## CYNSE70256 Network Search Engine

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### 1.0 Overview

Cypress Semiconductor Corporation's (Cypress's) CYNSE70256 network search engine (NSE) incorporates patent-pending Associative Processing Technology ${ }^{\text {TM }}$ (APT) and is designed to be a high-performance, pipelined, synchronous, 256K-entry NSE. The CYNSE70256 database entry size can be 72 bits, 144 bits, or 288 bits. In the 72 -bit entry mode, the size of the database is 128 K entries. In the 144 -bit mode, the size of the database is 64 K entries, and in the 288 -bit mode, the size of the database is 32 K entries. The CYNSE70256 is configurable to support multiple databases with different entry sizes. The 36-bit entry table can be implemented using the global mask registers (GMRs), building database size of 256 K entries with a single device.

The NSE can sustain 83 million transactions per second when the database is programmed or configured as 72 or 144 bits. When the database is programmed to have an entry size of 36 or 288 bits, the NSE will perform at 41.5 million transactions per second. Cypress's CYNSE70256 can be used to accelerate network protocols such as Longest-Prefix Match (CIDR), ARP, MPLS, and other layer 2, 3, and 4 protocols.

This high-speed, high-capacity NSE can be deployed in a variety of networking and communications applications. The performance and features of the CYNSE70256 make it attractive in applications such as Enterprise LAN switches and routers, and broadband switching and/or routing equipment that supports multiple data rates at OC-48 and beyond. The NSE is designed to be scalable in order to support network database sizes of up to 3840 K entries specifically for environments that require large network policy databases. Figure 2-1 on page 9 shows the block diagram for the CYNSE70256 device.

### 2.0 Features

## - 256K 36-bit entries in a single device

- 128K entries in 72-bit mode, 64K entries in 144-bit mode, 32K entries in 288-bit mode
- 83 million transactions per second in 72- and 144-bit configurations (CFGs)
- 41.5 million transactions in $\mathbf{3 6}$ - and 288 -bit configurations
- Searches any subfield in a single cycle
- Synchronous pipelined operation
- Up to fifteen NSEs can be cascaded
- When cascaded, database entries can range to 3840K 36-bit entries
- Multiple width tables in a single database bank
- Glueless interface to industry-standard SRAMs and/or SSRAMs
- Simple hardware instruction interface
- IEEE 1149.1 test access port
- 1.5 V core voltage supply
- $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ I/O voltage supply
- 388-pin BGA package.



### 3.0 Functional Description

The following subsections contain command (CMD) and DQ bus (command and databus), database entry, arbitration logic, pipeline and SRAM control, and full logic descriptions.

### 3.1 Command Bus and DQ Bus

CMD[10:0] carries the command and its associated parameters. $D Q[71: 0]$ is used for data transfer to and from the database entries, which comprise data and mask fields that are organized as data and mask arrays. The DQ bus carries the Search data (of the data and mask arrays and internal registers) during the Search command as well as the address and data during Read and/or Write operations. The DQ bus can also carry address information for the transparent accesses to the external SRAMs and/or SSRAMs.

### 3.2 Database Entry (Data and Mask Arrays)

Each database entry comprises data and mask fields. The resultant value of the entry is " 1, " " 0 ," or " $X$ (do not care)," depending on the value in the data mask bit. The on-chip priority encoder selects the first matching entry in the database that is nearest to location 0.

### 3.3 Arbitration Logic

When multiple NSEs are cascaded to create large databases, the data being searched is presented to all NSEs simultaneously in the cascaded system. If multiple matches occur, arbitration logic on the NSEs will enable the winning device (the one with a matching entry closest to address 0 of the cascaded database) to drive the SRAM bus.

### 3.4 Pipeline and SRAM Control

Pipeline latency is added to give enough time to a cascaded system's arbitration logic to determine the device that will drive the index of the matching entry on the SRAM bus. Pipeline logic adds latency to both the SRAM access cycles and the search successful flag (SSF) and search successful flag valid (SSV) signals to align them to the host ASIC receiving the associated data.

### 3.5 Full Logic

Bit[0] in each of the 72-bit entries has a special purpose for the Learn command ( $0=$ empty, $1=$ full). When all the data entries have bit[0] set to 1 , the database asserts the FULL flag, indicating that all the NSEs in the depth-cascaded array are full.

### 4.0 Signal Descriptions

Table 4-1 lists and describes all CYNSE70256 signals.
Table 4-1. CYNSE70256 Signal Description

| Pin Name | $\begin{gathered} \text { Pin }_{\text {rype }}{ }^{[1]} \end{gathered}$ | Pin Description |
| :---: | :---: | :---: |
| Clocks and Reset |  |  |
| CLK_MODE | I | Clock Mode. This signal allows the selection of clock input to the CLK1X/CLK2X pin. If the CLK_MODE pin is LOW, CLK2X must be supplied on that pin. PHS_L must also be supplied. If the CLK_MODE pin is HIGH, CLK1X must be supplied on the CLK2X/CLK1X pin, and the PHS_L signal is not required. When the CLK_mode is HIGH, PHS_L is unused and should be externally grounded. |
| CLK2X/CLK1X | 1 | Master Clock. Depending on the CLK MODE pin, either the CLK2X or the CLK1X must be supplied. CYNSE70256 samples control and data signals on both edges of CLK1X (if CLK1X is supplied). CYNSE70256 samples all data and control pins on the positive edge of CLK2X (if the CLK2X and PHS_L signals are supplied). All signals are driven out of the device on the rising edge of CLK1X (if CLK1X is supplied), and are driven on the rising edge of CLK2X when PHS_L is LOW (if CLK2X is supplied). |
| PHS_L | 1 | Phase. This signal runs at half the frequency of CLK2X and generates an internal CLK from CLK2X. See Section 5.0, "Clocks," on page 12. |
| RST_L | I | Reset. Driving RST_L LOW initializes the device to a known state. |
| CFG_L | 1 | Configuration. When CFG_L is LOW, CYNSE70256 will operate in backward compatibility mode with CYNSE70032 and CYNSE70064. When CFG_L is LOW, the CMD[10:9] should be externally grounded. With CFG_L LOW, the device will behave identically with CYNSE70032 and CYNSE70064, and the new feature added to CYNSE70256 will be disabled. When CFG_L is HIGH, the additional CMD[10:9] can be used and the following additional features will be supported: 1. sixteen pairs of global masks are supported instead of eight; 2. parallel Write to the data and mask arrays is supported (see Subsection 10.5, "Parallel Write," on page 24); and 3. configuring tables of up to three different widths does not require table identification bits in the data array, thus saving two bits from each 72-bit entry. |
| Command and DQ Bus |  |  |
| CMD[10:0] | 1 | Command Bus. [1:0] specifies the command; [10:2] contains the command parameters. The descriptions of individual CMDs explains the details of the parameters. The encoding of CMDs based on the [1:0] field are: <br> 00: PIO Read <br> 01: PIO Write <br> 10: Search <br> 11: Learn. |
| CMDV | 1 | Command Valid. This signal qualifies the command bus: <br> 0 : No command <br> 1: Command. |

Table 4-1. CYNSE70256 Signal Description (continued)

| Pin Name | Pin <br> Type $^{[1]}$ | Pin Description |
| :---: | :---: | :--- |
| DQ[71:0] | I/O | Address/Data Bus. This signal carries the Read and Write address and data during <br> register, data, and mask array operations. It carries the compare data during Search <br> operations. It also carries the SRAM address during SRAM PIO accesses. |
| ACK $^{[2]}$ | T | Read Acknowledge. This signal indicates that valid data is available on the DQ bus during <br> register, data, and mask array Read operations, or that the data is available on the SRAM <br> data bus during SRAM Read operations. |
| EOT ${ }^{[2]}$ | T | End of Transfer. This signal indicates the end of burst transfer to the data or mask array <br> during Read or Write burst operations. |
| SSF | T | Search Successful Flag. When asserted, this signal indicates that the device is the global <br> winner in a Search operation. |
| SSV | T | Search Successful Flag Valid. When asserted, this signal qualifies the SSF signal. |
| HIGH_SPEED | I | High Speed. This pin must be connected to ground $\left(\mathrm{V}_{\text {SS }}\right)$. It is provided for backward as <br> well as forward device compatibility. |

## SRAM Interface

| SADR[23:0] | T | SRAM Address. This bus contains address lines to access off-chip SRAMs that contain associative data. See Table 12-1 for the details of the generated SRAM address. In a database of multiple CYNSE70256s, each corresponding SADR bit from all cascaded devices must be connected. |
| :---: | :---: | :---: |
| CE_L | T | SRAM Chip Enable. This is the chip-enable (CE) control for external SRAMs. In a database of multiple CYNSE70256s, CE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices. |
| WE_L | T | SRAM Write Enable. This is the Write-enable control for external SRAMs. In a database of multiple CYNSE70256s, WE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices. |
| OE_L | T | SRAM Output Enable. This is the output-enable (OE) control for external SRAMs. Only the last device drives this signal (with the LRAM bit set). |
| ALE_L | T | Address Latch Enable. When this signal is LOW, the addresses are valid on the SRAM address bus. In a database of multiple CYNSE70256s, the ALE_L of all cascaded devices must be connected. This signal is then driven by only one of the devices. |
| Cascade Interface |  |  |
| LHI[6:0] | I | Local Hit In. These pins depth-cascade the device to form a larger table. One signal of this bus is connected to the LHO[1] or LHO[0] of each of the upstream devices in a block. All unused LHI pins are connected to a logic 0. For more information, see Section 11.0, "Depth Cascading," on page 84. LHI[0] stays unconnected. |
| LHO[1:0] | 0 | Local Hit Out. The LHO[1] and the LHO[0] are connected to one input on the LHI bus (from up to four downstream devices in a block totalling up to four). For more information, see Section 11.0, "Depth Cascading," on page 84. |
| BHI[2:0] | I | Block Hit In. Inputs from the previous block BHO[2:0] are tied to BHI[2:0] of the current device. In a four-block system, the last block can contain only seven devices because the identification code 11111 is used for broadcast access. |
| BHO[2:0] | 0 | Block Hit Out. These outputs from the last device in a block are connected to the BHI[2:0] inputs of the devices in the downstream blocks. |
| FULI[6:0] | I | Full In. Each signal in this bus is connected to FULO[0] or FULO[1] of an upstream device to generate the FULL flag for the depth-cascaded block. FULI[0] stays unconnected. |
| FULO[1:0] | 0 | Full Out. Both of these signals must be connected to the FULI of up to four downstream devices in a depth-cascaded table. Bit[0] in the data array indicates whether the entry is full (1) or empty (0). This signal is asserted if all bits in the data array are 1s. (Refer to Section 11.0, "Depth Cascading," on page 84, for information on how to generate the FULL flag.) |
| FULL | 0 | Full Flag. When asserted, this signal indicates that the table of multiple depth-cascaded devices is full. |

Table 4-1. CYNSE70256 Signal Description (continued)

| Pin Name | $\operatorname{Pin}_{\text {Type }^{[1]}}$ | Pin Description |
| :---: | :---: | :---: |
| Device Identification |  |  |
| ID[4:0] | I | Device Identification. The binary-encoded device identification (ID[4:1]) for a depth-cascaded system starts at 0000 and goes up to 1110.1111 is reserved for a special broadcast address that selects all cascaded NSEs in the system. On a broadcast Read-only, the device with the LDEV bit set to 1 responds. ID[0] stays unconnected. |
| Supplies |  |  |
| $V_{\text {DD }}$ | n/a | Chip Core Supply: 1.5V. |
| $\mathrm{V}_{\text {DDQ }}$ | n/a | Chip I/O Supply: $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$. |
| Test Access Port |  |  |
| TDI | I | Test access port's test data in. |
| TCK | 1 | Test access port's test clock. |
| TDO | T | Test access port's test data out. |
| TMS | I | Test access port's test mode select. |
| TRST_L | I | Test access port's reset. |

### 5.0 Clocks

If the CLK_MODE pin is LOW, CYNSE70256 receives the CLK2X and PHS_L signals. It uses the PHS_L signal to divide CLK2X and generate an internal clock (CLK ${ }^{[3,4]}$ ), as shown in Figure 5-1. The CYNSE70256 uses CLK2X and CLK for internal operations.


Figure 5-1. CYNSE70256 Clocks (CLK2X and PHS_L)
If the CLK_MODE pin is HIGH, CYNSE70256 receives CLK1X only. CYNSE70256 uses an internal phase-lock loop (PLL) to double the frequency of CLK1X and then divides that clock by two to generate a CLK for internal operations, as shown in Figure 5-2.


Figure 5-2. CYNSE70256 Clocks (CLK1X) ${ }^{[5]}$

## Notes:

1. $\quad I=$ Input only, $I / O=$ Input or Output, $O=$ Output only, $T=$ three-state output.
2. ACK and EOT require a weak external pulldown such as $47 \mathrm{~K} \Omega$ or $100 \mathrm{~K} \Omega$.
3. Any reference to "CLK" cycles means one cycle of CLK.
4. "CLK" is an internal clock signal.
5. For the purpose of showing timing diagrams, all such diagrams in this document will be shown in CLK2X mode. For a timing diagram in CLK1X mode, the following substitution can be made (see Figure 5-3).


Figure 5-3. CYNSE70256 Clocks for All Timing Diagrams

### 6.0 Phase-Lock Loop Usage

When the device first powers up, it takes 0.5 ms to lock the internal PLL. During this PLL, the RST_L must be held LOW for proper initialization of the device. It also takes 32 extra CLK1X cycles in CLK1X mode and 64 extra cycles in CLK2X mode. Setup and hold requirements will change in CLK1X mode if the duty cycle of the CLK1X is varied. All signals to the device in CLK1X mode are sampled by a clock that is generated by multiplying CLK1X by two. Since the PLL has a locking range, the device will only work between the range of frequencies specified in the timing specification section of this datasheet.

### 7.0 Registers

All registers in the CYNSE70256 device are 72 bits wide. The CYNSE70256 contains two banks of sixteen pairs of comparand storage registers, sixteen pairs of GMRs, eight search successful index registers, and one each of command, information, burst Read, burst Write, and next-free address registers. Table $7-1$ provides an overview of all the CYNSE70256 registers. The registers are in ascending address order; each register group is described in the following subsections.

Table 7-1. Register Overview (Bank0 and Bank1)

| Address | Abbreviation | Type | Name |
| :---: | :---: | :---: | :--- |
| $0-31$ | COMP0-31 | R | Sixteen pairs of comparand registers that store comparands from the <br> DQ bus for learning later. |
| $32-47$ |  |  |  |
| $96-111$ | MASKS | RW | Sixteen GMR pairs. |
| $48-55$ | SSR0-7 | R | Eight search successful index registers. |
| 56 | COMMAND | RW | Command register. |
| 57 | INFO | R | Information register. |
| 58 | RBURREG | RW | Burst Read register. |
| 59 | WBURREG | RW | Burst Write register. |
| 60 | NFA | R | Next-free address register. |
| $61-63$ | - | - | Reserved. |

### 7.1 Comparand Registers

The device contains two banks of 32 72-bit comparand registers (sixteen pairs) dynamically selected in every Search operation to store the comparand presented on the DQ bus. The Learn command will later use these registers when it is executed. The CYNSE70256 device stores the Search command's cycle A comparand in the even-numbered register and the cycle B comparand in the odd-numbered register, as shown in Figure 7-1 for each of the two banks of registers.


Figure 7-1. Comparand Register Selection during Search and Learn Instructions

### 7.2 Mask Registers

The device contains two banks of 3272 -bit GMRs (sixteen pairs) dynamically selected in every Search operation to select the Search subfield. The addressing of these registers is explained in Figure 7-2. The 4-bit GMR index supplied on the command bus can apply sixteen pairs of global masks during Search and Write operations, as shown below. ${ }^{[6]}$

| Index | 72 | $\xrightarrow{72}$ |
| :---: | :---: | :---: |
| 0 | 0 | 1 |
| 1 | 2 | 3 |
| 2 | 4 | 5 |
| 3 | 6 | 7 |
| 4 | 8 | 9 |
| 5 | 10 | 11 |
| 6 | 12 | 13 |
| 7 | 14 | 15 |
| 8 | 16 | 17 |
| 9 | 18 | 19 |
| 10 | 20 | 21 |
| 11 | 22 | 23 |
| 12 | 24 | 25 |
| 13 | 26 | 27 |
| 14 | 28 | 29 |
| 15 | 30 | 31 |

Search and Write Command Global Mask Selection
Figure 7-2. Addressing the GMR Array
Each mask bit in the GMRs is used during Search and Write operations. In Search operations, setting the mask bit to 1 enables compares; setting the mask bit to 0 disables compares at the corresponding bit position (forced match). In Write operations to the data or mask array, setting the mask bit to 1 enables Writes; setting the mask bit to 0 disables Writes at the corresponding bit position.

### 7.3 Search Successful Registers (SSR[0:7])

The device contains two banks of eight search successful registers (SSRs) to hold the index of the location at which a successful search occurred. The format of each register is described in Table 7-2. The Search command specifies which SSR stores the index of a specific Search command in cycle B of the Search instruction. Subsequently, the host ASIC can use this register to access that data array, mask array, or external SRAM using the index as part of the indirect access address (see Table 10-3 and Table 10-6).

The device with a valid bit set performs a Read or Write operation. All other devices suppress the operation.
Note:
6. In 72-bit Search and Write operations, the host ASIC must program both the even and odd mask registers with the same values for each of the banks.

