm{AVDD}=\mathrm{VDD} \leq 5.5 \mathrm{~V}$. Other settings are prohibited.
Cautions 1. The result of conversion performed immediately after bit 7 (ADCSO) is set is undefined.
2. The conversion result may be undefined after ADCSO has been cleared to 0 (For details, see 13.5 (5) Timing of undefined A/D conversion result).

Remark fx: Main system clock oscillation frequency
(2) A/D input selection register 0 (ADSO)

ADS0 specifies the port used to input the analog voltage to be converted to a digital signal.
ADSO is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears ADSO to 00 H .

Figure 13-3. Format of A/D Input Selection Register 0

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADSO | 0 | 0 | 0 | 0 | 0 | ADS02 | ADS01 | ADS00 | FF84H | OOH | R/W |


| ADS02 | ADS01 | ADS00 | Analog input channel specification |
| :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | ANI0 |
| 0 | 0 | 1 | ANI1 |
| 0 | 1 | 0 | ANI2 |
| 0 | 1 | 1 | ANI3 |
| 1 | 0 | 0 | ANI4 |
| 1 | 0 | 1 | ANI5 |
| 1 | 1 | 0 | ANI6 |
| 1 | 1 | 1 | ANI7 |

## Caution Bits $\mathbf{3}$ to $\mathbf{7}$ must all be set to 0 .

### 13.4 10-Bit A/D Converter Operation

### 13.4.1 Basic operation of 10-bit A/D converter

<1> Select a channel for A/D conversion, using A/D input selection register 0 (ADSO).
<2> The voltage supplied to the selected analog input channel is sampled using the sample and hold circuit.
$<3>$ After sampling continues for a certain period of time, the sample and hold circuit is put on hold to keep the input analog voltage until A/D conversion is completed.
<4> Bit 9 of the successive approximation register (SAR) is set. The series resistor string tap voltage at the tap selector is set to half of $A V_{\text {ref. }}$
<5> The series resistor string tap voltage is compared with the analog input voltage using the voltage comparator. If the analog input voltage is higher than half of $A V_{r e f, ~ t h e ~ M S B ~ o f ~ S A R ~ i s ~ l e f t ~ s e t . ~ I f ~ i t ~ i s ~}^{\text {s }}$ lower than half of $A V_{\text {REF, }}$ the MSB is reset.
<6> Bit 8 of SAR is set automatically, and comparison shifts to the next stage. The next tap voltage of the series resistor string is selected according to bit 9, which reflects the previous comparison result, as follows:

- Bit 9 =1: Three quarters of $A V_{\text {ref }}$
- Bit $9=0$ : One quarter of $A V_{\text {ref }}$

The tap voltage is compared with the analog input voltage. Bit 8 is set or reset according to the result of comparison.

- Analog input voltage $\geq$ tap voltage: Bit $8=1$
- Analog input voltage < tap voltage: Bit $8=0$
$<7>$ Comparison is repeated until bit 0 of SAR is reached.
<8> When comparison is completed for all of the 10 bits, a significant digital result is left in SAR. This value is sent to and latched in A/D conversion result register 0 (ADCRO). At the same time, it is possible to generate an A/D conversion end interrupt request (INTADO).

Cautions 1. The first A/D conversion value immediately after $A / D$ conversion has been started may be undefined.
2. In standby mode, A/D converter operation is stopped.

Figure 13-4. Basic Operation of 10-Bit A/D Converter


A/D conversion continues until bit 7 (ADCSO) of $A / D$ converter mode register 0 (ADMO) is reset to 0 by software.
If an attempt is made to write to $A D M O$ or $A / D$ input selection register 0 (ADSO) during $A / D$ conversion, the ongoing $A / D$ conversion is canceled. In this case, $A / D$ conversion is restarted from the beginning, if $A D C S 0$ is set to 1.
$\overline{\text { RESET }}$ input makes A/D conversion result register 0 (ADCRO) undefined.

### 13.4.2 Input voltage and conversion result

The relationships between the analog input voltage at the analog input pins (ANIO to ANI7) and the A/D conversion result (A/D conversion result register 0 (ADCRO)) are represented by:
$\mathrm{ADCRO}=\mathrm{INT}\left(\frac{\mathrm{V} \operatorname{IN}}{\mathrm{AV} \mathrm{V}_{\text {REF }}} \times 1,024+0.5\right)$
or
$($ ADCR0 -0.5$) \times \frac{A V_{\text {REF }}}{1,024} \leq \mathrm{V}_{\mathrm{IN}}<(\mathrm{ADCRO}+0.5) \times \frac{\mathrm{A} \mathrm{V}_{\text {REF }}}{1,024}$
INT( ): Function that returns the integer part of a parenthesized value
VIn: Analog input voltage
AVref: Voltage of AVref pin
ADCRO: Value in A/D conversion result register 0 (ADCRO)

Figure 13-5 shows the relationship between the analog input voltage and the $A / D$ conversion result.

Figure 13-5. Relationship Between Analog Input Voltage and A/D Conversion Result


### 13.4.3 Operation mode of 10-bit A/D converter

The A/D converter is initially in select mode. In this mode, A/D input selection register 0 (ADSO) is used to select an analog input channel from ANIO to ANI7 for A/D conversion.
$A / D$ conversion can be started only by software; that is, by setting A/D converter mode register 0 (ADMO).
The $A / D$ conversion result is saved to $A / D$ conversion result register 0 (ADCRO). At the same time, an interrupt request signal (INTADO) is generated.

## - Software-started A/D conversion

Setting bit 7 (ADCSO) of $A / D$ converter mode register 0 (ADMO) to 1 triggers $A / D$ conversion for the voltage applied to the analog input pin specified in $A / D$ input selection register 0 (ADSO). Upon completion of $A / D$ conversion, the conversion result is saved to $A / D$ conversion result register 0 (ADCRO). At the same time, an interrupt request signal (INTADO) is generated. Once A/D conversion is activated, and completed, another session of $A / D$ conversion is started. A/D conversion is repeated until new data is written to ADMO. If data where ADCS0 is 1 is written to ADM0 again during A/D conversion, the ongoing session of $A / D$ conversion is discontinued, and a new session of $A / D$ conversion begins for the new data. If data where ADCSO is 0 is written to ADM0 again during A/D conversion, A/D conversion is stopped immediately.

Figure 13-6. Software-Started A/D Conversion


Remarks 1. $\mathrm{n}=0$ to 7
2. $m=0$ to 7

### 13.5 Cautions Related to 10-Bit A/D Converter

## (1) Current consumption in standby mode

In standby mode, the A/D converter stops operating. Stopping conversion (bit 7 (ADCSO) of A/D converter mode register $0($ ADMO $)=0)$ can reduce the current consumption.
Figure 13-7 shows how to reduce the current drain in standby mode.

Figure 13-7. How to Reduce Current Consumption in Standby Mode

(2) Input range for ANIO to ANI7 pins

Be sure to keep the input voltage at ANIO to ANI7 within its rating. If a voltage greater than or equal to $A V_{\text {ref }}$ or less than or equal to $A V s s$ (even within the absolute maximum rating) is input into a conversion channel, the conversion output of the channel becomes undefined, and the conversion output of the other channels may also be affected.
(3) Conflict
$<1>$ Conflict between writing to A/D conversion result register 0 (ADCRO) at the end of conversion and reading from ADCRO using instruction
Reading from ADCR0 takes precedence. After reading, the new conversion result is written to ADCRO.
<2> Conflict between writing to ADCR0 at the end of conversion and writing to A/D converter mode register 0 (ADM0) or A/D input selection register 0 (ADSO)
Writing to ADMO or ADS0 takes precedence. ADCR0 is not written to. No A/D conversion end interrupt request signal (INTADO) is generated.
(4) Conversion result immediately after start of A/D conversion

The first $A / D$ conversion value immediately after $A / D$ conversion has been started is undefined. Poll the $A / D$ conversion end interrupt request (INTADO) and drop the first conversion result.
(5) Timing of undefined A/D conversion result

The $A / D$ conversion value may become undefined if the timing of the completion of $A / D$ conversion and that to stop the $A / D$ conversion operation conflict. Therefore, read the $A / D$ conversion result while the $A / D$ conversion operation is in progress. To read the $A / D$ conversion result after the $A / D$ conversion operation has been stopped, stop the A/D conversion operation before the next conversion operation is completed. Figures $13-8$ and $13-9$ show the timing at which the conversion result is read.

Figure 13-8. Conversion Result Read Timing (If Conversion Result Is Undefined)


Figure 13-9. Conversion Result Read Timing (If Conversion Result Is Normal)

(6) Noise prevention

To maintain a resolution of 10 bits, watch out for noise on the AVref and ANIO to ANI7 pins. The higher the output impedance of the analog input source, the larger the effect from noise. To reduce noise, attach an external capacitor to the relevant pins as shown in Figure 13-10.

Figure 13-10. Analog Input Pin Handling


## (7) ANIO to ANI7

The analog input pins (ANIO to ANI7) are alternate-function pins. They are used also as port pins (P60 to P67).
If any of ANIO to ANI7 has been selected for A/D conversion, do not execute input instructions for the ports. Otherwise, the conversion resolution may become lower.
If a digital pulse is applied to a pin adjacent to the analog input pins during A/D conversion, coupling noise may occur which prevents an A/D conversion result from being obtained as expected. Avoid applying a digital pulse to pins adjacent to the analog input pins during A/D conversion.
(8) Input impedance of ANIO to ANI7 pins

This $A / D$ converter charges the internal sampling capacitor for about $1 / 10$ of the conversion time, and performs sampling.
Therefore at times other than sampling, only the leak current is output. During sampling, the current for charging the capacitor is also output, so the input impedance fluctuates and has no meaning.
However, to ensure adequate sampling, it is recommended that the output impedance of the analog input source be set to below $10 \mathrm{k} \Omega$, or a 100 pF capacitor be connected to the ANIO to ANI7 pins (see Figure 1310).
(9) Input impedance of the AVref pin

A series resistor string of several $10 \mathrm{k} \Omega$ is connected across the $A V_{\text {ref }}$ and $A V s s$ pins.
If the output impedance of the reference voltage source is high, this high impedance is eventually connected in series with the series resistor string across the $A V_{\text {ref }}$ and $A V$ ss pins, leading to a higher reference voltage error.
(10) Interrupt request flag (ADIFO)

Changing the content of $A / D$ converter mode register 0 (ADMO) does not clear the interrupt request flag (ADIFO).
If the analog input pins are changed during $A / D$ conversion, therefore, the $A / D$ conversion result and the conversion end interrupt request flag may reflect the previous analog input immediately before writing to ADM0 occurs. In this case, ADIF0 may already be set if it is read-accessed immediately after ADM0 is writeaccessed, even when A/D conversion has not been completed for the new analog input. In addition, when A/D conversion is restarted, ADIFO must be cleared beforehand.

Figure 13-11. A/D Conversion End Interrupt Request Generation Timing


Remarks 1. $\mathrm{n}=0$ to 7
2. $\mathrm{m}=0$ to 7

## (11) AVdd pin

The AVDD pin is used to supply power to the analog circuit. It is also used to supply power to the ANIO to ANI7 input circuit.
If your application is designed to be changed to backup power, the $A V D D$ pin must be supplied with the same voltage level as for the Vddo pin, as shown in Figure 13-12.

Figure 13-12. AVdd Pin Treatment


## CHAPTER 14 SERIAL INTERFACE 20

### 14.1 Functions of Serial Interface 20

Serial interface 20 has the following three modes.

- Operation stop mode
- Asynchronous serial interface (UART) mode
- 3-wire serial I/O mode


## (1) Operation stop mode

This mode is used when serial transfer is not performed. Power consumption is minimized in this mode.
(2) Asynchronous serial interface (UART) mode

This mode is used to send and receive the one byte of data that follows a start bit. It supports full-duplex communication.

Serial interface 20 contains an UART-dedicated baud rate generator, enabling communication over a wide range of baud rates. It is also possible to define baud rates by dividing the frequency of the clock input to the ASCK20 pin.
(3) 3-wire serial I/O mode (switchable between MSB-first and LSB-first transmission)

This mode is used to transmit 8-bit data, using three lines: a serial clock ( $\overline{\mathrm{SCK20}}$ ) line and two serial data lines (SI20 and SO20).
As it supports simultaneous transmission and reception, 3-wire serial I/O mode requires less processing time for data transmission than asynchronous serial interface mode.
Because, in 3-wire serial I/O mode, it is possible to select whether 8-bit data transmission begins with the MSB or LSB, serial interface 20 can be connected to any device regardless of whether that device is designed for MSB-first or LSB-first transmission.
3-wire serial I/O mode is useful for connecting peripheral I/O circuits and display controllers having conventional synchronous serial interfaces, such as those of the 75XL, 78K, and 17K Series devices.

### 14.2 Configuration of Serial Interface 20

Serial interface 20 consists of the following hardware.

Table 14-1. Configuration of Serial Interface 20

| Item |  |
| :--- | :--- |
| Registers | Transmission shift register 20 (TXS20) <br> Reception shift register 20 (RXS20) <br> Reception buffer register 20 (RXB20) |
| Control registers | Serial operation mode register 20 (CSIM20) <br> Asynchronous serial interface mode register 20 (ASIM20) <br> Asynchronous serial interface status register 20 (ASIS20) <br> Baud rate generator control register 20 (BRGC20) <br> Port mode register 2 (PM2) <br> Port 2 (P2) |

Figure 14-1. Block Diagram of Serial Interface 20


Note See Figure 14-2 for the configuration of the baud rate generator.
$\stackrel{\rightharpoonup}{\oplus}$

Figure 14-2. Block Diagram of Baud Rate Generator 20


## (1) Transmission shift register 20 (TXS20)

TXS20 is a register in which transmission data is prepared. The transmission data is output from TXS20 bit serially.
When the data length is seven bits, bits 0 to 6 of the data in TXS20 will be transmission data. Writing data to TXS20 triggers transmission.
TXS20 can be written with an 8-bit memory manipulation instruction, but cannot be read.
RESET input sets TXS20 to FFH.

## Caution Do not write to TXS20 during transmission.

TXS20 and reception buffer register 20 (RXB20) are mapped at the same address, so any attempt to read from TXS20 results in a value being read from RXB20.
(2) Reception shift register 20 (RXS20)

RXS20 is a register in which serial data, received at the RxD20 pin, is converted to parallel data. Once one entire byte has been received, RXS20 feeds the reception data to reception buffer register 20 (RXB20).
RXS20 cannot be manipulated directly by a program.
(3) Reception buffer register 20 (RXB20)

RXB20 holds reception data. A new reception data is transferred from reception shift register 20 (RXS20) every 1-byte data reception.
When the data length is seven bits, the reception data is sent to bits 0 to 6 of RXB20, in which the MSB is always fixed to 0 .
RXB20 can be read with an 8-bit memory manipulation instruction, but cannot be written.
$\overline{\text { RESET input makes RXB20 undefined. }}$

## Caution RXB20 and transmission shift register 20 (TXS20) are mapped at the same address, so any attempt to write to RXB20 results in a value being written to TXS20.

## (4) Transmission controller

The transmission controller controls transmission. For example, it adds start, parity, and stop bits to the data in transmission shift register 20 (TXS20), according to the setting of asynchronous serial interface mode register 20 (ASIM20).

## (5) Reception controller

The reception controller controls reception according to the setting of asynchronous serial interface mode register 20 (ASIM20). It also checks for errors, such as parity errors, during reception. If an error is detected, asynchronous serial interface status register 20 (ASIS20) is set according to the status of the error.

### 14.3 Control Registers of Serial Interface 20

Serial interface 20 is controlled by the following registers.

- Serial operation mode register 20 (CSIM20)
- Asynchronous serial interface mode register 20 (ASIM20)
- Asynchronous serial interface status register 20 (ASIS20)
- Baud rate generator control register 20 (BRGC20)
- Port mode register 2 (PM2)
- Port 2 (P2)
(1) Serial operation mode register 20 (CSIM20)

CSIM20 is used to make the settings related to 3 -wire serial I/O mode.
CSIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears CSIM20 to 00H.
Figure 14-3. Format of Serial Operation Mode Register 20

| Symbol | <7> | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSIM20 | CSIE20 | SSE20 | 0 | 0 | DAP20 | DIR20 | CSCK20 | CKP20 | FF72H | 00H | R/W |


| CSIE20 |  | 3-wire serial I/O mode operation control |
| :---: | :--- | :--- |
| 0 | Operation disabled |  |
| 1 | Operation enabled |  |


| SSE20 | $\overline{\text { SS20-pin selection }}$ | Function of $\overline{\mathrm{SS20}} / \mathrm{P} 25$ pin | Communication status |
| :---: | :--- | :--- | :--- |
| 0 | Not used | Port function | Communication enabled |
| 1 | Used | 0 | Communication enabled |
|  |  | 1 | Communication disabled |


| DAP20 | 3-wire serial I/O mode data phase selection |
| :---: | :--- |
| 0 | Output at the falling edge of $\overline{\text { SCK20. }}$ |
| 1 | Output at the rising edge of $\overline{\text { SCK20. }}$. |


| DIR20 |  | First-bit specification |
| :---: | :--- | :--- |
| 0 | MSB |  |
| 1 | LSB |  |


| CSCK20 | 3-wire serial I/O mode clock selection |
| :---: | :--- |
| 0 | External clock input to the $\overline{\mathrm{SCK20}}$ pin |
| 1 | Output of the dedicated baud rate generator |


| CKP20 | 3-wire serial I/O mode clock phase selection |
| :---: | :--- |
| 0 | Clock is active-low, and $\overline{\text { SCK20 }}$ is high level in the idle state. |
| 1 | Clock is active-high, and $\overline{\text { SCK20 }}$ is low level in the idle state. |

Cautions 1. Bits 4 and 5 must be set to 0 .
2. CSIM20 must be cleared to 00 H , if UART mode is selected.
3. Switch operating modes after halting the serial transmit/receive operation.
4. When the external input clock is selected in 3 -wire serial I/O mode, set input mode by setting bit 0 of port mode register 2 (PM2) to 1.
(2) Asynchronous serial interface mode register 20 (ASIM20)

ASIM20 is used to make the settings related to the asynchronous serial interface mode.
ASIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ASIM20 to 00 H .

Figure 14-4. Format of Asynchronous Serial Interface Mode Register 20

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIM20 | TXE20 | RXE20 | PS201 | PS200 | CL20 | SL20 | 0 | 0 | FF70H | OOH | R/W |


| TXE20 |  | Transmit operation control |
| :---: | :--- | :--- |
| 0 | Transmit operation stop |  |
| 1 | Transmit operation enable |  |


| RXE20 |  | Receive operation control |
| :---: | :--- | :--- |
| 0 | Receive operation stop |  |
| 1 | Receive operation enable |  |


| PS201 | PS200 | Parity bit specification |
| :---: | :---: | :--- |
| 0 | 0 | No parity |
| 0 | 1 | Always add 0 parity at transmission. <br> Parity check is not performed at reception (No parity error is generated). |
| 1 | 0 | Odd parity |
| 1 | 1 | Even parity |


| CL20 | Transmit data character length specification |  |
| :---: | :--- | :--- |
| 0 | 7 bits |  |
| 1 | 8 bits |  |


| SL20 |  | Transmit data stop bit length |
| :---: | :--- | :--- |
| 0 | 1 bit |  |
| 1 | 2 bits |  |

## Cautions 1. Bits 0 and 1 must be set to 0 .

2. If 3-wire serial I/O mode is selected, ASIM20 must be cleared to 00 H .
3. Switch operating modes after halting serial transmit/receive operation.

Table 14-2. Operating Mode Settings of Serial Interface 20
(1) Operation stop mode

| ASIM20 |  | CSIM20 |  |  | PM22 | P22 | PM21 | P21 | PM20 | P20 | $\begin{gathered} \text { First } \\ \text { Bit } \end{gathered}$ | Shift <br> Clock | $\begin{aligned} & \text { P22/SI20/ } \\ & \text { RxD20 Pin } \\ & \text { Function } \end{aligned}$ | P21/SO20/ <br> TxD20 Pin <br> Function | $\begin{gathered} \text { P20/ } \overline{\text { SCK20 }} / \\ \text { ASCK20 Pin } \\ \text { Function } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TXE20 | RXE20 | CSIE20 | DIR20 | CSCK20 |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 0 | 0 | $\times$ | $\times$ | $x^{\text {Note } 1}$ | $x^{\text {Note } 1}$ | $\times^{\text {Note } 1}$ | $x^{\text {Note } 1}$ | $x^{\text {Note } 1}$ | $\times^{\text {Note } 1}$ | - | - | P22 | P21 | P20 |
| Other than above |  |  |  |  |  |  |  |  |  |  | Setting prohibited |  |  |  |  |

(2) 3-wire serial I/O mode

| ASIM20 |  | CSIM20 |  |  | PM22 | P22 | PM21 | P21 | PM20 | P20 | First <br> Bit | Shift <br> Clock | $\begin{gathered} \text { P22/SI20/ } \\ \text { RxD20 Pin } \\ \text { Function } \end{gathered}$ | P21/SO20/ <br> TxD20 Pin <br> Function | $\begin{gathered} \text { P20/ } \overline{\text { SCK20 }} / \\ \text { ASCK20 Pin } \\ \text { Function } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TXE20 | RXE20 | CSIE20 | DIR20 | CSCK20 |  |  |  |  |  |  |  |  |  |  |  |
| 0 | 0 | 1 | 0 | 0 | $x^{\text {Note } 2}$ | $x^{\text {Note } 2}$ | 0 | 1 | 1 | $\times$ | MSB | External <br> clock | $\mathrm{SI} 20{ }^{\text {Note } 2}$ | SO20 <br> (CMOS output) | $\begin{aligned} & \overline{\text { SCK20 }} \\ & \text { input } \end{aligned}$ |
|  |  |  |  | 1 |  |  |  |  | 0 | 1 |  | Internal <br> clock |  |  | $\begin{aligned} & \overline{\text { SCK20 }} \\ & \text { output } \end{aligned}$ |
|  |  | 1 | 1 | 0 |  |  |  |  | 1 | $\times$ | LSB | External clock |  |  | $\begin{aligned} & \overline{\text { SCK20 }} \\ & \text { input } \end{aligned}$ |
|  |  |  |  | 1 |  |  |  |  | 0 | 1 |  | Internal <br> clock |  |  | $\begin{aligned} & \overline{\text { SCK20 }} \\ & \text { output } \end{aligned}$ |
| Other than above |  |  |  |  |  |  |  |  |  |  | Setting prohibited |  |  |  |  |

(3) Asynchronous serial interface mode

| ASIM20 |  | CSIM20 |  |  | PM22 | P22 | PM21 | P21 | PM20 | P20 | First <br> Bit | Shift Clock | P22/SI20/ <br> RxD20 Pin <br> Function | P21/SO20/ <br> TxD20 Pin <br> Function | $\begin{gathered} \text { P20/SCK20/ } \\ \text { ASCK20 Pin } \\ \text { Function } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TXE20 | RXE20 | CSIE20 | DIR20 | CSCK20 |  |  |  |  |  |  |  |  |  |  |  |
| 1 | 0 | 0 | 0 | 0 | $\times^{\text {Note } 1}$ | $x^{\text {Note } 1}$ | 0 | 1 | 1 | $\times$ | LSB | External <br> clock | P22 | TxD20 <br> (CMOS output) | ASCK20 <br> input |
|  |  |  |  |  |  |  |  |  | $x^{\text {Note } 1}$ | $x^{\text {Note } 1}$ |  | Internal clock |  |  | P20 |
| 0 | 1 | 0 | 0 | 0 | 1 | $\times$ | $x^{\text {Note } 1}$ | $x^{\text {Note } 1}$ | 1 | $\times$ |  | External <br> clock | RxD20 | P21 | ASCK20 <br> input |
|  |  |  |  |  |  |  |  |  | $\times^{\text {Note } 1}$ | $\times^{\text {Note } 1}$ |  | Internal clock |  |  | P20 |
| 1 | 1 | 0 | 0 | 0 | 1 | $\times$ | 0 | 1 | 1 | $\times$ |  | External <br> clock |  | TxD20 <br> (CMOS output) | ASCK20 input |
|  |  |  |  |  |  |  |  |  | $\times^{\text {Note } 1}$ | $\times^{\text {Note } 1}$ |  | Internal <br> clock |  |  | P20 |
| Other than above |  |  |  |  |  |  |  |  |  |  | Setting prohibited |  |  |  |  |

Notes 1. These pins can be used for port functions.
2. When only transmission is used, this pin can be used as P22 (CMOS I/O).

Remark $\times$ : Don’t care.

## (3) Asynchronous serial interface status register 20 (ASIS20)

ASIS20 indicates the type of a reception error, if an error occurs while asynchronous serial interface mode is set.
ASIS20 is set with a 1-bit or 8-bit memory manipulation instruction.
The contents of ASIS20 are undefined in 3-wire serial I/O mode.
RESET input clears ASIS20 to 00H.

Figure 14-5. Format of Asynchronous Serial Interface Status Register 20

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIS20 | 0 | 0 | 0 | 0 | 0 | PE20 | FE20 | OVE20 | FF71H | OOH | R |


| PE20 | Parity error flag |
| :---: | :--- |
| 0 | No parity error has occurred. |
| 1 | A parity error has occurred (when the transmit parity and receive parity do not match). |


| FE20 | Flaming error flag |
| :---: | :--- |
| 0 | No framing error has occurred. |
| 1 | A framing error has occurred (when stop bit is not detected). Note 1 |


| OVE20 | Overrun error flag |
| :---: | :--- |
| 0 | No overrun error has occurred. |
| 1 | An overrun error has occurred. <br> (whete 2 |
| (when the next receive operation is completed before the data is read from reception buffer register 20) |  |

Notes 1. Even when the stop bit length is set to 2 bits by setting bit 2 (SL20) of asynchronous serial interface mode register 20 (ASIM20), the stop bit detection at reception is performed with 1 bit.
2. Be sure to read reception buffer register 20 (RXB20) when an overrun error occurs. If not, every time the data is received an overrun error occurs.
(4) Baud rate generator control register 20 (BRGC20)

BRGC20 is used to specify the serial clock for serial interface 20.
BRGC20 is set with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears BRGC20 to 00H.

Figure 14-6. Format of Baud Rate Generator Control Register 20

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BRGC20 | TPS203 | TPS202 | TPS201 | TPS200 | 0 | 0 | 0 | 0 | FF73H | 00 H | R/W |


| TPS203 | TPS202 | TPS201 | TPS200 |  | -bit counter source cloc | election | n |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | $\begin{gathered} \text { At } \mathrm{fx}=10.0 \mathrm{MHz} \\ \text { operation }^{\text {Note } 1} \end{gathered}$ | At fX $=5.0 \mathrm{MHz}$ operation |  |
| 0 | 0 | 0 | 0 | $\mathrm{fx} / 2$ | 5.00 MHz | 2.50 MHz | 1 |
| 0 | 0 | 0 | 1 | $\mathrm{fx} / 2^{2}$ | 2.50 MHz | 1.25 MHz | 2 |
| 0 | 0 | 1 | 0 | $\mathrm{fx} / 2^{3}$ | 1.25 MHz | 625 kHz | 3 |
| 0 | 0 | 1 | 1 | $\mathrm{fx} / 2^{4}$ | 625 kHz | 313 kHz | 4 |
| 0 | 1 | 0 | 0 | $\mathrm{fx} / 2^{5}$ | 313 kHz | 156 kHz | 5 |
| 0 | 1 | 0 | 1 | fx/2 ${ }^{6}$ | 156 kHz | 78.1 kHz | 6 |
| 0 | 1 | 1 | 0 | $\mathrm{fx} / 2{ }^{7}$ | 78.1 kHz | 39.1 kHz | 7 |
| 0 | 1 | 1 | 1 | $\mathrm{fx} / 2^{8}$ | 39.1 kHz | 19.5 kHz | 8 |
| 1 | 0 | 0 | 0 | External clock input to the ASCK20 pin ${ }^{\text {Note } 2}$ |  |  | - |
| Other than above |  |  |  | Setting prohibited |  |  |  |

Notes 1. Expanded-specification products only.
2. An external clock can be used only in UART mode.

Cautions 1. When writing to BRGC00 is performed during a communication operation, the output of baud rate generator is disrupted and communications cannot be performed normally. Be sure not to write to BRGC00 during communication operations.
2. Be sure not to select $\mathrm{n}=1$ in UART mode when $\mathrm{fx}>2.5 \mathrm{MHz}$ because the baud rate will exceed the rated range.
3. Be sure not to select $\mathbf{n}=\mathbf{2}$ in UART mode when $\mathrm{fx}>5.0 \mathrm{MHz}$ because the baud rate will exceed the rated range.
4. Be sure not to select $\mathbf{n}=1$ in 3 -wire serial $\mathrm{I} / \mathrm{O}$ mode when $\mathrm{fx}>5.0 \mathrm{MHz}$ because the serial clock specification will be exceeded.
5. When the external input clock is selected, set input mode by setting bit 0 of port mode register 2 (PM2) to 1.

Remarks 1. fx: Main system clock oscillation frequency
2. $n$ : Values determined by the settings of TPS200 to TPS203 $(1 \leq n \leq 8)$

The baud rate transmit/receive clock to be generated is either a signal scaled from the system clock, or a signal scaled from the clock input to the ASCK20 pin.
(a) Generation of baud rate transmit/receive clock from system clock

The transmit/receive clock is generated by scaling the system clock. The baud rate of a clock generated from the system clock is estimated by using the following expression.
$\left[\right.$ Baud rate] $=\frac{\mathrm{fx}}{2^{n+1} \times 8}[\mathrm{bps}]$
fx: Main system clock oscillation frequency
n : Values in Figure 14-6, determined by the values of TPS200 to TPS203 $(2 \leq \mathrm{n} \leq 8)$

Table 14-3. Example of Relationship Between System Clock and Baud Rate

| Baud Rate (bps) | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ |  |  | At fx $=5.0 \mathrm{MHz}$ |  |  | At $\mathrm{fx}=4.9152 \mathrm{MHz}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | n | BRGC20 <br> Set Value | Error <br> (\%) | n | BRGC20 <br> Set Value | Error <br> (\%) | n | BRGC20 <br> Set Value | Error <br> (\%) |
| 1,200 | - | - | 1.73 | 8 | 70H | 1.73 | 8 | 70 H | 0 |
| 2,400 | 8 | 70 H |  | 7 | 60 H |  | 7 | 60 H |  |
| 4,800 | 7 | 60 H |  | 6 | 50 H |  | 6 | 50 H |  |
| 9,600 | 6 | 50 H |  | 5 | 40 H |  | 5 | 40 H |  |
| 19,200 | 5 | 40 H |  | 4 | 30 H |  | 4 | 30 H |  |
| 38,400 | 4 | 30 H |  | 3 | 20 H |  | 3 | 20 H |  |
| 76,800 | 3 | 20 H |  | 2 | 10H |  | 2 | 10H |  |

Note Expanded-specification products only.

Cautions 1. Be sure not to select $\mathbf{n}=1$ during operation at $\mathrm{fx}>2.5 \mathrm{MHz}$ because the baud rate will exceed the rated range.
2. Be sure not to select $\mathbf{n}=\mathbf{2}$ during operation at $\mathrm{fx}>5.0 \mathrm{MHz}$ because the baud rate will exceed the rated range.
(b) Generation of baud rate transmit/receive clock from external clock input to ASCK20 pin

The transmit/receive clock is generated by scaling the clock input from the ASCK20 pin. The baud rate of a clock generated from the clock input to the ASCK20 pin is calculated by using the following expression.
[Baud rate] $=\frac{\mathrm{f}_{\mathrm{ASCK}}}{16}[\mathrm{bps}]$
fasck: Frequency of clock input to the ASCK20 pin

Table 14-4. Relationship Between ASCK20 Pin Input Frequency and Baud Rate (When BRGC20 Is Set to 80H)

| Baud Rate (bps) | ASCK20 Pin Input Frequency (kHz) |
| :---: | :---: |
| 75 | 1.2 |
| 150 | 2.4 |
| 300 | 4.8 |
| 600 | 9.6 |
| 1,200 | 19.2 |
| 2,400 | 38.4 |
| 4,800 | 76.8 |
| 9,600 | 153.6 |
| 19,200 | 307.2 |
| 31,250 | 500.0 |
| 38,400 | 614.4 |

(c) Generation of 3-wire serial I/O mode serial clock from system clock

The serial clock is generated by scaling the system clock. The serial clock frequency is calculated by using the following expression. BRGC20 does not have to be set when the serial clock is input to the $\overline{\text { SCK20 }}$ pin externally.

Serial clock frequency $=\frac{\mathrm{fx}}{2^{n+1}}[\mathrm{~Hz}]$
fx: System clock oscillation frequency
n : Value determined by settings of TPS200 to TPS203 in Fig. 14-6.

### 14.4 Operation of Serial Interface 20

Serial interface 20 provides the following three modes.

- Operation stop mode
- Asynchronous serial interface (UART) mode
- 3-wire serial I/O mode


### 14.4.1 Operation stop mode

In operation stop mode, serial transfer is not executed; therefore, the power consumption can be reduced. The P20/SCK20/ASCK20, P21/SO20/TxD20, and P22/SI20/RxD20 pins can be used as normal I/O ports.

## (1) Register setting

Operation stop mode is set by serial operation mode register 20 (CSIM20) and asynchronous serial interface mode register 20 (ASIM20).
(a) Serial operation mode register 20 (CSIM20)

CSIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears CSIM20 to 00H.

| Symbol | <7> | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSIM20 | CSIE20 | SSE20 | 0 | 0 | DAP20 | DIR20 | CSCK20 | CKP20 | FF72H | OOH | R/W |


| CSIE20 |  | Operation control in 3-wire serial I/O mode |
| :---: | :--- | :--- |
| 0 | Operation disabled |  |
| 1 | Operation enabled |  |

## Caution Bits 4 and 5 must be set to 0 .

(b) Asynchronous serial interface mode register 20 (ASIM20)

ASIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ASIM20 to 00H.

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIM20 | TXE20 | RXE20 | PS201 | PS200 | CL20 | SL20 | 0 | 0 | FF70H | OOH | R/W |


| TXE20 |  | Transmit operation control |
| :---: | :--- | :--- |
| 0 | Transmit operation stopped |  |
| 1 | Transmit operation enabled |  |


| RXE20 |  | Receive operation control |
| :---: | :--- | :--- |
| 0 | Receive operation stopped |  |
| 1 | Receive operation enabled |  |

## Caution Bits 0 and 1 must be set to 0 .

### 14.4.2 Asynchronous serial interface (UART) mode

In this mode, the one-byte data following the start bit is transmitted/received and thus full-duplex communication is possible.

This device incorporates an UART-dedicated baud rate generator that enables communications at the desired baud rate from many options. In addition, the baud rate can also be defined by dividing the clock input to the ASCK20 pin.

The UART-dedicated baud rate generator also can output the 31.25 kbps baud rate that complies with the MIDI standard.

## (1) Register setting

UART mode is set by serial operation mode register 20 (CSIM20), asynchronous serial interface mode register 20 (ASIM20), asynchronous serial interface status register 20 (ASIS20), baud rate generator control register 20 (BRGC20), port mode register 2 (PM2), and port $2(\mathrm{P} 2)$.
(a) Serial operation mode register 20 (CSIM20)

CSIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears CSIM20 to 00H.
Set CSIM20 to 00H when UART mode is selected.

| Symbol $<7>$ | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSIM20 | CSIE20 | SSE20 | 0 | 0 | DAP20 | DIR20 | CSCK20 | CKP20 | FF72H | $00 H$ | R/W |


| CSIE20 |  | 3-wire serial I/O mode operation control |
| :---: | :--- | :--- |
| 0 | Operation disabled |  |
| 1 | Operation enabled |  |


| SSE20 | $\overline{\text { SS20-pin selection }}$ | Function of $\overline{\text { SS20 }} /$ P25 pin | Communication status |
| :---: | :--- | :---: | :--- |
| 0 | Not used | Port function | Communication enabled |
| 1 | Used | 0 | Communication enabled |
|  |  | 1 | Communication disabled |


| DAP20 | 3-wire serial I/O mode data phase selection |
| :---: | :--- |
| 0 | Outputs at the falling edge of $\overline{\text { SCK20. }}$ |
| 1 | Outputs at the rising edge of $\overline{\text { SCK20. }}$ |


| DIR20 |  | First-bit specification |
| :---: | :--- | :--- |
| 0 | MSB |  |
| 1 | LSB |  |


| CSCK20 | 3-wire serial I/O mode clock selection |
| :---: | :--- |
| 0 | External clock input to the $\overline{\text { SCK20 }}$ pin |
| 1 | Output of the dedicated baud rate generator |


| CKP20 | 3-wire serial I/O mode clock phase selection |
| :---: | :--- |
| 0 | Clock is active-low, and $\overline{\text { SCK20 }}$ is high level in the idle state. |
| 1 | Clock is active-high, and $\overline{\text { SCK20 }}$ is low level in the idle state. |

## Cautions 1. Bits 4 and 5 must be set to 0 .

2. CSIM20 must be cleared to 00 H , if UART mode is selected.
3. Switch operating modes after halting serial transmit/receive operation.
(b) Asynchronous serial interface mode register 20 (ASIM20)

ASIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ASIM20 to 00H.

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIM20 | TXE20 | RXE20 | PS201 | PS200 | CL20 | SL20 | 0 | 0 | FF70H | OOH | R/W |


| TXE20 |  | Transmit operation control |
| :---: | :--- | :--- |
| 0 | Transmit operation stop |  |
| 1 | Transmit operation enable |  |


| RXE20 |  | Receive operation control |
| :---: | :--- | :--- |
| 0 | Receive operation stop |  |
| 1 | Receive operation enable |  |


| PS201 | PS200 | Parity bit specification |
| :---: | :---: | :--- |
| 0 | 0 | No parity |
| 0 | 1 | Always add 0 parity at transmission. <br> Parity check is not performed at reception (no parity error is generated). |
| 1 | 0 | Odd parity |
| 1 | 1 | Even parity |


| CL20 |  | Character length specification |
| :---: | :--- | :--- |
| 0 | 7 bits |  |
| 1 | 8 bits |  |


| SL20 |  | Transmit data stop bit length specification |
| :---: | :--- | :--- |
| 0 | 1 bit |  |
| 1 | 2 bits |  |

Cautions 1. Bits 0 and 1 must be set to 0 .
2. Switch operating modes after halting the serial transmit/receive operation.
(c) Asynchronous serial interface status register 20 (ASIS20)

ASIS20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ASIS20 to 00H.

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIS20 | 0 | 0 | 0 | 0 | 0 | PE20 | FE20 | OVE20 | FF71H | 00H | R |


| PE20 | Parity error flag |
| :---: | :--- |
| 0 | Parity error not generated |
| 1 | Parity error generated (when the parity of transmit data does not match) |


| FE20 | Framing error flag |
| :---: | :--- |
| 0 | Framing error not generated |
| 1 | Framing error generated (when stop bit is not detected) ${ }^{\text {Note } 1}$ |


| OVE20 | Overrun error flag |
| :---: | :--- |
| 0 | Overrun error not generated |
| 1 | Overrun error generatedNote 2 <br> (when the next receive operation is completed before data is read from reception buffer register 20) |

Notes 1. Even when the stop bit length is set to 2 bits by setting bit 2 (SL20) of asynchronous serial interface mode register 20 (ASIM20), the stop bit detection at reception is performed with 1 bit.
2. Be sure to read reception buffer register 20 (RXB20) when an overrun error occurs. If not, every time the data is received an overrun error is generated.
(d) Baud rate generator control register 20 (BRGC20)

BRGC20 is set with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears BRGC20 to 00 H .

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BRGC20 | TPS203 | TPS202 | TPS201 | TPS200 | 0 | 0 | 0 | 0 | FF73H | OOH | R/W |


| TPS203 | TPS202 | TPS201 | TPS200 |  | 3 -bit counter source cla | selection | n |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | $\begin{gathered} \text { At } \mathrm{fx}=10.0 \mathrm{MHz} \\ \text { operation }^{\text {Note }} \end{gathered}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |  |
| 0 | 0 | 0 | 0 | fx/2 | 5.00 MHz | 2.50 MHz | 1 |
| 0 | 0 | 0 | 1 | $\mathrm{fx} / 2^{2}$ | 2.50 MHz | 1.25 MHz | 2 |
| 0 | 0 | 1 | 0 | $\mathrm{fx} / 2^{3}$ | 1.25 MHz | 625 kHz | 3 |
| 0 | 0 | 1 | 1 | $\mathrm{fx} / 2^{4}$ | 625 kHz | 313 kHz | 4 |
| 0 | 1 | 0 | 0 | $\mathrm{fx} / 2^{5}$ | 313 kHz | 156 kHz | 5 |
| 0 | 1 | 0 | 1 | $\mathrm{fx} / 2^{6}$ | 156 kHz | 78.1 kHz | 6 |
| 0 | 1 | 1 | 0 | $\mathrm{fx} / 2^{7}$ | 78.1 kHz | 39.1 kHz | 7 |
| 0 | 1 | 1 | 1 | $\mathrm{fx} / 2^{8}$ | 39.1 kHz | 19.5 kHz | 8 |
| 1 | 0 | 0 | 0 | External clock input to the ASCK20 pin |  |  | - |
| Other than above |  |  |  | Setting prohibited |  |  |  |

Note Expanded-specification products only.

Cautions 1. When writing to BRGC20 is performed during a communication operation, the output of the baud rate generator is disrupted and communications cannot be performed normally. Be sure not to write to BRGC2O during a communication operation.
2. Be sure not to select $\mathbf{n}=1$ during operation at $\mathrm{fx}>2.5 \mathrm{MHz}$ because the baud rate will exceed the rated range.
3. Be sure not to select $\mathbf{n}=2$ during operation at $\mathrm{f} x>5.0 \mathrm{MHz}$ because the baud rate will exceed the rated range.
4. When the external input clock is selected, set input mode by setting bit 0 of port mode register 2 (PM2) to 1.

Remarks 1. fx : Main system clock oscillation frequency
2. n : Values determined by the settings of TPS200 to TPS203 $(1 \leq \mathrm{n} \leq 8)$

The baud rate transmit/receive clock to be generated is either a signal scaled from the system clock, or a signal scaled from the clock input to the ASCK20 pin.
(i) Generation of baud rate transmit/receive clock from system clock

The transmit/receive clock is generated by scaling the system clock. The baud rate of a clock generated from the system clock is estimated by using the following expression.
$[$ Baud rate $]=\frac{f x}{2^{n+1} \times 8}[b p s]$
fx: Main system clock oscillation frequency
n : Values in the above table determined by the settings of TPS200 to TPS203 $(2 \leq \mathrm{n} \leq 8)$

Table 14-5. Example of Relationship Between System Clock and Baud Rate

| Baud Rate (bps) | At $\mathrm{fx}=10.0 \mathrm{MHz}^{\text {Note }}$ |  |  | At fx $=5.0 \mathrm{MHz}$ |  |  | At $\mathrm{fx}=4.9152 \mathrm{MHz}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | n | BRGC20 <br> Set Value | Error <br> (\%) | n | BRGC20 <br> Set Value | Error <br> (\%) | n | BRGC20 <br> Set Value | Error <br> (\%) |
| 1,200 | - | - | 1.73 | 8 | 70 H | 1.73 | 8 | 70 H | 0 |
| 2,400 | 8 | 70H |  | 7 | 60H |  | 7 | 60H |  |
| 4,800 | 7 | 60 H |  | 6 | 50 H |  | 6 | 50 H |  |
| 9,600 | 6 | 50 H |  | 5 | 40 H |  | 5 | 40 H |  |
| 19,200 | 5 | 40 H |  | 4 | 30 H |  | 4 | 30 H |  |
| 38,400 | 4 | 30 H |  | 3 | 20 H |  | 3 | 20 H |  |
| 76,800 | 3 | 20 H |  | 2 | 10H |  | 2 | 10H |  |

Note Expanded-specification products only.

Cautions 1. Be sure not to select $\mathbf{n}=1$ during operation at $\mathrm{fx}>2.5 \mathrm{MHz}$ because the baud rate will exceed the rated range.
2. Be sure not to select $\mathbf{n}=2$ during operation at $\mathrm{fx}>5.0 \mathrm{MHz}$ because the baud rate will exceed the rated range.
(ii) Generation of baud rate transmit/receive clock from external clock input to ASCK20 pin

The transmit/receive clock is generated by scaling the clock input from the ASCK20 pin. The baud rate of a clock generated from the clock input to the ASCK20 pin is estimated by using the following expression.
$\left[\right.$ Baud rate] $=\frac{\mathrm{f}_{\text {ASCK }}}{16}[\mathrm{bps}]$
$f_{\text {ASck: }}$ Frequency of clock input to ASCK20 pin

Table 14-6. Relationship Between ASCK20 Pin Input Frequency and Baud Rate (When BRGC20 Is Set to 80H)

| Baud Rate (bps) | ASCK20 Pin Input Frequency (kHz) |
| :---: | :---: |
| 75 | 1.2 |
| 150 | 2.4 |
| 300 | 4.8 |
| 600 | 9.6 |
| 1,200 | 19.2 |
| 2,400 | 38.4 |
| 4,800 | 76.8 |
| 9,600 | 153.6 |
| 19,200 | 307.2 |
| 31,250 | 500.0 |
| 38,400 | 614.4 |

## (2) Communication operation

## (a) Data format

The transmit/receive data format is as shown in Figure 14-7. One data frame consists of a start bit, character bits, parity bit, and stop bit(s).
The specification of character bit length in one data frame, parity selection, and specification of stop bit length is carried out with asynchronous serial interface mode register 20 (ASIM20).

Figure 14-7. Asynchronous Serial Interface Transmit/Receive Data Format


- Start bits $\qquad$ 1 bit
- Character bits............ 7 bits/8 bits
- Parity bits .................. Even parity/odd parity/0 parity/no parity
- Stop bit(s) $\qquad$ 1 bit/2 bits

When 7 bits are selected as the number of character bits, only the lower 7 bits (bits 0 to 6 ) are valid; in transmission the most significant bit (bit 7) is ignored, and in reception the most significant bit (bit 7) is always " 0 ".
The serial transfer rate is selected by baud rate generator control register 20 (BRGC20).
If a serial data receive error is generated, the receive error contents can be determined by reading the status of asynchronous serial interface status register 20 (ASIS20).

## (b) Parity types and operation

The parity bit is used to detect a bit error in the communication data. Normally, the same kind of parity bit is used on the transmitting side and the receiving side. With even parity and odd parity, a one-bit (odd number) error can be detected. With 0 parity and no parity, an error cannot be detected.

## (i) Even parity

- At transmission

The parity bit is determined so that the number of bits with a value of "1" in the transmit data including the parity bit is even. The parity bit value should be as follows.

The number of bits with a value of " 1 " is an odd number in transmit data: 1
The number of bits with a value of " 1 " is an even number in transmit data: 0

## - At reception

The number of bits with a value of " 1 " in the receive data including parity bit is counted, and if the number is odd, a parity error is generated.

## (ii) Odd parity

- At transmission

Conversely to even parity, the parity bit is determined so that the number of bits with a value of " 1 " in the transmit data including parity bit is odd. The parity bit value should be as follows.

The number of bits with a value of " 1 " is an odd number in transmit data: 0
The number of bits with a value of " 1 " is an even number in transmit data: 1

## - At reception

The number of bits with a value of " 1 " in the receive data including parity bit is counted, and if the number is even, a parity error is generated.

## (iii) 0 parity

When transmitting, the parity bit is set to " 0 " irrespective of the transmit data.
At reception, a parity bit check is not performed. Therefore, a parity error is not generated, irrespective of whether the parity bit is set to " 0 " or " 1 ".

## (iv) No parity

A parity bit is not added to the transmit data. At reception, data is received assuming that there is no parity bit. Since there is no parity bit, a parity error is not generated.
(c) Transmission

A transmit operation is started by writing transmit data to transmission shift register 20 (TXS20). The start bit, parity bit, and stop bit(s) are added automatically.
When the transmit operation starts, the data in TXS20 is shifted out, and when TXS20 is empty, a transmission completion interrupt (INTST20) is generated.

Figure 14-8. Asynchronous Serial Interface Transmission Completion Interrupt Timing
(a) Stop bit length: 1

(b) Stop bit length: 2

TxD20 (output)


Caution Do not rewrite asynchronous serial interface mode register 20 (ASIM20) during a transmit operation. If the ASIM20 register is rewritten to during transmission, subsequent transmission may not be performed (the normal state is restored by RESET input).
It is possible to determine whether transmission is in progress by software by using a transmission completion interrupt (INTST20) or the interrupt request flag (STIF20) set by INTST20.

## (d) Reception

When bit 6 (RXE20) of asynchronous serial interface mode register 20 (ASIM20) is set to 1, a receive operation is enabled and sampling of the RxD20 pin input is performed.
RxD20 pin input sampling is performed using the serial clock specified by BRGC20.
When the RxD20 pin input becomes low, the 3-bit counter starts counting, and at the time when half the time determined by the specified baud rate has passed, the data sampling start timing signal is output. If the RxD20 pin input sampled again as a result of this start timing signal is low, it is identified as a start bit, the 3-bit counter is initialized and starts counting, and data sampling is performed. When character data, a parity bit, and one stop bit are detected after the start bit, reception of one frame of data ends.
When one frame of data has been received, the receive data in the shift register is transferred to reception buffer register 20 (RXB20), and a reception completion interrupt (INTSR20) is generated.
If an error is generated, the receive data in which the error was generated is still transferred to RXB20, and INTSR20 is generated.
If the RXE20 bit is reset to 0 during the receive operation, the receive operation is stopped immediately. In this case, the contents of RXB20 and asynchronous serial interface status register 20 (ASIS20) are not changed, and INTSR20 is not generated.

Figure 14-9. Asynchronous Serial Interface Reception Completion Interrupt Timing


Caution Be sure to read reception buffer register 20 (RXB20) even if a receive error occurs. If RXB20 is not read, an overrun error will be generated when the next data is received, and the receive error state will continue indefinitely.

## (e) Receive errors

The following three errors may occur during a receive operation: a parity error, framing error, and overrun error. After data reception, an error flag is set in asynchronous serial interface status register 20 (ASIS20). Receive error causes are shown in Table 14-7.
It is possible to determine what kind of error occurred during reception by reading the contents of ASIS20 in the reception error interrupt servicing (see Figures 14-9 and 14-10).
The contents of ASIS20 are reset to 0 by reading reception buffer register 20 (RXB20) or receiving the next data (if there is an error in the next data, the corresponding error flag is set).

