

STGIPS30C60

Datasheet - preliminary data

SLLIMM[™] small low-loss intelligent molded module IPM, 3-phase inverter - 30 A, 600 V short-circuit rugged IGBT

SDIP-25L

Features

- IPM 30 A, 600 V 3-phase IGBT inverter bridge including control ICs for gate driving and freewheeling diodes
- Short-circuit rugged IGBTs
- 3.3 V, 5 V, 15 V CMOS/TTL inputs comparators with hysteresis and pull down / pull up resistors
- Undervoltage lockout
- Internal bootstrap diode
- Interlocking function
- Smart shutdown function
- Comparator for fault protection against over temperature and overcurrent
- DBC leading to low thermal resistance
- Isolation rating of 2500 V_{rms}/min
- UL recognized: UL1557 file E81734

Applications

- 3-phase inverters for motor drives
- Air conditioners

Description

This intelligent power module provides a compact, high performance AC motor drive in a simple, rugged design. Combining ST proprietary control ICs with the most advanced short-circuit-rugged IGBT system technology, this device is ideal for 3-phase inverters in applications such as motor drives and air conditioners. SLLIMM[™] is a trademark of STMicroelectronics.

Table 1. Device summary

Order code	Marking	Package	Packaging
STGIPS30C60	GIPS30C60	SDIP-25L	Tube

DocID022471 Rev 3

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This is preliminary information on a new product now in development or undergoing evaluation. Details are subject to change without notice.

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1 Internal block diagram and pin configuration

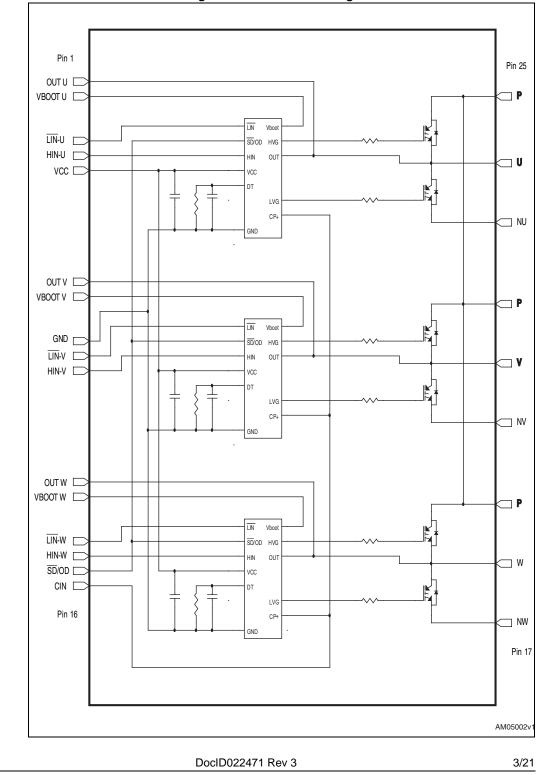


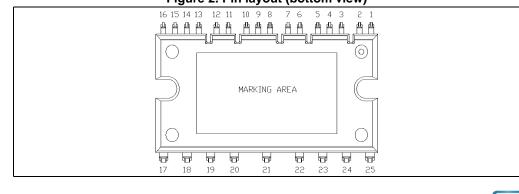
Figure 1. Internal block diagram

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Pin n°	Symbol	Description
1	OUT _U	High-side reference output for U phase
2	V _{bootU}	Bootstrap voltage for U phase
3	LINU	Low-side logic input for U phase
4	HINU	High-side logic input for U phase
5	V _{CC}	Low voltage power supply
6	OUT _V	High-side reference output for V phase
7	V _{boot V}	Bootstrap voltage for V phase
8	GND	Ground
9	LINV	Low-side logic input for V phase
10	HIN_V	High-side logic input for V phase
11	OUT _W	High-side reference output for W phase
12	V _{boot W}	Bootstrap voltage for W phase
13	LIN _W	Low-side logic input for W phase
14	HIN _W	High-side logic input for W phase
15	SD / OD	Shutdown logic input (active low) / open-drain (comparator output)
16	CIN	Comparator input
17	N _W	Negative DC input for W phase
18	W	W phase output
19	Р	Positive DC input
20	N _V	Negative DC input for V phase
21	V	V phase output
22	Р	Positive DC input
23	NU	Negative DC input for U phase
24	U	U phase output
25	Р	Positive DC input

Table 2. Pin description

Figure 2. Pin layout (bottom view)



