

# LC72131, 72131M

# **AM/FM PLL Frequency Synthesizer**



# **Overview**

The LC72131 and LC72131M are PLL frequency synthesizers for use in tuners in radio/cassette players. They allow high-performance AM/FM tuners to be implemented easily.

# **Applications**

PLL frequency synthesizer

# **Functions**

- High speed programmable dividers
  - FMIN: 10 to 160 MHz .....pulse swallower (built-in divide-by-two prescaler)
  - AMIN: 2 to 40 MHz .....pulse swallower
    - 0.5 to 10 MHz .....direct division
- IF counter
- IFIN: 0.4 to 12 MHz .....AM/FM IF counter
- Reference frequencies
  - Twelve selectable frequencies
    - (4.5 or 7.2 MHz crystal)
    - 1, 3, 5, 9, 10, 3.125, 6.25, 12.5, 15, 25, 50 and 100 kHz
- Phase comparator
  - Dead zone control
  - Unlock detection circuit
  - Deadlock clear circuit
- Built-in MOS transistor for forming an active low-pass filter
- I/O ports
  - Dedicated output ports: 4
  - Input or output ports: 2
  - Support clock time base output
- Serial data I/O
  - Support CCB format communication with the system controller.
- Operating ranges
  - Supply voltage......4.5 to 5.5 V
  - Operating temperature......-40 to +85°C
- Packages
  - DIP22S/MFP20

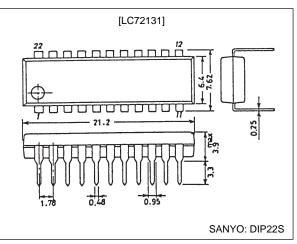
#### CCB is a trademark of SANYO ELECTRIC CO., LTD.

• CCB is SANYO's original bus format and all the bus addresses are controlled by SANYO.

# **Package Dimensions**

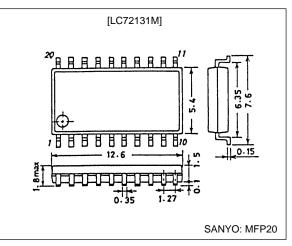
unit: mm

#### 3059-DIP22S



### unit: mm

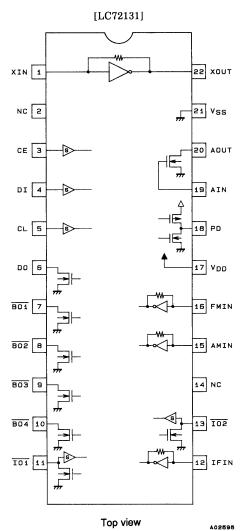
### 3036B-MFP20



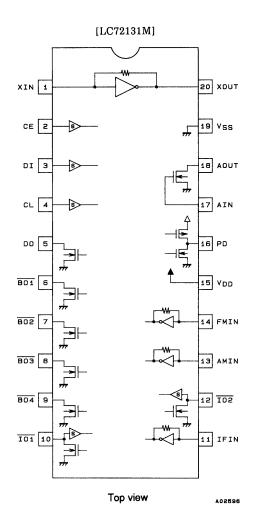
SANYO Electric Co., Ltd. Semiconductor Bussiness Headquarters TOKYO OFFICE Tokyo Bldg., 1-10, 1 Chome, Ueno, Taito-ku, TOKYO, 110 JAPAN

62096HA (OT)/O3195HA (OT)/92294TH (OT) No. 4921-1/23

# **Pin Assignments**

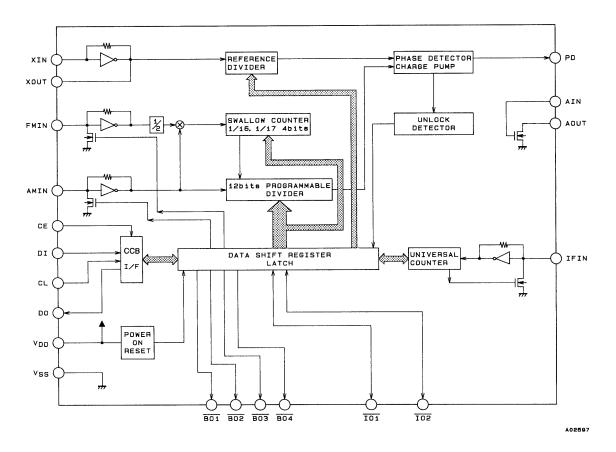


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## **Block Diagram**



# **Specifications**

# Absolute Maximum Ratings at $Ta=25^{\circ}C,\,V_{SS}$ = 0 V

Parameter	Symbol		Pins	Ratings	Unit
Supply voltage	V <sub>DD</sub> max	V <sub>DD</sub>	-0.3 to +7.0	V	
	V <sub>IN</sub> 1 max	CE, CL, DI, AIN	CE, CL, DI, AIN -0.3 to +7.0		
Maximum input voltage	V <sub>IN</sub> 2 max	XIN, FMIN, AMIN, IF	IN	-0.3 to V <sub>DD</sub> + 0.3	V
	V <sub>IN</sub> 3 max	101, 102		-0.3 to +15	V
	V <sub>O</sub> 1 max	DO		-0.3 to +7.0	V
Maximum output voltage	V <sub>O</sub> 2 max	XOUT, PD	-0.3 to V <sub>DD</sub> + 0.3	V	
	V <sub>O</sub> 3 max	BO1 to BO4, IO1, IC	02, AOUT	-0.3 to +15	V
	I <sub>O</sub> 1 max	BO1		0 to 3.0	mA
Maximum output current	I <sub>O</sub> 2 max	AOUT, DO		0 to 6.0	mA
	I <sub>O</sub> 3 max	BO2 to BO4, IO1, IC	02	0 to 10.0	mA
Allowable power dissinction	Pd max	Ta ≤ 85°C	LC72131: DIP22S	350	mW
Allowable power dissipation	Polmax	1a ≤ 85°C	LC72131M: MFP20	180	11100
Operating temperature	Topr			-40 to +85	°C
Storage temperature	Tstg			-55 to +125	°C

