

## FEATURES

- 16-bit resolution
- 500kHz sampling rate
- Functionally complete
- Excellent dynamic performance
- 83dB SNR, - 89 dB THD
- No missing codes
- Small, 40-pin, TDIP package
- 3.5 Watts power dissipation
- On-board FIFO


## PRODUCT OVERVIEW

The low-cost ADS-930 is a high-performance, 16 -bit, 500 kHz sampling A/D converter. This device accurately samples fullscale input signals up to Nyquist frequencies with no missing codes. The dynamic performance of the ADS-930 is optimized to achieve a THD of -89 dB and an SNR of 83dB.

Packaged in a small, 40-pin, ceramic TDIP, the functionally complete ADS-930 contains a fast-settling sample-hold amplifier, a subranging (three-pass) A/D converter, an internal reference, an on-board FIFO, timing and control logic, three-state outputs and errorcorrection circuitry. Digital inputs/ outputs are TTL.

Requiring $\pm 15 \mathrm{~V}$ and +5 V supplies, the ADS-930 typically dissipates 3.5 Watts. The unit is offered with a bipolar input range of $\pm 5 \mathrm{~V}$ or a unipolar input range of 0 to -10 V . Models are available for use in either commercial ( 0 to $+70^{\circ} \mathrm{C}$ ) or HI-REL ( -55 to $+125^{\circ} \mathrm{C}$ ) operating temperature ranges. Typical applications include radar, sonar, medical/graphic imaging, and FFT spectrum analysis.

| INPUT/OUTPUT CONNECTIONS |  |  |  |
| :---: | :---: | :---: | :---: |
| PIN | FUNCTION | PIN | FUNCTION |
| 1 | +10V REF. OUT | 40 | BIT 1 (MSB) |
| 2 | BIPOLAR | 39 | BIT 1 (MSB) |
| 3 | ANALOG INPUT | 38 | BIT 2 |
| 4 | ANALOG GROUND | 37 | BIT 3 |
| 5 | OFFSET ADJUST | 36 | BIT 4 |
| 6 | GAIN ADJUST | 35 | BIT 5 |
| 7 | +15V SUPPLY | 34 | BIT 6 |
| 8 | COMP. BITS | 33 | BIT 7 |
| 9 | ENABLE | 32 | BIT 8 |
| 10 | FIFO READ | 31 | BIT 9 |
| 11 | ANALOG GROUND | 30 | ANALOG GROUND |
| 12 | -15V SUPPLY | 29 | BIT 10 |
| 13 | ANALOG GROUND | 28 | BIT 11 |
| 14 | OVERFLOW | 27 | BIT 12 |
| 15 | EOC | 26 | BIT 13 |
| 16 | +5V SUPPLY | 25 | BIT 14 |
| 17 | START CONVERT | 24 | DIGITAL GROUND |
| 18 | DIGITAL GROUND | 23 | FIFO/DIR |
| 19 | FSTAT1 | 22 | BIT 15 |
| 20 | FSTAT2 | 21 | BIT 16 (LSB) |

## BLOCK DIAGRAM



Figure 1. ADS-930 Functional Block Diagram

| ABSULUIL MAXIMUIM RAIINGS |  |  | YHYSILAL/ENVIRUNMIENIAL |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| PARAMETERS | LIMITS | UNITS | PARAMETERS | MIN. | TYP. | MAX. | UNITS |
| +15V Supply (Pin 7) | 0 to +16 | Volts | Operating Temp. Range, Case |  |  |  |  |
| -15V Supply (Pin 12) | 0 to -16 | Volts | ADS-930MC | 0 | - | +70 | ${ }^{\circ} \mathrm{C}$ |
| +5V Supply (Pin 16) | 0 to +6 | Volts | ADS-930MM | -55 | - | +125 | ${ }^{\circ} \mathrm{C}$ |
| Digital Inputs (Pin 8, 9, 10, 17, 23) | -0.3 to + Vdd +0.3 | Volts | Thermal Impedance |  |  |  |  |
| Analog Input (Pin 3) |  |  | $\theta \mathrm{jc}$ | - | 4 | - | ${ }^{\circ} \mathrm{C} /$ Watt |
| Unipolar | -12.5 to +12.5 | Volts | $\theta$ ca | - | 18 | - | ${ }^{\circ} \mathrm{C} /$ Watt |
| Bipolar <br> Lead Temperature (10 seconds) | -7.5 to +12.5 +300 | Volts | Storage Temperature Range | -65 | - | +150 | ${ }^{\circ} \mathrm{C}$ |
| Lead Temperature (10 seconds) | +300 | ${ }^{\circ} \mathrm{C}$ | Package Type | 40-pin, metal-sealed, ceramic TDIP |  |  |  |
|  |  |  | Weight | 0.56 ounces (16 grams) |  |  |  |