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2 Electrical ratings

2.1 Absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{PN}	Supply voltage applied between P - $\rm N_{U}, \rm N_{V}, \rm N_{W}$	450	V
V _{PN(surge)}	Supply voltage (surge) applied between P - $\rm N_{U},$ $\rm N_{V},$ $\rm N_{W}$	500	V
V _{CES}	Each IGBT collector emitter voltage ($V_{IN}^{(1)} = 0$)	600	V
± I _C	Each IGBT continuous collector current at $T_{C} = 25^{\circ}C$	30	A
$\pm I_{CP}^{(2)}$	Each IGBT pulsed collector current	60	А
P _{TOT}	Each IGBT total dissipation at $T_{C} = 25^{\circ}C$	52	W
t _{scw}	Short circuit withstand time, $V_{CE} = 0.5 V_{(BR)CES}$ T _J = 125 °C, $V_{CC} = V_{boot} = 15 V$, $V_{IN (1)} = 0 - 5 V$	5	μs

Table 3. Inverter part

1. Applied between HIN_i , \overline{LIN}_i and G_{ND} for i = U, V, W

2. Pulse width limited by max junction temperature

Table 4. Control part

Symbol	Parameter	Value	Unit
V _{OUT}	Output voltage applied between OUT _{U,} OUT _V , OUT _W - GND	V_{boot} - 21 to V_{boot} + 0.3	V
V _{CC}	Low voltage power supply	-0.3 to +21	V
V _{CIN}	Comparator input voltage	-0.3 to V _{CC} +0.3	V
V _{boot}	Bootstrap voltage applied between $V_{boot i}$ - OUT _i for i = U, V, W	-0.3 to 620	V
V _{IN}	Logic input voltage applied between HIN, $\overline{\text{LIN}}$ and GND	-0.3 to 15	V
V _{SD/OD}	Open drain voltage	-0.3 to 15	V
dV _{OUT} /dt	Allowed output slew rate	50	V/ns

Table 5. Total system

Symbol	Parameter	Value	Unit
V _{ISO}	Isolation withstand voltage applied between each pin and heatsink plate (AC voltage, t = 60 sec.)	2500	V
Тj	Power chips operating junction temperature	-40 to 150	°C
Т _С	Module case operation temperature	-40 to 125	°C



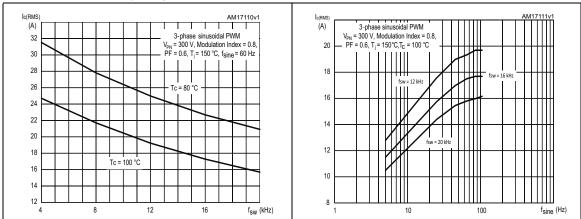
2.2 Thermal data

Symbol	Parameter	Value	Unit
R _{thJC}	Thermal resistance junction-case single IGBT	2.4	°C/W
	Thermal resistance junction-case single diode	5	°C/W

Table 6. Thermal data

Figure 3. Maximum I_{C(RMS)} current vs. switching frequency ⁽¹⁾

Figure 4. Maximum $|_{C(RMS)}$ current vs. $f_{sine}^{(1)}$



1. Simulated curves refer to typical IGBT parameters and maximum R_{thj-c.}



3 Electrical characteristics

 $T_J = 25$ °C unless otherwise specified.

Symbol	Baramatar	Test conditions	Value			Unit	
Symbol	Parameter	lest conditions	Min.	Min. Typ. Max.			
V	V _{CE(sat)} Collector-emitter saturation voltage	$V_{CC} = V_{boot} = 15 \text{ V}, V_{IN}^{(1)} = 0 \div 5 \text{ V},$ $I_{C} = 30 \text{ A}$	-	1.9		v	
Vor		$V_{CC} = V_{boot} = 15 \text{ V}, V_{IN}^{(1)} = 0 \div 5 \text{ V},$ $I_{C} = 30 \text{ A}, T_{J} = 125 \text{ °C}$	-	2.2		v	
I _{CES}	Collector-cut off current (V _{IN} ⁽¹⁾ = 0 "logic state")	$V_{CE} = 550$ V, $V_{CC} = V_{Boot} = 15$ V	-		100	μA	
V _F	Diode forward voltage	$V_{IN}^{(1)} = 0$ "logic state", $I_C = 30 \text{ A}$	-	2.0	2.3	V	
Inductive	load switching time and	energy			•	<u>.</u>	
t _{on}	Turn-on time		-	440	-		
t _{c(on)}	Crossover time (on)	V _{PN} = 300 V,	-	190	-		
t _{off}	Turn-off time	$V_{PN} = 500 \text{ V},$ $V_{CC} = V_{boot} = 15 \text{ V},$	-	780	-	ns	
t _{c(off)}	Crossover time (off)	$V_{IN}^{(1)} = 0 \div 5 V,$	-	135	-		
t _{rr}	Reverse recovery time	I _C = 30 A (see <i>Figure 5</i>)	-	100	-]	
Eon	Turn-on switching losses		-	870	-	1	
E _{off}	Turn-off switching losses		-	740	-	μJ	

