# Am79R79\*/Le79R79

# (Legerity.

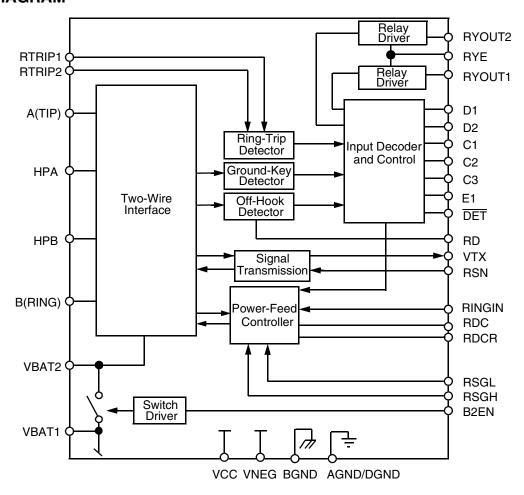
# **Ringing Subscriber Line Interface Circuit**

#### DISTINCTIVE CHARACTERISTICS

- Ideal for short-loop applications
- Ideal for ISDN terminal adaptor and fixed radio access applications
- On-chip ringing with on-chip ring-trip detector
- Low standby state power
- **■** Battery operation:
  - $V_{BAT1}$ : -40.5 V to -75 V
  - V<sub>BAT2</sub>: –19 V to V<sub>BAT1</sub>
- On-chip battery switching and feed selection
- On-hook transmission
- Two-wire impedance set by single external impedance

- Programmable constant-current feed
- **■** Programmable Open Circuit voltage
- Programmable loop-detect threshold
- Current gain = 1000
- Ground-key detector
- Tip Open state for ground-start lines
- Polarity reversal option available
- Internal V<sub>EE</sub> regulator (no external –5 V power supply required)
- Two on-chip relay drivers and snubber circuits
- Space Saving Package Options (8x8 QFN)

# **BLOCK DIAGRAM**



<sup>\*</sup>This product can be ordered using ordering part numbers Am79R79 or Le79R79. The Am79R79 ordering part number will be discontinued after 6/30/02, at which time the product will only be available using the Le79R79 ordering part number.

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# GENERAL DESCRIPTION

The Legerity family of subscriber line interface circuit (SLIC) products provide the telephone interface functions required throughout the worldwide market. Legerity SLIC devices address all major telephony markets including central office (CO), private branch exchange (PBX), digital loop carrier (DLC), fiber-in-the-loop (FITL), radio-in-the-loop (RITL), hybrid fiber coax (HFC), and video telephony applications.

The Legerity SLIC devices offer support of BORSHT (battery feed, overvoltage protection, ringing, supervision, hybrid, and test) functions with features including current limiting, on-hook transmission, polarity reversal, Tip Open, and loop-current detection. These features allow reduction of linecard cost by minimizing component count, conserving board space, and supporting automated manufacturing.

The Legerity SLIC devices provide the two- to four-wire hybrid function, DC-loop feed, and two-wire supervision. Two-wire termination is programmed by a scaled impedance network. Transhybrid balance can be achieved with an external balance circuit or simply programmed using a companion Legerity codec device, the Am79C02/03/031 DSLAC<sup>™</sup> device, the Am79Q02/021/03 Programmable Quad SLAC (QSLAC<sup>™</sup>) device, or the Am79Q5457/4457 Nonprogrammable QSLAC device.

The Le79R79 Ringing SLIC device is a bipolar monolithic SLIC that offers on-chip ringing. Now designers can achieve significant cost reductions at the system level for short-loop applications by integrating the ringing function on chip. Examples of such applications would be ISDN terminal adaptors, fiber-in-the-loop, radio-in-the-loop, hybrid fiber/coax and video telephony (home-side) boxes. The Le79R79 Ringing SLIC can provide sufficient voltage to meet the stringent LSSGR five-ringer equivalent specification. Using a CMOS-compatible input waveform and wave shaping R-C network, the Le79R79 Ringing SLIC can provide trapezoidal wave ringing to meet various design requirements.

In order to further enhance the suitability of this device in short-loop, distributed switching applications, Legerity has maximized power savings by incorporating battery switching on chip. The Le79R79 Ringing SLIC device switches between two battery supplies such that in the off-hook (active) state, a low battery is used to save power. In order to meet the Open Circuit voltage requirements of fax machines and maintenance termination units (MTU), the SLIC automatically switches to a higher voltage in the on-hook (standby) state.

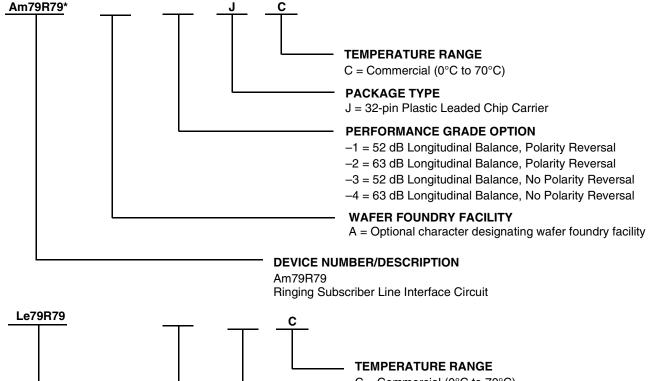
Like all of the Legerity SLIC devices, the Le79R79 Ringing SLIC device supports on-hook transmission, ringtrip detection, programmable loop-detect threshold, and is available with on-chip polarity reversal. The Le79R79 Ringing SLIC device is a programmable constant-current feed device with two on-chip relay drivers to operate external relays. Several performance grades are available to meet both CCITT and LSSGR requirements, including various longitudinal balance options. This unique device is available in the proven Legerity 75 V bipolar process.

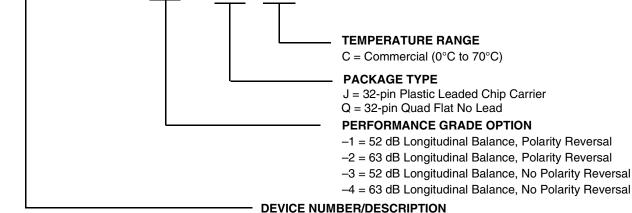


# ORDERING INFORMATION

# **Standard Products**

Legerity standard products are available in several packages and operating ranges. The order number (Valid Combination) is formed by a combination of the elements below





Le79R79

#### Ringing Subscriber Line Interface Circuit **Valid Combinations** -2 Am79R79 JC Am79R79A -3

\_4

Valid C	ombina	tions
Le79R79	-1 -2 -3 -4	JC QC**

#### **Valid Combinations**

Valid Combinations list configurations planned to be supported in volume for this device. Consult the local Legerity sales office to confirm availability of specific valid combinations, to check on newly released combinations, and to obtain additional data on Legerity's standard military-grade products.

