30W Isolated Wide-Range DC/DC Converters

Featuring a full 30 Watt output in 1.8 square inches of board area, the UEI series isolated DC/DC converter family offers efficient regulated DC power for printed circuit board mounting.

### **PRODUCT OVERVIEW**

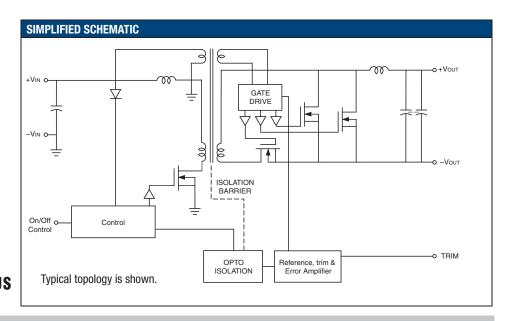
Wide range 4:1 inputs on the  $0.92" \times 1.92" \times 0.35"$ converter are either 9 to 36 Volts DC (Q12 models) or 18 to 75 Volts DC (Q48 models), ideal for battery-powered and telecom equipment. Fixed output voltages from 3.3 VDC to 15 VDC are tightly regulated and may be trimmed within  $\pm 10\%$ of nominal output. Applications include small instruments, computer-based systems, data communications equipment, remote sensor systems, vehicle and portable electronics.

The UEI 30W Series includes full magnetic and optical isolation up to 2250 Volts DC (basic insulation), Q48 models. For connection to digital systems, the outputs offer fast settling to current step loads and tolerance of higher capacitive loads. Excellent ripple and noise specifications assure compatibility to circuits using CPU's, ASIC's, programmable logic and FPGA's. No minimum load is required. For systems requiring controlled startup/shutdown, an external switch, transistor or digital logic may be used to activate the remote On/Off control. Remote Sense inputs compensate for resistive line drops at high currents.

A wealth of self-protection features avoid both converter and external circuit problems. These include input undervoltage lockout and overtemperature shutdown. The outputs current limit using the "hiccup" autorestart technique and the outputs may be short-circuited indefinitely. Additional features include output overvoltage and reverse conduction elimination.

The synchronous rectifier forward topology offers high efficiency for minimal heat buildup and "no fan" operation.

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muRata

**Murata Power Solutions** 

Typical unit

### FEATURES

- Small footprint DC/DC converter, ideal for high current applications
- 0.92" x 1.92" x 0.35" open frame package
- Wide range input voltages 9-36 and 18-75Vdc
- Assembly and attachment for RoHS-6 hazardous substance compliance
- Isolation up to 2250 VDC (basic), Q48 models
- Up to 30W total output power with overtemperature shutdown
- High efficiency synchronous rectifier forward topology
- Stable no-load operation with no required external components
- -40 to +85°C temperature range; see derating
- Certified to UL60950-1, CSA-C22.2 No. 234, EN60950-1 safety approvals
- Extensive self-protection shut down features
- RoHS-6 hazardous substance compliant

D

For full details go to

v.murata-ps.com/roh

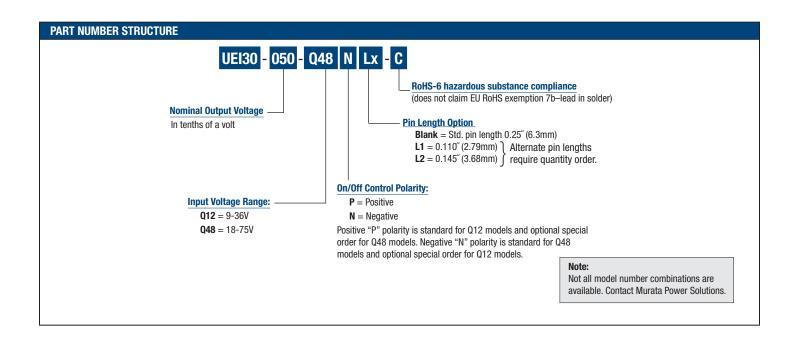
## **UEI30 Series**

30W Isolated Wide-Range DC/DC Converters

	Output								Ir	nput				Open Frame	
			Power	R/N (n	nVp-p)	Regulati	on (Max.)	VIN		lin,		Effic	iency	Packag	je, C80
Part Number $^{ extsf{D}}$	Vоит (V)	Іоит (А)	(W)	Тур.	Max.	Line	Load	Nom. (V)	Range (V)	no load (mA)	lin, full load (A)	Min.	Тур.	Case	Pinout
UEI30-033-Q12P-C	3.3	9	29.7	25	35	±0.2%	±0.25%	24	9-36	130	1.39	87.3%	89%	C80	P21
UEI30-033-Q48N-C	3.3	9	29.7	50	75	±0.2%	±0.25%	48	18-75	50	0.69	87%	89.5%	C80	P21
UEI30-050-Q12P-C	5	6	30	35	50	±0.2%	±0.2%	24	9-36	130	1.4	88%	89.5%	C80	P21
UEI30-050-Q48N-C	5	6	30	50	75	±0.2%	±0.2%	48	18-75	130	0.69	89.5%	91%	C80	P21
UEI30-120-Q12P-C	12	2.5	30	60	120	±0.2%	±0.1%	24	9-36	75	1.4	87.5%	89%	C80	P21
UEI30-120-Q48N-C	12	2.5	30	30	60	±0.2%	±0.1%	48	18-75	40	0.7	87.5%	89%	C80	P21
UEI30-150-Q12P-C	15	2	30	40	65	±0.2%	±0.1%	24	9-36	95	1.4	87.5%	89%	C80	P21
UEI30-150-Q48N-C	15	2	30	50	100	±0.2%	±0.1%	48	18-75	50	0.7	87.5%	89.5%	C80	P21

D Please refer to the part number structure for additional options and complete ordering part numbers.
All specifications are at nominal line voltage and full load, +25 deg.C. unless otherwise noted. See detailed specifications.

Output capacitors are 1 µF ceramic in parallel with 10 µF electrolytic. Input cap is 22 µF, low ESR. These I/O caps are necessary for our test equipment and may not be needed for your application. ③ Sense input is not included for 12 Vout and higher models. Sense is optional for 5 Vout and lower.



