

DSP Microcomputer

ADSP-2171/ADSP-2172/ADSP-2173

FEATURES

30 ns Instruction Cycle Time (33 MIPS) from 16.67 MHz Crystal at 5.0 V

50 ns Instruction Cycle Time (20 MIPS) from 10 MHz Crystal at 3.3 V

ADSP-2100 Family Code & Function Compatible with New Instruction Set Enhancements for Bit Manipulation Instructions, Multiplication Instructions, Biased Rounding, and Global Interrupt Masking

Bus Grant Hang Logic

2K Words of On-Chip Program Memory RAM

2K Words of On-Chip Data Memory RAM

8K Words of On-Chip Program Memory ROM (ADSP-2172)

8- or 16-Bit Parallel Host Interface Port 300 mW Typical Power Dissipation at 5.0 V at 30 ns 70 mW Typical Power Dissipation at 3.3 V at 50 ns Powerdown Mode Featuring Less than 0.55 mW (ADSP-2171/ADSP-2172) or 0.36 mW (ADSP-2173) CMOS

Standby Power Dissipation with 100 Cycle Recovery

from Powerdown

Dual Purpose Program Memory for Both Instruction
and Data Storage

Independent ALU, Multiplier/Accumulator, and Barrel Shifter Computational Units

Two Independent Data Address Generators

Powerful Program Sequencer Provides

Zero Overhead Looping

Conditional Instruction Execution

Two Double-Buffered Serial Ports with Companding Hardware and Automatic Data Buffering

Programmable 16-Bit Interval Timer with Prescaler Programmable Wait State Generation

Automatic Booting of Internal Program Memory from Byte-Wide External Memory, e.g., EPROM, or

Through Host Interface Port Stand-Alone ROM Execution (Optional)

Single-Cycle Instruction Execution

Single-Cycle Context Switch

Multifunction Instructions

Three Edge- or Level-Sensitive External Interrupts

Low Power Dissipation in Standby Mode

128-Lead TQFP and 128-Lead PQFP

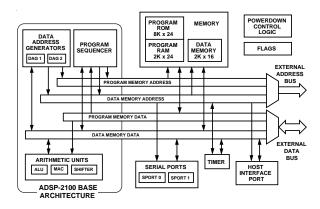
GENERAL DESCRIPTION

The ADSP-2171, ADSP-2172, and ADSP-2173 are single-chip microcomputers optimized for digital signal processing (DSP) and other high-speed numeric processing applications. The ADSP-2171 and ADSP-2172 are designed for 5.0 V applications. The ADSP-2173 is designed for 3.3 V applications. The ADSP-2172 also has 8K words (24-bit) of program ROM.

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FUNCTIONAL BLOCK DIAGRAM



The ADSP-217x combines the ADSP-2100 base architecture (three computational units, data address generators, and a program sequencer) with two serial ports, a host interface port, a programmable timer, extensive interrupt capabilities, and on-chip program and data memory.

In addition, the ADSP-217x supports new instructions, which include bit manipulations—bit set, bit clear, bit toggle, bit test—new ALU constants, new multiplication instruction (x squared), biased rounding, and global interrupt masking, for increased flexibility. The ADSP-217x also has a Bus Grant Hang Logic (\overline{BGH}) feature.

The ADSP-217x provides 2K words (24-bit) of program RAM and 2K words (16-bit) of data memory. The ADSP-2172 provides an additional 8K words (24-bit) of program ROM. Powerdown circuitry is also provided to meet the low power needs of battery operated portable equipment. The ADSP-217x is available in 128-pin TQFP and 128-pin PQFP packages.

Fabricated in a high-speed, double metal, low power, CMOS process, the ADSP-217X operates with a 30 ns instruction cycle time. Every instruction can execute in a single processor cycle.

The ADSP-217x's flexible architecture and comprehensive instruction set allow the processor to perform multiple operations in parallel. In one processor cycle the ADSP-217x can:

- generate the next program address
- fetch the next instruction
- · perform one or two data moves
- update one or two data address pointers
- perform a computational operation

This takes place while the processor continues to:

- receive and transmit data through the two serial ports
- receive and/or transmit data through the host interface port
- decrement timer

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Development System

The ADSP-2100 Family Development Software, a complete set of tools for software and hardware system development, supports the ADSP-217x. The System Builder provides a high-level method for defining the architecture of systems under development. The Assembler has an algebraic syntax that is easy to program and debug. The Linker combines object files into an executable file. The Simulator provides an interactive instruction-level simulation with a reconfigurable user interface to display different portions of the hardware environment. A PROM Splitter generates PROM programmer compatible files. The C Compiler, based on the Free Software Foundation's GNU C Compiler, generates ADSP-217x assembly source code. The Runtime Library includes over 100 ANSI-standard mathematical and DSP-specific functions.

EZ-Tools, low cost, easy-to-use hardware tools, also support the ADSP-217x.

The ADSP-217x EZ-ICE® Emulator aids in the hardware debugging of ADSP-217x systems. The emulator consists of hardware, host computer resident software, the emulator probe, and the pin adaptor. The emulator performs a full range of emulation functions including stand-alone operation or operation in the target, setting up to 20 breakpoints, single-step or full-speed operation in the target, examining and altering registers and memory values, and PC upload/download functions. If you plan to use the emulator, you should consider the emulator's restrictions (differences between emulator and processor operation).

The EZ-LAB® Evaluation Board is a PC plug-in card, but it can operate in stand-alone mode. The evaluation board/system development board executes EPROM-based or downloaded programs. Modular Analog Front End daughter cards with different codecs will be made available.

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Additional Information

This data sheet provides a general overview of ADSP-217x functionality. For additional information on the architecture and instruction set of the processor, refer to the *ADSP-2100 Family User's Manual*. For more information about the Development System and ADSP-217x programmer's reference information, refer to the *ADSP-2100 Family Assembler Tools & Simulator Manual*.

ARCHITECTURE OVERVIEW

Figure 1 is an overall block diagram of the ADSP-217x. The processor contains three independent computational units: the ALU, the multiplier/accumulator (MAC) and the shifter. The computational units process 16-bit data directly and have provisions to support multiprecision computations. The ALU performs a standard set of arithmetic and logic operations; division primitives are also supported. The MAC performs single-cycle multiply, multiply/add and multiply/subtract operations with 40 bits of accumulation. The shifter performs logical and arithmetic shifts, normalization, denormalization, and derive exponent operations. The shifter can be used to efficiently implement numeric format control including multiword and block floating-point representations.

The internal result (R) bus directly connects the computational units so that the output of any unit may be the input of any unit on the next cycle.

A powerful program sequencer and two dedicated data address generators ensure efficient delivery of operands to these computational units. The sequencer supports conditional jumps, subroutine calls and returns in a single cycle. With internal loop counters and loop stacks, the ADSP-217x executes looped code with zero overhead; no explicit jump instructions are required to maintain the loop.

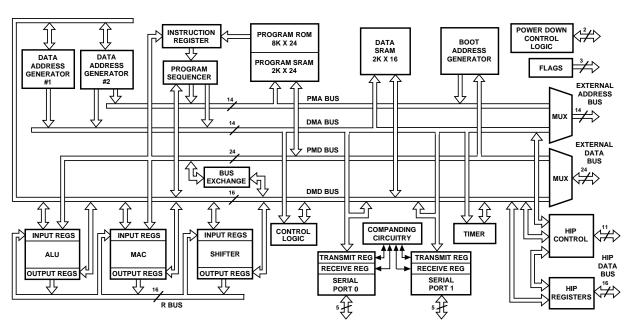


Figure 1. ADSP-217x Block Diagram

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Two data address generators (DAGs) provide addresses for simultaneous dual operand fetches (from data memory and program memory). Each DAG maintains and updates four address pointers. Whenever the pointer is used to access data (indirect addressing), it is post-modified by the value of one of four possible modify registers. A length value may be associated with each pointer to implement automatic modulo addressing for circular buffers.

Efficient data transfer is achieved with the use of five internal buses.

- Program Memory Address (PMA) Bus
- Program Memory Data (PMD) Bus
- Data Memory Address (DMA) Bus
- Data Memory Data (DMD) Bus
- Result (R) Bus

The two address buses (PMA and DMA) share a single external address bus, allowing memory to be expanded off-chip, and the two data buses (PMD and DMD) share a single external data bus.

Program memory can store both instructions and data, permitting the ADSP-217x to fetch two operands in a single cycle, one from program memory and one from data memory. The ADSP-217x can fetch an operand from on-chip program memory and the next instruction in the same cycle.

The memory interface supports slow memories and memory-mapped peripherals with programmable wait state generation. External devices can gain control of external buses with bus request/grant signals (\overline{BR} and \overline{BG}). One execution mode (Go Mode) allows the ADSP-217x to continue running from internal memory. Normal execution mode requires the processor to halt while buses are granted.

In addition to the address and data bus for external memory connection, the ADSP-217x has a configurable 8- or 16-bit Host Interface Port (HIP) for easy connection to a host processor. The HIP is made up of 16 data/address pins and 11 control pins. The HIP is extremely flexible and provides a simple interface to a variety of host processors. For example, the Motorola 68000 series, the Intel 80C51 series and the Analog Devices' ADSP-2101 can be easily connected to the HIP. The host processor can initialize the ASDP-217x's on-chip memory through the HIP.

The ADSP-217x can respond to eleven interrupts. There can be up to three external interrupts, configured as edge or level sensitive, and eight internal interrupts generated by the Timer, the Serial Ports ("SPORTs"), the HIP, the powerdown circuitry, and software. There is also a master RESET signal.

The two serial ports provide a complete synchronous serial interface with optional companding in hardware and a wide variety of framed or frameless data transmit and receive modes of operation. Each port can generate an internal programmable serial clock or accept an external serial clock.

Boot circuitry provides for loading on-chip program memory automatically from byte-wide external memory. After reset, seven wait states are automatically generated. This allows, for example, a 30 ns ADSP-217x to use an external 200 ns EPROM as boot memory. Multiple programs can be selected

and loaded from the EPROM with no additional hardware. The on-chip program memory can also be initialized through the HTP.

The ADSP-217x features three general-purpose flag outputs whose states can be simultaneously changed through software. You can use these outputs to signal an event to an external device. In addition, the data input and output pins on SPORT1 can be alternatively configured as an input flag and an output flag.

A programmable interval timer generates periodic interrupts. A 16-bit count register (TCOUNT) is decremented every *n* processor cycles, where *n-l* is a scaling value stored in an 8-bit register (TSCALE). When the value of the count register reaches zero, an interrupt is generated and the count register is reloaded from a 16-bit period register (TPERIOD).

The ADSP-217x instruction set provides flexible data moves and multifunction (one or two data moves with a computation) instructions. Every instruction can be executed in a single processor cycle. The ADSP-217x assembly language uses an algebraic syntax for ease of coding and readability. A comprehensive set of development tools supports program development.

Serial Ports

The ADSP-217x incorporates two complete synchronous serial ports (SPORT0 and SPORT1) for serial communications and multiprocessor communication.

Here is a brief list of the capabilities of the ADSP-217x SPORTs. Refer to the *ADSP-2100 Family User's Manual* for further details

- SPORTs are bidirectional and have a separate, doublebuffered transmit and receive section.
- SPORTs can use an external serial clock or generate their own serial clock internally.
- SPORTs have independent framing for the receive and transmit sections. Sections run in a frameless mode or with frame synchronization signals internally or externally generated.
 Frame sync signals are active high or inverted, with either of two pulse widths and timings.
- \bullet SPORTs support serial data word lengths from 3 to 16 bits and provide optional A-law and $\mu\text{-law}$ companding according to CCITT recommendation G.711.
- SPORT receive and transmit sections can generate unique interrupts on completing a data word transfer.
- SPORTs can receive and transmit an entire circular buffer of data with only one overhead cycle per data word. An interrupt is generated after a data buffer transfer.
- SPORT0 has a multichannel interface to selectively receive and transmit a 24 or 32 word, time-division multiplexed, serial bitstream.
- SPORT1 can be configured to have two external interrupts (IRQ0 and IRQ1) and the Flag In and Flag Out signals. The internally generated serial clock may still be used in this configuration.

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Pin Description

The ADSP-217x is available in 128-lead TQFP and 128-lead PQFP packages. Table I contains the pin descriptions.

Table I. ADSP-217x Pin List

Pin Group Name	# of Pins	Input/ Output	Function
Address	14	О	Address output for program, data and boot memory spaces
Data	24	I/O	Data I/O pins for program and data memories. Input only for boot memory space, with two MSBs used as boot space addresses.
RESET	1	I	Processor reset input
ĪRQ2	1	I	External interrupt request #2
\overline{BR}	1	I	External bus request input
$\overline{\mathrm{BG}}$	1	O	External bus grant output
BGH	1	O	External bus grant hang output
PMS	1	O	External program memory selec
DMS	1	O	External data memory select
BMS	1	O	Boot memory select
$\overline{\mathrm{RD}}$	1	O	External memory read enable
\overline{WR}	1	O	External memory write enable
MMAP	1	I	Memory map select
CLKIN, XTAL	2	I	External clock or quartz crystal input
CLKOUT	1	O	Processor clock output
\overline{HSEL}	1	I	HIP select input
HACK	1	O	HIP acknowledge output
HSIZE	1		8/16 bit host select input $0 = 16$ -bit; $1 = 8$ -bit
BMODE	1	I	Boot mode select input 0 = EPROM/data bus; 1 = HIP
HMD0	1	I	Bus strobe select input 0 = RD, WR; 1 = RW, DS
HMD1	1	I	HIP address/data mode select input 0 = separate; 1 = multiplexed
HRD/HRW	1	I	HIP read strobe/read/write select input
HWR/HDS	1	I	HIP write strobe/host data strobe select input
HD15-0/			
HAD15-0	16	I/O	HIP data/data and address
HA2/ALE	1	I	Host address 2/Address latch enable input
HA1-0/ Unused	2	I	Host addresses 1 and 0 invest-
SPORT0	ر 5	I/O	Host addresses 1 and 0 inputs
SPURIU	อ	1/U	Serial port 0 I/O pins (TFS0, RFS0, DT0, DR0, SCLK0)

SPORT1	5	I/O	Serial port 1 I/O pins
or			
IRQ1 (TFS1)	1	I	External interrupt request #1
IRQ0 (RFS1)	1	I	External interrupt request #0
SCLK1	1	O	Programmable clock output
FO (DT1)	1	O	Flag Output pin
FI (DR1)	1	I	Flag Input pin
FL2-0	3	O	General purpose flag output
			pins
V_{DD}	6		Power supply pins
GND	11		Ground pins
$\overline{\mathrm{PWD}}$	1	I	Powerdown pin
PWDACK	1	O	Powerdown acknowledge pin

Host Interface Port

The ADSP-217x host interface port is a parallel I/O port that allows for an easy connection to a host processor. Through the HIP, the ADSP-217x can be used as a memory-mapped peripheral to a host computer. The HIP can be thought of as an area of dual-ported memory, or mailbox registers, that allow communication between the computational core of the ADSP-217x and the host computer.

The HIP is completely asynchronous. The host processor can write data into the HIP while the ADSP-217x is operating at full speed.

The HIP can be configured with the following pins:

- HSIZE configures HIP for 8-bit or 16-bit communication with the host processor.
- BMODE (when MMAP = 0) determines whether the ADSP-217x boots from the host processor (through the HIP) or external EPROM (through the data bus).
- HMD0 configures the bus strobes as separate read and write strobes, or a single read/write select and a host data strobe.
- HMD1 selects separate address (3-bit) and data (16-bit) buses, or a multiplexed, 16-bit address/data bus with address latch enable.

Tying these pins to appropriate values configures the ADSP-217x for straight-wire interface to a variety of industry-standard microprocessors and microcomputers.

