SONY

CXA1082BQ/BS

Servo Signal Processor for CD Player

Description

CXA1082BQ/BS is a bipolar IC designed for the servo control of the compact disc player. The only difference from CXA1082AQ/AS is the FZC threshold.

Features

- Single power supply, 5 V
- Dual power supply, ±5 V
- Low power consumption (165 mW Typ.: ±5 V, 100 mW, 5 V)
- Servo functions same as the CX20108 (focus, tracking, and sled servo)
- Built-in auto sequencer
- · Built-in LPF for spindle servo
- Built-in loop filter and VCO for EFM clock reproduction PLL
- · Fewer external parts
- Built-in circuit for preventing sled runaway
- · Built-in circuit for disc defects
- Built-in anti-shock circuit
- High-speed access using a linear motor
- Sharing of the serial data bus of the microcomputer with the CX23035 or CXD1135Q
- Compatible in the upword with the CX20108 for microcomputer software
- The peaks of focus search, track jump, and sled kick pulse can be set with external resistors.

Functions

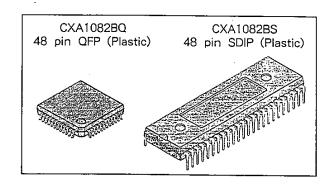
- · Focus servo control
- Tracking servo control
- Sled servo control
- Spindle servo

LPF, drive amplifier

- EFM clock reproduction PLL Loop filter, 8.64 MHz VCO
- Auto sequencer Built-in RAM

Structure

Bipolar silicon monolithic IC



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Absolute Maximum Ratings (Ta = 25°C)

 Supply voltage 	Vcc - VEE	12	V
 Operating temperature 	Topr	-20 to +75	°C
 Storage temperature 	Tstg	-55 to $+150$	°C
 Allowable power dissipation 	PD CXA1082E	3Q 833	mW
	CXA1082E	3S 1330	mW

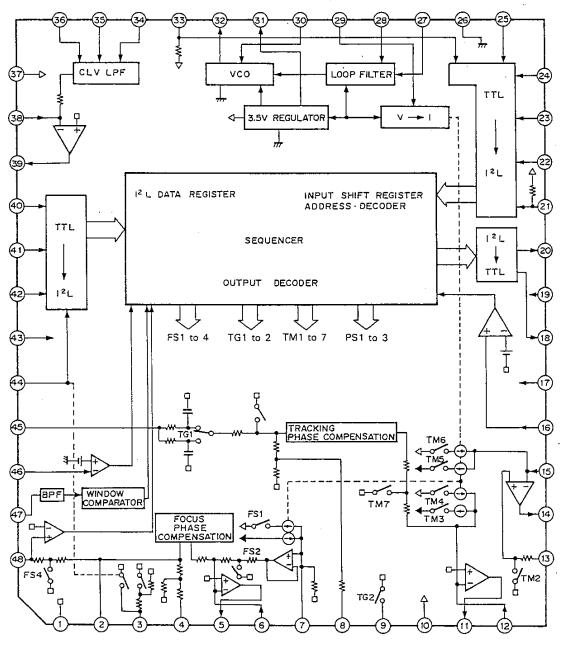
Recommended Operating Conditions

• Sup	pply voltage	
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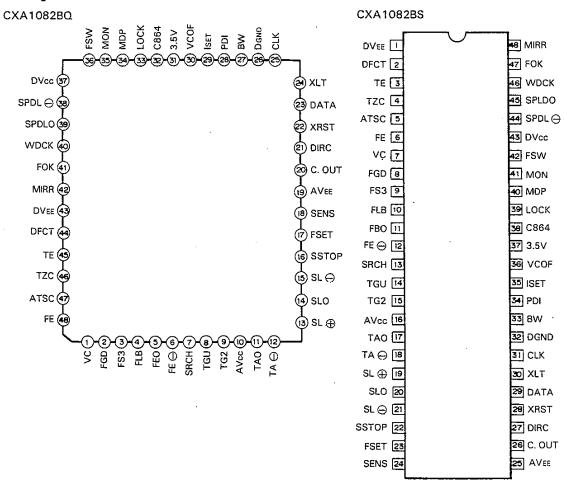
 Vcc — VEE
 4 to 11
 \

 Vcc — DGND
 4 to 5.5
 \

Block Diagram



Pin Configuration



Pin Description

Numbers in () show CXA1082BS

No.	Symbol	Equivalent Circuit	Description
2 (8)	FGD	20 N	Connect a capacitor between this pin and pin 3 (9) to reduce the high-frequency gain.
3 (9)	FS3	3 180 46K W	The high-frequency gain of the focus servo can be changed by switching FS3 ON or OFF.

No.	Symbol	Equivalent Circuit	Description
4 (10)	FLB	40K W W W	Time constant external pin for rising low bandwidth of the focus servo.
5 (11) 11 (19) 14 (20) 39 (45)	FEO TAO SLO SPDLO	9 Vcc 11 W 100 µА VEE	Focus drive output Tracking drive output Sled drive output Spindle drive output
6 (12)	FE —	6 ¥90К 2,5µА VEE	Inverse input pin for the focus amplifier.
7 (13)	SRCH	7 180 50K ₹3.5µA 1 11µA VEE	Pin for providing a time constant to generate the focus search waveform.
8 (14)	TGU	Vcc 20 K W W W 8 82 K	Pin for providing a time constant to switch the tracking gain of high-frequency.
9 (15)	TG2	9 180 W 470К 2µA VEE	Pin for providing a time constant to change the high-frequency tracking gain.