#### **Soldering Guidelines**

Murata Power Solutions recommends the specifications below when installing these converters. These specifications vary depending on the solder type. Exceeding these specifications may cause damage to the product. Your production environment may differ; therefore please thoroughly review these guidelines with your process engineers.

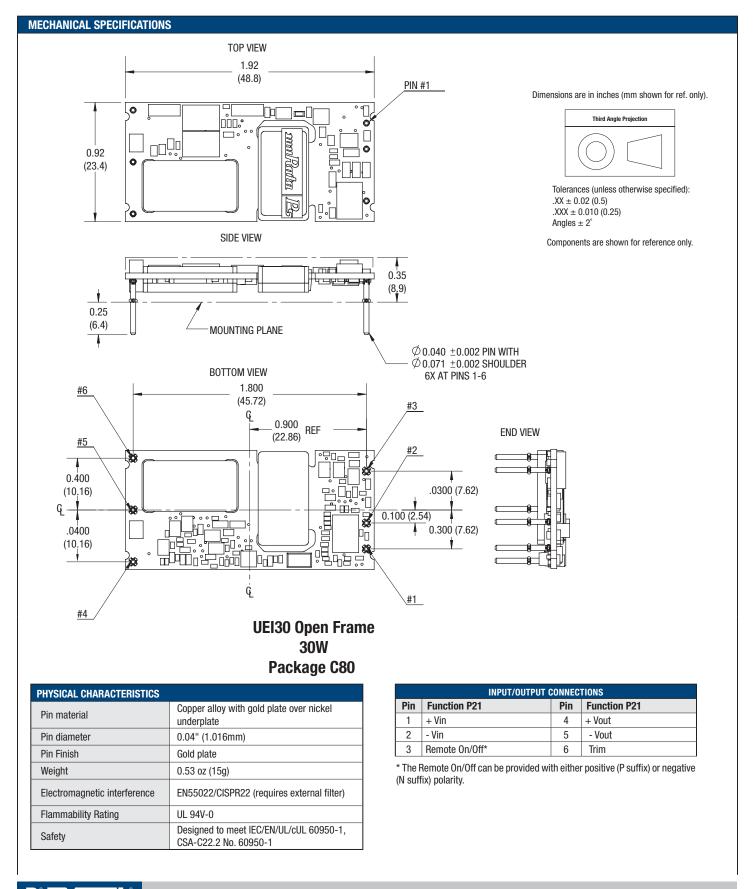
Wave Solder Operations for through-hole mounted products (THMT)								
For Sn/Ag/Cu based solders: For Sn/Pb based solders:								
Maximum Preheat Temperature 115° C.		Maximum Preheat Temperature	105° C.					
Maximum Pot Temperature	270° C.	Maximum Pot Temperature	250° C.					
Maximum Solder Dwell Time	7 seconds	Maximum Solder Dwell Time	6 seconds					



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### **UEI30 Series**

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INPUT CHARACTERIS	TICS													
	Start-up	Under- voltage	Reflected (back)		Input C	Current		Recom-	Internal	Reverse		Remote On/Off Control		
Model Family	threshold	Shut- down	Ripple Current <sup>2</sup>	Inrush Tran- sient	Output Short Circuit	Low Line	Standby Mode	mended Fuse	Input Filter Type	Polarity Protection	On/Off Current	Positive Logic	Negative Logic	
	V	V	mA pk-pk	A <sup>2</sup> sec	mA	Α	mA	Α			mA	"P" model suffix	"N" model suffix	
UEI30-033-Q12P-C	9.5	8.5				3.75		6						
UEI30-033-Q48N-C	17.3	16.0*				1.89	_	6			-			
UEI30-050-Q12P-C	9.5	8.0			50	3.75		6		None,	I	OFF=Gnd pin or	OFF=open pin	
UEI30-050-Q48N-C	17.0	16.2	00	0.05	50	1.89	1	6	L-C	install		-0.7 to +1.2V	or +10 to +15V	
UEI30-120-Q12P-C	9.5	8.3	30	0.05		3.75	3	6	L-0	external	3.5	max. ON=open pin or +10 to	max. ON=Gnd pin or –0.7 to	
UEI30-120-Q48N-C	17.0	16.5	1			1.85		4		fuse		+15V max.	+1.2V max.	
UEI30-150-Q12P-C	9.5	8.3	1		80	3.72	1	6			1			
UEI30-150-Q48N-C	16.9	16.3	1		50	1.83	1	4						

\*Specified at half load

OUTPUT CHARACTERISTICS										
	Vout Accuracy	Adjustment	Temperature	Capacitive Loading Max.	Overvoltage protection	ur 6	ion d	/Noise MHz vidth) <sup>8</sup>	oad tion	ncy
Model Family	50% Load	Range	Coefficient	Low ESR <0.02Ω Max, resistive load	Hiccup auto-start after fault removal	Minimum Ioading	0V protection method	Ripple/Noise (20 MHz bandwidth) <sup>8</sup>	Line/Load Regulation	Efficiency
	% of VNOM	% of Vnom	% of Vout /°C	μF	V	-	_ <u>a</u>	Ri ba		
UEI30-033-Q12P-C	2				5.0					
UEI30-033-Q48N-C					5.0					
UEI30-050-Q12P-C					7.0					
UEI30-050-Q48N-C		. 10	±0.02	2.000	7.3	No	Magnetic	O	da	
UEI30-120-Q12P-C	1	±10	±0.02	2,000	15.5	INO	feedback	See ordering guide		
UEI30-120-Q48N-C	]				14.1					
UEI30-150-Q12P-C	]				18.5	]				
UEI30-150-Q48N-C	]				24					

### ABSOLUTE MAXIMUM RATINGS

	Q12 models	Volts, max. continuous	0-36 VDC to rated specifications				
Input Voltage		Volts, transient, 100 mSec	50 VDC, no damage				
input voltage	Q48 models	Volts, max. continuous	0-75 VDC to rated specifications				
	Q46 III00EIS	Volts, transient, 100 mSec	100 VDC, no damage				
On/Off control, r	eferred to –Vin		-0.7 V. min to +15V max.				
Input Reverse Po	plarity Protection	on	None, install external fuse				
Output Overvolta	age		Vout nom. +20% max.				
Output Current			Current-limited. Devices can withstand sustained short circuit without damage. The outputs are not intended to accept appreci- able reverse current.				
Overtemperature	e Protection	Device includes electronic over- temperature shutdown protection under normal operation.					
Storage Tempera	ature		-55 to +125° C.				
Lead Temperatu	re	See soldering specifications					

#### **Absolute Maximum Ratings**

Absolute maximums are stress ratings. Exposure of devices to greater than any of these conditions may adversely affect long-term reliability. Proper operation under conditions other than those listed in the Performance/Functional Specifications Table is not implied nor recommended.

