

Single 12-/10-/8-Bit Rail-to- Rail DACs with 10ppm/°C Reference in SC70

FEATURES

- Integrated Precision Reference
 2.5V Full Scale 10ppm/°C (LTC2630-L)
 4.096V Full Scale 10ppm/°C (LTC2630-H)
- Maximum INL Error: 1 LSB (LTC2630A-12)
- Low Noise: 0.7mV_{P-P}, 0.1Hz to 200kHz
- Guaranteed Monotonic over Temperature
- Selectable Internal Reference or Supply as Reference
- 2.7V to 5.5V Supply Range (LTC2630-L)
- Low Power Operation: 180µA at 3V
- Power Down to 1.8µA Maximum (C and I Grades)
- Power-on Reset to Zero or Mid-Scale Options
- SPI Serial Interface
- Double-Buffered Data Latches
- Tiny 6-Lead SC70 Package

APPLICATIONS

- Mobile Communications
- Process Control and Industrial Automation
- Automatic Test Equipment
- Portable Equipment
- Automotive

DESCRIPTION

The LTC®2630 is a family of 12-, 10-, and 8-bit voltageoutput DACs with an integrated, high-accuracy, low-drift reference in a 6-lead SC70 package. It has a rail-to-rail output buffer and is guaranteed monotonic.

The LTC2630-L has a full-scale output of 2.5V, and operates from a single 2.7V to 5.5V supply. The LTC2630-H has a full-scale output of 4.096V, and operates from a 4.5V to 5.5V supply. Each DAC can also operate in supply as reference mode, which sets the full-scale output to the supply voltage.

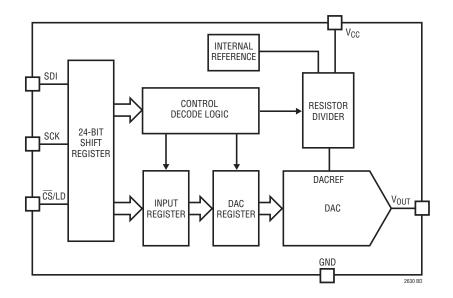
The parts use a simple SPI/MICROWIRE™ compatible 3-wire serial interface which operates at clock rates up to 50MHz.

The LTC2630 incorporates a power-on reset circuit. Options are available for reset to zero or reset to mid-scale after power-up.

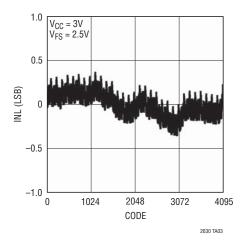
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BLOCK DIAGRAM



Integral Nonlinearity (LTC2630A-LZ12)



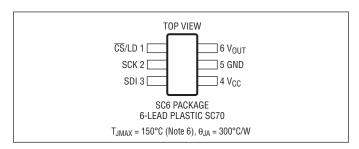


ABSOLUTE MAXIMUM RATINGS

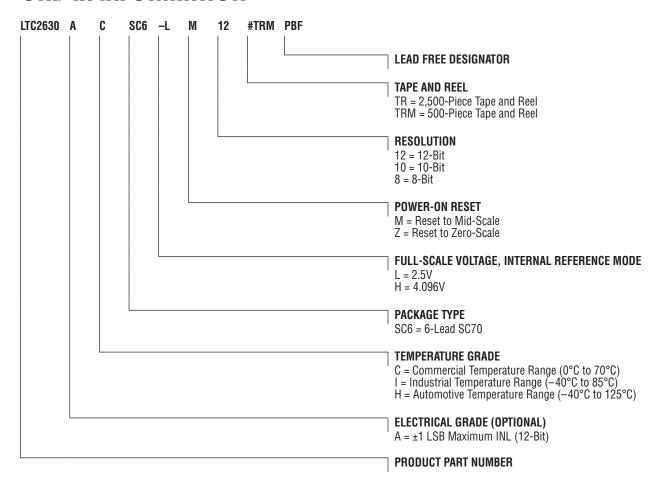
(Notes 1, 2)

(
Supply Voltage (V _{CC})	0.3V to 6V
CS/LD, SCK, SDI	0.3V to 6V
V _{OUT} 0.3V to m	$\sin(V_{CC} + 0.3V, 6V)$
Operating Temperature Range	
LTC2630C	0°C to 70°C
LTC26301	40°C to 85°C
LTC2630H (Note 3)	40°C to 125°C
Maximum Junction Temperature	150°C
Storage Temperature Range	65°C to 150°C
Lead Temperature (Soldering, 10 sec)	300°C

PIN CONFIGURATION



ORDER INFORMATION



Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/ For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

LINEAR

PRODUCT SELECTION GUIDE

PART NUMBER	PART MARKING*	V _{FS} WITH INTERNAL REFERENCE	POWER-ON RESET TO CODE	RESOLUTION	V _{CC}	MAXIMUM INL
LTC2630A-LM12	LCZB	2.5V • (4095/4096)	Mid-Scale	12-Bit	2.7V-5.5V	±1LSB
LTC2630A-LZ12	LCSB	2.5V • (4095/4096)	Zero	12-Bit	2.7V-5.5V	±1LSB
LTC2630A-HM12	LCWR	4.096V • (4095/4096)	Mid-Scale	12-Bit	4.5V-5.5V	±1LSB
LTC2630A-HZ12	LCZC	4.096V • (4095/4096)	Zero	12-Bit	4.5V-5.5V	±1LSB
LTC2630-LM12	LCZB	2.5V • (4095/4096)	Mid-Scale	12-Bit	2.7V-5.5V	±2LSB
LTC2630-LM10	LCZF	2.5V • (1023/1024)	Mid-Scale	10-Bit	2.7V-5.5V	±1LSB
LTC2630-LM8	LCYW	2.5V • (255/256)	Mid-Scale	8-Bit	2.7V-5.5V	±0.5LSB
LTC2630-LZ12	LCSB	2.5V • (4095/4096)	Zero	12-Bit	2.7V-5.5V	±2LSB
LTC2630-LZ10	LCZD	2.5V • (1023/1024)	Zero	10-Bit	2.7V-5.5V	±1LSB
LTC2630-LZ8	LCYV	2.5V • (255/256)	Zero	8-Bit	2.7V-5.5V	±0.5LSB
LTC2630-HM12	LCWR	4.096V • (4095/4096)	Mid-Scale	12-Bit	4.5V-5.5V	±2LSB
LTC2630-HM10	LCZH	4.096V • (1023/1024)	Mid-Scale	10-Bit	4.5V-5.5V	±1LSB
LTC2630-HM8	LCYY	4.096V • (255/256)	Mid-Scale	8-Bit	4.5V-5.5V	±0.5LSB
LTC2630-HZ12	LCZC	4.096V • (4095/4096)	Zero	12-Bit	4.5V-5.5V	±2LSB
LTC2630-HZ10	LCZG	4.096V • (1023/1024)	Zero	10-Bit	4.5V-5.5V	±1LSB
LTC2630-HZ8	LCYX	4.096V • (255/256)	Zero	8-Bit	4.5V-5.5V	±0.5LSB

