

# ISOFACE™

# ISO11813T

Isolated 8 Channel Digital Input with IEC61131-2 Type 1/2/3 Characteristics

# **Preliminary Data Sheet**

Revision 1.0, 2011-03-30

# Industrial & Multimarket

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# Isolated 8 Channel Digital Input with IEC61131-2 Type 1/2/3 Characteristics

## **Product Highlights**

- Minimization of power dissipation due to constant current characteristic
- Status LED output for each input
- Digital averaging of the input signals to suppress interference pulses
- Isolation between Input and Output using Coreless Transformer Technology

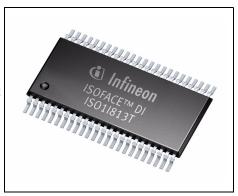


- Complete system integration (digital sensor or switch input, galvanic isolation and intelligent micro-controller or bus-ASIC interface
- 8-channel input according to IEC61131-2 (Type 1/2/3)
- Integrated galvanic isolation 500VAC (EN60664-1, UL508)
- 5/3.3V SPI and parallel micro-controller interface
- Adjustable deglitching filters
- Up to 500 kHz sampling frequency
- · Wire-break detection
- Vbb under-voltage detection
- Package: TSSOP 8 x 12.5 mm

#### **Typical Application**

Programmable Logic Controllers(PLC)
Industrial PC

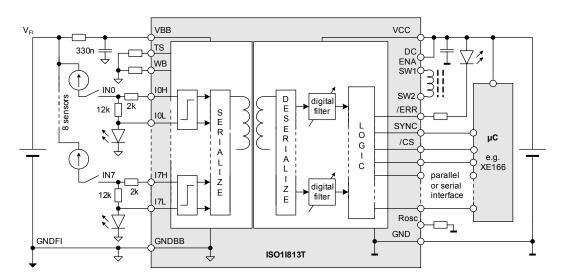
General Control Equipment



#### Description

The ISO1I813T is an electrically isolated 8 bit data input interface in TSSOP-48 package.

This part is used to detect the signal states of eight independent input lines according to IEC61131-2 Type 1/2/3 (e.g. two-wire proximity switches) with a common ground (GNDFI). For operating sensors of type 1/2/3 in accordance with IEC61131-2, it is necessary for the device to be wired with resistors  $R_V$  and  $R_{EXT}$  (it is recommended to use resistors with an accuracy of 2%, in any case < 5% - is mandatory, temperaturecoefficients < 200ppm are allowed). The figure below shows the typical application circuit for type 1/3. An 8 bit parallel µC compatible interface allows to connect the IC directly to a µC system. The input interface supports also a direct control mode and is designed to operate with 3.3/5V CMOS compatible levels. The data transfer from input to output side is realized by the integrated Coreless Transformer Technology.



Typical Application for Sensor of Type 1/3



#### **Pin Configuration and Functionality**

# 1 Pin Configuration and Functionality

The pin configuration slightly differs for the parallel or the serial interfaces.

# 1.1 Pin Configuration

The ordering, type and functions of the IC pins are listed in the Table 1.

Table 1 Pin Configuration

Pin	Parallel Ir	le	Serial Interface Mode					
	Symbol	Ctrl	Type <sub>2)</sub>	Function	Symbol	Ctrl.	Туре	Function
ASIC	/ Micro-con	troller	Side F	rins		-		-
1	GND		Α	Logic Ground	GND			
2	SEL	I	PD	Serial Parallel Mode Select	SEL			
3	SYNC	I	PU	Freeze Data & Diagnostics	SYNC			
4	Rosc		Α	Clock Frequency Adjustment	Rosc			
5	VCC		Α	Positive 5/3.3V logic supply	VCC			
6	ERR	0	D, PU	Fault Indication output	ERR			
7	GND		Α	Logic Ground	GND			
8	AD0	Ю	PPZ	Data output bit0	SDI	I	PD	SPI Data input
9	AD1	Ю	PPZ	Data output bit1	SSO	0	PPZ	SPI Status output
10	AD2	Ю	PPZ	Data output bit2	GND			
11	AD3	Ю	PPZ	Data output bit3	GND			
12	AD4	Ю	PPZ	Data output bit4	CRCERR	0	D, PU	CRC Error output
13	AD5	Ю	PPZ	Data output bit5	SCLK	I	PD	SPI Shift Clock input
14	AD6	Ю	PPZ	Data output bit6	SSI	I	PD	SPI Status input
15	AD7	Ю	PPZ	Data output bit7	SDO	О	PPZ	SPI Data output
16	CS	I	PU	Chip Select	CS			
17	RD	I	PU	Data Read	n.c.			
18	GND		Α	Logic Ground	GND		_	
19	WR	I	PU	Data Write	MS0	I	PD	SPI Mode Select bit 0
20	ALE	I	PD	Address Latch Enable	MS1	I	PD	SPI Mode Select bit 1
21	DC_ENA	I	PD	DC-DC Supply Enable	DC_ENA			
22	SW1		Α	DC-DC Switch Output 1	SW1			
23	SW2		Α	DC-DC Switch Output 2	SW2			
24	GND		Α	Logic Ground	GND			
	or Side Pins	3	1		<del> </del>	1		
25	GNDBB		Α	Input Ground	GNDBB			
26	VBB		Α	Positive input supply voltage	VBB			
27	IOL		Α	Input 0 Low, LED Out	IOL			
28	I0H		Α	Input 0 High	I0H			
29	I1L		Α	Input 1 Low, LED Out	I1L			



#### **Pin Configuration and Functionality**

Table 1 Pin Configuration

Pin	Parallel I	nterfa	се Мос	de	Serial Interface Mode				
	Symbol	Ctrl	Type <sub>2)</sub>	Function	Symbol	Ctrl.	Туре	Function	
30	I1H		Α	Input 1 High	I1H				
31	GNDBB		Α	Input Ground	GNDBB				
32	I2L		Α	Input 2 Low, LED Out	I2L				
33	I2H		Α	Input 2 High	I2H				
34	I3L		Α	Input 3 Low, LED Out	I3L				
35	I3H		Α	Input 3 High	I3H				
36	TS		Α	Sensor Type 1/2/3 Select	TS				
37	GNDBB		Α	Input Ground	GNDBB				
38	WB		Α	Wire Break Select	WB				
39	I4L		Α	Input 4 Low, LED Out	I4L				
40	I4H		Α	Input 4 High	I4H				
41	I5L		Α	Input 5 Low, LED Out	I5L				
42	I5H		Α	Input 5 High	I5H				
43	GNDBB		Α	Input Ground	GNDBB				
44	I6L		Α	Input 6 Low, LED Out	I6L				
45	I6H		Α	Input 6 High	I6H				
46	I7L		Α	Input 7 Low, LED Out	I7L				
47	I7H		Α	Input 7 High	I7H				
48	GNDBB		Α	Input Ground	GNDBB				

<sup>1)</sup> Direction of the pin: I = input, O = output, IO = Input/Output

<sup>2)</sup> Type of the pin: A = analog, D = Open-Drain, PU = internal Pull-Up resistor, PD = internal Pull-Down resistor, PPZ = Push-Pull pin with High-Impedance functionality



#### GND GNDBB GND 1 () GNDBB 48 48 SEL 47 17H SEL 47 I7H SYNC SYNC 3 I7L\_ 17L 46 I6H Rosc 16H Rosc 45 VCC I6L VCC I6L /ERR GNDBB /ERR 6 43 GNDBB GND I5H GND I5H AD0 15L SDI 41 I5L AD1 40 I4H SSO 40 I4H AD2 I4L GND I4L 11 WB GND 11 WB AD3 38 AD4 12 GNDBB 12 GNDBB CRCERR 37 Pinout for parallel 37 Pinout for serial Interface Interface TS TS AD5 13 SCLK 13 36 36 13H AD6 I3H SSI 35 34 I3L SDO 34 I3L I2H /CS 12H /RD 17 32 I2L 32 I2L nc GNDBB GNDBB GND 18 GND 31 31 WR 30 I1H MS0 30 I1H ALE 20 29 I1L MS1 20 29 I1L DC\_ENA 21 I0H DC\_ENA 21 10H SW1 IOL SW1 27 I0L SW2 23 VBB 23 26 VBB SW2 24 25 25 GNDBB GND GND

n.c. = Not Connected

#### Pin Configuration and Functionality

Figure 1 TSSOP-48 Pinout for Parallel and Serial Interface Modes

#### 1.2 Pin Functionality

#### 1.2.1 Pins of Sensor Interface

#### VBB (Positive supply 9.6-35V sensor supply)

VBB supplies the sensor input stage.

#### **GNDBB** (Ground for VBB domain)

This pin acts as the ground reference for the sensor input stage that is supplied by VBB.

#### I0H... I7H (Input channel 0 ... 7)

Sensor inputs with current sink characteristic according IEC61131-2 Type 1/2/3 which has been selected by pin TS

#### IOL... I7L (LED output channel 0 ... 7)

This pin provides the output signal to switch on the LED if the input voltage and current has been detected as "High" according to the selected Sensor Type.

#### WB (Wire-Break Select)

By connecting a resistor between WB and GNDBB, the level for the Wire-Break detection can be adjusted (refer to **Table 10**). This pin is for static configuration (pin-strapping). The input voltage must not change during operation.

#### TS (Type Select)

By connecting a resistor between TS and GNDBB the sensor type (Type 1/2/3) can be selected (refer to **Table 10** for corresponding resistor value). This pin is for static configuration (pin-strapping). The input voltage must not change during operation.

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#### Pin Configuration and Functionality

#### 1.2.2 Pins of Serial and Parallel logic Interface

Some pins are common for both interface types, some others are specific for the parallel or serial access.

#### VCC (Positive 3.3/5V logic supply)

VCC supplies the output interface that is electrically isolated from the sensor input stage. The interface can be supplied with 3.3/5V.

#### **GND (Ground for VCC domain)**

This pin acts as the ground reference for the uC-interface that is supplied by VCC.

#### **Rosc (Clock Adjustment)**

A high precision resistor has to be connected between Rosc and GND to set the frequency of the sampling clock.

#### DC\_ENA (DC-DC Converter Enable)

When the DC\_ENA pin is connected to VCC, the internal DC-DC driver is activated. When DC\_ENA is in the state Low, the switches are not driven. The input voltage must not change during operation. This pin has an internal Pull-Down resistor.

#### SW1, SW2 (DC-DC switch outputs 1/2)

When the DC\_ENA pin is connected to VCC, the outputs SW1 and SW2 switch at the clock-frequency determined by the resistor at Rosc to supply the external push-pull converter. The switching frequency can be divided by two by setting the responsible bit in the GLCFG register (see also section 6). Both outputs provide an open drain functionality.

#### ERR (Error)

The low active ERR signal contains the OR-wired information of the sensor input undervoltage and missing voltage detection, the internal data transmission failure detection unit and the overcurrent fault of the DC-DC-converter. The output pin ERR provides an open drain functionality. During Start Up this pin is pulled to High state. This pin has an internal Pull-Up resistor. In normal operation the signal ERR is high. See Section 3.5 for more details.

#### **SEL (Serial or Parallel Mode Select)**

When this pin is in a logic Low state, the IC operates in Parallel Mode. For Serial Mode operation the pin has to be pulled in logic High state. During Start Up the IC is operating in Parallel Mode. This pin has an internal Pull-Down resistor. This pin must not change during operation.

#### SYNC

When this pin is in a logic High state, the IC operates in continuous mode with the internal sampling clock. In isochronous mode, the internal data and diagnostics registers are synchronized on each falling edge detected at SYNC. In logic low state the internal data and diagnostic registers are not updated. During Start-Up this pin is pulled to high state. This pin has an internal Pull-Up resistor. (see also Section 3.9

#### CS (Chip Select)

When this pin is in a logic Low state, the IC interface is enabled and data can be transferred. This pin has an internal Pull-Up resistor.