Table 14-7. Receive Error Causes

| Receive Errors | Cause |
| :--- | :--- |
| Parity error | Transmission-time parity and reception data parity do not match |
| Framing error | Stop bit not detected |
| Overrun error | Reception of next data is completed before data is read from reception buffer register |

Figure 14-10. Receive Error Timing

## (a) Parity error occurred


(b) Framing error or overrun error occurred


Cautions 1. The contents of the ASIS20 register are reset to 0 by reading reception buffer register 20 (RXB20) or receiving the next data. To ascertain the error contents, read ASIS20 before reading RXB20.
2. Be sure to read reception buffer register 20 ( RXB 20 ) even if a receive error occurred. If RXB20 is not read, an overrun error will occur when the next data is received, and the receive error state will continue indefinitely.

## (f) Reading receive data

When the reception completion interrupt (INTSR20) occurs, receive data can be read by reading the value of reception buffer register 20 (RXB20).
To read the receive data stored in reception buffer register 20 (RXB20), read while reception is enabled (RXE20 = 1).

Remark However, if it is necessary to read receive data after reception has stopped (RXE20 $=0$ ), read using either of the following methods.
(a) Read after setting RXE20 $=0$ after waiting for one cycle or more of the source clock selected by BRGC20.
(b) Read after bit 2 (DIR20) of serial operation mode register 20 (CSIM20) is set (1).
Program example of (a) (BRGC20 $=00 \mathrm{H}$ (source clock $=\mathrm{fx} / 2)$ )
INTREX:

| NOP | $;$ Reception completion interrupt routine> |
| :--- | :--- |
| CLR1 RXE20 | $; 2$ clocks |
| MOV A, RXB20 | $;$ Reception stopped receive data |

Program example of (b)

| INTRXE: | ;<Reception completion interrupt routine> |  |
| :--- | :--- | :--- |
| SET1 | CSIM20.2 | ;DIR20 flag is set to LSB first |
| CLR1 | RXE20 | ;Reception stopped |
| MOV | A, RXB20 | ;Read receive data |

## (3) Cautions related to UART mode

(a) When bit 7 (TXE20) of asynchronous serial interface mode register 20 (ASIM20) is cleared during transmission, be sure to set transmission shift register 20 (TXS20) to FFH, then set TXE20 to 1 before executing the next transmission.
(b) When bit 6 (RXE20) of asynchronous serial interface mode register 20 (ASIM20) is cleared during reception, reception buffer register 20 (RXB20) and the reception completion interrupt (INTSR20) are as follows.


When RXE20 is set to 0 at the time indicated by $<1>$, RXB20 holds the previous data and INTSR20 is not generated.
When RXE20 is set to 0 at the time indicated by <2>, RXB20 renews the data and INTSR20 is not generated.
When RXE20 is set to 0 at the time indicated by <3>, RXB20 renews the data and INTSR20 is generated.

### 14.4.3 3-wire serial I/O mode

The 3 -wire serial I/O mode is useful for connection of peripheral I/Os and display controllers, etc., which incorporate a conventional synchronous serial interface, such as the 75XL Series, 78K Series, 17K Series, etc.

Communication is performed using three lines: the serial clock (SCK20), serial output (SO20), and serial input (SI20).

## (1) Register setting

3 -wire serial I/O mode settings are performed using serial operation mode register 20 (CSIM20), asynchronous serial interface mode register 20 (ASIM20), baud rate generator control register 20 (BRGC20), port mode register 2 (PM2), and port 2 (P2).
(a) Serial operation mode register 20 (CSIM20)

CSIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears CSIM20 to 00H.

| Symbol | <7> | 6 | 5 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| CSIM20 | CSIE20 | SSE20 | 0 | DAP20 | DIR20 | CSCK20 | CKP20 | FF72H | 00H | R/W |
|  | CSIE20 |  |  | 3-wire | erial I/O | mode ope | ration con |  |  |  |
|  | 0 | Operatio | isa |  |  |  |  |  |  |  |
|  | 1 | Operatio | na |  |  |  |  |  |  |  |


| SSE20 | $\overline{\text { SS20-pin selection }}$ | Function of $\overline{\text { SS20 }} / \mathrm{P} 25$ pin | Communication status |
| :---: | :--- | :--- | :--- |
| 0 | Not used | Port function | Communication enabled |
| 1 | Used | 0 | Communication enabled |
|  |  | 1 | Communication disabled |


| DAP20 | 3-wire serial I/O mode data phase selection |
| :---: | :--- |
| 0 | Outputs at the falling edge of SCK20. |
| 1 | Outputs at the rising edge of $\overline{\text { SCK20. }}$ |


| DIR20 | First-bit specification |  |
| :---: | :--- | :--- |
| 0 | MSB |  |
| 1 | LSB |  |


| CSCK20 | 3 -wire serial I/O mode clock selection |
| :---: | :--- |
| 0 | External clock input to the $\overline{\text { SCK20 }}$ pin |
| 1 | Output of the dedicated baud rate generator |


| CKP20 | 3-wire serial I/O mode clock phase selection |
| :---: | :--- |
| 0 | Clock is active-low, and $\overline{\text { SCK20 }}$ is at high level in the idle state. |
| 1 | Clock is active-high, and $\overline{\text { SCK20 }}$ is at low level in the idle state. |

Cautions 1. Bits 4 and 5 must be set to 0 .
2. Switch operating modes after halting the serial transmit/receive operation.
3. When the external input clock is selected in 3-wire serial I/O mode, set input mode by setting bit 0 of port mode register 2 (PM2) to 1.
(b) Asynchronous serial interface mode register 20 (ASIM20)

ASIM20 is set with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears ASIM20 to 00H.

| Symbol | <7> | <6> | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASIM20 | TXE20 | RXE20 | PS201 | PS200 | CL20 | SL20 | 0 | 0 | FF70H | OOH | R/W |


| TXE20 |  | Transmit operation control |
| :---: | :--- | :--- |
| 0 | Transmit operation stop |  |
| 1 | Transmit operation enable |  |


| RXE20 |  | Receive operation control |
| :---: | :--- | :--- |
| 0 | Receive operation stop |  |
| 1 | Receive operation enable |  |


| PS201 | PS200 | Parity bit specification |
| :---: | :---: | :--- |
| 0 | 0 | No parity |
| 0 | 1 | Always add 0 parity at transmission. <br> Parity check is not performed at reception (no parity error is generated). |
| 1 | 0 | Odd parity |
| 1 | 1 | Even parity |


| CL20 |  | Character length specification |
| :---: | :--- | :--- |
| 0 | 7 bits |  |
| 1 | 8 bits |  |


| SL20 |  | Transmit data stop bit length specification |
| :---: | :--- | :--- |
| 0 | 1 bit |  |
| 1 | 2 bits |  |

Cautions 1. Bits 0 and 1 must be set to 0 .
2. ASIM20 must be cleared to 00 H if 3 -wire serial I/O mode is selected.
3. Switch operating modes after halting serial transmit/receive operation.
(c) Baud rate generator control register 20 (BRGC20)

BRGC20 is set with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears BRGC20 to 00 H .

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| BRGC20 | TPS203 | TPS202 | TPS201 | TPS200 | 0 | 0 | 0 | 0 | FF73H | OOH | W |


| TPS203 | TPS202 | TPS201 | TPS200 | 3-bit counter source clock selection |  |  | n |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | $\begin{gathered} \text { At } \mathrm{fx}_{\mathrm{x}}=10.0 \mathrm{MHz} \\ \text { operation }^{\text {Note }} \end{gathered}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |  |
| 0 | 0 | 0 | 0 | fx/2 | 5.00 MHz | 2.50 MHz | 1 |
| 0 | 0 | 0 | 1 | $\mathrm{fx} / 2^{2}$ | 2.50 MHz | 1.25 MHz | 2 |
| 0 | 0 | 1 | 0 | $\mathrm{fx} / 2^{3}$ | 1.25 MHz | 625 kHz | 3 |
| 0 | 0 | 1 | 1 | $\mathrm{fx} / 2^{4}$ | 625 kHz | 313 kHz | 4 |
| 0 | 1 | 0 | 0 | $\mathrm{fx} / 2^{5}$ | 313 kHz | 156 kHz | 5 |
| 0 | 1 | 0 | 1 | $\mathrm{fx} / 2^{6}$ | 156 kHz | 78.1 kHz | 6 |
| 0 | 1 | 1 | 0 | $\mathrm{fx} / 2^{7}$ | 78.1 kHz | 39.1 kHz | 7 |
| 0 | 1 | 1 | 1 | $\mathrm{fx} / 2^{8}$ | 39.1 kHz | 19.5 kHz | 8 |
| Other than above |  |  |  | Setting prohibited |  |  |  |

Note Expanded-specification products only.

Cautions 1. When writing to BRGC20 is performed during a communication operation, the baud rate generator output is disrupted and communications cannot be performed normally. Be sure not to write to BRGC20 during a communication operation.
2. If $\mathrm{f} x>5.0 \mathrm{MHz}$ in the 3 -wire serial $\mathrm{I} / \mathrm{O}$ mode, this setting is prohibited because $\mathbf{n}=\mathbf{1}$ exceeds the rated range of the serial clock.

Remarks 1. fx: Main system clock oscillation frequency
2. $n$ : Values determined by the settings of TPS200 to TPS203 $(1 \leq n \leq 8)$

If the internal clock is used as the serial clock for 3-wire serial I/O mode, set the TPS200 to TPS203 bits to set the frequency of the serial clock. To obtain the frequency to be set, use the following expression. When an external clock is used, setting BRGC20 is not necessary.

$$
\text { Serial clock frequency }=\frac{f x}{2^{n+1}}[H z]
$$

fx: Main system clock oscillation frequency
n : Values in the above table determined by the settings of TPS200 to TPS203 $(1 \leq \mathrm{n} \leq 8)$

## (2) Communication operation

In 3-wire serial I/O mode, data transmission/reception is performed in 8-bit units. Data is transmitted/received bit by bit in synchronization with the serial clock.
Transmission shift register (TXS20/SIO20) and reception shift register (RXS20) shift operations are performed in synchronization with the fall of the serial clock ( $\overline{\mathrm{SCK2O}}$ ). Then transmit data is held in the SO20 latch and output from the SO20 pin. Also, receive data input to the SI20 pin is latched in the reception buffer register (RXB20/SIO20) on the rise of SCK20.
At the end of an 8-bit transfer, the operation of TXS20/SIO20 and RXS20 stops automatically, and the interrupt request signal (INTCSI20) is generated.

Figure 14-11. 3-Wire Serial I/O Mode Timing (1/7)
(i) Master operation timing (when DAP20 $=0, C K P 20=0, S S E 20=0$ )


Note The value of the last bit previously output is output.

Figure 14-11. 3-Wire Serial I/O Mode Timing (2/7)
(ii) Slave operation timing (when DAP20 $=0$, CKP20 $=0$, SSE20 $=0$ )


Note The value of the last bit previously output is output.


Notes 1. The value of the last bit previously output is output.
2. DOO is output until $\overline{\mathrm{SS} 20}$ rises.

When $\overline{\mathrm{SS} 20}$ is high, SO20 is in a high-impedance state.

Figure 14-11. 3-Wire Serial I/O Mode Timing (3/7)
(iv) Master operation (when DAP20 $=0$, CKP20 $=1$, SSE20 $=0$ )

(v) Slave operation (when DAP20 $=0$, CKP20 $=1$, SSE20 $=0$ )


Note The data of SI 20 is loaded at the first rising edge of $\overline{\text { SCK20 }}$. Make sure that the master outputs the first bit before the first rising of $\overline{\text { SCK20 }}$.

Figure 14-11. 3-Wire Serial I/O Mode Timing (4/7)
(vi) Slave operation (when DAP20 = 0, CKP20 = 1, SSE20 = 1)


Notes 1. The data of SI 20 is loaded at the first rising edge of $\overline{\mathrm{SCK2O}}$. Make sure that the master outputs the first bit before the first rising of $\overline{\text { SCK20 }}$.
2. SO20 is high until $\overline{\mathrm{SS} 20}$ rises after completion of DOO output. When $\overline{\mathrm{SS} 20}$ is high, SO20 is in a high-impedance state.
(vii) Master operation (when DAP20 $=1, C K P 20=0$, SSE20 $=0$ )


Figure 14-11. 3-Wire Serial I/O Mode Timing (5/7)
(viii) Slave operation (when DAP20 $=1$, CKP20 $=0$, SSE20 $=0$ )


Note The data of SI 20 is loaded at the first falling edge of SCK20. Make sure that the master outputs the first bit before the first falling of $\overline{\text { SCK20. }}$


Notes 1. The data of SI 20 is loaded at the first falling edge of $\overline{\mathrm{SCK2O}}$. Make sure that the master outputs the first bit before the first falling of $\overline{\text { SCK20. }}$
2. SO20 is high until $\overline{\mathrm{SS} 20}$ rises after completion of DOO output. When $\overline{\mathrm{SS2O}}$ is high, SO20 is in a high-impedance state.

Figure 14-11. 3-Wire Serial I/O Mode Timing (6/7)
(x) Master operation (when DAP20 = 1, CKP20 = 1, SSE20 = 0)


Note The value of the last bit previously output is output.
(xi) Slave operation (when DAP20 = 1, CKP20 = 1, SSE20 = 0)


Note The value of the last bit previously output is output.

Figure 14-11. 3-Wire Serial I/O Mode Timing (7/7)
(xii) Slave operation (when DAP20 = 1, CKP20 = 1, SSE20 = 1)


Notes 1. The value of the last bit previously output is output.
2. DOO is output until $\overline{\mathrm{SS} 20}$ rises.

When $\overline{\mathrm{SS} 20}$ is high, SO20 is in a high-impedance state.

## (3) Transfer start

Serial transfer is started by setting transfer data to the transmission shift register (TXS20/SIO20) when the following two conditions are satisfied.

- Serial operation mode register 20 (CSIM20) bit 7 (CSIE20) $=1$
- Internal serial clock is stopped or $\overline{\mathrm{SCK20}}$ is high after 8-bit serial transfer.


## Caution If CSIE20 is set to 1 after data is written to TXS20/SIO20, transfer does not start.

A termination of 8 -bit transfer stops the serial transfer automatically and generates the interrupt request signal (INTCSI20).

## CHAPTER 15 SMB0 ( $\mu$ PD789167Y AND 789177Y SUBSERIES)

### 15.1 SMBO Functions

SMB0 (system management bus) has the following two types of modes.

- Operation stop mode
- SMB mode (supporting multiple masters)
(a) Operation stop mode

This mode is used when serial transfer is not performed. Power consumption is minimized in this mode.
(b) SMB mode (supporting multiple masters)

This mode is used for performing 8-bit data transmission to several devices, using a serial clock (SCL0) line and a serial data bus (SDAO) line.
In this mode, which conforms to the SMB format, start conditions, data, and stop conditions can be output on the serial data bus during transmission. Moreover, these data can be automatically detected by hardware during reception.
In SMB0, SCLO and SDAO are open-drain outputs, and therefore a pull-up resistor is required for the serial clock line and serial data bus line.
$I^{2} \mathrm{C}$ (Inter IC) bus standard mode or high-speed mode can be specified by software in SMB mode.

Figure 15-1 shows the block diagram of SMB0.

Figure 15-1. Block Diagram of SMB0


### 15.2 SMBO Configuration

SMB0 consists of the following hardware.

Table 15-1. Configuration of SMB0

| Item |  |
| :--- | :--- |
| Registers | SMB shift register 0 (SMB0) <br> SMB slave address register 0 (SMBSVA0) |
| Control registers | SMB control register 0 (SMBC0) |
|  | SMB status register 0 (SMBSO) |
|  | SMB clock selection register 0 (SMBCLO) |
|  | SMB mode register 0 (SMBMO) |
|  | SMB input level setting register 0 (SMBVIO) |
|  | Port mode register 2 (PM2) |
|  | Port 2 (P2) |

(1) SMB shift register 0 (SMB0)

SMBO is a register that converts 8 -bit serial data to 8 -bit parallel data, and vice-versa. SMB0 is used both for transmitting and receiving data.
Write and read operations for SMB0 control actual send and receive operations.
SMB0 is manipulated with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears SMB0 to 00 H .
(2) SMB slave address register 0 (SMBSVAO)

This register is used to set a local address when used as a slave.
SMBSVAO is manipulated with an 8-bit memory manipulation instruction.
RESET input clears SMBSVAO to 00 H .
(3) SO latch

The SO latch is a latch that holds the SDAO pin output level.
(4) Wakeup controller

This circuit generates an interrupt request when the address value set in SMB slave address register 0 (SMBSVA0) and the received address match, or when an extension code is received.
(5) Clock selector

Selects the sampling clock to be used.
(6) Serial clock counter

Counts the serial clock output/input during send/receive operations, to check if 8 -bit data has been sent or received.
(7) Interrupt request signal generator

Controls the generation of interrupt request signals.
SMB interrupts are generated with the following two triggers.

- 8th clock or 9th clock of serial clock (set with WTIMO bit ${ }^{\text {Note }}$ )
- Generation of interrupt request at detection of stop condition (set with bit SPIEO ${ }^{\text {Note }}$ )

Note WTIMO bit: SMB control register 0 (SMBCO) bit 3 SPIEO bit: SMB control register 0 (SMBCO) bit 4
(8) Serial clock controller

In master mode, generates the clock to be output to the SCLO pin from the sampling clock.
(9) Serial clock wait controller

Controls the wait timing.
(10) Acknowledge output circuit, stop condition detector, start condition detector, acknowledge detector Perform output and detection of control signals.
(11) Data hold time corrector

Generates the data hold time from the falling edge of the serial clock.

### 15.3 SMBO Control Registers

The following five registers are used to control SMBO.

- SMB control register 0 (SMBC0)
- SMB status register 0 (SMBSO)
- SMB clock selection register 0 (SMBCLO)
- SMB mode register 0 (SMBMO)
- SMB input level setting register 0 (SMBVIO)

The following registers are also used.

- SMB shift register 0 (SMBO)
- SMB slave address register 0 (SMBSVAO)
(1) SMB control register 0 (SMBCO)

This register sets SMB operation enable/disable, the wait timing, and other SMB operations.
SMBCO is manipulated with a 1-bit or 8-bit memory manipulation instruction.
RESET input clears SMBCO to 00H.

Caution Set port mode register 2 (PM2x) as follows in SMB mode.
Reset the output latch to 0.

- Set P23 (SCLO) in output mode (PM23 = 0).
- Set P24 (SDA0) in output mode (PM24 = 0).

Figure 15-2. Format of SMB Control Register 0 (1/4)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBCO | SMBEO | LRELO | WRELO | SPIEO | WTIMO | ACKEO | STTO | SPT0 | FF78H | OOH | R/W |


| SMBE0 | SMB operation $^{\text {Note } 1}$ |  |
| :---: | :--- | :--- |
| 0 | Operation disabled. Presets SMB status register $0($ SMBSO $)$. | Internal operation also disabled. |
| 1 | Operation enabled | Set conditions $($ SMBEO $=1)$ |
| Clear conditions $(S M B E O=0)$ |  | - Set with instruction |
| - Cleared with instruction <br> - Cleared by $\overline{\text { RESET input }}$ |  |  |


| LRELO | Escape from transmission |  |
| :---: | :---: | :---: |
| 0 | Normal operation |  |
| 1 | Escape from the current transmission and enter the standby status. Automatically cleared after execution. <br> This bit is used when extension codes not relevant to the local station are received. <br> The SCLO and SDAO lines enter the high impedance status. <br> The following flags are cleared. <br> - STDO • STTO •SPTO • ACKDO •TRCO •COIO •EXCO •MSTSO |  |
| The standby status continues until the following communication participation conditions are met. <br> - Startup as master after detection of stop condition <br> - Matching addresses or extension code reception after start condition |  |  |
| Clear conditions (LRELO $=0)^{\text {Note } 2}$ |  | Set conditions (LRELO $=1$ ) |
| - Automatically cleared after execution <br> - Cleared by RESET input |  | - Set with instruction |


| WRELO | Wait cancel |  |
| :---: | :--- | :--- |
| 0 | Do not cancel wait. |  |
| 1 | Cancel wait. Automatically cleared after wait cancellation. |  |
| Clear conditions (WRELO $=0)^{\text {Note 2 }}$ |  | Set conditions (WRELO $=1$ ) |
| - Automatically cleared after execution <br> - Cleared by <br> RESET input | - Set with instruction |  |


| SPIEO | Interrupt request generation at stop condition detection |  |
| :---: | :---: | :---: |
| 0 | Disabled |  |
| 1 | Enabled |  |
| Clear conditions (SPIEO $=0)^{\text {Note } 2}$ |  | Set conditions (SPIE0 $=1$ ) |
| - Cleared with instruction <br> - Cleared by RESET input |  | - Set with instruction |

Notes 1. Before setting SMBEO to 1, fix the value of SMB clock selection register 0 (SMBCLO). To change the communication clock, clear SMBE0 to 0 first before rewriting SMBCLO.
2. This flag's signals are made invalid by setting SMBEO $=0$.

Figure 15-2. Format of SMB Control Register 0 (2/4)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBCO | SMBE0 | LRELO | WRELO | SPIEO | WTIM0 | ACKE0 | STTO | SPTO | FF78H | OOH | R/W |


| WTIM0 | Wait and interrupt request generation control |
| :---: | :---: |
| 0 | Generate interrupt request at falling edge of 8th clock. <br> In case of master: Wait with clock output at low level after 8 clocks have been output. <br> In case of slave: Wait master with clock set to low level after 8 clocks have been input. |
| 1 | Generate interrupt request at falling edge of 9th clock. <br> In case of master: Wait with clock at low level after 9 clocks have been output. <br> In case of slave: Wait master with clock set to low level after 9 clocks have been input. |

The setting of this bit becomes invalid during address transmission, and becomes effective at the end of transmission. During operation as master, a wait is inserted at the falling edge of the 9th clock during address transmission. A slave that receives a local address enters the wait status at the falling edge of the 8th or 9th clock according to the setting of AWTIMO. A slave that receives an extension code enters the wait status at the falling edge of the 8th clock.

| Clear conditions $(W T I M 0=0)^{\text {Note }}$ | Set conditions $($ WTIM0 $=1)$ |
| :--- | :--- |
| - Cleared with instruction | • Set with instruction |
| - Cleared by RESET input |  |


| ACKEO | Acknowledge control |  |
| :---: | :--- | :--- |
| 0 | Acknowledge disabled |  |
| 1 | Acknowledge enabled. SDAO line set to low level during 9 clocks. However, invalid during address <br> transmission, and valid when EXC0 $=1$. |  |
| Clear conditions $(\text { ACKE0 }=0)^{\text {Note }}$ |  | Set conditions $($ ACKEO $=1)$ |
| - Cleared with instruction <br> - Cleared by $\overline{\text { RESET input }}$ | - Set with instruction |  |

Note This flag's signals are made invalid by setting SMBE0 $=0$.

Figure 15-2. Format of SMB Control Register 0 (3/4)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBCO | SMBEO | LRELO | WRELO | SPIEO | WTIMO | ACKEO | STTO | SPTO | FF78H | OOH | R/W |


| STTO | Start condition trigger |  |
| :---: | :---: | :---: |
| 0 | Do not generate start condition. |  |
| 1 | When bus is released (stop status): <br> Generate start conditions (activation as master). Change SDAO line from high level to low level and generate start condition. Then secure rated time and sets SCLO to low level. <br> When not participating on bus: <br> Functions as start condition reservation flag. When set, automatically generate start condition after bus is released. |  |
| Cautions regarding set timing <br> - Master receive operation: Setting during transmission is prohibited. <br> Set ACKEO = 0; Can be set only after end of receive operation has been reported to slave. <br> - Master transmit operation: Note that start condition may not be generated normally during $\overline{\mathrm{ACK}}$ period. <br> - Setting at the same time as SPTO is prohibited. <br> - After setting STTO, resetting is prohibited if the clear conditions have not been met. |  |  |
| Clear conditions (STTO $=0)^{\text {Note }}$ |  | Set conditions ( STTO = 1) |
| - Cleared with instruction <br> - Cleared upon defeat in arbitration <br> - Cleared after generation of start condition by master <br> - Cleared when LRELO $=1$ <br> - Cleared when SMBEO $=0$ <br> - Cleared by RESET input |  | - Set with instruction |

Note This flag's signals are made invalid by setting SMBE $=0$.

Figure 15-2. Format of SMB Control Register 0 (4/4)


| SPT0 | Stop condition trigger |  |
| :---: | :---: | :---: |
| 0 | Do not generate stop condition. |  |
| 1 | Generate stop condition (end transmission as master). <br> After setting SDAO line to low level, set SCLO line to high level, or maintain SCLO line at high level. <br> Then, secure rated time, change SDAO line from low level to high level, and generate stop condition. |  |
| Cautions regarding set timing <br> - Master receive operation: Setting during transmission is prohibited. <br> Set ACKEO = 0 ; Can be set only after end of receive operation has been notified to slave. <br> - Master send operation: Note that stop condition may not be generated normally during $\overline{A C K}$ period. <br> - Setting at the same time as STTO is prohibited. <br> - Set SPTO only during operation as master. Note 1 <br> - After setting SPTO, resetting is prohibited if the clear conditions have not been met. <br> - Note that when WTIMO $=0$, if SPTO is set during the wait period after 8 -clock output, a stop condition is generated during the high-level period of the 9th clock following wait release. <br> If it is necessary to output a 9th clock, change the setting of WTIMO from 0 to 1 during the wait period following 8-clock output, and set SPTO during the wait period following the 9th clock output. |  |  |
| Clear co | ditions (SPTO $=0)^{\text {Note } 2}$ | Set conditions (SPT0 = 1) |
| - Clear <br> - Clear <br> - Clear <br> - Clear <br> - Clear <br> - Clear | with instruction <br> upon defeat in arbitration <br> automatically after detection of stop condition <br> when LRELO $=1$ <br> when SMBEO $=0$ <br> by RESET input | - Set with instruction |

Notes 1. Set SPTO only during operation as master. However, for master operation by the time a stop condition is detected for the first time following operation enable, SPT0 must be set once to generate a stop condition.
2. This flag's signals are made invalid by setting $\operatorname{SMBE} 0=0$.

Caution While SMB status register 0 (SMBSO) bit 3 (TRCO) = 1, when WRELO is set at the 9th clock and wait is released, TRCO is cleared and the SDAO line is set to high impedance.

Remarks 1. STDO: SMB status register 0 (SMBSO) bit 1
ACKD0: SMB status register 0 (SMBSO) bit 2
TRC0: SMB status register 0 (SMBSO) bit 3
COIO: SMB status register 0 (SMBSO) bit 4
EXC0: SMB status register 0 (SMBSO) bit 5
MSTS0: SMB status register 0 (SMBSO) bit 7
2. Bits 0 and 1 (SPTO, STTO) are 0 if read after data setting.

## (2) SMB status register 0 (SMBSO)

This register indicates the SMB status.
SMBS0 is manipulated with a 1-bit or 8-bit memory manipulation instruction. SMBS0 is a read-only register. RESET input clears SMBSO to 00 H .

Figure 15-3. Format of SMB Status Register 0 (1/3)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBS0 | MSTS0 | ALD0 | EXC0 | COIO | TRC0 | ACKD0 | STD0 | SPD0 | FF79H | 00 H | R |


| MSTS0 | Master status |  |
| :---: | :--- | :--- |
| 0 | Slave status or communication wait status |  |
| 1 | Master transmission status | Set conditions (MSTS0 = 1) |
| Clear conditions (MSTS0 $=0)$ | - Set during generation of start condition |  |
| - Cleared upon detection of stop condition |  |  |
| - Cleared when ALD0 $=1$ |  |  |
| - Cleared when LREL0 $=1$ |  |  |
| - Cleared when SMBE0 changes from 1 to 0 |  |  |
| - Cleared by RESET input |  |  |


| ALD0 | Arbitration defeat detection |  |
| :---: | :---: | :---: |
| 0 | No arbitration, or won in arbitration. |  |
| 1 | Defeated in arbitration. MSTS0 cleared. |  |
| Clear conditions ( $\mathrm{ALDO}=0$ ) |  | Set conditions (ALD0 = 1) |
| - Automatically cleared after reading SMBSONote <br> - Cleared when SMBEO changes from 1 to 0 <br> - Cleared by RESET input |  | - Set upon defeat in arbitration |


| EXC0 | Extension code receive detection |  |
| :---: | :--- | :--- |
| 0 | Do not receive extension code. |  |
| 1 | Receive extension code. | Set conditions (EXC0 $=1$ ) |
| Clear conditions (EXC0 $=0)$ | $\bullet$ Set when high 4 bits of received address are |  |
| - Cleared upon detection of start condition |  |  |
| - Cleared upon detection of stop condition |  |  |
| - Cleared when LREL0 $=1$ |  |  |
| - Cleared when SMBE0 changes from 1 to 0 |  |  |
| - Cleared by $\overline{R E S E T}$ input |  |  |

Note The bit is also cleared when a bit manipulation instruction is executed for any other bit SMBSO.

Figure 15-3. Format of SMB Status Register 0 (2/3)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBSO | MSTSO | ALDO | EXCO | COIO | TRCO | ACKDO | STDO | SPDO | FF79H | OOH | R |


| COIO | Matching address detection |  |
| :---: | :--- | :--- |
| 0 | Address does not match. |  |
| 1 | Address matches. | Set conditions (COIO $=1$ ) |
| Clear conditions $($ COIO $=0)$ |  | - Set when received address matches local address |
| - Cleared upon detection of start condition | (SVAO) (set at rising edge of 8th clock) |  |
| - Cleared upon detection of stop condition |  |  |
| - Cleared when LRELO $=1$ |  |  |
| - Cleared when SMBEO changes from 1 to 0 |  |  |
| - Cleared by $\overline{\text { RESET input }}$- |  |  |


| TRCO | Receive/send status detection |  |
| :---: | :---: | :---: |
| 0 | Receive status (when not in send status). Sets SDAO line to high impedance. |  |
| 1 | Send status. Sets so that SO latch value can be output to SDAO line (valid from falling edge of 9th clock of 1st byte). |  |
| Clear conditions ( $\mathrm{TRC0}=0$ ) |  | Set conditions (TRCO = 1) |
| - Cleared upon detection of stop condition <br> - Cleared when LRELO = 1 <br> - Cleared SMBEO changes from 1 to 0 <br> - Cleared when WRELO $=1^{\text {Note }}$ <br> - Cleared when ALDO changes from 0 to 1 <br> - Cleared by RESET input <br> In case of master: <br> - When "1" is output to 1st byte LSB (transmission direction specification bit) In case of slave: <br> - Upon detection of start condition <br> In case of non-participation in communication |  | In case of master: <br> - Upon generation of start condition <br> In case of slave: <br> - When "1" is input to 1st byte LSB (transmission direction specification bit) |


| ACKDO | Acknowledge output |  |
| :---: | :---: | :---: |
| 0 | Do not detect acknowledge. |  |
| 1 | Detect acknowledge. |  |
| Clear conditions ( $\mathrm{ACKDO}=0$ ) |  | Set conditions (ACKDO $=1$ ) |
| - Cleared upon detection of stop condition <br> - Cleared at rising edge of 1 st clock of following byte <br> - Cleared when LRELO = 1 <br> - Cleared when SMBE0 changes from 1 to 0 <br> - Cleared by RESET input |  | - Set when SDAO line is low level at rising edge of 9th clock of SCLO |

Note When bit 3 (TRCO) of SMB status register 0 (SMBSO) is set to 1 , bit 5 (WRELO) of SMB control register $0(S M B C O)$ is set during the ninth clock and wait is canceled, after which TRCO is cleared and the SDA0 line is set to high impedance.

Figure 15-3. Format of SMB Status Register 0 (3/3)

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBSO | MSTS0 | ALDO | EXCO | COIO | TRCO | ACKDO | STDO | SPD0 | FF79H | 00H | R |


| STDO | Start condition detection |  |
| :---: | :--- | :--- |
| 0 | Do not detect start condition. |  |
| 1 | Detect start condition. Indicates that address transmission is in progress. |  |
| Clear conditions (STD0 $=0$ ) |  | Set conditions (STD0 $=1$ ) |
| - Cleared upon detection of stop condition <br> - Cleared at rising edge of 1st clock of byte following <br> address transmission | - Set upon detection of start condition |  |
| - Cleared when LRELO $=1$ |  |  |
| - Cleared when SMBEO changes from 1 to 0 |  |  |
| - Cleared by RESET input |  |  |


| SPDO | Stop condition detection |  |
| :---: | :--- | :--- |
| 0 | Do not detect stop condition. | Set conditions (SPD0 $=1$ ) |
| 1 | Detect stop condition. Transmission by master is completed and bus is released. |  |
| Clear conditions (SPDO $=0)$ | Set upon detection of stop condition |  |
| - Cleared at rising edge of 1st clock of address transfer <br> byte following detection of start condition after this bit has <br> been set |  |  |
| - Cleared when SMBE0 changes from 1 to 0 |  |  |
| - Cleared by RESET input |  |  |

Remark LRELO: SMB control register 0 (SMBCO) bit 6
SMBEO: SMB control register 0 (SMBCO) bit 7

## (3) SMB clock selection register 0 (SMBCLO)

This register sets the SMB transmission clock.
SMBCLO is manipulated with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears SMBCL0 to 00 H . Table $15-2$ shows the SMB communication clocks.

Figure 15-4. Format of SMB Clock Selection Register 0 (1/2)

| Symbol | 7 | 6 | <5> | <4> | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBCLO | 0 | 0 | CLDO | DADO | SMC0 | DFC0 | CL01 | CLOO | FF7AH | OOH | R/W $W^{\text {Note }} 1$ |


| CLDO | SCLO line level detection (valid only when SMBEO $=1$ ) |  |
| :---: | :--- | :---: |
| 0 | Detect that SCLO line is low level. |  |
| 1 | Detect that SCLO line is high level. |  |
| Clear conditions (CLDO $=0$ ) |  |  |
| - Cleared when SCLO line is low level <br> - Cleared when SMBEO $=0$ <br> - Cleared by $\overline{\text { RESET }}$ input | Set conditions (CLD0 $=1$ ) |  |


| DADO | SDAO line level detection (valid only when SMBEO $=1$ ) |  |
| :---: | :--- | :---: |
| 0 | Detect that SDAO line is low level. |  |
| 1 | Detect that SDAO line is high level. |  |
| Clear conditions (DADO $=0$ ) |  |  |
| - Cleared when SDAO line is low level <br> - Cleared when SMBEO $=0$ <br> - Cleared by $\overline{\text { RESET input }}$ | Set conditions (DADO $=1$ ) |  |


| SMCO | Operating mode switching |  |
| :---: | :--- | :---: |
| 0 | IIC standard mode or SMB mode operation |  |
| 1 | IIC high-speed mode |  |
| Clear conditions $(S M C 0=0)$ |  |  |
| - Cleared with instruction <br> - Cleared by $\overline{\text { RESET input }}$ | Set conditions $(S M C 0=1)$ |  |


| DFC0 |  | Digital filter operation control ${ }^{\text {Note } 2}$ |
| :---: | :--- | :--- |
| 0 | Digital filter off |  |
| 1 | Digital filter on |  |

Notes 1. Bits 4 and 5 are read-only.
2. The digital filter can be used in high-speed mode. When used in high-speed mode, the digital filter provides a slower response.

## Caution Bits 6 and 7 must be set to 0 .

Figure 15-4. Format of SMB Clock Selection Register 0 (2/2)

| Symbol | 7 | 6 | <5> | <4> | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBCLO | 0 | 0 | CLDO | DADO | SMC0 | DFC0 | CL01 | CL00 | FF7AH | OOH | R/W ${ }^{\text {Note }}$ |


| CL01 | CL00 | Communication clock |  |
| :---: | :---: | :--- | :--- |
|  |  | SMB/IIC standard mode $(S M C 0=0)$ | IIC high-speed mode $(S M C 0=1)$ |
| 0 | 0 | $f_{x} / 44$ | $f_{x} / 24$ |
| 0 | 1 | $f_{x} / 86$ |  |
| 1 | 0 | $f_{x} / 172$ | $f_{x} / 48$ |
| 1 | 1 | Setting prohibited |  |

Note Bits 4 and 5 are read-only.

Caution To change the communication clock, stop operations (SMBEO $=0$ ) first before rewriting SMBCLO.

Remark fx: Main system clock oscillation frequency

Table 15-2. SMB0 Communication Clock

| SMCO | CL01 | CLOO | Communication Clock |  | Digital Filter Input Delay |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $\begin{gathered} \text { At } \mathrm{fx}=10.0 \mathrm{MHz} \\ \text { operation }^{\text {Note } 1} \end{gathered}$ | $\text { At } \mathrm{fx}=5.0 \mathrm{MHz}$ <br> operation |  |
| 0 | 0 | 0 | 227.2 kHz ${ }^{\text {Note } 2}$ | 113.6 kHz ${ }^{\text {Note } 2}$ | 250 ns |
| 0 | 0 | 1 | $116.2 \mathrm{kHz}^{\text {Note } 2}$ | 58.13 kHz | 250 ns |
| 0 | 1 | 0 | 58.13 kHz | 29.06 kHz | 500 ns |
| 1 | 0 | 0 | 416.6 kHz ${ }^{\text {Note } 3}$ | 208.3 kHz | 250 ns |
| 1 | 0 | 1 | 416.6 kHz ${ }^{\text {Note } 3}$ | 208.3 kHz | 250 ns |
| 1 | 1 | 0 | 208.3 kHz | 104.1 kHz | 500 ns |
| Other than above |  |  | Setting prohibited |  |  |

Notes 1. Expanded-specification products only.
2. Since the $\mathrm{SMB} / I I C$ standard mode standards specify a range of 10 to 100 kHz , this communication clock falls outside the specifications.
3. Since the standards of the IIC high-speed mode specify a range of 0 to 400 kHz , this communication clock falls outside the specifications.

## (4) SMB mode register 0 (SMBMO)

SMBMO is used to specify SCLO level control and interrupt control.
SMBMO is manipulated with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input sets SMBM0 to 20 H .

Figure 15-5. Format of SMB Mode Register 0 (1/2)

| Symbol | 7 | 6 | <5> | <4> | <3> | <2> | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBMO | 0 | 0 | SCLCTLO | AWTIMO | STIE0 | TOENO | TOCL01 | TOCL00 | FF7CH | 20 H | R/W |
|  | SCLCTLO | SCL level control ${ }^{\text {Note }} 1$ |  |  |  |  |  |  |  |  |  |
|  | 0 | SCLO is held low. <br> When SCLO is high, SCLO is held low after waiting until SCLO is made low. |  |  |  |  |  |  |  |  |  |
|  | 1 | Normal operation |  |  |  |  |  |  |  |  |  |


| AWTIMO | Wait and interrupt control when an address match is foundNotes 2,3 |
| :---: | :--- |
| 0 | At the slave, an interrupt request is generated on the falling edge of the 9th clock period when an address <br> match $($ COIO $=1)$ is found during address data reception. <br> The clock is pulled low to cause the master to wait. |
| 1 | At the slave, an interrupt request is generated on the falling edge of the 8th clock period when an address <br> match $(C O I O=1)$ is found during address data reception. <br> The clock is pulled low to cause the master to wait. |


| STIEO | Start condition interrupt enable |
| :---: | :--- |
| 0 | Start condition interrupt generation is disabled. |
| 1 | Normal operation |


| TOENO | Time out count enable bitNote 4 |
| :---: | :--- |
| 0 | The time out count is cleared to 0, then count operation is disabled. |
| 1 | Time out count operation is enabled. |

Notes 1. If SCLO is made low by SCLCTLO, the wait state cannot be released by WRELO.
2. When an extension code is received ( $\mathrm{EXCO}=1$ ), a wait state is forcibly set in the 8th clock period.
3. During address transfer, the master waits in the 9th clock period.
4. An interrupt (INTSMBOVO) is generated when the time out counter overflows. The hardware does not reset the SMB operation. Ensure that SMB operation is reset by software after INTSMBOVO generation.

## Caution Bits 6 and 7 must be set to 0 .

Figure 15-5. Format of SMB Mode Register 0 (2/2)

| Symbol | 7 | 6 | <5> | <4> | <3> | <2> | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBMO | 0 | 0 | SCLCTLO | AWTIMO | STIE0 | TOENO | TOCL01 | TOCL00 | FF7CH | 20 H | R/W |


| TOCL01 | TOCLOO | Time out clock fтo selection bits |  |  |
| :---: | :---: | :--- | :--- | :---: |
| 0 | 0 | $\mathrm{fx} / 2^{6}(78.1 \mathrm{kHz})$ |  |  |
| 0 | 1 | $\mathrm{fx} / 2^{7}(39.1 \mathrm{kHz})$ |  |  |
| 1 | 0 | $\mathrm{fx} / 2^{8}(19.5 \mathrm{kHz})$ |  |  |
| 1 | 1 | $\mathrm{fxT} \quad(32.768 \mathrm{kHz})$ |  |  |

Remarks 1. fx: Main system clock oscillation frequency
2. fx : Subsystem clock oscillation frequency

## (5) SMB input level setting register 0 (SMBVIO)

SMBVIO is manipulated with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears SMBVIO to 00 H .

Figure 15-6. Format of SMB Input Level Setting Register 0


| TOSO2 | TOS01 | TOS00 | Time out time selection bits |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | At $\mathrm{fx}=$ 10.0 MHz operation ${ }^{\text {Note } 1}$ |  |  | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |  |  | $\text { At } f_{x T}=$ $32.768 \text { kHz }$ <br> operation |
|  |  |  |  | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{fx} / 2^{6} \end{aligned}$ | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{f} / / 2^{7} \end{aligned}$ | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{fx} / 2^{8} \end{aligned}$ | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{fx} / 2^{6} \end{aligned}$ | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{fx} / 2^{7} \end{aligned}$ | $\begin{aligned} & \mathrm{fTO}= \\ & \mathrm{fx} / 2^{8} \end{aligned}$ | $\mathrm{f}_{\text {TO }}=\mathrm{f}_{\mathrm{x}}$ T |
| 0 | 0 | 0 | 1024/fтo | 6.55 ms | 13.1 ms | 26.2 ms | 13.1 ms | 26.2 ms | 52.4 ms | 31.2 ms |
| 0 | 0 | 1 | 896/fто | 5.73 ms | 11.4 ms | 22.9 ms | 11.4 ms | 22.9 ms | 45.8 ms | 27.3 ms |
| 0 | 1 | 0 | 768/fто | 4.91 ms | 9.83 ms | 19.6 ms | 9.83 ms | 19.6 ms | 39.3 ms | 23.4 ms |
| 0 | 1 | 1 | 640/fтo | 4.09 ms | 8.19 ms | 16.3 ms | 8.19 ms | 16.3 ms | 32.7 ms | 19.5 ms |
| 1 | 0 | 0 | 512/fтo | 3.27 ms | 6.55 ms | 13.1 ms | 6.55 ms | 13.1 ms | 26.2 ms | 15.6 ms |
| 1 | 0 | 1 | 384/fто | 2.45 ms | 4.91 ms | 9.83 ms | 4.91 ms | 9.83 ms | 19.6 ms | 11.7 ms |
| 1 | 1 | 0 | 256/fто | 1.63 ms | 3.27 ms | 6.55 ms | 3.27 ms | 6.55 ms | 13.1 ms | 7.81 ms |
| 1 | 1 | 1 | 128/fто | $819 \mu \mathrm{~s}$ | 1.63 ms | 3.27 ms | 1.63 ms | 3.27 ms | 6.55 ms | 3.90 ms |


| SVINO | Input level selection bit |
| :---: | :--- |
| 0 | Same input level as the ordinary hysteresis |
| 1 | The voltage set with LVL01 and LVLOOI is used as the SCLO and SDAO input level threshold |


| LVL01 | LVL00 |  | Input level selection bits ${ }^{\text {Note } 2}$ |
| :---: | :---: | :--- | :--- |
| 0 | 0 | The input level is $0.1875 \times \mathrm{VDD}$. |  |
| 0 | 1 | The input level is $0.25 \times \mathrm{VDD}$ |  |
| 1 | 0 | The input level is $0.375 \times \mathrm{VDD}_{\mathrm{DD}}$. |  |
| 1 | 1 | The input level is $0.5 \times \mathrm{VDD}_{\mathrm{DD}}$. |  |

Notes 1. Expanded-specification products only.
2. Set an input level from 0.75 to 1.25 V .

## Caution Bits 2 and 7 must be set to 0 .

Remarks 1. fx : Main system clock oscillation frequency
2. fx : Subsystem clock oscillation frequency
3. fto: Clock selected using bits 0 and 1 (TOCLO0, TOCL01) of SMB mode register 0 (SMBMO)
(6) SMB shift register 0 (SMB0)

This register is used to perform serial transmit/receive (shift operation) in synchronization with the serial clock.
Read/write operations can be performed in 8-bit units, but do not write data to SMB0 during transmission.

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMB0 |  |  |  |  |  |  |  |  | FF7EH | OOH | R/W |

(7) SMB slave address register 0 (SMBSVA0)

This register stores the SMB slave address.
It can be read/written in 8 -bit units, but bit 0 is fixed to 0 .

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SMBSVA0 |  |  |  |  |  |  |  | 0 | FF7BH | OOH | R/W |

### 15.4 SMBO Definition and Control Methods

The SMB0 serial data transmission format and the meanings of the signals used are described below.
The transmission timing of the start condition, data, and stop condition output to the serial data bus of the SMBO is shown in Figure 15-7.

Figure 15-7. SMB0 Serial Data Transmission Timing


The start condition, slave address, and stop condition are output by the master.
SDAO of only the start condition and stop condition can be changed when SCLO is high.
The acknowledge signal ( $\overline{\mathrm{ACK}}$ ) can be output by either the master or slave (the slave outputs $\overline{\mathrm{ACK}}$ when an address is transferred. The receiver of the data outputs $\overline{A C K}$ when 8-bit data is transferred).

The master continuously outputs the serial clock (SCLO). However, it is possible to prolong the low-level period of the SCLO and insert a wait in the case of the slave.

### 15.4.1 Start condition

A start condition is generated when the SDAO pin changes from high level to low level while the SCLO pin is high level (serial clock is not output). The start condition of the SCLO and SDAO pins is output at the start of serial transmission from the master to the slave. The slave incorporates hardware that detects the start condition.

Figure 15-8. Start Condition


The start condition is output when SMB control register 0 (SMBCO) bit 1 (STTO) is set to 1 in the stop condition detection status (STDO: SMB status register 0 (SMBSO) bit $1=1$ ). Moreover, when the start condition is detected, SMBSO bit 1 (STDO) is set to 1 , and when bit 3 (STIEO) of SMBM0 is set to 1 , INTSMB0 is generated.

### 15.4.2 Address

The 7-bit data following the start condition is defined as an address.
An address consists of 7 bits of data output to select a particular slave among several slaves connected to the master via the bus line. Therefore, slaves on the bus line must each have a different address.

Slaves detect start conditions via hardware and check if the 7-bit data matches the value of SMB slave address register 0 (SMBSVAO). If the 7 -bit data and the SMBSVA0 value match, that slave is selected, and communication between the master and that slave is performed until the master issues a start condition or a stop condition.

Figure 15-9. Address


Note If other than a local address or extension code is received during slave operation, INTSMBO is not issued.

Addresses are written and output to SMB shift register 0 (SMB0) as 8 bits consisting of the slave address and the transmission direction (see 15.4.3). Moreover, received addresses are written to SMB0.

Slave addresses are allocated to the high 7 bits of SMB0.

### 15.4.3 Specification of transmission direction

The master sends a 1-bit data following the 7-bit address to specify the transmission direction. When this transmission direction bit is 0 , the master sends data to the slave.
When this bit is 1 , the slave sends data to the master.

Figure 15-10. Specification of Transmission Direction


Note If other than a local address or extension code is received during slave operation, INTSMBO is not issued.

### 15.4.4 Acknowledge signal ( $\overline{\mathrm{ACK}}$ )

The acknowledge signal ( $\overline{\mathrm{ACK}})$ is used to confirm reception of serial data on the transmitting and receiving sides.
On the receiving side, an acknowledge signal is returned each time 8 bits of data are received. On the sending side, an acknowledge signal is normally received following transmission of 8 bits of data. However, when the master is receiving, no acknowledge signal is output after the final data has been received. The transmitting side detects whether an acknowledge signal is returned following transmission of 8 bits of data. If an acknowledge signal is returned, processing is continued assuming that the data was successfully received. If no acknowledge signal is returned by the slave, the master outputs a stop condition or a restart condition, and stops transmission. An acknowledge signal is not returned for the following two reasons.
<1> Reception was not performed normally.
<2> The final data was received.

If the receiving side sets the SDAO line to low level at the 9th clock, the acknowledge signal becomes active (normal reception response).

When SMB control register $0(S M B C O)$ bit $2(A C K E O)=1$, the acknowledge signal automatic generation enable state is entered.

SMB status register 0 (SMBSO) bit 3 (TRCO) is set by the 8th bit following the 7-bit address. However, when the TRC0 bit value is 0 , receive status is selected, therefore set ACKE0 to 1 .

During a slave receive operation (TRCO $=0$ ), if the slave side receives several bytes and does not require subsequent data, ACKEO can be set to 0 so that the master does not start the next transmission.

In the same way, if, during a master receive operation (TRCO = 0), subsequent data is not required and you want to output a restart condition or a stop condition, set ACKEO to 0 so that no $\overline{A C K}$ signal is output. This must be done so that the data's MSB is not output to the SDAO line during the slave transmission operation (transmission stop).

Figure 15-11. Acknowledge Signal


When a slave receives a local address, it automatically outputs an acknowledge signal in synchronization with the falling edge of the 8th clock of SCLO, regardless of the value of ACKEO. If a slave receives other than a local address, no acknowledge signal is output.

The acknowledge signal output method during data reception depends on the wait timing setting, as follows.

- 8 -clock wait: Acknowledge signal is output when the value of ACKEO becomes 1 before wait cancellation is performed.
- 9-clock wait: Acknowledge signal is automatically output in synchronization with the falling edge of the 8th clock of SCLO by setting ACKEO to 1 beforehand.