In 8-bit reads, the ADSP-217x three-states the upper eight bits of the bus. When the host processor writes an 8-bit value to the HIP, the upper eight bits are all zeros. For additional information refer to the *ADSP-2100 Family User's Manual*.

HIP Operation

The HIP contains six data registers (HDR5–0) and two status registers (HSR7–6) with an associated HMASK register for masking interrupts from individual HIP data registers. All HIP data registers are memory-mapped into the internal data memory of the ADSP-217x. HIP transfers can be managed using either interrupts or a polling scheme. These registers are shown in the section "ADSP-217x Registers."

The HIP allows a software reset to be performed by the host processor. The internal software reset signal is asserted for five ADSP-217x processor cycles.

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Interrupts

The interrupt controller allows the processor to respond to the eleven possible interrupts and reset with minimum overhead. The ADSP-217x provides up to three external interrupt input pins, $\overline{IRQ0}$, $\overline{IRQ1}$ and $\overline{IRQ2}$. $\overline{IRQ2}$ is always available as a dedicated pin; SPORT1 may be reconfigured for $\overline{IRQ0}$, $\overline{IRQ1}$, and the flags. The ADSP-217x also supports internal interrupts from the timer, the host interface port, the two serial ports, software, and the powerdown control circuit. The interrupt levels are internally prioritized and individually maskable (except powerdown and reset). The input pins can be programmed to be either level- or edge-sensitive. The priorities and vector addresses of all interrupts are shown in Table II, and the interrupt registers are shown in Figure 2.

Interrupts can be masked or unmasked with the IMASK register. Individual interrupt requests are logically ANDed with the bits in IMASK; the highest priority unmasked interrupt is then selected. The powerdown interrupt is nonmaskable.

The ADSP-217x masks all interrupts for one instruction cycle following the execution of an instruction that modifies the IMASK register. This does not affect autobuffering.

The interrupt control register, ICNTL, allows the external interrupts to be either edge- or level-sensitive. Interrupt routines can either be nested with higher priority interrupts taking precedence or processed sequentially.

The IFC register is a write-only register used to force and clear interrupts generated from software.

Table II. Interrupt Priority & Interrupt Vector Addresses

Source of Interrupt	Interrupt Vector Address (Hex)
Reset (or Power-Up with $PUCR = 1$)	0000 (Highest Priority)
Powerdown (Nonmaskable)	002C
IRQ2	0004
HIP Write	0008
HIP Read	000C
SPORT0 Transmit	0010
SPORT0 Receive	0014
Software Interrupt 1	0018
Software Interrupt 0	001C
SPORT1 Transmit or IRQ1	0020
SPORT1 Receive or IRQ0	0024
Timer	0028 (Lowest Priority)

On-chip stacks preserve the processor status and are automatically maintained during interrupt handling.

The stacks are twelve levels deep to allow interrupt nesting.

The following instructions allow global enable or disable servicing of the interrupts (including powerdown), regardless of the state of IMASK. Disabling the interrupts does not affect autobuffering.

ENA INTS; DIS INTS;

When you reset the processor, the interrupt servicing is enabled.

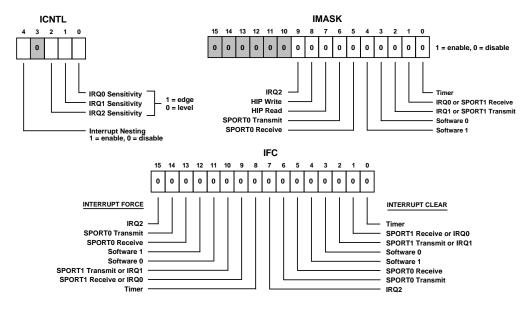


Figure 2. Interrupt Registers

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LOW POWER OPERATION

The ADSP-217x has three low power modes that significantly reduce the power dissipation when the device operates under standby conditions. These modes are:

- Powerdown
- Idle
- Slow Idle

The CLKOUT pin may also be disabled to reduce external power dissipation. The CLKOUT pin is controlled by Bit 14 of SPORTO Autobuffer Control Register, DM[0x3FF3].

Powerdown

The ADSP-217x processor has a low power feature that lets the processor enter a very low power dormant state through hardware or software control. Here is a brief list of powerdown features. Refer to the *ADSP-2100 Family User's Manual*, Chapter 9 "System Interface" for detailed information about the powerdown feature.

- Powerdown mode holds the processor in CMOS standby with a maximum current of less than 100 μA in some modes.
- Quick recovery from powerdown. The processor begins executing instructions in as few as 100 CLKIN cycles.
- Support for an externally generated TTL or CMOS processor clock. The external clock can continue running during powerdown without affecting the lowest power rating and 100 CLKIN cycle recovery.
- Support for crystal operation includes disabling the oscillator to save power (the processor automatically waits 4096 CLKIN cycles for the crystal oscillator to start and stabilize), and letting the oscillator run to allow 100 CLKIN cycle startup.
- Powerdown is initiated by either the powerdown pin (PWD) or the software powerdown force bit.
- Interrupt support allows an unlimited number of instructions to be executed before optionally powering down. The powerdown interrupt also can be used as a non-maskable, edge sensitive interrupt.
- Context clear/save control allows the processor to continue where it left off or start with a clean context when leaving the powerdown state.
- The RESET pin also can be used to terminate powerdown, and the host software reset feature can be used to terminate powerdown under certain conditions.
- Powerdown acknowledge pin indicates when the processor has entered powerdown.

Idle

When the ADSP-217x is in the Idle Mode, the processor waits indefinitely in a low power state until an interrupt occurs. When an unmasked interrupt occurs, it is serviced; execution then continues with the instruction following the *IDLE* instruction.

Slow Idle

The IDLE instruction is enhanced on the ADSP-217x to let the processor's internal clock signal be slowed during IDLE, further reducing power consumption. The reduced clock frequency, a

programmable fraction of the normal clock rate, is specified by a selectable divisor given in the IDLE instruction. The format of the instruction is

IDLE (n);

where n=16, 32, 64, or 128. This instruction keeps the processor fully functional, but operating at the slower clock rate. While it is in this state, the processor's other internal clock signals, such as SCLK, CLKOUT, and timer clock, are reduced by the same ratio. The default form of the instruction, when no clock divisor is given, is the standard IDLE instruction.

When the *IDLE* (n) instruction is used, it effectively slows down the processor's internal clock and thus its response time to incoming interrupts—the 1-cycle response time of the standard idle state is increased by n, the clock divisor. When an enabled interrupt is received, the ADSP-217x will remain in the idle state for up to a maximum of n processor cycles (n = 16, 32, 64, or 128) before resuming normal operation.

When the *IDLE* (*n*) instruction is used in systems that have an externally generated serial clock (SCLK), the serial clock rate may be faster than the processor's reduced internal clock rate. Under these conditions, interrupts must not be generated at a faster rate than can be serviced, due to the additional time the processor takes to come out of the idle state (a maximum of *n* processor cycles).

SYSTEM INTERFACE

Figure 3 shows a basic system configuration with the ADSP-217x, two serial devices, a host processor, a boot EPROM, and optional external program and data memories. Up to 14K words of data memory and 16K words of program memory can be supported. Programmable wait state generation allows the processor to interface easily to slow memories. The ADSP-217x also provides one external interrupt and two serial ports or three external interrupts and one serial port.

Clock Signals

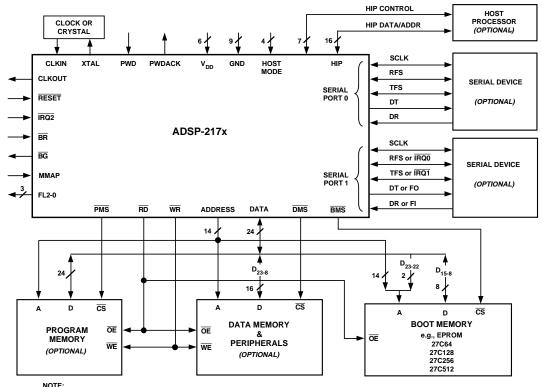
The ADSP-217x can be clocked by either a crystal or by a TTL-compatible clock signal.

The CLKIN input cannot be halted, changed during operation, or operated below the specified frequency during normal operation. The only exception is while the processor is in the Powerdown State. For additional information, refer to Chapter 9, *ADSP-2100 Family User's Manual* for detailed information on this powerdown feature.

If an external clock is used, it should be a TTL-compatible signal running at half the instruction rate. The signal is connected to the processor's CLKIN input. When an external clock is used, the XTAL input *must* be left unconnected.

The ADSP-217x uses an input clock with a frequency equal to half the instruction rate; a 16.67 MHz input clock yields a 30 ns processor cycle (which is equivalent to 33 MHz). Normally, instructions are executed in a single processor cycle. All device timing is relative to the internal instruction clock rate, which is indicated by the CLKOUT signal when enabled.

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THE TWO MSBs OF THE DATA BUS ARE USED AS THE MSBs OF THE BOOT EPROM ADDRESS. THIS IS ONLY REQUIRED FOR THE 27C256 AND 27C512.

Figure 3. ADSP-217x Basic System Configuration

Because the ADSP-217x includes an on-chip oscillator circuit, an external crystal may be used. The crystal should be connected across the CLKIN and XTAL pins, with two capacitors connected as shown in Figure 4. A parallel-resonant, fundamental frequency, microprocessor-grade crystal should be used.

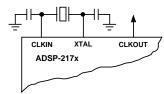


Figure 4. External Crystal Connections

A clock output (CLKOUT) signal is generated by the processor at the processor's cycle rate. This can be enabled and disabled by the CLKODIS bit in the SPORT0 Autobuffer Control Register, DM[0x3FF3].

Reset

The RESET signal initiates a master reset of the ADSP-217x. The RESET signal must be asserted during the power-up sequence to assure proper initialization. RESET during initial power-up must be held long enough to allow the internal clock

to stabilize. If RESET is activated any time after power-up, the clock continues to run and does not require stabilization time.

The power-up sequence is defined as the total time required for the crystal oscillator circuit to stabilize after a valid $V_{\rm DD}$ is applied to the processor, and for the internal phase-locked loop (PLL) to lock onto the specific crystal frequency. A minimum of 2000 CLKIN cycles ensures that the PLL has locked but does not include the crystal oscillator start-up time. During this power-up sequence the $\overline{\rm RESET}$ signal should be held low. On any subsequent resets, the $\overline{\rm RESET}$ signal must meet the minimum pulse width specification, $t_{\rm RSP}$.

The RESET input contains some hysteresis; however, if you use an RC circuit to generate your RESET signal, the use of an external Schmidt trigger is recommended.

The master reset sets all internal stack pointers to the empty stack condition, masks all interrupts and clears the MSTAT register. When \overline{RESET} is released, if there is no pending bus request and the chip is configured for booting (MMAP = 0), the boot-loading sequence is performed. Then the first instruction is fetched from internal program memory location 0x0000.

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Program Memory Interface

The on-chip program memory address bus (PMA) and the on-chip program memory data bus (PMD) are multiplexed with on-chip DMA and DMD buses, creating a single external data bus and a single external address bus. The 14-bit address bus directly addresses up to 16K words. 10K words of memory for ADSP-217x with optional 8K ROM and 2K words of memory for the non-ROM version are on-chip. The data bus is bidirectional and 24 bits wide to external program memory. Program memory may contain code and data.

The program memory data lines are bidirectional. The program memory select (\overline{PMS}) signal indicates access to the program memory and can be used as a chip select signal. The write (\overline{WR}) signal indicates a write operation and is used as a write strobe.

The read (\overline{RD}) signal indicates a read operation and is used as a read strobe or output enable signal.

The ADSP-217x writes data from its 16-bit registers to the 24-bit program memory using the PX register to provide the lower eight bits. When it reads data (not instructions) from 24-bit program memory to a 16-bit data register, the lower eight bits are placed in the PX register.

Program Memory Maps ADSP-217x

Program memory can be mapped in two ways, depending on the state of the MMAP pin. Figure 5 shows the different configurations. When MMAP = 0, internal RAM occupies 2K words beginning at address 0x0000. In this configuration, the boot loading sequence (described in "Boot Memory Interface") is automatically initiated when \overline{RESET} is released.

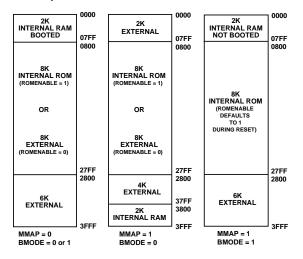


Figure 5. ADSP-217x Memory Maps

When MMAP = 1, words of external program memory begin at address 0x0000 and internal RAM is located in the upper 2K words, beginning at address 0x3800. In this configuration, program memory is not loaded although it can be written to and read from under program control.

The optional ROM always resides at locations PM[0x0800] through PM[0x27FF] regardless of the state of the MMAP pin. The ROM is enabled by setting the ROMENABLE bit in the Data Memory Wait State control register, DM[0x3FFE]. When the ROMENABLE bit is set to 1, addressing program memory in this range will access the on-chip ROM. When set to zero, addressing program memory in this range will access external program memory. The ROMENABLE bit is set to 0 on chip reset unless MMAP and BMODE = 1.

The program memory interface can generate 0 to 7 wait states for external memory devices; default is to 7 wait states after RESET.

Boot Memory Interface

The ADSP-217x can load on-chip memory from external boot memory space. The boot memory space consists of 64K by 8-bit space, divided into eight separate 8K by 8-bit pages. Three bits in the system control register select which page is loaded by the boot memory interface. Another bit in the system control register allows the user to force a boot loading sequence under software control. Boot loading from page 0 after \overline{RESET} is initiated automatically if MMAP=0.

The boot memory interface can generate 0 to 7 wait states; it defaults to 7 wait states after RESET. This allows the ADSP-217x to boot from a single low cost EPROM such as a 27C256. Program memory is booted one byte at a time and converted to 24-bit program memory words.

The \overline{BMS} and \overline{RD} signals are used to select and to strobe the boot memory interface. Only 8-bit data is read over the data bus, on pins D8–D15. To accommodate addressing up to eight pages of boot memory, the two MSBs of the data bus are used in the boot memory interface as the two MSBs of the boot space address.

The ADSP-2100 Family Assembler and Linker support the creation of programs and data structures requiring multiple boot pages during execution.

RD and WR must always be qualified by PMS, DMS, or BMS to ensure the correct program, data, or boot memory accessing.

HIP Booting

The ADSP-217x can also boot programs through its Host Interface Port. If BMODE = 1 and MMAP = 0, the ADSP-217x boots from the HIP. If BMODE = 0, the ADSP-217x boots through the data bus (in the same way as the ADSP-2101), as described above in "Boot Memory Interface." For additional information about HIP booting, refer to the *ADSP-2100 Family User's Manual*, Chapter 7, "Host Interface Port."

The ADSP-2100 Family Development Software includes a utility program called the HIP Splitter. This utility allows the creation of programs that can be booted via the ADSP-217x's HIP, in a similar fashion as EPROM-bootable programs generated by the PROM Splitter utility.

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Stand-Alone ROM Execution

When the MMAP and BMODE pins both are set to 1, the ROM is automatically enabled and execution commences from program memory location 0x0800 at the start of ROM. This feature lets an embedded design operate without external memory components. To operate in this mode, the ROM coded program must copy an interrupt vector table to the appropriate locations in program memory RAM. In this mode, the ROM enable bit defaults to 1 during reset.

