General Description

The MAX941/MAX942/MAX944 are single/dual/quad high-speed comparators optimized for systems powered from a 3V or 5V supply. These devices combine high speed, low power, and Rail-to-Rail[®] inputs. Propagation delay is 80ns, while supply current is only 350µA per comparator.

The input common-mode range of the MAX941/ MAX942/MAX944 extends beyond both power-supply rails. The outputs pull to within 0.4V of either supply rail without external pullup circuitry, making these devices ideal for interface with both CMOS and TTL logic. All input and output pins can tolerate a continuous shortcircuit fault condition to either rail.

Internal hysteresis ensures clean output switching, even with slow-moving input signals. The MAX941 features latch enable and device shutdown.

The single MAX941 and dual MAX942 are offered in a tiny μ MAX package. Both the single and dual MAX942 are available in 8-pin DIP and SO packages. The quad MAX944 comes in 14-pin DIP and narrow SO packages.

Applications

3V/5V Systems

Battery-Powered Systems

Threshold Detectors/Discriminators

Line Receivers

Zero-Crossing Detectors

Sampling Circuits

Features

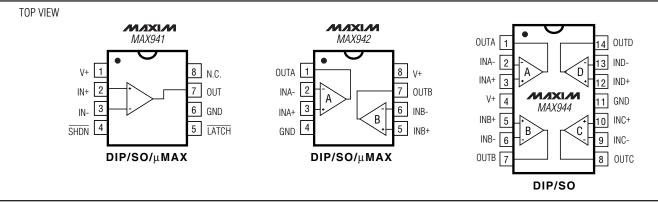
- ♦ Available in µMAX Package
- Optimized for 3V and 5V Applications (operation down to 2.7V)
- ♦ Fast, 80ns Propagation Delay (5mV overdrive)
- ♦ Rail-to-Rail Input Voltage Range
- Low Power: 1mW Power Dissipation per Comparator (3V) 350µA Supply Current
- Low, 1mV Offset Voltage
- Internal Hysteresis for Clean Switching
- Outputs Swing 200mV of Power Rails
- ♦ CMOS/TTL-Compatible Outputs
- Output Latch (MAX941 only)
- Shutdown Function (MAX941 only)

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX941CPA	0°C to +70°C	8 Plastic DIP
MAX941CSA	0°C to +70°C	8 SO
MAX941EPA	-40°C to +85°C	8 Plastic DIP
MAX941ESA	-40°C to +85°C	8 SO
MAX941EUA	-40°C to +85°C	8 µMAX

Ordering Information continued at end of data sheet.

Pin Configurations



Rail-to-Rail is a registered trademark of Nippon Motorola Ltd.

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ABSOLUTE MAXIMUM RATINGS

Power-Supply Ranges

i ener euppij i langee	
Supply Voltage V+ to GND	+6.5V
Differential Input Voltage	
Common-Mode Input Voltage	0.3V to (V+ + 0.3V)
LATCH Input (MAX941 only)	0.3V to (V+ + 0.3V)
SHDN Control Input (MAX941 only).	0.3V to (V+ + 0.3V)
Continuous Power Dissipation ($T_A = +7$	70°C)
8-Pin Plastic DIP (derate 9.09mW/°C	above +70°C)727mW
8-Pin SO (derate 5.88mW/°C above	+70°C)471mW

14-Pin Plastic DIP (derate 10.00mW/°C above +70°C)..800mW 14-Pin SO (derate 8.33mW/°C above +70°C)......667mW Operating Temperature Ranges

MAX94_C	0°C to +70°C
MAX94_E	-40°C to +85°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

(V+ = 2.7V to 6.0V, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 14)

