

## **OVERVIEW**

The SM9103M is a photodiode photoelectric current-to-voltage conversion head amplifier LSI for optical disk pickups in DVDRAM/DVDROM equipment. It sums the photodiode current data signals and then converts the signals to a differential signal for output. The output tracking servo and focusing servo signals are derived from built-in sum and difference circuits, and the gain for these servo signals can be adjusted using serial interface controls. Each of the signals from the photodiodes, used to generate DPD (Differential Phase Detection) tracking servo signal, is current-to-voltage converted and then also output. It operates from a single 5 V supply, and is available in 36-pin plastic SSOP packages.

## **FEATURES**

- RAM/ROM gain switching, low-noise RF signal generator (differential output)
- ROM tracking DPD signal output
- Variable-gain RAM tracking push-pull signal output
- Address signal, high-speed push-pull signal output
- Variable-gain focus error signal output
- Tracking PD sum signal output
- Focus PD sum signal output
- Offset correction timing output (logic)
- Temperature monitor function
- Serial interface to control internal parameter settings
- Sleep-mode function
- Single 5 V supply
- 36-pin plastic SSOP

### **TYPICAL APPLICATIONS**

- Double-speed DVDROM equipment
- Double-speed DVDRAM equipment

### **ORDERING INFORMATION**

Device	Package
SM9103M	36-pin SSOP

## PINOUT

#### 36-pin SSOP

#### (Top view)



## PACKAGE DIMENSIONS

(Unit: mm)



## **BLOCK DIAGRAM**



# **PIN DESCRIPTION**

Number	Name	I/O <sup>1</sup>	Function
1	MODE	Ipd	Mode switching/offset correction control input 1
2	WRITE	Ipd	Mode switching/offset correction control input 2
3	DGND	-	Logic circuit ground. Connect to the analog ground if there is no dedicated pickup or logic ground.
4	DVCC	-	Logic circuit supply. Connect to the analog supply if there is no dedicated pickup or logic supply.
5	TEMPO	0	Thermal sensor test output. Leave open for normal operation
6	TEMPI	I	Thermal sensor test input. Leave open for normal operation
7	T1	I	Tracking PD input A
8	T2	I	Tracking PD input B
9	T3	I	Tracking PD input C
10	T4	I	Tracking PD input D
11	F1	I	Focus PD input E
12	F2	I	Focus PD input F
13	AGND	-	Analog circuit ground
14	VREF	I	2.0 V reference voltage input
15	FSUBB	I	Focus error signal feedback input
16	FSUB	0	Focus error signal output
17	FADDB	I	Focus sum signal feedback input
18	FADD	0	Focus sum signal output
19	AGND	-	Analog circuit ground
20	AVCC	-	Analog circuit supply
21	DPDD	0	Buffered tracking signal output D for DPD servo
22	DPDC	0	Buffered tracking signal output C for DPD servo
23	DPDB	0	Buffered tracking signal output B for DPD servo
24	DPDA	0	Buffered tracking signal output A for DPD servo
25	DATAN	0	Phase-modulated data signal differential inverting output
26	DATAP	0	Phase-modulated data signal differential non-inverting output
27	TSUBB	I	Tracking push-pull signal feedback input
28	TSUB	0	Tracking push-pull signal output
29	CAPAN	0	ID data signal differential inverting output
30	CAPAP	0	ID data signal differential non-inverting output
31	TADDB	I	Tracking PD sum signal feedback input
32	TADD	0	Tracking PD sum signal output
33	CALREQ	0	Offset correction status/request output
34	SENB	I	Serial interface enable input
35	SDATA	I/O	Serial interface data input/acknowledge output
36	SCLK	I	Serial interface clock input

1. I = input, Ipd = Input with built-in pull-down resistor, I/O = input/output, O = output

# SPECIFICATIONS

# Absolute Maximum Ratings

GND = 0 V

Parameter	Symbol	Condition	Rating	Unit
Supply voltage range	V <sub>CC</sub>		-0.5 to 7.0	V
Input voltage range	V <sub>IN</sub>		-0.5 to V <sub>CC</sub> + 0.5	V
Input current range	I <sub>IN</sub>		- 3.0 to +3.0	mA
Operating temperature range	T <sub>opr</sub>		0 to 70	°C
Storage temperature range	T <sub>stg</sub>		-40 to 125	°C
Power dissipation	PD		250	mW
Soldering temperature	T <sub>sld</sub>		260	°C
Soldering time	t <sub>sld</sub>		10	S

