

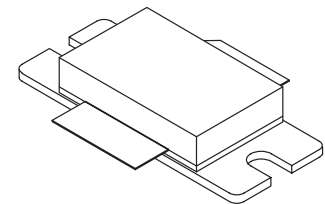
The RF MOSFET Line
RF Power Field Effect Transistors
N-Channel Enhancement-Mode Lateral MOSFETs

MRF18090A
MRF18090AS

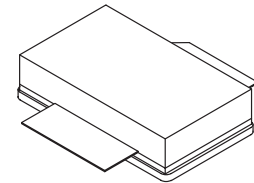
Designed for GSM and EDGE base station applications with frequencies from 1.8 to 2.0 GHz. Suitable for FM, TDMA, CDMA and multicarrier amplifier applications. To be used in class AB for GSM and EDGE cellular radio applications.

- GSM and EDGE Performances, Full Frequency Band
Power Gain — 13.5 dB (Typ) @ 90 Watts (CW)
Efficiency — 52% (Typ) @ 90 Watts (CW)
- Internally Matched, Controlled Q, for Ease of Use
- High Gain, High Efficiency and High Linearity
- Integrated ESD Protection
- Designed for Maximum Gain and Insertion Phase Flatness
- Capable of Handling 10:1 VSWR, @ 26 Vdc, 90 Watts (CW) Output Power
- Excellent Thermal Stability
- Characterized with Series Equivalent Large-Signal Impedance Parameters

1.80 – 1.88 GHz, 90 W, 26 V
LATERAL N-CHANNEL
RF POWER MOSFETS



CASE 465B-03, STYLE 1
(NI-880)
(MRF18090A)



CASE 465C-02, STYLE 1
(NI-880S)
(MRF18090AS)

MAXIMUM RATINGS

| Rating | Symbol | Value | Unit |
|--|-----------|-------------|------------------------------|
| Drain-Source Voltage | V_{DSS} | 65 | Vdc |
| Gate-Source Voltage | V_{GS} | +15, -0.5 | Vdc |
| Total Device Dissipation @ $T_C = 25^\circ\text{C}$ Derate above 25°C | P_D | 250 1.43 | Watts W/ $^\circ\text{C}$ |
| Storage Temperature Range | T_{stg} | -65 to +200 | $^\circ\text{C}$ |
| Operating Junction Temperature | T_J | 200 | $^\circ\text{C}$ |

ESD PROTECTION CHARACTERISTICS

| Test Conditions | Class |
|------------------|--------------|
| Human Body Model | 2 (Minimum) |
| Machine Model | M3 (Minimum) |

THERMAL CHARACTERISTICS

| Characteristic | Symbol | Max | Unit |
|--------------------------------------|-----------------|-----|---------------------------|
| Thermal Resistance, Junction to Case | $R_{\theta JC}$ | 0.7 | $^\circ\text{C}/\text{W}$ |

NOTE – **CAUTION** – MOS devices are susceptible to damage from electrostatic charge. Reasonable precautions in handling and packaging MOS devices should be observed.

ELECTRICAL CHARACTERISTICS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

| Characteristic | Symbol | Min | Typ | Max | Unit |
|----------------|--------|-----|-----|-----|------|
|----------------|--------|-----|-----|-----|------|

OFF CHARACTERISTICS

| | | | | | |
|--|---------------|----|---|----|-----------------|
| Drain–Source Breakdown Voltage ($V_{GS} = 0\text{ Vdc}$, $I_D = 100\ \mu\text{Adc}$) | $V_{(BR)DSS}$ | 65 | — | — | Vdc |
| Zero Gate Voltage Drain Current ($V_{DS} = 26\text{ Vdc}$, $V_{GS} = 0\text{ Vdc}$) | I_{DSS} | — | — | 10 | μAdc |
| Gate–Source Leakage Current ($V_{GS} = 5\text{ Vdc}$, $V_{DS} = 0\text{ Vdc}$) | I_{GSS} | — | — | 1 | μAdc |

ON CHARACTERISTICS

| | | | | | |
|--|--------------|-----|-----|-----|-----|
| Gate Quiescent Voltage ($V_{DS} = 26\text{ Vdc}$, $I_D = 750\text{ mAdc}$) | $V_{GS(Q)}$ | 2.5 | 3.7 | 4.5 | Vdc |
| Drain–Source On–Voltage ($V_{GS} = 10\text{ Vdc}$, $I_D = 1\text{ Adc}$) | $V_{DS(on)}$ | — | 0.1 | — | Vdc |
| Forward Transconductance ($V_{DS} = 10\text{ Vdc}$, $I_D = 3\text{ Adc}$) | g_{fs} | — | 7.2 | — | S |