#### The following pins are provided in the parallel interface mode

#### AD7:AD0 (AddressData input / output bit7 ... bit0)

The pins  $\overline{AD0}$  ..  $\overline{AD7}$  are the bidirectional input / outputs for data write and read. Depending on the state of the ALE,  $\overline{RD}$ ,  $\overline{WR}$  pins, register addresses or data can be transferred between the internal registers and e.g. the microcontroller.

#### RD, WR (Read / Write)

By pulling one of these pins down, a read or write transaction is initiated on the AddressData bus and the data becomes valid. These pins have internal Pull-Up resistors.



#### Pin Configuration and Functionality

#### **ALE (Address Latch Enable)**

The pin ALE is used to select between address (ALE is in a logic High state) or data (ALE is in a logic Low state). When ALE is pulled high, addresses are transferred and latched over the bit AD0 to AD7. During the Low State of ALE all read or write transactions hit the same adress. This pin has an internal Pull-Down resistor.

#### The following pins are provided in the serial interface mode

#### MS0, MS1 (Serial Mode Select)

By driving these pins to Logic High or Low the Serial Interface Mode can be selected. These pins have internal Pull-Down resistors. The mode of the Serial Interface can be changed during operation.

#### SCLK (Serial interface shift clock)

Input data are sampled with the rising edge and output data are updated with the falling edge of this input clock signal. This pin has an internal Pull-Down resistor.

#### SDI, SSI (Serial interface data/status input)

SDI/SSI data is put into a dedicated FIFO to program the filtering time and mask the Wire-Break diagnostic bits of each channel (SPI Mode 2 and 3) or to program register adresses the contents of which to be read subsequently. This pin has an internal Pull-Down resistor.

#### SDO, SSO (Serial interface data/status outputs)

SDO provides typically the sensor data bits, SSO provides the sensor diagnostic bits.

#### **CRCERR (CRC Error output)**

This pin is in a logic Low state when CRC errors or Shift-Clock errors are detected internally. This pin has an internal Pull-Up resistor.

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**Blockdiagram** 

# 2 Blockdiagram

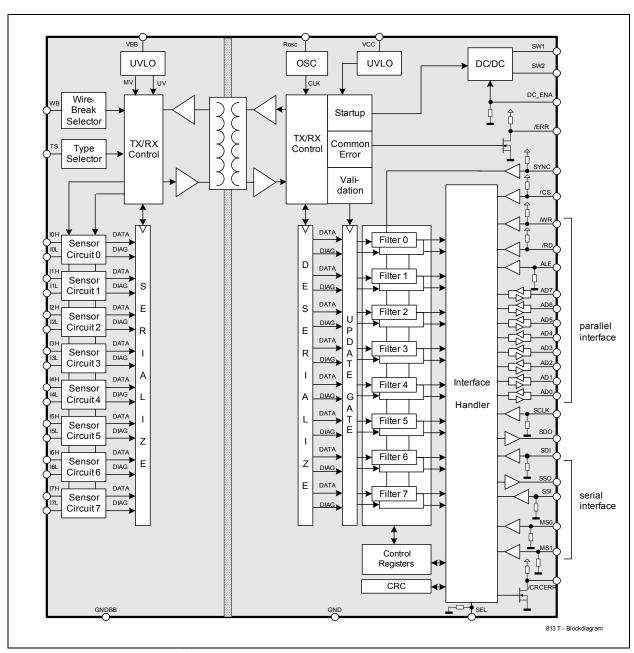


Figure 2 Block Diagram ISO1I813T



# 3 Functional Description

The ISO1I813T is an electrically isolated 8 bit data input interface. This part is used to detect the signal states of eight independent input lines according to IEC61131-2 Type 1/2/3 (e.g. two-wire proximity switches) with a common ground (GNDBB).

#### 3.1 Introduction

The current in the input circuit is determined by the switching element in state "0" and by characteristics of the input stage in state "1".

The octal input device is intended for a configuration comprising two specified external resistors per channel, as shown in the block diagram. As a result the power dissipation within the package is at a minimum.

The voltage dependent current through the external resistor  $R_{EXT}$  is compensated by a negative differential resistance of the current sink across pins IxH and IxL, therefore input INx behaves like a constant current sink.

The comparator assigns level 1 or 0 to the voltage present at input I. To improve interference protection, the comparator is provided with hysteresis. A status LED is connected in series with the input circuit ( $R_{\text{EXT}}$  and current sink).

If no LED is used an external resistor of 2 k $\Omega$  has to be connected between IxL and GNDBB. The specified switching thresholds may change if the LED is replaced by a resistor.

The LED drive short-circuits the status LED if the comparator detects "0". A constant current sink in parallel with the LED reduces the operating current of the LED, and a voltage limiter ensures that the input circuit remains operational if the LED is interrupted but the switching thresholds may change.

For each channel an adjustable digital filter is provided which samples the comparator signal at a rate configured by programming internal registers. The digital filter is designed to provide averaging characteristics. If the input value remains the same for the selected number of sampling values than, the output changes to the corresponding state.

The  $\mu$ C compatible interfaces allow a direct connection to the ports of a microcontroller without the need for other components. The diagnostic logic on the chip monitors the internal data transfer as well as the sensor input supply. The information is send via the internal coreless transformer to the pin  $\overline{\text{ERR}}$  at the input interface

#### 3.2 Power Supply

The IC contains 2 electrically isolated voltage domains that are independent from each other. The microcontroller interface is supplied via pin VCC, GND and the input stage is supplied via pin VBB, GNDBB. The different voltage domains can be switched on at different time. **Figure 4** shows the Start Up behaviour if both voltage domains are powered by an external power supply. If the VCC and VBB voltage have reached their operating range and the internal data transmission has been started successfully, the IC indicates the end of the Start Up procedure by setting the pin  $\overline{\text{ERR}}$  to logic high. In the situation of a supply voltage drop at VBB on the Sense Side - even short - the Sense Chip requires a proper restart and therefore the  $\mu$ Controller Side control unit needs to react accordingly, especially to guarantee the integrity of the sensor data provided to the filter stage.



## 3.2.1 Voltage Limits on VBB

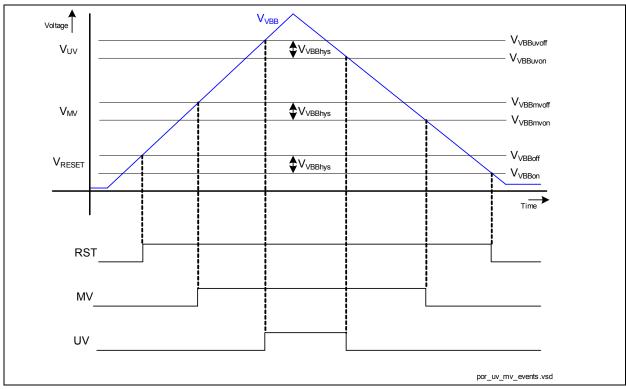


Figure 3 Start Up Procedure with external Power Supply

During UVLO, all registers are reset to their reset values as specified in the Chapter 6.2. As a result, the flags TE, UV as well as MV are High and the ERR pin is Low (error condition). Immediately after the reset is released, the chip is first configured by "reading" the logic level of the SEL, MS1, MS0 (when available). The IC powers up as a parallel device i.e. the AD0-7 pins are high-impedance until the IC configuration is over.

The supply voltage VBB is monitored during operation by two internal comparators (with typ. 8  $\mu$ s blanking time @ 500kHz  $f_{scantyp}$ ) detecting:

- VBB Undervoltage: If the voltage drops below the UV threshold (see Table 7), the UV-bit in the GLERR register is set High. The IC operates normally.
- VBB Missing Voltage: If the voltage further drops below the MV threshold, lower than the previous threshold, the MV-bit in the GLERR register is set, the Sense Side of the IC is turned off when reaching the V<sub>RESET</sub> threshold whereas the Micro-Controller Side remains active.

These 2 thresholds are inactive when the IC operates in Self Power Mode i.e. when the DC\_ENA pin is High.

Note: In case DC\_ENA ist high the integrated DC/DC driver is active. The driver stage is self-protected in overload condition: the internal switches will be turned off as long as the overcurrent condition is detected and the IC will automatically restart once the overload condition disappears.

Important: Since the UV and MV (as well as the TE and W4S) bits used for generating the  $\overline{ERR}$  signal are preset to High during UVLO, the  $\overline{ERR}$  pin is Low after power up. Therefore the  $\overline{ERR}$  signal requires to be explicitly cleared after power up. At least one read access to the GLERR and INTERR registers is needed to update those status bits and thus release the  $\overline{ERR}$  pin.



# 3.2.2 External Supply

Figure 4 shows the Start Up behaviour if both voltage domains are powered by an external power supply. If the VCC and VBB voltage have reached their operating range and the internal data transmission has been started successfully, the IC indicates the end of the Start Up procedure by setting the pin ERR to logic high

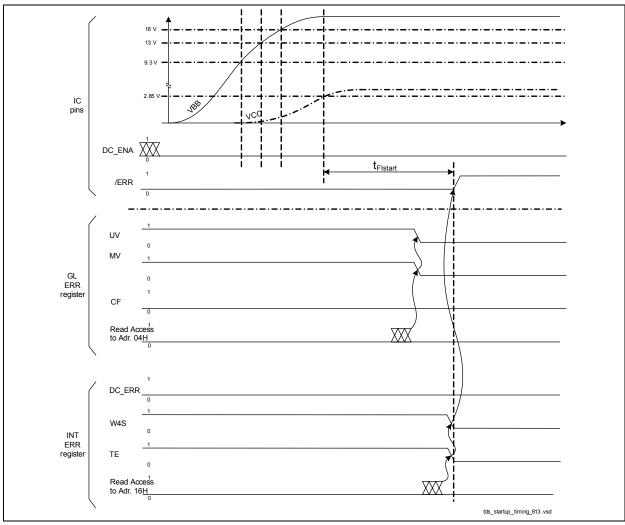


Figure 4 Start Up Procedure with external Power Supply



## 3.2.3 DC/DC Supply

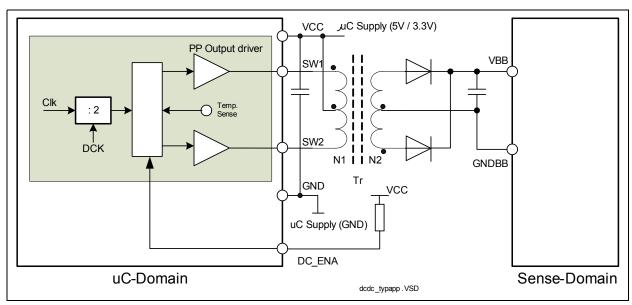


Figure 5 Typical Circuitry for Self Powered Mode with Push-Pull Converter

The IC can as well operate in self powered mode. In this case, the Process Side can be supplied at VBB with an isolated push-pull converter connected to the Micro-controller Side and driven by the pins SW1 and SW2. The internal driver stage at SW1 and SW2 is designed to power up two ISO1I813T (refer to Table 8). The DC/DC-Converter is driven by the internal clock. Parameters are calculated with the internal clock = 500 kHz. By setting the DCK Bit in the GLCFG register a prescaler by 2 can be activated. Should the user adjusts another different frequency the transformer has to be adjusted accordingly.

