

# Ultracompact, Precision 10.0 V/5.0 V/2.5 V/3.0 V Voltage References

# Data Sheet **ADR01/ADR02/ADR03/ADR06**

### **FEATURES**

**Ultracompact SC70 and TSOT packages Low temperature coefficient 8-lead SOIC: 3 ppm/°C 5-lead SC70: 9 ppm/°C 5-lead TSOT: 9 ppm/°C Initial accuracy ±0.1% No external capacitor required Low noise 10 µV p-p (0.1 Hz to 10.0 Hz) Wide operating range ADR01: 12.0 V to 36.0 V ADR02: 7.0 V to 36.0 V ADR03: 4.5 V to 36.0 V ADR06: 5.0 V to 36.0 V High output current 10 mA Wide temperature range: −40°C to +125°C ADR01/ADR02/ADR03 pin compatible to industrystandard REF01/REF02/REF03 ADR01, ADR02, ADR03 and ADR06 SOIC qualified for automotive applications**

# **PIN CONFIGURATIONS**



*Figure 1. 5-Lead, SC70/TSOT Surface-Mount Packages*



*Figure 2. 8-Lead, SOIC Surface-Mount Package*

### **APPLICATIONS**

**Precision data acquisition systems High resolution converters Industrial process control systems Precision instruments Auto battery monitoring PCMCIA cards**

## **GENERAL DESCRIPTION**

The ADR01, ADR02, ADR03, and ADR06 are precision 10.0 V, 5.0 V, 2.5 V, and 3.0 V band gap voltage references featuring high accuracy, high stability, and low power consumption. The parts are housed in tiny, 5-lead SC70 and TSOT packages, as well as in 8-lead SOIC versions. The SOIC versions of the ADR01, ADR02, and ADR03 are drop-in replacements<sup>1</sup> to the industrystandard REF01, REF02, and REF03. The small footprint and wide operating range make the ADR0x references ideally suited for general-purpose and space-constrained applications.

With an external buffer and a simple resistor network, the TEMP terminal can be used for temperature sensing and approximation. A TRIM terminal is provided on the devices for fine adjustment of the output voltage.

The ADR01, ADR02, ADR03, and ADR06 are compact, low drift voltage references that provide an extremely stable output voltage from a wide supply voltage range. They are available in 5-lead SC70 and TSOT packages, and 8-lead SOIC packages with A, B, and C grade selections. All parts are specified over the extended industrial (–40°C to +125°C) temperature range. The ADR01, ADR02, ADR03, and ADR06 A grade in 8-lead SOIC are qualified for automotive applications.

#### **Table 1. Selection Guide**



<sup>1</sup> ADR01, ADR02, and ADR03 are component-level compatible with REF01, REF02, and REF03, respectively. No guarantees for system-level compatibility are implied. SOIC versions of ADR01/ADR02/ADR03 are pin-to-pin compatible with 8-lead SOIC versions of REF01/REF02/REF03, respectively, with the additional temperature monitoring function.

#### **Rev. R**

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# **SPECIFICATIONS**

### **ADR01 ELECTRICAL CHARACTERISTICS**

 $V_{\text{IN}}$  = 12.0 V to 36.0 V,  $V_{\text{IN}}$  = 12.0 V to 24.0 V for ADR01WARZ, T<sub>A</sub> = 25°C, unless otherwise noted.

# **Table 2.**



<sup>1</sup> The long-term stability specification is noncumulative. The drift in subsequent 1000 hour periods is significantly lower than in the first 1000 hour period.

## **ADR02 ELECTRICAL CHARACTERISTICS**

 $V_{\text{IN}} = 7.0$  V to 36.0 V,  $V_{\text{IN}} = 7.0$  V to 26.0 V for ADR02WARZ,  $T_A = 25^{\circ}$ C, unless otherwise noted.





<sup>1</sup> The long-term stability specification is noncumulative. The drift in subsequent 1000 hour periods is significantly lower than in the first 1000 hour period.

