General Description

The MAX814/MAX815/MAX816 are high-accuracy microprocessor (μ P) supervisory circuits that provide power-on reset, watchdog, and power-fail functions. They eliminate manual trimming and improve reliability in critical applications needing high-accuracy reset thresholds. The RESET output is guaranteed to be in the correct state for V_{CC} down to 1V. The reset comparator is designed to ignore fast transients on V_{CC}. Reset thresholds are available for operation with a variety of 3V and 5V supply voltages.

A 75 μ A maximum supply current makes the MAX814/ MAX815/MAX816 ideal for use in portable equipment. All three devices are available in 8-pin DIP and SO packages. See the *Selector Table* below for a review of features.

FEATURE	MAX814	MAX815	MAX816
RESET Output	~	~	~
RESET Output	~		~
Manual Reset	~	~	~
V _{CC} Reset Voltage	K, L, N, T	K, L, N, T	Adjustable
Power-Fail Monitor	~	~	~
Low-Line Detector	 ✓ 		
Watchdog Circuit		~	

Selector Table

Applications

Medical Equipment Controllers Intelligent Instruments Critical µP Power Monitoring Portable/Battery-Powered Equipment Set-Top Boxes

__Features

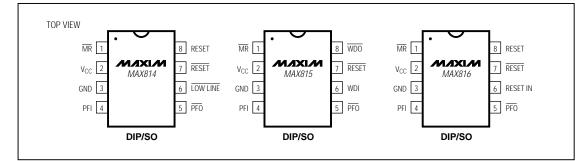
- ±1% Worst-Case Reset Threshold Accuracy
- 4.8V, 4.7V, 4.55V, 3.03V, or Adjustable Reset Thresholds
- ±1% Low-Line Threshold Accuracy (MAX814) 60mV Above Reset Threshold
- 200ms Reset Time Delay
- Active-Low RESET Output Active-High RESET Output (MAX814/MAX816)
- ♦ 75µA Max Supply Current
- ♦ Guaranteed RESET Valid to Vcc = 1V
- + Manual Reset Input
- + ±2% Power-Fail Comparator
- Independent Watchdog with 1.56sec Timeout (MAX815)
- Power-Supply Glitch Immunity
- 8-Pin SO and DIP Packages

__Reset Trip Thresholds

	=					
MAX814/MAX815						
SUFFIX	RESET TRIP	THRESHOLD				
SUFFIX	MIN (V)	MAX (V)				
K	4.75	4.85				
L	4.65	4.75				
N	4.50	4.60				
Т	3.00	3.06				
MAX816						
_	Adjustable					
adarina Informatio	doring Information appears at and of data sheat					

Ordering Information appears at end of data sheet.

Pin Configurations



///XI*//*/

Maxim Integrated Products 1

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ABSOLUTE MAXIMUM RATINGS

Terminal Voltage (with respect GND)

Terminal Voltage (with respect OND)	
V _{CC} 0.3V to +6.0V	
All Other Pins (Note 1)0.3V to (V _{CC} + 0.3V)	
Terminal Current (PFI, RESET IN, MR)10mA	
Terminal Current (all other pins)	
Continuous Power Dissipation ($T_A = +25^{\circ}C$)	
Plastic DIP (derate 9.09mW/°C above +70°C)727mW	
SO (derate 5.88mW/°C above +70°C)471mW	

Operating Temperature Ranges	
Commercial0°C to +70°C	
Extended40°C to +85C	
Storage Temperature Range65°C to +125°C	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS, +5V Parts (MAX814/MAX815K, L, N) ($V_{CC} = 4.85V$ to 5.5V for MAX814K/MAX815K, $V_{CC} = 4.75V$ to 5.5V for MAX814L/MAX815L, $V_{CC} = 4.60V$ to 5.5V for MAX814N/MAX815N, TA = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS
	N/	MAX814_C, MAX815_C	C	1.0		5.5	v
Operating Voltage Range	Vcc	MAX814_E, MAX815_E		1.2		5.5	v
Currently Current	1	MAX814_C, MAX815_C	2			75	
Supply Current	ISUPPLY	MAX814_E, MAX815_E				85	μΑ
		MAX814K, MAX815K		4.75		4.85	
Reset Threshold	VRT	MAX814L, MAX815L		4.65		4.75	V
		MAX814N, MAX815N		4.50		4.60	
Reset Threshold Hysteresis	V _{RT}				0		mV
Reset Pulse Width	trs			140	200	250	ms
	VOH	MAX014	I _{SOURCE} = 800µA	V _{CC} -1.5			V
RESET Output Voltage	Vol	MAX814	ISINK = 3.2mA			0.4	v
	Voh	ISOURCE = 800µA		Vcc -1.5			
RESET, WDO, PFO, LOW LINE		I _{SINK} = 3.2mA				0.4	v
Output Voltage	Vol	MAX814_C/MAX815_C, V _{CC} = 1.0V, I _{SINK} = 50µA				0.3	
		MAX814_E/MAX815_E, V _{CC} = 1.2V, I _{SINK} = 100µA				0.3	
Watchdog Timeout Period	twp	MAX815		1.12		2.00	sec
WDI Pulse Width	t _{WP}	MAX815		50			ns
WDI Input Threshold	Maria	MAX815, Vcc = 5.0V	Low			0.8	v
wbi input inieshold	Vwdi	VIAA015, VCC = 5.0V	High	2.4			
WDI Input Current	Iwdi	$WDI = V_{CC} \text{ or } WDI = 0V$		-1.0		1.0	μA
MR to WDO High Delay	twdo	MAX815 (Note 1)			1		μs
LOW LINE to RESET Differential	ΔV _{LL}	MAX814_C, V _{CC} falling MAX814_E, V _{CC} falling		50		70	mV
Threshold				48		73	IIIV
		MAX814K, V _{CC} rising		1	4.93	V	
LOW LINE Threshold	VLLT	MAX814L, V _{CC} rising					4.83
		MAX814N, V _{CC} rising				4.68	

