

ML2003, ML2004

Logarithmic Gain/Attenuator

Features

- Low noise: 0 dB_{rnc} max with +24dB gain
- Low harmonic distortion: -60dB max
- Gain range: -24 to +24dB
- Resolution: 0.1dB steps
- Flat frequency response:
 - ±0.05dB from .3–4 kHz
 - ±0.10dB from .1–20 kHz
- Low supply current 4mA max from ±5V supplies
- TTL/CMOS compatible digital interface
- ML2003 has pin selectable serial or parallel interface; ML2004 serial interface only

General Description

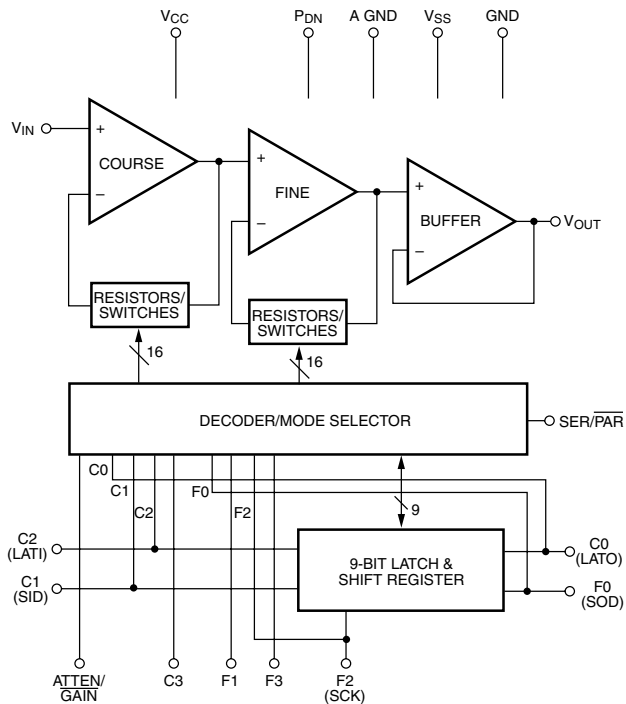
The ML2003 and ML2004 are digitally controlled logarithmic gain/attenuators with a range of -24 to +24 dB in 0.1 dB steps.

The gain settings are selected by a 9-bit digital word. The ML2003 digital interface is either parallel or serial. The ML2004 is packaged in a 14-pin DIP with a serial interface only.

Absolute gain accuracy is 0.05dB max over supply tolerance of ±10% and temperature range.

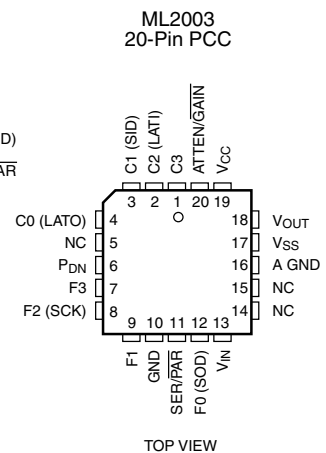
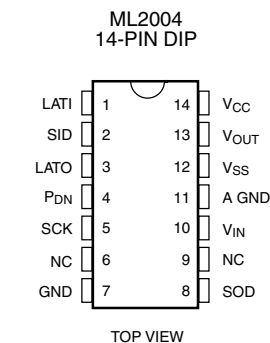
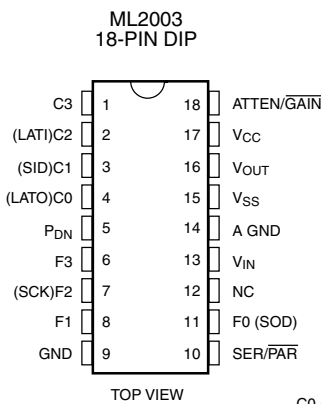
These CMOS logarithmic gain/attenuators are designed for a wide variety of applications in telecom, audio, sonar, or general purpose function generation. One specific intended application is analog telephone lines.

Block Diagram



NOTE: SERIAL MODE FUNCTIONS INDICATED BY PARENTHESES.

Pin Connections



Pin Description

Name	Function
C3	In serial mode, pin is unused. In parallel mode, coarse gain select bit. Pin has internal pulldown resistor to GND.
(LATI) C2	In serial mode, input latch clock which loads the data from the shift register into the latch. In parallel mode, coarse gain select bit. Pin has internal pulldown resistor to GND.
(SID) C1	In serial mode, serial data input that contains serial 9 bit data word which controls the gain setting. In parallel mode, coarse gain select bit. Pin has internal pulldown resistor to GND.
(LATO) C0	In serial mode, output latch clock which loads the 9 bit data word back into the shift register from the latch. In parallel mode, coarse gain select bit. Pin has internal pulldown resistor to GND.
PDN	Powerdown input. When PDN = 1, device is in powerdown mode. When PDN = 0, device is in normal operation. Pin has internal pulldown resistor to GND.
F3	In serial mode, pin is unused. In parallel mode, fine gain select bit. Pin has internal pulldown resistor to GND.
(SCK) F2	In serial mode, shift register clock which shifts the serial data on SID into the shift register on rising edges and out on SOD on falling edges. In parallel mode, fine gain select bit. Pin has internal pulldown resistor to GND.
F1	In serial mode, pin is unused. In parallel mode, fine gain select bit. Pin has internal pulldown resistor to GND.
GND	Digital ground. 0 volts. All digital inputs and outputs are referenced to this ground.
SER/PAR	Serial or parallel select input. When SER/PAR = 1, device is in serial mode. When SER/PAR = 0, device is in parallel mode. Pin has internal pullup resistor to VCC.
(SOD) F0	In serial mode, serial output data which is the output of the shift register. In parallel mode, fine gain select bit. Pin has internal pulldown resistor to GND.
VIN	Analog input.
AGND	Analog ground. 0 volts. Analog input and output are referenced to this ground.
VSS	Negative supply. -5 volts $\pm 10\%$.
VOUT	Analog output.
VCC	Positive supply. +5 volts $\pm 10\%$.
ATTEN/ $\overline{\text{GAIN}}$	In serial mode, pin is unused. In parallel mode, attenuation/gain select bit. Pin has internal pulldown resistor to GND.