### **ADR03 ELECTRICAL CHARACTERISTICS**

 $V_{\text{IN}} = 4.5$  V to 36.0 V,  $V_{\text{IN}} = 4.5$  V to 26.0 V for ADR03WARZ,  $T_A = 25$ °C, unless otherwise noted.

#### **Table 4.**



<sup>1</sup> The long-term stability specification is noncumulative. The drift in subsequent 1000 hour periods is significantly lower than in the first 1000 hour period.

### **ADR06 ELECTRICAL CHARACTERISTICS**

 $V_{\text{IN}} = 5.0$  V to 36.0 V,  $V_{\text{IN}} = 5.0$  V to 26.0 V for ADR06WARZ,  $T_A = 25^{\circ}$ C, unless otherwise noted.





<sup>1</sup> The long-term stability specification is noncumulative. The drift in subsequent 1000 hour periods is significantly lower than in the first 1000 hour period.

# ABSOLUTE MAXIMUM RATINGS

Ratings are at 25°C, unless otherwise noted.

#### **Table 6.**



Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **THERMAL RESISTANCE**

 $\theta_{JA}$  is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages.

#### **Table 7. Thermal Resistance**



### **ESD CAUTION**



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

# **TERMINOLOGY**

#### **Dropout Voltage (V<sub>DO</sub>)**

Dropout voltage, sometimes referred to as supply voltage headroom or supply output voltage differential, is defined as the minimum voltage differential between the input and output necessary for the device to operate, such as

 $V_{DO} = (V_{IN} - V_{OUT}) \min | I_L = Constant$ 

Because the dropout voltage depends upon the current passing through the device, it is always specified for a given load current.

#### **Temperature Coefficient (TCV<sub>O</sub>)**

The temperature coefficient relates the change in output voltage to the change in ambient temperature of the device, as normalized by the output voltage at 25°C. This parameter is expressed in ppm/°C and can be determined by the following equation:

$$
TCV_{O} = \frac{V_{OUT}(T_2) - V_{OUT}(T_1)}{V_{OUT}(25^{\circ}C) \times (T_2 - T_1)} \times 10^6 \text{ [ppm/°C]}
$$

where:

*VOUT(25°C)* is the output voltage at 25°C.  $V_{OUT}(T_1)$  is the output voltage at Temperature 1.  $V_{\text{OUT}}(T_2)$  is the output voltage at Temperature 2.

#### **Output Voltage Hysteresis (ΔVOUT\_HYS)**

Output voltage hysteresis represents the change in output voltage after the device is exposed to a specified temperature cycle. This may be expressed as either a shift in voltage or a difference in parts per million from the nominal output as follows:

 $V_{OUT\_HYS} = V_{OUT}(25^{\circ}C) - V_{OUT\_TC}$  [V]

$$
V_{OUT\_HYS} = \frac{V_{OUT}(25^{\circ}C) - V_{OUT\_TC}}{V_{OUT}(25^{\circ}C)} \times 10^{6} \text{ [ppm]}
$$

#### where:

*VOUT(25°C)* is the output voltage at 25°C.

*VOUT\_TC* is the output voltage after temperature cycling.

Thermal hysteresis occurs as a result of forces exhibited upon the internal die by its packaging. The effect is more pronounced in parts with smaller packages.

#### **Long-Term Stability (ΔV<sub>OUT\_LTD</sub>)**

Long-term stability refers to the shift in output voltage at 25°C after 1000 hours of operation in a 25°C environment. This may also be expressed as either a shift in voltage or a difference in parts per million from the nominal output as follows:

$$
\Delta V_{OUT\_LTD} = |\text{V}_{OUT}(t_1) - V_{OUT}(t_0)| \text{ [V]}
$$

$$
\Delta V_{OUT\_LTD} = \left| \frac{V_{OUT}(t_1) - V_{OUT}(t_0)}{V_{OUT}(t_0)} \right| \times 10^6 \text{ [ppm]}
$$

where,

 $V_{\text{OUT}}(t_0)$  is the  $V_{\text{OUT}}$  at 25°C at Time 0.

