General Description

The MAX8863T/S/R and MAX8864T/S/R low-dropout linear regulators operate from a +2.5V to +6.5V input range and deliver up to 120mA. A PMOS pass transistor allows the low, 80µA supply current to remain independent of load, making these devices ideal for battery-operated portable equipment such as cellular phones, cordless phones, and modems.

The devices feature Dual Mode[™] operation: their output voltage is preset (at 3.15V for the T versions, 2.84V for the S versions, or 2.80V for the R versions) or can be adjusted with an external resistor divider. Other features include low-power shutdown, short-circuit protection, thermal shutdown protection, and reverse battery protection. The MAX8864 also includes an auto-discharge function, which actively discharges the output voltage to ground when the device is placed in shutdown mode. Both devices come in a miniature 5-pin SOT23-5 package.

	Applications
Cordless Telephones	Modems
PCS Telephones	Hand-Held Instruments
Cellular Telephones	Palmtop Computers
PCMCIA Cards	Electronic Planners

BATTERY - LIN OUT CIN OUT CIN OUT MAX8863 MAX8864 SHDN GND SET - COUT 1µF

Dual Mode is a trademark of Maxim Integrated Products.

__ Maxim Integrated Products 1

For free samples & the latest literature: http://www.maxim-ic.com, or phone 1-800-998-8800. For small orders, phone 1-800-835-8769.

Typical Operating Circuit

Low Cost

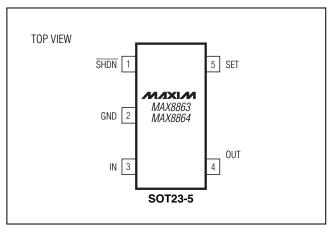
- ♦ Low, 55mV Dropout Voltage @ 50mA IOUT
- Low, 68µA No-Load Supply Current Low, 80µA Operating Supply Current (even in dropout)
- ◆ Low, 350µVRMS Output Noise
- Miniature External Components
- Thermal Overload Protection
- ♦ Output Current Limit
- ♦ Reverse Battery Protection
- Dual Mode[™] Operation: Fixed or Adjustable (1.25V to 6.5V) Output
- Low-Power Shutdown

_Ordering Information

PART	TEMP RANGE	PIN- PACKAGE	SOT TOP MARK*
MAX8863TEUK+T	-40°C to +85°C	5 SOT23-5	AABE
MAX8863SEUK+T	-40°C to +85°C	5 SOT23-5	AABF
MAX8863REUK+T	-40°C to +85°C	5 SOT23-5	AABV
MAX8864TEUK+T	-40°C to +85°C	5 SOT23-5	AABG
MAX8864SEUK+T	-40°C to +85°C	5 SOT23-5	AABH
MAX8864REUK+T	-40°C to +85°C	5 SOT23-5	AABW

*Alternate marking information: CY_ = MAX8863T, CZ_ = MAX8863S, DA_ = MAX8864T, DB_ = MAX8864S +Denotes a lead(Pb)-free/RoHS-compliant package.

Pin Configuration



Features

ABSOLUTE MAXIMUM RATINGS

V _{IN} to GND	7V to +7V
Output Short-Circuit Duration	Infinite
SET to GND	0.3V to +7V
SHDN to GND	7V to +7V
SHDN to IN	7V to +0.3V
OUT to GND	0.3V to (V _{IN} + 0.3V)
Continuous Power Dissipation ($T_A = +70$)°C)
SOT23-5 (3.1mW/°C above +70°C)	