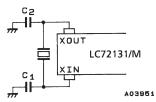
# Allowable Operating Ranges at Ta = -40 to $+85^{\circ}$ C, V<sub>SS</sub> = 0 V

Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
Supply voltage	V <sub>DD</sub>	V <sub>DD</sub>		4.5		5.5	V
lanut high laugh ugtana	V <sub>IH</sub> 1	CE, CL, DI		0.7 V <sub>DD</sub>		6.5	V
Input high-level voltage	V <sub>IH</sub> 2	101, 102		0.7 V <sub>DD</sub>		13	V
Input low-level voltage	VIL	CE, CL, DI, 101, 102		0		0.3 V <sub>DD</sub>	V
	V <sub>O</sub> 1	DO		0		6.5	V
Output voltage	V <sub>O</sub> 2	$\overline{BO1}$ to $\overline{BO4}$ , $\overline{IO1}$ , $\overline{IO2}$ , AOUT		0		13	V
	f <sub>IN</sub> 1	XIN	V <sub>IN</sub> 1	1		8	MHz
	f <sub>IN</sub> 2	FMIN	V <sub>IN</sub> 2	10		160	MHz
Input frequency	f <sub>IN</sub> 3	AMIN	V <sub>IN</sub> 3, SNS = 1	2		40	MHz
	f <sub>IN</sub> 4	AMIN	$V_{IN}4$ , SNS = 0	0.5		10	MHz
	f <sub>IN</sub> 5	IFIN	V <sub>IN</sub> 5	0.4		12	MHz
	V <sub>IN</sub> 1	XIN	f <sub>IN</sub> 1	400		1500	mVrms
	V <sub>IN</sub> 2-1	FMIN	f = 10 to 130 MHz	40		1500	mVrms
	V <sub>IN</sub> 2-2	FMIN	f = 130 to 160 MHz	70		1500	mVrms
Input amplitude	V <sub>IN</sub> 3	AMIN	f <sub>IN</sub> 3, SNS = 1	40		1500	mVrms
	V <sub>IN</sub> 4	AMIN	$f_{IN}4$ , SNS = 0	40		1500	mVrms
	V <sub>IN</sub> 5-1	IFIN	f <sub>IN</sub> 5, IFS = 1	40		1500	mVrms
	V <sub>IN</sub> 5-2	IFIN	$f_{IN}6$ , IFS = 0	70		1500	mVrms
Supported crystals	Xtal	XIN, XOUT	*	4.0		8.0	MHz

<Sample Oscillator Circuit>

Crystal oscillator: HC-49/U (manufactured by Kinseki, Ltd.), CL = 12 pF C1 = C2 = 15 pF

The circuit constants for the crystal oscillator circuit depend on the crystal used, the printed circuit board pattern, and other items. Therefore we recommend consulting with the manufacturer of the crystal for evaluation and reliability.



Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
	Rf1	XIN			1.0		MΩ
Built-in feedback resistance	Rf2	FMIN			500		kΩ
Duilt-III leeuback lesistalice	Rf3	AMIN			500		kΩ
	Rf4	IFIN			250		kΩ
Built-in pull-down resistor	Rpd1	FMIN			200		kΩ
Built-in pull-down resistor	Rpd2	AMIN			200		kΩ
Hysteresis	V <sub>HIS</sub>	CE, CL, DI, IO1, IO2			0.1 V <sub>DD</sub>		V
Output high level voltage	V <sub>OH</sub> 1	PD	I <sub>O</sub> = -1 mA	V <sub>DD</sub> – 1.0			V
	V <sub>OL</sub> 1	PD	I <sub>O</sub> = 1 mA			1.0	V
			l <sub>O</sub> = 0.5 mA			0.5	V
	V <sub>OL</sub> 2	BO1	I <sub>O</sub> = 1 mA			1.0	V
			I <sub>O</sub> = 1 mA			0.2	V
Output low level voltage	V <sub>OL</sub> 3	DO	I <sub>O</sub> = 5 mA			1.0	V
			I <sub>O</sub> = 1 mA			0.2	V
	V <sub>OL</sub> 4	$\overline{BO2}$ to $\overline{BO4}$ , $\overline{IO1}$ , $\overline{IO2}$	I <sub>O</sub> = 5 mA			1.0	V
			I <sub>O</sub> = 8 mA			1.6	V
	V <sub>OL</sub> 5	AOUT	I <sub>O</sub> = 1 mA, AIN = 1.3 V			0.5	V
	I <sub>IH</sub> 1	CE, CL, DI	V <sub>I</sub> = 6.5 V			5.0	V
	I <sub>IH</sub> 2	<u>101, 102</u>	V <sub>I</sub> = 13 V			5.0	μA
Input high level current	I <sub>IH</sub> 3	XIN	$V_{I} = V_{DD}$	2.0		11	μA
	I <sub>IH</sub> 4	FMIN, AMIN	$V_{I} = V_{DD}$	4.0		22	μA
	I <sub>IH</sub> 5	IFIN	$V_{I} = V_{DD}$	8.0		44	μΑ
	I <sub>IH</sub> 6	AIN	V <sub>I</sub> = 6.5 V			200	nA
	 I <sub>IL</sub> 1	CE, CL, DI	V <sub>1</sub> = 0 V			5.0	μA
	I <sub>IL</sub> 2	<u>101, 102</u>	V <sub>1</sub> = 0 V			5.0	μΑ
	I <sub>IL</sub> 3	XIN	V <sub>1</sub> = 0 V	2.0		11	μΑ
Input low level current	I <sub>IL</sub> 4	FMIN, AMIN	V <sub>I</sub> = 0 V	4.0		22	μΑ
	I <sub>IL</sub> 5	IFIN	V <sub>1</sub> = 0 V	8.0		44	μΑ
	I <sub>IL</sub> 6	AIN	V <sub>1</sub> = 0 V			200	nA
Output off leakage current	I <sub>OFF</sub> 1	BO1 to BO4, AOUT, IO1, IO2	V <sub>O</sub> = 13 V			5.0	μA
ouput on leakage current	I <sub>OFF</sub> 2	DO	V <sub>O</sub> = 6.5 V			5.0	μA
High level three-state off leakage current	I <sub>OFFH</sub>	PD	V <sub>O</sub> = V <sub>DD</sub>		0.01	200	nA
Low level three-state off leakage current	I <sub>OFFL</sub>	PD	V <sub>O</sub> = 0 V		0.01	200	nA
Input capacitance	CIN	FMIN			6		pF
	I <sub>DD</sub> 1	V <sub>DD</sub>	$\begin{array}{l} Xtal = 7.2 \text{ MHz}, \\ f_{IN}2 = 130 \text{ MHz}, \\ V_{IN}2 = 40 \text{ mVrms} \end{array}$		5	10	mA
Current drain	I <sub>DD</sub> 2	V <sub>DD</sub>	PLL block stopped (PLL INHIBIT), Xtal oscillator operating (Xtal = 7.2 MHz)		0.5		mA
	I <sub>DD</sub> 3	V <sub>DD</sub>	PLL block stopped Xtal oscillator stopped			10	μA