 $V_{\text{OUT}}(t_1)$  is the  $V_{\text{OUT}}$  at 25°C after 1000 hours of operation at 25°C.

#### **Line Regulation**

Line regulation refers to the change in output voltage in response to a given change in input voltage, and is expressed in either percent per volt, parts per million per volt, or microvolt per volt change in input voltage. This parameter accounts for the effects of self-heating.

#### **Load Regulation**

Load regulation refers to the change in output voltage in response to a given change in load current, and is expressed in either microvolts per milliampere, parts per million per milliampere, or ohms of dc output resistance. This parameter accounts for the effects of self-heating.

# TYPICAL PERFORMANCE CHARACTERISTICS



*Figure 3. ADR01 Typical Output Voltage vs. Temperature*



*Figure 4. ADR02 Typical Output Voltage vs. Temperature*



**3.002 3.001 VOUT (V) 3.000 2.999 2.998**  $\overline{\rm g}$ **–40 –25 –10 5 20 35 50 65 80 95 110 125** 02747-007 02747 **TEMPERATURE (°C)**

*Figure 6. ADR06 Typical Output Voltage vs. Temperature*







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*Figure 9. ADR03 Supply Current vs. Input Voltage*



*Figure 10. ADR06 Supply Current vs. Input Voltage*



*Figure 11. ADR01 Load Regulation vs. Temperature*



*Figure 12. ADR02 Load Regulation vs. Temperature*



*Figure 13. ADR03 Load Regulation vs. Temperature*



*Figure 14. ADR06 Load Regulation vs. Temperature*











*Figure 23. ADR01 Quiescent Current vs. Load Current*



*Figure 24. ADR02 Typical Noise Voltage 0.1 Hz to 10.0 Hz*



*Figure 25. ADR02 Typical Noise Voltage 10 Hz to 10 kHz* 



*Figure 26. ADR02 Line Transient Response*



*Figure 27. ADR02 Load Transient Response* 



*Figure 28. ADR02 Load Transient Response*



*Figure 29. ADR02 Turn-Off Response*



*Figure 30. ADR02 Turn-On Response*



*Figure 31. ADR02 Turn-Off with No Input Capacitor*



*Figure 32. ADR02 Turn-Off with No Input Capacitor*

# APPLICATIONS INFORMATION **OVERVIEW**

The ADR01/ADR02/ADR03/ADR06 are high precision, low drift 10.0 V, 5.0 V, 2.5 V, and 3.0 V voltage references available in an ultracompact footprint. The 8-lead SOIC versions of the devices are drop-in replacements of the REF01/REF02/REF03 sockets with improved cost and performance.

These devices are standard band gap references (see Figure 34). The band gap cell contains two NPN transistors (Q18 and Q19) that differ in emitter area by  $2 \times$ . The difference in their  $V_{BE}$ produces a proportional-to-absolute temperature current (PTAT) in R14, and, when combined with the  $V_{BE}$  of Q19, produces a band gap voltage,  $V_{BG}$ , that is almost constant in temperature. With an internal op amp and the feedback network of R5 and R6,  $V<sub>o</sub>$  is set precisely at 10.0 V, 5.0 V, 2.5 V, and 3.0 V for the ADR01, ADR02, ADR06, and ADR03, respectively. Precision laser trimming of the resistors and other proprietary circuit techniques are used to further enhance the initial accuracy, temperature curvature, and drift performance of the ADR01/ ADR02/ADR03/ADR06.

The PTAT voltage is made available at the TEMP pin of the ADR01/ADR02/ADR03/ADR06. It has a stable 1.96 mV/°C temperature coefficient, such that users can estimate the temperature change of the device by knowing the voltage change at the TEMP pin.