The short-circuit protection uses a temperature sensor located close to the drivers and disables the driver stages when a predefined temperature is reached (**Figure 7**, **Figure 5**). The target value for the switch-off-temperature is 160°C with a hysteresis of < 10°C. That means that the drivers are switched off at a temperature of 160 °C and switched on at a temperature of <=150°C



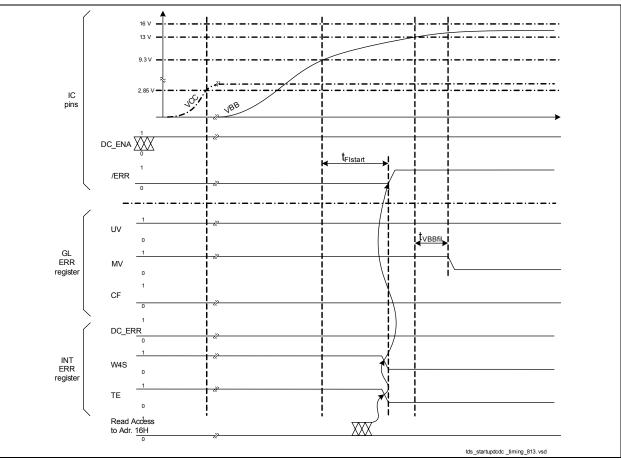


Figure 6 Start Up Procedure with DC/DC Supply



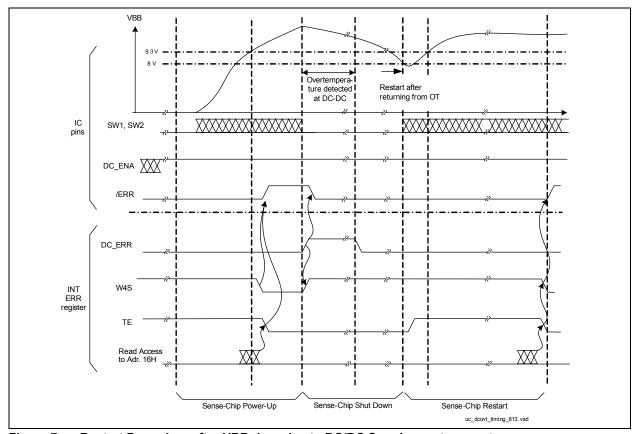


Figure 7 Restart Procedure after VBB drop due to DC/DC Supply overtemperature



#### 3.3 Internal Oscillator

An external resistor has to be connected to Rosc and allows the adjustment of the frequency as shown in Figure 8.

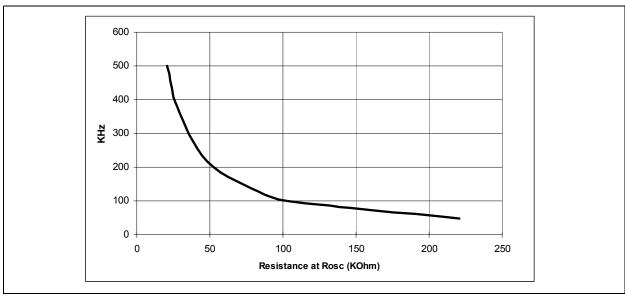


Figure 8 Internal Frequency Setting at Rosc

The internal oscillator provides the scan clock for the sampling of the sensor data and diagnostics as well as the internal digital averaging filters. Therefore the filter times as defined in the **Table 11** for the typical frequency of 500 KHz will change accordingly. As an example, it is possible to define filter time longer than 20 ms by reducing the internal oscillator frequency.

Moreover, in the applications where the IC current consumption is critical, it is possible to reduce the internal oscillator frequency by increasing the  $R_{OSC}$  (see **Figure 9**).

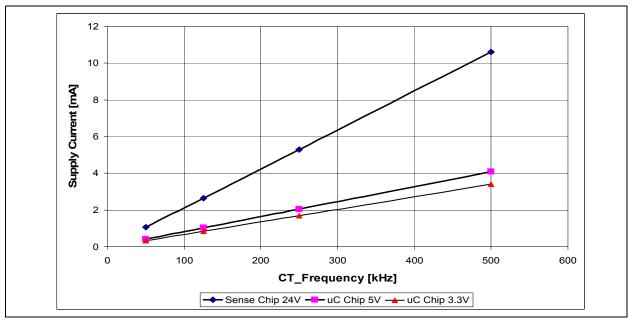


Figure 9 IC Current Consumption in function of the internal frequency



# 3.4 Sensor Input

## 3.4.1 Input Type Select

The sensor input structures are shown in **Figure 10**. Due to its active current a V-l-characteristic as shown in **Figure 11** is maintained. This V-l-curve is well within the IEC 61131 standard requirements of Type 1, Type 2 and Type 3 sensors, respectively. The **Figure 12** shows the typical application for sensor of type 2. It is recommended to choose for the external resistors  $R_{\text{EXT}}$ ,  $R_{\text{V}}$ ,  $R_{\text{LED}}$  an accuracy of 2 % (< 5% is mandatory) otherwise the V/l-characteristic shown in **Figure 11** cannot be guaranteed.

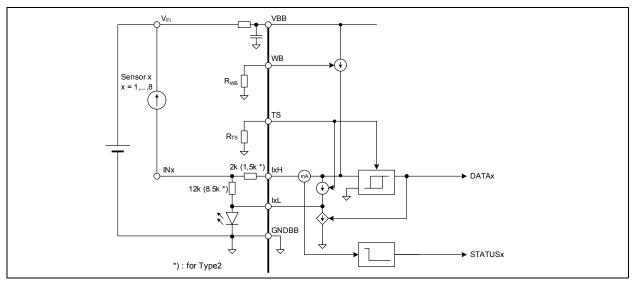


Figure 10 Typical Application for Sensor Input Type 1, 2 and 3



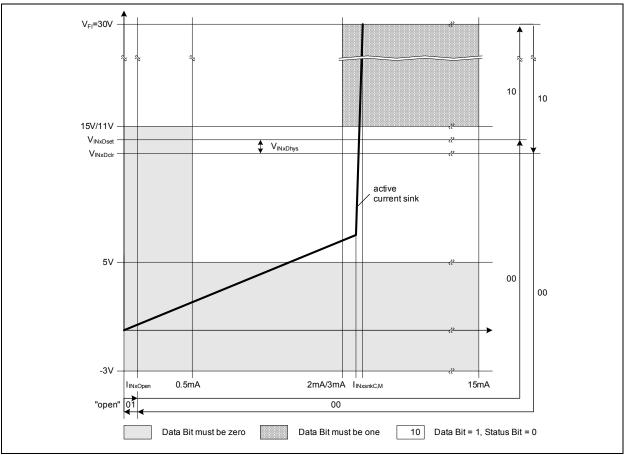


Figure 11 Sensor Input Characteristics

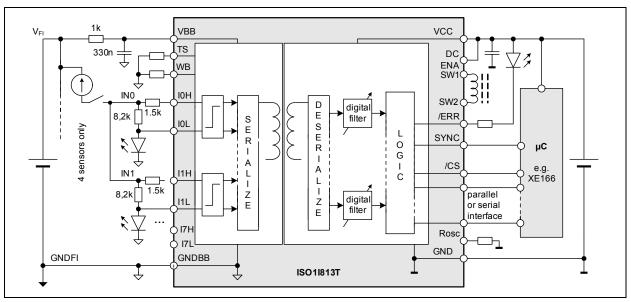


Figure 12 Typical Application for Sensor Type 2



#### 3.4.2 Wire Break Detection

The wire-break current can be adjusted by the  $R_{WB}$ -resistor value connected to the pin WB (**Figure 13**). The minimum wirebreak-current can be choosen only when a LED- or Zener-Diode is connected to the pin IxL with a forward current in the range of few uA in the voltage range below 1 V. In the case of a connected resistor at IxL a great current is flowing across the external resistor Rext and the IxL-resistor ( $R_{LED}$ ). This part cannot be measured internally and has to be added to the internal current part. In this case the minimum adjustable current is greater by about 150uA ( $R_{LED}$  = 2kOhm). The WB bits in the status register have a sticky (latched) property and remains set as long as they are not cleared by a read access and the fault condition is not detected anymore

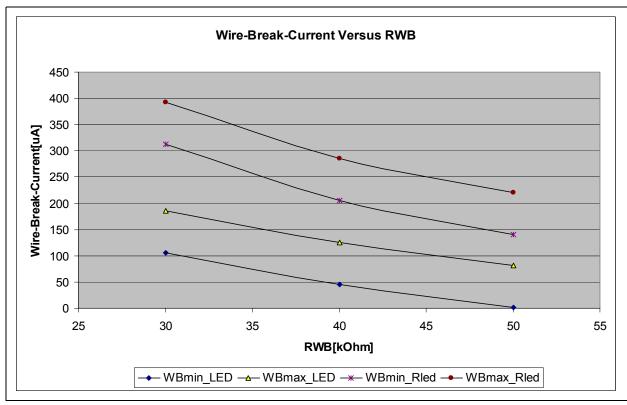


Figure 13 Wire Break Detection for Type 1,3 (typ. @ 25°C)



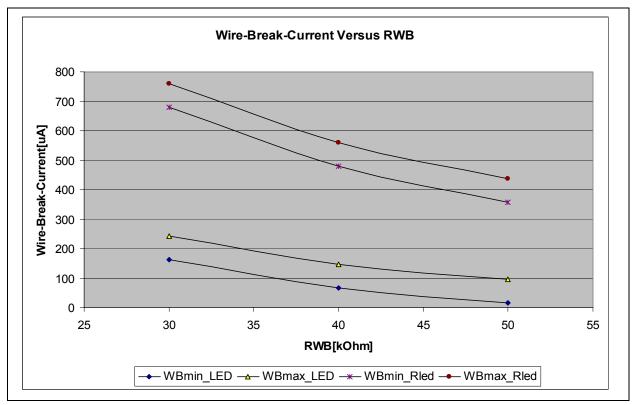


Figure 14 Wire Break Detection for Type 2 (typ. @ 25°C)

In the case of Type 2 two sense inputs are switched in parallel to achieve 2 \* 3 mA (**Figure 12**). In each sense input a mimimum wirebreak current of 60 uA can be measured which means in sum a minimum wirebreak current of 120 uA. It is not recommended to use external resistors at the pins IxL in case of wirebreak measurements. The recommended value would be  $R_{LED}$  = 1.2 kOhm which has been choosen in order not to produce a large voltage drop between IxL and GNDBB which in turn would limit the voltage drop across the sink. But the low value of  $R_{LED}$  would cause a high external current in case of wirebreak-measurements which has to be multiplied by two due to the parallel circuitry of the sense inputs.

#### 3.5 Common Error Output

The input (VBB) undervoltage and missing voltage status which are transmitted via the integrated coreless transformer to the output block and the internal data transmission monitoring information are evaluated in the common error output block, see **Figure 15**. In self-powered mode, an extra information in case of over-current at SW1/2 is evaluated as well.

In case of an internal data transmission error the data and status bits are replaced by the last valid transmission. Moreover, if four consecutive erroneous data transmissions (TE1=1) occur, an internal error signal (TE4=1) is set. The averaging filters are reset and this status is held until four consecutive error-free transmissions (TE1=0) occur. An example timing diagram is shown in **Figure 15**.

This internal error signal is OR-wired with the current VBB undervoltage and missing voltage status. Additionally in the ISO1I813T, the Collective Diagnostics flag is combined in the ERR. Since the output error signal is low-active, the OR-wired result is negated.

In the Self Powered mode, the UV and MV are masked out. Instead the DC\_ERR bit of the **INTERR** is combined with the Transmission Error signal and output at the ERR pin.



The output stage at pin  $\overline{\mathsf{ERR}}$  has an open drain functionality with a pull-up resistor. See **Table 13** for the electrical characteristics.

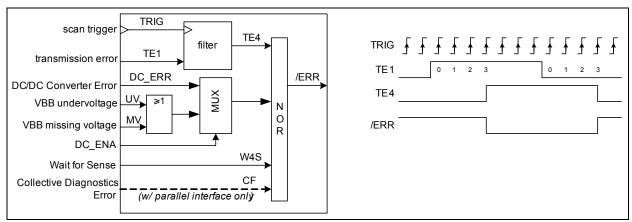


Figure 15 Common Error Output



#### 3.6 Programmable Digital Input Filter

The sensor data and diagnose bits of each input channel can be filtered by a configurable digital input filter. If selected, the filter changes its output according to an averaging rule with a selectable average length. When the sensor state changes without any spikes and noise the change is delayed by the averaging length. Sensor spikes that are shorter than the averaging length are suppressed. **Figure 16** shows the behavior of the filter.