Table 7. Inverter part

1. Applied between HIN_i, $\overline{\text{LIN}}_{i \text{ and }} G_{ND}$ for i = U, V, W. ($\overline{\text{LIN}}$ inputs are active-low).

Note: t_{ON} and t_{OFF} include the propagation delay time of the internal drive. $t_{C(ON)}$ and $t_{C(OFF)}$ are the switching time of IGBT itself under the internally given gate driving condition.



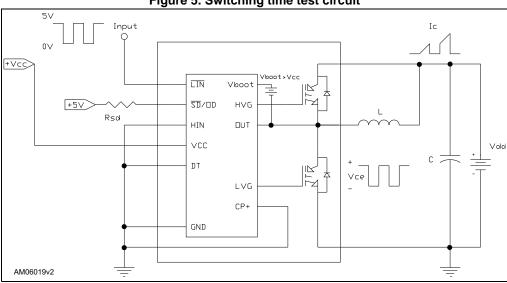


Figure 5. Switching time test circuit



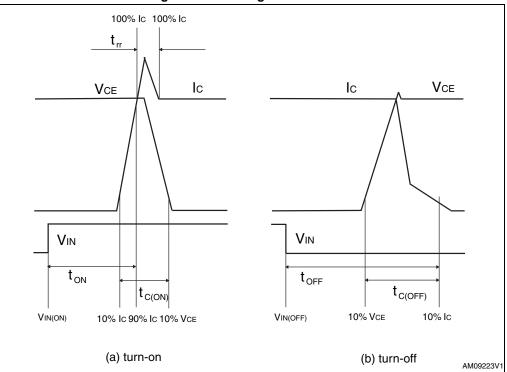


Figure 4 "Switching time definition" refers to HIN inputs (active high). For LIN inputs (active low), V_{IN} polarity must be inverted for turn-on and turn-off.



3.1 Control part

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{CC_hys}	V _{cc} UV hysteresis		1.2	1.5	1.8	V
V _{CC_thON}	V _{cc} UV turn ON threshold		11.5	12	12.5	V
$V_{\text{CC_thOFF}}$	V_{cc} UV turn OFF threshold		10	10.5	11	V
I _{qccu}	Undervoltage quiescent supply current				450	μA
I _{qcc}	Quiescent current				3.5	mA
V _{ref}	Internal comparator (CIN) reference voltage		0.5	0.54	0.58	V

Table 9. Bootstrapped voltage (V_{CC} = 15 V unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{BS_hys}	V _{BS} UV hysteresis		1.2	1.5	1.8	V
V _{BS_thON}	V _{BS} UV turn ON threshold		11.1	11.5	12.1	V
$V_{BS_{thOFF}}$	V _{BS} UV turn OFF threshold		9.8	10	10.6	V
I _{QBSU}	Undervoltage V _{BS} quiescent current	$V_{BS} < 9 V$ SD/OD = 5 V; LIN and HIN = 5 V; C _{IN} = 0		70	110	μA
I _{QBS}	V _{BS} quiescent current	$\frac{V_{BS} = 15 \text{ V}}{\text{SD/OD} = 5 \text{ V}; \text{ LIN and}}$ HIN = 5 V; C _{IN} = 0		200	300	μA
R _{DS(on)}	Bootstrap driver on resistance	LVG ON		120		W

Table 10. Logic inputs (V_{CC} = 15 V unless otherwise specified)

Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit
V _{il}	Low logic level voltage		0.8		1.1	V
V _{ih}	High logic level voltage		1.9		2.25	V
I _{HINh}	HIN logic "1" input bias current	HIN = 15 V	110	175	260	μA
I _{HINI}	HIN logic "0" input bias current	HIN = 0 V			1	μA
I _{LINI}	LIN logic "1" input bias current	$\overline{\text{LIN}} = 0 \text{ V}$	3	6	20	μA
I _{LINh}	LIN logic "0" input bias current	LIN = 15 V			1	μA
I _{SDh}	SD logic "0" input bias current	<u>SD</u> = 15 V	30	120	300	μA
I _{SDI}	SD logic "1" input bias current	$\overline{SD} = 0 V$			3	μA
Dt	Dead time	see Figure 7		600		ns

