



FLEX 10K

Embedded Programmable Logic Device Family

January 2003, ver. 4.2

Data Sheet

Features...

- The industry's first embedded programmable logic device (PLD) family, providing System-on-a-Programmable-Chip (SOPC) integration
 - Embedded array for implementing megafunctions, such as efficient memory and specialized logic functions
 - Logic array for general logic functions
- High density
 - 10,000 to 250,000 typical gates (see Tables 1 and 2)
 - Up to 40,960 RAM bits; 2,048 bits per embedded array block (EAB), all of which can be used without reducing logic capacity
- System-level features
 - MultiVoltTM I/O interface support
 - 5.0-V tolerant input pins in FLEX® 10KA devices
 - Low power consumption (typical specification less than 0.5 mA in standby mode for most devices)
 - FLEX 10K and FLEX 10KA devices support peripheral component interconnect Special Interest Group (PCI SIG) PCI Local Bus Specification, Revision 2.2
 - FLEX 10KA devices include pull-up clamping diode, selectable on a pin-by-pin basis for 3.3-V PCI compliance
 - Select FLEX 10KA devices support 5.0-V PCI buses with eight or fewer loads
 - Built-in Joint Test Action Group (JTAG) boundary-scan test (BST) circuitry compliant with IEEE Std. 1149.1-1990, available without consuming any device logic

246

189

310

Table 1. FLEX 10K Device Features					
Feature	EPF10K10 EPF10K10A	EPF10K20	EPF10K30 EPF10K30A	EPF10K40	EPF10K50 EPF10K50V
Typical gates (logic and RAM) (1)	10,000	20,000	30,000	40,000	50,000
Maximum system gates	31,000	63,000	69,000	93,000	116,000
Logic elements (LEs)	576	1,152	1,728	2,304	2,880
Logic array blocks (LABs)	72	144	216	288	360
Embedded array blocks (EABs)	3	6	6	8	10
Total RAM bits	6,144	12,288	12,288	16,384	20,480

Altera Corporation 1

189

150

Maximum user I/O pins

Table 2. FLEX 10K Device Features					
Feature	EPF10K70	EPF10K100 EPF10K100A	EPF10K130V	EPF10K250A	
Typical gates (logic and RAM) (1)	70,000	100,000	130,000	250,000	
Maximum system gates	118,000	158,000	211,000	310,000	
LEs	3,744	4,992	6,656	12,160	
LABs	468	624	832	1,520	
EABs	9	12	16	20	
Total RAM bits	18,432	24,576	32,768	40,960	
Maximum user I/O pins	358	406	470	470	

Note to tables:

(1) The embedded IEEE Std. 1149.1 JTAG circuitry adds up to 31,250 gates in addition to the listed typical or maximum system gates.

...and More Features

- Devices are fabricated on advanced processes and operate with a 3.3-V or 5.0-V supply voltage (see Table 3
- In-circuit reconfigurability (ICR) via external configuration device, intelligent controller, or JTAG port
- ClockLockTM and ClockBoostTM options for reduced clock delay/skew and clock multiplication
- Built-in low-skew clock distribution trees
- 100% functional testing of all devices; test vectors or scan chains are not required

Table 3. Supply Voltages for FLEX 10K & FLEX 10KA Devices				
5.0-V Devices	3.3-V Devices			
EPF10K10	EPF10K10A			
EPF10K20	EPF10K30A			
EPF10K30	EPF10K50V			
EPF10K40	EPF10K100A			
EPF10K50	EPF10K130V			
EPF10K70	EPF10K250A			
EPF10K100				

- Flexible interconnect
 - FastTrack[®] Interconnect continuous routing structure for fast, predictable interconnect delays
 - Dedicated carry chain that implements arithmetic functions such as fast adders, counters, and comparators (automatically used by software tools and megafunctions)
 - Dedicated cascade chain that implements high-speed, high-fan-in logic functions (automatically used by software tools and megafunctions)
 - Tri-state emulation that implements internal tri-state buses
 - Up to six global clock signals and four global clear signals
- Powerful I/O pins
 - Individual tri-state output enable control for each pin
 - Open-drain option on each I/O pin
 - Programmable output slew-rate control to reduce switching noise
 - FLEX 10KA devices support hot-socketing
- Peripheral register for fast setup and clock-to-output delay
- Flexible package options
 - Available in a variety of packages with 84 to 600 pins (see Tables 4 and 5)
 - Pin-compatibility with other FLEX 10K devices in the same package
 - FineLine BGATM packages maximize board space efficiency
- Software design support and automatic place-and-route provided by Altera development systems for Windows-based PCs and Sun SPARCstation, HP 9000 Series 700/800 workstations
- Additional design entry and simulation support provided by EDIF 2 0 0 and 3 0 0 netlist files, library of parameterized modules (LPM), DesignWare components, Verilog HDL, VHDL, and other interfaces to popular EDA tools from manufacturers such as Cadence, Exemplar Logic, Mentor Graphics, OrCAD, Synopsys, Synplicity, VeriBest, and Viewlogic

Device	84-Pin PLCC	100-Pin TQFP	144-Pin TQFP	208-Pin PQFP RQFP	240-Pin PQFP RQFP
EPF10K10	59		102	134	
EPF10K10A		66	102	134	
EPF10K20			102	147	189
EPF10K30				147	189
EPF10K30A			102	147	189
EPF10K40				147	189
EPF10K50					189
EPF10K50V					189
EPF10K70					189
EPF10K100					
EPF10K100A					189
EPF10K130V					
EPF10K250A					

Table 5. FLEX 1	OK Packa	ge Options	s & I/O Pin Count	(Continue	d) Note (1)		
Device	503-Pin PGA	599-Pin PGA	256-Pin FineLine BGA	356-Pin BGA	484-Pin FineLine BGA	600-Pin BGA	403-Pin PGA
EPF10K10							
EPF10K10A			150		150 (2)		
EPF10K20							
EPF10K30				246			
EPF10K30A			191	246	246		
EPF10K40							
EPF10K50				274			310
EPF10K50V				274			
EPF10K70	358						
EPF10K100	406						
EPF10K100A				274	369	406	
EPF10K130V		470				470	
EPF10K250A		470				470	

Notes to tables:

- (1) FLEX 10K and FLEX 10KA device package types include plastic J-lead chip carrier (PLCC), thin quad flat pack (TQFP), plastic quad flat pack (PQFP), power quad flat pack (RQFP), ball-grid array (BGA), pin-grid array (PGA), and FineLine BGATM packages.
- (2) This option is supported with a 256-pin FineLine BGA package. By using SameFrame pin migration, all FineLine BGA packages are pin compatible. For example, a board can be designed to support both 256-pin and 484-pin FineLine BGA packages. The Altera software automatically avoids conflicting pins when future migration is set.

General Description

Altera's FLEX 10K devices are the industry's first embedded PLDs. Based on reconfigurable CMOS SRAM elements, the Flexible Logic Element MatriX (FLEX) architecture incorporates all features necessary to implement common gate array megafunctions. With up to 250,000 gates, the FLEX 10K family provides the density, speed, and features to integrate entire systems, including multiple 32-bit buses, into a single device.

FLEX 10K devices are reconfigurable, which allows 100% testing prior to shipment. As a result, the designer is not required to generate test vectors for fault coverage purposes. Additionally, the designer does not need to manage inventories of different ASIC designs; FLEX 10K devices can be configured on the board for the specific functionality required.

Table 6 shows FLEX 10K performance for some common designs. All performance values were obtained with Synopsys DesignWare or LPM functions. No special design technique was required to implement the applications; the designer simply inferred or instantiated a function in a Verilog HDL, VHDL, Altera Hardware Description Language (AHDL), or schematic design file.

Table 6. FLEX 10K & FLEX 10KA Performance							
Application	Application Resources Performance Used					Units	
	LEs	EABs	-1 Speed Grade	-2 Speed Grade	-3 Speed Grade	-4 Speed Grade	
16-bit loadable counter (1)	16	0	204	166	125	95	MHz
16-bit accumulator (1)	16	0	204	166	125	95	MHz
16-to-1 multiplexer (2)	10	0	4.2	5.8	6.0	7.0	ns
256 × 8 RAM read cycle speed (3)	0	1	172	145	108	84	MHz
256 × 8 RAM write cycle speed (3)	0	1	106	89	68	63	MHz

Notes:

- (1) The speed grade of this application is limited because of clock high and low specifications.
- (2) This application uses combinatorial inputs and outputs.
- (3) This application uses registered inputs and outputs.

The FLEX 10K architecture is similar to that of embedded gate arrays, the fastest-growing segment of the gate array market. As with standard gate arrays, embedded gate arrays implement general logic in a conventional "sea-of-gates" architecture. In addition, embedded gate arrays have dedicated die areas for implementing large, specialized functions. By embedding functions in silicon, embedded gate arrays provide reduced die area and increased speed compared to standard gate arrays. However, embedded megafunctions typically cannot be customized, limiting the designer's options. In contrast, FLEX 10K devices are programmable, providing the designer with full control over embedded megafunctions and general logic while facilitating iterative design changes during debugging.

Each FLEX 10K device contains an embedded array and a logic array. The embedded array is used to implement a variety of memory functions or complex logic functions, such as digital signal processing (DSP), microcontroller, wide-data-path manipulation, and data-transformation functions. The logic array performs the same function as the sea-of-gates in the gate array: it is used to implement general logic, such as counters, adders, state machines, and multiplexers. The combination of embedded and logic arrays provides the high performance and high density of embedded gate arrays, enabling designers to implement an entire system on a single device.

FLEX 10K devices are configured at system power-up with data stored in an Altera serial configuration device or provided by a system controller. Altera offers the EPC1, EPC2, EPC16, and EPC1441 configuration devices, which configure FLEX 10K devices via a serial data stream. Configuration data can also be downloaded from system RAM or from Altera's BitBlasterTM serial download cable or ByteBlasterMVTM parallel port download cable. After a FLEX 10K device has been configured, it can be reconfigured in-circuit by resetting the device and loading new data. Because reconfiguration requires less than 320 ms, real-time changes can be made during system operation.

FLEX 10K devices contain an optimized interface that permits microprocessors to configure FLEX 10K devices serially or in parallel, and synchronously or asynchronously. The interface also enables microprocessors to treat a FLEX 10K device as memory and configure the device by writing to a virtual memory location, making it very easy for the designer to reconfigure the device.



For more information, see the following documents:

- Configuration Devices for APEX & FLEX Devices Data Sheet
- BitBlaster Serial Download Cable Data Sheet
- ByteBlasterMV Parallel Port Download Cable Data Sheet
- Application Note 116 (Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices)

FLEX 10K devices are supported by Altera development systems; single, integrated packages that offer schematic, text (including AHDL), and waveform design entry, compilation and logic synthesis, full simulation and worst-case timing analysis, and device configuration. The Altera software provides EDIF 2 0 0 and 3 0 0, LPM, VHDL, Verilog HDL, and other interfaces for additional design entry and simulation support from other industry-standard PC- and UNIX workstation-based EDA tools.

The Altera software works easily with common gate array EDA tools for synthesis and simulation. For example, the Altera software can generate Verilog HDL files for simulation with tools such as Cadence Verilog-XL. Additionally, the Altera software contains EDA libraries that use device-specific features such as carry chains which are used for fast counter and arithmetic functions. For instance, the Synopsys Design Compiler library supplied with the Altera development systems include DesignWare functions that are optimized for the FLEX 10K architecture.