Absolute Maximum Ratings¹

Parameter	Min.	Max.	Units
Supply Voltage			
VCC		+6.5	V
VSS		-6.5	V
AGND with respect to GND		± 0.5	V
Analog Input and Output	VSS -0.3V	VCC +0.3	V
Digital Input and Outputs	GND -0.3	VCC +0.3	V
Input Current Per Pin		± 25	mA
Power Dissipation		750	mW
Storage Temperature Range	-65	+150	°C
Lead Temperature (Soldering, 10 sec)		300	°C

Operating Conditions

Parameter	Min.	Max.	Units
Temperature Range ² ML2003CX, ML2004CX ML2003IX, ML2004IX	0 -40	70 85	°C °C
Supply Voltage VCC VSS	4 -4	6 -6	V V

Electrical Characteristics

Unless otherwise specified TA = TMIN to TMAX, VCC = 5V ± 10%, VSS = -5V ± 10%, Data Word: $\overline{\text{ATTEN}}/\overline{\text{GAIN}} = 1$, Other Bits = 0 (0dB Ideal Gain), CL = 100pF, RL = 600Ω, SCK = LATI = LATO = 0, dBm measurements use 600Ω as reference load, digital timing measured at 1.4 V, CL = 100pF or SOD.

Symbol	Parameter	Notes	Conditions	Min.	Typ. ³	Max.	Units
Analog							
AG	Absolute gain accuracy	4	VIN=8dBm, 1 kHz	-0.05		+0.05	dB
RG	Relative gain accuracy	4	100000001	-0.05		+0.05	dB
			000000000	-0.05		+0.05	dB
			000000001	-0.05		+0.05	dB
			All other gain settings	-0.1		+0.1	dB
			All values referenced to 100000000 gain when $\overline{\text{ATTEN}}/\overline{\text{GAIN}} = 1$, VIN = 8dBm when $\overline{\text{ATTEN}}/\overline{\text{GAIN}} = 0$ VIN = (8dBm – Ideal Gain) in dB				
FR	Frequency response	4	300-4000 Hz	-0.05		+0.05	dB
			100-20,000 Hz Relative to 1 kHz	-0.1		+0.1	dB
VOS	Output Offset Voltage	4	VIN = 0, +24dB gain			±100	mV
ICN	Idle Channel Noise	4	VIN = 0, +24dB gain, C msg. Weighted		-6	0	dBnc
		5	VIN = 0, +24dB gain, 1kHz	450	900	nv/√Hz	
HD	Harmonic Distortion	4	VIN = 8dBm gain, 1kHz Measure 2nd, 3rd harmonic relative to fundamental			-60	dB
SD	Signal to Distortion	4	VIN = 8dBm, 1 kHz C msg. weighted	+60			dB
PSRR	Power Supply Rejection	4	200mVp-p, 1 kHz sine, VIN = 0 on VCC on VSS		-60 -60	-40 -40	dB dB
ZIN	Input impedance, VIN	4		1			Meg
VINR	Input Voltage Range	4		±3.0			V
VOSW	Output Voltage Swing	4		±3.0			V
Digital and DC							
VIL	Digital Input Low Voltage	4				0.8	V
VIH	Digital Input High Voltage	4		2.0			V
VOL	Digital Output Low Voltage	4	IOL = 2mA			0.4	V

Electrical Characteristics (continued)

