

**LB1836M**

Low-saturation, Bidirectional Motor Driver for Low-voltage Applications

Overview

The LB1836M is a low-saturation two-channel bidirectional motor driver IC for use in low-voltage applications. The LB1836M is a bipolar stepper-motor driver IC that is ideal for use in printers, FDDs, cameras and other portable devices.

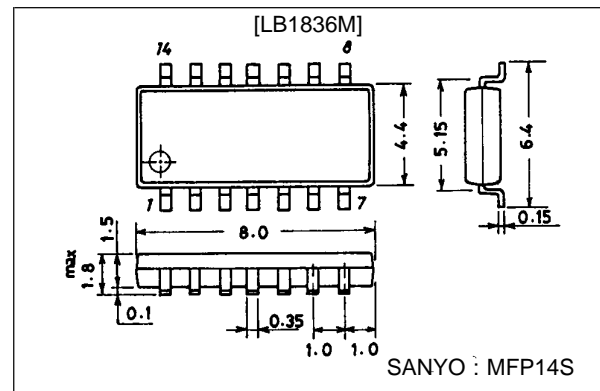
Features

- Low voltage operation (2.5 V min)
- Low saturation voltage (upper transistor + lower transistor residual voltage; 0.40 V typ at 400 mA).
- Parallel connection
(Upper transistor + lower transistor residual voltage; 0.5 V typ at 800 mA).
- Separate logic power supply and motor power supply
- Brake function
- Spark killer diodes built in
- Thermal shutdown circuit built in
- Compact package (14-pin MFP)

Package Dimensions

unit : mm

3111-MFP14S



Specifications

Absolute Maximum Ratings at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V _{CC} max		-0.3 to +10.5	V
	V _S max		-0.3 to +10.5	V
Output applied voltage	V _{OUT}		V _S + V _{SF}	V
Input applied voltage	V _{IN}		-0.3 to +10	V
Ground pin flow-out current	IGND	Per channel	1.0	A
Allowable power dissipation	Pd max	* With board	800	mW
Operating temperature	Topr		-20 to +75	°C
Storage temperature	Tstg		-40 to +125	°C

*Note: Mounted on 30 × 30 × 1.5 mm³ glass epoxy PCB

Allowable Operating Ranges at Ta = 25°C

Parameter	Symbol	Conditions	Ratings	Unit
Supply voltage	V _{CC}		2.5 to 9.0	V
	V _S		1.8 to 9.0	V
Input high-level voltage	V _{IH}		1.8 to 9.0	V
Input low-level voltage	V _{IL}		-0.3 to +0.7	V

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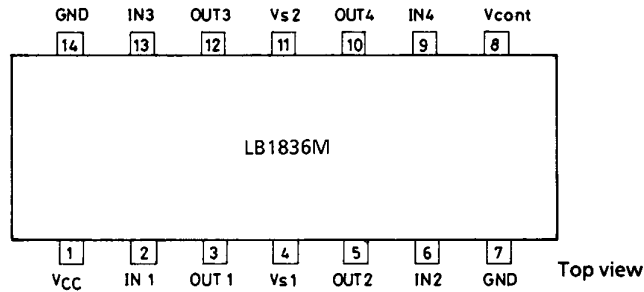
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LB1836M

Electrical Characteristics at $T_a = 25^\circ\text{C}$, $V_{CC} = V_S = 3\text{ V}$

Parameter	Symbol	Conditions	min	typ	max	Unit
Supply current	I_{CC0}	$V_{IN1, 2, 3, 4} = 0\text{ V}$, $I_{CC} + I_S$		0.1	10	μA
	I_{CC1}	$V_{IN1} = 3\text{ V}$, $V_{IN2, 3, 4} = 0\text{ V}$, $I_{CC} + I_S$		14	20	mA
	I_{CC2}	$V_{IN1, 2} = 3\text{ V}$, $V_{IN3, 4} = 0\text{ V}$, $I_{CC} + I_S$		22	35	mA
Output saturation voltage (upper + lower)	V_{OUT1}	$I_{OUT} = 200\text{ mA}$		0.2	0.28	V
	V_{OUT2}	$I_{OUT} = 400\text{ mA}$		0.4	0.6	V
	V_{OUT3}	$I_{OUT} = 400\text{ mA}$, parallel connection		0.25	0.35	V
	V_{OUT4}	$I_{OUT} = 800\text{ mA}$, parallel connection		0.5	0.7	V
Output sustaining voltage	V_O (sus)	$I_{OUT} = 400\text{ mA}$	9			V
Input current	I_{IN}	$V_{IN} = 2\text{ V}$, $V_{CC} = 6\text{ V}$			80	μA
Spark killer diode reverse current	I_S (leak)	$V_{CC1, 2} = 9\text{ V}$			30	μA
Spark killer diode forward voltage	V_{SF}	$I_{OUT} = 400\text{ mA}$			1.7	V

Pin Assignment



Note) Both GNDs must be connected. P-GND of OUT2 and OUT4 and S-GND of the control section are connected to the pin 7 GND within the IC, and P-GND of OUT1 and OUT3 is connected to the pin 14 GND.