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ELECTRICAL CHARACTERISTICS, +5V Parts (MAX814/MAX815K, L, N) (continued)

(V_{CC} = 4.85V to 5.5V for MAX814K/MAX815K, V_{CC} = 4.75V to 5.5V for MAX814L/MAX815L, V_{CC} = 4.60V to 5.5V for MAX814N/MAX815N, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
MR Pull-Up Current	IMR	$\overline{MR} = 0V$	70		240	μA
MR Pulse Width	t _{MR}		150			ns
MR Input Threshold	V MRLO	Low			1.1	V
Mix input mieshold	V mrhi	High	0.7 x Vcc			v
MR to RESET Out Delay	t _{MD}	(Note 3)			250	ns
PFI Input Threshold	VPFI	$V_{\rm CC} = 5.0 V$	2.45	2.50	2.55	V
PFI Input Current	IPFI		-15.00	6.0	35.00	nA
LOW LINE, PFO, WDO Assertion Delay		(Note 2)		200		μs

ELECTRICAL CHARACTERISTICS, +3V Parts (MAX814/MAX815T, MAX816)

(V_{CC} = 3.06V to 5.5V for MAX814T/MAX815T and MAX816, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

PARAMETER	SYMBOL	COND	TIONS	MIN	TYP	MAX	UNITS
Operating Voltage Dange	Maa	0°C to +70°C		1.0		5.5	v
Operating Voltage Range	Vcc	-40°C to +85°C		1.2		5.5	V
Supply Current	lauppuv	0°C to +70°C				75	
Supply Current	ISUPPLY	-40°C to +85°C				85	μA
Reset Threshold	VRT	0°C to +70°C		3.00		3.06	v
Reset Infeshold	VRI	-40°C to +85°C		3.00		3.08	
RESET IN Threshold	VRT	MAX816C		1.683	1.700	1.717	V
RESET IN THRESHOLD	VRI	MAX816E		1.678	1.700	1.722	V
RESET IN Input Current	IRT	MAX816		-15	6	35	nA
Reset Threshold Hysteresis	V _{RT}				0		mV
Reset Pulse Width	t _{RS}			140	200	250	ms
	Voh	V _{RT} (max) <v<sub>CC<3.6V;</v<sub>	ISOURCE = 500µA	0.8 x Vcc	;		
RESET Output Voltage	Vol	MAX814T, MAX816	Isink = 1.2mA			0.3	V
RESET Output Voltage	VOH	4.5V <v<sub>CC<5.5V;</v<sub>	I _{SOURCE} = 800µA	V _{CC} -1.5			
	Vol	MAX814T, MAX816	Isink = 3.2mA			0.4	
	Voh	V _{RT} (max) <v<sub>CC<3.6V</v<sub>	ISOURCE = 500µA	0.8 x Vcc	;		
	Vol		$I_{SINK} = 1.2 mA$			0.3	
RESET, WDO, PFO, LOW LINE	Voh	4.5V <vcc<5.5v< td=""><td>ISOURCE = 800µA</td><td>Vcc -1.5</td><td></td><td></td><td rowspan="2">V</td></vcc<5.5v<>	ISOURCE = 800µA	Vcc -1.5			V
Output Voltage	Vol	4.57<700<5.57	Isink = 3.2mA			0.4	
	Vol	$T_A = 0^{\circ}C \text{ to } + 70^{\circ}C, V_C$	_C = 1.0V, I _{SINK} = 50µA			0.3	
	VOL	T_A = -40°C to +85°C, V_{CC} = 1.2V, I_{SINK} = 100 μA				0.3	
Watchdog Timeout Period	twp	MAX815T		1.12		2.00	sec

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ELECTRICAL CHARACTERISTICS, +3V Parts (MAX814/MAX815T, MAX816) (continued) (V_{CC} = 3.06V to 5.5V for MAX814T/MAX815T and MAX816, T_A = T_{MIN} to T_{MAX}, unless otherwise noted.)

