

## Description

The Convert Select front end series represents a family of DIN-rail mountable DC-DC and AC-DC converters with power factor correction. The converters have been designed according to the latest industry requirements and standards. The converters are ideal for use in outdoor and other demanding applications to power building control systems, factory automation, industrial controls, instrumentation, electromagnetic drives, fans, and other DC loads. Different models are available with a single output or two independently regulated, electrically isolated outputs with 12, 24,36 , or 48 V . Special models for battery charging are available. The EW models are particularly suitable for 110 V railway applications; they have been designed in accordance with the railway standards EN 50155 and EN 50121.
Key features of the Convert Select line include power factor correction with low harmonic distortion, negligibly low inrush current, high immunity to transients and surges, and low
electromagnetic emissions. Internal protection circuits such as input over- and undervoltage lockout, thermal protection, as well as output overvoltage protection by a second control loop ensure safe operation of the final system.
The outputs deliver an electrically-isolated Safety Extra Low Voltage, SELV, (except models LWR/LWN1740) and low output noise. They are no-load, overload, and short-circuit proof. The electronically controlled short-term peak power capability of up to $150 \%$ of the rated output power enables the front end converters to deliver additional power to start-up motors or to safely operate subsequent circuit breakers. Built-in large sized output capacitors absorb possible reverse energy, which may be caused by quick deceleration of electromagnetic drives connected directly to the output. A green LED at the front cover displays the status of the output(s).

The Convert Select Series was designed according to all relevant international safety standards. The converters are approved by TÜV and UL, and are UL 508 listed. Adequate
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Safety according to IEC/EN 60950-1, UL/CSA 60950-1, IEC/EN 50178, IEC 61010-1, UL 508

## Features

- RoHS lead-free-solder and lead-solder-exempted products are available
- Rugged 35 mm DIN-rail snap-fit design
- Class I equipment
- Universal AC-input with single stage conversion AC to DC, or DC input of 66 to 350 V
- Power factor correction, harmonics IEC/EN 61000-3-2
- Virtually no inrush current
- Immunity to IEC/EN 61000-4-2, -3, -4, -5, -6, -11
- Emissions according to EN 55011/022
- Very high efficiency; up to 89\%
- Short-term output peak power capability, rectangular current limiting characteristic
- Single or two independently regulated outputs with 12 , 24,36 , or 48 V
- Outputs no-load, overload, and short-circuit proof
- PCBs protected by lacquer
- Very high reliability
clearances and creepage distances allow operation in pollution degree 3 environment (with AC input). All board assemblies are coated with a protective lacquer.
The thermal concept allows operation at full load up to an ambient temperature of $60^{\circ} \mathrm{C}$ (LW models) or $70^{\circ} \mathrm{C}$ (EW models) in free air without forced cooling. A rugged DIN snapfit device allows easy and reliable fixing onto the various 35 mm DIN rail models. The converters are fitted with cage clamp
terminals easily accessible from the front. System connectors with screw terminals for use with pre-assembled harnesses, external adjustment of the output voltage as well as various auxiliary functions are available as options.
The letter E stands for improved EMC performance of LW models.


## Model Selection

Table 1: Standard models

| Output 1 |  | Output 2 |  | Output Power $P_{\text {o nom }}$ [W] | Operating Input Voltage$V_{i \min }-V_{i \max }$ | Type Designation ${ }^{6}$ | Effic.$\begin{gathered} \eta_{\min }{ }^{8} \\ {[\%]} \end{gathered}$ | Options ${ }^{3,5}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} V_{\text {o1 nom }}{ }^{1} \\ {[V D C]} \end{gathered}$ | $I_{01}$ nom [A] | $\begin{aligned} & V_{\text {o2 nom }}{ }^{1} \\ & {[\text { VDC] }} \end{aligned}$ | $I_{02 \text { nom }}$ [A] |  |  |  |  |  |
| 12.35 | 7.5* | - | - | 93* | $\begin{gathered} 85^{2}-264 \mathrm{VAC} \\ 47-63 \mathrm{~Hz}^{4} \\ 90^{2}-350 \mathrm{VDC}^{7} \end{gathered}$ | LWR1301-6E 3 | 83* | $\begin{gathered} \mathrm{R} \\ \text { D1, D2, D5 } \\ \text { M1, M2 } \\ \text { F } \\ \text { K2, G } \end{gathered}$ |
| 12.35 | 14* | - | - | 173* |  | LWN1301-6E ${ }^{3}$ | 83* |  |
| 24.7 | 5 | - | - | 124 |  | LWR1601-6E | 87 |  |
| 24.7 | 10 | - | - | 247 |  | LWN1601-6E | 87 |  |
| 37 | 3.3 | - | - | 122 |  | LWR1701-6E ${ }^{3}$ | 88 |  |
| 37 | 6.6 | - | - | 244 |  | LWN1701-6E ${ }^{3}$ | 88 |  |
| 49.4 | 2.5 | - | - | 124 |  | LWR1801-6E | 88 |  |
| 49.4 | 5 | - | - | 247 |  | LWN1801-6E | 88 |  |
| 12.35 | 7* | 12.35 | 7* | 173* |  | LWN2320-6E ${ }^{3}$ | 83* |  |
| 24.7 | 5 | 24.7 | 5 | 247 |  | LWN2660-6E | 87 |  |
| 37 | 3.3 | 37 | 3.3 | 244 |  | LWN2770-6E ${ }^{3}$ | 89 |  |
| 49.4 | 2.5 | 49.4 | 2.5 | 247 |  | LWN2880-6E | 89 |  |
| 24.7 | 5 | - | - | 120 | $66-150$ VDC | EWR1601-0 ${ }^{9}$ | 87 | $\begin{gathered} \mathrm{R}, \mathrm{M} 1, \mathrm{M} 2 \\ \mathrm{Q}, \mathrm{~K} 2, \mathrm{G} \end{gathered}$ |
| 24.7 | 5 | 24.7 | 5 | 240 |  | EWN2660-0 ${ }^{9}$ | 87 |  |

* Version 106 or higher

1 R-input not connected.
2 For derating at low input voltage see section Output Power Derating.
3 For minimum quantity and lead times contact Power-One.
4 The converters have been tested up to 440 Hz ; for operating frequencies $<47 \mathrm{~Hz}$ or $>63 \mathrm{~Hz}$ contact Power-One.
5 On double-output models the options R, M2, D1, D2, D5 are related to the second output only.
6 Improved EMC performance for LWN/LWR models. Former models without E are still available on request.
$7 V_{i} \leq 250$ VDC for models with option $F$
8 Min. efficiency at $V_{\mathrm{i} \text { nom }}, I_{\mathrm{onom}}$, and $T_{\mathrm{A}}=25^{\circ} \mathrm{C}$. Typical values are approx. $2 \%$ better.
9 EWN and EWR models are designed for railway applications according to EN 50155 and EN 50121.

Table 2: Battery charger models (M1 included)

| Output Voltage |  |  | Nominal Output Values |  |  | Operating Input Voltage $V_{i \text { min }}-V_{i \text { max }}$ | Type Designation ${ }^{6}$ | $\begin{array}{\|c} \hline \text { Effic. } \\ \eta_{\text {min }} 8 \\ {[\%]} \\ {[\%]} \end{array}$ | Options ${ }^{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} V_{\text {Bat }} \\ {[\mathrm{VDC}]} \end{gathered}$ | $\begin{aligned} & V_{\text {o safe }^{1}} \\ & {[\mathrm{VDCD}]} \end{aligned}$ | $v_{o \text { max }}$ [VDC] | $\begin{aligned} & V_{\text {onom }^{5}} \\ & {[V D C]} \end{aligned}$ | $I_{0}$ nom $^{5}$ <br> [A] | $P_{o n o m}{ }^{5}$ <br> [W] |  |  |  |  |
| 12 | $13.8{ }^{1}$ | 14.65 | 13.8 | 7.5* | 104* | $\begin{gathered} 85^{2}-264 \mathrm{VAC}, \\ 47-63 \mathrm{~Hz}^{4}, \\ 90^{2}-350 \mathrm{VDC}^{7} \end{gathered}$ | LWR1140-6EM1 ${ }^{3}$ | 83* | $\begin{gathered} \mathrm{F} \\ \mathrm{~K} 2, \mathrm{G} \end{gathered}$ |
|  |  |  |  | 14* | 194* |  | LWN1140-6EM1 ${ }^{3}$ | 85* |  |
| 24 | $25.68{ }^{1}$ | 29.3 | 27.3 | 4.2 | 115 |  | LWR1240-6EM1 | 86 |  |
|  |  |  |  | 8.4 | 230 |  | LWN1240-6EM1 | 85 |  |
| 36 | $38.52^{1}$ | 43.95 | 40.88 | 2.8 | 115 |  | LWR1840-6EM1 ${ }^{3}$ | 86 |  |
|  |  |  |  | 5.6 | 230 |  | LWN1840-6EM1 ${ }^{3}$ | 86 |  |
| 48 | $51.36^{1}$ | 58.6 | 54.5 | 2.1 | 115 |  | LWR1740-6EM1 | 86 |  |
|  |  |  |  | 4.2 | 230 |  | LWN1740-6EM1 | 87 |  |

* Version 106 or higher

1 Setting voltage (typ.) with open R-input
2 For derating at low input voltage see section Output Power Derating.
3 For minimum quantity and lead times consult Power-One.
4 The converters have been tested up to 440 Hz ; for operating frequency $<47 \mathrm{~Hz}$ or $>63 \mathrm{~Hz}$ contact Power-One.
5 Nominal output figures, calculated with a cell voltage of 2.27 V at $20^{\circ} \mathrm{C}$.
6 Improved EMC performance. Former models without $E$ are still available on request.
$7 V_{\mathrm{i}} \leq 250$ VDC for models with option $F$.
8 Min. efficiency at $V_{\text {inom }}, V_{\text {o nom }}, I_{\text {o nom }}$, and $T_{\mathrm{A}}=25^{\circ} \mathrm{C}$. Typical values are approx. $2 \%$ better.

## Part Number Description


${ }^{1} \mathrm{G}$ is always placed at the end of the part number. Consult Power-One for availability!
${ }^{2}$ Only one of these options is possible.

[^0]
## Product Marking

Basic type designation, applicable safety approval and recognition marks, CE mark, warnings, pin designation, Power-One company logo.

Specific type designation, input voltage range, nominal output voltages and currents, degree of protection, batch no., serial no., and data code including production site, version, and date of production.

## Functional Description

The W Series converters are primary controlled AC-DC or DCDC flyback converters with a constant switching frequency of 130 kHz . The power-factor-corrected single-step conversion of the input voltage to a low output voltage results in extremely high efficiency. Depending upon the output power, the converters are fitted with one (125 W) or two (250 W) powertrains. Models with two powertrains have one or two outputs. Double-output models exhibit indiviually regulated powertrains.
The input voltage is fed via fuse, filter, and rectifier to the main transformer, designed in planar technique. The input filter with very small input capacitance generates virtually no inrush current. An input transient suppressor protects the converter against high voltage peaks and surges. Input over- and undervoltage lockout as well as input current limitation protect the converter from operation outside of its specification. The input voltage waveform is sensed by the primary control logic to allow active power factor correction, forcing the input current to follow the input voltage waveform.

