Monolithic Digital IC



LB1693

3-Phase Brushless Motor Driver

Overview

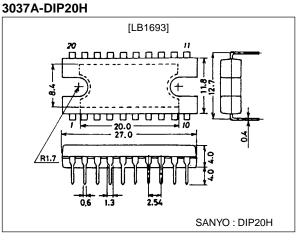
The LB1693 is a driver IC for 3-phase brushless motors. It is ideally suited for office automation equipment and DC fan motors.

Features

- 3-Phase brushless motor driver.
- 45V withstand voltage and 2.5A output current.
- PWM switch regulator control section.
- Current limitter.
- Overvoltage and overcurrent protection circuit.
- Thermal shutdown cirucit.
- Hall amp with hysteresis characteristic.

Package Dimensions

unit:mm



Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V _{CC} max		45	V
	V _M max		45	V
Maximum Output current	IO		2.5	A
Allowable power dissipation	Pd max	Independent IC	3	W
		With infinte heat sink	20	W
Operating temperature	Topr		-20 to +80	°C
Storage temperature	Tstg		-55 to +150	°C

Allowable Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V _{CC}		9 to 36	V
	VM		V _H to 41	V
Voltage regulator output current	IVH		0 to 20	mA
V _H supply voltage	VH		4.5 to 5.5	V
Comparator output current	losc		0 to 30	mA

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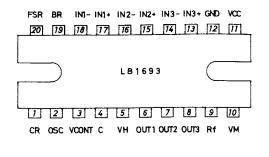
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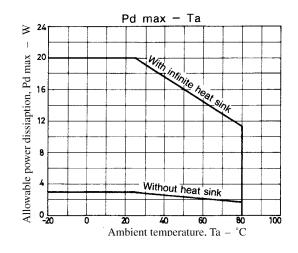
83198HA (KT)/4250TA, TS(GTPS) No.3295-1/9

Electrical Characteristics at Ta = 25 $^{\circ}C, V_{CC}$ = V_{M} = 24V

Parameter	Symbol	Conditions		Ratings		
	5,		min	typ	max	Unit
Supply current	ICC1	Stop mode		5	8	mA
	I _{CC²}	Hall current=5mA		15	21	mA
Output saturation voltage	V _O sat1	IO=1A, VO(sink)+VO(source)		2.1	3.0	V
	V _O sat2	IO=2A, VO(sink)+VO(source)		3.0	4.2	V
Output leakage current	I _O leak				100	μA
Voltage regulator output voltage	VH	I _{VH} =10mA	6.5	7.0	7.5	V
Voltage regulator load fluctuation	ΔV_{H1}	V _{CC} =9.5 to 36V		70	200	mV
Voltage regulator load fluctuation	ΔV_{H2}	I _{VH} =0 to 20mA		140	250	mV
Voltage Regulator temperature coefficient				-2		mV/°C
[Hall amp]	ł	•				
Input bias current	I _{HB}			1	4	μA
Common-mode input voltage range			1.5		V _H -1.8	V
Hysteresis width	ΔVIN		28	38	46	mV
Low to high input voltage	V _{SLH}		8	20	32	mV
High to low input voltage	VSHL		-32	-20	-8	mV
Oscillator		1				
High-level output voltage				3.45		V
Low-level output voltage				1.0		V
Oscillation frequency	f	R=36kΩ, C=4700pF		10		kHz
Amplitude			2.1	2.45	2.8	Vp-p
Temperature coefficient	Δf			0.1		%/°C
Comparator						
Output voltage	VOSC	I _{OSC} =30mA		1.1	1.5	V
Rising time	tr			0.5		μs
Falling time	tf			0.5		μs
Forward/Stop/Reverse	I	ł				
Forward	V _{FSR} 1			0	0.8	V
Stop	V _{FSR} ²		2.1	2.5	2.9	V
Reverse	V _{FSR} 3		4.2	5.0		V
Brake operation off	V _{BR} 1				0.8	V
Brake operation on	V _{BR} 2		2.0			V
Current limiter	DR					
Limiter1	VR _f 1		0.42	0.5	0.6	V
Lmiter2	VR _f 2		0.34	0.4	0.48	V
Overvoltage protection votlage	VOVSD		38	42	44.5	V
Hysteresis width			0.8	1.3	1.8	V
Thermal shutdown temperature	TSD	Design target	150	180		°C
Hysteresis width	ΔTSD			25		°C
Low-voltage protection voltage	VLVSD		3.6	4.0	4.4	V
Hysteresis width	∆VLVSD		0.04	0.11	0.18	V
Upper diode voltage	V _F	I _O =1A	0.8	2.8	4.7	v