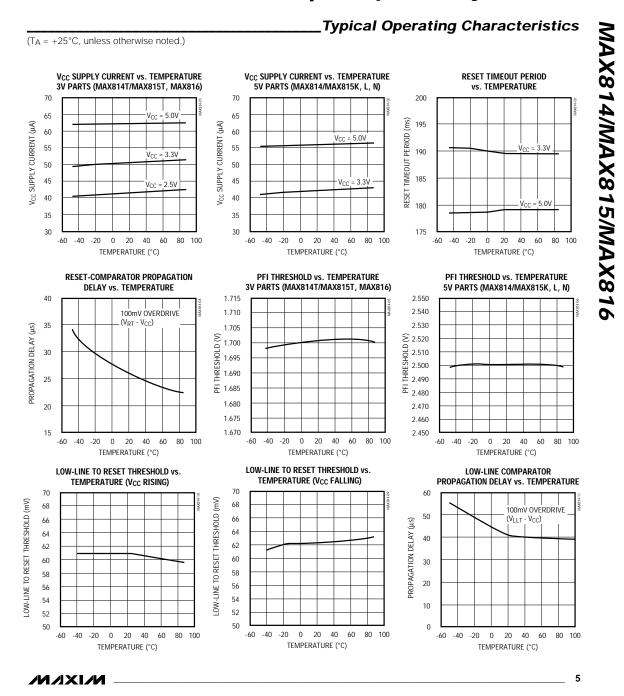
PARAMETER	SYMBOL	COND	ITIONS	MIN	TYP	MAX	UNITS	
WDI Pulse Width	÷	MAX815T	V _{RT} (max) <v<sub>CC<3.6V</v<sub>	100				
WDI Pulse Width	t _{WP}	ICT 87 AIN	4.5V <v<sub>CC<5.5V</v<sub>	50			ns	
		V _{RT} (max) <v<sub>CC<3.6V;</v<sub>	Low			0.8		
WDI Input Threshold	V _{WDI}	MAX815T	High	0.7 x Vcc	:			
WDI Input Mieshold		V _{CC} = 5.0V; MAX815T	Low			0.8		
		VCC = 5.0V, IVIAX6151	High	2.4			1	
WDI Input Current	Iwdi	WDI = V _{CC} or 0V, MAX	815T	-1.0		1.0	μΑ	
MR to WDO High Delay	t _{WDO}	MAX815T (Note 1)			1		μs	
LOW LINE to RESET Differential	A. \ (V _{CC} falling, MAX814TC		50		70	mV	
Threshold	ΔVLL	V _{CC} falling, MAX814TE		48		73		
LOW LINE Threshold	VLLT	V _{CC} rising				3.163	V	
MR Pull-Up Current	lu in	$\overline{MR} = 0V$	V _{RT} (max) <v<sub>CC<3.6V</v<sub>	70		240		
MR Pull-Op Current	IMR	I _{MR} MR = 0V	4.5V <v<sub>CC<5.5V</v<sub>	110		370	μA	
MR Pulse Width	ti in	V _{RT} (max) <v<sub>CC<3.6V</v<sub>		500				
INR Pulse Width	tMR	4.5V <v<sub>CC<5.5V</v<sub>		150			ns	
MR Input Threshold	V MRLO	Low				1.1	v	
WR Input Mieshold	V MRHI	High		0.7 x Vcc	:		ľ	
	ture	V _{RT} (max) <v<sub>CC<3.6V;</v<sub>	(Note 3)			750		
MR to RESET Out Delay	tMD	4.5V <v<sub>CC<5.5V, (Note 3)</v<sub>				250	ns	
PFI Input Threshold	Vpfi	V _{CC} = 3.3V, 5V		1.666	1.700	1.734	V	
PFI Input Current	IPFI			-15.00	6.0	35.00	nA	

Note 1: Applies if $\overline{\text{WDO}}$ is externally connected to $\overline{\text{MR}}$ or if $\overline{\text{MR}}$ is externally driven. **Note 2:** On power-up, delay from reset trip threshold crossing to valid outputs. **Note 3:** Applies to both RESET and RESET.