The secondary side of the main transformer supplies via the rectifier diode a large electrolytic output storage capacitor providing for the hold-up time. Double-output models exhibit an individual control logic each. The output voltage and the output current are measured and fed back to the primary control logic via an optocoupler. A second control loop monitors the output voltage. It disables the output in the case of a failure in the control logic and limits the output voltage.
Built-in temperature sensors monitor the internal temperature of each powertrain. If the temperature exceeds the limit, the converter reduces the output power continuously to keep the temperature below its limit. A green LED on the front cover confirms the presence of the output voltage(s).
The R input (option $\mathrm{R}, \mathrm{M} 1$, or M 2 ) allows for external adjustment of the output voltage by means of a resistor or an external voltage source. An external sensor can be connected to the R input and allows for temperature-controlled battery charging (see Accessories).


Fig. 1
Single-output converters (125 W).


Fig. 2
250 W converters.
The figure shows EWN double-output models. For the pinout of 250 W single-output models see fig. 1 or table 13.

## Electrical Input Data

General conditions:
$T_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless $T_{\mathrm{C}}$ is specified.
Table 4a: Input data of LW models

| Input |  |  | LWR |  | LWN |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | AC-Input | DC-Input | AC-Input | DC-Input |  |
| Characteristic |  | Conditions | min typ max | min typ max | min typ max | min typ max |  |
| $V_{i}$ | Operating input voltage range | $\begin{aligned} & I_{\mathrm{o}}=0-I_{\mathrm{onom}} \\ & T_{\mathrm{c}}-T_{\mathrm{c} \max } \end{aligned}$ | $85^{2} \quad 264$ | $90^{2} \quad 3504$ | $85^{2} \quad 264$ | $90^{2} \quad 3504$ | V |
| $V_{\text {inom }}$ | Rated input volt. range |  | 100 (230) 240 | 220 | 100 (230) 240 | 220 |  |
| $f_{i}$ | Rated input frequency ${ }^{1}$ |  | 50-60 | -- | 50-60 | -- | Hz |
| $I_{i}$ | Input current | $I_{\text {o nom, }}, V_{\mathrm{i}}=V_{\text {i nom }}$ | 0.63 | 0.65 | 1.25 | 1.3 | A |
|  |  | $I_{\text {onom, }}, V_{i}=V_{\text {i min }}$ | 1.75 | 1.67 | 3.5 | 3.3 |  |
| $P_{\text {i }}$ | No-load input power | $V_{i \text { min }}-V_{\text {i max }}$ | 1.2 | 0.9 | 1.3 | 1 | w |
| $I_{\text {inrush }}$ | Inrush current | $V_{\text {i max }}, t>0.1 \mathrm{~ms}$ | 3 | 3 | 5 | 5 | A |
| $C_{i}$ | Input capacitance |  | 5 | 5 | 6 | 6 | $\mu \mathrm{F}$ |
| PF | Power factor | $V_{\text {i nom }}=230 \mathrm{~V}, I_{\text {o nom }}$ | $0.86{ }^{5}$ | -- | $0.86{ }^{5}$ | -- |  |
| $V_{\text {i RFI }}$ | Conducted input RFI | $\begin{aligned} & \text { EN 55011/55022 } \\ & V_{\text {i nom, }} I_{\text {o nom }} \end{aligned}$ | A, $\mathrm{B}^{3}$ | A, $\mathrm{B}^{3}$ | A, $\mathrm{B}^{3}$ | A, $\mathrm{B}^{3}$ |  |
|  | Radiated input RFI |  | $\mathrm{B}^{3}$ | $\mathrm{B}^{3}$ | $\mathrm{B}^{3}$ | $\mathrm{B}^{3}$ |  |
| $f_{\text {switch }}$ | Switching frequency |  | 130 | 130 | 130 | 130 | kHz |

1 For operating frequencies $<47 \mathrm{~Hz}$ and $>63 \mathrm{~Hz}$ contact Power-One. The converters have been tested up to 440 Hz .
2 Output power derating at low input voltage and/or high case temperature $T_{\mathrm{C}}$ (see Output power derating).
3 Only valid for models with Option E (type test with LWN1801-6E)
$4 V_{i} \leq 250$ VDC for models with option F .
5 Models with 12 V output: $\geq 0.70$ for LWR, $\geq 0.75$ for LWN

Table 4b: Input data of EW models

| Input |  |  |  | EWR |  |  | W N |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | DC-Input |  |  | DC-Input |  |  |  |
| Characteristic |  | Conditions | min | typ | max | min | typ | max |  |
| $V_{\mathrm{i}}$ | Operating input voltage range | $\begin{aligned} & I_{\mathrm{o}}=0-I_{\mathrm{onom}} \\ & T_{\mathrm{c}} \text { to } T_{\mathrm{c} \max } \end{aligned}$ | 66 |  | $150{ }^{1}$ | 66 |  | $150{ }^{1}$ | V |
| $V_{\text {inom }}$ | Nominal input voltage |  |  | 110 |  |  | 110 |  |  |
| $V_{\text {UVT }}$ | Undervoltage trigger |  | 54 |  | 60 | 54 |  | 60 |  |
| $I_{\text {i }}$ | Input current | $I_{\text {o nom, }}, V_{\mathrm{i}}=V_{\text {i nom }}$ | 1.25 |  |  | 2.5 |  |  | A |
|  |  | $I_{\text {o nom, }}, V_{\mathrm{i}}=66 \mathrm{~V}$ | 2.2 |  |  | 4.4 |  |  |  |
| $P_{\text {i }}$ | No-load input power | $V_{i \text { min }}-V_{i \text { max }}$ | 0.8 |  |  | 1.3 |  |  | W |
| $I_{\text {inrush }}$ | Inrush current | $V_{\text {i max }}, t>0.1 \mathrm{~ms}$ | 6 |  |  | 12 |  |  | A |
| $\mathrm{C}_{\mathrm{i}}$ | Input capacitance |  | 2.5 |  |  | 4.5 |  |  | $\mu \mathrm{F}$ |
| $V_{\text {i RFI }}$ | Conducted input RFI | EN 55011/55022 <br> $V_{\text {i nom, }} I_{\text {o nom }}$ | A |  |  | A |  |  |  |
|  | Radiated input RFI |  | -- |  |  | -- |  |  |  |
| $f_{\text {switch }}$ | Switching frequency |  | 130 |  |  | 130 |  |  | kHz |

$1 V_{\mathrm{i}} \leq 168 \mathrm{VDC}$ for 3 s . Overvoltage trigger adjusted to $170-182 \mathrm{~V}$.

## Output Power Derating

The output power of LW models must be decreased at low input voltage and/or powertrain temperature above $125^{\circ} \mathrm{C}$.
The powertrain temperature depends on the output power, the input voltage, and the cooling method. At low input voltage the losses increase. At the maximum specified environment temperature $T_{\mathrm{A}}$ free air convection cooling might be insufficient approaching maximum ambient conditions. As a result, the output power has to be reduced according to the tables below.

Note: The measurements have been made by the approval boards with free air convection cooling according to UL 60950 specified ambient temperature $T_{\mathrm{A}}$ and with the converter built in a cardboard box according to UL 508 and a specified temperature outside the box $T_{\text {out. }}$. The tables give a correlation between $T_{\mathrm{A}}$ or $T_{\text {out }}$ and the case temperature $T_{C}$ (measuring point $T_{C}$ see Mechanical Data). For models not specified, please contact Power-One.
EW models have no derating.

Table 5a: $P_{0}$ derating according to UL 60950 at $T_{A}=60^{\circ} \mathrm{C}$, or according to UL 508 at $T_{\text {out }}=50^{\circ} \mathrm{C}$

| Model | $P_{\text {o nom }}$ [W] | $T_{\mathrm{C} \text { max }}$ [ ${ }^{\circ} \mathrm{C}$ ] | Derate below |  | derate by |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $V_{i}$ [VAC] | $V_{\mathrm{i}}$ [VDC] | [W/V] |
| LWR1601-6E | 124 | 80 | 108 | 98 | -0.67 |
| LWN1601/2660-6E | 247 | 89 | 125 | 115 | -1.25 |
| LWR1701-6E | 122 | 80 | 125 | 115 | -1.25 |
| LWN1701-6E | 244 | 90 | 125 | 115 | -1.25 |
| LWR1801-6E | 124 | 80 | 98 | 93 | -0.67 |
| LWN1801/2880-6E | 247 | 89 | 125 | 115 | -1.25 |

Table 5b: $P_{0}$ derating according to UL 60950 at $T_{\mathrm{A}}=50^{\circ} \mathrm{C}$, or according to $U L 508$ at $T_{\text {out }}=40^{\circ} \mathrm{C}$

| Model | $P_{\text {onom }}$ [W] | $T_{\mathrm{C} \text { max }}$ [ ${ }^{\circ} \mathrm{C}$ ] | Derate below |  | derate by |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | $V_{i}$ [VAC] | $V_{\text {i }}$ [VDC] | [W/V] |
| LWR1601-6E | 124 | 76 | 98 | no derating | -0.67 |
| LWN1601/2660-6E | 247 | 86 | 115 | 105 | -1.25 |
| LWR1801-6E | 124 | 76 | 93 | no derating | -0.67 |
| LWN1801/2880-6E | 247 | 86 | 105 | 95 | -1.25 |

## Input Fuse and Protection

A fast-blow fuse (Schurter F $6.3 \mathrm{~A}, 5 \times 20 \mathrm{~mm}$ ), protected by a sleeve, is connected to the input $\mathrm{L} \approx$ or $\mathrm{Vi}+$. EW models have a smaller fuse ( $250 \mathrm{~V}, 4 \times 9 \mathrm{~mm}$, SOC NT3 6.3A V009, ULrecognized E-39265). For DC input voltages above 250 V consult the Installation Instructions.
Converters with option $F$ have 2 small fuses, one in each input line. Converters with option EF ( E and F ) have 2 large fuses (F6.3A, $5 \times 20 \mathrm{~mm}$ ). The DC input voltage for converters with option $F$ is limited to 250 V .
A VDR and a symmetrical input filter form an effective protection against input transients.
An under- and an overvoltage lockout protect the converter, which is disabled below $V_{\mathrm{i} \text { min }}$ and above $V_{\mathrm{i} \max }$ by an internally generated inhibit signal.
The built-in bridge rectifier (LW models) provides reverse polarity protection at the input if operated from DC.
EW models are protected by the (blowing) input fuse in connection with the body diode of the main transistor. Option Q offers a serial diode, but this reduces the efficiency by approx. 1\%.

## Efficiency



Fig. 3
Efficiency versus load (LWN2660-6)

125, 250 Watt AC-DC and DC-DC DIN-Rail Converters

## Power Factor, Harmonics

All converters feature active power factor correction.


Fig. 4
Harmonic currents at input current, measured at $V_{i}=230$
VAC, $I_{0}=I_{0 \text { nom ( }}$ (LWN1701-6E).