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
θJA	324.3°C/Watt
Storage Temperature Range	
Lead Temperature (soldering, 10sec)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{IN} = +3.6V, GND = 0V, T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS	
Input Voltage (Note 2)	VIN				2.5		6.5	V
		0.4.4.		MAX886_T	3.05	3.15	3.25	
Output Voltage	Vout	$0mA \le I_{OUT} \le 50mA$, SET = GND		MAX886_S	2.75	2.84	2.93	V
				MAX886_R	2.70	2.80	2.88	
Adjustable Output Voltage Range (Note 3)	V _{OUT}				V _{SET}		6.5	V
Maximum Output Current					120			mA
Current Limit (Note 4)	ILIM					280		mA
Ground Pin Current	10	SET = GND		I _{LOAD} = 0mA		68	150	
	lQ	SET = GND		$I_{LOAD} = 50 mA$		80		μA
Dropout Voltage (Note 5)		I _{OUT} = 1mA			1.1		mV	
Diopour voltage (Note 3)		I _{OUT} = 50mA			55	120		
Line Regulation	ΔV_{LNR}	$V_{IN} = 2.5V$ to 6.5V, SET tied to OUT, $I_{OUT} = 1mA$		-0.15	0	0.15	%/V	
Load Regulation		IOUT = 0mA to 50mA		SET = GND		0.011	0.040	%/mA
	AVLDR	DR 1001 = 0 mA to 50 mA		SET tied to OUT		0.006		- /0/IIIA
Output Voltage Noise		10Hz to 1MHz		$C_{OUT} = 1\mu F$		350		μV _{RMS}
				$C_{OUT} = 100 \mu F$		220		µvrivis
SHUTDOWN								
SHDN Input Threshold	VIH				2.0			v
	VIL						0.4	v
SHDN Input Bias Current	I <u>shdn</u> \	V SHDN = VIN	T _A = ·	+25°C		0	100	nA
	1311010		T _A =			0.05		10.0
Shutdown Supply Current	IOSHDN	Vout = 0V	T _A = ·	+25°C		0.0001	1	μA
			$T_A = T_A$	Гмах	0.02			P., ,
Shutdown Discharge Resistance (MAX8864)						300		Ω

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{IN} = +3.6V, GND = 0V, T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
SET INPUT							
SET Reference Voltage (Note 3)	VSET	$V_{IN} = 2.5V$ to 6.5V,	$T_A = +25^{\circ}C$	1.225	1.25	1.275	V
SET Reference voltage (Note 5)	VSEI	IOUT = 1mA	$T_A = T_{MIN}$ to T_{MAX}	1.215	1.25	1.285	
SET Input Leakage Current	loct	1/2 = 1.2	$T_A = +25^{\circ}C$		0.015	2.5	nA
(Note 3)	ISET	$V_{SET} = 1.3V$	$T_A = T_{MAX}$		0.5		
THERMAL PROTECTION							
Thermal Shutdown Temperature	T _{SHDN}				170		°C
Thermal Shutdown Hysteresis	ΔTSHDN				20		°C

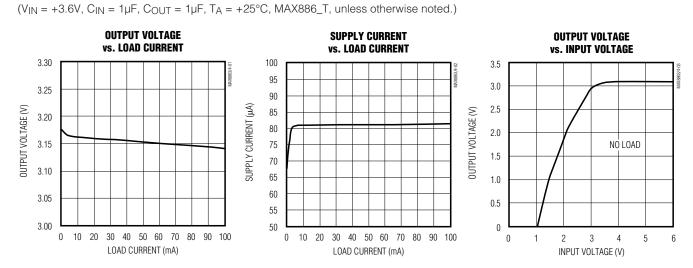
Note 1: Limits are 100% production tested at T_A = +25°C. Limits over the operating temperature range are guaranteed through correlation using Statistical Quality Control (SQC) Methods.

Note 2: Guaranteed by line regulation test.

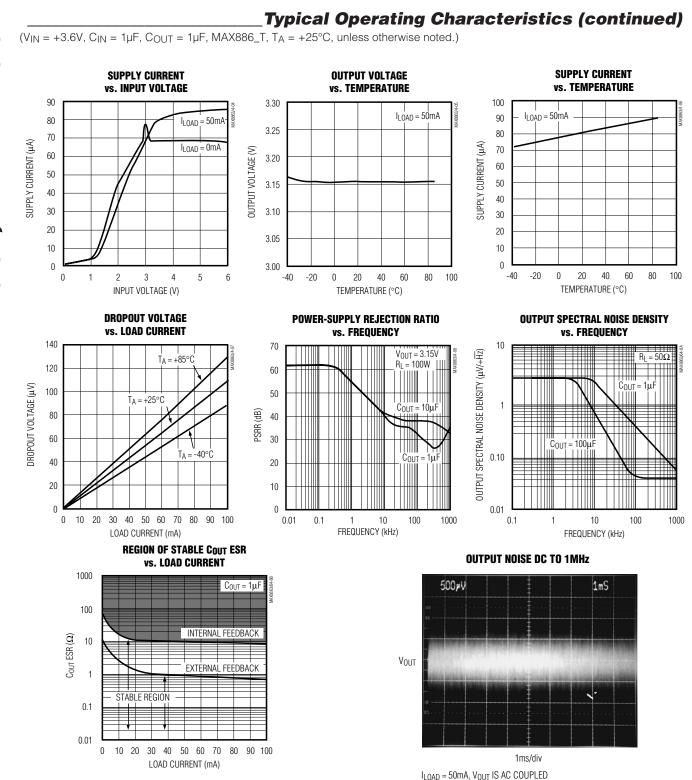
Note 3: Adjustable mode only.