### 15.4.5 Stop condition

When the SDAO pin changes from low level to high level while the SCLO pin is at high level, a stop condition is generated.

A stop condition is the signal that is output when serial transfer from the master to a slave is completed. Slaves incorporate hardware for the detection of stop conditions.

Figure 15-12. Stop Condition


A stop condition is generated when bit 0 (SPTO) of SMB control register 0 (SMBCO) is set to 1 . If, when a stop condition is detected, bit 0 (SPDO) of SMB status register 0 (SMBSO) and bit 4 (SPIEO) of SMBC0 are set to 1 , INTSMB0 is generated.

### 15.4.6 Wait signal (WAIT)

A wait signal ( $\overline{\text { WAIT }})$ indicates to the other party that the master or slave is getting ready (wait status) to send or receive data.

A wait status is notified by making the SCLO pin low level. When the wait status of both the master and slave is canceled, the next transmission starts.

Figure 15-13. Wait Signal (1/2)
(1) When master $=9$-clock wait, slave $=8$-clock wait
(Master: send, Slave: receive, ACKE0 = 1)


Slave


ACKEO H

Transmission Line


Figure 15-13. Wait Signal (2/2)
(2) Master, slave = 9-clock wait (Master: send, Slave: receive, ACKE0 = 1)


Remark ACKEO: SMB control register 0 (SMBCO) bit 2
WRELO: SMB control register 0 (SMBCO) bit 5

Waits are automatically generated by setting bit 3 (WTIMO) of SMB control register 0 (SMBCO).
Normally, the receive side cancels the wait status when SMBC0 bit 5 (WRELO) $=1$ or SMB shift register (SMB0)
$\leftarrow$ FFH write, and the transmit side cancels the wait status when data is written to SMB0.
In the case of the master, wait status can be canceled by the following methods.

- Setting SMBCO bit 1 (STTO) to 1
- Setting SMBCO bit 0 (SPTO) to 1


### 15.4.7 SMBO interrupt (INTSMBO)

The following section shows the values of SMB status register 0 (SMBSO) using the INTSMB0 interrupt request generation timing and INTSMBO interrupt timing.

## Caution The case when AWTIMO = 0 is described here.

(1) Master operation
(a) Start - Address - Data — Data — Stop (normal send/receive)
<1> When WTIMO = 0

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | $\Delta 1$ |  |  | $\Delta 2$ | -3 |  | $\triangle 4 \quad 5$ |

$\diamond 0: S M B S O=10001010 B$
$\triangle 1: S M B S 0=1000 \times 110 B$
4 2: $S M B S 0=1000 \times 000 B$
4 3: $\mathrm{SMBSO}=1000 \times 000 \mathrm{~B}$
44: SMBSO $=1000 \times \times 00 B$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

$\diamond 0: S M B S O=10001010 B$
A 1: $S M B S 0=1000 \times 110 B$
A 2: $\mathrm{SMBSO}=1000 \times 100 \mathrm{~B}$
$\triangle$ 3: SMBSO $=1000 \times \times 00 B$
$\triangle 4: ~ S M B S O=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(b) Start - Address — Data — Start — Address — Data — Stop (restart)
<1> When WTIMO = 0

$\diamond 0: S M B S 0=10001010 B$
A 1: SMBSO $=1000 \times 110 B$

- 2: $\mathrm{SMBSO}=1000 \times 000 \mathrm{~B}$
- 3: SMBSO $=1000 \times 110 B$

44: $\mathrm{SMBSO}=1000 \times 000 \mathrm{~B}$
45: SMBSO $=1000 \times \times 00 B$
$\triangle 6: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | SP | P |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | -1 |  | $\triangle 2$ |  |  |  | 43 |  | 44 |  |  |

$\diamond 0: S M B S 0=10001010 B$
^1: SMBSO = $1000 \times 110 \mathrm{~B}$
42: $\mathrm{SMBSO}=1000 \times \times 00 \mathrm{~B}$
$\triangle$ 3: $\mathrm{SMBSO}=1000 \times 110 \mathrm{~B}$
44: SMBSO $=1000 \times \times 00 B$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(c) Start - Code - Data - Data - Stop (extension code transmission)
<1> When WTIM0 = 0

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | $\Delta 1$ |  |  | $\Delta 2$ | $\Delta 3$ |  | -4 |

$\diamond 0:$ SMBSO $=10001010 \mathrm{~B}$
^1: SMBSO $=1010 \times \times 10 B$
4 2: SMBSO $=1010 \times 000 \mathrm{~B}$
3: SMBSO $=1010 \times 000 \mathrm{~B}$
44: SMBSO $=1010 \times \times 00 B$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIEO = 1
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

$\diamond 0: S M B S O=10001010 B$
-1: $\mathrm{SMBSO}=0010 \times 110 \mathrm{~B}$

- 2: SMBSO $=0010 \times 100 B$
- 3: $\mathrm{SMBSO}=0010 \times \times 00 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$
Remark $\diamond$ Generate only when STIE0 = 1
$\triangle$ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(2) Slave operation (during slave address data reception (matching SVA0))
(a) Start - Address - Data - Data - Stop
<1> When WTIMO $=0$

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\diamond 0$ | $\Delta 1$ |  |  | $\Delta 2$ | 43 |  |  |  |

$\diamond 0:$ SMBSO $=00000010 \mathrm{~B}$
^1: $\mathrm{SMBSO}=0001 \times 110 \mathrm{~B}$
4 2: $\mathrm{SMBSO}=0001 \times 000 \mathrm{~B}$

- 3: SMBSO $=0001 \times 000 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ | $\mathbf{\Delta 1}$ |  |  |  | $\Delta 2$ | $\Delta 3$ |  |  |

$\diamond 0:$ SMBSO $=00000010 \mathrm{~B}$
^1: SMBSO $=0001 \times 110 B$
A 2: SMBSO $=0001 \times 100 B$

- 3: SMBSO $=0001 \times \times 00 \mathrm{~B}$
$\triangle 4:$ SMBS0 $=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIEO $=1$

- Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(b) Start - Address - Data — Start — Address — Data - Stop
<1> When WTIMO = 0 (matching SVAO after restart)

$\diamond 0:$ SMBSO $=00000010 \mathrm{~B}$
-1: SMBS0 $=0001 \times 110 B$
42: SMBSO $=0001 \times 000 \mathrm{~B}$
- 3: $\mathrm{SMBSO}=0001 \times 110 \mathrm{~B}$

44: SMBSO $=0001 \times 000 \mathrm{~B}$
$\triangle 5: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIEO = 1

- Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1 (matching SVAO after restart)

$\diamond 0:$ SMBSO $=00000010 \mathrm{~B}$
$\triangle 1$ : $\mathrm{SMBSO}=0001 \times 110 \mathrm{~B}$
- 2: $\mathrm{SMBSO}=0001 \times \times 00 \mathrm{~B}$
- 3: $\mathrm{SMBSO}=0001 \times 110 \mathrm{~B}$

44: SMBSO $=0001 \times \times 00 B$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
A Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(c) Start - Address - Data — Start — Code - Data - Stop
<1> When WTIM0 = 0 (extension code reception after restart)

$\diamond 0: S M B S 0=00000010 B$

- 1: $\operatorname{SMBSO}=0001 \times 110 B$
- 2: $\mathrm{SMBSO}=0001 \times 000 \mathrm{~B}$
- 3: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$
-4: $\mathrm{SMBSO}=0010 \times 000 \mathrm{~B}$
$\triangle 5: S M B S O=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1 (extension code reception after restart)

$\diamond 0: S M B S 0=00000010 \mathrm{~B}$
-1: $\mathrm{SMBSO}=0001 \times 110 \mathrm{~B}$
42: SMBSO $=0001 \times \times 00 B$

- 3: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$

44: $\mathrm{SMBSO}=00100110 \mathrm{~B}$

- 5: $\mathrm{SMBSO}=0010 \times \times 00 \mathrm{~B}$
$\triangle 6: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE0 $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(d) Start - Address - Data — Start — Address — Data — Stop
<1> When WTIM0 = 0 (unmatching address (except extension code) after restart)

$\diamond 0: S M B S O=00000010 B$
$\triangle$ 1: $S M B S 0=0001 \times 110 B$

- 2: $\mathrm{SMBSO}=0001 \times 000 \mathrm{~B}$
- 3: SMBSO $=0000 \times \times 10 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1 (unmatching address (except extension code) after restart)

$\diamond 0: S M B S 0=00000010 \mathrm{~B}$
-1: SMBSO $=0001 \times 110 B$
4 2: SMBSO $=0001 \times \times 00 B$

- 3: $\mathrm{SMBSO}=0000 \times \times 10 \mathrm{~B}$
$\triangle 4:$ SMBSO $=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(3) Slave operation (during extension code reception)
(a) Start - Code - Data - Data - Stop
$<1>$ When WTIMO $=0$

$\diamond 0: S M B S 0=00000010 B$
-1: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$
42: $\mathrm{SMBSO}=0010 \times 000 \mathrm{~B}$

- 3: SMBSO $=0010 \times 000 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE0 = 1

- Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

| St | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ | $\Delta^{1} \Delta^{2}$ | $\Delta^{3} 3$ | $\Delta^{4} \quad \Delta 5$ |  |  |  |  |  |

$\diamond 0: S M B S 0=00000010 \mathrm{~B}$
-1: SMBSO $=0010 \times 010 B$
42: SMBSO $=0010 \times 110 B$

- 3: $\mathrm{SMBSO}=0010 \times \times 00 \mathrm{~B}$

44: SMBSO $=0010 \times \times 00 B$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(b) Start — Code - Data - Start — Address — Data - Stop
<1> When WTIMO = 0 (matching SVAO after restart)


Remark $\diamond$ Generate only when STIE $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1 (matching SVAO after restart)

(c) Start — Code - Data — Start — Code — Data — Stop
<1> When WTIMO = 0 (extension code reception after restart)

$\diamond 0: S M B S 0=00000010 \mathrm{~B}$

- 1: $\operatorname{SMBSO}=0010 \times 010 B$

4 2: SMBSO $=0010 \times 000 \mathrm{~B}$
③: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$
44: SMBSO $=0010 \times 000 \mathrm{~B}$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1 (extension code reception after restart)

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | $\Delta 1$ |  | $\triangle 3$ |  |  |  | -4 45 |  |  | $\triangle 6$ |  |

$\diamond 0: S M B S 0=00000010 \mathrm{~B}$
$\triangle 1$ : $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$

- 2: $\mathrm{SMBSO}=0010 \times 110 \mathrm{~B}$
$\triangle$ 3: $\mathrm{SMBSO}=0010 \times \times 00 \mathrm{~B}$
- 4: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$
- 5: SMBSO $=0010 \times 110 B$
- 6: SMBSO $=0010 \times \times 00 \mathrm{~B}$
$\triangle 7: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(d) Start — Code - Data - Start — Address — Data - Stop
<1> When WTIMO = 0 (unmatching address (except extension code) after restart)

$\diamond 0: S M B S 0=00000010 B$

- 1: SMBSO $=0010 \times 010 B$
- 2: SMBSO = 0010×000B

3: $\mathrm{SMBSO}=00000 \times 10 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
A Always generate
$\triangle$ Generate only when SPIEO $=1$
$\times$ Don't care
<2> When WTIMO = 1 (unmatching address (except extension code) after restart)

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\diamond 0$ |  | 1 |  |  |  |  |  | ^4 |  |  |  |

$\diamond 0:$ SMBSO $=00000010 \mathrm{~B}$
-1: $\mathrm{SMBSO}=0010 \times 010 \mathrm{~B}$
$\Delta$ 2: $S M B S 0=0010 \times 110 B$

- 3: SMBSO = 0010x×00B

④: $\mathrm{SMBSO}=00000 \times 10 \mathrm{~B}$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(4) Non-participation in communication
(a) Start - Code - Data - Data - Stop

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  |  |  |  |  |  |  |  |

$\diamond 0: S M B S O=00000010 \mathrm{~B}$
$\triangle 1: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 $=1$
$\triangle$ Generate only when SPIE0 $=1$
(5) Arbitration defeat operation (operation as slave after arbitration defeat)
(a) In case of arbitration defeat during slave address data transmission
<1> When WTIMO $=0$

$\diamond 0: S M B S O=10001010 B$
$\triangle$ 1: $\mathrm{SMBSO}=0101 \times 110 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)

- 2: $\mathrm{SMBSO}=0001 \times 000 \mathrm{~B}$
$\triangle$ 3: $\mathrm{SMBSO}=0001 \times 000 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | $\Delta 1$ |  |  |  | $\Delta 2$ | $\Delta 3$ |  |

$\diamond 0: S M B S 0=10001010 \mathrm{~B}$
-1: $\mathrm{SMBSO}=0101 \times 110 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)

- 2: SMBSO $=0001 \times 100 \mathrm{~B}$
- 3: $\mathrm{SMBSO}=0001 \times \times 00 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(b) In case of arbitration defeat during extension code transmission
<1> When WTIMO = 0

$\diamond 0: S M B S 0=10001010 \mathrm{~B}$
4 1: SMBS0 $=0110 \times 010 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
②: $\mathrm{SMBSO}=0010 \times 000 \mathrm{~B}$
$\triangle$ 3: $\mathrm{SMBSO}=0010 \times 000 \mathrm{~B}$
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP | P |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | ^1 |  | ③ |  |  |  |  |  |

$\diamond 0:$ SMBSO $=10001010 \mathrm{~B}$
© 1: SMBS0 $=0110 \times 010 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
A 2: SMBSO $=0010 \times 110 B$
3: SMBSO $=0010 \times 100 \mathrm{~B}$
44: $\mathrm{SMBSO}=0010 \times \times 00 \mathrm{~B}$
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(6) Arbitration defeat operation (non-participation after arbitration defeat)
(a) In case of arbitration defeat during slave address data transmission

$\diamond 0: S M B S O=10001010 B$
^1: SMBSO $=01000110 B$ (Example: Read ALD0 during interrupt processing)
$\triangle 2: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
(b) In case of arbitration defeat during extension code transmission

$\diamond 0: S M B S 0=10001010 B$
^ 1: SMBSO $=0110 \times 010 B$ (Example: Read ALD0 during interrupt processing,
LREL0 $=1$ set by software)
$\triangle 2: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(c) In case of arbitration defeat during data transmission
<1> When WTIM0 = 0

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | ^1 |  |  | $\triangle 2$ |  |  |  |

$\diamond 0: S M B S O=10001010 B$
^1: $S M B S 0=10001110 B$
© 2: SMBS0 $=01000000 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 3: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
<2> When WTIMO = 1

| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | $\begin{array}{ll}\text { AK } & \text { SP } \\ & \\ \end{array}$ |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\diamond 0$ |  | 41 |  |  | 42 |  |  |  |  |

$\diamond 0: S M B S 0=10001010 \mathrm{~B}$
^1: $S M B S 0=10001110 B$
A 2: $\mathrm{SMBSO}=01000100 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 3: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE $0=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
(d) In case of defeat by restart condition during data transmission
<1> Other than extension code (Example: Matching SVAO)

<2> Extension code

(e) In case of defeat by stop condition during data transmission

| ST | AD6 to AD0 | RW | AK | D7 to Dn | SP |
| :--- | :--- | :--- | :--- | ---: | ---: |
| $\diamond 0$ |  | $\mathbf{\Delta 1}$ |  |  |  |

$\diamond 0: S M B S 0=10001010 B$
A 1: $S M B S 0=1000 \times 110 B$
$\triangle 2: S M B S 0=01000001 B$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
Dn = D6 to D0
(f) In case of arbitration defeat by data low level while attempting to generate restart condition <1> When WTIMO = 0

|  |  |  |  |  |  | ${ }_{1}=1$ |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ST | AD6 to AD0 | RW | AK | D7 to D0 | AK | D7 to D0 | AK | D7 to D0 | AK | SP |
| $\checkmark 0$ |  | $\Delta 1$ |  |  | $\Delta^{2}$ ③ |  | $\triangle 4$ |  |  |  |

$\diamond 0: S M B S O=10001010 B$
A 1: $S M B S 0=1000 \times 110 B$
4 2: SMBSO $=1000 \times 000 B$
$\triangle$ 3: SMBSO $=1000 \times \times 00 \mathrm{~B}$
4 4: SMBSO $=10000000 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 5: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIEO $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

$\diamond 0: S M B S O=10001010 B$
^1: $S M B S 0=1000 \times 110 B$
A 2: $S M B S 0=1000 \times \times 00 B$
© 3: SMBS0 $=01000100 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 4: S M B S 0=00000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIE0 = 1
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(g) In case of arbitration defeat by stop condition while attempting to generate restart condition <1> When WTIM0 = 0

$\diamond 0: S M B S 0=10001010 B$
^1: SMBSO $=1000 \times 110 B$
$\Delta$ 2: $\mathrm{SMBSO}=1000 \times 000 \mathrm{~B}$

- 3: SMBSO $=1000 \times \times 00 B$
$\triangle 4: S M B S 0=01000001 \mathrm{~B}$

Remark $\diamond$ Generate only when STIEO $=1$
© Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

$\diamond 0: S M B S 0=10001010 \mathrm{~B}$
-1: $S M B S 0=1000 \times 110 B$
A 2: SMBSO $=1000 \times \times 00 B$
$\triangle 3: S M B S 0=01000001 B$

Remark $\diamond$ Generate only when STIEO $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(h) In case of arbitration defeat by data low level while attempting to generate a stop condition <1> When WTIMO = 0

$\diamond 0: S M B S 0=10001010 B$
$\triangle$ 1: SMBSO $=1000 \times 110 B$
42: $\mathrm{SMBSO}=1000 \times 000 \mathrm{~B}$
4 3: SMBSO $=1000 \times \times 00 B$
44: $\mathrm{SMBS} 0=01000000 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 5: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIEO $=1$
^ Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
<2> When WTIMO = 1

$\diamond 0: S M B S O=10001010 B$
A 1: $S M B S 0=1000 \times 110 B$

- 2: SMBSO $=1000 \times \times 00 B$
© 3: $\mathrm{SMBS} 0=01000000 \mathrm{~B}$ (Example: Read ALD0 during interrupt processing)
$\triangle 4: S M B S 0=00000001 B$

Remark $\diamond$ Generate only when STIE0 = 1

- Always generate
$\triangle$ Generate only when SPIE0 $=1$
$\times$ Don't care
(7) Slave operation (after STOP mode is released)
(a) Start - Address - Data - Data - Stop
<1> When WTIMO = 0

| ST | AD6-AD0 | RW | AK | D7-D0 | AK | D7-D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\boldsymbol{\Delta}$ - $\mathbf{\Delta 2}^{2}$ |  |  |  |  |  |  |  |  |

^1: SMBS0 $=0001$ X010B
4 2: $\mathrm{SMBSO}=0001 \mathrm{X000B}$
3: $\mathrm{SMBSO}=0001 \mathrm{X000B}$

Remark $\Delta$ Always generate
$\times$ Don't care
<2> When WTIMO = 1

| ST | AD6-AD0 | RW | AK | D7-D0 | AK | D7-D0 | AK | SP |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{\Delta 1}$ 边2 $\mathbf{\Delta 3}^{\text {2 }}$ |  |  |  |  |  |  |  |  |

Oscillation stabilization time

①: SMBS0 $=0001 \mathrm{X010B}$

- 2: $\mathrm{SMBS} 0=0001 \mathrm{X} 100 \mathrm{~B}$
- 3: SMBSO $=0001$ XX00B

Remark $\Delta$ Always generate
$\times$ Don't care

Cautions 1. Be sure to set STIEO = SPIEO $=0$ when releasing STOP mode upon address match. In this case however, the timeout count operation or stop operation cannot be controlled, because an interrupt is not generated even if a start or stop condition is output by another device during STOP mode operation.
2. When releasing STOP mode, the timeout count operation cannot be performed in the period from the start condition to oscillation stabilization, because an interrupt is not generated when a start condition is generated.

### 15.4.8 Interrupt request (INTSMBO) generation timing and wait control

INTSMBO generation and wait control can be performed at the timing indicated in Table 15-3 by setting bit 3 (WTIMO) of SMB control register 0 (SMBCO).

Table 15-3. INTSMB0 Generation Timing and Wait Control

| WTIM0 | AWTIMO | During Slave Operation |  |  | During Master Operation |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Address | Data Receive | Data Transmit | Address | Data Receive | Data Transmit |
| 0 | 0 | $9^{\text {Notes } 1,2}$ | $8^{\text {Note } 2}$ | $8^{\text {Note } 2}$ | 9 | 8 | 8 |
|  | 1 | $8^{\text {Notes } 1,2}$ |  |  |  |  |  |
| 1 | 0 | $9^{\text {Notes } 1,2}$ | $9^{\text {Note } 2}$ | $9^{\text {Note } 2}$ | 9 | 9 | 9 |
|  | 1 | $8^{\text {Notes } 1,2}$ |  |  |  |  |  |

Notes 1. INTSMBO and wait signals are generated by a slave at the falling edge of the 8th or 9th clock according to the setting of AWTIMO only when matching with the address of the SMB slave address register (SMBSVA0) occurs.
Moreover, at this time, an $\overline{A C K}$ signal is output regardless of the setting of bit 2 (ACKE0) of SMBCO. A slave that receives an extension code generates INTSMBO at the falling edge of the 8th clock.
2. If the address received does not match the address set in SMB slave address register 0 (SMBSVA0), the slave does not generate INTSMB0 and wait signals.

Remark Figures listed in Table 15-3 above indicate the number of serial clocks. Interrupt requests and wait control are synchronized with the falling edge of the serial clock.
(1) During address transmission/reception

- During slave operation: Interrupt and wait timing is set based on the conditions described in notes 1 and 2 above regardless of the WTIMO bit setting.
- During master operation: Interrupt and wait signals are generated at the falling edge of the 9th clock regardless of the WTIMO bit setting.


## (2) During data reception

- During master/slave operation: Interrupt and wait timing is set by the WTIMO bit.


## (3) During data transmission

- During master/slave operation: Interrupt and wait timing is set by the WTIMO bit.


## (4) Wait cancellation method

Waits can be canceled by one of the following four methods.

- Setting SMB control register 0 (SMBCO) bit 5 (WRELO) to 1
- Performing SMB shift register 0 (SMB0) write operation
- Setting a start condition (by setting SMBC0 bit 1 (STTO) to 1 )
- Setting a stop condition (by setting SMBCO bit 0 (SPTO) to 1 )

When 8-clock wait is selected $(W T I M O=0)$, the $\overline{A C K}$ output level must be determined before the wait status is released.
(5) Stop condition detection

An INTSMB0 signal is output when a stop condition is detected (only when SPIE0 $=1$ ).
(6) Start condition detection

An INTSMB0 signal is output when a start condition is detected (only when STIEO = 1).

### 15.4.9 Matching address detection method

In SMB mode, a particular slave device can be selected by sending that slave address to the master.
The detection of matching addresses is performed automatically by hardware. If a local address has been set in SMB slave address register 0 (SMBSVA0), and the slave address sent from the master matches the address set in SMBSVA0, or if the extension code is received, an INTSMBO interrupt request is generated.

### 15.4.10 Error detection

In SMB mode, because the status of the serial data bus (SDAO) during transmission is also input to SMB shift register 0 (SMB0), transmission errors can be detected by comparing the SMB0 data before transmission start and at transmission end. If the two data do not match, a transmission error is considered to have occurred.

### 15.4.11 Extension code

(1) An extension code is considered to have been received when the high four bits of the receive address are 0000 or 1111, and in this case the extension code receive flag (EXCO) is set and an interrupt request (INTSMBO) is generated at the falling edge of the 8th clock.
The local address stored in SMB slave address register 0 (SMBSVA0) is not affected.
(2) When $111110 \times \times$ is set to SMBSVAO and $111110 \times \times 0$ is transferred from the master during transfer of a 10-bit address, the following occurs. However, INTSMB0 is generated at the falling edge of the 8th clock.

- Matching high 4 bits: EXCO $=1^{\text {Note }}$
- Matching 7-bit data: $\mathrm{COIO}=1^{\text {Note }}$

Note EXC0: SMB status register 0 (SMBSO) bit 5
COIO: SMB status register 0 (SMBSO) bit 4
(3) Because the processing after an interrupt request is generated differs depending on the data that follows the extension code, it is performed by software. For instance, if operation as a slave is not desired following the reception of an extension code, set LRELO to 1 , in which case the following communication standby status is entered.

Table 15-4. Extension Code Bit Definition

| Slave Address | R/W Bit |  |
| :---: | :---: | :--- |
| 0000000 | 0 | General call address |
| 0000000 | 1 | Start byte |
| 0000001 | $\times$ | CBUS address |
| 0000010 | $\times$ | Address reserved for different bus format |
| $11110 \times \times$ | $\times$ | 10 -bit slave address specification |

Addresses reserved for system management bus are described below.

| Slave Address |  |
| :---: | :--- |
| 0001000 | SMB host |
| 0001100 | Response address for SMB alert |
| 1010001 | Default address of SMB device |
| $10010 \times \times$ | Free address |

### 15.4.12 Arbitration

If several masters output a start condition simultaneously (when STTO is set to 1 before STDO is set to $1^{\text {Note }}$ ), master communication is performed while adjusting the clock until data differs. This operation is referred to as arbitration.

A master defeated in arbitration sets the arbitration defeat flag (ALDO) of SMB status register 0 (SMBSO), and sets the SCLO and SDAO lines to $\mathrm{Hi}-\mathrm{Z}$ to release the bus.

Arbitration defeat is detected by software when ALDO = 1 at the next interrupt request generation timing (8th or 9th clock, stop condition detection, etc.).

For the interrupt generation timing, see 15.4.7 SMB0 interrupt (INTSMB0).

Note STDO: SMB status register 0 (SMBSO) bit 1
STTO: SMB control register 0 (SMBCO) bit 1

Figure 15-14. Arbitration Timing Examples


Table 15-5. Status at Arbitration and Interrupt Request Generation Timing

| Status at Arbitration | Interrupt Request Generation Timing |
| :---: | :---: |
| Address transmission in progress | Falling edge of 8th or 9th clock following byte transmission ${ }^{\text {Note } 1}$ |
| Read/write information following address transmission |  |
| Extension code transmission in progress |  |
| Read/write information following extension code transmission |  |
| Data transmission in progress |  |
| $\overline{\text { ACK }}$ transmission in progress following data transmission |  |
| Data transmission in progress, restart condition detection |  |
| Data transmission in progress, stop condition detection | During stop condition output (SPIE0 $=1)^{\text {Note } 2}$ |
| Attempt to output restart condition was made, but data was low level | Falling edge of 8th or 9th clock following byte transfer ${ }^{\text {Note } 1}$ |
| Attempt to output restart condition was made, but stop condition was detected | During stop condition output (SPIE0 $=1)^{\text {Note } 2}$ |
| Attempt to output stop condition was made, but data was low level | Falling edge of 8th or 9th clock following byte transfer ${ }^{\text {Note } 1}$ |
| Attempt to output restart condition was made, but SCLO was low level |  |

Notes 1. If WTIMO (bit 3 of SMB control register $0(S M B C O)=1$, an interrupt request is generated at the falling edge of the 9th clock. During reception of an extension code slave address when WTIMO $=0$, an interrupt request is generated at the falling edge of the 8th clock.
2. If there is a possibility of arbitration occurring, set SPIEO to 1 for master operation.

Remark SPIEO: SMB control register 0 (SMBCO) bit 4

### 15.4.13 Wakeup function

The SMB0 slave function generates an interrupt request (INTSMB0) when a local address and extension code are received. This interrupt enables release of STOP mode and HALT mode.

When the address does not match, no unnecessary interrupt request is generated, allowing greater processing efficiency.

When a start condition is detected, the wakeup standby status is entered. Because even a master (when a start condition is output) may become a slave if defeated in arbitration, the wakeup standby status is entered while address transmission is performed.

However, when a stop condition is detected, interrupt request enable/disable is determined by setting bit 4 (SPIE0) of SMB control register 0 (SMBC0) regardless of the wakeup function.

### 15.4.14 Communication reservation

If, during non-participation on the bus, the next master communication is desired, a start condition can be made to be sent at bus release by performing communication reservation. Non-participation on the bus includes the following two statuses.

- When unit neither master nor slave during bus arbitration
- When extension code is received and unit does not operate as slave (released bus with SMB control register 0 $(\mathrm{SMBCO})$ bit $6($ LRELO $)=1$, without returning $\overline{\mathrm{ACK}})$.

When bit 1 (STTO) of SMBCO is set during non-participation on the bus, a start condition is generated automatically after the bus is released (following detection of stop condition), and the wait status is entered.

When bus release is detected (detection of stop condition), address transmission as master is started through a SMB shift register 0 (SMB0) write operation. At this time, set SMBCO bit 4 (SPIEO).

When STTO is set, whether operation as a start condition or operation as communication reservation is selected depends on the bus status.

- If bus is released $\qquad$ Start condition generation
- If bus is not released (standby status).... Communication reservation

The method to detect which operation is selected by STTO is to set STTO, and reconfirm the STTO bit after the wait time elapses.

Secure the wait time by software as shown in Table 15-6. The wait time is set by bit 3 (SMC0) of SMB clock selection register 0 (SMBCLO).

Table 15-6. Wait Time

| SMC0 | Wait Time |
| :---: | :--- |
| 0 | 46 clocks |
| 1 | 16 clocks |

The communication reservation timing is shown in Figure 15-15.

Figure 15-15. Communication Reservation Timing


SMB0: SMB shift register 0
STTO: SMB control register 0 (SMBCO) bit 1
STD0: SMB status register 0 (SMBSO) bit 1
SPDO: SMB status register 0 (SMBSO) bit 0

Communication reservations are received at the following timing. After bit 1 (STDO) of SMB status register 0 (SMBSO) becomes 1, communication reservation is done by setting bit 1 (STTO) of SMB control register 0 (SMBC0) to "1" before detection of a stop condition.

Figure 15-16. Communication Reservation Reception Timing


Figure 15-17 shows the communication reservation procedure.

Figure 15-17. Communication Reservation Procedure


Note During the communication reservation operation, execute writing to SMB shift register 0 (SMB0) using a stop condition interrupt.

### 15.4.15 Additional cautions

If, after reset, master communication is attempted from a status where no stop condition is detected (bus is not released), a stop condition must be generated and the bus released before performing master communication.

In the case of multiple masters, master communication cannot be performed while the bus is not released (stop condition not detected).

A stop condition is generated in the following sequence.
<1> Setting of SMB clock selection register 0 (SMBCLO)
$<2>$ Setting of SMB control register 0 (SMBCO) bit 7 (SMBEO)
$<3>$ Setting of SMBC0 bit 0 (SPT0)

### 15.4.16 Communication operation

(1) Master operation

The master operation sequence is illustrated below.
Figure 15-18. Master Operation Sequence


## (2) Slave operation

The slave operation sequence is illustrated below.

Figure 15-19. Slave Operation Sequence


### 15.5 Timing Charts

In SMB mode, a master can select for communication a slave device from among many such devices by outputting an address to the serial bus.

After the slave address, the master sends the TRC0 bit (bit 3 of SMB status register 0 (SMBSO)) indicating the data transmission direction and starts the serial communication with the slave.

The timing charts for data transmission are shown in Figures 15-20 and 15-21.
The shift operation of SMB shift register 0 (SMBO) is performed in synchronization with the falling edge of the serial clock (SCLO), send data is transmitted to the SOO latch and output MSB first from the SDA0 pin.

Data input to the SDA0 pin at the rising edge of SCLO is read by SMBO.

Figure 15-20. Master $\rightarrow$ Slave Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (1/3)
(1) Start condition - Address


Notes 1. An interrupt signal is output only when $\operatorname{STIEO}=1$.
2. An interrupt signal is output only when $E X C 0=1$.
3. An interrupt signal is output only when SPIEO $=1$.
4. Perform slave wait cancellation by either changing SMBO $\leftarrow \mathrm{FFH}$, or setting WRELO.

Figure 15-20. Master $\rightarrow$ Slave Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (2/3)
(2) Data

Master Device Processing


Note Perform slave wait cancellation by either changing SMBO $\leftarrow \mathrm{FFH}$, or setting WRELO.

Figure 15-20. Master $\rightarrow$ Slave Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (3/3)

## (3) Stop condition



Notes 1. Perform slave wait cancellation by either changing SMBO $\leftarrow F F H$, or setting WRELO.
2. An interrupt signal is output only when SPIE $0=1$.
3. An interrupt signal is output only when $\mathrm{STIEO}=1$.

Figure 15-21. Slave $\rightarrow$ Master Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (1/3)
(1) Start condition - Address


Notes 1. Only when STIEO $=1$.
2. Perform slave wait cancellation by either changing SMBO $\leftarrow \mathrm{FFH}$, or setting WRELO.

Figure 15-21. Slave $\rightarrow$ Master Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (2/3)
(2) Data


Note Perform slave wait cancellation by either changing SMBO $\leftarrow \mathrm{FFH}$, or setting WRELO.

Figure 15-21. Slave $\rightarrow$ Master Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) (3/3)
(3) Stop condition


Notes 1. Perform slave wait cancellation by either changing SMBO $\leftarrow \mathrm{FFH}$, or setting WRELO.
2. An interrupt signal is output only when SPIE $0=1$.
3. An interrupt signal is output only when $\operatorname{STIEO}=1$.

## CHAPTER 16 MULTIPLIER

### 16.1 Multiplier Function

The multiplier has the following function.

- Calculation of 8 bits $\times 8$ bits $=16$ bits


### 16.2 Multiplier Configuration

(1) 16-bit multiplication result storage register 0 (MULO)

This register stores the 16 -bit result of multiplication.
This register holds the result of multiplication after 16 CPU clocks have elapsed.
MULO is set with a 16-bit memory manipulation instruction.
RESET input makes this register undefined.

Caution MULO is designed to be manipulated with a 16-bit memory manipulation instruction. It can also be manipulated with 8-bit memory manipulation instructions, however. When an 8-bit memory manipulation instruction is used to manipulate MULO, it must be accessed using direct addressing.
(2) Multiplication data registers $A$ and $B$ (MRAO and MRBO)

These are 8 -bit multiplication data storage registers. The multiplier multiplies the values of MRAO and MRBO.
MRAO and MRB0 are set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input makes these registers undefined.
Figure $16-1$ shows a block diagram of the multiplier.

Figure 16-1. Block Diagram of Multiplier


### 16.3 Multiplier Control Register

The multiplier is controlled by the following register.

- Multiplier control register 0 (MULCO)

MULC0 indicates the operating status of the multiplier, as well as controls the multiplier.
MULCO is set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears this register to 00 H .

Figure 16-2. Format of Multiplier Control Register 0


## Caution Bits $\mathbf{1}$ to $\mathbf{7}$ must all be set to $\mathbf{0}$.

### 16.4 Multiplier Operation

The multiplier of the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries can execute calculation of 8 bits $\times 8$ bits $=16$ bits.

Figure $16-3$ shows the operation timing of the multiplier where MRA0 is set to AAH and MRB0 is set to D3H.
<1> Counting is started by setting MULSTO.
<2> The data generated by the selector is added to the data of MULO at each CPU clock, and the counter value is incremented by one.
<3> If MULST0 is cleared when the counter value is 111B, the operation is stopped. At this time, MULO holds the data.
<4> While MULST0 is low, the counter and slave are cleared.

Figure 16-3. Multiplier Operation Timing (Example of AAH $\times$ D3H)


## CHAPTER 17 INTERRUPT FUNCTIONS

### 17.1 Interrupt Function Types

The following two types of interrupt functions are used.
(1) Non-maskable interrupt

This interrupt is acknowledged unconditionally. It does not undergo interrupt priority control and is given top priority over all other interrupt requests.
A standby release signal is generated.
The non-maskable interrupt has one interrupt source the watchdog timer.

## (2) Maskable interrupt

These interrupts undergo mask control. If two or more interrupts are simultaneously generated, each interrupt has a predetermined priority as shown in Table 17-1.
A standby release signal is generated.
For the $\mu$ PD789167 and 789177 Subseries, maskable interrupts have four external interrupt sources and ten internal interrupts sources. For the $\mu$ PD789167Y and 789177Y Subseries, maskable interrupts have four external interrupt sources of and 12 internal interrupt sources.

### 17.2 Interrupt Sources and Configuration

There are a total of 15 non-maskable and maskable interrupt sources for the $\mu$ PD789167 and 789177 Subseries, and a total of 17 non-maskable and maskable interrupt sources for the $\mu \mathrm{PD} 789167 \mathrm{Y}$ and 789177Y Subseries (see Table 17-1).

Table 17-1. Interrupt Sources

| Interrupt Type | Priority ${ }^{\text {Note } 1}$ | Interrupt Source |  | Internal/External | Vector Table <br> Address | Basic <br> Configuration <br> Type ${ }^{\text {Note } 2}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Name | Trigger |  |  |  |
| Non-maskable interrupt | - | INTWDT | Watchdog timer overflow (when watchdog timer mode 1 is selected) | Internal | 0004H | (A) |
| Maskable interrupt | 0 | INTWDT | Watchdog timer overflow (when interval timer mode is selected) |  |  | (B) |
|  | 1 | INTPO | Pin input edge detection | External | 0006H | (C) |
|  | 2 | INTP1 |  |  | 0008H |  |
|  | 3 | INTP2 |  |  | 000AH |  |
|  | 4 | INTP3 |  |  | 000 CH |  |
|  | 5 | INTSR20 | End of UART reception on serial interface 20 | Internal | 000EH | (B) |
|  |  | INTCSI20 | End of three-wire SIO transfer reception on serial interface 20 |  |  |  |
|  | 6 | INTST20 | End of UART transmission on serial interface 20 |  | 0010H |  |
|  | 7 | INTWT | Watch timer interrupt |  | 0012H |  |
|  | 8 | INTWTI | Interval timer interrupt |  | 0014H |  |
|  | 9 | INTTM80 | Generation of match signal for 8-bit timer/event counter 80 |  | 0016H |  |
|  | 10 | INTTM81 | Generation of match signal for 8-bit timer/event counter 81 |  | 0018H |  |
|  | 11 | INTTM82 | Generation of match signal for 8-bit timer 82 |  | 001 AH |  |
|  | 12 | INTTM90 | Generation of match signal for 16-bit timer 90 |  | 001 CH |  |
|  | 13 | INTSMB0 ${ }^{\text {Note } 3}$ | SMB interrupt |  | 001EH |  |
|  | 14 | INTSMBOV0 | SMB timeout interrupt |  | 0020H |  |
|  | 15 | INTADO | A/D conversion completion signal |  | 0022H |  |

Notes 1. The priority regulates which maskable interrupt is higher when two or more maskable interrupts are generated simultaneously. Zero signifies the highest priority, and 15 is the lowest.
2. Basic configuration types (A), (B), and (C) correspond to (A), (B), and (C) in Figure 17-1, respectively.
3. For the $\mu$ PD789167Y and 789177Y Subseries only

Figure 17-1. Basic Configuration of Interrupt Function
(A) Internal non-maskable interrupt

(B) Internal maskable interrupt

(C) External maskable interrupt


IF: Interrupt request flag
IE: Interrupt enable flag
MK: Interrupt mask flag

### 17.3 Interrupt Function Control Registers

The interrupt functions are controlled by the following registers.

- Interrupt request flag registers 0 and 1 (IF0 and IF1)
- Interrupt mask flag registers 0 and 1 (MK0 and MK1)
- External interrupt mode registers 0 and 1 (INTM0 and INTM1)
- Program status word (PSW)

Table 17-2 lists interrupt requests, the corresponding interrupt request flags, and interrupt mask flags.

Table 17-2. Interrupt Request Signals and Corresponding Flags

| Interrupt Request Signal | Interrupt Request Flag | Interrupt Mask Flag |
| :---: | :---: | :---: |
| INTWDT | TMIF4 | TMMK4 |
| INTPO | PIFO | PMKO |
| INTP1 | PIF1 | PMK1 |
| INTP2 | PIF2 | PMK2 |
| INTP3 | PIF3 | PMK3 |
| INTSR20/INTCSI20 | SRIF20 | SRMK20 |
| INTST20 | STIF20 | STMK20 |
| INTWT | WTIF | WTMK |
| INTWTI | WTIIF | WTIMK |
| INTTM80 | TMIF80 | TMMK80 |
| INTTM81 | TMIF81 | TMMK81 |
| INTTM82 | TMIF82 | TMMK82 |
| INTTM90 | TMIF90 | TMMK90 |
| INTSMB0 ${ }^{\text {Note }}$ | SMBIFO ${ }^{\text {Note }}$ | SMBMK0 ${ }^{\text {Note }}$ |
| INTSMBOV0 ${ }^{\text {Note }}$ | SMBOVIFO ${ }^{\text {Note }}$ | SMBOVMK0 ${ }^{\text {Note }}$ |
| INTADO | ADIFO | ADMKO |

Note For the $\mu$ PD789167Y and 789177Y Subseries only
(1) Interrupt request flag registers (IF0 and IF1)

An interrupt request flag is set to 1 when the corresponding interrupt request is issued, or when the related instruction is executed. It is cleared to 0 when the interrupt request is acknowledged, when a $\overline{\text { RESET }}$ signal is input, or when a related instruction is executed.

IF0 and IF1 are set with a 1-bit or 8-bit memory manipulation instruction.


Figure 17-2. Format of Interrupt Request Flag Register

| Symbol | <7> | <6> <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| IF0 | WTIF | STIF20SRIF20 | PIF3 | PIF2 | PIF1 | PIFO | TMIF4 | FFEOH | OOH | R/W |
|  | <7> | <6> <5> | <4> | <3> | <2> | <1> | <0> |  |  |  |
| IF1 | ADIFO | $\begin{array}{r} \text { Notel } \\ \text { SMBOVIFOS } \\ \text { NMBIFO } \end{array}$ | TMIF90 | TMIF82\| | TMIF81 | TMIF80 | WTIIF | FFE1H | 00H | R/W |
|  | $x \times$ IF $\times$ |  |  |  |  |  | Interrupt | quest flag |  |  |
|  | 0 | No interrupt requ | quest | signal has | as been | issued |  |  |  |  |
|  | 1 | An interrupt req | quest | signal ha | as been | issued | d; an inter | pt request | as been mad |  |

Note This flag is provided for the $\mu$ PD789167Y and 789177Y Subseries only. For the $\mu$ PD789167 and 789177 Subseries, the flag must be set to 0 .

Cautions 1. The TMIF4 flag can be read- and write-accessed only when the watchdog timer is being used as an interval timer. It must be cleared to 0 if the watchdog timer is used in watchdog timer mode 1 or 2.
2. When port 3 is being used as an output port, and its output level is changed, an interrupt request flag is set, because this port is also used as an external interrupt input. To use port 3 in output mode, therefore, the interrupt mask flag must be set to 1 in advance.
3. When an interrupt is acknowledged, the interrupt routine is entered after the interrupt request flag has been automatically cleared.
(2) Interrupt mask flag registers (MK0 and MK1)

The interrupt mask flags are used to enable and disable the corresponding maskable interrupts.
MK0 and MK1 are set with a 1-bit or 8-bit memory manipulation instruction.
$\overline{\text { RESET input sets MKO and MK1 to FFH. }}$

Figure 17-3. Format of Interrupt Mask Flag Register

| Symbol | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| MKO | WTMK | STMK20 | SRMK20 | PMK3 | PMK2 | PMK1 | PMK0 | TMMK4 | FFE4H | FFH | R/W |
|  | <7> | <6> | <5> | <4> | <3> | <2> | <1> | <0> |  |  |  |
| MK1 | ADMKO | \| $\begin{gathered}\text { Note } \\ \text { SMBOVMK0 } \\ \text { \| }\end{gathered}$ | SMBMK0 | TMMK90 | TMMK82 | TMMK81 | TMMK80 | WTIMK | FFE5H | FFH | R/W |
|  | x $\times$ MK $\times$ | Interrupt handling control |  |  |  |  |  |  |  |  |  |
|  | 0 | Enable interrupt handling. |  |  |  |  |  |  |  |  |  |
|  | 1 | Disable interrupt handling. |  |  |  |  |  |  |  |  |  |

Note This flag is provided for the $\mu$ PD789167Y and 789177Y Subseries only. For the $\mu$ PD789167 and 789177 Subseries, the flag must be set to 1.

Cautions 1. When the watchdog timer is being used in watchdog timer mode 1 or 2 , any attempt to read TMMK4 flag results in an undefined value being detected.
2. When port 3 is being used as an output port, and its output level is changed, an interrupt request flag is set, because this port is also used as an external interrupt input. To use port 3 in output mode, therefore, the interrupt mask flag must be set to 1 in advance.

## (3) External interrupt mode register 0 (INTMO)

INTM0 is used to specify the valid edge for INTPO to INTP2.
INTMO is set with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input clears INTMO to 00 H .

Figure 17-4. Format of External Interrupt Mode Register 0

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTMO | ES21 | ES20 | ES11 | ES10 | ES01 | ES00 | 0 | 0 |  | FFECH | $00 H$ | R/W


| ES21 | ES20 |  | INTP2 valid edge selection |
| :---: | :---: | :--- | :--- |
| 0 | 0 | Falling edge |  |
| 0 | 1 | Rising edge |  |
| 1 | 0 | Setting prohibited |  |
| 1 | 1 | Both rising and falling edges |  |


| ES11 | ES10 |  | INTP1 valid edge selection |
| :---: | :---: | :--- | :--- |
| 0 | 0 | Falling edge |  |
| 0 | 1 | Rising edge |  |
| 1 | 0 | Setting prohibited |  |
| 1 | 1 | Both rising and falling edges |  |


| ES01 | ES00 |  | INTP0 valid edge selection |
| :---: | :---: | :--- | :--- |
| 0 | 0 | Falling edge |  |
| 0 | 1 | Rising edge |  |
| 1 | 0 | Setting prohibited |  |
| 1 | 1 | Both rising and falling edges |  |

## Cautions 1. Bits 0 and 1 must be set to 0 .

2. Before setting INTMO, set the corresponding interrupt mask flag register ( $\times \times \mathrm{MK} \times$ ) to 1 to disable interrupts.
To enable interrupts, clear the corresponding interrupt request flag ( $x \times 1 \mathrm{~F} \times$ ) to 0 , then clear the corresponding interrupt mask flag register ( $x \times \mathrm{MK} \times$ to 0 ).

## (4) External interrupt mode register 1 (INTM1)

INTM1 is used to specify the valid edge for INTP3.
INTM1 is set with an 8-bit memory manipulation instruction.
RESET input clears INTM1 to 00 H .

Figure 17-5. Format of External Interrupt Mode Register 1

| Symbol | 7 | 6 | 5 |  | 3 | 2 | 1 | 0 | Address <br> FFEDH | After reset$\mathrm{OOH}$ | $\begin{aligned} & \text { R/W } \\ & \text { R/W } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| INTM1 | 0 | 0 | 0 |  | 0 | 0 | ES31 | ES30 |  |  |  |
|  | ES31 | ES30 |  |  |  | INT | d edge | ection |  |  |  |
|  | 0 | 0 | Falling |  |  |  |  |  |  |  |  |
|  | 0 | 1 | Rising |  |  |  |  |  |  |  |  |
|  | 1 | 0 | Setting |  |  |  |  |  |  |  |  |
|  | 1 | 1 | Both ris | d |  |  |  |  |  |  |  |

## Cautions 1. Bits $\mathbf{2}$ to $\mathbf{7}$ must be set to 0 .

2. Before setting INTM1, set PMK3 to 1 to disable interrupts.

To enable interrupts, clear PIF3 to 0 , then clear PMK3 to 0.
(5) Program status word (PSW)

The program status word is used to hold the instruction execution result and the current status of the interrupt requests. The IE flag, used to enable and disable maskable interrupts, is mapped to the PSW.
The PSW can be read- and write-accessed in 8-bit units, as well as using bit manipulation instructions and dedicated instructions (El and DI). When a vector interrupt is acknowledged, the PSW is automatically saved to a stack, and the IE flag is reset to 0 .
RESET input sets PSW to 02H.
Figure 17-6. Program Status Word Configuration


| IE | Whether to enable/disable interrupt acknowledgment |  |
| :---: | :--- | :--- |
| 0 | Disable |  |
| 1 | Enable |  |

### 17.4 Interrupt Processing Operation

### 17.4.1 Non-maskable interrupt request acknowledgment operation

A non-maskable interrupt request is unconditionally acknowledged even when interrupts are disabled. It is not subject to interrupt priority control and takes precedence over all other interrupts.

When a non-maskable interrupt request is acknowledged, the PSW and PC are saved to the stack in that order, the IE flag is reset to 0 , the contents of the vector table are loaded to the PC, and then program execution branches.

Figure 17-7 shows the flowchart from non-maskable interrupt request generation to acknowledgment. Figure 17-8 shows the timing of non-maskable interrupt request acknowledgment. Figure 17-9 shows the acknowledgment operation if multiple non-maskable interrupts are generated.

Caution During non-maskable interrupt service program execution, do not input another non-maskable interrupt request; if it is input, the service program will be interrupted and the new interrupt request will be acknowledged.

Figure 17-7. Flowchart from Non-Maskable Interrupt Request Generation to Acknowledgment


WDTM: Watchdog timer mode register
WDT: Watchdog timer

Figure 17-8. Timing of Non-Maskable Interrupt Request Acknowledgment


Figure 17-9. Acknowledgment Non-Maskable Interrupt Request


### 17.4.2 Maskable interrupt request acknowledgment operation

A maskable interrupt request can be acknowledged when the interrupt request flag is set to 1 and the corresponding interrupt mask flag is cleared to 0 . A vectored interrupt request is acknowledged in the interrupt enabled status (when the IE flag is set to 1).

The time required to start the interrupt servicing after a maskable interrupt request has been generated is shown in Table 17-3.

See Figures 17-11 and 17-12 for the interrupt request acknowledging timing.

Table 17-3. Time from Generation of Maskable Interrupt Request to Processing

| Minimum Time | Maximum Time ${ }^{\text {Note }}$ |
| :--- | :--- |
| 9 clocks | 19 clocks |

Note The wait time is maximum when an interrupt request is generated immediately before the BT and BF instruction.

Remark 1 clock: $\frac{1}{\text { fcPu }}$ (fcPu: CPU clock)
When two or more maskable interrupt requests are generated at the same time, they are acknowledged starting from the interrupt request assigned the highest priority.