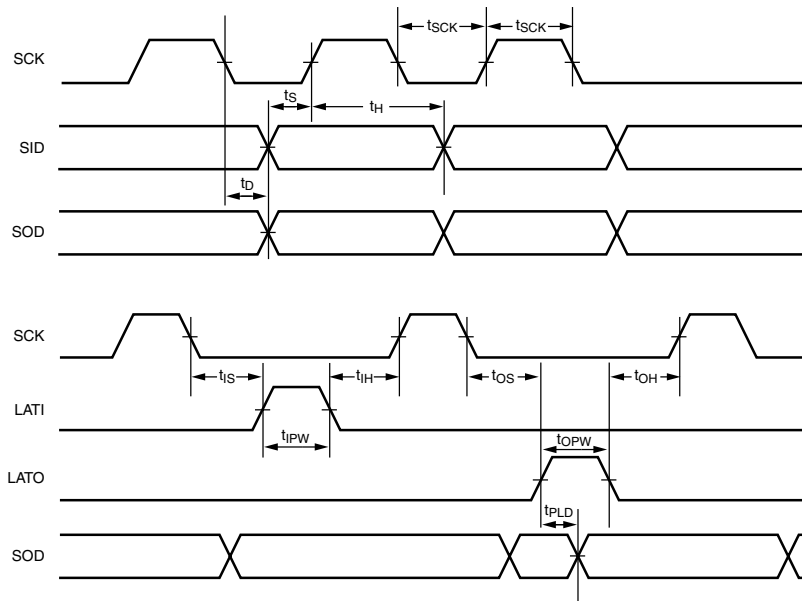
Unless otherwise specified $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = 5V \pm 10\%$, $V_{SS} = -5V \pm 10\%$, Data Word: $ATTEN/\overline{GAIN} = 1$, Other Bits = 0(0dB Ideal Gain), $C_L = 100pF$, $R_L = 600\Omega$, $SCK = LATI = LATO = 0$, dBm measurements use 600Ω as reference load, digital timing measured at 1.4 V, $C_L = 100pF$ or SOD.

Symbol	Parameter	Notes	Conditions	Min.	Typ. ³	Max.	Units
VOH	Digital Output High Voltage	4	$I_{OH} = -1mA$	4.0			V
INS	Input Current, SER/ PAR	4	$V_{IH} = GND$	-5		-100	μA
IND	Input Current, All Digital Inputs Except SER/PER	4	$V_{IH} = V_{CC}$	5		100	μA
ICC	VCC Supply Current	4	No output load, $V_{IL} = GND$, $V_{IH} = V_{CC}$, $V_{IN} = 0$			4	mA
ISS	VSS Supply Current	4	No output load, $V_{IL} = GND$, $V_{IH} = V_{CC}$, $V_{IN} = 0$			-4	mA
ICCP	VCC Supply Current, Powerdown Mode	4	No output load, $V_{IL} = GND$, $V_{IH} = V_{CC}$			0.5	mA
ISSP	VSS Supply Current Powerdown Mode	4	No output load, $V_{IL} = GND$, $V_{IH} = V_{CC}$			-0.1	mA
AC Characteristics							
tSET	VOUT Settling Time	4	$V_{IN} = 0.185V$. Change gain from -24 to $+24dB$. Measure from LATI rising edge to when VOUT settles to within 0.05dB of final value.			20	μs
tSTEP	VOUT Step Response	4	Gain = $+24dB$. $V_{IN} = -0.185$ to $+0.185V$ step. Measure when VOUT settles to within 0.05dB of final value.			20	μs
tSCK	SCK On/Off Period	4		250			ns
tS	SID Data Setup Time	4		50			ns
tH	SID Data Hold Time	4		50			ns
tD	SOD Data Delay	4		0		125	ns
tIPW	LATI Pulse Width	4		50			ns
tOPW	LATO Pulse Width	4		50			ns
tIS, tOS	LATI, LATO Setup Time	4		50			ns
tIH, tOH	LATI, LATO Hold Time	5		50			ns
tPLD	SOD Parallel Load Delay	4		0		125	ns

Notes:

1. Absolute maximum ratings are limits beyond which the life of the integrated circuit may be impaired. All voltages unless otherwise specified are measured with respect to ground.
2. $0^{\circ}C$ to $+70^{\circ}C$ and $-40^{\circ}C$ to $+85^{\circ}C$ operating temperature range devices are 100% tested with temperature limits guaranteed by 100% testing, sampling, or by correlation with worst-case test conditions.
3. Typicals are parametric norm at $25^{\circ}C$.
4. Parameter guaranteed and 100% production tested.
5. Parameter guaranteed. Parameters not 100% tested are not in outgoing quality level calculation.

Timing Diagram



TIMING PARAMETERS ARE REFERENCED TO THE 1.4 VOLT MIDPOINT.

Figure 1. Serial Mode Timing Diagram

Typical Performance Curves

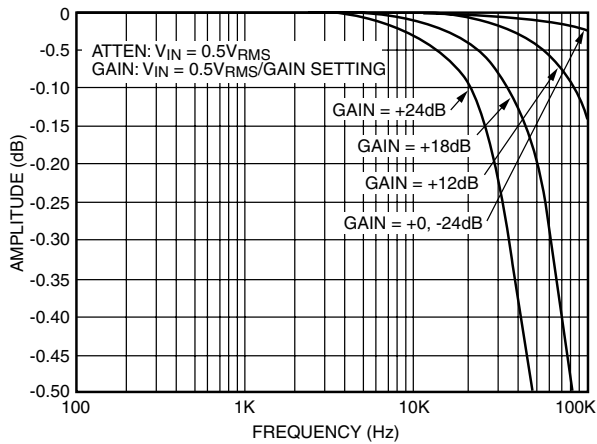


Figure 2. Amplitude vs Frequency ($V_{IN}/V_{OUT} = .5V_{RMS}$)