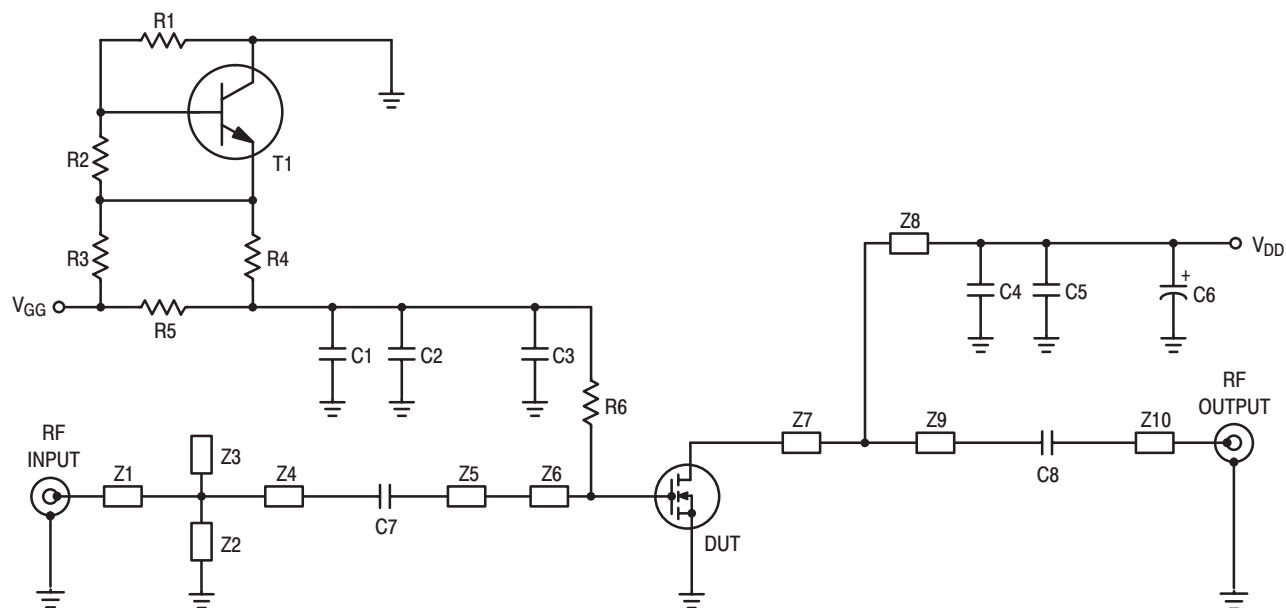
DYNAMIC CHARACTERISTICS

| | | | | | |
|---|-----------|---|-----|---|----|
| Reverse Transfer Capacitance ($V_{DS} = 26\text{ Vdc} \pm 30\text{ mV(rms)ac}$ @ 1 MHz, $V_{GS} = 0\text{ Vdc}$) | C_{rss} | — | 4.2 | — | pF |
|---|-----------|---|-----|---|----|

FUNCTIONAL TESTS (In Motorola Test Fixture)

| | | | | | |
|--|----------|---|------|-----|----|
| Common–Source Amplifier Power Gain @ 90 W (1) ($V_{DD} = 26\text{ Vdc}$, $I_{DQ} = 750\text{ mA}$, $f = 1805 - 1880\text{ MHz}$) | G_{ps} | 12.0 | 13.5 | — | dB |
| Drain Efficiency @ 90 W (1) ($V_{DD} = 26\text{ Vdc}$, $I_{DQ} = 750\text{ mA}$, $f = 1805 - 1880\text{ MHz}$) | η | 47 | 52 | — | % |
| Input Return Loss (1) ($V_{DD} = 26\text{ Vdc}$, $P_{out} = 90\text{ W CW}$, $I_{DQ} = 750\text{ mA}$, $f = 1805 - 1880\text{ MHz}$) | IRL | — | — | -10 | dB |
| Output Mismatch Stress ($V_{DD} = 26\text{ Vdc}$, $P_{out} = 90\text{ W CW}$, $I_{DQ} = 750\text{ mA}$ VSWR = 10:1, All Phase Angles at Frequency of Tests) | Ψ | No Degradation In Output Power Before and After Test | | | |

