

DS90LV032A 3V LVDS Quad CMOS Differential Line Receiver

3V LVDS Quad CMOS Differential Line Receiver

General Description

The DS90LV032A is a quad CMOS differential line receiver designed for applications requiring ultra low power dissipation and high data rates. The device is designed to support data rates in excess of 400 Mbps (200 MHz) utilizing Low Voltage Differential Signaling (LVDS) technology.

The DS90LV032A accepts low voltage (350 mV typical) differential input signals and translates them to 3V CMOS output levels. The receiver supports a TRI-STATE[®] function that may be used to multiplex outputs. The receiver also supports open, shorted and terminated (100 Ω) input Fail-safe. The receiver output will be HIGH for all fail-safe conditions.

The DS90LV032A and companion LVDS line driver (eg. DS90LV031A) provide a new alternative to high power PECL/ ECL devices for high speed point-to-point interface applications.

Features

- >400 Mbps (200 MHz) switching rates
- 0.1 ns channel-to-channel skew (typical)
- 0.1 ns differential skew (typical)
- 3.3 ns maximum propagation delay
- 3.3V power supply design
- Power down high impedance on LVDS inputs
- Low Power design (40mW @ 3.3V static)
- Interoperable with existing 5V LVDS networks
- Accepts small swing (350 mV typical) VID
- Supports open, short and terminated input fail-safe
- Compatible with ANSI/TIA/EIA-644
- Industrial temp. operating range (-40°C to +85°C)
- Available in SOIC and TSSOP Packaging

Functional Diagram

Connection Diagram



Order Number DS90LV032ATM or DS90LV032ATMTC See NS Package Number M16A or MTC16



10006702

Truth Table

ENABLES		INPUTS	OUTPUT
EN	EN*	R _{IN+} – R _{IN-}	R _{OUT}
L	Н	Х	Z
All other combinations		$V_{ID} \ge 0.1V$	Н
of ENABLE inputs		$V_{\rm ID} \leq -0.1 V$	L
		Full Fail-safe	
		OPEN/SHORT	н
		or Terminated	

TRI-STATE® is a registered trademark of National Semiconductor Corporation.

© 2010 National Semiconductor Corporation 100067

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage (V _{CC})	–0.3V to +4V			
Input Voltage (R _{IN+} , R _{IN-})	-0.3V to +3.9V			
Enable Input Voltage (EN, EN*)	-0.3V to (V _{CC} + 0.3V)			
Output Voltage (R _{OUT})	–0.3V to (V _{CC} + 0.3V)			
Maximum Package Power Dissipation @ +25°C				
M Package	1025 mW			
MTC Package	866 mW			
Derate M Package	8.2 mW/°C above +25°C			
Derate MTC Package	6.9 mW/°C above +25°C			
Storage Temperature Bange	65°C to 150°C			

Lead Temperature Range	
(Soldering 4 sec.)	+260°C
Maximum Junction Temperature	+150°C
ESD Rating (Note 10)	
(HBM 1.5 kΩ, 100 pF)	≥ 4.5 kV
(EIAJ 0 Ω, 200 pF)	≥ 250 V

Recommended Operating Conditions Min Typ May

	Min	Тур	Max	Units
Supply Voltage (V _{CC})	+3.0	+3.3	+3.6	V
Receiver Input Voltage	GND		+3.0	V
Operating Free Air				
Temperature (T _A)	-40	25	+85	°C

Electrical Characteristics

Over Supply Voltage and Operating Temperature ranges, unless otherwise specified. (Note 2)