Table 7-2. Search Successful Register Description

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :--- |
| INDEX | $[15: 0]$ | X | Index. This is the address of the 72-bit entry where a successful search <br> occurs. The device updates this field only when the search is successful. <br> If a hit occurs in a 144-bit entry-size quadrant, the least-significant bit (LSB) <br> is 0. If a hit occurs in a 288-bit entry-size quadrant, the two LSBs are 00. <br> This index updates if the device is either a local or global winner in <br> a Search operation. |
| - | $[30: 16]$ | 0 | Reserved. |
| VALID | $[31]$ | 0 | Valid. During Search operation in a depth-cascaded configuration, this ban <br> of the device that is a global winner in a match sets this bit to 1. This bit <br> updates only when the device is a global winner in a Search operation. |
| - | $[71: 32]$ | 0 | Reserved. |

### 7.4 Command Register

Table 7-3 describes the command registers' fields for each of the two banks; Bank 0 and Bank 1.
Table 7-3. Command Register Description

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :---: |
| SRST | [0] | 0 | Software Reset. If 1, this bit resets the bank with the same effect as a hardware reset. Internally, it generates a reset pulse lasting for eight CLK cycles. This bit automatically resets to a 0 after the reset has completed. |
| DEVE | [1] | 0 | Device Enable. If 0, it keeps the SRAM bus (SADR, WE_L, CE_L, OE_L and ALE_L), SSF, and SSV signals in a three-state condition and forces the cascade interface output signals LHO[1:0] and $\mathrm{BHO}[2: 0]$ to 0 . It also keeps the DQ bus in input mode. The purpose of this bit is to make sure that there are no bus contentions when the devices power up in the system. |
| TLSZ | [3:2] | 01 | Table Size. The host ASIC must program this field to configure each bank into a table of a certain size. This field affects the pipeline latency of the Search and Learn operations as well as the Read and Write accesses to the SRAM (SADR[23:0], CE_L, OE_L, WE_L, ALE_L, SSV, SSF, and ACK). Once programmed, the Search latency stays constant. <br> Latency in number of CLK cycles with HIGH_SPEED LOW: <br> 01: Up to four devices $5$ <br> 10: Up to fifteen devices <br> 11: Reserved. |
| HLAT | [6:4] | 000 | Latency of Hit Signals. This field adds further latency to the SSF and SSV signals during Search, and ACK signal during SRAM Read access by the following number of CLK cycles. |
| LDEV | [7] | 0 | Last Device in the Cascade. When set, this is the last bank of the last device in the depth-cascaded table and is the default driver for the SSF and SSV signals. In the event of a Search failure, the bank with this device with this bit set drives the hit signals as follows: $S S F=0, S S V=1$. During non-Search cycles, the device with this bit set drives the signals as follows: SSF $=0, \mathrm{SSV}=0$. |
| LRAM | [8] | 0 | Last Device on the SRAM Bus. When set, this is the last bank of the last device on the SRAM bus in the depth-cascaded table and is the default driver for the SADR, CE_L, WE_L, and ALE_L signals. In cycles where no CYNSE70256 device in a depth-cascaded table drives these signals, this drives the signals as follows: SADR $=23$ 'hFFFFFFF, CE_L $=1, W E \_L=1$, and ALE_L = 1 . OE_L is always driven by the device for which this bit is set. |

Table 7-3. Command Register Description (continued)

| Field | Range | Initial Value |  |
| :---: | :---: | :---: | :--- |
| CFG | $[24: 9]$ | 0000000000 <br> 000000 | Database Configuration. The device is divided internally into two banks each consisting of sixteen <br> partitions of 8K $\times 72$, each of which can be configured as $8 \mathrm{~K} \times 72,4 \mathrm{~K} \times 144$, or $2 \mathrm{~K} \times 288$, as <br> follows. <br> 00: $8 \mathrm{~K} \times 72$ <br> 01: $4 \mathrm{~K} \times 144$ <br> $10: 2 \mathrm{~K} \times 288$ <br> $11: \mathrm{LOW}$ power, partition not used for Search. <br> Bits[10:9] apply to configuring the first partition in the address space. <br> Bits[12:11] apply to configuring the second partition in the address space. <br> Bits[14:13] apply to configuring the third partition in the address space. <br> Bits[16:15] apply to configuring the fourth partition in the address space. <br> Bits[18:17] apply to configuring the fifth partition in the address space. <br> Bits[20:19] apply to configuring the sixth partition in the address space. <br> Bits[22:21] apply to configuring the seventh partition in the address space. <br> Bits[24:23] apply to configuring the eighth partition in the address space. |
|  |  |  |  |
|  |  |  |  |
|  |  |  |  |

### 7.5 Information Register

## Table 7-4 describes the information register fields for both banks.

Table 7-4. Information Register Description

| Field | Range | Initial Value | Description |
| :--- | :--- | :--- | :--- |
| Revision | $[3: 0]$ | 0001 | Revision Number. This is the current device revision number. <br> Numbers start at one and increment by one for each revision of <br> the device. |
| Implementation | $[6: 4]$ | 001 | This is the CYNSE70256 implementation number. |
| Reserved | $[7]$ | 0 | Reserved. |
| Device ID | $[15: 8]$ | 00000100 | This is the device identification number. |
| MFID | $[31: 16]$ | $1101 \_1100 \_0111 \_1111$ | Manufacturer ID. This field is the same as the manufacturer identi- <br> fication number and continuation bits in the TAP controller. |
| Reserved | $[71: 32]$ |  | Reserved. |

### 7.6 Read Burst Address Register

Table 7-5 shows the Read burst address register (RBURREG) fields that must be programmed before a burst Read when a Read burst transfer is done from any bank.
Table 7-5. Read Burst Register Description

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :--- |
| ADR | $[15: 0]$ | 0 | Address. This is the starting address of the data or mask array during <br> a burst-Read operation from a bank. It automatically increments by one <br> for each successive Read of the data or mask array. Once the operation <br> is complete, the contents of this field must be reinitialized for the next <br> operation. |
| BLEN | $[27: 19]$ | 0 | Reserved. |
|  |  | Length of Burst Access. The device provides the capability to Read from <br> $4-511$ locations in a single burst from each bank. The BLEN decre- <br> ments automatically. Once the operation is complete, the contents of <br> this field must be reinitialized for the next operation. |  |
|  | $[71: 28]$ |  | Reserved. |

### 7.7 Write Burst Address Register Description

Table 7-6 describes the Write burst address register (WBURREG) fields that must be programmed before a burst Write.
Table 7-6. Write Burst Register Description

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :--- |
| ADR | $[15: 0]$ | 0 | Address. This is the starting address of the data or mask array during a burst-Write <br> operation from a bank. It automatically increments by one for each successive <br> Write of the data or mask array. Once the operation is complete, the contents of <br> this field must be reinitialized for the next operation. |
| BLEN | $[27: 19]$ | 0 | Reserved. |
|  | $[71: 28]$ |  | Length of Burst Access. The device provides the capability to Write from 4-511 <br> locations in a single burst. The BLEN decrements automatically. Once the <br> operation is complete, the contents of this field must be <br> reinitialized for the next operation. |

### 7.8 NFA Register

Bit[0] of each 72-bit data entry is specially designated for use in the operation of the Learn command in each of the banks. For 72-bit-configured quadrants, this bit indicates whether a location is full (bit set to 1 ) or empty (bit set to 0). Every Write and/or Learn command loads the address of the first 72-bit location that contains a 0 in the entry's bit[0]. This is stored in the NFA register(see Table 7-7). If all the bits[0] in a device for both the banks within the device are set to 1 , the CYNSE70256 asserts FULO[1:0] to 1.

For 144-bit-configured quadrants, the LSB of the NFA register is always set to 0 . The host ASIC must set both bit[0] and bit[72] in a 144 -bit word to either 0 or 1 to indicate full or empty status. Both bit[0] and bit[72] must be set to either 0 or 1 , (that is, the 10 or 01 settings are invalid).
Table 7-7. NFA Register

| Address | 71-16 | 15-0 |
| :---: | :---: | :---: |
| 60 | Reserved | Index |

### 8.0 NSE Architecture and Operation Overview

The CYNSE70256 device consists of two banks of $64 \mathrm{~K} \times 72$-bit storage cells referred to as data bits. There is a mask cell corresponding to each data cell. Figure 8-1 shows the three organizations of the device based on the value of the CFG bits in the command register.


Figure 8-1. CYNSE70256 Database Width Configuration for Each of the Two Banks

During a Search operation, the search data bit (S), data array bit (D), mask array bit (M), and global mask bit (G) are used in the following manner to generate a match at that bit position (see Table 8-1). The entry with a match on every bit position results in a successful Search.

Table 8-1. Bit Position Match

| $\mathbf{G}$ | $\mathbf{M}$ | $\mathbf{D}$ | $\mathbf{S}$ | $\mathbf{X}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | $\mathbf{X}$ | $\mathbf{X}$ | X | 1 |
| 1 | 0 | 0 | 0 | 1 |
| 1 | 1 | 1 | 0 | 1 |
| 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 | 0 |
| 1 | 1 | $X$ | 1 |  |

In order for a successful Search within a device to make the device the local winner, all 72-bit positions must generate a match for a 72-bit entry in 72-bit-configured quadrants, or all 144-bit positions must generate a match for two consecutive even and odd 72 -bit entries in quadrants configured as 144 bits, or all 288-bit positions must generate a match for four consecutive entries aligned to four entry-page boundaries of 72-bit entries in quadrants configured as 288 bits.

An arbitration mechanism using a cascade bus determines the global winning device among the local winning devices in a Search cycle. The global winning device drives the SRAM bus, the SSV, and the SSF signals. In case of a Search failure, the device(s) with a bank with the LDEV and LRAM bits set drives the SRAM bus, SSF, and SSV signals.

The CYNSE70256 device can be configured to contain tables of different widths, even within the same chip. Figure 8-2 shows a sample configuration of different widths.


Figure 8-2. Multiwidth Database Configurations Example

### 9.0 Data and Mask Addressing

Figure 9-1 shows CYNSE70256 data and mask array addressing for both Bank 0 and Bank 1.


Figure 9-1. Addressing the CYNSE70256 Data and Mask Arrays

### 10.0 Commands

A master device such as an ASIC controller issues commands to the CYNSE70256 device using the CMDV signal and the command bus. The following subsections describe the operation of these commands.

### 10.1 Command Codes

The CYNSE70256 device implements four basic commands, as shown in Table 10-1. The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for two CLK2X cycles (cycles A and B) when the CLK_MODE pin is LOW. In CLK2X mode, the controller ASIC must align the instructions using the PHS_L signal. The command code must be presented to CMD[1:0] while keeping the CMDV signal HIGH for one CLK1X cycle when the CLK_MODE pin is HIGH. In CLK1X mode the HIGH phase is cycle A and the LOW phase is cycle B. The CMD[10:2] field passes command parameters in cycles A and B.
Table 10-1. Command Codes

| Command Code | Command | Description |
| :---: | :---: | :--- |
| 00 | Read | Reads one of the following: data array, mask array, device registers, or external SRAM. |
| 01 | Write | Writes one of the following: data array, mask array, device registers, or external SRAM. |
| 10 | Search | Searches the data array for a desired pattern using the specified register from the GMR <br> array and local mask associated with each data cell. |
| 11 | The device has internal storage for up to sixteen comparands that it can learn. The <br> device controller can insert these entries at the next-free address (as specified by the <br> NFA register) using the Learn instruction. |  |

### 10.2 Commands and Command Parameters

Table 10-2 lists the command bus fields that contain the CYNSE70256 command parameters and their respective cycles. Each command is described separately in the subsections that follow.

Table 10-2. Command Parameters

| CMD ${ }^{[7,8]}$ | CYC | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Read | A | X | X | SADR[23] | SADR[22] | SADR[21] | 0 | 0 | 0 | $\begin{gathered} 0=\text { Single } \\ 1=\text { Burst } \end{gathered}$ | 0 | 0 |
|  | B | X | X | 0 | 0 | 0 | 0 | 0 | 0 | $\begin{gathered} 0=\text { Single } \\ 1=\text { Burst } \end{gathered}$ | 0 | 0 |
| Write | A | $\begin{gathered} \hline \text { GMR } \\ \text { Index }{ }^{[9]} \end{gathered}$ | 0: Normal Write <br> 1: Parallel Write | SADR[23] | SADR[22] | SADR[21] | $\begin{gathered} \text { GMR Index } \\ {[2: 0]} \end{gathered}$ |  |  | $\begin{gathered} 0=\text { Single } \\ 1=\text { Burst } \end{gathered}$ | 0 | 1 |
|  | B | $\begin{gathered} \text { GMR } \\ \text { Index }^{[9]} \end{gathered}$ | 0: Normal Write <br> 1: Parallel Write | 0 | 0 | 0 | $\begin{gathered} \text { GMR Index } \\ {[2: 0]} \end{gathered}$ |  |  | $\begin{gathered} 0=\text { Single } \\ 1=\text { Burst } \end{gathered}$ | 0 | 1 |
| Search | A | $\begin{array}{\|c\|} \hline \text { GMR } \\ \text { Index }^{[9]} \end{array}$ | $\begin{array}{\|l\|} \hline 72 \text { bits: } 0 \\ 144 \text { bits: } 1 \\ 288 \text { bits: } X \end{array}$ | SADR[23] | SADR[22] | SADR[21] | $\begin{gathered} \text { GMR Index } \\ {[2: 0]} \end{gathered}$ |  |  | 72 bits or 144 bits: 0 288 bits: <br> 1 in first cycle <br> 0 in second cycle | 1 | 0 |
|  | B | X |  | SSR Index[2:0] |  |  | Comparand Register Index |  |  |  | 1 | 0 |
| Learn ${ }^{[9]}$ | A | X | X | SADR[23] | SADR[22] | SADR[21] | Comparand Register Index |  |  |  | 1 | 1 |
|  | B | X | X | 0 | 0 | Mode 0: 72 bits 1: 144 bits | Comparand Register Index |  |  |  | 1 | 1 |

### 10.3 Read Command

The Read can be a single Read of a data array, a mask array, an SRAM, or a register location (CMD[2] = 0). It can be a burst Read of the data (CMD[2] = 1) or mask array locations using an internal auto-incrementing address register (RBURADR). A description of each type is provided in Table 10-3. A single-location Read operation lasts six cycles, as shown in Figure 10-1. The burst Read adds two cycles for each successive Read. The SADR[23:21] bits supplied in Read instruction cycle A drives SADR[23:21] signals during a Read of an SRAM location.
Table 10-3. Read Command Parameters

| CMD Parameter <br> CMD[2] | Read Command |  |
| :---: | :---: | :--- |
| 0 | Single Read | Reads a single location of the data array, mask array, external SRAM, or device registers. <br> All access information is applied on the DQ bus. |
| 1 | Burst Read | Reads a block of locations from the data or mask array as a burst. RBURADR specifies <br> the starting address and the length of the data transfer from the data or mask array; it also <br> auto-increments the address for each access. All other access information is applied on <br> the DQ bus. ${ }^{[10]}$ |

## Notes:

7. Use CMD[8:0] only and connect CMD[10:9] to ground with CFG_L LOW.
8. For a description of CMD[9] and CMD[2], see Search 288-bit-configured tables and mixed-size searches with CFG_L HIGH.
9. The 288-bit-configured devices or 288-bit-configured quadrants within devices do not support the Learn instruction.
10. The device registers and external SRAM can only be read in single-Read mode.


Figure 10-1. Single-Location Read Cycle Timing
The single Read operation takes six CLK cycles that operate in the following sequence.

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] $=0$ ) using CMDV $=1$, and the DQ bus supplies the address, as shown in Table 10-4 and Table 10-5. The host ASIC selects the CYNSE70256 device for which ID[4:1] matches the DQ[25:22] lines. The DQ[21] specifies the bank of the device. If DQ[25:21] = 11111, the host ASIC selects the CYNSE70256 with the LDEV bit set. The host ASIC also supplies SADR[23:21] on CMD[8:6] in cycle A of the Read instruction if the Read is directed to the external SRAM.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives the ACK signal from $Z$ to LOW.
- Cycle 5: The selected device drives the Read data from the addressed location on the $D Q[71: 0]$ bus, and drives the $A C K$ signal HIGH.
- Cycle 6: The selected device floats the DQ[71:0] to a three-state condition and drives the ACK signal LOW.