# **Recommended Operating Conditions**

GND = 0 V

Parameter	Symbol	Condition	Rating	Unit
Specs-guaranteed supply voltage range	V <sub>CC</sub>		4.75 to 5.25	V
Operating supply voltage range	V <sub>CC</sub>		4.5 to 5.5	V
Reference voltage input	V <sub>REF</sub>		1.89 to 2.11	V
Operating temperature range	T <sub>opr</sub>		0 to 70	°C

## **DC Electrical Characteristics**

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70  $^{\circ}C$ 

Daramotor	Symbol	Condition		Rating			
Parameter	Symbol	Condition	min	typ	max	Unit	
Current concumption <sup>1</sup>	I <sub>CC1</sub>	Operating mode	-	24	30	m۸	
	I <sub>CC2</sub>	Sleep mode	-	-	1	IIIA	
MODE, WRITE, SENB, SDATA, SCLK HIGH-level input voltage	V <sub>IH</sub>		0.8V <sub>CC</sub>	-	-	V	
MODE, WRITE, SENB, SDATA, SCLK LOW-level input voltage	V <sub>IL</sub>		-	-	0.2V <sub>CC</sub>	V	
MODE, WRITE HIGH-level input current	I <sub>IH1</sub>	$V_{IN} = V_{CC}$	50	100	200	μA	
SENB, SDATA, SCLK HIGH-level input current	I <sub>IH2</sub>	$V_{IN} = V_{CC}$	-	-	3	μA	
MODE, WRITE, SENB, SDATA, SCLK LOW-level input current	Ι <sub>ΙL</sub>	V <sub>IN</sub> = 0 V	-3	-	-	μA	
CALREQ HIGH-level output voltage	V <sub>OH</sub>	I <sub>OH</sub> = -0.2 mA	$V_{CC} - 0.2$	-	-	V	
CALREQ LOW-level output voltage	V <sub>OL1</sub>	I <sub>OL</sub> = 0.8 mA	-	-	0.4	V	
SDATA LOW-level output voltage	V <sub>OL2</sub>	I <sub>OL</sub> = 7 mA	-	-	1.0	V	
VREF input current	I <sub>REF</sub>	V <sub>REF</sub> = 2.0 V	-	-	250	μA	

1. 18 k  $\!\Omega$  resistor connected between TSUB and TSUBB

47 k $\!\Omega$  resistor connected between TADD and TADDB

 $22 \text{ k}\Omega$  resistor connected between FSUB and FSUBB

27 kΩ resistor connected between FADD and FADDB SENB, SDATA, SCLK connected to GND; All other pins (excluding supply and ground pins) open circuit.

# Tracking PD Input Characteristics (T1, T2, T3, T4)

 $V_{CC} = 5 V \pm 5\%$ , GND = 0 V,  $T_a = 0$  to 70 °C

Daramatar	Condition			Unit		
Falanielei			min	typ	max	Unit
Input impedance	No signal		-	-	250	Ω
Input conversion noise current	100 kHz to 10 MHz	RAM read <sup>1</sup>	-	0.035	-	
		ROM read <sup>1</sup>	-	0.27	-	μΑ <sub>rms</sub>
Pin voltage	No signal		-	-	1.5	V