Table III. Boot Summary Table

	BMODE = 0	BMODE = 1
MMAP = 0	Boot from EPROM, then execution starts at internal RAM location 0x0000	Boot from HIP, then execution starts at internal RAM location 0x0000
MMAP = 1	No booting, execution starts at external memory location 0x0000	Stand-Alone Mode, execution starts at internal ROM location 0x0800

Ordering Procedure for ADSP-2172 Processors

To place an order for a custom ROM-coded ADSP-2172 processor, you must:

 Complete the following forms contained in the ADSP ROM Ordering Package, available from your Analog Devices sales representative:

ADSP-2172 ROM Specification Form ROM Release Agreement ROM NRE Agreement & Minimum Quantity Order (MQO) Acceptance Agreement for Pre-production ROM Products.

- 2. Return the forms to Analog Devices along with two copies of the Memory Image File (.EXE file) of your ROM code. The files must be supplied on two 3.5" or 5.25" floppy disks for IBM PC (DOS 2.01 or higher).
- Place a purchase order with Analog Devices for nonrecurring engineering charges (NRE) associated with ROM product development.

After this information is received, it is entered into Analog Devices' ROM Manager System which assigns a custom ROM model number to the product. This model number will be branded on all prototype and production units manufactured to these specifications.

To minimize the risk of code being altered during this process, Analog Devices verifies that the .EXE files on both floppy disks are identical, and recalculates the checksums for the .EXE file entered into the ROM Manager System. The checksum data, in the form of a ROM memory map, a hard copy of the .EXE file, and a ROM Data Verification Form are returned to you for inspection.

A signed ROM Verification Form and a purchase order for production units are required prior to any product being manufactured. Prototype units may be applied toward the minimum order quantity.

Upon completion of the prototype manufacture, Analog Devices will ship prototype units and a delivery schedule update for production units. An invoice against your purchase order for the NRE charges is issued at this time.

There is a charge for each ROM mask generated and a minimum order quantity. Consult your sales representative for details. A separate order must be placed for parts of a specific package type, temperature range, and speed grade.

Data Memory Interface

The data memory address (DMA) bus is 14 bits wide. The bidirectional external data bus is 24 bits wide, with the upper 16 bits (D8–D23) used for data memory data (DMD) transfers.

The data memory select (\overline{DMS}) signal indicates access to the data memory and can be used as a chip select signal. The write (\overline{WR}) signal indicates a write operation and can be used as a write strobe. The read (\overline{RD}) signal indicates a read operation and can be used as a read strobe or output enable signal.

The ADSP-217x supports memory-mapped I/O, with the peripherals memory mapped into the data or program memory address spaces and accessed by the processor in the same manner.

Data Memory Map

The on-chip data memory RAM resides in the 2K words of data memory beginning at address 0x3000, as shown in Figure 6. In addition, data memory locations from 0x3800 to the end of data memory at 0x3FFF are reserved. Control registers for the system, timer, wait state configuration, host interface port, and serial port operations are located in this region of memory.

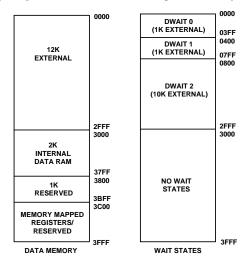


Figure 6. ADSP-217x Data Memory Map

The remaining 12K of data memory is external. External data memory is divided into three zones, each associated with its own wait state generator. By mapping peripherals into different zones, you can accommodate peripherals with different wait state requirements. All zones default to 7 wait states after RESET. For compatibility with other ADSP-2100 Family processors, bit definitions for DWAIT 3 and DWAIT4 are shown in the Data Memory Wait State Control Register, but they are not used by the ADSP-217x.

Bus Request & Bus Grant

The ADSP-217x can relinquish control of the data and address buses to an external device. When the external device requires access to memory, it asserts the bus request (\overline{BR}) signal. If the ADSP-217x is not performing an external memory access, then it responds to the active \overline{BR} input in the following processor cycle by:

- three-stating the data and address buses and the \overline{PMS} , \overline{DMS} , \overline{BMS} , \overline{RD} , \overline{WR} output drivers,
- asserting the bus grant (BG) signal, and
- halting program execution.

If the Go Mode is enabled, the ADSP-217x will not halt program execution until it encounters an instruction that requires an external memory access.

If the ADSP-217x is performing an external memory access when the external device asserts the \overline{BR} signal, then it will not three-state the memory interfaces or assert the \overline{BG} signal until the processor cycle after the access completes, which can be up to eight cycles later depending on the number of wait states. The instruction does not need to be completed when the bus is granted. If a single instruction requires two external memory accesses, the bus will be granted between the two accesses.

When the \overline{BR} signal is released, the processor releases the \overline{BG} signal, reenables the output drivers and continues program execution from the point where it stopped.

The bus request feature operates at all times, including when the processor is booting and when \overline{RESET} is active.

The new Bus Grant Hang logic and associated \overline{BGH} pin allow the ADSP-217x to operate in a multiprocessor environment with a minimal number of "wasted" processor cycles. The bus grant hang pin is asserted when the ADSP-217x desires a cycle,

but cannot execute it because the bus is granted to some other processor. With the \overline{BGH} signal, the other processor(s) in the system can be alerted that the ADSP-217x is hung and release the bus by deasserting bus request. Once the bus is released the ADSP-217x executes the external access and deasserts \overline{BGH} . This is a signal to the other processors that external memory is now available.

ADSP-217X REGISTERS

Figure 7 summarizes all the registers in the ADSP-217x. Some registers store values. For example, AX0 stores an ALU operand; I4 stores a DAG2 pointer. Other registers consist of control bits and fields, or status flags. For example, ASTAT contains status flags from arithmetic operations, and fields in DWAIT control the numbers of wait states for different zones of data memory.

A secondary set of registers in all computational units allows a single-cycle context switch.

The bit and field definitions for control and status registers are given in the rest of this section, except for IMASK, ICNTL and IFC, which are defined earlier in this data sheet. The system control register, DWAIT register, timer registers, HIP control registers, HIP data registers, and SPORT control registers are all mapped into data memory; that is, registers are accessed by reading and writing data memory locations rather than register names. The particular data memory address is shown with each memory-mapped register.

Register bit values shown on the following pages are the default bit values after reset. If no values are shown, the bits are indeterminate at reset. Reserved bits are shown in gray; these bits should always be written with zeros.

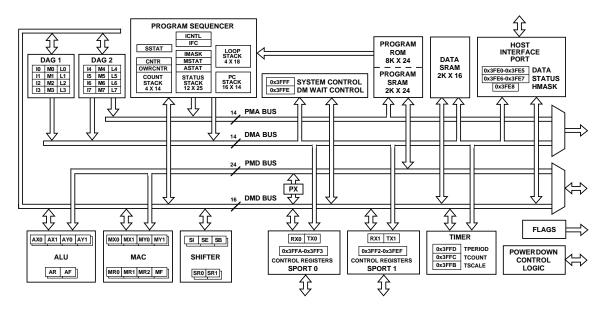
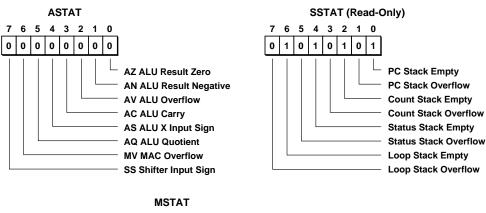
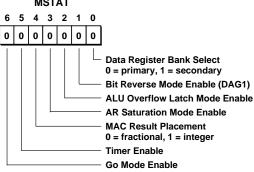
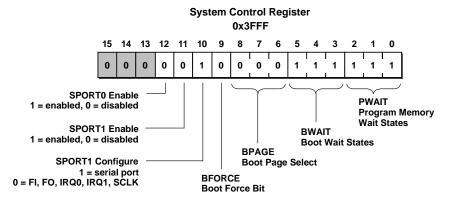


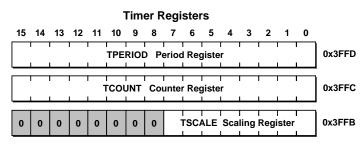
Figure 7. ADSP-217x Registers Control Register

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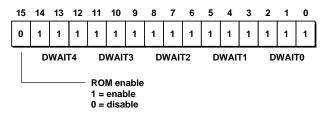




Control Registers

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ROM Enable/Data Memory Wait State Control Register 0x3FFE



SPORTO Multichannel Receive Word Enable Registers

1 = Channel Enabled

0x3FFA

0 = Channel Ignored

SPORT0 Multichannel Transmit Word Enable Registers

1 = Channel Enabled

0 = Channel Ignored

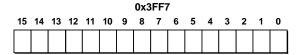
0x3FF8

6	25	24	23	22	21	20	19	18	17	16	
											l

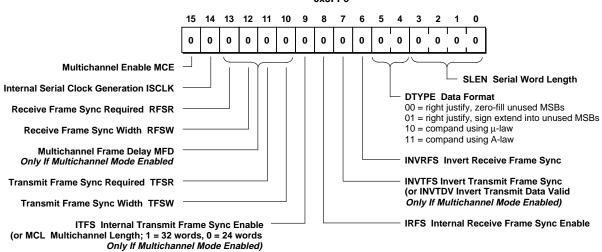
31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16
						<u> </u>	<u> </u>			<u> </u>	<u> </u>				ш

31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16

15 14 13 12 11 10 9 8 7 6 5 4 3	2 1	0



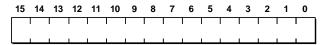
SPORT0 Control Register 0x3FF6



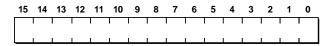
Control Registers

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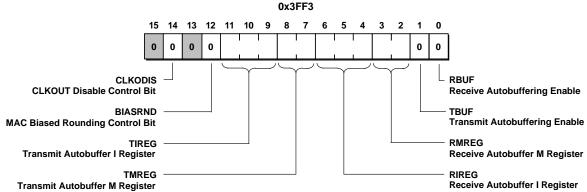
SPORT0 SCLKDIV Serial Clock Divide Modulus 0x3FF5



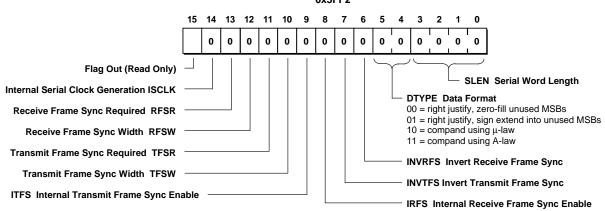
SPORT0 RFSDIV Receive Frame Sync Divide Modulus 0x3FF4



SPORT0 Autobuffer Control Register



SPORT1 Control Register 0x3FF2

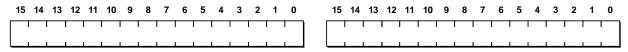


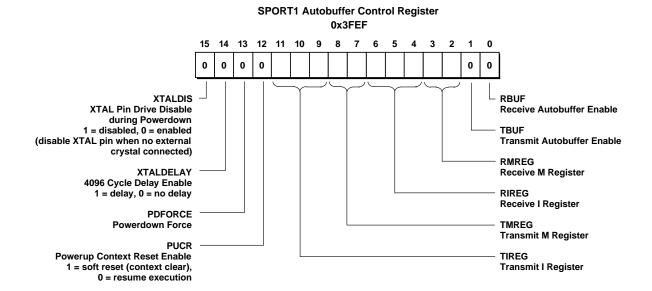
Control Registers

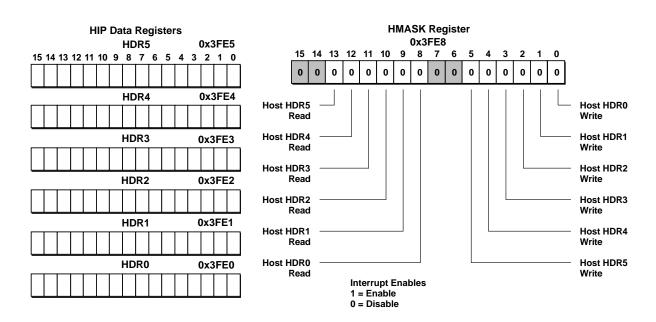
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SPORT1 SCLKDIV Serial Clock Divide Modulus 0x3FF1

SPORT1 RFSDIV Receive Frame Sync Divide Modulus 0x3FF0

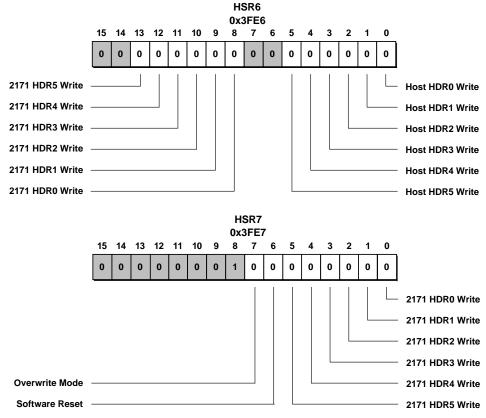






Control Registers

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Control Registers

Biased Rounding

A new mode allows biased rounding in addition to the normal unbiased rounding. When the BIASRND bit is set to 0, the normal unbiased rounding operations occur. When the BIASRND bit is set to 1, biased rounding occurs instead of the normal unbiased rounding. When operating in biased rounding mode all rounding operations with MR0 set to 0x8000 will round up, rather than only rounding odd MR1 values up. For example:

MR value before RND	biased RND result	unbiased RND result
00-0000-8000	00-0001-8000	00-0000-8000
00-0001-8000	00-0002-8000	00-0002-8000
00-0000-8001	00-0001-8001	00-0001-8001
00-0001-8001	00-0002-8001	00-0002-8001
00-0000-7FFF	00-0000-7FFF	00-0000-7FFF
00-0001-7FFF	00-0001-7FFF	00-0001-7FFF

This mode only has an effect when the MR0 register contains 0x8000, all other rounding operation work normally. This mode was added to allow more efficient implementation of bit specified algorithms which specify biased rounding such as the GSM speech compression routines. Unbiased rounding is preferred for most algorithms.

Note: BIASRND bit is Bit 12 of the SPORT0 Autobuffer Control register.

INSTRUCTION SET DESCRIPTION

The ADSP-217x assembly language instruction set has an algebraic syntax that was designed for ease of coding and readability. The assembly language, which takes full advantage of the processor's unique architecture, offers the following benefits:

- ullet The algebraic syntax eliminates the need to remember cryptic assembler mnemonics. For example, a typical arithmetic add instruction, such as AR = AX0 + AY0, resembles a simple equation.
- Every instruction assembles into a single, 24-bit word that can
 execute in a single instruction cycle.
- The syntax is a superset ADSP-2100 Family assembly language and is completely source and object code compatible with other family members. Programs may need to be relocated to utilize internal memory and conform to the ADSP-217x's interrupt vector and reset vector map.
- Sixteen condition codes are available. For conditional jump, call, return, or arithmetic instructions, the condition can be checked and the operation executed in the same instruction cycle.
- Multifunction instructions allow parallel execution of an arithmetic instruction with up to two fetches or one write to processor memory space during a single instruction cycle.

Consult the *ADSP-2100 Family User's Manual* for a complete description of the syntax and an instruction set reference.

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Example Code

The following example is a code fragment that performs the filter tap update for an adaptive (least-mean-squared algorithm) filter. Notice that the computations in the instructions are written like algebraic equations.

```
\label{eq:mass} \begin{split} MF=&MX0*MY1~(RND),~MX0=DM~(I2,M1);~/*~MF=error*beta~*/\\ MR=&MX0*MF~(RND),~AY0=PM~(I6,MS); \end{split}
```

DO adapt UNTIL CE;

AR=MR1 + AY0, MX0=DM (I2,M1), AY0=PM (I6,M7);

adapt: PM(I6,M6) =AR, MR=MX0*MF (RND);

MODIFY (I2, M3); /* Point to oldest data */
MODIFY (I6, M7); /* Point to start of data */

Interrupt Enable

The ADSP-217x supports an interrupt enable instruction. Interrupts are enabled by default at reset. The instruction source code is specified as follows:

Syntax: ENA INTS;

Description: Executing the ENA INTS instruction allows all

unmasked interrupts to be serviced again.