FUNCTIONAL SPECIFICATIONS
( $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}, \pm \mathrm{VCC}= \pm 15 \mathrm{~V}+\mathrm{VDD}=+5 \mathrm{~V}, 500 \mathrm{kHz}$ sampling rate, and a minimum 5 minute warmup (1) unless otherwise specified.)

| ANALOG INPUTS | $+25^{\circ} \mathrm{C}$ |  |  | $0 \mathrm{TO}+70^{\circ} \mathrm{C}$ |  |  | $-55 \mathrm{TO}+125^{\circ} \mathrm{C}$ |  |  | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |  |
| Input Voltage Range |  |  |  |  |  |  |  |  |  |  |
| Bipolar | - | $\pm 5$ | - | - | $\pm 5$ | - | - | $\pm 5$ | - | Volts |
| Unipolar | - | 0 to -10 | - | - | 0 to -10 | - | - | 0 to -10 | - | Volts |
| Input Resistance | 1.4 | 1.5 | 1.7 | 1.4 | 1.5 | 1.7 | 1.4 | 1.5 | 1.7 | k $\Omega$ |
| Input Capacitance | - | 7 | 15 | - | 7 | 15 | - | 7 | 15 | pF |
| DIGITAL INPUTS |  |  |  |  |  |  |  |  |  |  |
| Logic Levels |  |  |  |  |  |  |  |  |  |  |
| Logic "1" | +2.0 | - | - | +2.0 | - | - | +2.0 | - | - | Volts |
| Logic "0" | - | - | +0.8 | - | - | +0.8 | - | - | +0.8 | Volts |
| Logic Loading "1" | - | - | +20 | - | - | +20 | - | - | +20 | $\mu \mathrm{A}$ |
| Logic Loading "0" ② | - | - | -20 | - | - | -20 | - | - | -20 | $\mu \mathrm{A}$ |
| Start Convert Positive Pulse Width (3) | 175 | 200 | 215 | 175 | 200 | 215 | 175 | 200 | 215 | ns |
| STATIC PERFORMANCE |  |  |  |  |  |  |  |  |  |  |
| Resolution | - | 16 | - | - | 16 | - | - | 16 | - | Bits |
| Integral Nonlinearity (fin $=10 \mathrm{kHz}$ ) | - | $\pm 1.0$ | - | - | $\pm 1.5$ | - | - | $\pm 2.0$ | - | LSB |
| Differential Nonlinearity ( $\mathrm{fin}=10 \mathrm{kHz}$ ) | - | $\pm 0.75$ | - | - | $\pm 1.0$ | - | - | $\pm 1.5$ | - | LSB |
| Full Scale Absolute Accuracy | - | $\pm 0.05$ | $\pm 0.18$ | - | $\pm 0.2$ | $\pm 0.5$ | - | $\pm 0.5$ | $\pm 0.8$ | \%FSR |
| Unipolar Zero Error (Tech Note 2) | - | $\pm 0.05$ | $\pm 0.085$ | - | $\pm 0.1$ | $\pm 0.2$ | - | $\pm 0.2$ | $\pm 0.3$ | \%FSR |
| Bipolar Zero Error (Tech Note 2) | - | $\pm 0.05$ | $\pm 0.085$ | - | $\pm 0.15$ | $\pm 0.25$ | - | $\pm 0.25$ | $\pm 0.5$ | \%FSR |
| Bipolar Offset Error (Tech Note 2) | - | $\pm 0.05$ | $\pm 0.15$ | - | $\pm 0.1$ | $\pm 0.25$ | - | $\pm 0.25$ | $\pm 0.5$ | \%FSR |
| Gain Error (Tech Note 2) | - | $\pm 0.1$ | $\pm 0.15$ | - | $\pm 0.15$ | $\pm 0.35$ | - | $\pm 0.25$ | $\pm 0.65$ | \%FSR |
| No Missing Codes (fin $=10 \mathrm{kHz}$ ) | 16 | - | - | 16 | - | - | 16 | - | - | Bits |
| DYNAMIC PERFORMANCE |  |  |  |  |  |  |  |  |  |  |
| Peak Harmonics ( -0.5 dB ) dc to 100 kHz | - | -91 | - | - | -91 | - | - | -87 | - | dB |
| 100 kHz to 250 kHz | - | -86 | - | - | -86 | - | - | -84 | - | dB |
| Total Harmonic Distortion (-0.5dB) dc to 100 kHz | - | -89 | -81 | - | -89 | -81 | - | -85 | -76 | dB |
| 100 kHz to 250kHz | - | -84 | - | - | -84 | - | - | -82 | - | dB |
| Signal-to-Noise Ratio (w/o distortion, -0.5 dB ) |  |  |  |  |  |  |  |  |  |  |
| dc to 100 kHz | 81 | 83 | - | 81 | 83 | - | 75 | 80 | - | dB |
| 100 kHz to 250kHz | - | 80 | - | - | 80 | - | - | 79 | - | dB |
| Signal-to-Noise Ratio (4) |  |  |  |  |  |  |  |  |  |  |
| (\& distortion, -0.5 dB ) |  |  |  |  |  |  |  |  |  |  |
| dc to 100 kHz | 78 | 81 | - | 77 | 81 | - | 72 | 78 | - | dB |
| 100 kHz to 250 kHz | - | 78 | - | - | 78 | - | - | 76 | - | dB |
| Two-tone Intermodulation |  |  |  |  |  |  |  |  |  |  |
| Distortion (fin $=100 \mathrm{kHz}$, <br> $240 \mathrm{kHz}, \mathrm{fs}=500 \mathrm{kHz},-0.5 \mathrm{~dB}$ ) | - | -82 | - | - | -82 | - | - | -81 | - | dB |
| Noise | - | 150 | - | - | 150 | - | - | 150 | - | $\mu \mathrm{Vrms}$ |
| Input Bandwidth (-3dB) |  |  |  |  |  |  |  |  |  |  |
| Small Signal (-20dB input) | - | 2 | - | - | 2 | - | - | 2 | - | MHz |
| Large Signal ( -0.5 dB input) | - | 1.1 | - | - | 1.1 | - | - | 1.1 | - | MHz |
| Feedthrough Rejection (fin $=250 \mathrm{kHz}$ ) | - | -96 | - | - | -96 | - | - | -96 | - | dB |
| Slew Rate | - | $\pm 80$ | - | - | $\pm 80$ | - | - | $\pm 80$ | - | V/ L |
| Aperture Delay Time | - | $\pm 10$ | - | - | $\pm 10$ | - | - | $\pm 10$ | - | ns |
| Aperture Uncertainty | - | 50 | - | - | 50 | - | - | 50 | - | ps rms |
| S/H Acquisition Time |  |  |  |  |  |  |  |  |  | ns |
| Overvoltage Recovery Time (5) | - | 600 | 1000 | - | 600 | 1000 | - | 600 | 1000 | ns |
| A/D Conversion Rate | 500 | - | - | 500 | - | - | 500 | - | - | kHz |