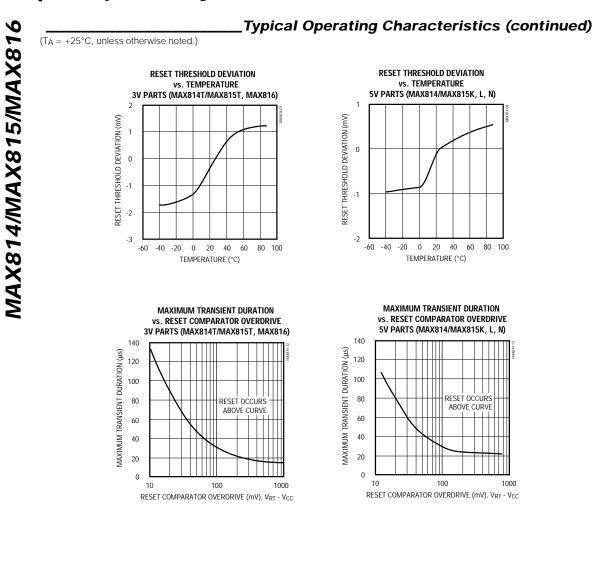
MAX814/MAX815/MAX816

M/IXI/M

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_Pin Description

	PIN		NAME	FUNCTION	
MAX814	MAX815	MAX816	NAME	FUNCTION	
1	1	1	MR	Manual-Reset Input. Triggers a reset when pulled below 1.10V. This active- low input has an internal 150 μ A pull-up current to V _{CC} , and can be driven with CMOS logic or shorted to GND with a switch or transistor.	
2	2	_	Vcc	Positive Power-Supply Input. When V _{CC} is below the reset threshold voltage*, RESET is low, and remains low for a minimum of 140ms after it rises above the threshold.	
—	_	2		Positive Power-Supply Input. On the MAX816, $\overline{\text{RESET}}$ is controlled by RESET IN, not V_CC.	
3	3	3	GND	Ground	
4	4	4	PFI	Power-Fail Input. The PFI threshold voltage is 1.70V on the MAX816 and on MAX814/MAX815 parts with the T suffix. It is 2.50V on MAX814/MAX815 parts with K, L, and N suffixes. Connect PFI to GND or V _{CC} when not used.	
5	5	5	PFO	Power-Fail Output. When PFI is below its threshold, PFO is low; otherwise it is high.	
_	6	_	WDI	Watchdog CMOS Input. If WDI remains high or low for more than 1.56sec, the watchdog timer times out, and \overline{WDO} goes low. The timer is reset to zero on each WDI transition.	
6	_	_	LOW LINE	Low-Line Output. Normally high, $\overline{\text{LOW LINE}}$ goes low when V_{CC} falls 60mV above the reset threshold. It returns high as soon as V_{CC} rises above the low-line threshold.	
_	_	6	RESET IN	Reset Comparator Input. Reference is 1.70V. When RESET IN is below 1.70V, RESET is low, and remains low for a minimum of 140ms after it rises above the reference.	
7	7	7	RESET	ESET Reset Output. Normally high, active low. Controlled by MR and reset comparator.	
_	8	_	WDO	Watchdog Output. Normally high, $\overline{\text{WDO}}$ goes low whenever the V _{CC} reset threshold comparator input voltage is low or when the watchdog timer times out. There is no appreciable delay going either direction when the V _{CC} threshold comparator toggles.	
8	—	8	RESET	Reset Output. Active high. The inverse of RESET.	

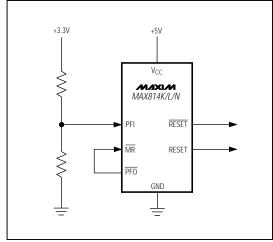
*Reset Threshold Voltage is determined by part number suffix: K = 4.80V, L = 4.70V, N = 4.55V, T = 3.03V.

_Detailed Description

The MAX814/MAX815/MAX816 are high-accuracy, lowpower microprocessor (μ P) supervisory circuits. They have μ P-reset, watchdog-timer, and power-fail functions. Typical applications illustrating their similarities and differences are shown in Figures 1, 2, and 3. Figures 4, 5, and 6 show the block diagrams of these parts.

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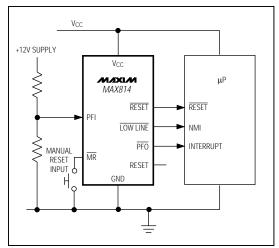


Figure 1a. Typical Application for Dual +3.3V and +5V Systems

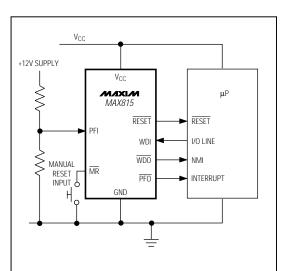
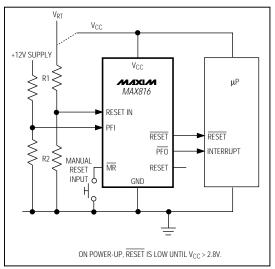
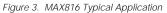


Figure 2. MAX815 Typical Application

Figure 1b. MAX814 Typical Application





Μ/ΙΧΙ/Ν



WDI

MR

Vcc

PFI

WATCHDOG

TRANSITION

DETECTOR

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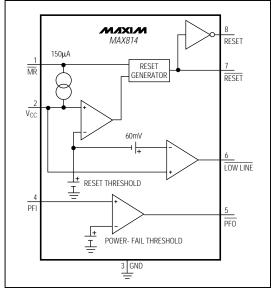


