

LA70011, 70011M

Recording/Playback Amplifier for VHS VCRs

Overview

The LA70011 and LA70011M are recording/playback amplifiers for VHS VCR video signals. When used in combination with the LA71000M and LA71500M Series of video signal processing ICs, they permit Y/C recording without current adjustment.

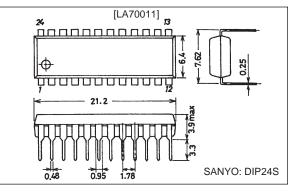
Features

- Connecting the playback amplifier input directly to the head reduces the number of external elements required.
- The recording amplifiers use a fixed-current drive configuration that yields stable recording characteristics even under changing loads. They include built-in automatic gain control circuits.
- Using the same dimensions and pin assignments as the LA70001 and LA70001M permits the use of the same circuit boards as these earlier chips. The LA70011 can also be mounted at the right end of an LA70020 socket.

Package Dimensions

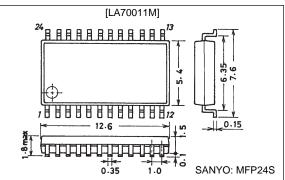
unit: mm

3067-DIP24S



unit: mm

3112-MFP24S



Specifications

Maximum Ratings at $Ta = 25^{\circ}C$

| Parameter | Symbol | Conditions | Ratings | Unit |
|------------------------------|--------------|--|-------------|------|
| Maximum power supply voltage | V_{CC} max | | 7.0 | V |
| Maximum power dissipation | Pd max | Ta ≤ 65°C [LA70011] | 600 | mW |
| | | Ta \leq 65°C [LA70011M] 114.3 \times 76.1 \times 1.6 mm: glass epoxy | 500 | mW |
| Operating temperature | Topr | | -10 to +65 | °C |
| Storage temperature | Tstg | | -40 to +150 | °C |

Operating Conditions at $Ta = 25^{\circ}C$

| Parameter | Symbol | Conditions | Ratings | Unit |
|--------------------------------------|--------------------|------------|------------|------|
| Recommended power supply voltage | V _{CC} | | 5.0 | V |
| Operating power supply voltage range | V _{CC} op | | 4.8 to 5.5 | V |

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Operating Characteristics at Ta = $25^{\circ}C$