| ANALOG OUTPUT | $+25^{\circ} \mathrm{C}$ |  |  | 0 to $+70^{\circ} \mathrm{C}$ |  |  | -55 to $+125^{\circ} \mathrm{C}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. | UNITS |
| Internal Reference |  |  |  |  |  |  |  |  |  |  |
| Voltage | +9.95 | +10.0 | +10.05 | +9.95 | +10.0 | +10.05 | +9.95 | +10.0 | +10.05 | Volts |
| Drift | - | $\pm 10$ | - | - | $\pm 10$ | - | - | $\pm 10$ | - | ppm/ ${ }^{\circ} \mathrm{C}$ |
| External Current | - | - | 1 | - | - | 1 | - | - | 1 | mA |
| DIGITAL OUTPUTS |  |  |  |  |  |  |  |  |  |  |
| Logic Levels |  |  |  |  |  |  |  |  |  |  |
| Logic "1" | +2.4 | - | - | +2.4 | - | - | +2.4 | - | - | Volts |
| Logic "0" | - | - | +0.4 | - | - | +0.4 | - | - | +0.4 | Volts |
| Logic Loading "1" | - | - | -4 | - | - | -4 | - | - | -4 | mA |
| Logic Loading "0" | - | - | +4 | - | - | +4 | - | - | +4 | mA |
| Delay, Falling Edge of ENABLE to |  |  |  |  |  |  |  |  |  |  |
| Output Coding |  | Com | mentary 0 | Binary; | plemen | Two's Con | ment, Off | Binary, T | Complen |  |
| POWER REQUIREMENTS |  |  |  |  |  |  |  |  |  |  |
| Power Supply Ranges |  |  |  |  |  |  |  |  |  |  |
| +15V Supply | +14.5 | +15.0 | +15.5 | +14.5 | +15.0 | +15.5 | +14.5 | +15.0 | +15.5 | Volts |
| -15V Supply | -14.5 | -15.0 | -15.5 | -14.5 | -15.0 | -15.5 | -14.5 | -15.0 | -15.5 | Volts |
| +5V Supply | +4.75 | +5.0 | +5.25 | +4.75 | +5.0 | +5.25 | +4.75 | +5.0 | +5.25 | Volts |
| Power Supply Currents |  |  |  |  |  |  |  |  |  |  |
| +15V Supply | - | +110 | +130 | - | +110 | +130 | - | +110 | +130 | mA |
| -15V Supply | - | -100 | -125 | - | -100 | -125 | - | -100 | -125 | mA |
| +5V Supply | - | +80 | +90 | - | +80 | +90 | - | +80 | +90 | mA |
| Power Dissipation | - | 3.5 | 4.25 | - | 3.5 | 4.25 | - | 3.5 | 4.25 | Watts |
| Power Supply Rejection | - | - | $\pm 0.02$ | - | - | $\pm 0.02$ | - | - | $\pm 0.02$ | \%FSR/\%V |