At the termination of cycle 6, the selected device releases the ACK line to a three-state condition. The Read instruction is complete, and a new operation can begin. ${ }^{[11]}$

Table 10-5 describes the Read address format for the internal registers. Figure 10-2 illustrates the timing diagram for the burst Read of the data or mask array.
Table 10-4. Read Address Format for Data Array, Mask Array, or SRAM

| $\begin{gathered} \text { DQ } \\ {[71: 30]} \end{gathered}$ | $\begin{aligned} & \hline \text { DQ } \\ & \text { [291 } \end{aligned}$ | $\begin{gathered} \text { DQ } \\ {[28: 26]} \end{gathered}$ | $\begin{array}{\|c\|} \hline \text { DQ } \\ {[25: 22]} \end{array}$ | $\begin{gathered} \hline \text { DQ } \\ \text { [21] } \end{gathered}$ | $\begin{gathered} \text { DQ } \\ {[20: 19]} \end{gathered}$ | $\begin{gathered} \text { DQ } \\ {[18: 16]} \end{gathered}$ | $\begin{gathered} \text { DQ } \\ {[15: 0]} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Reserved | 0: Direct <br> 1: Indirect | SSR Index (applicable if DQ[29] is indirect) | ID | Bank 0 or 1 | $\begin{aligned} & \text { 00: Data } \\ & \text { Array } \end{aligned}$ | Reserved | If DQ[29] is 0 , this field carries the address of the data array location. If DQ[29] is 1 , the SSR index specified on $D Q[28: 26]$ is used to generate the address of the data array location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{\text {[12] }}$ |
| Reserved | 0 : Direct <br> 1: Indirect | SSR Index (applicable if DQ[29] is indirect) | ID | Bank 0 or 1 | 01: Mask Array | Reserved | If DQ[29] is 0 , this field carries the address of the mask array location. If $D Q[29]$ is 1 , the SSR index specified on DQ[28:26] is used to generate the address of the mask array location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{[12]}$ |
| Reserved | 0: Direct <br> 1: Indirect | SSR Index (applicable if $\mathrm{DQ}[29]$ is indirect) | ID | Bank 0 or 1 | 10: <br> External SRAM | Reserved | If DQ[29] is 0 , this field carries the address of the SRAM location. If DQ[29] is 1 , the SSR index specified on $D Q[28: 26]$ is used to generate the address of the SRAM location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{12]}$ |

## Notes:

11. The latency of the SRAM Read will be different than the one described above (see Subsection 12.1 , "SRAM PIO Access," on page 86 ). Table $10-4$ lists and describes the format of the Read address for a data array, mask array, or SRAM.
12. "|" stands for logical OR operation. " $\}$ " stands for concatenation operator.

Table 10-5. Read Address Format for Internal Registers

| DQ[71:26] | DQ[25:22] | DQ[21] | DQ[20:19] | DQ[18:7] | DQ[6:0] |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Reserved | ID | Bank 0 or 1 | 11: Register | Reserved | Register Address |



Figure 10-2. Burst Read of the Data and Mask Arrays (BLEN = 4)
The Read operation lasts $4+2 n$ CLK cycles (where $n$ is the number of accesses in the burst specified by the BLEN field of the RBURREG) in the sequence shown below. This operation assumes that the host ASIC has programmed the RBURREG of the approprritae bank of the device with the starting ADR and the BLEN before initiating the burst Read command for the appropriate bank in the appropriate device.

- Cycle 1: The host ASIC applies the Read instruction on CMD[1:0] (CMD[2] =1) using CMDV $=1$, and the address supplied on the DQ bus as shown in Table 10-6. The host ASIC selects the bank 0 or 1 (Based on Bank Bit) of the CYNSE70256 device where ID[4:1] matches the DQ[25:22] lines. DQ[21] specfies the bank of the device that is written. If DQ[25:21] = 11111 the host ASIC selects the bank of the CYNSE70256 device with the LDEV bit set.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive the DQ[71:0] bus and drives ACK and EOT from Z to LOW.
- Cycle 5: The selected device drives the Read data from the address location on the DQ[71:0] bus and drives the ACK signal HIGH.

Cycles 4 and 5 repeat for each additional access until all the accesses specified in the BLEN field of RBURREG are complete. On the last transfer, the CYNSE70256 device drives the EOT signal HIGH.
 At the termination of cycle $(4+2 n)$, the selected device floats the ACK line to a three-state condition. The burst Read instruction is complete, and a new operation can begin. Table 10-6 describes the Read address format for data and mask arrays for burst Read operations.

Table 10-6. Read Address Format for Data and Mask Arrays

| DQ[71:26] | DQ[25:22] | DQ[21] | DQ[20:19] | DQ[18:16] | DQ[15:0] |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Reserved | ID | Bank 0 or 1 | 00: Data Array | Reserved | Do not care. These seventeen bits come from the RBURADR, <br> which increments for each access. |
| Reserved | ID | Bank 0 or 1 | 01: Mask Array | Reserved | Do not care. These seventeen bits come from the RBURADR, <br> which increments for each access. |

### 10.4 Write Command

The Write command can be a single Write of a data array, mask array, register, or external SRAM location (CMD[2] = 0). It can be a burst Write (CMD[2] = 1) using an internal auto-incrementing address registers (WBURADR) of the data or mask array
locations. A single-location Write is a three-cycle operation, as shown in Figure 10-3. The burst Write adds one extra cycle for each successive location Write.


Figure 10-3. Single Write Cycle Timing
The following is the Write operation sequence. Table 10-7 shows the Write address format for the data array, the mask array, or single-Write SRAM. Table 10-8 shows the Write address format for the internal registers.

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC also supplies the GMR index to mask the Write to the data or mask array location on \{CMD[10], CMD[5:3]\}. For SRAM Writes, the host ASIC must supply the SADR[23:21] on CMD[8:6]. The host ASIC sets CMD[9] to 0 for a normal Write.
- Cycle 1B:The host ASIC continues to apply the Write instruction to CMD[1:0] (CMD[2] = 0) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC selects the device where ID[4:1] matches the DQ[25:22] lines and the bank within the device using value on DQ[21], or it selects both banks of all the devices when DQ[25:21] = 11111.
- Cycle 2: The host ASIC drives DQ[71:0] with the data to be written to the data array, mask array, or register location of the selected device.
- Cycle 3: Idle cycle.

At the termination of cycle 3, another operation can begin. ${ }^{[13]}$
Table 10-7. Write Address Format for Data Array, Mask Array, or SRAM (Single Write)

| $\begin{gathered} \hline \text { DQ } \\ {[71: 30]} \end{gathered}$ | $\begin{aligned} & \hline \mathrm{DQ} \\ & {[29]} \end{aligned}$ | $\begin{gathered} \text { DQ } \\ {[28: 26]} \end{gathered}$ | $\begin{array}{\|c\|} \hline \text { DQ } \\ {[25: 22]} \end{array}$ | $\begin{aligned} & \hline \mathrm{DQ} \\ & \text { [211 } \end{aligned}$ | $\begin{gathered} \hline \text { DQ } \\ {[20: 19]} \end{gathered}$ | $\begin{gathered} \hline \text { DQ } \\ {[18: 16]} \end{gathered}$ | $\begin{gathered} \text { DQ } \\ {[15: 0]} \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Reserved | 0: Direct 1: Indirect | SSR (applicable if DQ[29] is indirect) | ID | Bank 0 or 1 | 00: Data Array | Reserved | If DQ[29] is 0 , this field carries the address of the data array location. <br> If DQ[29] is 1 , the SSR specified on DQ[28:26] is used to generate the address of data array location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{[14]}$ |
| Reserved | 0: Direct <br> 1: Indirect | SSR <br> (applicable if $D Q[29]$ is indirect) | ID | Bank 0 or 1 | 01: Mask Array | Reserved | If $D Q[29]$ is 0 , this field carries the address of the mask array location. <br> If DQ[29] is 1 , the SSR specified on DQ[28:26] is used to generate the address of the mask array location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{\text {[14] }}$ |
| Reserved | 0: Direct 1: Indirect | SSR (applicable if $D Q[29]$ is indirect) | ID | Bank 0 or 1 | 10: External SRAM | Reserved | If DQ[29] is 0 , this field carries the address of the SRAM location. <br> If DQ[29] is 1 , the SSR specified on DQ[28:26] is used to generate the address of SRAM location: \{SSR[15:2], SSR[1] \| DQ[1], SSR[0] | DQ[0]\}. ${ }^{14]}$ |

Table 10-8. Write Address Format for Internal Registers

| DQ[71:26] | DQ[25:22] | DQ[21] | DQ[20:19] | DQ[18:7] | DQ[6:0] |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Reserved | ID | Bank 0 or 1 | 11: Register | Reserved | Register address |

## Notes:

13. The latency of the SRAM Write will be different than the one described above (see Subsection 12.1, "SRAM PIO Access," on page 86).
14. "|" stands for logical OR operation. " $\}$ " stands for concatenation operator.

Figure $10-4$ shows the timing diagram of a burst Write operation of the data or mask array.


The burst Write operation lasts for $(\mathrm{n}+2)$ CLK cycles. n signifies the number of accesses in the burst as specified in the BLEN field of the WBURREG register. The following is the block Write operation sequence. This operation assumes that the host ASIC has programmed the WBURREG of the appropriate bank with the starting ADR and BLEN before initiating a burst Write command.

- Cycle 1A: The host ASIC applies the Write instruction to CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus as shown in Table 10-9. The host ASIC also supplies the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC sets CMD[9] to 0 for the normal Write.
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] (CMD[2] = 1) using CMDV = 1, and the address supplied on the DQ bus. The host ASIC continues to supply the GMR index to mask the Write to the data or mask array locations in \{CMD[10], CMD[5:3]\}. The host ASIC selects the device for which ID[4:1] matches the $\operatorname{DQ}[25: 22]$ lines and the bank of the device using DQ[21] lines. It selects all devices when $D Q[25: 21]=11111$.
- Cycle 2: The host ASIC drives the DQ[71:0] with the data to be written to the data or mask array location of the selected device. The CYNSE70256 device writes the data from the DQ[71:0] bus only to the subfield with the corresponding mask bit set to 1 in the GMR that is specified by the index \{CMD[10],CMD[5:3]\} supplied in cycle 1.
- Cycles $\mathbf{3}$ to $\mathbf{n + 1}$ : The host ASIC drives the DQ[71:0] with the data to be written to the next data or mask array location of the selected device (addressed by the auto-increment ADR field of the WBURREG register).
The CYNSE70256 device writes the data on the DQ[71:0] bus only to the subfield that has the corresponding mask bit set to 1 in the GMR specified by the index supplied in cycle 1 \{CMD[10],CMD[5:3]\}. The CYNSE70256 device drives the EOT signal LOW from cycle 3 to cycle $n$; the CYNSE70256 device drives the EOT signal HIGH in cycle $n+1$ ( $n$ is specified in the BLEN field of the WBURREG).
- Cycle n+2: The CYNSE70256 device drives the EOT signal LOW.

At the termination of cycle $n+2$, the CYNSE70256 device floats the EOT signal to a three-state operation, and a new instruction can begin.
Table 10-9. Write Address Format for Data and Mask Array (Burst Write)

| DQ <br> [71:26] | DQ <br> [25:22] | DQ <br> $[21]$ | DQ <br> $[20: 19]$ | DQ <br> $[18: 16]$ | DQ <br> [15:0] |
| :---: | :---: | :---: | :---: | :---: | :--- |
| Reserved | ID | Bank <br> 0 or 1 | 00: Data array | Reserved | Do not care. These seventeen bits come from <br> WBURADR, which increments with each access. |
| Reserved | ID | Bank <br> 0 or 1 | 01: Mask array | Reserved | Do not care. These seventeen bits come from <br> WBURADR, which increments with each access. |

### 10.5 Parallel Write

In order to write the data and mask arraysof both banks faster for initialization, testing, or diagnostics, many locations can be written simultaneously in the CYNSE70256 device. When CMD[9] is set in cycles A and B of the Write command during a Write to the data or mask arrays, the address present on DQ[10:1] that specifies 128 locations in the device is used, and 64 72-bit locations are simultaneously written in either the data or mask array. Setting $\mathrm{DQ}[21]$ to 0 will cause a write to the addresses, specified by $\operatorname{DQ}[10: 1]$, closer to address 0 , while setting $D Q[21]$ to 1 will cause a write to the addresses, also specified by DQ[10:1], further from address 0 .

### 10.6 Search Command

This subsection describes the following.

- 72-bit search on tables configured as $\times 72$ using up to four devices
- 72-bit search on tables configured as $\times 72$ using up to fifteen devices
- 144-bit search on tables configured as $\times 144$ using up to four devices
- 144-bit search on tables configured as $\times 144$ using up to fifteen device
- 288-bit search on tables configured as $\times 288$ using up to four devices
- 288 -bit search on tables configured as $\times 288$ using up to fifteen devices
- Mixed-size search on tables configured with different widths using an CYNSE70256 with CFG_L LOW
- Mixed-size searches on tables configured with different widths using an CYNSE70256 with CFG_L HIGH.


### 10.6.1 72-bit Search on Tables Configured as $\times 72$ using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in Figure 10-5. The following are the parameters programmed into each bank of the the four devices.

- First three devices (devices $0-2$, both banks): $C F G=0000000000000000, T L S Z=01, H L A T=010, L R A M=0$, and LDEV $=0$.
- Fourth device (device 3, Bank 0): CFG $=0000000000000000, T L S Z=01, H L A T=010$, LRAM $=0$, and LDEV $=0$
- Fourth device (device 3, Bank 1): CFG $=0000000000000000, T L S Z=01$, HLAT $=010$, LRAM $=1$, and LDEV $=1$. ${ }^{[15]}$

Figure 10-6 shows the timing diagram for a Search command in the 72-bit-configured table of four devices for device number 0. Figure $10-7$ shows the timing diagram for a Search command in the 72 -bit-configured table of four devices for device number 1. Figure $10-8$ shows the timing diagram for a Search command in the 72 -bit-configured table of four devices for device number 7 (the last device in this specific table). For these timing diagrams four 72-bit searches are performed sequentially. HIT/MISS assumptions were made as shown below in Table 10-10.
Table 10-10. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Hit | Miss |
| Device 1 | Miss | Hit | Hit | Miss |
| Device 2 | Miss | Miss | Miss | Miss |
| Device 3 | Miss | Miss | Hit | Hit |



Figure 10-5. Hardware Diagram for a Table with Four Devices
Note:
15. Each bank of the four devices must be programmed with the same values for TLSZ and HLAT. Only the the last bank of the last device in the table (device number 3 in this case) must be programmed with LRAM $=1$ and LDEV $=1$.


Figure 10-6. Timing Diagram for 72-bit Search Device Number 0



Figure 10-8. Timing Diagram for 72-bit Search Device Number 3 (Last Device)
The following is the sequence of operation for a single 72-bit Search command (also refer to Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) to CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data to be compared. The CMD[2] signal must be driven to logic 0 .
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and hit flag (see page 14 for a description of SSR[0:7]). The DQ[71:0] continues to carry the 72-bit data to be compared.
Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. Also, the even and odd pairs of GMRs selected for the comparison must be programmed with the same value.

The logical 72-bit Search operation is shown in Figure 10-9. The entire table of 72-bit entries (four devices) is compared to a 72 -bit word K (presented on the DQ bus in both cycles A and B of the command) using the GMR and local mask bits. The effective GMR is the 72-bit word specified by the identical value in both even and odd GMR pairs, in each of two banks of the four devices, and selected by the GMR Index in the command's cycle A. The 72-bit word K (presented on the DQ bus in both cycles A and B of the command) is also stored in both even and odd comparand register pairs (selected by the comparand register index in
command cycle $B$ ) in each of the four devices. In the $\times 72$ configuration, only the even comparand register can subsequently be used by the Learn command in one of the devices (the first non-full device only). The word K (presented on the DQ bus in both cycles $A$ and $B$ of the command) is compared with each entry in the table, starting at location 0 . The first matching entry's location address $L$ is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see Section 12.0, "SRAM Addressing," on page 86). The global winning device will drive the bus in a specific cycle. On a global miss cycle, the device with LRAM $=1$ (default driving device for the SRAM bus) and LDEV $=1$ (default driving device for SSF and SSV signals) will be the default driver for such missed cycles.


Figure 10-9. $\times 72$ Table with Four Devices
The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 72-bit searches in $\times 72$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 72-bit Search command cycle (two CLK2X cycles) is shown in Table 10-11.

Table 10-11. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Cycles |
| :---: | :---: | :---: |
| $1-4(T L S Z=01)$ | $512 \mathrm{~K} \times 72$ bits | 5 |
| $1-15(\mathrm{TLSZ}=10)$ | $1920 \mathrm{~K} \times 72$ bits | 6 |

The latency of the Search from command to SRAM access cycle is 5 for up to four devices in the table (TLSZ = 01). SSV and SSF also shift further to the right for different values of HLAT, as specified in Table 10-12.
Table 10-12. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

10.6.2 72-bit Search on Tables Configured as $\times 72$ using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in Figure 10-10. Each of the four blocks in the diagram represents four CYNSE70256 devices (except the last, which has three devices). The diagram for a block of four devices is shown in Figure 10-11. The following are the parameters programmed into the fifteen devices.

- First thirty devices (devices $0-13$, both banks): $C F G=0000000000000000, T L S Z=10, H L A T=001, L R A M=0$, $a n d$ LDEV $=0$.
- First thirty devices (device 14, Bank 0): CFG $=000000000000000$, $T L S Z=10, H L A T=001, L R A M=0$, and LDEV $=0$.
- Thirty-first device (device 14, Bank 1): CFG $=0000000000000000, T L S Z=10, H L A T=001, L R A M=1$, and LDEV $=1$.

Note. All fifteen devices must be programmed with the same values for TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 14 in this case).

The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in Table 10-13. For the purpose of illustrating the timings, it is further assumed that the there is only one device with a matching entry in each of the blocks. Figure 1012 shows the timing diagram for a Search command in the 72-bit-configured table of fifteen devices for each of the four devices in block number 0 . Figure $10-13$ shows the timing diagram for a Search command in the 72 -bit-configured table of fifteen devices for the all the devices in block number 1 (above the winning device in that block). Figure 10-14 shows the timing diagram for the globally winning device (defined as the final winner within its own and all blocks) in block number 1. Figure 10-15 shows the timing diagram for all the devices below the globally winning device in block number 1. Figure 10-16, Figure 10-17, and Figure 10-18 show the timing diagrams of the devices above the globally winning device, the globally winning device, and the devices below the globally winning device, respectively, for block number 2. Figure 10-19, Figure 10-20, Figure 10-21, and Figure 10-22 show the timing diagrams of the devices above globally winning device, the globally winning device, and the devices below the globally winning device except the last device (device 14), respectively, for block number 3.

