Monolithic Digital IC



LB1693

# **3-Phase Brushless Motor Driver**

## **Overview**

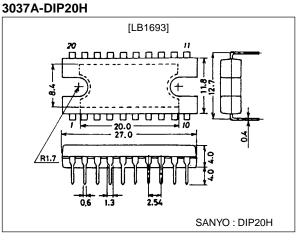
The LB1693 is a driver IC for 3-phase brushless motors. It is ideally suited for office automation equipment and DC fan motors.

## Features

- 3-Phase brushless motor driver.
- 45V withstand voltage and 2.5A output current.
- PWM switch regulator control section.
- Current limitter.
- Overvoltage and overcurrent protection circuit.
- Thermal shutdown cirucit.
- Hall amp with hysteresis characteristic.

## Package Dimensions

unit:mm



## **Specifications**

#### Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V <sub>CC</sub> max		45	V
	V <sub>M</sub> max		45	V
Maximum Output current	IO		2.5	A
Allowable power dissipation	Pd max	Independent IC	3	W
		With infinte heat sink	20	W
Operating temperature	Topr		-20 to +80	°C
Storage temperature	Tstg		-55 to +150	°C

#### Allowable Operating Conditions at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V <sub>CC</sub>		9 to 36	V
	VM		V <sub>H</sub> to 41	V
Voltage regulator output current	IVH		0 to 20	mA
V <sub>H</sub> supply voltage	VH		4.5 to 5.5	V
Comparator output current	losc		0 to 30	mA

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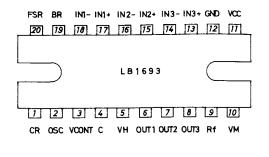
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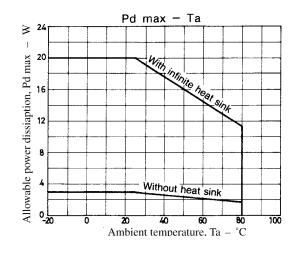
83198HA (KT)/4250TA, TS(GTPS) No.3295-1/9

## Electrical Characteristics at Ta = 25 $^{\circ}C, V_{CC}$ = $V_{M}$ = 24V

Parameter	Symbol	Conditions		Ratings		
	5,		min	typ	max	Unit
Supply current	ICC1	Stop mode		5	8	mA
	I <sub>CC<sup>2</sup></sub>	Hall current=5mA		15	21	mA
Output saturation voltage	V <sub>O</sub> sat1	IO=1A, VO(sink)+VO(source)		2.1	3.0	V
	V <sub>O</sub> sat2	IO=2A, VO(sink)+VO(source)		3.0	4.2	V
Output leakage current	I <sub>O</sub> leak				100	μA
Voltage regulator output voltage	VH	I <sub>VH</sub> =10mA	6.5	7.0	7.5	V
Voltage regulator load fluctuation	$\Delta V_{H1}$	V <sub>CC</sub> =9.5 to 36V		70	200	mV
Voltage regulator load fluctuation	$\Delta V_{H2}$	I <sub>VH</sub> =0 to 20mA		140	250	mV
Voltage Regulator temperature coefficient				-2		mV/°C
[Hall amp]	ł	•				
Input bias current	I <sub>HB</sub>			1	4	μA
Common-mode input voltage range			1.5		V <sub>H</sub> -1.8	V
Hysteresis width	ΔVIN		28	38	46	mV
Low to high input voltage	V <sub>SLH</sub>		8	20	32	mV
High to low input voltage	VSHL		-32	-20	-8	mV
Oscillator		1				
High-level output voltage				3.45		V
Low-level output voltage				1.0		V
Oscillation frequency	f	R=36kΩ, C=4700pF		10		kHz
Amplitude			2.1	2.45	2.8	Vp-p
Temperature coefficient	Δf			0.1		%/°C
Comparator						
Output voltage	VOSC	I <sub>OSC</sub> =30mA		1.1	1.5	V
Rising time	tr			0.5		μs
Falling time	tf			0.5		μs
Forward/Stop/Reverse	I	ł				
Forward	V <sub>FSR</sub> 1			0	0.8	V
Stop	V <sub>FSR</sub> <sup>2</sup>		2.1	2.5	2.9	V
Reverse	V <sub>FSR</sub> 3		4.2	5.0		V
Brake operation off	V <sub>BR</sub> 1				0.8	V
Brake operation on	V <sub>BR</sub> 2		2.0			V
Current limiter	DR					
Limiter1	VR <sub>f</sub> 1		0.42	0.5	0.6	V
Lmiter2	VR <sub>f</sub> 2		0.34	0.4	0.48	V
Overvoltage protection votlage	VOVSD		38	42	44.5	V
Hysteresis width			0.8	1.3	1.8	V
Thermal shutdown temperature	TSD	Design target	150	180		°C
Hysteresis width	ΔTSD			25		°C
Low-voltage protection voltage	VLVSD		3.6	4.0	4.4	V
Hysteresis width	∆VLVSD		0.04	0.11	0.18	V
Upper diode voltage	V <sub>F</sub>	I <sub>O</sub> =1A	0.8	2.8	4.7	v