^{*}The temperature grade is identified by a label on the shipping container.

ELECTRICAL CHARACTERISTICS The ullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{CC} = 2.7V$ to 5.5V, V_{OUT} unloaded unless otherwise specified.

 $\label{eq:ltc2630-lm12-lm10-lm8-ltc2630a-lm12-ltc2630a-ltc263$

				LT(C2630)-8	LT	C2630	1-10	LT	C2630)-12	LTC2630A-12			
SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
DC Perfo	rmance															
	Resolution		•	8			10			12			12			Bits
	Monotonicity	V _{CC} = 3V, Internal Ref. (Note 4)	•	8			10			12			12			Bits
DNL	Differential Nonlinearity	V _{CC} = 3V, Internal Ref. (Note 4)	•			±0.5			±0.5			±1			±1	LSB
INL	Integral Nonlinearity	V _{CC} = 3V, Internal Ref. (Note 4)	•	±	0.05	±0.5		±0.2	±1		±1	±2		±0.5	±1	LSB
ZSE	Zero Scale Error	V _{CC} = 3V, Internal Ref., Code = 0	•		0.5	5		0.5	5		0.5	5		0.5	5	mV
$\overline{V_{OS}}$	Offset Error	V _{CC} = 3V, Internal Ref. (Note 5)	•	=	±0.5	±5		±0.5	±5		±0.5	±5		±0.5	±5	mV
V _{OSTC}	V _{OS} Temperature Coefficient	V _{CC} = 3V, Internal Ref. (Note 5)		±	:10			±10			±10			±10		μV/°C
FSE	Full Scale Error	V _{CC} = 3V, Internal Ref.	•	±	0.2	±0.8		±0.2	±0.8		±0.2	±0.8		±0.2	±0.8	%FSR
V _{FSTC}	Full Scale Voltage Temperature Coefficient	V _{CC} = 3V, Internal Ref. (Note 10) C-Grade I-Grade H-Grade		±	:10 :10 :10			±10 ±10 ±10			±10 ±10 ±10			±10 ±10 ±10		ppm/°C ppm/°C ppm/°C
	Load Regulation	$ \begin{array}{l} \text{Internal Ref., Mid-Scale,} \\ V_{CC} = 3V \pm 10\%, -5\text{mA} \leq I_{OUT} \leq 5\text{mA} \\ V_{CC} = 5V \pm 10\%, -10\text{mA} \leq I_{OUT} \leq 10\text{mA} \end{array} $	•	-		0.016 0.016			0.064 0.064			0.256 0.256			0.256 0.256	
R _{OUT}		$\label{eq:local_local} \begin{split} &\text{Internal Ref., Mid-Scale,} \\ &\text{$V_{CC} = 3V \pm 10\%, -5\text{mA} \le I_{OUT} \le 5\text{mA}$} \\ &\text{$V_{CC} = 5V \pm 10\%, -10\text{mA} \le I_{OUT} \le 10\text{mA}$} \end{split}$		-		0.156 0.156			0.156 0.156			0.156 0.156			0.156 0.156	Ω

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OUT}	DAC Output Span	Supply as Reference Internal Reference			0V to V _{CC} 0V to 2.5		V
PSR	Power Supply Rejection	V _{CC} = 3V ±10% or 5V ±10%			-80		dB
I _{SC}	Short Circuit Output Current (Note 6) Sinking Sourcing	V _{FS} = V _{CC} = 5.5V Zero Scale; V _{OUT} Shorted to V _{CC} Full Scale; V _{OUT} Shorted to GND	•		27 -28	50 -50	mA mA
Power S	upply	·					
V_{CC}	Power Supply Voltage	For Specified Performance	•	2.7		5.5	V
I _{CC}	Supply Current (Note 7)	V _{CC} = 3V, Supply as Reference V _{CC} = 3V, Internal Reference V _{CC} = 5V, Supply as Reference V _{CC} = 5V, Internal Reference	•		160 180 180 190	220 240 250 260	Ац Ац Ац Ац
I _{SD}	Supply Current in Power-Down Mode (Note 7)	V _{CC} = 5V, C-Grade, I-Grade V _{CC} = 5V, H-Grade	•		0.36 0.36	1.8 5	μA μA
Digital I	0	·					
V _{IH}	Digital Input High Voltage	V _{CC} = 3.6V to 5.5V V _{CC} = 2.7V to 3.6V	•	2.4 2.0			V
V_{IL}	Digital Input Low Voltage	V _{CC} = 4.5V to 5.5V V _{CC} = 2.7V to 4.5V	•			0.8 0.6	V
I _{LK}	Digital Input Leakage	V _{IN} = GND to V _{CC}	•			±1	μА
C _{IN}	Digital Input Capacitance	(Note 8)	•			2.5	pF



ELECTRICAL CHARACTERISTICS The ullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{CC} = 2.7V$ to 5.5V, V_{OUT} unloaded unless otherwise specified.