The 72-bit Search operation is pipelined and executes as follows. Four cycles from the Search command, each of the devices knows the outcome internal to it for that operation. In the fifth cycle after the Search command, the devices in a block arbitrate for a winner among them (a "block" being defined as less than or equal to four devices resolving the winner within them using the LHI[6:0] and LHO[1:0] signalling mechanism). In the sixth cycle after the Search command, the blocks (of devices) resolve the winning block through the $\mathrm{BHI}[2: 0]$ and $\mathrm{BHO}[2: 0]$ signalling mechanism. The winning device within the winning block is the global winning device for a Search operation.
Table 10-13. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Block 0 | Miss | Miss | Miss | Miss |
| Block 1 | Miss | Miss | Hit | Miss |
| Block 2 | Miss | Hit | Hit | Miss |
| Block 3 | Hit | Hit | Miss | Miss |



Figure 10-10. Hardware Diagram for a Table with Fifteen Devices


Figure 10-11. Hardware Diagram for a Block of up to Four Devices
cycle cycle cycle cycle cycle cycle cycle cycle cycle cycle
 this device.this device.)
Figure 10-12. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)


Figure 10-13. Timing Diagram for Each Device Above the Winning Device in Block Number 1


Figure 10-14. Timing Diagram for Globally Winning Device in Block Number 1


Figure 10-15. Timing Diagram for Devices Below the Winning Device in Block Number 1


Figure 10-16. Timing Diagram for Devices Above the Winning Device in Block Number 2


Figure 10-17. Timing Diagram for Globally Winning Device in Block Number 2


Figure 10-18. Timing Diagram for Devices Below the Winning Device in Block Number 2


Figure 10-19. Timing Diagram for Devices Above the Winning Device in Block Number 3


Figure 10-20. Timing Diagram for Globally Winning Device in Block Number 3


Figure 10-21. Timing Diagram for Devices Below the Winning Device in Block Number 3
(Except the Last Device [Device 14])


Figure 10-22. Timing Diagram for Device Number 3 in Block Number 3 (Device 14 in Depth-Cascaded Table)
The following is the sequence of operation for a single 72-bit Search command (also refer to Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data to be compared. The CMD[2] signal must be driven to a logic 0 .
- Cycle B: The host ASIC continues to drive CMDV HIGH and applies Search command (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles $A$ and B. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] continues to carry the 72-bit data to be compared. Note. For 72-bit searches, the host ASIC must supply the same 72-bit data on DQ[71:0] during both cycles A and B. The even and odd pair of GMRs selected for the compare must be programmed with the same value.
The logical 72-bit Search operation is shown in Figure 10-23. The entire table (fifteen devices of 72-bit entries) is compared to a 72-bit word K (presented on the DQ bus in both cycles A and B of the command) using the GMR and local mask bits. The effective GMR is the 72-bit word specified by the identical value, in both even and odd GMR pairs in each of banks in each of the fifteen devices, and selected by the GMR index in the command's cycle A. The 72-bit word K (presented on the DQ bus in both
cycles $A$ and $B$ of the command) is also stored in both even and odd comparand register pairs, in each of the banks of all fifteen devices, and selected by the comparand register index in command cycle B. In the $\times 72$ configuration, the even comparand register can be subsequently used by the Learn command only in the first non-full device. The word K (presented on the DQ bus in both cycles $A$ and $B$ of the command) is compared with each entry in the table, starting at location 0 . The first matching entry's location address $L$ is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles, the device with LRAM $=1$ and LDEV = 1 will be the default driver for such missed cycles.


Figure 10-23. $\times 72$ Table with Fifteen Devices
The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 72-bit searches in $\times 72$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 72-bit Search command cycle (two CLK2X cycles) is shown in Table 10-14.

Table 10-14. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Cycles |
| :---: | :---: | :---: |
| $1-4(\mathrm{TLSZ}=01)$ | $512 \mathrm{~K} \times 72$ bits | 5 |
| $1-15(\mathrm{TLSZ}=10)$ | $1920 \mathrm{~K} \times 72$ bits | 6 |

For up to fifteen devices in the table (with TLSZ = 10), the latency of the Search from command to SRAM access cycle is 6 . In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-15.
Table 10-15. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

10.6.3 144-bit Search on Tables Configured as $\times 144$ using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in Figure 10-24. The following are parameters that are programmed into the four devices.

- First three devices (devices $0-2$, banks 0 and 1): CFG $=0101010101010101$, TLSZ $=01$, HLAT $=010$, LRAM $=0$, and LDEV $=0$.
- First three devices (devices 3, Bank 0): CFG $=0101010101010101$, $\mathrm{TLSZ}=01, \mathrm{HLAT}=010$, LRAM $=0$, and LDEV $=0$.
- Fourth device (device 3, Bank 1): CFG $=0101010101010101, T L S Z=01, H L A T=010, L R A M=1$, and LDEV $=1$.

Note. All four devices must be programmed with the same value of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 3 in this case).

Figure 10-25 shows the timing diagram for a Search command in the 144-bit-configured table of four devices for device number 0. Figure $10-26$ shows the timing diagram for a Search command in the 144-bit-configured table consisting of four devices for device number 1. Figure $10-27$ shows the timing diagram for a Search command in the 144-bit configured table consisting of four devices for device number 3 (the last device in this specific table). For these timing diagrams, the four 144-bit searches are performed sequentially, and the following Hit/Miss assumptions are made (see Table 10-16).

Table 10-16. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Hit | Miss |
| Device 1 | Miss | Hit | Hit | Miss |
| Devices 2-6 | Miss | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Hit | Hit |



Figure 10-24. Hardware Diagram for a Table with Four Devices


Figure 10-25. Timing Diagram for 144-bit Search Device Number 0


Figure 10-26. Timing Diagram for 144-bit Search Device Number 1


Figure 10-27. Timing Diagram for 144-bit Search Device Number 7 (Last Device)
The following is the sequence of operation for a single 144-bit Search command (also see Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the same bits that will be driven by this device on SADR[23:21] if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) in order to be compared against all even locations. The CMD[2] signal must be driven to logic 0 .
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the SSR index that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] is driven with 72-bit data ([71:0]) compared against all odd locations.

The logical 144-bit Search operation is shown in Figure 10-28. The entire table (four devices of 144-bit entries) is compared to a 144-bit word $K$ (presented on the DQ bus in cycles $A$ and $B$ of the command) using the GMR and local mask bits. The GMR is the 144-bit word specified by the even and odd global mask pair selected by the GMR index in the command's cycle A. The 144bit word $K$ (presented on the $D Q$ bus in cycles $A$ and $B$ of the command) is also stored in the even and odd comparand registers specified by the comparand register index in command cycle B. In $\times 144$ configurations, the even and odd comparand registers can be subsequently used by the Learn command in only one of the devices (the first non-full device). The word K (presented on the DQ bus in cycles $A$ and $B$ of the command) is compared to each entry in the table starting at location 0 . The first matching entry's location, address $L$, is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles the device with LRAM = 1 (the default driving device for the SRAM bus) and LDEV $=1$ (the default driving device for SSF and SSV signals) will be the default driver for such missed cycles. ${ }^{[16]}$


Figure 10-28. $\times 144$ Table with Four Devices
The Search command is a pipelined operation and executes a Search at half the rate of the frequency of CLK2X for 144-bit searches in $\times 144$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 144-bit Search command cycle (two CLK2X cycles) is shown in Table 10-17.

Table 10-17. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Cycles |
| :---: | :--- | :---: |
| $1-4(\mathrm{TLSZ}=01)$ | $256 \mathrm{~K} \times 144$ bits | 5 |
| $1-15(\mathrm{TLSZ}=10)$ | $960 \mathrm{~K} \times 144$ bits | 6 |

For one to four devices in the table and TLSZ $=01$, Search latency from command to SRAM access cycle is 5 . In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-18.
Table 10-18. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

Note:
16. During 144-bit searches of 144 -bit-configured tables, the Search hit will always be at an even address.

### 10.6.4 144-bit Search on Tables Configured as $\times 144$ using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in Figure 10-29. Each of the four blocks in the diagram represents a block of four CYNSE70256 devices (except the last, which has three devices). The diagram for a block of four devices is shown in Figure 10-30. The following are the parameters programmed into the fifteen devices.

- First fourteen devices (devices $0-13$, Bank 0 and 1): CFG $=0101010101010101$, TLSZ $=10$, HLAT $=001$, LRAM $=0$, and LDEV $=0$.
- Fifteenth device (device 14, Bank 0): CFG = 0101010101010101, TLSZ = 10, HLAT $=001$, LRAM $=0$, and LDEV $=0$.
- Fifteenth device (device 14, Bank 1): CFG $=0101010101010101, \mathrm{TLSZ}=10, \mathrm{HLAT}=001$, LRAM $=1$, and LDEV $=1 .{ }^{[17]}$

The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in Table 10-19. For the purpose of illustrating the timings, it is further assumed that the there is only one device with a matching entry in each of the blocks. Figure 1031 shows the timing diagram for a Search command in the 144-bit-configured table (fifteen devices) for each of the four devices in block number 0 . Figure $10-32$ shows the timing diagram for a Search command in the 144 -bit-configured table (fifteen devices) for all the devices above the winning device in block number 1. Figure 10-33 shows the timing diagram for the globally winning device (the final winner within its own and all blocks) in block number 1. Figure 10-34 shows the timing diagram for all the devices below the globally winning device in block number 1. Figure 10-35, Figure 10-36, and Figure 10-37 show the timing diagrams of the devices above globally winning device, the globally winning device, and the devices below the globally winning device, respectively, for block number 2. Figure 10-38, Figure 10-39, Figure 10-40, and Figure 10-41 show, respectively, the timing diagrams of the devices above the globally winning device, the globally winning device, and devices below the globally winning device except the last device (device 14), and then the last device (device 14) for block number 3.

The 144-bit Search operation is pipelined and executes as follows. In the fifth cycle after the Search command, the devices in a block (being less than or equal to four devices resolving the winner within them using the LHI[6:0] and LHO[1:0] signalling mechanism) arbitrate for a winner among them. In the sixth cycle after the Search command, the blocks (of devices) resolve the winning block through the $\mathrm{BHI}[2: 0]$ and $\mathrm{BHO}[2: 0]$ signalling mechanism. The winning device in the winning block is the global winning device for a Search operation.

Table 10-19. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ | $\mathbf{4}$ |
| :---: | :---: | :---: | :---: | :---: |
| Block 0 | Miss | Miss | Miss | Miss |
| Block 1 | Miss | Miss | Hit | Miss |
| Block 2 | Miss | Hit | Hit | Miss |
| Block 3 | Hit | Hit | Miss | Miss |

## Note:

17. All fifteen devices must be programmed with the same value of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 30 in this case).


Figure 10-29. Hardware Diagram for a Table with Fifteen Devices


Figure 10-30. Hardware Diagram for a Table with Four Devices


Figure 10-31. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)


Figure 10-32. Timing Diagram for Each Device Above the Winning Device in Block Number 1


Figure 10-33. Timing Diagram for Globally Winning Device in Block Number 1


Figure 10-34. Timing Diagram for Devices Below the Winning Device in Block Number 1


Figure 10-35. Timing Diagram for Devices Above the Winning Device in Block Number 2


Figure 10-36. Timing Diagram for Globally Winning Device in Block Number 2


Figure 10-37. Timing Diagram for Devices Below the Winning Device in Block Number 2


Figure 10-38. Timing Diagram for Devices Above the Winning Device in Block Number 3


Figure 10-39. Timing Diagram for Globally Winning Device in Block Number 3


Figure 10-40. Timing Diagram for Devices Below the Winning Device in Block Number 3 Except Device 14 (the Last Device)


Figure 10-41. Timing Diagram for Device Number 2 in Block Number 3 (Device 14 in a Depth-Cascaded Table)
The following is the sequence of operation for a single 144-bit Search command (also refer to "Command and Command Parameters," Section 10.2 on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair for use in this Search operation. CMD[8:6] signals must be driven with the bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) in order to be compared against all even locations. The CMD[2] signal must be driven to logic 0 .
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[5:2] must be driven by the index of the comparand register pair for storing the 144-bit word presented on the DQ bus during cycles $A$ and $B$. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of $\operatorname{SSR}[0: 7]$ ). The $\operatorname{DQ}[71: 0]$ is driven with 72-bit data ([71:0]) to be compared against all odd locations.
The logical 144-bit Search operation is shown in Figure 10-42. The entire table of fifteen devices (consisting of 144-bit entries) is compared against a 144 -bit word $K$ that is presented on the DQ bus in cycles $A$ and $B$ of the command using the GMR and local mask bits. The GMR is the 144-bit word specified by the even and odd global mask pair selected by the GMR Index in the command's cycle A.

The 144 -bit word $K$ that is presented on the $D Q$ bus in cycles $A$ and $B$ of the command is also stored in the even and odd comparand registers specified by the comparand register index in command cycle B. In $\times 144$ configurations, the even and odd comparand registers can subsequently be used by the Learn command in only the first non-full device. ${ }^{[18,19]}$


Figure 10-42. $\times 144$ Table with Fifteen Devices
The Search command is a pipelined operation. It executes a Search at half the rate of the frequency of CLK2X for 144-bit searches in $\times 144$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 144-bit Search command cycle (two CLK2X cycles) is shown in Table 10-20.

Table 10-20. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Cycles |
| :---: | :---: | :---: |
| $1-4(\mathrm{TLSZ}=01)$ | $256 \mathrm{~K} \times 144$ bits | 5 |
| $1-14(\mathrm{TLSZ}=10)$ | $960 \mathrm{~K} \times 144$ bits | 6 |

Search latency from command to the SRAM access cycle is 6 for $1-15$ devices in the table and TLSZ $=10$. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-21.
Table 10-21. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

### 10.6.5 288-bit Search on $\times 288$-configured Tables using up to Four CYNSE70256 Devices

The hardware diagram of the Search subsystem of four devices is shown in Figure 10-43. The following are the parameters programmed into the four devices.

- First seven devices (devices 0-6, Bank 0 and 1): CFG $=1010101010101010$, TLSZ $=01$, HLAT $=000$, LRAM $=0$, and LDEV $=0$.
- Fourth device (device 7, Bank 0): CFG $=1010101010101010, T L S Z=01, H L A T=000$, LRAM $=0$, and LDEV $=0$.
- Fourth device (device 7, Bank 1): CFG $=1010101010101010, T L S Z=01$, HLAT $=000$, LRAM $=1$, and LDEV $=1 .[20]$


## Notes:

18. The Learn command is supported for only one of the blocks consisting of up to four devices in a depth-cascaded table of more than one block. The word K that is presented on the DQ bus in cycles A and B of the command is compared with each entry in the table, starting at location 0 . The first matching entry's location address $L$ is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see Section 12.0, "SRAM Addressing," on page 86). The global winning device will drive the bus in a specific cycle. On global miss cycles the device with LRAM $=1$ (the default driving device for the SRAM bus) and LDEV $=1$ (the default driving device for SSF and SSV signals) will be the default driver for such missed cycles.
19. During 144-bit searches of 144-bit-configured tables, the Search hit will always be at an even address.
20. All four devices must be programmed with the same values of TLSZ and HLAT. Only the last bank of the last device in the table must be programmed with LRAM $=1$ and LDEV = 1 (device number 3 in this case).

Figure $10-44$ shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 0 . Figure $10-45$ shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 1. Figure $10-46$ shows the timing diagram for a Search command in the 288-bit-configured table of four devices for device number 3 (the last device in this table). In these timing diagrams, three 288 -bit searches are performed sequentially. The Hit/Miss assumptions were made as shown in Table 10-22.

Table 10-22. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ |
| :---: | :---: | :---: | :---: |
| Device 0 | Hit | Miss | Miss |
| Device 1 | Miss | Hit | Miss |
| Devices 2-6 | Miss | Miss | Miss |
| Device 7 | Miss | Miss | Miss |



Figure 10-43. Hardware Diagram for a Table with Four Devices


Figure 10-44. Timing Diagram for 288-bit Search Device Number 0


Figure 10-45. Timing Diagram for 288-bit Search Device Number 1


Figure 10-46. Timing Diagram for 288-bit Search Device Number 3 (Last Device)
The following is the sequence of operation for a single 288-bit Search command (see also Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair used for bits [287:144] of the data being searched in this operation.
DQ[71:0] must be driven with the 72-bit data ([287:216]) to be compared against all locations 0 in the four-word 72-bit page. The CMD[2] signal must be driven to logic 1. [21]
- Cycle B: The host ASIC continues to drive CMDV HIGH and applies Search command code (10) on CMD[1:0]. The DQ[71:0] is driven with the 72-bit data ([215:144]) to be compared against all locations 1 in the four 72-bits-word pages.
- Cycle C: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair used for bits [143:0] of the data being searched. CMD[8:6] signals must be driven with the bits that will be driven on SADR[23:21] by this device if it has a hit. DQ[71:0] must be driven with the 72-bit data ([143:72]) to be compared against all locations 2 in the four 72-bits-word page. The CMD[2] signal must be driven to logic 0 .
- Cycle D: The host ASIC continues to drive CMDV HIGH and applies Search command code (10) on CMD[1:0]. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for the description of SSR[0:7]). The DQ[71:0] is driven with the 72-bit data ([71:0]) to be compared to all locations 3 in the four 72-bits-word pages. CMD[5:2] is ignored because the Learn instruction is not supported for $\times 288$ tables. ${ }^{[22]}$
The logical 288 -bit Search operation is shown in Figure 10-47. The entire table of 288 -bit entries is compared to a 288 -bit word $K$ that is presented on the DQ bus in cycles A, B, C, and D of the command using the GMR and the local mask bits. The GMR is the 288-bit word specified by the two pairs of GMRs selected by the GMR indexes in command cycles $A$ and $C$ in each of the four devices. The 288-bit word $K$ that is presented on the $D Q$ bus in cycles $A, B, C$, and $D$ of the command is compared to each entry in the table starting at location 0 . The first matching entry's location address $L$ is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (Section 12.0, "SRAM Addressing," on page 86). ${ }^{[23]}$


## Notes:

21. $\mathrm{CMD}[2]=1$ signals that the Search is a 288-bit search. $\mathrm{CMD}[8: 3]$ in this cycle is ignored.
22. For 288-bit searches, the host ASIC must supply four distinct 72-bit data words on DQ[71:0] during cycles A, B, C, and D. The GMR index in cycle A selects a pair of GMRs in each of the banks of each of the four devices that apply to DQ data in cycles A and B. The GMR index in cycle C selects a pair of GMRs in each of the banks of each of the four devices that apply to DQ data in cycles $C$ and $D$.