No.	Symbol	Equivalent Circuit	Description
12 (18)	TA —	Vcc 180 ≥90K 180 111µA VEE	Inverse input pin for the tracking amplifier.
13 (19)	SL +	10K 10K W	Non-inverse input pin for the sled amplifier.
15 (21)	SL	180 15 3 µА 122 µА VEE	Inverse input pin for the sled amplifier.
16 (22)	SSTOP	180 Vcc 180 VEE	Pin for detecting a signal for the ON/OFF limit switch of the innermost part of the disc.
17 (23)	FSET	Vcc 17 180 15K ₹15K VEE	Pin for setting the peak frequency of the focus, tracking phase compensation and f0 of the CLV LPF.
18 (24)	SENS	180 ¥20K Vcc	Pin to output FZC, AS, TZC, SSTOP and BUSY by command from CPU.
20 (26)	C. OUT	(2) 100K	Track number count signal output

No.	Symbol	Equivalent Circuit	Description
21 (27) 22 (28) 23 (29) 24 (30) 25 (31) 33 (39)	DIRC XRST DATA XLT CLK LOCK	21 Vcc 23 47 К 23 M 180 Vee	Pin for one-track jump Contains a 47 kΩ pull-up resistor. Reset input pin, reset at "L" Serial data input from CPU Latch input from CPU Serial data transfer clock input from CPU Pin for the operation of the sled runaway prevention circuit at "L" Contains a 47 kΩ pull-up resistor.
27 (33)	BW	2.7 K 4.7 K ₹ VEE	Pin for providing a time constant for the loop filter.
28 (34)	PDI	4.7 K M 100 P	Input pin for the CX23035/CXD1135 phase comparator output PDO.
29 (35)	ISET	180 10K W VEE	Current is input, determining the peaks of focus search, track jump, and sled kick.
30 (36)	VCOF	39 TOK	The free-running frequency of VCO is almost proportional to the resistance value between this pin and pin 31 (37).
32 (38)	C864	₹DVcc ₹ 180	Output pin of 8.64 MHz VCO.

No.	Symbol	Equivalent Circuit	Description
34 (40)	MDP	180 W 20K 14.5K	Pin for connecting the CX23035/CXD1135 MDP pin.
35 (41)	MON	Vcc 180 35 W 20.5K VEE	Pin for connecting the CX23035/CXD1135 MON pin.
36 (42)	FSW	Vcc 180 36 W 220K VEE	Pin for providing an external LPF time constant of the CLV servo error signal.
38 (44)	SPDL —	180 VEE	Inverse input pin for the spindle drive amplifier.
40 (46) 41 (47) 42 (48) 44 (2)	WDCK FOK MIRR DFCT	180 41 42 44 VEE	Clock input for auto sequence, usually 88.2 kHz is input FOK signal input pin Mirror signal input pin Defect signal input pin for the operation of the defect countermeasure circuit at "H"
45 (3)	ΤE	680K 680K W	Input pin for tracking error signals.

No.	Symbol	Equivalent Circuit	Description
46 (4)	TZC	Vcc √7 ji A 180 √75 K ₹	Input pin for the zero-cross tracking comparator.
47 (5)	ATSC	Vcc 470K 470K W 47P VEE 47P	Input pin of the window comparator for ATSC detection.
48 (6)	FE	Vcc 7μ 180 ¥20K VEE	Input pin for focus error signals.

Electrical Characteristics

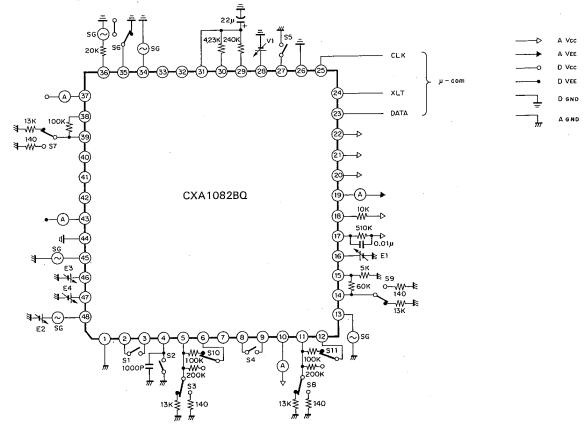
			T					sw	condi	tions							Bias o					1	= 2.5V AV _{εε} , DV _{εε} Description of output			GND=	
Test No.		Test items	Symbol	Sì	S2	S3	S4	S5	S6	S7	S8	S9	S10	Sii	SD	E1	E	$\overline{}$	E3	E4	Input point	Test point	waveform	Min.	Тур.	Max.	Unit
ł	Sup	oply current 1	VICC												00	0	0	T	0	0		10	Measure after resetting	2, 8	5, 5	8, 2	· mA
2	Sup	oply current 2	Dicc													П	П	1	П			37		10.8	. 15.0	19.2	mА
3	Sup	oply current 3	A, DIEE													\top	Ħ	\top	\Box	1		19 43		9.8	13, 0	16, 2	mА
4	Sup	oply current 4	IDGND													Ħ	П		T			26		4.8	7.5	10, 2	mΑ
5		DC voltage gain	Grao												08	П	П				48	5	SG == 10 Hz, 200mVp-p	18.0	21.0	24.0	dB
6		Feedthrough	VEROF	0	0														\prod	Τ	П		SG = 10kHz, 40mVp-p, Gain differ- ence between 08 and 00 of SD			-35	dB
7	Į,	Max. output voltage I	VFEOI										0		08	П	П				П		SG = 0.5VDC	1, 98			٧
8	Focus servo	Max. output voltage 2	VFEO2										0		08		П					l i	SG = -0.5VDC			-1. 98	٧
9	OV	Max. output voltage 3	VFE03			0							0		08-		П					П	SG = 0.5VDC	1, 18			٧
10		Max. output voltage 4	VITEO4			0							0		08				П				SG = -0.5VDC			-i.18	٧
n		Search output voltage 1	VSRCIII												02		П	T	П					-0. 64	-0, 55	-0, 36	٧
12		Search output voltage 2	VSRC112												03		П		П					0.36	0, 55	0, 64	٧
13		DC voltage gain	GTEO				0								25		П				45	11	SG = 10 Hz, 500mVp-p	11.6	14.6	17, 6	dB
14		Feedthrough	VTEOF				0								13		П		П		П		SG=10kHz,100mVp-p,Gain differ- ence between 25 and 20 of SD			~39	dB
15	-3	Max, output voltage i	VTEPI											0	25								SG = -1.5VDC	1. 98			٧
16	Tracking	Max. output voltage 2	VTEP2											0	25				П		Т		SG = 1.5VDC			~1.98	v
17	g servo	Max. output voltage 3	VTEP3								0			0	25								SG = -1.5VDC	1, 18			٧
18	•	Max. output voltage 4	VTEP4								0			0	25	П							SG = 1.5VDC			-l, 18	٧
19		Jump output voltage I	VJUMPI												2C	П	П		П					-0, 64	-0, 55	-0, 36	٧
20		Jump output voltage 2	VJUMP2												28		П		П					0, 36	0,55	0, 64	٧
21		DC voltage gain	GSLO												25		П		П		13	14	SG = 101fz Openloop gain	50	56	62	dB
22	Sled	Max. output voltage 1	VSLPI												25		П		П				SG = 0.4VDC	1, 98			٧
23	ed servo	Max. output voltage 2	VSLP2												25		П	T	П				SG = -0.4VDC			-1, 98	V
24	ò	Max. output voltage 3	VSLP3									0			25	\sqcap	\prod	T	\prod			П	SG = 0.4VDC	i, 18			٧