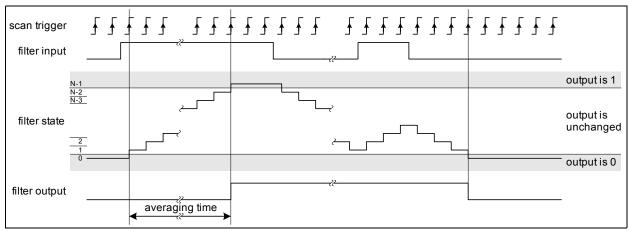


Figure 16 Digital Filter Behavior

The averaging length is selected for each channel individually using the configuration registers **COEFIL0-7**. The programmed filter time apply for both the data and the diagnostics of one channel. See **Table 11** for the different setting options including filter bypass.

In this case, a minimal processing time occurs until the new configuration and the filtered data are valid and can e.g. be frozen with the pin SYNC (described in Figure 17 and Table 17).

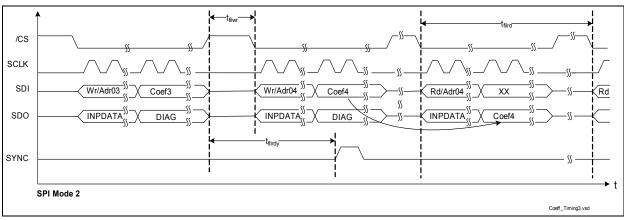


Figure 17 Filter Time Programming and Update Timing

Whereas the absolute filter time depends on the internal oscillator frequency accuracy, the maximal jitter per channel of the IC is 1.5 %. The channel jitter defined in the **Figure 18** is due to the sampling error of the sensor data with the internal clock and applies equally for all the channels.

Furthermore, a fixed propagation delay has to be taken into account due to the data transmission over the Coreless Transformer.



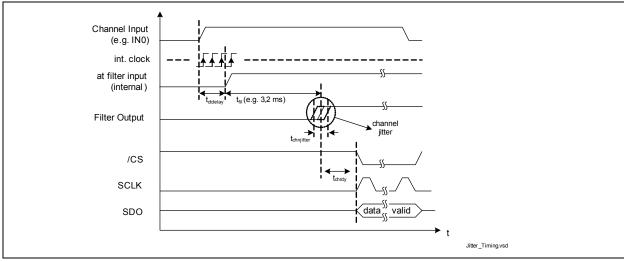


Figure 18 Channel Jitter Definition

#### 3.7 Parallel Interface Mode

The ISO1I813T contains a parallel interface that can be selected by pulling the SEL Pin to logic low state. The interface can be directly controlled by the microcontroller output ports. (**Figure 19**). The output pins AD7:AD0 are in state "Z" as long as  $\overline{CS}$ =1. Otherwise, new sensor data bits or status bits are sampled with the rising edge of  $\overline{RD}$ , and driven by the falling edge of  $\overline{WR}$  and provided at pins AD7:AD0.

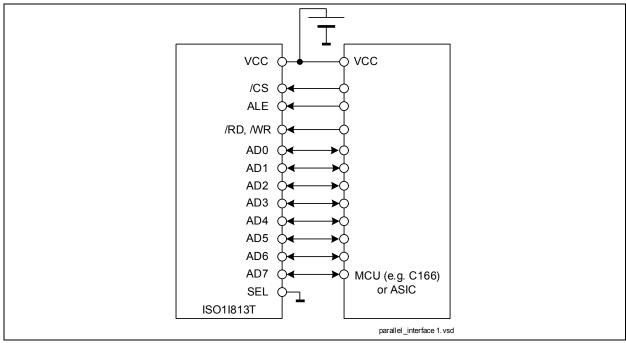


Figure 19 Bus Configuration for Parallel Mode

The timing requirements for the parallel interface are shown in Figure 20, Figure 21 and Table 15.



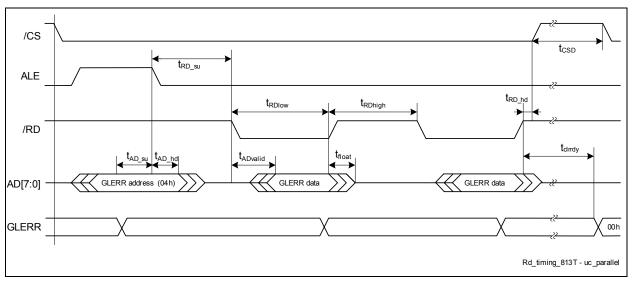


Figure 20 Parallel Bus Timing Read

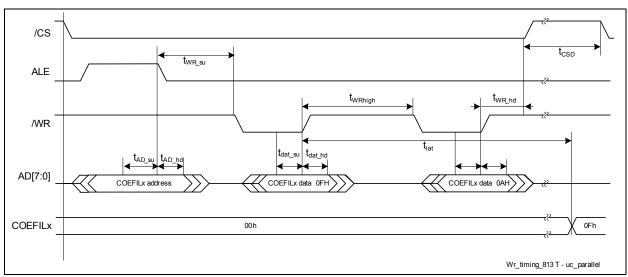


Figure 21 Parallel Bus Timing Write



#### 3.8 Serial Interface Mode

The ISO1I813T contains two serial interfaces that can be activated by pulling the SEL pin to logic High state. The interface can be directly controlled by the microcontroller output ports. The output pins SDO and SSO are in state "Z" as long as  $\overline{\text{CS}}$ =1. Otherwise, the bits are sampled with the falling edge of  $\overline{\text{CS}}$ . With every falling edge of SCLK the bits are provided serially to the pin SDO and SSO, respectively. At the same time, the inputs to SDI, SSI are registered into input-FIFO buffers (sampled with the rising edge of SCLK). When all internally sampled bits have been transferred to SDO/SSO, the buffered bits from the inputs SDI/SSI are provided to these pins (daisy-chain support).

The timing requirements for the serial interface are shown in Figure 22 and in Table 16.

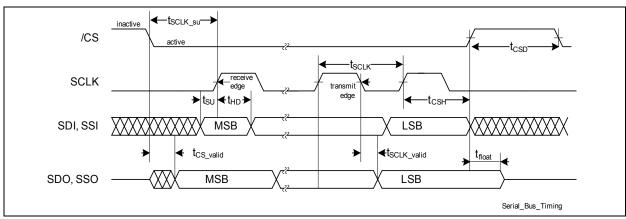


Figure 22 Serial Bus Timing

Several SPI topologies are supported: pure bus topology, daisy-chain and any combinations (Figure 23). Of course independent individual control with dedicated SPI controller interfaces for each slave IC is possible, as well.

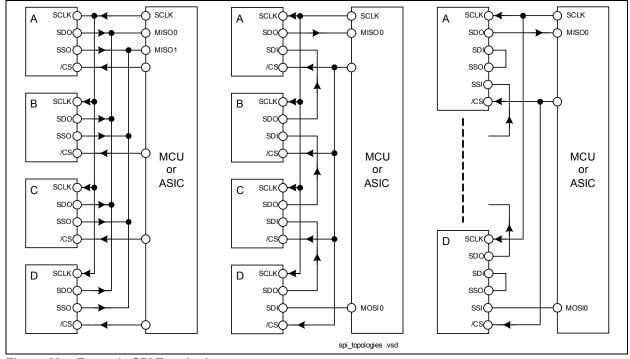


Figure 23 Example SPI Topologies



#### 3.8.1 SPI Modes

#### 3.8.1.1 Switching Serial Modes

All serial modes MS1, MS0 = 11, 01, 10, 00 are switchable during operation but not within a serial transfer frame. No internal registers are affected. Only multiplexers and CRC-engines can be activated or deactivated. Internal FSMs are reset. The user has to run one dummy serial process after switching of serial modes to clear the serial shift registers and reset the internal FSMs. For example: switching from MS1, MS0 = 00 to MS1, MS0 = 11 means the 24 bit serial shift registers and the CRC-engines will be activated. To guarantee proper operation one dummy read sequence has to be processed means "shift in 24 bits with read address, zeros and CRC within a  $\overline{\text{CS}}$ = low frame" to operate the serial interface in the new mode. A reliable output is not guaranteed for the first serial process. The same is true for changing the serial mode in the reverse direction : from MS1, MS0 = 11 to MS1, MS0 = 00. Here at least one serial access (8 SCLK-cycles) within a  $\overline{\text{CS}}$ =low frame is necessary.

Be aware that in Mode01 read access the data at SDO/SSO corresponds to the adress which has been written in the frame before. Mode00 and Mode 01 support the daisy-chain application.