| Parameter | | Symbol | Conditions | | Ratings | | Unit |
|--|--------------------------|--|---|----------|---------|------|------|
| Dieu de cele Manda | | | | min | typ | max | |
| Playback Mode | | | | | | | |
| Current drain SP-L CH1 | | ICCP | Current flowing into pin 13 | 44 | 53 | 60 | mA |
| | | G _{VP} 1 | - | 56 | 59 | 62 | dB |
| Voltage gain | SP-H CH2 | G _{VP} 2 | V _{IN} = 38 mVp-p, f = 1 MHz | 56 | 59 | 62 | dB |
| | EP-L CH3 | G _{VP} 3 | - | 56 | 59 | 62 | dB |
| | EP-H CH4 | G _{VP} 4 | | 56 | 59 | 62 | dB |
| Voltage gain difference | | ∆G _{VP} 1 | $G_{VP}1 - G_{VP}2$ | -1 | 0 | +1 | dB |
| laternada acia difference | | ∆G _{VP} 2 | $G_{VP3} - G_{VP4}$ | -1 -1 | 0 | +1 | dB |
| Intermode gain difference | CH1 | ∆G _{VP} 3 | G _{VP} 3 — G _{VP} 1 | -1 | 0 | +1 | dB |
| Converted input noise voltage | CH2 CH3 CH4 | V _{NIN1} V _{NIN2} V _{NIN3} V _{NIN4} | Ratio of the output from a 1.1 MHz low pass filter to the output with no input under the same conditions as those used for measuring voltage gain. | | 1.0 | 1.5 | µVrm |
| Frequency characteristic CH3 CH3 CH4 | | $\begin{array}{c} \Delta V_{fp} 1 \\ \Delta V_{fp2} \\ \Delta V_{fp3} \\ \Delta V_{fp4} \end{array}$ | Ratios of the output for V _{IN} = 38 mVp-p and f = 7 MHz to the voltage gains G_{VP} 1, G_{VP} 2, G_{VP} 3, and G_{VP} 4. | -2.5 | 0 | | dB |
| Secondary harmonic distortion | CH1 CH2 CH3 CH4 | $\Delta V_{HDP} 1$ ΔV_{HDP2} ΔV_{HDP3} ΔV_{HDP4} | Ratio of the 8 MHz (secondary) component of the output to its 4 MHz (primary) component for V_{IN} = 38 mVp-p and f = 4 MHz. | | -40 | -35 | dB |
| Maximum output level | CH1 CH2 CH3 CH4 | ΔV _{OMP} 1 ΔV _{OMP2} ΔV _{OMP3} ΔV _{OMP4} | Output level, for f = 1 MHz, at which the ratio of the 3 MHz (tertiary) component to the 1 MHz (primary) component is -30 dB. | 1.0 | 1.2 | | Vp-p |
| Crosstalk SP | | V _{CR} 1 | Ratio of the output for V _{IN} = 38 mVp-p and f = 4 MHz to G_{VP} 1. | | -40 | -35 | dB |
| | | V _{CR} 2 | Ratio of the output for V_{IN} = 38 mVp-p and f = 4 MHz to G_{VP} 2. | | -40 | -35 | dB |
| | | V _{CR} 3 | Ratio of the output for V_{IN} = 38 mVp-p and f = 4 MHz to G _{VP} 3. | | -40 | -35 | dB |
| Crosstalk EP | | V _{CR} 4 | Ratio of the output for V _{IN} = 38 mVp-p and $f = 4$ MHz to G _{VP} 4. | | -40 | -35 | dB |
| | | ΔV_{ODC} 1 | CH1 — CH2 | | | | |
| | | $\Delta V_{ODC} 2$ | CH3 — CH4 | | | | |
| | | ΔV _{ODC} 3 | CH1 — CH3 | 100 | 0 | .100 | mV |
| Output DC offset | | $\Delta V_{ODC}4$ | CH2 — CH4 | -100 | 0 | +100 | |
| | | $\Delta V_{ODC} 5$ | CH1 — CH4 | | | | |
| | | $\Delta V_{ODC} 6$ | CH2 — CH3 | | | | |
| Envelope detector output pin vol | tage | V _{ENV} | T6 DC level with no signal input. | 0.0 | 08 | 1.3 | V |
| | | V _{ENVSP} 1 | T6 DC level at which T7A output level is 175 mVp-p for f = 4 MHz. | 2.0 | 2.5 | 3.0 | v |
| Envelope detector output pin vol | tage SP | V _{ENVSP} 2 | T6 DC level at which T7A output level is 400 mVp-p for $f = 4$ MHz. | 4.0 | 4.5 | 5.0 | v |
| | to go ED | V _{ENVEP} 1 | T6 DC level at which T7A output level is 125 mVp-p for $f = 4$ MHz. | 2.0 | 2.5 | 3.0 | V |
| Envelope detector output pin vol | | V _{ENVEP} 2 | T6 DC level at which T7A output level is 300 mVp-p for f = 4 MHz. | 4.0 | 4.5 | 5.0 | V |
| | | V _{COMP} 1 | T2 DC level for V_{IN} = 38 mVp-p and f = 4 MHz. | | 0.4 | 0.7 | V |
| Comparator output voltage | | V _{COMP} 2 | T2 DC level for V_{IN} = 38 mVp-p and f = 4 MHz. | 4.5 | 4.8 | | V |
| SW-Tr on resistance during play | back | R _{PON} 17 R _{PON} 22 | DC difference for 1 and 2 mA current inputs. | | 4 | 6 | Ω |
| | | TR1-1 | Normal \rightarrow Trick1 : *1 | 3.2 | | 5.0 | V |
| Trick threshold lovel | | TR1-2 | $Trick1 \rightarrow Normal$ | 1.2 | | 2.8 | V |
| Trick threshold level | | TR2-1 | Normal \rightarrow Trick2 : *1 | 0.0 | | 0.8 | V |
| | | TR2-2 | $Trick2 \rightarrow Normal$ | 1.2 | | 2.8 | V |