A pending interrupt is acknowledged when the status where it can be acknowledged is set.
Figure 17-10 shows the algorithm of acknowledging interrupt requests.
When a maskable interrupt request is acknowledged, the contents of the PSW and PC are saved to the stack in that order, the IE flag is reset to 0 , and the data in the vector table determined for each interrupt request is loaded to the PC, and execution branches.

To return from interrupt processing, use the RETI instruction.

Figure 17-10. Interrupt Request Acknowledgment Processing Algorithm

$\times \times$ IF: $\quad$ Interrupt request flag
$x \times$ MK: Interrupt mask flag
IE: $\quad$ Flag to control maskable interrupt request acknowledgment ( $1=$ Enabled, $0=$ Disabled $)$

Figure 17-11. Interrupt Request Acknowledgment Timing (Example of MOV A,r)


Interrupt $\qquad$ $\square$

If an interrupt request flag ( $x \times \mathrm{IF}$ ) is set before instruction clock $\mathrm{n}(\mathrm{n}=4$ to 10 ) under execution becomes $\mathrm{n}-1$, the interrupt is acknowledged after the instruction under execution's completed. Figure 17-11 shows an example of the interrupt request acknowledgment timing for an 8-bit data transfer instruction MOV A,r. Since this instruction is executed in 4 clocks, if an interrupt occurs within 3 clocks after the execution starts, the interrupt acknowledgment processing is performed after the MOV A,r instruction is completed.

Figure 17-12. Interrupt Request acknowledgment Timing (When Interrupt Request Flag Is Generated at Last Clock During Instruction Execution)


Interrupt $\qquad$几

If an interrupt request flag ( $x \times \mathrm{IF}$ ) is set at the last clock of the instruction, the interrupt acknowledgment processing starts after the next instruction is executed. Figure 17-12 shows an example of the interrupt acknowledgment timing for an interrupt request flag that is set at the second clock of NOP (2-clock instruction). In this case, the MOV A,r instruction after the NOP instruction is executed, and then the interrupt acknowledgment processing is performed.

## Caution Interrupt requests are reserved while interrupt request flag register 0 or 1 (IF0 or IF1) or the interrupt mask flag register 0 or 1 (MK0 or MK1) is being accessed.

### 17.4.3 Multiple interrupt processing

Multiple interrupt processing in which another interrupt is acknowledged while an interrupt is being serviced can be processed by priority. When two or more interrupts are generated at once, interrupt servicing is performed according to the priority assigned to each interrupt request in advance (see Table 17-1).

Figure 17-13. Example of Multiple Interrupts
Example 1. Multiple interrupt is acknowledged


During interrupt INTxx servicing, interrupt request INTyy is acknowledged, and multiple interrupts are generated. An El instruction is issued before each interrupt request acknowledgment, and the interrupt request acknowledgment enabled state is set.

Example 2. Multiple interrupts are not generated because interrupts are not enabled


Because interrupts are not enabled in interrupt INTxx servicing (the El instruction was not issued), interrupt request INTyy is not acknowledged, and multiple interrupts are not generated. The INTyy request is held pending and acknowledged after INTxx servicing is performed.

## IE = 0: Interrupt request acknowledgment disabled

### 17.4.4 Interrupt request hold

Some instructions may hold the acknowledgment of an instruction request pending until the completion of the execution of the next instruction even if the interrupt request (maskable interrupt, non-maskable interrupt, and external interrupt) is generated during the execution. The following shows such instructions (interrupt request hold instructions).

- Manipulation instruction for interrupt request flag registers 0 and 1 (IF0 and IF1)
- Manipulation instruction for interrupt mask flag registers 0 and 1 (MK0 and MK1)


## CHAPTER 18 STANDBY FUNCTION

### 18.1 Standby Function and Configuration

### 18.1.1 Standby function

The standby function is to used reduce the power consumption of the system and can be effected in the following two modes.

## (1) HALT mode

This mode is set when the HALT instruction is executed. HALT mode stops the operation clock of the CPU. The system clock oscillator continues oscillating. This mode does not reduce the current consumption as much as the STOP mode, but is useful for resuming processing immediately when an interrupt request is generated, or for intermittent operations.

## (2) STOP mode

This mode is set when the STOP instruction is executed. STOP mode stops the main system clock oscillator and stops the entire system. The current consumption of the CPU can be substantially reduced in this mode.
The low voltage ( $\mathrm{V}_{\mathrm{DD}}=1.8 \mathrm{~V} \mathrm{~min}$.) of the data memory can be retained. Therefore, this mode is useful for retaining the contents of the data memory at an extremely low current consumption.

STOP mode can be released by an interrupt request, so this mode can be used for intermittent operations. However, some time is required until the system clock oscillator stabilizes after STOP mode has been released. If processing must be resumed immediately by using an interrupt request, therefore, use the HALT mode.

In both modes, the previous contents of the registers, flags, and data memory before setting standby mode are all retained. In addition, the statuses of the output latches of the I/O ports and output buffers are also retained.

## Caution To set STOP mode, be sure to stop the operations of the peripheral hardware, and then execute

 the STOP instruction.
### 18.1.2 Standby function control register

The wait time after STOP mode is released upon interrupt request until the oscillation stabilizes is controlled with the oscillation stabilization time selection register (OSTS).

OSTS is set with an 8-bit memory manipulation instruction.
$\overline{\text { RESET }}$ input sets OSTS to 04 H . However, the oscillation stabilization time after $\overline{\text { RESET }}$ input is $2^{15} / \mathrm{fx}$, instead of $2^{17} / f x$.

Figure 18-1. Format of Oscillation Stabilization Time Selection Register

| Symbol | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 | Address | After reset | R/W |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OSTS | 0 | 0 | 0 | 0 | 0 | OSTS2 | OSTS1 | OSTS0 | FFFAH | 04H | R/W |


| OSTS2 | OSTS1 | OSTS0 | Oscillation stabilization time selection |  |  |
| :---: | :---: | :---: | :--- | :--- | :--- |
|  |  |  |  | At $\mathrm{fx}=10.0 \mathrm{MHz}$ operation ${ }^{\text {Note }}$ | At $\mathrm{fx}=5.0 \mathrm{MHz}$ operation |
| 0 | 0 | 0 | $2^{12} / \mathrm{fx}$ | $409 \mu \mathrm{~s}$ | $819 \mu \mathrm{~s}$ |
| 0 | 1 | 0 | $2^{15} / \mathrm{fx}$ | 3.27 ms | 6.55 ms |
| 1 | 0 | 0 | $2^{17} / \mathrm{fx}$ | 13.1 ms | 26.2 ms | | Other than above |
| :--- |

Note Expanded-specification products only.

Caution The wait time after STOP mode is released does not include the time from STOP mode release to clock oscillation start (" $a$ " in the figure below), regardless of release by $\overline{\text { RESET }}$ input or by interrupt generation.


Remark fx: Main system clock oscillation frequency

### 18.2 Operation of Standby Function

### 18.2.1 HALT mode

## (1) HALT mode

HALT mode is set by executing the HALT instruction. The operation status in HALT mode is shown in the following table.

Table 18-1. Operation Statuses in HALT Mode

| Item | HALT Mode Operation Status While Main System Clock Is Operating |  | HALT Mode Operation Status While Subsystem Clock Is Operating |  |
| :---: | :---: | :---: | :---: | :---: |
|  | While Subsystem Clock Is Operating | While Subsystem Clock Is Not Operating | While Main System Clock Is Operating | While Main System Clock Is Not Operating |
| Main system clock generator | Main system clock oscillation enabled |  |  | Does not operate |
| CPU | Operation disabled |  |  |  |
| Port (output latch) | Remains in the state existing before the selection of HALT mode |  |  |  |
| 16-bit timer (TM90) | Operation enabled | Operation enabled ${ }^{\text {Note } 1}$ | Operation enabled | Operation enabled ${ }^{\text {Note } 2}$ |
| 8 -bit timer/event counter (TM80) | Operation enabled |  |  | Operation enabled ${ }^{\text {Nooe } 3}$ |
| 8-bit timer/event counter (TM81) | Operation enabled |  |  | Operation enabled ${ }^{\text {Note } 4}$ |
| 8-bit timer (TM82) | Operation enabled | Operation enabled ${ }^{\text {Note } 1}$ | Operation enabled | Operation enabled ${ }^{\text {Note } 5}$ |
| Watch timer | Operation enabled | Operation enabled ${ }^{\text {Note } 1}$ | Operation enabled | Operation enabled ${ }^{\text {Note } 5}$ |
| Watchdog timer | Operation enabled |  | Operation disabled |  |
| Serial interface 20 | Operation enabled |  |  | Operation enabled ${ }^{\text {Note } 6}$ |
| SMB0 | Operation enabled |  |  | Operation enabled ${ }^{\text {Note } 7}$ |
| A/D converter | Operation disabled |  |  |  |
| Multiplier | Operation disabled |  |  |  |
| External interrupt | Operation enabled ${ }^{\text {Note }}$ 8 |  |  |  |

Notes 1. Operation is enabled when the main system clock is selected.
2. Operation is enabled when the subsystem clock is selected and when buzzer output is enabled (for details, see 8.5 Notes on 16-Bit Timer 90).
3. Operation is enabled only when TI80 is selected as the count clock.
4. Operation is enabled only when TI81 is selected as the count clock.
5. Operation is enabled when the subsystem clock is selected.
6. Operation is enabled in both 3-wire serial I/O and UART modes while an external clock is being used.
7. An interrupt can be generated when addresses match during the slave operation.
8. Maskable interrupt that is not masked

## (2) Releasing HALT mode

HALT mode can be released by the following three sources.

## (a) Releasing by unmasked interrupt request

HALT mode is released by an unmasked interrupt request. In this case, if the interrupt request is enabled to be acknowledged, vectored interrupt servicing is performed. If interrupts are disabled, the instruction at the next address is executed.

Figure 18-2. Releasing HALT Mode by Interrupt


Remarks 1. The broken lines indicate the case where the interrupt request that has released standby mode is acknowledged.
2. The wait time is as follows:

- When vectored interrupt servicing is performed: 9 to 10 clocks
- When vectored interrupt servicing is not performed: 1 to 2 clocks
(b) Releasing by non-maskable interrupt request

HALT mode is released regardless of whether interrupts are enabled or disabled, and vectored interrupt servicing is performed.
(c) Releasing by $\overline{\text { RESET }}$ input

When HALT mode is released by the RESET signal, execution branches to the reset vector address in the same manner as the ordinary reset operation, and program execution is started.

Figure 18-3. Releasing HALT Mode by RESET Input


Note 3.27 ms (at $\mathrm{fx}=10.0 \mathrm{MHz}$ operation), 6.55 ms (at $\mathrm{fx}=5.0 \mathrm{MHz}$ operation)

Table 18-2. Operation after Release of HALT Mode

| Releasing Source | MK $\times \times$ | IE |  |
| :--- | :---: | :---: | :--- |
| Maskable interrupt request | 0 | 0 | Operation |
|  | 0 | 1 | Executes next address instruction |
|  | 1 | $\times$ | Retains HALT mode |
| Non-maskable interrupt request | - | $\times$ | Executes interrupt servicing |
| $\overline{\text { RESET input }}$ | - | - | Reset processing |

$\times$ : Don't care

### 18.2.2 STOP mode

## (1) Setting and operation status of STOP mode

STOP mode is set by executing the STOP instruction.
Caution Because standby mode can be released by an interrupt request signal, standby mode is released as soon as it is set if there is an interrupt source whose interrupt request flag is set and interrupt mask flag is reset. When STOP mode is set, therefore, HALT mode is set immediately after the STOP instruction has been executed, the wait time set by the oscillation stabilization time selection register (OSTS) elapses, and then operation mode is set.

The operation status in STOP mode is shown in the following table.
Table 18-3. Operation Statuses in STOP Mode

| Item | STOP Mode Operation Status While Main System Clock Is Operating |  |
| :---: | :---: | :---: |
|  | While Subsystem Clock Is Operating | While Subsystem Clock Is Not Operating |
| Main system clock generator | Main system clock oscillation stopped |  |
| CPU | Operation disabled |  |
| Port (output latch) | Remains in the state existing before the selection of STOP mode |  |
| 16-bit timer (TM90) | Operation enabled ${ }^{\text {Note } 1}$ | Operation disabled |
| 8-bit timer/event counter (TM80) | Operation enabled ${ }^{\text {Note } 2}$ |  |
| 8 -bit timer/event counter (TM81) | Operation enabled ${ }^{\text {Note } 3}$ |  |
| 8-bit timer (TM82) | Operation enabled ${ }^{\text {Note } 4}$ | Operation disabled |
| Watch timer | Operation enabled ${ }^{\text {Note } 4}$ | Operation disabled |
| Watchdog timer | Operation disabled |  |
| Serial interface 20 | Operation enabled ${ }^{\text {Note } 5}$ |  |
| SMB0 | Operation enabled ${ }^{\text {Note } 6}$ |  |
| A/D converter | Operation disabled |  |
| Multiplier | Operation disabled |  |
| External interrupt | Operation enabled ${ }^{\text {Note } 7}$ |  |

Notes 1. Operation is enabled when the subsystem clock is selected and when buzzer output is enabled.
2. Operation is enabled only when TI80 is selected as the count clock.
3. Operation is enabled only when TI81 is selected as the count clock.
4. Operation is enabled when the subsystem clock is selected.
5. Operation is enabled in both 3-wire serial I/O and UART modes while an external clock is being used.
6. An interrupt can be generated when addresses match during the slave operation.
7. Maskable interrupt that is not masked

## (2) Releasing STOP mode

STOP mode can be released by the following two sources.

## (a) Releasing by unmasked interrupt request

STOP mode can be released by an unmasked interrupt request. In this case, if the interrupt is enabled to be acknowledged, vectored interrupt servicing is performed, after the oscillation stabilization time has elapsed. If the interrupt acknowledgment is disabled, the instruction at the next address is executed.

Figure 18-4. Releasing STOP Mode by Interrupt


Remark The broken lines indicate the case where the interrupt request that has released standby mode is acknowledged.
(b) Releasing by RESET input

When STOP mode is released by the RESET signal, the reset operation is performed after the oscillation stabilization time has elapsed.

Figure 18-5. Releasing STOP Mode by RESET Input


Note 3.27 ms (at $\mathrm{fx}=10.0 \mathrm{MHz}$ operation), 6.55 ms (at $\mathrm{fx}=5.0 \mathrm{MHz}$ operation)

Table 18-4. Operation After Release of STOP Mode

| Releasing Source | MK $\times x$ | IE |  |
| :--- | :---: | :---: | :--- |
| Maskable interrupt request | 0 | 0 | Executes next address instruction |
|  | 0 | 1 | Executes interrupt servicing |
|  | 1 | $\times$ | Retains STOP mode |
|  | - | - | Reset processing |

$\times$ : Don't care

## CHAPTER 19 RESET FUNCTION

The following two operations are available to generate reset signals.
(1) External reset input via RESET pin
(2) Internal reset by program loop time detected by watchdog timer

External and internal reset have no functional differences. In both cases, program execution starts at the address at 0000 H and 0001 H by reset signal input.

When a low level is input to the $\overline{\text { RESET }}$ pin or the watchdog timer overflows, a reset is applied and each hardware is set to the status shown in Table 19-1. Each pin is high impedance during reset input or during oscillation stabilization time just after reset release.

When a high level is input to the RESET pin, the reset is released and program execution is started after the oscillation stabilization time has elapsed. The reset applied by the watchdog timer overflow is automatically released after reset, and program execution is started after the oscillation stabilization time has elapsed (see Figures 19-2 through 19-4).

Cautions 1. For an external reset, input a low level for $10 \mu \mathrm{~s}$ or more to the RESET pin.
2. When STOP mode is released by reset, the STOP mode contents are held during reset input. However, the port pins become high impedance.

Figure 19-1. Block Diagram of Reset Function


Figure 19-2. Reset Timing by RESET Input


Figure 19-3. Reset Timing by Overflow in Watchdog Timer


Figure 19-4. Reset Timing by RESET Input in STOP Mode


Table 19-1. State of Hardware After Reset (1/2)

| Hardware |  | State After Reset |
| :---: | :---: | :---: |
| Program counter (PC) ${ }^{\text {Note } 1}$ |  | Loaded with the contents of the reset vector table (0000H, 0001H) |
| Stack pointer (SP) |  | Undefined |
| Program status word (PSW) |  | 02H |
| RAM | Data memory | Undefined ${ }^{\text {Note } 2}$ |
|  | General-purpose register | Undefined ${ }^{\text {Note } 2}$ |
| Ports (P0 to P3, P5, P6) (output latch) |  | 00H |
| Port mode registers (PM0 to PM3, PM5) |  | FFH |
| Pull-up resistor option registers (PU0, PUB2, PUB3) |  | 00H |
| Processor clock control register (PCC) |  | 02H |
| Suboscillation mode register (SCKM) |  | 00H |
| Subclock control register (CSS) |  | OOH |
| Oscillation stabilization time selection register (OSTS) |  | 04H |
| 16-bit timer 90 | Timer counter (TM90) | 0000H |
|  | Compare register (CR90) | FFFFH |
|  | Capture register (TCP90) | Undefined |
|  | Mode control register (TMC90) | 00H |
|  | Buzzer output control register (BZC90) | OOH |
| 8 -bit timer/event counters 80 to 82 | Timer counters (TM80 to TM82) | OOH |
|  | Compare registers (CR80 to CR82) | Undefined |
|  | Mode control registers (TMC80 to TMC82) | 00H |
| Watch timer | Mode control register (WTM) | OOH |
| Watchdog timer | Timer clock selection register (TCL2) | OOH |
|  | Mode register (WDTM) | 00H |
| A/D converter | Mode register (ADM0) | OOH |
|  | A/D input selection register (ADS0) | OOH |
|  | A/D conversion result register (ADCRO) | Undefined |
| Serial interface 20 | Mode register (CSIM20) | 00H |
|  | Asynchronous serial interface mode register (ASIM20) | 00H |
|  | Asynchronous serial interface status register (ASIS20) | OOH |
|  | Baud rate generator control register (BRGC20) | OOH |
|  | Transmission shift register (TXS20) | FFH |
|  | Reception buffer register (RXB20) | Undefined |

Notes 1. While a reset signal is being input, and during the oscillation stabilization period, the contents of the PC will be undefined, while the remainder of the hardware will be the same as after the reset.
2. In standby mode, the RAM enters the hold state after a reset.

Table 19-1. State of Hardware After Reset (2/2)

| Hardware |  | State After Reset |
| :---: | :---: | :---: |
| SMB0 | Control register (SMBCO) | OOH |
|  | Status register (SMBSO) | OOH |
|  | Clock selection register (SMBCLO) | OOH |
|  | Slave address register (SMBSVAO) | OOH |
|  | Mode register (SMBMO) | 2 H |
|  | Input level setting register (SMBVIO) | OOH |
|  | Shift register (SMBO) | OOH |
| Multiplier | 16-bit multiplication result storage register (MULO) | Undefined |
|  | Multiplication data registers (MRAO, MRB0) | Undefined |
|  | Multiplier control register (MULCO) | OOH |
| Interrupts | Request flag registers (IF0, IF1) | OOH |
|  | Mask flag registers (MK0, MK1) | FFH |
|  | External interrupt mode registers (INTM0, INTM1) | OOH |

## CHAPTER 20 FLASH MEMORY VERSION

The $\mu$ PD78F9177, $\mu$ PD78F9177A, $\mu$ PD78F9177A(A), and $\mu$ PD78F9177A(A1) are flash memory versions of the $\mu$ PD789177 Subseries.

The $\mu$ PD78F9177Y, $\mu$ PD78F9177AY, and $\mu$ PD78F9177AY(A) are flash memory versions of the $\mu$ PD789177Y Subseries.

The $\mu$ PD78F9177, $\mu$ PD78F9177A, $\mu$ PD78F9177A(A) , and $\mu$ PD78F9177A(A1) replace the internal ROM of the $\mu$ PD789167 and 789177 Subseries with flash memory, while the $\mu$ PD78F9177Y, $\mu$ PD78F9177AY, and $\mu$ PD78F9177AY(A) replace the internal ROM of the $\mu$ PD789167Y and 789177 Y Subseries with flash memory. The differences between the flash memory and mask ROM versions are shown in Table 20-1.

Table 20-1. Differences Between Flash Memory and Mask ROM Versions

| Item |  | Flash Memory |  | Mask ROM |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | $\mu$ PD78F9177A $\mu$ PD78F9177AY $\mu$ PD78F9177A(A) $\mu$ PD78F9177AY(A) $\mu$ PD78F9177AY(A1) | $\mu$ PD78F9177 <br> $\mu$ PD78F9177Y | $\mu$ PD789166 <br> $\mu$ PD789166Y $\mu$ PD789166(A) $\mu \mathrm{PD789166Y(A)}$ $\mu$ PD789166(A1) $\mu$ PD789166(A2) | $\mu$ PD789167 <br> $\mu$ PD789167Y <br> $\mu$ PD789167(A) <br> $\mu$ PD789167Y(A) <br> $\mu$ PD789167(A1) <br> $\mu$ PD789167(A2) | $\mu$ PD789176 $\mu$ PD789176Y $\mu$ PD789176(A) $\mu$ PD789176Y(A) $\mu$ PD789176(A1) $\mu$ PD789176(A2) | $\mu$ PD789177 <br> $\mu$ PD789177Y <br> $\mu$ PD789177(A) <br> $\mu$ PD789177Y(A) <br> $\mu$ PD789177(A1) <br> $\mu$ PD789177(A2) |
| Internal memory | ROM structure | Flash Memory |  | Mask ROM |  |  |  |
|  | ROM capacity | 24 KB |  | 16 KB | 24 KB | 16 KB | 24 KB |
|  | High- <br> speed <br> RAM | 512 bytes |  |  |  |  |  |
| Minimum instruction execution time |  | $0.2 \mu \mathrm{~s}$ <br> (at 10 MHz <br> operation) | $0.4 \mu \mathrm{~s}$ <br> (at 5 MHz <br> operation) | Expanded-specification products: (A1) products, (A2) products, and conventional products: |  | $0.2 \mu \mathrm{~s}$ (at 10 MHz operation) <br> $0.4 \mu \mathrm{~s}$ (at 5 MHz operation) |  |
| A/D converter resolution |  | 10 bits |  | 8 bits |  | 10 bits |  |
| Specification of on-chip pull-up resistors for P50 to P53 by mask option |  | Disabled |  | Enabled |  |  |  |
| Vpp pin |  | Provided |  | Not provided |  |  |  |
| Electrical specifications |  | Refer to the ELECTRICAL SPECIFICATIONS chapters. |  |  |  |  |  |

Cautions 1. There are differences in the amount of noise immunity and noise radiation between the flash memory and mask ROM versions. When pre-producing an application set with the flash memory version and then mass producing it with the mask ROM version, be sure to conduct sufficient evaluations on the commercial samples (CS) (not engineering sample, ES) of the mask ROM version.
2. When using $A / D$ conversion result register $O$ (ADCRO) with an 8-bit A/D converter ( $\mu$ PD789167 and 789167Y Subseries), manipulate with an 8-bit memory manipulation instruction; when using it with a 10-bit A/D converter ( $\mu$ PD789177 and 789177Y Subseries), use a 16-bit memory manipulation instruction.

When the flash memory version of the $\mu$ PD789167 or 789167 Y Subseries, is used, however, ADCRO can be manipulated with an 8-bit memory manipulation instruction. In this case, use an object file assembled with the $\mu$ PD789167 or 789167 Y Subseries.

### 20.1 Flash Memory Characteristics

Flash memory programming is performed by connecting a dedicated flash programmer (Flashpro III (part no. FLPR3, PG-FP3) or Flashpro IV (part no. FL-PR4, PG-FP4)) to the target system with the $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY mounted on the target system (on-board). A flash memory program adapter (FA adapter), which is a target board used exclusively for programming, is also provided.

Remark FL-PR3, FL-PR4, and the program adapter are the products made by Naito Densei Machida Mfg. Co., Ltd. (TEL +81-45-475-4191).

Programming using flash memory has the following advantages.

- Software can be modified after the microcontroller is solder-mounted on the target system.
- Distinguishing software facilities low-quantity, varied model production
- Easy data adjustment when starting mass production


### 20.1.1 Programming environment

The following shows the environment required for $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY flash memory programming.

When Flashpro III or Flashpro IV is used as a dedicated flash programmer, a host machine is required to control the dedicated flash programmer. Communication between the host machine and flash programmer is performed via RS-232C/USB (Rev. 1.1).

For details, refer to the manual for Flashpro III or Flashpro IV.

Remark USB is supported by Flashpro IV only.

Figure 20-1. Environment for Writing Program to Flash Memory


### 20.1.2 Communication mode

Use the communication mode shown in Table 20-2 to perform communication between the dedicated flash programmer and $\mu \mathrm{PD} 78 \mathrm{~F} 9177, \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{Y}, ~ \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{~A}$, and $\mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{AY}$.

Table 20-2. Communication Mode List

| Communication Mode | TYPE Setting ${ }^{\text {Note } 1}$ |  |  |  |  | Pins Used | Number of VPP Pulses |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | COMM PORT | SIO Clock | CPU Clock ${ }^{\text {Notes } 2,3}$ |  | Multiple Rate |  |  |
|  |  |  | In Flashpro | On Target Board |  |  |  |
| 3-wire serial I/O (SIO3) | SIO ch-0 <br> (3-wired, sync.) | $\begin{aligned} & 100 \mathrm{~Hz} \text { to } \\ & 1.25 \mathrm{MHz}^{\text {Note } 3} \end{aligned}$ | $\begin{aligned} & 1,2,4,5,6,8, \\ & 10 \mathrm{MHz}^{\text {Note } 4} \end{aligned}$ | 1 to 10 MHz | 1.0 | SI20/RxD20/P22 SO20/TxD20/P21 <br> SCK20/ASCK20/ P20 | 0 |
|  | SIO ch- 1 <br> (3-wired, sync. $)^{\text {Note } 8}$ |  |  |  |  | $\begin{array}{\|l} \mathrm{P} 02 \\ \mathrm{P} 01 \\ \mathrm{P} 00 \end{array}$ | $1^{\text {Note } 8}$ |
| SMB ${ }^{\text {Note }} 5$ | I2C ch-0 | 10 to 100 kHz | $\begin{aligned} & 1,2,4,5,6,8, \\ & 10 \mathrm{MHz}^{\text {Note } 4} \end{aligned}$ | 1 to 10 MHz | 1.0 | SCLO/P23 <br> SDAO/P24 | 4 |
| UART | UART ch-0 (Async.) | $\begin{array}{\|l\|} \hline 4,800 \text { to } \\ 76,800 \text { bps }^{\text {Noteses.6.6 }} \end{array}$ | 5, $10 \mathrm{MHz}^{\text {Note } 7}$ | $\begin{aligned} & 4.91,5, \\ & 10 \mathrm{MHz} \end{aligned}$ | 1.0 | RxD20/SI20/P22 TxD20/SO20/P21 | 8 |
| Pseudo $\text { 3-wire }{ }^{\text {Note } 9}$ | Port A <br> (Pseudo <br> 3-wire) | $\begin{aligned} & 100 \mathrm{~Hz} \text { to } \\ & 1 \mathrm{kHz} \end{aligned}$ | 1, 2, 4, 5 MHz | 1 to 5 MHz | 1.0 | $\begin{array}{\|l} \mathrm{P} 02 \\ \mathrm{P} 01 \\ \mathrm{P} 00 \\ \hline \end{array}$ | $12^{\text {Note } 9}$ |

Notes 1. Selection items for TYPE settings on the dedicated flash programmer (Flashpro III or Flashpro IV).
2. Setting a frequency 5 MHz or higher for the $\mu \mathrm{PD} 78 \mathrm{~F} 9177$ and $\mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{Y}$ is prohibited.
3. The possible setting range differs depending on the voltage. For details, refer to chapters related to the ELECTRICAL SPECIFICATIONS chapters.
4. Only $2,4,8 \mathrm{MHz}$ in Flashpro III.
5. For the $\mu$ PD78F9177Y and $\mu$ PD78F9177AY only. Set SLAVE ADDRESS to 10 H .
6. Because signal wave slew also affects UART communication, in addition to the baud rate error, thoroughly evaluate the slew and baud rate error.
7. Available only in Flashpro IV. When using Flashpro III, be sure to select the clock of the resonator on the board. UART cannot be used with the clock supplied by Flashpro III.
8. In the $\mu$ PD78F9177A and $\mu$ PD78F9177AY only.
9. In the $\mu$ PD78F9177 and $\mu$ PD78F9177Y only. Serial transfer is performed by controlling the ports by software.

Figure 20-2. Communication Mode Selection Format


RESET
$V_{D D}$
$V_{S S}$


Figure 20-3. Example of Connection with Dedicated Flash Programmer (1/2)
(a) 3-wire serial I/O (SIO-ch0)

(b) 3-wire serial I/O (SIO-ch1) or pseudo 3-wire mode

(c) SMB


Note Connect this pin when the system clock is supplied from the dedicated flash programmer. If a resonator is already connected to the X1 pin, the CLK pin does not need to be connected.

Caution The VDD pin, if already connected to the power supply, must be connected to the VDD pin of the dedicated flash programmer. Before using the power supply connected to the VdD pin, supply voltage before starting programming.

Figure 20-3. Example of Connection with Dedicated Flash Programmer (2/2)
(d) UART


Notes 1. Connect this pin when the system clock is supplied from the dedicated flash programmer. If a resonator is already connected to the X1 pin, the CLK pin does not need to be connected.
2. When using UART with Flashpro III, the clock of the resonator connected to the X1 pin must be used, so connection to the CLK pin is not necessary.

Caution The Vod pin, if already connected to the power supply, must be connected to the VDD pin of the dedicated flash programmer. Before using the power supply connected to the Vod pin, supply voltage before starting programming.

If Flashpro III or Flashpro IV is used as a dedicated flash programmer, the following signals are generated for the $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY. For details, refer to the manual of Flashpro III or Flashpro IV.

Table 20-3. Pin Connection List

| Signal <br> Name | I/O | Pin Function | Pin Name | 3-Wire <br> Serial I/O <br> (SIO-ch0) | $\left\|\begin{array}{c} \text { 3-Wire } \\ \text { Serial I/O } \\ \left(\text { SIO-ch11) }{ }^{\text {Note 1 }}\right. \end{array}\right\|$ | SMB ${ }^{\text {Note } 2}$ | UART | $\begin{gathered} \text { Pseudo } \\ \text { 3-Wire }{ }^{\text {Note } 3} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VPP1 | Output | Write voltage | Vpp | O | O | O | O | O |
| VPP2 | - | - | - | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |
| VDD | I/O | Vod voltage generation/ voltage monitoring | $\mathrm{V}_{\mathrm{DD}} / \mathrm{V}_{\text {DD } 1 / A V} \mathrm{~V}_{\text {DD }}$ | () ${ }^{\text {Note } 4}$ | (0) ${ }^{\text {Note } 4}$ | (o) Note 4 | () Note 4 | (0) Note 4 |
| GND | - | Ground | Vsso/Vss1/AVss | O | O | O | O | $\bigcirc$ |
| CLK | Output | Clock output | X1 | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ | $\bigcirc$ |
| RESET | Output | Reset signal | RESET | O | O | O | O | O |
| SI | Input | Receive signal | SO20/P01/ <br> SDA0/TxD20 | O | © | O | O | O |
| SO | Output | Transmit signal | SI20/P02/SCLO <br> /RxD20 | O | © | O | O | O |
| SCK | Output | Transfer clock | SCK20/P00 | O | O | $\times$ | $\times$ | O |
| HS | Input | Handshake signal | - | $\times$ | $\times$ | $\times$ | $\times$ | $\times$ |

Notes 1. In the $\mu$ PD78F9177A and $\mu$ PD78F9177AY only
2. In the $\mu$ PD78F9177Y and $\mu$ PD78F9177AY only
3. In the $\mu$ PD78F9177 and $\mu$ PD78F9177Y only
4. VDD voltage must be supplied before programming is started.

Remark ©: Pin must be connected.
O: If the signal is supplied on the target board, pin does not need to be connected.
$x$ : Pin does not need to be connected.

### 20.1.3 On-board pin processing

When performing programming on the target system, provide a connector on the target system to connect the dedicated flash programmer.

An on-board function that allows switching between normal operation mode and flash memory programming mode may be required in some cases.
<VPP pin>
In normal operation mode, input 0 V to the VPP pin. In flash memory programming mode, a write voltage of 10.0 V (TYP.) is supplied to the VPP pin, so perform one of the following (1) or (2).
(1) Connect a pull-down resistor $(\mathrm{RVPP}=10 \mathrm{k} \Omega)$ to the VPP pin.
(2) Use the jumper on the board to switch the VPP pin input to either the writer or directly to GND.

A VPP pin connection example is shown below.

Figure 20-4. VPP Pin Connection Example

<Serial interface pin>
The following shows the pins used by the serial interface.

| Serial Interface |  | Pins Used |
| :--- | :--- | :--- |
| 3-wire serial I/O | SIO-ch0 | SI20, SO20, $\overline{\text { SCK20 }}$ |
|  | SIO-ch1 |  |
| SMBte 1 $^{\text {Note 2 }}$ | P00, P01, P02 |  |
| UART | SCL0, SDA0 |  |
| Pseudo 3-wire ${ }^{\text {Note 3 }}$ | RxD20, TxD20 |  |

Notes 1. In the $\mu$ PD78F9177A and $\mu$ PD78F9177AY only
2. In the $\mu$ PD78F9177Y and $\mu$ PD78F9177AY only
3. In the $\mu$ PD78F9177 and $\mu$ PD78F9177Y only

When connecting the dedicated flash programmer to a serial interface pin that is connected to another device onboard, signal conflict or abnormal operation of the other device may occur. Care must therefore be taken with such connections.

## (1) Signal conflict

If the dedicated flash programmer (output) is connected to a serial interface pin (input) that is connected to another device (output), a signal conflict occurs. To prevent this, isolate the connection with the other device or set the other device to the output high impedance status.

Figure 20-5. Signal Conflict (Input Pin of Serial Interface)


In the flash memory programming mode, the signal output by another device and the signal sent by the dedicated flash programmer conflict, therefore, isolate the signal of the other device.

## (2) Abnormal operation of other device

If the dedicated flash programmer (output or input) is connected to a serial interface pin (input or output) that is connected to another device (input), a signal is output to the device, and this may cause an abnormal operation. To prevent this abnormal operation, isolate the connection with the other device or set so that the input signals to the other device are ignored.

Figure 20-6. Abnormal Operation of Other Device
$\mu$ PD78F9177, 78F9177Y,
78F9177A, 78F9177AY

If the signal output by the $\mu$ PD78F9177, 78F9177Y, 78F9177A, and 78F9177AY affects another device in the flash memory programming mode, isolate the signals of the other device.
\muPD78F9177, 78F9177Y,
\muPD78F9177, 78F9177Y,
78F9177A, 78F9177AY
78F9177A, 78F9177AY

If the signal output by the dedicated flash programmer affects another device in the flash memory programming mode, isolate the signals of the other device.
< $\overline{\text { RESET }}$ pin>
If the reset signal of the dedicated flash programmer is connected to the $\overline{\text { RESET }}$ pin connected to the reset signal generator on-board, a signal conflict occurs. To prevent this, isolate the connection with the reset signal generator.
If the reset signal is input from the user system in the flash memory programming mode, a normal programming operation cannot be performed. Therefore, do not input reset signals from other than the dedicated flash programmer.

Figure 20-7. Signal Conflict ( $\overline{\text { RESET }}$ Pin)


The signal output by the reset signal generator and the signal output from the dedicated flash programmer conflict in the flash memory programming mode, so isolate the signal of the reset signal generator.
<Port pins>
When the $\mu$ PD78F9177, $\mu$ PD78F9177Y, $\mu$ PD78F9177A, and $\mu$ PD78F9177AY enter the flash memory programming mode, all the pins other than those that communicate in flash memory programmer are in the same status as immediately after reset.
If the external device does not recognize initial statuses such as the output high impedance status, therefore, connect the external device to Vddo, VdD1, Vsso, or Vss1 via a resistor.

## <Oscillator>

When using the on-board clock, connect $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 as required in the normal operation mode.
When using the clock output of the flash programmer, connect it directly to X 1 , disconnecting the main oscillator on-board, and leave the X2 pin open. The subclock conforms to the method in the normal operation mode.
<Power supply>
To use the power output from the flash programmer, connect the Vodo and Vodi pins to VDD of the flash programmer, and the Vsso and Vss1 pins to GND of the flash programmer.
To use the on-board power supply, make connections that accord with the normal operation mode. However, because the voltage is monitored by the flash programmer, be sure to connect VDD of the flash programmer. Supply the same power as in the normal operation mode to the other power supply pins (AVdd, AVref, and AVss).

### 20.1.4 Connection of adapter for flash writing

The following figures show the examples of recommended connection when the adapter for flash writing is used.

Figure 20-8. Wiring Example for Flash Writing Adapter in 3-Wire Serial I/O Mode (SIO-ch0) (1/2)
(a) 44-pin plastic LQFP $(10 \times 10)$


Figure 20-8. Wiring Example for Flash Writing Adapter in 3-Wire Serial I/O Mode (SIO-ch0) (2/2)
(b) 48-pin plastic TQFP (fine pitch) $(7 \times 7)$


Figure 20-9. Wiring Example for Flash Writing Adapter in 3-Wire Serial I/O Mode (SIO-ch1) or Pseudo 3-Wire Mode (1/2)
(a) 44-pin plastic LQFP $(10 \times 10)$


Figure 20-9. Wiring Example for Flash Writing Adapter in 3-Wire Serial I/O Mode (SIO-ch1) or Pseudo 3-Wire Mode (2/2)
(b) 48-pin plastic TQFP (fine pitch) $(7 \times 7)$


Figure 20-10. Wiring Example for Flash Writing Adapter in SMB Mode (1/2)
(a) 44-pin plastic LQFP $(10 \times 10)$


Figure 20-10. Wiring Example for Flash Writing Adapter in SMB Mode (2/2)
(b) 48-pin plastic TQFP (fine pitch) $\mathbf{( 7 \times 7 )}$


Figure 20-11. Wiring Example for Flash Writing Adapter in UART Mode (1/2)
(a) 44-pin plastic LQFP $(10 \times 10)$


Figure 20-11. Wiring Example for Flash Writing Adapter in UART Mode (2/2)
(b) 48-pin plastic TQFP (fine pitch) $(7 \times 7)$


## CHAPTER 21 MASK OPTION

Table 21-1. Selection of Mask Option for Pins

| Pin |  |
| :--- | :--- |
| P50 to P53 | Whether a pull-up resistor is to be incorporated can be specified in 1-bit units. |

For P50 to P53 (port 5), whether a pull-up resistor is to be incorporated can be specified by a mask option. The mask option is specified in 1-bit units.

Caution The flash memory versions do not provide a mask option pull-up resistor incorporation function.

## CHAPTER 22 INSTRUCTION SET

This chapter lists the instruction set of the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries. For details of the operation and machine language (instruction code) of each instruction, refer to 78K/0S Series Instruction User's Manual (U11047E).

### 22.1 Operation

### 22.1.1 Operand identifiers and description methods

Operands are described in "Operand" column of each instruction in accordance with the description method of the instruction operand identifier (refer to the assembler specifications for details). When there are two or more description methods, select one of them. Uppercase letters the and symbols, \#, !, \$, and [ ] are keywords and must be described as they are. Each symbol has the following meaning.

- \#: Immediate data specification
- !: Absolute address specification
- \$: Relative address specification
- [ ]: Indirect address specification

In the case of immediate data, describe an appropriate numeric value or a label. When using a label, be sure to describe the \#, !, \$ and [ ] symbols.

For operand register identifiers, $r$ and $r p$, either function names ( $X, A, C$, etc.) or absolute names (names in parentheses in the table below, R0, R1, R2, etc.) can be used for description.

Table 22-1. Operand Identifiers and Description Methods

| Identifier | $\quad$ Description Method |
| :--- | :--- |
| $r$ | X (R0), A (R1), C (R2), B (R3), E (R4), D (R5), L (R6), H (R7) |
| rp |  |
| sfr |  |\(\left.\quad \begin{array}{l}AX (RP0), BC (RP1), DE (RP2), HL (RP3) <br>


Special-function register symbol\end{array}\right]\)| saddr |  |
| :--- | :--- |
| saddrp | FE20H to FF1FH Immediate data or labels |
| FE20H to FF1FH Immediate data or labels (even addresses only) <br> addr5 | 0000H to FFFFH Immediate data or labels (only even addresses for 16-bit data transfer instructions) <br> 0040H to 007FH Immediate data or labels (even addresses only) |
| word <br> byte <br> bit | 16-bit immediate data or label <br> 8-bit immediate data or label <br> 3-bit immediate data or label |

Remark See Table 5-3 Special-Function Registers for symbols of special-function registers.

### 22.1.2 Description of "Operation" column

A: A register; 8-bit accumulator
X: $\quad \mathrm{X}$ register
B: B register
C: $\quad$ C register
D: D register
E: E register
H: H register
$\mathrm{L}: \quad \mathrm{L}$ register
AX: $\quad$ AX register pair; 16-bit accumulator
$B C$ : $\quad B C$ register pair
DE: DE register pair
HL: HL register pair
PC: Program counter
SP: Stack pointer
PSW: Program status word
CY: $\quad$ Carry flag
AC: Auxiliary carry flag
Z: Zero flag
IE: Interrupt request enable flag
( ): Memory contents indicated by address or register contents in parenthesis
×H, $\times$ L: $\quad$ Higher 8 bits and lower 8 bits of 16-bit register
$\wedge$ : Logical product (AND)
v: Logical sum (OR)
$\forall$ : Exclusive logical sum (exclusive OR)
—: Inverted data
addr16: 16-bit immediate data or label
jdisp8: $\quad$ Signed 8-bit data (displacement value)

### 22.1.3 Description of "Flag" column

(Blank): Unchanged
0: $\quad$ Cleared to 0
1: $\quad$ Set to 1
$\times$ : Set/cleared according to the result
$R$ : $\quad$ Previously saved value is stored

### 22.2 Operation List

| Mnemonic | Operands | Bytes | Clocks | Operation | Flag |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Z AC CY |
| MOV | r, \#byte | 3 | 6 | $r \leftarrow$ byte |  |
|  | saddr, \#byte | 3 | 6 | (saddr) $\leftarrow$ byte |  |
|  | sfr, \#byte | 3 | 6 | $\mathrm{sfr} \leftarrow$ byte |  |
|  | A, $\mathrm{r}^{\text {Note } 1}$ | 2 | 4 | $A \leftarrow r$ |  |
|  | $\mathrm{r}, \mathrm{A}^{\text {Note } 1}$ | 2 | 4 | $r \leftarrow A$ |  |
|  | A, saddr | 2 | 4 | $\mathrm{A} \leftarrow$ ( saddr) |  |
|  | saddr, A | 2 | 4 | (saddr) $\leftarrow \mathrm{A}$ |  |
|  | A, sfr | 2 | 4 | $\mathrm{A} \leftarrow \mathrm{sfr}$ |  |
|  | sfr, A | 2 | 4 | $\mathrm{sfr} \leftarrow \mathrm{A}$ |  |
|  | A, !addr16 | 3 | 8 | $\mathrm{A} \leftarrow$ (addr16) |  |
|  | !addr16, A | 3 | 8 | (addr16) $\leftarrow$ A |  |
|  | PSW, \#byte | 3 | 6 | PSW $\leftarrow$ byte | $\times \times$ |
|  | A, PSW | 2 | 4 | $\mathrm{A} \leftarrow \mathrm{PSW}$ |  |
|  | PSW, A | 2 | 4 | $\mathrm{PSW} \leftarrow \mathrm{A}$ | $\times \times$ |
|  | A, [DE] | 1 | 6 | $\mathrm{A} \leftarrow(\mathrm{DE})$ |  |
|  | [DE], A | 1 | 6 | $(\mathrm{DE}) \leftarrow \mathrm{A}$ |  |
|  | A, [HL] | 1 | 6 | $\mathrm{A} \leftarrow(\mathrm{HL})$ |  |
|  | [HL], A | 1 | 6 | $(\mathrm{HL}) \leftarrow \mathrm{A}$ |  |
|  | A, [HL + byte] | 2 | 6 | $\mathrm{A} \leftarrow(\mathrm{HL}+$ byte $)$ |  |
|  | [HL + byte], A | 2 | 6 | $(\mathrm{HL}+$ byte) $\leftarrow \mathrm{A}$ |  |
| XCH | A, X | 1 | 4 | $\mathrm{A} \leftrightarrow \mathrm{X}$ |  |
|  | A, $\mathrm{r}^{\text {Note } 2}$ | 2 | 6 | $A \leftrightarrow r$ |  |
|  | A, saddr | 2 | 6 | A $\leftrightarrow$ (saddr) |  |
|  | A, sfr | 2 | 6 | $\mathrm{A} \leftrightarrow \mathrm{sfr}$ |  |
|  | A, [DE] | 1 | 8 | A $\leftrightarrow$ (DE) |  |
|  | A, [HL] | 1 | 8 | A $\leftrightarrow(\mathrm{HL})$ |  |
|  | A, [HL + byte] | 2 | 8 | A $\leftrightarrow(\mathrm{HL}+$ byte $)$ |  |

Notes 1. Except $r=A$.
2. Except $r=A, X$.

Remark One instruction clock cycle is one CPU clock cycle (fCPU) selected by the processor clock control register (PCC).

| Mnemonic | Operands | Bytes | Clocks | Operation | Flag |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Z | AC CY |
| MOVw | rp, \#word | 3 | 6 | $\mathrm{rp} \leftarrow$ word |  |  |
|  | AX, saddrp | 2 | 6 | $\mathrm{AX} \leftarrow$ (saddrp) |  |  |
|  | saddrp, AX | 2 | 8 | (saddrp) $\leftarrow$ AX |  |  |
|  | AX, rp ${ }^{\text {Note }}$ | 1 | 4 | $A X \leftarrow r p$ |  |  |
|  | $\mathrm{rp}, \mathrm{AX}{ }^{\text {Note }}$ | 1 | 4 | $\mathrm{rp} \leftarrow \mathrm{AX}$ |  |  |
| XCHW | AX, rp ${ }^{\text {Note }}$ | 1 | 8 | $A X \leftrightarrow r p$ |  |  |
| ADD | A, \#byte | 2 | 4 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+$ byte | $\times$ | $\times \times$ |
|  | saddr, \#byte | 3 | 6 | (saddr), $\mathrm{CY} \leftarrow$ (saddr) + byte | $\times$ | $\times \times$ |
|  | A, r | 2 | 4 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+\mathrm{r}$ | $\times$ | $\times \times$ |
|  | A, saddr | 2 | 4 | A, CY $\leftarrow \mathrm{A}+$ (saddr) | $\times$ | $\times \times$ |
|  | A, !addr16 | 3 | 8 | A, CY $\leftarrow \mathrm{A}+($ addr16) | $\times$ | $\times \times$ |
|  | A, [HL] | 1 | 6 | $A, C Y \leftarrow A+(H L)$ | $\times$ | $\times \times$ |
|  | A, [HL + byte] | 2 | 6 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+(\mathrm{HL}+$ byte $)$ | $\times$ | $\times \times$ |
| ADDC | A, \#byte | 2 | 4 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+$ byte + CY | $\times$ | $\times \times$ |
|  | saddr, \#byte | 3 | 6 | (saddr), $\mathrm{CY} \leftarrow$ (saddr) + byte + CY | $\times$ | $\times \times$ |
|  | A, r | 2 | 4 | $A, C Y \leftarrow A+r+C Y$ | $\times$ | $\times \times$ |
|  | A, saddr | 2 | 4 | A, CY $\leftarrow \mathrm{A}+$ ( saddr) +CY | $\times$ | $\times \times$ |
|  | A, !addr16 | 3 | 8 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+($ addr16 $)+\mathrm{CY}$ | $\times$ | $\times \times$ |
|  | A, [HL] | 1 | 6 | A, $\mathrm{CY} \leftarrow \mathrm{A}+(\mathrm{HL})+\mathrm{CY}$ | $\times$ | $\times \times$ |
|  | A, [HL + byte] | 2 | 6 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}+(\mathrm{HL}+$ byte $)+\mathrm{CY}$ | $\times$ | $\times \times$ |
| SUB | A, \#byte | 2 | 4 | A, CY $\leftarrow \mathrm{A}$ - byte | $\times$ | $\times \times$ |
|  | saddr, \#byte | 3 | 6 | (saddr), $\mathrm{CY} \leftarrow$ (saddr) - byte | $\times$ | $\times \times$ |
|  | A, r | 2 | 4 | A, CY $\leftarrow \mathrm{A}-\mathrm{r}$ | $\times$ | $\times \times$ |
|  | A, saddr | 2 | 4 | A, CY $\leftarrow \mathrm{A}$ - (saddr) | $\times$ | $\times \times$ |
|  | A, !addr16 | 3 | 8 | $\mathrm{A}, \mathrm{CY} \leftarrow \mathrm{A}-$ (addr16) | $\times$ | $\times \times$ |
|  | A, [HL] | 1 | 6 | $A, C Y \leftarrow A-(H L)$ | $\times$ | $\times \times$ |
|  | A, [HL + byte] | 2 | 6 | A, CY $\leftarrow \mathrm{A}-(\mathrm{HL}+$ byte $)$ | $\times$ | $\times \times$ |

Note Only when $\mathrm{rp}=\mathrm{BC}, \mathrm{DE}$, or HL .