(1) To meet application requirements, Motorola test fixtures have been designed to cover the full GSM1800 band, ensuring batch-to-batch consistency.



| | | | |
|------------|--|-----|--------------------------------------|
| C1, C3 | 1.0 μ F Chip Capacitors (0805) | Z2 | 0.197" x 0.087" Microstrip |
| C2 | 1.0 nF Chip Capacitor (0805) | Z3 | 0.819" x 0.087" Microstrip |
| C4, C5 | 6.8 pF, 100B Chip Capacitors, ATC | Z4 | 0.181" x 0.144" Microstrip |
| C6 | 220 μ F, 50 V Electrolytic Capacitor | Z5 | 0.383" x 1.148" Microstrip |
| C7, C8 | 12 pF, 100B Chip Capacitors, ATC | Z6 | 0.400" x 1.380" Microstrip |
| R1 | 2.2 k Ω Chip Resistor (0805) | Z7 | 0.351" x 0.351" Microstrip |
| R2, R3, R6 | 1.0 k Ω Chip Resistors (0805) | Z8 | 0.126" x 0.087" Microstrip |
| R4 | 10 k Ω Chip Resistor (0805) | Z9 | 1.280" x 0.087" Microstrip |
| R5 | 6.8 k Ω Chip Resistor (0805) | Z10 | \approx 1.275" x 0.055" Microstrip |
| T1 | BC847 SOT-23 | PCB | Teflon [®] Glass |
| Z1 | 0.697" x 0.087" Microstrip | | |

Figure 1. 1.80 – 1.88 GHz Test Fixture Schematic

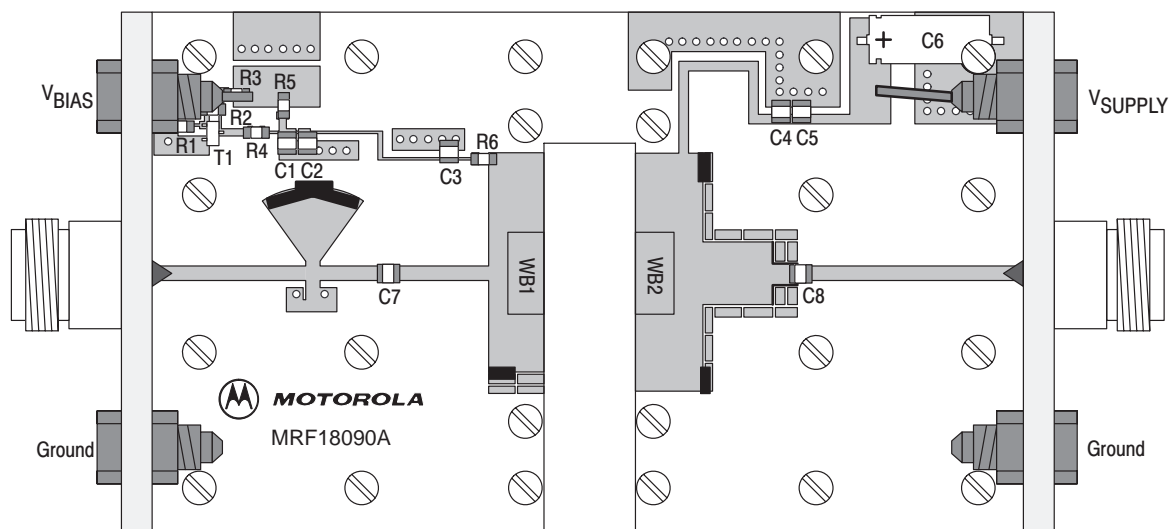
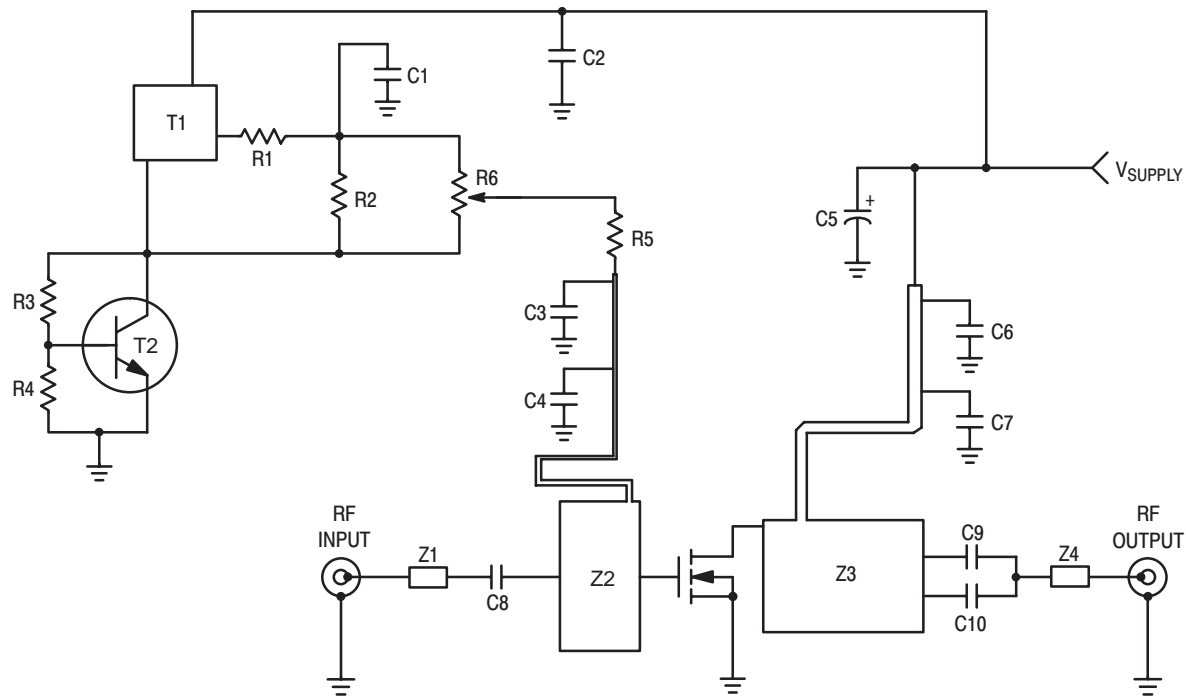