*Serial data (hex)

Test		Test items	Symbol					sw	cond	itions					. *	В	ias co	nditi	ons	lnou	Test	Description of output				
No.		rest tterns	3,111001	Sı	S2	S3	S4	S5	S6	S7	S8	S9	S10	Sil	SD	E1	E2	E3	Ε4	point	poin	waveform and test method	Min.	Тур.	Max.	Unit
25		Max. output voltage 4	VSLP4									0			25	0	0	0	0	13	14	SG = -0.4VDC		l —	-1.18	v
26	Sled s	Feed through	VSLOF															Т		\top	П	SG=10kHz,200mVp-p,Gain difference between 25 and 20 of SD			-34	dB
27	Servo	Kick output voltage l	VKICKI												22			П				7.7.	-0, 75	-0, 6	-0, 45	V
28		Kick output voltage 2	VKICK2												23			П	11		Ħ	<u></u>	0, 45	0, 6	0.75	V
29	ļ	Spindle servo gain	GSPO						0											34	39	SG = 10Hz, 200mVp-p	14	16,5	19	dВ
30	Spi	Max. output voltage'1	VSPP1						0							\top			\top	П	П	SG = 1.0VDC	1.78			v
31	Spindle so	Max. output voltage 2	VSPP2						0									П	 	\sqcap	\sqcap	SG = 1.0VDC			-1.78	V
32	Servo	Max, output voltage 3	VSPP3			-			0	0									11	П		SG = 1.0VDC	1, 13			v
33		Max. output voltage 4	VSPP4						0	0						T		T	11	\prod	Ħ	SG = ~1.0VDC			-1, 13	v
34		PLL Reg. output voltage	Vreg													\top			11	Ė	31	DC voltage	3, 3	3, 5	3, 85	v
35	PLL	Self-running frequency	Fvco														0	T			32	$V_1 = 0 \text{mV}$	7.4	8.6	9.7	MHz
36	į.	Frequency deviation !	ΔFı																11			Frequency deviation from FVCO, V ₁ =148mV	7	11	15	%
37		Frequency deviation 2	ΔFz																			V ₁ =-148mV	-15	-11	-7	%
38	SENS	low level	VSENS																П		18				-1, 98	v
39	COUT	low level	VCOUT															1	П		20				-1, 98	v
40	FZC value	threshold	VTZC												00		•	T	\prod	48	18		39	50	61	mV
41	ATSC value	threshold	Vatsci												10		0	T		47		* Value of E when SENS	-45	-26	-7	mV
42	ATSC	threshold	VATSC2												10		\top	Ť		47		becomes High (=1.1V) by E1 to E4 varying	7	26	45	мV
43	TZC i	threshold	VTZC						İ						20	П	\Box	•	0	46		SG = 0V	-20	0	20	mV
44	SSTO value	P threshold	VSSTOP								\neg				30	-	\top	0	$\top \top$	16		- WT	-65	-50	-35	mV

Electrical Characteristics Test Circuit

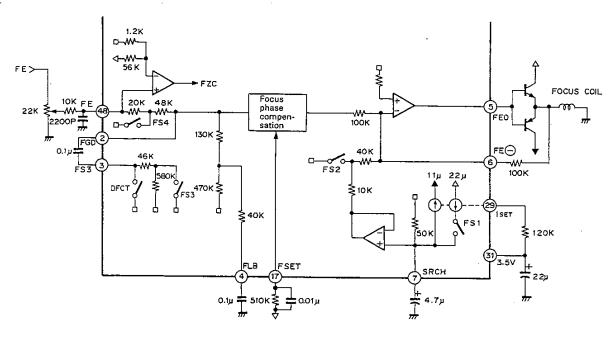
(See the Pin Configuration for CXA1082BS)



Description of Functions

Focus servo system

(See the Pin Configuration for CXA1082BS)



The above is a block diagram of the focus servo system.

When FS3 is switched on, the high frequency gain can be reduced by forming a low frequency time constant through a capacitor connected across pins 2 and 3 and the internal resistor.

The capacitor across the pin 4 and GND has a time constant to raise the low frequency usually playback condition.

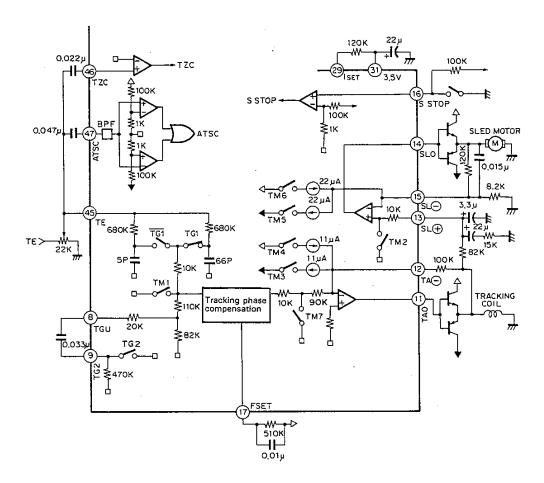
The peak frequency of the focus phase compensation is inversely proportional to the resistor connected to pin 17 (about 1.2 kHz when the resistor is 510 k Ω).

The focus search peak becomes about ± 1.1 Vp-p with the above constant. The peak is inversely proportional to the resistor connected across the pins 29 and 30. However, when this resistor is varied, the peaks of track jump and sled kick also vary.

The FZC comparator invert input is set to 2% of the difference between the reference voltage VCC and VC (pin 1): $2\% \times (VCC - VC)$.

Note) A resistor of 510 k Ω is recommended for pin 17.

Tracking sled servo system



The above is a block diagram of the tracking sled servo system.

The capacitor across pins 8 and 9 has a time constant to lower the high frequency when TG2 is switched off. The tracking phase compensation peak frequency is inversely proportional to the resistor connected to pin 17 (about 1.2 kHz when the resistor is 510 k Ω).