Symbol	Parameter	Conditions	Pin	Min	Тур	Max	Units
V _{TH}	Differential Input High Threshold	$V_{CM} = +1.2V$	R _{IN+} ,		+20	+100	mV
V _{TL}	Differential Input Low Threshold	(Note 13)	R _{IN-}	-100	-20		mV
VCMR	Common-Mode Voltage Range	VID = 200 mV peak to peak (Note 5)]	0.1		2.3	V
I _{IN}	Input Current	$V_{IN} = +2.8V$ $V_{CC} = 3.6V \text{ or } 0V$		-10	±1	+10	μA
		$V_{IN} = 0V$		-10	±1	+10	μA
		$V_{IN} = +3.6V$ $V_{CC} = 0V$]	-20		+20	μA
V _{OH}	Output High Voltage	$I_{OH} = -0.4 \text{ mA}, V_{ID} = +200 \text{ mV}$	R _{OUT}	2.7	3.0		V
		$I_{OH} = -0.4$ mA, Input terminated	1	2.7	3.0		V
		$I_{OH} = -0.4$ mA, Input shorted		2.7	3.0		V
V _{OL}	Output Low Voltage	$I_{OL} = 2 \text{ mA}, V_{ID} = -200 \text{ mV}$			0.1	0.25	V
I _{os}	Output Short Circuit Current	Enabled, V _{OUT} = 0V (<i>Note 11</i>)		-15	-48	-120	mA
I _{oz}	Output TRI-STATE Current	Disabled, $V_{OUT} = 0V$ or V_{CC}		-10	±1	+10	μA
V _{IH}	Input High Voltage		EN,	2.0		V _{cc}	V
V _{IL}	Input Low Voltage		EN*	GND		0.8	V
I _I	Input Current	$V_{IN} = 0V \text{ or } V_{CC}$, Other Input = V_{CC} or GND]	-10	±1	+10	μA
V _{CL}	Input Clamp Voltage	$I_{CL} = -18 \text{ mA}$]	-1.5	-0.8		V
I _{CC}	No Load Supply Current	EN, EN* = V _{CC} or GND, Inputs Open	V _{cc}		10	15	mA
	Receivers Enabled	EN, EN* = 2.4V or 0.5V, Inputs Open]		10	15	mA
I _{CCZ}	No Load Supply Current Receivers Disabled	EN = GND, EN* = V _{CC} , Inputs Open			3	5	mA

Switching Characteristics

Over Supply Voltage and Operating Temperature ranges, unless otherwise specified. (Note 3, Note 4, Note 7, Note 8)

Symbol	Parameter	Conditions	Min	Тур	Max	Units
t _{PHLD}	Differential Propagation Delay High to Low	C _L = 10 pF	1.8		3.3	ns
t _{PLHD}	Differential Propagation Delay Low to High	V _{ID} = 200 mV	1.8		3.3	ns
t _{SKD1}	Differential Pulse Skew It _{PHLD} – t _{PLHD} I (Note 6)	(<i>Figure 1</i> and <i>Figure 2</i>)	0	0.1	0.35	ns
t _{SKD2}	Differential Channel-to-Channel Skew-same device (Note 7)		0	0.1	0.5	ns
t _{SKD3}	Differential Part to Part Skew (Note 8)				1.0	ns
t _{SKD4}	Differential Part to Part Skew (Note 9)				1.5	ns
t _{TLH}	Rise Time			0.35	1.2	ns
t _{THL}	Fall Time			0.35	1.2	ns
t _{PHZ}	Disable Time High to Z	$R_L = 2 k\Omega$		8	12	ns
t _{PLZ}	Disable Time Low to Z	C _L = 10 pF		6	12	ns
t _{PZH}	Enable Time Z to High	(Figure 3 and Figure 4)		11	17	ns
t _{PZL}	Enable Time Z to Low			11	17	ns
f _{MAX}	Maximum Operating Frequency (Note 14)	All Channels Switching	200	250		MHz

Note 1: "Absolute Maximum Ratings" are those values beyond which the safety of the device cannot be guaranteed. They are not meant to imply that the devices should be operated at these limits. The table of "Electrical Characteristics" specifies conditions of device operation.

Note 2: Current into device pins is defined as positive. Current out of device pins is defined as negative. All voltages are referenced to ground unless otherwise specified.

Note 3: All typicals are given for: $V_{CC} = +3.3V$, $T_A = +25^{\circ}C$.

Note 4: Generator waveform for all tests unless otherwise specified: f = 1 MHz, $Z_0 = 50\Omega$, t_r and t_r (0% to 100%) $\leq 3 \text{ ns for } R_{IN}$.