The Altera development systems run on Windows-based PCs and Sun SPARCstation, and HP 9000 Series 700/800 workstations.



See the MAX+PLUS II Programmable Logic Development System & Software Data Sheet for more information.

Functional Description

Each FLEX 10K device contains an embedded array to implement memory and specialized logic functions, and a logic array to implement general logic.

The embedded array consists of a series of EABs. When implementing memory functions, each EAB provides 2,048 bits, which can be used to create RAM, ROM, dual-port RAM, or first-in first-out (FIFO) functions. When implementing logic, each EAB can contribute 100 to 600 gates towards complex logic functions, such as multipliers, microcontrollers, state machines, and DSP functions. EABs can be used independently, or multiple EABs can be combined to implement larger functions.

The logic array consists of logic array blocks (LABs). Each LAB contains eight LEs and a local interconnect. An LE consists of a 4-input look-up table (LUT), a programmable flipflop, and dedicated signal paths for carry and cascade functions. The eight LEs can be used to create medium-sized blocks of logic—8-bit counters, address decoders, or state machines—or combined across LABs to create larger logic blocks. Each LAB represents about 96 usable gates of logic.

Signal interconnections within FLEX 10K devices and to and from device pins are provided by the FastTrack Interconnect, a series of fast, continuous row and column channels that run the entire length and width of the device.

Each I/O pin is fed by an I/O element (IOE) located at the end of each row and column of the FastTrack Interconnect. Each IOE contains a bidirectional I/O buffer and a flipflop that can be used as either an output or input register to feed input, output, or bidirectional signals. When used with a dedicated clock pin, these registers provide exceptional performance. As inputs, they provide setup times as low as 1.6 ns and hold times of 0 ns; as outputs, these registers provide clock-to-output times as low as 5.3 ns. IOEs provide a variety of features, such as JTAG BST support, slew-rate control, tri-state buffers, and open-drain outputs.

Figure 1 shows a block diagram of the FLEX 10K architecture. Each group of LEs is combined into an LAB; LABs are arranged into rows and columns. Each row also contains a single EAB. The LABs and EABs are interconnected by the FastTrack Interconnect. IOEs are located at the end of each row and column of the FastTrack Interconnect.

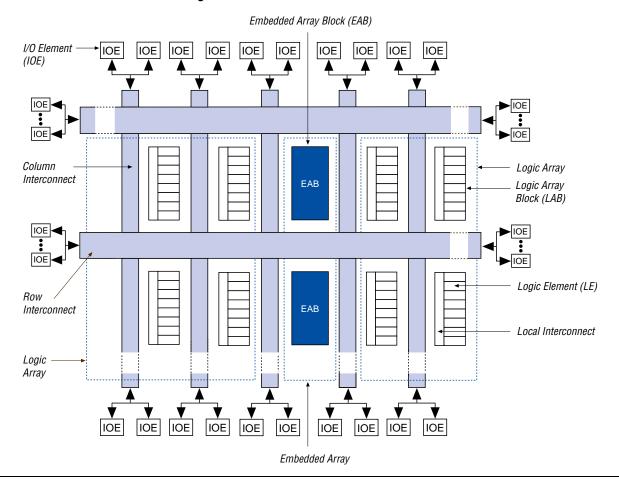


Figure 1. FLEX 10K Device Block Diagram

FLEX 10K devices provide six dedicated inputs that drive the flipflops' control inputs to ensure the efficient distribution of high-speed, low-skew (less than 1.5 ns) control signals. These signals use dedicated routing channels that provide shorter delays and lower skews than the FastTrack Interconnect. Four of the dedicated inputs drive four global signals. These four global signals can also be driven by internal logic, providing an ideal solution for a clock divider or an internally generated asynchronous clear signal that clears many registers in the device.

Embedded Array Block

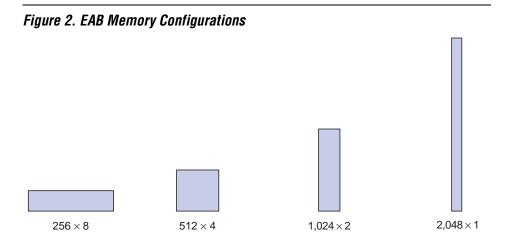
The EAB is a flexible block of RAM with registers on the input and output ports, and is used to implement common gate array megafunctions. The EAB is also suitable for functions such as multipliers, vector scalars, and error correction circuits, because it is large and flexible. These functions can be combined in applications such as digital filters and microcontrollers.

Logic functions are implemented by programming the EAB with a readonly pattern during configuration, creating a large LUT. With LUTs, combinatorial functions are implemented by looking up the results, rather than by computing them. This implementation of combinatorial functions can be faster than using algorithms implemented in general logic, a performance advantage that is further enhanced by the fast access times of EABs. The large capacity of EABs enables designers to implement complex functions in one logic level without the routing delays associated with linked LEs or field-programmable gate array (FPGA) RAM blocks. For example, a single EAB can implement a 4×4 multiplier with eight inputs and eight outputs. Parameterized functions such as LPM functions can automatically take advantage of the EAB.

The EAB provides advantages over FPGAs, which implement on-board RAM as arrays of small, distributed RAM blocks. These FPGA RAM blocks contain delays that are less predictable as the size of the RAM increases. In addition, FPGA RAM blocks are prone to routing problems because small blocks of RAM must be connected together to make larger blocks. In contrast, EABs can be used to implement large, dedicated blocks of RAM that eliminate these timing and routing concerns.

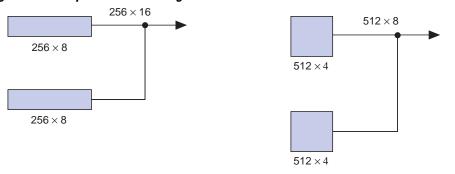
EABs can be used to implement synchronous RAM, which is easier to use than asynchronous RAM. A circuit using asynchronous RAM must generate the RAM write enable (WE) signal, while ensuring that its data and address signals meet setup and hold time specifications relative to the WE signal. In contrast, the EAB's synchronous RAM generates its own WE signal and is self-timed with respect to the global clock. A circuit using the EAB's self-timed RAM need only meet the setup and hold time specifications of the global clock.

When used as RAM, each EAB can be configured in any of the following sizes: 256×8 , 512×4 , $1,024 \times 2$, or $2,048 \times 1$. See Figure 2.



Larger blocks of RAM are created by combining multiple EABs. For example, two 256×8 RAM blocks can be combined to form a 256×16 RAM block; two 512×4 blocks of RAM can be combined to form a 512×8 RAM block. See Figure 3.

Figure 3. Examples of Combining EABs



If necessary, all EABs in a device can be cascaded to form a single RAM block. EABs can be cascaded to form RAM blocks of up to 2,048 words without impacting timing. Altera's software automatically combines EABs to meet a designer's RAM specifications.

EABs provide flexible options for driving and controlling clock signals. Different clocks can be used for the EAB inputs and outputs. Registers can be independently inserted on the data input, EAB output, or the address and WE inputs. The global signals and the EAB local interconnect can drive the WE signal. The global signals, dedicated clock pins, and EAB local interconnect can drive the EAB clock signals. Because the LEs drive the EAB local interconnect, the LEs can control the WE signal or the EAB clock signals.

Each EAB is fed by a row interconnect and can drive out to row and column interconnects. Each EAB output can drive up to two row channels and up to two column channels; the unused row channel can be driven by other LEs. This feature increases the routing resources available for EAB outputs. See Figure 4.

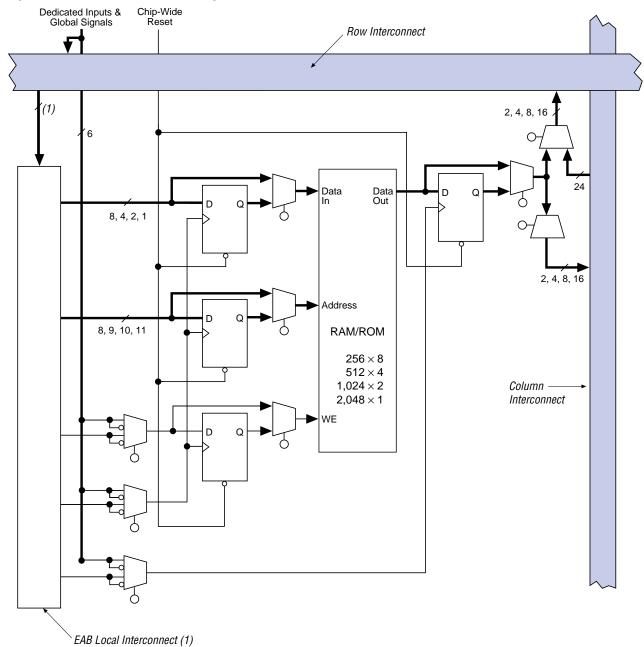


Figure 4. FLEX 10K Embedded Array Block

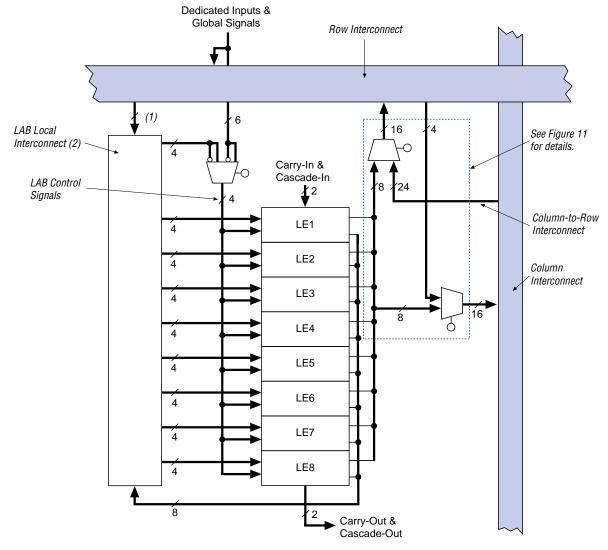
Note

(1) EPF10K10, EPF10K10A, EPF10K20, EPF10K30, EPF10K30A, EPF10K40, EPF10K50, and EPF10K50V devices have 22 EAB local interconnect channels; EPF10K70, EPF10K100, EPF10K100A, EPF10K130V, and EPF10K250A devices have 26.

Logic Array Block

Each LAB consists of eight LEs, their associated carry and cascade chains, LAB control signals, and the LAB local interconnect. The LAB provides the coarse-grained structure to the FLEX 10K architecture, facilitating efficient routing with optimum device utilization and high performance. See Figure 5.