Fig. 5
Power factor versus load (LWN2660-6)

## Electrical Output Data

Table 6a: Output data of 125 Watt standard models. General conditions: $T_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless $T_{\mathrm{A}}$ is specified; $R$ input open-circuit

| Model |  |  | LWR1301 |  |  | EWR/LWR1601 |  |  | LWR1701 |  |  | LWR1801 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristic |  | Conditions | min | typ | max | min | typ | max | min | typ | max | min | typ max |  |
| $V_{\text {o nom }}$ | Output voltage nominal ${ }^{1}$ | $V_{\text {i nom }}, I_{\text {o nom }}$ |  | 12.0 |  | 24.25 | 24.7 | 25.2 | 36.4 | 37 | 37.8 | 48.5 | 49.3650 .4 | V |
|  |  |  | 12.2 | 12.35 | 12.5 | 24.55 | 24.7 | 24.85 | 36.8 | 37 | 37.2 | 49.06 | 49.3649 .66 |  |
| $V_{\text {o worst }}$ | Output voltage range of tolerance | $\begin{aligned} & V_{\mathrm{i} \min }-V_{\mathrm{i} \max }, \\ & I_{0}=(0.1-1) I_{0 \text { nom }} \end{aligned}$ | 11.9 |  | 13.0 | 24.0 |  | 25.8 | 36.0 |  | 38.7 | 48.0 | 51.6 |  |
| $V_{\text {OL }}$ | Overvoltage protection |  | 14* |  | 15 * | 28.5 |  | 30 | 42.7 |  | 45 | 57 | 60 |  |
| $P_{\text {o nom }}$ | Nominal output power | $V_{\mathrm{i}}=100 \mathrm{~V}-V_{\mathrm{i} \text { max }}$ | 105* |  |  | 124 |  |  | 122 |  |  | 124 |  | W |
| $I_{\text {o nom }}$ | Output current nominal |  | 7.5 * |  |  | 5.0 |  |  | 3.3 |  |  | 2.5 |  | A |
| $I_{0 L}$ | Output current limit ${ }^{3}$ | $V_{\text {i min }}-V_{\text {i max }}$ | 7.6 |  | 8.5 | 5.1 |  | 5.7 | 3.4 |  | 3.8 | 2.53 | 2.9 |  |
| $I_{\text {op }}$ | Output current boost ${ }^{4}$ | typ. 1 s |  | 11.3 |  |  | 7.5 |  |  | 5.0 |  |  | 3.75 |  |
| $V_{0}$ | Ripple and noise $\frac{\text { EWR }}{\text { LWR }}$ | $V_{\mathrm{i}}=110 \mathrm{VDC}, I_{\text {o nom }}$ | - |  |  | 500 |  |  | - |  |  | - |  | mV pp |
|  |  | $\begin{aligned} & V_{\mathrm{i}}=230 \mathrm{VAC}, \\ & f_{\mathrm{i}}=50 \mathrm{~Hz}, I_{\mathrm{onom}} \end{aligned}$ | 100 |  |  | 100 |  |  | 100 |  |  | 100 |  |  |
|  |  |  | $1100{ }^{2}$ |  |  | $1100{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  |
| $\Delta V_{\text {ou }}$ | Static line regulation | $100 \mathrm{~V}-V_{\text {i max }}, I_{\text {o nom }}$ |  | $\pm 0.08$ |  | $\pm 0.1$ |  |  | $\pm 0.15$ |  |  |  | $\pm 0.15$ | V |
| $\Delta V_{01}$ | Static load regulation (droop) | $\begin{aligned} & V_{\text {i nom }} \\ & I_{0}=(0.1-1) I_{\mathrm{onom}} \end{aligned}$ |  | -0.2 |  |  | -0.4 |  |  | -0.6 |  |  | -0.8 |  |
| $V_{\text {od }}$ | Dynamic load regulation <br> Voltage deviation <br> Recovery time | $\begin{aligned} & V_{\text {i nom }} \\ & I_{\mathrm{o}}=(0.5 \leftrightarrow 1) I_{\mathrm{onom}} \end{aligned}$ | $\pm 1$ |  |  | $\pm 1.2$ |  |  | $\pm 1.5$ |  |  | $\pm 1.8$ |  |  |
|  |  |  | 40 |  |  | 40 |  |  | 80 |  |  | 80 |  | ms |
| $\alpha V_{0}$ | Temperature coefficient | $T_{\text {C min }}-T_{\text {C max }}$ | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  | \%/K |
| $t_{\text {or }}$ | Start-up time | $V_{\mathrm{i}}=0 \rightarrow V_{\text {i nom, }} I_{\text {o nom }}$ | 700 |  |  | 700 |  |  | 700 |  |  | 700 |  | ms |
| $t_{\text {oh min }}$ | Hold-up time | $I_{0}$ nom, $V_{\text {o nom }} \rightarrow 0.8 V_{\text {o nom }}$ | 10 |  |  | 6/15 |  |  | 20 |  |  | 25 |  |  |

* Converters with feature E and version $\geq 106$

1 Setting voltage with open $R$-input
2 Superimposed low frequency ripple at $2 \cdot f_{i}$
3 Rectangular current limit characteristic (continuous operation)
4 Short-term peak power capability $150 \%$ of $P_{\mathrm{o} \text { nom }}$ for approx. 1 s

Table 6b: Output data of 250 Watt single-output standard models. General conditions as in table 6a

| Model |  |  | LWN1301 |  |  | LWN1601 |  |  | LWN1701 |  |  | LWN1801 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristic |  | Conditions | min | typ | max | min | typ | max | min | typ | max | min | typ | max |  |
| $V$ onom | Output voltage nominal ${ }^{1}$ | $V_{\text {i nom }}, I_{\text {o nom }}$ |  |  |  | 24.25 | 24.7 | 25.2 | 36.4 | 37 | 37.8 | 48.5 | 49.36 | 50.4 | V |
|  |  |  | 12.2 | 12.35 | 12.5 | 24.55 | 24.7 | 24.85 | 36.8 | 37 | 37.2 | 49.06 | 49.36 | 49.66 |  |
| $V$ oworst | Output voltage range of tolerance | $\begin{aligned} & V_{\mathrm{i} \min }-V_{\mathrm{i} \text { max }} \\ & I_{\mathrm{o}}=(0.1-1) I_{\mathrm{on} \text { nom }} \end{aligned}$ | 11.9 |  | 13.0 | 24.0 |  | 25.8 | 36.0 |  | 38.7 | 48.0 |  | 51.6 |  |
| $V_{\text {o L }}$ | Overvoltage protection |  | 14* |  | 15 * | 28.5 |  | 30 | 42.7 |  | 45 | 57 |  | 60 |  |
| $P_{\text {o nom }}$ | Nominal output power | $V_{i}=100 \mathrm{~V}-V_{\mathrm{i} \text { max }}$ | 173 * |  |  | 247 |  |  | 244 |  |  | 247 |  |  | W |
| $I_{0}$ nom | Output current nominal |  | 14 * |  |  | 10 |  |  | 6.6 |  |  | 5.0 |  |  | A |
| $I_{\text {OL }}$ | Output current limit ${ }^{3}$ | $V_{\text {i min }}-V_{\text {i max }}$ | 14.1 * |  | 16* | 10.2 |  | 11.4 | 6.7 |  | 7.6 | 5.1 |  | 5.6 |  |
| $l_{\text {op }}$ | Output current boost ${ }^{4}$ | typ. 1 s |  | 21 |  |  | 15 |  |  | 10 |  |  | 7.5 |  |  |
| $v_{0}$ | Ripple and noise | $\begin{aligned} & V_{\mathrm{i}}=230 \mathrm{VAC}, \\ & f_{\mathrm{i}}=50 \mathrm{~Hz}, I_{\mathrm{onom}} \end{aligned}$ | 100 |  |  | 100 |  |  | 100 |  |  | 100 |  |  | mV pp |
|  |  |  | $1100{ }^{2}$ |  |  | $1100{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  |  |
| $\Delta V_{\text {ou }}$ | Static line regulation | $100 \mathrm{~V}-V_{\text {i max }}, I_{\text {o nom }}$ | $\pm 0.08$ |  |  | $\pm 0.1$ |  |  | $\pm 0.15$ |  |  | $\pm 0.15$ |  |  | V |
| $\Delta V_{\text {ol }}$ | Static load regul. (droop) | $V_{\text {i nom }},(0.1-1) I_{0 \text { nom }}$ | -0.2 |  |  | -0.4 |  |  | -0.6 |  |  | -0.8 |  |  |  |
| $V_{\text {od }}$ | Dynamic load regulation | $\begin{aligned} & V_{\text {i nom }}, \\ & I_{0}=(0.5 \leftrightarrow 1) I_{0} \text { nom } \end{aligned}$ | $\pm 1$ |  |  | $\pm 1.2$ |  |  | $\pm 1.5$ |  |  | $\pm 1.8$ |  |  |  |
|  | Recovery time |  | 40 |  |  | 40 |  |  | 80 |  |  | 80 |  |  | ms |
| $\alpha V_{0}$ | Temperature coefficient | $T_{\text {C min }}-T_{\text {C max }}$ | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | \%/K |
| $t_{\text {or }}$ | Start-up time | $V_{\mathrm{i}}=0 \rightarrow V_{\text {i nom }}, l_{\mathrm{o}}$ nom | 700 |  |  | 700 |  |  | 700 |  |  | 700 |  |  | ms |
| $t_{\text {oh min }}$ | Hold-up time | $I_{\mathrm{on}}$ nom, <br> $V_{\text {o nom }} \rightarrow 0.8 V_{\text {o nom }}$ | 10 |  |  | 15 |  |  | 20 |  |  | 25 |  |  |  |

Table 6c: Output data of 250 Watt double-output standard models. General conditions as in table 6a