1. DATAP - DATAN output difference operation when 10 pF capacitors are connected to T1, T2, T3, T4

## **Data Signal Processor Characteristics**

 $V_{CC}$  = 5 V ± 5%, GND = 0 V,  $T_a$  = 0 to 70 °C

Daramatar	Condition		Unit		
Palameter			typ	max	Unit
DATAP–DATAN current-to-voltage	RAM read	10.0	12.5	15.0	kO
converter coefficient <sup>1</sup>	ROM read	2.50	3.12	3.74	K22
CAPAP–CAPAN current-to-voltage converter coefficient <sup>2</sup>	RAM read	11.3	14.1	16.9	kΩ
DATAP, DATAN, CAPAP, CAPAN output impedance		-	-	100	Ω
DATAP, DATAN, CAPAP, CAPAN output center voltage <sup>3</sup>	No signal	0.9V <sub>REF</sub>	-	1.1V <sub>REF</sub>	V
CAPAP, CAPAN output center voltage difference <sup>3</sup>	No signal	-	-	±50	mV
DATAP, DATAN, CAPAP, CAPAN output operating output voltage	10 k $\Omega$ load, output center voltage reference	-0.7	-	+0.7	V
Variable coefficient switching time	$RAM \leftrightarrow ROM \text{ read}$	-	-	10	ms
Saturation output reset time <sup>4</sup>	RAM write $\rightarrow$ RAM read	-	-	500	ns
DATAP, DATAN signal bandwidth <sup>5</sup>	f = 100 kHz - 3 dB frequency	19	-	-	MHz
CAPAP, CAPAN signal bandwidth <sup>5</sup>	f = 100 kHz - 3 dB frequency	20	-	-	MHz
DATAP–DATAN, CAPAP–CAPAN gain peaking <sup>5</sup>	f = 100 kHz –3 dB frequency	-3	-	+0.5	dB
DATAP-DATAN, CAPAP-CAPAN group delay time <sup>5</sup>	f = 1 to 10 MHz	-	-	±1.0	ns

1.  $[DATAP - DATAN] = K \times [I_{T1} + I_{T2} + I_{T3} + I_{T4}]$ 2.  $[CAPAP - CAPAN] = K \times \{[I_{T1} + I_{T2}] - [I_{T3} + I_{T4}]\}$ 3. 5 k\Omega load connected to ground to prevent abnormal operation

4. Converging to within final value  $\pm$  10%

5. 10 pF input load capacitors connected to T1, T2, T3, T4. DATAP, DATAN, CAPAP, CAPAN output load conditions shown below.



# **Tracking Signal Processor Characteristics**

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70  $^{\circ}C$ 

Parameter Condition		ondition	Rating			Unit
Falancici			min	typ	max	Unit
	RAM read		10.64	11.95	13.26	
TSUB current-to-voltage converter coefficient <sup>1</sup>	ROM read	$R_f = 18 k\Omega$ , Vour = Vprr + 0.8 V	2.67	2.99	9.92	kΩ
	RAM write		1.78	1.99	2.20	
	RAM read		27.82	31.25	34.68	
TADD current-to-voltage converter coefficient <sup>2</sup>	ROM read	$R_{f} = 47 \text{ k}\Omega$	6.95	7.80	8.65	kΩ
	RAM write		4.63	5.20	5.77	
DPDA, DPDB, DPDC, DPDD	RAM read		40.0	50.0	60.0	ko
current-to-voltage converter coefficient <sup>3</sup>	ROM read		10.0	12.5	15.0	K22
T1, T2, T3, T4 converter coefficient relative error	TSUB output, RAI	M/ROM read	-	-	±2	%
TSUB, TADD, DPDA, DPDB, DPDC, DPDD output impedance			-	-	100	Ω
TSUB operating output voltage	10 kΩ load conne	cted to VREF	1	-	3	V
TADD, DPDA, DPDB, DPDC, DPDD operating output voltage	10 kΩ load conne	cted to VREF	V <sub>REF</sub>	-	3	V
Converter coefficient quitching time	$RAM \text{ read} \leftrightarrow ROM \text{ read}$		-	-	10	ms
	RAM write $\leftrightarrow$ RAI	M read	-	-	3	μs
TSUB, TADD signal bandwidth <sup>4</sup>	DC to -3 dB frequency		1	-	-	MHz
DPDA, DPDB, DPDC, DPDD signal bandwidth <sup>4</sup>	f = 100 kHz to -3 dB frequency		5	-	-	MHz
TSUB, TADD gain peaking <sup>4</sup>	f = 10 kHz to3 dB frequency		-3	-	+0.5	dB
DPDA, DPDB, DPDC, DPDD gain peaking <sup>4</sup>	f = 100 kHz to -3 dB frequency		-3	-	+4.0	dB
TSUB phase response <sup>4</sup>	@ f = 100 kHz		-	-	10	0
DPDA, DPDB, DPDC, DPDD group	f = 1 to 5 MHz grou absolute value	up delay differential	-	-	5	ns
bandwidth <sup>4</sup> TSUB, TADD gain peaking <sup>4</sup> DPDA, DPDB, DPDC, DPDD gain peaking <sup>4</sup> TSUB phase response <sup>4</sup> DPDA, DPDB, DPDC, DPDD group delay <sup>4</sup>	Relative error betw	veen 4 pins	-	-	1.0	
		RAM read/write max gain	-	-	±10.0	
	No input signal, V <sub>REF</sub> reference,	RAM read, min to max gain	-	-	±26	
I SUB onset voltage	post-correction, $T_a = 25$ °C, $R_f = 18 k\Omega$	RAM read/write differential gain max.	-	-	±4	mv
		ROM read, gain min/max	-	-	±100	
	No input signal,	RAM read	-	-	±30	m\/
TADD OTSEL VOLAGE	V <sub>REF</sub> reference	ROM read	-	-	±300	
DPDA, DPDB, DPDC, DPDD offset voltage	No input signal, V <sub>REF</sub> reference	RAM/ROM read	-550	-	+50	mV
TSUB offset voltage temperature coefficient	R <sub>f</sub> = 18 kΩ		-	-	±0.4	mV/°C
TSUB variable gain range			-16	-	+14	dB
TSUB variable gain step width			_	2	-	dB