Figure 5. FLEX 10K LAB



Notes:

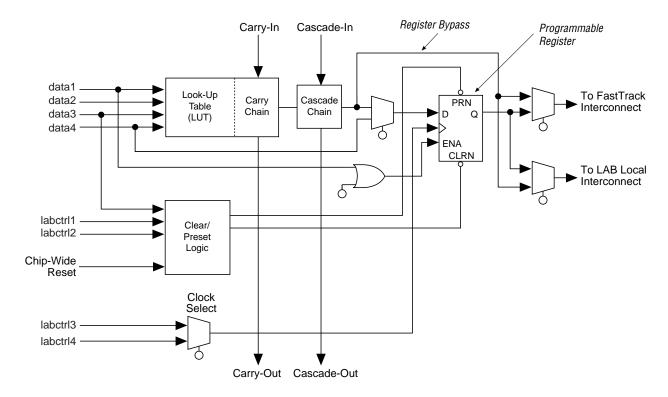
- (1) EPF10K10, EPF10K10A, EPF10K20, EPF10K30, EPF10K30A, EPF10K40, EPF10K50, and EPF10K50V devices have 22 inputs to the LAB local interconnect channel from the row; EPF10K70, EPF10K100, EPF10K100A, EPF10K130V, and EPF10K250A devices have 26.
- (2) EPF10K10, EPF10K10A, EPF10K20, EPF10K30, EPF10K30A, EPF10K40, EPF10K50, and EPF10K50V devices have 30 LAB local interconnect channels; EPF10K70, EPF10K100, EPF10K100A, EPF10K130V, and EPF10K250A devices have 34 LABs.

Each LAB provides four control signals with programmable inversion that can be used in all eight LEs. Two of these signals can be used as clocks; the other two can be used for clear/preset control. The LAB clocks can be driven by the dedicated clock input pins, global signals, I/O signals, or internal signals via the LAB local interconnect. The LAB preset and clear control signals can be driven by the global signals, I/O signals, or internal signals via the LAB local interconnect. The global control signals are typically used for global clock, clear, or preset signals because they provide asynchronous control with very low skew across the device. If logic is required on a control signal, it can be generated in one or more LEs in any LAB and driven into the local interconnect of the target LAB. In addition, the global control signals can be generated from LE outputs.

Logic Element

The LE, the smallest unit of logic in the FLEX 10K architecture, has a compact size that provides efficient logic utilization. Each LE contains a four-input LUT, which is a function generator that can quickly compute any function of four variables. In addition, each LE contains a programmable flipflop with a synchronous enable, a carry chain, and a cascade chain. Each LE drives both the local and the FastTrack Interconnect. See Figure 6.

Figure 6. FLEX 10K Logic Element



The programmable flipflop in the LE can be configured for D, T, JK, or SR operation. The clock, clear, and preset control signals on the flipflop can be driven by global signals, general-purpose I/O pins, or any internal logic. For combinatorial functions, the flipflop is bypassed and the output of the LUT drives the output of the LE.

The LE has two outputs that drive the interconnect; one drives the local interconnect and the other drives either the row or column FastTrack Interconnect. The two outputs can be controlled independently. For example, the LUT can drive one output while the register drives the other output. This feature, called register packing, can improve LE utilization because the register and the LUT can be used for unrelated functions.

The FLEX 10K architecture provides two types of dedicated high-speed data paths that connect adjacent LEs without using local interconnect paths: carry chains and cascade chains. The carry chain supports high-speed counters and adders; the cascade chain implements wide-input functions with minimum delay. Carry and cascade chains connect all LEs in an LAB and all LABs in the same row. Intensive use of carry and cascade chains can reduce routing flexibility. Therefore, the use of these chains should be limited to speed-critical portions of a design.

Carry Chain

The carry chain provides a very fast (as low as 0.2 ns) carry-forward function between LEs. The carry-in signal from a lower-order bit drives forward into the higher-order bit via the carry chain, and feeds into both the LUT and the next portion of the carry chain. This feature allows the FLEX 10K architecture to implement high-speed counters, adders, and comparators of arbitrary width efficiently. Carry chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry. Parameterized functions such as LPM and DesignWare functions automatically take advantage of carry chains.

Carry chains longer than eight LEs are automatically implemented by linking LABs together. For enhanced fitting, a long carry chain skips alternate LABs in a row. A carry chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB. For example, the last LE of the first LAB in a row carries to the first LE of the third LAB in the row. The carry chain does not cross the EAB at the middle of the row. For instance, in the EPF10K50 device, the carry chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB.

Figure 7 shows how an n-bit full adder can be implemented in n+1 LEs with the carry chain. One portion of the LUT generates the sum of two bits using the input signals and the carry-in signal; the sum is routed to the output of the LE. The register can either be bypassed for simple adders or be used for an accumulator function. The carry chain logic generates the carry-out signal, which is routed directly to the carry-in signal of the next-higher-order bit. The final carry-out signal is routed to an LE, where it can be used as a general-purpose signal.

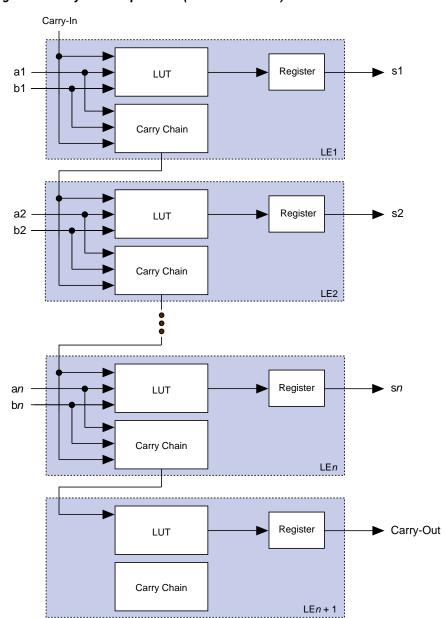


Figure 7. Carry Chain Operation (n-bit Full Adder)

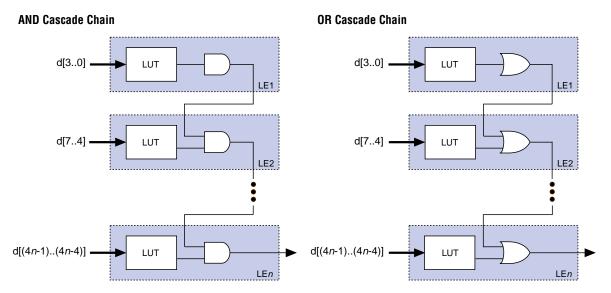
Cascade Chain

With the cascade chain, the FLEX 10K architecture can implement functions that have a very wide fan-in. Adjacent LUTs can be used to compute portions of the function in parallel; the cascade chain serially connects the intermediate values. The cascade chain can use a logical AND or logical OR (via De Morgan's inversion) to connect the outputs of adjacent LEs. Each additional LE provides four more inputs to the effective width of a function, with a delay as low as 0.7 ns per LE. Cascade chain logic can be created automatically by the Compiler during design processing, or manually by the designer during design entry.

Cascade chains longer than eight bits are implemented automatically by linking several LABs together. For easier routing, a long cascade chain skips every other LAB in a row. A cascade chain longer than one LAB skips either from even-numbered LAB to even-numbered LAB, or from odd-numbered LAB to odd-numbered LAB (e.g., the last LE of the first LAB in a row cascades to the first LE of the third LAB). The cascade chain does not cross the center of the row (e.g., in the EPF10K50 device, the cascade chain stops at the eighteenth LAB and a new one begins at the nineteenth LAB). This break is due to the EAB's placement in the middle of the row.

Figure 8 shows how the cascade function can connect adjacent LEs to form functions with a wide fan-in. These examples show functions of 4n variables implemented with n LEs. The LE delay is as low as 1.6 ns; the cascade chain delay is as low as 0.7 ns. With the cascade chain, 3.7 ns is needed to decode a 16-bit address.

Figure 8. Cascade Chain Operation



LE Operating Modes

The FLEX 10K LE can operate in the following four modes:

- Normal mode
- Arithmetic mode
- Up/down counter mode
- Clearable counter mode

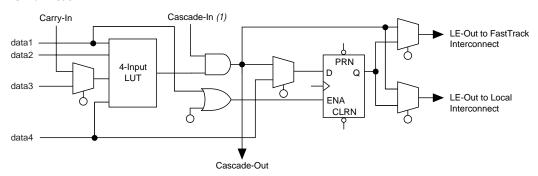
Each of these modes uses LE resources differently. In each mode, seven available inputs to the LE—the four data inputs from the LAB local interconnect, the feedback from the programmable register, and the carry-in and cascade-in from the previous LE—are directed to different destinations to implement the desired logic function. Three inputs to the LE provide clock, clear, and preset control for the register. The Altera software, in conjunction with parameterized functions such as LPM and DesignWare functions, automatically chooses the appropriate mode for common functions such as counters, adders, and multipliers. If required, the designer can also create special-purpose functions which use a specific LE operating mode for optimal performance.

The architecture provides a synchronous clock enable to the register in all four modes. The Altera software can set DATA1 to enable the register synchronously, providing easy implementation of fully synchronous designs.

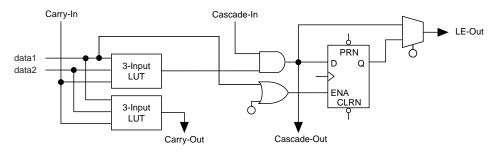
Figure 9 shows the LE operating modes.

Figure 9. FLEX 10K LE Operating Modes

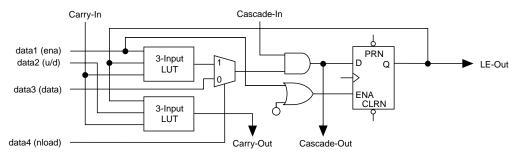
Normal Mode



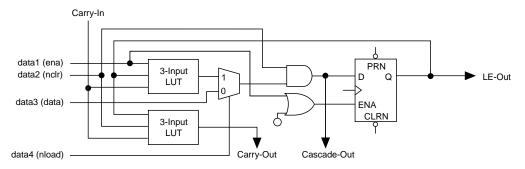
Arithmetic Mode



Up/Down Counter Mode



Clearable Counter Mode



Note:

(1) Packed registers cannot be used with the cascade chain.

Normal Mode

The normal mode is suitable for general logic applications and wide decoding functions that can take advantage of a cascade chain. In normal mode, four data inputs from the LAB local interconnect and the carry-in are inputs to a four-input LUT. The Compiler automatically selects the carry-in or the DATA3 signal as one of the inputs to the LUT. The LUT output can be combined with the cascade-in signal to form a cascade chain through the cascade-out signal. Either the register or the LUT can be used to drive both the local interconnect and the FastTrack Interconnect at the same time.

The LUT and the register in the LE can be used independently; this feature is known as register packing. To support register packing, the LE has two outputs; one drives the local interconnect and the other drives the FastTrack Interconnect. The DATA4 signal can drive the register directly, allowing the LUT to compute a function that is independent of the registered signal; a three-input function can be computed in the LUT, and a fourth independent signal can be registered. Alternatively, a four-input function can be generated, and one of the inputs to this function can be used to drive the register. The register in a packed LE can still use the clock enable, clear, and preset signals in the LE. In a packed LE, the register can drive the FastTrack Interconnect while the LUT drives the local interconnect, or vice versa.

Arithmetic Mode

The arithmetic mode offers 2 three-input LUTs that are ideal for implementing adders, accumulators, and comparators. One LUT computes a three-input function, and the other generates a carry output. As shown in Figure 9 on page 19, the first LUT uses the carry-in signal and two data inputs from the LAB local interconnect to generate a combinatorial or registered output. For example, in an adder, this output is the sum of three signals: a, b, and carry-in. The second LUT uses the same three signals to generate a carry-out signal, thereby creating a carry chain. The arithmetic mode also supports simultaneous use of the cascade chain.