Interrupt Disable

The ADSP-217x supports an interrupt disable instruction. The instruction source code is specified as follows:

Syntax: DIS INTS;

Description: Reset enables interrupt servicing. Executing the

DIS INTS instruction causes all interrupts to be masked without changing the contents of the IMASK register. Disabling interrupts does not affect the autobuffer circuitry, which will operate normally whether or not interrupts are enabled. The disable interrupt instruction masks all user interrupts including the powerdown interrupt.

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ADSP-2171/ADSP-2172—SPECIFICATIONS

RECOMMENDED OPERATING CONDITIONS

		K G	rade	B G	rade	
Parameter		Min	Max	Min	Max	Unit
$V_{ m DD} \ T_{ m AMB}$	Supply Voltage Ambient Operating Temperature	4.5 0	5.5 +70	4.5 -40	5.5 +85	V °C

ELECTRICAL CHARACTERISTICS

			K/B Gr	ades	
Parameter		Test Conditions	Min	Max	Unit
$\overline{V_{IH}}$	Hi-Level Input Voltage ^{1, 2}	@ V _{DD} = max	2.0		V
V_{IH}	Hi-Level CLKIN Voltage	$@V_{DD} = max$	2.2		V
V_{IH}	Hi-Level RESET Voltage	$@V_{DD} = max$	2.2		V
V_{IL}	Lo-Level Input Voltage ^{I, 3}	$@V_{DD} = min$		0.8	V
V_{OH}	Hi-Level Output Voltage ^{1, 4, 5}	$@V_{DD} = min$			
0.1	1 0	$I_{OH} = -0.5 \text{ mA}$	2.4		V
		$@V_{DD} = min$			
		$I_{OH} = -100 \ \mu A^6$	$V_{ m DD} - 0.3$		V
V_{OL}	Lo-Level Output Voltage ^{1, 4, 5}	$@V_{DD} = min$			
	. 0	$I_{OL} = 2 \text{ mA}$		0.4	V
I_{IH}	Hi-Level Input Current ³	@ V _{DD} = max			
	•	$V_{IN} = V_{DD} \max$		10	μΑ
I_{IL}	Lo-Level Input Current ³	$@V_{DD} = max$			
	_	$V_{IN} = 0 V$		10	μΑ
I_{OZH}	Tristate Leakage Current ⁷	$@V_{DD} = max,$			
	_	$V_{\rm IN} = V_{\rm DD} \text{max}^8$		10	μΑ
I_{OZL}	Tristate Leakage Current ⁷	$@V_{DD} = max,$			
		$V_{IN} = 0 V^8$		10	μΑ
I_{DD}	Supply Current (Idle) ^{9, 10}	$@V_{DD} = max$		18	mA
I_{DD}	Supply Current (Dynamic) ¹⁰	$@V_{DD} = max$			
		$t_{\rm CK} = 30 \; {\rm ns}^{11}$		75	mA
I_{DD}	Supply Current (Powerdown) ¹⁰	Lowest Power Mode ¹²		100	μΑ
C_{I}	Input Pin Capacitance ^{3, 6, 13}	$@V_{IN} = 2.5 \text{ V},$			
		$f_{IN} = 1.0 \text{ MHz},$			
	0.7	$T_{AMB} = 25^{\circ}C$		8	pF
C_{O}	Output Pin Capacitance ^{6, 7, 13, 14}	$@V_{IN} = 2.5 \text{ V},$			
		$f_{IN} = 1.0 \text{ MHz},$			
		$T_{AMB} = 25^{\circ}C$		8	pF

Specifications subject to change without notice.

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Bidirectional pins: D0-D23, RFS0, RFS1, SCLK0, SCLK1, TFS0, TFS1, HD0-HD15/HAD0-HAD15.

Input only pins: RESET, IRQ2, BR, MMAP, DR0, DR1, HSEL, HSIZE, BMODE, HMD0, HMD1, HRD/HWR, HWR/HDS, PWD, HA2/ALE, HA1-0.

Input only pins: CLKIN, RESET, IRQ2, BR, MMAP, DR0, DR1, HSEL, HSIZE, BMODE, HMD0, HMD1, HRD/HWR, HWR/HDS, PWD, HA2/ALE, HA1-0.

⁴Output pins: BG, PMS, DMS, BMS, RD, WR, PWDACK, A0-A13, DTO, DT1, CLKOUT, HACK, FL2-0, BGH.
⁵Although specified for TTL outputs, all ADSP-2171/ADSP-2172 outputs are CMOS-compatible and will drive to V_{DD} and GND, assuming no dc loads. ⁶Guaranteed but not tested.

⁷Three-statable pins: A0-A13, D0-D23, PMS, DMS, BMS, RD, WR, DT0, DT1, SCLK0, SCLK1, TFS0, TFS1, RFS0, RSF1, HD0-HD15/HAD0-HAD15.

⁸0 V on BR, CLKIN Active (to force three-state condition).

⁹Idle refers to ADSP-2171/ADSP-2172 state of operation during execution of IDLE instruction. Deasserted pins are driven to either V_{DD} or GND. Current reflects device operation with CLKOUT disabled.

 $^{^{10}\}text{Current}$ reflects device operating with no output loads. $^{11}\text{V}_{\text{IN}}=0.4~\text{V}$ and 2.4 V. For typical figures for supply currents, refer to "Power Dissipation" section.

¹²See Chapter 9, of the ADSP-2100 Family User's Manual for details.

¹³Applies to TQFP and PQFP package types.

¹⁴Output pin capacitance is the capacitive load for any three-state output pin.

ADSP-2171/ADSP-2172

ABSOLUTE MAXIMUM RATINGS*

Supply Voltage
Input Voltage0.3 V to V _{DD} + 0.3 V
Output Voltage Swing0.3 V to V _{DD} + 0.3 V
Operating Temperature Range (Ambient)40°C to +85°C
Storage Temperature Range65°C to +150°C
Lead Temperature (5 sec) TQFP+280°C
Lead Temperature (5 sec) PQFP +280°C

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ESD SENSITIVITY

The ADSP-217x is an ESD (electrostatic discharge) sensitive device. Electrostatic charges readily accumulate on the human body and equipment and can discharge without detection. Permanent damage may occur to devices subjected to high energy electrostatic discharges.

The ADSP-217x features proprietary ESD protection circuitry to dissipate high energy discharges (Human Body Model). Per method 3015 of MIL-STD-883, the ADSP-217x has been classified as a Class 1 device.

Proper ESD precautions are recommended to avoid performance degradation or loss of functionality. Unused devices must be stored in conductive foam or shunts, and the foam should be discharged to the destination before devices are removed.



ADSP-2171/ADSP-2172 TIMING PARAMETERS

GENERAL NOTES

Use the exact timing information given. Do not attempt to derive parameters from the addition or subtraction of others. While addition or subtraction would yield meaningful results for an individual device, the values given in this data sheet reflect statistical variations and worst cases. Consequently, you cannot meaningfully add up parameters to derive longer times.

TIMING NOTES

Switching characteristics specify how the processor changes its signals. You have no control over this timing; it is dependent on the internal design. Timing requirements apply to signals that are controlled outside the processor, such as the data input for a read operation.

Timing requirements guarantee that the processor operates correctly with another device. Switching characteristics tell you what the device will do under a given circumstance. Also, use the switching characteristics to ensure any timing requirement of a device connected to the processor (such as memory) is satisfied.

ADSP-2171/ADSP-2172

MEMORY REQUIREMENTS

This chart links common memory device specification names and ADSP-2171/ADSP-2172 timing parameters for your convenience.

Parameter Name	Function	Common Memory Device Specification Name
t _{ASW}	A0-A13, DMS, PMS	Address Setup to
	Setup before WR Low	Write Start
t_{AW}	A0-A13, $\overline{\rm DMS}$, $\overline{\rm PMS}$	Setup Address Setup
	before WR Deasserted	to Write End
t_{WRA}	A0-A13, DMS, PMS	Address Hold Time
	Hold after WR Deasserted	
t_{DW}	Data Setup before WR High	Data Setup Time
t_{DH}	Data Hold after WR High	Data Hold Time
$t_{ m RDD}$	RD Low to Data Valid	OE to Data Valid
t_{AA}	A0-A13, $\overline{\rm DMS}$, $\overline{\rm PMS}$,	Address Access Time
	BMS to Data Valid	

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ADSP-2171/ADSP-2172

Parameter	•	Min	Max	Unit
input clock clock (whice 16.67 MHz range of 0.5 timing para	ed as $0.5~t_{\rm CKI}$. The ADSP-2171/ADSP-2172 uses an with a frequency equal to half the instruction rate; a h is equivalent to 60 ns) yields a 30 ns processor cycle input (equivalent to 33 MHz). $t_{\rm CK}$ values within the $t_{\rm CKI}$ period should be substituted for all relevant meters to obtain specification value. $t_{\rm CKI} = 0.5t_{\rm CK} - 7~t_{\rm NS} = 0.5~(30~t_{\rm NS}) - 7~t_{\rm NS} = 8~t_{\rm NS}$.			
Timing Red	quirement:			
$t_{\mathrm{CKI}} \ t_{\mathrm{CKIL}} \ t_{\mathrm{CKIH}}$	CLKIN Period CLKIN Width Low CLKIN Width High	60 20 20	150	ns ns ns
Switching 0	Characteristic:			
$\begin{array}{l} t_{CKL} \\ t_{CKH} \\ t_{CKOH} \end{array}$	CLKOUT Width Low CLKOUT Width High CLKIN High to CLKOUT High	$\begin{array}{c} 0.5t_{CK} - 7 \\ 0.5t_{CK} - 7 \\ 0 \end{array}$	20	ns ns ns
Control Si	gnals			
Timing Rec t_{RSP}	quirement: RESET Width Low	5t _{CK} ¹		ns

¹Applies after power-up sequence is complete. Internal phase lock loop requires no more than 2000 CLKIN cycles assuming stable CLKIN (not including crystal oscillator start-up time).

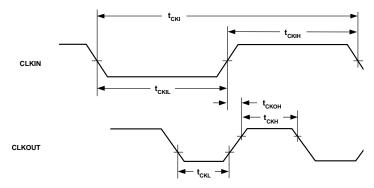


Figure 8. Clock Signals

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ADSP-2171/ADSP-2172

Parameter		Min	Max	Unit
Interrupts	and Flags			
Timing Re	quirement:			
$t_{IFS} \\ t_{IFH}$	IRQx or FI Setup before CLKOUT Low ^{1, 2, 3} IRQx or FI Hold after CLKOUT High ^{1, 2, 3}	$\begin{array}{c} 0.25t_{CK} + 15 \\ 0.25t_{CK} \end{array}$		ns ns
Switching 6	Characteristic:			
$\begin{array}{c} t_{FOH} \\ t_{FOD} \end{array}$	Flag Output Hold after CLKOUT Low ⁴ Flag Output Delay from CLKOUT Low ⁴	0.5t _{CK} - 7	$0.5t_{CK} + 5$	ns ns

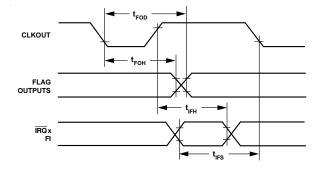


Figure 9. Interrupts and Flags

NOTES

If \overline{IRQx} and FI inputs meet t_{IFS} and t_{IFH} setup/hold requirements, they will be recognized during the current clock cycle; otherwise the signals will be recognized on the following cycle. (Refer to "Interrupt Controller Operation" in the Program Control chapter of the User's Manual for further information on interrupt servicing.)

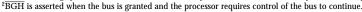
Edge-sensitive interrupts require pulse widths greater than 10 ns; level-sensitive interrupts must be held low until serviced. $\overline{IRQx} = \overline{IRQ0}, \overline{IRQ1}, \text{ and } \overline{IRQ2}.$ Flag Output = FL0, FL1, FL2, and FO.

ADSP-2171/ADSP-2172

Paramete	r	Min	Max	Unit
Bus Requ	est/Grant			
Timing Re	quirement:			
t_{BH} t_{BS}	\overline{BR} Hold after CLKOUT High 1 \overline{BR} Setup before CLKOUT Low 1	$\begin{array}{c} 0.25t_{CK} + 2 \\ 0.25t_{CK} + 17 \end{array}$		ns ns
Switching	Characteristic:			
$t_{\rm SD}$	CLKOUT High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Disable		$0.25t_{CK} + 16$	ns
t_{SDB}	$\overline{\rm DMS}$, $\overline{\rm PMS}$, $\overline{\rm BMS}$, $\overline{\rm RD}$, $\overline{\rm WR}$ Disable to $\overline{\rm BG}$ Low	0		ns
t_{SE}	\overline{BG} High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Enable	0		ns
t _{SEC}	DMS, PMS, BMS, RD, WR Enable to CLKOUT High	0.25t _{CK} - 7		ns
t _{SDBH}	DMS, PMS, BMS, RD, WR Disable to BGH Low ²	0		ns
t_{SEH}	\overline{BGH} High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Enable ²	0		ns

NOTES

 $^{^{1}\}overline{BR}$ is an asynchronous signal. If \overline{BR} meets the setup/hold requirements, it will be recognized during the current clock cycle; otherwise the signal will be recognized on the following cycle. Refer to the *ADSP-2100 Family User's Manual* for $\overline{BR}/\overline{BG}$ cycle relationships. $^{2}\overline{BGH}$ is asserted when the bus is granted and the processor requires control of the bus to continue.



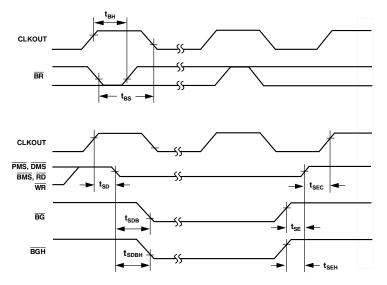


Figure 10. Bus Request-Bus Grant

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ADSP-2171/ADSP-2172

Parameter	Parameter		Max	Unit
Memory Re	ead			
Timing Requ	uirement:			
t _{RDD} t _{AA} t _{RDH}	RD Low to Data Valid A0-A13, PMS, DMS, BMS to Data Valid Data Hold from RD High	0	$\begin{array}{l} 0.5t_{CK} - 9 + w \\ 0.75t_{CK} - 10.5 + w \end{array}$	ns ns ns
Switching Cl	haracteristic:			
t_{RP} t_{CRD} t_{ASR} t_{RDA} t_{RWR}	RD Pulse Width CLKOUT High to RD Low A0-A13, PMS, DMS, BMS Setup before RD Low A0-A13, PMS, DMS, BMS Hold after RD Deasserted RD High to RD or WR Low	$\begin{array}{c} 0.5t_{CK} - 5 + w \\ 0.25t_{CK} - 5 \\ 0.25t_{CK} - 6 \\ 0.25t_{CK} - 3 \\ 0.5t_{CK} - 5 \end{array}$	$0.25t_{\rm CK} + 7$	ns ns ns ns

 $w = wait \ states \ x \ t_{CK}.$

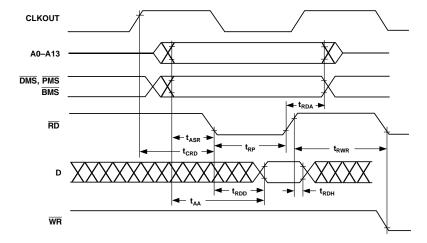


Figure 11. Memory Read

ADSP-2171/ADSP-2172

Parameter	r	Min	Max	Unit
Memory V	Vrite			
Switching 0	Characteristic:			
t_{DW}	Data Setup before WR High	$0.5 t_{CK} - 7 + w$		ns
t _{DH}	Data Hold after WR High	$0.25t_{CK} - 2$		ns
t_{WP}	WR Pulse Width	$0.5t_{CK} - 5 + w$		ns
t_{WDE}	WR Low to Data Enabled	0		ns
t_{ASW}	A0-A13, DMS , PMS Setup before WR Low	$0.25t_{\rm CK} - 6$		ns
t_{DDR}	Data Disable before WR or RD Low	$0.25t_{\rm CK} - 7$		ns
t_{CWR}	CLKOUT High to WR Low	$0.25t_{\rm CK} - 5$	$0.25 t_{\rm CK} + 7$	ns
t_{AW}	A0-A13, DMS, PMS, Setup before WR Deasserted	$0.75t_{CK} - 9 + w$		ns
t_{WRA}	A0-A13, DMS, PMS Hold after WR Deasserted	$0.25t_{\rm CK} - 3$		ns
t_{WWR}	\overline{WR} High to \overline{RD} or \overline{WR} Low	$0.5t_{\rm CK} - 5$		ns

 $w = wait states x t_{CK}$.

