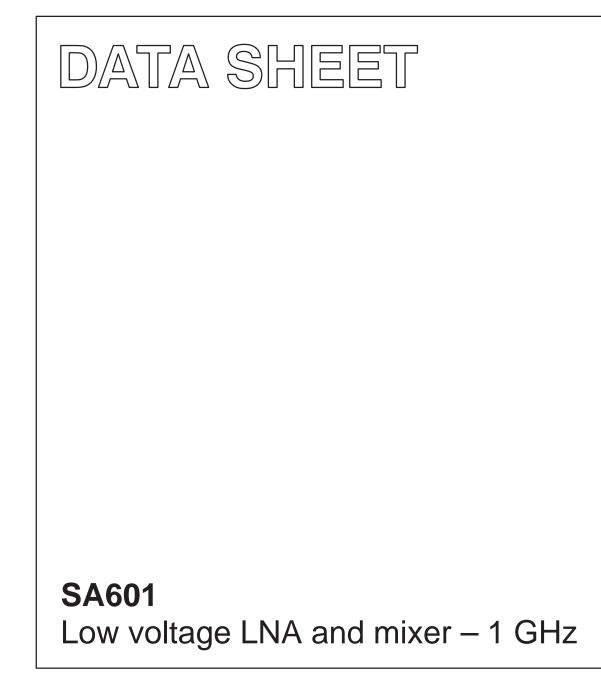
# INTEGRATED CIRCUITS



Product data Supersedes data of 1994 Dec 15 2004 Dec 14



PHILIPS



2004 Dec 14

# 1GHz low voltage LNA and mixer

#### DESCRIPTION

The SA601 is a combined RF amplifier and mixer designed for high-performance low-power communication systems from 800-1200MHz. The low-noise preamplifier has a 1.6dB noise figure at 900MHz with 11.5dB gain and an IP3 intercept of -2dBm at the input. The gain is stabilized by on-chip compensation to vary less than  $\pm 0.2dB$  over -40 to  $\pm 85^{\circ}$ C temperature range. The wide-dynamic-range mixer has a 9.5dB noise figure and IP3 of -2dBm at the input at 900MHz. The nominal current drawn from a single 3V supply is 7.4mA. The Mixer can be powered down to further reduce the supply current to 4.4mA.

#### FEATURES

- Low current consumption: 7.4mA nominal, 4.4mA with the mixer powered-down
- Outstanding LNA noise figure: 1.6dB at 900MHz
- High system power gain: 18dB (LNA + Mixer) at 900MHz
- Excellent gain stability versus temperature and supply voltage
- External >-7dBm LO can be used to drive the mixer

#### **PIN CONFIGURATION**

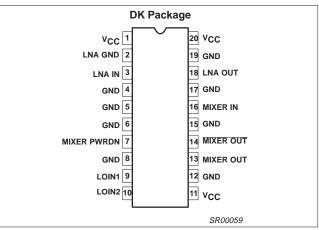


Figure 1. Pin Configuration

#### **APPLICATIONS**

- 900MHz cellular front-end (NADC, GSM, AMPS, TACS)
- 900MHz cordless front-end (CT1, CT2)
- 900MHz receivers

#### **ORDERING INFORMATION**

DESCRIPTION	TEMPERATURE RANGE	ORDER CODE	DWG #	
20-Pin Plastic Shrink Small Outline Package (Surface-mount, SSOP)	-40 to +85°C	SA601DK	SOT266-1	