Remark One instruction clock cycle is one CPU clock cycle (fcpu) selected by the processor clock control register (PCC).

| Mnemonic | Operands | Bytes | Clocks | Operation | Flag |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Z | AC CY |
| SUBC | A, \#byte | 2 | 4 | A, CY $\leftarrow$ A - byte - CY | $\times$ | $\times \times$ |
|  | saddr, \#byte | 3 | 6 | (saddr), $\mathrm{CY} \leftarrow$ (saddr) - byte - CY | $\times$ | $\times \times$ |
|  | A, r | 2 | 4 | A, $\mathrm{CY} \leftarrow \mathrm{A}-\mathrm{r}-\mathrm{CY}$ | $\times$ | $\times \times$ |
|  | A, saddr | 2 | 4 | A, CY $\leftarrow \mathrm{A}-$ (saddr) - CY | $\times$ | $\times \times$ |
|  | A, laddr16 | 3 | 8 | A, CY $\leftarrow \mathrm{A}-($ addr16 $)-\mathrm{CY}$ | $\times$ | $\times \times$ |
|  | A, [HL] | 1 | 6 | A, CY $\leftarrow \mathrm{A}-(\mathrm{HL})-\mathrm{CY}$ | $\times$ | $\times \times$ |
|  | A, [HL + byte] | 2 | 6 | $A, C Y \leftarrow A-(H L+$ byte $)-C Y$ | $\times$ | $\times \times$ |
| AND | A, \#byte | 2 | 4 | $\mathrm{A} \leftarrow \mathrm{A} \wedge$ byte | $\times$ |  |
|  | saddr, \#byte | 3 | 6 | (saddr) $\leftarrow$ (saddr) ^ byte | $\times$ |  |
|  | A, r | 2 | 4 | $A \leftarrow A \wedge r$ | $\times$ |  |
|  | A, saddr | 2 | 4 | $A \leftarrow A \wedge$ (saddr) | $\times$ |  |
|  | A, !addr16 | 3 | 8 | $\mathrm{A} \leftarrow \mathrm{A} \wedge($ addr16) | $\times$ |  |
|  | A, [HL] | 1 | 6 | $A \leftarrow A \wedge(H L)$ | $\times$ |  |
|  | A, [HL + byte] | 2 | 6 | $A \leftarrow A \wedge(H L+$ byte $)$ | $\times$ |  |
| OR | A, \#byte | 2 | 4 | $A \leftarrow A \vee$ byte | $\times$ |  |
|  | saddr, \#byte | 3 | 6 | (saddr) $\leftarrow$ (saddr) $\vee$ byte | $\times$ |  |
|  | A, r | 2 | 4 | $A \leftarrow A \vee r$ | $\times$ |  |
|  | A, saddr | 2 | 4 | $A \leftarrow A \vee$ (saddr) | $\times$ |  |
|  | A, !addr16 | 3 | 8 | $A \leftarrow A \vee($ addr16 $)$ | $\times$ |  |
|  | A, [HL] | 1 | 6 | $A \leftarrow A \vee(H L)$ | $\times$ |  |
|  | A, [HL + byte] | 2 | 6 | $A \leftarrow A \vee(H L+$ byte $)$ | $\times$ |  |
| XOR | A, \#byte | 2 | 4 | $A \leftarrow A \forall$ byte | $\times$ |  |
|  | saddr, \#byte | 3 | 6 | $($ saddr $) \leftarrow$ (saddr) $\vee$ byte | $\times$ |  |
|  | A, r | 2 | 4 | $A \leftarrow A \nabla r$ | $\times$ |  |
|  | A, saddr | 2 | 4 | $A \leftarrow A \nabla$ (saddr) | $\times$ |  |
|  | A, !addr16 | 3 | 8 | $A \leftarrow A \nabla$ (addr16) | $\times$ |  |
|  | A, [HL] | 1 | 6 | $\mathrm{A} \leftarrow \mathrm{A} \nabla(\mathrm{HL})$ | $\times$ |  |
|  | A, [HL + byte] | 2 | 6 | $A \leftarrow A \nabla(H L+$ byte $)$ | $\times$ |  |

Remark One instruction clock cycle is one CPU clock cycle (fcpu) selected by the processor clock control register (PCC).

| Mnemonic | Operands | Bytes | Clocks | Operation | Flag |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  | Z | AC CY |
| CMP | A, \#byte | 2 | 4 | A - byte | $\times$ | $\times \times$ |
|  | saddr, \#byte | 3 | 6 | (saddr) - byte | $\times$ | $\times \times$ |
|  | A, r | 2 | 4 | A - r | $\times$ | $\times \times$ |
|  | A, saddr | 2 | 4 | A - (saddr) | $\times$ | $\times \times$ |
|  | A, !addr16 | 3 | 8 | A - (addr16) | $\times$ | $\times \times$ |
|  | A, [HL] | 1 | 6 | A - (HL) | $\times$ | $\times \times$ |
|  | A, [HL + byte] | 2 | 6 | A - (HL + byte) | $\times$ | $\times \times$ |
| ADDW | AX, \#word | 3 | 6 | $\mathrm{AX}, \mathrm{CY} \leftarrow \mathrm{AX}+$ word | $\times$ | $\times \times$ |
| SUBW | AX, \#word | 3 | 6 | $A X, C Y \leftarrow A X$ - word | $\times$ | $\times \times$ |
| CMPW | AX, \#word | 3 | 6 | AX - word | $\times$ | $\times \times$ |
| INC | r | 2 | 4 | $r \leftarrow r+1$ | $\times$ | $\times$ |
|  | saddr | 2 | 4 | (saddr) $\leftarrow$ (saddr) +1 | $\times$ | $\times$ |
| DEC | r | 2 | 4 | $r \leftarrow r-1$ | $\times$ | $\times$ |
|  | saddr | 2 | 4 | (saddr) $\leftarrow$ (saddr) - 1 | $\times$ | $\times$ |
| INCW | rp | 1 | 4 | $\mathrm{rp} \leftarrow \mathrm{rp}+1$ |  |  |
| DECW | rp | 1 | 4 | $\mathrm{rp} \leftarrow \mathrm{rp}-1$ |  |  |
| ROR | A, 1 | 1 | 2 | $\left(C Y, A_{7} \leftarrow A_{0}, A_{m-1} \leftarrow A_{m}\right) \times 1$ |  | $\times$ |
| ROL | A, 1 | 1 | 2 | $\left(C Y, A_{0} \leftarrow A_{7}, A_{m+1} \leftarrow A_{m}\right) \times 1$ |  | $\times$ |
| RORC | A, 1 | 1 | 2 | $\left(C Y \leftarrow A_{0}, A_{7} \leftarrow C Y, A_{m-1} \leftarrow A_{m}\right) \times 1$ |  | $\times$ |
| ROLC | A, 1 | 1 | 2 | $\left(C Y \leftarrow A_{7}, A_{0} \leftarrow C Y, A_{m+1} \leftarrow A_{m}\right) \times 1$ |  | $\times$ |
| SET1 | saddr.bit | 3 | 6 | (saddr.bit) $\leftarrow 1$ |  |  |
|  | sfr.bit | 3 | 6 | sfr.bit $\leftarrow 1$ |  |  |
|  | A.bit | 2 | 4 | A.bit $\leftarrow 1$ |  |  |
|  | PSW.bit | 3 | 6 | PSW.bit $\leftarrow 1$ | $\times$ | $\times \times$ |
|  | [HL].bit | 2 | 10 | (HL). $\mathrm{bit} \leftarrow 1$ |  |  |
| CLR1 | saddr.bit | 3 | 6 | (saddr.bit) $\leftarrow 0$ |  |  |
|  | sfr.bit | 3 | 6 | sfr.bit $\leftarrow 0$ |  |  |
|  | A.bit | 2 | 4 | A.bit $\leftarrow 0$ |  |  |
|  | PSW.bit | 3 | 6 | PSW.bit $\leftarrow 0$ | $\times$ | $\times \times$ |
|  | [HL].bit | 2 | 10 | (HL).bit $\leftarrow 0$ |  |  |
| SET1 | CY | 1 | 2 | $\mathrm{CY} \leftarrow 1$ |  | 1 |
| CLR1 | CY | 1 | 2 | $\mathrm{CY} \leftarrow 0$ |  | 0 |
| NOT1 | CY | 1 | 2 | $\mathrm{CY} \leftarrow \overline{\mathrm{CY}}$ |  | $\times$ |

Remark One instruction clock cycle is one CPU clock cycle (fcru) selected by the processor clock control register (PCC).

| Mnemonic | Operands | Bytes | Clocks | Operation | Flag |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  | AC CY |
| CALL | !addr16 | 3 | 6 | $\begin{aligned} & (\mathrm{SP}-1) \leftarrow(\mathrm{PC}+3) \mathrm{H},(\mathrm{SP}-2) \leftarrow(\mathrm{PC}+3)\llcorner, \\ & \mathrm{PC} \leftarrow \text { addr16, } \mathrm{SP} \leftarrow \mathrm{SP}-2 \end{aligned}$ |  |  |
| CALLT | [addr5] | 1 | 8 | $\begin{aligned} & (\mathrm{SP}-1) \leftarrow(\mathrm{PC}+1) \mathrm{H},(\mathrm{SP}-2) \leftarrow(\mathrm{PC}+1) \mathrm{L}, \\ & \mathrm{PCH} \leftarrow(00000000, \text { addr5 + 1), } \\ & \mathrm{PCL} \leftarrow(00000000, \text { addr5 }), \mathrm{SP} \leftarrow \mathrm{SP}-2 \end{aligned}$ |  |  |
| RET |  | 1 | 6 | $\mathrm{PCH} \leftarrow(\mathrm{SP}+1), \mathrm{PCL} \leftarrow(\mathrm{SP}), \mathrm{SP} \leftarrow \mathrm{SP}+2$ |  |  |
| RETI |  | 1 | 8 | $\begin{aligned} & \mathrm{PC} \mathrm{H} \leftarrow(\mathrm{SP}+1), \mathrm{PCL} \leftarrow(\mathrm{SP}), \\ & \mathrm{PSW} \leftarrow(\mathrm{SP}+2), \mathrm{SP} \leftarrow \mathrm{SP}+3 \end{aligned}$ | R | R R |
| PUSH | PSW | 1 | 2 | $(\mathrm{SP}-1) \leftarrow \mathrm{PSW}, \mathrm{SP} \leftarrow \mathrm{SP}-1$ |  |  |
|  | rp | 1 | 4 | $(S P-1) \leftarrow \mathrm{rpH},(\mathrm{SP}-2) \leftarrow \mathrm{rpL}, \mathrm{SP} \leftarrow \mathrm{SP}-2$ |  |  |
| POP | PSW | 1 | 4 | $\mathrm{PSW} \leftarrow(\mathrm{SP}), \mathrm{SP} \leftarrow \mathrm{SP}+1$ | R | R R |
|  | rp | 1 | 6 | $\mathrm{rpH} \leftarrow(\mathrm{SP}+1), \mathrm{rpL} \leftarrow(\mathrm{SP}), \mathrm{SP} \leftarrow \mathrm{SP}+2$ |  |  |
| MOVW | SP, AX | 2 | 8 | $\mathrm{SP} \leftarrow \mathrm{AX}$ |  |  |
|  | AX, SP | 2 | 6 | $A X \leftarrow S P$ |  |  |
| BR | !addr16 | 3 | 6 | $\mathrm{PC} \leftarrow$ addr16 |  |  |
|  | \$addr16 | 2 | 6 | $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 |  |  |
|  | AX | 1 | 6 | $\mathrm{PC}+\leftarrow \mathrm{A}, \mathrm{PCL} \leftarrow \mathrm{X}$ |  |  |
| BC | \$saddr16 | 2 | 6 | $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 if $\mathrm{CY}=1$ |  |  |
| BNC | \$saddr16 | 2 | 6 | $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 if $\mathrm{CY}=0$ |  |  |
| BZ | \$saddr16 | 2 | 6 | $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 if $\mathrm{Z}=1$ |  |  |
| BNZ | \$saddr16 | 2 | 6 | $\mathrm{PC} \leftarrow \mathrm{PC}+2+\mathrm{jdisp} 8$ if $\mathrm{Z}=0$ |  |  |
| BT | saddr.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+\mathrm{jdisp8}$ if (saddr.bit) $=1$ |  |  |
|  | sfr.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+$ jdisp8 if sfr.bit $=1$ |  |  |
|  | A.bit, \$addr16 | 3 | 8 | $\mathrm{PC} \leftarrow \mathrm{PC}+3+$ jdisp8 if $\mathrm{A} . \mathrm{bit}=1$ |  |  |
|  | PSW.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+$ jdisp8 if PSW.bit $=1$ |  |  |
| BF | saddr.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+$ jdisp8 if (saddr.bit) $=0$ |  |  |
|  | sfr.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+$ jdisp8 if sfr.bit $=0$ |  |  |
|  | A.bit, \$addr16 | 3 | 8 | $\mathrm{PC} \leftarrow \mathrm{PC}+3+$ jdisp8 if A.bit $=0$ |  |  |
|  | PSW.bit, \$addr16 | 4 | 10 | $\mathrm{PC} \leftarrow \mathrm{PC}+4+$ jdisp8 if PSW.bit $=0$ |  |  |
| DBNZ | B, \$addr16 | 2 | 6 | $\mathrm{B} \leftarrow \mathrm{B}-1$, then $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 if $\mathrm{B} \neq 0$ |  |  |
|  | C, \$addr16 | 2 | 6 | $\mathrm{C} \leftarrow \mathrm{C}-1$, then $\mathrm{PC} \leftarrow \mathrm{PC}+2+$ jdisp8 if $\mathrm{C} \neq 0$ |  |  |
|  | saddr, \$addr16 | 3 | 8 | $\begin{aligned} & (\text { saddr }) \leftarrow(\text { saddr })-1, \text { then } \\ & \mathrm{PC} \leftarrow \mathrm{PC}+3+\text { jdisp8 if }(\text { saddr }) \neq 0 \end{aligned}$ |  |  |
| NOP |  | 1 | 2 | No Operation |  |  |
| El |  | 3 | 6 | $\mathrm{IE} \leftarrow 1$ (Enable Interrupt) |  |  |
| DI |  | 3 | 6 | $\mathrm{IE} \leftarrow 0$ (Disable Interrupt) |  |  |
| HALT |  | 1 | 2 | Set HALT Mode |  |  |
| STOP |  | 1 | 2 | Set STOP Mode |  |  |

Remark One instruction clock cycle is one CPU clock cycle (fcPu) selected by the processor clock control register (PCC).

### 22.3 Instructions Listed by Addressing Type

(1) 8-bit instructions

MOV, XCH, ADD, ADDC, SUB, SUBC, AND, OR, XOR, CMP, INC, DEC, ROR, ROL, RORC, ROLC, PUSH, POP, DBNZ

| 2nd Operand <br> 1st Operand | \#byte | A | $r$ | sfr | saddr | !addr16 | PSW | [DE] | [HL] | [HL + byte] | \$addr16 | 1 | None |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A | ADD <br> ADDC <br> SUB <br> SUBC <br> AND <br> OR <br> XOR <br> CMP |  | $\begin{aligned} & \text { MOV }^{\text {Note }} \\ & \mathrm{XCH} \\ & \text { Note } \\ & \text { ADD } \\ & \text { ADDC } \\ & \text { SUB } \\ & \text { SUBC } \\ & \text { AND } \\ & \text { OR } \\ & \text { XOR } \\ & \text { CMP } \end{aligned}$ | $\begin{aligned} & \mathrm{MOV} \\ & \mathrm{XCH} \end{aligned}$ | MOV <br> XCH <br> ADD <br> ADDC <br> SUB <br> SUBC <br> AND <br> OR <br> XOR <br> CMP | MOV <br> ADD <br> ADDC <br> SUB <br> SUBC <br> AND <br> OR <br> XOR <br> CMP | MOV | $\begin{aligned} & \mathrm{MOV} \\ & \mathrm{XCH} \end{aligned}$ | MOV <br> XCH <br> ADD <br> ADDC <br> SUB <br> SUBC <br> AND <br> OR <br> XOR <br> CMP | $\begin{aligned} & \text { MOV } \\ & \text { XCH } \\ & \text { ADD } \\ & \text { ADDC } \\ & \text { SUB } \\ & \text { SUBC } \\ & \text { AND } \\ & \text { OR } \\ & \text { XOR } \\ & \text { CMP } \end{aligned}$ |  | ROR <br> ROL <br> RORC <br> ROLC |  |
| r | MOV | MOV |  |  |  |  |  |  |  |  |  |  | $\begin{aligned} & \text { INC } \\ & \text { DEC } \end{aligned}$ |
| B, C |  |  |  |  |  |  |  |  |  |  | DBNZ |  |  |
| sfr | MOV | MOV |  |  |  |  |  |  |  |  |  |  |  |
| saddr | MOV <br> ADD <br> ADDC <br> SUB <br> SUBC <br> AND <br> OR <br> XOR <br> CMP | MOV |  |  |  |  |  |  |  |  | DBNZ |  | $\begin{aligned} & \text { INC } \\ & \text { DEC } \end{aligned}$ |
| !addr16 |  | MOV |  |  |  |  |  |  |  |  |  |  |  |
| PSW | MOV | MOV |  |  |  |  |  |  |  |  |  |  | PUSH POP |
| [DE] |  | MOV |  |  |  |  |  |  |  |  |  |  |  |
| [HL] |  | MOV |  |  |  |  |  |  |  |  |  |  |  |
| [HL + byte] |  | MOV |  |  |  |  |  |  |  |  |  |  |  |

Note Except $r=A$.
(2) 16-bit instructions

MOVW, XCHW, ADDW, SUBW, CMPW, PUSH, POP, INCW, DECW

| 1st Operand | \#word | AX | rpote | saddrp | SP | None |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| AX | ADDW SUBW <br> CMPW |  | MOVW <br> XCHW | MOVW | MOVW |  |
| rp | MOVW | MOVW $^{\text {Note }}$ |  |  | INCW <br> DECW <br> PUSH <br> POP |  |
| saddrp |  |  |  |  |  |  |
| sp |  | MOVW |  |  |  |  |

Note Only when $r p=B C, D E$, or HL.
(3) Bit manipulation instructions

SET1, CLR1, NOT1, BT, BF

| 1st Operand | \$addr16 | None |
| :--- | :--- | :--- |
| A.bit | BT | SET1 <br> CLR1 <br> BF |
| sfr.bit | BT | SET1 |
|  | BF | CLR1 |
| saddr.bit | BT | SET1 |
|  | BF | CLR1 |
| PSW.bit | BT | SET1 |
|  | BF | CLR1 |
| [HL].bit |  | SET1 |
|  |  | CLR1 |
| CY |  | SET1 |
|  |  | CLR1 |
|  |  | NOT1 |

(4) Call instructions/branch instructions

CALL, CALLT, BR, BC, BNC, BZ, BNZ, DBNZ

|  | AX | !addr16 | [addr5] | \$addr16 |
| :--- | :--- | :--- | :--- | :--- |
| 1st Operand |  |  |  |  |
| Basic instructions | BR | CALL | CALLT | BR |
|  |  | BR |  | BC |
|  |  |  |  | BNC |
|  |  |  |  | BZ |
|  |  |  |  | BNZ |
| Compound instructions |  |  |  |  |

(5) Other instructions

RET, RETI, NOP, EI, DI, HALT, STOP

## CHAPTER 23 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78916x, 17x, 16xY, 17xY, 16x(A), 17x(A), 16xY(A), 17xY(A))

Remark The values listed in this chapter are for expanded-specification products.

Absolute Maximum Ratings ( $\mathrm{T}_{\mathrm{A}}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions |  | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | VdD | $\begin{aligned} & \mathrm{A} \mathrm{~V}_{\mathrm{DD}}-0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\mathrm{REF}} \leq \mathrm{AV} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\text {REF }} \leq \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \end{aligned}$ |  | $-0.3 \text { to }+6.5$ | V |
|  | AVDD |  |  | V |
|  | AVref |  |  | V |
| Input voltage | $V_{11}$ | Pins other than P50 to P53, P23, P24 |  |  | -0.3 to $V_{\text {DD }}+0.3$ | V |
|  | $\mathrm{V}_{12}$ | P23, P24 |  |  | -0.3 to +5.5 | V |
|  | $\mathrm{V}_{13}$ | P50 to P53 | N -ch open drain | -0.3 to +13 | V |
|  |  |  | On-chip pull-up resistor | -0.3 to $V_{D D}+0.3$ | V |
| Output voltage | Vo |  |  | -0.3 to $V_{\text {dD }}+0.3$ | V |
| Output current, high | IOH | Per pin | $\begin{aligned} & \mu \text { PD78916x, 78917x, } \\ & 78916 \mathrm{Y}, 78917 \mathrm{xY} \end{aligned}$ | -10 | mA |
|  |  | Total for all pins |  | -30 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A}), \\ & 78917 \mathrm{x}(\mathrm{~A}), 78916 \mathrm{xY}(\mathrm{~A}), \\ & 78917 \mathrm{xY}(\mathrm{~A}) \end{aligned}$ | -7 | mA |
|  |  | Total for all pins |  | -22 | mA |
| Output current, low | lot | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78916 \mathrm{x}, 78917 \mathrm{x} \\ & 78916 \mathrm{xY}, 78917 \mathrm{xY} \end{aligned}$ | 30 | mA |
|  |  | Total for all pins |  | 160 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A}), \\ & 78917 \mathrm{x}(\mathrm{~A}), 78916 \mathrm{xY}(\mathrm{~A}), \\ & 78917 \mathrm{YY}(\mathrm{~A}) \end{aligned}$ | 10 | mA |
|  |  | Total for all pins |  | 120 | mA |
| Operating ambient temperature | TA |  |  | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ |  |  | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |

Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## Main System Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VdD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Ceramic resonator |  | Oscillation frequency (fx) ${ }^{\text {Note } 1}$ | $\mathrm{V}_{D D}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | $V_{\text {DD }}=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $V_{\text {DD }}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | After Vod reaches oscillation voltage range MIN. |  |  | 4 | ms |
| Crystal resonator |  | Oscillation frequency (fx) ${ }^{\text {Note } 1}$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | V DD $=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $V_{D D}=4.5$ to 5.5 V |  |  | 10 | ms |
|  |  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 30 | ms |
| External clock |  | X 1 input frequency (fx) ${ }^{\text {Note } 1}$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | $\mathrm{V}_{\text {DD }}=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | X1 input high-/low-level width ( $\mathrm{txh}, \mathrm{txL}$ ) | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 45 |  | 500 | ns |
|  |  |  | $V_{\text {DD }}=3.0$ to 5.5 V | 75 |  | 500 | ns |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 85 |  | 500 | ns |
|  | $\begin{array}{\|ll\|} \hline \mathrm{X} 1 & \mathrm{X} 2 \\ \hline \end{array}$ | X1 input frequency (fx) ${ }^{\text {Note } 1}$ | $V_{\text {DD }}=2.7$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  | OPEN | X1 input high-/low-level width (txh, txL) | $V_{D D}=2.7$ to 5.5 V | 85 |  | 500 | ns |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the main system clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. When the main system clock is stopped and the device is operating on the subsystem clock, wait until the oscillation stabilization time has been secured by the program before switching back to the main system clock.

Recommended Oscillator Constant ( $\mu$ PD78916x, 17x, 16xY, and 17xY)

Ceramic resonator ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $+85^{\circ} \mathrm{C}$ )

| Manufacturer | Part Number | Frequency (MHz) | Recommended Circuit Constant (pF) |  | Oscillation Voltage <br> Range (Vdo) |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C1 | C2 | MIN. | MAX. |  |
| Murata Mfg. Co., Ltd. (Standard type) | CSBLA1M00J58-B0 | 1.000 | 150 | 150 | 2.2 | 5.5 | Without on-chip capacitor |
|  | CSBFB1M00J58-R1 |  |  |  |  |  |  |
|  | CSTCC2M00G56-R0 | 2.000 | - | - | 1.8 | 5.5 | With on-chip capacitor |
|  | CSTLS2M00G56-B0 |  |  |  |  |  |  |
|  | CSTCR4M00G53-R0 | 4.000 |  |  |  |  |  |
|  | CSTLS4M00G53-B0 |  |  |  |  |  |  |
|  | CSTCR4M19G53-R0 | 4.195 |  |  |  |  |  |
|  | CSTLS4M19G53-B0 |  |  |  |  |  |  |
|  | CSTCR4M91G53-R0 | 4.915 |  |  |  |  |  |
|  | CSTLS4M91G53-B0 |  |  |  |  |  |  |
|  | CSTCR5M00G53-R0 | 5.000 |  |  |  |  |  |
|  | CSTLS5M00G53-B0 |  |  |  |  |  |  |
|  | CSTCR6M00G53-R0 | 6.000 |  |  |  |  |  |
|  | CSTLS6M00G53-B0 |  |  |  |  |  |  |
|  | CSTCE8M00G52-R0 | 8.000 |  |  |  |  |  |
|  | CSTLS8M00G53-B0 |  |  |  |  |  |  |
|  | CSTCE8M38G52-R0 | 8.388 |  |  |  |  |  |
|  | CSTLS8M38G53-B0 |  |  |  |  |  |  |
|  | CSTCE10M0G52-R0 | 10.000 |  |  |  |  |  |
|  | CSTLS10M0G53-B0 |  |  |  | 1.9 | 5.5 |  |
| Murata Mfg. Co., Ltd. (Low-voltage drive type) | CSTLS10M0G53093- <br> B0 | 10.000 | - | - | 1.8 | 5.5 | With on-chip capacitor |

Caution The oscillator constant is a reference value based on evaluation in specific environments by the resonator manufacturer. If the oscillator characteristics need to be optimized in the actual application, request the resonator manufacturer for evaluation on the implementation circuit. Note that the oscillation voltage and oscillation frequency merely indicate the characteristics of the oscillator. Use the internal operation conditions of the $\mu$ PD78916x, 17x, 16xY, and 17xY within the specifications of the DC and AC characteristics.

Recommended Oscillator Constant ( $\mu$ PD78916x(A), 17x(A), 16xY(A), and 17xY(A))

Ceramic resonator ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $+85^{\circ} \mathrm{C}$ )

| Manufacturer | Part Number | Frequency <br> (MHz) | Recommended Circuit Constant (pF) |  | Oscillation Voltage Range (VdD) |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C1 | C2 | MIN. | MAX. |  |
| Murata Mfg. Co., Ltd. | CSTCC2M00G56A-R0 | 2.000 | - | - | 1.8 | 5.5 | On-chip capacitor |
|  | CSTCR4M00G53A-R0 | 4.000 |  |  |  |  |  |
|  | CSTCR4M19G53A-R0 | 4.195 |  |  |  |  |  |
|  | CSTCR4M91G53A-R0 | 4.915 |  |  |  |  |  |
|  | CSTCR5M00G53A-R0 | 5.000 |  |  |  |  |  |
|  | CSTCR6M00G53A-R0 | 6.000 |  |  |  |  |  |
|  | CSTCE8M00G52A-R0 | 8.000 |  |  |  |  |  |
|  | CSTCE8M38G52A-R0 | 8.388 |  |  |  |  |  |
|  | CSTCE10M0G52A-R0 | 10.000 |  |  |  |  |  |

Caution The oscillator constant is a reference value based on evaluation in specific environments by the resonator manufacturer. If the oscillator characteristics need to be optimized in the actual application, request the resonator manufacturer for evaluation on the implementation circuit. Note that the oscillation voltage and oscillation frequency merely indicate the characteristics of the oscillator. Use the internal operation conditions of the $\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A}), 17 \times(A), 16 \times Y(A)$, and $17 \times Y(A)$ within the specifications of the DC and AC characteristics.

Subsystem Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal resonator |  | Oscillation frequency (fxt) ${ }^{\text {Note } 1}$ |  | 32 | 32.768 | 35 | kHz |
|  |  |  |  |  |  |  |  |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $V_{\text {DD }}=4.5$ to 5.5 V |  | 1.2 | 2 | s |
|  |  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 10 | s |
| External clock | XT1 XT 2 | XT1 input frequency (fxT) ${ }^{\text {Note } 1}$ |  | 32 |  | 35 | kHz |
|  |  | XT1 input high-/low-level width (tхтн, tхть) |  | 14.3 |  | 15.6 | $\mu \mathrm{s}$ |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the subsystem clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. The subsystem clock oscillator is designed as a low-amplitude circuit for reducing current consumption, and is more prone to malfunction due to noise than the main system clock oscillator. Particular care is therefore required with the wiring method when the subsystem clock is used.

Remark For the resonator selection and oscillator constant, customers are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (1/3)

| Parameter | Symbol | Conditions |  |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output current, high | Іон | Per pin | $\mu$ PD78916x, 78917x, 78916xY,78917xY |  |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  |  | -15 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A}), 78917 \mathrm{x}(\mathrm{~A}), \\ & 78916 \mathrm{xY}(\mathrm{~A}), 78917 \mathrm{xY}(\mathrm{~A}) \end{aligned}$ |  |  |  | -1 | mA |
|  |  | Total for all p |  |  |  |  | -11 | mA |
| Output current, low | IoL | Per pin |  |  |  |  | 10 | mA |
|  |  | Total for all pins |  |  |  |  | 80 | mA |
|  |  | Per pin |  | $\mu$ PD78916x(A), 78917x(A), <br> 78916xY(A), 78917xY(A) |  |  | 3 | mA |
|  |  | Total for all p | pins 78916 |  |  |  | 60 | mA |
| Input voltage, high | $\mathrm{V}_{1+1}$ | $\begin{aligned} & \text { P00 to P05, P10, P11, } \\ & \text { P60 to P67 } \end{aligned}$ |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.7 VDD |  | VDD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | $0.9 \mathrm{~V}_{\mathrm{DD}}$ |  | VDD | V |
|  | $\mathrm{V}_{\mathbf{1 + 2}}$ | $\begin{aligned} & \hline \text { P50 to } \\ & \text { P53 } \end{aligned}$ | N -ch open drain | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.7 VDD |  | 12 | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 VDD |  | 12 | V |
|  |  |  | On-chip pullup resistor | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.7 V do |  | VDD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 VDD |  | VDD | V |
|  | Vінз | RESET, <br> P20 to P26, P30 to P33 |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.8 V DD |  | VDD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | $0.9 \mathrm{~V}_{\mathrm{DD}}$ |  | VDD | V |
|  | $\mathrm{V}_{1+4}$ | X1, X2, XT1, XT2 |  | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | VDD-0.5 |  | VDD | V |
|  |  |  |  | $V_{\text {DD }}=1.8$ to 5.5 V | VDD - 0.1 |  | VDD | V |
| Input voltage, low | VIL1 | $\begin{aligned} & \text { P00 to P05, P10, P11, } \\ & \text { P60 to P67 } \end{aligned}$ |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.3VDD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | 0.1 VDD | V |
|  | VIL2 | P50 to P53 |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.3 VdD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  | Vוъ | RESET, <br> P20 to P26, P30 to P33 |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.2 VDD | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{VDD}^{\text {d }}$ | V |
|  | VIL4 | X1, X2, XT1, XT2 |  | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 0 |  | 0.4 | V |
|  |  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | 0.1 | V |
| Output voltage, high | Vон | Pins other than P23, <br> P24, P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , $\mathrm{IOH}=-1 \mathrm{~mA}$ |  | VDD-1.0 |  |  | V |
|  |  |  | $\mathrm{V} \mathrm{DD}=1.8$ to $5.5 \mathrm{~V}, \mathrm{loH}=-100 \mu \mathrm{~A}$ |  | VDD - 0.5 |  |  | V |
| Output voltage, low | Volı | Pins other than P50 to P53 | $\begin{aligned} & \hline \mathrm{VD}=4.5 \mathrm{t} \\ & (\mu \mathrm{PD} 7891 \\ & 78917 \mathrm{xY}) \\ & \hline \end{aligned}$ | $\begin{aligned} & 5.5 \mathrm{~V} \text {, loL }=10 \mathrm{~mA} \\ & , 78917 \mathrm{x}, 78916 \mathrm{x}, \end{aligned}$ |  |  | 1.0 | V |
|  |  |  | $V_{D D}=4.5 t$ $\text { ( } \mu \text { PD7891 }$ <br> 78916xY( | .5 V , loL $=3 \mathrm{~mA}$ <br> (A), $78917 x(\mathrm{~A})$, <br> $78917 \mathrm{xY}(\mathrm{A}))$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ | $5 \mathrm{~V}, \mathrm{loL}=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
|  | Vol2 | P50 to P53 | $V_{D D}=4.5 t$ $\text { ( } \mu \text { PD7891 }$ <br> 78917xY) | $\begin{aligned} & 5.5 \mathrm{~V}, \mathrm{loL}=10 \mathrm{~mA} \\ & , 78917 \mathrm{x}, 78916 \mathrm{Y} \mathrm{Y} \end{aligned}$ |  |  | 1.0 | V |
|  |  |  | $V_{D D}=4.5 \mathrm{tc}$ $\text { ( } \mu \text { PD7891 }$ 78916xY(A | .5 V , loL $=3 \mathrm{~mA}$ <br> (A), 78917x(A), <br> $78917 x Y(A))$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to | .5 V, loL $=1.6 \mathrm{~mA}$ |  |  | 0.4 | V |

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## DC Characteristics $\left(\mathrm{T}_{\mathrm{A}}=-\mathbf{4 0}\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V$)(2 / 3)$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input current leakage, high | ІІн1 | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  | 3 | $\mu \mathrm{A}$ |
|  | ІІн2 |  | $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1, \mathrm{XT} 2$ |  |  | 20 | $\mu \mathrm{A}$ |
|  | ІІнз | $\mathrm{V}_{\mathrm{I}}=12 \mathrm{~V}^{\text {Note } 1}$ | P50 to P53 <br> ( N -ch open drain) |  |  | 20 | $\mu \mathrm{A}$ |
| Input current leakage, low | lıL1 | $\mathrm{V}_{\mathrm{I}}=0 \mathrm{~V}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  | -3 | $\mu \mathrm{A}$ |
|  | ILı2 |  | X1, X2, XT1, XT2 |  |  | -20 | $\mu \mathrm{A}$ |
|  | Іııз |  | P50 to P53 <br> (N-ch open drain) |  |  | $-3^{\text {Note } 2}$ | $\mu \mathrm{A}$ |
| Output current leakage, high | ІІон | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  |  |  | 3 | $\mu \mathrm{A}$ |
| Output current leakage, low | ILoL | V o $=0 \mathrm{~V}$ |  |  |  | -3 | $\mu \mathrm{A}$ |
| Software pull-up resistor | R1 | $\mathrm{V}_{1}=0 \mathrm{~V}$, for pins other than P23, P24, and P50 to P53 |  | 50 | 100 | 200 | k $\Omega$ |
| Mask option pullup resistor | R2 | V $=0 \mathrm{~V}$, P50 to P53 |  | 15 | 30 | 60 | k $\Omega$ |

Notes 1. When pull-up resistors are not connected to P50 to P53 (specified by the mask option).
2. A low-level input leakage current of $-60 \mu \mathrm{~A}$ (MAX.) flows only during the 1 -cycle time after a read instruction is executed to P50 to P53 when on-chip pull-up resistors are not connected to P50 to P53 (specified by the mask option) and P50 to P53 are set to input mode. At times other than this, a $-3 \mu \mathrm{~A}$ (MAX.) current flows.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (3/3)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply current | IDD1 ${ }^{\text {Note } 1}$ | 10.0 MHz crystal oscillation operating mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 3.2 | 8.0 | mA |
|  |  | 6.0 MHz crystal oscillation operating mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 2.0 | 4.7 | mA |
|  |  | 5.0 MHz crystal oscillation operating mode$(\mathrm{C} 1=\mathrm{C} 2=22 \mathrm{pF})$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 1.8 | 4.0 | mA |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 0.6 | 1.2 | mA |
|  |  |  | $\mathrm{VDD}=2.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 0.35 | 0.7 | mA |
|  | IDD2 ${ }^{\text {Note } 1}$ | 10.0 MHz crystal oscillation HALT mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 1.5 | 3.0 | mA |
|  |  | 6.0 MHz Crystal oscillation HALT mode | $\mathrm{V} D \mathrm{LD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 0.9 | 1.8 | mA |
|  |  | 5.0 MHz crystal oscillation HALT mode$(C 1=C 2=22 p F)$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 0.75 | 1.5 | mA |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 0.4 | 0.8 | mA |
|  |  |  | $\mathrm{VDD}=2.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 0.25 | 0.5 | mA |
|  | IDD3 ${ }^{\text {Note } 1}$ | 32.768 kHz crystal oscillation operating mode ${ }^{\text {Note } 3}$$\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 25 | 90 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%$ |  | 7.0 | 50 | $\mu \mathrm{A}$ |
|  |  |  | $V D D=2.0 \mathrm{~V} \pm 10 \%$ |  | 3.5 | 30 | $\mu \mathrm{A}$ |
|  | IDD4 ${ }^{\text {Note } 1}$ | 32.768 kHz crystal oscillatior HALT mode ${ }^{\text {Note } 3}$$\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & R=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 16 | 75 | $\mu \mathrm{A}$ |
|  |  |  | $V \mathrm{DD}=3.0 \mathrm{~V} \pm 10 \%$ |  | 4.5 | 35 | $\mu \mathrm{A}$ |
|  |  |  | $V D D=2.0 \mathrm{~V} \pm 10 \%$ |  | 2.3 | 18 | $\mu \mathrm{A}$ |
|  | IDD5 ${ }^{\text {Note } 1}$ | 32.768 kHz crystal stop STOP mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 0.1 | 10 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 5.0 | $\mu \mathrm{A}$ |
|  |  |  | $V \mathrm{DD}=2.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 3.0 | $\mu \mathrm{A}$ |
|  | IDD6 ${ }^{\text {Note } 2}$ | 10.0 MHz crystal oscillation A/D operating mode | $V_{D D}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 4.0 | 10.0 | mA |
|  |  | 6.0 MHz crystal oscillation A/D operating mode | VDD $=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 2.8 | 6.7 | mA |
|  |  | 5.0 MHz crystal oscillation A/D operating mode (C1 = C2 = 22 pF ) | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 2.6 | 6.0 | mA |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 1.4 | 3.2 | mA |
|  |  |  | $\mathrm{VDD}=2.0 \mathrm{~V} \pm 10 \%^{\text {Note } 5}$ |  | 1.15 | 2.7 | mA |

Notes 1. The $A V_{\text {ref }} O N$ (ADCSO (bit 7 of ADMO; $A / D$ converter mode register 0 ) $=1$ ), $A V D d$, and the port current (including the current flowing through the internal pull-up resistors) are not included.
2. The $A V_{\text {REF }} O N(A D C S O=1)$ and port current (including the current flowing through the internal pull-up resistors) are not included. Refer to the A/D converter characteristics for the current flowing through AVref.
3. When the main system clock is stopped.
4. During high-speed mode operation (when the processor clock control register (PCC) is set to 00 H .)
5. During low-speed mode operation (when PCC is set to 02H)

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## AC Characteristics

(1) Basic operation ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Cycle time (minimum instruction execution time) | Tcy | Operation based on the main system clock | $V_{\text {DD }}=4.5$ to 5.5 V | 0.2 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=3.0$ to 5.5 V | 0.33 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.4 |  | 8 | $\mu \mathrm{S}$ |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 1.6 |  | 8 | $\mu \mathrm{s}$ |
|  |  | Operation based on the subsystem clock |  | 114 | 122 | 125 | $\mu \mathrm{s}$ |
| TI80 and TI81 input frequency | $\mathrm{f}_{\text {T }}$ | V DD $=2.7$ to 5.5 V |  | 0 |  | 4 | MHz |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 0 |  | 275 | kHz |
| TI80 and TI81 input high-/low-level width | ttin, ttil | $V_{\text {dD }}=2.7$ to 5.5 V |  | 0.1 |  |  | $\mu \mathrm{s}$ |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  | 1.8 |  |  | $\mu \mathrm{s}$ |
| Interrupt input high-/low-level width | tinth, tintl | INTP0 to INTP3 |  | 10 |  |  | $\mu \mathrm{s}$ |
| RESET input lowlevel width | trsL |  |  | 10 |  |  | $\mu \mathrm{s}$ |
| CPT90 input high-/low-level width | tcph, tcPL |  |  | 10 |  |  | $\mu \mathrm{s}$ |

Tcy vs. Vdd (main system clock)

(2) Serial interface $\operatorname{SIO20}\left(\mathrm{T}_{\mathrm{A}}=-40\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V$)$
(a) 3-wire serial I/O mode (SCK20...Internal clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy 1 | $V_{D D}=2.7$ to 5.5 V |  | 800 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 3200 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/lowlevel width | $\mathrm{t}_{\text {KH1 }}$, $\mathrm{tkL1}^{1}$ | $V_{D D}=2.7$ to 5.5 V |  | $\mathrm{tkcyı}_{1 / 2}$ - 50 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | $\mathrm{tkcy1}_{1 / 2-150}$ |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik1 | $V_{D D}=2.7$ to 5.5 V |  | 150 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 500 |  |  | ns |
| SI20 hold time (from $\overline{\text { SCK2O } \uparrow \text { ) }}$ | tksı11 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tksO1 | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega, \\ & \mathrm{C}=100 \mathrm{pF}^{\mathrm{Note}} \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 250 | ns |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 0 |  | 1000 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(b) 3-wire serial I/O mode (SCK20...External clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tксү2 | $V_{D D}=2.7$ to 5.5 V |  | 900 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 3500 |  |  | $n s$ |
| SCK20 high-/lowlevel width | tкH2, tkL2 | $V_{\text {DD }}=2.7$ to 5.5 V |  | 400 |  |  | $n \mathrm{~s}$ |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 1600 |  |  | $n s$ |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik2 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 100 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 150 |  |  | $n s$ |
| SI20 hold time (from SCK20 $\uparrow$ ) | tks12 | $V_{\text {dD }}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tkso2 | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega \\ & \mathrm{C}=100 \mathrm{pF}^{\text {Note }} \end{aligned}$ | $V_{\text {dD }}=2.7$ to 5.5 V | 0 |  | 300 | $n s$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=1.8$ to 5.5 V | 0 |  | 1000 | ns |
| SO20 setup time (when using SS20, to $\overline{\mathrm{SS} 20} \downarrow$ ) | tKAS2 | $V_{D D}=2.7$ to 5.5 V |  |  |  | 120 | ns |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  |  |  | 400 | $n s$ |
| SO20 disable time (when using $\overline{\text { SS20 }}$, from $\overline{\mathrm{SS} 2 \mathrm{O}} \uparrow$ ) | tkDS2 | $V_{D D}=2.7$ to 5.5 V |  |  |  | 240 | ns |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  |  |  | 800 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(c) UART mode (dedicated baud rate generator output)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :--- | :---: | :---: | :---: | :---: |
| Transfer rate |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  |  | 78125 | bps |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |  | 19531 | bps |

(d) UART mode (external clock input)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASCK20 cycle time | tксуз | $\mathrm{V} \mathrm{DD}=2.7$ to 5.5 V | 900 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 3500 |  |  | ns |
| ASCK20 high-/lowlevel width | tкн3, tкL3 | $V_{\text {DD }}=2.7$ to 5.5 V | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 1600 |  |  | ns |
| Transfer rate |  | $V_{D D}=2.7$ to 5.5 V |  |  | 39063 | bps |
|  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 9766 | bps |
| ASCK20 rise time, fall time | $\mathrm{t}_{\mathrm{R}, \mathrm{t}} \mathrm{t}$ |  |  |  | 1 | $\mu \mathrm{s}$ |

(3) Serial interface SMB0 ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) ( $\mu \mathrm{PD} 78916 \mathrm{xY}, 78917 \mathrm{xY}, 78916 \mathrm{xY}(\mathrm{A}), 78917 \mathrm{xY}(\mathrm{A})$ only)
(a) DC characteristics

| Parameter | Symbol |  | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage, high | V ${ }_{\text {H }}$ | SCLO, SDAO (at hysteresis) |  | 0.8VDD |  | VDD | V |
| Input voltage, low | VIL | SCLO, SDAO (at hysteresis) |  | 0 |  | 0.2 Vdo | V |
| Output voltage, low | VoL | SCLO, SDAO | $\mathrm{V} \mathrm{DD}=4.5$ to 5.5 V , $\mathrm{IoL}=10 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | V DD $=1.8$ to 5.5 V , loL $=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
| Input current leakage, high | ІІІ | SCLO, SDAO | $V_{1}=V_{D D}$ |  |  | 3 | $\mu \mathrm{A}$ |
| Input current leakage, low | ILIL | SCLO, SDAO | V I $=0 \mathrm{~V}$ |  |  | -3 | $\mu \mathrm{A}$ |

(b) DC characteristics (when using comparator)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input range | VsdA, <br> Vscl | $V_{D D}=1.8$ to 5.5 V | 0 |  | 5.5 | V |
| Transfer level | VISDA, <br> Viscl | $4.5 \leq \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ | $0.72 \mathrm{~V}_{\text {Ismb }}$ | Vismb | 1.28 V Ismb | V |
|  |  | $3.3 \leq \mathrm{VDD}^{2} 4.5 \mathrm{~V}$ | $0.78 \mathrm{~V}_{\text {Ismb }}$ | $V_{\text {ISmb }}$ | $1.22 \mathrm{~V}_{\text {ISM }}$ | V |
|  |  | $2.7 \leq \mathrm{VDD}<3.3 \mathrm{~V}$ | 0.75 V Ismb | VIsmb | 1.25 V Ismb | V |
|  |  | $1.8 \leq \mathrm{VDD}^{2} \times 2.7 \mathrm{~V}$ | $0.90 \mathrm{~V}_{\text {Ismb }}$ | $V_{\text {ISmb }}$ | $1.45 \mathrm{~V}_{\text {ISM }}$ | V |
| Input level threshold value ${ }^{\text {Note }}$ | VIsmb | LVL01, LVL00 = 0, 1 |  | $0.25 \times$ VD |  | V |
|  |  | LVL01, LVL00 $=1,0$ |  | $0.375 \times \mathrm{VoD}$ |  | V |
|  |  | LVL01, LVL00 = 1, 1 |  | $0.5 \times \mathrm{VDD}$ |  | V |

Note VIsmb is an input level threshold value selected by bits LVLOO and LVL01 (bits 0 and 1 of SMB input level setting register 0 (SMBVIO)).
According to the SMB standard (V1.1), the maximum value of low-level input voltage is 0.8 V , and the minimum value of high-level input voltage, 2.1 V . To satisfy these conditions, set LVL01 and LVL00 as follows.

- When Vod = 1.8 to $3.3 \mathrm{~V}: \quad$ LVL01, $\mathrm{LVL00}=1,1(0.5 \times \mathrm{V} D \mathrm{D})$
- When VDD $=3.3$ to $4.5 \mathrm{~V}: \operatorname{LVL01,~LVL00~}=1,0(0.375 \times \mathrm{VDD})$
- When VDd $=4.5$ to 5.5 V : LVL01, LVLOO $=0,1(0.25 \times \mathrm{VDD})$
"LVL01, LVL00 $=0,0$ " is not possible since this setting does not satisfy the SMB standard (V1.1).
(c) AC characteristics

| Parameter |  | Symbol | SMB Mode |  | Standard Mode $I^{2} \mathrm{C}$ Bus |  | High-Speed Mode $I^{2} \mathrm{C}$ Bus |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | MAX. | MIN. | MAX. | MIN. | MAX. |  |
| SCL0 clock frequency |  |  | fclk | 10 | 100 | 0 | 100 | 0 | 400 | kHz |
| Bus free time <br> (between stop and start condition) |  | tbuF | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| Hold time ${ }^{\text {Note } 1}$ |  | thd:STA | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Start/restart condition setup time |  | tsu:STA | 4.7 | - | 4.7 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Stop condition setup time |  | tsu:sto | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Data hold time | When using CBUScompatible master | thd:dat | - | - | 5 | - | - | - | $\mu \mathrm{s}$ |
|  | When using SMB/IIC bus |  | 300 | - | $0^{\text {Note } 2}$ | - | $0^{\text {Note } 2}$ | $900^{\text {Note } 3}$ | ns |
| Data setup time |  | tsu:DAT | 250 | - | 250 | - | $100{ }^{\text {Note } 4}$ | - | ns |
| SCLO clock low-level width |  | tow | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| SCL0 clock high-level width |  | thigh | 4.0 | 50 | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| SCL0 and SDA0 signal fall time |  | $\mathrm{tF}_{F}$ | - | 300 | - | 300 | - | 300 | ns |
| SCL0 and SDA0 signal rise time |  | $\mathrm{tR}_{R}$ | - | 1000 | - | 1000 | - | 300 | ns |
| Spike pulse width controlled by input filter |  | tsp | - | - | - | - | 0 | 50 | ns |
| Timeout |  | ttimeout | 25 | 35 | - | - | - | - | ms |
| Total extended time of SCLO clock low-level period (slave) |  | tıow:SEXT | - | 25 | - | - | - | - | ms |
| Total extended time of cumulative clock low-level period (master) |  | tLow:MEXT | - | 10 | - | - | - | - | ms |
| Capacitive load per each bus line |  | Cb | - | - | - | 400 | - | 400 | pF |

Notes 1. In the start condition, the first clock pulse is generated after this hold time.
2. To fill in the undefined area of the SCLO falling edge, it is necessary for the device to internally provide at least 300 ns of hold time for the SDA0 signal (which is ViHmin. of the SCLO signal).
3. If the device does not extend the SCLO signal low hold time (toow), only maximum data hold time thd:Dat needs to be fulfilled.
4. The high-speed mode $I^{2} C$ bus is available in the $S M B$ mode and the standard mode $I^{2} C$ bus system. At this time, the conditions described below must be satisfied.

- If the device extends the SCLO signal low state hold time
tsu:DAT $\geq 250 \mathrm{~ns}$
- If the device extends the SCLO signal low state hold time

Be sure to transmit the next data bit to the SDAO line before the SCLO line is released (trmax.+ tsu:DAT $=1000+250=1250 \mathrm{~ns}$ by the SMB mode or the standard mode $\mathrm{I}^{2} \mathrm{C}$ bus specification).