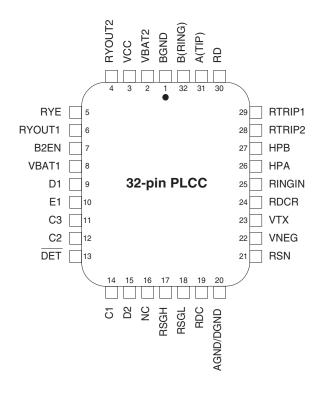
<sup>\*</sup>This product can be ordered using ordering part numbers Am79R79 or Le79R79. The Am79R79 ordering part number will be discontinued after 6/30/02, at which time the product will only be available using the Le79R79 ordering part number.

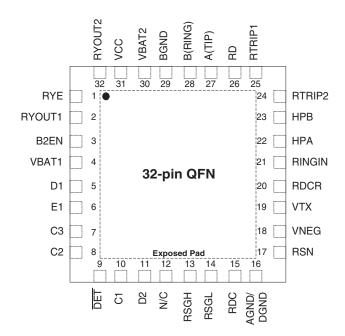
<sup>\*\*</sup>Due to size constraints, QFN devices are marked by omitting the "Le" prefix and the performance grade dash character. For example, Le79R79-2QC is marked 79R792QC.



# **CONNECTION DIAGRAMS**

# **Top View**





#### Notes:

- 1. Pin 1 is marked for orientation.
- 2. NC = No connect
- 3. The thermally enhanced QFN package features an exposed pad on the underside which must be electrically tied to VBAT1.



# **PIN DESCRIPTIONS**

Pin Names	Туре	Description
AGND/DGND	Gnd	Analog and Digital ground
A(TIP)	Output	Output of A(TIP) power amplifier
B2EN	Input	$V_{BAT2}$ Enable. Logic Low enables operation from $V_{BAT2}$ . Logic High enables operation from $V_{BAT1}$ . TTL compatible.
BGND	Gnd	Battery (power) ground
B(RING)	Output	Output of B(RING) power amplifier
C3-C1	Input	Decoder. TTL compatible. C3 is MSB and C1 is LSB.
D1	Input	Relay1 Control. TTL compatible. Logic Low activates the Relay1 relay driver.
D2	Input	(Option) Relay2 Control. TTL compatible. Logic Low activates the Relay2 relay driver.
DET	Output	Switchhook Detector. Logic Low indicates that the selected detector is tripped. Logic inputs C3–C1 and E1 select the detector. Open-collector with a built-in 15 k $\Omega$ pull-up resistor.
E1	Input	(Option) Ground-Key Enable. A logic High selects the off-hook detector. A logic Low selects the ground-key detector. TTL compatible.
HPA	Capacitor	High-pass filter capacitor. A(TIP) side of high-pass filter capacitor.
НРВ	Capacitor	High-pass filter capacitor. B(RING) side of high-pass filter capacitor.
RD	Resistor	Detector resistor. Detector threshold set and filter pin.
RDC	Resistor	DC feed resistor. Connection point for the DC feed current programming network, which also connects to the receiver summing node (RSN). V <sub>RDC</sub> is negative for normal polarity and positive for reverse polarity.
RDCR	_	Connection point for feedback during ringing.
RINGIN	Input	Ring Signal Input. Pin for ring signal input. Square-wave shaped by external RC filter. Requires 50% duty cycle. CMOS-compatible input.
RSGH	Input	Saturation Guard High. Pin for resistor to adjust Open Circuit voltage when operating from $V_{\text{BAT1}}$ .
RSGL	Input	Saturation Guard Low. Pin for resistor to adjust the anti-saturation cut-in voltage when operating from both $V_{BAT1\ and}\ V_{BAT2}.$
RSN	Input	Receive Summing Node. The metallic current (AC and DC) between A(TIP) and B(RING) is equal to 1000 x the current into this pin. The networks that program receive gain, two-wire impedance, and feed resistance all connect to this node.
RTRIP1	Input	Ring-trip detector. Ring-trip detector threshold set and filter pin.
RTRIP2	Input	Ring-trip detector. Ring-trip detector threshold offset (switch to $V_{BAT1}$ ). For power conservation in any nonringing state, this switch is open.
RYE	Output	Common Emitter of RYOUT1/RYOUT2. Emitter output of RYOUT1 and RYOUT2. Normally connected to relay ground.
RYOUT1	Output	Relay/switch driver. Open-collector driver with emitter internally connected to RYE.
RYOUT2	Output	(Option) Relay/switch driver. Open-collector driver emitter internally connected to RYE.
VBAT1	Battery	Battery supply and connection to substrate.
VBAT2	Battery	Power supply to output amplifiers. Connect to off-hook battery through a diode.
VCC	Power	Positive analog power supply.
VNEG	Power	Negative analog power supply. This pin is the return for the intern VEE regulator.
VTX	Output	Transmit Audio. This output is 0.5066 gain version of the A(TIP) and B(RING) metallic voltage. VTX also sources the two-wire input impedance programming network.
Exposed Pad	Battery	This must be electrically tied to VBAT1.



# **ABSOLUTE MAXIMUM RATINGS**

Storage temperature55°C to +150°C
$V_{CC}$ with respect to AGND/DGND0.4 V to +7 V
$\rm V_{NEG}$ with respect to AGND/DGND0.4 V to $\rm V_{BAT2}$
V <sub>BAT2</sub> V <sub>BAT1</sub> to GND
V <sub>BAT1</sub> with respect to AGND/DGND:
Continuous+0.4 V to -80 V 10 ms+0.4 V to -85 V
BGND with respect to AGND/DGND+3 V to $-3$ V
A(TIP) or B(RING) to BGND:
ContinuousV <sub>BAT1</sub> –5 V to +1 V
10 ms (f = 0.1 Hz)
1 $\mu$ s (f = 0.1 Hz) $V_{BAT1}$ –15 V to +8 V 250 ns (f = 0.1 Hz) $V_{BAT1}$ –20 V to +12 V
Current from A(TIP) or B(RING)±150 mA
RYOUT1, RYOUT2 current
RYOUT1, RYOUT2 voltageRYE to +7 V
RYOUT1, RYOUT2 transient
RYE voltageBGND to V <sub>BAT1</sub>
C3–C1, D2–D1, E1, B2EN, and RINGIN
Input voltage
Maximum continuous power dissipation*
T <sub>A</sub> = 70°C
In 32-pin PLCC package 1.67 W
In 32-pin QFN package 3.00 W
$T_A = 85^{\circ}C$
In 32-pin PLCC package 1.33 W In 32-pin QFN package 2.40 W
Thermal data:θ,IA
In 32-pin PLCC package45°C/W typ
In 32-pin QFN package**25°C/W typ

<sup>\*</sup> Thermal limiting circuitry on chip will shut down the circuit at a junction temperature of about 165°C. Continuous operation above 145°C junction temperature may degrade device reliability.