Pin Assignment





Pin Description

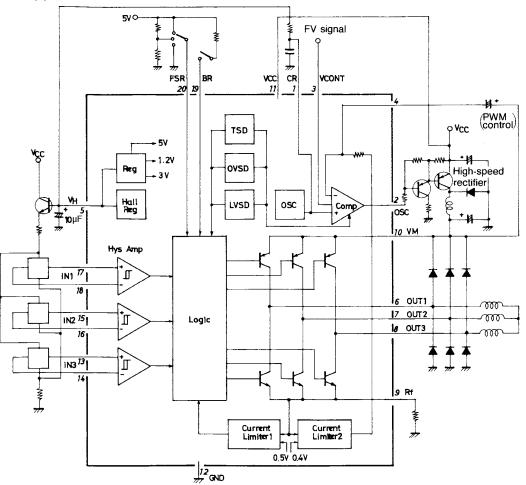
Pin Name	Pin No.	Description			
IN1 ⁺ , IN1 ⁻	17, 18	OUT1:Hall element input pins for Phase 1. High logic is the state when IN1 ⁺ > IN1 ⁻ .			
IN2+, IN2-	15, 16	OUT2: Hall element input pins for Phase 2. High" logic is the state when IN1 ⁺ > IN1 [−] .			
IN3+, IN3 ⁻	13, 14	OUT3: Hall element input pins for Phase 3. High logic is the state when IN1 ⁺ > IN1 ⁻ .			
OUT1	6	Output pin for Phase 1.			
OUT2	7	Output pin for Phase 2.			
OUT3	8	Output pin for Phase 3.			
V _{CC}	11	Power supply pin for applying voltage to each section other than output section.			
VM	10	Power supply for output section.			
R _f	9	Output current detect pin; R _f is inserted between this pin and ground to detect the output current as a voltage.			
GND	12	Ground for other output The minimum potential of output transistor is at the ${\sf R}_f$ pin.			
B _R	19	Brake pin The brake is switched on/off by setting this pin high (2V or more)/low (0.8V or less).			
FSR	20	Forward/Stop/Reverse control pin. The motor is driven forward, stopped, or driven in reverse according to the voltage at this pin. Forward : 0 to 0.8V Stop : 2.1 to 2.9V Reverse : 4.2 to 5.0V			
VH	5	Power pin for Hall elements When using the internal (stabilized) power supply : V_H =7V typ. When using the external (stabilized) power supply : V_H =5V typ.			
CR	1	Sets the oscillation frequency for the switching regulator/			
OSC	2	Outputs duty-controlled pulsed ; open collector output.			
VCONT	3	Speed control pin ; varies the swithcing regulator output votlage.			
С	4	Suppresses ripples in the motor current during operation of current limiter 2.			