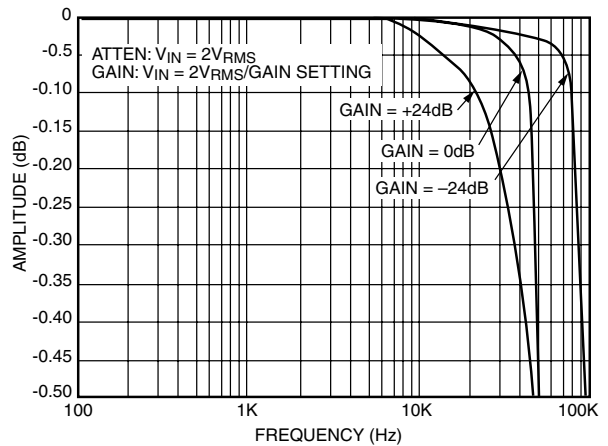


Figure 3. Amplitude vs Frequency ($V_{IN}/V_{OUT} = 2V_{RMS}$)

Typical Performance Curves (continued)

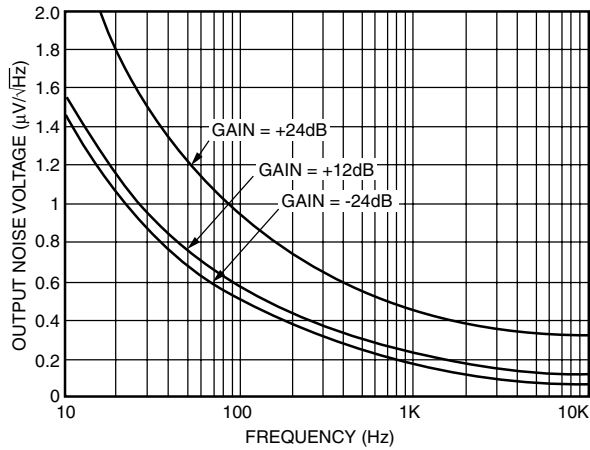


Figure 4. Output Noise Voltage vs Frequency

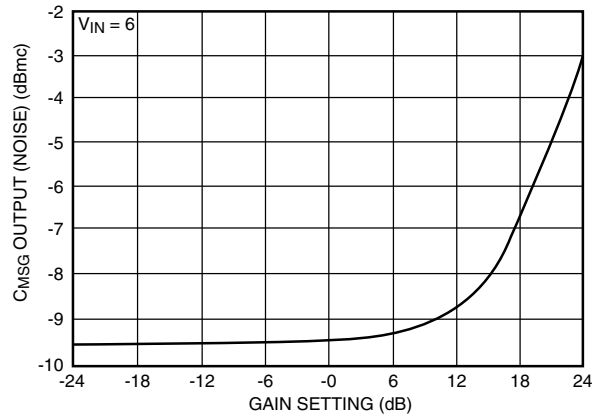


Figure 5. CMSG Output Noise vs Gain Settings

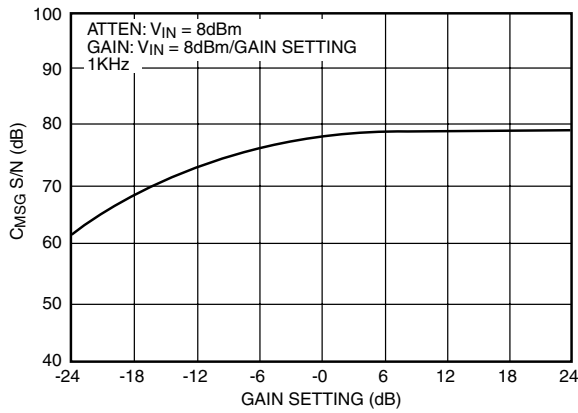


Figure 6. CMSG S/N vs Gain Setting

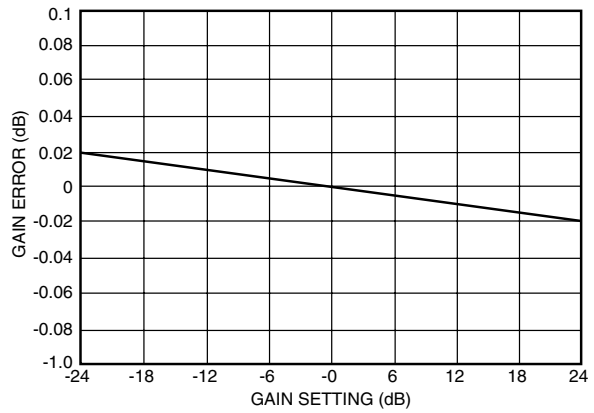


Figure 7. Gain Error vs Gain Setting

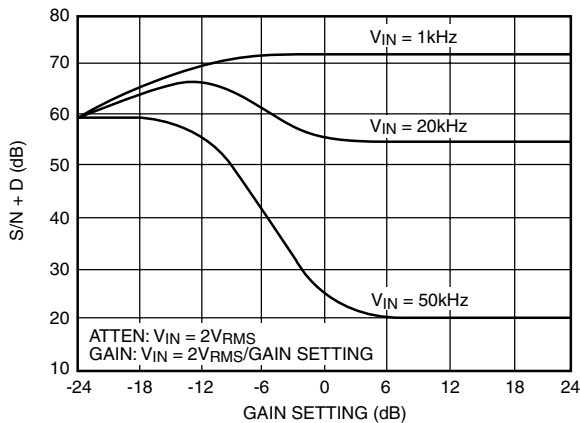


Figure 8. S/N + D vs Gain Setting ($V_{IN}/V_{OUT} = 2V_{RMS}$)

