# **Audio ICs**

# Fluorescent display tube level meter driver, 16-point × 2 channel, VU scale, bar display BA6803S

The BA6803S is a two-channel, 16-point fluorescent display tube driver for VU-scale bar-level meters. It uses a dynamic-drive system and has both AC and DC inputs.

The AC input mode has a peak hold circuit. The IC features a power-on mute, and the output block can directly drive fluorescent display tubes, so few external components are required.

The grid output duty cycle is 1 / 8.

### Applications

Level meters for all types of AV equipment

### Features

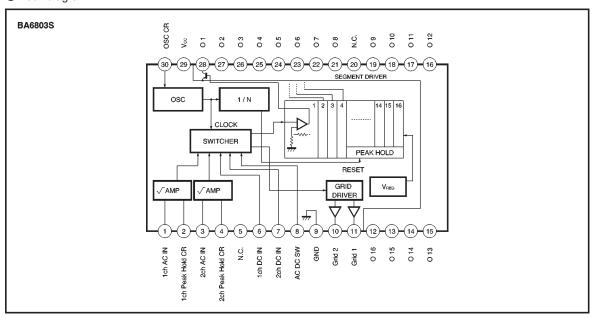
- 1) Uses dynamic-drive system to display two 16-point channels. 30-pin SDIP package.
- 2) AC and DC inputs provided. Switching function allows two-mode display.
- 3) Upper 12 points have peak hold function in AC mode (two seconds typ.).
- 4) Power-on mute function.
- Dynamic-drive system reduces the power dissipation of the fluorescent display tube power supply.
- 6) Square root compression amplifier built in.

### ●Absolute maximum ratings (Ta = 25°C)

Parameter	Symbol	Limits	Unit
Power supply voltage	Vcc	7.0	V
Power dissipation	Pd	1200*1	mW
Operating temperature	Topr	<b>−20~+70</b>	°C
Storage temperature	Tstg	<b>−55</b> ~ <b>+125</b>	°C
Output voltage	V <sub>CC</sub> +V <sub>EE</sub>	36	V

\*1 Reduced by 12mW for each increase in Ta of 1°C over 25°C.

# Block diagram



# ●Electrical characteristics (unless otherwise noted, Ta = 25°C)

Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Operating voltage range	Vcc	4.5	5.0	5.8	٧	
Quiescent current	lα	_	17	24	mA	_
AC input resistance	RINAC	175	250	325	Ω	Pins 1 and 3
DC input resistance	RINDC	7	10	13	kΩ	_
Oscillator frequency	fosc	1.7	2.0	2.3	kHz	C=0.015 $\mu$ F, R=30kΩ
Peak hold time	Thold	_	2	_	s	fosc=2kHz
Output duty cycle	Duty	_	1/8	_	_	fosc=2kHz
Grid output low level voltage	VgL	_	0.4	0.8	٧	I <sub>G</sub> =5mA
Grid output leakage current	lG leak	_	_	10	μA	Vcc=5V