Truth Table

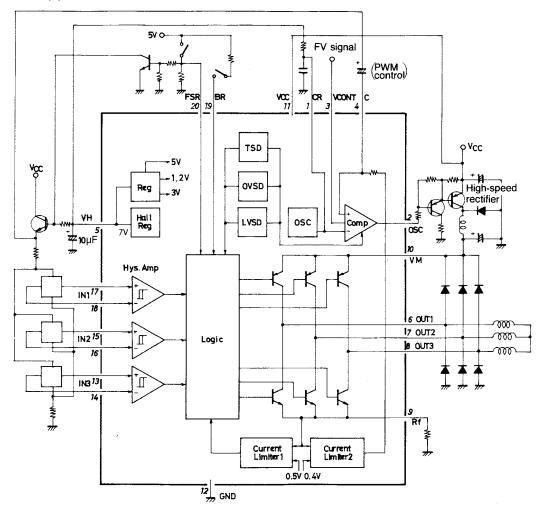
ltem	Source Sink	Input			Forward/Reverse	
	Source Slink	IN1	IN2	IN3	Control	
1	OUT3 → OUT2	н	н	L	L	
	OUT2 → OUT3				Н	
2	OUT3 → OUT1	н	L	L	L	
2	OUT1 → OUT3				Н	
3	OUT2 → OUT3	1	L	н	L	
	OUT3 → OUT2	Ŀ			Н	
4	OUT1 → OUT2		L H	L	L	
	OUT2 → OUT1	L			н	
5	OUT2 → OUT1	Ц	H L	н	L	
	OUT1 → OUT2	11			н	
6	OUT1 → OUT3		н	н	L	
	OUT3 → OUT1				н	

Block Diagram and Peripheral Circuit Diagram

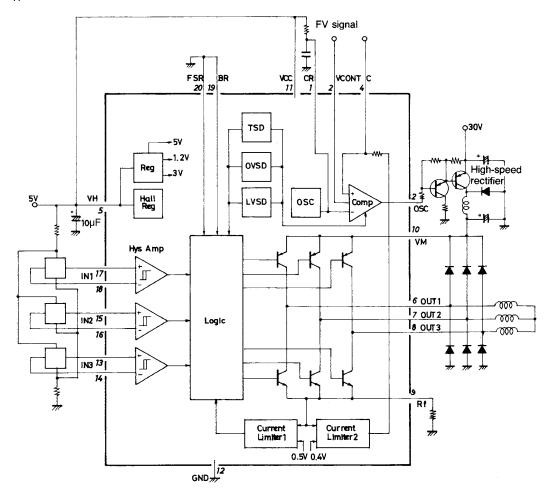
PWM control (1)



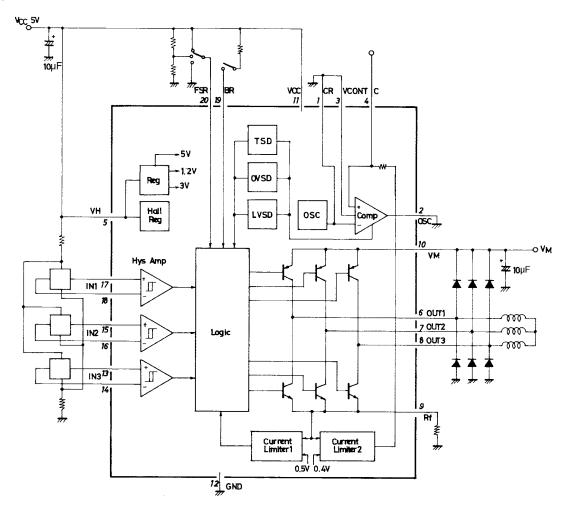
PWM control (2)



V_{CC}=V_H=5V PMW control



 $V_{CC}{=}5V\!,\,V_M$ are No speed control



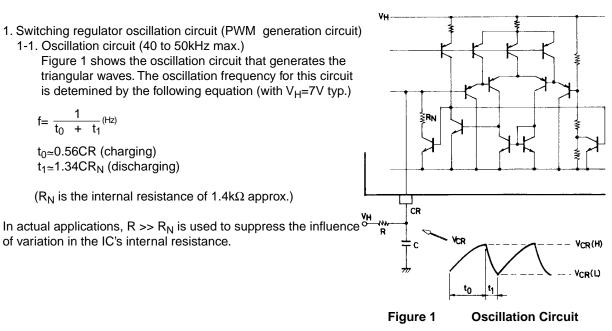
- 1. Switching regulator oscillation circuit (PWM generation circuit) 1-1. Oscillation circuit (40 to 50kHz max.)
 - Figure 1 shows the oscillation circuit that generates the triangular waves. The oscillation frequency for this circuit is detemined by the following equation (with $V_{H}=7V$ typ.)