# Electrical Characteristics for the Allowable Operating Ranges at Ta = –40 to +85°C, $V_{SS}$ = 0 V

# **Pin Functions**

Symbol	Pin No. (MFP pin Nos. are in parentheses.)	Туре	Functions	Circuit configuration
XIN XOUT	1 (1) 22 (20)	Xtal OSC	Crystal resonator connection (4.5/7.2 MHz)	A02598
FMIN	16 (14)	Local oscillator signal input	<ul> <li>FMIN is selected when the serial data input DVS bit is set to 1.</li> <li>The input frequency range is from 10 to 160 MHz.</li> <li>The input signal passes through the internal divide-by-two prescaler and is input to the swallow counter.</li> <li>The divisor can be in the range 272 to 65535. However, since the signal has passed through the divide-by-two prescaler, the actual divisor is twice the set value.</li> </ul>	A02599
AMIN	15 (13)	Local oscillator signal input	<ul> <li>AMIN is selected when the serial data input DVS bit is set to 0.</li> <li>When the serial data input SNS bit is set to 1: <ul> <li>The input frequency range is 2 to 40 MHz.</li> <li>The signal is directly input to the swallow counter.</li> <li>The divisor used will be the value set.</li> </ul> </li> <li>When the serial data input SNS bit is set to 0: <ul> <li>The input frequency range is 0.5 to 10 MHz.</li> <li>The signal is directly input to a 12-bit programmable divider.</li> <li>The divisor can be in the range 4 to 4095, and the divisor used will be the value set.</li> </ul> </li> </ul>	A02599
CE	3 (2)	Chip enable	Set this pin high when inputting (DI) or outputting (DO) serial data.	□ \$ A02500
CL	5 (4)	Clock	Used as the synchronization clock when inputting (DI) or outputting (DO) serial data.	D
DI	4 (3)	Data input	Inputs serial data transferred from the controller to the LC72131.	<b>S</b> A02600
DO	6 (5)	Data output	<ul> <li>Outputs serial data transferred from the LC72131 to the controller.</li> <li>The content of the output data is determined by the serial data DOC0 to DOC2.</li> </ul>	
V <sub>DD</sub>	17 (15)	Power supply	<ul> <li>The LC72131 power supply pin (V<sub>DD</sub> = 4.5 to 5.5 V)</li> <li>The power on reset circuit operates when power is first applied.</li> </ul>	

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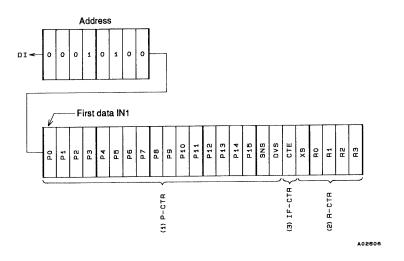
Symbol	Pin No. (MFP pin Nos. are Type Functions in parentheses.)		Functions	Circuit configuration
V <sub>SS</sub>	21 (19)	Ground	The LC72131 ground	_
BO1 BO2 BO3 BO4	7 (6) 8 (7) 9 (8) 10 (9)	Output port	<ul> <li>Dedicated output pins</li> <li>The output states are determined by BO1 to BO4 bits in the serial data. Data: 0 = open, 1 = low</li> <li>A time base signal (8 Hz) can be output from the BO1 pin. (When the serial data TBC bit is set to 1.)</li> <li>Care is required when using the BO1 pin, since it has a higher on impedance that the other output ports (pins BO2 to BO4).</li> <li>All output ports are set to the open state following a power on reset.</li> </ul>	
101 102	11 (10) 13 (12)	I/O port	<ul> <li>I/O dual-use pins</li> <li>The direction (input or output) is determined by bits IOC1 and IOC2 in the serial data. Data: 0 = input port, 1 = output port</li> <li>When specified for use as input ports: The state of the input pin is transmitted to the controller over the DO pin. Input state: Iow = 0 data value high = 1 data value</li> <li>When specified for use as output ports: The output states are determined by the IO1 and IO2 bits in the serial data. Data: 0 = open, 1 = Iow</li> <li>These pins function as input pins following a power on reset.</li> </ul>	
PD	18 (16)	Charge pump output	• PLL charge pump output When the frequency generated by dividing the local oscillator frequency by N is higher than the reference frequency, a high level is output from the PD pin. Similarly, when that frequency is lower, a low level is output. The PD pin goes to the high impedance state when the frequencies match.	A02603
AIN AOUT	19 (17) 20 (18)	LPF amplifier transistor	The n-channel MOS transistor used for the PLL active low-pass filter.	A02504
IFIN	12 (11)	IF counter	<ul> <li>Accepts an input in the frequency range 0.4 to 12 MHz.</li> <li>The input signal is directly transmitted to the IF counter.</li> <li>The result is output starting the MSB of the IF counter using the DO pin.</li> <li>Four measurement periods are supported: 4, 8, 32, and 64 ms.</li> </ul>	

# Serial Data I/O Methods

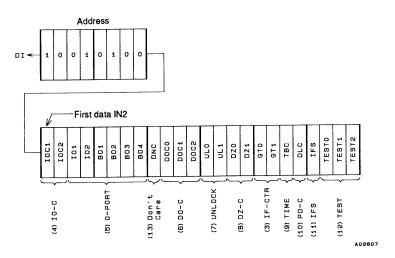
The LC72131 inputs and outputs data using the Sanyo CCB (computer control bus) audio LSI serial bus format. This LSI adopts an 8-bit address format CCB.