| Parameter | Symbol | Conditions | Ratings | | | Unit |
|---|--|--|---------|------|-----|-------|
| Parameter | Symbol | Conditions | min | typ | max | |
| | HAP-1 | $SP \rightarrow EP$: *1 | 1.7 | | 5.0 | V |
| HA playback threshold level | HAP-2 | EPSP | 0.0 | | 1.3 | V |
| | SW30-1 | $Lch \rightarrow Hch$: *1 | 1.2 | | 5.0 | V |
| SW30 threshold level | SW30-2 | $Hch\toLch$ | 0.0 | | 0.8 | V |
| Recording Mode | | · · · · · | | | | |
| Current drain | ICCR | Current input at pin 13. | 52 | 59 | 66 | mA |
| REC AGC AMP output level | V _{RSP} | Output level for V_{IN} = 400 mVp-p and f = 4 MHz. | 127 | 135 | 143 | mVp-p |
| | V _{REP} | | 104 | 111 | 119 | mVp-p |
| Intermode gain difference | ∆GVR | VRSP/VREP | 1.4 | 1.7 | 2.0 | dB |
| | ΔV_{AGC} 1-SP ΔV_{AGC} 1-EP | Output level divided by V_{RSP} or V_{REP} for $f = 4$ MHz and $V_{IN} = 700$ mVp-p. | | 0.5 | 1.0 | dB |
| REC AGC AMP control characteristic | ΔV_{AGC} 2-SP ΔV_{AGC} 2-EP | Output level divided by V_{RSP} or V_{REP} for $f = 4$ MHz and $V_{IN} = 100$ mVp-p. | -1.0 | -0.5 | | dB |
| REC AGC AMP frequency characteristic | $\Delta V_{FRS} \\ \Delta V_{FRE}$ | Ratio of f = 7 MHz output to f = 1 MHz output for V_{IN} = 400 mVp-p. *2 | -1 | 0 | +1 | dB |
| REC AGC AMP secondary primary distortion | ΔV_{HDRS} ΔV_{HDRE} | Ratio of the 8 MHz (secondary) component of the output to its 4 MHz (primary) component for V_{IN} = 400 mVp-p and f = 4 MHz. | | -45 | -40 | dB |
| REC AGC AMP maximum output level | $\Delta V_{MOSP} \\ \Delta V_{MOEP}$ | Output level, for $f = 4$ MHz, at which the secondary distortion is -35 dB. | 20 | 22 | | mApp |
| REC AGC AMP muting attenuation | ΔV_{MRS} ΔV_{MRE} | Output level divided by V_{RSP} or V_{REP} for $f = 4$ MHz and $V_{IN} = 400$ mVp-p. | | -45 | -40 | dB |
| REC AGC AMP cross modulation relative level | $\Delta V_{CYS} \Delta V_{CYE}$ | Output ratio (4M \pm 629k)/4M for V _{IN} = 400 mVp-p and f = 4 MHz at T9A and V _{IN} = 2.4 Vp-p and f = 629 kHz at T10A. | | -45 | -40 | dB |
| HA REC threshold level | H _{AR} -1 | $SP \rightarrow EP:*1$ | 1.7 | | 5.0 | V |
| | H _{AR} -2 | $EP\toSP$ | 0.0 | | 1.3 | V |
| REC MUTE threshold level | MUTE-1 | MUTE OFF \rightarrow MUTE ON *1 | 1.2 | | 2.8 | V |
| | MUTE-2 | $MUTE\;ON\toMUTE\;OFF$ | 3.2 | | 5.0 | V |
| REC PB threshold level | PB-REC | $PB \rightarrow REC *1$ | 1.2 | | 5.0 | V |
| | REC-PB | $REC \to PB$ | 0.0 | | 0.8 | V |

Notes:* Before measuring the items under Playback Mode, input a 0 to 5.0 V trigger pulse to T5 (H-SYNC), the pin from which the LA70011 takes its T3 (HA) control switch timing.
* The resistance between pins 13 and 14 must be accurate to within 1.0%.
*1. These are voltage application points.
*2. Apply a DC voltage of approximately 1.8 V to the AGC wave detector filter pin (pin 15) to fix the AGC amplifier gain.
*3. Apply a DC voltage to the REC-CUR-Adj pin (pin 12) and adjust the output level.