#### **BLOCK DIAGRAM**

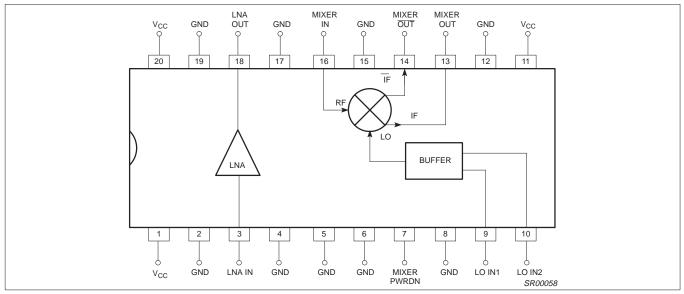


Figure 2. Block Diagram

SA601

#### **ABSOLUTE MAXIMUM RATINGS<sup>3</sup>**

SYMBOL	PARAMETER	RATING	UNITS
V <sub>CC</sub>	Supply voltage <sup>1</sup>	-0.3 to +6	V
V <sub>IN</sub>	Voltage applied to any other pin	-0.3 to (V <sub>CC</sub> + 0.3)	V
P <sub>D</sub>	Power dissipation, T <sub>A</sub> = 25°C (still air) <sup>2</sup> 20-Pin Plastic SSOP	980	mW
T <sub>JMAX</sub>	Maximum operating junction temperature	150	°C
P <sub>MAX</sub>	Maximum power input/output	+20	dBm
T <sub>STG</sub>	Storage temperature range	-65 to +150	۵°

NOTE:

Transients exceeding 8V on V<sub>CC</sub> pin may damage product.
 Maximum dissipation is determined by the operating ambient temperature and the thermal resistance, θ<sub>JA</sub>: 20-Pin SSOP = 110°C/W
 Pins 9 and 10 are sensitive to electrostatic discharge (ESD).

#### **RECOMMENDED OPERATING CONDITIONS**

SYMBOL	PARAMETER	RATING	UNITS
V <sub>CC</sub>	Supply voltage	2.7 to 5.5	V
T <sub>A</sub>	Operating ambient temperature range	-40 to +85	°C
TJ	Operating junction temperature	-40 to +105	°C

#### **DC ELECTRICAL CHARACTERISTICS**

 $V_{CC}$  = +3V,  $T_A$  = 25°C; unless otherwise stated.

SYMBOL	PARAMETER TEST CONDITIONS		LIMITS				
		TEST CONDITIONS	MIN TYP MAX				
100	Supply current			7.4		mA	
Icc		Mixer power-down input low		4.4			
V <sub>LNA–IN</sub>	LNA input bias voltage			0.78		V	
V <sub>LNA-OUT</sub>	LNA output bias voltage			2.1		V	
V <sub>MX–IN</sub>	Mixer RF input bias voltage			0.94		V	

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# 1GHz low voltage LNA and mixer

#### AC ELECTRICAL CHARACTERISTICS

 $V_{CC}$  = +3V,  $\ T_A$  = 25°C; LO\_{IN} = -7dBm @ 964MHz; unless otherwise stated.

SYMBOL	PARAMETER	TEST CONDITIONS		LIMITS		
	PARAMETER	TEST CONDITIONS	<b>-3</b> σ	TYP	<b>+3</b> σ	
S <sub>21</sub>	Amplifier gain	881MHz	10	11.5	13	dB
$\Delta S_{21}/\Delta T$	Gain temperature sensitivity	881MHz		0.003		dB/°C
$\Delta S_{21}/\Delta f$	Gain frequency variation	800MHz - 1.2GHz		0.01		dB/MHz
S <sub>12</sub>	Amplifier reverse isolation	881MHz		-20		dB
S <sub>11</sub>	Amplifier input match <sup>1</sup>	881MHz		-10		dB
S <sub>22</sub>	Amplifier output match <sup>1</sup>	881MHz		-10		dB
P <sub>-1dB</sub>	Amplifier input 1dB gain compression	881MHz		-16		dBm
IP3	Amplifier input third order intercept	f <sub>2</sub> – f <sub>1</sub> = 25kHz, 881MHz	-3.5	-2	-0.5	dBm
NF	Amplifier noise figure	881MHz	1.3	1.6	1.9	dB
VG <sub>C</sub>	Mixer voltage conversion gain: $R_P = R_L = 1k\Omega$	$f_S = 881MHz, f_{LO} = 964MHz, f_{IF} = 83MHz$	18.0	19.5	21.0	dB
PG <sub>C</sub>	Mixer power conversion gain: $R_P = R_L = 1k\Omega$	$f_S = 881MHz, f_{LO} = 964MHz, f_{IF} = 83MHz$	5.0	6.5	8.0	dB
S <sub>11M</sub>	Mixer input match <sup>1</sup>	881MHz		-10		dB
NFM	Mixer SSB noise figure	881MHz	8.0	9.5	11.0	dB
P <sub>-1dB</sub>	Mixer input 1dB gain compression	881MHz		-13		dBm
IP3 <sub>M</sub>	Mixer input third order intercept	f <sub>2</sub> – f <sub>1</sub> = 25kHz, 881MHz	-3.5	-2	-0.5	dBm
IP <sub>2INT</sub>	Mixer input second order intercept	881MHz		12		dBm
P <sub>RFM-IF</sub>	Mixer RF feedthrough	881MHz		-7		dB
P <sub>LO-IF</sub>	LO feedthrough to IF	881MHz		-25		dB
P <sub>LO-RFM</sub>	LO to mixer input feedthrough	881MHz		-38		dB
P <sub>LO-RF</sub>	LO to LNA input feedthrough 881MHz		-40		dB	
P <sub>LNA-RFM</sub>	LNA output to mixer input	881MHz		-40		dB
P <sub>RFM-LO</sub>	Mixer input to LO feedthrough	881MHz	881MHz -23			dB
LOIN	LO drive level	964MHz	İ	-7		dBm