 $LTC2630-LM12/-LM10/-LM8/-LZ12/-LZ10/-LZ8, \quad LTC2630A-LM12/-LZ12 \; (V_{FS}=2.5V)$

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
AC Perfor	mance					
t _S	Settling Time	V _{CC} = 3V (Note 9) ±0.39% (±1LSB at 8 Bits) ±0.098% (±1LSB at 10 Bits) ±0.024% (±1LSB at 12 Bits)		3.2 3.9 4.4		µs µs µs
	Voltage Output Slew Rate			1.0		V/µs
	Capacitive Load Driving			500		pF
	Glitch Impulse	At Mid-Scale Transition		2		nV∙s
e _n	Output Voltage Noise Density	At f = 1kHz, Supply as Reference At f = 10kHz, Supply as Reference At f = 1kHz, Internal Reference At f = 10kHz, Internal Reference		140 130 160 150		$\begin{array}{c c} nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \end{array}$
	Output Voltage Noise	0.1Hz to 10Hz, Supply as Reference 0.1Hz to 10Hz, Internal Reference 0.1Hz to 200kHz, Supply as Reference 0.1Hz to 200kHz, Internal Reference		20 20 650 700		μV _{P-P} μV _{P-P} μV _{P-P}

TIMING CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25\,^{\circ}\text{C}$. $V_{CC} = 2.7V$ to 5.5V. (See Figure 1) (Note 8).

 $LTC2630-LM12/-LM10/-LM8/-LZ12/-LZ10/-LZ8, \quad LTC2630A-LM12/-LZ12 \; (V_{FS}=2.5V)$

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
t ₁	SDI Valid to SCK Setup		•	4			ns
t ₂	SDI Valid to SCK Hold		•	4			ns
t ₃	SCK High Time		•	9			ns
t ₄	SCK Low Time		•	9			ns
t ₅	CS/LD Pulse width		•	10			ns
t ₆	SCK High to CS/LD High		•	7			ns
t ₇	CS/LD Low to SCK High		•	7			ns
t ₁₀	CS/LD High to SCK Positive Edge		•	7			ns
	SCK Frequency	50% Duty Cycle	•			50	MHz

ELECTRICAL CHARACTERISTICS The ullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}C$. $V_{CC} = 4.5V$ to 5.5V, V_{OUT} unloaded unless otherwise specified.

 $\label{eq:ltc2630-hm12-hm10-hm8-hz12-hz10-hz8} \text{LTC2630A-hm12-hz12 (V}_{FS} = 4.096\text{V})$

				LTC	2630)-8	LT	C2630	-10	LT	C2630	-12	LT	C2630/	A-12	
SYMBOL	PARAMETER	CONDITIONS		MIN T	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	UNITS
DC Perfo	rmance															
	Resolution		•	8			10			12			12			Bits
	Monotonicity	V _{CC} = 5V, Internal Ref. (Note 4)	•	8			10			12			12			Bits
DNL	Differential Nonlinearity	V _{CC} = 5V, Internal Ref. (Note 4)	•			±0.5			±0.5			±1			±1	LSB
INL	Integral Nonlinearity	V _{CC} = 5V, Internal Ref. (Note 4)	•	±(0.05	±0.5		±0.2	±1		±1	±2		±0.5	±1	LSB
ZSE	Zero Scale Error	V _{CC} = 5V, Internal Ref., Code = 0	•	(0.5	5		0.5	5		0.5	5		0.5	5	mV
$\overline{V_{0S}}$	Offset Error	V _{CC} = 5V, Internal Ref. (Note 5)	•	±	±0.5	±5		±0.5	±5		±0.5	±5		±0.5	±5	mV
V _{OSTC}	V _{OS} Temperature Coefficient	V _{CC} = 5V, Internal Ref. (Note 5)		±	±10			±10			±10			±10		mV/°C
FSE	Full Scale Error	V _{CC} = 5V, Internal Ref.	•	±(0.2	±0.8		±0.2	±0.8		±0.2	±0.8		±0.2	±0.8	%FSR
V _{FSTC}	Full Scale Voltage Temperature Coefficient	V _{CC} = 5V, Internal Ref. (Note 10) C-Grade I-Grade H-Grade		±.	10 10 10			±10 ±10 ±10			±10 ±10 ±10			±10 ±10 ±10		ppm/°C ppm/°C ppm/°C
	Load Regulation	V_{CC} = 5V ±10%, Internal Ref., Mid-Scale, -10mA \leq I _{OUT} \leq 10mA	•	0.	.006	0.01		0.025	0.04		0.10	0.16		0.10	0.16	LSB/ mA
R _{OUT}	DC Output Impedance	V_{CC} = 5V ±10%, Internal Ref., Mid-Scale, -10mA \leq I _{OUT} \leq 10mA	•	0.	.1	0.156		0.1	0.156		0.1	0.156		0.1	0.156	Ω

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V _{OUT}	DAC Output Span	Supply as Reference Internal Reference			0V to V _{CC} 0V to 4.096		V
PSR	Power Supply Rejection	V _{CC} = 5V ±10%			-80		dB
I _{SC}	Short Circuit Output Current (Note 6) Sinking Sourcing	V _{FS} = V _{CC} = 5.5V Zero Scale; V _{OUT} Shorted to V _{CC} Full Scale; V _{OUT} Shorted to GND	•		27 –28	50 –50	mA mA
Power S	upply	·					
V _{CC}	Power Supply Voltage	For Specified Performance	•	4.5		5.5	V
I _{CC}	Supply Current (Note 7)	V _{CC} = 5V, Supply as Reference V _{CC} = 5V, Internal Reference	•		180 200	260 280	μA μA
I _{SD}	Supply Current in Power-Down Mode (Note 7)	V _{CC} = 5V, C-Grade, I-Grade V _{CC} = 5V, H-Grade	•		0.36 0.36	1.8 5	μA μA
Digital I/	0	·					
V _{IH}	Digital Input High Voltage		•	2.4			V
V_{IL}	Digital Input Low Voltage		•			0.8	V
I _{LK}	Digital Input Leakage	V _{IN} = GND to V _{CC}	•			±1	μА
C _{IN}	Digital Input Capacitance	(Note 8)	•			2.5	pF

ELECTRICAL CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25 \,^{\circ}\text{C}$. $V_{CC} = 4.5 \,^{\circ}\text{V}$ to $5.5 \,^{\circ}\text{V}$, V_{OUT} unloaded unless otherwise specified.