PARAMETER	SYMBOL	CONDITIONS			MIN	TYP	MAX	UNITS
Positive Supply Voltage	V+				2.7		6.0	V
Input Voltage Range	VCMR	(Note 1)			-0.2		V+ + 0.2	V
Input-Referred Trip Points	Vtrip	V _{CM} = 0 or V _{CM} = V+ (Note 2)	T _A = +25°C	MAX94_C, MAX94_EP_, MAX94_ES_		1	3	mV
				MAX941EUA/MAX942EUA		1	4	
			TA = TMIN to T _{MAX}	MAX94_C, MAX94_EP_, MAX94_ES_			4	mV
				MAX941EUA/MAX942EUA			6	
	Vos V	VCM = 0 or VCM = V+ (Note 3)	T _A = +25°C	MAX94_C, MAX94_EP_, MAX94_ES_		1	2	mV
Input Offset Voltage				MAX941EUA/MAX942EUA		1	3	
input Onset voltage			$T_A = T_{MIN}$ to T_{MAX}	MAX94_C, MAX94_EP_, MAX94_ES_			3	mV
				MAX941EUA/MAX942EUA			5.5	
Input Bias Current	Ι _Β	V _{IN} = V _{OS} , V	/CM = 0 or	MAX94_C		150	300	nA
		V _{CM} = V+ (Note 4) MAX94_E			150	400	ПА	
Input Offset Current	los	$V_{IN} = V_{OS}, V_{CM} = 0 \text{ or } V_+$				10	100	nA
Common-Mode Rejection Ratio	CMRR			MAX94_C, MAX94_EP_, MAX94_ES_		80	300	μV/V
				MAX941EUA/MAX942EUA		80	800	
Power-Supply Rejection Ratio	PSRR	$2.7V \le V + \le V_{CM} = 0$	2.7V \leq V+ \leq 6.0V, MAX94_C_ , MAX94_EP_, MAX94_ES_			80	300	μV/V
		VCM = 0		MAX941EUA/MAX942EUA		80	350	
Output High Voltage	Mou	ISOURCE = 400µA		V+ - 0.4	V+ - 0.2	2	V	
	Vон	Isource = 4mA			V+ - 0.4	V+ - 0.3	3	v
Output Low Voltage	Vol	ISINK = 400µA			0.2	0.4	- V	
	VOL	I _{SINK} = 4mA				0.3		0.4
Output Leakage Current	ILEAK	(Note 6)					1	μA

ELECTRICAL CHARACTERISTICS (continued)

 $(V + = 2.7V \text{ to } 6.0V, T_A = T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A = +25^{\circ}C.)$ (Note 14)

PARAMETER	SYMBOL		CONDITIONS	MIN	TYP	MAX	UNITS
Supply Current per Comparator	Icc	V+ = 3V	MAX941		380	600	μΑ
		V+=3V	MAX942/MAX944		350	500	
		V+ = 5V	MAX941		430	700	
			MAX942/MAX944		400	600	
		MAX941 only, shutdown mode $(V + = 3V)$			12	60	
Power Dissipation per		(Note 7)	MAX941		1.0	4.2	- mW
Comparator	PD		MAX942/MAX944		1.0	3.6	
Propagation Dalay	tPD+,	(Note 8)	MAX94_C		80	150	– ns
Propagation Delay	t _{PD-}		MAX94_E		80	200	
Differential Propagation Delay	dtpD	(Note 9)			10		ns
Propagation Delay Skew		(Note 10)			10		ns
Logic Input Voltage High	VIH	(Note 11)		$\frac{V_{+}}{2}$ + 0.4	$\frac{V+}{2}$		V
Logic Input Voltage Low	VIL	(Note 11)			<u>V+</u> 2	$\frac{V_{+}}{2}$ - 0.4	V
Logic Input Current	IIL, IIH	V _{LOGIC} = 0 or V+ (Note 11)			2	10	μA
Data-to-Latch Setup Time	ts	(Note 12)			20		ns
Latch-to-Data Hold Time	tH	(Note 12)			30		ns
Latch Pulse Width	tLPW	MAX941 only			50		ns
Latch Propagation Delay	tLPD	MAX941 only			70		ns
Shutdown Time		(Note 13)		3		μs	
Shutdown Disable Time		(Note 13)		10		μs	

Note 1: Inferred from the CMRR test. Note also that either or both inputs can be driven to the absolute maximum limit (0.3V beyond either supply rail) without damage or false output inversion.

Note 2: The input-referred trip points are the extremities of the differential input voltage required to make the comparator output change state. The difference between the upper and lower trip points is equal to the width of the input-referred hysteresis zone (see Figure 1).

Note 3: Vos is defined as the center of the input-referred hysteresis zone (see Figure 1).

Note 4: The polarity of I_B reverses direction as V_{CM} approaches either supply rail. See *Typical Operating Characteristics* for more detail.

Note 5: Specified over the full common-mode range (V_{CMR}).