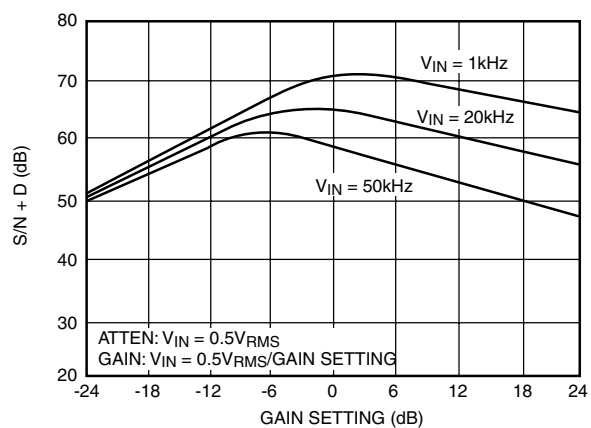


Figure 9. S/N + D vs Gain Setting ($V_{IN}/V_{OUT} = 0.5V_{RMS}$)

Functional Description

The ML2003 consists of a coarse gain stage, a fine gain stage, an output buffer, and a serial/parallel digital interface.

Gain Stages

The analog input, V_{IN} , goes directly into the op amp input in the coarse gain stage. The coarse gain stage has a gain range of 0 to 22.5dB in 1.5dB steps.

The fine gain stage is cascaded onto the coarse section. The fine gain stage has a gain range of 0 to 1.5dB in 0.1dB steps.

In addition, both sections can be programmed for either gain or attenuation, thus doubling the effective gain range.

The logarithmic steps in each gain stage are generated by placing the input signal across a resistor string of 16 series resistors. Analog switches allow the voltage to be tapped from the resistor string at 16 points. The resistors are sized such that each output voltage is at the proper logarithmic ratio relative to the input signal at the top of the string. Attenuation is implemented by using the resistor string as a simple voltage divider, and gain is implemented by using the resistor string as a feedback resistor around an internal op amp.

Gain Settings

Since the coarse and fine gain stages are cascaded, their gains can be summed logarithmically. Thus, any gain from -24dB to +24dB in 0.1dB steps can be obtained by combining the coarse and fine gain settings to yield the desired gain setting. The relationship between the digital select bits and the corresponding analog gain values is shown in Tables 1 and 2. Note that C3-C0 selects the coarse gain, F3-F0 selects the fine gain, and $\overline{ATTEN/GAIN}$ selects either attenuation or gain.

Output Buffer

The final analog stage is the output buffer. This amplifier has internal gain of 1 and is designed to drive 600 ohms and 100pF loads. Thus, it is suitable for driving a telephone hybrid circuit directly without any external amplifier.

Power Supplies

The digital section is powered between V_{CC} and GND , or 5 volts. The analog section is powered between V_{CC} and V_{SS} and uses $AGND$ as the reference point, or ± 5 volts.

GND and $AGND$ are totally isolated inside the device to minimize coupling from the digital section into the analog section. However, $AGND$ and GND should be tied together physically near the device and ideally close to the common power supply ground connection.

Typically, the power supply rejection of V_{CC} and V_{SS} to the analog output is greater than -60dB at 1 kHz. If decoupling of the power supplies is still necessary in a system, V_{CC} and V_{SS} should be decoupled with respect to $AGND$.

Powerdown Mode

A powerdown mode can be selected with pin P_{DN} . When $P_{DN} = 1$, the device is powered down. In this state, the power consumption is reduced by removing power from the analog section and forcing the analog output, V_{OUT} , to a high impedance state. While the device is in powerdown mode, the digital section is still functional and the current data word remains stored in the latch when in serial mode. When $P_{DN} = 0$, the device is in normal operation.

Digital Section

The ML2003 can be operated with a serial or parallel interface. The SER/\overline{PAR} pin selects the desired interface. When $SER/\overline{PAR} = 1$, the serial mode is selected. When $SER/\overline{PAR} = 0$, the parallel mode is selected. The ML2004 digital interface is serial only.

Serial Mode

Serial mode is selected by setting SER/\overline{PAR} pin high. The serial interface allows the gain settings to be set from a serial data word.

The timing for the serial mode is shown in Figure 10. The serial input data, SID , is loaded into a shift register on rising edges of the shift clock, SCK . The data can be parallel loaded into a latch when the input latch signal, $LATI$, is high. The $LATI$ pulse must occur when SCK is low. In this way, a new data word can be loaded into the shift register without disturbing the existing data word in the latch.

The parallel outputs of the latch control the attenuation/gain setting. The order of the data word bits in the latch is shown in Figure 11. Note that bit 0 is the first bit of the data word clocked into the shift register. Tables 1 and 2 describe how the data word programs the gain.