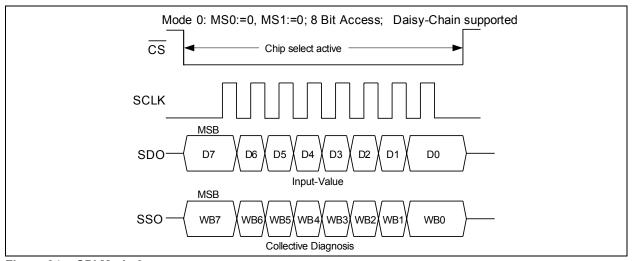


Figure 24 SPI Mode 0



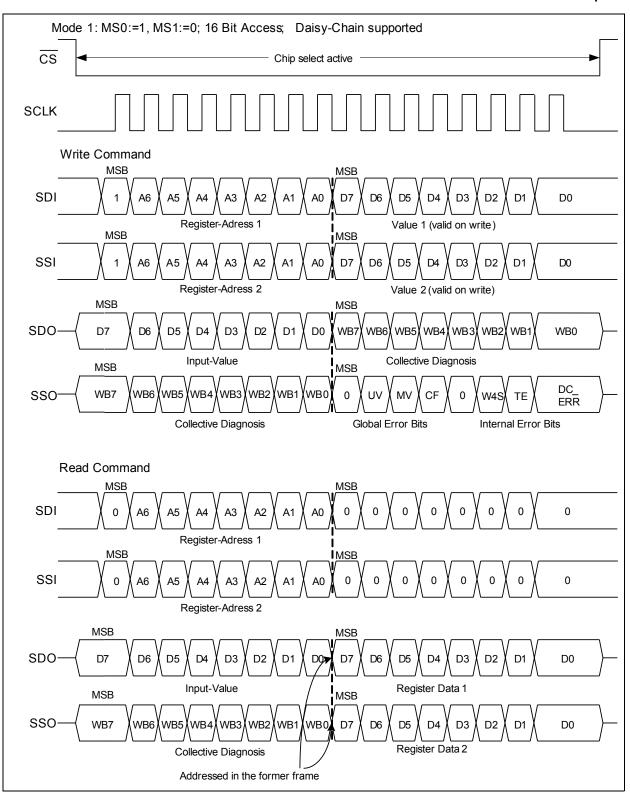


Figure 25 SPI Mode 1



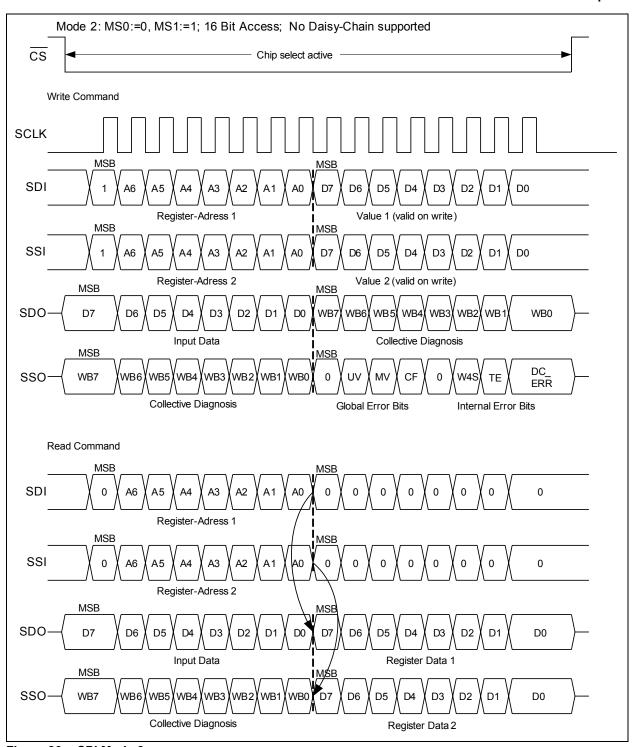


Figure 26 SPI Mode 2



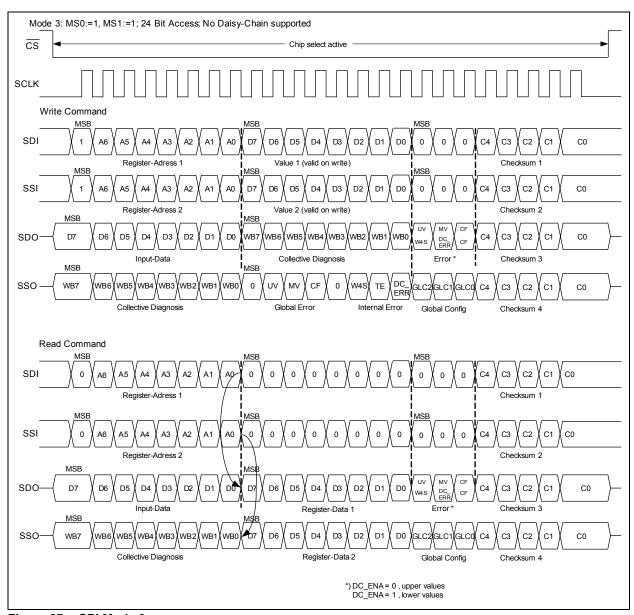


Figure 27 SPI Mode 3

The error values in the SDO-segment depend on the setting of DC\_ENA. If DC\_ENA is set to 1 the IC is supplied by the integrated DC/DC converter and the error information W4S, DC\_ERR, CF is valid. If DC\_ENA is set to 0 the error information UV, MV, CF is valid

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#### 3.8.2 Architecture of CRC-Engines

For writing serial data into the uC-interface chip one serial-SPI-mode (MS1, MS0 = 11) delivers with the pure input data bit stream (write by an uC, 19 bits) also the CRC-signature (5 bits). The total bitstream is fed into the CRC-input engines and processed according to the underlying CRC-algorithm serially.

The CRC is a 5-Bit-checksum and will be calculated with the polynom X5+ X4+ X2+1 and is calculated from Bit [23:5]. The checksum is transferred to Bit [4:0]. After totally processed 24 serially shifted in-bits (including the CRC-signature) the total result of the CRC-algorithm processing has to be zero. In the case of another result different from zero the delivered signature is not consistent with the delivered bit stream. This will be indicated by setting the CRC ERR Pin to low

For reading of registers by a uC a CRC-signature (5 bits) (MS1, MS0 = 11) will be delivered with the pure data bit stream (19 bits): data output (read by a uC). The read bitstream has to be processed according to the CRC-algorithm serially. After totally processed 19 serially shifted out-bits the CRC-signature has been calculated and delivered to the output pins SDO, SSO.

#### 3.9 SYNC Operation

The filtered data and diagnostics can be synchronized on the falling edge of the SYNC pin or "frozen" by holding SYNC low (see Figure 28 and Table 17).

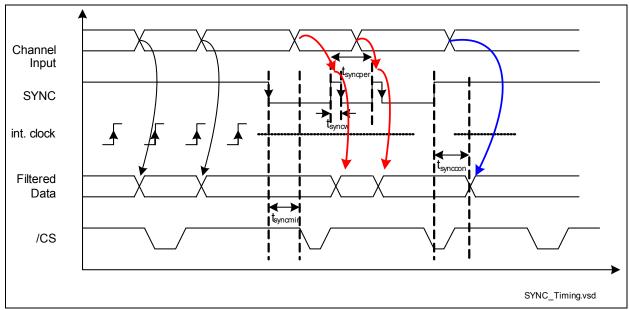


Figure 28 SYNC Operation Timing



#### **Standard Compliance**

# 4 Standard Compliance

The ISO1I813T allows the design of a sensor interface compliant with the standard requirements listed below:

System Insulation Characteristics as shown in Table 3,

System Immunity Characteristics as shown in Table 5.

These requirements are valid for an application using the ISO1I813T including external circuitry (as proposed in **Figure 29**), not for the IC alone.

Note: When the IC is not supplied, probing of the sensor input interface is still possible due to the external circuitry, i.e. the 12k resistor and the LED. In addition to the current through the LED a small current  $I_{IxH}$  flows through the pins IxH and IxL.

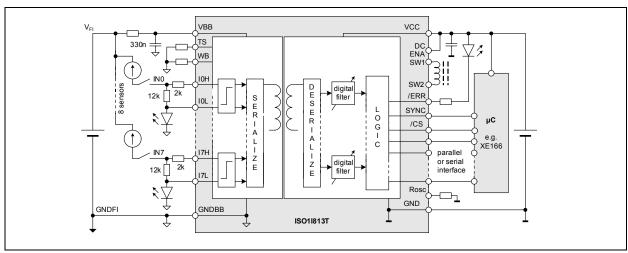


Figure 29 Recommended Application Circuit

Table 2 System Absolute Maximum Ratings

Parameter	Symbol	Values			Unit	Note /
		Min.	Тур.	Max.		Test Condition
Field Input Voltage Overvoltage 1300 ms	V <sub>Flov</sub>	-45		+45	V	
Input Voltage INx	$V_{INx}$	-45		+45	V	

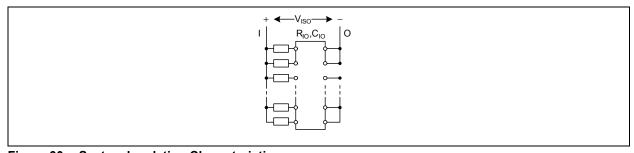


Figure 30 System Insulation Characteristics



## **Standard Compliance**

 Table 3
 System Insulation Characteristics

Parameter	Symbol	Values			Unit	Note /	
		Min.	Тур.	Max.		<b>Test Condition</b>	
Pollution Degree (DIN VDE 0110/1.89, DIN EN 60664-1)			2				
Minimum External Clearance	CLR	6.7			mm		
Minimum External Creepage	CPG	6.2			mm		
Comparative Tracking Index	CTI	550			V		
Maximum Working Insulation Voltage	V <sub>ISO</sub>	500			V <sub>AC</sub>	1 min duration <sup>1)</sup>	
Approval UL1577						Pending	
Approval CSA						Pending	
Approval EN61131-2						Pending	

<sup>1)</sup> not subject to production test, verified by characterization



## 5 Electrical Characteristics

This section comprises:

- Operating Conditions and Power Supply (see Section 5.2)
- Electrical Characteristics Input Side (see Section 5.3)
- Electrical Characteristics Microcontroller Interface (see Section 5.4)

Tolerance values always contain the sum of process-related tolerance values and tolerance-values based on the temperature drift within the specified temperature range.

## 5.1 Absolute Maximum Ratings

All voltages at pins 25 to 48 are measured with respect to ground GNDBB. All voltages at pins 1 to 24 are measured with respect to GND. The voltage levels are valid if other ratings are not violated. The two voltage domains VCC, GND and VBB, GNDBB are internally electrically isolated.

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Table 4 Absolute Maximum Ratings

Parameter	Symbol	Value		Unit	Note /
		Min.	Max.		<b>Test Condition</b>
Continuous Voltage at pin VBB	$V_{VBB}$	-0.3	45	V	Power Dissipation must not exceed max-value
Peak Voltage VBB, Overvoltage 500 ms	$V_{VBB}$	-0.3	45	V	
Supply Voltage VCC	V <sub>VCC</sub>	-0.3	6.5	V	
Continuous Voltage at logic pins 1 - 24 (except VCC and GND pins)	$V_{LOG}$	-0.3	6.5	V	
Continuous Voltage at pin TS, WB		-0.3	6.5	V	
Junction Temperature	$T_J$	-40	150	°C	
Storage Temperature	T <sub>S</sub>	-50	150	°C	
Power Dissipation	P <sub>tot</sub>		800	mW	
Input Voltage Range	V <sub>IxH</sub>	-45	45	V	
Input Voltage Range	V <sub>IxL</sub>	-0.3	5	V	
Error Pin Sink Current (ERR=0)	I <sub>ERRsink</sub>		5	mA	$V_{ERR} < 0.25 \cdot V_{VCC}$
Error Pin Sink Current (CRCERR=0)	I <sub>CRCsink</sub>		5	mA	$V_{ERR} < 0.25 \cdot V_{VCC}$
DC-DC switch outputs 1/2	SW1/2		20	V	
Electrostatic discharge voltage (Human Body Model) according to JESD22-A114-B	$V_{ESD}$	_	-	2.5	kV
Electrostatic discharge voltage (Charge Device Model) according to ESD STM5.3.1 - 1999	$V_{ESD}$	_	-	1.5	kV



## 5.2 Operating Conditions and Power Supply

For proper operation of the device, absolute maximum rating (Section 4) and the parameter ranges in Table 5 must not be violated. Exceeding the limits of operating condition parameters may result in device malfunction or spec violations. The power supply pins VBB and VCC have the characteristics given in Table 7.

Table 5 Operating Range

Symbol	Value	lue	Unit	Note /
	Min.	Max.		<b>Test Condition</b>
$V_{VCC}$	2.85	5.5	V	related to GND
$V_{VBB}$	9.6	35	V	related to GNDBB
$V_{VBBDC}$	12	16	V	see <b>Figure 4</b> for operation points) <sup>1)</sup>
T <sub>A</sub>	-40	85	°C	
T <sub>J</sub>	-40	125	°C	
dV <sub>ISO</sub> /dt	-25	25	kV/μs	
H <sub>IM</sub>	30		A/m	IEC61000-4-8
	$\begin{array}{c} V_{VCC} \\ V_{VBB} \\ V_{VBBDC} \\ \end{array}$ $\begin{array}{c} T_A \\ T_J \\ dV_{ISO}/dt \end{array}$	$\begin{tabular}{lllllllllllllllllllllllllllllllllll$	$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$	$\begin{tabular}{ c c c c c c c c c c c c c c c c c c c$

<sup>1)</sup> recommended for operation

Table 6 Thermal Characteristics

Parameter	Symbol	Limit V	Values	Unit	Note / Test Condition
at T $_{\rm j}$ = -40 125°C, V $_{\rm bb}$ =9.635V, V $_{\rm CC}$ =2.855.5V, unless otherwise specified		Min.	Max.		
Thermal resistance junction - case top	R <sub>thJC_Top</sub>		15.0	K/W	measured on top side) <sup>1)</sup>
Thermal resistance junction - case bottom	R <sub>thJC_Bot</sub>		13.8	K/W	)1)
Thermal resistance junction - pin	R <sub>thJP</sub>		11.8	K/W	)1)
Thermal resistance @ 2 cm² cooling area²) (thermal conductance only by radiation and free convection)	R <sub>th(JA)</sub>		88.6	K/W	)1)

<sup>1)</sup> not subject to production test, specified by design

Table 7 Electrical Characteristics of the Power Supply Pins

Parameter	Symbol		Value	S	Unit	Note /
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
VBB UVLO startup threshold	$V_{VBBon}$			9.6	V	
VBB UVLO shutdown threshold	$V_{VBBoff}$	8.0			V	) <sup>1)</sup>
VBB UVLO Hysteresis	$V_{VBBhys}$		1		V	
	$V_{VBBmvoff}$			13.9	V	
VBB missing voltage ON (MV) threshold	$V_{VBBmvon}$	12.1			V	

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<sup>2)</sup> Device on 50 mm x 50 mm x 1.5 mm epoxy PCB FR4 with 2 cm $^2$  (one layer, 35  $\mu$ m thick) copper area. PCB is vertical without blow air.