## Footnotes:

(1) All power supplies must be on before applying a start convert pulse. All supplies and the clock (START CONVERT) must be present during warmup periods. The device must be continuously converting during this time.
(2) When COMP. BITS (pin 8 ) is low, logic loading " 0 " will be $-350 \mu \mathrm{~A}$.
(3) A 200 ns wide start convert pulse is used for all production testing. For applications requiring less than a 500 kHz sampling rate, wider start convert pulses can be used.
(4) Effective bits is equal to:
$\frac{(\text { SNR }+ \text { Distortion })-1.76+\left[20 \log \frac{\text { Full Scale Amplitude }}{\text { Actual Input Amplitude }}\right]}{6.02}$

TECHNICAL NOTES

1. Obtaining fully specified performance from the ADS-930 requires careful attention to pc-card layout and power supply decoupling. The device's analog and digital ground systems are connected to each other internally. For optimal performance, tie all ground pins ( $4,11,13,18,24$ and 30 ) directly to a large analog ground plane beneath the package.
Bypass all power supplies and the +10 V reference output to ground with $4.7 \mu \mathrm{~F}$ tantalum capacitors in parallel with $0.1 \mu \mathrm{~F}$ ceramic capacitors. Locate the bypass capacitors as close to the unit as possible.
2. The ADS-930 achieves its specified accuracies without the need for external calibration. If required, the device's small initial offset and gain errors can be reduced to zero using the adjustment circuitry shown in Figure 2. When using this circuitry, or any similar offset and gain calibration hardware, make adjustments following warmup. To avoid interaction, always adjust offset before gain. Tie pins 5 and 6 to ANALOG GROUND (pin 4) if not using offset and gain adjust circuits.
3. Pin 8 (COMP. BITS) is used to select the digital output coding format of the ADS-930. See Tables 3a and 3b. When this pin has a TTL logic "0" applied, it complements all of the ADS-930's digital outputs.

When pin 8 has a logic "1" applied and the ADS-930 is operated within its unipolar ( 0 to -10 V ) input range, the output coding is straight binary. Apply-
ing a logic " 0 " to pin 8 under these conditions changes the output coding to complementary binary.
When pin 8 has a logic " 1 " applied and the ADS-930 is operated within its bipolar ( $\pm 5 \mathrm{~V}$ ) input range, the output coding is offset binary. Applying a logic " 0 " to pin 8 under these conditions changes the coding to complementary offset binary. Using the MSB output (pin 40) instead of the MSB output (pin 39) under these conditions changes the respective output codings to two's complement and complementary two's complement.