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Symbol	Parameter	Test conditions	Min.	Тур.	Max.	Unit	
l _{ib(i)}	Input bias current	V _{CIN(i)} = 1 V, i = U, V o W			3	μA	
V _{ol}	Open-drain low-level output voltage	I _{od} = 3 mA			0.5	V	
t _{d_comp}	Comparator delay	\overline{SD} /OD pulled to 5 V through 100 k Ω resistor		90	130	ns	
SR	Slew rate	C_{L} = 180 pF; R_{pu} = 5 kΩ		60		V/µsec	
t _{sd}	Shut down to high / low side driver propagation delay		50	125	200		
t _{isd}	Comparator triggering to high / low side driver turn-off propagation delay	Measured applying a voltage step from 0 V to 3.3 V to pin CIN _i	50	200	250	ns	

Table 12. Truth table

Condition	Logic input (V _I)			Output		
Condition	SD/OD	LIN	HIN	LVG	HVG	
Shutdown enable half-bridge tri-state	L	х	х	L	L	
Interlocking half-bridge tri-state	н	L	н	L	L	
0 ''logic state" half-bridge tri-state	н	н	L	L	L	
1 "logic state" low side direct driving	н	L	L	н	L	
1 "logic state" high side direct driving	н	н	н	L	н	

Note: X: don't care

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3.2 Waveforms definitions

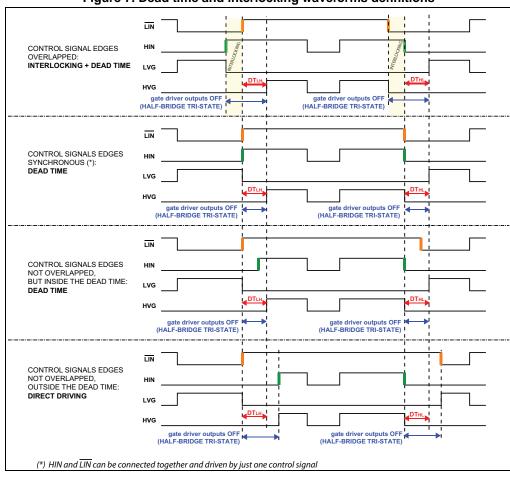


Figure 7. Dead time and interlocking waveforms definitions



4 Smart shutdown function

The STGIPS30C60 integrates a comparator for fault sensing purposes. The comparator has an internal voltage reference V_{ref} connected to the inverting input, while the non-inverting input, available on pin (C_{IN}), can be connected to an external shunt resistor in order to implement a simple over-current protection function. When the comparator triggers, the device is set in shutdown state and both its outputs are set to low-level leading the halfbridge in tri-state. In the common overcurrent protection architectures the comparator output is usually connected to the shutdown input through a RC network, in order to provide a mono-stable circuit, which implements a protection time that follows the fault condition. Our smart shutdown architecture allows to immediately turn-off the output gate driver in case of overcurrent, the fault signal has a preferential path which directly switches off the outputs. The time delay between the fault and the outputs turn-off is no more dependent on the RC values of the external network connected to the shutdown pin. At the same time the DMOS connected to the open-drain output (pin SD/OD) is turned on by the internal logic which holds it on until the shutdown voltage is lower than the logic input lower threshold (V_{il}). Finally the smart shutdown function provides the possibility to increase the real disable time without increasing the constant time of the external RC network.