Table 7 Electrical Characteristics of the Power Supply Pins (cont'd)

Parameter	Symbol		Value	S	Unit	Note /
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
VBB undervoltage OFF (UV) threshold	$V_{VBBuvoff}$			17.0	V	
VBB undervoltage voltage ON (UV) threshold	$V_{VBBuvon}$	15.0			V	
Glitch filters for VBB missing voltage and undervoltage	$T_{VBBfil}$		8		μs	) <sup>2)</sup>
Undervoltage Current for VBB	I <sub>VBBuv</sub>		3.5		mA	V <sub>VBB</sub> < V <sub>VBBon</sub>
Quiescent Current VBB	$I_{VBBq}$		5		mA	V <sub>VBB</sub> = 24 V, I <sub>INx</sub> = 0, VCC = 2.5V
Startup Delay (time between VBBon/VCCon and first data output)	t <sub>VXXon</sub>		26		μs	Digital Filter bypassed) <sup>)2 3)</sup>
VCC UVLO startup threshold	$V_{VCCon}$			2.85	V	
VCC UVLO shutdown threshold	$V_{VCCoff}$	2.5			V	)4)
VCC UVLO threshold hysteresis	V <sub>VCChys</sub>		0.1		V	) <sup>)2 5)</sup>
Quiescent Current VCC	I <sub>VCCq</sub>		3.1		mA	$V_{VCC} = 5 \text{ V})^{2)5} V_{VBB} = 0 \text{ V}$
Quiescent Current VCC	I <sub>VCCq</sub>		2.3		mA	V <sub>VCC</sub> = 3.3 V V <sub>VBB</sub> = 0V

<sup>1)</sup> Note that the specified operation of the IC requires  $V_{VBB}$  as given in Table 5

Table 8 Self-Powered Supply Operation

Parameter	Symbol		Value	s	Unit Note /	Note /
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
ON Resistance at SW1/2	R <sub>DSON</sub>			2.3	Ω	140 mA
Current Rating	I <sub>SW</sub>			140	mA	
Thermal overload trip temperature	T <sub>jt</sub>	157		165	°C	) <sup>1)</sup>
Thermal hysteresis	$\Delta T_{jt}$		5		K	) <sup>1)</sup>

<sup>1)</sup> not subject to production test, specified by design

<sup>2)</sup> defined for  $f_{scantyp}$  500kHz

<sup>3)</sup> not subject to production test, specified by design

<sup>4)</sup> Note that the specified operation of the IC requires  $V_{VCC}$  as given in Table 5

<sup>5)</sup> No Push-Pull Converter connected at SW1/2



# 5.3 Electrical Characteristics Input Side

The electrical characteristics of the input side (pins 25-48) are given in **Table 9**. Note that some parameters refer to INO to IN7 which are nodes of external circuitry (see **Figure 10** or **Figure 29**). Electrical characteristics with respect to these nodes are given for the system including the external circuitry and not for the IC alone.

See also Figure 11 for the different threshold parameters.

Table 9 Sensors Inputs

<b>Parameter</b> at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified	Symbol		Value	s	Unit	Note / Test Condition
		Min.	Тур.	Max.		
Sink Current Limit at Saturation Edge Type 1/3	I <sub>INxsnkC13</sub>	2.3			mA	$V_{VBB} = V_{VBBon},$ $V_{INX} = 6.7V, V_{IXL} = 1.2V$
Sink Current Limit at Saturation Edge Type 2	I <sub>INxsnkC2</sub>	3.3			mA	$V_{VBB} = V_{VBBon}$ , $V_{INX} = 6.7V$ , $V_{IXL} = 1.2V$
Sink Current Limit at Maximum Input Voltage Type 1/3	I <sub>INxsnkM13</sub>			3.4	mA	V <sub>VBB</sub> =35V, V <sub>INx</sub> =30V V <sub>IxL</sub> =2.5V
Sink Current Limit at Maximum Input Voltage Type 2	I <sub>INxsnkM2</sub>			4.8	mA	V <sub>VBB</sub> =35V, V <sub>INx</sub> =30V, V <sub>IxL</sub> =2.5V
LED Supply Current at Maximum Input Voltage, Type 1/3	I <sub>IxLmax</sub>	2.1		3.1	mA	V <sub>VBB</sub> =35V, V <sub>INx</sub> =30V, V <sub>IxL</sub> =2.5V
LED Supply Current at Maximum Input Voltage, Type 2	I <sub>IxLmax</sub>	3.1		4.5	mA	V <sub>VBB</sub> =35V, V <sub>INx</sub> =30V, V <sub>IxL</sub> =2.5V
LED Supply Current at High Threshold Type 3	I <sub>IxL1</sub>	1.5		2.5	mA	V <sub>VBB</sub> =V <sub>VBBon</sub> , V <sub>INX</sub> =11V, V <sub>IxL</sub> =2.5V
LED Supply Current at High Threshold Type 2	I <sub>IxL2</sub>	2.3		3.6	mA	V <sub>VBB</sub> =V <sub>VBBon</sub> , V <sub>INX</sub> =11V, V <sub>IXL</sub> =2.5V
LED Supply Current at High Threshold Type 1	I <sub>IxL3</sub>	1.6		2.6	mA	$V_{VBB} = V_{VBBon},$ $V_{INX} = 15V, V_{IXL} = 2.5V$
LED Voltage recommended	V <sub>FLED</sub>	1.9		3.0	V	)1)
Sense Voltage Switching Threshold, L→H (Type 1)	V <sub>INxDset(1)</sub>			15		$V_{VBB} = 24V$ $V_{IxL} = 2.5V)^{2}$
Sense Voltage Switching Threshold H→L (Type 1)	V <sub>INxDclr(1)</sub>	11				$V_{VBB} = 24V$ $V_{IxL} = 2.5V)^{2}$
Hysteresis H↔L (Type 1)	V <sub>INxDhys(1)</sub>		1			
Sense Voltage Switching Threshold L→H (Type 2)	V <sub>INxDset(2)</sub>			11		$V_{VBB} = 24V$ $V_{IxL} = 2.5V)^{2}$
Sense Voltage Switching Threshold H→L (Type 2)	V <sub>INxDclr(2)</sub>	7				V <sub>VBB</sub> =24V V <sub>IxL</sub> =2.5V) <sup>2)</sup>
Hysteresis H↔L (Type 2)	V <sub>INxDhys(2)</sub>		0.65			
Sense Voltage Switching Threshold L→H (Type 3)	V <sub>INxDset(3)</sub>			11		V <sub>VBB</sub> =24V V <sub>IxL</sub> =2.5V) <sup>2)</sup>
Sense Voltage Switching Threshold H→L (Type 3)	V <sub>INxDclr(3)</sub>	7				V <sub>VBB</sub> =24V V <sub>IxL</sub> =2.5V) <sup>2)</sup>



Table 9 Sensors Inputs (cont'd)

Parameter at T <sub>j</sub> = -40 125°C, V <sub>bb</sub> =9.635V, V <sub>CC</sub> =2.855.5V, unless otherwise specified	Symbol		Value	S	Unit	Note /
		Min.	Тур.	Max.		Test Condition
Hysteresis H↔L (Type 3)	V <sub>INxDhys(3)</sub>		0.7			
Input Sink Current when V <sub>VBB</sub> =0	I <sub>IxHq</sub>		300		μΑ	$V_{VBB}$ =0V $V_{IxH}$ =30V , IxI = open

<sup>1)</sup> not subject to production test, specified by design

Table 10 Setting at the Configuration Pins (TS, WB)

Parameter	Symbol		Value	S	Unit	Note /
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
TS Pull-Down Resistance for Type 1 Selection	R <sub>TSpd1</sub>		33		Ω	)1)
TS Pull-Down Resistance for Type 2 Selection	R <sub>TSpd2</sub>		33		kΩ	2) 1)
TS Pull-Down Resistance for Type 3 Selection	R <sub>TSpd3</sub>		330		kΩ	)1)
WB pin source current	I <sub>WBsource</sub>		12.5		μΑ	$R_{WB} = 40k\Omega$
WB pin detection current	I <sub>WB</sub>		80		μΑ	$R_{WB} = 40k\Omega$
Wirebreak detection blanking time	t <sub>WB_blank</sub>		1		μs	) <sup>3) 4)</sup>
Type selection blanking time	t <sub>TS_blank</sub>		2		μs	) <sup>3) 4)</sup>
Max. WB Pin Load Capacitance	C <sub>WBmax</sub>			5	pF	)1)
Max. TS Pin Load Capacitance	C <sub>TSmax</sub>			20	pF	)1)

<sup>1)</sup> required for operation

<sup>2)</sup> clamped to 2.5V if "logic 1", internally limited if logic "0"

<sup>2)</sup> Only 4 channels can be used for this case.

<sup>3)</sup> not subject to production test, specified by design

<sup>4)</sup> defined for  $f_{scantyp}$  500kHz



## **5.4** Electrical Characteristics Microcontroller Interface

For the Parallel Mode see Table 11, Table 12, Table 14 and Table 15, For the Serial Mode see Table 11, Table 12, Table 14 and Table 16.

Timing characteristics refer to  $C_L$  < 50 pF and  $R_L$  > 10 k $\Omega$ .

Table 11 Sensor Scanning and Averaging 1)

Parameter	Symbol		Values		Unit	Note /
at T <sub>j</sub> = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
Typical Scan Frequency	f <sub>scantyp</sub>	440		510	kHz	$R_{OSC}$ = 22.1 k $\Omega$ without tolerance
Scan Frequency Range	f <sub>scanrge</sub>	50		500	kHz	) <sup>2)</sup> refer to <b>Figure 8</b>
Input Scan Propagation Delay	t <sub>ctdelay</sub>		8		μs	) <sup>1)</sup> applies equally to all channels
Filter Bypass delay	t <sub>bypass</sub>		2		μs	) <sup>1)</sup>
Minimal Filter Output valid time (until Readout i.e. CS falling edge)	t <sub>csrdy</sub>		1.2		μs	including maximum channel jitter) <sup>1)</sup>
Channel Jitter <sup>3)</sup>	t <sub>chnjitter</sub>	0		2 µs	μs	for t <sub>FILT00</sub> and t <sub>FILT01</sub> ) <sup>1)</sup>
Channel Jitter	t <sub>chnjitter</sub>	0		1.5	%	for t <sub>FILT02</sub> to t <sub>FILT07</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT00</sub>		0.050		ms	FT=00 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT01</sub>		0.100		ms	FT=01 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT02</sub>		0.400		ms	FT=02 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT03</sub>		0.800		ms	FT=03 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT04</sub>		1.600		ms	FT=04 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT05</sub>		3.200		ms	FT=05 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT06</sub>		10.000		ms	FT=06 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILT07</sub>		20.000		ms	FT=07 <sub>H</sub> ) <sup>1)</sup>
Digital Filter Monitoring Time	t <sub>FILToff</sub>		2.0		μs	FT=08 <sub>H</sub> 0F <sub>H</sub> ) <sup>1)</sup>