LTC2630-HM12/-HM10/-HM8/-HZ12/-HZ10/-HZ8, LTC2630A-HM12/-HZ12 ($V_{FS} = 4.096V$)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
AC Perfo	rmance		, , , , , , , , , , , , , , , , , , ,				
t _S	Settling Time	V _{CC} = 5V (Note 9) ±0.39% (±1LSB at 8 Bits) ±0.098% (±1LSB at 10 Bits) ±0.024% (±1LSB at 12 Bits)			3.7 4.4 4.8		μs μs μs
	Voltage Output Slew Rate				1.0		V/µs
	Capacitive Load Driving				500		pF
	Glitch Impulse	At Mid-Scale Transition			2.4		nV∙s
e _n	Output Voltage Noise Density	At f = 1kHz, Supply as Reference At f = 10kHz, Supply as Reference At f = 1kHz, Internal Reference At f = 10kHz, Internal Reference			140 130 210 200		$\begin{array}{c} nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \\ nV/\sqrt{\underline{Hz}} \end{array}$
	Output Voltage Noise	0.1Hz to 10Hz, Supply as Reference 0.1Hz to 10Hz, Internal Reference 0.1Hz to 200kHz, Supply as Reference 0.1Hz to 200kHz, Internal Reference			20 20 650 750		μV _{P-P} μV _{P-P} μV _{P-P} μV _{P-P}

TIMING CHARACTERISTICS The \bullet denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}$ C. $V_{CC} = 4.5$ V to 5.5V. (See Figure 1) (Note 8).

LTC2630-HM12/-HM10/-HM8/-HZ12/-HZ10/-HZ8, LTC2630A-HM12/-HZ12 ($V_{FS} = 4.096V$)

SYMBOL	PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
t ₁	SDI Valid to SCK Setup		•	4			ns
t ₂	SDI Valid to SCK Hold		•	4			ns
t ₃	SCK High Time		•	9			ns
t ₄	SCK Low Time		•	9			ns
t ₅	CS/LD Pulse width		•	10			ns
t ₆	SCK High to CS/LD High		•	7			ns
t ₇	CS/LD Low to SCK High		•	7			ns
t ₁₀	CS/LD High to SCK Positive Edge		•	7			ns
	SCK Frequency	50% Duty Cycle	•			50	MHz

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: All voltages are with respect to GND.

Note 3: High temperatures degrade operating lifetimes. Operating lifetime is derated at temperatures greater than 105°C.

Note 4: Linearity and monotonicity are defined from code k_L to code 2^N-1 , where N is the resolution and k_L is given by $k_L = 0.016 \bullet (2^N/V_{FS})$, rounded to the nearest whole code. For $V_{FS} = 2.5V$ and N = 12, $k_L = 26$ and linearity is defined from code 26 to code 4,095. For $V_{FS} = 4.096V$ and N = 12, $k_L = 16$ and linearity is defined from code 16 to code 4,095.

Note 5: Inferred from measurement at code 16 (LTC2630-12), code 4 (LTC2630-10) or code 1 (LTC2630-8).

Note 6: This IC includes current limiting that is intended to protect the device during momentary overload conditions. Junction temperature can exceed the rated maximum during current limiting. Continuous operation above the specified maximum operating junction temperature may impair device reliability.

Note 7: Digital inputs at 0V or V_{CC} .

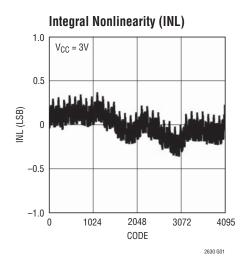
Note 8: Guaranteed by design and not production tested.

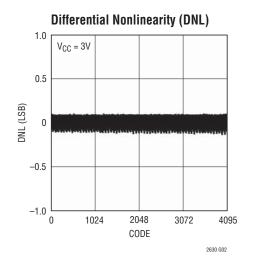
Note 9: Internal Reference mode. DAC is stepped 1/4 scale to 3/4 scale and 3/4 scale to 1/4 scale. Load is $2k\Omega$ in parallel with 100pF to GND.

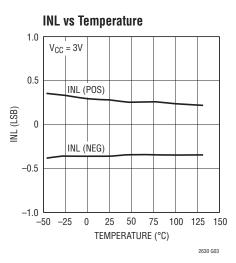
Note 10: Temperature coefficient is calculated by dividing the maximum change in output voltage by the specified temperature range.