Stresses above those listed under Absolute Maximum Ratings may cause permanent device failure. Functionality at or above these limits is not implied. Exposure to Absolute Maximum Ratings for extended periods may affect device reliability.

# OPERATING RANGES Commercial (C) Devices

Ambient temperature	0°C to +70°C*
V <sub>CC</sub>	4.75 V to 5.25 V
V <sub>NEG</sub>	4.75 V to V <sub>BAT2</sub>
V <sub>BAT1</sub>	40.5 V to -75 V
V <sub>BAT2</sub>	–19 V to V <sub>BAT1</sub>
AGND/DGND	0 V
BGND with respect to AGND/DGND	–100 mV to +100 mV
Load resistance on VTX to	ground20 k $\Omega$ min

The Operating Ranges define those limits between which the functionality of the device is guaranteed.

<sup>\*\*</sup> The thermal performance of a thermally enhanced package is assured through optimized printed circuit board layout. Specified performance requires that the exposed thermal pad be soldered to an equally sized exposed copper surface, which, in turn, conducts heat through multiple vias to a large internal copper plane.

<sup>\*</sup> Legerity guarantees the performance of this device over commercial (0°C to 70°C) and industrial (-40°C to 85°C) temperature ranges by conducting electrical characterization over each range and by conducting a production test with single insertion coupled to periodic sampling. These characterization and test procedures comply with section 4.6.2 of Bellcore TR-TSY-000357 Component Reliability Assurance Requirements for Telecommunications Equipment.



# **ELECTRICAL CHARACTERISTICS**

Description	Test Conditions (See Note 1)	Min	Тур	Max	Unit	Note	
Transmission Performance							
2-wire return loss	200 Hz to 3.4 kHz (Test Circuit D)	26			dB	1, 4, 6	
Z <sub>VTX</sub> , analog output impedance	3 20		20	Ω	4		
V <sub>VTX</sub> , analog output offset voltage	0°C to +70°C	-35		+35	mV		
	-40°C to +85°C	-40		+40	IIIV	4	
Z <sub>RSN</sub> , analog input impedance	1 20		Ω	4			
Overload level, 2-wire and 4-wire, off hook	Active state	2.5			Vpk	2a	
Overload level, 2-wire	On hook, $R_{LAC} = 600 \Omega$	0.88			Vrms	2b	
THD (Total Harmonic Distortion)	+3 dBm, BAT2 = -24 V		-64	-50			
THD, on hook, OHT state	0 dBm, $R_{LAC}$ = 600 $\Omega$ BAT1 = -75 V			-40	dB	5	
Longitudinal Performance (See Test	Circuit C)					I	
Longitudinal to metallic	200 Hz to 1 kHz	52					
L-T, L-4 balance	normal polarity -2, -4	63					
	reverse polarity -2	54					
	normal polarity, -40°C to +85°C -2, -4	58				4	
	1 kHz to 3.4 kHz	52			dB		
	normal polarity -2, -4	58					
	reverse polarity –2	54					
	normal polarity, -40°C to +85°C -2, -4	54				4	
Longitudinal signal generation 4-L	200 Hz to 800 Hz normal polarity	42					
Longitudinal current per pin (A or B)	Active or OHT state	12	28		mArms	4	
Longitudinal impedance at A or B	0 to 100 Hz, T <sub>A</sub> = +25°C		25		Ω/pin		
Idle Channel Noise							
C-message weighted noise	0°C to +70°C		+7	+11	dBrnC		
	-40°C to +85°C			+12	dBITIC		
Psophometric weighted noise	0°C to +70°C		-83	-79	dBmp	4	
	-40°C to +85°C			-78	иыпр		
Insertion Loss and Four- to Four-Wire	e Balance Return Signal (See Test Cir	cuits A a	nd B)			T	
Gain accuracy 4- to 2-wire	0 dBm, 1 kHz	-0.20	0	+0.20			
Gain accuracy 2- to 4-wire and 4- to 4-wire	0 dBm, 1 kHz	-6.22	-6.02	<i>–</i> 5.82			
Gain accuracy 4- to 2-wire	OHT state, on hook	-0.35	0	+0.35		3	
Gain accuracy 2- to 4-wire and 4- to 4-wire	OHT state, on hook	-6.37	-6.02	<i>–</i> 5.77			
Gain accuracy over frequency	300 to 3400 Hz 0°C to +70°C	-0.10		+0.10	dB		
	relative to 1 kHz -40°C to +85°C	-0.15		+0.15			
Gain tracking	+3 dBm to -55 dBm 0°C to +70°C	-0.10		+0.10			
	relative to 0 dBm -40°C to +85°C	-0.15		+0.15		3, 4	
Gain tracking	0 dBm to -37 dBm 0°C to +70°C	-0.10		+0.10			
OHT state, on hook	−40°C to +85°C	-0.15		+0.15			
	+3 dBm to 0 dBm	-0.35		+0.35		3	
Group delay	0 dBm, 1 kHz		3		μs	1, 4, 6	