 $<sup>\</sup>overline{1)}$  valid for  $f_{\text{scantyp}} = 500 \text{kHz}$ 

Table 12 Setting at the Configuration Pin (Rosc) see also Figure 8

Parameter	Symbol	nbol Values		Unit	Note /	
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
Rosc Pin Source Current	I <sub>Roscsrc</sub>		50		μΑ	$R_{OSC}$ = 22.1 k $\Omega$
Rosc Resistance to GND	R <sub>Rosc</sub>	18.4	22.1	221	kΩ	E96 resistor
Rosc Pin Regulated Voltage	V <sub>Roscreg</sub>		1.2		V	
Max. Rosc Pin Load Capacitance	C <sub>Roscmax</sub>			5	pF	)1)

<sup>1)</sup> required for operation

<sup>2)</sup> not subject to production test, specified by design

<sup>3)</sup> the channel jitter is defined in Figure 18



Table 13 Error Pins (ERR, CRCERR)

Parameter	Symbol		Value	S	Unit	Note / Test Condition
at T <sub>j</sub> = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		
Error Pin Pull-Up Resistance (ERR=1)	R <sub>ERRpu</sub>		50		kΩ	
Maximum Switching Frequency (ERR, CRCERR)	f <sub>SW</sub>	10		500	kHz	)1)
Error Pin low voltage	V <sub>ERROL</sub>			0.25·V <sub>VCC</sub>	V	$I_{ERROL} = 5mA$

<sup>1)</sup> not subject to production test, specified by design

Table 14 Logical Pins (RD, WR, ALE, MS0/1, CS, AD7:AD0, SCLK, SDO, SSO, SDI, SSI, SEL)

Parameter	Symbol		Values	Unit	Note /	
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
Input Voltage High Level	V <sub>IH</sub>	0.7·V <sub>VCC</sub>		V <sub>VCC</sub> +0.3	V	
Input Voltage Low Level	V <sub>IL</sub>	-0.3		0.3·V <sub>VCC</sub>	V	
Input Voltage Hysteresis	V <sub>Ihys</sub>		100		mV	
Output Voltage High Level	V <sub>OH</sub>	0.75·V <sub>VCC</sub>		1·V <sub>VCC</sub>	V	I <sub>OH</sub> = 5mA
Output Voltage Low Level	V <sub>OL</sub>	0		0.25·V <sub>VCC</sub>	V	I <sub>OL</sub> = 5mA

#### Table 15 Parallel Interface

Parameter	Symbol		Value	s	Unit	Note /	
at T $_{\rm j}$ = -40 125°C, V $_{\rm bb}$ =9.635V, V $_{\rm CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition	
Input Pull Up Resistance (RD, WR, CS)	R <sub>PU</sub>		50		kΩ		
Input Pull Down Resistance (ALE)	R <sub>PD</sub>		50		kΩ		
Read Request Frequency	$f_{RD}$	0.06		9	MHz		
Read Request Period (1/f <sub>RD</sub> )	t <sub>RD</sub>	110		15000	ns		
CS Disable time (minimum CS high time between two accesses)	t <sub>CSD</sub>	7			μs		
Valid AD0-7 Output	t <sub>ADout</sub>			55 55	ns	VCC = 3.3V VCC = 5.0V	
AD0-7 setup time related to WR	t <sub>dat_su</sub>		25		ns		
AD0-7 hold time related to WR	t <sub>dat_hd</sub>		25		ns		
RD setup time	t <sub>RD_su</sub>	55			ns		
WR setup time	t <sub>WR_su</sub>	55			ns		
RD Low duration	t <sub>RDlow</sub>	55			ns	VCC = 3.3V VCC = 5.0V	

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Table 15 Parallel Interface (cont'd)

Parameter	Symbol		Values	S	Unit	Note /	
at $T_j$ = -40 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition	
WR Low duration	t <sub>WRlow</sub>	55			ns	VCC = 3.3V VCC = 5.0V	
RD hold time	t <sub>RD_hd</sub>	0	20		ns		
WR hold time	t <sub>WR_hd</sub>	0	20		ns		
RD Pad to DIAG, GLERR and INTERR Registers Update (Bits Clearing)	t <sub>clrrdy</sub>	4		6.2	μs		
AD0-7 Output disable time	t <sub>float</sub>		30		ns	VCC = 5.0V	
AD0-7 Data bus setup time	t <sub>AD_su</sub>	40			ns		
AD0-7 Data bus hold time	t <sub>AD_hd</sub>	50			ns		
WR latency time	tlat	600		4800	ns		

## Table 16 Serial Interface

Parameter	Symbol		Value	s	Unit	Note / Test Condition	
at T $_{\rm j}$ = -25 125°C, V $_{\rm bb}$ =9.635V, V $_{\rm CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.			
Input Pull Up Resistance ( CS)	R <sub>PU</sub>		50		kΩ		
Input Pull Down Resistance (SCLK, SDI)	R <sub>PD</sub>		50		kΩ		
Serial Clock Frequency	f <sub>SCLK</sub>			9 10	MHz	VCC = 3.3V VCC = 5.0V	
Serial Clock Period (1/f <sub>SCLK</sub> )	t <sub>SCLK</sub>	110 100			ns	VCC = 3.3V VCC = 5.0V	
Serial Clock High Period	t <sub>sclkh</sub>	55 50			ns	VCC = 3.3V VCC = 5.0V	
Serial Clock Low Period	t <sub>SCLKL</sub>	55 50			ns	VCC = 3.3V VCC = 5.0V	
Minimum CS Hold time (rising edge of SCLK to rising edge of CS)	t <sub>CSH</sub>	40			ns		
Minimum CS Disable time (CS high time between two accesses)	t <sub>CSD</sub>	4.8			μs	)1)	
Minimum Data setup time (required time SDI to rising edge of SCLK)	t <sub>SU</sub>	5			ns		
Minimum Data hold time (rising edge of SCLK to SDI)	t <sub>HD</sub>	15			ns		
Minimum valid time CS falling edge to output SDO/SSO	t <sub>CS_valid</sub>	50			ns		
CS falling edge to first rising SCLK edge	t <sub>SCLK_su</sub>	80			ns		



Table 16 Serial Interface (cont'd)

Parameter	Symbol		Values			Note /
at $T_j$ = -25 125°C, $V_{bb}$ =9.635V, $V_{CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition
Minimum valid time SCLK falling edge to output SDO/SSO	t <sub>SCLK_valid</sub>			80 70	ns	VCC = 3.3V VCC = 5.0V
Minimum SDO/SSO Output disable time	t <sub>float</sub>			50 65	ns	VCC = 3.3V VCC = 5.0V
New serial mode activation time (MS0/MS1 change to earliest interface access)	t <sub>MS_rdy</sub>		4		μs	no μController access allowed during the change) <sup>1) 2)</sup>

<sup>1)</sup> valid for  $f_{\text{scantyp}} = 500 \text{kHz}$ 

Table 17 Sync and Coefficient Update Timing

Parameter	Symbol		Values	S	Unit	Note /	
at T $_{\rm j}$ = -40 125°C, V $_{\rm bb}$ =9.635V, V $_{\rm CC}$ =2.855.5V, unless otherwise specified		Min.	Тур.	Max.		Test Condition	
Minimum time interval for μC- Read-Access after falling edge of SYNC-signal	t <sub>syncmin</sub>		500		ns		
Minimum time interval for switching from sync mode into the continuous mode	t <sub>synccon</sub>	3			μs		
Minimum width of SYNC-signal	t <sub>sync</sub>	200			ns		
SYNC-period	t <sub>syncper</sub>	500			ns		
Minimum time interval between 2 write cycles for filter time programming	t <sub>filwr</sub>	4			μs	)1)	
Minimum time interval between a write cycle and a read back cycle for filter time programming	t <sub>filrr</sub>	4			μs	)1)	
Minimum time interval between a filter time write cycle and updated filter data freeze		4			μs	)1)	

<sup>1)</sup> valid for  $f_{\text{scantyp}} = 500 \text{kHz}$ 

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<sup>2)</sup> not subject to production test, specified by design



# 6 Registers of Microcontroller-Interface-Chip

This chapter describes the  $\mu Controller$  Chip registers.

Table 6-1 Register Bit Type Definition

Туре	Symbol	Description
Read	r	The bit can be read
Read only, updated by hardware	h	The bit is updated by the device itself (for instance: sticky bit)
Write	W	The bit can be written

## 6.1 μController Chip Registers Overview

The **Table 6-2** gives an overview of the  $\mu$ Controller Chip registers and their address.

Table 6-2 Registers Summary

Short Name	Description	Access Rights <sup>1)</sup>	Address A6-A0, R/W
DIAG	Collective Diagnostics Register (Wire-Break Detection)	rh	00 <sub>H</sub>
INPDATA	Input Data Register (Input Channel Data)	r	02 <sub>H</sub>
GLERR	Global Error Register	rh	04 <sub>H</sub>
COEFILO (COEFILO-7)	Filter Time for the Data and the Diagnostics of Channel 0	rw	06 <sub>H</sub> , 86 <sub>H</sub>
COEFIL1	Filter Time for the Data and the Diagnostics of Channel 1	rw	08 <sub>H</sub> , 88 <sub>H</sub>
COEFIL2	Filter Time for the Data and the Diagnostics of Channel 2	rw	0A <sub>H</sub> , 8A <sub>H</sub>
COEFIL3	Filter Time for the Data and the Diagnostics of Channel 3	rw	0C <sub>H</sub> , 8C <sub>H</sub>
COEFIL4	Filter Time for the Data and the Diagnostics of Channel 4	rw	0E <sub>H</sub> , 8E <sub>H</sub>
COEFIL5	Filter Time for the Data and the Diagnostics of Channel 5	rw	10 <sub>H</sub> , 90 <sub>H</sub>
COEFIL6	Filter Time for the Data and the Diagnostics of Channel 6	rw	12 <sub>H</sub> , 92 <sub>H</sub>
COEFIL7	Filter Time for the Data and the Diagnostics of Channel 7	rw	14 <sub>H</sub> , 94 <sub>H</sub>
INTERR	Internal Error Register	rh	16 <sub>H</sub>
GLCFG	Global Configuration Register	rw	1C <sub>H</sub>
	Reserved	n.a.	other

<sup>1)</sup> r=read-only, rw=read-write (timing restrictions apply), rh=read update by hardware



### 6.2 Presentation of the Registers

The  $\mu$ Controller side chip provides several 8-bit registers which can be accessed by the  $\mu$ Controller over the serial or parallel interface. Since those registers are located in the chip internal clock domain, the access is controlled by an internal arbiter processing the read / write requests as well as the synchronization requirements especially to freeze the internal registers when the isochronous mode is used (SYNC pin).

Some timing requirements apply to guarantee the data consistency provided to the  $\mu$ Controller (see Electrical Characterisitcs).

## 6.2.1 Sensor Registers

The sensor data and status (Wire-Break) detected at the channel inputs IxH/L by the sensor side chip are available in the INPDATA and DIAG registers respectively. The bits of the DIAG register have a sticky property i.e.once a wire-break condition has been detected (after the filter time), the respective bits remain set. A read access resets the sticky bits under the condition, that no wirebreak is detected anymore. In the serial modes, both registers are per default driven out at the SPO/SSO outputs.

## 6.2.2 Status Registers

The **GLERR** and **INTERR** registers contains the status of the IC. **GLERR** monitors the application relevant parameters: undervoltage (UV), missing voltage (MV) and collective fault (CF) whereas **INTERR** indicates the status of internal signals important for the proper operation of the IC: wait for sense chip (W4S), transmission error (TE) and DC-DC error (DC\_ERR) in case of self powered mode. Those registers can be read over the serial or parallel interface especially to identify the fault causing the error pin (ERR) to be pulled down. There are different options to read those registers: either through direct addressing (e.g. parallel mode) or through the telegram mode when the serial interface is selected where the bits are shifted out during the transaction.

The bits of the **GLERR** and **INTERR** registers have a sticky property and remains set as long as they are not cleared by a read access and the fault condition is not detected anymore. The **Table 6-3** presents which bits are cleared depending on the serial mode and the SPI channel. In the case of the parallel interface, the bits cleared are the ones whose address is contained in the internal ALE register. Only the bits having been read can be cleared. Since the bits are frozen when a read access is detected, it is guaranted that only these bits read over the serial or parallel interface can be cleared: if the status of the bits changes during the transaction, they will not be cleared.