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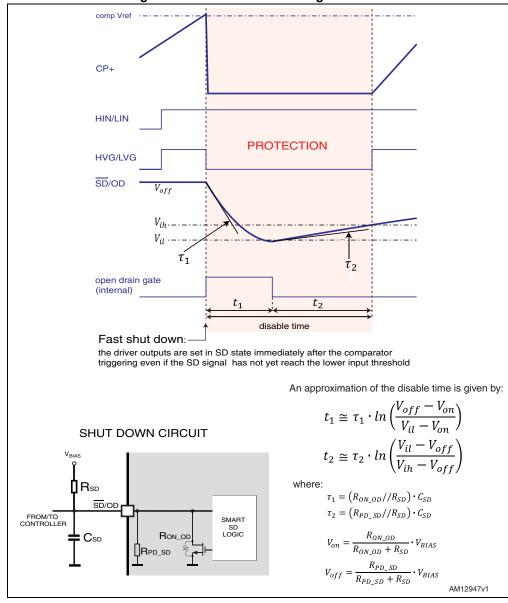
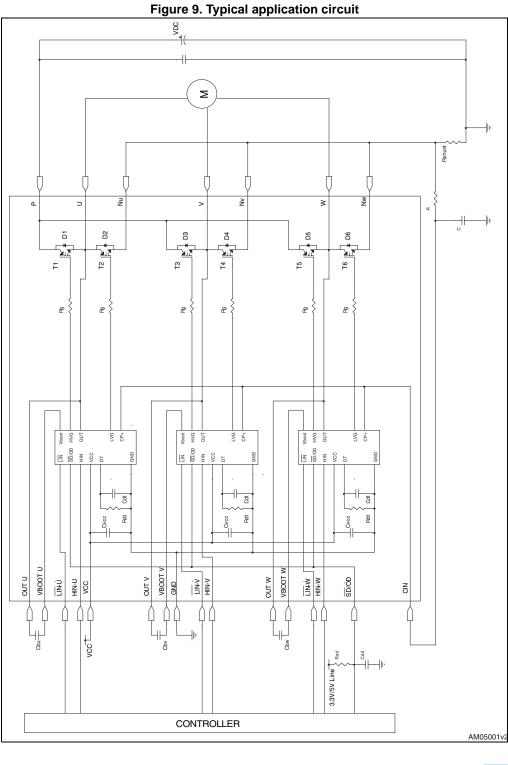


Figure 8. Smart shutdown timing waveforms

Pls refer to Table 11 for internal propagation delay time details.



5 Applications information





5.1 Recommendations

- Input signal HIN is active high logic. A 85 kΩ (typ.) pull down resistor is built-in for each high side input. If an external RC filter is used, for noise immunity, pay attention to the variation of the input signal level.
- Input signal LIN is active low logic. A 720 kΩ (typ.) pull-up resistor, connected to an internal 5 V regulator through a diode, is built-in for each low side input.
- To prevent the input signals oscillation, the wiring of each input should be as short as possible.
- By integrating an application specific type HVIC inside the module, direct coupling to MCU terminals without any opto-coupler is possible.
- Each capacitor should be located as nearby the pins of IPM as possible.
- Low inductance shunt resistors should be used for phase leg current sensing.
- Electrolytic bus capacitors should be mounted as close to the module bus terminals as possible. Additional high frequency ceramic capacitor mounted close to the module pins will further improve performance.
- The SD/OD signal should be pulled up to 5 V / 3.3 V with an external resistor (see *Section 4: Smart shutdown function* for detailed info).

Symbol	Parameter	Conditions	Value			Unit
	Farameter	Conditions	Min.	Тур.	Max.	Unit
V _{PN}	Supply Voltage	Applied between P-Nu,Nv,Nw		300	400	V
V _{CC}	Control supply voltage	Applied between V _{CC} -GND	13.5	15	18	V
V _{BS}	High side bias voltage	Applied between V_{BOOTi} -OUT _i for i=U,V,W	13		18	V
t _{dead}	Blanking time to prevent Arm-short	For each input signal	1			μs
f _{PWM}	PWM input signal	-40°C < T _c < 100°C -40°C < T _j < 125°C			20	kHz
т _с	Case operation temperature				100	°C

Table 13. Recommended operating conditions

Note: For further details refer to AN3338.

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6 Package mechanical data

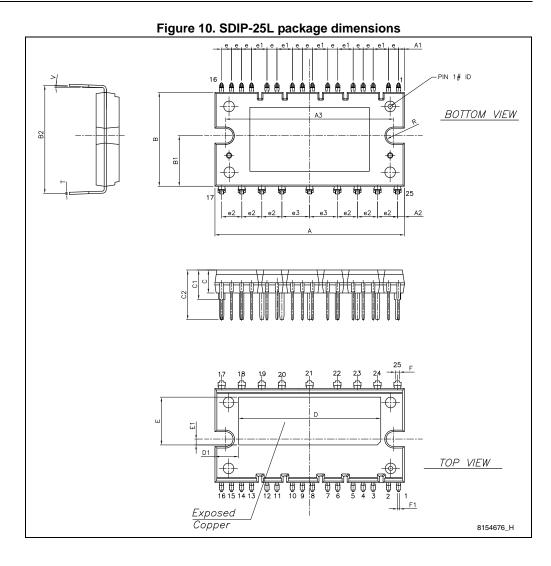
In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: *www.st.com*. ECOPACK[®] is an ST trademark.