| Model |  |  | LWN2320 |  |  | EWN/LWN2660 |  |  | LWN2770 |  |  | LWN2880 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristic |  | Conditions | min | typ | max | min | typ | max | min | typ | max | min | typ | $\max$ |  |
| $V_{01}$ nom <br> $V_{\text {o2 nom }}$ | Output voltage nominal ${ }^{1}$ | $V_{\text {i nom }}, I_{\text {o nom }}$ |  |  |  | $24.25 \quad 24.7 \quad 25.2$ |  |  | 37 |  |  | $\begin{array}{llll}48.5 & 49.36 & 50.4\end{array}$ |  |  | V |
|  |  |  | 12.2 | 12.35 | 12.5 | 24.55 | 24.7 | 24.85 | 36.8 | 37 | 37.2 | 49.06 | 49.36 | 49.66 |  |
| $V_{\text {o worst }}$ | Output voltage range of tolerance | $\begin{aligned} & V_{\mathrm{i} \min }-V_{\mathrm{i} \max }, \\ & I_{0}=(0.1-1) I_{\mathrm{onom}} \end{aligned}$ | 11.9 |  | 13.0 | 24.0 |  | 25.8 | 36.0 |  | 38.7 | 48.0 |  | 51.6 |  |
| $V_{\text {oL }}$ | Overvoltage protection |  | 14* |  | 15 * | 28.5 |  | 30 | 42.7 |  | 45 | 57 |  | 60 |  |
| $P_{\text {o nom }}$ | Nominal output power | $V_{\mathrm{i}}=100 \mathrm{~V}-V_{\mathrm{i} \text { max }}$ |  | 173* |  |  | 247 |  |  | 244 |  |  | 247 |  | W |
| $I_{\text {o nom }}$ | Output current nominal |  |  | $2 \times 7$ * |  |  | $2 \times 5$ |  |  | $2 \times 3.3$ |  |  | $2 \times 2.5$ |  | A |
| $I_{\text {OL }}$ | Output current limit ${ }^{3}$ | $V_{\mathrm{imin}}-V_{\mathrm{i} \text { max }}$ | 7.1* |  | 7.8* | 5.1 |  | 5.7 | 3.4 |  | 3.8 | 2.53 |  | 2.9 |  |
| $I_{\text {op }}$ | Output current boost ${ }^{4}$ | typ. 1 s |  | 2×11.3* |  |  | $2 \times 7.5$ |  |  | $2 \times 5.0$ |  |  | $2 \times 3.75$ |  |  |
| $v_{0}$ | Ripple and noise | $V_{\mathrm{i}}=230 \mathrm{VAC},$ |  | 100 |  |  | $100{ }^{5}$ |  |  | 100 |  |  | 100 |  | mV pp |
|  |  | $f_{\mathrm{i}}=50 \mathrm{~Hz}, I_{\text {o nom }}$ |  | $1100^{2}$ |  |  | $110{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  |
| $\Delta V_{\text {ou }}$ | Static line regulation | $100 \mathrm{~V}-V_{\text {i max }}, I_{\text {onom }}$ |  | $\pm 0.08$ |  |  | $\pm 0.1$ |  |  | $\pm 0.15$ |  |  | $\pm 0.15$ |  | V |
| $\Delta V_{\text {ol }}$ | Static load regul. (droop) | $V_{\text {i nom, }}(0.1-1) I_{\text {o nom }}$ |  | -0.2 |  |  | -0.4 |  |  | -0.6 |  |  | -0.8 |  |  |
| $V_{\text {od }}$ | Dynamic load regulation | $V_{\text {i nom }}$, |  | $\pm 1$ |  |  | $\pm 1.2$ |  |  | $\pm$ |  |  | $\pm 1.8$ |  |  |
|  | Voltage deviation <br> Recovery time | $I_{\mathrm{O}}=(0.5 \leftrightarrow 1) I_{\mathrm{o}} \mathrm{nom}$ |  | 40 |  |  | 40 |  |  | 80 |  |  | 80 |  | ms |
| $\alpha V_{0}$ | Temperature coefficient | $T_{\text {C min }}-T_{\text {C max }}$ |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  | \%/K |
| $t_{\text {or }}$ | Start-up time | $V_{\mathrm{i}}=0 \rightarrow V_{\text {i nom },} I_{\mathrm{o}}$ nom |  | 700 |  |  | 700 |  |  | 700 |  |  | 700 |  | ms |
| $t_{\text {oh min }}$ | Hold-up time | $I_{0}$ nom, <br> $V_{\text {o nom }} \rightarrow 0.8 V_{\text {o nom }}$ |  | 10 |  |  | 6/15 |  |  | 20 |  |  | 25 |  |  |

* Converters with feature E and version $\geq 106 \quad 4$ Short-term peak power capability $150 \%$ of $P_{\text {o nom }}$ for approx. 1 s

1 Setting voltage with open R-input $\quad 5_{2}$ EWN2660: 500 mV @ $V_{\mathrm{i}}=110$ VDC
2 Superimposed low frequency ripple at $2 \cdot f_{\mathrm{i}}$
${ }^{3}$ Rectangular current limit characteristic (continuous operation)

Table 7a: Output data of 125 Watt battery charger models. General conditions: $T_{\mathrm{A}}=25{ }^{\circ} \mathrm{C}$, unless $T_{\mathrm{A}}$ is specified; $R$ input left open-circuit, unless otherwise specified

| Model |  |  | LWR1140-6EM1 |  | LWR1240-6EM1 |  |  | LWR1840-6EM1 |  |  | LWR1740-6EM1 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristic |  | Conditions | min typ | max | min | typ | max | min | typ | max | min | typ | max |  |
| $V_{\text {o safe }}$ | Output setting voltage ${ }^{1}$ | $V_{i n \text { nom, }} I_{\text {o nom }}$ | 13.4813 .8 | 14.15 | 24.5 | 25.68 | 26.3 | 36.75 | 38.52 | 39.5 | 49 | 51.36 | 52.6 | V |
| $V_{\text {Bat }}$ | Output voltage (max.) controlled by R input | $\begin{aligned} & V_{\text {imin }}-V_{i \max ,} \\ & I_{0}=(0.1-1) I_{0 \text { nom }} \end{aligned}$ | 14.65 |  | 29.3 |  |  | 43.95 |  |  | 58.6 |  |  |  |
| $V_{\text {oL }}$ | Overvoltage protection |  | 15.4* | 16.3* | 30.9 |  | 32.5 | 46 |  | 48.8 | 61.8 |  | 65 |  |
| $P_{\text {o nom }}$ | Nominal output power | $V_{i}=100 \mathrm{~V}-V_{\text {i max }}$ | 104* |  | 115 |  |  | 115 |  |  | 115 |  |  | W |
| $I_{\text {o nom }}$ | Output current nominal |  | 7.5* |  | 4.2 |  |  | 3.1 |  |  | 2.1 |  |  | A |
| $I_{0}$ L | Output current limit ${ }^{3}$ | $V_{i \text { min }}-V_{\text {i max }}$ | 7.6* | $8.4 *$ | 4.3 |  | 4.8 | 3.2 |  | 3.7 | 2.2 |  | 2.5 |  |
| $l_{\text {op }}$ | Output current boost ${ }^{4}$ | typ. 1 s | 11.2* |  | 7 |  |  | 4.7 |  |  | 3.5 |  |  |  |
| $v_{0}$ | Ripple and noise | $\begin{aligned} & V_{\mathrm{i}}=230 \mathrm{VAC}, \\ & f_{\mathrm{i}}=50 \mathrm{~Hz}, I_{\mathrm{onom}} \end{aligned}$ | 100 |  | 100 |  |  | 100 |  |  | 100 |  |  | mV pp |
|  |  |  | $1100{ }^{2}$ |  | $1100{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  |  |
| $\Delta V_{\text {ou }}$ | Static line regulation | $100 \mathrm{~V}-V_{\text {i max }}$, $I_{\text {o nom }}$ | $\pm 0.08$ |  | $\pm 0.1$ |  |  | $\pm 0.15$ |  |  | $\pm 0.15$ |  |  | V |
| $\Delta V_{\text {ol }}$ | Static load regulation (droop) | $\begin{aligned} & V_{\text {i nom, }} \\ & I_{0}=(0.1-1) I_{0} \text { nom } \end{aligned}$ | -0.2 |  | -0.4 |  |  | -0.6 |  |  | -0.8 |  |  |  |
| $v_{\text {od }}$ | Dynamic load regulation Voltage deviation Recovery time | $\begin{aligned} & V_{\text {i nom, }}, \\ & I_{0}=(0.5 \leftrightarrow 1) I_{\mathrm{onom}} \end{aligned}$ | $\pm 1.2$ |  | $\pm 1.2$ |  |  | $\pm 1.6$ |  |  | $\pm 1.9$ |  |  |  |
|  |  |  | 40 |  | 40 |  |  | 80 |  |  | 80 |  |  | ms |
| $\alpha v_{0}$ | Temperature coefficient | $T_{\mathrm{C}_{\text {min }}}-T_{\mathrm{C}_{\text {max }}}$ | $\pm 0.02$ |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | \%/K |
| $t_{\text {or }}$ | Start-up time | $V_{\mathrm{i}}=0 \rightarrow V_{\text {i nom, }} I_{\text {o nom }}$ | 700 |  | 700 |  |  | 700 |  |  | 700 |  |  | ms |

Table 7b: Output data of 250 Watt battery charger models. General conditions as in table 7a

| Model |  |  | LWN1140-6EM1 |  |  | LWN1240-6EM1 |  |  | LWN1840-6EM1 |  |  | LWN1740-6EM1 |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristic |  | Conditions | min | typ | max | min | typ | max | min | typ | max | min | typ | max |  |
| $V_{\text {o safe }}$ | Output setting voltage ${ }^{1}$ | $V_{\mathrm{i} \text { nom }}, I_{\text {o nom }}$ | 13.48 | 13.8 | 14.15 | 24.5 | 25.68 | 26.3 | 36.75 | 38.52 | 39.5 | 49 | 51.36 | 52.6 | V |
| $V_{\text {Bat }}$ | Output voltage (max.) controlled by R input | $\begin{aligned} & V_{\mathrm{i} \min }-V_{\mathrm{i} \text { max }}, \\ & I_{0}=(0.1-1) I_{0 \text { nom }} \end{aligned}$ | 14.65 |  |  | 29.3 |  |  | 43.95 |  |  | 58.6 |  |  |  |
| $V_{\text {OL }}$ | Overvoltage protection |  | 15.4* |  | 16.3* | 30.9 |  | 32.5 | 46 |  | 48.8 | 61.8 |  | 65 |  |
| $P_{\text {onom }}$ | Nominal output power | $V_{\mathrm{i}}=100 \mathrm{~V}-V_{\mathrm{i} \text { max }}$ | 194* |  |  | 230 |  |  | 230 |  |  | 230 |  |  | W |
| $I_{\text {o nom }}$ | Output current nominal |  | 14* |  |  | 10 |  |  | 6.2 |  |  | 4.2 |  |  | A |
| $I_{0}$ L | Output current limit ${ }^{3}$ | $V_{\text {i min }}-V_{\text {i max }}$ | 14.2* |  | 15.6* | 8.6 |  | 9.6 | 6.4 |  | 7.4 | 4.4 |  | 5.0 |  |
| $l_{\text {op }}$ | Output current boost ${ }^{4}$ | typ. 1 s | 22.6* |  |  | 14 |  |  | 9.4 |  |  | 7.0 |  |  |  |
| $v_{0}$ | Ripple and noise | $V_{\mathrm{i}}=230 \mathrm{VAC}$, | 100 |  |  | 100 |  |  | 100 |  |  | 100 |  |  | mV pp |
|  |  | $\mathrm{Hz}, I_{0}$ nom | $1100{ }^{2}$ |  |  | $1100{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  | $1200{ }^{2}$ |  |  |  |
| $\Delta V_{\text {ou }}$ | Static line regulation | $100 \mathrm{~V}-V_{\text {i max }}, I_{\text {o nom }}$ | $\pm 0.08$ |  |  | $\pm 0.1$ |  |  | $\pm 0.15$ |  |  | $\pm 0.15$ |  |  | V |
| $\Delta V_{\text {ol }}$ | Static load regulation (droop) | $\begin{aligned} & V_{\mathrm{i} \text { nom, }} \\ & I_{\mathrm{o}}=(0.1-1) I_{\mathrm{o} \text { nom }} \end{aligned}$ | $-0.2$ |  |  | -0.4 |  |  | -0.6 |  |  | -0.8 |  |  |  |
| $V_{\text {od }}$ | Dynamic load regulation <br> Voltage deviation <br> Recovery time | $\begin{aligned} & V_{\mathrm{i} \text { nom, }} \\ & I_{\mathrm{o}}=(0.5 \leftrightarrow 1) I_{\mathrm{o} \text { nom }} \end{aligned}$ | $\pm 1.2$ |  |  | $\pm 1.2$ |  |  | $\pm 1.6$ |  |  | $\pm 1.9$ |  |  |  |
|  |  |  | 40 |  |  | 40 |  |  | 80 |  |  | 80 |  |  | ms |
| $\alpha V_{0}$ | Temperature coefficient | $T_{\mathrm{C}_{\text {min }}}-T_{\mathrm{C} \text { max }}$ | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | $\pm 0.02$ |  |  | \%/K |
| $t_{\text {or }}$ | Start-up time | $V_{\mathrm{i}}=0 \rightarrow V_{\text {i nom, }} I_{\text {o nom }}$ | 700 |  |  | 700 |  |  | 700 |  |  | 700 |  |  | ms |

* Converters with feature E and version $\geq 106$

1 Setting voltage with open R -input $=V_{\mathrm{o} \text { safe }}$
2 Superimposed low frequency ripple at 2 - $f_{\mathrm{i}}$
3 Rectangular current limit characteristic (continuous operation)
4 Short-term peak power capability $150 \%$ of $P_{\text {o nom }}$ for approx. 1 s

## Parallel Operation

Double-output models exhibit an independant control logic each. Both outputs can be connected in parallel, provided that the options S (included in M1) and R are not used, since they influence only the $2^{\text {nd }}$ output. The two power trains share the current due to their output voltage droop characteristic.
Up to 3 converters with the same output voltage may be operated in parallel. It is possible to parallel W Series with $X$ Series converters.