Table 6-3	Clear of t	Clear of the Sticky Bits by Serial Interface									
	Mode 0	Mode 1		Mode 2		Mode 3					
Read / Write	Read	Read	Write	Read	Write	Read	Write				

	1					1110000		
Read / Write	Read	Read	Write	Read	Write	Read	Write	
SPI channel-0	n.a.	RDREG <sup>1)</sup>	DIAG	RDREG <sup>1)</sup>	DIAG	RDREG <sup>1)</sup> UV, MV, W4S, DC_ERR) <sup>2)</sup>	DIAG UV, MV, W4S, DC_ERR) <sup>2)</sup>	
SPI channel-1	DIAG	DIAG RDREG <sup>1)</sup>	DIAG UV, MV, W4S, TE, DC ERR	DIAG RDREG <sup>1)</sup>	DIAG UV, MV, W4S, TE, DC ERR	DIAG RDREG <sup>1)</sup>	DIAG UV, MV, W4S, TE, DC ERR	

<sup>1)</sup> The bits of register which is being read (Direct addressing)

<sup>2)</sup> depends on setting of DC\_ENA



## 6.2.3 Configuration Registers

The filter times of each channel can be programmed with the **COEFILO-7** registers. Since the write access requires some time to update the internal registers, specific timing requirements apply especially between 2 successive programming operations. The **COEFILO-7** registers define as well if the wire break detection should be masked or not in the **DIAG** register.

Only one of the COEFILx registers can be written at the same time (in serial mode only one SPI channel can be used). It is possible to program a filter time and simultaneously to read out another register e.g. another channel filter time.

Furthermore, the behavior of the IC can be customized with the **GLCFG** register:

- The ratio of the switching frequency of the DC-DC ouput stage over the internal clock frequency set at the CLKADJ pin can be changed from 1:1 (default) to 2:1.
- A soft reset can be generated to clear the filter stages and reinitialize the data transmission between Sense side and μController side chips.
- The automatical clearing of the DIAG when the register is read without direct addressing can be disabled.



## 6.3 μController Registers Description

## 6.3.1 Collective Diagnostics Register

This register contains the filtered values of the Wire-Break detection of the channels 0 to 7.

This register can be read by the  $\mu$ Controller. The WB[x] are set with the occurence of a wire break at input line x and can only be cleared by a read operation of this register if the wire break is not detected anymore (sticky bits). As soon as one of those bits is set, the CF-bit of the **GLERR** is set as well. The **Chapter 6.2.2** explains the way the sticky bits are cleared.

## DIAG Collective Diagnostics Register

(Address: 00<sub>H</sub>) Reset Value: 00<sub>H</sub> 7 6 5 1 0 WB7 WB6 WB5 WB4 WB3 WB2 WB1 WB0 rh

Field	Bits	Type	Description			
WB[x]	7-0	rh	nannel Wire-Break Detected			
			This bit indicates if a Wire-Break has been detected at the channel x.			
			0 <sub>B</sub> No wire-break signal detected at channel x.			
			1 <sub>B</sub> A wire-break condition has been detected at channel x.			

#### 6.3.2 Input Channel Data Register

This register contains the filtered values of the input data detected at the channels 0 to 7.

This register can be read by the µController.

When the parallel interface is selected, the default address contained in the internal ALE register is the address of this register.

# INPDATA Input Data Register

(Address: 02<sub>H</sub>) Reset Value: 00<sub>H</sub> 7 6 5 3 2 0 4 D6 D3 D2 D1 D0 D7 D5 D4

Field	Bits	Type	Description
D[x]	7-0	rh	Input Channel Data
			This bit represents the input data detected at the pins IxH of the channel x
			depending on the sensor type selected.
			0 <sub>B</sub> Input Data below the input threshold.
			1 <sub>B</sub> Input Data above the input threshold.



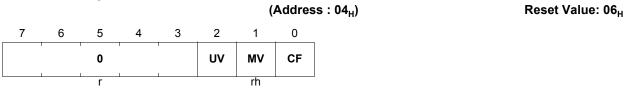
## 6.3.3 Global Error Register

This register contains the status of the IC parameters monitored during operation.

This register can only be read by the  $\mu$ Controller. The CF-bit is the OR-combination of all the bits of the **DIAG** register. The bits of this register are sticky and can only be cleared when the bits are read out and the faults are not detected anymore (refer to **Chapter 6.2.2** for more details).

The UV and MV bits are reset to 1 when the VBB voltage is below the UVLO threshold or during transmission error between the sensor side and  $\mu$ Controller side. The bits of the **GLERR** register are used in the generation of the signal of the error pin (ERR) and shifted out in some of the serial modes when the SPI interface is selected.

## GLERR Global Error Register



Field	Bits	Туре	Description	
CF	0	rh		
MV	1	rh	VBB Missing Voltage  This bit indicates if a missing voltage condition has been detected at the VBB pin.  0 <sub>B</sub> No missing voltage detected at VBB.  1 <sub>B</sub> A missing voltage condition has been detected at VBB.	
UV	2	rh	VBB Under Voltage This bit indicates if an undervoltage condition has been detected at the VBB pin.  0 <sub>B</sub> No undervoltage detected at VBB.  1 <sub>B</sub> An undervoltage has been detected at VBB.	
0	[7:3]	r	Reserved returns 0 if read.	



# 6.3.4 Filter Time of Channel 0-7 Register

This register defines the filter time for both the data and diagnostics for each channel IN0-7. It configures as well if the wire-break bit must be masked in the **DIAG** register.

This register can be modified and read by the  $\mu Controller$ . .

#### COEFIL0-7

## **Channel 0-7 Filter Time Register**

(Address:  $06_H$  -  $14_H$  for read access,  $86_H$  -  $94_H$  for write access,) Reset Value:  $1F_H$ 

7	6	5	4	3	2	1	0
0	0	0	MWB		F.	Γ	
	r		rw		rv	٧	

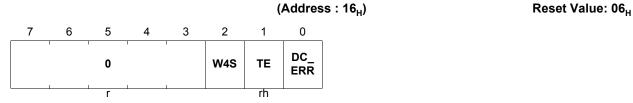
Field	Bits	Type	Description	
FT	[3:0]	rw	Filter Time This bit field configures the filter time for averaging the Data and the Wire-Break signals detected at channels IN0-7. $ 00_{H} 50 \mu s $ $ 01_{H} 100 \mu s $ $ 02_{H} 400 \mu s $ $ 03_{H} 800 \mu s $ $ 04_{H} 1,6 ms $ $ 05_{H} 3,2 ms $ $ 06_{H} 10 ms $ $ 07_{H} 20 ms $ $ 08_{H} - 0F_{H} bypassed (default)$	
MWB	4	rw	Mask Wire-Break Detection This bit masks the filtered signal of the Wire-Break detection.  O <sub>B</sub> The wire-break signal is masked and is not visible in the DIAG register.  The wire-break signal is not masked and appears in the DIAG register. (default).	
0	[7:5]	r	Reserved returns 0 if read; should be written with 0.	



## 6.3.5 Internal Error Register

This register contains the status of the internal errors monitored for safe IC operation. The bits are sticky and remain set once the fault condition is detected until a read operation occurs and the faults are resolved. The bits of the **INTERR** register are used in the generation of the error pin (ERR) and shifted out in some of the serial modes when the SPI interface is selected. On power up (UVLO), the bits W4S and TE are preset to High and will have to be cleared by a read access during the startup phase.

## INTERR IC Status Register



Field	Bits	Type	Description	
DC_ERR	0	rh	DC-DC Converter Error This bit indicates if overload condition has been detected at the SW1 or SW2 switches.  0 <sub>B</sub> No overload detected. 1 <sub>B</sub> Overload detected.	
TE	1	rh	Transmission Error This bit indicates if a transmission error has been detected over the Coreless Transformer between the Sense side chip and the $\mu$ Controller side chip. $0_B$ No transmission error. $1_B$ Transmission error.	
W4S	2	rh	Wait for Sense Chip This bit indicates the Sense side chip is correctly supplied and ready for transmission.  0 <sub>B</sub> Sense Side is ready.  1 <sub>B</sub> Sense Side is not ready because of insufficient supply or long transmission error.	
00000	[7:3]	r	Reserved returns 0 if read.	

Reset Value: 00<sub>H</sub>

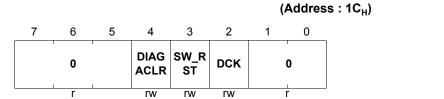


## Registers of Microcontroller-Interface-Chip

# 6.3.6 Global Configuration Register

This register contains configuration parameters for the sensor type selection as well as the DC-DC driver.

## GLCFG Global Configuration Register



Field	Bits	Type	Description		
00	1:0	rw	Reserved		
			returns 0 if read; should be written with 0.		
DCK	2	rw DC-DC Driver Switching Frequency Ratio			
			This bit indicates the ratio between the sampling clock frequency set at Rosc and		
			the switching frequency of the DC-DC driver (pins SW1/2).		
			0 <sub>B</sub> DC-DC switching frequency is equal to the sampling frequency (1:1) (default).		
			1 <sub>B</sub> DC-DC switching frequency is half to the sampling frequency (2:1).		
SW_RST	3	rw	Soft-Reset for the Filtering Stage		
			This bit triggers the reset of the Filter registers		
			0 <sub>B</sub> No Reset		
			1 <sub>B</sub> Reset is generated for the Filter stage		
DIAG_ACLR	4	rw	Diagnostics Automatical Clear		
			This bit selects if the <b>DIAG</b> register is automatically cleared after any access to		
			the <b>DIAG</b> register (especially for the second SPI channel at the SSO pin). The		
			diagnostics remain in both case sticky.		
			0 <sub>B</sub> Automatical clear after any access to the <b>DIAG</b> register (default)		
			1 <sub>B</sub> Automatical clear disabled		
000	[7:5]	r	Reserved		
			returns 0 if read; should be written with 0.		

Confidential Package Outline

# 7 Package Outline

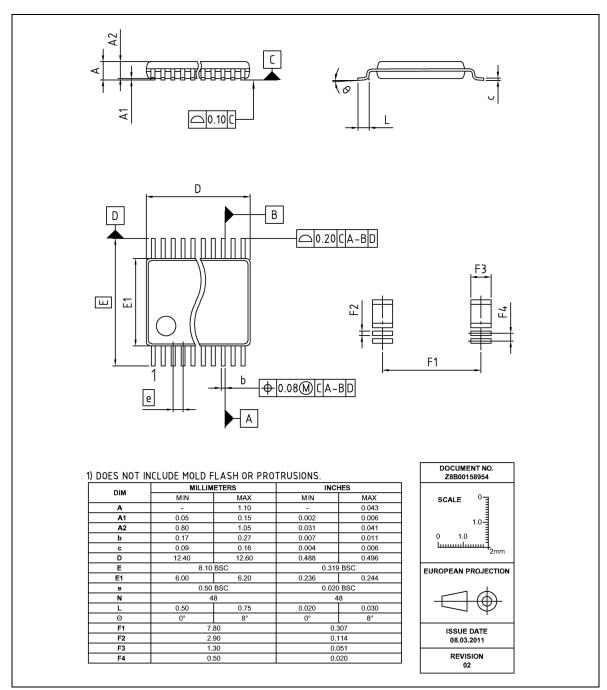


Figure 7-1 Package Outline TSSOP-48 (tie bar not drawn in outline)

#### **Notes**

- 1. You can find all of our packages, sorts of packing and others in our Infineon Internet Page "Products": <a href="http://www.infineon.com/products">http://www.infineon.com/products</a>.
- 2. Dimensions in mm.