# Note:

<sup>\*</sup> Performance Grade



# ELECTRICAL CHARACTERISTICS (continued)

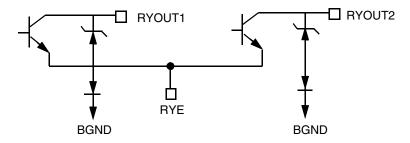
Description	Test Conditions (See Note 1)	Min	Тур	Max	Unit	Note
Line Characteristics						
I <sub>L</sub> , Loop-current accuracy	I <sub>L</sub> in constant-current region, B2EN = 0	0.915I <sub>L</sub>	ال	1.085I <sub>L</sub>		
I <sub>L</sub> , Long loops, Active state	$R_{LDC} = 600 \Omega$ , RSGL = open	20	21.7			
	$R_{LDC} = 750 \Omega$ , RSGL = short	20				
I <sub>L</sub> , Accuracy, Standby state	$V_{\rm BAT1} = 10 V$	0.8I <sub>L</sub>	ΙL	1.2l <sub>L</sub>		
	$I_{L} = \frac{\left  V_{BATI} \right  - 10V}{R_{L} + 400}$				mA	
	I <sub>L</sub> = constant-current region	18	27	39		
	T <sub>A</sub> = 25°C					
	-40°C to +85°C	18	27			4
I <sub>L</sub> LIM	Active, A and B to ground		55	110		
	OHT, A and B to ground		55			4
I <sub>L</sub> , Loop current, Open Circuit state	$R_L = 0$			100	μA	
I <sub>A</sub> , Pin A leakage, Tip Open state	$R_L = 0$			100	μΛ	
I <sub>B</sub> , Pin B current, Tip Open state	B to ground		34		mA	
VA, Standby, ground-start signaling	A to $-48 \text{ V} = 7 \text{ k}\Omega$ ,	-7.5	<b>-</b> 5			4
	B to ground = 100 $\Omega$				V	
V <sub>AB</sub> , Open Circuit voltage		42.8				8
	PLE = 100 mVrms), Active Normal State					
V <sub>CC</sub>	50 Hz to 3400 Hz	33	50			
V <sub>NEG</sub>	50 Hz to 3400 Hz	30	40		dB	
$V_{BAT1}$	50 Hz to 3400 Hz	30	50		uБ	5
$V_{BAT2}$	50 Hz to 3400 Hz	30	50			
Power Dissipation						
On hook, Open Circuit state	V <sub>BAT1</sub>		48	100		
On hook, Standby state	V <sub>BAT2</sub>		55	80		10
On hook, OHT state	V <sub>BAT1</sub>		200	300		
On hook, Active state	V <sub>BAT1</sub>		220	350	mW	
Off hook, Standby state	$V_{BAT1}$ or $V_{BAT2}$ $R_L = 300 \Omega$		2000	2800		10
Off hook, OHT state	$V_{BAT1}$ $R_L = 300 \Omega$		2000	2200		
Off hook, Active state	$V_{BAT2}$ $R_L = 300 \Omega$		550	750		
Supply Currents						
I <sub>CC</sub> ,	Open Circuit state		3.0	4.5		
On-hook V <sub>CC</sub> supply current	Standby state		3.2	5.5		
	OHT state		6.2	8.0		
	Active state-normal		6.5	9.0		
I <sub>NEG</sub> ,	Open Circuit state		0.1	0.2		
On-hook V <sub>NEG</sub> supply current	Standby state		0.1	0.2		
	OHT state		0.7	1.1	mA	
	Active state-normal		0.7	1.1		
I <sub>BAT</sub> ,	Open Circuit state		0.45	1.0		
On-hook V <sub>BAT</sub> supply current	Standby state		0.6	1.5		
DAI SEED SAME	OHT state		2.0	4.0		
	Active state-normal		2.7	5.0		



# **ELECTRICAL CHARACTERISTICS (continued)**

Description	Test Conditions (See Note 1)	Min	Тур	Max	Unit	Note
Logic Inputs (C3-C1, D2-D1, E1, and	B2EN)	•				•
V <sub>IH</sub> , Input High voltage		2.0			V	
V <sub>IL</sub> , Input Low voltage				0.8	V	
I <sub>IH</sub> , Input High current		-75		40	^	
I <sub>IL</sub> , Input Low current		-400			μA	
Logic Output DET						
V <sub>OL</sub> , Output Low voltage	$I_{OUT}$ = 0.8 mA, 15 k $\Omega$ to $V_{CC}$			0.40	٧	
V <sub>OH</sub> , Output High voltage	$I_{OUT} = -0.1$ mA, 15 k $\Omega$ to $V_{CC}$	2.4			V	
Ring-Trip Detector Input						
Ring detect accuracy	$IRTD = \left(\frac{ BAT1  - 1}{RRT1} + 24 \ \mu A\right) \bullet 335$	-10		+10	%	
Ring Signal		•	•			•
V <sub>AB</sub> , Ringing	Bat1 = $-75$ V, ringload = $1570 \Omega$	66	69		Vpk	7
V <sub>AB</sub> Ringing offset	V <sub>RINGIN</sub> = 2.5 V	-10	0	10	V	
ΔV <sub>AB</sub> /ΔV <sub>RINGIN</sub> (RINGIN gain)		150	180	210		
Ground-Key Detector Thresholds						
Ground-key resistive threshold	B to ground	2	5	10	kΩ	
Ground-key current threshold	B to ground		11		mA	
Loop Detector						
R <sub>LTH</sub> , Loop-resistance detect threshold	Active, V <sub>BAT1</sub>	-20		20		
	Active, V <sub>BAT2</sub>	-20		20	%	9
	Standby	-12		12		
Relay Driver Output (RELAY1 and 2)						
V <sub>OL</sub> , On voltage (each output)	I <sub>OL</sub> = 30 mA		+0.25	+0.4	V	
V <sub>OL</sub> , On voltage (each output)	$I_{OL} = 40 \text{ mA}$		+0.30	+0.8	v	4
I <sub>OH</sub> , Off leakage (each output)	V <sub>OH</sub> = +5 V			100	μΑ	
Zener breakover (each output)	I <sub>Z</sub> = 100 μA	6.6	7.9		V	
Zener on voltage (each output)	$I_Z = 30 \text{ mA}$		11		v	

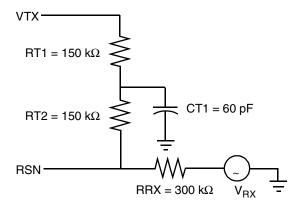
# **RELAY DRIVER SCHEMATIC**





#### Notes

1. Unless otherwise noted, test conditions are BAT1 = -75 V, BAT2 = -24 V,  $V_{CC}$  = +5 V,  $V_{NEG}$  = -5 V,  $R_L$  = 600  $\Omega$ ,  $R_{DC1}$  = 80 k $\Omega$ ,  $R_{DC2}$  = 20 k $\Omega$ ,  $R_D$  = 75 k $\Omega$ , no fuse resistors,  $C_{HP}$  = 0.018  $\mu$ F,  $C_{DC}$  = 1.2  $\mu$ F,  $D_1$  =  $D_2$  = 1N400x, two-wire AC input impedance (ZSL) is a 600  $\Omega$  resistance synthesized by the programming network shown below.  $R_{SGL}$  = open,  $R_{SGH}$  = open,  $R_{DCR}$  = 2 k $\Omega$ ,  $R_{RT1}$  = 430 k $\Omega$ ,  $R_{RT2}$  = 12 k $\Omega$ ,  $C_{RT}$  = 1.5  $\mu$ F,  $R_{SLEW}$  = 100 k $\Omega$ ,  $C_{SLEW}$  = 0.33  $\mu$ F.