Reasonable current sharing is achieved by the droop characteristic. Correct mode of operation is highly dependent upon the wiring of the converters and the impedance of these wires. Use wires with equal length and equal cross sections of $\min .1 .5 \mathrm{~mm}^{2}$. The best results for parallel operation can be achieved with the wiring shown in fig. 6.
Parallel operation of single-output models using the option R (output voltage adjust) is possible, but not recommended. Refer to fig. 6 ; the connections between the pins 8 and 9 (both Vo-) should be as short as possible.

Note: Parallel operation is not possible, if a temperature sensor is connected, as the sensor eliminates the output voltage droop.
Note: For ORing diodes, we recommend to use Schottkydiodes, mounted on a commun heatsink to avoid thermal run away (or the use of double diodes).

## Series Connection

Series connection of several outputs up to 150 V is possible. Exceeding an output voltage of 60 V , the output is not SELV.

## Output Characteristic and Protection

The output characteristic, individual for each powertrain, is rectangular with a droop to ease parallel operation; see fig. 7.

However, a $50 \%$ higher output current is possible for a short time, such allowing startup of loads or charging of capacitors; see fig. 8.

Each output is independently protected against internal overvoltage by means of a second control loop. When the output voltage exceeds $V_{o \mathrm{~L}}$, the respective output is disabled.

## Overtemperature Protection

A built-in temperature sensor protects each powertrain is independently protected against overtemperature. When a


Fig. 6
Wiring for single-output converters connected in parallel. Additional wiring for higher output currents and with the use of option $R$ is shown.


Fig. 7
$V_{0}$ versus $I_{0}$ (single-output model, typical values).


Fig. 8
Short term peak power characteristic: overcurrent versus time (typical values).
certain temperature is reached, the concerned powertrain reduces its output power continuously.

## Thermal Considerations

The thermal conditions are influenced by input voltage, output current, airflow, and temperature of surrounding components. $T_{\text {A max }}$ is therefore, contrary to $T_{\mathrm{C} \text { max }}$, an indicative value only.

Caution: The installer must ensure that under all operating conditions $T_{C}$ remains within the limits stated in the table Temperature specifications.

Note: Sufficient forced cooling allows $T_{\mathrm{A}}$ to be higher than $T_{\mathrm{A} \text { max }}$ provided that $T_{\mathrm{C} \text { max }}$ is not exceeded. It is recommended that continuous operation under worst case conditions of the following 3 parameters be avoided: Minimum input voltage, maximum output power, and maximum temperature.

## Battery Charging and Temperature Sensor



Fig. 9
Trickle charge voltage versus temperature for different temperature coefficients ( $V_{0}$ safe with disconnected sensor)

The battery charger models exhibit the option M1 and have been designed to charge lead-acid batteries. The R-input allows for connecting a battery-specific temperature sensor, which provides temperature controlled adjust of the trickle charge voltage. This optimizes charging as well as battery life time. Depending upon the cell voltage and the temperature coefficient of the battery, different sensor types are available; see Accessories.

Note: Parallel operation is not possible, if the temperature sensor is connected to the paralleled outputs Vo+, as the sensor eliminates the output voltage droop.
However, it is possible to insert bleeding resistors in the Vo+ output lines of each converter in order to create a droop of approx. $0.6 \mathrm{~V} @ I_{\mathrm{o}}$ nom for 24 V outputs (1.2 $\mathrm{V} @ I_{\mathrm{o}}$ nom for 48 V outputs), but this creates considerable power losses.


Fig. 10
Schematic circuit diagram of a system with battery backup and temperature-controlled charging.

## Electromagnetic Compatibility (EMC)

## Electromagnetic Immunity

The W Series has been successfully tested to the following specifications:
Table 8: Electromagnetic immunity (type tests)

| Phenomenon | Standard | Level | Coupling mode ${ }^{1}$ | Value applied | Waveform | Source imped. | Test procedure | In oper. | Perform. ${ }^{2}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Electrostatic discharge (to case) | $\begin{aligned} & \text { IEC/EN } \\ & 61000-4-2 \end{aligned}$ | $4^{3}$ | contact discharge | $8000 \mathrm{~V}_{\mathrm{p}}$ | 1/50 ns | $330 \Omega$ | 10 positive and 10 negative discharges | yes | A |
|  |  |  | air discharge | $15000 \mathrm{~V}_{\mathrm{p}}$ |  |  |  |  |  |
| Electromagnetic field RF | $\begin{array}{\|l\|} \hline \text { IEC/EN } \\ 61000-4-3 \end{array}$ | $3^{4}$ | antenna | $10 \mathrm{~V} / \mathrm{m}^{4}$ | AM 80\% <br> 1 kHz | n.a. | $80-1000 \mathrm{MHz}$ | yes | A |
|  | ENV 50204 | 3 | antenna | $10 \mathrm{~V} / \mathrm{m}$ | 50\% duty cycle, 200 Hz repet. frequ. | n.a. | $900 \pm 5 \mathrm{MHz}$ | yes | A |
|  | IEC/EN 61000-4-3 <br> (EW models) | 5 | antenna | $20 \mathrm{~V} / \mathrm{m}$ | 80\% AM, 1 kHz | n.a. | $800-1000 \mathrm{MHz}$ | yes | A |
|  |  |  |  | 10 Vm |  |  | $1400-2100 \mathrm{MHz}$ |  |  |
|  |  |  |  | $5 \mathrm{~V} / \mathrm{m}$ |  |  | $2100-2500 \mathrm{MHz}$ |  |  |
| Electrical fast transients/burst | $\begin{array}{\|l\|} \hline \text { IEC/EN } \\ 61000-4-4 \end{array}$ | $3^{6}$ | capacitive, o/c | $\pm 2000 \mathrm{~V}_{\mathrm{p}}$ | bursts of $5 / 50 \mathrm{~ns}$, 5 kHz over 15 ms , burst period: 300 ms | $50 \Omega$ | 60 s positive + 60 s negative transients per coupling mode | yes | A |
|  |  | $4^{6}$ | $\pm i / c,+i-i$ <br> direct coupling | $\pm 4000 \mathrm{Vp}^{6}$ |  |  |  |  |  |
| Surges | $\begin{array}{\|l\|} \hline \text { IEC/EN } \\ 61000-4-5 \end{array}$ | $3^{7}$ | +i/c, -i/c | $\pm 2000 \mathrm{~V}_{\mathrm{p}}$ | 1.2/50 $\mu \mathrm{s}$ | $12 \Omega$ | 5 pos. and 5 neg. surges per coupling mode | yes | B |
|  |  | $2^{7}$ | +i/-i | $\pm 1000 \mathrm{~V}_{\mathrm{p}}$ | 1.2/50 $\mu \mathrm{s}$ | $2 \Omega$ |  |  |  |
| Conducted disturbances | $\begin{array}{\|l\|} \mid \text { IEC/EN } \\ 61000-4-6 \end{array}$ | $3^{8}$ | i, o, signal wires | $\begin{gathered} 10 \mathrm{VAC} \\ (140 \mathrm{~dB} \mu \mathrm{~V}) \end{gathered}$ | $\begin{gathered} \text { AM 80\% } \\ 1 \text { kHz } \end{gathered}$ | $150 \Omega$ | $0.15-80 \mathrm{MHz}$ | yes | A |
| Surges <br> (EW models) | $\begin{array}{\|l\|} \hline \text { IEC/EN } \\ \text { 50155:2001 } \end{array}$ | wave $A^{9}$ | +i/c, -i/c | $1800 \mathrm{~V}_{\mathrm{p}}$ | 5/50 $\mu \mathrm{s}$ | $5 \Omega$ | 5 pos. and 5 neg. pulses | yes | B |

1 i = input, o = output, c = case.
$2 A=$ Normal operation, no deviation from specifications, $B=$ Normal operation, temporary loss of function or deviation from specs. possible.
3 Exceeds EN 50121-3-2:2006 table 9.3 and EN 50121-4:2006 table 1.4.
4 EW models: $20 \mathrm{~V} / \mathrm{m}$, which corresponds to EN 50121-3-2:2006 table 9.1 and exceeds EN 50121-4:2006 table 1.1.
5 EW models only. Corresponds to EN 50121-3-2:2006 table 9.2 and EN 50121-4:2006 table 1.2 (compliance with digital mobile phones).
6 EW models are tested to level $3\left( \pm 2000 V_{p}\right.$ ), corresponding to EN 50121-3-2:2006 table 7.2 and EN 50121-4:2006 table 2.2.
7 Complies with EN 50121-3-2:2006 table 7.3 and EN 50121-4:2006 table 2.3.
8 Corresponds to EN 50121-3-2:2006 table 8.1 and EN 50121-4:2006 table 3.1 (radio frequency common mode).
9 Corresponds to EN 50121-3-2:2000. Covers EN 50155:1995, RIA12, direct transients, wafeform D (EW models only).

## Emissions

Table 9: Electromagnetic emissions for LW models with feature E: (type tests with LWN1801-6E)

| Phenomenon | Standards | Conditions | Results |
| :--- | :--- | :--- | :--- |
| Harmonics | EN 61000-3-2:2000 | $V_{\mathrm{i}}=230 \mathrm{~V}, V_{\text {o nom }}, I_{\text {onom }}$ | Class A, D |
| Voltage fluctuation and flicker | EN 61000-3-3 $+\mathrm{A1:2001}$ | $V_{\mathrm{i}}=230 \mathrm{~V}, V_{\text {o nom }}, I_{\text {onom }}$ | Complied |



Fig. 11a
Conducted emissions for LW models without feature E: Typical disturbances (quasi-peak) at the input according to EN 55022, measured at $V_{\mathrm{i} \text { nom }}$ and $I_{\text {o nom. }}$.