Figure 10-47. $\times 288$ Table with Four Devices
The Search command is a pipelined operation and executes a Search at one-fourth the rate of the frequency of CLK2X for 288bit searches in $\times 288$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 288-bit Search command (measured in CLK cycles) from the CLK2X cycle that contains the C and D cycles is shown in Table 10-23.

Table 10-23. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Dycles |
| :---: | :---: | :---: |
| $1-4(T L S Z=01)$ | $128 \mathrm{~K} \times 288$ bits | 5 |
| $1-15(\mathrm{TLSZ}=10)$ | $480 \mathrm{~K} \times 288$ bits | 6 |

Search latency from command to SRAM access cycle is 5 for only a single device in the table and TLSZ $=01$. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-24.

Table 10-24. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

### 10.6.6 288-bit Search on Tables Configured as $\times 288$ Using up to Fifteen CYNSE70256 Devices

The hardware diagram of the Search subsystem of fifteen devices is shown in Figure 10-48. Each of the four blocks in the diagram represents a block of four CYNSE70256 devices except the last, which has three devices. The diagram for a block of four devices is shown in Figure 10-49. The following are the parameters programmed into the fifteen devices.

- First thirty devices (devices $0-13$, banks 0 and 1 ): $C F G=1010101010101010$, TLSZ $=10$, HLAT $=000$, LRAM $=0$, and LDEV $=0$.
- Thirty-first device (device 30, Bank 0): CFG $=1010101010101010$, $\mathrm{TLSZ}=10, \mathrm{HLAT}=000$, LRAM $=0$, and LDEV $=0$.
- Thirty-first device (device 30, Bank 1): CFG $=1010101010101010, T L S Z=10, H L A T=000, L R A M=1$, and LDEV $=1 .{ }^{[24]}$


## Note:

23. The matching address is always going to be location 0 in a four-entry page for 288 -bit Search (two LSBs of the matching index will be 00 ).
24. All fifteen devices must be programmed with the same value for TLSZ and HLAT. Only the last bank in the last device in the table must be programmed with LRAM = 1 and LDEV = 1 (device number 14 in this case).

The timing diagrams referred to in this paragraph reference the Hit/Miss assumptions defined in Table 10-25. For the purpose of illustrating the timings, it is further assumed that there is only one device with the matching entry in each block. Figure 10-50 shows the timing diagram for a Search command in the 288-bit-configured table of fifteen devices for each of the four devices in block number 0 . Figure $10-51$ shows the timing diagram for a Search command in the 288 -bit-configured table of fifteen devices for all devices above the winning device in block number 1. Figure $10-52$ shows the timing diagram for the globally winning device (the final winner within its own and all blocks) in block number 1. Figure 10-53 shows the timing diagram for the devices below the globally winning device in block number 1. Figure 10-54, Figure 10-55, and Figure 10-56, respectively, show the timing diagrams of the devices above the globally winning device, the globally winning device, and the devices below the globally winning device in block number 2. Figure 10-57, Figure 10-58, Figure 10-59, and Figure 10-60, respectively, show the timing diagrams of the devices above the globally winning device, the globally winning device, the devices below the globally winning device (except device 14), and the last device (device 14) in block number 3.

The 288-bit Search operation is pipelined and executes as follows. Four cycles from the last cycle of the Search command, each of the devices knows the outcome internal to it for that operation. In the fifth cycle from the Search command, the devices in a block (less than or equal to four devices resolving the winner among them using LHI[6:0] and LHO[1:0] signalling mechanisms) arbitrate for a winner. In the sixth cycle after the Search command, the blocks of devices resolve the winning block through $\mathrm{BHI}[2: 0]$ and $\mathrm{BHO}[2: 0]$ signalling mechanisms. The winning device within the winning block is the global winning device for the Search operation.

Table 10-25. Hit/Miss Assumptions

| Search Number | $\mathbf{1}$ | $\mathbf{2}$ | $\mathbf{3}$ |
| :---: | :---: | :---: | :---: |
| Block 0 | Miss | Miss | Miss |
| Block 1 | Miss | Miss | Hit |
| Block 2 | Miss | Hit | Hit |
| Block 3 | Hit | Hit | Miss |




Figure 10-49. Hardware Diagram for a Block of up to Four Devices


Figure 10-50. Timing Diagram for Each Device in Block Number 0 (Miss on Each Device)


Figure 10-51. Timing Diagram for Each Device Above the Winning Device in Block Number 1


Figure 10-52. Timing Diagram for Globally Winning Device in Block Number 1


Figure 10-53. Timing Diagram for Devices Below the Winning Device in Block Number 1


Figure 10-54. Timing Diagram for Devices Above the Winning Device in Block Number 2


Figure 10-55. Timing Diagram for Globally Winning Device in Block Number 2
 Note: |LHO[1:0] is logical "or" of LHO[0] and LHO[1].

Figure 10-56. Timing Diagram for Devices Below the Winning Device in Block Number 2


Figure 10-57. Timing Diagram for Devices Above the Winning Device in Block Number 3


Figure 10-58. Timing Diagram for Globally Winning Device in Block Number 3


Figure 10-59. Timing Diagram for Devices Below the Winning Device in Block Number 3 Except Device 14 (Last Device)


Figure 10-60. Timing Diagram of the Last Device in Block Number 3 (Device 14 in the Table)
The following is the sequence of operation for a single 288-bit Search command (see also Subsection 10.2, "Commands and Command Parameters," on page 19).

- Cycle A: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair used for bits [287:144] of the data being searched. DQ[71:0] must be driven with the 72-bit data ([287:216]) to be compared to all locations 0 in the four 72-bits-word page. The CMD[2] signal must be driven to logic 1. ${ }^{[25]}$
- Cycle B: The host ASIC continues to drive CMDV HIGH and to apply Search command (10) on CMD[1:0]. The DQ[71:0] is driven with the 72-bit data ([215:144]) to be compared to all locations 1 in the four 72-bits-word pages.
- Cycle C: The host ASIC drives CMDV HIGH and applies Search command code (10) on CMD[1:0] signals. \{CMD[10],CMD[5:3]\} signals must be driven with the index to the GMR pair used for the bits [143:0] of the data being searched. CMD[8:6] signals must be driven with the bits that will be driven by this device on $\operatorname{SADR}[23: 21]$ if it has a hit. DQ[71:0] must be driven with the 72 -bit data ([143:72]) to be compared to all locations 2 in the four 72-bits-word pages. The CMD[2] signal must be driven to logic 0 .

Note:
25. $\mathrm{CMD}[2]=1$ signals that the Search is a $\times 288$-bit Search. $\mathrm{CMD}[8: 6]$ is ignored in this cycle.

- Cycle D: The host ASIC continues to drive CMDV HIGH and to apply Search command code (10) on CMD[1:0]. CMD[8:6] signals must be driven with the index of the SSR that will be used for storing the address of the matching entry and the hit flag (see page 14 for a description of SSR[0:7]). The DQ[71:0] is driven with the 72-bit data ([71:0]) to be compared to all locations 3 in the four 72-bits-word pages. CMD[5:2] is ignored because the Learn instruction is not supported for $\times 288$ tables. ${ }^{[26]}$
The logical 288 -bit Search operation is shown in Figure 10-61. The entire table of 288-bit entries is compared to a 288 -bit word $K$ that is presented on the DQ bus in cycles $A, B, C$, and $D$ of the command using the GMR and local mask bits. The GMR is the 288-bit word specified by the two pairs of GMRs selected by the GMR Indexes in command cycles A and C in each of the fifteen devices. The 288-bit word $K$ that is presented on the $D Q$ bus in cycles $A, B, C$, and $D$ of the command is compared to each entry in the table starting at location 0 . The first matching entry's location address $L$ is the winning address that is driven as part of the SRAM address on the SADR[23:0] lines (see "SRAM Addressing" on page 86). ${ }^{[27]}$


Figure 10-61. $\times 288$ Table with Fifteen Devices
The Search command is a pipelined operation and executes a Search at one-fourth the rate of the frequency of CLK2X for 288bit searches in $\times 288$-configured tables. The latency of SADR, CE_L, ALE_L, WE_L, SSV, and SSF from the 288-bit Search command (measured in CLK cycles) from the CLK2X cycle that contains the $\bar{C}$ and $\bar{D}$ cycles is shown in Table 10-26.

Table 10-26. Search Latency from Instruction to SRAM Access Cycle

| Number of Devices | Max Table Size | Latency in CLK Cycles |
| :---: | :---: | :---: |
| $1-8(T L S Z=01)$ | $128 \mathrm{~K} \times 288$ bits | 5 |
| $1-15(\mathrm{TLSZ}=10)$ | $480 \mathrm{~K} \times 288$ bits | 6 |

Search latency from command to SRAM access cycle is 6 for a single device in the table and TLSZ $=10$. In addition, SSV and SSF shift further to the right for different values of HLAT, as specified in Table 10-27.
Table 10-27. Shift of SSF and SSV from SADR

| HLAT | Number of CLK Cycles |
| :---: | :---: |
| 000 | 0 |
| 001 | 1 |
| 010 | 2 |
| 011 | 3 |
| 100 | 4 |
| 101 | 5 |
| 110 | 6 |
| 111 | 7 |

## Notes:

26. For 288-bit searches, the host ASIC must supply four distinct 72-bit data words on DQ[71:0] during cycles A, B, C, and D. The GMR index in cycle A selects a pair of GMRs in each of the fifteen devices that apply to DQ data in cycles $A$ and $B$. The GMR index in cycle $C$ selects a pair of GMRs in each of the fifteen devices that apply to DQ data in cycles C and D.
27. The matching address is always going to be location 0 in a four-entry page for 288 -bit search (two LSBs of the matching index will be 00 ).

### 10.6.7 Mixed-size Searches on Tables Configured with Different Widths Using a CYNSE70256 With CFG_L LOW

This subsection will cover mixed searches ( $\times 72, \times 144$, and $\times 288$ ) with tables of different widths $(\times 72, \times 144, \times 288)$. The sample operation shown is for a single device with CFG $=1010010100000000$ that contains three tables of $\times 72, \times 144$, and $\times 288$ widths. The operation can be generalized to a block of $4-15$ devices using four blocks; the timing and pipeline operation is the same as described previously for fixed searches on a table of one-width size.

Figure 10-62 shows three sequential searches: first, a 72 -bit Search on a $\times 72$-configured table; a 144 -bit Search on a $\times 144$ configured table; finally, a 288 -bit Search on a $\times 288$-configured table that each results in a hit. ${ }^{[28]}$

Figure $10-63$ shows the sample table. Two bits in each 72-bit entry will need to designated as the table number bits. One choice can be the 00 values for the table configured as $\times 72,01$ values for tables configured as $\times 144$, and 10 values for tables configured as $\times 288$. For the above explanation, it is further assumed that bits [71:70] for each entry will be designed as such table designation bits. cycle cycle cycle cycle cycle cycle cycle cycle cycle cycle


Figure 10-62. Timing Diagram for Mixed Search (One Device)


Figure 10-63. Multiwidth Configurations Example
Note:
28. The DQ[71:70] will be 00 in each of the two $A$ and $B$ cycles of the $\times 72$-bit Search (Search1). DQ[71:70] is 01 in each of the $A$ and $B$ cycles of the $\times 144$-bit Search (Search2). DQ[71:70] is 10 in each of the A, B, C, and D cycles of the $\times 288$-bit Search (Search3). By having table designation bits, the CYNSE70256 device enables the creation of many tables in a bank of NSEs of different widths.

### 10.6.8 Mixed-size Searches on Tables Configured to Different Widths using a CYNSE70256 Device with CFG_L HIGH

This subsection will cover the mixed-size searches $(\times 72, \times 144$ and $\times 288$ ) with tables of different widths $(\times 72, \times 144, \times 288)$ with CFG_L set HIGH. The previous subsection described searches on tables of different widths using table designation bits in the data array, which can be wasteful. In order to avoid the waste of these bits and yet support up to three tables of $\times 72, \times 144$ and $\times 288$ widths, CMD[2] and CMD[9] (in CFG_L HIGH mode) in cycle A of the command can be used as shown in Table 10-28.

Table 10-28. Searches with CFG_L Set HIGH

| CMD[9] | CMD[2] | Search |
| :---: | :---: | :--- |
| 0 | 0 | Search 72-bit-configured partitions only. |
| 1 | 0 | Search 144-bit-configured partitions only. |
| $X$ | 1 | Cycles A and B for searching 288-bit-configured partitions. |
| $X$ | 0 | Cycles C and D for searching 288-bit-configured partitions. |

### 10.7 LRAM and LDEV Description

When NSEs are cascaded using multiple CYNSE70256 devices, the SADR, CE_L, and WE_L (three-state signals) are all tied together. In order to eliminate external pull-up and pull downs, one device in a bank is designated the default driver. For nonSearch or nonLearn cycles (see Subsection 10.8, "Learn Command") or Search cycles with a global miss, the SADR, CE_L, and WE_L signals are driven by the device with the LRAM bit set. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SADR, CE_L, and WE_L, and can potentially cause damage to the device(s).

Similarly, when NSEs using multiple CYNSE70256 devices are cascaded, SSF and SSV (also three-state signals) are tied together. In order to eliminate external pull-up and pull downs, one device in a bank is designated as the default driver. For nonSearch cycles or Search cycles with a global miss, the SSF and SSV signals are driven by the device with the LDEV bit set in Bank 1. It is important that only one device in a bank of cascaded NSEs have this bit set. Failure to do so will cause contention on the SSV and SSF, and can potentially cause damage to the device(s).

### 10.8 Learn Command

Bit[0] of each 72-bit data location specifies whether an entry in the database is occupied. If all the entries in a device are occupied, the device asserts a FULO signal to inform the downstream devices that it is full. The result of this communication between depthcascaded devices determines the global FULL signal for the entire table. The FULL signal in the last device determines the fullness of the depth-cascaded table.

The device contains sixteen pairs of internal 72-bit-wide comparand registers that store the comparands as the device executes searches. On a miss by the Search that is signalled to the ASIC through the SSV and SSF signals (SSV = 1, SSF = 0), the host ASIC can apply the Learn command to learn the entry from a comparand register as to the next-free location (see Subsection 7.8, "NFA Register," on page 17).

The NFA register is updated with the new next-free location information following each Write or Learn command. In a depthcascaded table, only a single device will Learn the entry through the application of a Learn instruction. The determination as to which device will Learn is based on the FULI and FULO signals between the devices. The first non-full device learns the entry by storing the contents of the specified comparand registers to the location(s) pointed to by the NFA register. In a $\times 72$-configured table, the Learn command writes a single 72 -bit location. In a $\times 144$-configured table, the Learn command writes the next even and odd 72 -bit locations. In 144-bit mode, bit[0] of the even and odd 72 -bit locations is 0 , indicating that they are cascaded empty, or 1, indicating that they are occupied. The global FULL signal indicates to the table controller (the host ASIC) that all entries within a block are occupied and that no more entries can be learned. The CYNSE70256 device updates the signal after each Write or Learn command to a data array. The Learn command generates a Write cycle to the external SRAM, also using the NFA register as part of the SRAM address (see Section 12.0, "SRAM Addressing," on page 86).
The Learn command is supported on a single block containing up to four devices if the table is configured either as $\times 72$ or $\times 144$; it is not supported for $\times 288$-configured tables. The Learn command is a pipelined operation and lasts for two CLK cycles, as shown in Figure 10-64 and Figure 10-65 where TLSZ $=01$. Figure $10-64$ and Figure $10-65$ assume that the device performing the Learn operation is not the last device in the table and will therefore have its LRAM bit set to 0. ${ }^{[29]}$

Note:
29. The OE L for the device with the LRAM bit set goes HIGH for two cycles for each Learn (one during the SRAM Write cycle and one during the cycle before). The SRAM Write cycle latency from the second cycle of the instruction is shown in Table 10-29.


Figure 10-64. Timing Diagram of Learn (Except on the Last Device [TLSZ = 01])


Figure 10-65. Timing Diagram of Learn on Device Number 3 (TLSZ = 01)

Table 10-29. SRAM Write Cycle Latency from Second Cycle of Learn Instruction

| Number of Devices | Latency in CLK Cycles |
| :---: | :---: |
| $1-4(\mathrm{TLSZ} \mathrm{=} \mathrm{01)}$ | 5 |
| $1-15(\mathrm{TLSZ} \mathrm{=} \mathrm{10)}$ | 6 |

The Learn operation lasts two CLK cycles. The sequence of operation is as follows.