- 2. a. Overload level is defined when THD = 1%.
- b. Overload level is defined when THD = 1.5%.
- 3. Balance return signal is the signal generated at  $V_{TX}$  by  $V_{RX}$ . This specification assumes that the two-wire AC load impedance matches the programmed impedance.
- 4. Not tested in production. This parameter is guaranteed by characterization or correlation to other tests.
- 5. This parameter is tested at 1 kHz in production. Performance at other frequencies is guaranteed by characterization.
- 6. Group delay can be greatly reduced by using a  $Z_T$  network such as that shown in Note 1 above. The network reduces the group delay to less than 2  $\mu$ s and increases 2WRL. The effect of group delay on linecard performance may also be compensated for by synthesizing complex impedance with the QSLAC or DSLAC device.
- 7. 70 Vpk provides 50 Vrms with a crest factor of 1.25 to a load of 1400  $\Omega$  with 2 Rf = 100, and Rline = 70  $\Omega$  (1570  $\Omega$ ).
- 8. Open Circuit V<sub>AB</sub> can be modified using RSGH.
- 9.  $R_D$  must be greater than 56  $k\Omega$ . Refer to Table 2 for typical value of  $R_{ITH}$ .
- 10. Lower power is achieved by switching into low-battery state in standby. Standby loop current is returned to  $V_{BAT1}$  regardless of the battery selected.

**Table 1. SLIC Decoding** 

			(DET) Output		
State	C3 C2 C1	2-Wire Status	E1 = 1	E1 = 0	Battery Selection
0	0 0 0	Open Circuit	Ring trip	Ring trip	
1	0 0 1	Ringing	Ring trip	Ring trip	B2EN
2	0 1 0	Active	Loop detector	Ground key	DZEIN
3	0 1 1	On-hook TX (OHT)	Loop detector	Ground key	
4	1 0 0	Tip Open	Loop detector	Ground key	B2EN = 1**
5	1 0 1	Standby	Loop detector	Ground key	V <sub>BAT1</sub>
6*	1 1 0	Active Polarity Reversal	Loop detector	Ground key	B2EN
7*	1 1 1	OHT Polarity Reversal	Loop detector	Ground key	DZEN

# Notes:

<sup>\*</sup> Only -1 and -2 performance grade devices support polarity reversal.

<sup>\*\*</sup> For correct ground-start operation using Tip Open, V<sub>BAT1</sub> on-hook battery must be used.



# **Table 2. User-Programmable Components**

	Trogrammasio componente
$Z_{\rm T} = 500(Z_{\rm 2WIN} - 2R_{\rm F})$	$Z_T$ is connected between the VTX and RSN pins. The fuse resistors are $R_F, \ \mbox{and} \ Z_{2WIN}$ is the desired 2-wire AC input impedance. When computing $Z_T,$ the internal current amplifier pole and any external stray capacitance between VTX and RSN must be taken into account.
$Z_{\rm RX} = \frac{Z_{\rm L}}{G_{42\rm L}} \bullet \frac{1000 \bullet Z_{\rm T}}{Z_{\rm T} + 500(Z_{\rm L} + 2R_{\rm F})}$	$Z_{RX}$ is connected from $V_{RX}$ to $R_{SN}. \ Z_T$ is defined above, and $G_{42L}$ is the desired receive gain.
$R_{DC1} + R_{DC2} = \frac{2500}{I_{LOOP}}$	$R_{DC1}$ , $R_{DC2}$ , and $C_{DC}$ form the network connected to the RDC pin. $I_{LOOP}$ is the desired loop current in the constant-current region.
$R_{DCR1} + R_{DCR2} = \frac{3000}{Iringlim}$	$R_{DCR1}$ , $R_{DCR2}$ , and $C_{DCR}$ form the network connected to the RDCR pin. See Applications Circuit for these components.
$C_{DC} = 19 \text{ ms} \bullet \frac{R_{DC1} + R_{DC2}}{R_{DC1}R_{DC2}}$	
$C_{\text{DCR}} = \frac{R_{\text{DCR1}} + R_{\text{DCR2}}}{R_{\text{DCR1}} R_{\text{DCR2}}} \bullet 150  \mu \text{s}$	$C_{\text{DCR}}$ sets the ringing time constant, which can be between 15 $\mu s$ and 150 $\mu s.$
$R_{\rm D}$ = $R_{\rm LTH}$ • 12.67 for high battery state	$R_D$ is the resistor connected from the RD pin to GND and $R_{LTH}$ is the loop-resistance threshold between on-hook and off-hook detection. $R_D$ should be greater than 56 $k\Omega$ to guarantee detection occurs in the Standby state. Choose the value of $R_D$ for high battery state; then use the equation for $R_{LTH}$ to find where the threshold is for low battery.
Loop-Threshold Detect Equations	
$R_{LTH} = \frac{R_D}{12.67}$ for high battery	This is the same equation as for ${\rm R}_{\rm D}$ above, except solved for ${\rm R}_{\rm LTH}.$
$R_{LTH} = \frac{R_D}{11.37}$ for low battery	For low battery, the detect threshold is slightly higher, which avoids oscillating between states.
$R_{LTH} = \frac{ V_{BATI}  - 10}{915} \bullet R_{D} - 400 - 2R_{F}$	$R_{LTH}$ standby < $R_{LTH}$ active $V_{BAT1}$ < $R_{LTH}$ active $V_{BAT2}$ , which guarantees no unstable states under all operating conditions. This equation shows at what resistance the standby threshold is; it is actually a current threshold rather than a resistance threshold, which is shown by the Vbat dependency.



# DC FEED CHARACTERISTICS

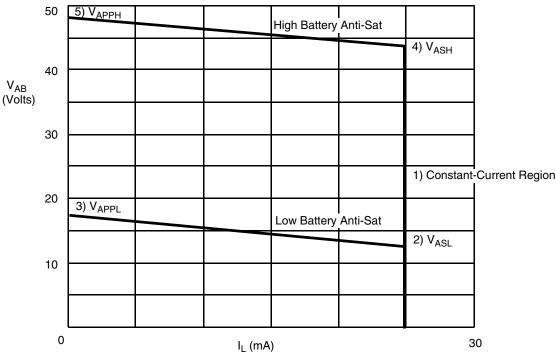


Figure 1. Typical V<sub>AB</sub> vs. I<sub>L</sub> DC Feed Characteristics

$$R_{DC} = R_{DC1} + R_{DC2} = 20 \hspace{0.2cm} k\Omega + 80 \hspace{0.2cm} k\Omega = 100 \hspace{0.2cm} k\Omega$$

$$(V_{\mathrm{BAT1}} = -75~\mathrm{V,} V_{\mathrm{BAT2}} = -24~\mathrm{V})$$

#### Notes

- 1. Constant-current region:  $V_{AB} = I_L R_L = \frac{2500}{RDC} R_L$ ; where  $R_L = R_L + 2R_F$ ,
- 2. Low battery  $V_{ASL} = \frac{1000 \bullet (104 \bullet 10^3 + R_{SGL})}{6720 \bullet 10^3 + (80 \bullet R_{SGL})}$ ; where  $R_{SGL}$  = resistor to GND, B2EN = logic Low.