Note 6: Applies to the MAX941 only when in shutdown mode. Specification is for current flowing into or out of the output pin for V_{OUT} driven to any voltage from V+ to GND.

Note 7: Typical power dissipation specified with V + = 3V; maximum with V + = 6V.

Note 8: Parameter is guaranteed by design and specified with $V_{OD} = 5mV$ and $C_{LOAD} = 15pF$ in parallel with 400μ A of sink or source current. V_{OS} is added to the overdrive voltage for low values of overdrive (see Figure 2).

Note 9: Specified between any two channels in the MAX942/MAX944.

Note 10: Specified as the difference between $\underline{t_{PD+}}$ and $\underline{t_{PD-}}$ for any one comparator.

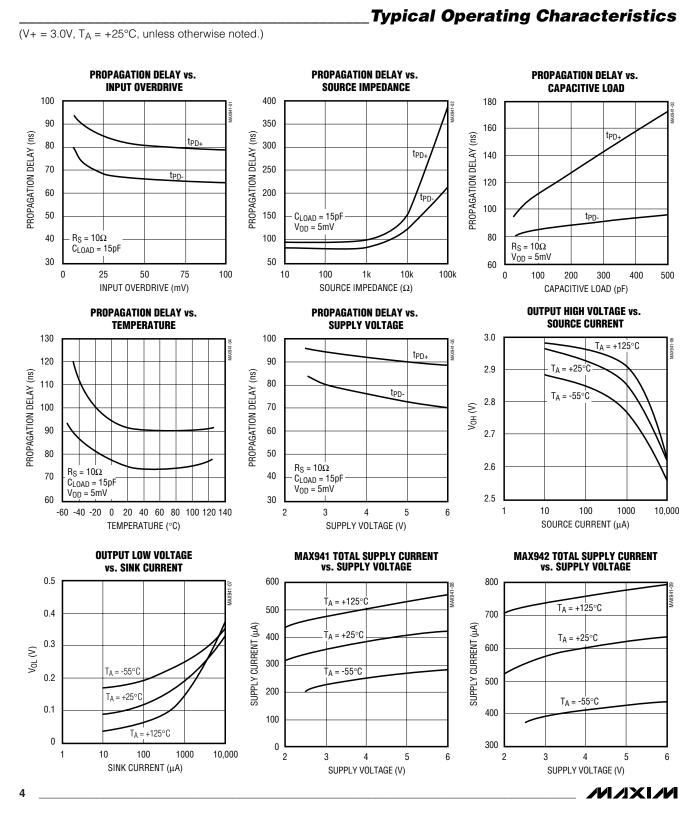
Note 11: Applies to the MAX941 only for both SHDN and LATCH pins.

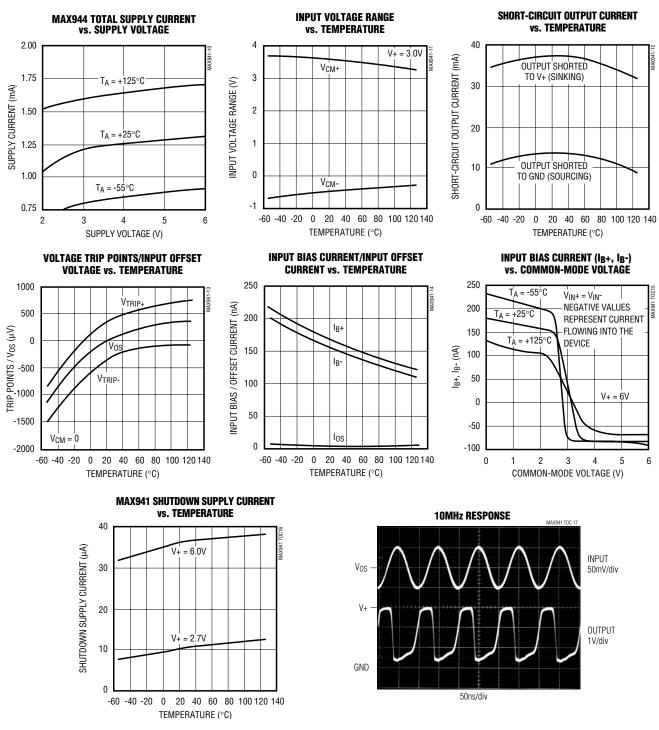
Note 12: Applies to the MAX941 only. Comparator is active with LATCH pin driven high and is latched with LATCH pin driven low (see Figure 2).