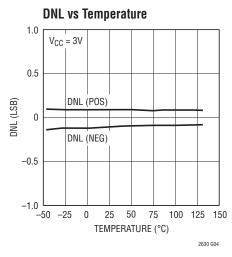


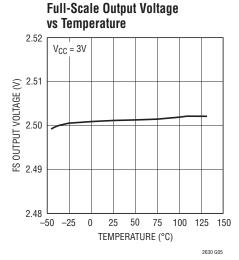
 $LTC2630-LM12/-LZ12 (V_{FS} = 2.5V)$

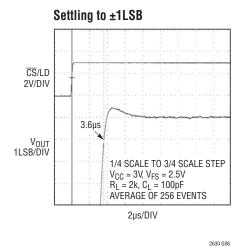


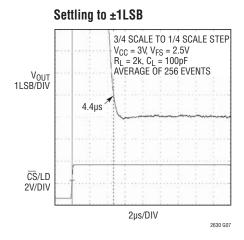




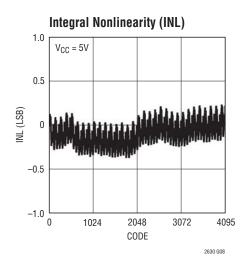


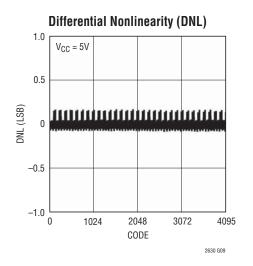


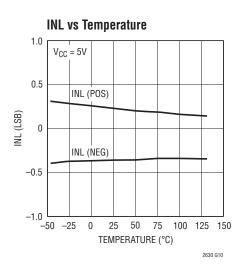


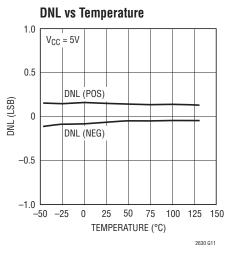


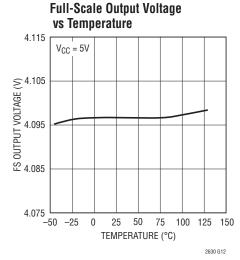
LTC2630-HM12/-HZ12 ($V_{FS} = 4.096V$)

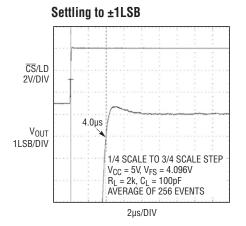


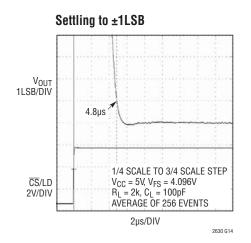






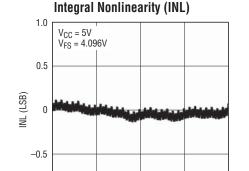




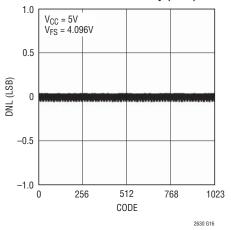


2630 G13

LTC2630-10



Differential Nonlinearity (DNL)



LTC2630-8

-1.0

0

Integral Nonlinearity (INL)

512

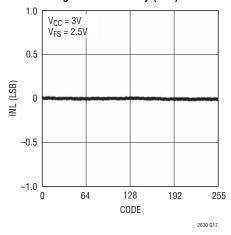
CODE

768

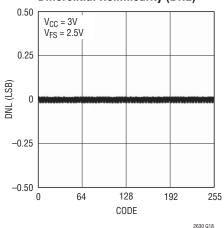
1023

2630 G15

256

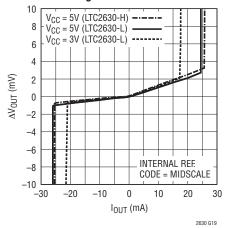


Differential Nonlinearity (DNL)

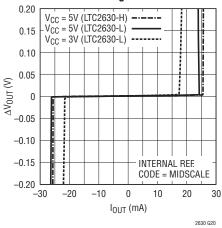


LTC2630

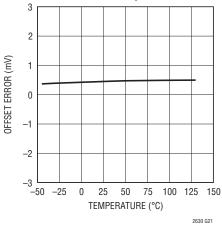
Load Regulation







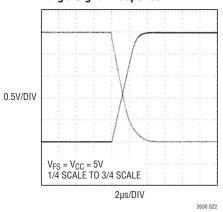
Offset Error vs Temperature



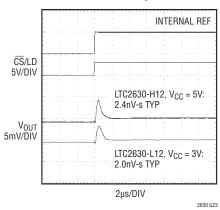


LTC2630

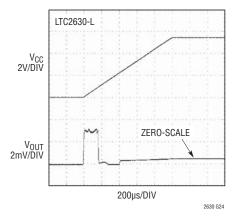
Large-Signal Response



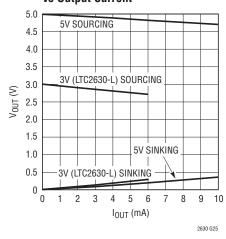
Mid-Scale-Glitch Impulse



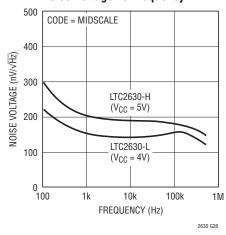
Power-On Reset Glitch



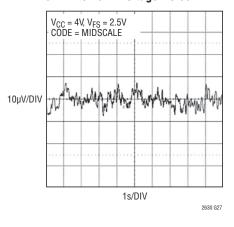
Headroom at Rails vs Output Current



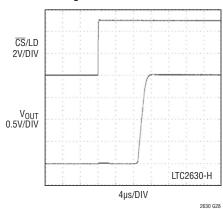
Noise Voltage vs Frequency



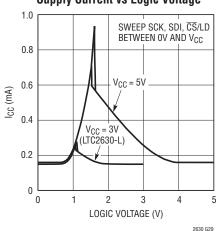
0.1Hz to 10Hz Voltage Noise



Exiting Power-Down to Mid-Scale



Supply Current vs Logic Voltage



PIN FUNCTIONS

CS/LD (**Pin 1**): Serial Interface Chip Select/Load Input. When \overline{CS}/LD is low, SCK is enabled for shifting data on SDI into the register. When \overline{CS}/LD is taken high, SCK is disabled and the specified command (see Table 1) is executed.

SCK (Pin 2): Serial Interface Clock Input. CMOS and TTL compatible.

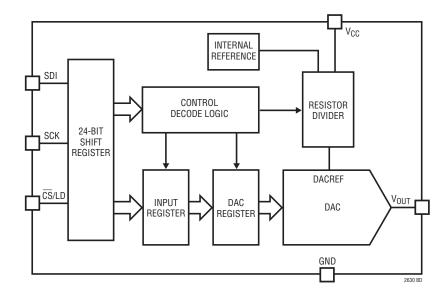
SDI (Pin 3): Serial Interface Data Input. Data on SDI is clocked into the DAC on the rising edge of SCK. The LTC2630 accepts input word lengths of either 24 or 32 bits.