## **APPLYING THE ADR01/ADR02/ADR03/ADR06** *Input and Output Capacitors*

Although the ADR01/ADR02/ADR03/ADR06 are designed to function stably without any external components, connecting a 0.1 μF ceramic capacitor to the output is highly recommended to improve stability and filter out low level voltage noise. An additional 1 μF to 10 μF electrolytic, tantalum, or ceramic capacitor can be added in parallel to improve transient performance in response to sudden changes in load current; however, the designer should keep in mind that doing so increases the turn-on time of the device.

A 1 μF to 10 μF electrolytic, tantalum or ceramic capacitor can also be connected to the input to improve transient response in applications where the supply voltage may fluctuate. An additional 0.1 μF ceramic capacitor should be connected in parallel to reduce supply noise. Mount both input and output capacitors as close to the device pins as possible.

### *Output Adjustment*

The ADR01/ADR02/ADR03/ADR06 trim terminal can be used to adjust the output voltage over a nominal voltage. This feature allows a system designer to trim system errors by setting the reference to a voltage other than 10.0 V/5.0 V/2.5 V/3.0 V. For finer adjustment, add a series resistor of 470 kΩ. With the configuration shown in Figure 35, the ADR01 can be adjusted from 9.70 V to 10.05 V, the ADR02 can be adjusted from 4.95 V to 5.02 V, the ADR06 can be adjusted from 2.8 V to 3.3 V, and the

ADR03 can be adjusted from 2.3 V to 2.8 V. Adjustment of the output does not significantly affect the temperature performance of the device, provided the temperature coefficients of the resistors are relatively low.











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### *Temperature Monitoring*

As described at the end of the Overview section, the ADR01/ ADR02/ADR03/ADR06 provide a TEMP output (Pin 1 in Figure 1 and Pin 3 in Figure 2) that varies linearly with temperature. This output can be used to monitor the temperature change in the system. The voltage at  $V<sub>TEMP</sub>$  is approximately 550 mV at 25°C, and the temperature coefficient is approximately 1.96 mV/°C (see Figure 36). A voltage change of 39.2 mV at the TEMP pin corresponds to a 20°C change in temperature.

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The TEMP function is provided as a convenience rather than a precise feature. Because the voltage at the TEMP node is acquired from the band gap core, current pulling from this pin has a significant effect on V<sub>OUT</sub>. Care must be taken to buffer the TEMP output with a suitable low bias current op amp, such as the AD8601, AD820, or OP1177, all of which result in less than a 100 µV change in ∆V<sub>OUT</sub> (see Figure 37). Without buffering, even tens of microamps drawn from the TEMP pin can cause V<sub>OUT</sub> to fall out of specification.





#### **LOW COST CURRENT SOURCE**

Unlike most references, the ADR01/ADR02/ADR03/ADR06 employ an NPN Darlington in which the quiescent current remains constant with respect to the load current, as shown in Figure 23*.* As a result, a current source can be configured as shown in Figure 38 where  $I_{\text{SET}} = (V_{\text{OUT}} - V_{\text{L}})/R_{\text{SET}}$ . I<sub>L</sub> is simply the sum of Iser and I<sub>Q</sub>. Although simple, I<sub>Q</sub> varies typically from 0.55 mA to 0.65 mA, limiting this circuit to general-purpose applications.



*Figure 38. Low Cost Current Source*

### **PRECISION CURRENT SOURCE WITH ADJUSTABLE OUTPUT**

Alternatively, a precision current source can be implemented with the circuit shown in Figure 39. By adding a mechanical or digital potentiometer, this circuit becomes an adjustable current source. If a digital potentiometer is used, the load current is simply the voltage across Terminal B to Terminal W of the digital potentiometer divided by RSET.

$$
I_L = \frac{V_{REF} \times D}{R_{SET}} \tag{1}
$$

where *D* is the decimal equivalent of the digital potentiometer input code.



*Figure 39. Programmable 0 mA to 5 mA Current Source*

To optimize the resolution of this circuit, dual-supply op amps should be used because the ground potential of ADR02 can swing from -5.0 V at zero scale to VL at full scale of the potentiometer setting.