Please refer to dedicated technical note TN0107 for mounting instructions.

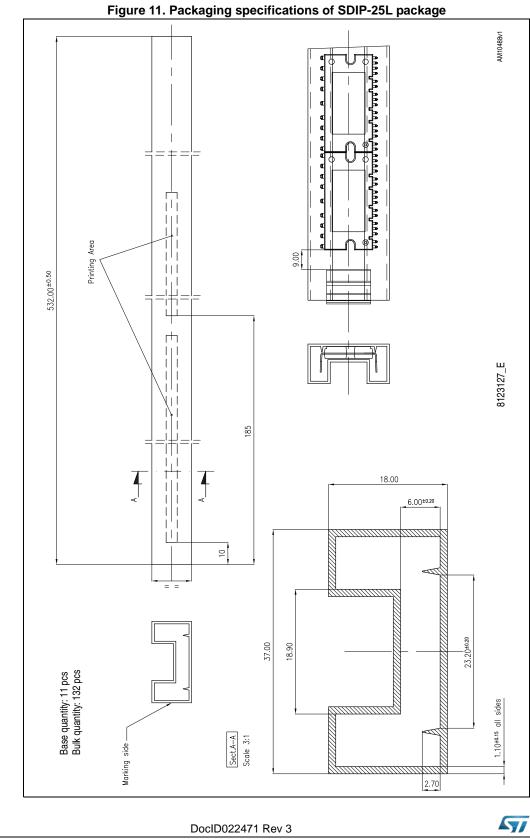
Dim.		(mm.)	
	Min.	Тур.	Max.
А	43.90	44.40	44.90
A1	1.15	1.35	1.55
A2	1.40	1.60	1.80
A3	38.90	39.40	39.90
В	21.50	22.00	22.50
B1	11.25	11.85	12.45
B2	24.83	25.23	25.63
С	5.00	5.40	6.00
C1	6.50	7.00	7.50
C2	11.20	11.70	12.20
e	2.15	2.35	2.55
e1	3.40	3.60	3.80
e2	4.50	4.70	4.90
e3	6.30	6.50	6.70
D		33.30	
D1		5.55	
E		11.20	
E1		1.40	
F	0.85	1.00	1.15
F1	0.35	0.50	0.65
R	1.55	1.75	1.95
Т	0.45	0.55	0.65
V	0°		6°

Table 14. SDIP-25L package mechanical data









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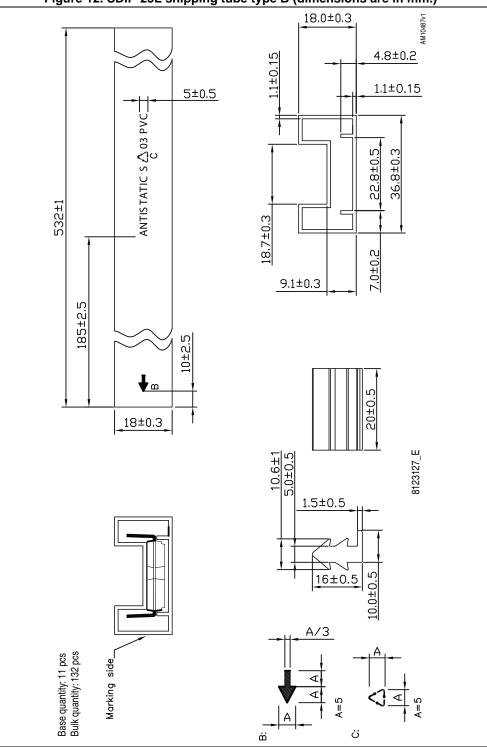


Figure 12. SDIP-25L shipping tube type B (dimensions are in mm.)



7 Revision history

Date	Revision	Changes
07-Dec-2011	1	Initial release
04-Dec-2012	2	Added: P _{TOT} value <i>Table 3 on page 5</i> , R _{thJC} values <i>Table 6 on page 6</i> , typical values <i>Table 7 on page 7</i> and <i>Figure 12 on page 19</i> . Removed: note 2 <i>Table 3 on page 5</i> .
02-Apr-2013	3	Added <i>Figure 3</i> and <i>Figure 4 on page 6.</i> Updated: <i>Figure 8 on page 13.</i>

Table 15. Document revision history

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