Deremeter	0	Condition		Rating			
Falanielei				typ	max	Unit	
TSUB gain switching absolute accuracy	V <sub>OUT</sub> = V <sub>REF</sub> ±	-16 to +8 dB	-	-	±0.5	dP	
	0.8 V	+10 to +14 dB	-	-	±1.0	- ab	

1. TSUB = K × { $[I_{T1} + I_{T2}] - [I_{T3} + I_{T4}]$ }, gain = 0 dB 2. TADD = K × { $[I_{T1} + I_{T2} + I_{T3} + I_{T4}]$ 3. DPDA = K ×  $I_{T1}$ , DPDB = K ×  $I_{T2}$ , DPDC = K ×  $I_{T3}$ , DPDD = K ×  $I_{T4}$ 4. T1, T2, T3, T4: 10 pF input load capacitance

TSUB, TADD, DPDA, DPDB, DPDC, DPDD: 10 pF output load capacitance TSUB, TADD: 10 k $\Omega$  load resistance

DPDA, DPDB, DPDC, DPDD: 100 k $\Omega$  load resistance

### Focus PD Input Characteristics (F1, F2)

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70 °C

Daramotor	Condition			Unit		
Falancici			min	typ	max	Onit
Input impedance	No signal		-	-	250	Ω
	DC to 10 kHz	RAM read <sup>1</sup>	-	-	24	
Input conversion noise current		ROM read <sup>1</sup>	-	-	96	nA <sub>rms</sub>
		RAM write <sup>1</sup>	-	-	150	
Pin voltage	No signal, V <sub>REF</sub> reference	No signal, V <sub>REF</sub> reference		-	±50	mV

1. Conversion from FSUB output noise value when 14 pF capacitors connected to F1 and F2

### **Focus Signal Processor Characteristics**

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70 °C

Daramotor	Condition			Unit		
Falancici			min	typ	max	Unit
FSUB current-to-voltage converter coefficient <sup>1</sup>	RAM read		370	415	460	
	ROM read	$R_f = 22 k\Omega$ , VOUT = VREF ± 0.35 V	94	105	116	kΩ
	RAM write		58	65	72	
FADD current-to-voltage converter coefficient <sup>2</sup>	RAM read		223	250	277	
	ROM read	R <sub>f</sub> = 27 kΩ	56.1	63	69.9	kΩ
	RAM write		35.6	40	44.1	
F1, F2 converter coefficient relative error	FSUB output, RAM/ROM read		-	-	±2	%
FSUB, FADD output impedance				-	100	Ω
FSUB operating output voltage	10 k $\Omega$ load connec	ted to VREF	1	-	3	V
FADD operating output voltage	10 k $\Omega$ load connec	ted to VREF	V <sub>REF</sub>	-	3	V
Converter coefficient quitching time	$RAM \text{ read} \leftrightarrow ROM \text{ read}$		-	-	10	ms
Converter coefficient switching time	$RAM\;write\leftrightarrowRAM\;read$		-	-	3	μs
FSUB, FADD signal bandwidth <sup>3</sup>	DC to -3 dB freque	ency	200	-	-	kHz
FSUB, FADD gain peaking <sup>3</sup>	f = 10 kHz to -3 dE	3 frequency	-3	-	+0.5	dB
FSUB, FADD phase response <sup>3</sup>	@ f = 10 kHz		-	-	5	0