 $\begin{aligned} &\textit{Anti-sat region:} V_{ASL} = \frac{1000 \bullet (R_{SGL} - 56 \bullet 10^3)}{6720 \bullet 10^3 + (80 \bullet R_{SGL})} \;\; ; \;\; \textit{where } R_{SGL} = \textit{resistor to } V_{CC} \; \textit{B2EN} = \textit{logic Low.} \\ &\textit{R}_{SGL} \; \textit{to } V_{CC} \; \textit{must be greater than } \frac{1000 \, \text{k} \Omega \cdot R_{SGL}}{1000 \, \text{k} \Omega \cdot R_{SGL}} \end{aligned} \; ; \;\; \textit{where } R_{SGL} = \textit{resistor to } V_{CC} \; \textit{B2EN} = \textit{logic Low.}$ 

3.  $V_{APPL} = 4.17 + V_{ASL}$ 

$$I_{\text{LOOPL}} = \frac{V_{\text{APPL}}}{\frac{(R_{\text{DC1}} + R_{\text{DC2}})}{600} + 2R_{\text{F}} + R_{\text{LOOP}}}$$

4. High battery $V_{ASH} = V_{ASHH} + V_{ASL}$ 

Anti-sat region:  $V_{ASHH} = \frac{1000 \bullet (70 \bullet 10^3 + R_{SGH})}{1934 \bullet 10^3 + (31.75 \bullet R_{SGH})}$ ; where  $R_{SGH}$  = resistor to GND, B2EN = logic High.

 $\begin{aligned} \mathbf{V}_{\mathrm{ASHH}} &= \frac{1000 \bullet (\mathbf{R}_{\mathrm{SGH}} + 2.75 \bullet 10^{3})}{1934 \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})} \;\; ; \;\; \textit{where $R_{\mathrm{SGH}}$ = resistor to $V_{\mathrm{CC}}$, B2EN = logic High.} \\ \mathbf{R}_{\mathrm{SGH}} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{SGH}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm{CC}})$} \; \text{to $V_{\mathrm{CC}}^{1934} \bullet 10^{3} + (31.75 \bullet \mathbf{R}_{\mathrm$ 

5.  $V_{APPH} = 4.17 + V_{ASH}$ 

$$I_{\text{LOOPH}} = \frac{V_{\text{APPH}}}{\frac{(R_{\text{DC1}} + R_{\text{DC2}})}{600} + 2R_{\text{F}} + R_{\text{LOOP}}}$$



# **RING-TRIP COMPONENTS**

$$R_{RT2} = 12 \text{ k}\Omega$$

$$C_{RT} = 1.5 \mu F$$

$$R_{RT1} = 320 \bullet \text{CF} \bullet \frac{V_{\text{BAT1}}}{V_{\text{BAT1}} - 5 - (24 \ \mu\text{A} \bullet 320 \bullet \text{CF} \bullet (R_{\text{LRT}} + 150 + 2R_{\text{F}}))} \bullet (R_{\text{LRT}} + 150 + 2R_{\text{F}})$$

where  $R_{LRT}$  = Loop-detection threshold resistance for ring trip and CF = Crest factor of ringing signal ( $\approx 1.25$ )

# R<sub>SLEW</sub>, C<sub>SLEW</sub>

Ring waveform rise time  $\approx 0.214 \bullet (R_{SLEW} \bullet C_{SLEW}) \approx tr.$ 

For a 1.25 crest factor @ 20 Hz,  $tr \approx 10$  mS.

 $\therefore$  (R<sub>SLEW</sub> = 150 k $\Omega$ , C<sub>SLEW</sub> = 0.33  $\mu$ F.)

 $C_{SLEW}$  should be changed if a different crest factor is desired.

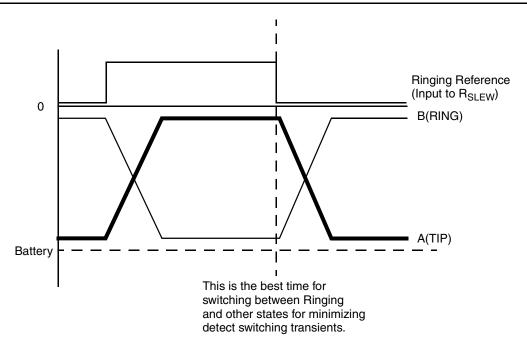


Figure 2. Ringing Waveforms



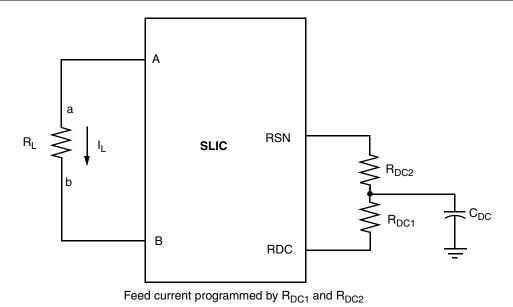
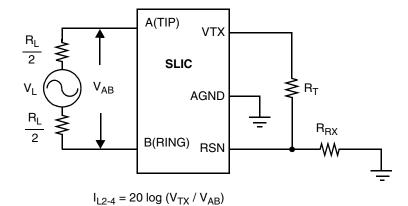
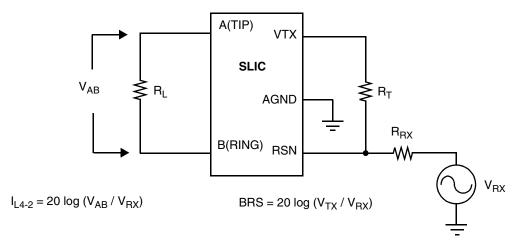