#### **Maximum Ratings Notes**

The transient specifications indicate that sample lots were successfully tested for 100 mS at the transient stress voltage and were not damaged. As a practical matter in your application, it is often difficult to determine how long an input overvoltage was applied. Therefore, do not exceed the continuous voltage rating.



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ISOLATION CHARACTERISTICS									
Model Family	Input to Output.	lsolation Resistance	Isolation Capacitance	Insulation Safety					
	Min	Min	-	Rating					
	VDC	MΩ	pF	Rating					
UEI30-033-Q12P-C	2000		1000						
UEI30-033-Q48N-C	2250		1000						
UEI30-050-Q12P-C	2000		1000						
UEI30-050-Q48N-C	2250	10	1000	Basic					
UEI30-120-Q12P-C	2000	10	1500	Dasic					
UEI30-120-Q48N-C	2250		1500						
UEI30-150-Q12P-C	2000		1500						
UEI30-150-Q48N-C	2250		2000						

DYNAMIC CHARACTERISTICS								
	Dynamic Load	Sta	rt-up Time					
Model Family	Response (50-75-50% load step)	Vin to Vout regulated (Max.)		Switching Frequency				
	µsec	mSec	mSec	KHz				
UEI30-033-Q12P-C	120 to 2%			275				
UEI30-033-Q48N-C	180 to 2%			280				
UEI30-050-Q12P-C	80 to 2%			275				
UEI30-050-Q48N-C	100 to 1%	50	50	275				
UEI30-120-Q12P-C	200 to 1%	50	50	275				
UEI30-120-Q48N-C	150 to 1%			275				
UEI30-150-Q12P-C	150 to 1%	]		275				
UEI30-150-Q48N-C	150 to 1%	]		275				

MISCELLANEOUS CH	ARACTERISTICS								
	Output Current Limit Inception	Output Short	Output Short	Output Short	Output		Storage	Thermal	Relative Humidity,
Model Family	98% of Vout, after warmup	Circuit Protection Method	Circuit Current	Circuit Duration (output shorted to ground)	Pre-biased setup	Operating Temperature Range	temperature range	shutdown	non- condensing
	Α	INCLIOU	Α				°C	°C	°C
UEI30-033-Q12P-C	11.5		0.3 max.						
UEI30-033-Q48N-C	11.2		0.3 max.						
UEI30-050-Q12P-C	7.9	Current	3.0						
UEI30-050-Q48N-C	7.0	limiting,	0.3 max.	Quattana	Monotonic	-40 to +85°C;	FF to 10500	C 115	To +85⁰C/
UEI30-120-Q12P-C	4.1	hiccup auto	1.5	Continuous	(external Vout < Vset)	with Derating (see Notes)	–55 to 125°C		85% RH
UEI30-120-Q48N-C	3.65	restart	0.75			(000 110100)			
UEI30-150-Q12P-C	3.0		0.5	1					
UEI30-150-Q48N-C	3.25		0.1 max.	1					

#### **Specification Notes:**

(1) All models are tested and specified with external 1 µF and 10 µF parallel output capacitors and a 22 µF external input capacitor. All capacitors are low ESR types. These capacitors are necessary to accommodate our test equipment and may not be required to achieve specified performance in your applications. All models are stable and regulate within spec under no-load conditions.

All specifications are typical unless noted. General conditions for Specifications are +25 deg.C, Vin=nominal, Vout=nominal, full load. Adequate airflow must be supplied for extended testing under power.

(2) Input Back Ripple Current is tested and specified over a 5 Hz to 20 MHz bandwidth. Input filtering is Cin=33 µF, 100V, Cbus=220 µF, 100V, Lbus=12 µH.

(3) Note that Maximum Power Derating curves indicate an *average* current at nominal input voltage. At higher temperatures and/or lower airflow, the DC/DC converter will tolerate brief full current outputs if the total RMS current over time does not exceed the Derating curve. All Derating curves are presented at sea level altitude. Be aware of reduced power dissipation with increasing density altitude.

(4) Refer to page 10 for MTBF values.

(5) The On/Off Control is normally selected by a switch or an open collector or open drain transistor. But it may also be driven with external logic or by applying appropriate external voltages which are referenced to Input Common and do not exceed the On/Off voltage specifications.

(6) Output current limiting begins when the output voltage degrades approximately 2% from the selected setting.

(7) The outputs are not intended to sink appreciable reverse current.

(8) Output noise may be further reduced by adding an external filter. Low voltage logic circuits may have a small voltage margin between logic ZERO and logic ONE, requiring noise suppression. Use only as much output filtering as needed to achieve your noise requirements. Excessive output capacitance can retard transient response or possibly cause instability. Low ESR ceramic capacitors may degrade dynamic performance. Be sure to thoroughly test your system under full load with all components installed.

(9) All models are fully operational and meet published specifications, including "cold start" at -40°C. (10) Regulation specifications describe the deviation as the line input voltage or output load current is varied from a nominal midpoint value to either extreme.

(11) The output overvoltage protection is automatic recovery. The overvoltage may occur either from internal failure or from an external forcing voltage as in a shared power system.