	Address									
	I/O mode	B0	B1	B2	B3	A0	A1	A2	A3	- Function
1	IN1 (82)	0	0	0	1	0	1	0	0	<ul> <li>Control data input mode (serial data input)</li> <li>24 data bits are input.</li> <li>See the "DI Control Data (serial data input) Structure" item for details on the meaning of the input data.</li> </ul>
2	IN2 (92)	1	0	0	1	0	1	0	0	<ul> <li>Control data input mode (serial data input)</li> <li>24 data bits are input.</li> <li>See the "DI Control Data (serial data input) Structure" item for details on the meaning of the input data.</li> </ul>
3	OUT (A2)	0	1	0	1	0	1	0	0	<ul> <li>Data output mode (serial data output)</li> <li>The number of bits output is equal to the number of clock cycles.</li> <li>See the "DO Output Data (serial data output) Structure" item for details on the meaning of the output data.</li> </ul>
			mal higi	<u>л</u>	 					A3 First Data IN1/2 First Data OUT First Data OUT First Data OUT A02805

- 1. DI Control Data (Serial Data Input) Structure
  - IN1 Mode



• IN2 Mode



### 2. DI Control Data Functions

No.	Control block/data				Functio	ins			Related data
	Programmable divider data	Data that	sets the d	ivisor of th	e programmable	e divider.			
	P0 to P15	A binary v DVS and S				_SB char	ges depending on		
		DVS	SNS	LSB	Divisor settir	ng (N)	Actual divisor	]	
		1	*	P0	272 to 655	35	Twice the value of the setting	]	
		0	1	P0	272 to 655	35	The value of the setting	]	
		0	0	P4	4 to 40		The value of the setting		
(1)		Note: P0	to P3 are	ignored w	hen P4 is the LS	SB.			
	DVS, SNS		-		MIN or FMIN) fo don't care)	or the pro	grammable divider, switches		
		DVS	SNS	Input	pin	l	nput frequency range		
		1	*	FMI	N		10 to 160 MHz	]	
		0	1	AM	IN		2 to 40 MHz	]	
		0	0	AM			0.5 to 10 MHz		
				-		ire" item	for more information.		
	Reference divider data	Reference	e frequenc	y (fref) sel	ection data.				
	R0 to R3	R3	R2	R1	R0	Re	ference frequency (kHz)	]	
		0	0	0	0		100	]	
		0	0 0	0	1		50 25		
		0	0		1		25		
		0	1	0	0		12.5		
			1 0		6.25 3.125				
		1	0	0	0		10	11	
		1	0	0	1		9		
(2)		1	0 0	1	0		5 1		
		1	1	0	0		3		
		1	1	0	1		15		
		1	1	1	0	PLL I	NHIBIT + Xtal OSC STOP	]	
		1	1	1	1		PLL INHIBIT		
		Note: PLI		oblo divido	r block and the l	Ecoupto	block are stapped the EMIN		
		The programmable divider block and the IF counter block are stopped, the FMIN, AMIN, and IFIN pins are set to the pull-down state (ground), and the charge pump goes to the high impedance state.							
	XS	Crystal res	sonator se	election					
		XS = 0: 4.							
		XS = 1: 7. The 7.2 M		ncv is sele	ected after the p	ower-on	reset.		
	IF counter control data	IF counter							
	CTE	CTE = 1:	Counter st	art					
		CTE = 0: 0							
	GT0, GT1	Determine	es the IF c	ounter me	asurement perio	od		IFS	
(2)		GT1	GT0	Mea	surement time (	ms)	Wait time (ms)		
(3)		0	0		4		3 to 4		
		0	1		8		3 to 4		
		1	0		32		7 to 8		
		1	1		64		7 to 8		
					ucture" item for				
(4)	I/O port specification data IOC1, IOC2	Data: 0 =	input mod	e, 1 = outp		<u> </u>			
	Output port data				ut from the $\overline{BO1}$	to BO4,	IO1 and IO2 output ports	IOC1	
(5)	BO1 to BO4, IO1, IO2	Data: 0 =	open, 1 =	IOW				1002	