- Cycle 1A: The host ASIC applies the Learn instruction on CMD[1:0] using CMDV = 1. The CMD[5:2] field specifies the index of the comparand register pair that will be written in the data array in the 144-bit-configured table. For a Learn in a 72-bitconfigured table, the even-numbered comparand specified by this index will be written. CMD[8:6] carries the bits that will be driven on SADR[23:21] in the SRAM Write cycle.
- Cycle 1B: The host ASIC continues to drive CMDV to 1, CMD[1:0] to 11, and CMD[5:2] with the comparand pair index. CMD[6] must be set to 0 if the Learn is being performed on a 72-bit-configured table, and to 1 if the Learn is being performed on a 144-bit-configured table.
- Cycle 2: The host ASIC drives CMDV to 0.

At the end of cycle 2, a new instruction can begin. SRAM Write latency is the same as the Search to the SRAM Read cycle. It is measured from the second cycle of the Learn instruction.

### 11.0 Depth Cascading

The NSE application can depth-cascade the devices to various table sizes of different widths ( 72 bits, 144 bits, or 288 bits). The devices perform all the necessary arbitration to decide which device will drive the SRAM bus. Search latency increases as the table size increases; the Search rate itself remains constant.

### 11.1 Depth Cascading up to Four Devices (One Block)

Figure $11-1$ shows that up to four devices can be cascaded to form $512 \mathrm{~K} \times 72,256 \mathrm{~K} \times 144$, or $128 \mathrm{~K} \times 288$ tables. It also shows the interconnection between devices for depth cascading. Each NSE asserts the LHO[1] and LHO[0] signals to inform downstream devices of its result. LHI[6:0] signals for a device are connected to LHO signals of the upstream devices. The host ASIC must program the TLSZ to 01 for each of up to four devices in a block. Only a single device drives the SRAM bus in any single cycle.


Figure 11-1. Hardware Diagram for a Block of up to Four Devices

### 11.2 Depth Cascading up to Fifteen Devices (Four Blocks)

Figure 11-2 shows the cascading of up to four blocks. Each block except the last contains up to four CYNSE70256 devices, and the interconnection within each with the cascading of up to four devices in a block was shown in the previous subsection. ${ }^{[30]}$

## Note:

30. The interconnection between blocks for depth cascading is important. For each Search, a block asserts $\mathrm{BHO}[2]$, $\mathrm{BHO}[1]$, and $\mathrm{BHO}[0]$. The $\mathrm{BHO}[2: 0]$ signals for a block are taken only from the last device in that block. For all other devices within that block, these signals stay open and floating. The host ASIC must program TLSZ to 10 in each of the devices for cascading up to fifteen devices (in up to four blocks).


Figure 11-2. Depth Cascading four Blocks

### 11.3 Depth Cascading for a FULL Signal

Bit[0] of each of the 72-bit entries is designated as a special bit ( $1=$ occupied, $0=$ empty). For each Learn or PIO Write to the data array, each device asserts FULO[1] or FULO[0] depending on whether or not it has any empty locations within it (see Figure 11-3). Each device combines the FULO signals from the devices above it with its own full status to generate a FULL signal that gives the full status of the table up to the device asserting the FULL signal. Figure 11-3 shows the hardware connection diagram for generating the FULL signal that goes back to the ASIC. In a depth-cascaded block of up to four devices, the FULL signal from the last device should be fed back to the ASIC controller to indicate the fullness of the table. The FULL signal of the other devices should be left open. ${ }^{[31]}$

## Note:

31. The Learn instruction is supported for only up to four devices, whereas FULL cascading is allowed only for one block in tables containing more than four devices. In tables for which a Learn instruction is not going to be used, the bit[0] of each 72-bit entry should always be set to 1 .


Figure 11-3. FULL Signal Generation in a Cascaded Table

### 12.0 SRAM Addressing

Table 12-1 describes the commands used to generate addresses on the SRAM address bus. The index [15:0] field contains the address of a 72-bit entry that results in a hit in 72-bit-configured quadrant. It is the address of the 72-bit entry that lies at the 144bit page, and the 288-bit page boundaries in 144-bit- and 288-bit-configured quadrants, respectively.

Section 7.0, "Registers," on page 13 of this datasheet describes the NFA and SSR registers. ADR[15:0] contains the address supplied on the DQ bus during PIO access to the CYNSE70256. Command bits 8, 7, and $6\{C M D[8: 6]\}$ are passed from the command to the SRAM address bus. See Section 10.0, "Commands," on page 19, for more information. ID[4:0] is the ID of the device driving the SRAM bus (see Section 18.0, "Pinout Descriptions and Package Diagrams," on page 103, for more information).
Table 12-1. SRAM Bus Address

| Command | SRAM Operation | $\mathbf{2 3}$ | $\mathbf{2 2}$ | $\mathbf{2 1}$ | $[\mathbf{2 0 : 1 7 ]}$ | $\mathbf{1 6}$ | [15:0] |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Search | Read | C 8 | C 7 | C 6 | $\mathrm{ID}[4: 1]$ | bank | Index[15:0] |
| Learn | Write | C 8 | C 7 | C 6 | $\mathrm{ID}[4: 1]$ | bank | NFA[15:0] |
| PIO Read | Read | C 8 | C 7 | C 6 | $\mathrm{ID}[4: 1]$ | bank | ADR16:0] |
| PIO Write | Write | C 8 | C 7 | C 6 | $\mathrm{ID}[4: 1]$ | bank | ADR[15:0] |
| Indirect Access | Write/Read | C 8 | C 7 | C 6 | $\mathrm{ID}[4: 1]$ | bank | SSR[15:0] |

### 12.1 SRAM PIO Access

The remainder of Section 12.0 describes SRAM Read and Write operations.
SRAM Read enables Read access to the off-chip SRAM containing associative data. The latency from the issuance of the Read instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the value programmed for the TLSZ parameter in the device configuration register. The latency of the ACK from the Read instruction is the same as that from the Search instruction to the SRAM address latency, plus the HLAT programmed in the configuration register. ${ }^{[32,33]}$

## Notes:

32. SRAM Read is a blocking operation-no new instruction can begin until the ACK is returned by the selected device performing the access. SRAM Write enables Write access to the off-chip SRAM containing associative data. The latency from the second cycle of the Write instruction to the appearance of the address on the SRAM bus is the same as the Search instruction latency, and will depend on the TLSZ value parameter programmed in the device configuration register.
33. SRAM Write is a pipelined operation.

### 12.2 SRAM Read with a Table of up to Four Devices

The following explains the SRAM Read operation completed through a table of up to four devices using the following parameter: TLSZ $=01$. Figure 12-1 diagrams a block of four devices. The following assumes that SRAM access is successfully achieved through CYNSE70256 device number 0. Figure 12-2 and Figure 12-3 show timing diagrams for device number 0 and device number 3, respectively.

- Cycle 1A: The host ASIC applies the Read instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10 , to select the SRAM address. The host ASIC selects the device for which ID[4:0] matches the DQ[25:21] lines. During this cycle the host ASIC also supplies SADR[23:21] on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction on CMD[1:0], using CMDV $=1$. The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive $D Q[71: 0]$.
- Cycle 5: The selected device continues to drive DQ[71:0] and drives ACK from High-Z to LOW.
- Cycle 6: The selected device drives the Read address on SADR[23:0], and drives ACK HIGH, CE_L LOW, WE_L HIGH, and ALE_L LOW.
- Cycle 7: The selected device drives CE_L, ALE_L, WE_L, and the DQ bus in a three-state condition. It continues to drive ACK LOW.

At the end of cycle 7, the selected device floats ACK in a three-state condition. A new command can begin.


Figure 12-1. Hardware Diagram of a Block of Four Devices


Figure 12-2. SRAM Read Through Device Number 0 in a Block of Four Devices


TLSZ $=01$, HLAT $=000, \operatorname{LRAM}=1$, LDEV $=1$
Figure 12-3. SRAM Read Timing for Device Number 7 in a Block of Four Devices

### 12.3 SRAM Read with a Table of up to Fifteen Devices

The following explains the SRAM Read operation accomplished through a table of up to fifteen devices, using the following parameter: TLSZ = 10. The hardware diagram is shown in Figure 12-4. The following assumes that SRAM access is being accomplished through CYNSE70256 device number 0 , and that device number 0 is the selected device. Figure 12-5 and Figure 12-6 show the timing diagrams for device number 0 and device number 14, respectively.

- Cycle 1A: The host ASIC applies the Read instruction to CMD[1:0] using CMDV = 1. The DQ bus supplies the address, with DQ[20:19] set to 10 , to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. During this cycle, the host ASIC also supplies SADR[23:21] on CMD[8:6].
- Cycle 1B: The host ASIC continues to apply the Read instruction to CMD[1:0], using CMDV = 1 . The DQ bus supplies the address, with DQ[20:19] set to 10, to select the SRAM address.
- Cycle 2: The host ASIC floats DQ[71:0] to a three-state condition.
- Cycle 3: The host ASIC keeps DQ[71:0] in a three-state condition.
- Cycle 4: The selected device starts to drive DQ[71:0].
- Cycles 5 to 6: The selected device continues to drive DQ[71:0].
- Cycle 7: The selected device continues to drive DQ[71:0], and drives an SRAM Read cycle.
- Cycle 8: The selected device drives ACL from $Z$ to LOW.
- Cycle 9: The selected device drives ACK to HIGH.
- Cycle 10: The selected device drives ACK from HIGH to LOW.

At the end of cycle 10, the selected device floats ACK in a three-state condition.


Figure 12-4. Hardware Diagram of Fifteen Devices Using Four Blocks


Figure 12-5. SRAM Read Through Device Number 0 in a Bank of Fifteen Devices (Device Number 0 Timing)


TLSZ = 10, HLAT = 010, LRAM = 1, LDEV = 1
Figure 12-6. SRAM Read Through Device Number 0 in a Bank of Fifteen Devices (Device Number 14 Timing)

### 12.4 SRAM Write with a Table of up to Four Devices

The following explains the SRAM Write operation accomplished through a table(s) of up to four devices with the following parameters ( TLSZ = 01). The hardware diagram for this table is shown in Figure 12-7. The following assumes that SRAM access is achieved through CYNSE70256 device number 0. Figure 12-8 and Figure 12-9 show the timing diagram for device number 0 and device number 3 , respectively.

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[23:21] on CMD[8:6] in this cycle. ${ }^{[34]}$
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV $=1$. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. ${ }^{[34]}$
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.

At the end of cycle 3, a new command can begin. Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.


Figure 12-7. Hardware Diagram of a Block of Four Devices


Figure 12-8. SRAM Write Through Device Number 0 in a Block of Four Devices
Note:
34. CMD[2] must be set to 0 for SRAM Write because burst Writes into the SRAM are not supported.


Figure 12-9. SRAM Write Timing for Device Number 3 in Block of Four Devices

### 12.5 SRAM Write with Table(s) Consisting of up to Fifteen Devices

The following explains the SRAM Write operation accomplished through a table of up to fifteen devices with the following parameter: TLSZ = 10. The hardware diagram is shown in Figure 12-10. The following assumes that SRAM access is accomplished through the selected device: CYNSE70256 device number 0. Figure 12-11 and Figure 12-12 show timing diagrams for device number 0 and device number 14, respectively.

- Cycle 1A: The host ASIC applies the Write instruction on CMD[1:0] using CMDV = 1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. The host ASIC selects the device for which the ID[4:0] matches the DQ[25:21] lines. The host ASIC also supplies SADR[23:21] on CMD[8:6] in this cycle. ${ }^{[34]}$
- Cycle 1B: The host ASIC continues to apply the Write instruction on CMD[1:0] using CMDV =1. The DQ bus supplies the address with DQ[20:19] set to 10 to select the SRAM address. ${ }^{[34]}$
- Cycle 2: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.
- Cycle 3: The host ASIC continues to drive DQ[71:0]. The data in this cycle is not used by the CYNSE70256 device.

At the end of cycle 3, a new command can begin. The Write is a pipelined operation, but the Write cycle appears at the SRAM bus with the same latency as that of a Search instruction, as measured from the second cycle of the Write command.


Figure 12-10. Table of Fifteen Devices (Four Blocks)


Figure 12-11. SRAM Write Through Device Number 0 in Bank of Fifteen Devices (Device 0 Timing)


Figure 12-12. SRAM Write Through Device Number 0 in Bank of Fifteen CYNSE70256 Devices (Device Number 14 Timing)

### 13.0 Power

CYNSE70256 has two separate power supplies, one for the core $\left(\mathrm{V}_{\mathrm{DD}}\right)$ and another for the $\mathrm{IOs}\left(\mathrm{V}_{\mathrm{DDQ}}\right)$.

### 13.1 Power-up Sequence

Proper power-up sequence is required to correctly initialize the Cypress Network Search Engines before functional access to the device can begin. RST_L and TRST_L should be held low before the power supplies ramp up. RST_L must be set low for a duration of time afterward and then sèt high. The following steps describe the proper power-up sequence.

1. Set RST_L and TRST_L low.
2. Power up $V_{D D}, V_{D D Q}$ and start running CLK1X when operating in CLK1X mode or CLK2X and PHS_L when operating in CLK2X mode. The order in which these signals (including $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{DDQ}}$ ) are applied is not critical.
3. RST_L should be held low for 0.5 ms (PLL lock time requirement). In CLK1X mode, the counting starts on the first rising edge of CLK1X after both $V_{D D}$ and $V_{D D Q}$ have reached their steady state voltages. In CLK2X mode, the counting starts on the first rising edge of CLK2X when PHS_L is high, after both $V_{D D}$ and $V_{D D Q}$ have reached their steady state voltages.
4. Continue to hold RST_L low for a minimum of 32 CLK1X cycles (when operating in CLK1X mode) or 64 CLK2X cycles (when operating in CLK2X mode). Set RST_L to high afterward to complete the power-up sequence. For JTAG reset, TRST_L can be brought high after $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{DDQ}} \overline{\text { have both reached their steady state voltages. }}$

Figure 13-1 and Figure 13-2 illustrate the proper sequences of the power-up operation.


Figure 13-1. Power-up Sequence (CLK2x)


Figure 13-2. Power-up Sequence (CLK1x)

### 13.2 Power Consumption

The following figure depicts expected power consumption over a range of frequencies. The calculations assume $100 \%$ of the operations will be SEARCH operations. If an application includes other operations such as READ or WRITE, then power consumption will be lower. The worst case line indicates power consumption when the I/Os switch 100 percent of the time. The other lines (All Search Hit and All Search Miss) assume the I/Os switch 50\% of the time.


Figure 13-3. Power Consumption of CYNSE70256

### 14.0 Application

Figure 14-1 shows how an NSE subsystem can be formed using a host ASIC and a CYNSE70256 bank. It also shows how this NSE subsystem is integrated in a switch or router. The CYNSE70256 device can access synchronous and asynchronous SRAMs by allowing the host ASIC to set the same HLAT parameter in all NSEs within a bank of NSEs.