Note 5: The VCMR range is reduced for larger VID. Example: if VID = 400mV, the VCMR is 0.2V to 2.2V. The fail-safe condition with inputs shorted is valid over a common-mode range of 0V to 2.3V. A VID up to $V_{CC} - 0V$ may be applied to the R_{IN+}/R_{IN-} inputs with the Common-Mode voltage set to $V_{CC}/2$. Propagation delay and Differential Pulse skew decrease when VID is increased from 200mV to 400mV. Skew specifications apply for 200mV \leq VID \leq 800mV over the common-mode range .

Note 6: t_{SKD1} is the magnitude difference in differential propagation delay time between the positive going edge and the negative going edge of the same channel Note 7: t_{SKD2}, Channel-to-Channel Skew, is defined as the difference between the propagation delay of one channel and that of the others on the same chip with any event on the inputs.

Note 8: t_{SKD3} , part to part skew, is the differential channel-to-channel skew of any event between devices. This specification applies to devices at the same V_{CC} , and within 5°C of each other within the operating temperature range.

Note 9: t_{SKD4} , part to part skew, is the differential channel-to-channel skew of any event between devices. This specification applies to devices over recommended operating temperature and voltage ranges, and across process distribution. t_{SKD4} is defined as IMax – Minl differential propagation delay.

Note 10: ESD Rating:

HBM (1.5 kΩ, 100 pF) ≥ 4.5kV

EIAJ (0Ω, 200 pF) ≥ 250V

Note 11: Output short circuit current (I_{OS}) is specified as magnitude only, minus sign indicates direction only. Only one output should be shorted at a time, do not exceed maximum junction temperature specification.

Note 12: C_L includes probe and jig capacitance.

Note 13: V_{CC} is always higher than R_{IN+} and R_{IN-} voltage. R_{IN-} and R_{IN+} are allowed to have a voltage range –0.2V to V_{CC} – VID/2. However, to be compliant with AC specifications, the common voltage range is 0.1V to 2.3V

Note 14: f_{MAX} generator input conditions: $t_r = t_f < 1ns$ (0% to 100%), 50% duty cycle, differential (1.05V to 1.35V peak to peak). Output Criteria: 60%/40% duty cycle, V_{OL} (max 0.4V), V_{OH} (min 2.7V), Load = 10 pF (stray plus probes)

Parameter Measurement Information



FIGURE 1. Receiver Propagation Delay and Transition Time Test Circuit



FIGURE 2. Receiver Propagation Delay and Transition Time Waveforms



 \mathbf{C}_{L} includes load and test jig capacitance.

 $S_1 = V_{CC}$ for t_{PZL} , and t_{PLZ} measurements. $S_1 = GND$ for t_{PZH} and t_{PHZ} measurements.



4





Typical Application





Applications Information

General application guidelines and hints for LVDS drivers and receivers may be found in the following application notes: www.national.com/lvds.

LVDS drivers and receivers are intended to be primarily used in an uncomplicated point-to-point configuration as is shown in Figure 5. This configuration provides a clean signaling environment for the fast edge rates of the drivers . The receiver is connected to the driver through a balanced media which may be a standard twisted pair cable, a parallel pair cable, or simply PCB traces. Typically the characteristic impedance of the media is in the range of 100Ω . A termination resistor of 100Ω should be selected to match the media, and is located as close to the receiver input pins as possible. The termination resistor converts the driver output (current mode) into a voltage that is detected by the receiver. Other configurations are possible such as a multi-receiver configuration, but the effects of a mid-stream connector(s), cable stub(s), and other impedance discontinuities as well as ground shifting, noise margin limits, and total termination loading must be taken into account.

The DS90LV032A differential line receiver is capable of detecting signals as low as 100 mV, over a \pm 1V common-mode range centered around +1.2V. This is related to the driver offset voltage which is typically +1.2V. The driven signal is centered around this voltage and may shift \pm 1V around this center point. The \pm 1V shifting may be the result of a ground potential difference between the driver's ground reference and the receiver's ground reference, the common-mode effects of coupled noise, or a combination of the two. Both receiver input pins have a recommended operating input voltage range of 0V to +2.4V (measured from each pin to ground), exceeding these limits may turn on the ESD protection circuitry which will clamp the bus voltages.