Truth Table

IN 1, 3	IN 2, 4	OUT 1, 3	OUT 2, 4	Mode
H	L	H	L	Forward
L	H	L	H	Reverse
H	H	L	L	Brake
L	L	OFF	OFF	Standby

Design Notes

If large current flows on the power supply (V_S) line and the GND line, then in some applications and layouts, misoperation due to line oscillation may result.

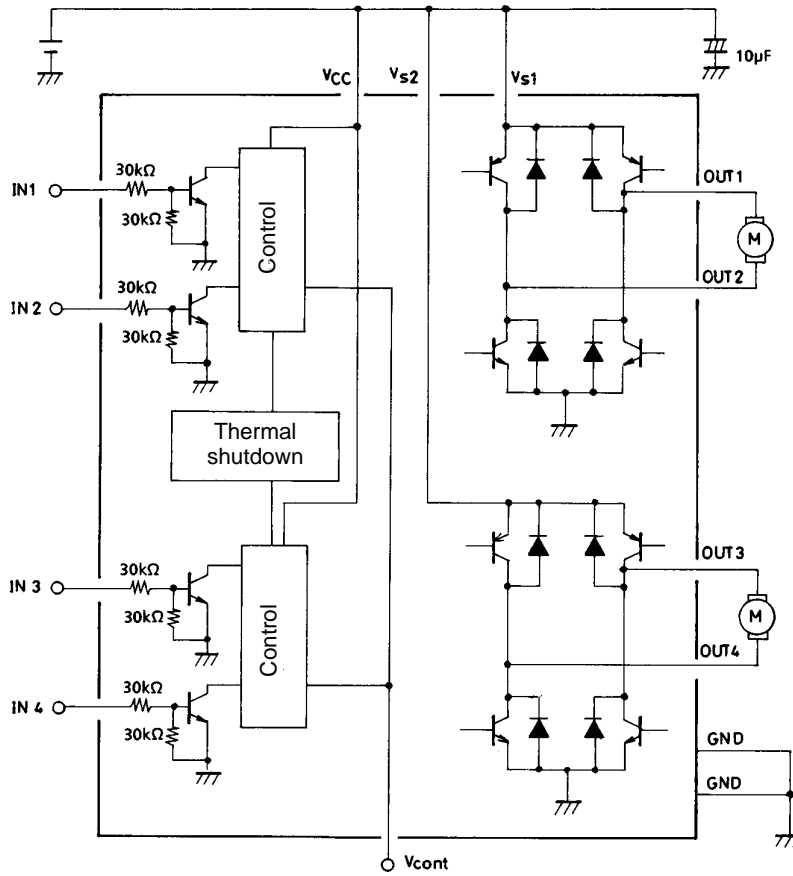
The modes during which large current flows are as follows:

- Motor surge current when the DC motor starts up or when it shifts rotation directions (forward \rightleftharpoons reverse).
- Passthrough current generated within the IC when shifting rotation directions (forward \rightleftharpoons reverse) or when shifting from forward/reverse rotation to braking, or vice versa.

The following points should be kept in mind regarding the pattern layout:

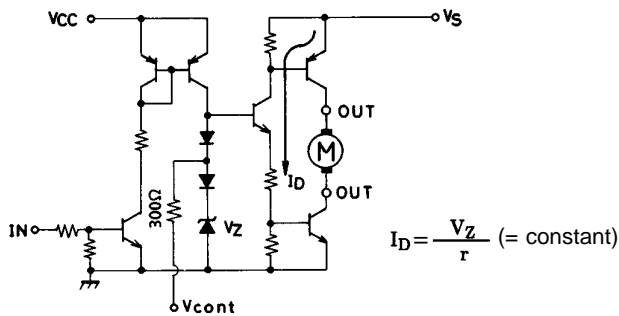
- Keep the wiring lines thick and short in order to reduce wiring inductance between the power supply (V_S) and GND.
- Insert a passthrough capacitor near the IC. (Maximum effect is obtained by inserting the passthrough capacitor between V_S and the pin 7 GND at the closest distance possible.
- If the CPU and the LB1836M are mounted on separate boards and the difference between the ground potential of each board is large, install resistors of about $10\text{ k}\Omega$ in series between the CPU and the LB1836M inputs.