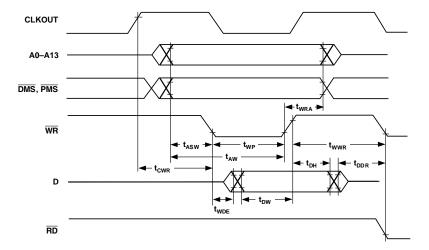


Figure 12. Memory Write

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ADSP-2171/ADSP-2172

Parameter	Parameter		Max	Unit
Serial Por	ts			
Timing Red	quirement:			
t_{SCK}	SCLK Period	50		ns
t_{SCS}	DR/TFS/RFS Setup before SCLK Low	4		ns
t_{SCH}	DR/TFS/RFS Hold after SCLK Low	7		ns
t_{SCP}	SCLK _{IN} Width	20		ns
Switching (Characteristic:			
t_{CC}	CLKOUT High to SCLK _{OUT}	$0.25t_{ m CK}$	$0.25t_{CK} + 10$	ns
t _{SCDE}	SCLK High to DT Enable	0		ns
t_{SCDV}	SCLK High to DT Valid		15	ns
t_{RH}	TFS/RFS _{OUT} Hold after SCLK High	0		ns
t_{RD}	TFS/RFS _{OUT} Delay from SCLK High		15	ns
t_{SCDH}	DT Hold after SCLK High	0		ns
$t_{ m TDE}$	TFS(Alt) to DT Enable	0		ns
t_{TDV}	TFS(Alt) to DT Valid		15	ns
t_{SCDD}	SCLK High to DT Disable		15	ns
t_{RDV}	RFS (Multichannel, Frame Delay Zero) to DT Valid		15	ns

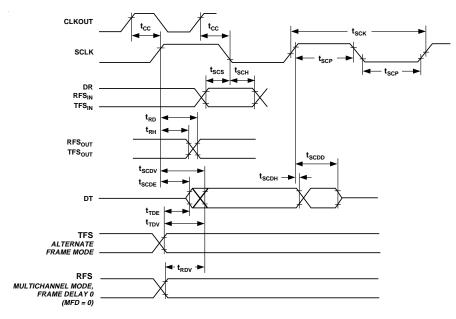


Figure 13. Serial Ports

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ADSP-2171/ADSP-2172

Parameter	,	Min	Max	Unit
	face Port ata and Address (HMD1 = 0) e and Write Strobe (HMD0 = 0)			
Timing Red	quirement:			
$t_{ m HSU}$ $t_{ m HDSU}$ $t_{ m HWDH}$ $t_{ m HH}$ $t_{ m HRWP}$	HA2-0 Setup before Start of Write or Read ^{1, 2} Data Setup before End of Write ³ Data Hold after End of Write ³ HA2-0 Hold after End of Write or Read ^{3, 4} Read or Write Pulse Width ⁵	5 5 3 3 20		ns ns ns ns
Switching (Characteristic:			
t _{HSHK} t _{HKH} t _{HDE} t _{HDD}	HACK Low after Start of Write or Read ^{1, 2} HACK Hold after End of Write or Read ^{3, 4} Data Enabled after Start of Read ² Data Valid after Start of Read ⁴ Data Hold after End of Read ⁴	0 0 0	15 15 18	ns ns ns ns
$t_{ m HRDD}$	Data Disabled after End of Read ⁴		7	ns

NOTES

[|] Start of Write = \overline{HWR} Low and \overline{HSEL} Low.
| Start of Read = \overline{HRD} Low and \overline{HSEL} Low.
| Send of Write = \overline{HWR} High or \overline{HSEL} High.
| Send of Read = \overline{HRD} High or \overline{HSEL} High.
| Send Pulse Width = \overline{HRD} Low and \overline{HSEL} Low, Write Pulse Width = \overline{HWR} Low and \overline{HSEL} Low.

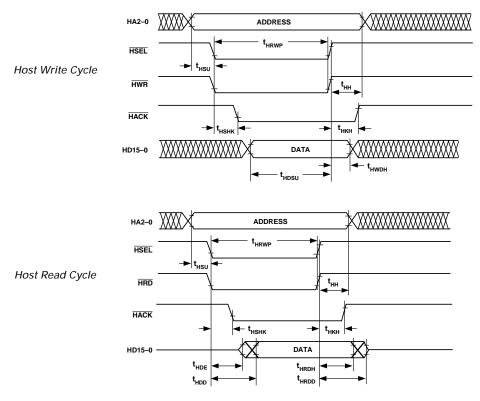


Figure 14. Host Interface Port (HMD1 = 0, HMD0 = 0)

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Parameter	Parameter		Max	Unit
	e Port and Address (HMD1 = 0) and Write Strobe (HMD0 = 1)			
Timing Requir	ement:			
t _{HSU} t _{HDSU} t _{HWDH} t _{HH} t _{HRWP}	HA2-0, HRW Setup before Start of Write or Read ¹ Data Setup before End of Write ² Data Hold after End of Write ² HA2-0, HRW Hold after End of Write or Read ² Read or Write Pulse Width ³	5 5 3 3 20		ns ns ns ns
Switching Cha	racteristic:			
t _{HSHK} t _{HKH} t _{HDE} t _{HDD} t _{HRDH}	HACK Low after Start of Write or Read ¹ HACK Hold after End of Write or Read ² Data Enabled after Start of Read ¹ Data Valid after Start of Read ¹ Data Hold after End of Read ²	0 0 0	15 15 18	ns ns ns ns
t _{HRDD}	Data Disabled after End of Read ²		7	ns

¹Start of Write or Read = HDS Low and HSEL Low.

²End of Write or Read = HDS High and HSEL High.

³Read or Write Pulse Width = HDS Low and HSEL Low.

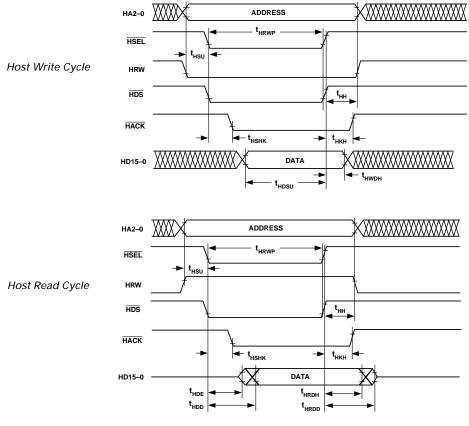


Figure 15. Host Interface Port (HMD1 = 0, HMD0 = 1)

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ADSP-2171/ADSP-2172

Parameter	r	Min	Max	Unit
	face Port d Data and Address (HMD1 = 1) e and Write Strobe (HMD0 = 0)			
Timing Re	quirement:			
thalp thasu thah thals thdsu thwdh thrwp	ALE Pulse Width HAD15-0 Address Setup, before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low ^{1, 2} HAD15-0 Data Setup before End of Write ³ HAD15-0 Data Hold after End of Write ³ Read or Write Pulse Width ⁴	10 5 2 10 5 3 20		ns ns ns ns ns ns
Switching 6	Characteristic:			
t _{HSHK} t _{HKH} t _{HDE} t _{HDD}	HACK Low after Start of Write or Read ^{1, 2} HACK Hold after End of Write or Read ^{3, 5} HAD15-0 Data Enabled after Start of Read ² HAD15-0 Data Valid after Start of Read HAD15-0 Data Hold after End of Read	0 0 0	15 15 18	ns ns ns ns
t_{HRDD}	HAD15-0 Data Disabled after End of Read ⁵		7	ns

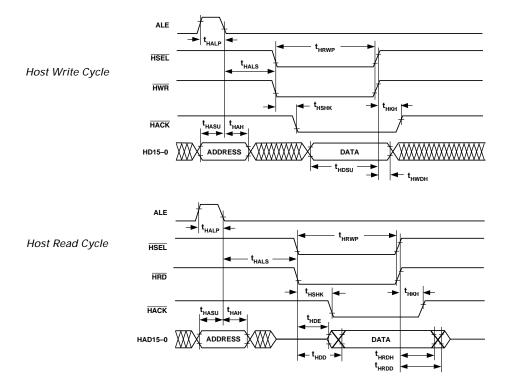


Figure 16. Host Interface Port (HMD1 = 1, HMD0 = 0)

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NOTES

Start of Write = \overline{HWR} Low and \overline{HSEL} Low.

Start of Read = \overline{HRD} Low and \overline{HSEL} Low.

Tend of Write = \overline{HWR} High or \overline{HSEL} High.

Read Pulse Width = \overline{HRD} Low and \overline{HSEL} Low, Write Pulse Width = \overline{HWR} Low and \overline{HSEL} Low.

Tend of Read = \overline{HRD} High or \overline{HSEL} High.

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ice Port			
Data and Address (HMD1 = 1) and Write Strobe (HMD0 = 1)			
irement:			
ALE Pulse Width HAD15-0 Address Setup before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low HRW Setup before Start of Write or Read¹ HAD15-0 Data Setup before End of Write² HAD15-0 Data Hold after End of Write² HRW Hold after End of Write or Read² Read or Write Pulse Width³	10 5 2 10 5 5 3 3 20		ns ns ns ns ns ns ns
naracteristic:			
HACK Low after Start of Write or Read ¹ HACK Hold after End of Write or Read ² HAD15-0 Data Enabled after Start of Read ¹ HAD15-0 Data Valid after Start of Read ¹ HAD15-0 Data Hold after End of Read ² HAD15-0 Data Disabled after End of Read ²	0 0 0	15 15 18	ns ns ns ns ns
ľ	and Write Strobe (HMD0 = 1) irement: ALE Pulse Width HAD15-0 Address Setup before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low¹ HRW Setup before Start of Write or Read¹ HAD15-0 Data Setup before End of Write² HAD15-0 Data Hold after End of Write² HRW Hold after End of Write or Read² Read or Write Pulse Width³ aracteristic: HACK Low after Start of Write or Read¹ HACK Hold after End of Write or Read² HAD15-0 Data Enabled after Start of Read¹ HAD15-0 Data Valid after Start of Read¹	and Write Strobe (HMD0 = 1) irement: ALE Pulse Width HAD15-0 Address Setup before ALE Low 5 HAD15-0 Address Hold after ALE Low 2 Start of Write or Read after ALE Low HRW Setup before Start of Write or Read HAD15-0 Data Setup before End of Write HAD15-0 Data Hold after End of Write HRW Hold after End of Write or Read Read or Write Pulse Width Taken Low after Start of Write or Read HAD15-0 Data Enabled after Start of Read HAD15-0 Data Hold after End of Read HAD15-0 Data Hold after End of Read	and Write Strobe (HMD0 = 1) irement: ALE Pulse Width HAD15-0 Address Setup before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low HRW Setup before Start of Write or Read ¹ HAD15-0 Data Setup before End of Write ² HAD15-0 Data Hold after End of Write ² HRW Hold after End of Write or Read ² Read or Write Pulse Width ³ aracteristic: HACK Low after Start of Write or Read ¹ HAD15-0 Data Enabled after Start of Read ¹ HAD15-0 Data Valid after Start of Read ¹ HAD15-0 Data Hold after End of Read ² HAD15-0 Data Hold after Start of Read ¹ HAD15-0 Data Hold after End of Read ² O 18

NOTES

¹Start of Write or Read = <u>HDS</u> Low and <u>HSEL</u> Low.

²End of Write or Read = <u>HDS</u> High and <u>HSEL</u> High.

³Read or Write Pulse Width = <u>HDS</u> Low and <u>HSEL</u> Low.

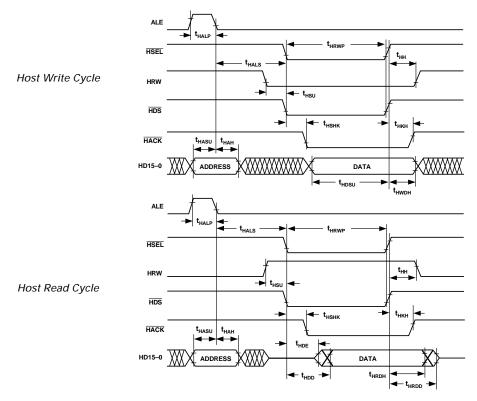


Figure 17. Host Interface Port (HMD1 = 1, HMD0 = 1)

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ADSP-2171/ADSP-2172

ENVIRONMENTAL CONDITIONS

Ambient Temperature Rating:

$$\begin{split} T_{AMB} &= T_{CASE} - (PD \times \theta_{CA}) \\ T_{CASE} &= Case \ Temperature \ in \ ^{\circ}C \end{split}$$

PD = Power Dissipation in W

 θ_{CA} = Thermal Resistance (Case-to-Ambient)

 θ_{JA} = Thermal Resistance (Junction-to-Ambient)

 θ_{IC} = Thermal Resistance (Junction-to-Case)

Package	$\theta_{ exttt{JA}}$	θ_{JC}	θ_{CA}
TQFP	50°C/W	2°C/W	48°C/W
PQFP	41°C/W	10°C/W	31°C/W

POWER DISSIPATION

To determine total power dissipation in a specific application, the following equation should be applied for each output:

$$C \times V_{DD}^2 \times f$$

C = load capacitance, f = output switching frequency.

Example:

In an application where external data memory is used and no other outputs are active, power dissipation is calculated as follows:

Assumptions:

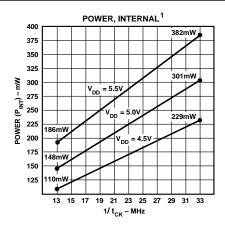
- External data memory is accessed every cycle with 50% of the address pins switching.
- External data memory writes occur every other cycle with 50% of the data pins switching.
- Each address and data pin has a 10 pF total load at the pin.
- The application operates at V_{DD} = 5.0 V and t_{CK} = 30 ns. *Total Power Dissipation* = $P_{INT} + (C \times V_{DD}^2 \times f)$

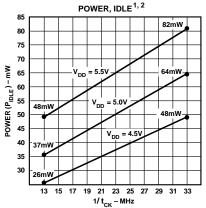
 P_{INT} = internal power dissipation from Power vs. Frequency graph (Figure 18).