Pin 8 is TTL-compatible and can be directly driven with digital logic in applications requiring dynamic control over its function. There is an internal pull-up resistor on pin 8 allowing it to be either connected to +5 V or left open when a logic " 1 " is required.
4. To enable the three-state outputs, connect ENABLE (pin 9) to a logic "0" (low). To disable, connect pin 9 to a logic "1" (high).
5. Applying a start convert pulse while a conversion is in progress ( $\overline{\mathrm{EOC}}=$ logic "1") will initiate a new and probably inaccurate conversion cycle.
6. Do not enable/disable or complement the output bits or read from the FIFO during the conversion process (from the falling edge of START CONVERT to the falling edge of EOC).

INNOVATION and EXCELLENCE

## INTERNAL FIFO OPERATION

The ADS-930 contains an internal, user-initiated, 18-bit, 16- word FIFO memory. Each word in the FIFO contains the 16 data bits as well as the MSB and OVERFLOW bits. Pins 23 (FIFO/DIR) and 10 (FIFO READ) control the FIFO's operation. The FIFO's status can be monitored by reading pins 19 (FSTAT1) and 20 (FSTAT2).

When pin 23 (FIFO/DIR) has a logic "1" applied, the FIFO is inserted into the digital data path. When pin 23 has a logic " 0 " applied, the FIFO is transparent, and the output data goes directly to the output three-state register (whose operation is controlled by pin 9 (ENABLE)). Read and write commands to the FIFO are ignored when the ADS-930 is operated in the "direct" mode. It takes a maximum of 20 ns to switch the FIFO in or out of the ADS-930's operation.

## FIFO WRITE and READ Modes

Once the FIFO has been enabled (pin 23 high), digital data is automatically written to it, regardless of the status of FIFO READ (pin 10). Assuming the FIFO is initially empty, it will accept data (18-bit words) from the next 16 consecutive A/D conversions. As a precaution, pin 10 (which controls the FIFO's READ function) should not be low when data is first written to an empty FIFO.

When the FIFO is initially empty, digital data from the first conversion (the "oldest" data) appears at the output of the FIFO immediately after the first conversion has been completed and remains there until the FIFO is read.
If the output three-state register has been enabled (logic "0" applied to pin 9), data from the first conversion will appear at the output of the ADS-930. Attempting to write a 17th word to a full FIFO will result in that data, and any subsequent conversion data, being lost.

Once the FIFO is full (indicated by FSTAT1 and FSTAT2 both = "1"), it can be read by dropping the FIFO READ line (pin 10) to a logic " 0 " and then applying a series of 15 rising edges to the read line. Since the first data word is already present at the FIFO output, the first read command (the first rising edge applied to FIFO READ) will bring data from the second conversion to the output. Each subsequent read command/rising edge brings the next word to the output lines.

If a read command is issued after the FIFO has been emptied, the last word (the 16th conversion) will remain present at the outputs.

## FIFO Reset Feature

At any time, the FIFO can be reset to an empty state by putting the ADS-930 into its "direct" mode (logic "0" applied to pin 23, FIFO/DIR) and also applying a logic "0" to the FIFO READ line (pin 10). The empty status of the FIFO will be indicated by FSTAT1 going to a "0" and FSTAT2 going to a "1". The status outputs will change 40 ns after the control signals have been applied.

## FIFO Status, FSTAT1 and FSTAT2

The status of the data in the FIFO can be monitored by reading the two status pins, FSTAT1 (pin 19) and FSTAT2 (pin 20).
CONTENTS
Empty ( 0 words)
<half full ( $\leq 7$ words)
half-full or more ( $\geq 8$ words)
Full ( 16 words)