Figure 2. 1.80 – 1.88 GHz Test Fixture Component Layout



| | | | |
|-------------|--|----|--|
| C1, C3 | 1 μ F Chip Capacitors (0805) | R5 | 10 k Ω Chip Resistor (0603) |
| C2 | 0.1 μ F Chip Capacitor (0805) | R6 | 5 k Ω , SMD Potentiometer |
| C4 | 1 nF Chip Capacitor (0805) | T1 | LP2951 Micro-8 Voltage Regulator |
| C5 | 220 μ F, 50 V Electrolytic Capacitor | T2 | BC847 SOT-23 NPN Transistor |
| C6, C7 | 8.2 pF, 100A Chip Capacitors | Z1 | 0.210" x 0.055" Microstrip |
| C8, C9, C10 | 22 pF, 100A Chip Capacitors | Z2 | 0.419" x 0.787" Microstrip |
| R1 | 10 Ω Chip Resistor (0805) | Z3 | 0.836" x 0.512" Microstrip |
| R2, R3 | 1 k Ω Chip Resistors (0805) | Z4 | 0.164" x 0.055" Microstrip |
| R4 | 2.2 k Ω Chip Resistor (0805) | | Substrate = 0.5 mm Teflon [®] Glass |

Figure 3. 1.80 – 1.88 GHz Demo Board Schematic

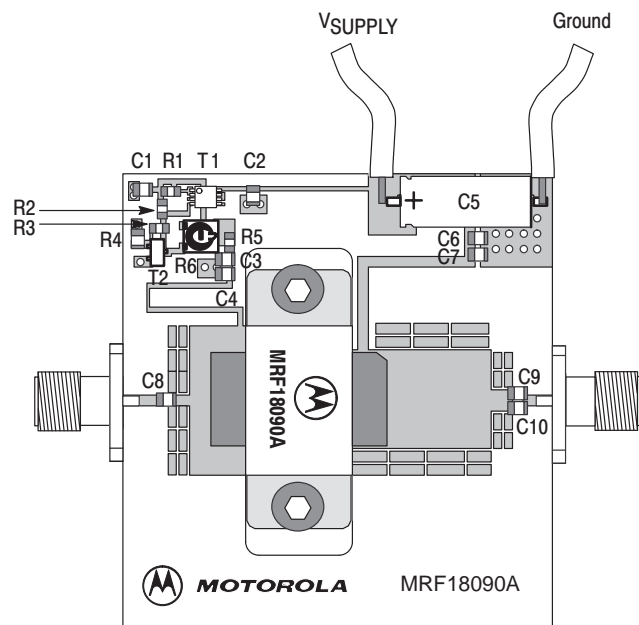


Figure 4. 1.80 – 1.88 GHz Demo Board Component Layout