Figure 4. MAX814 Block Diagram

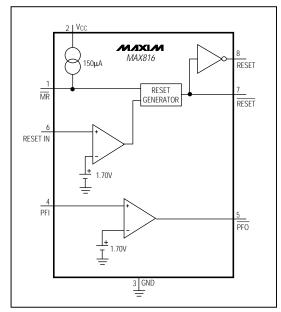
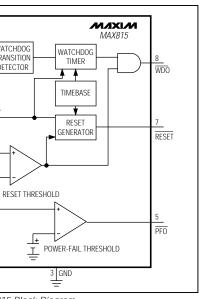


Figure 6. MAX816 Block Diagram





MAX814/MAX815/MAX816

Figure 5. MAX815 Block Diagram

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Reset Output

A µP's reset input starts the µP in a known state. Whenever the µP is in an unknown state, it should be held in reset. The MAX814/MAX815/MAX816 assert reset during power-up, power-down, or brownout conditions.

On power-up, once V_{CC} reaches 1V, RESET is a guaranteed logic low of 0.4V or less. As VCC rises, RESET stays low. As VCC rises above the reset threshold, an internal timer releases RESET after 200ms. RESET also pulses low whenever V_{CC} dips below the reset threshold (i.e., brownout condition). If brownout occurs in the middle of a previously initiated reset, the internal timer is reset and the output remains low for at least another 140ms after the brownout ends. On power-down, once Vcc falls below the reset threshold, RESET stays low and is guaranteed to be less than 0.3V until Vcc drops below 1V.

The MAX814 and MAX816 also offer active-high RESET outputs. They are the inverse of the RESET outputs.

Reset Threshold

±1% Accuracy, Low-Power, +3V and +5V

The MAX814/MAX815 have fixed, factory-set reset thresholds, signified by the first suffix letter in the part number (see Figure 7 for more information on reset ranges). The MAX816 has an adjustable reset threshold.

μP Supervisory Circuits

MAX814/MAX815 K-suffix parts have a minimum reset threshold set to 4.75V, worst case. They are intended for 5.0V systems with a $\pm 4\%$ or better power-supply tolerance design that must meet worst-case system parameters over time, temperature, line, and load variations. Typically, the reset threshold (V_{RT}) is greater than or equal to the minimum IC operating voltage (V_{ICMIN}). The "K" series 1%-tolerance reset threshold allows a larger range of power-supply tolerance. System ICs

that have a tight operating supply range, like the 386/486 $\mu\text{Ps},$ need a RESET initiated at a minimum threshold of 4.75V, worst case.

L-suffix parts have a minimum reset threshold set to 4.65V, worst case. They are intended for 5.0V systems with a ±5% power-supply tolerance. Typically, the reset threshold is less than or equal to the minimum powersupply voltage, allowing system operation over the complete power-supply range. A reset is initiated at 4.75V maximum. The 1% "L" version maximizes the System IC Guard-Band Range.

N-suffix parts have a minimum reset threshold set to 4.50V, worst case. They are intended for 5.0V systems with a $\pm 10\%$ IC system. Typically, the reset threshold

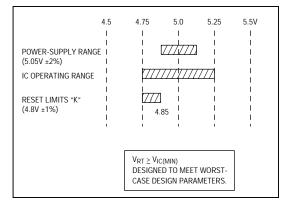
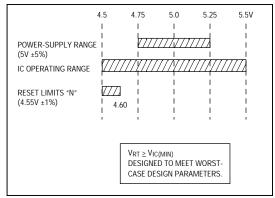


Figure 7a. K Suffix Design Range





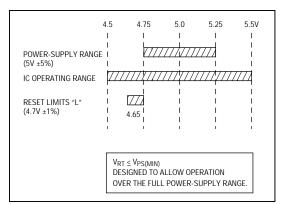


Figure 7b. L Suffix Design Range

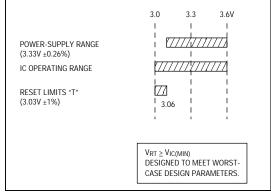


Figure 7d. T Suffix Design Range

M/IXI/M

Early Power-Fail Warning

Critical systems often require early warning to indicate when power is failing. This warning provides time for the μ P to store vital data and take care of any additional "housekeeping" before the power supply gets too far out of tolerance for the μ P to operate reliably.

Power-Fail Comparator

The power-fail comparator is intended as an undervoltage detector to signal a failing power supply. However, the comparator does not need to be dedicated to this function, because it is completely separate from the rest of the circuitry. To build an early-warning circuit for power failure, connect the PFI pin to a voltage divider (see Figures 1, 2, and 3). Choose the voltage divider ratio, so the voltage at PFI falls below V_{PFI} just before the monitored voltage drops out. Use PFO to interrupt the μ P, so it can prepare for an orderly power-down.