Power Decoupling Recommendations:

Bypass capacitors must be used on power pins. High frequency ceramic (surface mount is recommended) 0.1μ F in parallel with 0.01μ F, in parallel with 0.001μ F at the power supply pin as well as scattered capacitors over the printed circuit board. Multiple vias should be used to connect the decoupling capacitors to the power planes A 10μ F (35V) or greater solid tantalum capacitor should be connected at the power entry point on the printed circuit board.

PC Board considerations:

Use at least 4 PCB layers (top to bottom); LVDS signals, ground, power, TTL signals.

Isolate TTL signals from LVDS signals, otherwise the TTL may couple onto the LVDS lines. It is best to put TTL and LVDS signals on different layers which are isolated by a power/ground plane(s).

Keep drivers and receivers as close to the (LVDS port side) connectors as possible.

Differential Traces:

Use controlled impedance traces which match the differential impedance of your transmission medium (ie. cable) and termination resistor. Run the differential pair trace lines as close together as possible as soon as they leave the IC (stubs should be < 10mm long). This will help eliminate reflections and ensure noise is coupled as common-mode. Lab experiments show that differential signals which are 1mm apart radiate far less noise than traces 3mm apart since magnetic field cancellation is much better with the closer traces. Plus, noise induced on the differential lines is much more likely to appear as common-mode which is rejected by the receiver.

Match electrical lengths between traces to reduce skew. Skew between the signals of a pair means a phase difference between signals which destroys the magnetic field cancellation benefits of differential signals and EMI will result. (Note the velocity of propagation, v = c/Er where c (the speed of light) = 0.2997mm/ps or 0.0118 in/ps). Do not rely solely on the autoroute function for differential traces. Carefully review dimensions to match differential impedance and provide isolation for the differential lines. Minimize the number of vias and other discontinuities on the line.

Avoid 90° turns (these cause impedance discontinuities). Use arcs or 45° bevels.

Within a pair of traces, the distance between the two traces should be minimized to maintain common-mode rejection of the receivers. On the printed circuit board, this distance should remain constant to avoid discontinuities in differential impedance. Minor violations at connection points are allowable.

Termination:

Use a resistor which best matches the differential impedance of your transmission line. The resistor should be between 90Ω and 130Ω . Remember that the current mode outputs need the termination resistor to generate the differential voltage. LVDS will not work without resistor termination. Typically, connect a single resistor across the pair at the receiver end.

Surface mount 1% to 2% resistors are best. PCB stubs, component lead, and the distance from the termination to the receiver inputs should be minimized. The distance between the termination resistor and the receiver should be <10mm (12mm MAX)

Probing LVDS Transmission Lines:

Always use high impedance (> $100k\Omega$), low capacitance (< 2 pF) scope probes with a wide bandwidth (1 GHz) scope. Improper probing will give deceiving results.

Cables and Connectors, General Comments:

When choosing cable and connectors for LVDS it is important to remember:

Use controlled impedance media. The cables and connectors you use should have a matched differential impedance of about 100Ω . They should not introduce major impedance discontinuities.

Balanced cables (e.g. twisted pair) are usually better than unbalanced cables (ribbon cable, simple coax.) for noise reduction and signal quality. Balanced cables tend to generate less EMI due to field canceling effects and also tend to pick up electromagnetic radiation as common-mode (not differential mode) noise which is rejected by the receiver. For cable distances < 0.5M, most cables can be made to work effectively. For distances $0.5M \le d \le 10M$, CAT 3 (category 3) twisted pair cable works well, is readily available and relatively inexpensive.

Fail-Safe Feature:

The LVDS receiver is a high gain, high speed device that amplifies a small differential signal (20mV) to CMOS logic levels. Due to the high gain and tight threshold of the receiver, care should be taken to prevent noise from appearing as a valid signal.