### **Pin Assignment**





### **Pin Description**

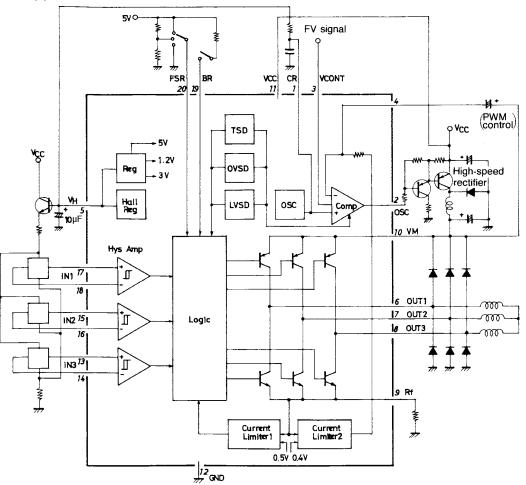
Pin Name	Pin No.	Description			
IN1 <sup>+</sup> , IN1 <sup>-</sup>	17, 18	OUT1:Hall element input pins for Phase 1. High logic is the state when IN1 <sup>+</sup> > IN1 <sup>-</sup> .			
IN2+, IN2-	15, 16	OUT2: Hall element input pins for Phase 2. High" logic is the state when IN1 <sup>+</sup> > IN1 <sup>−</sup> .			
IN3+, IN3 <sup>-</sup>	13, 14	OUT3: Hall element input pins for Phase 3. High logic is the state when IN1 <sup>+</sup> > IN1 <sup>-</sup> .			
OUT1	6	Output pin for Phase 1.			
OUT2	7	Output pin for Phase 2.			
OUT3	8	Output pin for Phase 3.			
V <sub>CC</sub>	11	Power supply pin for applying voltage to each section other than output section.			
VM	10	Power supply for output section.			
R <sub>f</sub>	9	Output current detect pin; R <sub>f</sub> is inserted between this pin and ground to detect the output current as a voltage.			
GND	12	Ground for other output The minimum potential of output transistor is at the ${\sf R}_f$ pin.			
B <sub>R</sub>	19	Brake pin The brake is switched on/off by setting this pin high (2V or more)/low (0.8V or less).			
FSR	20	Forward/Stop/Reverse control pin. The motor is driven forward, stopped, or driven in reverse according to the voltage at this pin. Forward : 0 to 0.8V Stop : 2.1 to 2.9V Reverse : 4.2 to 5.0V			
VH	5	Power pin for Hall elements When using the internal (stabilized) power supply : $V_H$ =7V typ. When using the external (stabilized) power supply : $V_H$ =5V typ.			
CR	1	Sets the oscillation frequency for the switching regulator/			
OSC	2	Outputs duty-controlled pulsed ; open collector output.			
VCONT	3	Speed control pin ; varies the swithcing regulator output votlage.			
С	4	Suppresses ripples in the motor current during operation of current limiter 2.			