For a tracking jump in the FWD or REV direction, TM3 or TM4 are set to ON. At this time, the peak voltage fed to the tracking coil is determined by the TM3 and TM4 current values and the feedback resistor from pin 12. That is:

Track jump peak voltage = TM3 (TM4) current value x feedback resistor value
The FWD or REV sled kick is done by setting TM5 or TM6 to ON. At this time, the peak voltage added
to the sled motor is determined by the TM5 or TM6 current value and the feedback resistor from pin 15.

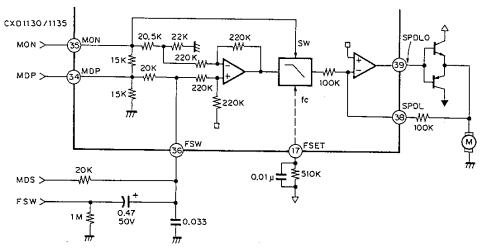
Sled jump peak voltage = TM5 (TM6) current value x feedback resistor value Each SW current value is determined by the resistor connected to pins 29 and 31. When the resistor is at about 120 $k\Omega$,

TM3 or TM4 is \pm 11 μ A and TM5 or TM6 is \pm 22 μ A.

This current value is almost inversely proportional to the resistor, variable within a range of about 5 to 40 μ A for TM3.

S STOP is the ON/OFF detection signal for the limit SW of the sled motor's innermost circumference.

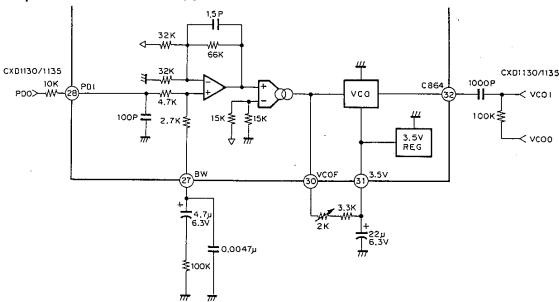
Spindle servo and LPF



The 200 Hz LPF is formed with 0.033 μ F and 20 k Ω connected to pin 36 and the secondary LPF is formed with the built-in LPF (fc up to 200 Hz with 510 k Ω for pin 17), and the carrier component of the CLV servo error signals MDS and MDP is eliminated.

In the CLV-S mode, FSW becomes L and the pin 36 LPF fc lowers, strengthening the filter further. With the pin 17 resistor connected to Vcc, fc does not vary with power supply voltage fluctuations. **Note**) Use the phase compensation instead of MDS when the CX23035 is used.

VCO loop filter and 8.64 MHz VCO



The phase compensation output PDO input from pin 28 has its PWM carrier component removed in the loop filter. Then, the V-I conversion is made and the free-running frequency setting current from pin 30 is added to control the VCO frequency. The VCO self-running frequency is almost inversely proportional to the resistor across pins 30 and 31. This resistor is set so that the PLL capture range center matches the 4.3218 MHz at pin 70 of the CXD1135/1130.

Commands

The input data to activate this IC consists of 8-bits. It shall be represented as \$XX in two hexadecimal digits. (X denotes 0 to F). Instructions for the CXA1082BQ are classified into 8 types — \$0X to 7X.

1. \$0X [SENSE Pin 18 is "FZC"]

This instruction is related to the focus servo control.

The bit configuration is as follows:

D7	D6	D5	D4	D3	D2	D1	DO
0	0	0	0	FS4	FS3	FS2	FS1

The four switches FS1 to FS4 are related to focusing, and correspond to D0 to D3.

\$00 At FS1 = 0, Pin 7 is charged to $(22\mu A - 11\mu A) \times 50 \text{ k}\Omega = 0.55 \text{ V}$). If FS2 = 0, this voltage is not output and the output of Pin 5 remains 0 V.

\$02 From the above state, FS2 only becomes 1 and a negative output is output to Pin 5. This voltage level is stipulated as follows:

(22 μA
$$-$$
 11 μA) x 50 kΩ x Resistance value between Pin 5 and Pin 6 (1)

\$03 From the above state, FS1 becomes 1 and the current supply of $+22 \,\mu\text{A}$ is separated. Then, the CR charge/discharge circuit is formed and Pin 7 voltage decreases as time passes, as shown in Fig. 1.

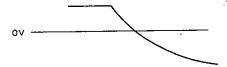


Fig. 1 Voltage of pin 7 as FS1 changes from 0 to 1

The time constant is formed by 50 $k\Omega$ and an external capacitor.

By giving instructions \$02 and \$03 alternately, the focus search voltage is produced (Fig. 2).

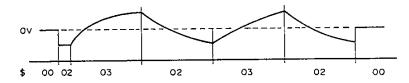


Fig. 2 Production of search voltage by \$02 and \$03 (Voltage of Pin 5)

1) Description of FS4

This switch is placed between the focus error input 48 and the focus phase compensation, serving to switch on and off the focus servo.

$$$00 \rightarrow $08$$
Focus off \leftarrow Focus on

2) Focusing procedure

Assume the polarity as follows:

- a) The lens moves away or toward the disc in searching.
- b) At this time, the output voltage of Pin 5 varies from negative to positive.
- c) Further, the focus S-curve changes as follows:

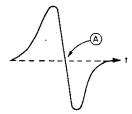


Fig. 3 S-curve

The focus servo is activated at the operating point of (A) as shown in Fig. 3. Generally, a logical product (AND) with the Focus-OK signal is used to switch on the focus servo in focus searching and to prevent a malfunction while passing the (A) point in Fig. 3.

This IC is designed so FZC (Focus Zero Cross) is output from the Sense Pin 18 as the A point passing signal.

The Focus-OK indicates a signal is in focus (focus is enabled in this case). In summary, the following time chart shows how to obtain the focus.

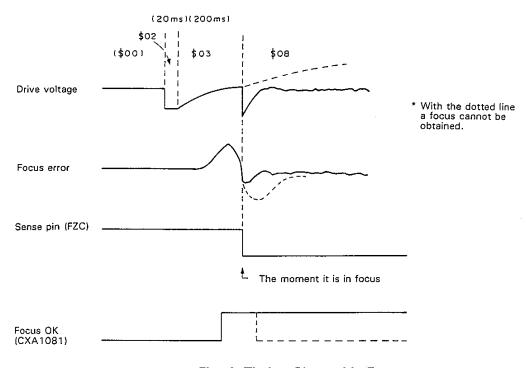


Fig. 4 Timing Chart of In-Focus

Care should be taken here that \$08 is instructed in the shortest time after FZC changes from H to L.