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Deremeter		Condition		Rating			
Parameter		onation	min	typ	max	Unit	
FSUB offset voltage	No input signal	RAM read/write, ROM read max gain	-	-	±7.0		
	$V_{REF}$ reference, post-correction, $T_a = 25^{\circ}C$	RAM read, min to max gain	-	-	±21	mV	
		RAM read/write differential gain max.	-	-	±4		
FADD offset voltage	No input signal, V <sub>REF</sub> reference		-	-	±50	mV	
FSUB offset voltage temperature coefficient			-	-	±0.22	mV/°C	
FSUB variable gain range			-16	-	+14	dB	
FSUB variable gain step width			-	2	-	dB	
FSUB gain switching absolute accuracy	V <sub>OUT</sub> = V <sub>REF</sub> ± 0.35 V	-16 to +8 dB	-	-	±0.5	dB	
		+10 to +14 dB	_	_	±1.0		

 $\begin{array}{ll} 1. & \text{FSUB} = K \times [I_{F1} - I_{F2}], \text{ gain } = 0 \text{ dB} \\ 2. & \text{FADD} = K \times [I_{F1} + I_{F2}] \\ 3. & \text{F1}, \text{F2}: 14 \text{ pF input load capacitance} \\ & \text{FSUB}, \text{FADD}: 10 \text{ pF output load capacitance}, 10 \text{ k}\Omega \text{ load resistance} \\ \end{array}$ 

## **Mode Control Logic**

Contro	ol input	Operating mode	Offset correction	
WRITE	MODE	Operating mode		
LOW or open	LOW or open	RAM read		
LOW or open	HIGH	ROM read	Active	
HIGH	LOW or open	DAM write		
HIGH	HIGH	KAW WILE	Inactive	

## **Offset Correction Characteristics**

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70  $^{\circ}C$ 

Darameter	Symbol	Condition		Unit			
Falameter	Symbol	Condition	min	typ	max	Gill	
TSUB offset residual		$V_{REF}$ reference, $T_a = 25 \ ^{\circ}C$	-	-	±8.5	mV	
FSUB offset residual		$V_{REF}$ reference, $T_a = 25 \ ^{\circ}C$	-	-	±5.5	mV	
Supply voltage droop detect level	V <sub>1</sub>		1.9	2.8	3.7	V	
Correction circuit startup supply voltage	V <sub>2</sub>		3.2	3.8	4.4	V	
$V_1$ and $V_2$ difference	V <sub>2</sub> -V <sub>1</sub>		0.7	1.0	1.3	V	
Correction thermal sensor detect temperature			15	20	25	°C	
Offset correction time			-	-	150	ms	

## **Serial Interface Characteristics**

 $V_{CC}$  = 5 V  $\pm$  5%, GND = 0 V,  $T_a$  = 0 to 70  $^{\circ}C$ 

Darameter	Symbol	Condition		Unit		
Falameter	Symbol		min	typ	max	Unit
SCLK pulse cycle	t <sub>суSCK</sub>		100	-	-	ns
SCLK HIGH-level pulsewidth	t <sub>whSCK</sub>		40	-	-	ns
SCLK LOW-level pulsewidth	t <sub>wisck</sub>		40	-	-	ns
SENB setup time	t <sub>sSEN</sub>		20	-	-	ns
SENB hold time	t <sub>hSEN</sub>		40	-	-	ns
SDATA setup time	t <sub>sSDA</sub>		15	-	-	ns
SDATA hold time	<sup>t</sup> hSDA		15	-	-	ns
ACK setup time <sup>1</sup>	t <sub>sACK</sub>		0	-	20	ns
ACK hold time <sup>1</sup>	t <sub>hACK</sub>		-	-	50	ns
SENB interval	t <sub>inSEN</sub>		100	-	-	ns

1. ACK is the acknowledge output (n-channel open-drain). LOW-level output when the data received is valid. SDATA load capacitance is 15 pF.