### **PROGRAMMABLE 4 mA TO 20 mA CURRENT TRANSMITTER**

Because of their precision, adequate current handling, and small footprint, the devices are suitable as the reference sources for many high performance converter circuits. One of these applications is the multichannel 16-bit, 4 mA to 20 mA current transmitter in the industrial control market (see Figure 40). This circuit employs a Howland current pump at the output to yield better efficiency, a lower component count, and a higher voltage compliance than the conventional design with op amps and MOSFETs. In this circuit, if the resistors are matched such that  $R1 = R1'$ ,  $R2 = R2'$ ,  $R3 = R3'$ , the load current is

$$
I_L = \frac{(R2 + R3)/R1}{R3'} \times \frac{V_{REF} \times D}{2^N}
$$
 (2)

where *D* is similarly the decimal equivalent of the DAC input code and *N* is the number of bits of the DAC.

According to Equation 2, R3′ can be used to set the sensitivity. R3′ can be made as small as necessary to achieve the current needed within U4 output current driving capability. Alternatively, other resistors can be kept high to conserve power.

In this circuit, the AD8512 is capable of delivering 20 mA of current, and the voltage compliance approaches 15.0 V.



*Figure 40. Programmable 4 mA to 20 mA Transmitter*

The Howland current pump yields a potentially infinite output impedance, that is highly desirable, but resistance matching is critical in this application. The output impedance can be determined using Equation 3. As shown by this equation, if the resistors are perfectly matched,  $Z<sub>O</sub>$  is infinite. Alternatively, if they are not matched,  $Z<sub>O</sub>$  is either positive or negative. If the latter is true, oscillation can occur. For this reason, connect Capacitor C1 in the range of 1 pF to 10 pF between VP and the output terminal of U4 to filter any oscillation.

$$
Z_O = \frac{V_t}{I_t} = \frac{RI'}{\left(\frac{RI'R2}{RIR2'} - 1\right)}
$$
(3)

In this circuit, an ADR01 provides the stable 10.000 V reference for the AD5544 quad 16-bit DAC. The resolution of the adjustable current is 0.3 µA/step; the total worst-case INL error is merely 4 LSBs. Such error is equivalent to 1.2 µA or a 0.006% system error, which is well below most systems' requirements. The result is shown in Figure 41 with measurement taken at 25°C and 70°C; total system error of 4 LSBs at both 25°C and 70°C.



*Figure 41. Result of Programmable 4 mA to 20 mA Current Transmitter*

### **PRECISION BOOSTED OUTPUT REGULATOR**

A precision voltage output with boosted current capability can be realized with the circuit shown in Figure 42. In this circuit, U2 forces  $V_0$  to be equal to  $V_{REF}$  by regulating the turn-on of N1, thereby making the load current furnished by  $V_{IN}$ . In this configuration, a 50 mA load is achievable at  $V_{IN}$  of 15.0 V. Moderate heat is generated on the MOSFET, and higher current can be achieved with a replacement of a larger device. In addition, for a heavy capacitive load with a fast edging input signal, a buffer should be added at the output to enhance the transient response.



*Figure 42. Precision Boosted Output Regulator*

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# OUTLINE DIMENSIONS



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# **ORDERING GUIDES**

### *ADR01 Ordering Guide*



 $1 Z =$  RoHS Compliant Part.

 $2$  W = Qualified for automotive applications.

### *ADR02 Ordering Guide*



 $1 Z =$  RoHS Compliant Part.

 $2$  W = Qualified for automotive applications.

#### *ADR03 Ordering Guide*



 $1 Z =$  RoHS Compliant Part.

 $2 W =$  Qualified for automotive applications.

#### *ADR06 Ordering Guide*



 $1 Z =$  RoHS Compliant Part.

 $2$  W = Qualified for automotive applications.

## **AUTOMOTIVE PRODUCTS**

The ADR01W, ADR02W, ADR03W and ADR06W models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models; therefore, designers should review the Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific Automotive Reliability reports for these models.

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