Up/Down Counter Mode

The up/down counter mode offers counter enable, clock enable, synchronous up/down control, and data loading options. These control signals are generated by the data inputs from the LAB local interconnect, the carry-in signal, and output feedback from the programmable register. The Up/down counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. A 2-to-1 multiplexer provides synchronous loading. Data can also be loaded asynchronously with the clear and preset register control signals, without using the LUT resources.

Clearable Counter Mode

The clearable counter mode is similar to the up/down counter mode, but supports a synchronous clear instead of the up/down control. The clear function is substituted for the cascade-in signal in the up/down counter mode. Clearable counter mode uses 2 three-input LUTs: one generates the counter data, and the other generates the fast carry bit. Synchronous loading is provided by a 2-to-1 multiplexer. The output of this multiplexer is ANDed with a synchronous clear signal.

Internal Tri-State Emulation

Internal tri-state emulation provides internal tri-stating without the limitations of a physical tri-state bus. In a physical tri-state bus, the tri-state buffers' output enable (OE) signals select which signal drives the bus. However, if multiple OE signals are active, contending signals can be driven onto the bus. Conversely, if no OE signals are active, the bus will float. Internal tri-state emulation resolves contending tri-state buffers to a low value and floating buses to a high value, thereby eliminating these problems. The Altera software automatically implements tri-state bus functionality with a multiplexer.

Clear & Preset Logic Control

Logic for the programmable register's clear and preset functions is controlled by the DATA3, LABCTRL1, and LABCTRL2 inputs to the LE. The clear and preset control structure of the LE asynchronously loads signals into a register. Either LABCTRL1 or LABCTRL2 can control the asynchronous clear. Alternatively, the register can be set up so that LABCTRL1 implements an asynchronous load. The data to be loaded is driven to DATA3; when LABCTRL1 is asserted, DATA3 is loaded into the register.

During compilation, the Compiler automatically selects the best control signal implementation. Because the clear and preset functions are active-low, the Compiler automatically assigns a logic high to an unused clear or preset.

The clear and preset logic is implemented in one of the following six modes chosen during design entry:

- Asynchronous clear
- Asynchronous preset
- Asynchronous clear and preset
- Asynchronous load with clear
- Asynchronous load with preset
- Asynchronous load without clear or preset

In addition to the six clear and preset modes, FLEX 10K devices provide a chip-wide reset pin that can reset all registers in the device. Use of this feature is set during design entry. In any of the clear and preset modes, the chip-wide reset overrides all other signals. Registers with asynchronous presets may be preset when the chip-wide reset is asserted. Inversion can be used to implement the asynchronous preset. Figure 10 shows examples of how to enter a section of a design for the desired functionality.

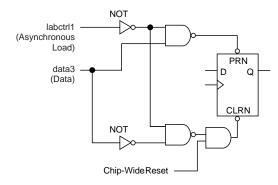
Figure 10. LE Clear & Preset Modes

Asynchronous Clear Asynchronous Preset Asynchronous Clear & Preset labctrl1 <u>vcc</u> Chip-Wide Reset PRN labctrl1 or PRN Q labctrl2 Q PRN CLRN CLRN labctrl1 or labctrl2 labctrl2 Chip-Wide Reset CLRN Chip-Wide Reset VCC

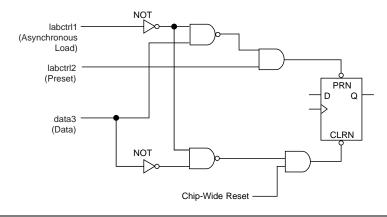
Asynchronous Load with Clear

labctrl1 (Asynchronous Load) data3 (Data) NOT PRN PRN CLRN Chip-Wide Reset

Asynchronous Load without Clear or Preset



Asynchronous Load with Preset



Asynchronous Clear

The flipflop can be cleared by either LABCTRL1 or LABCTRL2. In this mode, the preset signal is tied to V_{CC} to deactivate it.

Asynchronous Preset

An asynchronous preset is implemented as either an asynchronous load, or with an asynchronous clear. If DATA3 is tied to V_{CC} , asserting LABCTRL1 asynchronously loads a one into the register. Alternatively, the Altera software can provide preset control by using the clear and inverting the input and output of the register. Inversion control is available for the inputs to both LEs and IOEs. Therefore, if a register is preset by only one of the two LABCTRL signals, the DATA3 input is not needed and can be used for one of the LE operating modes.

Asynchronous Preset & Clear

When implementing asynchronous clear and preset, LABCTRL1 controls the preset and LABCTRL2 controls the clear. DATA3 is tied to V_{CC} , therefore, asserting LABCTRL1 asynchronously loads a one into the register, effectively presetting the register. Asserting LABCTRL2 clears the register.

Asynchronous Load with Clear

When implementing an asynchronous load in conjunction with the clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear. LABCTRL2 implements the clear by controlling the register clear; LABCTRL2 does not have to feed the preset circuits.

Asynchronous Load with Preset

When implementing an asynchronous load in conjunction with preset, the Altera software provides preset control by using the clear and inverting the input and output of the register. Asserting LABCTRL2 presets the register, while asserting LABCTRL1 loads the register. The Altera software inverts the signal that drives DATA3 to account for the inversion of the register's output.

Asynchronous Load without Preset or Clear

When implementing an asynchronous load without preset or clear, LABCTRL1 implements the asynchronous load of DATA3 by controlling the register preset and clear.

FastTrack Interconnect

In the FLEX 10K architecture, connections between LEs and device I/O pins are provided by the FastTrack Interconnect, which is a series of continuous horizontal and vertical routing channels that traverse the device. This global routing structure provides predictable performance, even in complex designs. In contrast, the segmented routing in FPGAs requires switch matrices to connect a variable number of routing paths, increasing the delays between logic resources and reducing performance.

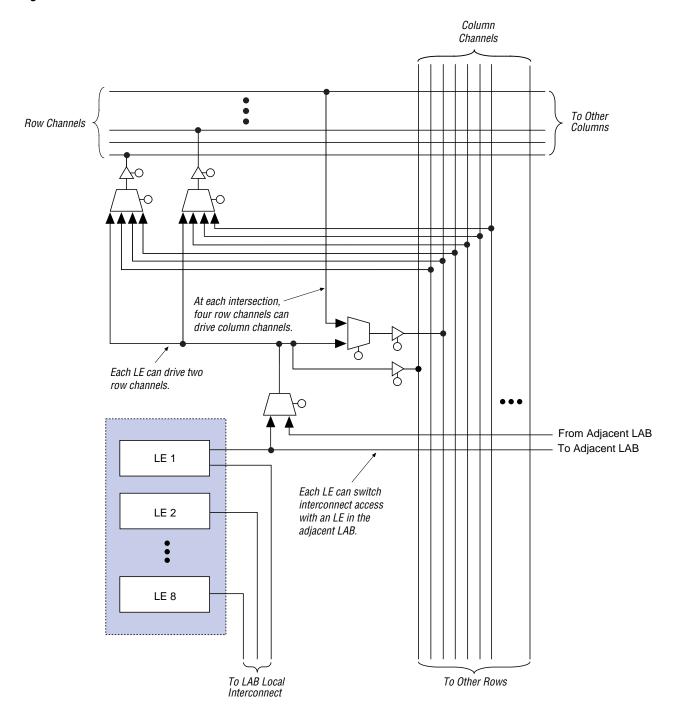
The FastTrack Interconnect consists of row and column interconnect channels that span the entire device. Each row of LABs is served by a dedicated row interconnect. The row interconnect can drive I/O pins and feed other LABs in the device. The column interconnect routes signals between rows and can drive I/O pins.

A row channel can be driven by an LE or by one of three column channels. These four signals feed dual 4-to-1 multiplexers that connect to two specific row channels. These multiplexers, which are connected to each LE, allow column channels to drive row channels even when all eight LEs in an LAB drive the row interconnect.

Each column of LABs is served by a dedicated column interconnect. The column interconnect can then drive I/O pins or another row's interconnect to route the signals to other LABs in the device. A signal from the column interconnect, which can be either the output of an LE or an input from an I/O pin, must be routed to the row interconnect before it can enter an LAB or EAB. Each row channel that is driven by an IOE or EAB can drive one specific column channel.

Access to row and column channels can be switched between LEs in adjacent pairs of LABs. For example, an LE in one LAB can drive the row and column channels normally driven by a particular LE in the adjacent LAB in the same row, and vice versa. This routing flexibility enables routing resources to be used more efficiently. See Figure 11.

Figure 11. LAB Connections to Row & Column Interconnect



For improved routing, the row interconnect is comprised of a combination of full-length and half-length channels. The full-length channels connect to all LABs in a row; the half-length channels connect to the LABs in half of the row. The EAB can be driven by the half-length channels in the left half of the row and by the full-length channels. The EAB drives out to the full-length channels. In addition to providing a predictable, row-wide interconnect, this architecture provides increased routing resources. Two neighboring LABs can be connected using a half-row channel, thereby saving the other half of the channel for the other half of the row.

Table 7 summarizes the FastTrack Interconnect resources available in each FLEX 10K device.

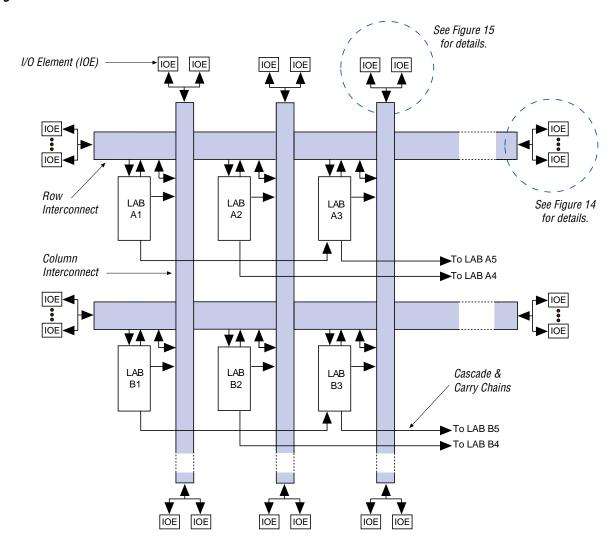
Table 7. FLEX 10K FastTrack Interconnect Resources					
Device	Rows	Channels per Row	Columns	Channels per Column	
EPF10K10 EPF10K10A	3	144	24	24	
EPF10K20	6	144	24	24	
EPF10K30 EPF10K30A	6	216	36	24	
EPF10K40	8	216	36	24	
EPF10K50 EPF10K50V	10	216	36	24	
EPF10K70	9	312	52	24	
EPF10K100 EPF10K100A	12	312	52	24	
EPF10K130V	16	312	52	32	
EPF10K250A	20	456	76	40	

In addition to general-purpose I/O pins, FLEX 10K devices have six dedicated input pins that provide low-skew signal distribution across the device. These six inputs can be used for global clock, clear, preset, and peripheral output enable and clock enable control signals. These signals are available as control signals for all LABs and IOEs in the device.

The dedicated inputs can also be used as general-purpose data inputs because they can feed the local interconnect of each LAB in the device. However, the use of dedicated inputs as data inputs can introduce additional delay into the control signal network.