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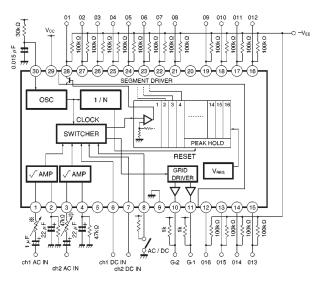
Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Segment output high level voltage	Vон	3.7	4.0	_	٧	lo=2mA
Segment output leakage current	Oleak	_	_	10	μΑ	V <sub>EE</sub> =-31V
Input switching threshold	Vтн	2.2	2.5	2.8	V	
AC sensitivity	VINAC	250	400	630	mV	R <sub>IN</sub> = 1.5kΩ AC comparator 11 on level
Maximum grid output current	Ідм	5	_	_	mA	V <sub>OL</sub> =0.8V
Maximum segment output current	Іом	2	_	_	mA	V <sub>OH</sub> =3.7V
AC comparator level 16	VC16AC	9.0	10	12	dB	Pin 12 output
AC comparator level 15	VC15AC	7.0	8	9	dB	Pin 13 output
AC comparator level 14	VC14AC	5.0	6	7.0	dB	Pin 14 output
AC comparator level 13	VC13AC	3.0	4	5.0	dB	Pin 15 output
AC comparator level 12	VC12AC	1.0	2	3.0	dB	Pin 16 output
AC comparator level 11	VC11AC	_	0	_	dB	Pin 17 output
AC comparator level 10	VC10AC	-3.0	-2	-1.0	dB	Pin 18 output
AC comparator level 9	V <sub>C9AC</sub>	-5.0	-4	-3.0	dB	Pin 19 output
AC comparator level 8	VC8AC	-7.0	-6	-5.0	dB	Pin 21 output
AC comparator level 7	V <sub>C7AC</sub>	-9.0	-8	-7.0	dB	Pin 22 output
AC comparator level 6	V <sub>C6AC</sub>	-13	-11	-9.0	dB	Pin 23 output
AC comparator level 5	V <sub>C5AC</sub>	-17.5	<b>—15</b>	-13	dB	Pin 24 output
AC comparator level 4	V <sub>C4AC</sub>	-22.5	-20	-17.5	dB	Pin 25 output
AC comparator level 3	Vсзас	-27	-25	-22.5	dB	Pin 26 output
AC comparator level 2	Vc2AC	-34	-30	-27	dB	Pin 27 output
AC comparator level 1	VC1AC	-50	-40	-34	dB	Pin 28 output
DC comparator level 16	VC16DC	2.84	3.18	3.52	٧	Pin 12 output
DC comparator level 15	Vc15DC	2.55	2.87	3.19	٧	Pin 13 output
DC comparator level 14	VC14DC	2.27	2.55	2.83	٧	Pin 14 output
DC comparator level 13	V <sub>C13DC</sub>	1.98	2.26	2.52	٧	Pin 15 output
DC comparator level 12	VC12DC	1.79	2.03	2.27	٧	Pin 16 output
DC comparator level 11	VC11DC	1.56	1.82	2.08	٧	Pin 17 output
DC comparator level 10	Vc10DC	1.31	1.64	1.97	٧	Pin 18 output
DC comparator level 9	Vcenc	1.15	1.49	1.83	٧	Pin 19 output
DC comparator level 8	Vcsdc	1.00	1.33	1.66	٧	Pin 21 output
DC comparator level 7	Vc7DC	0.95	1.19	1.43	٧	Pin 22 output
DC comparator level 6	Vcedc	0.80	1.03	1.26	٧	Pin 23 output
DC comparator level 5	Vcsdc	0.60	0.82	1.02	٧	Pin 24 output
DC comparator level 4	V <sub>C4DC</sub>	0.45	0.63	0.81	٧	Pin 25 output
DC comparator level 3	VсзDC	0.33	0.48	0.63	٧	Pin 26 output
DC comparator level 2	V <sub>C2DC</sub>	0.25	0.37	0.49	٧	Pin 27 output
DC comparator level 1	Vc1DC	0.10	0.20	0.30	٧	Pin 28 output

Notes: (1) The value given above for AC comparator level is measured after 0dB point adjustment.

(2) DC comparator level Vc (n) DC (Max.) > Vc (n+1) DC (Min.), but when the nth comparator is off, the (n+1)th comparator is never on.



### Measurement circuit



#VR of  $3k\Omega$  or more.

Fig. 1

# Application example

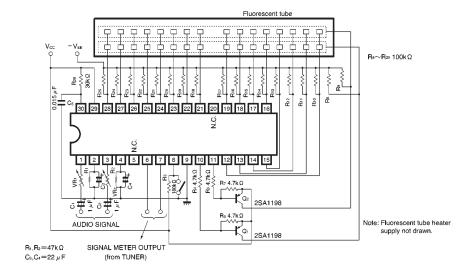


Fig. 2

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### Circuit operation

### (1) Input block

The AC input pins are pins 1 and 3, and the DC input pins are pins 6 and 7. Pin 8 is used to switch between the AC and DC inputs. When the input to pin 8 is "H", AC input is selected (pins 1 and 3).

Therefore, by using pin 8 to switch between the AC and DC modes, the IC can do two jobs, using one fluorescent tube. For example, pins 1 and 3 can be used for audio signal input, and pins 6 and 7 can be used as the input for the signal meter output from a tuner (DC). The AC input impedance of pins 1 and 3 is a low 250 $\Omega$  (typ.), so connect potentiometers (VR<sub>1</sub> and VR<sub>2</sub>) in series with the inputs to adjust the sensitivity and ch1 and ch2 balance.

### (2) Peak hold circuit

The BA6803S features a peak hold circuit that temporarily holds peak signal levels in AC input mode.

The peak hold function can be used with the upper 12 points (5 to 16). The peak hold time depends on the oscillator frequency. It is 2 sec. (typ.) for an oscillator frequency of 2kHz.

DC mode does not have a peak hold function.

### (3) Grid output

The pin 10 and 11 grid outputs are open-collector NPN transistors. The logic is active low (the fluorescent tube lights when the output is "L"), so connect two PNP transistors  $Q_1$  and  $Q_2$  as shown in the application example circuit to drive the fluorescent tubes (see Fig. 3).

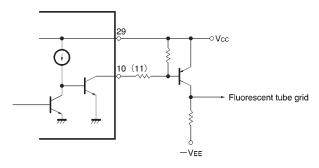


Fig. 3

## (4) Segment output block

Pins 12 to 28 are the segment outputs. The output circuits are open-collector PNP transistors. When grid 1 is "L", the ch1 level is output (pin 1 or 6 input level), and when grid 2 is "L", the ch2 level is output (pin 3 or 7 input level). Refer to Fig. 4.