Note 4: Not tested. For design purposes, the current limit should be considered 120mA minimum to 420mA maximum.

Note 5: The dropout voltage is defined as $(V_{IN} - V_{OUT})$ when V_{OUT} is 100mV below the value of V_{OUT} for $V_{IN} = V_{OUT} + 2V$.



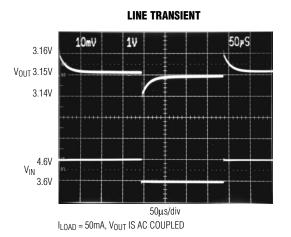
Typical Operating Characteristics

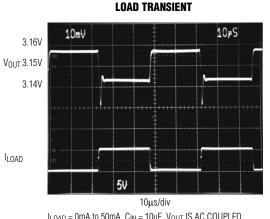


MAX8863T/S/R, MAX8864T/S/R

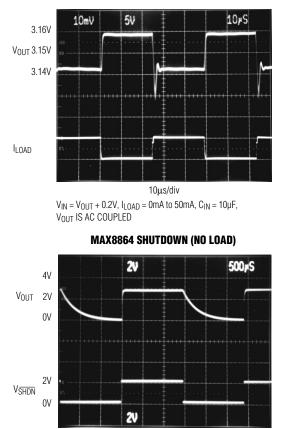
Typical Operating Characteristics (continued)

(VIN = +3.6V, CIN = 1µF, COUT = 1µF, MAX886_T, TA = +25°C, unless otherwise noted.)





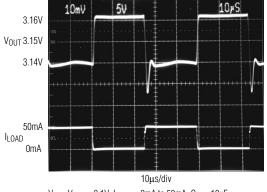
 I_{LOAD} = 0mA to 50mA, C_{IN} = 10 $\mu F,\,V_{OUT}$ IS AC COUPLED



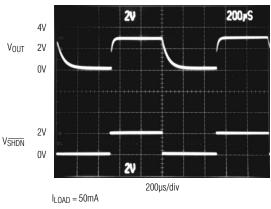
500µs/div

LOAD TRANSIENT

LOAD TRANSIENT



 $V_{IN}=V_{OUT}$ + 0.1V, I_{LOAD} = 0mA to 50mA, C_{IN} = 10 $\mu F,$ VOUT IS AC COUPLED



MAX8864 SHUTDOWN

MAX8863T/S/R, MAX8864T/S/R

NO LOAD

Pin Description

PIN	NAME	FUNCTION
1	SHDN	Active-Low Shutdown Input. A logic low reduces the supply current to 0.1nA. On the MAX8864, a logic low also causes the output voltage to discharge to GND. Connect to IN for normal operation.
2	GND	Ground. This pin also functions as a heatsink. Solder to large pads or the circuit board ground plane to max- imize thermal dissipation.
3	IN	Regulator Input. Supply voltage can range from +2.5V to +6.5V. Bypass with 1μ F to GND (see <i>Capacitor Selection and Regulator Stability</i>).
4	OUT	Regulator Output. Fixed or adjustable from 1.25V to +6.5V. Sources up to 120mA. Bypass with a 1µF, <0.2 Ω typical ESR capacitor to GND.
5	SET	Feedback Input for Setting the Output Voltage. Connect to GND to set the output voltage to the preset 2.80V (MAX886_R), 2.84V (MAX886_S), or 3.15V (MAX886_T). Connect to an external resistor divider for adjustable-output operation.

Detailed Description

The MAX8863/MAX8864 are low-dropout, low-quiescentcurrent linear regulators designed primarily for batterypowered applications. They supply an adjustable 1.25V to 6.5V output or a preselected 2.80V (MAX886_R), 2.84V (MAX886_S), or 3.15V (MAX886_T) output for load currents up to 120mA. As illustrated in Figure 1, these devices consist of a 1.25V reference, error amplifier, MOSFET driver, Pchannel pass transistor, Dual Mode[™] comparator, and internal feedback voltage divider.

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the selected feedback voltage and amplifies the difference. The MOSFET driver reads the error signal and applies the appropriate drive to the P-channel pass transistor. If the feedback voltage is lower than the reference, the pass-transistor gate is pulled lower, allowing more current to pass and increasing the output voltage. If the feedback voltage is too high, the pass-transistor gate is pulled up, allowing less current to pass to the output.