Figure 12 shows the interconnection of adjacent LABs and EABs with row, column, and local interconnects, as well as the associated cascade and carry chains. Each LAB is labeled according to its location: a letter represents the row and a number represents the column. For example, LAB B3 is in row B, column 3.

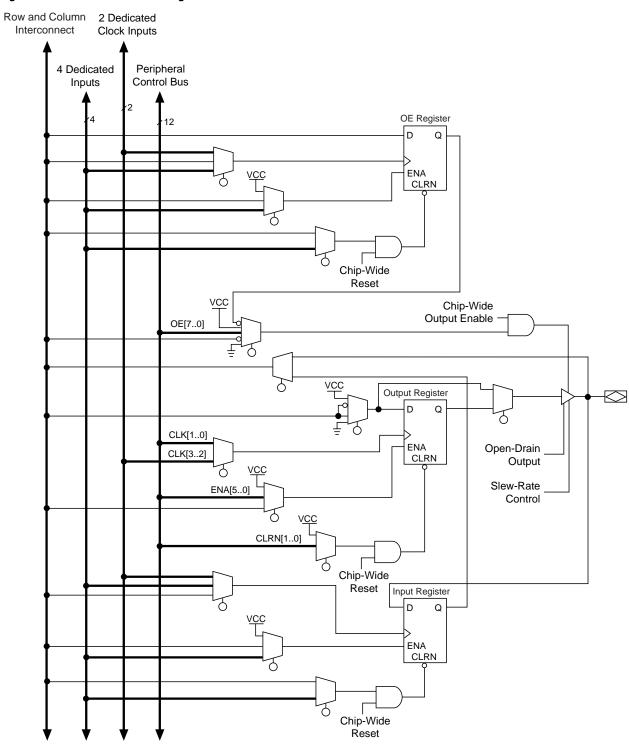
Figure 12. Interconnect Resources



I/O Element

An I/O element (IOE) contains a bidirectional I/O buffer and a register that can be used either as an input register for external data that requires a fast setup time, or as an output register for data that requires fast clock-to-output performance. In some cases, using an LE register for an input register will result in a faster setup time than using an IOE register. IOEs can be used as input, output, or bidirectional pins. For bidirectional registered I/O implementation, the output register should be in the IOE and, the data input and output enable register should be LE registers placed adjacent to the bidirectional pin. The Compiler uses the programmable inversion option to invert signals from the row and column interconnect automatically where appropriate. Figure 13 shows the bidirectional I/O registers.

Figure 13. Bidirectional I/O Registers



Each IOE selects the clock, clear, clock enable, and output enable controls from a network of I/O control signals called the peripheral control bus. The peripheral control bus uses high-speed drivers to minimize signal skew across devices; it provides up to 12 peripheral control signals that can be allocated as follows:

- Up to eight output enable signals
- Up to six clock enable signals
- Up to two clock signals
- Up to two clear signals

If more than six clock enable or eight output enable signals are required, each IOE on the device can be controlled by clock enable and output enable signals driven by specific LEs. In addition to the two clock signals available on the peripheral control bus, each IOE can use one of two dedicated clock pins. Each peripheral control signal can be driven by any of the dedicated input pins or the first LE of each LAB in a particular row. In addition, an LE in a different row can drive a column interconnect, which causes a row interconnect to drive the peripheral control signal. The chip-wide reset signal will reset all IOE registers, overriding any other control signals.

Tables 8 and 9 list the sources for each peripheral control signal, and the rows that can drive global signals. These tables also show how the output enable, clock enable, clock, and clear signals share 12 peripheral control signals.

able 8. EPF10K10, EPF10K20,	EPF10K30, EPF	10K40 & EPF10	0K50 Periphera	Bus Sources	
Peripheral Control Signal	EPF10K10 EPF10K10A	EPF10K20	EPF10K30 EPF10K30A	EPF10K40	EPF10K50 EPF10K50V
OE0	Row A	Row A	Row A	Row A	Row A
OE1	Row A	Row B	Row B	Row C	Row B
OE2	Row B	Row C	Row C	Row D	Row D
OE3	Row B	Row D	Row D	Row E	Row F
OE4	Row C	Row E	Row E	Row F	Row H
OE5	Row C	Row F	Row F	Row G	Row J
CLKENA0/CLK0/GLOBAL0	Row A	Row A	Row A	Row B	Row A
CLKENA1/OE6/GLOBAL1	Row A	Row B	Row B	Row C	Row C
CLKENA2/CLR0	Row B	Row C	Row C	Row D	Row E
CLKENA3/OE7/GLOBAL2	Row B	Row D	Row D	Row E	Row G
CLKENA4/CLR1	Row C	Row E	Row E	Row F	Row I
CLKENA5/CLK1/GLOBAL3	Row C	Row F	Row F	Row H	Row J

Table 9. EPF10K70, EPF10K100,	EPF10K130V & EF	PF10K250A Periphe	ral Bus Sources	
Peripheral Control Signal	EPF10K70	EPF10K100 EPF10K100A	EPF10K130V	EPF10K250A
OE0	Row A	Row A	Row C	Row E
OE1	Row B	Row C	Row E	Row G
OE2	Row D	Row E	Row G	Row I
OE3	Row I	Row L	Row N	Row P
OE4	Row G	Row I	Row K	Row M
OE5	Row H	Row K	Row M	Row O
CLKENA0/CLK0/GLOBAL0	Row E	Row F	Row H	Row J
CLKENA1/OE6/GLOBAL1	Row C	Row D	Row F	Row H
CLKENA2/CLR0	Row B	Row B	Row D	Row F
CLKENA3/OE7/GLOBAL2	Row F	Row H	Row J	Row L
CLKENA4/CLR1	Row H	Row J	Row L	Row N
CLKENA5/CLK1/GLOBAL3	Row E	Row G	Row I	Row K

Signals on the peripheral control bus can also drive the four global signals, referred to as GLOBALO through GLOBAL3 in Tables 8 and 9. The internally generated signal can drive the global signal, providing the same low-skew, low-delay characteristics for an internally generated signal as for a signal driven by an input. This feature is ideal for internally generated clear or clock signals with high fan-out. When a global signal is driven by internal logic, the dedicated input pin that drives that global signal cannot be used. The dedicated input pin should be driven to a known logic state (such as ground) and not be allowed to float.

When the chip-wide output enable pin is held low, it will tri-state all pins on the device. This option can be set in the Global Project Device Options menu. Additionally, the registers in the IOE can be reset by holding the chip-wide reset pin low.

Row-to-IOE Connections

When an IOE is used as an input signal, it can drive two separate row channels. The signal is accessible by all LEs within that row. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the row channels. Up to eight IOEs connect to each side of each row channel. See Figure 14.

Figure 14. FLEX 10K Row-to-IOE Connections

The values for m and n are provided in Table 10.

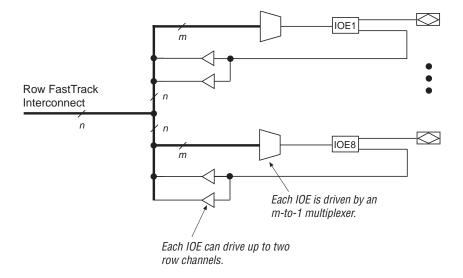


Table 10 lists the FLEX 10K row-to-IOE interconnect resources.

Table 10. FLEX 10K	Row-to-IOE Interconnect Re	sources
Device	Channels per Row (n)	Row Channels per Pin (m)
EPF10K10 EPF10K10A	144	18
EPF10K20	144	18
EPF10K30 EPF10K30A	216	27
EPF10K40	216	27
EPF10K50 EPF10K50V	216	27
EPF10K70	312	39
EPF10K100 EPF10K100A	312	39
EPF10K130V	312	39
EPF10K250A	456	57

Column-to-IOE Connections

When an IOE is used as an input, it can drive up to two separate column channels. When an IOE is used as an output, the signal is driven by a multiplexer that selects a signal from the column channels. Two IOEs connect to each side of the column channels. Each IOE can be driven by column channels via a multiplexer. The set of column channels that each IOE can access is different for each IOE. See Figure 15.

Figure 15. FLEX 10K Column-to-IOE Connections

The values for m and n are provided in Table 11.

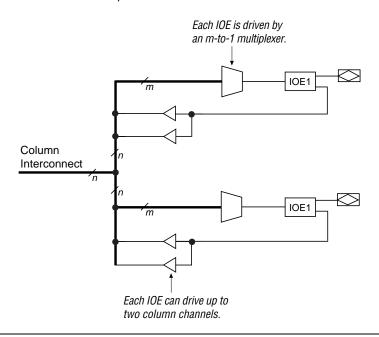


Table 11 lists the FLEX 10K column-to-IOE interconnect resources.

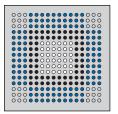
Table 11. FLEX 10K Column-to-IOE Interconnect Resources					
Device	Channels per Column (n)	Column Channel per Pin (<i>m</i>)			
EPF10K10	24	16			
EPF10K10A					
EPF10K20	24	16			
EPF10K30	24	16			
EPF10K30A					
EPF10K40	24	16			
EPF10K50	24	16			
EPF10K50V					
EPF10K70	24	16			
EPF10K100	24	16			
EPF10K100A					
EPF10K130V	32	24			
EPF10K250A	40	32			

SameFrame Pin-Outs

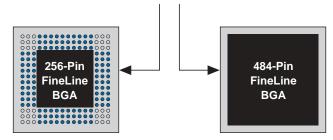
FLEX 10KE devices support the SameFrame pin-out feature for FineLine BGA packages. The SameFrame pin-out feature is the arrangement of balls on FineLine BGA packages such that the lower-ball-count packages form a subset of the higher-ball-count packages. SameFrame pin-outs provide the flexibility to migrate not only from device to device within the same package, but also from one package to another. A given printed circuit board (PCB) layout can support multiple device density/package combinations. For example, a single board layout can support a range of devices from an EPF10K10A device in a 256-pin FineLine BGA package to an EPF10K100A device in a 484-pin FineLine BGA package.

The Altera software provides support to design PCBs with SameFrame pin-out devices. Devices can be defined for present and future use. The Altera software generates pin-outs describing how to lay out a board to take advantage of this migration (see Figure 16).

Figure 16. SameFrame Pin-Out Example



Printed Circuit Board
Designed for 484-PinFineLine BGA Package



256-Pin FineLine BGA Package (Reduced I/O Count or Logic Requirements) 484-Pin FineLine BGA Package (Increased I/O Count or Logic Requirements)

ClockLock & ClockBoost Features

To support high-speed designs, selected FLEX 10K devices offer optional ClockLock and ClockBoost circuitry containing a phase-locked loop (PLL) that is used to increase design speed and reduce resource usage. The ClockLock circuitry uses a synchronizing PLL that reduces the clock delay and skew within a device. This reduction minimizes clock-to-output and setup times while maintaining zero hold times. The ClockBoost circuitry, which provides a clock multiplier, allows the designer to enhance device area efficiency by sharing resources within the device. The ClockBoost feature allows the designer to distribute a low-speed clock and multiply that clock on-device. Combined, the ClockLock and ClockBoost features provide significant improvements in system performance and bandwidth.

The ClockLock and ClockBoost features in FLEX 10K devices are enabled through the Altera software. External devices are not required to use these features. The output of the ClockLock and ClockBoost circuits is not available at any of the device pins.