For this purpose, the (b) sequence required for software is better than (a).

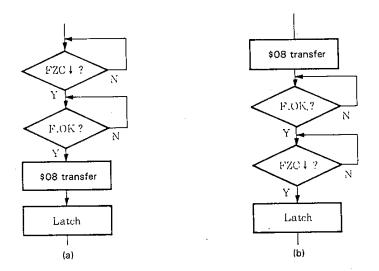


Fig. 5 Bad Sequence and Good Sequence

3) On the Sense Pin 18

What is output to the Sense Pin depends on input data as follows:

FZC is output with \$0X.

AS is output with \$1X.

TZC is output with \$2X.

SSTOP is output with \$3X.

BUSY is output with \$4X.

HIGH-Z is output with \$5X to 7X.

Higher instructions than \$7X are codes for the CXD1135 and several outputs are obtained by connecting to the CXD1135 "SENS" pin.

2. \$1X (SENS Pin 18 is "AS")

This instruction refers to ON/OFF of TG1, TG2 and the break circuit. The bit configuration is as follows:

D7	D6	D5	D4	.D3	D2	D1	DO
0	0	0	1	ANTI	Break	TG2	TG1
				SHOCK	Circuit		
				ON/OFF	ON/OFF		

TG1, TG2

The purpose of these switches is to switch on or off Up/Normal of the tracking servo gain. The break circuit refers to a mechanism which prevents a volatile actuator servo circuit. After jumps of 100 tracks or 10 tracks, the servo circuit is out of the linear range and the actuator often sets tracks wrong. Using a feature that the RF envelope and the tracking error are out of phase by 180° when the actuator crosses the tracks outwardly and vice versa, unwanted tracking errors are cut to break this undesirable jumping.

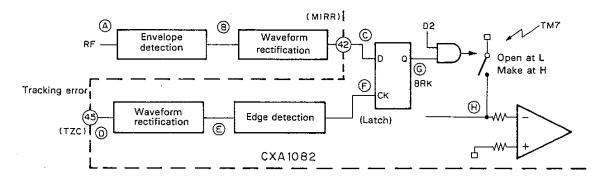


Fig. 6 TM7 Movement (Break Circuit)

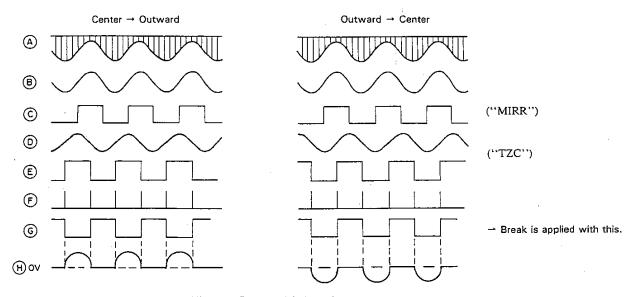


Fig. 7 External Waveform

3. \$2X (SENS Pin 18 is "TZC")

This instruction refers to ON/OFF of the tracking servo and thread servo as well as generation of the jump pulse and speed feeding pulse in accessing.

D7 .	D6	D5	D4	D3 D2	D1 D0
0	0	1	0	Tracking control	Sled control
-	_	·	•	00 off	00 off
				01 Servo ON	01 Servo ON
				10 F-JUMP	10 F-speed feed
				11 R-JUMP ↓	11 R-speed feed
				TM1,TM3,TM4	TM2,TM5,TM6

DIRC Pin 21 and 1 Track Jump

Generally, for a 1-track jump, an acceleration pulse is added and a deceleration pulse is given for a specified time from the moment a tracking error passes the 0 point; then the tracking servo is switched on again. For the 100-track jump to be explained in the next item, as long as the number of tracks is about 100 there is no problem, but the 1-track jump must be exactly, requiring the above complicated procedure. For a 1-track jump of a CD player, both the acceleration and deceleration take about 300 to 400 μ s. When software is used to execute this operation, it will be as in the flow chart of Fig. 9, but practically it takes time to transfer data.

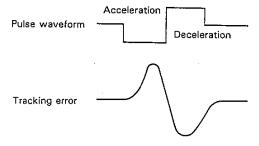


Fig. 8 Pulse Waveform and Tracking Error of 1-Track Jump

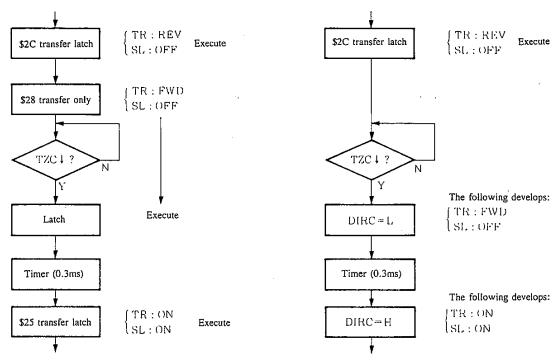


Fig. 9 1-Track Jump without DIRC 21

Fig. 10 1-Track Jump with DIRC 21

For this IC, the "DIRC" (Direct Control) Pin is provided for simple 1-track jumping. For 1-track jump using DIRC, the following is undertaken (DIRC = normal H).

- (a) Acceleration pulse is output. (\$2C for REV or \$28 for FWD).
- (b) With TZC ↓ (or TZC ↑), set DIRC to L. (SENS Pin 18 is "TZC"). As the jump pulse polarity is inverted, deceleration is applied.
- (c) Set DIRC to H for a specific time. Both the tracking servo and sled servo are switched on automatically. As a result, the track jump will be as shown in the flow chart of Fig. 10 and two serial data transfers are saved.

4. \$3X

This command is for switching the Focus search and Sled kick peak value.

DO, D1 Sled, NORMAL feed, high-speed feed

D2, D3 Focus search peak switching

	D6 *D5 D4 D3 (PS3) 0 1 1 1 1 1 1	earch peak	Sled					
D7	D6	•D5	D4	D3	D2	D1	D0	Relative value
				(PS3)	(PS2)	(PSI)	(PS0)	14140
				0	0	0	0	·± 1
0	Λ	1	1	0	1	0	1	± 2
"	U	1	1	1	0	1	0	± 3
				1	1	1	1	± 4

5. \$4X to \$7X

\$4X to \$7X are for the auto sequencer commands. Refer to the table and timing chart for the auto sequencer.