The power-fail input (PFI) is compared to an internal reference. If the voltage on PFI is less than the power-fail reference, $\overline{\text{PFO}}$ sinks at least 1.2mA to GND; otherwise it sources at least 300µA from V_{CC}. The reference is 2.50V in the MAX814/MAX815 with K, L, N suffixes, or 1.70V with the T suffix. It is also 1.70V in the MAX816.

LOW LINE Output (MAX814)

The low-line detector is a separate comparator that monitors V_{CC} with a typical threshold voltage of 60mV above the normal reset threshold, with 2mV of hysteresis (Figure 9). If V_{CC} rises faster than 10µs/V, insert a 100pF capacitor from LOW LINE to GND to ensure proper start-up. For normal operation (V_{CC} above the reset threshold), LOW LINE is pulled to V_{CC}. Use LOW LINE to provide an NMI to the µP when power begins to fall. In most battery-operated portable systems, reserve energy in the battery provides ample time to complete the shutdown routine once the low-line warning is encountered, and before reset asserts. If the system must also contend with a more rapid V_{CC} fall timesuch as when the main battery is disconnected or a high-side switch is opened during operation-use capacitance on the $V_{\mbox{CC}}$ line to provide time to execute the shutdown routine. First, calculate the worst-case time required for the system to perform its shutdown routine. Then use the worst-case shutdown time (tSHDN), worst-case load current (ILOAD), and minimum low-line to reset threshold (VLR) to calculate the amount of capacitance required to allow the shutdown routine to complete before reset is asserted.

$$C_{\text{HOLD}} = \frac{I_{\text{LOAD}} \times I_{\text{SHDN}}}{V_{\text{LR}}}$$

 (V_{RT}) is greater than or equal to the minimum IC operating voltages (V_{ICMIN}). The 1% "L" series allows the use of a 5V $\pm 5\%$ power supply, and guarantees system operation over worst-case conditions, maximizing the Power-Supply Guard-Band Range.

T-suffix parts have a minimum reset threshold set to 3.00V, worst case. They are intended for 3.3V systems (3.33V \pm 0.26V) with a 7.8% or better power-supply tolerance. Typically, the reset threshold (V_{RT}) is greater than or equal to the minimum IC operating voltages (V_{ICMIN}).

The MAX816 has an adjustable reset threshold, set with an external resistive divider (Figure 3). The voltage on the RESET IN pin is monitored, not the voltage on V_{CC}. The RESET IN threshold is 1.700V, and has very high impedance and 35nA maximum leakage. Calculate the trip point, V_{RT}, as follows:

$$V_{\rm RT} = \frac{V_{\rm RIT} \times (R1 + R2)}{R2}$$

where V_{RT} = the desired reset threshold, V_{RIT} is the RESET IN threshold (1.700V), R1 is the resistor connected between V_{RT} and RESET IN, and R2 is the resistor connected between RESET IN and GND.

Resistors R1 and R2 can have very high values. The usual procedure is to set R2 to some conveniently high value (100k Ω , for example) and calculate R1 based on the desired reset threshold, using the following formula:

$$R1 = R2 \times \left[\left(V_{RT} / V_{RIT} \right) - 1 \right]$$

The MAX816 can achieve $\pm 1.2\%$ accuracy with 0.1% resistors.

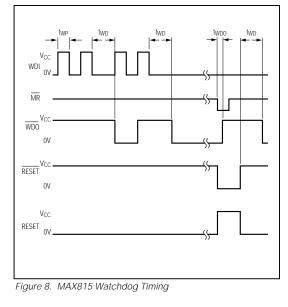
Watchdog Timer (MAX815)

The watchdog circuit monitors the μ P's activity. If the μ P does not toggle the watchdog <u>input</u> (WDI) within the watchdog timeout period (twp), WDO goes low (Figure 8). WDO also goes low during reset conditions. Whenever Vcc is below the reset threshold, WDO stays low; however, unlike RESET, WDO does not have a minimum pulse width. As soon as Vcc rises above the reset threshold, WDO goes high with no delay (Figure 9).

Typically, \overline{WDO} is connected to the non-maskable interrupt (NMI) of a μ P. When V_{CC} drops below the reset threshold, \overline{WDO} goes low whether or not the watchdog timer has timed out (Figure 9). This would normally trigger an NMI interrupt, but RESET goes low simultaneously and thus overrides the NMI interrupt.

Connecting \overline{WDO} to \overline{MR} enables the watchdog timeout to generate a reset in the MAX815.

MIXIM



where C_{HOLD} is the capacitance (in Farads), I_{LOAD} is the current being drained from the capacitor (in Amperes), and V_{LR} is the low-line to reset threshold difference (in Volts).