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No.	Control block/data				Functions		Related data		
	DO pin control data	Data that	determine	es the DO	pin output				
	DOC0, DOC1, DOC2	DOC2	DOC1	DOC0	Г	DO pin state			
		0	0	0	Open				
		0	0	1	Low when the unlock	state is detected			
		0	1	0	end-UC*1				
		0	1	1	Open				
			0	0	Open The IO1 pin state <sup>*2</sup>				
			0	1	The $\overline{102}$ pin state <sup>2</sup> The $\overline{102}$ pin state <sup>2</sup>				
		1	1	1	Open				
		The open	state is s	elected aft	er the power-on reset.				
					F counter measurement	completion			
(6)		UL0, UL1, CTE,							
		DO pin	/	N		¥{	IOC1, IOC2		
			① Cou	l Int start	<li>② Cour</li>	nt end ③ CE: High			
						started (i.e., when CTE is changed			
		from zero to one), the DO pin automatically goes to the open state. 2 When the IF counter measurement completes, the DO pin goes low to indicate the measurement completion state.							
		2.							
		=							
						ne DO control data (DOC0 to DOC2) an OUT mode period with CE high)	•		
		wi	ill output t	he content	s of the internal DO seri	al data in synchronization with the			
	Unlock detection data					C control data (DOC0 to DOC2).			
	UL0, UL1	<ul> <li>Selects the phase error (øE) detection width for checking PLL lock.</li> <li>A phase error in excess of the specified detection width is seen as an unlocked state.</li> </ul>							
		UL1	UL0	ø	E detection width	Detector output			
<i>(</i> )		0	0	Stopped		Open	DOC0,		
(7)		0	1	0		øE is output directly	DOC1, DOC2		
		1	0	±0.55 µs	i	øE is extended by 1 to 2 ms	0002		
		1	1	±1.11 µs		øE is extended by 1 to 2 ms			
		Note: In the unlocked state the DO pin goes low and the UL bit in the serial data becomes zero.							
	Phase comparator				or dead zone.				
	control data DZ0, DZ1	DZ1	DZ0		Dead zo	one mode			
	-,	0	0	DZA					
(6)		0	1	DZB					
(8)			0	DZD					
			1	DZD					
			'						
		Dead zon	e widths:	DZA < DZ	B < DZC < DZD				
(9)	Clock time base TBC	from the E	301 pin. (I	BO1 data	s invalid in this mode.)	me base signal to be output	BO1		
	Charge pump control data DLC	Forcibly c	ontrols the	e charge p	ump output.				
		D	LC		Charge p	ump output			
(10)			0		operation				
(10)	1 Forced low								
		oso	illator sto	pping, dea		je (Vtune) going to zero and the VCO forcing the charge pump output to lock clearing circuit.)			

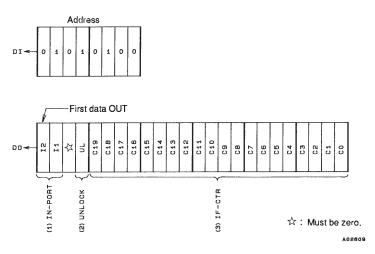
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No.	Control block/data	Functions	Related data			
(11)	IF counter control data IFS	<ul> <li>This data must be set 1 in normal mode. Though if this value is set to zero, the system enters input sensitivity degradation mode, and the sensitivity is reduced to 10 to 30 mVrms.</li> <li>* See the "IF Counter Operation" item for details.</li> </ul>				
(12)	LSI test data TEST 0 to TEST3	LSI test data     TEST0     TEST1     TEST2     These values must all be set to 0.     TEST2     These test data are set to 0 automatically after the power-on reset.				
(13)	DNC	Don't care. This data must be set to 0.				

## 3. DO Output Data (Serial Data Output)

# • OUT Mode



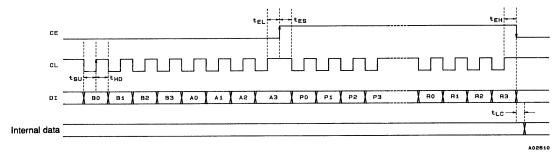
No.	Control block/data	Functions	Related data
(1)	I/O port data I2, I1	<ul> <li>Latched from the pin states of the IO1 and IO2 I/O ports.</li> <li>These values follow the pin states regardless of the input or output setting.</li> <li>Bits I2, I1 reflect the data latched into each input port when the device changes to OUT Mode. I1 ← IO1 pin state  High: 1 I2 ← IO2 pin state  Low: 0</li> </ul>	IOC1, IOC2
(2)	PLL unlock data UL	• Latched from the state of the unlock detection circuit. UL $\leftarrow$ 0: Unlocked UL $\leftarrow$ 1: Locked or detection stopped mode	UL0, UL1
(3)	IF counter binary data C19 to C0	• Latched from the value of the IF counter (20-bit binary counter). C19 $\leftarrow$ MSB of the binary counter C0 $\leftarrow$ LSB of the binary counter	CTE, GT0, GT1

## 4. DO Output Data

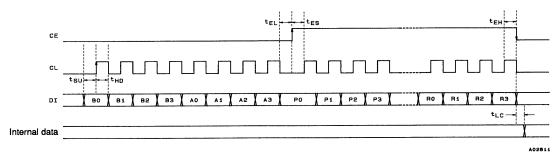
### LC72131, 72131M

5. Serial Data Input (IN1/IN2)  $t_{SU}$ ,  $t_{HD}$ ,  $t_{EL}$ ,  $t_{ES}$ ,  $t_{EH} \ge 0.75 \ \mu s$ ,  $t_{LC} \le 0.75 \ \mu s$ 

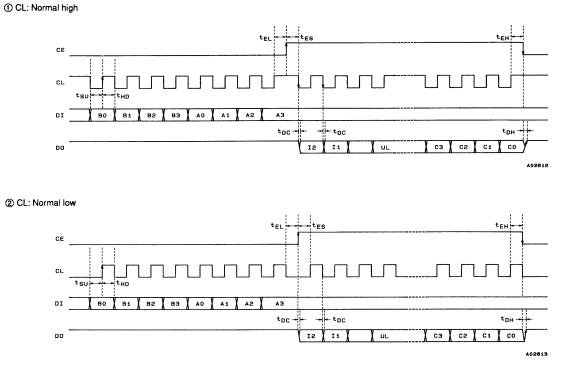
### ① CL: Normal high



② CL: Normal low

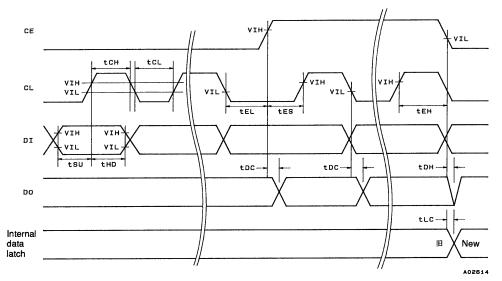


6. Serial Data Output (OUT)  $t_{SU}$ ,  $t_{HD}$ ,  $t_{EL}$ ,  $t_{ES}$ ,  $t_{EH} \ge 0.75 \ \mu s$ ,  $t_{DC}$ ,  $t_{DH} \le 0.35 \ \mu s$ 