V_{CC} (**Pin 4**): Supply Voltage Input. $2.7V \le V_{CC} \le 5.5V$ (LTC2630-L) or $4.5V \le V_{CC} \le 5.5V$ (LTC2630-H). Also used as the reference input when the part is programmed to operate in supply as reference mode. Bypass to GND with a $0.1\mu F$ capacitor.

GND (Pin 5): Ground.

V_{OUT} (Pin 6): DAC Analog Voltage Output.

BLOCK DIAGRAM



TIMING DIAGRAM

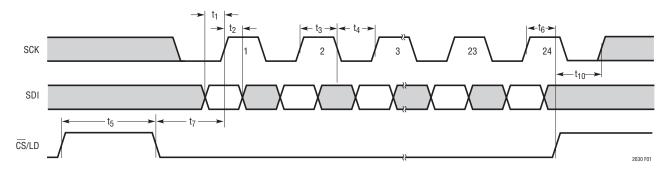


Figure 1. Serial Interface Timing

OPERATION

The LTC2630 is a family of single voltage output DACs in 6-lead SC70 packages. Each DAC can operate rail-to-rail referenced to the input supply, or with its full-scale voltage set by an integrated reference. Twelve combinations of accuracy (12-, 10-, and 8-bit), power-on reset value (zero or mid-scale), and full-scale voltage (2.5V or 4.096V) are available. The LTC2630 is controlled using a 3-wire SPI/MICROWIRE compatible interface.

Power-On Reset

The LTC2630-HZ/-LZ clear the output to zero scale when power is first applied, making system initialization consistent and repeatable.

For some applications, downstream circuits are active during DAC power-up, and may be sensitive to nonzero outputs from the DAC during this time. The LTC2630 contains circuitry to reduce the power-on glitch: the analog output typically rises less than 5mV above zero scale during power on if the power supply is ramped to 5V in 1ms or more. In general, the glitch amplitude decreases as the power supply ramp time is increased. See "Power-On Reset Glitch" in the Typical Performance Characteristics section.

The LTC2630-HM/-LM provide an alternative reset, setting the output to mid-scale when power is first applied.

Transfer Function

The digital-to-analog transfer function is

$$V_{OUT(IDEAL)} = \left(\frac{k}{2^N}\right) V_{REF}$$

where k is the decimal equivalent of the binary DAC input code, N is the resolution, and V_{REF} is either 2.5V (LTC2630-L) or 4.096V (LTC2630-H) in internal reference mode, and V_{CC} in Supply as reference mode.

Table 1. Command Codes

Con	ıman	d*		
C3	C2	C1	CO	
0	0	0	0	Write to Input Register
0	0	0	1	Update (Power up) DAC Register
0	0	1	1	Write to and Update (Power up) DAC Register
0	1	0	0	Power down
0	1	1	0	Select Internal Reference (Power-on Reset Default)
0	1	1	1	Select Supply as Reference (V _{REF} = V _{CC})

^{*}Command codes not shown are reserved and should not be used.

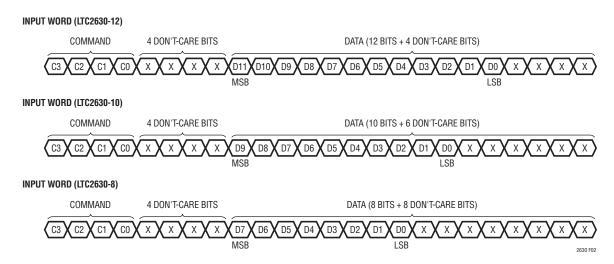


Figure 2. Command and Data Input Format

Serial Interface

The CS/LD input is level triggered. When this input is taken low, it acts as a chip-select signal, enabling the SDI and SCK buffers and the input shift register. Data (SDI input) is transferred at the next 24 rising SCK edges. The 4-bit command, C3-C0, is loaded first; then 4 don't-care bits; and finally the 16-bit data word. The data word comprises the 12-, 10- or 8-bit input code, ordered MSB-to-LSB, followed by 4, 6 or 8 don't-care bits (LTC2630-12, -10 and -8 respectively; see Figure 2). Data can only be transferred to the device when the $\overline{\text{CS}}/\text{LD}$ signal is low, beginning on the first rising edge of $\overline{\text{CS}}/\text{LD}$. The rising edge of $\overline{\text{CS}}/\text{LD}$ ends the data transfer and causes the device to execute the command specified in the 24-bit input sequence. The complete sequence is shown in Figure 3a.

The command (C3-C0) assignments are shown in Table 1. The first three commands in the table consist of write and update operations. A Write operation loads a 16-bit data word from the 24-bit shift register into the input register. In an Update operation, the input register is copied to the DAC register and converted to an analog voltage at the DAC output. Write to and Update combines the first two commands. The Update operation also powers up the DAC if it had been in power-down mode. The data path and registers are shown in the Block Diagram.

While the minimum input sequence is 24-bits, it may optionally be extended to 32-bits to accommodate microprocessors that have a minimum word width of 16-bits (2 bytes). To use the 32-bit width, 8 don't-care bits are transferred to the device first, followed by the 24-bit sequence described. Figure 3b shows the 32-bit sequence.

The 16-bit data word is ignored for all commands that do not include a Write operation.

Power-Down Mode

For power-constrained applications, power-down mode can be used to reduce the supply current whenever the DAC output is not needed. When in power-down, the buffer amplifier, bias circuit, and reference circuit are disabled and draw essentially zero current. The DAC output is put into a high-impedance state, and the output pin is passively pulled to ground through a $200 \mathrm{k}\Omega$ resistor. Input and DAC register contents are not disturbed during power-down.

The DAC can be put into power-down mode by using command 0100. The supply current is reduced to 1.8μA maximum when the DAC is powered down.

Normal operation resumes after executing any command that includes a DAC update, as shown in Table 1. The DAC is powered up and its voltage output is updated. Normal settling is delayed while the bias, reference, and amplifier circuits are re-enabled. The power-up delay time is 18µs for settling to 12-bits.