Table 1. Fine Gain Settings (C3-C0 = 0)

F3	F2	F1	F0	Ideal Gain (dB)	
				$\overline{ATTEN/GAIN} = 1$	$\overline{ATTEN/GAIN} = 0$
0	0	0	0	0	0
0	0	0	1	-.1	.1
0	0	1	0	-.2	.2
0	0	1	1	-.3	.3
0	1	0	0	-.4	.4
0	1	0	1	-.5	.5
0	1	1	0	-.6	.6
0	1	1	1	-.7	.7
1	0	0	0	-.8	.8
1	0	0	1	-.9	.9
1	0	1	0	-1.0	1.0
1	0	1	1	-1.1	1.1
1	1	0	0	-1.2	1.2
1	1	0	1	-1.3	1.3
1	1	1	0	-1.4	1.4
1	1	1	1	-1.5	1.5

Table 2. Coarse Gain Settings (F3-F0 = 0)

				Ideal Gain (dB)	
C3	C2	C1	C0	ATTEN/ $\overline{\text{GAIN}}$ = 1	ATTEN/ $\overline{\text{GAIN}}$ = 0
0	0	0	0	0	0
0	0	0	1	-1.5	1.5
0	0	1	0	-3.0	3.0
0	0	1	1	-4.5	4.5
0	1	0	0	-6.0	6.0
0	1	0	1	-7.5	7.5
0	1	1	0	-9.0	9.0
0	1	1	1	-10.5	10.5
1	0	0	0	-12.0	12.0
1	0	0	1	-13.5	13.5
1	0	1	0	-15.0	15.0
1	0	1	1	-16.5	16.5
1	1	0	0	-18.0	18.0
1	1	0	1	-19.5	19.5
1	1	1	0	-21.0	21.0
1	1	1	1	-22.5	22.5

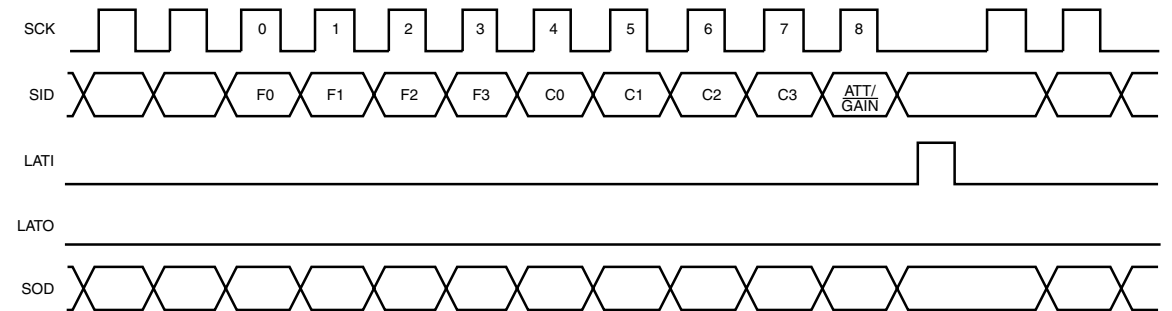
The device also has the capability to read out the data word stored in the latch. This can be done by parallel loading the data from the latch back into the shift register when the latch signal, LATI, is high. The LATO pulse must occur when SCK is low. Then, the data word can be shifted out of the shift register serially to the output, SOD, on falling edges of the shift clock, SCK.

The loading and reading of the data word can be done continuously or in burst. Since the shift register and latch circuitry inside the device is static, there are no minimum frequency requirements on the clocks or data pulses. However, there is coupling (typically less than 100 μ V) of the digital signals into the analog section. This coupling can be minimized by clocking the data bursts in during noncritical intervals or at a frequency outside the analog frequency range.

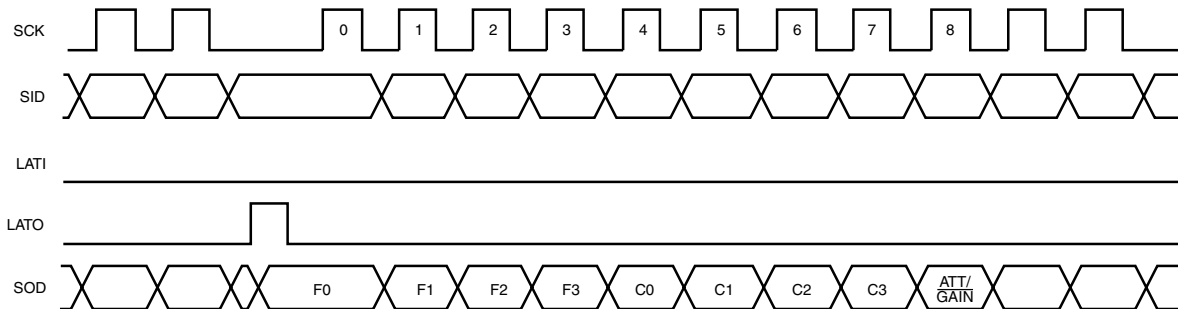
Parallel Mode

The parallel mode is selected by setting SER/ $\overline{\text{PAR}}$ pin low. The parallel interface allows the gain settings to be set with external switches or from a parallel microprocessor interface.