The ClockLock and ClockBoost circuitry locks onto the rising edge of the incoming clock. The circuit output can only drive the clock inputs of registers; the generated clock cannot be gated or inverted.

The dedicated clock pin (GCLK1) supplies the clock to the ClockLock and ClockBoost circuitry. When the dedicated clock pin is driving the ClockLock or ClockBoost circuitry, it cannot drive elsewhere in the device.

In designs that require both a multiplied and non-multiplied clock, the clock trace on the board can be connected to GCLK1. With the Altera software, GCLK1 can feed both the ClockLock and ClockBoost circuitry in the FLEX 10K device. However, when both circuits are used, the other clock pin (GCLK0) cannot be used. Figure 17 shows a block diagram of how to enable both the ClockLock and ClockBoost circuits in the Altera software. The example shown is a schematic, but a similar approach applies for designs created in AHDL, VHDL, and Verilog HDL. When the ClockLock and ClockBoost circuits are used simultaneously, the input frequency parameter must be the same for both circuits. In Figure 17, the input frequency must meet the requirements specified when the ClockBoost multiplication factor is two.

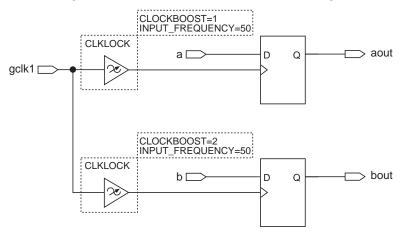


Figure 17. Enabling ClockLock & ClockBoost in the Same Design

To use both the ClockLock and ClockBoost circuits in the same design, designers must use Revision C EPF10K100GC503-3DX devices and MAX+PLUS II software versions 7.2 or higher. The die revision is indicated by the third digit of the nine-digit code on the top side of the device.

Output Configuration

This section discusses the peripheral component interconnect (PCI) pull-up clamping diode option, slew-rate control, open-drain output option, MultiVolt I/O interface, and power sequencing for FLEX 10K devices. The PCI pull-up clamping diode, slew-rate control, and open-drain output options are controlled pin-by-pin via Altera logic options. The MultiVolt I/O interface is controlled by connecting $V_{\rm CCIO}$ to a different voltage than $V_{\rm CCINT}.$ Its effect can be simulated in the Altera software via the **Global Project Device Options** dialog box (Assign menu).

PCI Clamping Diodes

The EPF10K10A and EPF10K30A devices have a pull-up clamping diode on every I/O, dedicated input, and dedicated clock pin. PCI clamping diodes clamp the transient overshoot caused by reflected waves to the $V_{\rm CCIO}$ value and are required for 3.3-V PCI compliance. Clamping diodes can also be used to limit overshoot in other systems.

Clamping diodes are controlled on a pin-by-pin basis via a logic option in the Altera software. When $V_{\rm CCIO}$ is 3.3 V, a pin that has the clamping diode turned on can be driven by a 2.5-V or 3.3-V signal, but not a 5.0-V signal. When $V_{\rm CCIO}$ is 2.5 V, a pin that has the clamping diode turned on can be driven by a 2.5-V signal, but not a 3.3-V or 5.0-V signal. However, a clamping diode can be turned on for a subset of pins, which allows devices to bridge between a 3.3-V PCI bus and a 5.0-V device.

Slew-Rate Control

The output buffer in each IOE has an adjustable output slew rate that can be configured for low-noise or high-speed performance. A slower slew rate reduces system noise and adds a maximum delay of approximately 2.9 ns. The fast slew rate should be used for speed-critical outputs in systems that are adequately protected against noise. Designers can specify the slew rate on a pin-by-pin basis during design entry or assign a default slew rate to all pins on a device-wide basis. The slow slew rate setting affects only the falling edge of the output.

Open-Drain Output Option

FLEX 10K devices provide an optional open-drain (electrically equivalent to an open-collector) output for each I/O pin. This open-drain output enables the device to provide system-level control signals (e.g., interrupt and write enable signals) that can be asserted by any of several devices. It can also provide an additional wired-OR plane. Additionally, the Altera software can convert tri-state buffers with grounded data inputs to open-drain pins automatically.

Open-drain output pins on FLEX 10K devices (with a pull-up resistor to the 5.0-V supply) can drive 5.0-V CMOS input pins that require a $V_{\rm IH}$ of 3.5 V. When the open-drain pin is active, it will drive low. When the pin is inactive, the trace will be pulled up to 5.0 V by the resistor. The open-drain pin will only drive low or tri-state; it will never drive high. The rise time is dependent on the value of the pull-up resistor and load impedance. The $I_{\rm OL}$ current specification should be considered when selecting a pull-up resistor.

Output pins on 5.0-V FLEX 10K devices with $V_{\rm CCIO} = 3.3$ V or 5.0 V (with a pull-up resistor to the 5.0-V supply) can also drive 5.0-V CMOS input pins. In this case, the pull-up transistor will turn off when the pin voltage exceeds 3.3 V. Therefore, the pin does not have to be open-drain.

MultiVolt I/O Interface

The FLEX 10K device architecture supports the MultiVolt I/O interface feature, which allows FLEX 10K devices to interface with systems of differing supply voltages. These devices have one set of V_{CC} pins for internal operation and input buffers (VCCINT) and another set for I/O output drivers (VCCIO).

Table 12 describes the FLEX 10K device supply voltages and MultiVolt I/O support levels.

Devices	ces Supply Voltage (V) MultiVolt I/O S		Supply Voltage (V)		port Levels (V)
	V _{CCINT}	V _{CCIO}	Input	Output	
FLEX 10K (1)	5.0	5.0	3.3 or 5.0	5.0	
	5.0	3.3	3.3 or 5.0	3.3 or 5.0	
EPF10K50V (1)	3.3	3.3	3.3 or 5.0	3.3 or 5.0	
EPF10K130V	3.3	3.3	3.3 or 5.0	3.3 or 5.0	
FLEX 10KA (1)	3.3	3.3	2.5, 3.3, or 5.0	3.3 or 5.0	
	3.3	2.5	2.5, 3.3, or 5.0	2.5	

Note

Power Sequencing & Hot-Socketing

Because FLEX 10K devices can be used in a multi-voltage environment, they have been designed specifically to tolerate any possible power-up sequence. The V_{CCIO} and V_{CCINT} power supplies can be powered in any order.

Signals can be driven into FLEX 10KA devices before and during power up without damaging the device. Additionally, FLEX 10KA devices do not drive out during power up. Once operating conditions are reached, FLEX 10KA devices operate as specified by the user.

IEEE Std. 1149.1 (JTAG) Boundary-Scan Support All FLEX 10K devices provide JTAG BST circuitry that complies with the IEEE Std. 1149.1-1990 specification. All FLEX 10K devices can also be configured using the JTAG pins through the BitBlaster serial download cable, or ByteBlasterMV parallel port download cable, or via hardware that uses the JamTM programming and test language. JTAG BST can be performed before or after configuration, but not during configuration. FLEX 10K devices support the JTAG instructions shown in Table 13.

^{(1) 240-}pin QFP packages do not support the MultiVolt I/O features, so they do not have separate V_{CCIO} pins.

JTAG Instruction	Description
SAMPLE/PRELOAD	Allows a snapshot of signals at the device pins to be captured and examined during normal device operation, and permits an initial data pattern output at the device pins.
EXTEST	Allows the external circuitry and board-level interconnections to be tested by forcing a test pattern at the output pins and capturing test results at the input pins.
BYPASS	Places the 1-bit bypass register between the TDI and TDO pins, which allows the BST data to pass synchronously through a selected device to adjacent devices during normal device operation.
USERCODE	Selects the user electronic signature (USERCODE) register and places it between the TDI and TDO pins, allowing the USERCODE to be serially shifted out of TDO.
IDCODE	Selects the IDCODE register and places it between TDI and TDO, allowing the IDCODE to be serially shifted out of TDO.
ICR Instructions	These instructions are used when configuring a FLEX 10K device via JTAG ports with a BitBlaster, or ByteBlasterMV or MasterBlaster download cable, or using a Jam File (.jam) or Jam Byte-Code File (.jbc) via an embedded processor.

The instruction register length of FLEX 10K devices is 10 bits. The USERCODE register length in FLEX 10K devices is 32 bits; 7 bits are determined by the user, and 25 bits are predetermined. Tables 14 and 15 show the boundary-scan register length and device IDCODE information for FLEX 10K devices.

Table 14. FLEX 10K Boundary-Scan Register Length		
Device	Boundary-Scan Register Length	
EPF10K10, EPF10K10A	480	
EPF10K20	624	
EPF10K30, EPF10K30A	768	
EPF10K40	864	
EPF10K50, EPF10K50V	960	
EPF10K70	1,104	
EPF10K100, EPF10K100A	1,248	
EPF10K130V	1,440	
EPF10K250A	1,440	

Table 15. 32-Bit FLEX 10K Device IDCODE Note (1)							
Device	IDCODE (32 Bits)						
	Version (4 Bits)	Part Number (16 Bits)	Manufacturer's Identity (11 Bits)	1 (1 Bit) (2)			
EPF10K10, EPF10K10A	0000	0001 0000 0001 0000	00001101110	1			
EPF10K20	0000	0001 0000 0010 0000	00001101110	1			
EPF10K30, EPF10K30A	0000	0001 0000 0011 0000	00001101110	1			
EPF10K40	0000	0001 0000 0100 0000	00001101110	1			
EPF10K50, EPF10K50V	0000	0001 0000 0101 0000	00001101110	1			
EPF10K70	0000	0001 0000 0111 0000	00001101110	1			
EPF10K100, EPF10K100A	0000	0000 0001 0000 0000	00001101110	1			
EPF10K130V	0000	0000 0001 0011 0000	00001101110	1			
EPF10K250A	0000	0000 0010 0101 0000	00001101110	1			

Notes:

- (1) The most significant bit (MSB) is on the left.
- (2) The least significant bit (LSB) for all JTAG IDCODEs is 1.

FLEX 10K devices include weak pull-ups on JTAG pins.



For more information, see the following documents:

- Application Note 39 (IEEE 1149.1 (JTAG) Boundary-Scan Testing in Altera Devices)
- BitBlaster Serial Download Cable Data Sheet
- ByteBlasterMV Parallel Port Download Cable Data Sheet
- Jam Programming & Test Language Specification

Figure 18. JTAG Waveforms

Figure 18 shows the timing requirements for the JTAG signals.

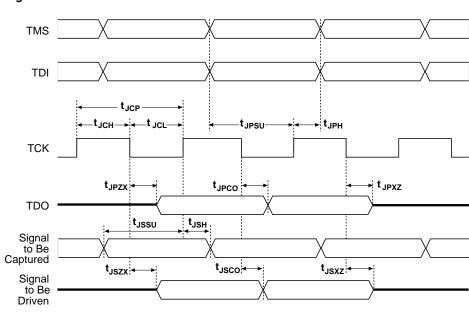


Table 16 shows the timing parameters and values for FLEX 10K devices.