Block Diagram



Note: As long as the voltages applied to V_{CC} , V_{S1} , V_{S2} , and $IN1$ through $IN4$ are within the limits set by the absolute maximum ratings, there are no restrictions on the relationship of each voltage level in comparison with the others (regarding which is higher or lower). (ex. $V_{CC} = 3\text{ V}$, $V_{S1, 2} = 2\text{ V}$, $IN1$ to $IN4 = 5\text{ V}$)

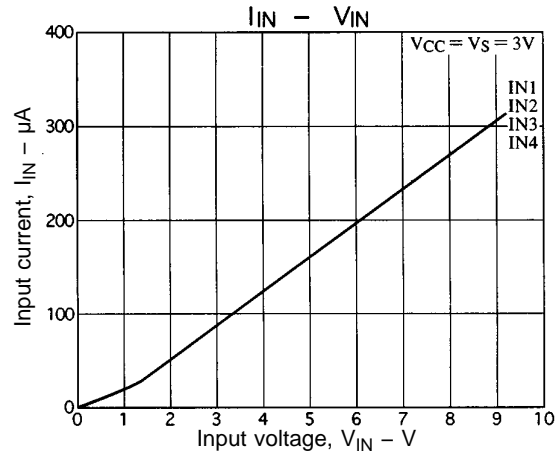
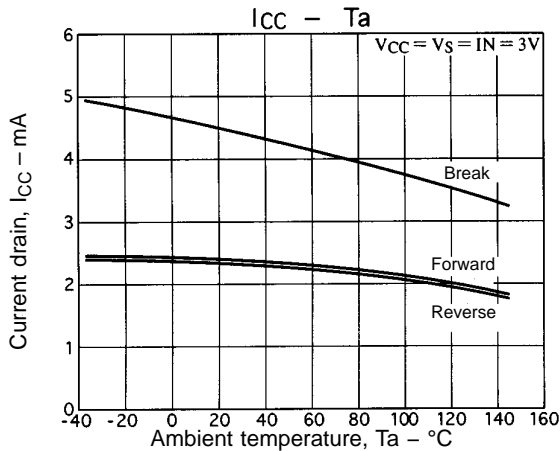
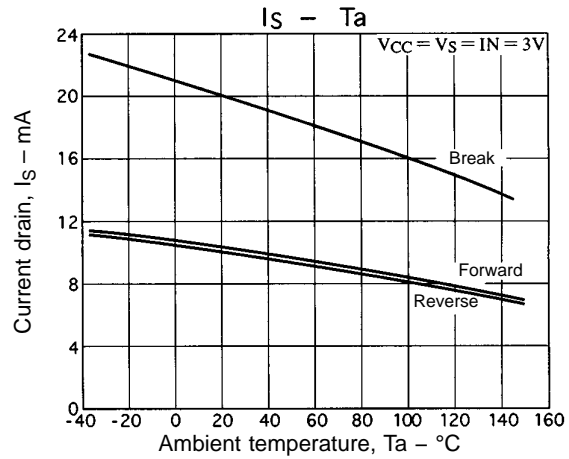
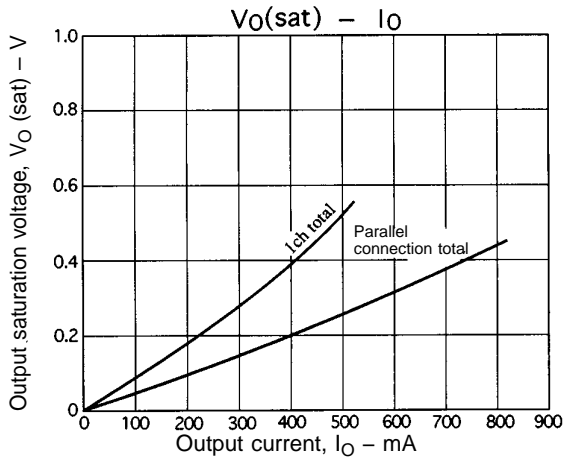
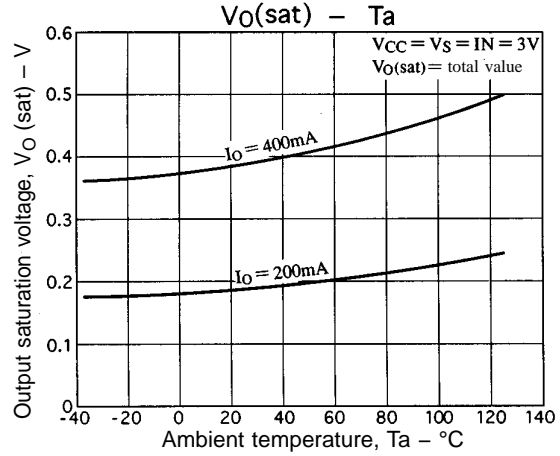
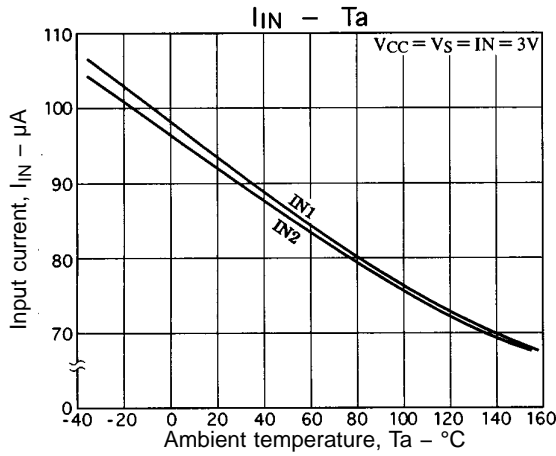
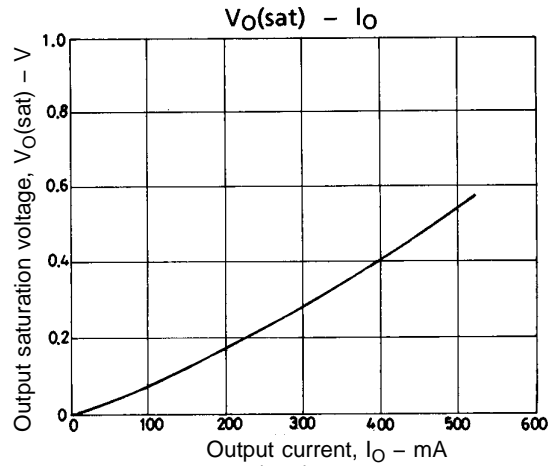
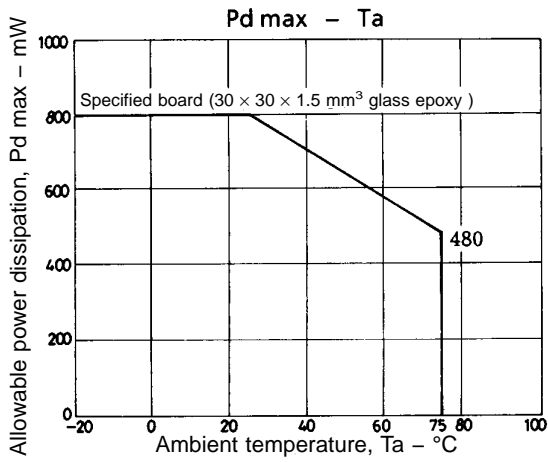
Vcont pin