In parallel mode, the shift register and latch are bypassed and connections are made directly to the gain select bits with external pins ATTEN/ $\overline{\text{GAIN}}$, C3-C0, and F3-F0. Tables 1 and 2 describe how these pins program the gain. The pins ATTEN/ $\overline{\text{GAIN}}$, C3-C0, and F3-F0 have internal pulldown resistors to GND. The typical value of these pulldown resistors is 100k Ω .



a) LOAD



b) READ

Figure 10. Serial Mode Timing

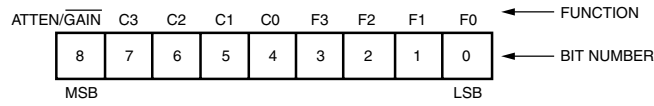


Figure 11. 9-Bit Latch

Applications

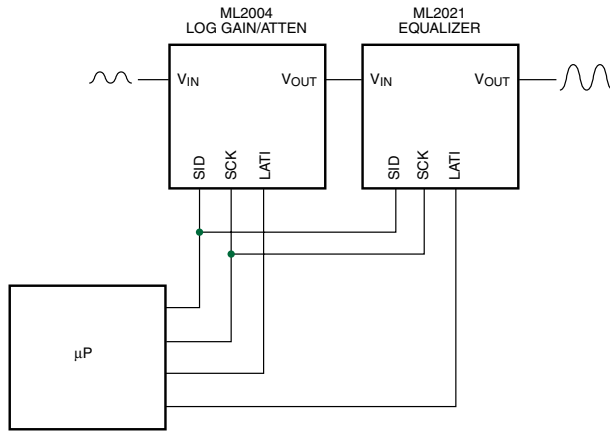


Figure 12. Typical Serial Interface

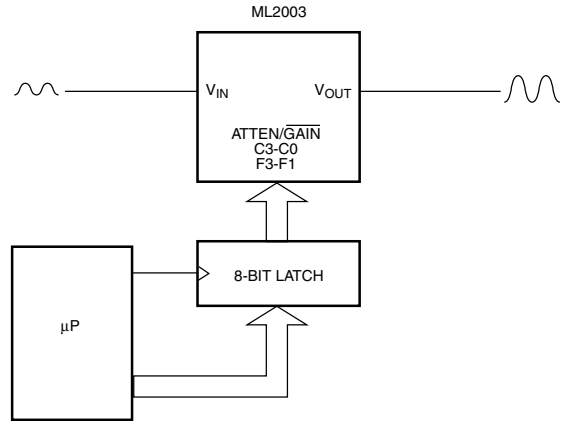


Figure 13. Typical μP Parallel Interface

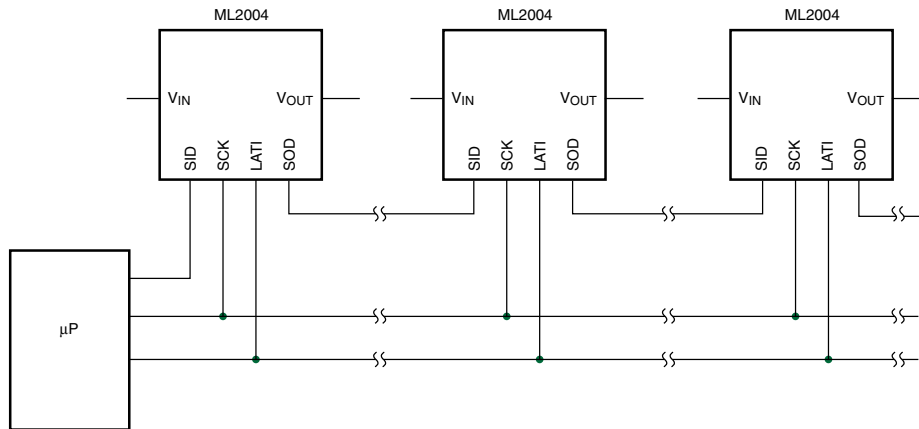


Figure 14. Controlling Multiple ML2004 with Only 3 Digital Lines Using One Long Data Word