Figure 3. Feed Programming

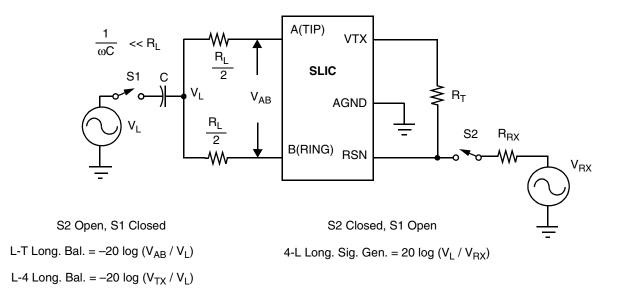
# **TEST CIRCUITS**



A. Two- to Four-Wire Insertion Loss



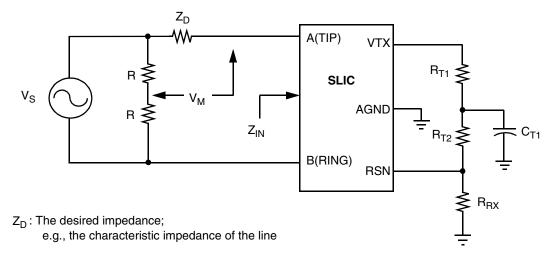
# B. Four- to Two-Wire Insertion Loss and Four- to Four-Wire Balance Return Signal



C. Longitudinal Balance

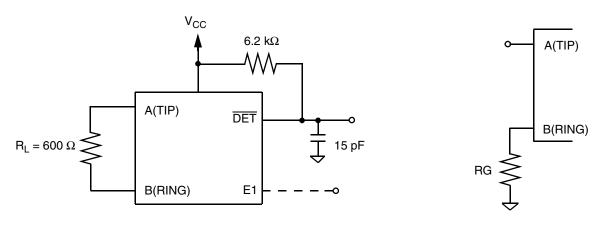


# **TEST CIRCUITS (continued)**



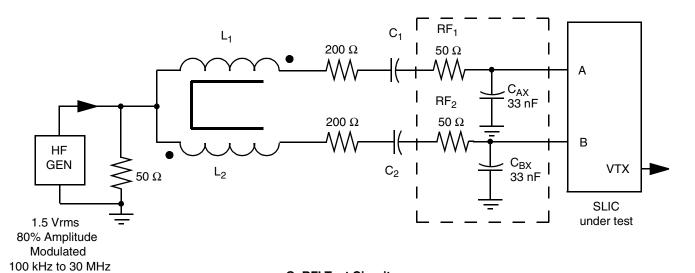
Return loss =  $-20 \log (2 V_M / V_S)$ 

# D. Two-Wire Return Loss Test Circuit



E. Loop-Detector Switching

# F. Ground-Key Switching



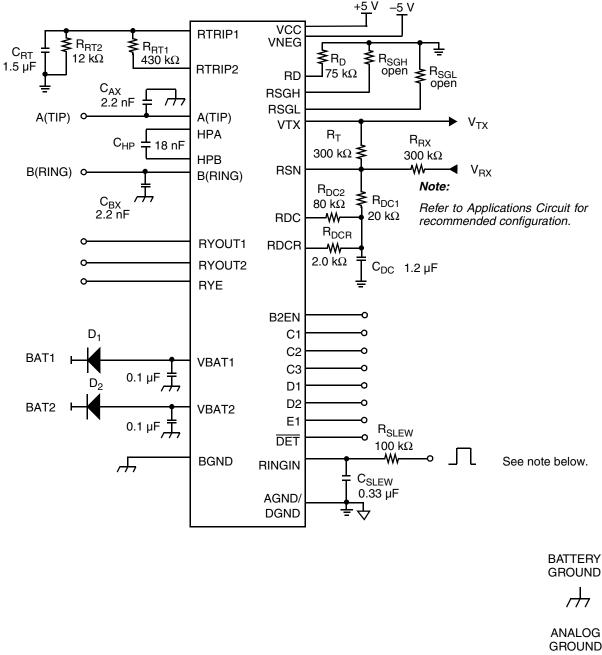
**G. RFI Test Circuit** 



# **TEST CIRCUITS (continued)**

Note:

The input should be 50% duty cycle CMOS-compatible input.



**GROUND** 

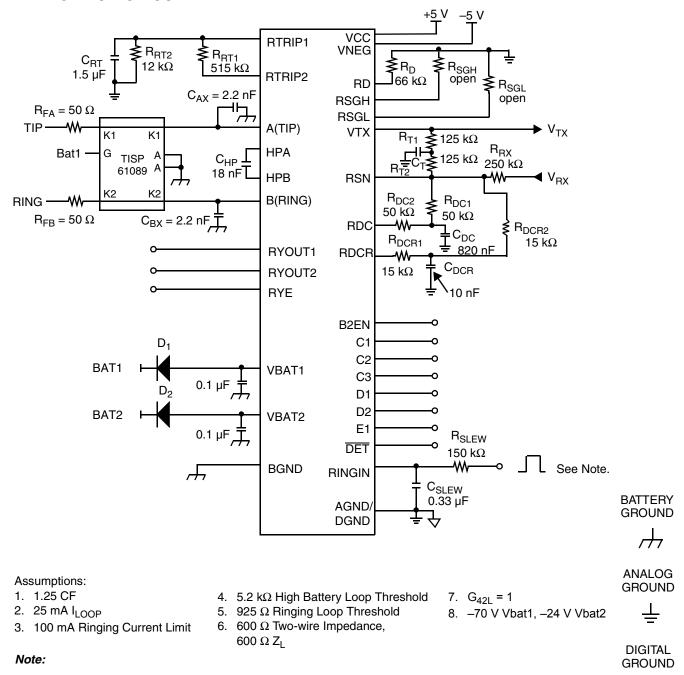
**GROUND** 

DIGITAL **GROUND** 

H. Le79R79 Test Circuit



# **APPLICATION CIRCUIT**



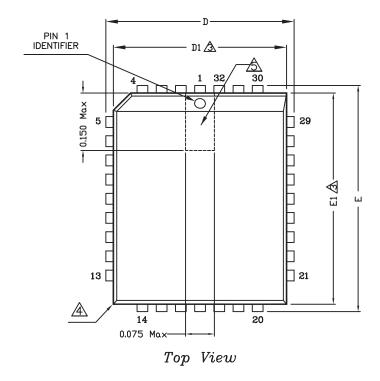
The input should be 50% duty cycle CMOS-compatible input.