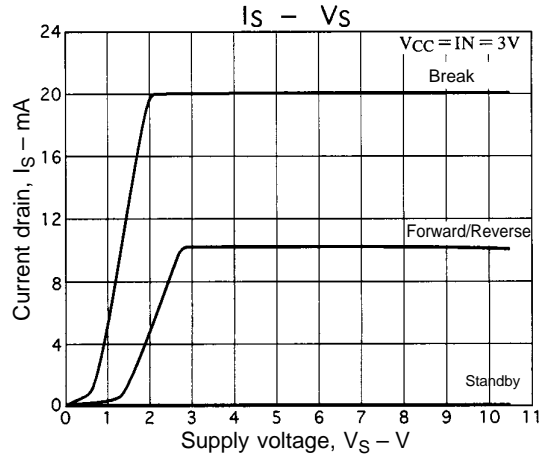
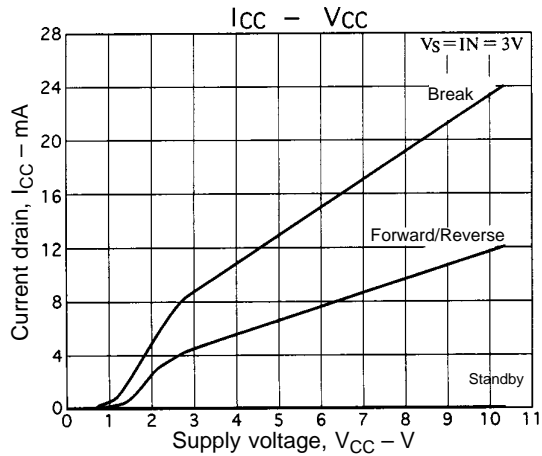


As shown in the above diagram, the V_{cont} pin outputs the voltage of the band gap Zener $V_Z + V_F (= 1.93\text{ V})$.

In normal use, this pin is left open.

The drive current I_D is varied by the V_{cont} voltage. However, because the band gap Zener is shared, it functions as a bridge.





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