Fig. 12a
Conducted emissions of LW models with feature E:
Disturbances (quasi-peak) at the phase input according to EN 55022, measured at $V_{\text {i nom }}$ and $I_{o n o m}$. (LWN1801-6E)


Fig. 11b
Radiated emissions for LW models without feature E:
Typical electromagnetic field strength (quasi-peak) according to EN 55014, measured at $V_{\mathrm{i} \text { nom }}$ and $I_{\mathrm{o} \text { nom. }}$.


Fig. 12b
Radiated emissions measured according to EN 55022:2001 for LW models with feature E (LWN1801-6E, antenna 3 m distance, horizontal polarized)


Fig. 13a
Conducted emissions of EW models:
Disturbances (peak) at the phase input according to EN 55011, measured at $V_{\mathrm{i} \text { nom }}$ and $I_{\mathrm{o}}$ nom. (EWN2660-0)


Fig. 13b
Radiated emissions of EW models, measured at $V_{\mathrm{i} \text { nom }}, I_{0}$ nom, accord. to EN 55011, antenna 3 m distance, (EWN2660-0)

## External EMC Filter for Models with Feature E

An external EMC filter can be connected to the inputs lines of the converter. However, a small choke has to be included in the phase line to avoid interferences between the internal and external filter, which would cause dramatically increased low harmonics.


Fig. 14a
Conducted emissions of LW models with external filter: Disturbances (peak) at the phase input according to EN 55011/55022, at $V_{\mathrm{i}}=230 \mathrm{VAC}, I_{\text {o nom }}$ (LWN1701-6E).


Fig. 14b
Conducted emissions of LW models with feature E: Disturbances (average) at the phase input according to EN 55011/55022, at $V_{\mathrm{i}}=230$ VAC, Io nom (LWN1701-6E).

Fig. 14a and 14b show the conducted emissions smoothed by an external filter. The standards EN 55011 and 55022 define limits for conducted (quasi)peak and conducted average emissions. In general the limits for average emissions are more difficult to meet.
The figure below shows the used external filter configuration consisting of the inlet filter KMF1.1241.11 (4 A, Schurter www.schurter.com) and the decopling choke EPCOS B82111B0000C018, $11 \mu \mathrm{H}, 4 \mathrm{~A}, 6 \times 20 \mathrm{~mm}$.
Note: This filter allows for connection of an IEC inlet and is available with 1 or 2 incorporated fuses.

A similar filter with AMP terminals ( $6.3 \times 20 \mathrm{~mm}$ ) is also available (Schurter FMLB 5500.2028).


Fig. 15a
External filter to reduce conducted emissions of LW models with feature $E(L 1=L 2=1.6 \mathrm{mH}, C x=47 \mathrm{nF}$, $C y=2.2 n F)$

Fig. 15b
External inlet filter

## Immunity to Environmental Conditions

Table 10: Mechanical stress and climatic

| Test method |  | Standard | Test conditions |  | Status |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Cab | Damp heat steady state | IEC/EN 60068-2-78 <br> MIL-STD-810D sect. 507.2 | Temperature: Relative humidity: Duration: | $\begin{aligned} & 40 \pm 2{ }^{\circ} \mathrm{C} \\ & 93^{+2 /-3} \% \\ & 56 \text { days } \end{aligned}$ | Converter <br> not operating |
| Kb | Salt mist, cyclic (sodium chloride NaCl solution) | IEC/EN 60068-2-52 | Concentration: <br> Duration: <br> Conditions: <br> Storage duration: | $5 \%\left(30^{\circ} \mathrm{C}\right)$ <br> 2 h per cycle $40^{\circ} \mathrm{C}, 93 \%$ rel. humidity 3 cycles of 22 h | Converter <br> not operating |
| Eb | Bump (half-sinusoidal) | IEC/EN 60068-2-29 <br> MIL-STD-810D sect. 516.3 | Acceleration amplitude: Bump duration: 6000 bumps: | $25 \mathrm{~g}_{\mathrm{n}}=245 \mathrm{~m} / \mathrm{s}^{2}$ <br> 11 ms <br> 1000 in each direction | Converter not operating, wall-mounted ${ }^{1}$ |
|  |  |  | Acceleration amplitude: <br> Bump duration: <br> 6000 bumps: | $\begin{aligned} & 10 \mathrm{~g}_{\mathrm{n}}=98.1 \mathrm{~m} / \mathrm{s}^{2} \\ & 11 \mathrm{~ms} \\ & 1000 \text { in each direction } \end{aligned}$ | Converter not operating, on DIN-rail ${ }^{2}$ |
| Fc | Vibration (sinusoidal) | IEC/EN 60068-2-6 <br> MIL-STD-810D sect. 514.3 | Acceleration amplitude and frequency (1 Octave/min): Test duration: | $\begin{aligned} & 0.35 \mathrm{~mm}(10-60 \mathrm{~Hz}) \\ & 5 \mathrm{~g}_{\mathrm{n}}=49 \mathrm{~m} / \mathrm{s}^{2}(60-2000 \mathrm{~Hz}) \\ & 7.5 \mathrm{~h}(2.5 \mathrm{~h} \text { each axis }) \end{aligned}$ | Converter operating, wall-mounted ${ }^{1}$ |
|  |  |  | Acceleration amplitude and frequency (1 Octave/min): Test duration: | $\begin{aligned} & 0.25 \mathrm{~mm}(10-60 \mathrm{~Hz}) \\ & 2 \mathrm{~g}_{\mathrm{n}}=19 \mathrm{~m} / \mathrm{s}^{2}(60-2000 \mathrm{~Hz}) \\ & 7.5 \mathrm{~h}(2.5 \mathrm{~h} \text { each axis }) \end{aligned}$ | Converter operating, on DIN-rail ${ }^{2}$ |
| Ea | Shock (half-sinusoidal) | IEC/EN 60068-2-27 <br> MIL-STD-810D sect. 516.3 | Acceleration amplitude: <br> Bump duration: <br> Number of bumps: | $\begin{aligned} & 50 \mathrm{~g} \mathrm{~g}_{\mathrm{n}}=490 \mathrm{~m} / \mathrm{s}^{2} \\ & 11 \mathrm{~ms} \\ & 18 \text { (3 in each direction) } \end{aligned}$ | Converter not operating, wall-mounted ${ }^{1}$ |
| -- | Shock | EN 50155/EN $61373^{3}$ sect. 10, class A and B body mounted ${ }^{4}$ | Acceleration amplitude: <br> Bump duration: <br> Number of bumps: | $\begin{aligned} & 5.1 \mathrm{~g}_{\mathrm{n}} \\ & 30 \mathrm{~ms} \\ & 18 \text { (3 in each direction) } \end{aligned}$ | Converter operating, on DIN-rail ${ }^{2}$ |
| Fh | Random vibration broad band, digital control and guidance | IEC/EN 60068-2-64 | Acceleration spectral density: <br> Frequency band: <br> Acceleration magnitude: <br> Test duration: | $\begin{aligned} & 0.05 \mathrm{~g}_{\mathrm{n}}{ }^{2} / \mathrm{Hz} \\ & 8-500 \mathrm{~Hz} \\ & 4.9 \mathrm{~g}_{\mathrm{n} \text { rms }} \\ & 3 \mathrm{~h}(1 \mathrm{~h} \text { each axis }) \end{aligned}$ | Converter operating, wall-mounted ${ }^{1}$ |
| Fda | Random vibration wide band, high reproducibility | IEC/EN 60068-2-35 | Acceleration spectral density: Frequency band: Acceleration magnitude: Test duration: | $\begin{aligned} & 0.01 \mathrm{~g}_{\mathrm{n}}{ }^{2} / \mathrm{Hz} \\ & 20-500 \mathrm{~Hz} \\ & 2.2 \mathrm{~g}_{\mathrm{n} \text { rms }} \\ & 1.5 \mathrm{~h}(0.5 \mathrm{~h} \text { each axis) } \end{aligned}$ | Converter operating, mounted on a DIN-rail ${ }^{2}$ |
| -- | Simulated long life time testing at increased random vibration levels | EN 50155/EN 61373 ${ }^{3}$ sect. 8 and 9 , class B body mounted ${ }^{3}$ | Acceleration spectral density: Frequency band: Acceleration magnitude: Test duration: | $\begin{aligned} & 0.01 \mathrm{~g}_{\mathrm{n}}{ }^{2} / \mathrm{Hz} \\ & 5-150 \mathrm{~Hz} \\ & 0.8 \mathrm{~g}_{\mathrm{n} \text { rms }} \\ & 1.5 \mathrm{~h}(0.5 \mathrm{~h} \text { each axis }) \end{aligned}$ | Converter operating, mounted on a DIN-rail ${ }^{2}$ |

1 Wall-mounted with brackets UMB-W [HZZ00618]; see Accessories
2 Fastened on a DIN-rail with 2 additional DIN-rail fixing brackets DMB-EWG, see Accessories. This covers also wall-mounting with brackets, because wall mounting performs better in vibration test.
4 EW models (railway standards)
4 Body mounted = chassis of a railway coach

## Temperatures

Table 11: Temperature specifications, valid for an air pressure of 800-1200 hPa (800-1200 mbar)

| Model |  |  | LW models -6 |  | EW models -0 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Characteristics |  | Conditions | min | max | min | max |  |
| $T_{\text {A }}$ | Ambient temperature | Converter operating ${ }^{1}$ | -40 | 60 | -40 | $70^{3}$ | ${ }^{\circ} \mathrm{C}$ |
| $T_{\text {c }}$ | Case temperature |  | -40 | $90^{2}$ | -40 | $90^{3}$ |  |
| $T_{\text {S }}$ | Storage temperature | Not operating | -40 | 100 | -40 | 100 |  |

1 See Thermal Considerations
${ }^{2}$ See table $5 P_{0}$ derating
${ }^{3}$ Mounted in vertical position

## 125, 250 Watt AC-DC and DC-DC DIN-Rail Converters

Changing the Shape of Power
Failure Rates
Table 12: MTBF

| Values at specified <br> case temperature | Module types | Ground benign | Ground fixed |  | Ground mobile | Unit |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| MTBF $^{1}$ |  | $40^{\circ} \mathrm{C}$ | $40^{\circ} \mathrm{C}$ | $70^{\circ} \mathrm{C}$ | $50^{\circ} \mathrm{C}$ |  |
|  | LWR1601 | 892000 | 180000 | 197000 | 68000 | h |
|  | LWN1601 | 644000 | 131000 | 72000 | 51000 |  |
|  | LWN2660 | 522000 | 101000 | 55000 | 38000 |  |

1 Calculated in accordance with MIL-HDBK-217E, notice 2.

## Mechanical Data



Fig. 16
Case W01
EWN/LWN: weight approx. 1400 g
EWR/LWR: weight approx. 1200 g
Case designed by ATP, Munich.

## Safety and Installation Instructions

## Terminal Allocation

The terminal allocation tables define the electrical potential of the converters.