## FUNCTIONAL DESCRIPTION

#### **Serial Interface**

The SM9103M uses a serial interface comprising 2 ports to control and set TSUB/FSUB output gain switching, sleep mode to reduce current consump-

tion, and TSUB/FSUB offset correction. The address and bit configuration of each port is shown in table 1.

Table 1. Port address and bit configuration<sup>1</sup>

Bit number															
15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
	Data Address														
MSB	ISB							LSB							
TG3	TG2	TG1	TG0	FG3	FG2	FG1	FG0	×	LOW	LOW	LOW	LOW	LOW	×	×
SL1	CS1	-	-	-	-	-	-	×	LOW	LOW	LOW	LOW	HIGH	×	×

1.  $\times$  = don't care, – = unassigned

TG3 to TG0: TSUB gain set bits. Default = 0111 (0 dB)

FG3 to FG0: FSUB gain set bits. Default = 0111 (0 dB)

SL1: sleep mode set bit. Sleep mode when 1, normal operation when 0. Default = 0.

CS1: offset correction control. Offset correction when 1, normal operation when 0. Default = 0.

Serial data is input on SDATA with the LSB first in sync with the falling edge of the SCLK clock. After the 16th SCLK falling edge and 16 bits of valid data has been input, the SDATA n-channel open-drain output goes LOW to perform the function of an acknowledge signal.

If the number of SCLK cycles which occur when SENB (serial interface enable) is HIGH is less than

#### **Data Signal Processor**

This stage creates the data signal and ID signal for output. The weak current from the tracking PD cells (T1, T2, T3, T4) are input to the front-end amplifier where the signals are current-to-voltage converted at fixed gain.

The gain setting is controlled by pins WRITE and MODE. WRITE switches between read/write, and MODE switches the gain between values corresponding to high-reflectivity and low-reflectivity discs. These signals control the settings for RAM (low-reflectivity disc) read/write and ROM (high-reflectivity disc) read. 16, the received data is ignored and the internal port is not updated. If the number of SCLK cycles is greater than 16, the data is still considered value up to the 16th SCLK falling edge, the data is latched into the internal port, and the acknowledge signal is output. The acknowledge signal is held until SENB goes LOW again.

The front-end amplifier outputs are processed by the signal processor block to generate intermediate signals. The data signal, (A + B + C + D), is converted to a difference signal by a differential output buffer and output on DATAP and DATAN. The ID signal, generated from the difference between 2 signals, (A + B) and (C + D), is converted to a difference signal by a differential output buffer and output on CAPAP and CAPAN. The data signal (DATAP, DATAN) and ID signal (CAPAP, CAPAN) DC components are removed using output stage capacitive networks.

T1, T2, T3 and T4 have a hold function to provide the appropriate reverse bias required by the tracking PD to ensure the data read bandwidth.

#### **Tracking Signal Processor**

The tracking stage generates the push-pull tracking error signal and output signal for DPD servo, as well as a push-pull sum signal used as an auxiliary signal.

The [(A + B) - (C + D)] signal from the common-data front-end amplifier and signal processor block is sent to the gain switching block. The gain switching block amplifies the difference signal using one of 16 preset gain settings in 2 dB steps to form a push-pull signal output on TSUB. A feedback resistor connected to TSUBB is used to ensure gain setting stability. The gain of the gain switching block is controlled by serial interface control bits as shown in table 2.

Each signal from T1, T2, T3, T4 is buffered and then output on DPDA, DPDB, DPDC, DPDD, respectively, for DPD servos.

The auxiliary signal is generated from the push-pull sum signal (A + B + C + D). This signal is buffered (TAB) and output on TADD. A feedback resistor connected to TADDB is used to ensure gain setting stability.