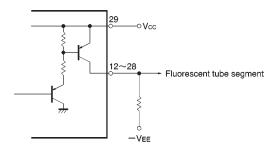


Fig. 4

(5) Grid and segment output timing chart. The grid and segment output timing is shown in Fig. 5.

### (6) Attack and release times

The response characteristic for AC input signals is set by resistor  $R_1$  and capacitor  $C_3$  for ch1 and resistor  $R_2$  and capacitor  $C_4$  for ch2 (pins 2 and 4). When  $R_1 = 47k\Omega$  and  $C_3 = 22\mu F$ , the attack time is about 4ms, and the release time is about 1sec. (same for ch2).

Attack time : Time for the voltage on pins 2 and 4

to rise from 1V to 2.5V when the input goes from no input to 2.6Vrms,

then back to no input.

Release time : Time for the voltage on pins 2 and 4  $\,$ 

to fall from 2.5V to 1V when the input goes from 2.6Vrms to no input.

# (7) Oscillator frequency

The resistor  $R_{26}$  and capacitor  $C_5$  connected to pin 30 determine the oscillator frequency. The oscillator frequency ( $f_{\text{osc}}$ ) and grid output period (T) are related as follows: T (ms) = 16 /  $f_{\text{osc}}$  (kHz)

### Timing chart

BA6803S (when oscillator frequency is 2kHz)

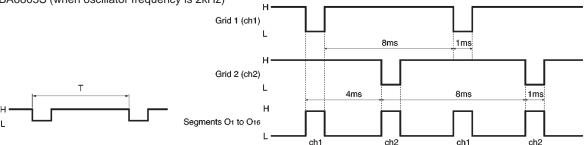


Fig. 5 (a)

Fig. 5 (b)

Attached components (refer to "Circuit operation")

C<sub>1</sub> and C<sub>2</sub>: input coupling capacitors.

VR1 and VR2: AC sensitivity adjustment and balance adjustment ( $3k\Omega$  recommended).

 $C_3$ ,  $R_1$ ,  $C_4$  and  $R_2$ : set the response characteristics with respect to the AC input signal. In the example given, the attack time is about 4mS and the release time is about

R<sub>3</sub>: Pullup resistor for the input switching terminal (pin 8).  $R_6$  and  $R_7$ : resistor for the grid leak current path (IGleak). Set so that IGleak  $x R_6 (R_7) < 0.6V$ .

 $R_4$  and  $R_5$ : base bias resistors for  $Q_1$  and  $Q_2$ .

Conditions for base bias current (IB) flow are Vcc = 5V and  $V_F = 0.6V$ :

$$\frac{R_4}{R_6} < \frac{5 - 0.6}{0.6} = 7.3$$

the base current is given by the following formula.

$$I_{B} \; (mA) \;\; \stackrel{\centerdot}{=} \;\; \frac{5-0.6}{R_{4} \; (k\Omega)} - \frac{0.6}{R_{6} \; (k\Omega)}$$

Set resistors R4 and R6 (R5 and R7) so that

Q1 and Q2: grid output inverting transistors. Use transistors for which  $V_{CEO} > V_{CC} + V_{EE}$ .

R<sub>8</sub> to R<sub>25</sub>: Resistors that reverse bias the segments and grid when the fluorescent tube is not lit. The application example given is for general cases. Select the resistors to suit the characteristics of the fluorescent tube used.

C₅ and R₂6: set the oscillator frequency. Capacitor C₅ should be a component with good temperature characteristics.

- Operation notes
- 1) Adjust the potentiometers  $VR_1$  and  $VR_2$  (connected to pins 1 and 3) to adjust the 0dB input level and the dispersion of ch1 and ch2.
- 2) The temperature characteristic for the lighting limit for the 16th LED is shown in Fig. 6.
- 3) The external resistor R<sub>26</sub> connected to the oscillator (pin 30) should be in the range of  $20k\Omega$  and  $100k\Omega$ . If it is outside this range, oscillation may stop due to the influence of temperature (see Fig. 7).

### Electrical characteristics curves

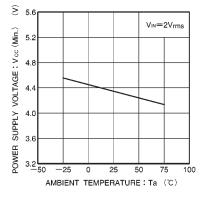


Fig. 6 16th point lighting limit power supply voltage vs. ambient temperature

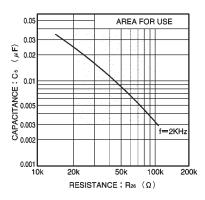


Fig. 7 Value of external components for oscillator  $(V_{cc} = 5.0V)$ 

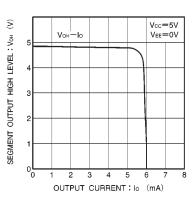


Fig. 8 Segment high level output vs. output current

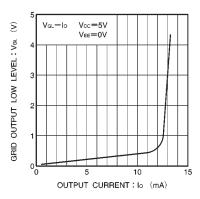


Fig. 9 Grid lowlevel output vs. output current

# External dimensions (Units: mm)