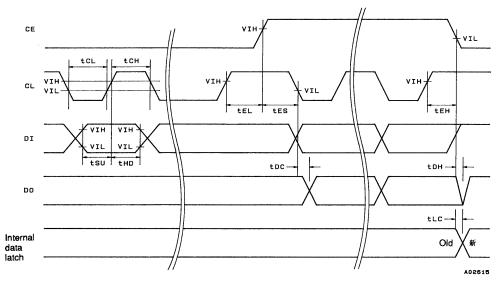


Note: Since the DO pin is an n-channel open-drain pin, the time for the data to change (t<sub>DC</sub> and t<sub>DH</sub>) will differ depending on the value of the pull-up resistor and printed circuit board capacitance.

## 7. Serial Data Timing



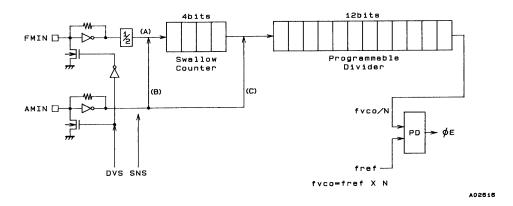
When stopped with CL low



When stopped with CL high

Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
Data setup time	t <sub>SU</sub>	DI, CL		0.75			μs
Data hold time	t <sub>HD</sub>	DI, CL		0.75			μs
Clock low-level time	t <sub>CL</sub>	CL		0.75			μs
Clock high-level time	t <sub>CH</sub>	CL		0.75			μs
CE wait time	t <sub>EL</sub>	CE, CL		0.75			μs
CE setup time	t <sub>ES</sub>	CE, CL		0.75			μs
CE hold time	t <sub>EH</sub>	CE, CL		0.75			μs
Data latch change time	t <sub>LC</sub>					0.75	μs
Data output time	t <sub>DC</sub>	DO, CL	Differs depending on the value of the pull-up resistor			0.35	
	t <sub>DH</sub>	DO, CE	and the printed circuit board capacitances.			0.35	μs

### **Programmable Divider Structure**



	DVS	SNS	Input pin	Set divisor	Actual divisor: N	Input frequency range (MHz)
A	1	*	FMIN	272 to 65535	Twice the set value	10 to 160
В	0	1	AMIN	272 to 65535	The set value	2 to 40
С	0	0	AMIN	4 to 4095	The set value	0.5 to 10
Nata & D						

Note: \* Don't care.

1. Programmable Divider Calculation Examples

 FM, 50 kHz steps (DVS = 1, SNS = \*, FMIN selected) FM RF = 90.0 MHz (IF = +10.7 MHz) FM VCO = 100.7 MHz
 PLL fref = 25 kHz (R0 to R1 = 1, R2 to R3 = 0)
 100.7 MHz (FM VCO) ÷ 25 kHz (fref) ÷ 2 (FMIN: divid

100.7 MHz (FM VCO)  $\div$  25 kHz (fref)  $\div$  2 (FMIN: divide-by-two prescaler) = 2014  $\rightarrow$  07DE (HEX)

	Ę							,					<u> </u>										
0	1	1	1	1	0	1	1	1	1	1	0	0	0	0	0	*	1			1	1	0	0
0 0	н 1	CU CL	m d.	Р 4	Б С	8	P7	8d	6 d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	xs	в	H1	R2	R3

A02617

SW, 5 kHz steps (DVS = 0, SNS = 1, AMIN high-speed side selected) SW RF = 21.75 MHz (IF = +450 kHz) SW VCO = 22.20 MHz
PLL fref = 5 kHz (R0 = R2 = 0, R1 = R3 = 1)
22.2 MHz (SW VCO) ÷ 5 kHz (fref) = 4440 → 1158 (HEX)

_		3		_		5		_	_	<u> </u>		<i></i>	;	L									
0	0	0	1	1	0	1	0	1	0	0	0	1	0	0	0	1	0			0	1	0	1
od	4	54	£Ч	4 4	5	9d	Р7	8d	6 d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	sx	в	в1	ВZ	R3

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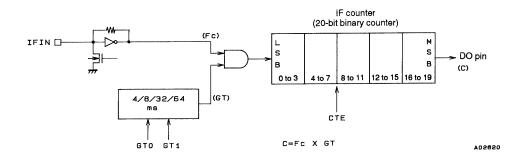
MW, 10 kHz steps (DVS = 0, SNS = 0, AMIN low-speed side selected) MW RF = 1000 kHz (IF = +450 kHz) MW VCO = 1450 kHz
PLL fref = 10 kHz (R0 to R2 = 0, R3 = 1)

1450 kHz (MW VCO)  $\div$  10 kHz (fref) = 145  $\rightarrow$  091 (HEX)

				<u></u>	;	<u> </u>		_		<u> </u>													
*	*	*	*	1	0	0	0	1	0	0	1	0	0	0	0	0	0			0	0	0	1
РО	1 d	ЪЗ	ЪЗ	P.4	ß	9 4	P7	8d	6d	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	sx	он	н1	В2	ВЗ

#### **IF Counter Structure**

The LC72131 IF counter is a 20-bit binary counter. The result, i.e., the counter's msb, can be read serially from the DO pin.



GT1	GT0	Measurement time									
GII	GIU	Measurement period (GT) (ms)	Wait time (twu) (ms)								
0	0	4	3 to 4								
0	1	8	3 to 4								
1	0	32	7 to 8								
1	1	64	7 to 8								

The IF frequency (Fc) is measured by determining how many pulses were input to an IF counter in a specified measurement period, GT.