TYPICAL CHARACTERISTICS

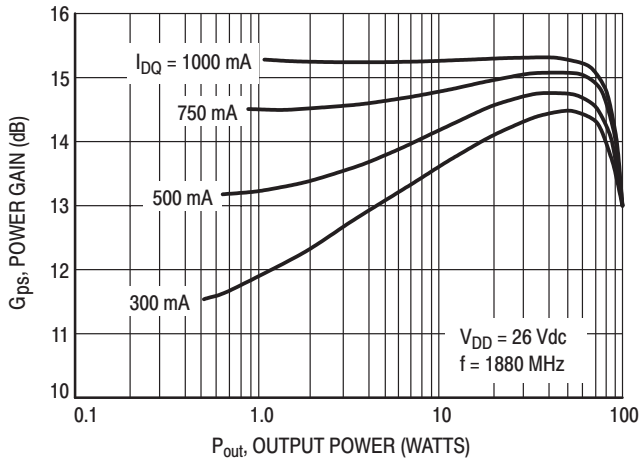


Figure 5. Power Gain versus Output Power

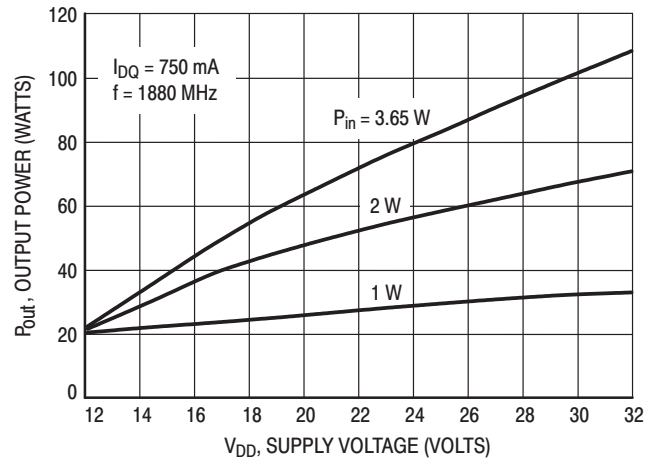


Figure 6. Output Power versus Supply Voltage

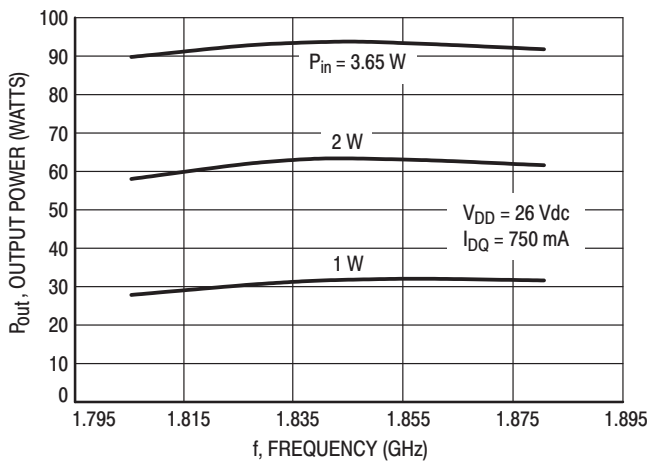


Figure 7. Output Power versus Frequency

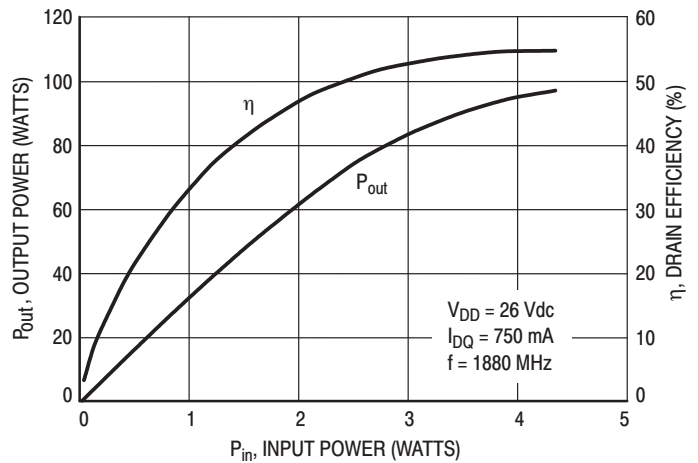


Figure 8. Output Power and Efficiency versus Input Power

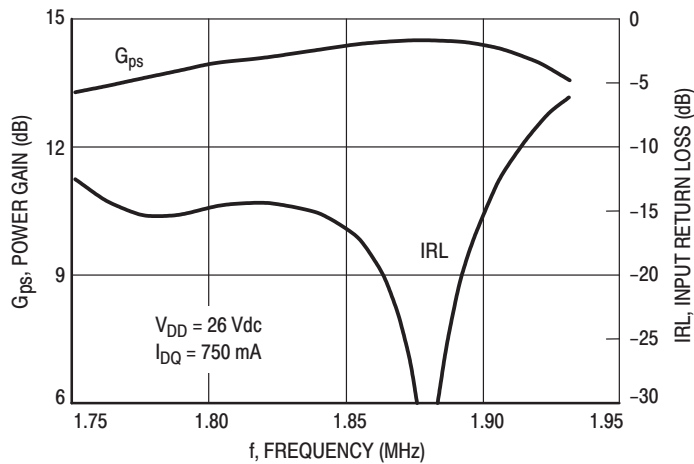
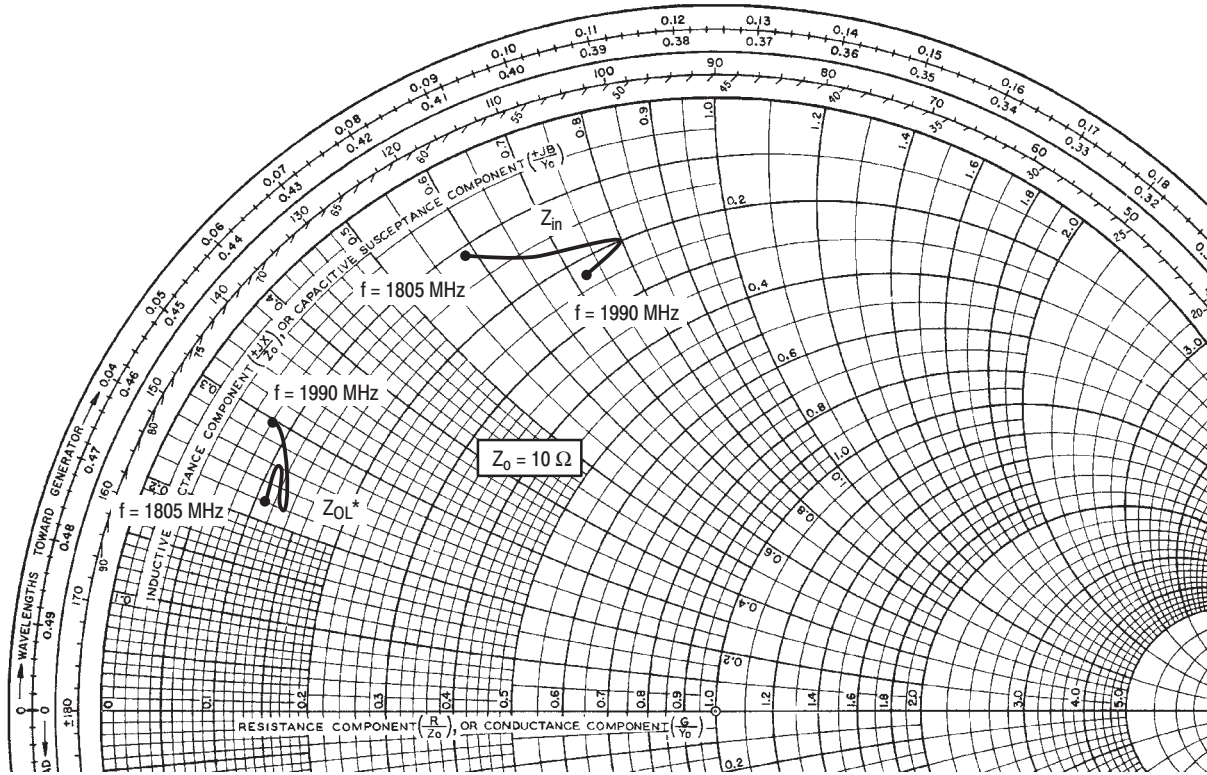