## AC Timing Measurement Points (Excluding X1 and XT1 Inputs)



## Clock Timing



## TI Timing



Interrupt Input Timing


## RESET Input Timing



## CPT90 Input Timing



## Serial Transfer Timing

## 3-wire serial I/O mode:



Remark $m=1,2$

3-wire serial I/O mode (when using SS20):


UART mode (external clock input):


## SMB mode:



8-Bit A/D Converter Characteristics ( $\mu$ PD78916x, 78916xY, 78916x(A), 78916xY(A))
( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, 1.8 \leq \mathrm{A} V_{\mathrm{ref}} \leq \mathrm{AV}$ dd $=\mathrm{V}_{\mathrm{dD}} \leq 5.5 \mathrm{~V}$, AV ss $=\mathrm{V}_{\mathrm{ss}}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 8 | 8 | 8 | bit |
| Overall error ${ }^{\text {Note }}$ |  | $2.7 \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.4$ | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \leq \mathrm{AV}_{\text {ref }} \leq \mathrm{AV}$ do $\leq 5.5 \mathrm{~V}$ |  | $\pm 0.8$ | $\pm 1.2$ | \%FSR |
| Conversion time | tconv | $4.5 \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ | 12 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $2.7 \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ | 14 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $1.8 \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ | 28 |  | 100 | $\mu \mathrm{s}$ |
| Analog input voltage | $V_{\text {IAN }}$ |  | 0 |  | $\mathrm{AV}_{\text {ReF }}$ | V |
| Reference voltage | AV ${ }_{\text {ref }}$ |  | 1.8 |  | AVDD | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{\text {ss }}$ | Radref |  | 20 | 40 |  | $\mathrm{k} \Omega$ |

Note Excludes quantization error ( $\pm 0.2 \% \mathrm{FSR}$ ).
Remark FSR: Full scale range

10-Bit A/D Converter Characteristics ( $\mu$ PD78917x, 78917xY, 78917x(A), 78917xY(A))
( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, 1.8 \leq \mathrm{A} V_{\mathrm{ref}} \leq \mathrm{AVdd}=\mathrm{V}_{\mathrm{dd}} \leq 5.5 \mathrm{~V}$, $\mathrm{AV} \mathrm{ss}=\mathrm{Vss}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 10 | 10 | 10 | bit |
| $\qquad$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.2$ | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.4$ | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.8$ | $\pm 1.2$ | \%FSR |
| Conversion time | tconv | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VdD}^{5} 5.5 \mathrm{~V}$ | 12 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ | 14 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VdD}^{5} 5.5 \mathrm{~V}$ | 28 |  | 100 | $\mu \mathrm{s}$ |
| Zero-scale error ${ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VdD} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| Full-scale error ${ }^{\text {N }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| Integral linearity error ${ }^{\text {Note }}$ | INL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 2.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 4.5$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 8.5$ | LSB |
| Differential linearity error ${ }^{\text {Note }}$ | DNL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {d }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VdD}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 2.0$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV}$ dD $\leq 5.5 \mathrm{~V}$ |  |  | $\pm 3.5$ | LSB |
| Analog input voltage | VIAN |  | 0 |  | AV ${ }_{\text {ref }}$ | V |
| Reference voltage | AVref |  | 1.8 |  | AVDD | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{s s}$ | Radref |  | 20 | 40 |  | $k \Omega$ |

Note Excludes quantization error ( $\pm 0.05 \%$ FSR).
Remark FSR: Full scale range

Data Memory STOP Mode Low Power Supply Voltage Data Retention Characteristics ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $\mathbf{+ 8 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :--- | :---: | :---: | :---: | :---: |
| Data retention power <br> supply voltage | VDDDR |  | 1.8 |  | 5.5 | V |
| Release signal set time | tsReL |  | 0 |  |  | $\mu \mathrm{~s}$ |
| Oscillation stabilization <br> wait time | twait 1 |  |  |  |  |  |

Notes 1. The oscillation stabilization time is the time the CPU operation is stopped to prevent unstable operation when oscillation starts.
2. By using bits 0 to 2 (OSTSO to OSTS2) of the oscillation stabilization time selection register (OSTS), $2^{12} / \mathrm{fx}, 2^{15} / \mathrm{fx}$, or $2^{17} / \mathrm{fx}$ can be selected.

Remark fx: Main system clock oscillation frequency

Data Retention Timing (STOP Mode Release by $\overline{\text { RESET }}$ )


Data Retention Timing (Standby Release Signal: STOP Mode Release by Interrupt Signal)


CHAPTER 24 CHARACTERISTICS CURVES ( $\mu$ PD78916x, 17x, 16xY, 17xY, 16x(A), 17x(A), 16xY(A), 17xY(A))




CHAPTER 25 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78916x(A1), 17x(A1), 16x(A2), 17x(A2))

Absolute Maximum Ratings ( $\mathrm{T}_{\mathrm{A}}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions |  | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | Vdo | $\begin{aligned} & A V_{D D}-0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\text {REF }} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\text {REF }} \leq \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \end{aligned}$ |  | -0.3 to +6.5 | V |
|  | AVDD |  |  | V |
|  | AV VEFF |  |  | V |
| Input voltage | $\mathrm{V}_{11}$ | Pins other than P50 to P53, P23, P24 |  |  | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
|  | $\mathrm{V}_{12}$ | P23, P24 |  |  | -0.3 to +5.5 | V |
|  | $V_{13}$ | P50 to P53 | N -ch open drain | -0.3 to +13 | V |
|  |  |  | On-chip pull-up resistor | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
| Output voltage | Vo |  |  | -0.3 to $\mathrm{V}_{\mathrm{DD}}+0.3$ | V |
| Output current, high | Іон | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A} 1), \\ & 78917 \mathrm{x}(\mathrm{~A} 1) \end{aligned}$ | -4 | mA |
|  |  | Total for all pins |  | -14 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \text { PD78916x(A2), } \\ & 78917 x(\text { A2) } \end{aligned}$ | -2 | mA |
|  |  | Total for all pins |  | -6 | mA |
| Output current, low | loL | Per pin | $\begin{aligned} & \mu \text { PD78916x(A1), } \\ & 78917 x(\text { A1) } \end{aligned}$ | 5 | mA |
|  |  | Total for all pins |  | 80 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \text { PD78916x(A2), } \\ & \text { 78917x(A2) } \end{aligned}$ | 2 | mA |
|  |  | Total for all pins |  | 40 | mA |
| Operating ambient temperature | TA | $\mu$ PD78916x(A1), 78917x(A1) |  | -40 to +110 | ${ }^{\circ} \mathrm{C}$ |
|  |  | $\mu \mathrm{PD} 78916 \times(\mathrm{A} 2), 78917 \times(\mathrm{A} 2)$ |  | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ |  |  | -65 to +150 | ${ }^{\circ} \mathrm{C}$ |

Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## Main System Clock Oscillator Characteristics

(VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1))$,
$=-40$ to $+125^{\circ} \mathrm{C}(\mu$ PD78916x(A2), 78917x(A2)))


Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the main system clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. When the main system clock is stopped and the device is operating on the subsystem clock, wait until the oscillation stabilization time has been secured by the program before switching back to the main system clock.
3. For ceramic resonator, use the part number for which the resonator manufacturer guarantees operation under the following conditions.

$$
\begin{aligned}
& \mu \text { PD78916x(A1), } 78917 x(\mathrm{~A} 1): \mathrm{T}_{A}=110^{\circ} \mathrm{C} \\
& \mu \text { PD78916x(A2), } 78917 x(\mathrm{~A} 2): \mathrm{T}_{\mathrm{A}}=125^{\circ} \mathrm{C}
\end{aligned}
$$

Remark For the resonator selection and oscillator constant, users are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

## Subsystem Clock Oscillator Characteristics

(VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 1), 78917 \times(\mathrm{A} 1)$ ),

$$
\left.=-40 \text { to }+125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A} 2), 78917 \mathrm{x}(\mathrm{~A} 2))\right)
$$

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal resonator | \|Vsso XT1 XT2 | Oscillation frequency ( $\mathrm{fxT}^{\text {Note } 1}$ |  | 32 | 32.768 | 35 | kHz |
|  |  |  |  |  |  |  |  |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ |  |  | 1.2 | 2 | S |
| External clock | XT1 $\quad$ XT2 | XT1 input frequency ( $\mathrm{fxx}^{\text {( }}$ Note 1 |  | 32 |  | 35 | kHz |
|  |  | XT1 input high-/low-level width (tхтн, tхтц) |  | 14.3 |  | 15.6 | $\mu \mathrm{S}$ |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the subsystem clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. The subsystem clock oscillator is designed as a low-amplitude circuit for reducing current consumption, and is more prone to malfunction due to noise than the main system clock oscillator. Particular care is therefore required with the wiring method when the subsystem clock is used.

Remark For the resonator selection and oscillator constant, users are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

## DC Characteristics

(VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1))$,

$$
=-40 \text { to }+125^{\circ} \mathrm{C}(\mu \text { PD78916x(A2), 78917x(A2))) (1/3) }
$$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output current, high | Іон | Per pin | $\mu \mathrm{PD78916x}(\mathrm{~A} 1), 78917 \times(\mathrm{A} 1)$ |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -7 | mA |
|  |  | Per pin | $\mu \mathrm{PD78916x}(\mathrm{~A} 2), 78917 \mathrm{x}(\mathrm{A} 2)$ |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -3 | mA |
| Output current, low | IoL | Per pin | $\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 1), 78917 \mathrm{x}(\mathrm{A} 1)$ |  |  | 1.6 | mA |
|  |  | Total for all pins |  |  |  | 40 | mA |
|  |  | Per pin | $\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 2), 78917 \mathrm{x}(\mathrm{A} 2)$ |  |  | 1.6 | mA |
|  |  | Total for all pins |  |  |  | 20 | mA |
| Input voltage, high | $\mathrm{V}_{\mathrm{HH}}$ | P00 to P05, P10, P11, P60 to P67 |  | 0.7 VdD |  | VDD | V |
|  | $\mathrm{V}_{1+2}$ | P50 to P53 | N -ch open drain | 0.7 VdD |  | 10 | V |
|  |  |  | On-chip pull-up resistor | 0.7 VDD |  | VDD | V |
|  | Vінз | RESET, P20 to P26, P30 to P33 |  | 0.8 VdD |  | VDD | V |
|  | $\mathrm{V}_{1+4}$ | X1, X2, XT1, XT2 |  | VDD - 0.1 |  | VDD | V |
| Input voltage, low | VIL1 | P00 to P05, P10, P11, P60 to P67 |  | 0 |  | 0.3VdD | V |
|  | VIL2 | P50 to P53 |  | 0 |  | 0.3VDD | V |
|  | Vוьз | RESET, P20 to P26, P30 to P33 |  | 0 |  | 0.2 VdD | V |
|  | VIL4 | X1, X2, XT1, XT2 |  | 0 |  | 0.1 | V |
| Output voltage, high | Vон | Pins other than P23, P24, P50 to P53 | $\mathrm{IOH}^{\text {a }}$ - 1 mA | VDD-2.0 |  |  | V |
|  |  |  | IOH $=-100 \mu \mathrm{~A}$ | VDD - 1.0 |  |  | V |
| Output voltage, low | Volı | Pins other than P50 to P53 | $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 2.0 | V |
|  |  |  | loL $=400 \mu \mathrm{~A}$ |  |  | 1.0 | V |
|  | Vol2 | P50 to P53 | $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 1.0 | V |

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## DC Characteristics

(Vdd $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1))$,

$$
=-40 \text { to }+125^{\circ} \mathrm{C}(\mu \text { PD78916x(A2), 78917x(A2))) (2/3) }
$$

| Parameter | Symbol | Conditions |  |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input leakage current, high | ІІн1 | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  |  | 10 | $\mu \mathrm{A}$ |
|  | lıн2 |  | X1, X2, XT1, XT2 |  |  |  | 20 | $\mu \mathrm{A}$ |
|  | ІІнз | $V_{l}=10 \mathrm{~V}^{\text {Notel }}$ | $\begin{array}{\|l\|} \hline \begin{array}{l} \text { P50 to P53 } \\ \text { (N-ch open drain) } \end{array} \\ \hline \end{array}$ |  |  |  | 80 | $\mu \mathrm{A}$ |
| Input leakage current, low | ILı1 | V I $=0 \mathrm{~V}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  |  | -10 | $\mu \mathrm{A}$ |
|  | ILı2 |  | X1, X2, XT1, XT2 |  |  |  | -20 | $\mu \mathrm{A}$ |
|  | lıเз |  | $\begin{aligned} & \text { P50 to P53 } \\ & \text { (N-ch open drain) } \end{aligned}$ |  |  |  | $-10^{\text {Note } 2}$ | $\mu \mathrm{A}$ |
| Output leakage current, high | ILoh | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  |  |  |  | 10 | $\mu \mathrm{A}$ |
| Output leakage current, low | ILoL | V o $=0 \mathrm{~V}$ |  |  |  |  | -10 | $\mu \mathrm{A}$ |
| Software pull-up resistor | R1 | $\mathrm{V}_{1}=0 \mathrm{~V}$, for pins other than P23, P24, and P50 to P53 |  |  | 50 | 100 | 300 | k $\Omega$ |
| Mask option pullup resistor | R2 | $\mathrm{V}_{1}=0 \mathrm{~V}$, P50 to P53 |  | $\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 1), 78917 \times(\mathrm{A} 1)$ | 15 | 30 | 100 | $\mathrm{k} \Omega$ |
|  |  |  |  | $\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 2), 78917 \times(\mathrm{A} 2)$ | 10 | 30 | 100 | k $\Omega$ |

Notes 1. When pull-up resistors are not connected to P50 to P53 (specified by mask option).
2. A low-level input leakage current of $-60 \mu \mathrm{~A}$ (MAX.) flows only during the 1 -cycle time after a read instruction is executed to P50 to P53 when on-chip pull-up resistors are not connected to P50 to P53 (specified by mask option) and P50 to P53 are set to input mode. At times other than this, a $-10 \mu \mathrm{~A}$ (MAX.) current flows.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## DC Characteristics

$\left(\mathrm{VDD}=4.5\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1))$,

$$
\left.=-40 \text { to }+125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{~A} 2), 78917 \mathrm{x}(\mathrm{~A} 2))\right)(3 / 3)
$$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply current | IDD1 ${ }^{\text {Note } 1}$ | 5.0 MHz crystal oscillation operating mode (C1 = C2 = 22 pF ) | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%^{\text {Nole } 4}$ |  | 2.0 | 8.0 | mA |
|  | IdD2 ${ }^{\text {Note } 1}$ | 5.0 MHz crystal oscillation HALT mode (C1 = C2 = 22 pF ) | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 1.0 | 5.0 | mA |
|  | $\mathrm{IDDa}^{\text {Note }} 1$ | 32.768 kHz crystal oscillation operating mode ${ }^{\text {Note } 3}$ $\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & R=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%$ |  | 30 | 1200 | $\mu \mathrm{A}$ |
|  | IDD4 ${ }^{\text {Note } 1}$ | 32.768 kHz crystal oscillation HALT mode ${ }^{\text {Note } 3}$ $\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & R=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \pm 10 \%$ |  | 25 | 1100 | $\mu \mathrm{A}$ |
|  | IdD5 $5^{\text {Note } 1}$ | 32.768 kHz crystal stop STOP mode | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%$ |  | 0.1 | 1000 | $\mu \mathrm{A}$ |
|  | Idob ${ }^{\text {Note } 2}$ | 5.0 MHz crystal oscillation A/D operating mode (C1 = C2 = 22 pF ) | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 3.0 | 10.0 | mA |

Notes 1. The $A V_{\text {refo }} O N$ (ADCSO (bit 7 of ADMO; A/D converter mode register 0 ) $=1$ ), $A V_{D D}$, and port current (including the current flowing through the internal pull-up resistors) is not included.
2. The $A V_{\text {ref }} O N(A D C S O=1)$ and port current (including the current flowing through the internal pull-up resistors) is not included. Refer to the A/D converter characteristics for the current flowing through AVref.
3. When the main system clock is stopped.
4. During high-speed mode operation (when the processor clock control register (PCC) is set to 00H.)

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## AC Characteristics

## (1) Basic operation

(VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1)$ ), $=-40$ to $+125^{\circ} \mathrm{C}(\mu$ PD78916x(A2), 78917x(A2)))

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. |
| :--- | :---: | :--- | :---: | :---: | :---: |
| Cycle time <br> (minimum instruction <br> execution time) | Tcy | Operation based on the main system clock | 0.4 |  | 8 |

Tcy vs. Vdd (main system clock)

(2) Serial interface 20
(VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 1), 78917 x(\mathrm{~A} 1))$,

$$
=-40 \text { to }+125^{\circ} \mathrm{C}(\mu \text { PD78916x(A2), 78917x(A2))) }
$$

(a) 3-wire serial I/O mode (SCK20...Internal clock)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy1 |  | 800 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/lowlevel width | tKH1, $^{\text {tkL1 }}$ |  | $\mathrm{tKCY}_{1} / 2-50$ |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik1 |  | 150 |  |  | $n s$ |
| SI20 hold time (from SCK20 $\uparrow$ ) | tksı11 |  | 400 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tKsO1 | $\mathrm{R}=1 \mathrm{k} \Omega, \mathrm{C}=100 \mathrm{pF}^{\text {Note }}$ | 0 |  | 250 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(b) 3-wire serial I/O mode (SCK20.. External clock)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy2 |  | 900 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/lowlevel width | tKH2, tKL2 |  | 400 |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsIK2 |  | 100 |  |  | ns |
| SI20 hold time (from SCK20 $\uparrow$ ) | tksI2 |  | 400 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tKsO2 | $\mathrm{R}=1 \mathrm{k} \Omega, \mathrm{C}=100 \mathrm{~F}^{\text {Note }}$ | 0 |  | 300 | ns |
| SO20 setup time (when using $\overline{\mathrm{SS2O}}$, to $\overline{S S 20} \downarrow$ ) | tKAS2 |  |  |  | 120 | ns |
| SO20 disable time <br> (when using $\overline{\mathrm{SS2O}}$, from $\overline{\text { SS20 }} \uparrow$ ) | tkDS2 |  |  |  | 240 | ns |

Note $R$ and $C$ are the load resistance and load capacitance of the SO20 output line.
(c) UART mode (dedicated baud rate generator output)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Transfer rate |  |  |  |  | 78125 | bps |

(d) UART mode (external clock input)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| ASCK20 cycle <br> time | tксҮ3 |  | 900 |  |  |  |
| ASCK20 high-/low- <br> level width | tкн3, tкL3 |  | 400 |  | ns |  |
| Transfer rate |  |  |  | ns |  |  |
| ASCK20 rise time, <br> fall time | $\mathrm{t}_{\mathrm{R}, \mathrm{t}} \mathrm{t}$ |  |  |  |  |  |

## AC Timing Measurement Points (Excluding X1 and XT1 Inputs)



## Clock Timing



TI Timing

T180, T181


Interrupt Input Timing

$\overline{\text { RESET Input Timing }}$


## CPT90 Input Timing



## Serial Transfer Timing

## 3-wire serial I/O mode:



Remark $m=1,2$

3-wire serial I/O mode (when using SS20):


UART mode (external clock input):


8-Bit A/D Converter Characteristics ( $\mu$ PD78916x(A1), 78916x(A2))
( $\mathrm{T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1)),-40$ to $+125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 2))$
$4.5 \leq \mathrm{AV}_{\mathrm{REF}} \leq \mathrm{AV} \mathrm{DD}^{2}=\mathrm{V}_{\mathrm{dD}} \leq 5.5 \mathrm{~V}$, AV Ss $=\mathrm{V}_{\mathrm{ss}}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 8 | 8 | 8 | bit |
| Overall error ${ }^{\text {Note }}$ |  |  |  | $\pm 0.4$ | $\pm 1.0$ | \%FSR |
| Conversion time | tconv |  | 14 |  | 28 | $\mu \mathrm{s}$ |
| Analog input voltage | Vian |  | 0 |  | AV ${ }_{\text {ref }}$ | V |
| Reference voltage | $\mathrm{AV}_{\text {REF }}$ |  | 4.5 |  | AV ${ }_{\text {do }}$ | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{\text {ss }}$ | Radref |  | 20 | 40 |  | $\mathrm{k} \Omega$ |

Note Excludes quantization error ( $\pm 0.2 \%$ FSR $)$.
Remark FSR: Full scale range

10-Bit A/D Converter Characteristics ( $\mu$ PD78917x(A1), 78917x(A2))
( $\mathrm{T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}\left(\mu \mathrm{PD} 78917 x(\mathrm{~A} 1)\right.$ ), -40 to $+125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78917 x(\mathrm{~A} 2))$


| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 10 | 10 | 10 | bit |
| Overall error ${ }^{\text {Note }}$ |  |  |  | $\pm 0.2$ | $\pm 0.6$ | \%FSR |
| Conversion time | tconv |  | 14 |  | 28 | $\mu \mathrm{s}$ |
| Zero-scale error ${ }^{\text {Note }}$ |  |  |  |  | $\pm 0.6$ | \%FSR |
| Full-scale error ${ }^{\text {Note }}$ |  |  |  |  | $\pm 0.6$ | \%FSR |
| Integral linearity error ${ }^{\text {Note }}$ | INL |  |  |  | $\pm 4.5$ | LSB |
| Differential linearity error ${ }^{\text {Note }}$ | DNL |  |  |  | $\pm 2.0$ | LSB |
| Analog input voltage | Vian |  | 0 |  | $\mathrm{AV}_{\text {ReF }}$ | V |
| Reference voltage | $\mathrm{AV}_{\text {ReF }}$ |  | 4.5 |  | $A V_{\text {DD }}$ | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{\text {ss }}$ | Radref |  | 20 | 40 |  | $\mathrm{k} \Omega$ |

Note Excludes quantization error ( $\pm 0.05 \%$ FSR $)$.
Remark FSR: Full scale range

Data Memory STOP Mode Low Power Supply Voltage Data Retention Characteristics
( $\mathrm{T}_{\mathrm{A}}=-40$ to $+110^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 x(\mathrm{~A} 1), 78917 x(\mathrm{~A} 1)$ ),
$=-40$ to $+125^{\circ} \mathrm{C}(\mu \mathrm{PD} 78916 \mathrm{x}(\mathrm{A} 2), 78917 \mathrm{x}(\mathrm{A} 2))$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. |
| :--- | :---: | :--- | :---: | :---: | :---: |
| Data retention power <br> supply voltage | VDDDR |  | 4.5 |  | 5.5 |
| Release signal set time | tsREL |  | 0 |  |  |
| Oscillation stabilization <br> wait time | twait 1 | Release by RESET |  |  |  |

Notes 1. The oscillation stabilization time is the time the CPU operation is stopped to prevent unstable operation when oscillation starts.
2. $2^{12} / \mathrm{fx}, 2^{15} / \mathrm{fx}$, or $2^{17} / \mathrm{fx}$ can be selected by using bits 0 to 2 (OSTSO to OSTS2) of the oscillation stabilization time selection register (OSTS).

Remark fx: Main system clock oscillation frequency

## Data Retention Timing (STOP Mode Release by $\overline{\text { RESET }}$ )



## Data Retention Timing (Standby Release Signal: STOP Mode Release by Interrupt Signal)



CHAPTER 26 CHARACTERISTICS CURVES ( $\mu$ PD78916x(A1), 17x(A1), 16x(A2), 17x(A2))




## CHAPTER 27 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177A, 78F9177AY, 78F9177A(A), 78F9177AY(A))

Absolute Maximum Ratings ( $\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions |  | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage | VDD | $\begin{aligned} & \mathrm{A} \mathrm{~V}_{\mathrm{DD}}-0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\text {REF }} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\text {REF }} \leq \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \end{aligned}$ |  | -0.3 to +6.5 | V |
|  | AV ${ }_{\text {dD }}$ |  |  | V |
|  | $\mathrm{AV}_{\text {REF }}$ |  |  | V |
|  | VPP | Note |  |  | -0.3 to +10.5 | V |
| Input voltage | $V_{11}$ | Pins other than P50 to P53, P23, P24 |  |  | -0.3 to VDD +0.3 | V |
|  | $\mathrm{V}_{12}$ | P23, P24 |  | -0.3 to +5.5 | V |
|  | $V_{13}$ | P50 to P53 |  | -0.3 to +13 | V |
| Output voltage | Vo |  |  | -0.3 to $\mathrm{VDD}^{\text {d }} 0.3$ | V |
| Output current, high | Іон | Per pin | $\mu$ PD78F9177A, $\mu$ PD78F9177AY | -10 | mA |
|  |  | Total for all pins |  | -30 | mA |
|  |  | Per pin | $\mu$ PD78F9177A(A), $\mu$ PD78F9177AY(A) | -7 | mA |
|  |  | Total for all pins |  | -22 | mA |
| Output current, low | lot | Per pin | $\mu$ PD78F9177A, $\mu$ PD78F9177AY | 30 | mA |
|  |  | Total for all pins |  | 160 | mA |
|  |  | Per pin | $\mu$ PD78F9177A(A), $\mu$ PD78F9177AY(A) | 10 | mA |
|  |  | Total for all pins |  | 120 | mA |
| Operating ambient temperature | TA | In normal operation mode |  | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
|  |  | During flash memory programming |  | +10 to +40 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  |  | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |

Note Make sure that the following conditions of the VPP voltage application timing are satisfied when the flash memory is written.

- When supply voltage rises

VPP must exceed VDD $10 \mu$ s or more after VDD has reached the lower-limit value ( 1.8 V ) of the operating voltage range (see a in the figure below).

- When supply voltage drops

Vdd must be lowered $10 \mu \mathrm{~s}$ or more after VPP falls below the lower-limit value ( 1.8 V ) of the operating voltage range of $V_{D D}$ (see $\mathbf{b}$ in the figure below).


Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark
Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

Main System Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VdD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Ceramic resonator |  | Oscillation frequency (fx) ${ }^{\text {Note } 1}$ | $V_{\text {DD }}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | V dD $=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $V_{\text {do }}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | After Vod reaches oscillation voltage range MIN. |  |  | 4 | ms |
| Crystal resonator |  | Oscillation frequency (fx) ${ }^{\text {Note } 1}$ | $V_{\text {DD }}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | V DD $=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $V_{\text {DD }}=4.5$ to 5.5 V |  |  | 10 | ms |
|  |  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 30 | ms |
| External <br> clock |  | X1 input frequency (fx) ${ }^{\text {Note } 1}$ | $V_{D D}=4.5$ to 5.5 V | 1.0 |  | 10.0 | MHz |
|  |  |  | $\mathrm{V}_{\text {dD }}=3.0$ to 5.5 V | 1.0 |  | 6.0 | MHz |
|  |  |  | $\mathrm{V} \mathrm{DD}=1.8$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  |  | X1 input high-/low-level width ( $\mathrm{t} x \mathrm{H}, \mathrm{txL}$ ) | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 45 |  | 500 | ns |
|  |  |  | V DD $=3.0$ to 5.5 V | 75 |  | 500 | ns |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 85 |  | 500 | ns |
|  |  | X 1 input frequency (fx) ${ }^{\text {Note } 1}$ | V dD $=2.7$ to 5.5 V | 1.0 | $5.0$ |  | MHz |
|  |  | X1 input high-/low-level width ( $\mathrm{txh}, \mathrm{txL}$ ) | $V_{D D}=2.7$ to 5.5 V | 85 |  | 500 | ns |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the main system clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. When the main system clock is stopped and the device is operating on the subsystem clock, wait until the oscillation stabilization time has been secured by the program before switching back to the main system clock.

## Recommended Oscillator Constant ( $\mu$ PD78F9177A and 78F9177AY)

Ceramic resonator ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$ )

| Manufacturer | Part Number | $\begin{gathered} \text { Frequency } \\ (\mathrm{MHz}) \end{gathered}$ | Recommended Circuit Constant (pF) |  | Oscillation Voltage Range (Vdo) |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C1 | C2 | MIN. | MAX. |  |
| Murata Mfg. Co., Ltd. (Standard type) | CSBLA1M00J58-B0 | 1.000 | 150 | 150 | 2.4 | 5.5 | Without on-chip capacitor |
|  | CSBFB1M00J58-R1 |  |  |  |  |  |  |
|  | CSTCC2M00G56-R0 | 2.000 | - | - | 1.8 | 5.5 | With on-chip capacitor |
|  | CSTLS2M00G56-B0 |  |  |  |  |  |  |
|  | CSTCR4M00G53-R0 | 4.000 |  |  |  |  |  |
|  | CSTLS4M00G53-B0 |  |  |  | 1.9 | 5.5 |  |
|  | CSTCR4M19G53-R0 | 4.195 |  |  | 1.8 | 5.5 |  |
|  | CSTLS4M19G53-B0 |  |  |  | 1.9 | 5.5 |  |
|  | CSTCR4M91G53-R0 | 4.915 |  |  |  |  |  |
|  | CSTLS4M91G53-B0 |  |  |  | 2.1 | 5.5 |  |
|  | CSTCR5M00G53-R0 | 5.000 |  |  | 1.9 | 5.5 |  |
|  | CSTLS5M00G53-B0 |  |  |  | 2.1 | 5.5 |  |
|  | CSTCR6M00G53-R0 | 6.000 |  |  | 1.9 | 5.5 |  |
|  | CSTLS6M00G53-B0 |  |  |  | 2.1 | 5.5 |  |
|  | CSTCE8M00G52-R0 | 8.000 |  |  | 1.8 | 5.5 |  |
|  | CSTLS8M00G53-B0 |  |  |  | 2.0 | 5.5 |  |
|  | CSTCE8M38G52-R0 | 8.388 |  |  | 1.8 | 5.5 |  |
|  | CSTLS8M38G53-B0 |  |  |  | 2.0 | 5.5 |  |
|  | CSTCE10M0G52-R0 | 10.000 |  |  |  |  |  |
|  | CSTLS10M0G53-B0 |  |  |  | 2.2 | 5.5 |  |
| Murata Mfg. Co., Ltd. (Low-voltage drive type) | CSTLS4M00G53093-B0 | 4.000 | - | - | 1.8 | 5.5 | With on-chip capacitor |
|  | CSTLS4M19G53093-B0 | 4.195 |  |  |  |  |  |
|  | CSTCR4M91G53093-R0 | 4.915 |  |  |  |  |  |
|  | CSTLS4M91G53U-B0 |  |  |  |  |  |  |
|  | CSTCR5M00G53093-R0 | 5.000 |  |  |  |  |  |
|  | CSTLS5M00G53U-B0 |  |  |  |  |  |  |
|  | CSTCR6M00G53093-R0 | 6.000 |  |  |  |  |  |
|  | CSTLS6M00G53U-B0 |  |  |  |  |  |  |
|  | CSTLS8M00G53U-B0 | 8.000 |  |  |  |  |  |
|  | CSTLS8M38G53U-B0 | 8.388 |  |  |  |  |  |
|  | CSTLS10M0G53U-B0 | 10.000 |  |  |  |  |  |

Caution The oscillator constant is a reference value based on evaluation in specific environments by the resonator manufacturer. If the oscillator characteristics need to be optimized in the actual application, request the resonator manufacturer for evaluation on the implementation circuit.
Note that the oscillation voltage and oscillation frequency merely indicate the characteristics of the oscillator. Use the internal operation conditions of the $\mu$ PD78F9177A and 78F9177AY within the specifications of the DC and AC characteristics.

Recommended Oscillator Constant ( $\mu$ PD78F9177A(A) and 78F9177AY(A))

Ceramic resonator ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $+85^{\circ} \mathrm{C}$ )

| Manufacturer | Part Number | Frequency (MHz) | Recommended Circuit Constant (pF) |  | Oscillation Voltage Range (Vod) |  | Remarks |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | C1 | C2 | MIN. | MAX. |  |
| Murata Mfg. Co., Ltd. (Standard type) | CSTCC2M00G56A-R0 | 2.000 | - | - | 1.8 | 5.5 | With on-chip capacitor |
|  | CSTCR4MO0G53A-R0 | 4.000 |  |  |  |  |  |
|  | CSTCR4M19G53A-R0 | 4.195 |  |  |  |  |  |
|  | CSTCR4M91G53A-R0 | 4.915 |  |  | 1.9 | 5.5 |  |
|  | CSTCR5M00G53A-R0 | 5.000 |  |  |  |  |  |
|  | CSTCR6M00G53A-R0 | 6.000 |  |  |  |  |  |
|  | CSTCE8M00G52A-R0 | 8.000 |  |  | 1.8 | 5.5 |  |
|  | CSTCE8M38G52A-R0 | 8.388 |  |  |  |  |  |
|  | CSTCE10M0G52A-R0 | 10.000 |  |  | 2.0 | 5.5 |  |
| Murata Mfg. Co., Ltd. (Low-voltage drive type) | CSTCR4M91G53A093-R0 | 4.915 | - | - | 1.8 | 5.5 | With on-chip capacitor |
|  | CSTCR5M00G53A093-R0 | 5.000 |  |  |  |  |  |
|  | CSTCR6M00G53A093-R0 | 6.000 |  |  |  |  |  |

Caution The oscillator constant is a reference value based on evaluation in specific environments by the resonator manufacturer. If the oscillator characteristics need to be optimized in the actual application, request the resonator manufacturer for evaluation on the implementation circuit. Note that the oscillation voltage and oscillation frequency merely indicate the characteristics of the oscillator. Use the internal operation conditions of the $\mu$ PD78F9177A(A) and 78F9177AY(A) within the specifications of the DC and AC characteristics.

Subsystem Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal <br> resonator |  | $\text { Oscillation frequency (fxT) }{ }^{\text {Note } 1}$ |  | 32 | 32.768 | 35 | kHz |
|  |  |  |  |  |  |  |  |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V |  | 1.2 | 2 | s |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |  | 10 | s |
| External clock | XT1 XT2 | XT1 input frequency (fxT) ${ }^{\text {Note } 1}$ |  | 32 |  | 35 | kHz |
|  | $\Delta$ | XT1 input high-/low-level width (tхтн, tхтL) |  | 14.3 |  | 15.6 | $\mu \mathrm{s}$ |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation stabilization wait time.

Cautions 1. When using the subsystem clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. The subsystem clock oscillator is designed as a low-amplitude circuit for reducing current consumption, and is more prone to malfunction due to noise than the main system clock oscillator. Particular care is therefore required with the wiring method when the subsystem clock is used.

Remark For the resonator selection and oscillator constant, customers are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (1/2)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output current, high | Ioh | Per pin | $\mu$ PD78F9177A, $\mu$ PD78F9177AY |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -15 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{~A}(\mathrm{~A}), \\ & \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{AY}(\mathrm{~A}) \end{aligned}$ |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -11 | mA |
| Output current, low | IoL | Per pin | $\mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{~A}, \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{AY}$ |  |  | 10 | mA |
|  |  | Total for all pins |  |  |  | 80 | mA |
|  |  | Per pin | $\begin{aligned} & \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{~A}(\mathrm{~A}), \\ & \mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{AY}(\mathrm{~A}) \end{aligned}$ |  |  | 3 | mA |
|  |  | Total for all pins |  |  |  | 60 | mA |
| Input voltage, high | V $\mathrm{HH}_{1}$ | P00 to P05, P10, P11,P60 to P67 | $V_{\text {do }}=2.7$ to 5.5 V | 0.7 V do |  | VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 V DD |  | VDD | V |
|  | VIH2 | P50 to P53 | $\mathrm{V} D \mathrm{~L}=2.7$ to 5.5 V | 0.7 V DD |  | 12 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 V DD |  | 12 | V |
|  | VIH3 | $\begin{aligned} & \hline \text { RESET, } \\ & \text { P20 to P26, P30 } \\ & \text { to P33 } \\ & \hline \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.8 V do |  | VDD | V |
|  |  |  | $\mathrm{V} D \mathrm{D}=1.8$ to 5.5 V | 0.9 VDD |  | VDD | V |
|  | VIH4 | X1, X2, XT1, XT2 | $V_{D D}=4.5$ to 5.5 V | VDD -0.5 |  | VDD | V |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | $V_{D D}-0.1$ |  | VDD | V |
| Input voltage, low | VIL1 | P00 to P05, P10, P11, P60 to P67 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.3Vdd | V |
|  |  |  | $V_{\text {dD }}=1.8$ to 5.5 V | 0 |  | 0.1 Vdd | V |
|  | VIL2 | P50 to P53 | $V_{D D}=2.7$ to 5.5 V | 0 |  | 0.3 V DD | V |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\text {dD }}$ | V |
|  | VIL3 | $\begin{aligned} & \hline \overline{\text { RESET,P20 to }} \\ & \text { P26, P30 to P33 } \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.2Vdd | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  | VIL4 | X1, X2, XT1, XT2 | VDD $=4.5$ to 5.5 V | 0 |  | 0.4 | V |
|  |  |  | $\mathrm{V}_{\text {DD }}=1.8$ to 5.5 V | 0 |  | 0.1 | V |
| Output voltage, high | Vor | Pins other than P23, P24, P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , І I ( $=-1 \mathrm{~mA}$ | VdD - 1.0 |  |  | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V , $\mathrm{IOH}=-100 \mu \mathrm{~A}$ | VDD - 0.5 |  |  | V |
| Output voltage, low | Vol1 | Pins other than P50 to P53 | V DD $=4.5$ to 5.5 V , loL $=10 \mathrm{~mA}$ ( $\mu$ PD78F9177A, $\mu$ PD78F9177AY) $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , lol $=3 \mathrm{~mA}$ ( $\mu$ PD78F9177(A), $\mu$ PD78F9177AY(A)) |  |  | 1.0 | V |
|  |  |  | V DD $=1.8$ to 5.5 V , IOL $=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
|  | Vol2 | P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , loL $=10 \mathrm{~mA}$ ( $\mu$ PD78F9177A, $\mu$ PD78F9177AY) $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , lol $=3 \mathrm{~mA}$ ( $\mu$ PD78F9177(A), $\mu$ PD78F9177AY(A)) |  |  | 1.0 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V , $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 0.4 | V |
| Input leakage current, high | ILIH1 | $\mathrm{V}_{1}=\mathrm{V}_{\text {DD }}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  | 3 | $\mu \mathrm{A}$ |
|  | ILIH2 |  | X1, X2, XT1, XT2 |  |  | 20 | $\mu \mathrm{A}$ |
|  | ІІІн3 | $\mathrm{V}_{\mathrm{I}}=12 \mathrm{~V}$ | P50 to P53 (N-ch open drain) |  |  | 20 | $\mu \mathrm{A}$ |
| Input leakage current, low | ILLL1 | $\mathrm{V}_{1}=0 \mathrm{~V}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT2 |  |  | -3 | $\mu \mathrm{A}$ |
|  | ILIL2 |  | X1, X2, XT1, XT2 |  |  | -20 | $\mu \mathrm{A}$ |
|  | ILIL3 |  | P50 to P53 (N-ch open drain) |  |  | $-3^{\text {Note }}$ | $\mu \mathrm{A}$ |
| Output leakage current, high | ILOH | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  |  |  | 3 | $\mu \mathrm{A}$ |
| Output leakage current, low | Ilol | $\mathrm{Vo}=0 \mathrm{~V}$ |  |  |  | -3 | $\mu \mathrm{A}$ |
| Software pull-up resistor | R1 | VI $=0$ V, for pins other than P23, P24, and P50 to P53 |  | 50 | 100 | 200 | k $\Omega$ |

Note A low-level input leakage current of $-60 \mu \mathrm{~A}$ (MAX.) flows only during the 1-cycle time after a read instruction is executed to P50 to P53 and P50 to P53 are set to input mode. At times other than this, $-3 \mu \mathrm{~A}$ (MAX.) current flows.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (2/2)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply current | $\text { IdD1 }^{\text {Note } 1}$ | 10.0 MHz crystal oscillation operating mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 10.0 | 20.0 | mA |
|  |  | 6.0 MHz crystal oscillation operating mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 6.0 | 12.0 | mA |
|  |  | 5.0 MHz crystal oscillation operating mode$(\mathrm{C} 1=\mathrm{C} 2=22 \mathrm{pF})$ | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 5.0 | 10.0 | mA |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 1.2 | 2.5 | mA |
|  |  |  | $\text { VDD }=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 1.0 | 2.0 | mA |
|  | $\text { IDD2 }^{\text {Note } 1}$ | 10.0 MHz crystal oscillation HALT mode | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 1.2 | 6.0 | mA |
|  |  | 6.0 MHz crystal oscillation HALT mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 0.9 | 2.8 | mA |
|  |  | 5.0 MHz crystal oscillation HALT mode (C1 = $\mathrm{C} 2=22 \mathrm{pF}$ ) | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 0.8 | 2.5 | mA |
|  |  |  | $\mathrm{V} D \mathrm{D}=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 0.4 | 2.0 | mA |
|  |  |  | $\mathrm{V} D \mathrm{D}=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 0.25 | 1.5 | mA |
|  | $\mathrm{IDD3}^{\text {Note } 1}$ | 32.768 kHz crystal oscillation operating Note 3 mode$\begin{aligned} & (\mathrm{C} 3=\mathrm{C} 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | VDD $=5.0 \mathrm{~V} \pm 10 \%$ |  | 100 | 320 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=3.0 \mathrm{~V} \pm 10 \%$ |  | 80 | 240 | $\mu \mathrm{A}$ |
|  |  |  | $V D D=2.0 \mathrm{~V} \pm 10 \%$ |  | 65 | 210 | $\mu \mathrm{A}$ |
|  | ${ }_{\text {IDD4 }}$ Note 1 | 32.768 kHz crystal oscillation HALT mode <br> Note 3 $\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 18 | 120 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=3.0 \mathrm{~V} \pm 10 \%$ |  | 5.0 | 50 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=2.0 \mathrm{~V} \pm 10 \%$ |  | 2.5 | 30 | $\mu \mathrm{A}$ |
|  | ${ }_{\text {IDD5 }}$ Note 1 | 32.768 kHz crystal stop STOP mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 0.1 | 30 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=3.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 10 | $\mu \mathrm{A}$ |
|  |  |  | $V \mathrm{DD}=2.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 10 | $\mu \mathrm{A}$ |
|  | $\begin{array}{\|l\|l\|} \hline \text { Note 2 } \end{array}$ | 10.0 MHz crystal oscillation A/D operating mode | V DD $=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 10.8 | 22.0 | mA |
|  |  | 6.0 MHz crystal oscillation A/D operating mode | $\mathrm{V} D \mathrm{LD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 6.8 | 14.0 | mA |
|  |  | 5.0 MHz crystal oscillation A/D operating mode$(\mathrm{C} 1=\mathrm{C} 2=22 \mathrm{pF})$ | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 5.8 | 12.0 | mA |
|  |  |  | $V_{D D}=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 2.0 | 4.5 | mA |
|  |  |  | $V_{D D}=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 1.8 | 4.0 | mA |

Notes 1. The $A V_{\text {ref }} O N$ (ADCSO (bit 7 of $A D M O ; A / D$ converter mode register 0 ) $=1$ ), $A V D d$, and the port current (including the current flowing through the internal pull-up resistors) are not included.
2. The $A V_{\text {refo }} O N(A D C S O=1)$ and port current (including the current flowing through the internal pull-up resistors) are not included. Refer to the A/D converter characteristics for the current flowing through AVref.
3. When the main system clock is stopped.
4. During high-speed mode operation (when the processor clock control register (PCC) is set to 00 H .)
5. During low-speed mode operation (when PCC is set to 02 H )

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## AC Characteristics

(1) Basic operation ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Cycle time (minimum instruction execution time) | Tcy | Operation based on the main system clock | $\mathrm{V}_{\text {DD }}=4.5$ to 5.5 V | 0.2 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $\mathrm{V}_{\text {DD }}=3.0$ to 5.5 V | 0.333 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $V_{\text {DD }}=2.7$ to 5.5 V | 0.4 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 1.6 |  | 8 | $\mu \mathrm{s}$ |
|  |  | Operation based on the subsystem clock |  | 114 | 122 | 125 | $\mu \mathrm{s}$ |
| TI80 and TI81 input frequency | $\mathrm{f}_{\mathrm{T}}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 0 |  | 4 | MHz |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 0 |  | 275 | kHz |
| TI80 and TI81 input high-/low-level width | ttin, till | $V_{\text {DD }}=2.7$ to 5.5 V |  | 0.1 |  |  | $\mu \mathrm{s}$ |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 1.8 |  |  | $\mu \mathrm{s}$ |
| Interrupt input high-/low-level width | tinth, tintl | INTP0 to INTP3 |  | 10 |  |  | $\mu \mathrm{S}$ |
| $\overline{\text { RESET }}$ input lowlevel width | trsL |  |  | 10 |  |  | $\mu \mathrm{s}$ |
| CPT90 input high-/low-level width | tcPe, <br> tcPL |  |  | 10 |  |  | $\mu \mathrm{s}$ |

Tcy vs. VdD (main system clock)


CHAPTER 27 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177A, 78F9177AY, 78F9177A(A), 78F9177AY(A))
(2) Serial interface $\operatorname{SIO20}\left(\mathrm{T}_{\mathrm{A}}=-40\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V$)$
(a) 3-wire serial I/O mode (SCK20...Internal clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy1 | $V_{\text {do }}=2.7$ to 5.5 V |  | 800 |  |  | ns |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  | 3200 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/lowlevel width | $\mathrm{t}_{\text {KH1 }}$, tKL1 | $V_{D D}=2.7$ to 5.5 V |  | $\mathrm{tkcyı}_{1} / 2-50$ |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | tkcrı $/ 2-150$ |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik1 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 150 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 500 |  |  | ns |
| SI20 hold time (from $\overline{\text { SCK20 } \uparrow \text { ) }}$ | tksI1 | $V_{D D}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tKsO1 | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega, \\ & \mathrm{C}=100 \mathrm{pF}^{\text {Note }} \end{aligned}$ | $V_{\text {DD }}=2.7$ to 5.5 V | 0 |  | 250 | ns |
|  |  |  | $V_{\text {DD }}=1.8$ to 5.5 V | 0 |  | 1000 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(b) 3-wire serial I/O mode (SCK20...External clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy2 | $V_{D D}=2.7$ to 5.5 V |  | 900 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 3500 |  |  | ns |
| SCK20 high-/lowlevel width | tKH2, $\mathrm{tkL2}^{\text {2 }}$ | $V_{D D}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 1600 |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik2 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 100 |  |  | ns |
|  |  | $\mathrm{V} D \mathrm{D}=1.8$ to 5.5 V |  | 150 |  |  | ns |
| SI20 hold time (from SCK20 $\uparrow$ ) | tksı2 | $V_{\text {dD }}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tksO2 | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega, \\ & \mathrm{C}=100 \mathrm{pF}^{\text {Note }} \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 300 | ns |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 0 |  | 1000 | ns |
| SO20 setup time <br> (when using $\overline{\text { SS20 }}$, to $\overline{\mathrm{SS} 20} \downarrow$ ) | tKAS2 | $V_{\text {dD }}=2.7$ to 5.5 V |  |  |  | 120 | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  |  |  | 400 | ns |
| SO20 disable time (when using $\overline{\mathrm{SS20}}$, from $\overline{\mathrm{SS} 2 \mathrm{O}} \uparrow$ ) | tkDS2 | $\mathrm{V}_{\text {do }}=2.7$ to 5.5 V |  |  |  | 240 | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  |  |  | 800 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(c) UART mode (dedicated baud rate generator output)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. |
| :---: | :---: | :--- | :--- | :---: | :---: |
| Transfer rate |  | VDD $=2.7$ to 5.5 V |  |  | 78125 |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | bps |  |

(d) UART mode (external clock input)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASCK20 cycle time | tксуз | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 900 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 3500 |  |  | ns |
| ASCK20 high-/low- <br> level width | tкнз, ткıз | VDD $=2.7$ to 5.5 V | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 1600 |  |  | ns |
| Transfer rate |  | V DD $=2.7$ to 5.5 V |  |  | 39063 | bps |
|  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 9766 | bps |
| ASCK20 rise time, fall time | $\mathrm{t}_{\mathrm{R}, \mathrm{t}} \mathrm{t}$ |  |  |  | 1 | $\mu \mathrm{s}$ |

(3) Serial interface SMB0 ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VdD}=1.8$ to 5.5 V$)(\mu \mathrm{PD} 78 \mathrm{F9177AY}, 78 \mathrm{F9177AY}(\mathrm{~A})$ only)
(a) DC characteristics

| Parameter | Symbol |  | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage, high | $\mathrm{V}_{\mathrm{H}}$ | SCLO, SDAO (at hysteresis) |  | 0.8 VdD |  | VDD | V |
| Input voltage, low | VIL | SCLO, SDAO (at hysteresis) |  | 0 |  | 0.2 Vdo | V |
| Output voltage, low | VoL | SCLO, SDAO | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , $\mathrm{loL}=10 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{V} \mathrm{DD}=1.8$ to 5.5 V , $\mathrm{loL}=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
| Input leakage current, high | ІІІ | SCLO, SDAO | $V_{1}=V_{D D}$ |  |  | 3 | $\mu \mathrm{A}$ |
| Input leakage current, low | ILIL | SCLO, SDAO | $\mathrm{V},=0 \mathrm{~V}$ |  |  | -3 | $\mu \mathrm{A}$ |

(b) DC characteristics (when using comparator)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input range | Vsda, <br> VscL | $V_{D D}=1.8$ to 5.5 V | 0 |  | 5.5 | V |
| Transfer level | VISDA, <br> Viscl | $4.5 \leq \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ | $0.72 \mathrm{~V}_{\text {Ism }}$ | VIsmb | 1.28 V ISM | V |
|  |  | $3.3 \leq \mathrm{VDD}^{<} 4.5 \mathrm{~V}$ | 0.78 V Ism | $V_{\text {Ismb }}$ | $1.22 \mathrm{~V}_{\text {IsM }}$ | V |
|  |  | $2.7 \leq \mathrm{VDD}^{2} 3.3 \mathrm{~V}$ | 0.75 V Ismb | Vismb | 1.25 V Ismb | V |
|  |  | $1.8 \leq \mathrm{VDD}^{2} 2.7 \mathrm{~V}$ | 0.90 V Ismb | $V \mathrm{I}$ mm | 1.45 V Ism | V |
| Input level threshold value ${ }^{\text {Note }}$ | VIsmb | LVL01, LVL00 = 0, 1 |  | $0.25 \times V_{\text {DD }}$ |  | V |
|  |  | LVL01, LVL00 = 1, 0 |  | $0.375 \times \mathrm{VDO}$ |  | V |
|  |  | LVL01, LVL00 = 1, 1 |  | $0.5 \times \mathrm{VDD}$ |  | V |

Note Vismb is an input level threshold value selected by bits LVLOO and LVL01 (bits 0 and 1 of SMB input level setting register 0 (SMBVIO)).
According to the SMB standard (V1.1), the maximum value of low-level input voltage is 0.8 V , and the minimum value of high-level input voltage, 2.1 V . To satisfy these conditions, set LVL01 and LVL00 as follows;

- When $V_{D D}=1.8$ to 3.3 V : LVL01, LVLOO $=1,1(0.5 \times \mathrm{VDD})$
- When Vdd $=3.3$ to 4.5 V : LVL01, LVL00 $=1,0(0.375 \times \mathrm{V} D)$
- When Vdd $=4.5$ to 5.5 V : LVL01, LVLOO $=0,1(0.25 \times \mathrm{VDD})$
"LVL01, LVL00 $=0,0$ " is not available since this setting does not satisfy the SMB standard (V1.1).
(c) AC characteristics

| Parameter |  | Symbol | SMB Mode |  | Standard Mode $I^{2} \mathrm{C}$ Bus |  | High-speed Mode $I^{2} \mathrm{C}$ Bus |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | MAX. | MIN. | MAX. | MIN. | MAX. |  |
| SCL0 clock frequency |  |  | fclk | 10 | 100 | 0 | 100 | 0 | 400 | kHz |
| Bus free time <br> (between stop and start condition) |  | tbuF | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| Hold time ${ }^{\text {Note } 1}$ |  | thd:STA | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Start/restart condition setup time |  | tsu:STA | 4.7 | - | 4.7 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Stop condition setup time |  | tsu:sto | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Data hold time | When using CBUScompatible master | thd:dat | - | - | 5 | - | - | - | $\mu \mathrm{s}$ |
|  | When using SMB/IIC bus |  | 300 | - | $0{ }^{\text {Note } 2}$ | - | $0^{\text {Note } 2}$ | $900^{\text {Note } 3}$ | ns |
| Data setup time |  | tsu:DAT | 250 | - | 250 | - | $100{ }^{\text {Note } 4}$ | - | ns |
| SCLO clock low-level width |  | tıow | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| SCLO clock high-level width |  | thigh | 4.0 | 50 | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| SCL0 and SDA0 signal fall time |  | $\mathrm{tF}_{F}$ | - | 300 | - | 300 | - | 300 | ns |
| SCL0 and SDA0 signal rise time |  | tr | - | 1000 | - | 1000 | - | 300 | ns |
| Spike pulse width controlled by input filter |  | tsp | - | - | - | - | 0 | 50 | ns |
| Timeout |  | ttimeout | 25 | 35 | - | - | - | - | ms |
| Total extended time of SCLO clock low-level period (slave) |  | tıow:SExt | - | 25 | - | - | - | - | ms |
| Total extended time of cumulative clock low-level period (master) |  | tıow:mext | - | 10 | - | - | - | - | ms |
| Capacitive load per each bus line |  | Cb | - | - | - | 400 | - | 400 | pF |

Notes 1. In the start condition, the first clock pulse is generated after this hold time.
2. To fill in the undefined area of the SCLO falling edge, it is necessary for the device to internally provide at least 300 ns of hold time for the SDA0 signal (which is VIHmin. of the SCLO signal).
3. If the device does not extend the SCLO signal low hold time (thow), only maximum data hold time thd:DAT needs to be fulfilled.
4. The high-speed mode $I^{2} C$ bus is available in the $S M B$ mode and the standard mode $I^{2} C$ bus system. At this time, the conditions described below must be satisfied.