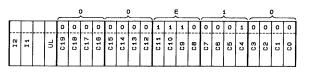
$$Fc = \frac{C}{GT}$$
 (C = Fc × GT)

C: Count value (number of pulses)

- 1. IF Counter Frequency Calculation Examples
  - When the measurement period (GT) is 32 ms, the count (C) is 53980 hexadecimal (342400 decimal): IF frequency (Fc) =  $342400 \div 32 \text{ ms} = 10.7 \text{ MHz}$

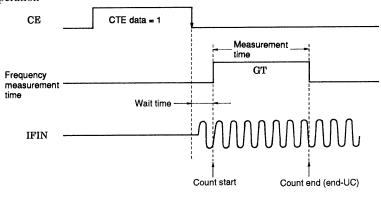
				_		5		_	:	<u> </u>				₹			_	<u> </u>				2	
				0	1	0	1	0	0	1	1	1	0	0	1	1	0	0	0	0	0	0	0
I2	11		Ч	C19	C18	C17	C16	C15	C14	C13	C12		C10	60	CB	C7	CG	CS	C 4	сэ	25	C1	о С
		<b>.</b>									L												A0

• When the measurement period (GT) is 8 ms, the count (C) is E10 hexadecimal (3600 decimal): IF frequency (Fc) =  $3600 \div 8$  ms = 450 kHz



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2. IF Counter Operation



Before starting the IF count, the IF counter must be reset in advance by setting CTE in the serial data to 0.

The IF count is started by changing the CTE bit in the serial data from 0 to 1. The serial data is latched by the LC72131 when the CE pin is dropped from high to low. The IF signal must be supplied to the IFIN pin in the period between the point the CE pin goes low and the end of the wait time at the latest. Next, the value of the IF counter at the end of the measurement period must be read out during the period that CTE is 1. This is because the IF counter is reset when CTE is set to 0.

Note: When operating the IF counter, the control microprocessor must first check the state of the IF-IC SD (station detect) signal and only after determining that the SD signal is present turn on IF buffer output and execute an IF count operation. Autosearch techniques that use only the IF counter are not recommended, since it is possible for IF buffer leakage output to cause incorrect stops at points where there is no station.

#### IFIN minimum input sensitivity standard

			f (MHz)
IFS	$0.4 \le f < 0.5$	0.5 ≤ f < 8	$8 \le f \le 12$
1: Normal mode	40 mVrms (0.1 to 3 mVrms)	40 mVrms	40 mVrms (1 to 10 mVrms)
0: Degradation mode	70 mVrms (10 to 15 mVrms)	70 mVrms	70 mVrms (30 to 40 mVrms)

Note: Values in parentheses are actual performance values presented as reference data.

#### **Unlock Detection Timing**

1. Unlock Detection Determination Timing

Unlocked state detection is performed in the reference frequency (fref) period (interval). Therefore, in principle, unlock determination requires a time longer than the period of the reference frequency. However, immediately after changing the divisor N (frequency) unlock detection must be performed after waiting at least two periods of the reference frequency.

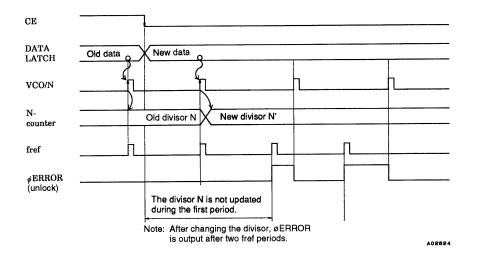
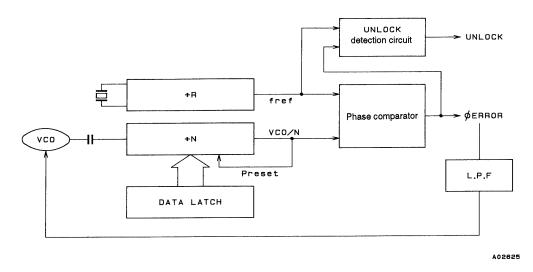
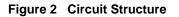


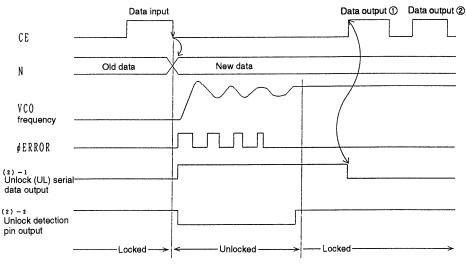
Figure 1 Unlocked State Detection Timing

For example, if fref is 1 kHz, i.e., the period is 1 ms, after changing the divisor N, the system must wait at least 2 ms before checking for the unlocked state.





#### 2. Unlock Detection Software

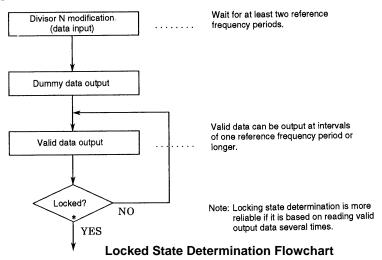




#### 3. Unlocked State Data Output Using Serial Data Output

In the LC72131, once an unlocked state occurs, the unlocked state serial data (UL) will not be reset until a data input (or output) operation is performed. At the data output ① point in Figure 3, although the VCO frequency has stabilized (locked), since no data output has been performed since the divisor N was changed the unlocked state data remains in the unlocked state. As a result, even though the frequency has stabilized (locked), the system remains (from the standpoint of the data) in the unlocked state.

Therefore, the unlocked state data acquired at data output <sup>(1)</sup>, which occurs immediately after the divisor N was changed, should be treated as a dummy data output and ignored. The second data output (data output <sup>(2)</sup>) and following outputs are valid data.

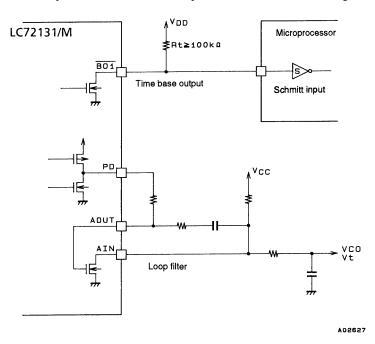


4. Directly Outputting Unlocked State Data from the DO Pin (Set by the DO pin control data)

Since the unlocked state (high = locked, low = unlocked) is output directly from the DO pin, the dummy data processing described in section 3 above is not required. After changing the divisor N, the locking state can be checked after waiting at least two reference frequency periods.