(12) Output overvoltage and short circuit protection is non-latching. When the overvoltage fault is removed, the converter will immediately recover. After an output overcurrent or short circuit, "hiccup' operation repeatedly attempts to restart the converter with a brief, full-current output. If the overcurrent condition still exists, the restart current will be removed and then tried again. This short current pulse prevents overheating and damaging the converter. Once the fault is removed, the converter immediately resumes normal operation.

(13) Do not exceed maximum power specifications when adjusting the output trim.

(14) At zero output current, the output may contain low frequency components which exceed the ripple specification. The output may be operated indefinitely with no load.

(15) If reverse polarity is accidentally applied to the input, to ensure reverse input protection with full output load, always connect an external input *fuse* in series with the +Vin input. Use approximately twice the full input current rating with nominal input voltage.

**CAUTION:** This product is not internally fused. To comply with safety agency certifications and to avoid injury to personnel or equipment, the user **must** connect an external fast-blow fuse to the input terminals. See fuse information.



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#### APPLICATION NOTES

#### **Input Fusing**

Certain applications and/or safety agencies may require fuses at the inputs of power conversion components. Fuses should also be used when there is the possibility of sustained input voltage reversal which is not current-limited. For greatest safety, we recommend a fast blow fuse installed in the ungrounded input supply line.

The installer must observe all relevant safety standards and regulations. For safety agency approvals, install the converter in compliance with the end-user safety standard.

#### **Input Reverse-Polarity Protection**

If the input voltage polarity is reversed, an internal diode will become forward biased and likely draw excessive current from the power source. If this source is not current-limited or the circuit appropriately fused, it could cause permanent damage to the converter.

#### Input Under-Voltage Shutdown and Start-Up Threshold

Under normal start-up conditions, converters will not begin to regulate properly until the rising input voltage exceeds and remains at the Start-Up Threshold Voltage (see Specifications). Once operating, converters will not turn off until the input voltage drops below the Under-Voltage Shutdown Limit. Subsequent restart will not occur until the input voltage rises again above the Start-Up Threshold. This built-in hysteresis prevents any unstable on/off operation at a single input voltage.

Users should be aware however of input sources near the Under-Voltage Shutdown whose voltage decays as input current is consumed (such as capacitor inputs), the converter shuts off and then restarts as the external capacitor recharges. Such situations could oscillate. To prevent this, make sure the operating input voltage is well above the UV Shutdown voltage AT ALL TIMES.

#### **Start-Up Delay**

Assuming that the output current is set at the rated maximum, the Vin to Vout Start-Up Delay (see Specifications) is the time interval between the point when the rising input voltage crosses the Start-Up Threshold and the fully loaded regulated output voltage enters and remains within its specified regulation band. Actual measured times will vary with input source impedance, external input capacitance, input voltage slew rate and final value of the input voltage as it appears at the converter.

These converters include a soft start circuit to moderate the duty cycle of the PWM controller at power up, thereby limiting the input inrush current.

The On/Off Remote Control interval from inception to Vour regulated assumes that the converter already has its input voltage stabilized above the Start-Up Threshold before the On command. The interval is measured from the On command until the output enters and remains within its specified accuracy band. The specification assumes that the output is fully loaded at maximum rated current.

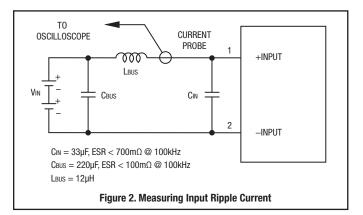
#### **Input Source Impedance**

These converters will operate to specifications without external components, assuming that the source voltage has very low impedance and reasonable input voltage regulation. Since real-world voltage sources have finite impedance, performance is improved by adding external filter components. Sometimes only a small ceramic capacitor is sufficient. Since it is difficult to totally characterize all applications, some experimentation may be needed. Note that external input capacitors must accept high speed switching currents.

Because of the switching nature of DC/DC converters, the input of these converters must be driven from a source with both low AC impedance and adequate DC input regulation. Performance will degrade with increasing input inductance. Excessive input inductance may inhibit operation. The DC input regulation specifies that the input voltage, once operating, must never degrade below the Shut-Down Threshold under all load conditions. Be sure to use adequate trace sizes and mount components close to the converter.

#### I/O Filtering, Input Ripple Current and Output Noise

All models in this converter series are tested and specified for input reflected ripple current and output noise using designated external input/output components, circuits and layout as shown in the figures below. External input capacitors (C<sub>N</sub> in the figure) serve primarily as energy storage elements, minimizing line voltage variations caused by transient IR drops in the input conductors. Users should select input capacitors for bulk capacitance (at appropriate frequencies), low ESR and high RMS ripple current ratings. In the figure below, the CBUS and LBUS components simulate a typical DC voltage bus. Your specific



system configuration may require additional considerations. Please note that the values of C<sub>IN</sub>, L<sub>BUS</sub> and C<sub>BUS</sub> will vary according to the specific converter model.

In critical applications, output ripple and noise (also referred to as periodic and random deviations or PARD) may be reduced by adding filter elements such as multiple external capacitors. Be sure to calculate component temperature rise from reflected AC current dissipated inside capacitor ESR. In figure 3, the two copper strips simulate real-world printed circuit impedances between the power supply and its load. In order to minimize circuit errors and standardize tests between units, scope measurements should be made using BNC connectors or the probe ground should not exceed one half inch and soldered directly to the fixture.

#### **Floating Outputs**

Since these are isolated DC/DC converters, their outputs are "floating" with respect to their input. The essential feature of such isolation is ideal ZERO CURRENT FLOW between input and output. Real-world converters however do exhibit tiny leakage currents between input and output (see Specifications).



### COPPER STRIP +OUTPUT SCOPE RLOAD C1 C2 -OUTPUT COPPER STRIP $C1 = 0.1 \mu F CERAMIC$ $C2 = 10\mu F LOW ES$ LOAD 2-3 INCHES (51-76mm) FROM MODULE Figure 3 – Measuring Output Ripple and Noise (PARD)

These leakages consist of both an AC stray capacitance coupling component and a DC leakage resistance. When using the isolation feature, do not allow the isolation voltage to exceed specifications. Otherwise the converter may be damaged. Designers will normally use the negative output (-Output) as the ground return of the load circuit. You can however use the positive output (+Output) as the ground return to effectively reverse the output polarity.