 $\label{eq:hamiltonian} \begin{array}{c} \textbf{Manual Reset} \\ \text{Many } \mu \text{P-based products require manual-reset capabil-}\\ \text{ity, allowing the operator, a test technician, or external}\\ \text{logic circuitry to initiate a reset. A logic low on MR}\\ \text{asserts reset. Reset remains asserted while MR is low, and for tRs (200ms) after MR returns high. This input has an internal pull-up resistor, so it can be left open if not used. MR can be driven with TTL/CMOS-logic levels or with open-drain/collector outputs. \end{array}$

Connect a normally open momentary switch from MR to GND to create a manual-reset function; external debounce circuitry is not required.

The watchdog circuit can be used to force a reset in the MAX815 by connecting WDO to MR. If MR is driven from long cables, or the device is used in a noisy environment, connect a 0.1μ F capacitor to ground to provide additional noise immunity.

2
-

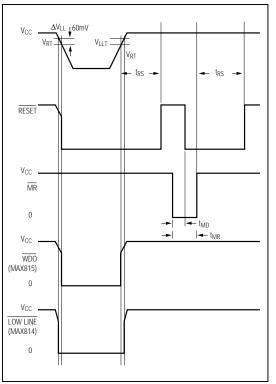


Figure 9. Timing Diagram

Applications Information

Low-Voltage Operation

The $\overline{\text{LOW LINE}}$, $\overline{\text{PFO}}$, and $\overline{\text{WDO}}$ outputs will be locked to logic low when the power supply drops below the lockout threshold (typically 1V below the reset threshold).

Ensuring a Valid RESET Output Down to VCC = 0V

When V_{CC} falls below 1V, the RESET output no longer sinks current, but becomes an open circuit. High-impedance CMOS-logic inputs can drift to undetermined voltages if left undriven. If a pull-down resistor is added to the RESET pin as shown in Figure 10, any stray charge or leakage currents will be drained to ground, holding RESET low. Resistor value R1 is not critical. It should be about 100kΩ—large enough not to load RESET, and small enough to pull RESET to ground.

M/IXI/N



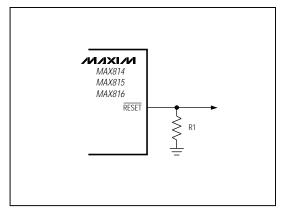


Figure 10. RESET Valid to Ground Circuit



Monitor voltages other than the V_{CC} by connecting a voltage divider to PFI and adjusting the ratio appropriately. If required, add hysteresis by connecting a resistor (with a value approximately 10-times the sum of the two resistors in the potential divider network) between PFI and PFO. A capacitor between PFI and GND will reduce the power-fail circuit's sensitivity to high-frequency noise on the line being monitored. RESET can be asserted on other voltages in addition to the +5V V_{CC} line. Connect PFO to MR to initiate a reset when PFI drops below 2.50V (K, L, N suffix) or 1.70V (T suffix or MAX816). Figure 11 shows the MAX814K/L/N/ MAX815K/L/N configured to assert RESET when the +5V supply falls below the reset threshold, or when the +12V supply falls below approximately 11V.

Monitoring a Negative Voltage

The power-fail comparator can also monitor a negative supply rail (Figure 12). When the negative rail is good (a negative voltage of large magnitude), PFO is low. When the negative rail is degraded (a negative voltage of lesser magnitude), PFO is high. By adding the resistors and transistor as shown, a high PFO triggers reset. As long as PFO remains high, the MAX814/MAX815/MAX816 will keep reset asserted (RESET = low, RESET = high). Note that this circuit's accuracy depends on the PFI threshold tolerance, the V_{CC} line, and the resistor.

+12V +5V RESET ΤΟ μΡ Vcc 340k 1% MIXIM MAX814K/L/N mR MAX815K/L/N 100k 5 DEI PFC 1% GND Ŧ PARAMETER MIN TYP MAX UNIT +12V Rese 10.57 11.00 11.45 V Threshold at +25°C

Figure 11. Monitoring Both +5V and +12V

Watchdog Software Considerations

A way to help the watchdog timer keep closer tabs on software execution involves setting and resetting the watchdog input at different points in the program, rather than pulsing the watchdog input high-low-high or low-high-low. This technique avoids a stuck loop where the watchdog timer continues to be reset within the loop, keeping the watchdog from timing out. Figure 13 shows an example flow diagram where the I/O driving the watchdog input is set low at the beginning of the program, set high at the beginning of every subroutine, then set low at the end of every subroutine. If the program should hang in any subroutine, the I/O is continually set high and the watchdog timer is allowed to time out, causing a reset to be issued.