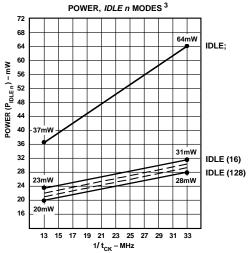
 $(C \times V_{DD}^2 \times f)$ is calculated for each output:

	# of				
	Pins	× C	$\times V_{DD}^2$	×f	
Address, DMS	8	× 10 pF	$\times 5^2 \text{ V}$	× 33.3 MHz =	66.6 mW
Data Output, WR	9	× 10 pF	$\times 5^2 \text{ V}$	× 16.67 MHz =	37.5 mW
$\overline{\text{RD}}$	1	× 10 pF	$\times 5^2 \text{ V}$	× 16.67 MHz =	4.2 mW
CLKOUT	1	× 10 pF	$\times 5^2 \text{ V}$	× 33.3 MHz =	8.3 mW
					116.6 mW

Total power dissipation for this example is P_{INT} + 116.6 mW.







VALID FOR ALL TEMPERATURE GRADES.

³ TYPICAL POWER DISSIPATION AT 5.0V V_{DD} DURING EXECUTION OF *IDLE N* INSTRUCTION (CLOCK FREQUENCY REDUCTION). POWER REFLECTS DEVICE OPERATING WITH CLKOUT DISABLED.

Figure 18. Power vs. Frequency

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¹ POWER REFLECTS DEVICE OPERATING WITH NO OUTPUT LOADS.

EXECUTION OF IDLE INSTRUCTION. DEASSERTED PINS ARE DRIVEN TO EITHER \mathbf{V}_{DD} OR GND. POWER REFLECTS DEVICE OPERATING WITH CLKOUT DISABLED.

ADSP-2171/ADSP-2172

CAPACITIVE LOADING

Figures 19 and 20 show the capacitive loading characteristics of the ADSP-2171/ADSP-2172.

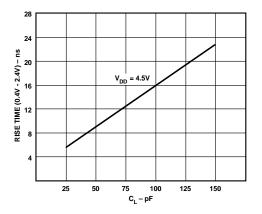


Figure 19. Typical Output Rise Time vs. Load Capacitance, C_L (at Maximum Ambient Operating Temperature)

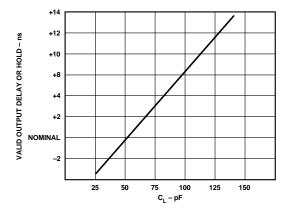


Figure 20. Typical Output Valid Delay or Hold vs. Load Capacitance, C_L (at Maximum Ambient Operating Temperature)

TEST CONDITIONS

Output Disable Time

Output pins are considered to be disabled when they have stopped driving and started a transition from the measured output high or low voltage to a high impedance state. The output disable time ($t_{\rm DIS}$) is the difference of $t_{\rm MEASURED}$ and $t_{\rm DECAY}$, as shown in the Output Enable/Disable diagram. The time is the interval from when a reference signal reaches a high or low voltage level to when the output voltages have changed by 0.5 V from the measured output high or low voltage. The decay time,

 $t_{\rm DECAY}$, is dependent on the capacitative load, $C_{\rm L}$, and the current load, $i_{\rm L}$, on the output pin. It can be approximated by the following equation:

$$t_{DECAY} = \frac{C_L \bullet 0.5 \, V}{i_L}$$

from which

$$t_{DIS} = t_{MEASURED} - t_{DECAY}$$

is calculated. If multiple pins (such as the data bus) are disabled, the measurement value is that of the last pin to stop driving.

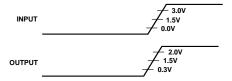
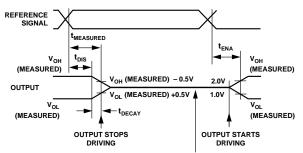


Figure 21. Voltage Reference Levels for AC Measurements (Except Output Enable/Disable)

Output Enable Time

Output pins are considered to be enabled when that have made a transition from a high-impedance state to when they start driving. The output enable time ($t_{\rm ENA}$) is the interval from when a reference signal reaches a high or low voltage level to when the output has reached a specified high or low trip point, as shown in the Output Enable/Disable diagram. If multiple pins (such as the data bus) are enabled, the measurement value is that of the first pin to start driving.



HIGH-IMPEDANCE STATE. TEST CONDITIONS CAUSE
THIS VOLTAGE LEVEL TO BE APPROXIMATELY 1.5V.

Figure 22. Output Enable/Disable

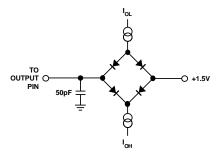


Figure 23. Equivalent Device Loading for AC Measurements (Including All Fixtures)

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ADSP-2173—SPECIFICATIONS

RECOMMENDED OPERATING CONDITIONS

		K Grade		B Grade			
Parameter		Min	Max	Min	Max	Unit	
$egin{array}{c} \overline{V_{ m DD}} \ T_{ m AMB} \end{array}$	Supply Voltage Ambient Operating Temperature	3.0 0	3.6 +70	3.0 -40	3.6 +85	V °C	

ELECTRICAL CHARACTERISTICS

			K/B Gra	ades	
Parameter		Test Conditions	Min	Max	Unit
$\overline{V_{\text{IH}}}$	Hi-Level Input Voltage ^{1, 2}	@ V _{DD} = max	2.0		V
V_{IH}	Hi-Level CLKIN Voltage	$@V_{DD} = max$	2.0		V
V_{IH}	Hi-Level RESET Voltage	$@V_{DD} = max$	2.2		V
V_{IL}	Lo-Level Input Voltage ^{Y, 3}	@ V _{DD} = min		0.4	V
V_{OH}	Hi-Level Output Voltage ^{1, 4, 5}	$@V_{DD} = min$			
		$I_{OH} = -0.5 \text{ mA}$	2.4		V
		$@V_{DD} = min$			
		$I_{OH} = -100 \text{ mA}^6$	V_{DD} – 0.3		V
V_{OL}	Lo-Level Output Voltage ^{1, 4, 5}	$@V_{DD} = min$			
		$I_{OL} = 2 \text{ mA}$		0.4	V
I_{IH}	Hi-Level Input Current ³	$@V_{DD} = max$			
		$V_{IN} = V_{DD} \max$		10	μA
I_{IL}	Lo-Level Input Current ³	$@V_{DD} = max$			
		$V_{IN} = 0 V$		10	μA
I_{OZH}	Tristate Leakage Current ⁷	$@V_{DD} = max,$			
	- 7	$V_{\rm IN} = V_{\rm DD} \text{max}^8$		10	μΑ
I_{OZL}	Tristate Leakage Current ⁷	$@V_{DD} = max,$			
_		$V_{IN} = 0 V^8$		10	μA
I_{DD}	Supply Current (Idle) ^{9, 10}	$@V_{DD} = max$		7	mA
I_{DD}	Supply Current (Dynamic) ¹⁰	$@V_{DD} = \max_{11}$			
	G 1 G . (B 1)10	$t_{\rm CK} = 50 \text{ ns}^{11}$		27	mA
I_{DD}	Supply Current (Powerdown) ¹⁰	Lowest Power Mode ¹²		100	μΑ
C_{I}	Input Pin Capacitance ^{3, 6, 13}	@ $V_{IN} = 2.5 \text{ V}$,			
		$f_{IN} = 1.0 \text{ MHz},$		0	
C	O + + P: C + + 6.7.13.14	$T_{AMB} = 25^{\circ}C$		8	pF
C_{O}	Output Pin Capacitance ^{6, 7, 13, 14}	$@V_{IN} = 2.5 \text{ V},$			
		$f_{IN} = 1.0 \text{ MHz},$		o	, T.
		$T_{AMB} = 25^{\circ}C$		8	pF

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¹Bidirectional pins: D0-D23, RFS0, RFS1, SCLK0, SCLK1, TFS0, TFS1, HD0-HD15/HAD0-HAD15.

²Input only pins: RESET, IRQ2, BR, MMAP, DR0, DR1, HSEL, HSIZE, BMODE, HMD0, HMD1, HRD/HWR, HWR/HDS, PWD, HA2/ALE, HA1-0.

³Input only pins: CLKIN, RESET, IRQ2, BR, MMAP, DR0, DR1, HSEL, HSIZE, BMODE, HMD0, HMD1, HRD/HWR, HWR/HDS, PWD, HA2/ALE, HA1-0. ⁴Output pins: \overline{BG} , \overline{PMS} , \overline{DMS} , \overline{BMS} , \overline{RD} , \overline{WR} , PWDACK, A0-A13, DT0, DT1, CLKOUT, \overline{HACK} , FL2-0, \overline{BGH} .

⁵Although specified for TTL outputs, all ADSP-2173 outputs are CMOS-compatible and will drive to V_{DD} and GND, assuming no dc loads.

⁶Guaranteed but not tested.

⁷Three-statable pins: A0-A13, D0-D23, PMS, DMS, BMS, RD, WR, DT0, DT1, SCLK0, SCLK1, TFS0, TFS1, RFS0, RSF1, HD0-HD15/HAD0-HAD15.

⁸0 V on BR, CLKIN Active (to force three-state condition).

 $^{^9}$ Idle refers to ADSP-2173 state of operation during execution of IDLE instruction. Deasserted pins are driven to either V $_{
m DD}$ or GND. Current reflects device operation with CLKOUT disabled.

¹⁰Current reflects device operating with no output loads.

 $^{^{11}}V_{IN} = 0.4 \text{ V}$ and 2.4 V. For typical figures for supply currents, refer to "Power Dissipation" section.

¹²See Chapter 9, of the ADSP-2100 Family User's Manual for details.

¹³Applies to TQFP and PQFP package types.
14Output pin capacitance is the capacitve load for any three-state output pin.

Specifications subject to change without notice.

ADSP-2173 TIMING PARAMETERS

GENERAL NOTES

Use the exact timing information given. Do not attempt to derive parameters from the addition or subtraction of others. While addition or subtraction would yield meaningful results for an individual device, the values given in this data sheet reflect statistical variations and worst cases. Consequently, you cannot meaningfully add up parameters to derive longer times.

TIMING NOTES

Switching characteristics specify how the processor changes its signals. You have no control over this timing; it is dependent on the internal design. Timing requirements apply to signals that are controlled outside the processor, such as the data input for a read operation.

Timing requirements guarantee that the processor operates correctly with another device. Switching characteristics tell you what the device will do under a given circumstance. Also, use the switching characteristics to ensure any timing requirement of a device connected to the processor (such as memory) is satisfied.

MEMORY REQUIREMENTS

This chart links common memory device specification names and ADSP-2173 timing parameters for your convenience.

Parameter Name	Function	Common Memory Device Specification Name
t _{ASW}	A0-A13, DMS, PMS	Address Setup to
	Setup before WR Low	Write Start
t_{AW}	A0-Â13, DMS, PMS	Setup Address Setup
	before WR Deasserted	to Write End
t_{WRA}	A0-A13, $\overline{\rm DMS}$, $\overline{\rm PMS}$	Address Hold Time
	Hold after WR Deasserted	
t_{DW}	Data Setup before WR High	Data Setup Time
t_{DH}	Data Hold after WR High	Data Hold Time
$t_{ m RDD}$	RD Low to Data Valid	OE to Data Valid
t_{AA}	A0-A13, $\overline{\rm DMS}$, $\overline{\rm PMS}$,	Address Access Time
	BMS to Data Valid	

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Parameter	Parameter		Max	Unit
a frequency clock (which (equivalent period shou to obtain sp	eads ed as $0.5~t_{CKI}$. The ADSP-2173 uses an input clock with equal to half the instruction rate; a $10.0~MHz$ input h is equivalent to $100~ns$) yields a $50~ns$ processor cycle to $20~MHz$). t_{CK} values within the range of $0.5~t_{CKI}$ ald be substituted for all relevant timing parameters pecification value. $t_{CKH} = 0.5t_{CK} - 10~ns = 0.5~(50~ns) - 10~ns = 15~ns$.			
Timing Rec	quirement:			
$\begin{array}{l} t_{CKI} \\ t_{CKIL} \\ t_{CKIH} \end{array}$	CLKIN Period CLKIN Width Low CLKIN Width High	100 20 20	160	ns ns ns
Switching C	Characteristic:			
$\begin{array}{l} t_{CKL} \\ t_{CKH} \\ t_{CKOH} \end{array}$	CLKOUT Width Low CLKOUT Width High CLKIN High to CLKOUT High	$\begin{array}{c} 0.5t_{CK} - 10 \\ 0.5t_{CK} - 10 \\ 0 \end{array}$	25	ns ns ns
Control Si	gnals			
Timing Rec	quirement: RESET Width Low	5t _{CK} ¹		ns

¹Applies after power-up sequence is complete. Internal phase lock loop requires no more than 2000 CLKIN cycles assuming stable CLKIN (not including crystal oscillator start-up time).

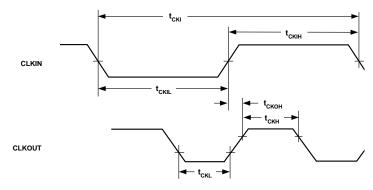


Figure 24. Clock Signals

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Parameter		Min	Max	Unit
Interrupts a	and Flags			
Timing Requ	uirement:			
$t_{\rm IFS} \\ t_{\rm IFH}$	\overline{IRQx} or FI Setup before CLKOUT Low ^{1, 2, 3} \overline{IRQx} or FI Hold after CLKOUT High ^{1, 2, 3}	$\begin{array}{c} 0.25t_{CK} + 23 \\ 0.25t_{CK} \end{array}$		ns ns
Switching Cl	haracteristic:			
$t_{ m FOH} \ t_{ m FOD}$	Flag Output Hold after CLKOUT Low ⁴ Flag Output Delay from CLKOUT Low ⁴	0.5t _{CK} - 10	$0.5t_{CK} + 5$	ns ns

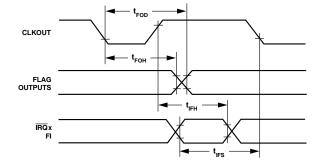


Figure 25. Interrupts and Flags

NOTES

If \overline{IRQx} and FI inputs meet t_{IFS} and t_{IFH} setup/hold requirements, they will be recognized during the current clock cycle; otherwise the signals will be recognized on the following cycle. (Refer to "Interrupt Controller Operation" in the Program Control chapter of the User's Manual for further information on interrupt servicing.)

Edge-sensitive interrupts require pulse widths greater than 10 ns; level-sensitive interrupts must be held low until serviced. $\overline{IRQx} = \overline{IRQ0}, \overline{IRQ1}, \text{ and } \overline{IRQ2}.$ Flag Output = FL0, FL1, FL2, and FO.

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Parameter		Min	Max	Unit
Bus Requ	est/Grant			
Timing Re	quirement:			
t_{BH} t_{BS}	\overline{BR} Hold after CLKOUT High \overline{BR} Setup before CLKOUT Low 1	$\begin{array}{c} 0.25t_{CK} + 2 \\ 0.25t_{CK} + 22 \end{array}$		ns ns
Switching 6	Characteristic:			
t_{SD}	CLKOUT High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Disable		$0.25t_{CK} + 16$	ns
t_{SDB}	$\overline{\rm DMS}$, $\overline{\rm PMS}$, $\overline{\rm BMS}$, $\overline{\rm RD}$, $\overline{\rm WR}$ Disable to $\overline{\rm BG}$ Low	0		ns
t_{SE}	\overline{BG} High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Enable	0		ns
t_{SEC}	DMS, PMS, BMS, RD, WR Enable to CLKOUT High	0.25t _{CK} - 10		ns
t_{SDBH}	DMS, PMS, BMS, RD, WR Disable to BGH Low ²	0		ns
t_{SEH}	\overline{BGH} High to \overline{DMS} , \overline{PMS} , \overline{BMS} , \overline{RD} , \overline{WR} Enable ²	0		ns

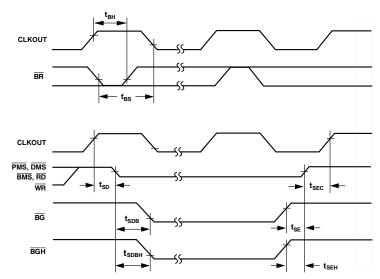


Figure 26. Bus Request-Bus Grant

NOTES

**\bar{BR}\$ is an asynchronous signal. If \$\overline{BR}\$ meets the setup/hold requirements, it will be recognized during the current clock cycle; otherwise the signal will be recognized on the following cycle. Refer to the \$ADSP-2100 Family User's Manual for \$\overline{BR}/\overline{BG}\$ cycle relationships.