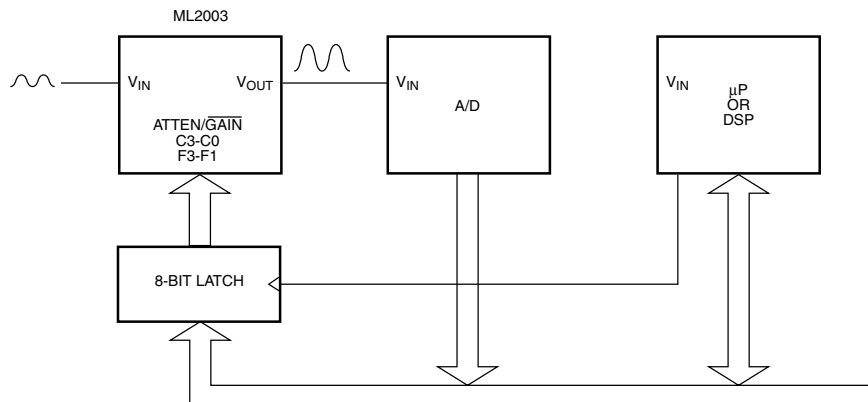


Figure 15. AGC for DSP or Modem Front End

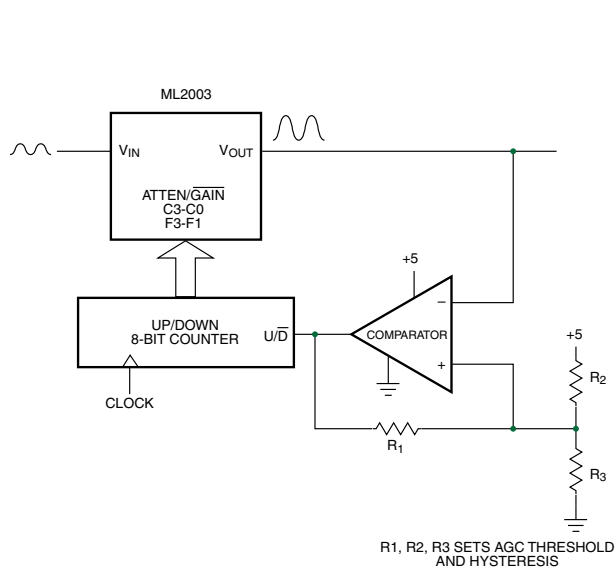


Figure 16. Analog AGC

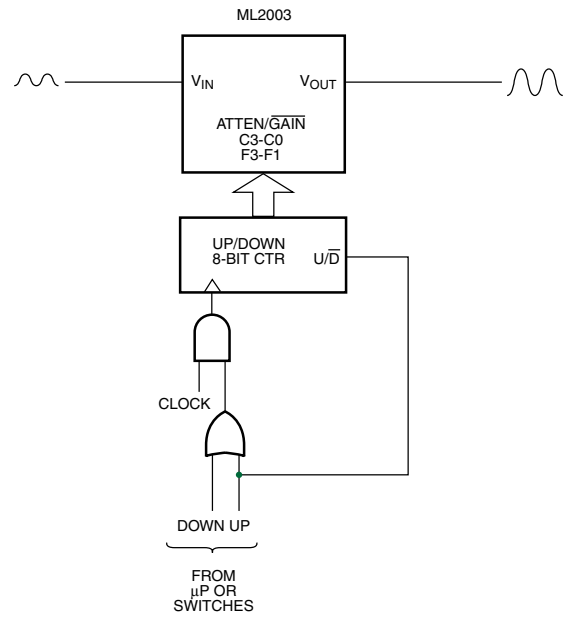


Figure 17. Digitally Controlled Volume Control

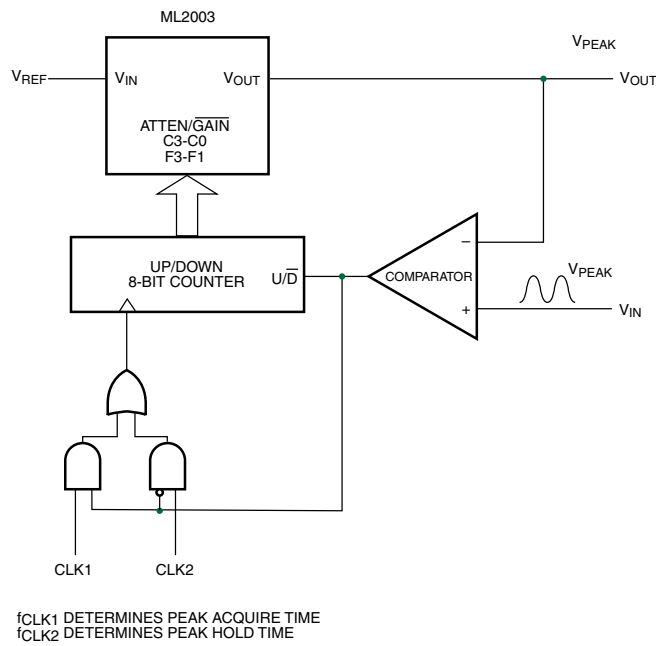


Figure 18. Precision Peak Detector ($\pm 1\%$) with Controllable Acquire and Hold Time

Ordering Information

Part Number	Temperature Range	Package
ML2003IQ	-40°C to 85°C	Molded PCC (Q20)
ML2003CP	0°C to 70°C	Molded DIP (P18)
ML2003CQ	0°C to 70°C	Molded PCC (Q20)
ML2004IP	-40°C to 85°C	Molded DIP (P14)
ML2004CP	0°C to 70°C	Molded DIP (P14)

DISCLAIMER

FAIRCHILD SEMICONDUCTOR RESERVES THE RIGHT TO MAKE CHANGES WITHOUT FURTHER NOTICE TO ANY PRODUCTS HEREIN TO IMPROVE RELIABILITY, FUNCTION OR DESIGN. FAIRCHILD DOES NOT ASSUME ANY LIABILITY ARISING OUT OF THE APPLICATION OR USE OF ANY PRODUCT OR CIRCUIT DESCRIBED HEREIN; NEITHER DOES IT CONVEY ANY LICENSE UNDER ITS PATENT RIGHTS, NOR THE RIGHTS OF OTHERS.

LIFE SUPPORT POLICY

FAIRCHILD'S PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS WITHOUT THE EXPRESS WRITTEN APPROVAL OF THE PRESIDENT OF FAIRCHILD SEMICONDUCTOR CORPORATION. As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.
2. A critical component in any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.