| FSTAT1 | FSTAT2 |
| :---: | :---: |
| 0 | 1 |
| 0 | 0 |
| 1 | 0 |
| 1 | 1 |


| DELAY | PIN | TRANSITION | MIN. | TYP. | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Direct mode to FIFO enabled | 23 | $0-1$ | - | 10 | 20 | ns |
| FIFO enabled to direct mode | 23 | $1 \downharpoonright 0$ | - | 10 | 20 | ns |
| FIFO READ to output data valid | 10 | $0-1$ | - | - | 40 | ns |
| FIFO READ to status update when changing from <half full (1 word) to empty | 10 | $1 \longrightarrow 0$ | - | - | 28 | ns |
| FIFO READ to status update when changing from >half full ( 8 words) to <half full ( 7 words) | 10 | $0-1$ | - | - | 110 | ns |
| FIFO READ to status update when changing from full ( 16 words) to $\geq$ half full ( 15 words) | 10 | $0<1$ | - | - | 190 | ns |
| Falling edge of $\overline{\mathrm{EOC}}$ to status update when writing first word into empty FIFO | 15 | $1 \smile 0$ | - | - | 190 | ns |
| Falling edge of $\overline{E O C}$ to status update when changing FIFO from <half full (7 words) to 2half full (8 words) | 15 | $1 \longrightarrow 0$ | - | - | 110 | ns |
| Falling edge of $\overline{\mathrm{EOC}}$ to status update when filling FIFO with 16th word | 15 | $1 \smile 0$ | - | - | 28 | ns |

Table 1. FIFO Delays

## CALIBRATION PROCEDURE (Refer to Figure 2 and Tables 3a and 3b)

Connect the converter per Table 2 for the appropriate input voltage range. Any offset/gain calibration procedures should not be implemented until the device is fully warmed up. To avoid interaction, adjust offset before gain. The ranges of adjustment for the circuits in Figure 2 are guaranteed to compensate for the ADS-930's initial accuracy errors and may not be able to compensate for additional system errors.
A/D converters are calibrated by positioning their digital outputs exactly on the transition point between two adjacent digital output codes. This is accomplished by connecting LED's to the digital outputs and performing adjustments until certain LED's "flicker" equally between on and off. Other approaches employ digital comparators or microcontrollers to detect when the outputs change from one code to the next.

For the ADS-930, offset adjusting is normally accomplished when the ana$\log$ input is 0 minus $1 / 2$ LSB $(-76 \mu \mathrm{~V})$. See Table 3b for the proper bipolar and unipolar output coding.
Gain adjusting is accomplished when the analog input is at nominal full scale minus $11 / 2$ LSB's ( -9.999771 V for unipolar and +4.999771 V for bipolar).
Note: Connect pin 5 to ANALOG GROUND (pin 4) for operation without zero/offset adjustment. Connect pin 6 to pin 4 for operation without gain adjustment.

| INPUT RANGE | INPUT PIN | TIE TOGETHER |
| :---: | :---: | :---: |
| 0 to -10 V | Pin 3 | Pins 2 and 4 |
| $\pm 5 \mathrm{~V}$ | Pin 3 | Pins 1 and 2 |
| Table 2. Input Connections |  |  |

Table 2. Input Connections

| OUTPUT FORMAT | PIN 8 LOGIC LEVEL |
| :--- | :---: |
| Straight Binary | 1 |
| Complementary Binary | 0 |
| Complementary Offset Binary | 0 |
| Offset Binary | 1 |
| Complementary Two's Complement (Using $\overline{\mathrm{MSB}}$, pin 40) | 0 |
| Two's Complement (Using $\overline{\mathrm{MSB}}$, pin 40) | 1 |

Table 3a. Setting Output Coding Selection (Pin 8)

## Zero/Offset Adjust Procedure

1. Apply a train of pulses to the START CONVERT input (pin 17) so that the converter is continuously converting.
2. For unipolar or bipolar zero/offset adjust, apply $-76.3 \mu \mathrm{~V}$ to the ANALOG INPUT (pin 3).
3. For a bipolar input - adjust the offset potentiometer until the code flickers between 1000000000000000 and 0111111111111111 with pin 8 tied high (offset binary) or between 0111111111111111 and 10000000 00000000 with pin 8 tied low (complementary offset binary).
For a unipolar input - adjust the offset potentiometer until all output bits are 0's and the LSB flickers between 0 and 1 with pin 8 tied high (straight binary) or until all output bits are 1's and the LSB flickers between 0 and 1 with pin 8 tied low (complementary binary).
4. Two's complement coding requires using BIT 1 (MSB) (pin 40). With pin 8 tied high, adjust the trimpot until the output code flickers between all 0's and all 1's.