The output voltage is fed back through either an internal resistor voltage divider connected to the OUT pin, or an external resistor network connected to the SET pin. The Dual Mode comparator examines the SET voltage and selects the feedback path. If SET is below 60mV, internal feedback is used and the output voltage is regulated to the preset output voltage. Additional blocks include a current limiter, reverse battery protection, thermal sensor, and shutdown logic.

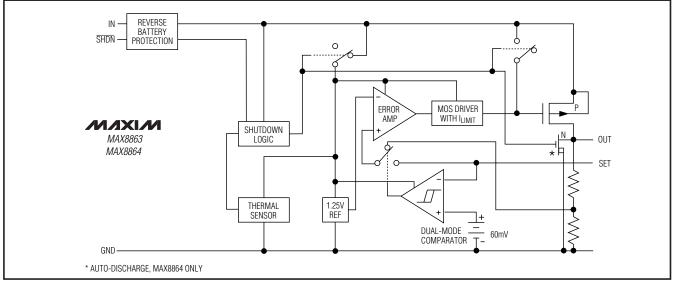


Figure 1. Functional Diagram

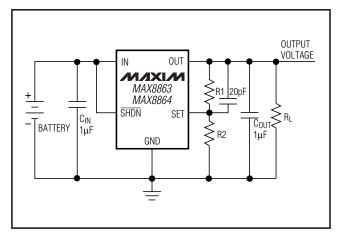


Figure 2. Adjustable Output Using External Feedback Resistors

Internal P-Channel Pass Transistor

The MAX8863/MAX8864 feature a 1.1Ω typical P-channel MOSFET pass transistor. This provides several advantages over similar designs using PNP pass transistors, including longer battery life.

The P-channel MOSFET requires no base drive current, which reduces quiescent current considerably. PNPbased regulators waste considerable amounts of current in dropout when the pass transistor saturates. They also use high base-drive currents under large loads. The MAX8863/MAX8864 do not suffer from these problems, and consume only 80µA of quiescent current, whether in dropout, light load, or heavy load applications (see *Typical Operating Characteristics*).

Output Voltage Selection

The MAX8863/MAX8864 feature Dual Mode operation: they operate in either a preset voltage mode or an adjustable mode.

In preset voltage mode, internal, trimmed feedback resistors set the MAX886_R output to 2.80V, the MAX886_S output to 2.84V, and the MAX886_T output to 3.15V. Select this mode by connecting SET to ground.

In adjustable mode, select an output between 1.25V and 6.5V using two external resistors connected as a voltage divider to SET (Figure 2). The output voltage is set by the following equation:

$$VOUT = VSET (1 + R1 / R2)$$

where $V_{SET} = 1.25V$. To simplify resistor selection:

$$R1 = R2 \left(\frac{V_{OUT}}{V_{SET}} - 1 \right)$$

Choose R2 = $100k\Omega$ to optimize power consumption, accuracy, and high-frequency power-supply rejection. The total current through the external resistive feedback and load resistors should not be less than 10μ A. Since the V_{SET} tolerance is typically less than ± 25 mV, the output can be set using fixed resistors instead of trim pots. Connect a 10pF to 25pF capacitor across R1 to compensate for layout-induced parasitic capacitances.

In preset voltage mode, impedances between SET and ground should be less than $100k\Omega$. Otherwise, spurious conditions could cause the voltage at SET to exceed the 60mV Dual Mode threshold.

Shutdown

A low input on the SHDN pin shuts down the MAX8863/MAX8864. In shutdown mode, the pass transistor, control circuit, reference, and all biases are turned off, reducing the supply current to typically 0.1nA. Connect SHDN to IN for normal operation. The MAX8864 output voltage is actively discharged to ground when the part is placed in shutdown (see *Typical Operating Characteristics*).

Current Limit

The MAX8863/MAX8864 include a current limiter that monitors and controls the pass transistor's gate voltage, estimating the output current and limiting it to about 280mA. For design purposes, the current limit should be considered 120mA (min) to 420mA (max). The output can be shorted to ground for an indefinite time period without damaging the part.

Thermal Overload Protection

Thermal overload protection limits total power dissipation in the MAX8863/MAX8864. When the junction temperature exceeds $T_J = +170^{\circ}$ C, the thermal sensor sends a signal to the shutdown logic, turning off the pass transistor and allowing the IC to cool. The thermal sensor will turn the pass transistor on again after the IC's junction temperature typically cools by 20°C, resulting in a pulsed output during continuous thermal overload conditions.