Pin Descriptions

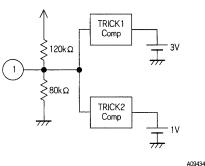
| Pin Number | Pin Name | Stand | dard DC Voltage (V) | Equivalent Circuit | Notes |
|------------|------------|-------|--|---|--|
| 1 | TRICK-H | | | VCC 120kΩ 120kΩ Trick1 3V Comp Trick2 1V J J A0941B | Trick1 3.0 V NORMAL 1.0 V Trick2 |
| 2 | COMP-OUT | PB | H: min. 4.5 V L: max. 0.7 V Open | | EP > SP ENV High |
| 3 | HA (EP/SP) | | | 3 1kΩ HA Comp 1.5V 100kΩ 777 Λ09420 | EP 1.0 V SP |
| 4 | SW30 | | | 4 1kΩ 50kΩ 50kΩ 1V 1V 1V 777 A09421 | Hch Lch |
| 5 | H-SYNC | | | 5 20kΩ 80kΩ 777 777 409422 | SYNC H L |

| Pin Number | Pin Name | Stan | dard DC Voltage (V) | Equivalent Circuit | Notes |
|--------------|-------------|-------------------------|-----------------------------|--|-------------------|
| 6 ENVDET-OUT | PB | See relevant documents. | | | |
| | REC | 0 | 6 20kΩ 7777 A09423 | | |
| 7 | | РВ | 1.7 | 100Ω ↓ | |
| | PB-OUT REC | REC | 2.1 | (7) → ↓ 1mA → ↓ 1mA → ↓ A09424 | |
| 8 20 | GND | | | | |
| 9 | | PB REC-Y-IN | 4.0 | 300Ω 300Ω 5kΩ | |
| | | REC | 3.7 | | |
| 10 | REC-C-IN | PB | 4.0 | 10 25kΩ 5kΩ 5kΩ 77 Α09426 | |
| | | REC | 3.7 | | |
| 11 | REC/MUTE/PB | | | 20k Ω | REC MUTE 9B |

| Pin Number | Pin Name | Stan | dard DC Voltage (V) | Equivalent Circuit | Notes |
|-------------------------|--------------------|-------|---------------------|---|-------|
| 12 REC-CURRENT- ADJ2 | РВ | 2.5 V | | | |
| | REC | 2.5 V | 100kΩ | | |
| 13 | V _{CC} | | | | |
| 14 | REC-CURRENT- | PB | 5.0 | | |
| | 14 ADJ1 | REC | 4.5 | | |
| 15 | 15 REC-AGC-FILT | PB | 0 | 15-W- 1 | |
| | | REC | 1.6 | 10kΩ | |
| 16 19 | SP L-IN SP H-IN | PB | 2.1 | REC-ON VCC | |
| 21 24 | EP L-IN EP H-IN | REC | 4.1 | (2)(2) → PB-ON → 2.4mA 777 A09431 | |
| 17 REC | REC SP OUT | | | | |
| 22 | 22 EP OUT | REC | 4.1 | γ< | |

| Pin Number | Pin Name | Stand | dard DC Voltage (V) | Equivalent Circuit | Notes |
|------------|----------|-------|---------------------|-----------------------------|-------|
| 18 | | РВ | 0 | ⁽¹⁸ €3) ≥20kΩ | |
| 23 | PB FILT | REC | 2.5 | PB-ON 20k Ω Λ09433 | |

Usage Notes Control Pin Logic Switching Trick Mode with Pin 1

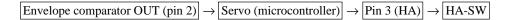


GND < pin 1 level - DC < 1.0 V: TRICK2 1.0 V < pin 1 level - DC < 3.0 V: NORMAL 3.0 V < pin 1 level - DC < 5.0 V: TRICK1

NORMAL Mode Two channels selected with pin 3 (EP/SP): ON Envelope comparator: OFF

TRICK1 and TRICK2 Modes All four channels: ON Envelope comparator: OFF

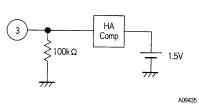
Difference between TRICK1 and TRICK2 modes (See the Block Diagram.) TRICK1 is a special playback mode using the following path