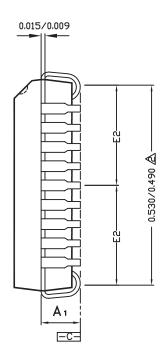
# I. Application Circuit

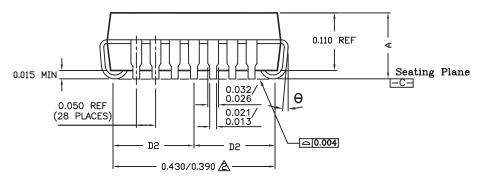


# **PHYSICAL DIMENSIONS**

### 32-Pin PLCC





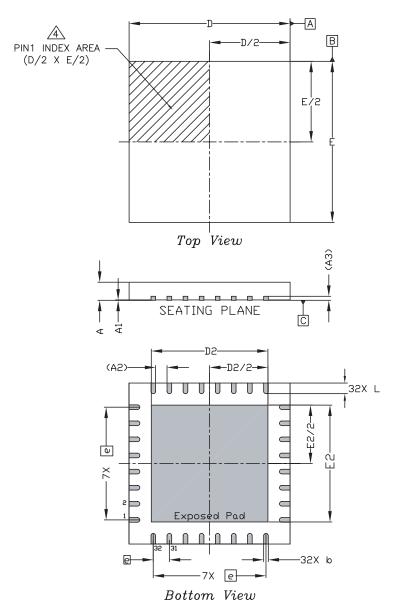


#### NOTE:

- DIMENSIONING AND TOLERANCING CONFORM TO ASME Y14,5M-1994.
- ⚠ TO BE MEASURED AT SEATING PLANE —C— CONTACT POINT.
- △ DIMENSIONS D1 AND E1 DO NOT INCLUDE MOLD PROTRUSION. ALLOWABLE MOLD PROTURSION IS 0.010 IN PER SIDE. DIMENSIONS D, AND E, INCLUDE MOLD MISMATCH AND DETERMINED AT THE PARTING LINE; THAT IS D1 AND E1 ARE MEASURED AT THE EXTREME MATERIAL CONDITION AT THE UPPER OR LOWER PARTING LINE.
- A EXACT SHAPE OF THIS FEATURE IS OPTIONAL.
- △ DETAILS OF PIN 1 IDENTIFIER ARE OPTIONAL BUT MUST BE LOCATED WITHIN THE ZONE INDICATED.
- 6. SUM OF DAM BAR PROTRUSIONS TO BE 0.007 MAX PER LEAD.
- 7. CONTROLLING DIMENSION: INCH.
- 8. REFERENCE DOCUMENT : JEDEC MS-016

Symbol	Dimension in inch			
Syllibol	Min	Nom	Max	
Α	0.125		0.140	
A1	0.075	0.090	0.095	
D	0.485	0.490	0.495	
D1	0.447	0.450	0.453	
D2	0	.205 RE	F	
Ε	0.585	0.590	0.595	
E1	0.547	0.550	0.553	
E2	0.255 REF			
θ	0,		10°	

#### 32-Pin QFN



Symbol	Dime	Dimension in mm			
Syllibol	Min	Nom	Max		
Α	0.80	0.90	1.00		
A2	C	).57 REI	F		
b	0.18	0.23	0.28		
D	8	3.00 BS	0		
D2	5.70	5.80	5.90		
Ε	8	3.00 BS	C		
E2	5.70	5.80	5.90		
е	C	.80 BS			
L	0.43	0.53	0.63		
N	32				
A1	0.00	0.02	0.05		
A3	0.20 REF				

#### NOTES:

- DIMENSIONING AND TOLERANCING CONFORM TO ASME Y14,5M-1994.
- 2. ALL DIMENSIONS ARE IN MILLIMETERS,  $\boldsymbol{\theta}$  IS IN DEGREES.
- 3. N IS THE TOTAL NUMBER OF TERMINALS.
- THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JEP 95-1 SSP-012. DETAILS OF THE TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE.
- 5. COPLANARITY APPLIES TO EXPOSED PAD AS WELL AS THE TERMINALS.
- 6. REFERENCE DOCUMENT : JEDEC MO-220



# **REVISION SUMMARY**

# **Revision B to Revision C**

- Minor changes were made to the data sheet style and format to conform to Legerity standards.
- Electrical Characteristics; Last row under Ring Signal, min changed from 130 to 150, typ changed from 160 to 180, and max changed from 190 to 210.
- SLIC Decoding Table; Added B2EN reference to the Battery Selection column and its corresponding note to the notes section.
- · Applications Circuit; Revised

## **Revision C to Revision D**

Minor changes were made to the data sheet style and format to conform to Legerity standards.

# **Revision D to Revision E**

On pages 17 and 18, R<sub>DC1</sub> and R<sub>DC2</sub> were switched.

### Revision E to Revision F

- The physical dimensions (PL032) were added to the Physical Dimensions section.
- Deleted the Ceramic DIP and Plastic DIP packages and references to them.
- Updated the Pin Description table to correct inconsistencies.

# **Revision F to Revision G**

• The equation on page 13 was changed:

from: 
$$R_{RT1} = 300 \bullet CF \bullet \frac{V_{BAT1}}{Vbat - 3.5 - (15 \ \mu A \bullet 300 \bullet CF \bullet (R_{LRT} + 150 + 2R_F))} \bullet (R_{LRT} + 150 + 2R_F)$$
 to: 
$$R_{RT1} = 320 \bullet CF \bullet \frac{V_{BAT1}}{Vbat - 5 - (24 \ \mu A \bullet 320 \bullet CF \bullet (R_{LRT} + 150 + 2R_F))} \bullet (R_{LRT} + 150 + 2R_F)$$

### Revision G to Revision H

In "Ordering Information" section, added description for wafer foundry facility optional character.

# Revision H to I

- Updated device name from "Am79R79" to "Le79R79" throughout document.
- Added QFN package to "Connection Diagram," "Absolute Maximum Ratings," and "Physical Dimensions."
- Absolute Maximum Ratings: Notes updated to standard.
- Operating Ranges: Temperature statement updated to standard.
- Removed obsolete "Sales Office Listing."
- Removed reference to PLCC package type in "General Description."

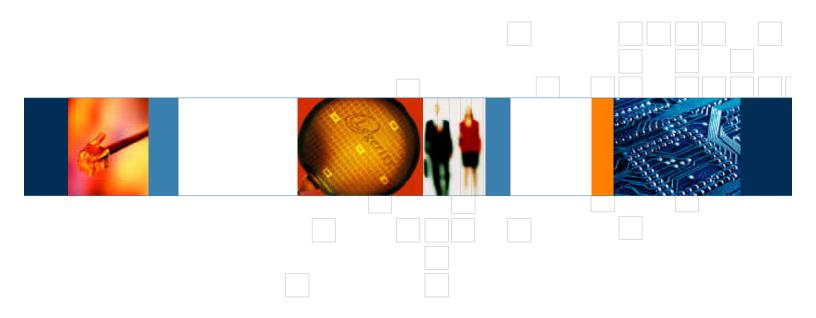
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Legerity P.O. Box 18200 Austin, Texas 78760-8200

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