#### **Minimum Output Loading Requirements**

These converters employ a synchronous rectifier design topology. All models regulate within specification and are stable under no load to full load conditions. Operation under no load might however slightly increase output ripple and noise.

#### **Thermal Shutdown**

To protect against thermal over-stress, these converters include thermal shutdown circuitry. If environmental conditions cause the temperature of the DC/ DC's to rise above the Operating Temperature Range up to the shutdown temperature, an on-board electronic temperature sensor will power down the unit. When the temperature decreases below the turn-on threshold, the converter will automatically restart. There is a small amount of hysteresis to prevent rapid on/off cycling. CAUTION: If you operate too close to the thermal limits, the converter may shut down suddenly without warning. Be sure to thoroughly test your application to avoid unplanned thermal shutdown.

#### **Temperature Derating Curves**

The graphs in the next section illustrate typical operation under a variety of conditions. The Derating curves show the maximum continuous ambient air temperature and decreasing maximum output current which is acceptable under increasing forced airflow measured in Linear Feet per Minute ("LFM"). Note that these are AVERAGE measurements. The converter will accept brief increases in temperature and/or current or reduced airflow as long as the average is not exceeded.

Note that the temperatures are of the ambient airflow, not the converter itself which is obviously running at higher temperature than the outside air. Also note that "natural convection" is defined as very flow rates which are not using fan-forced airflow. Depending on the application, "natural convection" is usually about 30-65 LFM but is not equal to still air (0 LFM).

Murata Power Solutions makes Characterization measurements in a closed cycle wind tunnel with calibrated airflow. We use both thermocouples and an

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infrared camera system to observe thermal performance. As a practical matter, it is guite difficult to insert an anemometer to precisely measure airflow in most applications. Sometimes it is possible to estimate the effective airflow if you thoroughly understand the enclosure geometry, entry/exit orifice areas and the fan flowrate specifications.

CAUTION: If you exceed these Derating guidelines, the converter may have an unplanned Over Temperature shut down. Also, these graphs are all collected near Sea Level altitude. Be sure to reduce the derating for higher altitude.

#### **Output Overvoltage Protection (OVP)**

This converter monitors its output voltage for an over-voltage condition using an on-board electronic comparator. The signal is optically coupled to the primary side PWM controller. If the output exceeds OVP limits, the sensing circuit will power down the unit, and the output voltage will decrease. After a time-out period, the PWM will automatically attempt to restart, causing the output voltage to ramp up to its rated value. It is not necessary to power down and reset the converter for the this automatic OVP-recovery restart.

If the fault condition persists and the output voltage climbs to excessive levels, the OVP circuitry will initiate another shutdown cycle. This on/off cycling is referred to as "hiccup" mode.

#### **Output Fusing**

The converter is extensively protected against current, voltage and temperature extremes. However, your application circuit may need additional protection. In the extremely unlikely event of output circuit failure, excessive voltage could be applied to your circuit. Consider using an appropriate external protection.

#### **Output Current Limiting**

As soon as the output current increases to approximately its overcurrent limit, the DC/DC converter will enter a current-limiting mode. The output voltage will decrease proportionally with increases in output current, thereby maintaining a somewhat constant power output. This is commonly referred to as power limiting.

Current limiting inception is defined as the point at which full power falls below the rated tolerance. See the Performance/Functional Specifications. Note particularly that the output current may briefly rise above its rated value. This enhances reliability and continued operation of your application. If the output current is too high, the converter will enter the short circuit condition.

#### **Output Short Circuit Condition**

When a converter is in current-limit mode, the output voltage will drop as the output current demand increases. If the output voltage drops too low, the magnetically coupled voltage used to develop PWM bias voltage will also drop, thereby shutting down the PWM controller. Following a time-out period, the PWM will restart, causing the output voltage to begin rising to its appropriate value. If the short-circuit condition persists, another shutdown cycle will initiate. This on/off cycling is called "hiccup mode." The hiccup cycling reduces the average output current, thereby preventing excessive internal temperatures.

#### **Trimming the Output Voltage**

The Trim input to the converter allows the user to adjust the output voltage over the rated trim range (please refer to the Specifications). In the trim equations and circuit diagrams that follow, trim adjustments use either a trimpot or a single fixed resistor connected between the Trim input and either the +Sense or -Sense terminals. (On some converters, an external user-supplied precision DC voltage

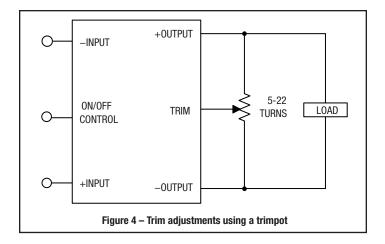


### may also be used for trimming). Trimming resistors should have a low temperature coefficient ( $\pm 100$ ppm/deg.C or less) and be mounted close to the converter. Keep leads short. If the trim function is not used, leave the trim unconnected. With no trim, the converter will exhibit its specified output voltage accuracy.

There are two CAUTIONs to observe for the Trim input:

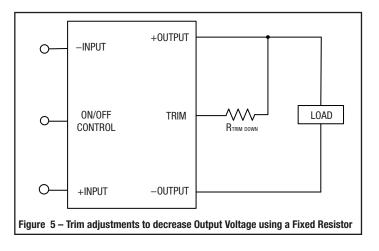
<u>CAUTION</u>: To avoid unplanned power down cycles, do not exceed EITHER the maximum output voltage OR the maximum output power when setting the trim. Be particularly careful with a trimpot. If the output voltage is excessive, the OVP circuit may inadvertantly shut down the converter. If the maximum power is exceeded, the converter may enter current limiting. If the power is exceeded for an extended period, the converter may overheat and encounter overtemperature shut down.

<u>CAUTION</u>: Be careful of external electrical noise. The Trim input is a sensitve input to the converter's feedback control loop. Excessive electrical noise may



cause instability or oscillation. Keep external connections short to the Trim input. Use shielding if needed.