Fig. 17a
View of the input terminals (cage clamp style)


Fig. 17b
View of the output terminals (cage clamp style)
Table 13a: Input terminals of LW models

| Pin no. | Pin designation | Electrical determination |
| :--- | :--- | :--- |
| 1 | $\oplus$ | Protective earth PE |
| 2 | $\mathrm{~N} \bar{\sim}$ | Input neutral, DC negative |
| 3 | $\mathrm{~L} \bar{\sim}$ | Input phase, DC positive |

Table 13b: Input terminals of EW models

| Pin no. | Pin designation | Electrical determination |
| :--- | :--- | :--- |
| 1 | $\cap$ | Protective earth PE |
| 2 | Vi- | Input negative |
| 3 | Vi+ | Input positive |

Table 13c: Terminal allocation output side

| Pin no. | Pin des. | Single output | Double output |
| :--- | :--- | :--- | :--- |
| 1 |  | Functional <br> earth to load | Functional <br> earth to load |
| 2 | + | Output positive | Output 1 positive |
| 3 | + | Output positive | Output 1 positive |
| 4 | - | Output negative | Output 1 negative |
| 5 | - | Output negative | Output 1 negative |
| 6 | + | Output positive | Output 2 positive |
| 7 | + | Output positive | Output 2 positive |
| 8 | - | Output negative | Output 2 negative |
| 9 | - | Output negative | Output 2 negative |
| 10 | AUX | Option | Option |
| 11 |  | Functional <br> earth to load | Functional <br> earth to load |

## Installation Instructions

The converters of the W Series are components, intended exclusively for inclusion within other equipment by professional installers. Installation must strictly follow the national safety regulations in compliance with the enclosure, mounting, creepage, clearance, casualty, markings and segregation requirements of the end-use application.
DIN-rail mounting is possible with the built-in snap-fit device on a DIN-rail. This fulfills the mechanical transport requirements as per ETSI 300019-1-2, class 2 (vertical).
To fulfill the requirements of IEC 721-3-2, class 2.1 (vertical), 2 additional fixing brackets DMB-EWG [formerly HZZO0624] (see Accessories) must be fitted on the bottom side of the DIN-rail. For heavy duty railway applications, we recommend installing all 4 fixing brackets DMB-EWG.
Wall mounting is possible with the wall-mounting brackets UMB-W [HZZ00618] (see Accessories). This complies with IEC 721-3-2, class 2.2 (vertical and horizontal).

Caution: Install the converters vertically, and make sure that there is sufficient airflow available for convection cooling. The minimum space to the next device should be: top/bottom: 30 mm , left/right: 20 mm .
The converters of the W Series are class I equipment: Input terminal $1(\Theta)$ and the output terminals 1 and 11 ( ) are reliably connected to the case. For safety reasons it is


Fig. 18a
Snap-fit mounting to DIN-Rail.


Fig. 18b
Dismounting from DIN-rail. Use proper tool (min. 3 mm screwdriver) and adequate force.
essential to connect the input terminal $1\left({ }^{-}\right)$to the protective earth of the supply system. Output terminals 1 and 11 can be used to connect the output voltage(s) or the load to functional earth.

The phase input ( $\mathrm{L} \approx$ or $\mathrm{Vi}+$ ) is internally fused; see Input Fuse.


Fig. 19
Cage clamp terminals. Use 0.5 to $2.5 \mathrm{~mm}^{2}$ (AWG 20 to 12) solid or stranded wires depending on local requirements.

This fuse is designed to break an overcurrent in case of a malfunction of the converter and is not customer-accessible.
External fuses in the wiring to one or both input lines ( $\mathrm{L} \bar{\sim}$ and/ or $\mathrm{N} \bar{\sim}$ ) may be necessary to ensure compliance with local requirements. A built-in second fuse in the neutral path is available as option $F$.

A second fuse in the wiring to the neutral terminal $\mathrm{N} \approx$ or option $F$ is needed if:

- Local requirements demand an individual fuse in each source line
- Neutral and earth impedance is high or undefined
- Phase and neutral of the mains are not defined or cannot be assigned to the corresponding terminals ( $\mathrm{L} \approx$ to phase and $\mathrm{N} \sim$ to neutral).
Models with Option F: Caution! Double-pole/neutral fusing.
If the converters operate at source voltages above 250 VDC , an external fuse or a circuit breaker at system level should be installed.


## Caution:

- Installation must strictly follow the national safety regulations.
- Do not open this apparatus!


## Protection Degree

The protection degree of the converters is IP 20. Protective covers over input and output terminals are available on request; see Accessories.

## Cleaning Agents

Any penetration of liquid or foreign solid objects is to be prevented, since the converters are not hermetically sealed.

## Standards and Approvals

The converters of the LW Series with feature E were approved according to IEC/EN 60950-1:2001 (IEC/EN 60950 for models without E), IEC 61010-1:C11:2002 (models without E: IEC 61010-1), and EN 50178:1997 (with and without E).
The converters are UL508-listed and approved according to UL 60950-1 and CAN/CSA C22.2 No. 60950-1.

Safety approvals for EW models are in process.
The converters have been designed in accordance with said standards for:

- Class I equipment
- Power supply for building-in, vertical mounting on 35 mm DIN-rail or on a wall
- Overvoltage category II (III for 110 VAC supply)
- Basic insulation between input and case, based on 250 VAC
- Double or reinforced insulation between input and output, based on 250 VAC and 350 VDC.
- Functional insulation between outputs and case.
- Functional insulation between outputs.
- Pollution degree 3 environment (AC-input) and degree 2 (DC input).
The converters are subject to manufacturing surveillance in accordance with the above mentioned UL standards and with ISO9001:2000.


## Railway Applications

The W Series converters have been designed observing the railway standards EN 50155 and EN 50121. All boards are coated with a protective lacquer.
The EW Series is particularily suitable for connection to 110 V railway batteries.

Table 14: Isolation

| Characteristic |  | Input to case <br> and output(s) | Output(s) to <br> case | Output 1 to <br> output 2 and AUX | Unit <br> Electric <br> strength <br> test |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  | Factory test $\geq 1 \mathrm{~s}$ <br> to factory test | $2.8^{1}$ | 1.4 | 0.5 | kVDC |
| Insulation resistance | 2.0 | 1.0 | 0.35 | kVAC |  |

1 In accordance with EN 50116 and IEC/EN 60950-1, subassemblies are pretested with 4.2 kVDC.
2 Tested at 500 VDC.

## Isolation

The electric strength test is performed in the factory as routine test in accordance with EN 50116 and IEC/EN 60950 and should not be repeated in the field. Power-One will not honor any warranty claims resulting from electric strength field tests.

## Leakage Currents in AC-DC Operation

Leakage currents flow due to internal leakage capacitance and RFI suppression Y-capacitors. The current values are proportional to the mains voltage and nearly proportional to the mains frequency. They are specified at maximum operating input voltage where phase, neutral, and protective earth are correctly connected as required for class I equipment.
Leakage current may exceed 3.5 mA , if $f_{\mathrm{i}}>63 \mathrm{~Hz}$.

## Safety of Operator-Accessible Output Circuits

If the output circuit of a converter is operator accessible, it shall be a SELV circuit according to IEC/EN 60950 related safety standards.
The converters have SELV output circuits up to an output voltage of 57.5 V . However, if the isolated outputs are connected to another voltage source or connected in series with a total of $>57.5 \mathrm{~V}$ the outputs are hazardous.
It is the sole responsibility of the installer to ensure the compliance with the relevant and applicable safety regulations.

## LED Indicator

A green LED is activated, when the output voltage $V_{0}$ is within the normal operating tolerance band.

Note: This LED is also activated, when the converter is not powered by the input, but a loaded battery is connected to the output.

## Description of Options

E designates LW models with improved EMC performance. Refer to the EC Declaration of Conformity (last page). Feature $E$ is standard for new designs.
Single options D1, D2, D5, R are available (as single choice options) on the AUX terminal (10), referenced to Vo-.

Option M1 and M2 designate a combination of several options accessible via a D-SUB connector. Option M1 includes the function SD (shutdown).
Note: In double-output models, the options D1, D5, R, and SD concern only output 2 .

## Single Options Using the AUX Pin

The connection is shown in the figure below. For the description refer to Adjustment of $V_{0}$ or $V_{02}$ (next section).


Fig. 20
Connection of adjust resistors or an external voltage source to adjust the output voltage $V_{0}$ or $V_{02}$ (option M1 or M2 not fitted)

## Multiple Options M1 or M2 via D-SUB Connector

The option board is suitable for applications, where several options are needed. Option M1 is standard for battery charger models, option M2 is suitable for applications without battery or for simple applications with battery.

Table 15: Pin allocation of the 9 pin D-SUB connector

| Pin | Designation | Description |
| :---: | :---: | :---: |
| 1 | GND1 ${ }^{1}$ | System ground / common signal return |
| 2 | R | R input ${ }^{3}$ |
| 3 | VCC ${ }^{2}$ | Positive supply voltage ( $\approx$ output 2) |
| 4 | D1 | Output voltage monitor $V_{0 \text { olow D1 }}{ }^{3}$ |
| 5 | D5 | Output 2 voltage monitor $V_{\text {o low D5 }}{ }^{3}$ |
| 6 | SD | Shutdown ${ }^{3}$ |
| 7 | D-adj | Adjustment of threshold values of D1 or D5 |
| 8 | D2 | Input voltage monitor $V_{\text {i low }}$ |
| 9 | Sys-OK | System okay (all outputs are okay) |

1 Do not connect GND1 (pin 1) with the neg. output (-)
2 Do not connect VCC (pin 3) with the positive output (+)
${ }^{3}$ In double-output models, R, D1, D5, SD concern output 2 only.

Table 16a: Option board M1

| Function | Description |
| :---: | :--- |
| R | Output voltage adjust ${ }^{1}$ |
| D 1 | Output voltage monitor $V_{\text {o low D1 }}{ }^{1}$ |
| D2 | Input voltage monitor $V_{\text {i low }}$ |
| D5 | Output 2 voltage monitor <br>  <br> (battery deep discharged): $V_{\text {o low }}$ D5 |
| Sys-OK | System okay |
| SD | Shutdown ${ }^{1}$ |
| D-adj | Adjustment of trigger values D1 and D5 |

1 In double-output models, only output 2 is concerned.
Table 16b: Option board M2

| Function | Description |
| :---: | :--- |
| $R$ | Output voltage adjust $^{1}$ |
| D2 | Input voltage monitor $V_{\text {i low }}$ |
| D5 | Output voltage monitor <br>  <br> (battery deep discharged): $V_{0}$ low D5 |
| D-adjust | Adjustment of trigger values D1 and D5 |

1 In double-output models, only output 2 is concerned.

## D2: Input Voltage Monitor (Power Fail)

D2 monitors the input voltage $V_{\mathrm{i}}$. When $V_{\mathrm{i}}$ drops below $65 \pm 3$ VAC or 92 VDC, the D2 signal output is high impedance (opencollector, max. 50 V ).
When $V_{i}$ is greater then said level, the signal output D 2 is conducting: $V_{\mathrm{D} 2}<1.5 \mathrm{~V}, I_{\mathrm{D} 2 \max }<50 \mathrm{~mA}$. D1: Output Voltage Monitor


## D1: Output Voltage Monitor

D1 is intended for monitoring the bus voltage of a batterybuffered system. It indicates that the system is powered from the battery and can for instance be used as a warning signal or to switch off a part of the load. When the output voltage $V_{0}$ (or $V_{02}$ ) is greater than $V_{0}$ low D1 specified in table 17, the D1 signal output is conducting: $V_{\mathrm{D} 1}<1.5 \mathrm{~V}, I_{\mathrm{D} 1 \text { max }}<50 \mathrm{~mA}$.
When $V_{0}$ is lower, the D1 signal output is high impedance (open-collector, max. 58.6 V). In double-output models, D1 monitors only output $2\left(V_{02}\right)$.
In applications without battery-buffering the D1 signal may not be suitable, since smaller dynamic load changes may cause D1 to trigger. For such applications, D5 with a trigger level of approx. $85 \%$ of $V_{\text {o nom }}$ should be chosen (e.g., for a bus voltage of 24.7 V : trigger level at 21 V ).