#### Clock Time Base Usage Notes

The pull-up resistor used on the clock time base output pin ( $\overline{BO1}$ ) should be at least 100 k $\Omega$ . This is to prevent degrading the VCO C/N characteristics when a loop filter is formed using the built-in low-pass filter transistor. Since the clock time base output pin and the low-pass filter have a common ground internal to the IC, it is necessary to minimize the time base output pin current fluctuations and to suppress their influence on the low-pass filter. Also, to prevent chattering we recommend using a Schmitt input at the controller (microprocessor) that receives this signal.



#### Other Items

DZ1	DZ0	Dead zone mode	Charge pump	Dead zone
0	0	DZA	ON/ON	0 s
0	1	DZB	ON/ON	-0 s
1	0	DZC	OFF/OFF	+0 s
1	1	DZD	OFF/OFF	+ +0 s

1. Notes on the Phase Comparator Dead Zone

Since correction pulses are output from the charge pump even if the PLL is locked when the charge pump is in the ON/ON state, the loop can easily become unstable. This point requires special care when designing application circuits.

The following problems may occur in the ON/ON state.

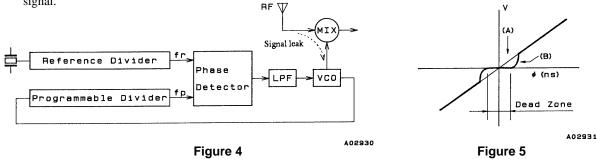
- Side band generation due to reference frequency leakage
- Side band generation due to both the correction pulse envelope and low frequency leakage

Schemes in which a dead zone is present (OFF/OFF) have good loop stability, but have the problem that acquiring a high C/N ratio can be difficult. On the other hand, although it is easy to acquire a high C/N ratio with schemes in which there is no dead zone, it is difficult to achieve high loop stability. Therefore, it can be effective to select DZA or DZB, which have no dead zone, in applications which require an FM S/N ratio in excess of 90 to 100 dB, or in which an increased AM stereo pilot margin is desired. On the other hand, we recommend selecting DZC or DZD, which provide a dead zone, for applications which do not require such a high FM signal-to-noise ratio and in which either AM stereo is not used or an adequate AM stereo pilot margin can be achieved.

#### Dead Zone

The phase comparator compares fp to a reference frequency (fr) as shown in Figure 4. Although the characteristics of this circuit (see Figure 5) are such that the output voltage is proportional to the phase difference  $\emptyset$  (line A), a region (the dead zone) in which it is not possible to compare small phase differences occurs in actual ICs due to internal circuit delays and other factors (line B). A dead zone as small as possible is desirable for products that must provide a high S/N ratio.

However, since a larger dead zone makes this circuit easier to use, a larger dead zone is appropriate for popularlypriced products. This is because it is possible for RF signals to leak from the mixer to the VCO and modulate the VCO in popularly-priced products in the presence of strong RF inputs. When the dead zone is narrow, the circuit outputs correction pulses and this output can further modulate the VCO and generate beat frequencies with the RF signal.



2. Notes on the FMIN, AMIN, and IFIN Pins

Coupling capacitors must be placed as close as possible to their respective pin. A capacitance of about 100 pF is desirable. In particular, if a capacitance of 1000 pF or over is used for the IF pin, the time to reach the bias level will increase and incorrect counting may occur due to the relationship with the wait time.

3. Notes on IF Counting  $\rightarrow$  SD must be used in conjunction with the IF counting time

When using IF counting, always implement IF counting by having the microprocessor determine the presence of the IF-IC SD (station detect) signal and turn on the IF counter buffer only if the SD signal is present. Schemes in which auto-searches are performed with only IF counting are not recommended, since they can stop at points where there is no signal due to leakage output from the IF counter buffer.

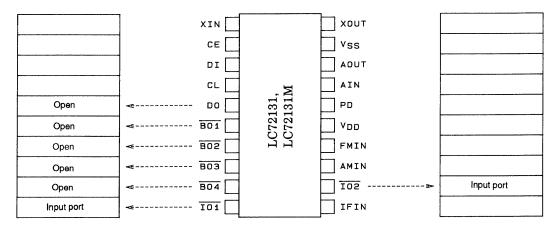
4. DO Pin Usage Techniques

In addition to data output mode times, the DO pin can also be used to check for IF counter count completion and for unlock detection output. Also, an input pin state can be output unchanged through the DO pin and input to the controller.

5. Power Supply Pins

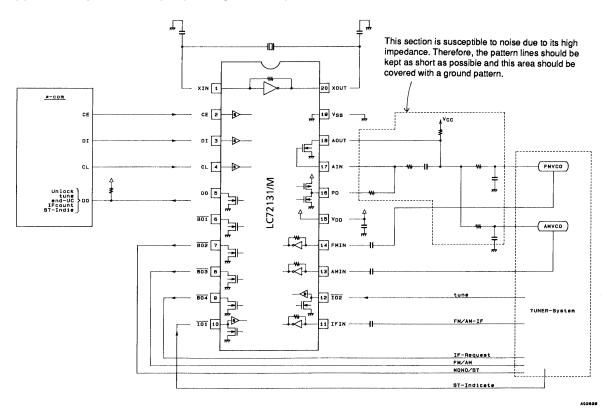
A capacitor of at least 2000 pF must be inserted between the power supply  $V_{DD}$  and  $V_{SS}$  pins for noise exclusion. This capacitor must be placed as close as possible to the  $V_{DD}$  and  $V_{SS}$  pins.

### Pin States After the Power ON Reset



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### Application System Example (Package: MFP20)



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