#### **Trim Equations**



Trim Up <Connect trim resistor between Trim and –Sense> Trim Down <Connect trim resistor between Trim and +Sense>

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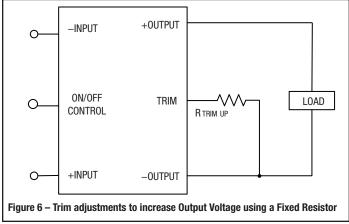
UEI30-033	UEI30-033-Q12/-Q48									
$R_{T_{UP}}(\Omega) = \frac{12775}{V_O - 3.3} - 2050$	$R_{T_{DOWN}}(\Omega) = \frac{5110 (Vo - 2.5)}{3.3 - Vo} - 2050$									
UEI30-050	-Q12/-Q48									
$R_{T_{UP}}(\Omega) = \frac{12775}{V_0 - 5} - 2050$	$R_{T_{DOWN}}(\Omega) = \frac{5110 (V_0 - 2.5)}{5 - V_0} - 2050$									
UEI30-120	-Q12/-Q48									
$R_{T_{UP}}(\Omega) = \frac{25000}{V_0 - 12} - 5110$	$R_{T_{\text{DOWN}}}(\Omega) = \frac{10000 \text{ (Vo-2.5)}}{12 - \text{Vo}} - 5110$									
UEI30-150	-Q12/-Q48									
$R_{T_{UP}}(\Omega) = \frac{25000}{V_0 - 15} - 5110$	$R_{T_{\text{DOWN}}}(\Omega) = \frac{10000 \text{ (Vo-2.5)}}{15 - \text{Vo}} - 5110$									

Where Vo = Desired output voltage. Adjustment accuracy is subject to resistor tolerances and factory-adjusted output accuracy. Mount trim resistor close to converter. Use short leads.

#### **Remote On/Off Control**

On the input side, a remote On/Off Control can be specified with either positive or negative logic as follows:

<u>Positive</u>: Models equipped with Positive Logic are enabled when the On/Off pin is left open or is pulled high to  $+V_{IN}$  with respect to  $-V_{IN}$ . An internal bias current causes the open pin to rise to  $+V_{IN}$ . Some models will also turn on at lower intermediate voltages (see Specifications). Positive-polarity devices are disabled when the On/Off is grounded or brought to within a low voltage (see Specifications) with respect to  $-V_{IN}$ .



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<u>Negative:</u> Models with negative polarity are on (enabled) when the On/Off is grounded or brought to within a low voltage (see Specifications) with respect to  $-V_{IN}$ . The device is off (disabled) when the On/Off is left open or is pulled high to  $+15V_{DC}$  Max. with respect to  $-V_{IN}$ .

Dynamic control of the On/Off function should be able to sink the specified signal current when brought low and withstand appropriate voltage when brought high. Be aware too that there is a finite time in milliseconds (see Specifications) between the time of On/Off Control activation and stable, regulated output. This time will vary slightly with output load type and current and input conditions.

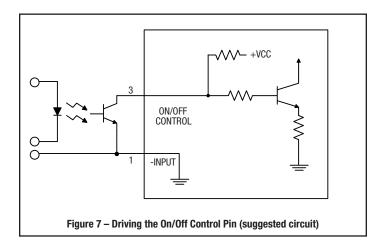
There are two CAUTIONs for the On/Off Control:

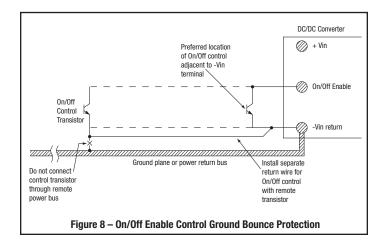
<u>CAUTION:</u> While it is possible to control the On/Off with external logic if you carefully observe the voltage levels, the preferred circuit is either an open drain/open collector transistor or a relay (which can thereupon be controlled by logic). The On/Off prefers to be set at approx. +15V (open pin) for the ON state, assuming positive logic.

<u>CAUTION</u>: Do not apply voltages to the On/Off pin when there is no input power voltage. Otherwise the converter may be permanently damaged.

#### **On/Off Enable Control Ground Bounce Protection**

To improve reliability, if you use a small signal transistor or other external circuit to select the Remote On/Off control, make sure to return the LO side directly to the –Vin power input on the DC/DC converter. To avoid ground bounce errors, do not connect the On/Off return to a distant ground plane or current-carrying bus. If necessary, run a separate small return wire directly to the –Vin terminal. There is very little current (typically 1-5 mA) on the On/Off control however, large current changes on a return ground plane or ground bus can accidentally trigger the converter on or off. If possible, mount the On/Off transistor or other control circuit adjacent to the converter.







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#### Mean Time Before Failure (MTBF) Table

These figures use a standard MTBF probability calculation as an indication of component parts stress and life derating. The calculaton is based on separate MTBF values for all internal parts in addition to stated environmental conditions. Two MTBF values are presented. The Telcordia method is widely used in industry, particularly telecom. The United States MIL-HDBK method is for military and industrial applications. Please refer to a qualified reliability engineer for more background.