Reference Modes

For applications where an accurate external reference is not available, the LTC2630 has a user-selectable, integrated reference. The LTC2630-LM and LTC2630-LZ provide a full-scale output of 2.5V. The LTC2630-HM and LTC2630-HZ provide a full-scale output of 4.096V.

The internal reference can be useful in applications where the supply voltage is poorly regulated. Internal Reference mode can be selected by using command 0110, and is the power-on default.

The DAC can also operate in supply as reference mode using command 0111. In this mode, V_{CC} supplies the DAC's reference voltage and the supply current is reduced.

Voltage Output

The LTC2630's integrated rail-to-rail amplifier has guaranteed load regulation when sourcing or sinking up to 10mA at 5V, and 5mA at 3V.

Load regulation is a measure of the amplifier's ability to maintain the rated voltage accuracy over a wide range of load current. The measured change in output voltage per change in forced load current is expressed in LSB/mA.

DC output impedance is equivalent to load regulation, and may be derived from it by simply calculating a change in units from LSB/mA to ohms. The amplifier's DC output impedance is 0.1Ω when driving a load well away from the rails.

When drawing a load current from either rail, the output voltage headroom with respect to that rail is limited by the 50Ω typical channel resistance of the output devices (e.g., when sinking 1mA, the minimum output voltage is $50\Omega \cdot 1$ mA, or 50mV). See the graph "Headroom at Rails vs. Output Current" in the Typical Performance Characteristics section.

The amplifier is stable driving capacitive loads of up to 500pF.



Rail-to-Rail Output Considerations

In any rail-to-rail voltage output device, the output is limited to voltages within the supply range.

Since the analog output of the DAC cannot go below ground, it may limit for the lowest codes as shown in Figure 4b. Similarly, limiting can occur near full scale when using the supply as reference. If $V_{FS} = V_{CC}$ and the DAC full-scale error (FSE) is positive, the output for the highest codes limits at V_{CC} , as shown in Figure 4. No full-scale limiting can occur if V_{FS} is less than V_{CC} –FSE.

Offset and linearity are defined and tested over the region of the DAC transfer function where no output limiting can occur.

Board Layout

The PC board should have separate areas for the analog and digital sections of the circuit. A single, solid ground plane should be used, with analog and digital signals carefully routed over separate areas of the plane. This keeps digital signals away from sensitive analog signals and minimizes the interaction between digital ground currents and the analog section of the ground plane. The resistance from

the LTC2630 GND pin to the ground plane should be as low as possible. Resistance here will add directly to the effective DC output impedance of the device (typically 0.1Ω). Note that the LTC2630 is no more susceptible to this effect than any other parts of this type; on the contrary, it allows layout-based performance improvements to shine rather than limiting attainable performance with excessive internal resistance.

Another technique for minimizing errors is to use a separate power ground return trace on another board layer. The trace should run between the point where the power supply is connected to the board and the DAC ground pin. Thus the DAC ground pin becomes the common point for analog ground, digital ground, and power ground. When the LTC2630 is sinking large currents, this current flows out the ground pin and directly to the power ground trace without affecting the analog ground plane voltage.

It is sometimes necessary to interrupt the ground plane to confine digital ground currents to the digital portion of the plane. When doing this, make the gap in the plane only as long as it needs to be to serve its purpose and ensure that no traces cross over the gap.

Optoisolated 4mA to 20mA Process Controller

Figure 5 shows how to use an LTC2630HZ to make an optoisolated, digitally-controlled 4mA to 20mA transmitter. The transmitter circuitry, including optoisolation, is powered by the loop voltage which has a wide range of 5.4V to 80V. The 5V output of the LT $^{\circ}$ 3010-5 is used to set the 4mA offset current and V_{OUT} is used to digitally control the 0mA to 16mA signal current. The supply current for the regulator, DAC, and op amp is well below the 4mA budget at zero scale. R_S senses the total loop current, which includes the quiescent supply current and

additional current through Q1. Note that at the maximum loop voltage of 80V, Q1 will dissipate 1.6W when I_{OUT} = 20mA and must have an appropriate heat sink.

 R_{OFFSET} and R_{GAIN} are the closest 0.1% values to ideal for controlling a 4mA to 20mA output as the digital input varies from zero scale to full scale. Alternatively, R_{OFFSET} can be a 365k, 1% resistor in series with a 20k trim pot and R_{GAIN} can be a 75.0k, 1% resistor in series with a 5k trim pot. The optoisolators shown will limit the speed of the serial bus; the 6N139 is an alternative that will allow higher data rates.

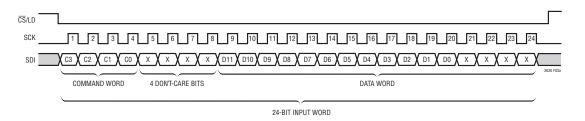


Figure 3a. 24-Bit Load Sequence (Minimum Input Word)
LTC2630-12 SDI Data Word: 12-Bit Input Code + 4 Don't-Care Bits (Shown);
LTC2630-10 SDI Data Word: 10-Bit Input Code + 6 Don't-Care Bits;

LTC2630-10 SDI Data Word: 10-Bit Input Code + 6 Don't-Care Bit LTC2630-8 SDI Data Word: 8-Bit Input Code + 8 Don't-Care Bits

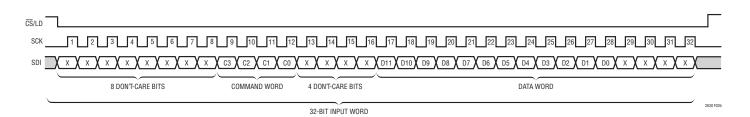


Figure 3b. 32-Bit Load Sequence

LTC2630-12 SDI Data Word: 12-Bit Input Code + 4 Don't-Care Bits (Shown);

LTC2630-10 SDI Data Word: 10-Bit Input Code + 6 Don't-Care Bits; LTC2630-8 SDI Data Word: 8-Bit Input Code + 8 Don't-Care Bits