Note 13: Applicable to the MAX941 only. Comparator is active with SHDN pin driven high and is in shutdown with SHDN pin driven low. Shutdown disable time is the delay when SHDN is driven high to the time the output is valid.

Note 14: The MAX941EUA and MAX942EUA are 100% production tested at $T_A = +25^{\circ}C$. Specifications over temperature are guaranteed by design.

///XI///

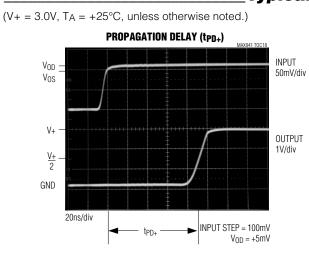


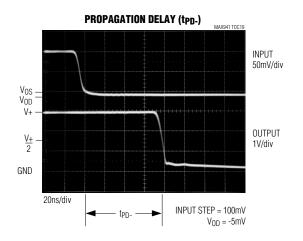


Typical Operating Characteristics (continued)

(V+ = 3.0V, T_A = +25°C, unless otherwise noted.)

MAX941/MAX942/MAX944





Typical Operating Characteristics (continued)

Pin Description

PIN			FUNCTION			
MAX941	MAX942	MAX944	NAME	FUNCTION		
	1	1	OUTA	Comparator A Output		
_	2	2	INA-	Comparator A Inverting Input		
_	3	3	INA+	Comparator A Noninverting Input		
1	8	4	V+	Positive Supply (V+ to GND must be ≤ 6.5 V)		
_	5	5	INB+	Comparator B Noninverting Input		
_	6	6	INB-	Comparator B Inverting Input		
	7	7	OUTB	Comparator B Output		
	_	8	OUTC	Comparator C Output		
_		9	INC-	Comparator C Inverting Input		
_		10	INC+	Comparator C Noninverting Input		
6	4	11	GND	Ground		
		12	IND+	Comparator D Noninverting Input		
_		13	IND-	Comparator D Inverting Input		
	_	14	OUTD	Comparator D Output		
2	_	—	IN+	Noninverting Input		
3	_	—	IN-	Inverting Input		
4	_	_	SHDN	Shutdown: MAX941 is active when SHDN is driven high; MAX941 is in shutdown when SHDN is driven low.		
5	_	_	LATCH	The output is latched when LATCH is low. The latch is transparent when LATCH is high.		
7	_	—	OUT	Comparator Output		
8	—	—	N.C.	No Connection. Not internally connected.		

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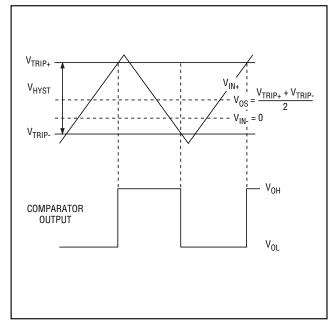


Figure 1. Input and Output Waveform, Noninverting Input Varied

Detailed Description

The MAX941/MAX942/MAX944 single-supply comparators feature internal hysteresis, high speed, and low power. Their outputs are guaranteed to pull within 0.4V of either supply rail without external pullup or pulldown circuitry. Rail-to-rail input voltage range and low-voltage single-supply operation make these devices ideal for portable equipment. The MAX941/MAX942/ MAX944 interface directly to CMOS and TTL logic.

Timing

Most high-speed comparators oscillate in the linear region because of noise or undesired parasitic feedback. This tends to occur when the voltage on one input is at or equal to the voltage on the other input. To counter the parasitic effects and noise, the MAX941/ MAX942/MAX944 have internal hysteresis.

The hysteresis in a comparator creates two trip points: one for the rising input voltage and one for the falling input voltage (Figure 1). The difference between the trip points is the hysteresis. When the comparator's input voltages are equal, the hysteresis effectively causes one comparator input voltage to move quickly past the other, thus taking the input out of the region where oscillation occurs. Standard comparators require hysteresis to be added with external resistors. The MAX941/MAX942/MAX944's fixed internal hysteresis

eliminates these resistors and the equations needed to determine appropriate values.

Figure 1 illustrates the case where IN- is fixed and IN+ is varied. If the inputs were reversed, the figure would look the same, except the output would be inverted.