Model Number	MTBF (Hours)	Method [1,2]
UEI30-033-Q12N-C	2,676,902	Telcordia
UEI30-033-Q12N-C	2,123,124	MIL-HDBK
UEI30-033-Q12P-C	2,733,781	Telcordia
UEI30-033-Q12P-C	2,142,206	MIL-HDBK
UEI30-033-Q48N-C	3,416,592	Telcordia
UEI30-033-Q48N-C	3,172,548	MIL-HDBK
UEI30-033-Q48P-C	3,427,027	Telcordia
UEI30-033-Q48P-C	3,193,652	MIL-HDBK
UEI30-050-Q12N-C	2,531,509	Telcordia
UEI30-050-Q12N-C	2,207,508	MIL-HDBK
UEI30-050-Q12P-C	2,554,127	Telcordia
UEI30-050-Q12P-C	2,229,031	MIL-HDBK
UEI30-120-Q48N-C	3,072,461	Telcordia
UEI30-120-Q48N-C	2,510,927	MIL-HDBK
UEI30-120-Q48P-C	2,900,319	Telcordia
UEI30-120-Q48P-C	2,495,846	MIL-HDBK
UEI30-150-Q48N-C	2,833,366	Telcordia
UEI30-150-Q48N-C	2,408,836	MIL-HDBK
UEI30-150-Q48P-C	2,776,615	Telcordia
UEI30-150-Q48P-C	2,421,938	MIL-HDBK

Notes:

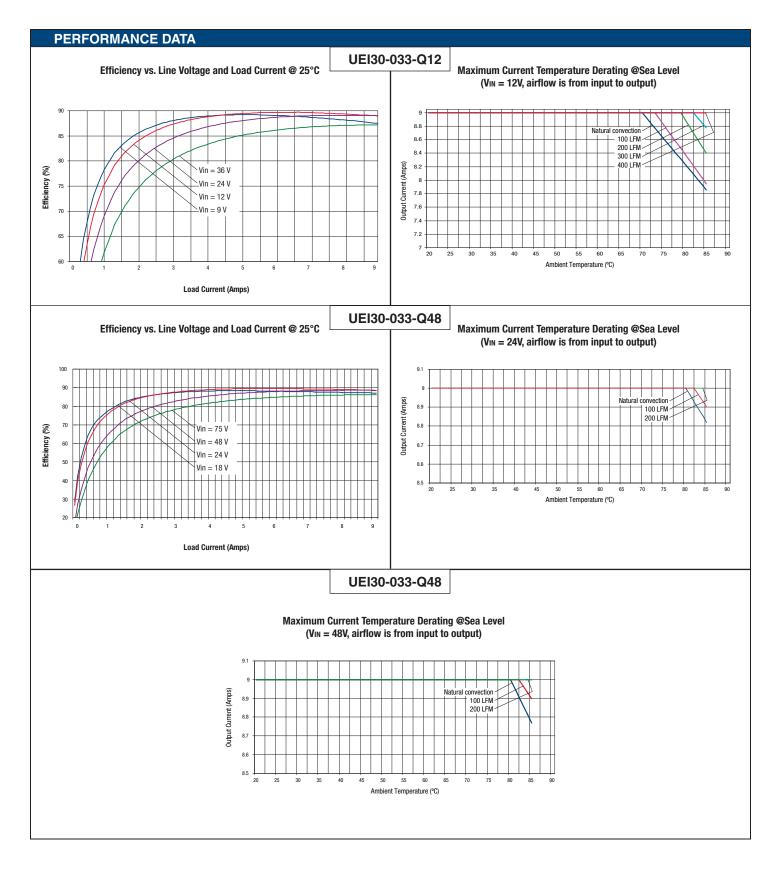
 Mean Time Before Failure is calculated using the Telcordia (Belcore) SR-332 Method 1, Case 3, ISSUE 2, ground fixed controlled conditions, Tambient=+25°C, full output load, natural air convection.

[2] Mean Time Before Failure is calculated using MIL-HDBK-217FN2, GB ground benign, Tambient=+25°C, full output load, natural air convection.





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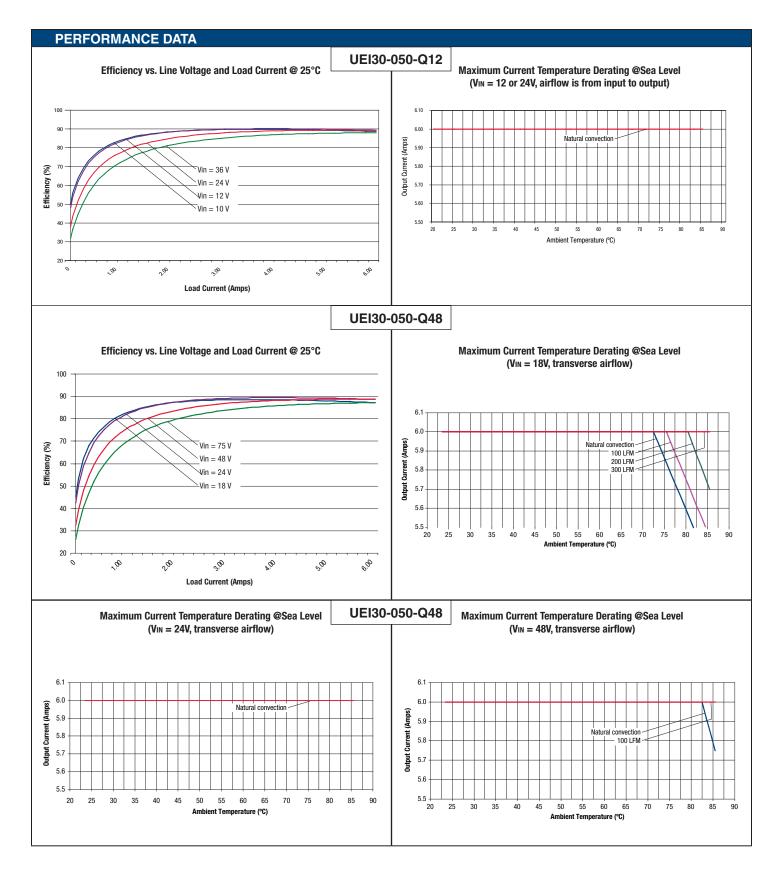