## **Truth Table**

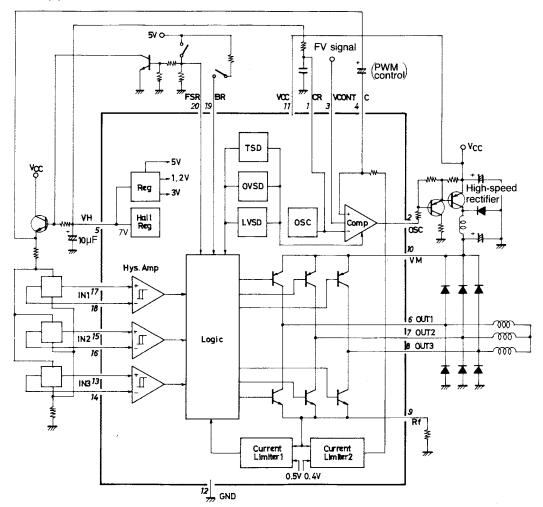
ltem	Source Sink	Input			Forward/Reverse	
	Source Slink	IN1	IN2	IN3	Control	
1	OUT3 → OUT2	н	н	L	L	
	OUT2 → OUT3				Н	
2	OUT3 → OUT1	н	L	L	L	
2	OUT1 → OUT3				Н	
3	OUT2 → OUT3	1	L	н	L	
	OUT3 → OUT2	Ŀ			Н	
4	OUT1 → OUT2		L H	L	L	
	OUT2 → OUT1	L			н	
5	OUT2 → OUT1	Ц	H L	н	L	
	OUT1 → OUT2	11			н	
6	OUT1 → OUT3		н	н	L	
	OUT3 → OUT1				н	

## Block Diagram and Peripheral Circuit Diagram

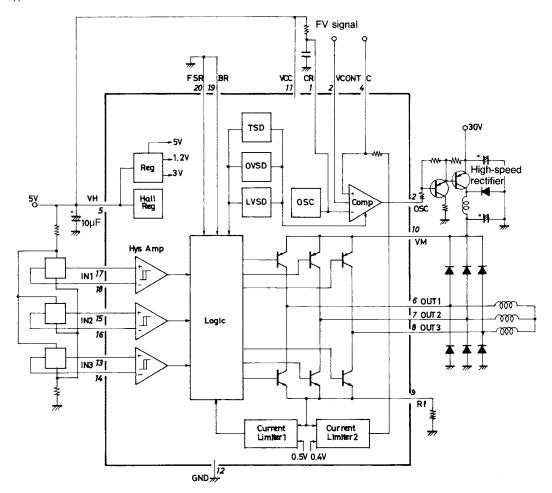
PWM control (1)



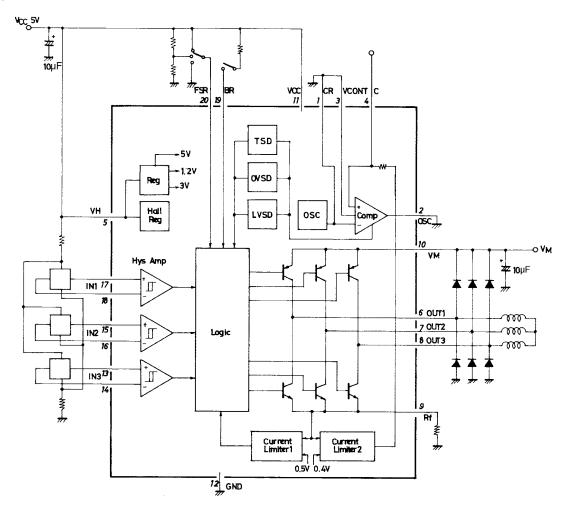
## PWM control (2)



V<sub>CC</sub>=V<sub>H</sub>=5V PMW control



 $V_{CC}{=}5V\!,\,V_M$  are No speed control



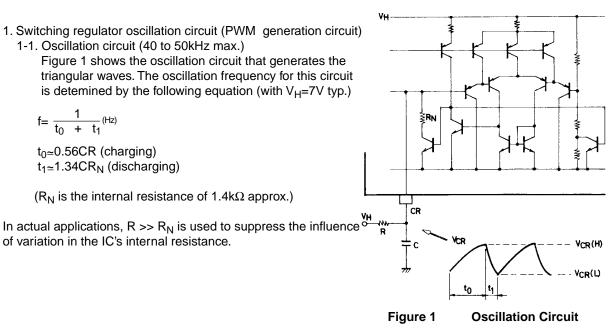
- 1. Switching regulator oscillation circuit (PWM generation circuit) 1-1. Oscillation circuit (40 to 50kHz max.)
  - Figure 1 shows the oscillation circuit that generates the triangular waves. The oscillation frequency for this circuit is detemined by the following equation (with  $V_{H}=7V$  typ.)

$$f = \frac{1}{t_0 + t_1}$$
(Hz)

t<sub>0</sub>≃0.56CR (charging) t<sub>1</sub>~1.34CR<sub>N</sub> (discharging)

of variation in the IC's internal resistance.