TRICK2 provides SP searching

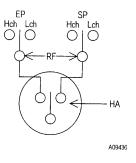
Envelope comparator OUT \rightarrow HA-SW

HA-SW (EP/SP mode switch): Pin 3



GND < pin 3 level - DC < 1.5 V: SP mode 1.5 V < pin 3 level - DC < 5 V: EP mode Synchronization of HA Switching Timing during Playback with H-SYNC Signal

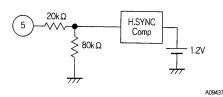
During playback, the LA70011's video circuits synchronize the HA-SW switching timing shown in the following figure with the H-SYNC signal from pin 5. (Other EP/SP switching takes place in real time.)



Comparator Output: Pin 2 EP envelope > SP envelope: High (min. 4.0 V)

EP envelope $\langle SP$ envelope: Low (max. 0.7 V)

H-SYNC Input: Pin 5



Pin 5 level - DC > 1.5 V: H-SYNC interval

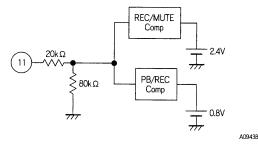
Playback:

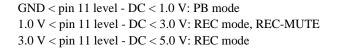
- Determines timing of HA switching (EP/SP)
- Determines timing of special playback

Recording:

- Serves as gate pulse for REC-AGC-AMP SYNC unit

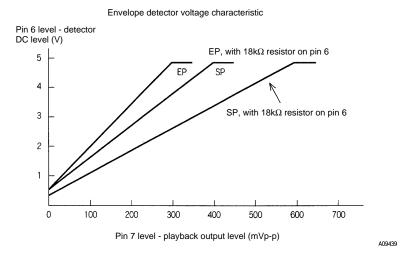






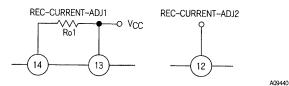
Envelope Detector Characteristic: Pin 6

The LA70011 includes a built-in playback signal envelope detector circuit for use in automating tracking adjustment.



REC AMP Gain Control

The LA70011 eliminates recording current adjustment by adding an automatic gain control circuit to the recording amplifier. It is also possible to change the recording current with the following methods.



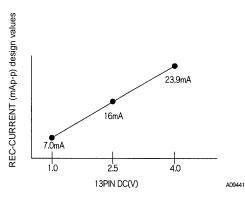
REC-CURRENT-ADJ2 Open

The internal bias forces the DC level at pin 12 to $1/2 V_{CC}$ (that is, approximately 2.5 V), and R_01 determines the recording current.

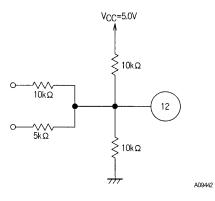
Design values $R_O 1 = 1.5 \text{ k}\Omega = 16.0 \text{ mA (SP) (per channel)}$ $R_O 1 = 1.5 \text{ k}\Omega = 12.7 \text{ mA (EP)}$

REC-CURRENT-ADJ2 Used

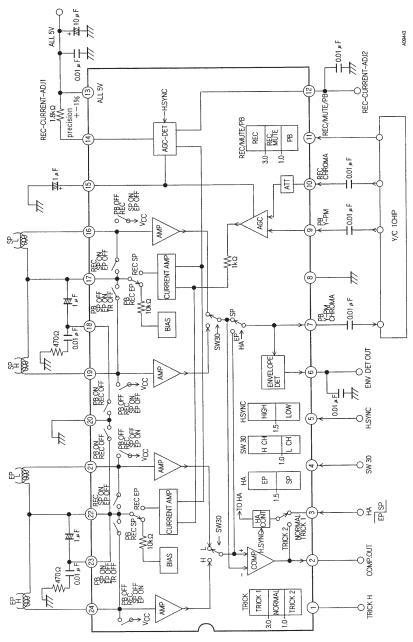
Applying a DC control voltage between 1 and 4 V to pin 12 adjusts the figure determined by R_01 between -6.0 dB and +3.5 dB.



Note: One possible circuit for applying this voltage is the following, which provides 9 modes between 1 and 4 V.



Block Diagram



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