NOTE:

1. Simple L/C elements are needed to achieve specified return loss.

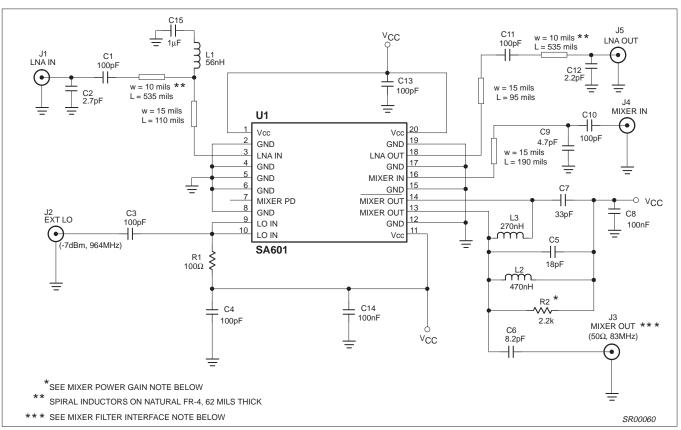


Figure 3. Application Circuit

# CIRCUIT TECHNOLOGY

#### LNA

**Impedance Match:** Intrinsic return loss at the input and output ports is 7dB and 9dB, respectively. With no external matching, the associated LNA gain is ~10dB and the noise figure is ~1.4dB. However, the return loss can be improved at 881MHz using suggested L/C elements (Figure 5) as the LNA is unconditionally stable.

Noise Match: The LNA achieves 1.6dB noise figure at 881MHz when  $S_{11} = -10$ dB. Further improvements in  $S_{11}$  will slightly decrease the NF and increase  $S_{21}$ .

Temperature Compensation: The LNA has a built-in temperature compensation scheme to reduce the gain drift to 0.003 dB/°C from  $-40^{\circ}$ C to  $+85^{\circ}$ C.

Supply Voltage Compensation: Unique circuitry provides gain stabilization over wide supply voltage range. The gain changes no more than 0.5dB when  $V_{CC}$  increases from 3V to 5V.

**LO Drive Level:** Resistor R1 can be replaced by an inductor of 4.7nH and C3 should be adjusted to achieve a good return loss at the LO port. Under this condition, the mixer will operate with less than -10dBm LO drive.

**IP3 Performance:** C9 between Pin 16 and ground can be removed to introduce 3dB mismatch loss, while improving the IP3 to +3dBm. The associated noise figure is 11dB.

#### Mixer

**Input Match:** The mixer is configured for maximum gain and best noise figure. The user needs to supply L/C elements to achieve this performance.

**Power Gain:** The gain can be increased by approximately 1.5dB by placing R2 across C7, instead of C5.

**Power Down:** The mixer can be disabled by connecting Pin 7 to ground. When the mixer is disabled, 3mA is saved.

**Power Combining:** The mixer output circuit features passive power combining (patent pending) to optimize conversion gain and noise figure performance without using extra DC current or degrading the IP3. For IF frequencies significantly different than 83MHz, the component values must be altered accordingly.

**Filter Interface:** For system integration where a high impedance filter of  $1k\Omega$  is to be cascaded at the mixer IF output, capacitors C5 and C6 need to be changed to 27pF and 1000pF, respectively.