Figure 14-1. Sample Switch/Router Using the CYNSE70256 Device

### 15.0 JTAG (1149.1) Testing

The CYNSE70256 device supports the Test Access Port and Boundary Scan Architecture as specified in IEEE JTAG Standard Number 1149.1. The pin interface to the chip consists of five signals with the standard definitions: TCK, TMS, TDI, TDO, and TRST_L. Table 15-1 describes the operations that the test access port controller supports, and Table 15-2 describes the TAP

Device ID Register. ${ }^{[35]}$ Note: To disable JTAG functionality, connect the TCK, TMS and TDI pins to VDDQ through a pull-up, and TRST_L to ground through a pull-down
Table 15-1. Supported Operations

| Instruction | Type | Description |
| :---: | :---: | :--- |
| SAMPLE/PRELOA <br> D | Mandatory | Sample/Preload. This operation loads the values of signals going to and from I/O pins into the <br> boundary scan shift register to provide a snapshot of the normal functional operation. |
| EXTEST | Mandatory | External Test. This operation uses boundary scan values shifted in from the TAP to test connec- <br> tivity external to the device. |
| BYPASS | Mandatory | Bypass. This operation bypasses the device in a JTAG chain by loading a single bit shift register <br> between TDI and TDO and provides a minimum-length serial path when no test operation is <br> required. |
| IDCODE | Optional | Device JTAG ID Code. This operation selects the JTAG Identification register and output the <br> IDCODE field serially through TDO. |
| CLAMP | Optional | Output Clamp. This operation drives preset values onto the outputs of the device. |
| HIGHZ | Optional | High-Z Output. This operation sets the device output signals in high impedance state. |

Table 15-2. TAP Device ID Register

| Field | Range | Initial Value | Description |
| :---: | :---: | :---: | :--- |
| Revision | $[31: 28]$ | 0001 | Revision Number. This is the current device revision number. Numbers <br> start from one and increment by one for each revision of the device. |
| Part Number | $[27: 12]$ | 0000000000000100 | This is the part number for the device. |
| MFID | $[11: 1]$ | $000 \_1101 \_1100$ | Manufacturer ID. This field is the same as the manufacturer ID used <br> in the TAP controller. |
| LSB | $[0]$ | 1 | Least significant bit. |


| Note: |
| :--- |
| 35. To disable JTAG functionality, connect the TCK, TMS, and TDI pins toVDDQ through a pull-up, and TRST_L to ground through a pull-down. |

### 16.0 Electrical Specifications

This section describes the electrical specifications, capacitance, operating conditions, DC characteristics, and AC timing parameters for the CYNSE70256 device (see Table 16-1 and Table 16-2).
Table 16-1. DC Electrical Characteristics for CYNSE70256

| Parameter | Description | Conditions | Min. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{I}_{\text {LI }}$ | Input leakage current | $\mathrm{V}_{\text {DDQ }}=\mathrm{V}_{\text {DDQ }}$ Max, $\mathrm{V}_{\text {IN }}=0$ to $\mathrm{V}_{\text {DDQ }}$ Max | -20 | 20 | $\mu \mathrm{A}$ |
| LLo | Output leakage current | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\text {DDQ }}$ Max, $\mathrm{V}_{\text {IN }}=0$ to $\mathrm{V}_{\text {DDQ }}$ Max | -20 | 20 | $\mu \mathrm{A}$ |
| $\mathrm{V}_{\text {IL }}$ | Input LOW voltage (2.5V) |  | -0.3 | 0.7 | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH voltage (2.5V) |  | 1.7 | $\begin{gathered} \mathrm{V}_{\mathrm{DDQ}}+ \\ 0.3 \end{gathered}$ | V |
| $\mathrm{V}_{\mathrm{OL}}$ | Output LOW voltage (2.5V) | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \mathrm{Min}, \mathrm{I}_{\mathrm{OL}}=8 \mathrm{~mA}$ |  | 0.4 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output HIGH voltage (2.5V) | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \mathrm{Min}, \mathrm{I}_{\mathrm{OH}}=8 \mathrm{~mA}$ | 2.0 |  | V |
| $\mathrm{V}_{\text {IL }}$ | Input LOW voltage (3.3V) |  | -0.3 | 0.8 |  |
| $\mathrm{V}_{\mathrm{IH}}$ | Input HIGH voltage (3.3V) |  | 2.0 | $\begin{gathered} \mathrm{V}_{\mathrm{DDQ}}+ \\ 0.3 \end{gathered}$ |  |
| $\mathrm{V}_{\mathrm{OL}}$ | Output LOW voltage (3.3V) | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \mathrm{Min}, \mathrm{I}_{\mathrm{OL}}=8 \mathrm{~mA}$ |  | 0.4 | V |
| $\mathrm{V}_{\mathrm{OH}}$ | Output HIGH voltage (3.3V) | $\mathrm{V}_{\mathrm{DDQ}}=\mathrm{V}_{\mathrm{DDQ}} \mathrm{Min}, \mathrm{I}_{\mathrm{OH}}=8 \mathrm{~mA}$ | 2.4 |  | V |
| $\mathrm{I}_{\text {DD2 }}$ | $3.3 \mathrm{~V} / 2.5 \mathrm{~V}$ supply current at $\mathrm{V}_{\mathrm{DD}} \mathrm{Max}$ | $83-\mathrm{MHz}$ search rate, $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}$ |  | 600 | mA |
| $\mathrm{I}_{\text {DD2 }}$ | $3.3 \mathrm{~V} / 2.5 \mathrm{~V}$ supply current at $\mathrm{V}_{\mathrm{DD}} \mathrm{Max}$ | $66-\mathrm{MHz}$ search rate, $\mathrm{I}_{\text {OUT }}=0 \mathrm{~mA}$ |  | 480 | mA |
| IDDI | 1.5 V supply current at $\mathrm{V}_{\mathrm{DD}}$ (Typical) | 83-MHz search rate |  | $7.4{ }^{[41]}$ | A |
| $\mathrm{I}_{\text {DDI }}$ | 1.5 V supply current at $\mathrm{V}_{\mathrm{DD}}$ (Typical) | 66-MHz search rate |  | $5.9{ }^{[41]}$ | A |


| Parameter | Description | Max. | Unit |
| :--- | :--- | :---: | :---: |
| $\mathrm{C}_{\mathrm{IN}}$ | Input capacitance | 12 | $\mathrm{pF}^{[36]}$ |
| $\mathrm{C}_{\text {OUT }}$ | Output capacitance | 12 | $\mathrm{pF}^{[37]}$ |

Table 16-2. Operating Conditions for CYNSE70256

| Parameter | Description | Min. | Max. | Unit |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{V}_{\mathrm{DDQ}}=3.3 \mathrm{~V}$ | Operating voltage for I/O | 3.135 | 3.465 | V |
| $\mathrm{~V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}$ | Operating voltage for I/O | 2.375 | 2.625 | V |
| $\mathrm{~V}_{\mathrm{DD}}$ | Operating supply voltage | 1.425 | 1.575 | V |
| $\mathrm{~V}_{\mathrm{IH}}$ | Input HIGH voltage ${ }^{[39]}(2.5 \mathrm{~V})$ | 1.7 | $\mathrm{~V}_{\mathrm{DDQ}}+0.3$ | V |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input LOW voltage ${ }^{[38]}(2.5 \mathrm{~V})$ | -0.3 | 0.7 | V |
| $\mathrm{~V}_{\mathrm{IH}}$ | Input HIGH voltage ${ }^{[40]}(3.3 \mathrm{~V})$ | 2.0 | $\mathrm{~V}_{\mathrm{DDQ}}+0.3$ | V |
| $\mathrm{~V}_{\mathrm{IL}}$ | Input LOW voltage ${ }^{[38]}(3.3 \mathrm{~V})$ | -0.3 | 0.8 | V |
| $\mathrm{~T}_{\mathrm{A}}$ | Ambient operating temperature | 0 | 70 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{J}}$ | Junction temperature | 0 | ${ }^{\circ} \mathrm{C}$ |  |
|  | Supply voltage tolerance | $-5 \%$ | $+5 \%$ |  |

## Notes:

36. $f=1 \mathrm{MHz}, \mathrm{V}_{\mathrm{IN}}=0 \mathrm{~V}$.
37. $\mathrm{f}=1 \mathrm{MHz}, \mathrm{V}_{\mathrm{OUT}}=0 \mathrm{~V}$.
38. Minimum allowable applies to undershoot only
39. Maximum allowable applies to overshoot only ( $\mathrm{V}_{\mathrm{DDQ}}$ is 2.5 V supply).
40. Maximum allowable applies to overshoot only ( $\mathrm{V}_{\mathrm{DDQ}}$ is 3.3 V supply).
41. Typical. 80\% Compare utilization.
42. Please refer to "CYNSE70256 Airflow and Heat Sink Requirements" application note.

### 17.0 AC Timing Waveforms

Table 17-1 and Table 17-2 show the AC timing parameters for the CYNSE70256 device. Table 17-3 shows the AC test conditions for the CYNSE70256 device. Figure 17-1 shows the input wave form for the CYNSE70256 device. Figure 17-2 and Figure 17-3 show the output load and output load equivalent of the CYNSE70256 device. Figure 17-4 shows timing wave form diagrams for CLK2X. Figure 17-5 details timing wave form diagrams for CLK1X.
Table 17-1. AC Timing Parameters with CLK2X

| Parameter | Description | CYNSE70256-066 |  | CYNSE70256-083 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min. | Max. | Min. | Max. |  |
| $\mathrm{f}_{\text {CLOCK }}$ | CLK2X frequency. | 40 | 133 | 40 | 166 | MHz |
| $\mathrm{t}_{\text {CLOK }}$ | PLL lock time. |  | 0.5 |  | 0.5 | ms |
| $\mathrm{t}_{\mathrm{CKHI}}$ | CLK2X HIGH pulse. ${ }^{43]}$ | 3.0 |  | 2.4 |  | ns |
| $\mathrm{t}_{\text {CKLO }}$ | CLK2X LOW pulse. ${ }^{[43]}$ | 3.0 |  | 2.4 |  | ns |
| $\mathrm{t}_{\text {ISCH }}$ | Input setup time to CLK2X rising edge. ${ }^{[43]}$ | 2.5 |  | 1.8 |  | ns |
| $\mathrm{t}_{\mathrm{IHCH}}$ | Input hold time to CLK2X rising edge. ${ }^{[43]}$ | 0.6 |  | 0.6 |  | ns |
| $\mathrm{t}_{\mathrm{ICSCH}}$ | C]ascaded input setup time to CLK2X rising edge. ${ }^{[43]}$ | 4.2 |  | 3.5 |  | ns |
| $\mathrm{t}_{\mathrm{ICHCH}}$ | Cascaded input hold time to CLK2X rising edge. ${ }^{\text {[43] }}$ | 2.0 |  | 2.0 |  | ns |
| $\mathrm{t}_{\text {CKHOV }}$ | Rising edge of CLK2X to LHO, FULO, BHO, FULL valid. ${ }^{[44]}$ |  | 8.5 |  | 7.0 | ns |
| $\mathrm{t}_{\text {CKHDV }}$ | Rising edge of CLK2X to DQ valid. ${ }^{[44]}$ |  | 9.0 |  | 7.5 | ns |
| $\mathrm{t}_{\text {CKHDZ }}$ | Rising edge of CLK2X to DQ HIGH-Z. ${ }^{45]}$ | 0.5 | 8.5 | 0.5 | 7.0 | ns |
| $\mathrm{t}_{\text {CKHSV }}$ | Rising edge of CLK2X to SRAM bus valid. ${ }^{[44]}$ |  | 9.0 |  | 7.5 | ns |
| $\mathrm{t}_{\text {CKHSHZ }}$ | Rising edge of CLK2X to SRAM bus HIGH-Z. ${ }^{45]}$ | 0.5 | 6.5 | 0.5 | 6.0 | ns |
| $\mathrm{t}_{\text {CKHSLZ }}$ | Rising edge of CLK2X to SRAM bus LOW-Z. ${ }^{[45]}$ | 7.0 |  | 6.5 |  | ns |

Table 17-2. AC Timing Parameters with CLK1X

| Parameter | Description | CYNSE70256-066 |  | CYNSE70256-083 |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  | Min | Max | Min | Max |  |
| $\mathrm{f}_{\text {CLOCK }}$ | CLK1X frequency. | 20 | 66 | 20 | 83 | MHz |
| $\mathrm{t}_{\text {CLOK }}$ | PLL lock time. |  | 0.5 |  | 0.5 | ms |
| $\mathrm{t}_{\mathrm{CKHI}}$ | CLK1X HIGH pulse; worst-case duty cycle. ${ }^{46]}$ | 6.75 |  | 5.4 |  | ns |
| $\mathrm{t}_{\text {CKLO }}$ | CLK1X LOW pulse; worst-case duty cycle. ${ }^{46]}$ | 6.75 |  | 5.4 |  | ns |
| $\mathrm{t}_{\text {ISCH }}$ | Input setup time to CLK1X edge. ${ }^{[46]}$ | 2.5 |  | 1.8 |  | ns |
| $\mathrm{t}_{\mathrm{IHCH}}$ | Input hold time to CLK1X edge. ${ }^{[46]}$ | 0.6 |  | 0.6 |  | ns |
| $\mathrm{t}_{\mathrm{ICSCH}}$ | Cascaded input setup time to CLK1X rising edge. ${ }^{[46]}$ | 4.2 |  | 3.5 |  | ns |
| $\mathrm{t}_{\text {ICHCH }}$ | Cascaded input hold time to CLK1X rising edge. ${ }^{[46]}$ | 2.0 |  | 2.0 |  | ns |
| $\mathrm{t}_{\text {CKHOV }}$ | Rising edge of CLK1X to LHO, FULO, BHO, FULL valid. ${ }^{[47]}$ |  | 8.5 |  | 7.0 | ns |
| $\mathrm{t}_{\text {CKHDV }}$ | Rising edge of CLK1X to DQ valid. ${ }^{[47]}$ |  | 9.0 |  | 7.5 | ns |
| $\mathrm{t}_{\text {CKHDZ }}$ | Rising edge of CLK1X to DQ HIGH-Z. ${ }^{48]}$ | 0.5 | 8.5 | 0.5 | 7.0 | ns |
| $\mathrm{t}_{\text {CKHSV }}$ | Rising edge of CLK1X to SRAM bus valid. ${ }^{[47]}$ |  | 9.0 |  | 7.5 | ns |
| $\mathrm{t}_{\text {CKHSHZ }}$ | Rising edge of CLK1X to SRAM bus HIGH-Z. ${ }^{48]}$ | 0.5 | 6.5 | 0.5 | 6.0 | ns |
| $\mathrm{t}_{\text {CKHSLZ }}$ | Rising edge of CLK1X to SRAM bus LOW-Z. ${ }^{[48]}$ | 7.0 |  | 6.5 |  | ns |

## Notes:

43. Values are based on $50 \%$ signal levels.
44. Based on an AC load of $C_{L}=30 \mathrm{pF}$ (see Figure 17-1, Figure 17-2, and Figure 17-3).
45. These parameters are sampled but not $100 \%$ tested, and are based on an AC load of 5 pF .
46. Values are based on $50 \%$ signal levels and a $50 \% / 50 \%$ duty cycle of CLK1X.
47. Based on an AC load of $C_{L}=30 \mathrm{pF}$ (see Figure 17-1, Figure 17-2, and Figure 17-3).
48. These parameters are sampled but not $100 \%$ tested, and are based on an AC load of 5 pF .

Table 17-3. AC Table for Test Condition of CYNSE70256

| Conditions | Results |
| :--- | :--- |
| Input pulse levels $\left(\mathrm{V}_{\mathrm{DDQ}}=3.3 \mathrm{~V}\right)$ | GND to 3.3 V |
| Input pulse levels $\left(\mathrm{V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}\right)$ | GND to 2.5V |
| Input rise and fall times measured at 0.3 V and 2.7V $\left(\mathrm{V}_{\mathrm{DDQ}}=3.3 \mathrm{~V}\right)$ | $\leq 2 \mathrm{~ns}$ (see Figure 17-1 $)$ |
| Input rise and fall times measured at 0.25 V and $2.25 \mathrm{~V}\left(\mathrm{~V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}\right)$ | $\leq 2 \mathrm{~ns}$ (see Figure 17-1) |
| Input timing reference levels $\left(\mathrm{V}_{\mathrm{DDQ}}=3.3 \mathrm{~V}\right)$ | 1.65 V |
| Input timing reference levels $\left(\mathrm{V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}\right)$ | 1.25 V |
| Output reference levels $\left(\mathrm{V}_{\mathrm{DDQ}}=3.3 \mathrm{~V}\right)$ | 1.65 V |
| Output reference levels $\left(\mathrm{V}_{\mathrm{DDQ}}=2.5 \mathrm{~V}\right)$ | 1.25 V |
| Output load | See Figure 17-2 and Figure 17-3 |



Figure 17-1. Input Wave Form for CYNSE70256


Figure 17-2. Output Load for CYNSE70256


Figure 17-3. 2.5 I/O Output Load Equivalent for CYNSE70256
Notes:
49. Output loading is specified with $\mathrm{CL}=5 \mathrm{pF}$ as in Figure 17-3. Transition is measured at $\pm 200 \mathrm{mV}$ from steady state voltage.
50. The load used for $\mathrm{V}_{\mathrm{OH}}, \mathrm{V}_{\mathrm{OL}}$ testing is shown in Figure 17-3.


Figure 17-4. AC Timing Wave Forms with CLK2X


Figure 17-5. AC Timing Wave Forms with CLK1X

### 18.0 Pinout Descriptions and Package Diagrams



Figure 18-1. Pinout Diagram
Table 18-1. Pinout Descriptions for Pinout Diagram

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A1 | $\mathrm{V}_{\text {DDQ }}{ }^{[51]}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AA26 | CMD[2] | Input |
| A10 | DQ[43] | I/O | AA3 | $V_{D D}$ | 1.5 V |
| A11 | DQ[41] | I/O | AA4 | $\mathrm{V}_{\text {SS }}$ | Ground |
| A12 | DQ[37] | I/O | AB1 | FULL | Output |
| A13 | DQ[35] | I/O | AB2 | ACK | Output-T |
| A14 | DQ[31] | I/O | AB23 | VSS | Ground |
| A15 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AB24 | $V_{\text {DD }}$ | 1.5 V |
| A16 | DQ[25] | I/O | AB25 | CMD[5] | Input |
| A17 | DQ[21] | I/O | AB26 | CMD[4] | Input |
| A18 | DQ[17] | I/O | AB3 | $V_{\text {DD }}$ | 1.5 V |
| A19 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AB4 | $V_{S S}$ | Ground |
| A2 | DQ[71] | I/O | AC1 | $V_{S S}$ | Ground |
| A20 | DQ[09] | I/O | AC10 | $V_{S S}$ | Ground |
| A21 | DQ[05] | I/O | AC11 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| A22 | DQ[03] | I/O | AC12 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| A23 | $\mathrm{V}_{\text {SS }}$ | Ground | AC13 | $V_{\text {DD }}$ | 1.5 V |
| A24 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AC14 | $V_{D D}$ | 1.5 V |
| A25 | HIGH_SPEED | Ground | AC15 | $V_{\text {DD }}$ | 1.5 V |
| A26 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AC16 | $V_{\text {DD }}$ | 1.5 V |