(R<sub>N</sub> is the internal resistance of  $1.4k\Omega$  approx.)



#### 1-2. Comparator circuit

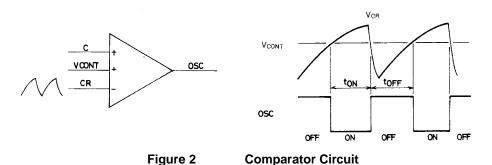
Figure 2 shows the comparator circuit for comparing the triangular wave output, the speed control signal, etc.

input terminals

CR Input the triangular wave output.

V<sub>CONT</sub> Input the speed control signal.

C Goes high when current limiter 2 is operating. (When  $V_{C(H)} > V_{CR(H)}$ , the OSC output is off.)



2. Position detection circuit (Hall element input circuit)

The position detection circuit is a differential amp with hysteresis (38mV typ.). For the operating DC level, use within the common-mode phase input voltage range (1.5 to  $V_H$ –1.8V). Also it is recommended that the input level is at least three times (150 to 200mVp-p) the hysteresis.

#### 3. V<sub>H</sub> power supply circuit

The V<sub>H</sub> power supply pins can be used to from the internal power supply or an external power supply. When using the internal power supply, the internal logic operates with V<sub>H</sub>=7V typical (V<sub>CC</sub>=24V). When using an external power supply, set V<sub>CC</sub>=V<sub>H</sub>=5V and operate the internal logic at 5V.

#### 4. Current limiter circuits

4-1. Current limiter 1

The current is limited by moving the sink side transistor from saturated to undaturated, so ASO can be a problem.

$$I = \frac{V_{Rf}1}{R_f} \quad (A)$$

Therefore, design so that as much as possible current limiter 1 is not triggered. Also, take particular care not to exceed the maximum output current (2.5A) when current limiter 1 is triggered.

4-2. Current limiter 2

This circuit limits the current by lowering the PWM output duty, thus lowering the  $V_M$  voltage. When current limiter 2 is triggered, the output current is no greater than 2A.

$$I = \frac{V_{Rf}2}{R_f}$$

When not controlling the PWM, add a current limiter to the  $V_M$  power supply. (A current setting no greater than 60% to 70% of the current value of current limiter 1 and a short delay time are recommended.)

#### 5. Protection circuits

- 5-1. Overvoltage protection circuit
  - If the voltage at the  $V_{CC}$  pin rises above the regulated votlage (38V), PWM output is inhibited and the sink side output driver is switched off.
- 5-2. Low-votlage protection circuit If the voltage at the V<sub>CC</sub> pin falls below the regulated voltage, just as in 5-1, PWM output is inhibited and the sink side output driver is switched off.
- 5-3. Thermal shutdown circuit
  - If the junction temperature rises above the regulated temperature, just as in 5-1, PMW output is inhibited and the sink output driver is switched off.

Use a voltage greater than the V<sub>H</sub> voltage for the V<sub>M</sub> power supply vollage

#### 6. Minimum voltage at V<sub>M</sub> power

V<sub>M</sub>≥V<sub>H</sub> f – R Duty Ratio - VCONT 100 V<sub>CC</sub> = 24V R=39kΩ ON OFF C=2200pF Vн Mr. . ≩Rx <sup>90%</sup> R₹ Oscillation Frequency, f - kHz CR OSC 80 % 100 60 Duty Cycle, Duty 330 pF 40 1000 pF 20 10 3300 pF 5 0 10 1.0 1.2 1.4 1.6 1.8 2.0 2.2 2.4 2.6 2.8 100 Resistance, R - kΩ Output Voltage, VCONT - V

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