## D5: System Voltage Monitor (Battery Deep)

D5 monitors the output voltage $V_{0}$ ( $V_{02}$ in double-output models) or the lowest admissible voltage of a connected battery (battery deep discharge). The definition of D5 is similar to D 1 , but the trigger level is lower. When $V_{0}$ (or $V_{02}$ ) is greater than $V_{0 \text { low D5 }}$ specified in table 17, the D2 signal output is conducting:

$$
V_{D 5}<1.5 \mathrm{~V}, I_{\mathrm{D} 5 \max }<50 \mathrm{~mA} .
$$

When $V_{0}$ is lower, the D5 signal output is high impedance (open-collector, max. 58.6 V ). In double-output models, D5 monitors only output $2\left(V_{02}\right)$.
In systems without battery support, D5 signals that $V_{0}$ (or $V_{02}$ ) is going to drop below a safe value.
In battery-buffered systems, D5 indicates that the battery has reached its deepest discharge level prior to getting damaged.

The D5 signal can be used for instance to disable loads, save data, or to start a controlled switch-off of running processes.

Table 17: Options D1 and D5: Trigger and switch-on levels

| Model | $\begin{array}{c}\text { Battery } \\ V_{\text {Bat }} \\ \text { [V] }\end{array}$ | $\begin{array}{c}V_{\text {o low D1 }} \\ \text { trigger } \\ \text { [V] }\end{array}$ |  | $\begin{array}{c}\text { switch on } \\ \text { [V] }\end{array}$ | $\begin{array}{c}V_{\text {o low D5 }} \\ \text { trigger } \\ {[\mathrm{V}]}\end{array}$ |
| :--- | :---: | :---: | :---: | :---: | :---: |
| switch on |  |  |  |  |  |
| [V] |  |  |  |  |  |$]$

## Adjustment of Threshold Levels (D1/D5)

Pin 7 of the D-SUB connector allows for adjustment of the threshold levels of D1 and D5. Both levels are influenced by the voltage divider $R x / R y$. Resistor $R x$ to pin 3 (VCC) lowers the levels, whereas Ry to pin 1 (GND1) increases them (see fig. 22).


Fig. 22
Wiring to adjust both threshold levels of option D1 or D5

## SD: Shutdown

Reduces the output power to approx. 1 W , but the converter is not fully disabled. In a no-load condition, $V_{0}$ drops below 6.2 V ; see fig. 23. In double-output models, only output 2 is influenced.


Fig. 23
Output voltage versus output current, while the shutdown is $\operatorname{activated}\left(V_{\mathrm{i}}=V_{\mathrm{i} \text { nom }}\right)$.

Table 18: Shutdown conditions

| Voltage $\boldsymbol{V}_{\text {sD }}$ on <br> shutdown pin | Result |
| :---: | :--- |
| $<0.7 \mathrm{~V}$ | Converter disabled ( $P_{\mathrm{o}}$ approx. 1 W ) |
| $\geq 2.0 \mathrm{~V}$ or open | Converter enabled |

## Sys-OK: Status

This function allows in a battery charger application for checking, whether the output is correctly following the external control signal at the R -input (coming for instance from the temperature sensor). The logic is shown in table 19.

Table 19: System OK (M1 with external battery sensor)

| System Status | Input | $V_{\text {control }}$ <br> sensor signal | $V_{\text {Bat }}$ <br> theoretical | $V_{\text {Bat }}$ <br> measured | Sys-OK <br> output |
| :--- | :---: | :---: | :---: | :---: | :---: |
| System OK | O.K. | 2.7 V | 27 V | 27 V | Low ohmic |
| Battery overchared / temp. sensor <br> defect / control voltage to high | O.K. | 2.7 V | 27 V | 28 V | High ohmic |
| Overload, converter cannot follow the <br> control signal | O.K. | 2.7 V | 27 V | 24 V | High ohmic |
| Output does not follow control signal, <br> since battery would be overcharged | O.K. | 3.0 V | 30 V | 27 V | High ohmic |
| System OK | O.K. | 2.5 V | 25 V | 25 V | Low ohmic |

## R: Adjustment of $V_{0}$ or $V_{o 2}$

The R input allows external adjustment of the output voltage in the range of $50 \%$ to $110 \% V_{\text {o nom }}$. Double-output models allow only adjustment of output 2 (connected to the terminals 6, 7, 8 and 9 ). This enables asymmetric output voltage configuration.
Adjustment can be achieved via a resistor or an external voltage source (in the range of $1.25-2.75 \mathrm{~V}$ ).

Note: If the R input is not connected: $V_{0}$ or $V_{02} \approx V_{0}$ nom.
a) Adjustment by an external resistor:

Resistor $R_{\text {ext1 }}$, connected between R (pin 2 ) and GND1 (pin 1) of the D-SUB connector or according to fig. 20.
$V_{0}=50-100 \% V_{\text {onom }} . \quad R_{\text {ext1 }} \approx 4 \mathrm{k} \Omega \cdot \frac{V_{0}}{V_{\text {onom }}-V_{0}}$
Resistor $R_{\text {ext2 }}$, connected between R (pin 2) and VCC (pin 3) of the D-SUB connector or according to fig. 20.

$$
V_{0}=100-110 \% V_{\text {o nom }} \cdot R_{\text {ext } 2} \approx 4 \mathrm{k} \Omega \cdot \frac{V_{0}-2.5 \mathrm{~V}}{2.5 \mathrm{~V} \cdot\left(V_{0} / V_{\text {onom }}-1\right)}
$$

Note: If the $R$ function is not included in M1 or M2, refer to figure 20 how to connect $R_{\text {ext1 }}$ or $R_{\text {ext2 }}$.
b) Adjustment by an external control voltage $V_{\text {ext }}(1.25-2.75$ V ), connected between R (pin 2) and GND (pin 1) of the DSUB connector or according to fig. 20.

$$
V_{\mathrm{ext}} \approx 2.5 \mathrm{~V} \cdot \frac{V_{\mathrm{o}}}{V_{\mathrm{onom}}} \quad V_{\mathrm{o}} \approx V_{\mathrm{onom}} \cdot \frac{V_{\mathrm{ext}}}{2.5 \mathrm{~V}}
$$

Caution: To prevent damage, $V_{\text {ext }}$ should not exceed 3 V , nor be negative.
Note: If longer wires are used to connect the R input at the D-SUB connector, the wiring to pin 1 (GND1) should be done as star point connection. If wired differently, the output voltage setting may be adversely affected.
In battery charging systems, an external battery temperature sensor (see Accessories) can be connected to optimize $V_{0}$. However, adjustment using the $R$ input (pin 2 of $D-S U B$ ) is possible as well. The above shown formulas are valid, but $V_{\text {o nom }}$ stands for the voltage with open R input ( $=V_{\text {o safe }}$ ).

## F: Built-in Second Fuse

A built-in second fuse in the neutral line provides safe phase-
to-phase connection at low mains voltages (e.g., USA $120 \mathrm{~V} /$ $208 \mathrm{~V} / 60 \mathrm{~Hz}$ systems).
The built-in second fuse also enables safe connection to the mains, where phase and neutral are not defined or cannot be identified, as e.g., in the case of plug and socket connection to the mains via German Schuko-plugs; see also Safety and Installation Instructions.
Option F limits the DC input voltage to $\leq 250 \mathrm{~V}$.

## Q: Reverse Polarity Protection

EW models have no bridge rectifier at the input. To provide reverse polarity protection, an additional diode can be fitted. However this lowers the efficiency by approximately $1 \%$.

## K2: System Connectors

For installation in systems using pre-assembled harnesses the converters are available with system connectors. They are ULlisted, approved for currents up to 15 A at -40 to $105^{\circ} \mathrm{C}$.
The mating system connectors with screw terminals and retainers are delivered together with every converter with option K2. Use max. $2.5 \mathrm{~mm}^{2}$ (AWG 12) solid or stranded wires, or max. $1.5 \mathrm{~mm}^{2}$ (AWG 14) stranded wires with crimp termination, stripped length 6 mm . Tightening torque of input/ output terminals: max. 0.79 Nm (7 lbs.in.).

## G: RoHS

RoHS compliant for all six substances.


Fig. 24
System connectors Option K2

## Accessories

UMB-W: Shock-Resistant Wall Mounting
Set of wall mounting brackets UMB-W [HZZOO618]
Content: 2 clamps, 4 countersunk screws M4, washers, and spring washers.


Fig. 25
Brackets UMB-W

Fig. 26
Wall mounting with
 mounting brackets UMB-W.

## DMB-EWG: DIN-Rail Fixing Brackets

For DIN-Rail vibration-proof fastening, use a set of brackets DMB-EWG (replacement for HZZOO624). For heavy-duty application 2 sets ( $=4$ brackets) are preferable.


Fig. 27
DIN-rail fixing bracket DMB-EWG.

## Protective Covers over Terminals

Set of plastic covers COVER-W [HZZ 01219]
Content: 2 covers to protect the input and output connector.


Fig. 28
Protective covers COVER-W

## Battery Temperature Sensor

To charge lead-acid batteries according to their temperature different types of temperature sensors are available, (see Battery Charging and Temperature Sensor in this data sheet and the Temperature Sensor data sheet at www.powerone.com).


Fig. 29
Temperature sensor

Table 20: Sensors for converters with standard $R$ input

| Battery <br> voltage <br> nom.[V] | Sensor <br> type | Cell <br> voltage <br> [V] | Cell temp. <br> coefficient <br> [mV/K] | Cable <br> length <br> [m] |
| :---: | :---: | :---: | :---: | :---: |
| 12 | S-KSMH12-2.27-30-2 | 2.27 | -3.0 | 2 |
| 24 | S-KSMH24-2.27-30-2 | 2.27 | -3.0 | 2 |
| 24 | S-KSMH24-2.27-35-2 | 2.27 | -3.5 | 2 |
| 24 | S-KSMH24-2.31-35-0 | 2.31 | -3.5 | 4.5 |
| 24 | S-KSMH24-2.35-35-2 | 2.35 | -3.5 | 2 |
| 48 | S-KSMH48-2.27-30-2 | 2.27 | -3.0 | 2 |
| 48 | S-KSMH48-2-27-35-2 | 2.27 | -3.5 | 2 |



Fig. 30
Connection of a temperature sensor

For additional information go to www.power-one.com.


[^0]:    Example: LWN2660-6ED2FK2G: Power factor corrected AC-DC converter, operating input voltage range $85-264$ VAC, 2 electrically isolated and individually regulated outputs, each providing $24.7 \mathrm{~V}, 5 \mathrm{~A}$, improved EMC performance, options D2, F, K2, and RoHS-compatible for all 6 substances.