The auto sequencer automates the troublesome routines of focus pull-in and track jump, eliminating any timing control of the microcomputer that is less than 10 ms and combining with the Q register of the CXD1135, CXD1125, CXD1130 and CX23035.

Auto focus

When a focus is pull-in during the \overline{BUSY} shifts $H \to L \to H$, the \$08 is automatically set in the register. Even when it is out of focus, no auto pull-in is done requiring FOK to be monitored.

Track jump

When the \overline{BUSY} shifts $H \to L \to H$ and the track jump is completed, the \$25 is automatically set in the register.

Sequencer malfunctions can be relieved with the \$40 anytime.

Others

1. Connection of the power supply pin

	Vcc	VEE	VC
±5 V dual power supply	+5V	-5V	0V
5 V single power supply	+ 5V	0V	VC*

* CXA1081

2. FSET pin

The FSET pin determines the characteristic of the high frequency phase compensation of Focus, Tracking servo, and cut-off frequency (fc) of CLV LPF.

3. ISET pin

ISET current = 1.27 V/R

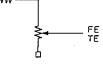
= Focus search current (\$30)

= Tracking kick current

= 1/2 sled kick current (\$30)

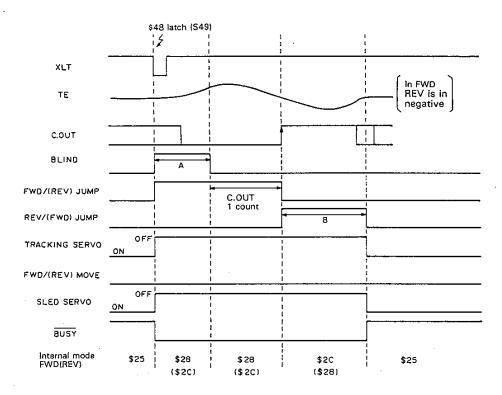
- 4. In the tracking amplifier, input is clamped at 1 VBE to prevent over input.
- 5. How to change the FE and TE gains
 - (1) To increase: Pins 5 and 6, pins 11 and 12 to more than 100 k Ω .
 - (2) To decrease: Divide the FE and TE resistor of input.
- 6. Tracking servo phase

From TE to TAO the phase is negative. (CXA20108 has a positive phase.)

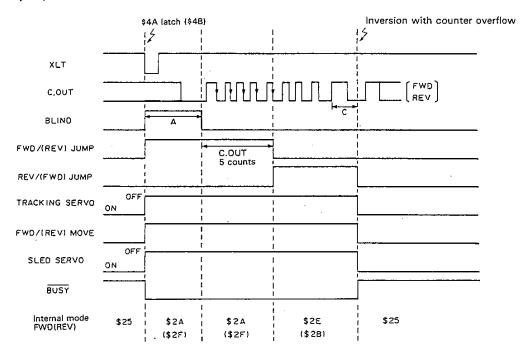


Auto Sequencer Timing Chart

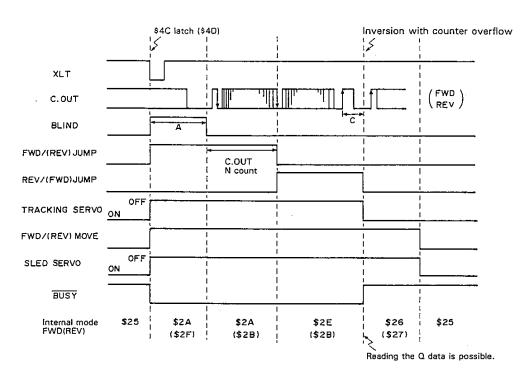
1. 1 track jump



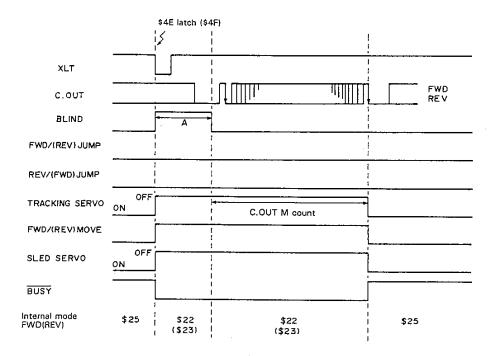
2. 10 track jump



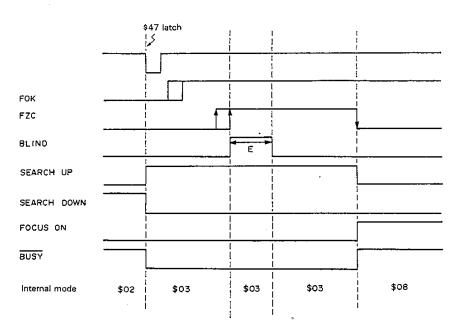
3. 2N track jump



4. M track movement



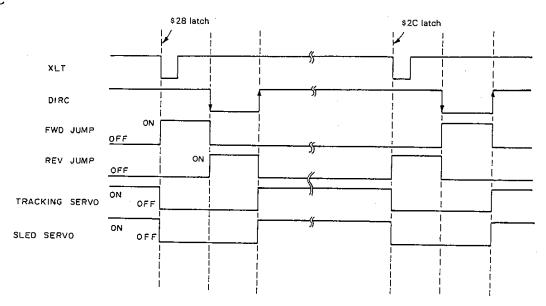
5. Auto focus



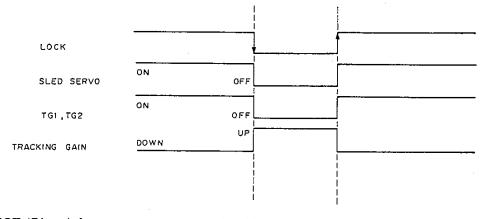
- 6. Notes on use of the auto sequence
 - 1) The horizontal axis of the timing chart is not always proportional to the actual time.
 - Use the auto focus only while the search is down.
 Use the track jump while the focus, tracking, and sled servo are switched on.
 - 3) The auto sequencer does not cover tracking gain up, brake, anti-shock, and focus gain down. Separate commands are required.
 - 4) BUSY does not tell the full status of the player. Monitor FOK and GFS, etc. using the microcomputer.
 - 5) When the sequencer hangs up, detect BUSY's max, excess time using the microcomputer and send \$40 (CANCEL) to reset to the preceding status.
 - 6) In all modes the auto sequence starts from the first WDCK falling edge right after XLT falls.