Thermal overload protection is designed to protect the MAX8863/MAX8864 in the event of fault conditions. Stressing the device with high load currents and high input-output differential voltages (which result in die temperatures above +125°C) may cause a momentary overshoot (2% to 8% for 200ms) when the load is completely removed. This can be remedied by raising the minimum load current from 0µA (+125°C) to 100µA (+150°C). For continuous operation, do not exceed the absolute maximum junction temperature rating of T_J = +150°C.

Operating Region and Power Dissipation Maximum power dissipation of the MAX8863/MAX8864 depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of air flow. The power dissipation across the device is P = IOUT (VIN -VOUT). The resulting maximum power dissipation is:

$P_{MAX} = (T_J - T_A) / \theta_{JA}$

where $(T_J - T_A)$ is the temperature difference between the MAX8863/MAX8864 die junction and the surrounding air, and θ_{JA} is the thermal resistance of the chosen package to the surrounding air.

The GND pin of the MAX8863/MAX8864 performs the dual function of providing an electrical connection to ground and channeling heat away. Connect the GND pin to ground using a large pad or ground plane.

Reverse Battery Protection

The MAX8863/MAX8864 have a unique protection scheme that limits the reverse supply current to less than 1mA when either V_{IN} or $V_{\overline{SHDN}}$ falls below ground. The circuitry monitors the polarity of these two pins, disconnecting the internal circuitry and parasitic diodes when the battery is reversed. This feature prevents the device from overheating and damaging the battery.

VIN > 5.5V Minimum Load Current

When operating the MAX8863/MAX8864 with an input voltage above 5.5V, a minimum load current of 20µA is required to maintain regulation in preset voltage mode. When setting the output with external resistors, the minimum current through the external feedback resistors and load must be 30µA.

Applications Information

Capacitor Selection and Regulator Stability

Normally, use a 1µF capacitor on the input and a 1µF capacitor on the output of the MAX8863/MAX8864. Larger input capacitor values and lower ESR provide better supply-noise rejection and transient response. A higher-value input capacitor (10µF) may be necessary

if large, fast transients are anticipated and the device is located several inches from the power source. Improve load-transient response, stability, and power-supply rejection by using large output capacitors. For stable operation over the full temperature range, with load currents up to 120mA, a minimum of 1µF is recommended.

Noise

The MAX8863/MAX8864 exhibit 350µV_{RMS} noise during normal operation. When using the MAX8863/MAX8864 in applications that include analog-to-digital converters of greater than 12 bits, consider the ADC's power-supply rejection specifications (see the Output Noise DC to 1MHz photo in the Typical Operating Characteristics).

Power-Supply Rejection and Operation from Sources Other than Batteries

The MAX8863/MAX8864 are designed to deliver low dropout voltages and low quiescent currents in batterypowered systems. Power-supply rejection is 62dB at low frequencies and rolls off above 300Hz. As the frequency increases above 20kHz, the output capacitor is the major contributor to the rejection of power-supply noise (see the Power-Supply Rejection Ratio vs. Ripple Frequency graph in the *Typical Operating* Characteristics).

When operating from sources other than batteries, improve supply-noise rejection and transient response by increasing the values of the input and output capacitors, and using passive filtering techniques (see the supply and load-transient responses in the *Typical* Operating Characteristics).

Load Transient Considerations

The MAX8863/MAX8864 load-transient response graphs (see *Typical Operating Characteristics*) show two components of the output response: a DC shift of the output voltage due to the different load currents, and the transient response. Typical overshoot for step changes in the load current from 0mA to 50mA is 12mV. Increasing the output capacitor's value and decreasing its ESR attenuates transient spikes.

Input-Output (Dropout) Voltage

A regulator's minimum input-output voltage differential (or dropout voltage) determines the lowest usable supply voltage. In battery-powered systems, this will determine the useful end-of-life battery voltage. Because the MAX8863/MAX8864 use a P-channel MOSFET pass transistor, their dropout voltage is a function of R_{DS(ON)} multiplied by the load current (see *Electrical Characteristics*).

Package Information

For the latest package outline information and land patterns (footprints), go to **www.maxim-ic.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
5 SOT23	U5+1	<u>21-0057</u>	<u>90-0174</u>



Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
3	5/11	Added lead-free designation and updated continuous power dissipation and θ_{JA} specs	1, 2

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

10

© 2011 Maxim Integrated Products

 Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600

 Products
 Maxim is a registered trademark of Maxim Integrated Products, Inc.