MAX814/MAX815/MAX816

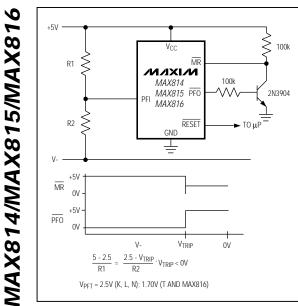


Figure 12. Monitoring a Negative Voltage

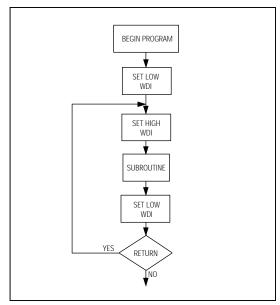


Figure 13. Flow Chart of WDI Implementation

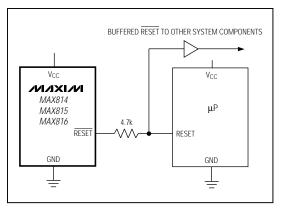


Figure 14. Interfacing to µPs with Bidirectional Reset I/O

Negative-Going Vcc Transients

In addition to issuing a reset to the µP during power-up, power-down, and brownout conditions, the MAX814/ MAX815/MAX816 series is relatively immune to short duration negative-going V_{CC} transients (glitches). The Typical Operating Characteristics show a graph of Maximum Transient Duration vs. Reset Comparator Overdrive, for which a reset is not generated. The graph was made using a negative-going pulse applied to Vcc, starting 1.5V above the actual reset threshold and ending below it by the magnitude indicated (reset comparator overdrive). The graph indicates the typical maximum pulse width a negative-going V_{CC} transient may have without causing a reset pulse. As the magnitude of the transient increases (goes further below the reset threshold), the maximum allowable pulse width decreases. Typically, a Vcc that goes 100mV below the reset threshold and lasts 30µs or less will not cause a reset pulse to be issued.

A 0.1µF bypass capacitor mounted as close as possible to pin 2 (Vcc) provides additional transient immunity.

Interfacing to µPs with Bidirectional Reset Pins

 μPs with bidirectional reset pins, such as the Motorola 68HC11 series, can cause a conflict with the RESET output. If, for example, the RESET output is driven high and the μP wants to pull it low, indeterminate logic levels may result. To correct this, connect a 4.7k Ω resistor between the RESET output and the μP reset I/O, as in Figure 14. Buffer the RESET output to other system components.

M/IXI/M

	Ordering	Information
PART*	TEMP. RANGE	PIN-PACKAGE
MAX814_CPA	0°C to +70°C	8 Plastic DIP
MAX814_CSA	0°C to +70°C	8 SO
MAX814_EPA	-40°C to +85°C	8 Plastic DIP
MAX814_ESA	-40°C to +85°C	8 SO
MAX815_CPA	0°C to +70°C	8 Plastic DIP
MAX815_CSA	0°C to +70°C	8 SO
MAX815_EPA	-40°C to +85°C	8 Plastic DIP
MAX815_ESA	-40°C to +85°C	8 SO
MAX816CPA	0°C to +70°C	8 Plastic DIP
MAX816CSA	0°C to +70°C	8 SO
MAX816EPA	-40°C to +85°C	8 Plastic DIP
MAX816ESA	-40°C to +85°C	8 SO

* The MAX814/MAX815 offer a choice of reset threshold voltage. From the Reset Trip Threshold table, select the suffix corresponding to the desired threshold and insert it into the blank to complete the part number.

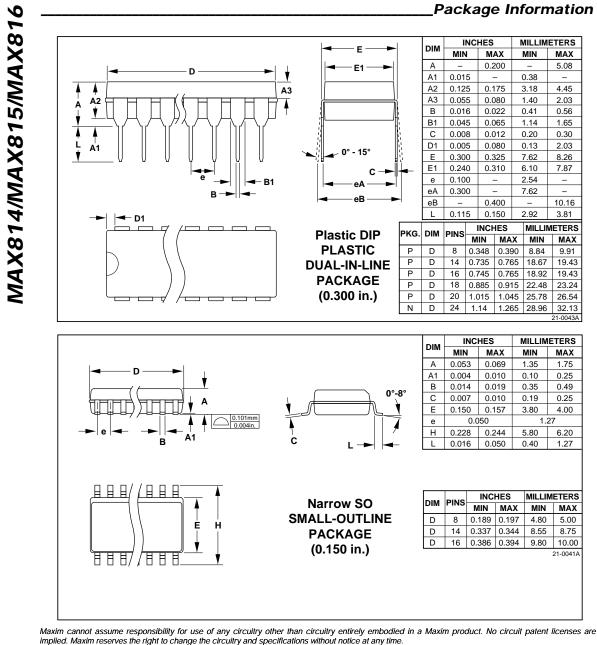
ete the part number.

_Reset Trip Thresholds

MAX814/MAX815					
SUFFIX	RESET TRIP	THRESHOLD			
JUFFIX	MIN (V)	MAX (V)			
К	4.75	4.85			
L	4.65	4.75			
N	4.50	4.60			
Т	3.00 3.06				
MAX816					
—	Adjustable				

____Chip Information

TRANSISTOR COUNT: 744



u. waxiin reserves the right to change the circulity and specifications without notice at any time. ______Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 (408) 737-7600

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