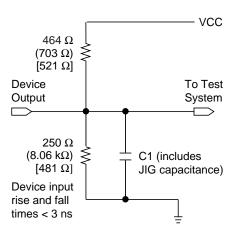
Table 16. JTAG Timing Parameters & Values					
Symbol	Parameter	Min	Max	Unit	
t _{JCP}	TCK clock period	100		ns	
t _{JCH}	TCK clock high time	50		ns	
t _{JCL}	TCK clock low time	50		ns	
t _{JPSU}	JTAG port setup time	20		ns	
t _{JPH}	JTAG port hold time	45		ns	
t _{JPCO}	JTAG port clock to output		25	ns	
t _{JPZX}	JTAG port high impedance to valid output		25	ns	
t _{JPXZ}	JTAG port valid output to high impedance		25	ns	
t _{JSSU}	Capture register setup time	20		ns	
t _{JSH}	Capture register hold time	45		ns	
t _{JSCO}	Update register clock to output		35	ns	
t _{JSZX}	Update register high-impedance to valid output		35	ns	
t _{JSXZ}	Update register valid output to high impedance		35	ns	

Generic Testing

Each FLEX 10K device is functionally tested. Complete testing of each configurable SRAM bit and all logic functionality ensures 100% yield. AC test measurements for FLEX 10K devices are made under conditions equivalent to those shown in Figure 19. Multiple test patterns can be used to configure devices during all stages of the production flow.

Figure 19. FLEX 10K AC Test Conditions

Power supply transients can affect AC measurements. Simultaneous transitions of multiple outputs should be avoided for accurate measurement. Threshold tests must not be performed under AC conditions. Large-amplitude, fast-ground-current transients normally occur as the device outputs discharge the load capacitances. When these transients flow through the parasitic inductance between the device ground pin and the test system ground, significant reductions in observable noise immunity can result. Numbers without parentheses are for 5.0-V devices or outputs. Numbers in parentheses are for 3.3-V devices or outputs. Numbers in brackets are for 2.5-V devices or outputs.



Operating Conditions

Tables 17 through 21 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 5.0-V FLEX 10K devices.

Symbol	Parameter	Conditions	Min	Max	Unit
Oyillbui	1 diameter	Conditions	141111	IVIAA	Oiiit
V_{CC}	Supply voltage	With respect to ground (2)	-2.0	7.0	V
V_{I}	DC input voltage		-2.0	7.0	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	° C
T _{AMB}	Ambient temperature	Under bias	-65	135	° C
TJ	Junction temperature	Ceramic packages, under bias		150	° C
		PQFP, TQFP, RQFP, and BGA		135	° C
		packages, under bias			

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
V _{CCIO}	Supply voltage for output buffers, 5.0-V operation	(3), (4)	4.75 (4.50)	5.25 (5.50)	V
	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
VI	Input voltage		-0.5	V _{CCINT} + 0.5	V
Vo	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	For commercial use	0	70	°C
		For industrial use	-40	85	°C
T _J	Operating temperature	For commercial use	0	85	°C
		For industrial use	-40	100	°C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Table 1	9. FLEX 10K 5.0-V Devi	ce DC Operating Conditions No	tes (5), (6)			
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IH}	High-level input voltage		2.0		V _{CCINT} + 0.5	V
V _{IL}	Low-level input voltage		-0.5		8.0	V
V _{OH}	5.0-V high-level TTL output voltage	$I_{OH} = -4 \text{ mA DC}, V_{CCIO} = 4.75 \text{ V}$ (7)	2.4			V
	3.3-V high-level TTL output voltage	$I_{OH} = -4 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V}$ (7)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V}$ (7)	V _{CCIO} - 0.2			V
V _{OL}	5.0-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 4.75 V (8)			0.45	V
	3.3-V low-level TTL output voltage	I_{OL} = 12 mA DC, V_{CCIO} = 3.00 V (8)			0.45	V
	3.3-V low-level CMOS output voltage	$I_{OL} = 0.1 \text{ mA DC}, V_{CCIO} = 3.00 \text{ V}$ (8)			0.2	V
I _I	Input pin leakage current	$V_I = V_{CC}$ or ground (9)	-10		10	μΑ
I _{OZ}	Tri-stated I/O pin leakage current	$V_O = V_{CC}$ or ground (9)	-40		40	μΑ
I _{CC0}	V _{CC} supply current (standby)	V _I = ground, no load		0.5	10	mA

Table 2	Table 20. 5.0-V Device Capacitance of EPF10K10, EPF10K20 & EPF10K30 Devices Note (10)					
Symbol	Parameter	Conditions	Min	Max	Unit	
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		8	pF	
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		12	pF	
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		8	pF	

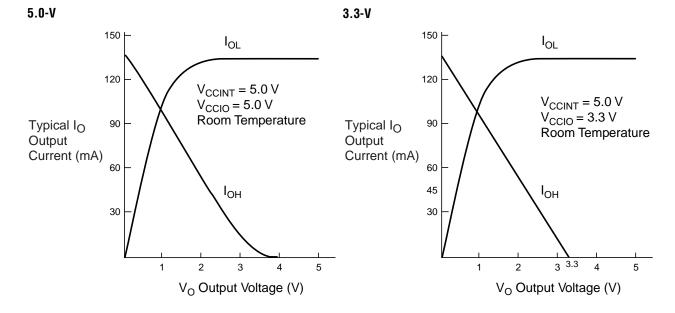
Table 21. 5.0-V Device Capacitance of EPF10K40, EPF10K50, EPF10K70 & EPF10K100 DevicesNote (10)						
Symbol	Parameter	Conditions	Min	Max	Unit	
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		10	pF	
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		15	pF	
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		10	pF	

Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC input voltage is –0.5 V. During transitions, the inputs may undershoot to –2.0 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V_{CC} rise time is 100 ms. V_{CC} must rise monotonically.
- (5) Typical values are for $T_A = 25^{\circ}$ C and $V_{CC} = 5.0$ V.
- (6) These values are specified under the Recommended Operation Condition shown in Table 18 on page 45.
- (7) The I_{OH} parameter refers to high-level TTL or CMOS output current.
- (8) The I_{OL} parameter refers to low-level TTL or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (9) This value is specified for normal device operation. The value may vary during power-up.
- (10) Capacitance is sample-tested only.

Figure 20 shows the typical output drive characteristics of FLEX 10K devices with 5.0-V and 3.3-V $V_{\rm CCIO}$. The output driver is compliant with the 5.0-V *PCI Local Bus Specification, Revision 2.2* (for 5.0-V $V_{\rm CCIO}$).

Figure 20. Output Drive Characteristics of FLEX 10K Devices



Tables 22 through 25 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for EPF10K50V and EPF10K130V devices.

Table 2	2. EPF10K50V & EPF10K130V	Note (1)			
Symbol	Parameter	Conditions	Min	Max	Unit
V _{CC}	Supply voltage	With respect to ground (2)	-0.5	4.6	V
V _I	DC input voltage		-2.0	5.75	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	° C
T _{AMB}	Ambient temperature	Under bias	-65	135	° C
TJ	Junction temperature	Ceramic packages, under bias		150	° C
		RQFP and BGA packages, under bias		135	° C

Table 23. EPF10K50V & EPF10K130V Device Recommended Operating Conditions						
Symbol	Parameter	Conditions	Min	Max	Unit	
V _{CCINT}	Supply voltage for internal logic and input buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V	
V _{CCIO}	Supply voltage for output buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V	
V _I	Input voltage	(5)	-0.5	5.75	V	
Vo	Output voltage		0	V _{CCIO}	V	
T _A	Ambient temperature	For commercial use	0	70	° C	
		For industrial use	-40	85	° C	
T _J	Operating temperature	For commercial use	0	85	° C	
		For industrial use	-40	100	° C	
t _R	Input rise time			40	ns	
t _F	Input fall time			40	ns	

Table 2	24. EPF10K50V & EPF10K130V L	Device DC Operating Condi	tions Notes	s (6), (7))	
Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IH}	High-level input voltage		2.0		5.75	V
V _{IL}	Low-level input voltage		-0.5		0.8	V
V _{OH}	3.3-V high-level TTL output voltage	I _{OH} = -8 mA DC (8)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC } (8)$	V _{CCIO} - 0.2			V
V _{OL}	3.3-V low-level TTL output voltage	I _{OL} = 8 mA DC (9)			0.45	V
	3.3-V low-level CMOS output voltage	I _{OL} = 0.1 mA DC (9)			0.2	V
I_{\parallel}	Input pin leakage current	$V_1 = 5.3 \text{ V to } -0.3 \text{ V } (10)$	-10		10	μА
I _{OZ}	Tri-stated I/O pin leakage current	$V_O = 5.3 \text{ V to } -0.3 \text{ V } (10)$	-10		10	μА
I _{CC0}	V _{CC} supply current (standby)	V _I = ground, no load		0.3	10	mA
		V_I = ground, no load (11)		10		mA

Table 2	25. EPF10K50V & EPF10K130V D	Device Capacitance (12)			
Symbol	Parameter	Conditions	Min	Max	Unit
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		10	pF
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		15	pF
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC input voltage is –0.5 V. During transitions, the inputs may undershoot to –2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V_{CC} rise time is 100 ms. V_{CC} must rise monotonically.
- (5) EPF10K50V and EPF10K130V device inputs may be driven before V_{CCINT} and V_{CCIO} are powered.
- (6) Typical values are for $T_A = 25^{\circ}$ C and $V_{CC} = 3.3$ V.
- (7) These values are specified under the EPF10K50V and EPF10K130V device Recommended Operating Conditions in Table 23 on page 48.
- (8) The I_{OH} parameter refers to high-level TTL or CMOS output current.
- (9) The I_{OL} parameter refers to low-level TTL or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (10) This value is specified for normal device operation. The value may vary during power-up.
- (11) This parameter applies to -1 speed grade EPF10K50V devices, -2 speed grade EPF10K50V industrial temperature devices, and -2 speed grade EPF10K130V devices.
- (12) Capacitance is sample-tested only.

Figure 21 shows the typical output drive characteristics of EPF10K50V and EPF10K130V devices.

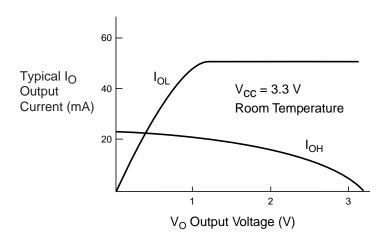


Figure 21. Output Drive Characteristics of EPF10K50V & EPF10K130V Devices

Tables 26 through 31 provide information on absolute maximum ratings, recommended operating conditions, DC operating conditions, and capacitance for 3.3-V FLEX 10K devices.