The receiver's internal fail-safe circuitry is designed to source/ sink a small amount of current, providing fail-safe protection (a stable known state of HIGH output voltage) for floating, terminated or shorted receiver inputs.

1. **Open Input Pins.** The DS90LV032A is a quad receiver device, and if an application requires only 1, 2 or 3 receivers, the unused channel(s) inputs should be left

OPEN. Do not tie unused receiver inputs to ground or any other voltages. The input is biased by internal high value pull up and pull down resistors to set the output to a HIGH state. This internal circuitry will guarantee a HIGH, stable output state for open inputs.

- Terminated Input. If the driver is disconnected (cable 2. unplugged), or if the driver is in a TRI-STATE or poweroff condition, the receiver output will again be in a HIGH state, even with the end of cable 100Ω termination resistor across the input pins. The unplugged cable can become a floating antenna which can pick up noise. If the cable picks up more than 10mV of differential noise, the receiver may see the noise as a valid signal and switch. To insure that any noise is seen as common-mode and not differential, a balanced interconnect should be used. Twisted pair cable will offer better balance than flat ribbon cable.
- Shorted Inputs. If a fault condition occurs that shorts З. the receiver inputs together, thus resulting in a 0V differential input voltage, the receiver output will remain in a HIGH state. Shorted input fail-safe is not supported across the common-mode range of the device (GND to 2.4V). It is only supported with inputs shorted and no external common-mode voltage applied.

External lower value pull up and pull down resistors (for a stronger bias) may be used to boost fail-safe in the presence of higher noise levels. The pull up and pull down resistors should be in the $5k\Omega$ to $15k\Omega$ range to minimize loading and waveform distortion to the driver. The common-mode bias point should be set to approximately 1.2V (less than 1.75V) to be compatible with the internal circuitry.

The footprint of the DS90LV032A is the same as the industry standard 26LS32 Quad Differential (RS-422) Receiver.

Pin No.	Name	Description	
2, 6,	R _{IN+}	Non-inverting receiver input pin	
10, 14			
1, 7,	R _{IN-}	Inverting receiver input pin	
9, 15			
3, 5,	R _{OUT}	Receiver output pin	
11, 13			
4	EN	Active high enable pin, OR-ed with EN*	
12	EN*	Active low enable pin, OR-ed with EN	
16	V _{CC}	Power supply pin, $+3.3V \pm 0.3V$	
8	GND	Ground pin	

Pin Descriptions

Ordering Information

Operating	Package Type/	Order Number
Temperature	Number	
-40°C to +85°C	SOP/M16A	DS90LV032ATM
-40°C to +85°C	TSSOP/MTC16	DS90LV032ATMTC



FIGURE 6. ICC vs Frequency, four channels switching



FIGURE 7. Typical Common-Mode Range variation with respect to amplitude of differential input



FIGURE 8. Typical Pulse Skew variation versus common-mode voltage



FIGURE 9. Variation in High to Low Propagation Delay versus VCM



FIGURE 10. Variation in Low to High Propagation Delay versus VCM

9



Notes

DS90LV032A

Notes

For more National Semiconductor product information and proven design tools, visit the following Web sites at: www.national.com

Pro	oducts	Design Support		
Amplifiers	www.national.com/amplifiers	WEBENCH® Tools	www.national.com/webench	
Audio	www.national.com/audio	App Notes	www.national.com/appnotes	
Clock and Timing	www.national.com/timing	Reference Designs	www.national.com/refdesigns	
Data Converters	www.national.com/adc	Samples	www.national.com/samples	
Interface	www.national.com/interface	Eval Boards	www.national.com/evalboards	
LVDS	www.national.com/lvds	Packaging	www.national.com/packaging	
Power Management	www.national.com/power	Green Compliance	www.national.com/quality/green	
Switching Regulators	www.national.com/switchers	Distributors	www.national.com/contacts	
LDOs	www.national.com/ldo	Quality and Reliability	www.national.com/quality	
LED Lighting	www.national.com/led	Feedback/Support	www.national.com/feedback	
Voltage References	www.national.com/vref	Design Made Easy	www.national.com/easy	
PowerWise® Solutions	www.national.com/powerwise	Applications & Markets	www.national.com/solutions	
Serial Digital Interface (SDI)	www.national.com/sdi	Mil/Aero	www.national.com/milaero	
Temperature Sensors	www.national.com/tempsensors	SolarMagic™	www.national.com/solarmagic	
PLL/VCO	www.national.com/wireless	PowerWise® Design University	www.national.com/training	