## Note:

51. All $\mathrm{V}_{\mathrm{DDQ}}$ pins should be set to either 2.5 V or 3.3 V .

Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A3 | $V_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AC17 | $\mathrm{V}_{\text {SS }}$ | Ground |
| A4 | DQ[67] | I/O | AC18 | $\mathrm{V}_{\mathrm{SS}}$ | Ground |
| A5 | DQ[63] | I/O | AC19 | $\mathrm{V}_{\text {SS }}$ | Ground |
| A6 | $\mathrm{V}_{\text {DDQ }}$ | 2.5 V 3.3 V | AC2 | EOT | Output-T |
| A7 | DQ[57] | I/O | AC20 | $\mathrm{V}_{\text {SS }}$ | Ground |
| A8 | DQ[53] | I/O | AC21 | $\mathrm{V}_{\text {SS }}$ | Ground |
| A9 | DQ[51] | I/O | AC22 | $\mathrm{V}_{\text {SS }}$ | Ground |
| AA1 | FULO[1] | Output | AC23 | $\mathrm{V}_{\mathrm{SS}}$ | Ground |
| AA2 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AC24 | $V_{\text {DD }}$ | 1.5V |
| AA23 | VSS | Ground | AC25 | CMD[6] | Input |
| AA24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AC26 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AA25 | CMD[3] | Input | AC3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| AC4 | $V_{\text {SS }}$ | Ground | AE10 | DQ[44] | I/O |
| AC5 | $\mathrm{V}_{\text {SS }}$ | Ground | AE11 | DQ[42] | I/O |
| AC6 | $V_{S S}$ | Ground | AE12 | DQ[38] | I/O |
| AC7 | $\mathrm{V}_{\text {SS }}$ | Ground | AE13 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AC8 | $V_{S S}$ | Ground | AE14 | DQ[32] | I/O |
| AC9 | $V_{S S}$ | Ground | AE15 | DQ[28] | I/O |
| AD1 | RST_L | Input | AE16 | DQ[26] | I/O |
| AD10 | DQ[46] | 1/0 | AE17 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AD11 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AE18 | DQ[18] | I/O |
| AD12 | $V_{\text {DD }}$ | 1.5 V | AE19 | DQ[12] | I/O |
| AD13 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AE2 | $\mathrm{V}_{\mathrm{SS}}$ | Ground |
| AD14 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AE20 | DQ[10] | I/O |
| AD15 | $V_{\text {DD }}$ | 1.5 V | AE21 | DQ[06] | I/O |
| AD16 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AE22 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AD17 | DQ[20] | I/O | AE23 | DQ[00] | I/O |
| AD18 | DQ[16] | I/O | AE24 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AD19 | NC | No Connect | AE25 | $V_{\text {SS }}$ | Ground |
| AD2 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | AE26 | $\mathrm{V}_{\text {SS }}$ | Ground |
| AD20 | $V_{\text {DD }}$ | 1.5 V | AE3 | DQ[70] | I/O |
| AD21 | $V_{D D}$ | 1.5 V | AE4 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AD22 | $V_{D D}$ | 1.5 V | AE5 | DQ[64] | I/O |
| AD23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AE6 | DQ[60] | I/O |
| AD24 | $V_{D D}$ | 1.5 V | AE7 | DQ[58] | I/O |
| AD25 | CMD[8] | Input | AE8 | DQ[54] | I/O |
| AD26 | CMD[7] | Input | AE9 | DQ[50] | I/O |
| AD3 | $V_{D D}$ | 1.5 V | AF1 | NC | No Connect |
| AD4 | $V_{D D}$ | 1.5 V | AF10 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AD5 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | AF11 | DQ[40] | I/O |
| AD6 | $V_{\text {DD }}$ | 1.5 V | AF12 | DQ[36] | I/O |
| AD7 | $V_{D D}$ | 1.5 V | AF13 | DQ[34] | I/O |
| AD8 | NC | No Connect | AF14 | DQ[30] | I/O |

Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AD9 | $\mathrm{V}_{\text {DDQ }}$ | 2.5V/3.3V | AF15 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AE1 | $\mathrm{V}_{\text {SS }}$ | Ground | AF16 | DQ[24] | I/O |
| AF17 | DQ[22] | I/O | B23 | $\mathrm{V}_{\text {SS }}$ | Ground |
| AF18 | DQ[14] | I/O | B24 | CFG_L | Input |
| AF19 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | B25 | $\mathrm{V}_{\text {SS }}$ | Ground |
| AF2 | $\mathrm{V}_{\text {SS }}$ | Ground | B26 | SADR[0] | Output-T |
| AF20 | DQ[08] | I/O | B3 | DQ[69] | I/O |
| AF21 | DQ[04] | I/O | B4 | DQ[65] | I/O |
| AF22 | DQ[02] | I/O | B5 | DQ[61] | I/O |
| AF23 | SSV | Output-T | B6 | DQ[59] | I/O |
| AF24 | SSF | Output-T | B7 | DQ[55] | I/O |
| AF25 | CMD[10] | Input | B8 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AF26 | CMD[9] | Input | B9 | DQ[47] | I/O |
| AF3 | DQ[68] | I/O | C1 | TCK | Input |
| AF4 | DQ[66] | I/O | C10 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| AF5 | DQ[62] | I/O | C11 | $V_{\text {DD }}$ | 1.5 V |
| AF6 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | C12 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| AF7 | DQ[56] | I/O | C13 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| AF8 | DQ[52] | I/O | C14 | $V_{\text {DD }}$ | 1.5 V |
| AF9 | DQ[48] | I/O | C15 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B1 | TDI | Input | C16 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B10 | DQ[45] | I/O | C17 | DQ[19] | I/O |
| B11 | DQ[39] | 1/O | C18 | DQ[13] | I/O |
| B12 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | C19 | NC | No Connect |
| B13 | DQ[33] | I/O | C2 | TMS | Input |
| B14 | DQ[29] | 1/O | C20 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B15 | DQ[27] | I/O | C21 | $V_{\text {DD }}$ | 1.5 V |
| B16 | DQ[23] | I/O | C22 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B17 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | C23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B18 | DQ[15] | I/O | C24 | $V_{\text {DD }}$ | 1.5 V |
| B19 | DQ[11] | 1/0 | C25 | SADR[1] | Output-T |
| B2 | $\mathrm{V}_{\text {SS }}$ | Ground | C26 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| B20 | DQ[07] | I/O | C3 | $V_{\text {DD }}$ | 1.5 V |
| B21 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | C4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| B22 | DQ[01] | I/O | C5 | $V_{\text {DD }}$ | 1.5 V |
| C6 | $V_{D D}$ | 1.5 V | E24 | $V_{\text {DD }}$ | 1.5 V |
| C7 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | E25 | SADR[5] | Output-T |
| C8 | NC | No Connect | E26 | SADR[4] | Output-T |
| C9 | DQ[49] | I/O | E3 | $V_{\text {DD }}$ | 1.5 V |
| D1 | TRST_L | Input | E4 | $\mathrm{V}_{\text {SS }}$ | Ground |
| D10 | VSS | Ground | F1 | ID[1] | Input |
| D11 | $V_{\text {DD }}$ | 1.5 V | F2 | ID[2] | Input |
| D12 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | F23 | $\mathrm{V}_{\text {SS }}$ | Ground |

Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| D13 | $V_{\text {DD }}$ | 1.5 V | F24 | $V_{\text {DD }}$ | 1.5 V |
| D14 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | F25 | SADR[6] | Output-T |
| D15 | $V_{\text {DD }}$ | 1.5 V | F26 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| D16 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | F3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| D17 | $V_{S S}$ | Ground | F4 | $\mathrm{V}_{S S}$ | Ground |
| D18 | $\mathrm{V}_{\mathrm{SS}}$ | Ground | G1 | ID[3] | Input |
| D19 | $\mathrm{V}_{\mathrm{SS}}$ | Ground | G2 | ID[4] | Input |
| D2 | TDO | Output-T | G23 | $\mathrm{V}_{S S}$ | Ground |
| D20 | $\mathrm{V}_{\mathrm{SS}}$ | Ground | G24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| D21 | $V_{S S}$ | Ground | G25 | SADR[8] | Output-T |
| D22 | $\mathrm{V}_{S S}$ | Ground | G26 | SADR[7] | Output-T |
| D23 | $\mathrm{V}_{S S}$ | Ground | G3 | $V_{\text {DD }}$ | 1.5 V |
| D24 | $V_{D D}$ | 1.5V | G4 | $V_{S S}$ | Ground |
| D25 | SADR[3] | Output-T | H1 | LHI[0] | No Connect |
| D26 | SADR[2] | Output-T | H2 | LHI[1] | Input |
| D3 | $V_{D D}$ | 1.5 V | H23 | $\mathrm{V}_{S S}$ | Ground |
| D4 | $\mathrm{V}_{\text {SS }}$ | Ground | H24 | NC | No Connect |
| D5 | $V_{S S}$ | Ground | H25 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| D6 | $V_{S S}$ | Ground | H26 | SADR[9] | Output-T |
| D7 | $\mathrm{V}_{\text {SS }}$ | Ground | H3 | NC | No Connect |
| D8 | $\mathrm{V}_{S S}$ | Ground | H4 | $\mathrm{V}_{S S}$ | Ground |
| D9 | $\mathrm{V}_{\mathrm{SS}}$ | Ground | J1 | LHI[2] | Input |
| E1 | ID[0] | No Connect | J2 | LHI[3] | Input |
| E2 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | J23 | $\mathrm{V}_{S S}$ | Ground |
| E23 | $V_{\text {SS }}$ | Ground | J24 | SADR[11] | Output-T |
| J25 | SADR[12] | Output-T | M2 | BHI[0] | Input |
| J26 | SADR[10] | Output-T | M23 | $V_{\text {DD }}$ | 1.5 V |
| J3 | $\mathrm{V}_{\text {DDQ }}$ | 2.5V/3.3V | M24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| J4 | $\mathrm{V}_{\text {SS }}$ | Ground | M25 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| K1 | LHI[6] | Input | M26 | SADR[17] | Output-T |
| K2 | LHI[4] | Input | M3 | $V_{\text {DD }}$ | 1.5 V |
| K23 | $\mathrm{V}_{\text {SS }}$ | Ground | M4 | $V_{\text {DD }}$ | 1.5 V |
| K24 | SADR[13] | Output-T | N1 | BHI[1] | Input |
| K25 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | N11 | $\mathrm{V}_{\text {SS }}$ | Ground |
| K26 | SADR[14] | Output-T | N12 | $V_{S S}$ | Ground |
| K3 | LHI[5] | Input | N13 | $V_{S S}$ | Ground |
| K4 | $\mathrm{V}_{\text {SS }}$ | Ground | N14 | $\mathrm{V}_{\text {SS }}$ | Ground |
| L1 | LHO[0] | Output | N15 | $V_{S S}$ | Ground |
| L11 | $V_{\text {SS }}$ | Ground | N16 | $V_{S S}$ | Ground |
| L12 | $\mathrm{V}_{\text {SS }}$ | Ground | N2 | BHI[2] | Input |
| L13 | $V_{S S}$ | Ground | N23 | $V_{\text {DD }}$ | 1.5 V |
| L14 | $V_{S S}$ | Ground | N24 | $V_{D D}$ | 1.5 V |
| L15 | $\mathrm{V}_{\text {SS }}$ | Ground | N25 | SADR[19] | Output-T |

Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

| Package Ball Number | Signal Name | Signal Type | Package Ball Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| L16 | $\mathrm{V}_{\text {SS }}$ | Ground | N26 | SADR[18] | Output-T |
| L2 | LHO[1] | Output | N3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| L23 | $V_{\text {DD }}$ | 1.5 V | N4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| L24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | P1 | BHO[0] | Output |
| L25 | SADR[15] | Output-T | P11 | $\mathrm{V}_{\text {SS }}$ | Ground |
| L26 | SADR[16] | Output-T | P12 | $\mathrm{V}_{\mathrm{SS}}$ | Ground |
| L3 | $V_{\text {DD }}$ | 1.5 V | P13 | $\mathrm{V}_{\text {SS }}$ | Ground |
| L4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | P14 | $\mathrm{V}_{S S}$ | Ground |
| M1 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | P15 | $\mathrm{V}_{\mathrm{SS}}$ | Ground |
| M11 | $V_{\text {SS }}$ | Ground | P16 | $\mathrm{V}_{\text {SS }}$ | Ground |
| M12 | $V_{S S}$ | Ground | P2 | NC | No Connect |
| M13 | $V_{S S}$ | Ground | P23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| M14 | $V_{S S}$ | Ground | P24 | $V_{\text {DD }}$ | 1.5 V |
| M15 | $\mathrm{V}_{S S}$ | Ground | P25 | SADR[21] | Output-T |
| M16 | $\mathrm{V}_{S S}$ | Ground | P26 | SADR[20] | Output-T |
| P3 | $V_{\text {DD }}$ | 1.5 V | U24 | OE_L | Output-T |
| P4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | U25 | PHS_L | Input |
| R1 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | U26 | CLK1X/CLK2X | Input |
| R11 | $V_{\text {SS }}$ | Ground | U3 | FULI[1] | Input |
| R12 | $\mathrm{V}_{\text {SS }}$ | Ground | U4 | $\mathrm{V}_{\text {SS }}$ | Ground |
| R13 | $\mathrm{V}_{S S}$ | Ground | V1 | FULI[2] | Input |
| R14 | $V_{S S}$ | Ground | V2 | FULI[3] | Input |
| R15 | $V_{S S}$ | Ground | V23 | $\mathrm{V}_{\text {SS }}$ | Ground |
| R16 | $\mathrm{V}_{\mathrm{SS}}$ | Ground | V24 | CE_L | Output-T |
| R2 | BHO[1] | Output | V25 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| R23 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | V26 | WE_L | Output-T |
| R24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | V3 | FULI[4] | Input |
| R25 | SADR[22] | Output-T | V4 | $\mathrm{V}_{\text {SS }}$ | Ground |
| R26 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | W1 | $\mathrm{V}_{\text {DDQ }}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |
| R3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | W2 | FULI[5] | Input |
| R4 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | W23 | $\mathrm{V}_{\text {SS }}$ | Ground |
| T1 | BHO[2] | Output | W24 | NC | Output-T |
| T11 | $V_{\text {SS }}$ | Ground | W25 | CMDV | Input |
| T12 | $\mathrm{V}_{S S}$ | Ground | W26 | ALE_L | Output-T |
| T13 | $\mathrm{V}_{\text {SS }}$ | Ground | W3 | NC | No Connect |
| T14 | $V_{S S}$ | Ground | W4 | $\mathrm{V}_{\text {SS }}$ | Ground |
| T15 | $V_{S S}$ | Ground | Y1 | FULI[6] | Input |
| T16 | $V_{S S}$ | Ground | Y2 | FULO[0] | Output |
| T2 | $V_{S S}$ | Ground | Y23 | $\mathrm{V}_{\text {SS }}$ | Ground |
| T23 | $V_{\text {DD }}$ | 1.5 V | Y24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |
| T24 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V | Y25 | CMD[1] | Input |
| T25 | CLK_MODE | Input | Y26 | CMD[0] | Input |
| T26 | SADR[23] | Output-T | Y3 | $\mathrm{V}_{\mathrm{DD}}$ | 1.5 V |

Table 18-1. Pinout Descriptions for Pinout Diagram (continued)

| Package Ball <br> Number | Signal Name | Signal Type | Package Ball <br> Number | Signal Name | Signal Type |
| :---: | :---: | :---: | :---: | :---: | :---: |
| T 3 | $\mathrm{~V}_{\mathrm{DD}}$ | 1.5 V | Y 4 | $\mathrm{~V}_{\mathrm{SS}}$ | Ground |
| T 4 | $\mathrm{~V}_{\mathrm{DD}}$ | 1.5 V |  |  |  |
| U 1 | $\mathrm{FULI}[0]$ | No Connect |  |  |  |
| U 2 | $\mathrm{~V}_{\mathrm{DDQ}}$ | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ |  |  |  |
| U 23 | $\mathrm{~V}_{\mathrm{SS}}$ | Ground |  |  |  |

### 19.0 Ordering Information

## Table 19-1. Ordering Information

| Part Number | Description | I/O Voltage | Frequency | Temperature Range |
| :---: | :---: | :---: | :---: | :---: |
| CYNSE70256-066BHC | NSE | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | 66 MHz | Commercial |
| CYNSE70256-083BHC | NSE | $2.5 \mathrm{~V} / 3.3 \mathrm{~V}$ | 83 MHz | Commercial |

### 20.0 Package Diagram

388-ball HSBGA ( $35 \times 35 \times 2.33 \mathrm{~mm}$ ) BH388


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## Document History Page

| Document Title: CYNSE70256 Network Search Engine Document Number: 38-02035 |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| REV. | ECN NO. | Issue Date | Orig. of Change | Description of Change |
| ** | 110448 | 11/29/01 | AFX | New Data Sheet |
| *A | 112905 | 03/22/02 | ED | Added 3.3 V I/O specs. Added package diagrams. |
| *B | 115995 | 08/27/02 | KHS | Updated AC Timing, DC Characteristics, JTAG, Pinout Diagram and Pinout Description. <br> Removed references to TEST signals from Pinout Diagram, Pinout Description and Signal Description. <br> Added Power section covering power-up sequence and power consumption. Removed all references to $1.8 \mathrm{~V} / / \mathrm{O}$. <br> Removed all references to CLK_TUNE[3:0] and set it to 100\% ("1001"). |
| *C | 119308 | 11/22/02 | KHS | Changed package from BGC to BHC. Added min hold timing to 0.5 ns for $\mathrm{t}_{\mathrm{CKHDZ}}$ and $\mathrm{t}_{\mathrm{CKHSHZ}}$. Added note to power-up sequence. <br> Updated power-up figures. <br> Remove alternative power-up sequence. <br> Added note to JTAG testing. |
| *D | 125508 | 05/08/03 | DCU | Added note for LHI[0] and FULI[0] being unconnected. Added Ground in empty pin description. <br> Clarified Parallel Write description. <br> Changed diagram for power-up sequence. <br> Changed from Preliminary to Final Data Sheet |
| *E | 131896 | 12/12/03 | FSG | Minor Change: Upload MPN to external website. No content change |