- If the device extends the SCLO signal low state hold time
tsu:DAt $\geq 250 \mathrm{~ns}$
- If the device extends the SCLO signal low state hold time

Be sure to transmit the next data bit to the SDAO line before the SCLO line is released (trmax.+ tsu:DAT $=1000+250=1250 \mathrm{~ns}$ by the SMB mode or the standard mode $\mathrm{I}^{2} \mathrm{C}$ bus specification).

AC Timing Measurement Points (Excluding X1 and XT1 Inputs)


## Clock Timing



## TI Timing



Interrupt Input Timing


RESET Input Timing


CPT90 Input Timing


## Serial Transfer Timing

## 3-wire serial I/O mode:



Remark $m=1,2$

3-wire serial I/O mode (when using SS20):


UART mode (external clock input):


## SMB mode:



10-Bit A/D Converter Characteristics ( $\mathrm{TA}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, 1.8 \leq \mathrm{AV}$ REF $\leq \mathrm{AVDD}=\mathrm{VDD} \leq 5.5 \mathrm{~V}, \mathrm{AV}$ Ss $=\mathrm{V} \mathrm{SS}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 10 | 10 | 10 | bit |
| $\text { Overall error }{ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDO} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.2$ | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\mathrm{REF}} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.4$ | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\mathrm{REF}} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.8$ | $\pm 1.2$ | \%FSR |
| Conversion time | tconv | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\mathrm{REF}} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ | 12 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ | 14 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ | 28 |  | 100 | $\mu \mathrm{s}$ |
| $\text { Zero-scale error }{ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDO}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| $\text { Full-scale error }{ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {D }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| Integral linearity error ${ }^{\text {Note }}$ | INL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{A} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 2.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 4.5$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 8.5$ | LSB |
| Differential linearity error | DNL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 2.0$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 3.5$ | LSB |
| Analog input voltage | $V_{\text {IAN }}$ |  | 0 |  | $\mathrm{AV}_{\text {REF }}$ | V |
| Reference voltage | $\mathrm{AV}_{\text {REF }}$ |  | 1.8 |  | AV ${ }_{\text {dD }}$ | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{\text {ss }}$ | Radref |  | 20 | 40 |  | $\mathrm{k} \Omega$ |

Note Excludes quantization error ( $\pm 0.05 \%$ FSR $)$.

Remark FSR: Full scale range

Flash Memory Write/Erase Characteristics ( $\mathrm{T}_{\mathrm{A}}=10$ to $40^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Write current (VdD pin) ${ }^{\text {Note } 1}$ | Idow | When VPP supply voltage $=\mathrm{V}_{\mathrm{PP} 1}$ (5.0 MHz operation) |  |  | 23 | mA |
| Write current (VPp pin) | IPPW | When VPP supply voltage $=\mathrm{V}_{\text {PP1 }}$ |  |  | 20 | mA |
| Erase current $(\text { Vod pin })^{\text {Note } 1}$ | IdDE | When VPP supply voltage $=\mathrm{V}_{\text {PP } 1}$ (5.0 MHz operation) |  |  | 23 | mA |
| Erase current (Vpp pin) | IPPE | When VPP supply voltage $=$ VPP1 |  |  | 100 | mA |
| Unit erase time ${ }^{\text {Note } 2}$ | ter |  | 0.2 | 0.2 | 0.2 | s |
| Total erase time | tera |  |  |  | 20 | S |
| Write count ${ }^{\text {Note } 3}$ |  | Erase/write is regarded as 1 cycle | 20 | 20 | 20 | Times |
| VPP supply voltage | Vppo | In normal operation | 0 |  | 0.2VDD | V |
|  | VPP1 | During flash memory programming | 9.7 | 10.0 | 10.3 | V |

Notes 1. The current flowing to the ports (including the current flowing through an on-chip pull-up resistor) and AVdd current are not included.
2. The prewrite time before erasure and the erase verify time (writeback time) is not included.
3. When a product is first written after shipment, "erase $\rightarrow$ write" is taken as one rewrite.

Data Memory STOP Mode Low Power Supply Voltage Data Retention Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $\mathbf{+ 8 5 ^ { \circ }} \mathbf{C}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data retention power supply voltage | Voddr |  | 1.8 |  | 5.5 | V |
| Release signal set time | tsrel |  | 0 |  |  | $\mu \mathrm{s}$ |
| Oscillation stabilization wait time ${ }^{\text {Note } 1}$ | twalt | Release by RESET |  | $2^{15} / \mathrm{fx}$ |  | S |
|  |  | Release by interrupt request |  | Note 2 |  | S |

Notes 1. The oscillation stabilization time is the time the CPU operation is stopped to prevent unstable operation when oscillation starts.
2. By using bits 0 to 2 (OSTS0 to OSTS2) of the oscillation stabilization time selection register (OSTS), $2^{12} / f x, 2^{15} / f x$, or $2^{17} / f x$ can be selected.

Remark fx: Main system clock oscillation frequency

Data Retention Timing (STOP Mode Release by RESET)


Data Retention Timing (Standby Release Signal: STOP Mode Release by Interrupt Signal)


## CHAPTER 28 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177, 78F9177Y)

Absolute Maximum Ratings $\left(\mathrm{T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right)$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Supply voltage | VDD | $\begin{aligned} & A V_{D D}-0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\mathrm{REF}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\mathrm{REF}} \leq \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \end{aligned}$ | -0.3 to +6.5 | V |
|  | AV ${ }_{\text {dD }}$ |  |  | V |
|  | $\mathrm{AV}_{\text {ref }}$ |  |  | V |
|  | Vpp | Note | -0.3 to +10.5 | V |
| Input voltage | $V_{11}$ | Pins other than P50 to P53, P23, P24 | -0.3 to VDD +0.3 | V |
|  | $\mathrm{V}_{12}$ | P23, P24 | -0.3 to +5.5 | V |
|  | $\mathrm{V}_{13}$ | P50 to P53 | -0.3 to +13 | V |
| Output voltage | Vo |  | -0.3 to VDD +0.3 | V |
| Output current, high | Іон | Per pin | -10 | mA |
|  |  | Total for all pins | -30 | mA |
| Output current, low | loL | Per pin | 30 | mA |
|  |  | Total for all pins | 160 | mA |
| Operating ambient temperature | TA | In normal operation mode | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
|  |  | During flash memory programming | +10 to +40 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |

Note Make sure that the following conditions of the VPP voltage application timing are satisfied when the flash memory is written.

- When supply voltage rises

VPP must exceed VDD $10 \mu$ s or more after VDD has reached the lower-limit value ( 1.8 V ) of the operating voltage range (see a in the figure below).

- When supply voltage drops
$V_{D D}$ must be lowered $10 \mu \mathrm{~s}$ or more after VPP falls below the lower-limit value ( 1.8 V ) of the operating voltage range of Vdo (see bin the figure below).


Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

Main System Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$, $\mathrm{VdD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Ceramic resonator | $\begin{array}{\|lll\|} \hline \text { Vsso } & \text { X1 } & \text { X2 } \\ \hline \end{array}$ | $\text { Oscillation frequency (fx) }{ }^{\text {Note } 1}$ | $\mathrm{V}_{\mathrm{DD}}=$ oscillation voltage range | 1.0 |  | 5.0 | MHz |
|  | $\begin{aligned} & \text { C1 } 1=2=1 \\ & \text { m- } \end{aligned}$ | Oscillation stabilization time ${ }^{\text {Note } 2}$ | After Vod reaches oscillation voltage range MIN. |  |  | 4 | ms |
| Crystal resonator |  | $\text { Oscillation frequency (fx) }{ }^{\text {Note } 1}$ |  | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V |  |  | 10 | ms |
|  |  |  | $V_{D D}=1.8$ to 5.5 V |  |  | 30 | ms |
| External <br> clock | X2 | X 1 input frequency (fx) ${ }^{\text {Note } 1}$ |  | 1.0 |  | 5.0 | MHz |
|  |  | X1 input high-llow-level width (txh, txL) |  | 85 |  | 500 | ns |
|  | $\mathrm{X} 1 \quad \mathrm{X} 2$ | X1 input frequency (fx) ${ }^{\text {Note } 1}$ | $V_{D D}=2.7$ to 5.5 V | 1.0 |  | 5.0 | MHz |
|  | $\triangle \quad \text { OPEN }$ | X1 input high-/low-level width (txh, txL) | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 85 |  | 500 | ns |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the main system clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. When the main system clock is stopped and the device is operating on the subsystem clock, wait until the oscillation stabilization time has been secured by the program before switching back to the main system clock.

Remark For the resonator selection and oscillator constant, customers are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

Subsystem Clock Oscillator Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal resonator |  | $\text { Oscillation frequency (fxT) }{ }^{\text {Note } 1}$ |  | 32 | 32.768 | 35 | kHz |
|  |  |  |  |  |  |  |  |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V |  | 1.2 | 2 | s |
|  |  |  | $\mathrm{V} D \mathrm{D}=1.8$ to 5.5 V |  |  | 10 | s |
| External clock | XT1 XT 2 | XT1 input frequency (fxT) ${ }^{\text {Note } 1}$ |  | 32 | - | 35 | kHz |
|  | $\wedge$ | XT1 input high-/low-level width ( $\mathbf{x т н , ~ t х т L ) ~}$ |  | 14.3 |  | 15.6 | $\mu \mathrm{s}$ |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation stabilization wait time.

Cautions 1. When using the subsystem clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. The subsystem clock oscillator is designed as a low-amplitude circuit for reducing current consumption, and is more prone to malfunction due to noise than the main system clock oscillator. Particular care is therefore required with the wiring method when the subsystem clock is used.

Remark For the resonator selection and oscillator constant, customers are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (1/2)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output current, high | Іон | Per pin |  |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -15 | mA |
| Output current, low | IoL | Per pin |  |  |  | 10 | mA |
|  |  | Total for all pins |  |  |  | 80 | mA |
| Input voltage, high | $\mathrm{V}_{\mathbf{H + 1}}$ | P00 to P05, P10, P11,P60 to P67 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.7 VDD |  | VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 VDD |  | VDD | V |
|  | $\mathrm{V}_{1+2}$ | P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | $0.7 \mathrm{~V}_{\mathrm{DD}}$ |  | 12 | V |
|  |  |  | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=1.8 \text { to } 5.5 \mathrm{~V}, \\ & T_{\mathrm{A}}=25 \text { to }+85^{\circ} \mathrm{C} \\ & \hline \end{aligned}$ | 0.9 VDD |  | 12 | V |
|  | $\mathrm{V}_{1+3}$ | $\begin{aligned} & \hline \text { RESET, } \\ & \text { P20 to P26, P30 } \end{aligned}$to P33 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0.8 V DD |  | VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0.9 VDD |  | VDD | V |
|  | $\mathrm{V}_{1+4}$ | $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1, \mathrm{XT} 2$ | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | VDD - 0.5 |  | VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | $V_{D D}-0.1$ |  | VDD | V |
| Input voltage, low | VIL1 | $\begin{aligned} & \text { P00 to P05, P10, } \\ & \text { P11, P60 to P67 } \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.3 VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  | VIL2 | P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.3 VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  | Vıı3 | RESET,P20 toP26, P30 to P33 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 0.2 VDD | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | $0.1 \mathrm{~V}_{\mathrm{DD}}$ | V |
|  | VIL4 | X1, X2, XT1, XT2 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V | 0 |  | 0.4 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | 0.1 | V |
| Output voltage, high | Vон | Pins other than P23, P24, P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , $\mathrm{I} \mathrm{OH}=-1 \mathrm{~mA}$ | $\mathrm{V}_{\text {DD }}-1.0$ |  |  | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V, І-H $=-100 \mu \mathrm{~A}$ | VDD - 0.5 |  |  | V |
| Output voltage, low | Vol1 | Pins other than P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , $\mathrm{loL}=10 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V , loL $=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
|  | Vol2 | P50 to P53 | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V , $\mathrm{loL}=10 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V , $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 0.4 | V |
| Input leakage current, high | Іıн1 | $\mathrm{V}_{\mathrm{I}}=\mathrm{V}_{\mathrm{DD}}$ | Pins other than P50 to P53 ( N -ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT2 |  |  | 3 | $\mu \mathrm{A}$ |
|  | lıн2 |  | X1, X2, XT1, XT2 |  |  | 20 | $\mu \mathrm{A}$ |
|  | ІІнз | $\mathrm{V}_{1}=12 \mathrm{~V}$ | P50 to P53 (N-ch open drain) |  |  | 20 | $\mu \mathrm{A}$ |
| Input leakage current, low | ILLL1 | $\mathrm{V}_{1}=0 \mathrm{~V}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT2 |  |  | -3 | $\mu \mathrm{A}$ |
|  | İL2 |  | X1, X2, XT1, XT2 |  |  | -20 | $\mu \mathrm{A}$ |
|  | Iıレ3 |  | P50 to P53 (N-ch open drain) |  |  | $-3^{\text {Note }}$ | $\mu \mathrm{A}$ |
| Output leakage current, high | ILoн | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  |  |  | 3 | $\mu \mathrm{A}$ |
| Output leakage current, low | ILoL | V o $=0 \mathrm{~V}$ |  |  |  | -3 | $\mu \mathrm{A}$ |
| Software pull-up resistor | R1 | $\mathrm{V}_{1}=0 \mathrm{~V}$, for pins other than P23, P24, and P50 to P53 |  | 50 | 100 | 200 | $\mathrm{k} \Omega$ |

Note A low-level input leakage current of $-60 \mu \mathrm{~A}$ (MAX.) flows only during the 1 -cycle time after a read instruction is executed to P50 to P53 and P50 to P53 are set to input mode. At times other than this, a -3 $\mu \mathrm{A}$ (MAX.) current flows.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

DC Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V ) (2/2)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply current | $\text { IDD } 1^{\text {Note } 1}$ | 5.0 MHz crystal oscillation operating mode$(\mathrm{C} 1=\mathrm{C} 2=22 \mathrm{pF})$ | VDD $=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 5.0 | 15.0 | mA |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 2.0 | 5.0 | mA |
|  |  |  | $V_{D D}=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 1.5 | 3.0 | mA |
|  | $\mathrm{IDD2}^{\text {Note }} 1$ | 5.0 MHz crystal oscillation HALT mode (C1 = C2 = 22 pF ) | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 2.0 | 6.0 | mA |
|  |  |  | $\mathrm{V} D \mathrm{~L}=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 1.0 | 2.5 | mA |
|  |  |  | $V_{D D}=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 0.75 | 1.5 | mA |
|  | ${ }_{\text {ldD3 }}$ Note 1 | 32.768 kHz crystal oscillation operating Note 3 mode$\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%$ |  | 250 | 750 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%$ |  | 200 | 600 | $\mu \mathrm{A}$ |
|  |  |  | $V D D=2.0 \mathrm{~V} \pm 10 \%$ |  | 150 | 450 | $\mu \mathrm{A}$ |
|  | $\text { IDD } 4^{\text {Note } 1}$ | 32.768 kHz crystal oscillation HALT mode <br> Note 3 $\begin{aligned} & (\mathrm{C} 3=\mathrm{C} 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 50 | 150 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{VDD}=3.0 \mathrm{~V} \pm 10 \%$ |  | 30 | 90 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{V} D \mathrm{D}=2.0 \mathrm{~V} \pm 10 \%$ |  | 20 | 60 | $\mu \mathrm{A}$ |
|  | ${ }_{\text {IDD5 }}$ Note 1 | 32.768 kHz crystal stop STOP mode | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%$ |  | 0.1 | 30 | $\mu \mathrm{A}$ |
|  |  |  | VDD $=3.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 10 | $\mu \mathrm{A}$ |
|  |  |  | $\mathrm{VDD}=2.0 \mathrm{~V} \pm 10 \%$ |  | 0.05 | 10 | $\mu \mathrm{A}$ |
|  | ${ }_{\text {IDD6 }}$ Note 2 | 5.0 MHz crystal oscillation A/D operating mode$(\mathrm{C} 1=\mathrm{C} 2=22 \mathrm{pF})$ | $\mathrm{VDD}=5.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 4}$ |  | 6.0 | 17.0 | mA |
|  |  |  | V DD $=3.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 3.0 | 7.0 | mA |
|  |  |  | $\mathrm{VDD}=2.0 \mathrm{~V} \pm 10 \%{ }^{\text {Note } 5}$ |  | 2.5 | 5.0 | mA |

Notes 1. The $A V_{\text {ref }} O N$ ( $A D C S 0$ (bit 7 of $A D M 0 ; A / D$ converter mode register 0 ) $=1$ ), $A V D D$, and the port current (including the current flowing through the internal pull-up resistors) are not included.
2. The $A V_{\text {refe }} O N(A D C S O=1)$ and port current (including the current flowing through the internal pull-up resistors) are not included. Refer to the A/D converter characteristics for the current flowing through $A V_{\text {ref. }}$
3. When the main system clock is stopped.
4. During high-speed mode operation (when the processor clock control register (PCC) is set to 00 H .)
5. During low-speed mode operation (when PCC is set to 02 H )

Remark
Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## AC Characteristics

(1) Basic operation ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V )

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Cycle time (minimum instruction execution time) | Tcy | Operation based on the main system clock | $\mathrm{V}_{\text {DD }}=2.7$ to 5.5 V | 0.4 |  | 8 | $\mu \mathrm{s}$ |
|  |  |  | $V_{D D}=1.8$ to 5.5 V | 1.6 |  | 8 | $\mu \mathrm{s}$ |
|  |  | Operation based on the subsystem clock |  | 114 | 122 | 125 | $\mu \mathrm{s}$ |
| TI80 and TI81 input frequency | $\mathrm{f}_{\mathrm{T}}$ | $V_{\text {DD }}=2.7$ to 5.5 V |  | 0 |  | 4 | MHz |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 0 |  | 275 | kHz |
| TI80 and TI81 input high-/low-level width | ttin, ttil | $V_{D D}=2.7$ to 5.5 V |  | 0.1 |  |  | $\mu \mathrm{S}$ |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 1.8 |  |  | $\mu \mathrm{S}$ |
| Interrupt input high-/low-level width | tinth, tintl | INTP0 to INTP3 |  | 10 |  |  | $\mu \mathrm{s}$ |
| RESET input lowlevel width | trsL |  |  | 10 |  |  | $\mu \mathrm{S}$ |
| CPT90 input high-/low-level width | tcPH, tcPL |  |  | 10 |  |  | $\mu \mathrm{s}$ |

Tcy vs. Vdd (main system clock)

(2) Serial interface $\mathrm{SIO20}\left(\mathrm{~T}_{\mathrm{A}}=-40\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V$)$
(a) 3-wire serial I/O mode ( $\overline{\text { SCK20 }}$...Internal clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | $\mathrm{tkcy}_{1}$ | $V_{D D}=2.7$ to 5.5 V |  | 800 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 3200 |  |  | ns |
| $\overline{\text { SCK20 high-/low- }}$ level width | $\mathrm{t}_{\text {KH1 }}$, tkL1 | $V_{D D}=2.7$ to 5.5 V |  | $\mathrm{tkcrı}_{1} / 2-50$ |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | $\mathrm{tkcy1}_{1} / 2-150$ |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik1 | $V_{D D}=2.7$ to 5.5 V |  | 150 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 500 |  |  | ns |
| SI20 hold time (from $\overline{\text { SCK20 } \uparrow \text { ) }}$ | tksı11 | $V_{D D}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from $\overline{\text { SCK20 }} \downarrow$ | tKsO1 | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega, \\ & \mathrm{C}=100 \mathrm{pF}^{\text {Note }} \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 250 | $n s$ |
|  |  |  | $V_{\text {DD }}=1.8$ to 5.5 V | 0 |  | 1000 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(b) 3-wire serial I/O mode (SCK20...External clock)

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy2 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 900 |  |  | ns |
|  |  | $V_{\text {DD }}=1.8$ to 5.5 V |  | 3500 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/low- <br> level width | tкH2, tкL2 | $V_{D D}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 1600 |  |  | ns |
| $\begin{aligned} & \text { SI20 setup time } \\ & \text { (to } \overline{\text { SCK20 } \uparrow \text { ) }} \end{aligned}$ | tsikz | VDD $=2.7$ to 5.5 V |  | 100 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 150 |  |  | ns |
| SI20 hold time <br> (from SCK20 $\uparrow$ ) | tksi2 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  | 400 |  |  | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  | 600 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tksoz | $\begin{aligned} & \mathrm{R}=1 \mathrm{k} \Omega, \\ & \mathrm{C}=100 \mathrm{pF} \end{aligned}$ | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V | 0 |  | 300 | ns |
|  |  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V | 0 |  | 1000 | ns |
| SO20 setup time (when using SS20, to $\overline{\mathrm{SS} 20} \downarrow$ ) | tkas2 | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  |  |  | 120 | ns |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |  |  | 400 | ns |
| SO20 disable time (when using $\overline{\text { SS20 }}$, from $\overline{\mathrm{SS} 20} \uparrow$ ) | tkDS2 | V DD $=2.7$ to 5.5 V |  |  |  | 240 | ns |
|  |  | V DD $=1.8$ to 5.5 V |  |  |  | 800 | ns |

Note R and C are the load resistance and load capacitance of the SO20 output line.
(c) UART mode (dedicated baud rate generator output)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :--- | :---: | :---: | :---: | :---: |
| Transfer rate |  | $\mathrm{VDD}_{\mathrm{DD}}=2.7$ to 5.5 V |  |  | 78125 | bps |
|  |  | $\mathrm{VDD}=1.8$ to 5.5 V |  |  | 19531 | bps |

(d) UART mode (external clock input)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASCK20 cycle time | tксуз | V DD $=2.7$ to 5.5 V | 900 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 3500 |  |  | ns |
| ASCK20 high-/lowlevel width | tкн3, tкL3 | V DD $=2.7$ to 5.5 V | 400 |  |  | ns |
|  |  | $V_{D D}=1.8$ to 5.5 V | 1600 |  |  | ns |
| Transfer rate |  | $\mathrm{V}_{\mathrm{DD}}=2.7$ to 5.5 V |  |  | 39063 | bps |
|  |  | $\mathrm{V}_{\mathrm{DD}}=1.8$ to 5.5 V |  |  | 9766 | bps |
| ASCK20 rise time, fall time | $\mathrm{t}_{\mathrm{R}, \mathrm{t}}$ |  |  |  | 1 | $\mu \mathrm{s}$ |

(3) Serial interface $\operatorname{SMBO}\left(\mathrm{T}_{\mathrm{A}}=-40\right.$ to $+85^{\circ} \mathrm{C}, \mathrm{VDD}=1.8$ to 5.5 V$)(\mu \mathrm{PD} 78 \mathrm{~F} 9177 \mathrm{Y}$ only)
(a) DC characteristics

| Parameter | Symbol |  | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input voltage, high | $\mathrm{V}_{\mathrm{H}}$ | SCLO, SDAO (at hysteresis) |  | 0.8 VdD |  | VDD | V |
| Input voltage, low | VIL | SCLO, SDAO (at hysteresis) |  | 0 |  | 0.2 VdD | V |
| Output voltage, low | VoL | SCLO, SDAO | $\mathrm{V} \mathrm{DD}=4.5$ to 5.5 V , $\mathrm{loL}=10 \mathrm{~mA}$ |  |  | 1.0 | V |
|  |  |  | V DD $=1.8$ to 5.5 V , loL $=400 \mu \mathrm{~A}$ |  |  | 0.5 | V |
| Input leakage current, high | ILIH | SCLO, SDAO | $V_{1}=V_{D D}$ |  |  | 3 | $\mu \mathrm{A}$ |
| Input leakage current, low | ILIL | SCLO, SDAO | $\mathrm{V}_{1}=0 \mathrm{~V}$ |  |  | -3 | $\mu \mathrm{A}$ |

(b) DC characteristics (when using comparator)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input range | VsdA, <br> Vscl | $V_{D D}=1.8$ to 5.5 V | 0 |  | 5.5 | V |
| Transfer level | Visda, <br> VISCL | $4.5 \leq \mathrm{VDD} \leq 5.5 \mathrm{~V}$ | 0.72 V ISMB | VIsmb | 1.28 V ISMB | V |
|  |  | $3.3 \leq \mathrm{VDD}^{2} 4.5 \mathrm{~V}$ | 0.78 V ISmb | VISmb | $1.22 \mathrm{~V}_{\text {ISmB }}$ | V |
|  |  | $2.7 \leq \mathrm{VDD}<3.3 \mathrm{~V}$ | 0.75 V Ismb | VISmb | $1.25 \mathrm{~V}_{\text {ISmb }}$ | V |
|  |  | $1.8 \leq \mathrm{V}_{\mathrm{DD}}<2.7 \mathrm{~V}$ | 0.90 V ISmB | VIsmb | 1.45 V ISMB | V |
| Input level threshold value ${ }^{\text {Note }}$ | Vismb | LVL01, LVL00 = 0, 1 |  | $0.25 \times \mathrm{VDD}$ |  | V |
|  |  | LVL01, LVL00 = 1, 0 |  | $0.375 \times$ VDD |  | V |
|  |  | LVL01, LVL00 = 1, 1 |  | $0.5 \times \mathrm{VDD}$ |  | V |

Note VISMB is an input level threshold value selected by bits LVL00 and LVL01 (bits 0 and 1 of SMB input level setting register 0 (SMBVIO)).
According to the SMB standard (V1.1), the maximum value of low-level input voltage is 0.8 V , and the minimum value of high-level input voltage, 2.1 V . To satisfy these conditions, set LVL01 and LVL00 as follows;

- When $V_{D D}=1.8$ to 3.3 V : LVL01, LVLOO $=1,1(0.5 \times \mathrm{VDD})$
- When Vdd $=3.3$ to 4.5 V : LVL01, LVLOO $=1,0(0.375 \times \mathrm{VDD})$
- When VDd $=4.5$ to 5.5 V : LVL01, LVLOO $=0,1(0.25 \times \mathrm{VDD})$
"LVL01, LVL00 $=0,0$ " is not available since this setting does not satisfy the SMB standard (V1.1).
(c) AC characteristics

| Parameter |  | Symbol | SMB Mode |  | Standard Mode ${ }^{2} \mathrm{C}$ Bus |  | High-speed Mode $I^{2} \mathrm{C}$ Bus |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | MIN. | MAX. | MIN. | MAX. | MIN. | MAX. |  |
| SCLO clock frequency |  |  | fclk | 10 | 100 | 0 | 100 | 0 | 400 | kHz |
| Bus free time <br> (between stop and start condition) |  | tbuF | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| Hold time ${ }^{\text {Note } 1}$ |  | thd:STA | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Start/restart condition setup time |  | tsu:STA | 4.7 | - | 4.7 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Stop condition setup time |  | tsu:sto | 4.0 | - | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| Data hold time | When using CBUScompatible master | thd:DAT | - | - | 5 | - | - | - | $\mu \mathrm{s}$ |
|  | When using SMB/IIC bus |  | 300 | - | $0^{\text {Note } 2}$ | - | $0^{\text {Note } 2}$ | 900 Note 3 | ns |
| Data setup time |  | tsu:dat | 250 | - | 250 | - | $100^{\text {Note } 4}$ | - | ns |
| SCLO clock low-level width |  | tıow | 4.7 | - | 4.7 | - | 1.3 | - | $\mu \mathrm{s}$ |
| SCLO clock high-level width |  | thigh | 4.0 | 50 | 4.0 | - | 0.6 | - | $\mu \mathrm{s}$ |
| SCL0 and SDA0 signal fall time |  | $\mathrm{t}_{\mathrm{F}}$ | - | 300 | - | 300 | - | 300 | ns |
| SCL0 and SDA0 signal rise time |  | $t_{R}$ | - | 1000 | - | 1000 | - | 300 | ns |
| Spike pulse width controlled by input filter |  | tsp | - | - | - | - | 0 | 50 | ns |
| Timeout |  | ttimeout | 25 | 35 | - | - | - | - | ms |
| Total extended time of SCLO clock low-level period (slave) |  | tıow:SEXT | - | 25 | - | - | - | - | ms |
| Total extended time of cumulative clock low-level period (master) |  | tLow:MEXT | - | 10 | - | - | - | - | ms |
| Capacitive load per each bus line |  | Cb | - | - | - | 400 | - | 400 | pF |

Notes 1. In the start condition, the first clock pulse is generated after this hold time.
2. To fill in the undefined area of the SCLO falling edge, it is necessary for the device to internally provide at least 300 ns of hold time for the SDA0 signal (which is VIHmin. of the SCLO signal).
3. If the device does not extend the SCLO signal low hold time (tlow), only maximum data hold time thd:DAT needs to be fulfilled.
4. The high-speed mode $I^{2} \mathrm{C}$ bus is available in the SMB mode and the standard mode $\mathrm{I}^{2} \mathrm{C}$ bus system. At this time, the conditions described below must be satisfied.

- If the device extends the SCLO signal low state hold time
tsu:DAt $\geq 250 \mathrm{~ns}$
- If the device extends the SCLO signal low state hold time

Be sure to transmit the next data bit to the SDAO line before the SCLO line is released (trmax.+ tsu:DAT $=1000+250=1250 \mathrm{~ns}$ by the SMB mode or the standard mode $\mathrm{I}^{2} \mathrm{C}$ bus specification).

## AC Timing Measurement Points (Excluding X1 and XT1 Inputs)



## Clock Timing



## TI Timing

TI80, TI81


Interrupt Input Timing


## RESET Input Timing



## CPT90 Input Timing



## Serial Transfer Timing

3-wire serial I/O mode:


Remark $m=1,2$

3-wire serial I/O mode (when using $\overline{\mathbf{S S 2 0}}$ ):


UART mode (external clock input):


## SMB mode:



10-Bit A/D Converter Characteristics ( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}, 1.8 \leq \mathrm{AV}$ REF $\leq \mathrm{AV} \mathrm{DD}=\mathrm{VDD} \leq 5.5 \mathrm{~V}, \mathrm{AV} \mathrm{SS}=\mathrm{Vss}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 10 | 10 | 10 | bit |
| $\text { Overall error }{ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {ref }} \leq \mathrm{AV} \mathrm{Vd} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.2$ | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV} \mathrm{V}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.4$ | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {ref }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ |  | $\pm 0.8$ | $\pm 1.2$ | \%FSR |
| Conversion time | tconv | $4.5 \mathrm{~V} \leq \mathrm{AV} \mathrm{V}_{\text {Ref }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ | 14 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {ref }} \leq \mathrm{AV}$ do $\leq 5.5 \mathrm{~V}$ | 14 |  | 100 | $\mu \mathrm{s}$ |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {Ref }} \leq \mathrm{AV} \mathrm{V}_{\text {do }} \leq 5.5 \mathrm{~V}$ | 28 |  | 100 | $\mu \mathrm{s}$ |
| Zero-scale error ${ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {ref }} \leq \mathrm{AV}$ do $\leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| Full-scale error ${ }^{\text {Note }}$ |  | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.4$ | \%FSR |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 0.6$ | \%FSR |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 1.2$ | \%FSR |
| Integral linearity error | INL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 2.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 4.5$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {ReF }} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 8.5$ | LSB |
| Differential linearity error Note | DNL | $4.5 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{VDD}^{5} 5.5 \mathrm{~V}$ |  |  | $\pm 1.5$ | LSB |
|  |  | $2.7 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\text {DD }} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 2.0$ | LSB |
|  |  | $1.8 \mathrm{~V} \leq \mathrm{AV}_{\text {REF }} \leq \mathrm{AV} \mathrm{V}_{\mathrm{DD}} \leq 5.5 \mathrm{~V}$ |  |  | $\pm 3.5$ | LSB |
| Analog input voltage | Vian |  | 0 |  | AVREF | V |
| Reference voltage | AV REFF |  | 1.8 |  | $A V_{\text {do }}$ | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{s s}$ | Radref |  | 20 | 40 |  | k $\Omega$ |

Note Excludes quantization error ( $\pm 0.05 \%$ FSR).

Remark FSR: Full scale range

Flash Memory Write/Erase Characteristics ( $\mathrm{TA}=10$ to $40^{\circ} \mathrm{C}, \mathrm{V} D=1.8$ to 5.5 V )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Write current $(\text { Vod pin })^{\text {Note } 1}$ | IDDW | When Vpp supply voltage $=$ VPP1 (5.0 MHz operation) |  |  | 18 | mA |
| Write current (VPP pin) | Ippw | When VPP supply voltage = VPP1 |  |  | 7.5 | mA |
| Erase current $(\text { Vod pin })^{\text {Note } 1}$ | Idde | When Vpp supply voltage $=$ VPp1 <br> (5.0 MHz operation) |  |  | 18 | mA |
| Erase current (VPP pin) | IPPE | When VPP supply voltage $=$ VPP1 |  |  | 100 | mA |
| Unit erase time ${ }^{\text {Note } 2}$ | ter |  | 0.5 | 1 | 1 | S |
| Total erase time | tera |  |  |  | 20 | S |
| $\text { Write count }^{\text {Note } 3}$ |  | Erase/write is regarded as 1 cycle | 20 | 20 | 20 | Times |
| VPP supply voltage | VPPO | In normal operation | 0 |  | 0.2 VdD | V |
|  | VPP1 | During flash memory programming | 9.7 | 10.0 | 10.3 | V |

Notes 1. The current flowing to the ports (including the current flowing through an on-chip pull-up resistor) and AVdd current are not included.
2. The prewrite time before erasure and the erase verify time (writeback time) is not included.
3. When a product is first written after shipment, "erase $\rightarrow$ write" is taken as one rewrite.

Data Memory STOP Mode Low Power Supply Voltage Data Retention Characteristics ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $\mathbf{+ 8 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :--- | :---: | :---: | :---: | :---: |
| Data retention power <br> supply voltage | VDDDR |  | 1.8 |  | 5.5 | V |
| Release signal set time | tsREL |  | 0 |  |  | $\mu \mathrm{~s}$ |
| Oscillation stabilization <br> wait time | twait 1 |  |  |  |  |  |

Notes 1. The oscillation stabilization time is the time the CPU operation is stopped to prevent unstable operation when oscillation starts.
2. By using bits 0 to 2 (OSTSO to OSTS2) of the oscillation stabilization time selection register (OSTS), $2^{12} / f x, 2^{15} / \mathrm{fx}$, or $2^{17} / \mathrm{fx}$ can be selected.

Remark fx: Main system clock oscillation frequency

## Data Retention Timing (STOP Mode Release by RESET)



Data Retention Timing (Standby Release Signal: STOP Mode Release by Interrupt Signal)


## CHAPTER 29 CHARACTERISTICS CURVES ( $\mu$ PD78F9177, 78F9177Y)



## CHAPTER 30 ELECTRICAL SPECIFICATIONS ( $\mu$ PD78F9177A(A1))

## Absolute Maximum Ratings ( $\mathrm{T}_{\mathrm{A}}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Supply voltage | VDD | $\begin{aligned} & \mathrm{A} \mathrm{~V}_{\mathrm{DD}}-0.3 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\mathrm{REF}} \leq \mathrm{A} \mathrm{~V}_{\mathrm{DD}}+0.3 \mathrm{~V} \\ & \mathrm{~A} \mathrm{~V}_{\mathrm{REF}} \leq \mathrm{V}_{\mathrm{DD}}+0.3 \mathrm{~V} \end{aligned}$ | -0.3 to +6.5 | V |
|  | AVDD |  |  | V |
|  | $\mathrm{AV}_{\text {ref }}$ |  |  | V |
|  | VPP | Note | -0.3 to +10.5 | V |
| Input voltage | $V_{11}$ | Pins other than P50 to P53, P23, P24 | -0.3 to VDD +0.3 | V |
|  | $\mathrm{V}_{12}$ | P23, P24 | -0.3 to +5.5 | V |
|  | $\mathrm{V}_{13}$ | P50 to P53 | -0.3 to +13 | V |
| Output voltage | Vo |  | -0.3 to VDD +0.3 | V |
| Output current, high | Іон | Per pin | -4 | mA |
|  |  | Total for all pins | -14 | mA |
| Output current, low | lot | Per pin | 5 | mA |
|  |  | Total for all pins | 80 | mA |
| Operating ambient temperature | TA | In normal operation mode | -40 to +105 | ${ }^{\circ} \mathrm{C}$ |
|  |  | During flash memory programming | 10 to 40 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | $\mathrm{T}_{\text {stg }}$ |  | -40 to +125 | ${ }^{\circ} \mathrm{C}$ |

Note Make sure that the following conditions of the VPP voltage application timing are satisfied when the flash memory is written.

- When supply voltage rises

VPP must exceed VDD $10 \mu \mathrm{~s}$ or more after VDD has reached the lower-limit value ( 4.5 V ) of the operating voltage range (see a in the figure below).

- When supply voltage drops

Vod must be lowered $10 \mu \mathrm{~s}$ or more after VPP falls below the lower-limit value (4.5 V ) of the operating voltage range of $V_{\text {do }}$ (see $\mathbf{b}$ in the figure below).


Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## Main System Clock Oscillator Characteristics

$\left(\mathrm{V}_{\mathrm{DD}}=4.5\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)$

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Ceramic resonator |  | Oscillation frequency (fx) ${ }^{\text {Note } 1}$ | $V_{D D}=$ oscillation voltage range | 1.0 |  | 5.0 | MHz |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ | After Vod reaches oscillation voltage range MIN. |  |  | 4 | ms |
| External clock | $\begin{array}{ll} \mathrm{X} 1 & \mathrm{X} 2 \\ \hline \end{array}$ | X1 input frequency (fx) ${ }^{\text {Note } 1}$ |  | 1.0 |  | 5.0 | MHz |
|  |  | X1 input high-/low-level width $\text { ( } \mathrm{txh}, \mathrm{txL} \text { ) }$ |  | 85 |  | 500 | ns |
|  | X2 | X 1 input frequency ( fx$)^{\text {Note } 1}$ |  | 1.0 |  | 5.0 | MHz |
|  | $\triangle \quad \text { OPEN }$ | X1 input high-/low-level width (txh, txL) |  | 85 |  | 500 | ns |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the main system clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. When the main system clock is stopped and the device is operating on the subsystem clock, wait until the oscillation stabilization time has been secured by the program before switching back to the main system clock.
3. For ceramic resonator, use the part number for which the resonator manufacturer guarantees operation under the condition of $\mathrm{T}_{\mathrm{A}}=105^{\circ} \mathrm{C}$.

Remark For the resonator selection and oscillator constant, users are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

## Subsystem Clock Oscillator Characteristics

$\left(\mathrm{VDD}=4.5\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)$

| Resonator | Recommended Circuit | Parameter | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Crystal resonator | Vsso XT1 XT2 | Oscillation frequency ( fxT$)^{\text {Note } 1}$ |  | 32 | 32.768 | 35 | kHz |
|  |  |  |  |  |  |  |  |
|  |  | Oscillation stabilization time ${ }^{\text {Note } 2}$ |  |  | 1.2 | 2 | s |
| External clock | $\mathrm{XT1} \quad \mathrm{XT2}$ | XT1 input frequency (fxT) ${ }^{\text {Note } 1}$ |  | 32 |  | 35 | kHz |
|  | $\lambda$ | XT1 input high-/low-level width (tхтн, tхтL) |  | 14.3 |  | 15.6 | $\mu \mathrm{s}$ |

Notes 1. Indicates only oscillator characteristics. Refer to AC Characteristics for instruction execution time.
2. Time required to stabilize oscillation after reset or STOP mode release. Use a resonator that stabilizes oscillation within the oscillation wait time.

Cautions 1. When using the subsystem clock oscillator, wire as follows in the area enclosed by the broken lines in the above figures to avoid an adverse effect from wiring capacitance.

- Keep the wiring length as short as possible.
- Do not cross the wiring with the other signal lines.
- Do not route the wiring near a signal line through which a high fluctuating current flows.
- Always make the ground point of the oscillator capacitor the same potential as Vsso.
- Do not ground the capacitor to a ground pattern through which a high current flows.
- Do not fetch signals from the oscillator.

2. The subsystem clock oscillator is designed as a low-amplitude circuit for reducing current consumption, and is more prone to malfunction due to noise than the main system clock oscillator. Particular care is therefore required with the wiring method when the subsystem clock is used.

Remark For the resonator selection and oscillator constant, users are requested to either evaluate the oscillation themselves or apply to the resonator manufacturer for evaluation.

DC Characteristics (VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)(1 / 3)$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output current, high | Іон | Per pin |  |  |  | -1 | mA |
|  |  | Total for all pins |  |  |  | -7 | mA |
| Output current, low | IoL | Per pin |  |  |  | 1.6 | mA |
|  |  | Total for all pins |  |  |  | 40 | mA |
| Input voltage, high | $\mathrm{V}_{1+1}$ | P00 to P05, P10, P11, P60 to P67 |  | 0.7 V DD |  | VDD | V |
|  | $\mathrm{V}_{1+2}$ | P50 to P53 |  | 0.7Vdd |  | 10 | V |
|  | Vінз | RESET, P20 to P26, P30 to P33 |  | 0.8 VDD |  | VDD | V |
|  | $\mathrm{V}_{1+4}$ | X1, X2, XT1, XT2 |  | V $\mathrm{DD}-0.1$ |  | VDD | V |
| Input voltage, low | VLL1 | P00 to P05, P10, P11, P60 to P67 |  | 0 |  | 0.3 VdD | V |
|  | VIL2 | P50 to P53 |  | 0 |  | 0.3 Vdo | V |
|  | VIL3 | RESET, P20 to P26, P30 to P33 |  | 0 |  | 0.2 Vdo | V |
|  | VIL4 | X1, X2, XT1, XT2 |  | 0 |  | 0.1 | V |
| Output voltage, high | Vон | Pins other than P23, P24, P50 to P53 | $\mathrm{IOH}=-1 \mathrm{~mA}$ | Vod - 2.0 |  |  | V |
|  |  |  | $\mathrm{IOH}=-100 \mu \mathrm{~A}$ | VDD - 1.0 |  |  | V |
| Output voltage, low | Vol1 | Pins other than P50 to P53 | $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 2.0 | V |
|  |  |  | $\mathrm{loL}=400 \mu \mathrm{~A}$ |  |  | 1.0 | V |
|  | Vol2 | P50 to P53 | $\mathrm{loL}=1.6 \mathrm{~mA}$ |  |  | 1.0 | V |

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

DC Characteristics (VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)(2 / 3)$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Input leakage current, high | ІІн1 | $\mathrm{V}_{1}=\mathrm{V}_{\mathrm{DD}}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  | 10 | $\mu \mathrm{A}$ |
|  | lılı2 |  | $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1, \mathrm{XT} 2$ |  |  | 20 | $\mu \mathrm{A}$ |
|  | ІІнз | $\mathrm{V}_{1}=10 \mathrm{~V}$ | P 50 to P53 <br> ( N -ch open drain) |  |  | 80 | $\mu \mathrm{A}$ |
| Input leakage current, low | İL1 | V I $=0 \mathrm{~V}$ | Pins other than P50 to P53 (N-ch open drain), $\mathrm{X} 1, \mathrm{X} 2, \mathrm{XT} 1$, and XT 2 |  |  | -10 | $\mu \mathrm{A}$ |
|  | ILL2 |  | X1, X2, XT1, XT2 |  |  | -20 | $\mu \mathrm{A}$ |
|  | lıı3 |  | P50 to P53 <br> ( N -ch open drain) |  |  | $-10^{\text {Note }}$ | $\mu \mathrm{A}$ |
| Output leakage current, high | ІІон | $\mathrm{V}_{\mathrm{O}}=\mathrm{V}_{\mathrm{DD}}$ |  |  |  | 10 | $\mu \mathrm{A}$ |
| Output leakage current, low | ILOL | V o $=0 \mathrm{~V}$ |  |  |  | -10 | $\mu \mathrm{A}$ |
| Software pull-up resistor | R1 | $\mathrm{V}_{1}=0 \mathrm{~V}$, for pins other than P23, P24, and P50 to P53 |  | 50 | 100 | 300 | k $\Omega$ |

Note A low-level input leakage current of $-60 \mu \mathrm{~A}$ (MAX.) flows only during the 1-cycle time after a read instruction is executed to P50 to P53 when P50 to P53 are set to input mode. At times other than this, a $-10 \mu \mathrm{~A}$ (MAX.) current flows.

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

DC Characteristics (VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)(3 / 3)$

| Parameter | Symbol | Conditions |  | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Power supply current | IDD1 ${ }^{\text {Note } 1}$ | 5.0 MHz crystal oscillation operating mode (C1 = C2 = 22 pF ) | $\mathrm{V} D=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 7.5 | 20.0 | mA |
|  | IDD2 ${ }^{\text {Note }} 1$ | 5.0 MHz crystal oscillation HALT mode (C1 = C2 = 22 pF ) | $\mathrm{V} D=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 3.0 | 6.0 | mA |
|  | $\mathrm{Idos}^{\text {Note } 1}$ | 32.768 kHz crystal oscillation operating mode ${ }^{\text {Note } 3}$ $\begin{aligned} & (\mathrm{C} 3=\mathrm{C} 4=22 \mathrm{pF}, \\ & \mathrm{R}=220 \mathrm{k} \Omega) \end{aligned}$ | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%$ |  | 30 | 3000 | $\mu \mathrm{A}$ |
|  | $\mathrm{ImD}^{\text {Note }} 1$ | 32.768 kHz crystal oscillation HALT mode ${ }^{\text {Note } 3}$ $\begin{aligned} & (C 3=C 4=22 \mathrm{pF}, \\ & R=220 \mathrm{k} \Omega) \end{aligned}$ | V DD $=5.0 \mathrm{~V} \pm 10 \%$ |  | 25 | 2500 | $\mu \mathrm{A}$ |
|  | IDDs ${ }^{\text {Note } 1}$ | 32.768 kHz crystal stop STOP mode | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%$ |  | 1.0 | 1000 | $\mu \mathrm{A}$ |
|  | $1 \mathrm{IDDG}^{\text {Note } 2}$ | 5.0 MHz crystal oscillation A/D operating mode (C1 = C2 = 22 pF ) | $\mathrm{V} D \mathrm{D}=5.0 \mathrm{~V} \pm 10 \%^{\text {Note } 4}$ |  | 8.7 | 22.3 | mA |

Notes 1. The $A V_{\text {refo }} O N$ (ADCSO (bit 7 of ADMO; $A / D$ converter mode register 0 ) $=1$ ), $A V_{D D}$, and port current (including the current flowing through the internal pull-up resistors) is not included.
2. The $A V_{\text {ref }} O N(A D C S O=1)$ and port current (including the current flowing through the internal pull-up resistors) is not included. Refer to the A/D converter characteristics for the current flowing through AVref.
3. When the main system clock is stopped.
4. During high-speed mode operation (when the processor clock control register (PCC) is set to 00 H .)

Remark Unless otherwise specified, the characteristics of alternate-function pins are the same as those of port pins.