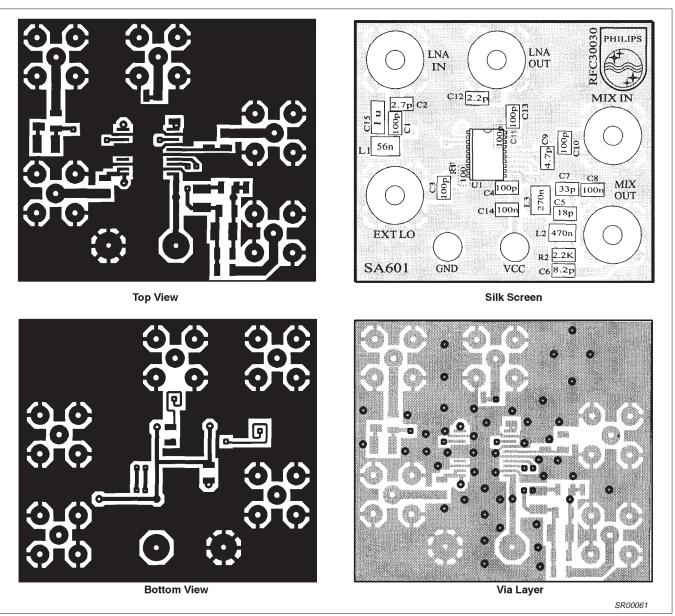


Figure 4. SA601 Demoboard Layout (Not Actual Size)

#### **TYPICAL PERFORMANCE CHARACTERISTICS**

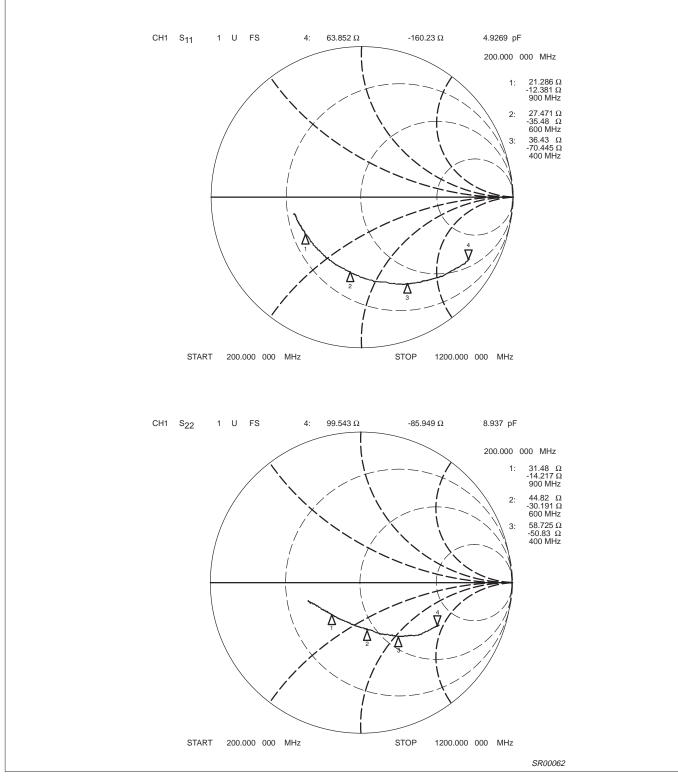


Figure 5. LNA Input and Output Match (at Device Pin)

SA601

#### TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

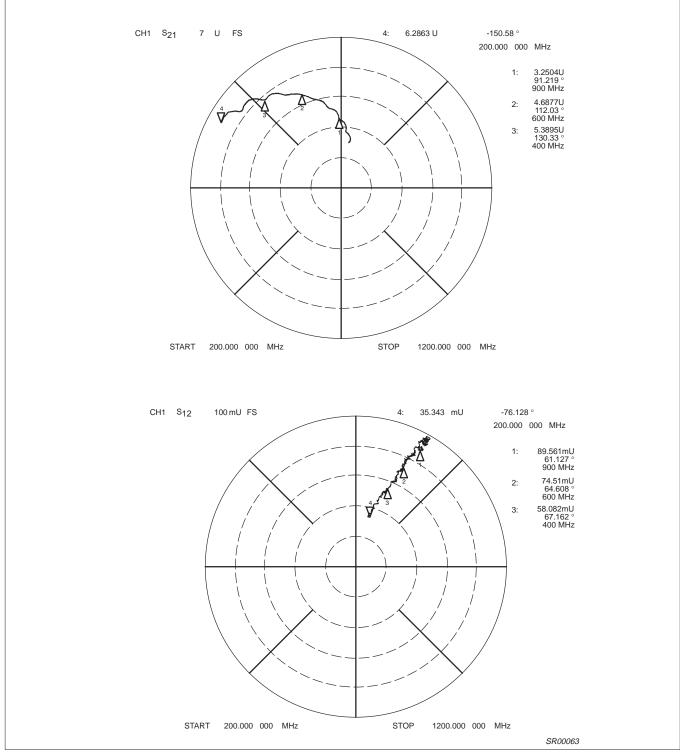


Figure 6. LNA Transmission and Isolation Characteristics (at Device Pin)