The MAX941 includes an internal latch that allows storage of comparison results. The LATCH pin has a high input impedance. If LATCH is high, the latch is transparent (i.e., the comparator operates as though the latch is not present). The comparator's output state is stored when LATCH is pulled low. All timing constraints must be met when using the latch function (Figure 2).

Shutdown Mode (MAX941 Only)

The MAX941 shuts down when \overline{SHDN} is low. When shut down, the supply current drops to less than 60µA, and the three-state output becomes high impedance. The \overline{SHDN} pin has a high input impedance. Connect \overline{SHDN} to V+ for normal operation. Exit shutdown with \overline{LATCH} high; otherwise, the output will be indeterminate.

Input Stage Circuitry

The MAX941/MAX942/MAX944 include internal protection circuitry that prevents damage to the precision input stage from large differential input voltages. This protection circuitry consists of two back-to-back diodes between IN+ and IN- as well as two 4.1k Ω resistors (Figure 3). The diodes limit the differential voltage applied to the internal circuitry of the comparators to be no more than 2V_F, where V_F is the forward voltage drop of the diode (about 0.7V at +25°C).

For a large differential input voltage (exceeding 2VF), this protection circuitry increases the input bias current at IN+ (source) and IN- (sink).

Input Current =
$$\frac{(IN+ - IN-) - 2V_F}{2 \times 4.1 k\Omega}$$

Input current with large differential input voltages should not be confused with input bias current (I_B). As long as the differential input voltage is less than $2V_F$, this input current is equal to I_B. The protection circuitry also allows for the input common-mode range of the MAX941/MAX942/MAX944 to extend beyond both power-supply rails. The output is in the correct logic state if one or both inputs are within the common-mode range.

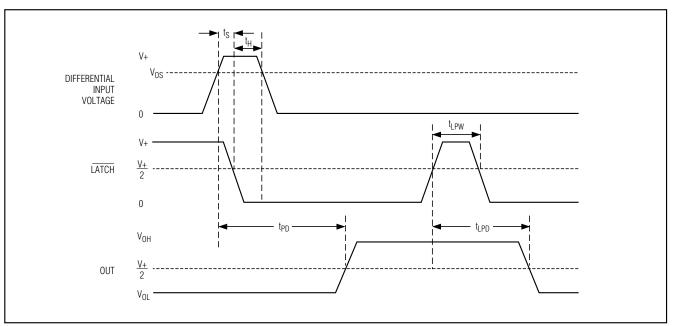


Figure 2. MAX941 Timing Diagram with Latch Operator

Output Stage Circuitry

The MAX941/MAX942/MAX944 contain a current-driven output stage as shown in Figure 4. During an output transition, ISOURCE or ISINK is pushed or pulled to the output pin. The output source or sink current is high during the transition, creating a rapid slew rate. Once the output voltage reaches V_{OH} or V_{OL} , the source or sink current decreases to a small value, capable of maintaining the V_{OH} or V_{OL} static condition. This significant decrease in current conserves power after an output transition has occurred.

One consequence of a current-driven output stage is a linear dependence between the slew rate and the load capacitance. A heavy capacitive load will slow down a voltage output transition. This can be useful in noisesensitive applications where fast edges may cause interference.

Applications Information

Circuit Layout and Bypassing

The high gain bandwidth of the MAX941/MAX942/ MAX944 requires design precautions to realize the comparators' full high-speed capability. The recommended precautions are:

- 1) Use a printed circuit board with a good, unbroken, low-inductance ground plane.
- 2) Place a decoupling capacitor (a 0.1μ F ceramic capacitor is a good choice) as close to V+ as possible.
- 3) Pay close attention to the decoupling capacitor's bandwidth, keeping leads short.
- 4) On the inputs and outputs, keep lead lengths short to avoid unwanted parasitic feedback around the comparators.
- 5) Solder the device directly to the printed circuit board instead of using a socket.