Figure 9. Wideband Gain and IRL (at Small Signal)



$V_{DD} = 26\text{ V}$, $I_{DQ} = 750\text{ mA}$, $P_{out} = 90\text{ Watts (CW)}$

| f MHz | Z_{in} Ω | Z_{OL}^* Ω |
|----------|----------------------|------------------------|
| 1805 | $1.10 + j5.85$ | $1.15 + j2.16$ |
| 1880 | $1.56 + j6.75$ | $1.13 + j2.60$ |
| 1930 | $2.05 + j8.00$ | $1.30 + j2.23$ |
| 1990 | $2.30 + j7.30$ | $0.82 + j2.90$ |

Z_{in} = Complex conjugate of the source impedance.

Z_{OL}^* = Complex conjugate of the optimum load at a given voltage, P1dB, gain, efficiency, bias current and frequency.

Note: Z_{OL}^* was chosen based on tradeoffs between gain, output power, and drain efficiency.

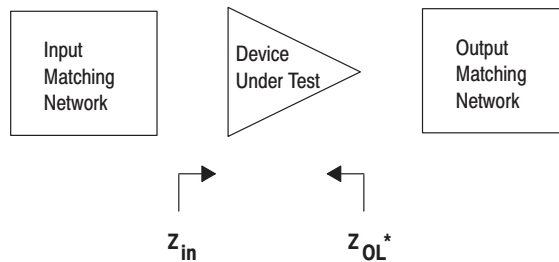
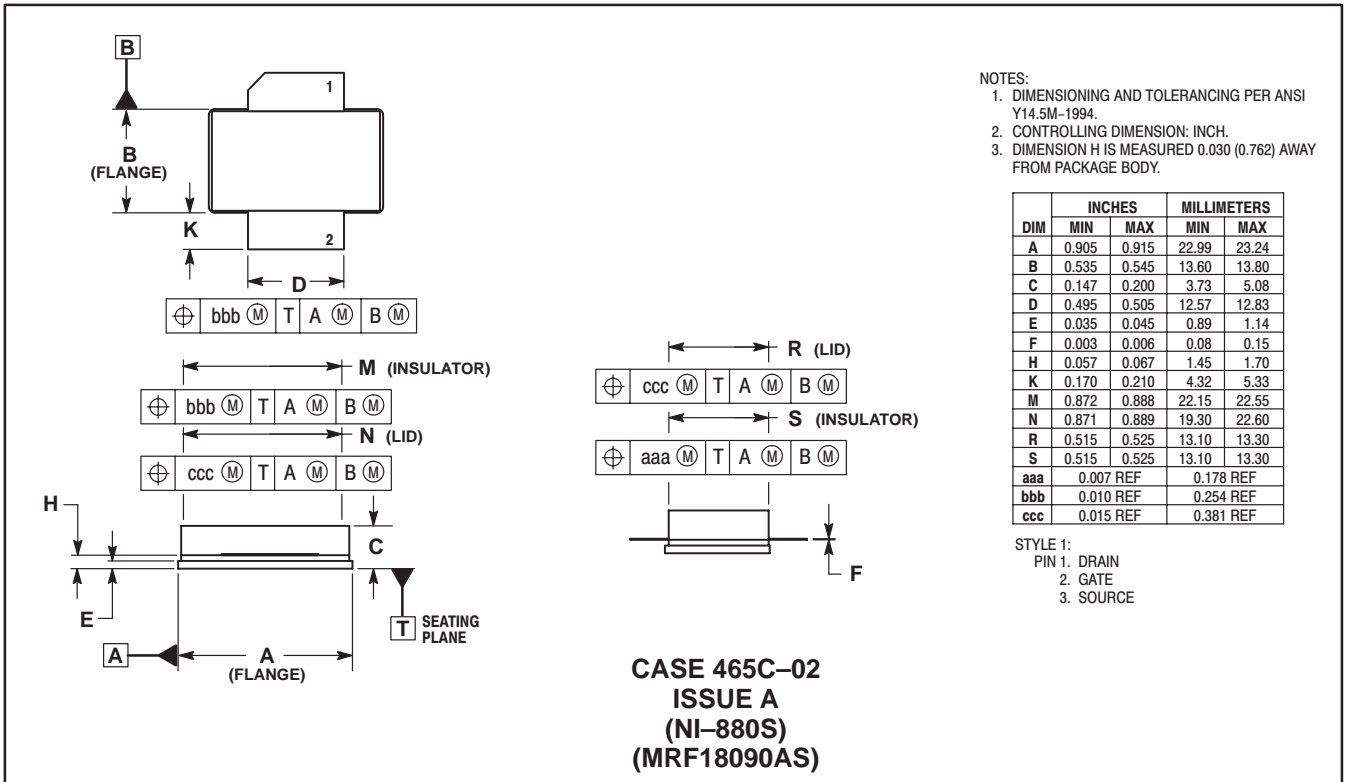
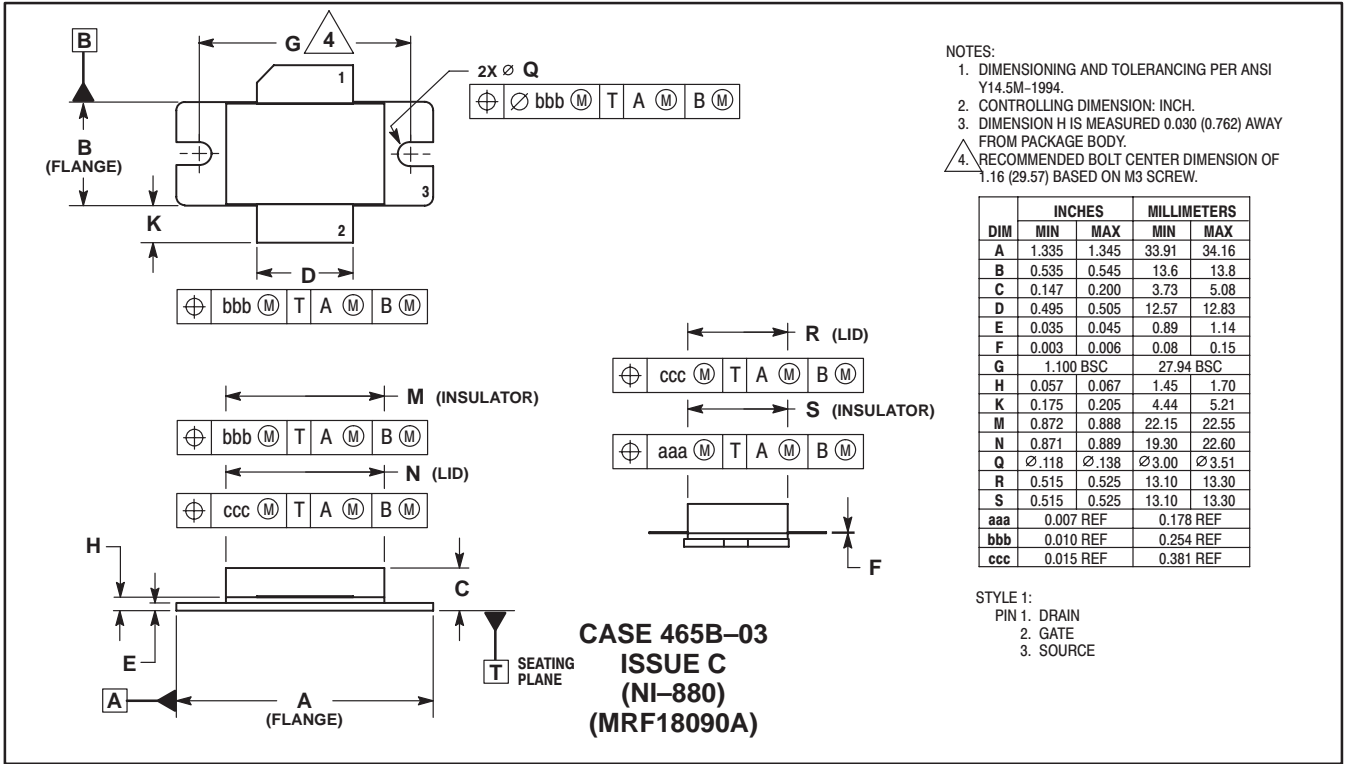



Figure 10. Large Signal Input and Output Impedance

PACKAGE DIMENSIONS



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