## AC Characteristics

(1) Basic operation (VDD $=4.5$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $+105^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Cycle time (minimum instruction execution time) | Tcy | Operation based on the main system clock | 0.4 |  | 8 | $\mu \mathrm{s}$ |
|  |  | Operation based on the subsystem clock | 114 | 122 | 125 | $\mu \mathrm{s}$ |
| TI80 and TI81 input frequency | $\mathrm{f}_{\mathrm{T}}$ |  | 0 |  | 4 | MHz |
| TI80 and TI81 input high-/low-level width | tтin, ttil |  | 0.1 |  |  | $\mu \mathrm{s}$ |
| Interrupt input high-/low-level width | tinth, tintl | INTP0 to INTP3 | 10 |  |  | $\mu \mathrm{s}$ |
| RESET input lowlevel width | trsL |  | 10 |  |  | $\mu \mathrm{s}$ |
| CPT90 input high-/low-level width | tcPe, tCPL |  | 10 |  |  | $\mu \mathrm{s}$ |

Tcy vs. Vdd (main system clock)

(2) Serial interface $20\left(\mathrm{VDD}=4.5\right.$ to $5.5 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=-40$ to $\left.+105^{\circ} \mathrm{C}\right)$
(a) 3-wire serial I/O mode ( $\overline{\text { SCK20 }}$...Internal clock)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 cycle time }}$ | tkcy1 |  | 800 |  |  | ns |
| $\overline{\text { SCK20 }}$ high-/lowlevel width | tkh1, tkL1 |  | tксү1/2-50 |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsik 1 |  | 150 |  |  | ns |
| SI20 hold time (from SCK20 $\uparrow$ ) | tksS1 |  | 400 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tksO1 | $R=1 \mathrm{k} \Omega, \mathrm{C}=100 \mathrm{pF}$ Note | 0 |  | 250 | ns |

Note $R$ and $C$ are the load resistance and load capacitance of the SO 20 output line.
(b) 3-wire serial I/O mode (SCK20...External clock)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\overline{\text { SCK20 }}$ cycle time | tkcy2 |  | 900 |  |  | ns |
| $\overline{\text { SCK20 high-/low- }}$ level width | tkh2, $_{\text {tkL2 }}$ |  | 400 |  |  | ns |
| SI20 setup time (to SCK20 $\uparrow$ ) | tsıK2 |  | 100 |  |  | ns |
| SI20 hold time (from SCK20 $\uparrow$ ) | tksI2 |  | 400 |  |  | ns |
| SO20 output delay time from SCK20 $\downarrow$ | tKSO2 | $\mathrm{R}=1 \mathrm{k} \Omega, \mathrm{C}=100 \mathrm{pF}^{\text {Note }}$ | 0 |  | 300 | ns |
| SO20 setup time (when using $\overline{\mathrm{SS20}}$, to $\overline{\mathrm{SS} 20} \downarrow$ ) | tKAS2 |  |  |  | 120 | ns |
| SO20 disable time <br> (when using $\overline{\mathrm{SS} 20}$, from $\overline{\mathrm{SS} 20} \uparrow$ ) | tKDS2 |  |  |  | 240 | ns |

Note $R$ and $C$ are the load resistance and load capacitance of the SO20 output line.
(c) UART mode (dedicated baud rate generator output)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Transfer rate |  |  |  |  | 78125 | bps |

(d) UART mode (external clock input)

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ASCK20 cycle time | tксү3 |  | 900 |  |  | ns |
| ASCK20 high-/lowlevel width | tıh3, $_{\text {tkL3 }}$ |  | 400 |  |  | ns |
| Transfer rate |  |  |  |  | 39063 | bps |
| ASCK20 rise time, fall time | $t_{R}, t_{F}$ |  |  |  | 1 | $\mu \mathrm{s}$ |

## AC Timing Measurement Points (Excluding X1 and XT1 Inputs)



## Clock Timing



TI Timing

TI80, TI81


Interrupt Input Timing

NTP0 to INTP3


RESET Input Timing


## CPT90 Input Timing



## Serial Transfer Timing

## 3-wire serial I/O mode:



Remark $m=1,2$

3-wire serial I/O mode (when using SS20):


UART mode (external clock input):


10-Bit A/D Converter Characteristics ( $\mathrm{T}_{\mathrm{A}}=-\mathbf{4 0}$ to $+105^{\circ} \mathrm{C}, 4.5 \leq \mathrm{AV}$ REF $\leq \mathrm{AV} \mathrm{DD}=\mathrm{VDD} \leq 5.5 \mathrm{~V}, \mathrm{AV} \mathrm{Ss}=\mathrm{V} s \mathrm{~s}=0 \mathrm{~V}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Resolution |  |  | 10 | 10 | 10 | bit |
| Overall error ${ }^{\text {Note }}$ |  |  |  | $\pm 0.2$ | $\pm 0.6$ | \%FSR |
| Conversion time | tconv |  | 14 |  | 28 | $\mu \mathrm{s}$ |
| Zero-scale error ${ }^{\text {Note }}$ |  |  |  |  | $\pm 0.6$ | \%FSR |
| Full-scale error ${ }^{\text {Note }}$ |  |  |  |  | $\pm 0.6$ | \%FSR |
| Integral linearity error ${ }^{\text {Note }}$ | INL |  |  |  | $\pm 4.5$ | LSB |
| Differential linearity error ${ }^{\text {Note }}$ | DNL |  |  |  | $\pm 2.0$ | LSB |
| Analog input voltage | Vian |  | 0 |  | AVREF | V |
| Reference voltage | $\mathrm{AV}_{\text {ref }}$ |  | 4.5 |  | AV ${ }_{\text {dD }}$ | V |
| Resistance between $A V_{\text {ref }}$ and $A V_{s s}$ | Radref |  | 20 | 40 |  | k $\Omega$ |

Note Excludes quantization error ( $\pm 0.05 \%$ FSR).

Remark FSR: Full scale range

Flash Memory Write/Erase Characteristics ( $\mathrm{T}_{\mathrm{A}}=10$ to $40^{\circ} \mathrm{C}, \mathrm{VDD}=4.5$ to 5.5 V )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Write current (Vid pin) ${ }^{\text {Note } 1}$ | Idow | When VPP supply voltage $=$ VPP1 (5.0 MHz operation) |  |  | 23 | mA |
| Write current (VPP pin) | Ippw | When VPP supply voltage $=\mathrm{V}_{\text {PP } 1}$ |  |  | 20 | mA |
| Erase current ${ }^{\text {Note } 1}$ (Vod pin) | Idde | When VPP supply voltage $=$ VPP1 (5.0 MHz operation) |  |  | 23 | mA |
| Erase current <br> (Vpp pin) | IPPE | When VPP supply voltage $=$ VPP1 |  |  | 100 | mA |
| Unit erase time ${ }^{\text {Note } 2}$ | ter |  | 0.2 | 0.2 | 0.2 | s |
| Total erase time | tera |  |  |  | 20 | S |
| Write count ${ }^{\text {Note } 3}$ |  | Erase/write is regarded as 1 cycle | 20 | 20 | 20 | Times |
| VPP supply voltage | Vppo | In normal operation mode | 0 |  | 0.2VDD | V |
|  | VPP1 | During flash memory programming | 9.7 | 10.0 | 10.3 | V |

Notes 1. The current flowing to the ports (including the current flowing through an on-chip pull-up resistor) and AVdd current are not included.
2. The prewrite time before erasure and the erase verify time (writeback time) are not included.
3. When a product is first written after shipment, "erase $\rightarrow$ write" is taken as one rewrite.

Data Memory STOP Mode Low Power Supply Voltage Data Retention Characteristics ( $\mathrm{T}_{\mathrm{A}}=\mathbf{- 4 0}$ to $\mathbf{+ 1 0 5}{ }^{\circ} \mathrm{C}$ )

| Parameter | Symbol | Conditions | MIN. | TYP. | MAX. | Unit |
| :--- | :---: | :--- | :---: | :---: | :---: | :---: |
| Data retention power <br> supply voltage | VDDDR |  | 4.5 |  | 5.5 | V |
| Release signal set time | tsReL |  | 0 |  |  | $\mu \mathrm{~s}$ |
| Oscillation stabilization <br> wait time | twait 1 |  |  |  |  |  |

Notes 1. The oscillation stabilization time is the time the CPU operation is stopped to prevent unstable operation when oscillation starts.
2. $2^{12} / \mathrm{fx}, 2^{15} / \mathrm{fx}$, or $2^{17} / \mathrm{fx}$ can be selected by using bits 0 to 2 (OSTSO to OSTS2) of the oscillation stabilization time selection register (OSTS).

Remark fx: Main system clock oscillation frequency

## Data Retention Timing (STOP Mode Release by $\overline{\text { RESET }}$



Data Retention Timing (Standby Release Signal: STOP Mode Release by Interrupt Signal)


## CHAPTER 31 PACKAGE DRAWINGS

## 44 PIN PLASTIC LQFP (10x10)



NOTE
Each lead centerline is located within 0.20 mm of its true position (T.P.) at maximum material condition.

| ITEM | MILLIMETERS |
| :---: | :--- |
| A | $12.0 \pm 0.2$ |
| B | $10.0 \pm 0.2$ |
| C | $10.0 \pm 0.2$ |
| D | $12.0 \pm 0.2$ |
| F | 1.0 |
| G | 1.0 |
| H | $0.37_{-0.07}^{+0.08}$ |
| I | 0.20 |
| J | 0.8 (T.P.) |
| K | $1.0 \pm 0.2$ |
| L | 0.5 |
| M | $0.17+0.03$ |
| N | 0.10 |
| P | $1.4 \pm 0.05$ |
| Q | $0.1 \pm 0.05$ |
| $R$ | $3_{-3}^{\circ+4{ }^{\circ}}$ |
| S | 1.6 MAX. |
| T | 0.25 (T.P.) |
| U | $0.6 \pm 0.15$ |
|  | S44GB-80-8ES-2 |

## 48-PIN PLASTIC TQFP (FINE PITCH) (7x7)


detail of lead end


## NOTE

Each lead centerline is located within 0.10 mm of its true position (T.P.) at maximum material condition.

| ITEM | MILLIMETERS |
| :---: | :--- |
| A | $9.0 \pm 0.2$ |
| B | $7.0 \pm 0.2$ |
| C | $7.0 \pm 0.2$ |
| D | $9.0 \pm 0.2$ |
| F | 0.75 |
| G | 0.75 |
| H | $0.22_{-0.04}^{+0.05}$ |
| I | 0.10 |
| J | $0.5($ T.P. $)$ |
| K | $1.0 \pm 0.2$ |
| L | $0.5 \pm 0.2$ |
| M | $0.145_{-0.055}^{+0.055}$ |
| N | 0.10 |
| P | $1.0 \pm 0.1$ |
| Q | $0.1 \pm 0.05$ |
| R | $3^{\circ+7^{\circ}}$ |
| S | 1.27 MAX. |
|  | S48GA-50-9EU-2 |

## CHAPTER 32 RECOMMENDED SOLDERING CONDITIONS

The $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries should be soldered and mounted under the following recommended conditions.

For technical information, see the following website.
Semiconductor Device Mount Manual (http://www.necel.com/pkg/en/mount/index.html).

Table 32-1. Surface Mounting Type Soldering Conditions (1/4)

| $\mu$ PD789166GB- $\times \times \times$-8ES: | 0) |
| :---: | :---: |
| $\mu$ PD789167GB- $\times \times \times$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176GB- $\times \times \times$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177GB-×××-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789166YGB-×Xx-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789167YGB- $\times \times \times$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176YGB- $\times \times \times$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177YGB-××x-8ES: | $44-$ pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789166GB(A)-×x ${ }^{\text {a }}$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789167GB(A)-×x ${ }^{\text {a }}$-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176GB(A)-×××-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177GB(A)-×××-8ES: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu \mathrm{PD} 789166 \mathrm{~GB}(\mathrm{~A} 1)-\times \times \times-8 \mathrm{E}$ | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789167GB(A1)-××× ${ }^{\text {( }}$ (ES | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176GB(A1)-××× ${ }^{\text {( }}$ - ${ }^{\text {ES }}$ | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177GB(A1)-××x-8ES | 44-pin plastic LQFP (10×10) |
| $\mu$ PD789166GB(A2)-××x-8ES | 44-pin plastic LQFP (10×10) |
| $\mu$ PD789167GB(A2)- $\times \times \times \times$-8ES | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176GB(A2)-×xx-8ES | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177GB(A2)-×xx-8ES | 44-pin plastic LQFP (10 $\times 10$ ) |


| Soldering Method | Soldering Conditions | Recommended Condition <br> Symbol |
| :--- | :--- | :--- |
| Infrared reflow | Package peak temperature: $235^{\circ} \mathrm{C}$, Time: 30 seconds max. <br> (at $210^{\circ} \mathrm{C}$ or higher), Count: Twice or less | IR35-00-2 |
| VPS | Package peak temperature: $215^{\circ} \mathrm{C}$, Time: 40 seconds max. <br> (at $200^{\circ} \mathrm{C}$ or higher), Count: Twice or less | VP15-00-2 |
| Wave soldering | Solder bath temperature: $260^{\circ} \mathrm{C}$ max., Time: 10 seconds max., Count: <br> Once, Preheating temperature: $120^{\circ} \mathrm{C}$ max. (package surface <br> temperature) | WS60-00-1 |
| Partial heating | Pin temperature: $350^{\circ} \mathrm{C}$ max., Time: 3 seconds max. (per pin row) | - |

Caution Do not use different soldering methods together (except for partial heating).

Remark For soldering methods and conditions other than those recommended above, contact an NEC Electronics sales representative.

Table 32-1. Surface Mounting Type Soldering Conditions (2/4)

| $\mu$ PD78F9177GB-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| :--- | :--- |
| $\mu$ PD78F9177YGB-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| $\mu$ PD78F9177AGB-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| $\mu$ PD78F9177AYGB-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| $\mu$ PD78F9177AGB(A)-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| $\mu$ PD78F9177AYGB(A)-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |
| $\mu$ PD78F9177AGB(A1)-8ES: | 44-pin plastic LQFP $(10 \times 10)$ |


| Soldering Method | Soldering Conditions | Recommended <br> Condition Symbol |
| :--- | :--- | :--- |
| Infrared reflow | Package peak temperature: $235^{\circ} \mathrm{C}$, Time: 30 seconds max. (at $210^{\circ} \mathrm{C}$ or <br> higher), Count: Twice or less, Exposure limit: 3 days Note (After that, <br> prebaking is necessary at $125^{\circ} \mathrm{C}$ for 10 to 72 hours) | IR35-103-2 |
| VPS | Package peak temperature: $215^{\circ} \mathrm{C}$, Time: 40 seconds max. (at $200^{\circ} \mathrm{C}$ or <br> higher), Count: Twice or less, Exposure limit: 3 days Note (After that, <br> prebaking is necessary at $125^{\circ} \mathrm{C}$ for 10 to 72 hours) | VP15-103-2 |
| Wave soldering | Solder bath temperature: $260^{\circ} \mathrm{C}$ max., Time: 10 seconds max., Count: <br> Once, Preheating temperature: $120^{\circ} \mathrm{C}$ max. (package surface <br> temperature), Exposure limit: 3 days Note (After that, prebaking is <br> necessary at $125^{\circ} \mathrm{C}$ for 10 to 72 hours) | WS60-103-1 |
| Partial heating | Pin temperature: $350^{\circ} \mathrm{C}$ max., Time: 3 seconds max. (per pin row) |  |

Note The number of days for storage at $25^{\circ} \mathrm{C}, 65 \%$ RH MAX after the dry pack has been opened.

Caution Do not use different soldering methods together (except for partial heating).

Remark For soldering methods and conditions other than those recommended above, contact an NEC Electronics sales representative.

Table 32-1. Surface Mounting Type Soldering Conditions (3/4)

|  |  |
| :---: | :---: |
| $\mu$ PD789167GA- $-\times \times$-9EU: |  |
| $\mu$ PD789176GA-×××-9EU: |  |
| $\mu$ PD789177GA-×××-9EU |  |
| $\mu$ PD789166YGA-×Xx-9EU: |  |
| 789 |  |
|  |  |
| $\mu$ PD789177YGA-××x-9EU: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789166YGA(A)-×X×-9EU: 48-pin plastic TQFP (fine pitch) ( $7 \times 7$ ) |  |
| $\mu$ PD789167YGA(A)-×××-9EU: 48-pin plastic TQFP (fine pitch) (7×7) |  |
| $\mu$ PD789176YGA(A)-×××-9EU: 48-pin plastic TQFP (fine pitch) ( $7 \times 7$ ) |  |
| $\mu$ PD789177YGA(A)- $\times \times \times$-9EU: 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |  |
| $\mu$ PD78F9177YGA-9EU: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ | 7) |
| $\mu$ PD78F9177AYGA-9EU: | 48-pin plastic TQFP (fine pitch) ( $7 \times 7$ ) |
| PD78F9177AYGA(A)-9E | 48-pin plastic T |


| Soldering Method | Soldering Conditions | Recommended Condition <br> Symbol |
| :--- | :--- | :--- |
| Infrared reflow | Package peak temperature: $235^{\circ} \mathrm{C}$, Time: 30 seconds max. <br> (at $210^{\circ} \mathrm{C}$ or higher), Count: Twice or less, Exposure limit: 3 days ${ }^{\text {Note }}$ <br> (After that, prebaking is necessary at $125^{\circ} \mathrm{C}$ for 10 to 72 hours) | IR35-103-2 |
| VPS | Package peak temperature: $215^{\circ} \mathrm{C}$, Time: 40 seconds max. <br> (at $200^{\circ} \mathrm{C}$ or higher), Count: Twice or less, Exposure limit: 3 days ${ }^{\text {Note }}$ <br> (After that, prebaking is necessary at $125^{\circ} \mathrm{C}$ for 10 to 72 hours) | VP15-103-2 |
| Partial heating | Pin temperature: $350^{\circ} \mathrm{C}$ max., Time: 3 seconds max. (per pin row) | - |

Note The number of days for storage at $25^{\circ} \mathrm{C}, 65 \%$ RH MAX after the dry pack has been opened.

Caution Do not use different soldering methods together (except for partial heating).

Remark For soldering methods and conditions other than those recommended above, contact an NEC Electronics sales representative.

Table 32-1. Surface Mounting Type Soldering Conditions (4/4)

| $\mu$ PD789166GB-×××-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| :---: | :---: |
| $\mu$ PD789167GB-×××-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu \mathrm{PD} 789176 \mathrm{~GB}-\times \times \times$-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu \mathrm{PD} 789177 \mathrm{~GB}-\times \times \times-8 \mathrm{ES}-\mathrm{A}$ : | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789166YGB-×XX-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789 | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789176YGB- $\times \times \times$-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789177YGB-×XX-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD78F9177GB-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD78F9177AGB-8ES-A: | 44-pin plastic LQFP (10 $\times 10$ ) |
| $\mu$ PD789166GA-XXX-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789167GA-×××-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789176GA-×××-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789177GA-XXX-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789166YGA-×××-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789167YGA- $\times \times \times$-9EU-A | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789176YGA-XXX-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD789177YGA-×XX-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD78F9177YGA-9EU-A: | 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ |
| $\mu$ PD78F9177YAGA-9EU-A: | 48-pin plastic TQFP (fine pitch) ( $7 \times 7$ ) |


| Soldering Method | Soldering Conditions | Recommended Condition <br> Symbol |
| :--- | :--- | :---: |
| Infrared reflow | Package peak temperature: $260^{\circ} \mathrm{C}$, Time: 60 seconds max. (at $220^{\circ} \mathrm{C}$ or <br> higher), Count: Three times or less, Exposure limit: 7 days Note (After that, <br> prebaking is necessary at $125^{\circ} \mathrm{C}$ for 20 to 72 hours) | IR60-207-3 |
| Wave soldering | When the pin pitch of the package is 0.65 mm or more, wave soldering <br> can also be performed. For details, ask an NEC Electronics sales <br> representative. | - |
| Partial heating | Pin temperature: $350^{\circ} \mathrm{C}$ max., Time: 3 seconds max. (per pin row) | - |

Caution Do not use different soldering methods together (except for partial heating).

Remarks 1. Products that have the part numbers suffixed by "-A" are lead-free products.
2. For soldering methods and conditions other than those recommended above, contact an NEC Electronics sales representative.

## APPENDIX A DEVELOPMENT TOOLS

The following development tools are available for development of systems using the $\mu$ PD789167, 789177, 789167Y, and 789177Y Subseries. Figure A-1 shows the development tools.

- Support to PC98-NX Series

Unless specified otherwise, the products supported by IBM PC/AT ${ }^{T M}$ compatibles can be used in PC98-NX Series. When using the PC98-NX Series, refer to the explanation of IBM PC/AT compatibles.

- Windows ${ }^{\text {TM }}$

Unless specified otherwise, "Windows" indicates the following operating systems.

- Windows 3.1
- Windows 95
- Windows 98
- Windows 2000
- Windows NT $^{\text {TM }}$ Ver. 4.0
- Windows XP

Figure A-1. Development Tools


Notes 1. The C library source file is not included in the software package.
2. The project manager is included in the assembler package.

The project manager is used only in the Windows environment.

## A. 1 Software Package

| SP78K0S <br> Software package | Software tools for development of the $78 \mathrm{~K} / 0$ S Series are combined in this package. <br> The following tools are included. <br> RA78K0S, CC78K0S, ID78K0S-NS, SM78K0S, and device files |
| :--- | :--- |
|  | Part number: $\mu$ SXXXXSP78K0S |

Remark $x \times x \times$ in the part number differs depending on the OS used
$\mu \mathrm{S} \times \times \times \times$ SP78K0S

| $\times X X \times$ | Host Machine | OS | Supply Medium |
| :---: | :--- | :--- | :---: |
| AB17 | PC-9800 series, IBM PC/AT <br> compatible | Japanese Windows | CD-ROM |
| BB17 | English Windows |  |  |

## A. 2 Language Processing Software

| RA78K0S <br> Assembler package | Program that converts program written in mnemonic into object codes that can be executed by a microcontroller. <br> In addition, automatic functions to generate a symbol table and optimize branch instructions are also provided. <br> Used in combination with a device file (DF789178) (sold separately). <br> <Caution when used in PC environment> <br> The assembler package is a DOS-based application but may be used in the Windows environment by using the project manager of Windows (included in the assembler package). |
| :---: | :---: |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ RA78K0S |
| CC78K0S <br> C compiler package | Program that converts program written in C language into object codes that can be executed by a microcontroller. <br> Used in combination with an assembler package (RA78K0S) and device file (DF789178) (both sold separately). <br> <Caution when used in PC environment> <br> The C compiler package is a DOS-based application but may be used in the Windows environment by using the project manager of Windows (included in the assembler package). |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ CC78K0S |
| DF789178 ${ }^{\text {Note } 1}$ <br> Device file | File containing the information inherent to the device. Used in combination with the RA78K0S, CC78K0S, ID78K0S-NS, and SM78KOS (all sold separately). |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ DF789178 |
| CC78K0S-L ${ }^{\text {Note } 2}$ <br> C library source file | Source file of functions for generating an object library included in the C compiler package. Necessary for changing the object library included in the C compiler package according to the customer's specifications. Since this is a source file, its working environment does not depend on any particular operating system. |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ CC78K0S-L |

Notes 1. DF789178 is a common file that can be used with the RA78K0S, CC78K0S, ID78K0S-NS, and SM78K0S.
2. CC78K0S-L is not included in the software package (SP78K0S).

Remark $\times x \times \times$ in the part number differs depending on the host machine and operating system to be used.
$\mu \mathrm{S} \times \times \times \times$ RA78K0S
$\mu \mathrm{S} \times \times \times \times$ CC78K0S

| xXxX | Host Machine | OS | Supply Medium |
| :---: | :---: | :---: | :---: |
| AB13 | PC-9800 series, IBM PC/AT compatible | Japanese Windows | 3.5" 2HD FD |
| BB13 |  | English Windows |  |
| AB17 |  | Japanese Windows | CD-ROM |
| BB17 |  | English Windows |  |
| 3P17 | HP9000 series 700 ${ }^{\text {TM }}$ | HP-UX ${ }^{\text {TM }}$ (Rel. 10.10) |  |
| 3K17 | SPARCstation ${ }^{\text {™ }}$ | $\begin{aligned} & \text { SunOS }^{\text {TM }} \text { (Rel. 4.1.4), } \\ & \text { Solaris }^{\text {TM }} \text { (Rel. 2.5.1) } \end{aligned}$ |  |

## $\mu \mathrm{Sx} \mathrm{\times x} \mathrm{\times DF} 789178$ <br> $\mu \mathrm{Sxx} \mathrm{\times xCC78K0S-L}$

| xxxx | Host Machine | OS | Supply Medium |
| :---: | :---: | :---: | :---: |
| AB13 | PC-9800 series, IBM PC/AT compatible | Japanese Windows | 3.5 " 2HD FD |
| BB13 |  | English Windows |  |
| 3P16 | HP9000 series 700 | HP-UX (Rel. 10.10) | DAT |
| 3K13 | SPARCstation | SunOS (Rel. 4.1.4), <br> Solaris (Rel. 2.5.1) | 3.5" 2HD FD |
| 3K15 |  |  | 1/4-inch CGMT |

## A. 3 Control Software

| Project manager | Control software created for efficient development of the user program in the Windows <br> environment. User program development operations such as editor startup, build, and <br> debugger startup can be performed from the project manager. <br> <Caution> <br> The project manager is included in the assembler package (RA78K0S). <br> The project manager is used only in the Windows environment. |
| :--- | :--- |

## A. 4 Flash Memory Writing Tools

| Flashpro III (FL-PR3, PG-FP3) <br> Flashpro IV (FL-PR4, PG-FP4) | Dedicated flash programmer for microcontrollers incorporating flash memory |
| :--- | :--- |
| Flash programmer |  |
| FA-44GB-8ES | Adapter for writing to flash memory and connected to Flashpro III or Flashpro IV. |
| FA-48GA | - FA-44GB-8ES: For 44-pin plastic LQFP (GB-8ES type) |
| Flash memory writing adapter | - FA-48GA: For 48-pin plastic TQFP (GA-9EU type) |

Remark The FL-PR3, FL-PR4, FA-44GB-8ES, and FA-48GA are products made by Naito Densei Machida Mfg. Co., Ltd. (TEL +81-45-475-4191).

## A. 5 Debugging Tools (Hardware)

| IE-78K0S-NS <br> In-circuit emulator | In-circuit emulator for debugging hardware and software of an application system using the <br> $78 \mathrm{~K} / 0$ S Series. Supports the integrated debugger (ID78K0S-NS). Used in combination with an <br> AC adapter, emulation probe, and interface adapter for connecting the host machine. |
| :--- | :--- |
| IE-78K0S-NS-A <br> In-circuit emulator | The IE-78K0S-NS-A provides a coverage function in addition to the IE-78K0S-NS functions, thus <br> enhancing the debug functions, including the tracer and timer functions. |
| IE-70000-MC-PS-B <br> AC adapter | Adapter for supplying power from an AC 100 to 240 V outlet. |
| IE-70000-98-IF-C <br> Interface adapter | Adapter necessary when using a PC-9800 series PC (except notebook type) as the host machine <br> (C bus supported) |
| IE-70000-CD-IF-A <br> PC card interface | PC card and interface cable necessary when using a notebook PC as the host machine (PCMCIA <br> socket supported) |
| IE-70000-PC-IF-C <br> Interface adapter | Interface adapter necessary when using an IBM PC/AT compatible as the host machine (ISA bus <br> supported) |
| IE-70000-PCI-IF-A <br> Interface adapter | Adapter necessary when using a personal computer incorporating a PCI bus as the host machine |
| IE-789177-NS-EM1 <br> Emulation board | Board for emulating the peripheral hardware inherent to the device. Used in combination with an <br> in-circuit emulator. |
| NP-44GB-TQ <br> NP-H44GB-TQ <br> Emulation probe | Probe to connect the in-circuit emulator and target system. <br> Used in combination with the TGB-044SAP. |
| TGB-044SAP <br> Conversion <br> adapter | Conversion socket to connect the NP-44GB-TQ, NP-H44GB-TQ and a target system board on <br> which a 44-pin plastic LQFP (GB-8ES type) can be mounted. |
| NP-48GA <br> Emulation probe <br> TGA-048SDP <br> Conversion <br> adapter | Cable to connect the in-circuit emulator and target system. <br> Used in combination with the TGA-048SDP. |
| TQFP (fine pitch) (GA-9EU type) can be mounted |  |

Remarks 1. The NP-44GB-TQ and NP-H44GB-TQ are products made by Naito Densei Machida Mfg. Co., Ltd. (TEL +81-45-475-4191).
2. The TGB-044SAP and TGA-048SDP are products made by TOKYO ELETECH CORPORATION. For further information, contact: Daimaru Kogyo, Ltd.

Tokyo Electronics Department (TEL +81-3-3820-7112)
Osaka Electronics Department (TEL +81-6-6244-6672)

## A. 6 Debugging Tools (Software)

| ID78K0S-NS Integrated debugger | This debugger supports the in-circuit emulators IE-78KOS-NS and IE-78KOS-NS-A for the 78K/OS Series. The ID78K0S-NS is Windows-based software. <br> It has improved C-compatible debugging functions and can display the results of tracing with the source program using an integrating window function that associates the source program, disassemble display, and memory display with the trace result. Used in combination with a device file (DF789178) (sold separately). |
| :---: | :---: |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ ID78K0S - NS |
| SM78K0S <br> System simulator | This is a system simulator for the $78 \mathrm{~K} / 0$ S Series. The SM78KOS is Windows-based software. It can be used to debug the target system at $C$ source level or assembler level while simulating the operation of the target system on the host machine. <br> Using SM78K0S, the logic and performance of the application can be verified independently of hardware development. Therefore, the development efficiency can be enhanced and the software quality can be improved. in combination with a device file (DF789178) (sold separately). |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ SM78K0S |
| DF789178 ${ }^{\text {Note }}$ <br> Device file | File containing the information inherent to the device. <br> Used in combination with the RA78KOS, CC78K0S, ID78KOS-NS, and SM78KOS (all sold separately). |
|  | Part number: $\mu \mathrm{S} \times \times \times \times$ DF789178 |

Note DF789178 is a common file that can be used with the RA78K0S, CC78K0S, ID78K0S-NS, and SM78K0S.

Remark $X X X X$ in the part number differs depending on the operating system and supply medium to be used.
$\mu \mathrm{S} \times \times \times \times$ ID78K0S-NS
$\mu \mathrm{S} \times \times \times \times$ SM78K0S

| $x>x x$ | Host Machine | OS | Supply Medium |
| :---: | :---: | :---: | :---: |
| AB13 | PC-9800 series IBM PC/AT compatible | Japanese Windows | 3.5" 2HD FD |
| BB13 |  | English Windows |  |
| AB17 |  | Japanese Windows | CD-ROM |
| BB17 |  | English Windows |  |

## APPENDIX B NOTES ON TARGET SYSTEM DESIGN

The following shows the conditions when connecting the emulation probe to the conversion connector or conversion socket. Follow the configuration below and consider the shape of parts to be mounted on the target system when designing a system.

Figure B-1. Distance Between In-Circuit Emulator and Conversion Socket (NP-44GB-TQ)


Note Distance when NP-44GB-TQ is used. When NP-H44GB-TQ is used, the distance is 370 mm .

Remarks 1. NP-44GB-TQ and NP-H44GB-TQ are products of Naito Densei Machida Mfg. Co., Ltd.
2. TGB-044SAP is a product of TOKYO ELETECH CORPORATION.

Figure B-2. Connection Condition of Target System (NP-H44GB-TQ)


Remarks 1. NP-H44GB-TQ is a product of Naito Densei Machida Mfg. Co., Ltd.
2. TGB-044SAP is a product of TOKYO ELETECH CORPORATION.

Figure B-3. Distance Between In-Circuit Emulator and Conversion Socket (NP-48GA)


Remarks 1. NP-48GA is a product of Naito Densei Machida Mfg. Co., Ltd.
2. TGA-048SDP is a product of TOKYO ELETECH CORPORATION.

Figure B-4. Connection Condition of Target System (NP-48GA)


Remarks 1. NP-48GA is a product of Naito Densei Machida Mfg. Co., Ltd.
2. TGA-048SDP is a product of TOKYO ELETECH CORPORATION.

## APPENDIX C REGISTER INDEX

## C. 1 Register Name Index

16-bit capture register 90 (TCP90) ..... 118
16-bit compare register 90 (CR90) ..... 118
16-bit multiplication result storage register 0 (MULO) ..... 289
16-bit timer counter 90 (TM90) ..... 118
16-bit timer mode control register 90 (TMC90) ..... 119
8 -bit compare registers 80, 81, 82 (CR80, CR81, CR82) ..... 137
8-bit timer counter 80, 81, 82 (TM80, TM81, TM82) ..... 137
8 -bit timer mode control register 80 (TMC80) ..... 138
8 -bit timer mode control register 81 (TMC81) ..... 139
8 -bit timer mode control register 82 (TMC82) ..... 140
[A]
A/D conversion result register 0 (ADCRO) ..... 165, 178
A/D converter mode register 0 (ADMO) ..... 167, 180
A/D input selection register 0 (ADSO) ..... 168, 181
Asynchronous serial interface mode register 20 (ASIM20) ..... 195, 202, 205, 218
Asynchronous serial interface status register 20 (ASIS20) ..... 197, 206
[B]
Baud rate generator control register 20 (BRGC20) ..... 198, 207, 219
Buzzer output control register 90 (BZC90) ..... 121
[E]
External interrupt mode register 0 (INTMO) ..... 299
External interrupt mode register 1 (INTM1) ..... 300
[I]
Interrupt mask flag registers 0,1 (MKO, MK1) ..... 298
Interrupt request flag registers 0,1 (IF0, IF1) ..... 297
[M]
Multiplication data registers A0, B0 (MRA0, MRB0) ..... 289
Multiplier control register 0 (MULCO) ..... 291
[0]
Oscillation stabilization time selection register (OSTS). ..... 309
[P]
Port 0 (PO) ..... 87
Port 1 (P1) ..... 88
Port 2 (P2) ..... 89
Port 3 (P3) ..... 94
Port 5 (P5) ..... 97
Port 6 (P6) ..... 98
Port mode register 0 (PMO) ..... 99
Port mode register 1 (PM1) ..... 88, 99
Port mode register 2 (PM2) ..... 89, 99, 141
Port mode register 3 (PM3) ..... 99, 122, 141
Port mode register 5 (PM5) ..... 99
Processor clock control register (PCC) ..... 105
Pull-up resistor option register 0 (PU0) ..... 100
Pull-up resistor option registers B2, B3 (PUB2, PUB3) ..... 101
[R]
Reception buffer register 20 (RXB20) ..... 193
Reception shift register 20 (RXS20) ..... 193
[S]
Serial operation mode register 20 (CSIM20) ..... 194, 201, 204, 217
SMB clock selection register 0 (SMBCLO) ..... 239
SMB control register 0 (SMBC0) ..... 231
SMB input level setting register 0 (SMBVIO) ..... 243
SMB mode register 0 (SMBMO) ..... 241
SMB shift register 0 (SMB0) ..... 229, 244
SMB slave address register 0 (SMBSVA0) ..... 229, 244
SMB status register 0 (SMBSO) ..... 236
Subclock control register (CSS) ..... 107
Suboscillation mode register (SCKM) ..... 106
[T]
Timer clock selection register 2 (TCL2) ..... 160
Transmission shift register 20 (TXS20) ..... 193
[W]
Watch timer mode control register (WTM) ..... 155
Watchdog timer mode register (WDTM) ..... 161

## C. 2 Register Symbol Index

## [A]

ADCR0: A/D conversion result register 0 ..... 165, 178
ADMO: A/D converter mode register 0 ..... 167, 180
ADS0: A/D input selection register 0 ..... 168, 181
ASIM20: Asynchronous serial interface mode register 20 ..... 195, 202, 205, 218
ASIS20: Asynchronous serial interface status register 20. ..... 197, 206
[B]
BRGC20: Baud rate generator control register 20 ..... 198, 207, 219
BZC90: Buzzer output control register 90 ..... 121
[C]
CR80: 8-bit compare register 80 ..... 137
CR81: 8-bit compare register 81 ..... 137
CR82: 8-bit compare register 82 ..... 137
CR90: 16-bit compare register 90 ..... 118
CSIM20: Serial operation mode register 20 ..... 194, 201, 204, 217
CSS: Subclock control register ..... 107
[I]
IF0: Interrupt request flag register 0 ..... 297
IF1: Interrupt request flag register 1 ..... 297
INTM0: External interrupt mode register 0 ..... 299
INTM1: External interrupt mode register 1 ..... 300
[M]
MKO: Interrupt mask flag register 0 ..... 298
MK1: Interrupt mask flag register 1 ..... 298
MRAO: Multiplication data register AO ..... 289
MRBO: Multiplication data register B0 ..... 289
MULO: $\quad$ 16-bit multiplication result storage register 0 ..... 289
MULCO: Multiplier control register 0 ..... 291
[O]
OSTS: Oscillation stabilization time selection register ..... 309
[P]
P0: $\quad$ Port 0 ..... 87
P1: Port 1 ..... 88
P2: $\quad$ Port 2 ..... 89
P3: Port 3 ..... 94
P5: Port 5 ..... 97
P6: Port 6 ..... 98
PCC: Processor clock control register ..... 105
PMO: Port mode register 0 ..... 99
PM1: Port mode register 1 ..... 88, 99
PM2: Port mode register 2 ..... 89, 99, 141
PM3: Port mode register 3 ..... 99, 132, 141
PM5: Port mode register 5 ..... 99
PU0: Pull-up resistor option register 0 ..... 100
PUB2: Pull-up resistor option register B2 ..... 101
PUB3: Pull-up resistor option register B3 ..... 101
[R]
RXB20: Reception buffer register 20 ..... 193
RXS20: Reception shift register 20 ..... 193
[S]
SCKM: Suboscillation mode register ..... 106
SMB0: SMB shift register 0 ..... 229, 244
SMBCO: SMB control register 0 ..... 231
SMBCLO: SMB clock selection register 0 ..... 239
SMBMO: SMB mode register 0 ..... 241
SMBSO: SMB status register 0 ..... 236
SMBSVAO: SMB slave address register 0 ..... 229, 244
SMBVIO: SMB input level setting register 0 ..... 243
[T]
TCL2: Timer clock selection register 2 ..... 160
TCP90: 16-bit capture register 90 ..... 118
TM80: $\quad 8$-bit timer counter 80 ..... 137
TM81: $\quad 8$-bit timer counter 81 ..... 137
TM82: $\quad 8$-bit timer counter 82 ..... 137
TM90: 16-bit timer counter 90 ..... 118
TMC80: 8 -bit timer mode control register 80 ..... 138
TMC81: 8-bit timer mode control register 81 ..... 139
TMC82: 8-bit timer mode control register 82 ..... 140
TMC90: 16-bit timer mode control register 90 ..... 119
TXS20: Transmission shift register 20 ..... 193
[W]
WDTM: Watchdog timer mode register ..... 161
WTM: Watch timer mode control register ..... 155

## APPENDIX D REVISION HISTORY

## D. 1 Major Revisions in This Edition

| Page | Description |
| :---: | :---: |
| pp. 21-24 <br> pp. 26-28 | CHAPTER 1 GENERAL ( $\mu$ PD789167 AND 789177 SUBSERIES) <br> - Addition of lead-free products $\mu$ PD789166GB- $\times \times \times-8 E S-A, \mu$ PD789166GA-×××-9EU-A, $\mu$ PD789167GB- $-\times \times$-8ES-A, $\mu$ PD789167GA- $\times \times x-9 E U-A$, $\mu$ PD789176GB- $x \times x-8 E S-A, \mu$ PD789176GA-××x-9EU-A, $\mu$ PD789177GB- $\times \times \times-8 E S-A, \mu$ PD789177GA-×××-9EU-A, $\mu$ PD78F9177GB-8ES-A, $\mu$ PD78F9177AGB-8ES-A, $\mu$ PD78F9177AGA-9EU-A <br> - Update of 1.5 78K/OS Series Lineup |
| pp. 35-38 <br> p. 40 | CHAPTER 2 GENERAL ( $\mu$ PD789167Y AND 789177Y SUBSERIES) <br> - Addition of lead-free products $\mu$ PD789166YGB- $\times \times \times-8 E S-A, \mu$ PD789166YGA- $\times \times \times-9 E U-A$, $\mu$ PD789167YGB-×××-8ES-A, $\mu$ PD789167YGA-×××-9EU-A, $\mu$ PD789176YGB- $x \times x-8 E S-A, \mu$ PD789176YGA-×××-9EU-A, $\mu$ PD789177YGB-××x-8ES-A, $\mu$ PD789177YGA-××x-9EU-A, $\mu$ PD78F9177YGB-8ES-A, $\mu$ PD78F9177AYGA-9EU-A, $\mu$ PD78F9177AYGB-8ES-A, $\mu$ PD78F9177AYGA-9EU-A <br> - Update of 2.5 78K/OS Series Lineup |
| p. 51 | CHAPTER 3 PIN FUNCTIONS ( $\mu$ PD789167 AND 789177 SUBSERIES) <br> - Addition of descriptions in 3.2.15 VPp (flash memory version only) |
| p. 59 | CHAPTER 4 PIN FUNCTIONS ( $\mu$ PD789167Y AND 789177Y SUBSERIES) <br> - Addition of descriptions in 4.2.15 VPP (flash memory version only) |
| p. 144 | CHAPTER 9 8-BIT TIMER/EVENT COUNTERS 80 TO 82 <br> - Modification of Figure 9-10 External Event Counter Operation Timing (with Rising Edge Specified) |
| p. 444 | CHAPTER 32 RECOMMENDED SOLDERING CONDITIONS <br> - Addition of recommended soldering conditions for the lead-free products |

The mark $\star$ shows major revised points.

## D. 2 Revision History up to Previous Edition

Revisions up to the previous edition are shown below. The "Applied to" column indicates the chapter in each edition to which the revision was applied.
$(1 / 3)$

| Edition | Revision from Previous Edition | Applied to: |
| :---: | :---: | :---: |
| Second edition | Addition of description of $\mu$ PD789166Y, $\mu$ PD789167Y, $\mu$ PD789176Y, and $\mu$ PD789177Y | Throughout |
|  | Change of status of $\mu$ PD789166, $\mu$ PD789167, $\mu$ PD789176, and $\mu$ PD789177 from "under development" to "developed" |  |
|  | Addition of description of SMBO special function registers to Table 5-3 Special Function Registers | CHAPTER 5 CPU ARCHITECTURE |
|  | Modification of Figure 6-5 Block Diagram of P21 | CHAPTER 6 PORT FUNCTIONS |
|  | Addition of 8.5 Notes on Using 16-Bit Timer | CHAPTER 8 16-BIT TIMER |
|  | Addition of 15 SMB0 ( $\mu$ PD789167Y AND 789177Y SUBSERIES) | CHAPTER 15 SMBO ( $\mu$ PD789167Y AND 789177Y SUBSERIES) |
|  | Addition of description of SMBO interrupt to 17 INTERRUPT FUNCTIONS | CHAPTER 17 INTERRUPT FUNCTIONS |
|  | Addition of Figure 20-3 Flashpro III Connection in SMB Mode | CHAPTER $20 \mu$ PD78F9177 AND $\mu$ PD78F9177Y |
|  | Addition of setting with SMB mode in Table 20-4 Setting with PG-FP3 |  |
|  | Addition of development tools for $\mu$ PD789166Y, $\mu$ PD789167Y, $\mu$ PD789176Y, and $\mu$ PD789177Y | APPENDIX A DEVELOPMENT TOOLS |
| Third edition | - Addition of $\mu$ PD789166(A), 789167(A), 789176(A), 789177(A), 789166Y(A), 789167Y(A), 789176Y(A), 789177Y(A), 789166(A1), 789167(A1), 789176(A1), 789177(A1), 789166(A2), 789167(A2), 789176(A2), 789177(A2), 78F9177A, 78F9177AY, 78F9177A(A), 78F9177AY(A), and 78F9177A(A1) <br> - Addition of description on expanded-specification products ( 10 MHz ) | Throughout |
|  | - Addition of description on generic terms used in this manual <br> - Change of Related Documents | INTRODUCTION |
|  | - Addition of 1.1 Expanded-Specification Products and Conventional Products <br> - Addition of 1.10 Differences Between Standard Quality Grade Products and (A) Products, (A1) Products, and (A2) Products | CHAPTER 1 GENERAL ( $\mu$ PD789167 AND 789177 SUBSERIES) |
|  | - Addition of 2.1 Expanded-Specification Products and Conventional Products <br> - Addition of 2.10 Differences Between Standard Quality Grade Products and (A) Products | CHAPTER 2 GENERAL ( $\mu$ PD789167Y AND 789177Y SUBSERIES) |
|  | - Modification of VPP pin connection in 3.2.15 VPP (flash memory version only) and Table 3-1 Types of I/O Circuits for Each Pin and Recommended Connection of Unused Pins | CHAPTER 3 PIN FUNCTIONS ( $\mu$ PD789167 AND 789177 SUBSERIES) |
|  | - Addition of Note to Figure 7-3 Format of Suboscillation Mode Register | CHAPTER 7 CLOCK GENERATOR |
|  | - Modification of description in 8.4.1 Operation as timer interrupt <br> - Modification of description in 8.4.2 Operation as timer output | CHAPTER 8 16-BIT TIMER 90 |


| Edition | Revision from Previous Edition | Applied to: |
| :---: | :---: | :---: |
| Third edition | - Addition of 9.5 (4) Cautions when set to STOP mode <br> - Addition of 9.5 (5) Start timing of external event counter | CHAPTER 9 8-BIT TIMER/EVENT COUNTERS 80 TO 82 |
|  | - Addition of 12.5 (8) Input impedance of ANIO to ANI7 pins | CHAPTER 12 8-BIT A/D CONVERTER ( $\mu$ PD789167 AND 789167Y SUBSERIES) |
|  | - Modification of description in 13.2 (2) A/D conversion result register 0 (ADCRO) <br> - Modification of Figure 13-4 Basic Operation of 10-Bit A/D Converter <br> - Addition of 13.5 (8) Input impedance of ANIO to ANI7 pins | CHAPTER 13 10-BIT A/D CONVERTER ( $\mu$ PD789177 AND 789177Y SUBSERIES) |
|  | - Modification of Figure 14-1 Block Diagram of Serial Interface 20 <br> - Modification of description on PE20 flag in Figure 14-5 Format of Asynchronous Serial Interface Status Register 20 <br> - Addition of 14.4.2 (2) (f) Reading receive data | CHAPTER 14 SERIAL INTERFACE 20 |
|  | - Overall revision of description on flash memory programming | CHAPTER 20 FLASH MEMORY VERSION |
|  | - Addition of electrical specifications | CHAPTER 23, 25, 27, 29, and 31 ELECTRICAL SPECIFICATIONS |
|  | - Addition of characteristics curves | CHAPTER 24, 26, 28, 30, and 32 CHARACTERISTICS CURVES |
|  | - Addition of package drawings | CHAPTER 33 PACKAGE DRAWINGS |
|  | - Addition of recommended soldering conditions | CHAPTER 34 RECOMMENDED SOLDERING CONDITIONS |
|  | - Overall revision of description on development tools <br> - Deletion of embedded software | APPENDIX A DEVELOPMENT TOOLS |
| Fourth edition | Change of Related Documents | INTRODUCTION |
|  | - Deletion of SMB from block diagram in 1.8 | CHAPTER 1 GENERAL ( $\mu$ PD789167 AND 789177 SUBSERIES) |
|  | - Deletion of P60 to P67 from Table 6-3 Port Mode Register and Output Latch Settings for Using Alternate Functions | CHAPTER 6 PORT FUNCTIONS |
|  | - Modification of Figures 8-6 Timing of Timer Interrupt Operation and 88 Timer Output Timing | CHAPTER 8 16-BIT TIMER 90 |
|  | - Change of description of Cautions in 9.5 Notes on Using 8-Bit Timer/Event Counters 80 to 82 | CHAPTER 9 8-BIT TIMER/EVENT COUNTERS 80 TO 82 |
|  | - Modification of Notes in Figure 12-2 Format of A/D Converter Mode Register 0 | CHAPTER 12 8-BIT A/D CONVERTER ( $\mu$ PD789167 AND 789167Y SUBSERIES) |
|  | - Modification of Notes in Figure 13-2 Format of A/D Converter Mode Register 0 | CHAPTER 13 10-BIT A/D CONVERTER ( $\mu$ PD789177 AND 789177Y SUBSERIES) |
|  | - Modification of Figure 14-1 Block Diagram of Serial Interface 20 <br> - Modification of description of Cautions in Figure 14-6 Format of Baud Rate Generator Control Register 20 <br> - Addition of 14.3 (4) (c) Generation of serial clock from system clock input to 3 -wire serial I/O mode | CHAPTER 14 SERIAL INTERFACE 20 |


| Edition | Revision from Previous Edition | Applied to: |
| :---: | :---: | :---: |
| Fourth edition | - Addition of description of SMBMO to 15.4.1 Start condition <br> - Addition of description of 15.4 .7 (7) Slave operation (after stop mode is released) <br> - Modification of Table 15-3 INTSMB0 Generation Timing and Wait Control <br> - Addition of 15.4 .8 (6) Start condition detection <br> - Addition and modification of description of Notes in Figure 15-20 Master $\rightarrow$ Slave Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) and Figure 15-21. Slave $\rightarrow$ Master Communication Example (When 9-Clock Wait Is Selected for Both Master and Slave) | CHAPTER 15 SMBO ( $\mu$ PD789167Y AND 789177Y SUBSERIES) |
|  | - Addition of Caution in Figure 17-2 Format of Interrupt Request Flag Register | CHAPTER 17 INTERRUPT FUNCTIONS |
|  | - Modification of Table 20-2 Communication Mode List and Table 203 Pin Connection List <br> - Addition of description of pseudo 3-wire mode to Figure 20-3 Example of Connection with Dedicated Flash Programmer and Figure 20-9 Wiring Example for Flash Writing Adapter in 3-Wire Serial I/O Mode (SIO-ch1) or Pseudo 3-Wire Mode | CHAPTER 20 FLASH MEMORY VERSION |
|  | - Modification of electrical specifications | CHAPTERS 23, 25, 27, 28, 30 ELECTRICAL SPECIFICATIONS |
|  | - Addition of chapter | APPENDIX B NOTES ON TARGET SYSTEM DESIGN |
| Fifth edition | Addition of 48-pin plastic TQFP (fine pitch) $(7 \times 7)$ to $\mu$ PD789167, 789177 Subseries <br> $\mu$ PD789166GA-×××-9EU, $\mu$ PD789167GA-×××-9EU, $\mu$ PD789176GA-×××-9EU, $\mu$ PD789177GA-×××-9EU, $\mu$ PD78F9177AGA-9EU | Throughout |
|  | - Addition of description on IC3 to 3.1 (2) Non-port pins <br> - Addition of 3.2.17 IC3 <br> - Addition of description on IC3 to Table 3-1 Types of I/O Circuits for Each Pin and Recommended Connection of Unused Pins | CHAPTER 3 PIN FUNCTIONS ( $\mu$ PD789167 AND 789177 SUBSERIES) |