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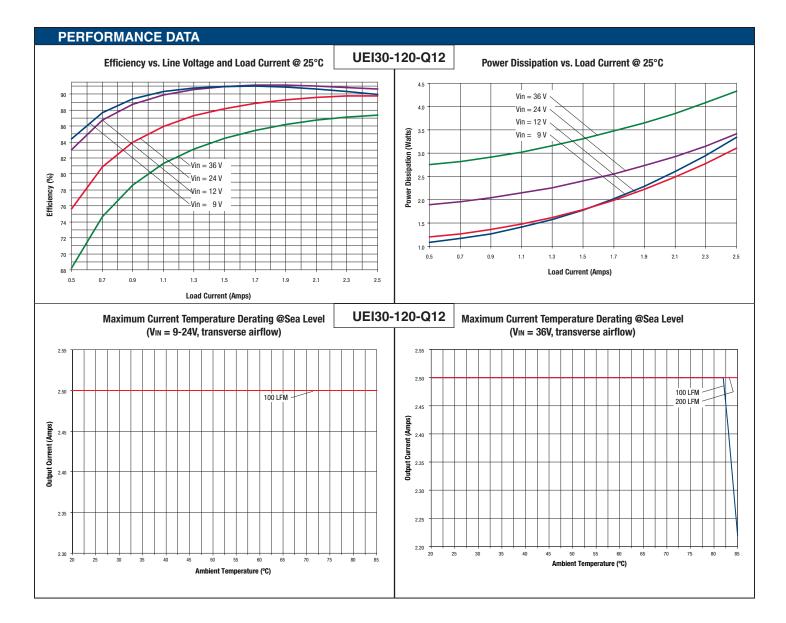
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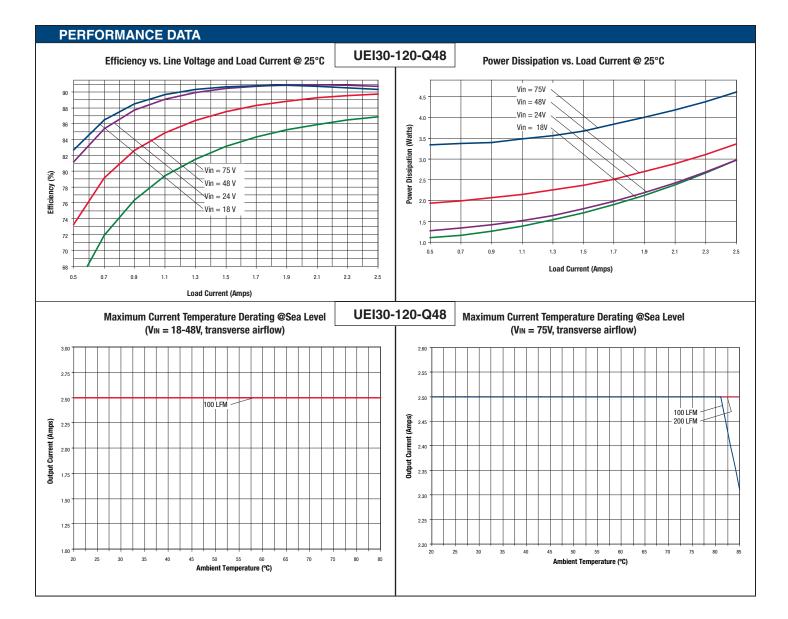




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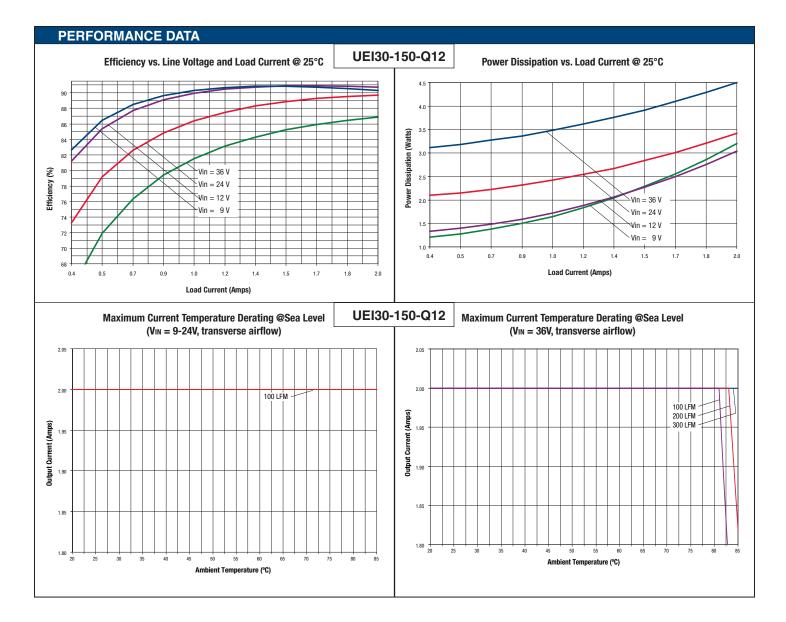




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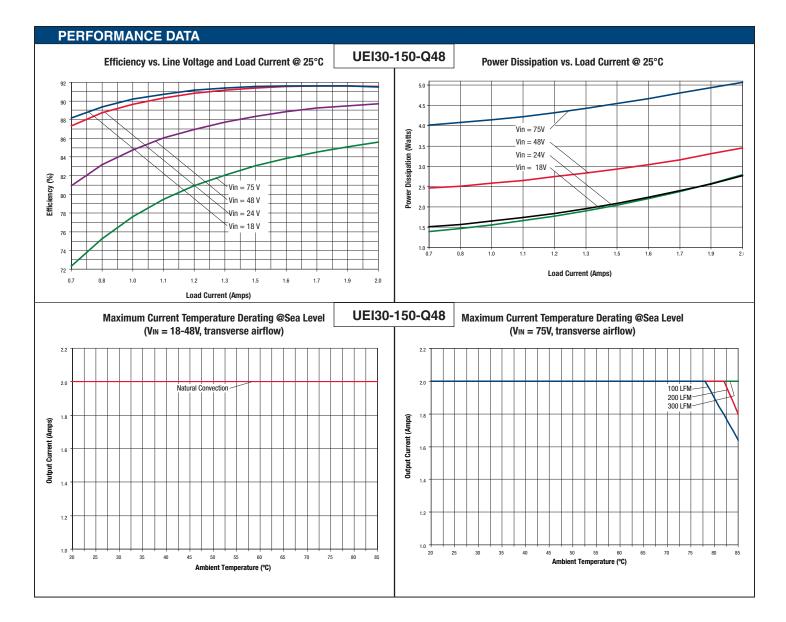




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