**\bar{BGH}\$ is asserted when the bus is granted and the processor requires control of the bus to continue.

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Parameter		Min	Max	Unit
Memory R	ead			
Timing Req	uirement:			
t _{RDD} t _{AA} t _{RDH}	RD Low to Data Valid A0-A13, PMS, DMS, BMS to Data Valid Data Hold from RD High	0	$\begin{array}{l} 0.5t_{CK} - 15 + w \\ 0.75t_{CK} - 20.5 + w \end{array}$	ns ns ns
Switching C	Characteristic:			
$t_{ m RP}$ $t_{ m CRD}$ $t_{ m ASR}$ $t_{ m RDA}$ $t_{ m RWR}$	RD Pulse Width CLKOUT High to RD Low A0-A13, PMS, DMS, BMS Setup before RD Low A0-A13, PMS, DMS, BMS Hold after RD Deasserted RD High to RD or WR Low	$\begin{array}{c} 0.5t_{CK} - 5 + w \\ 0.25t_{CK} - 5 \\ 0.25t_{CK} - 7 \\ 0.25t_{CK} - 3 \\ 0.5t_{CK} - 5 \end{array}$	$0.25t_{CK} + 10$	ns ns ns ns

 $w = wait \ states \ x \ t_{CK}.$

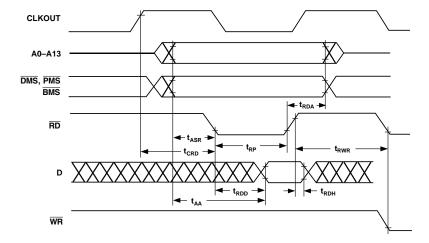


Figure 27. Memory Read

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Paramete	er	Min	Max	Unit
Memory	Write			
Switching	Characteristic:			
t_{DW}	Data Setup before WR High	$0.5 t_{CK} - 7 + w$		ns
t _{DH}	Data Hold after WR High	$0.25t_{CK} - 2$		ns
t_{WP}	WR Pulse Width	$0.5t_{\rm CK} - 5 + w$		ns
t_{WDE}	WR Low to Data Enabled	0		ns
t_{ASW}	A0–A13, $\overline{\rm DMS}$, $\overline{\rm PMS}$ Setup before $\overline{\rm WR}$ Low	$0.25t_{CK} - 7$		ns
t_{DDR}	Data Disable before WR or RD Low	$0.25t_{\rm CK} - 7$		ns
t_{CWR}	CLKOUT High to $\overline{ m WR}$ Low	$0.25t_{\rm CK} - 5$	$0.25 t_{CK} + 10$	ns
t_{AW}	A0-A13, DMS, PMS, Setup before WR Deasserted	$0.75t_{CK} - 11.5 + w$		ns
t_{WRA}	A0-A13, DMS, PMS Hold after WR Deasserted	$0.25t_{\rm CK} - 3$		ns
t_{WWR}	\overline{WR} High to \overline{RD} or \overline{WR} Low	$0.5t_{\rm CK} - 5$		ns

 $w = wait states x t_{CK}$.

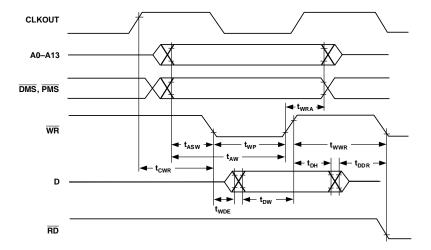


Figure 28. Memory Write

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Parameter		Min	Max	Unit
Serial Por	ts			
Timing Rec	uirement:			
t_{SCK} t_{SCS} t_{SCH}	SCLK Period DR/TFS/RFS Setup before SCLK Low DR/TFS/RFS Hold after SCLK Low SCLK _{IN} Width	76.9 8 10 28		ns ns ns
Switching C	Characteristic:			
$\begin{array}{c} t_{CC} \\ t_{SCDE} \\ t_{SCDV} \\ t_{RH} \end{array}$	CLKOUT High to SCLK _{OUT} SCLK High to DT Enable SCLK High to DT Valid TFS/RFS _{OUT} Hold after SCLK High	0.25t _{CK} 0	$0.25t_{CK} + 15$	ns ns ns
$t_{ m RD}$ $t_{ m SCDH}$ $t_{ m TDE}$ $t_{ m TDV}$ $t_{ m SCDD}$	TFS/RFS _{OUT} Delay from SCLK High DT Hold after SCLK High TFS(Alt) to DT Enable TFS(Alt) to DT Valid SCLK High to DT Disable	0 0	20 19 25	ns ns ns ns
t_{RDV}	RFS (Multichannel, Frame Delay Zero) to DT Valid		20	ns

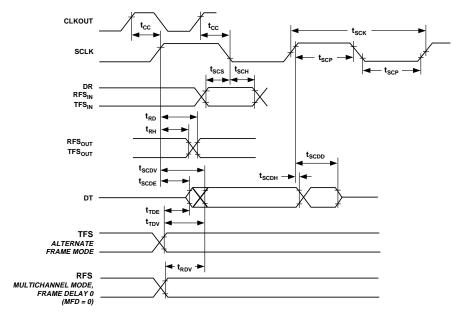


Figure 29. Serial Ports

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Paramete	r	Min	Max	Unit
	rface Port vata and Address (HMD1 = 0) ve and Write Strobe (HMD0 = 0)			
Timing Re	quirement:			
$t_{ m HSU}$ $t_{ m HDSU}$ $t_{ m HWDH}$ $t_{ m HH}$ $t_{ m HRWP}$	HA2-0 Setup before Start of Write or Read ^{1, 2} Data Setup before End of Write ³ Data Hold after End of Write ³ HA2-0 Hold after End of Write or Read ^{3, 4} Read or Write Pulse Width ⁵	8 8 3 3 30		ns ns ns ns
Switching	Characteristic:			
t _{HSHK} t _{HKH} t _{HDE} t _{HDD} t _{HRDH}	HACK Low after Start of Write or Read ^{1, 2} HACK Hold after End of Write or Read ^{3, 4} Data Enabled after Start of Read ² Data Valid after Start of Read ² Data Hold after End of Read ⁴ Data Disabled after End of Read ⁴	0 0 0	20 20 23 15	ns ns ns ns ns

NOTES

[|] Start of Write = \overline{HWR} Low and \overline{HSEL} Low.
| Start of Read = \overline{HRD} Low and \overline{HSEL} Low.
| Start of Write = \overline{HWR} High or \overline{HSEL} High.
| Start of Read = \overline{HRD} High or \overline{HSEL} High.
| Start of Read = \overline{HRD} High or \overline{HSEL} High.
| Start of Read Pulse Width = \overline{HRD} Low and \overline{HSEL} Low. Write Pulse Width = \overline{HWR} Low and \overline{HSEL} Low.

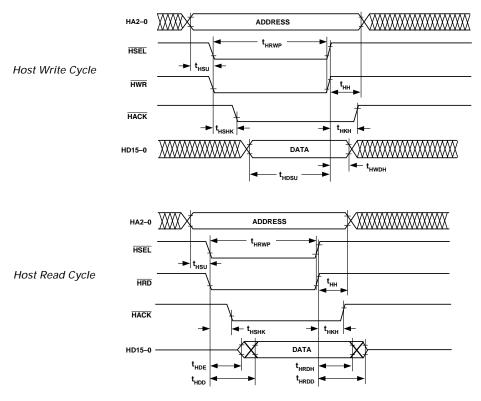


Figure 30. Host Interface Port (HMD1 = 0, HMD0 = 0)

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Parameter	r	Min	Max	Unit
	face Port ata and Address (HMD1 = 0) e and Write Strobe (HMD0 = 1)			
Timing Red	quirement:			
$t_{ m HSU}$ $t_{ m HDSU}$ $t_{ m HWDH}$ $t_{ m HH}$ $t_{ m HRWP}$	HA2-0, HRW Setup before Start of Write or Read ¹ Data Setup before End of Write ² Data Hold after End of Write ² HA2-0, HRW Hold after End of Write or Read ² Read or Write Pulse Width ³	8 8 3 3 30		ns ns ns ns
Switching (Characteristic:			
$t_{ m HSHK}$ $t_{ m HKH}$ $t_{ m HDE}$ $t_{ m HDD}$ $t_{ m HRDH}$	HACK Low after Start of Write or Read ¹ HACK Hold after End of Write or Read ² Data Enabled after Start of Read ¹ Data Valid after Start of Read ¹ Data Hold after End of Read ² Data Disabled after End of Read ²	0 0 0	20 20 23 15	ns ns ns ns ns

¹Start of Write or Read = HDS Low and HSEL Low.

²End of Write or Read = HDS High and HSEL High.

³Read or Write Pulse Width = HDS Low and HSEL Low.

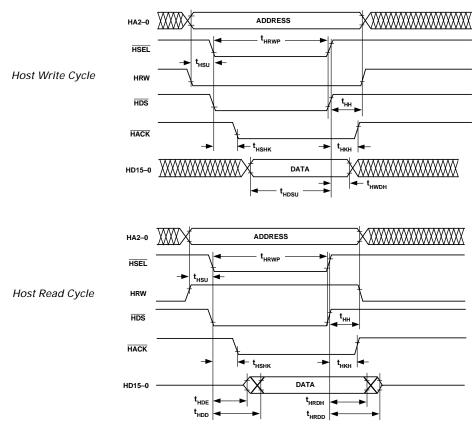


Figure 31. Host Interface Port (HMD1 = 0, HMD0 = 1)

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Parameter	Γ	Min	Max	Unit
	face Port d Data and Address (HMD1 = 1) e and Write Strobe (HMD0 = 0)			
Timing Red	quirement:			
t _{HALP} t _{HASU} t _{HAH} t _{HALS} t _{HDSU} t _{HWDH}	ALE Pulse Width HAD15-0 Address Setup, before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low ^{1, 2} HAD15-0 Data Setup before End of Write ³ HAD15-0 Data Hold after End of Write ³ Read or Write Pulse Width ⁵	15 5 2 15 8 3 30		ns ns ns ns ns ns
Switching (Characteristic:			
t _{HSHK} t _{HKH} t _{HDE} t _{HDD} t _{HRDH} t _{HRDD}	HACK Low after Start of Write or Read ^{1, 2} HACK Hold after End of Write or Read ^{3, 4} HAD15-0 Data Enabled after Start of Read ² HAD15-0 Data Valid after Start of Read ² HAD15-0 Data Hold after End of Read HAD15-0 Data Disabled after End of Read ⁴	0 0 0	20 20 23 15	ns ns ns ns ns

NOTES

NOTES

1Start of Write = \overline{HWR} Low and \overline{HSEL} Low.

2Start of Read = \overline{HRD} Low and \overline{HSEL} Low.

3End of Write = \overline{HWR} High or \overline{HSEL} High.

4End of Read = \overline{HRD} High or \overline{HSEL} High.

5Read Pulse Width = \overline{HRD} Low and \overline{HSEL} Low, Write Pulse Width = \overline{HWR} Low and \overline{HSEL} Low.

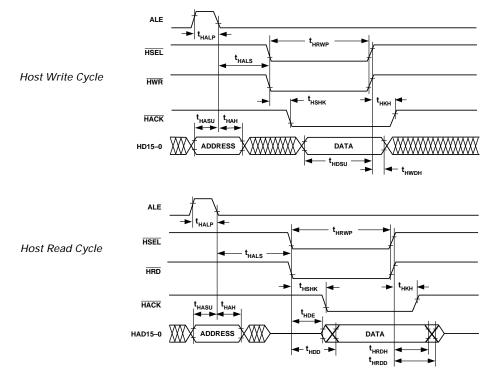


Figure 32. Host Interface Port (HMD1 = 1, HMD0 = 0)

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Parameter		Min	Max	Unit
	face Port Data and Address (HMD1 = 1) e and Write Strobe (HMD0 = 1)			
Timing Rec	uirement:			
t _{HALP} t _{HASU} t _{HAH} t _{HALS} t _{HSU} t _{HDSU} t _{HWDH} t _{HH}	ALE Pulse Width HAD15-0 Address Setup before ALE Low HAD15-0 Address Hold after ALE Low Start of Write or Read after ALE Low HRW Setup before Start of Write or Read¹ HAD15-0 Data Setup before End of Write² HAD15-0 Data Hold after End of Write² HRW Hold after End of Write or Read² Read or Write Pulse Width³	15 5 2 15 8 8 3 3 3		ns ns ns ns ns ns ns
Switching C	Characteristic:			
$t_{ m HSHK}$ $t_{ m HKH}$ $t_{ m HDE}$ $t_{ m HDD}$ $t_{ m HRDH}$ $t_{ m HRDD}$	HACK Low after Start of Write or Read ¹ HACK Hold after End of Write or Read ² HAD15-0 Data Enabled after Start of Read ¹ HAD15-0 Data Valid after Start of Read ¹ HAD15-0 Data Hold after End of Read ² HAD15-0 Data Disabled after End of Read ²	0 0 0	20 20 23 15	ns ns ns ns ns

NOTES

¹Start of Write or Read = \overline{HDS} Low and \overline{HSEL} Low.

²End of Write or Read = \overline{HDS} High and \overline{HSEL} High.

³Read or Write Pulse Width = \overline{HDS} Low and \overline{HSEL} Low.

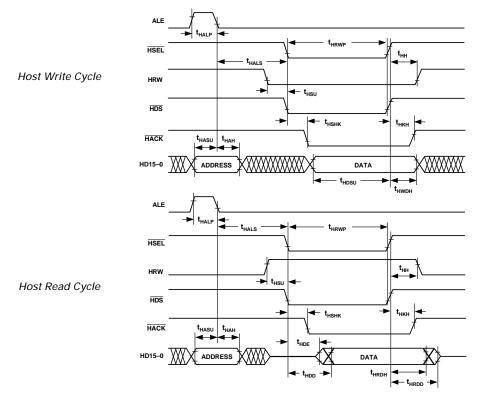


Figure 33. Host Interface Port (HMD1 = 1, HMD0 = 1)

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ENVIRONMENTAL CONDITIONS

Ambient Temperature Rating:

$$\begin{split} T_{AMB} &= T_{CASE} - (PD \times \theta_{CA}) \\ T_{CASE} &= Case \ Temperature \ in \ ^{\circ}C \\ PD &= Power \ Dissipation \ in \ W \end{split}$$

 θ_{CA} = Thermal Resistance (Case-to-Ambient)

 θ_{JA} = Thermal Resistance (Junction-to-Ambient)

 θ_{JC} = Thermal Resistance (Junction-to-Case)

Package	$\theta_{ exttt{JA}}$	θ_{JC}	$\theta_{\mathbf{CA}}$
TQFP	50°C/W	2°C/W	48°C/W
PQFP	41°C/W	10°C/W	31°C/W

POWER DISSIPATION

To determine total power dissipation in a specific application, the following equation should be applied for each output:

$$C \times V_{DD}^2 \times f$$

C = load capacitance, f = output switching frequency.