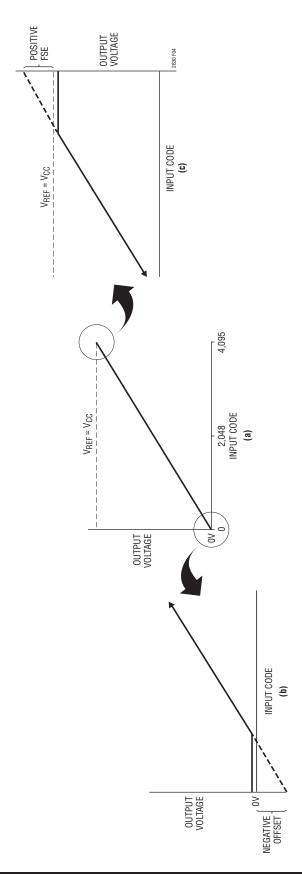
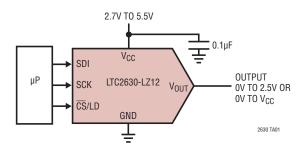


Figure 4. Effects of Rail-to-Rail Operation on a DAC Transfer Curve (Shown for 12-Bits).

(a) Overall Transfer Function (b) Effect of Negative Offset for Codes Near Zero (c) Effect of Positive Full-Scale Error for Codes Near Full Scale

TYPICAL APPLICATION

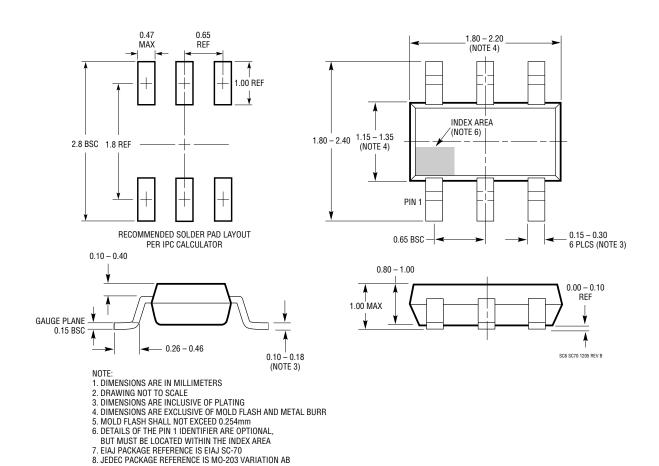
12-Bit, 2.7V to 5.5V Single Supply, Voltage Output DAC



PACKAGE DESCRIPTION

SC6 Package 6-Lead Plastic SC70

(Reference LTC DWG # 05-08-1638 Rev B)



TYPICAL APPLICATION

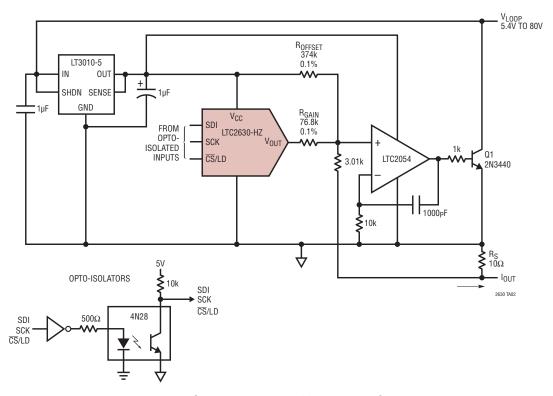


Figure 5. An Optoisolated 4mA to 20mA Process Controller

RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LTC1660/LTC1665	Octal 10-/8-Bit V _{OUT} DACs in 16-Pin Narrow SSOP	V _{CC} = 2.7V to 5.5V, Micropower, Rail-to-Rail Output
LTC1663	Single 10-Bit V _{OUT} DAC in SOT-23	V _{CC} = 2.7V to 5.5V, 60μA, Internal reference, SMBus Interface
LTC1664	Quad 10-Bit V _{OUT} DAC in 16-Pin Narrow SSOP	V _{CC} = 2.7V to 5.5V, Micropower, Rail-to-Rail Output
LTC1669	Single 10-Bit V _{OUT} DAC in SOT-23	V_{CC} = 2.7V to 5.5V, 60 μ A, Internal reference, I ² C Interface
LTC1821	Parallel 16-Bit Voltage Output DAC	Precision 16-Bit Settling in 2µs for 10V Step
LTC2600/LTC2610/LTC2620	Octal 16-/14-/12-Bit V _{OUT} DACs in 16-Lead SSOP	250μA per DAC, 2.5V to 5.5V Supply Range, Rail-to-Rail Output, SPI Serial Interface
LTC2601/LTC2611/LTC2621	Single 16-/14-/12-Bit V _{OUT} DACs in 10-Lead DFN	300μA per DAC, 2.5V to 5.5V Supply Range, Rail-to-Rail Output, SPI Serial Interface
LTC2602/LTC2612/LTC2622	Dual 16-/14-/12-Bit V _{OUT} DACs in 8-Lead MSOP	300μA per DAC, 2.5V to 5.5V Supply Range, Rail-to-Rail Output, SPI Serial Interface
LTC2604/LTC2614/LTC2624	Quad 16-/14-/12-Bit V _{OUT} DACs in 16-Lead SSOP	250μA per DAC, 2.5V to 5.5V Supply Range, Rail-to-Rail Output, SPI Serial Interface
LTC2631	Single 12-/10-/8-Bit I ² C V _{OUT} DACs with 10ppm/°C Reference in ThinSOT	180μA per DAC, 2.7V to 5.5V Supply Range, 10ppm/°C Reference, Selectable External Ref. Mode, Rail-to-Rail Output, I ² C Interface
LTC2640	Single 12-/10-/8-Bit SPI V _{OUT} DACs with 10ppm/°C Reference in ThinSOT	180µA per DAC, 2.7V to 5.5V Supply Range, 10ppm/°C Reference, Selectable External Ref. Mode, Rail-to-Rail Output, SPI Interface