THE CONTENTS OF THIS DOCUMENT ARE PROVIDED IN CONNECTION WITH NATIONAL SEMICONDUCTOR CORPORATION ("NATIONAL") PRODUCTS. NATIONAL MAKES NO REPRESENTATIONS OR WARRANTIES WITH RESPECT TO THE ACCURACY OR COMPLETENESS OF THE CONTENTS OF THIS PUBLICATION AND RESERVES THE RIGHT TO MAKE CHANGES TO SPECIFICATIONS AND PRODUCT DESCRIPTIONS AT ANY TIME WITHOUT NOTICE. NO LICENSE, WHETHER EXPRESS, IMPLIED, ARISING BY ESTOPPEL OR OTHERWISE, TO ANY INTELLECTUAL PROPERTY RIGHTS IS GRANTED BY THIS DOCUMENT.

TESTING AND OTHER QUALITY CONTROLS ARE USED TO THE EXTENT NATIONAL DEEMS NECESSARY TO SUPPORT NATIONAL'S PRODUCT WARRANTY. EXCEPT WHERE MANDATED BY GOVERNMENT REQUIREMENTS, TESTING OF ALL PARAMETERS OF EACH PRODUCT IS NOT NECESSARILY PERFORMED. NATIONAL ASSUMES NO LIABILITY FOR APPLICATIONS ASSISTANCE OR BUYER PRODUCT DESIGN. BUYERS ARE RESPONSIBLE FOR THEIR PRODUCTS AND APPLICATIONS USING NATIONAL COMPONENTS. PRIOR TO USING OR DISTRIBUTING ANY PRODUCTS THAT INCLUDE NATIONAL COMPONENTS, BUYERS SHOULD PROVIDE ADEQUATE DESIGN, TESTING AND OPERATING SAFEGUARDS.

EXCEPT AS PROVIDED IN NATIONAL'S TERMS AND CONDITIONS OF SALE FOR SUCH PRODUCTS, NATIONAL ASSUMES NO LIABILITY WHATSOEVER, AND NATIONAL DISCLAIMS ANY EXPRESS OR IMPLIED WARRANTY RELATING TO THE SALE AND/OR USE OF NATIONAL PRODUCTS INCLUDING LIABILITY OR WARRANTIES RELATING TO FITNESS FOR A PARTICULAR PURPOSE, MERCHANTABILITY, OR INFRINGEMENT OF ANY PATENT, COPYRIGHT OR OTHER INTELLECTUAL PROPERTY RIGHT.

LIFE SUPPORT POLICY

NATIONAL'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS PRIOR WRITTEN APPROVAL OF THE CHIEF EXECUTIVE OFFICER AND GENERAL COUNSEL OF NATIONAL SEMICONDUCTOR CORPORATION. As used herein:

Life support devices or systems are devices which (a) are intended for surgical implant into the body, or (b) support or sustain life and whose failure to perform when properly used in accordance with instructions for use provided in the labeling can be reasonably expected to result in a significant injury to the user. A critical component is any component in a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system or to affect its safety or effectiveness.

National Semiconductor and the National Semiconductor logo are registered trademarks of National Semiconductor Corporation. All other brand or product names may be trademarks or registered trademarks of their respective holders.

Copyright© 2010 National Semiconductor Corporation

For the most current product information visit us at www.national.com



National Semiconductor Americas Technical Support Center Email: support@nsc.com Tel: 1-800-272-9959

National Semiconductor Europe Technical Support Center Email: europe.support@nsc.com National Semiconductor Asia Pacific Technical Support Center Email: ap.support@nsc.com National Semiconductor Japan Technical Support Center Email: jpn.feedback@nsc.com