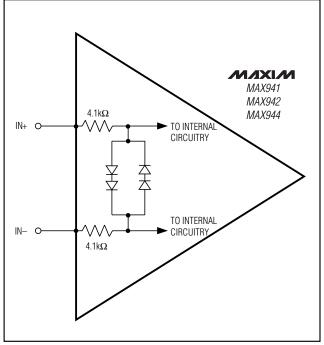


Figure 3. Input Stage Circuitry

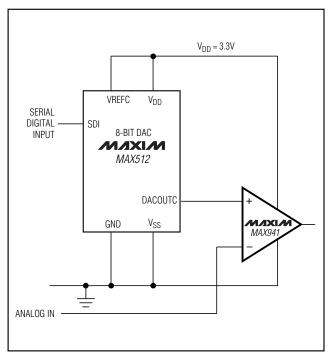


Figure 5. 3.3V Digitally Controlled Threshold Detector

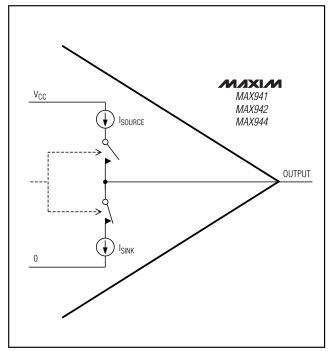


Figure 4. Output Stage Circuitry

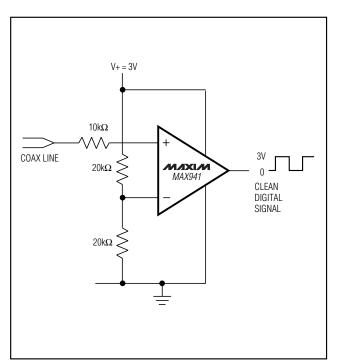


Figure 6. Line Transceiver Application

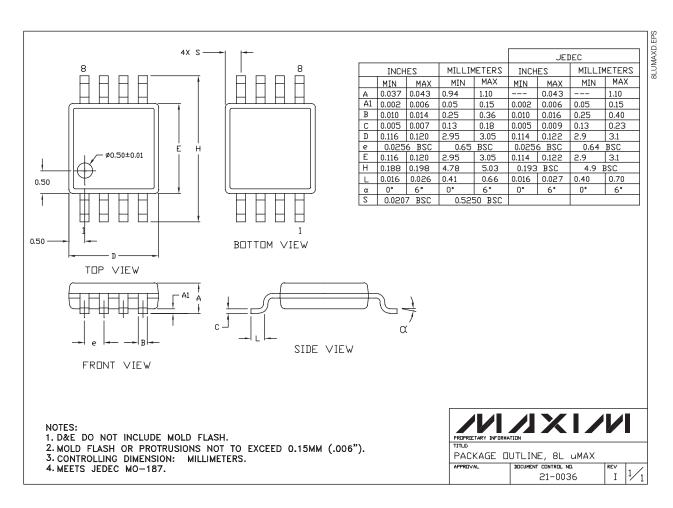
PART TEMP RANGE **PIN-PACKAGE** MAX942CPA 0°C to +70°C 8 Plastic DIP MAX942CSA 0°C to +70°C 8 SO MAX942EPA -40°C to +85°C 8 Plastic DIP MAX942ESA -40°C to +85°C 8 SO MAX942EUA -40°C to +85°C 8 μΜΑΧ 14 Plastic DIP MAX944CPD 0°C to +70°C 14 SO MAX944CSD 0°C to +70°C -40°C to +85°C MAX944EPD 14 Plastic DIP MAX944ESD -40°C to +85°C 14 SO

Ordering Information (continued)

Chip Information

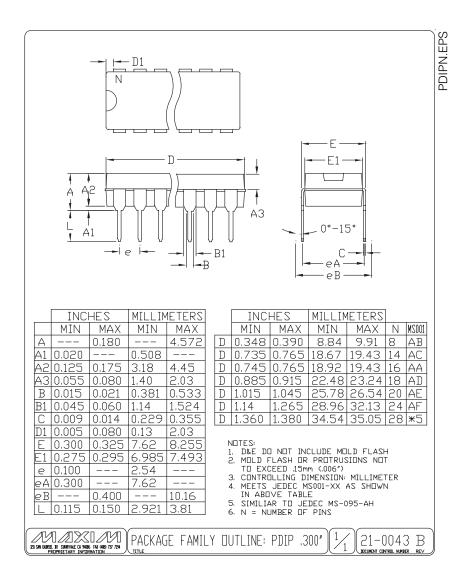
MAX941 TRANSISTOR COUNT: 192 MAX942 TRANSISTOR COUNT: 314 MAX944 TRANSISTOR COUNT: 620 PROCESS: BiPolar

Package Information



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Package Information (continued)



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