SA601

### TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

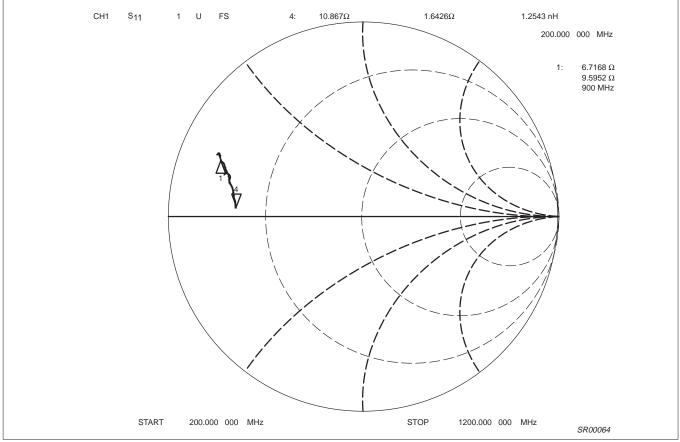
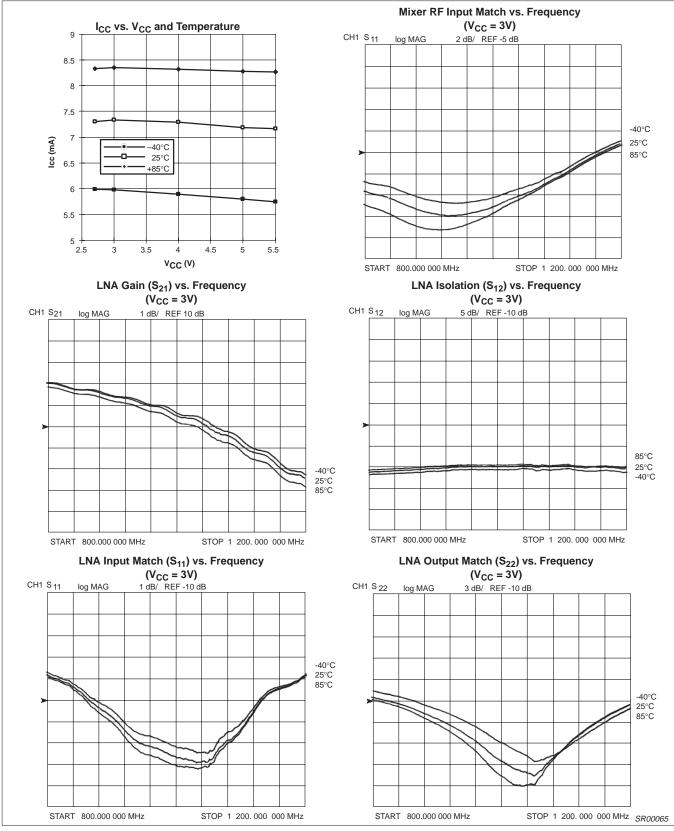


Figure 7. Mixer RF Input Match (at Device Pin)