Example:

In an application where external data memory is used and no other outputs are active, power dissipation is calculated as follows:

Assumptions:

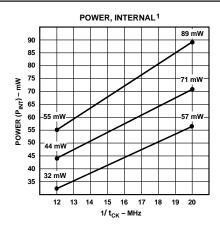
- External data memory is accessed every cycle with 50% of the address pins switching.
- · External data memory writes occur every other cycle with 50% of the data pins switching.
- Each address and data pin has a 10 pF total load at the pin.
- The application operates at V_{DD} = 3.3 V and t_{CK} = 50 ns. Total Power Dissipation = $P_{INT} + (C \times V_{DD}^2 \times f)$

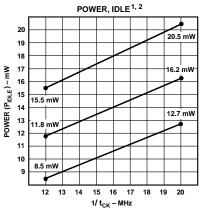
 P_{INT} = internal power dissipation from Power vs. Frequency graph (Figure 18).

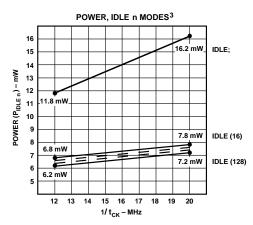
 $(C \times V_{DD}^2 \times f)$ is calculated for each output:

	# of Pins	× C	\times V_{DD}^2	×f		
Address, DMS Data Output, WR RD CLKOUT	9	× 10 pF × 10 pF	$\times 3.3^2 \text{ V}$ $\times 3.3^2 \text{ V}$	× 20 MHz × 10 MHz × 10 MHz × 20 MHz	=	9.8 mW 1.1 mW
						30.5 mW

Total power dissipation for this example is $P_{\rm INT}~+~30.5~mW$.







VALID FOR ALL TEMPERATURE GRADES.

¹ POWER REFLECTS DEVICE OPERATING WITH NO OUTPUT LOADS. ² IDLE REFERS TO ADSP-2173 STATE OF OPERATION DURING EXECUTION OF IDLE INSTRUCTION. DEASSERTED PINS ARE DRIVEN TO EITHER V_{DD} OR GND. POWER REFLECTS DEVICE

OPERATING WITH CLKOUT DISABLED. 3 TYPICAL POWER DISSIPATION AT 3.3V V_{DD} DURING EXECUTION OF IDLE n INSTRUCTION (CLOCK FREQUENCY REDUCTION).
POWER REFLECTS DEVICE OPERATING WITH CLKOUT DISABLED.

Figure 34. Power vs. Frequency

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CAPACITIVE LOADING

Figures 35 and 36 show the capacitive loading characteristics of the ADSP-2173.

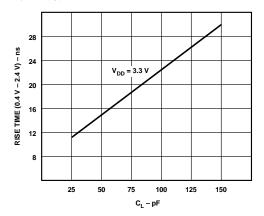


Figure 35. Typical Output Rise Time vs. Load Capacitance, C_L (at Maximum Ambient Operating Temperature)

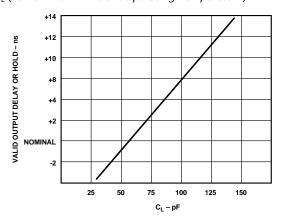


Figure 36. Typical Output Valid Delay or Hold vs. Load Capacitance, C_L (at Maximum Ambient Operating Temperature)

TEST CONDITIONS Output Disable Time

Output pins are considered to be disabled when they have stopped driving and started a transition from the measured output high or low voltage to a high impedance state. The output disable time ($t_{\rm DIS}$) is the difference of $t_{\rm MEASURED}$ and $t_{\rm DECAY}$, as shown in the Output Enable/Disable diagram. The time is the interval from when a reference signal reaches a high or low voltage level to when the output voltages have changed by 0.5 V from the measured output high or low voltage. The decay time,

 $t_{\rm DECAY}$, is dependent on the capacitative load, $C_{\rm L}$, and the current load, $i_{\rm L}$, on the output pin. It can be approximated by the following equation:

$$t_{DECAY} = \frac{C_L \cdot 0.5 \, V}{i_L}$$

from which

$$t_{DIS} = t_{MEASURED} - t_{DECAY}$$

is calculated. If multiple pins (such as the data bus) are disabled, the measurement value is that of the last pin to stop driving.

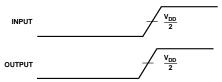
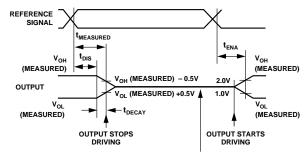


Figure 37. Voltage Reference Levels for AC Measurements (Except Output Enable/Disable)

Output Enable Time

Output pins are considered to be enabled when that have made a transition from a high-impedance state to when they start driving. The output enable time $(t_{\rm ENA})$ is the interval from when a reference signal reaches a high or low voltage level to when the output has reached a specified high or low trip point, as shown in the Output Enable/Disable diagram. If multiple pins (such as the data bus) are enabled, the measurement value is that of the first pin to start driving.



HIGH-IMPEDANCE STATE. TEST CONDITIONS CAUSE THIS VOLTAGE LEVEL TO BE APPROXIMATELY 1.5V.

Figure 38. Output Enable/Disable

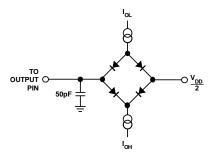
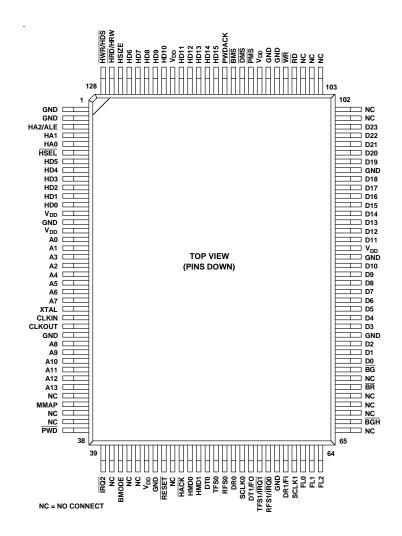


Figure 39. Equivalent Device Loading for AC Measurements (Including All Fixtures)

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128-Lead TQFP Package Pinout



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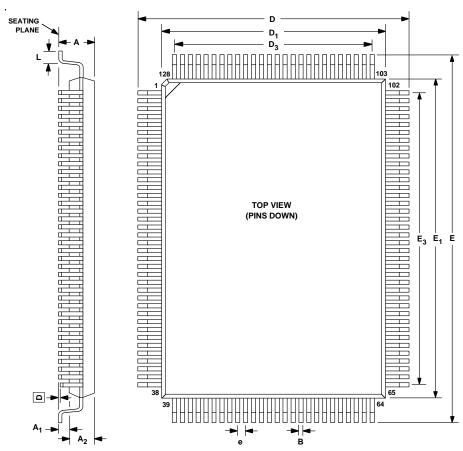
TQFP Pin Configurations

TQFP Number	Pin Name	TQFP Number	Pin Name	TQFP Number	Pin Name	TQFP Number	Pin Name
1	GND	33	A13	65	NC	97	D20
2	GND	34	NC	66	$\overline{\text{BGH}}$	98	D21
3	HA2/ALE	35	MMAP	67	NC	99	D22
4	HA1	36	NC	68	NC	100	D23
5	HA0	37	NC	69	NC	101	NC
6	HSEL	38	$\overline{ ext{PWD}}$	70	$\overline{\mathrm{BR}}$	102	NC
7	HD5	39	$\overline{\text{IRQ2}}$	71	NC	103	NC
8	HD4	40	NC	72	$\overline{\mathrm{BG}}$	104	NC
9	HD3	41	BMODE	73	D0	105	NC
10	HD2	42	NC	74	D1	106	$\overline{ ext{RD}}$
11	HD1	43	NC	75	D2	107	$\overline{\mathrm{WR}}$
12	HD0	44	$ m V_{DD}$	76	GND	108	GND
13	$V_{ m DD}$	45	GND	77	D3	109	GND
14	GND	46	RESET	78	D4	110	$V_{ m DD}$
15	$V_{ m DD}$	47	NC	79	D5	111	PMS
16	A0	48	HACK	80	D6	112	$\overline{\mathrm{DMS}}$
17	A1	49	HMD0	81	D7	113	BMS
18	A2	50	HMD1	82	D8	114	PWDACK
19	A3	51	DT0	83	D9	115	HD15
20	A4	52	TFS0	84	D10	116	HD14
21	A5	53	RFS0	85	GND	117	HD13
22	A6	54	DR0	86	$ m V_{DD}$	118	HD12
23	A7	55	SCLK0	87	D11	119	HD11
24	XTAL	56	DT1/FO	88	D12	120	$V_{ m DD}$
25	CLKIN	57	TFS1/IRQ1	89	D13	121	HD10
26	CLKOUT	58	RFS1/IRQ0	90	D14	122	HD9
27	GND	59	GND	91	D15	123	HD8
28	A8	60	DR1/F1	92	D16	124	HD7
29	A9	61	SCLK1	93	D17	125	HD6
30	A10	62	FL0	94	D18	126	HSIZE
31	A11	63	FL1	95	GND	127	HRD/HRW
32	A12	64	FL2	96	D19	128	HWR/HDS

NC = These pins MUST remain unconnected.

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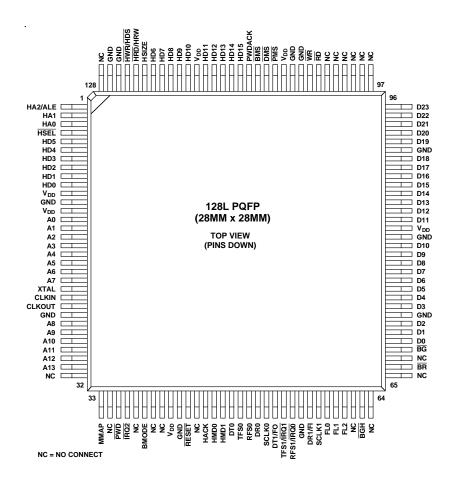
OUTLINE DIMENSIONS 128-Lead Metric Thin Plastic Quad Flatpack (TQFP)



	MILLIMETERS			INCHES		
SYMBOL	MIN	TYP	MAX	MIN	TYP	MAX
Α			1.60		0.063	
A ₁	0.05		0.15	0.002		0.006
A ₂	1.30	1.40	1.50	0.051	0.055	0.059
D	15.75	16.00	16.25	0.620	0.630	0.640
D ₁	13.90	14.00	14.10	0.547	0.551	0.555
D_3		12.50	12.58		0.492	0.495
E	21.75	22.00	22.25	0.856	0.866	0.876
E ₁	19.90	20.00	20.10	0.783	0.787	0.792
E ₃		18.50	18.58		0.728	0.731
L	0.45	0.60	0.75	0.018	0.024	0.030
е	0.42	0.50	0.58	0.017	0.019	0.023
В	0.17	0.22	0.27	0.007	0.009	0.011
۵			0.10			0.004

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128-Lead PQFP Package Pinout



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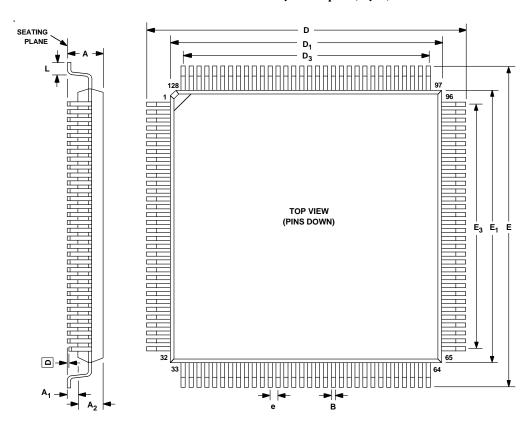
PQFP Pin Configurations

PQFP Number	Pin Name	PQFP Number	Pin Name	PQFP Number	Pin Name	PQFP Number	Pin Name
1	HA2/ALE	33	MMAP	65	NC	97	NC
2	HA1	34	NC NC	66	$\frac{NC}{BR}$	98	NC NC
3		35	PWD			98	NC NC
3	HA0 HSEL			67	$\frac{NC}{BG}$	100	
4		36	IRQ2	68			NC NC
5	HD5	37	NC	69	D0	101	NC
6	HD4	38	BMODE	70	D1	102	NC
7	HD3	39	NC	71	D2	103	RD
8	HD2	40	NC	72	GND	104	WR
9	HD1	41	$V_{ m DD}$	73	D3	105	GND
10	HD0	42	GND	74	D4	106	GND
11	$V_{ m DD}$	43	RESET	75	D5	107	$V_{\rm DD}$
12	GND	44	NC	76	D6	108	PMS
13	$V_{ m DD}$	45	HACK	77	D7	109	DMS
14	A0	46	HMD0	78	D8	110	BMS
15	A1	47	HMD1	79	D9	111	PWDACK
16	A2	48	DT0	80	D10	112	HD15
17	A3	49	TFS0	81	GND	113	HD14
18	A4	50	RFS0	82	$V_{ m DD}$	114	HD13
19	A5	51	DR0	83	D11	115	HD12
20	A6	52	SCLK0	84	D12	116	HD11
21	A7	53	DT1/FO	85	D13	117	$V_{ m DD}$
22	XTAL	54	TFS1/IRQ1	86	D14	118	HD10
23	CLKIN	55	RFS1/IRQ0	87	D15	119	HD9
24	CLKOUT	56	GND	88	D16	120	HD8
25	GND	57	DR1/F1	89	D17	121	HD7
26	A8	58	SCLK1	90	D18	122	HD6
27	A9	59	FL0	91	GND	123	HSIZE
28	A10	60	FL1	92	D19	124	HRD/HRW
29	A10 A11	61	FL2	93	D20	125	HWR/HDS
30	A11 A12	62	NC	94	D20	126	GND
31	A12 A13	63	BGH	95	D21 D22	127	GND
32	NC	64	NC	96	D22	127	NC NC
32	INC	04	INC	90	Das	120	NC

 $NC = These \ pins \ MUST \ remain \ unconnected.$

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OUTLINE DIMENSIONS 128-Lead Metric Thin Plastic Quad Flatpack (PQFP)



	MILLIMETERS			INCHES		
SYMBOL	MIN	TYP	MAX	MIN	TYP	MAX
Α			4.07			0.160
A ₁	0.25			0.010		
A ₂	3.17	3.49	3.67	0.125	0.137	0.144
D, E	30.95	31.20	31.45	1.219	1.228	1.238
D ₁ , E ₁	27.90	28.00	28.10	1.098	1.102	1.106
D ₃ , E ₃	24.73	24.80	24.87	0.974	0.976	0.979
L	0.65	0.88	1.03	0.031	0.035	0.041
е	0.73	0.80	0.87	0.029	0.031	0.034
В	0.30	0.35	0.45	0.012	0.014	0.018
Ω			0.10			0.004

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ORDERING GUIDE*

Part Number**	Ambient Temperature Range	Instruction Rate (MHz)	Package Description
ADSP-2171KST-133	0°C to +70°C	33.33	128-Lead TQFP
ADSP-2171BST-133	-40°C to +85°C	33.33	128-Lead TQFP
ADSP-2171KS-133	0°C to +70°C	33.33	128-Lead PQFP
ADSP-2171BS-133	-40°C to +85°C	33.33	128-Lead PQFP
ADSP-2171KST-104	0°C to +70°C	26	128-Lead TQFP
ADSP-2171BST-104	-40°C to +85°C	26	128-Lead TQFP
ADSP-2171KS-104	0°C to +70°C	26	128-Lead PQFP
ADSP-2171BS-104	-40°C to +85°C	26	128-Lead PQFP
ADSP-2173BST-80	-40°C to +85°C	20	128-Lead TQFP
ADSP-2173BS-80	-40°C to +85°C	20	128 Lead PQFP

^{*}Refer to section titled "Ordering Procedure for ADSP-2172 ROM Processors" for information about ordering ROM-coded parts.

**S = Plastic Quad Flatpack, ST = Plastic Thin Quad Flatpack.

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