TG3	TG2	TG1 TG0		Gain (dB) <sup>1</sup>
0	0	0	0 0	
0	0	0	1	+12
0	0	1	0	+10
0	0	1	1	+8
0	1	0	0	+6
0	1	0	1	+4
0	1	1	0	+2
0	1	1	1	0
1	0	0	0	-2
1	0	0	1	-4
1	0	1	0	-6
1	0	1	1	-8
1	1	0	0	-10
1	1	0	1	-12
1	1	1	0	-14
1	1	1	1	-16

1. Default is 0 dB

#### **Focus Signal Processor**

The focus stage generates the focus error signal from the focus PD, and a sum signal. The weak focus PD current signals (F1, F2) are input to the front-end amplifier and then current-to-voltage converted at fixed gain.

The front-end amplifier output is sent to the signal processor block where the focus error signal (F1 - F2) and the sum signal (F1 + F2) are generated.

The focus error signal is sent to the gain switching block. The gain switching block amplifies the difference signal using one of 16 preset gain settings in 2 dB steps with output on FSUB. A feedback resistor connected to FSUBB is used to ensure gain setting stability. The gain of the gain switching block is controlled by serial interface control bits as shown in table 3.

The sum is buffered and output on FADD. A feedback resistor connected to FADDB is used to ensure gain setting stability.

#### Table 3. FSUB gain setting

Table 2. TSUB gain setting

FG3	FG2	FG1	FG0	Gain (dB) <sup>1</sup>
0	0	0	0	+14
0	0	0	1	+12
0	0	1	0	+10
0	0	1	1	+8
0	1	0	0	+6
0	1	0	1	+4
0	1	1	0	+2
0	1	1	1	0
1	0	0	0	-2
1	0	0	1	-4
1	0	1	0	-6
1	0	1	1	-8
1	1	0	0	-10
1	1	0	1	-12
1	1	1	0	-14
1	1	1	1	-16

1. Default is 0 dB

### **Offset Correction**

The SM9103M has built-in offset correction circuits for tracking and focus. During offset correction, the internal the device operates in RAM read mode, and FSUB and TSUB operate at maximum gain (+14 dB). The outputs on FSUB, FADD and TSUB are indeterminate. Also, inputs T1, T2, T3, T4 and F1, F2 may be ignored. After correction is complete, the FSUB and TSUB gain settings return to their default values (0 dB).

Offset correction is performed under the following conditions:

- When power is applied.
- When the supply drops below  $2.8 \pm 0.9$  V and then rises to above  $3.8 \pm 0.6$  V.
- When sleep mode operation is cancelled.
- When the serial interface bit CS1 is 1. Note that if SL1 is also 1, then SL1 has priority.

Table 4. Offset correction setting

CS1	Offset correction <sup>1</sup>
0	No correction
1	Correction

1. Default is No correction

### Sleep Mode

The SM9103M features a sleep mode which can be used when the device is not operating to significantly reduce current consumption. The sleep mode is controlled by serial interface bit SL1.

#### **Preset Function**

When power is applied or after offset correction, all serial interface flags are reset to their default values.

If the voltage falls below  $2.8 \pm 0.9$  V during offset correction, then correction stops and does not restart until the supply recovers to above  $3.8 \pm 0.6$  V.

During offset correction, the CALREQ output is held HIGH. CALREQ goes LOW after correction stops.

The SM9103M also incorporates a temperature detect function which detects temperature changes of  $20 \pm 5$  °C from the time the initial correction is performed. If a temperature change is detected, CAL-REQ goes HIGH and the device waits for an offset correction instruction.

Note that when both WRITE and MODE are HIGH, offset correction is inactive and the output appears uncorrected. However, if a correction start condition occurs when correction is inactive, such as the correction flag CS1 set to 1, then correction operation is initiated internally but does not appear at the output unless correction is activated prior to correction operation finishing.

Once correction has been made inactive, the output remains uncorrected even if correction is subsequently reactivated. In this case, the output remains uncorrected until a valid correction start condition is detected.

Table 5. Sleep mode settings

SL1	Sleep mode <sup>1</sup>	Mode description
LOW	OFF	Normal operation
HIGH	ON	Sleep condition

1. Default is OFF

Flags TG3 to TG0 and FG3 to FG0 are also set to their default values in sleep mode.

SM9103M

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