Table 1.	Typical	LNA and	d Mixer	<b>S-Parameters</b>
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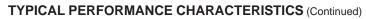
		Mixer			
f	S <sub>11</sub>	S <sub>22</sub>	S <sub>21</sub>	S <sub>12</sub>	S <sub>11</sub>
200MHz	63.852Ω – j 160.23Ω	99.543Ω – j 85.949Ω	6.2863U ∠ 150.58°	35.343mU ∠ 76.128°	10.867Ω + j 1.6426Ω
300MHz	44.879Ω – j 101.69Ω	73.387Ω – j 67.707Ω	5.8096U ∠ 140.47°	47.946mU∠71.169°	10.4Ω + j 3.4609Ω
400MHz	36.43Ω – j 70.445Ω	58.725Ω – j 50.83Ω	5.3895U ∠ 130.33°	58.082mU∠67.162°	10.067Ω + j 4.897Ω
500MHz	30.395Ω – j 48.393Ω	49.928Ω – j 38.813Ω	5.0428U ∠ 120.5°	66.44mU ∠ 66.388°	9.394Ω + j 6.0142Ω
600MHz	27.471Ω – j 35.48Ω	44.82Ω – j 30.191Ω	4.6877U∠112.03°	74.51mU ∠ 64.608°	8.8945Ω + j 7.2227Ω
700MHz	24.428Ω – j 25Ω	39.268Ω – j 24.502Ω	4.2409U ∠ 104.44°	82.235mU∠65.002°	8.1353Ω + j 8.1597Ω
800MHz	22.434Ω – j 17.255Ω	34.664Ω – j 18.59Ω	3.7491U ∠ 97.765°	86.582mU ∠ 62.743°	7.976Ω + j 9.1958Ω
900MHz	21.286Ω – j 12.381Ω	31.48Ω – j 14.217Ω	3.2504U ∠ 91.219°	89.561mU∠61.127°	6.7168Ω + j 9.5952Ω
1000MHz	20.261Ω – j 8.7109Ω	27.887Ω – j 10.77Ω	2.8785U ∠ 84.957°	95.135mU∠60.539°	6.2393Ω + j 10.271Ω
1100MHz	19.718Ω – j 6.252Ω	25.741Ω – j 8.2607Ω	2.5752U ∠ 82.893°	97.348mU ∠ 62.202°	6.0791Ω + j 10.571Ω
1200MHz	19.101Ω – j 4.9316Ω	23.584Ω – j 6.2715Ω	2.1386U ∠ 80.257°	96.558mU∠61.563°	5.8185Ω + j 10.288Ω

### SA601



#### TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

Figure 8. Typical Performance Characteristics (cont.)



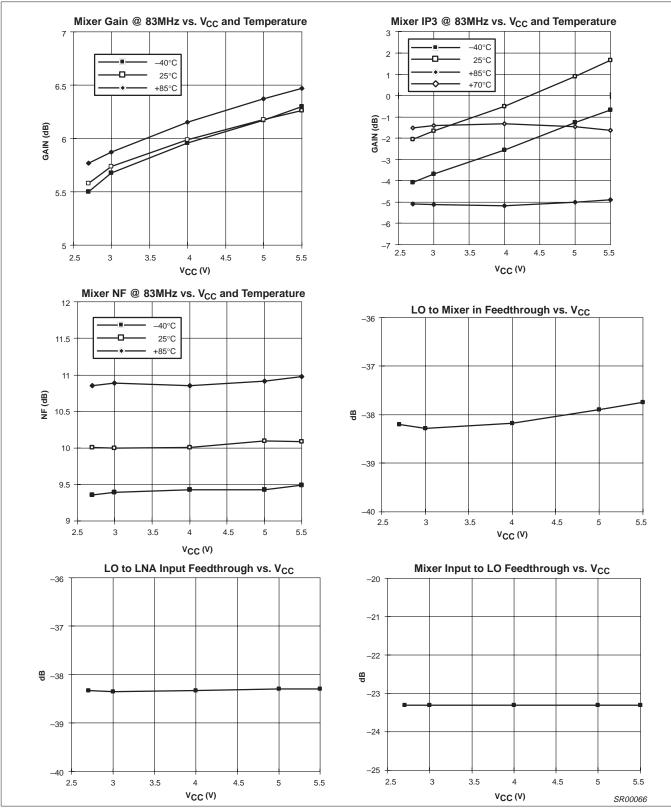


Figure 9. Typical Performance Characteristics (cont.)