$$f = \frac{1}{t_0 + t_1}$$
(Hz)

t₀≃0.56CR (charging) t₁~1.34CR_N (discharging)

of variation in the IC's internal resistance.

(R_N is the internal resistance of $1.4k\Omega$ approx.)



1-2. Comparator circuit

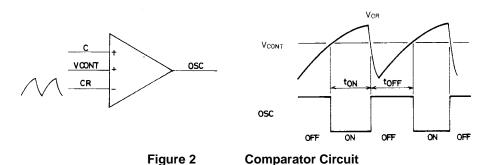
Figure 2 shows the comparator circuit for comparing the triangular wave output, the speed control signal, etc.

input terminals

CR Input the triangular wave output.

V_{CONT} Input the speed control signal.

C Goes high when current limiter 2 is operating. (When $V_{C(H)} > V_{CR(H)}$, the OSC output is off.)



2. Position detection circuit (Hall element input circuit)

The position detection circuit is a differential amp with hysteresis (38mV typ.). For the operating DC level, use within the common-mode phase input voltage range (1.5 to V_H –1.8V). Also it is recommended that the input level is at least three times (150 to 200mVp-p) the hysteresis.

3. V_H power supply circuit

The V_H power supply pins can be used to from the internal power supply or an external power supply. When using the internal power supply, the internal logic operates with V_H=7V typical (V_{CC}=24V). When using an external power supply, set V_{CC}=V_H=5V and operate the internal logic at 5V.

4. Current limiter circuits

4-1. Current limiter 1

The current is limited by moving the sink side transistor from saturated to undaturated, so ASO can be a problem.

$$I = \frac{V_{Rf}1}{R_f} \quad (A)$$

Therefore, design so that as much as possible current limiter 1 is not triggered. Also, take particular care not to exceed the maximum output current (2.5A) when current limiter 1 is triggered.

4-2. Current limiter 2

This circuit limits the current by lowering the PWM output duty, thus lowering the V_M voltage. When current limiter 2 is triggered, the output current is no greater than 2A.

$$I = \frac{V_{Rf}2}{R_f}$$

When not controlling the PWM, add a current limiter to the V_M power supply. (A current setting no greater than 60% to 70% of the current value of current limiter 1 and a short delay time are recommended.)

5. Protection circuits

- 5-1. Overvoltage protection circuit
 - If the voltage at the V_{CC} pin rises above the regulated votlage (38V), PWM output is inhibited and the sink side output driver is switched off.
- 5-2. Low-votlage protection circuit If the voltage at the V_{CC} pin falls below the regulated voltage, just as in 5-1, PWM output is inhibited and the sink side output driver is switched off.
- 5-3. Thermal shutdown circuit
 - If the junction temperature rises above the regulated temperature, just as in 5-1, PMW output is inhibited and the sink output driver is switched off.

Use a voltage greater than the V_H voltage for the V_M power supply vollage

6. Minimum voltage at V_M power

V_M≥V_H f – R Duty Ratio - VCONT 100 V_{CC} = 24V R=39kΩ ON OFF C=2200pF Vн Mr. . ≩Rx ^{90%} R₹ Oscillation Frequency, f - kHz CR OSC 80 % 100 60 Duty Cycle, Duty 330 pF 40 1000 pF 20 10 3300 pF 5 0 10 1.0 1.2 1.4 1.6 1.8 2.0 2.2 2.4 2.6 2.8 100 Resistance, R - kΩ Output Voltage, VCONT - V

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