|  |  | OUTPUT CO |  |  |  |  | BIPOLAR SCALE |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| UNIPOLAR | INPUT RANGE 0 to |  |  |  |  | INPUT RANGE$\pm 5 \mathrm{~V}$ |  |
| SCALE | -10V | MSB LSB | MSB LSB | $\overline{\text { MSB }}$ LSB | $\overline{\text { MSB }} \quad$ LSB |  |  |
| -FS +1 LSB | -9.999847 | 1111111111111111 | 0000000000000000 | 0111111111111111 | 1000000000000000 | +4.999847 | +FS -1 LSB |
| -FS +1 1/2 LSB | -9.999771 | LSB "1" to "0" | LSB "0" to "1" | LSB "1" to "0" | LSB "0" to "1" | +4.999771 | +FS -1 1/2 LSB |
| -7/8 FS | -8.750000 | 1110000000000000 | 0001111111111111 | 0110000000000000 | 1001111111111111 | +3.750000 | +3/4 FS |
| -3/4 FS | -7.500000 | 1100000000000000 | 0011111111111111 | 0100000000000000 | 1011111111111111 | +2.500000 | +1/2 FS |
| -1/2FS | -5.000000 | 1000000000000000 | 0111111111111111 | 0000000000000000 | 1111111111111111 | 0.000000 | $\begin{gathered} \frac{1}{0} \\ \hline-1 / 2 \text { LSB } \end{gathered}$ |
| -1/2FS -1/2LSB | -4.999924 | 0111111111111111 | 1000000000000000 | 1111111111111111 | 0000000000000000 | -0.000076 |  |
| -1/4FS | -2.500000 | 0100000000000000 | 1011111111111111 | 1100000000000000 | 0011111111111111 | $-2.500000$ | -1/2 FS |
| -1/8FS | -1.250000 | 0010000000000000 | 1101111111111111 | 1010000000000000 | 0101111111111111 | -3.750000 | -3/4 FS |
| -1 LSB | -0.000153 | 0000000000000001 | 1111111111111110 | 1000000000000001 | 0111111111111110 | -4.999847 | -FS +1 LSB |
| -1/2LSB | -0.000076 | LSB "0" to "1" | LSB "1" to "0" | LSB "0" to "1" | LSB "1" to "0" | -4.999924 | -FS + 1/2 LSB |
| 0 | 0.000000 | 0000000000000000 | 1111111111111111 | 1000000000000000 | 0111111111111111 | -5.000000 | -FS |
|  |  | COMP. OFF. BIN. | OFFSET BINARY | COMP. TWO'S COMP. | TWO'S COMP. |  |  |

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## Gain Adjust Procedure

1. Apply +4.999771 V to the ANALOG INPUT (pin 3) for bipolar gain adjust or apply -9.999771 V to pin 3 for unipolar gain adjust.
2. For a unipolar input - adjust the gain potentiometer until all output bits are 1 's and the LSB flickers between a 1 and 0 with pin 8 tied high (straight binary) or until all output bits are 0's and the LSB flickers between a 1 and 0 with pin 8 tied low (complementary binary).
For a bipolar input - adjust the gain potentiometer until all output bits are 1 's and the LSB flickers between a 1 and 0 with pin 8 tied low (complementary offset binary) or until all output bits are 0's and the LSB flickers between a 1 and 0 ith pin 8 tied high (offset binary).
3. Two's complement coding requires using pin 40 . With pin 8 tied high, adjust the gain trimpot until the output code flickers equally between 1000 000000000000 and 1000000000000001.

## THERMAL REQUIREMENTS

All DATEL sampling A/D converters are fully characterized and specified over operating temperature (case) ranges of 0 to $+70^{\circ} \mathrm{C}$ and -55 to $+125^{\circ} \mathrm{C}$. All room-temperature $\left(\mathrm{TA}=+25^{\circ} \mathrm{C}\right)$ production testing is performed without the use of heat sinks or forced-air cooling. Thermal impedance figures for each device are listed in their respective specification tables.

These devices do not normally require heat sinks, however, standard precautionary design and layout procedures should be used to ensure devices do not overheat. The ground and power planes beneath the package, as well as all pcb signal runs to and from the device, should be as heavy as possible to help conduct heat away from the package. Electrically insulating, thermally conductive "pads" may be installed underneath the package. Devices should be soldered to boards rather than "socketed", and of course, minimal air flow over the surface can greatly help reduce the package temperature.


Figure 3. ADS-930 Timing Diagram


Figure 4. FFT Analysis of ADS-930

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