### TYPICAL PERFORMANCE CHARACTERISTICS (Continued)

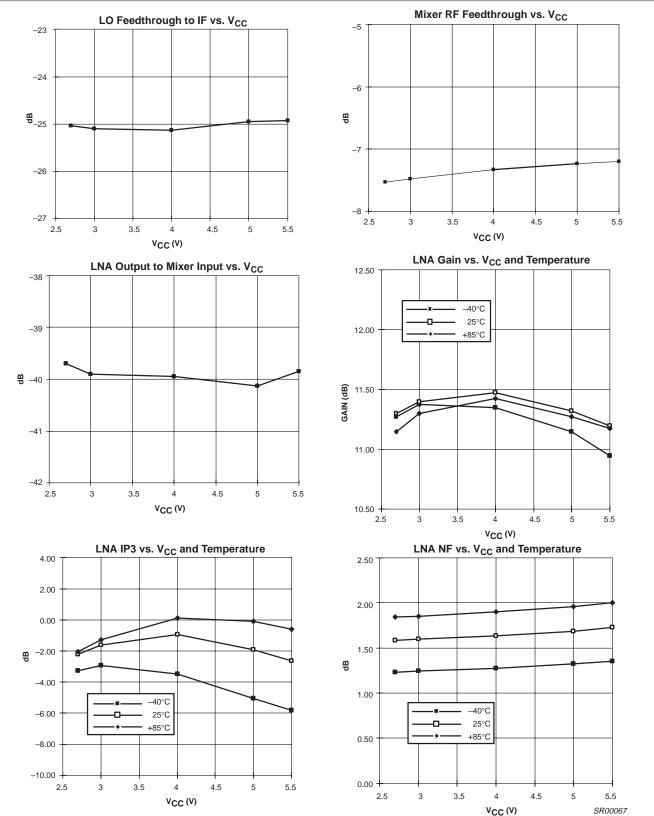
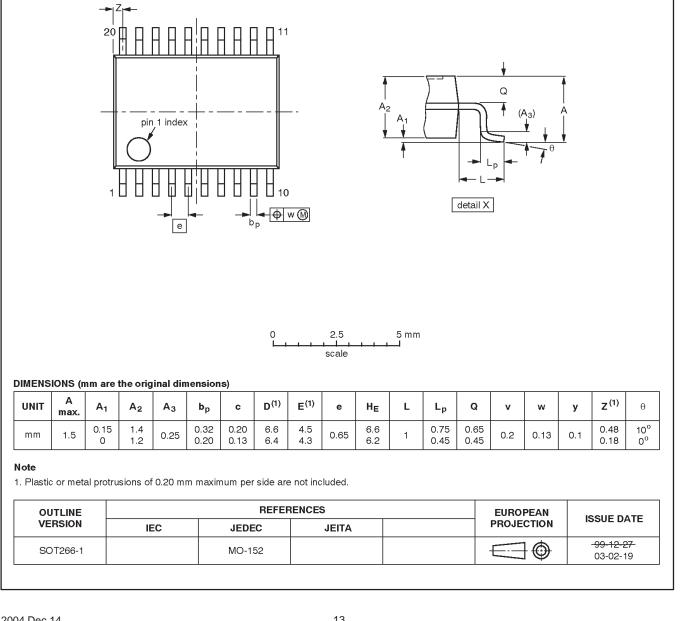


Figure 10. Typical Performance Characteristics (cont.)



SSOP20: plastic shrink small outline package; 20 leads; body width 4.4 mm

SOT266-1

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### **REVISION HISTORY**

Rev	Date	Description
_2	20041214	Product data (9397 750 14447); supersedes SA601 of 15 Dec 1994.
		Modifications:
		<ul> <li>Added package outline and legal information</li> </ul>
_1	19941215	Product specification

#### Data sheet status

Level	Data sheet status <sup>[1]</sup>	Product status <sup>[2] [3]</sup>	Definitions
I	Objective data	Development	This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice.
II	Preliminary data	Qualification	This data sheet contains data from the preliminary specification. Supplementary data will be published at a later date. Philips Semiconductors reserves the right to change the specification without notice, in order to improve the design and supply the best possible product.
111	Product data	Production	This data sheet contains data from the product specification. Philips Semiconductors reserves the right to make changes at any time in order to improve the design, manufacturing and supply. Relevant changes will be communicated via a Customer Product/Process Change Notification (CPCN).

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[2] The product status of the device(s) described in this data sheet may have changed since this data sheet was published. The latest information is available on the Internet at URL http://www.semiconductors.philips.com.

[3] For data sheets describing multiple type numbers, the highest-level product status determines the data sheet status.

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#### Contact information

For additional information please visit http://www.semiconductors.philips.com. Fax

sales.addresses@www.semiconductors.philips.com

For sales offices addresses send e-mail to:

Fax: +31 40 27 24825

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