Parallel Direct Interface

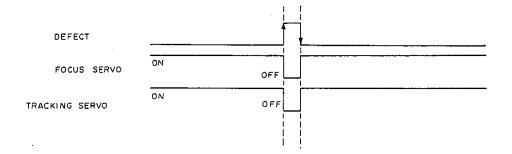
1. DIRC



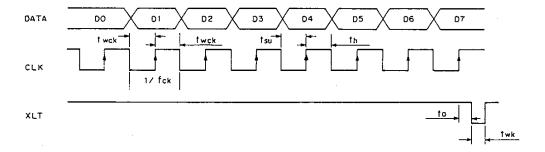
2. LOCK (Sled runaway prevention circuit)



3. DEFECT (Disc defect countermeasure circuit)



CPU Serial Interface Timing Chart



DVcc - DGND = 4.5 to 5.5V

Item	Symbol	Min.	Тур.	Max.	Unit
Clock frequency	fck			1	MHz
Clock pulse width	fwck	500			ns
Hold time	tsu	500			ns
Setup time	th	500	-		ns
Delay time	ŧ _D	500			ns
Latch pulse width	twL	1000			ns

System Control

	Τ,	ADDRESS		DA	TA		SENS	
	Item	D7D6D5D4	D3	D2	D1	D0	Output	
			FS4	FS3	FS2	FS1		
Focus Co	ontrol	0000	Focus	Gain	Search	Search	FZC	
			ОИ	Down	ON	Up		
Trooking	Control	0 0 0 1	Anti	Brake	TG2	TG1	A.S	
Tracking	Tracking Control		Shock	ON	Gain Set *.	1	A.S	
Tracking	Mode	0 0 1 0	Tracking N	lode *2	Sled Mode	TZC		
			PS4	PS3	PS2	PS1		
Select		0 0 1 1	Focus	Focus	Sled	Sled	SSTOP	
			Search+2	Search + 1	Kick+2	Kick+1		
Auto Seq	uence *4	0 1 0 0	AS3	AS2	AS1	AS0	BUSY	
	Blind (A, E)/Overflow (C)	0 1 0 1	0.18ms	0.09ms	0.045ms 0.022ms			
*5 Brake (B)		0 1 0 1	0.36ms	0.18ms	0.09ms	0.045ms		
RAM	Kick (D)	0 1 1 0	11.6ms	5.8ms	2.9ms	1.45ms	Hi-2	
SEI	Track Jump (N)	0 1 1 1	64	32	16	8].	
	Track Move (M)	0 1 1 1	128	64	32	16		

Note)*1. GAIN SET

It is possible to set TG1 and TG2 independently.

When the anti-shock is 1 (00011xxx), invert both TG1 and TG2 when the internal anti-shock is H.

* 2 TRACKING MODE

	D3	D2
OFF	0	0
ON	0	1
FWD JUMP	1	0
REV JUMP	1	1

* 3 SLED MODE

	D1	D0
OFF	0	0
ОИ	0	1
FWD MOVE	1	0
REV MOVE	1	1

* 4 AUTO SEQUENCE

	AS3	AS2	ASI	AS0
CANCEL	0	0	0	0
FOCUS ON	0	1	1	1.
1 TRACK JUMP	1	0	0	X
10 TRACK JUMP	1	0	1	X
2N TRACK JUMP	1	1	0	X
M TRACK MOVE	1	1	I	X

X = 0 FORWARD X = 1 REVERSE

- When CANCEL \$40 is sent, the status immediately preceding the auto sequence mode (just before \$4X is sent) is reset.
- The auto sequence mode starts with the first falling of the pin 40 input pulse (WDCK) after the \$4X transfer and the falling of latch pulse.

*5 RAM·SET

- Values \$1 to \$E (not \$0, \$F) can be set.
- The above set values are ones when WDCK (88.2 kHz) is input to pin 40.
- The RAM is preset when the power is switched on and the internal initial set values are as follows:

ADDRESS	DATA
0 1 0 1	0 1 0 1
0 1 1 0	0 1 1 1
0 1 1 1	1110

• The actual count values are slightly different from the set values.

A set value + 4 to 5 WDCK
B, D, E set value + 3 WDCK
C set value + 5 WDCK
N, M set value + 3 Count out

SONY CXA1082BQ/BS

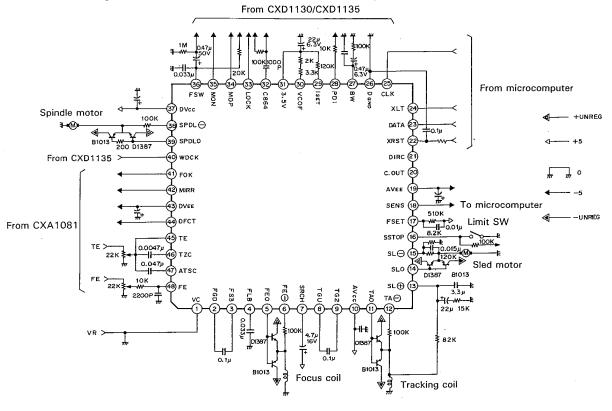
Serial Data Truth Table

Serial data	Hexa.	Function
FOCUS CONTROL		FS = 4321
00000000	\$00	0000
10000000	\$01	0001
00000010	\$02	0 0 1 0
00000011	\$03	0011
00000100	\$04	0100
0.0000101	\$05	0101
00000110	\$06	0 1 1 0
00000111	\$07	0 1 1 1
00001000	\$08	1000
00001001	\$09	1 0 0 1
0 0 0 0 1 0 1 0	\$0A	1010
0 0 0 0 1 0 1 1	\$0B	1011
00001100	\$0C	1100
00001101	\$0D	1 1 0 1
0 0 0 0 1 1 1 0	\$0E	1 1 1 0
00001111	\$0F	1 1 1 1
TRACKING CONTROL		AS= 0 AS= 1
		TG= 2 1 TG= 2 1
00010000	\$10	0 0 0 0
00010001	\$11	0 1 0 1
00010010	\$12	1 0 1 0
00010011	\$13	1 1 1 1
00010100	\$14	0 0 0 0
00010101	\$15	0 1 0 1
00010110	\$16	1 0 1 0
0 0 0 1 0 1 1 1	\$17	1 1 1 1
00011000	\$18	0 0 1 1
00011001	\$19	0 1 1 0
0 0 0 1 1 0 1 0	\$1A	1 0 0 1
0 0 0 1 1 0 1 1	\$1B	1 1 0 0
0 0 0 1 1 1 0 0	\$1C	0 0 1 1
0 0 0 1 1 1 0 1	\$ID	0 1 1 0
0 0 0 1 1 1 1 0	\$1E	1 0 0 1
00011111	\$1F	1 1 0 0

Serial data	Hexa.	Fun	ction	
TRACKING MODE		DIRC=1	DIRC=0	DIRC=1
11010101110111011101110111	_	TM = 654321	654321	654321
00100000	\$20	000000	001000	000011
00100001	\$21	000010	001010	000011
00100010	\$22	010000	011000	100001
00100011	\$23	100000	101000	100001
00100100	\$24	000001	000100	000011
00100101	\$25	000011	000110	000011
00100110	\$26	010001	010100	100001
00100111	\$27	100001	100100	100001
00101000	\$28	001000	001000	000011
00101001	\$29	000110	001010	000011
00101010	\$2A	010100	011000	100001
00101011	\$2B	100100	101000	100001
00101100	\$2C	001000	000100	110000
00101101	\$2D	001010	000110	000011
00101110	\$2E	011000	010100	100001
00101111	\$2F	101000	100100	100001

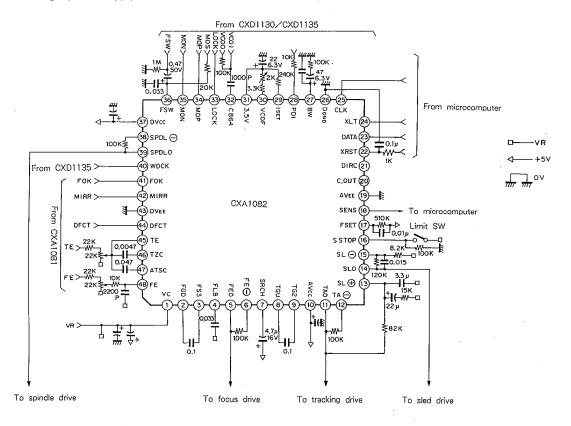
Application Circuit

1. ± 5 V dual power supply for CXA1082BQ (48 pin QFP) (See the Pin Configuration for CXA1082BS)



Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.

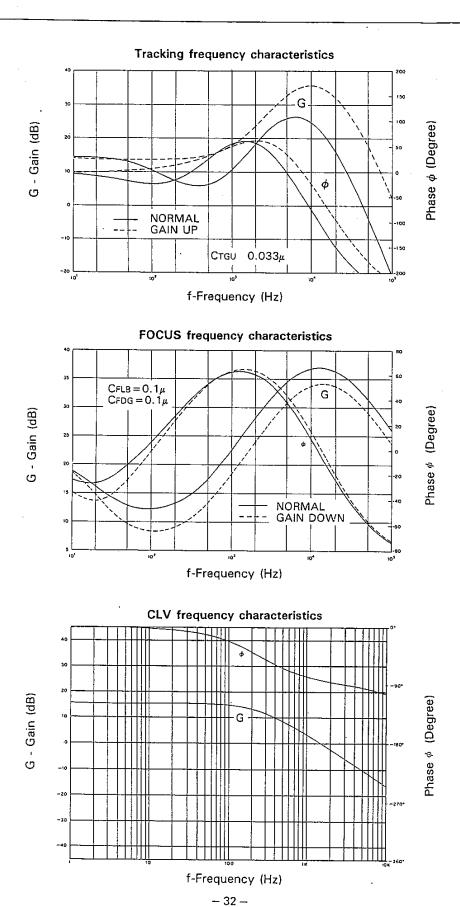
2. \pm 5 V single power supply for CXA1082BQ (See the Pin Configuration for CXA1082BS)



Application circuits shown are typical examples illustrating the operation of the devices. Sony cannot assume responsibility for any problems arising out of the use of these circuits or for any infringement of third party patent and other right due to same.

Internal Phase Compensation Standard Circuit Design Data

						SW 6	condi	itions	6				Bia	s co	ndition	ıs	2 5	모	Output waveform and description	N4:	T		Unit
Mode	Item	Symbol	SI	S2	S3	S4	S5	S6	S7	S8	S9	SD					Ë	Test	of test methods	Min.	Тур.	Max.	Onit
	1. 2 kHz gain		0	0								08				1	48	5			21, 5		dB
Focus	1. 2 kHz phase		0	0								08									. 63		deg
Sus	1. 2 kHz gain		0	0								0C					П				16		dB
	1. 2 kHz phase		0	0								00.					П				63		deg
	1. 2 kHz gain					0						25					45	П			13		dB
Tracking	1. 2 kHz phase					0						25					П				-125		deg
king	2. 7 kHz gain					0						25									26, 5		dB
	2. 7 kHz phase					0						25					Ţ	\Box			~130		deg
"	100 Hz phase																34	39			-30		deg
Spindle	2 kHz gain																П				-3, 5		dB
ē																		\Box					

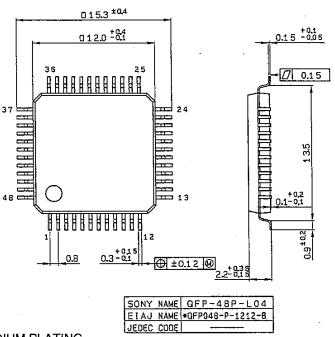


Package Outline

Unit: mm

CXA1082BQ

48pin QFP (Plastic) 0.7g

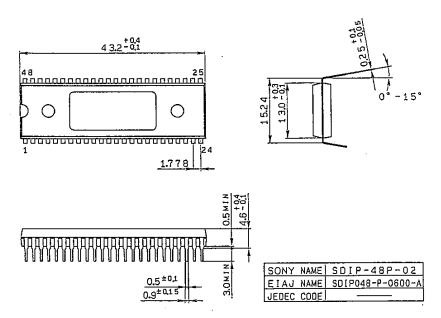


NOTE: PALLADIUM PLATING

This product uses S-PdPPF (Sony Spec.-Palladium Pre-Plated Lead Frame).

CXA1082BS

48pin SDIP (Plastic) 600mil 5.1g



NOTE: PALLADIUM PLATING

This product uses S-PdPPF (Sony Spec.-Palladium Pre-Plated Lead Frame).