Table 26. FLEX 10KA 3.3-V Device Absolute Maximum RatingsNote (1)					
Symbol	Parameter	Conditions	Min	Max	Unit
V _{CC}	Supply voltage	With respect to ground (2)	-0.5	4.6	V
V _I	DC input voltage		-2.0	5.75	V
I _{OUT}	DC output current, per pin		-25	25	mA
T _{STG}	Storage temperature	No bias	-65	150	°C
T _{AMB}	Ambient temperature	Under bias	-65	135	°C
TJ	Junction temperature	Ceramic packages, under bias		150	°C
		PQFP, TQFP, RQFP, and BGA packages, under bias		135	°C

Symbol	Parameter	Conditions	Min	Max	Unit
V _{CCINT}	Supply voltage for internal logic and input buffers	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
V _{CCIO}	Supply voltage for output buffers, 3.3-V operation	(3), (4)	3.00 (3.00)	3.60 (3.60)	V
	Supply voltage for output buffers, 2.5-V operation	(3), (4)	2.30 (2.30)	2.70 (2.70)	V
VI	Input voltage	(5)	-0.5	5.75	V
Vo	Output voltage		0	V _{CCIO}	V
T _A	Ambient temperature	For commercial use	0	70	°C
		For industrial use	-40	85	°C
T _J	Operating temperature	For commercial use	0	85	°C
		For industrial use	-40	100	°C
t _R	Input rise time			40	ns
t _F	Input fall time			40	ns

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{IH}	High-level input voltage		1.7 or 0.5 × V _{CCINT} , whichever is lower		5.75	V
V _{IL}	Low-level input voltage		-0.5		$0.3 \times V_{CCINT}$	V
V _{OH}	3.3-V high-level TTL output voltage	I _{OH} = -11 mA DC, V _{CCIO} = 3.00 V (8)	2.4			V
	3.3-V high-level CMOS output voltage	$I_{OH} = -0.1 \text{ mA DC},$ $V_{CCIO} = 3.00 \text{ V (8)}$	V _{CCIO} - 0.2			V
	3.3-V high-level PCI output voltage	$I_{OH} = -0.5 \text{ mA DC},$ $V_{CCIO} = 3.00 \text{ to } 3.60 \text{ V } (8)$	0.9 × V _{CCIO}			V
	2.5-V high-level output voltage	$I_{OH} = -0.1 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (8)$	2.1			V
		$I_{OH} = -1 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (8)$	2.0			V
		$I_{OH} = -2 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (8)$	1.7			V
V _{OL}	3.3-V low-level TTL output voltage	I _{OL} = 9 mA DC, V _{CCIO} = 3.00 V (9)			0.45	V
	3.3-V low-level CMOS output voltage	I _{OL} = 0.1 mA DC, V _{CCIO} = 3.00 V (9)			0.2	V
	3.3-V low-level PCI output voltage	I _{OL} = 1.5 mA DC, V _{CCIO} = 3.00 to 3.60 V (9)			0.1 × V _{CCIO}	V
	2.5-V low-level output voltage	$I_{OL} = 0.1 \text{ mA DC},$ $V_{CCIO} = 2.30 \text{ V } (9)$			0.2	V
		I _{OL} = 1 mA DC, V _{CCIO} = 2.30 V (9)			0.4	V
		I _{OL} = 2 mA DC, V _{CCIO} = 2.30 V (9)			0.7	V
I _I	Input pin leakage current	V _I = 5.3 V to -0.3 V (10)	-10		10	μΑ
I _{OZ}	Tri-stated I/O pin leakage current	$V_0 = 5.3 \text{ V to } -0.3 \text{ V } (10)$	-10		10	μΑ
I _{CC0}	V _{CC} supply current (standby)	V _I = ground, no load		0.3	10	mA
		V _I = ground, no load (11)		10		mA

Table 2	9. 3.3-V Device Capacitance of	EPF10K10A & EPF10K30A Devices	Note (12)		
Symbol	Parameter	Conditions	Min	Max	Unit
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		8	pF
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		12	pF
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		8	pF

Table 3	O. 3.3-V Device Capacitance of	EPF10K100A Devices Note (12)			
Symbol	Parameter	Conditions	Min	Max	Unit
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		10	pF
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		15	pF
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		10	pF

Table 31. 3.3-V Device Capacitance of EPF10K250A Devices Note (12)					
Symbol	Parameter	Conditions	Min	Max	Unit
C _{IN}	Input capacitance	V _{IN} = 0 V, f = 1.0 MHz		10	pF
C _{INCLK}	Input capacitance on dedicated clock pin	V _{IN} = 0 V, f = 1.0 MHz		15	pF
C _{OUT}	Output capacitance	V _{OUT} = 0 V, f = 1.0 MHz		10	pF

Notes to tables:

- (1) See the Operating Requirements for Altera Devices Data Sheet.
- (2) Minimum DC voltage input is -0.5 V. During transitions, the inputs may undershoot to -2.0 V or overshoot to 5.75 V for input currents less than 100 mA and periods shorter than 20 ns.
- (3) Numbers in parentheses are for industrial-temperature-range devices.
- (4) Maximum V_{CC} rise time is 100 ms, and V_{CC} must rise monotonically.
- (5) FLEX 10KA device inputs may be driven before V_{CCINT} and V_{CCIO} are powered.
- (6) Typical values are for $T_A = 25^{\circ}$ C and $V_{CC} = 3.3$ V.
- (7) These values are specified under the Recommended Operating Conditions shown in Table 27 on page 51.
- (8) The I_{OH} parameter refers to high-level TTL, PCI, or CMOS output current.
- (9) The I_{OL} parameter refers to low-level TTL, PCI, or CMOS output current. This parameter applies to open-drain pins as well as output pins.
- (10) This value is specified for normal device operation. The value may vary during power-up.
- (11) This parameter applies to all -1 speed grade commercial temperature devices and all -2 speed grade industrial-temperature devices.
- (12) Capacitance is sample-tested only.

Figure 22 shows the typical output drive characteristics of EPF10K10A, EPF10K30A, EPF10K100A, and EPF10K250A devices with 3.3-V and 2.5-V $V_{\rm CCIO}$. The output driver is compliant with the 3.3-V *PCI Local Bus Specification, Revision 2.2* (with 3.3-V $V_{\rm CCIO}$). Moreover, device analysis shows that the EPF10K10A, EPF10K30A, and EPF 10K100A devices can drive a 5.0-V PCI bus with eight or fewer loads.

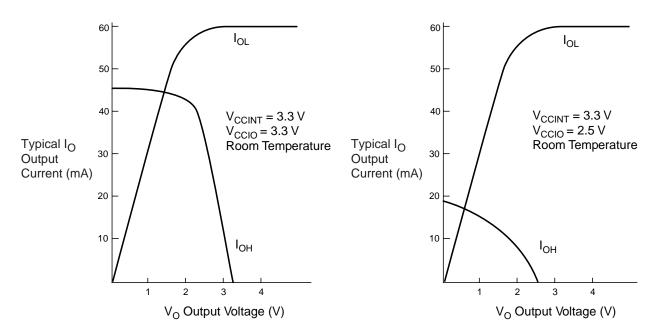


Figure 22. Output Drive Characteristics for EPF10K10A, EPF10K30A & EPF10K100A Devices

Figure 23 shows the typical output drive characteristics of the EPF10K250A device with 3.3-V and 2.5-V $V_{\rm CCIO}$.

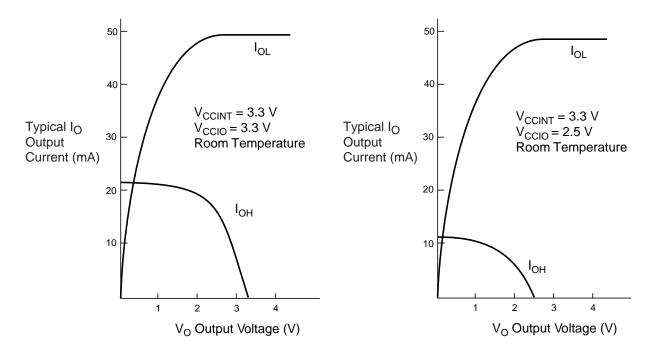


Figure 23. Output Drive Characteristics for EPF10K250A Device

Timing Model

The continuous, high-performance FastTrack Interconnect routing resources ensure predictable performance and accurate simulation and timing analysis. This predictable performance contrasts with that of FPGAs, which use a segmented connection scheme and therefore have unpredictable performance.

Device performance can be estimated by following the signal path from a source, through the interconnect, to the destination. For example, the registered performance between two LEs on the same row can be calculated by adding the following parameters:

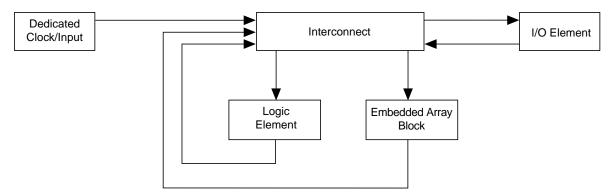
- LE register clock-to-output delay (t_{CO})
- Interconnect delay ($t_{SAMEROW}$)
- LE look-up table delay (t_{LUT})
- LE register setup time (t_{SU})

The routing delay depends on the placement of the source and destination LEs. A more complex registered path may involve multiple combinatorial LEs between the source and destination LEs.

Timing simulation and delay prediction are available with the MAX+PLUS II Simulator and Timing Analyzer, or with industry-standard EDA tools. The Simulator offers both pre-synthesis functional simulation to evaluate logic design accuracy and post-synthesis timing simulation with 0.1-ns resolution. The Timing Analyzer provides point-to-point timing delay information, setup and hold time analysis, and device-wide performance analysis.

Figure 24 shows the overall timing model, which maps the possible paths to and from the various elements of the FLEX 10K device.

Figure 24. FLEX 10K Device Timing Model



Figures 25 through 27 show the delays that correspond to various paths and functions within the LE, IOE, and EAB timing models.

Figure 25. FLEX 10K Device LE Timing Model

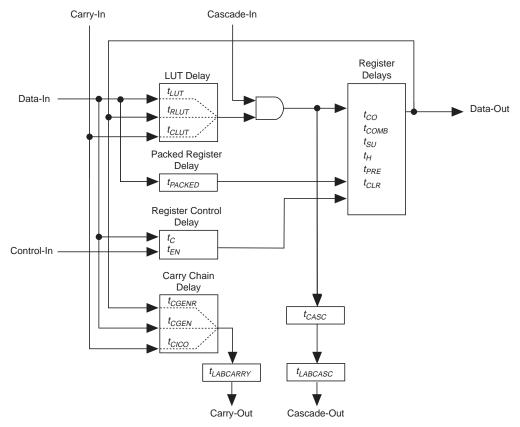


Figure 26. FLEX 10K Device IOE Timing Model

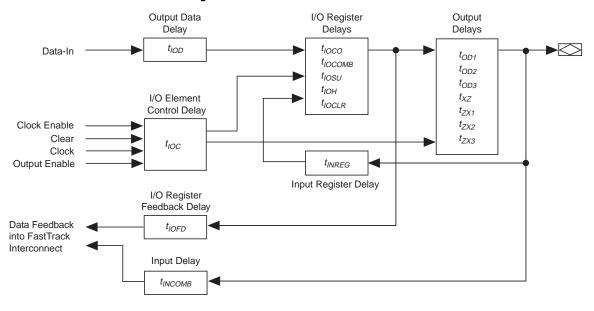
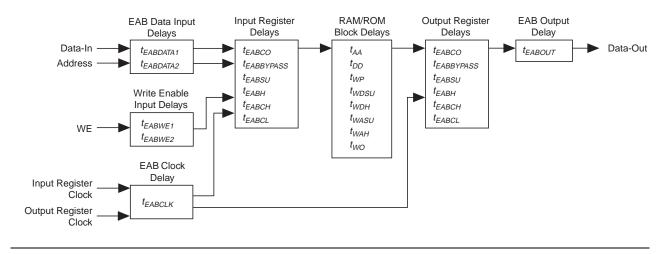


Figure 27. FLEX 10K Device EAB Timing Model



Figures 28 shows the timing model for bidirectional I/O pin timing.

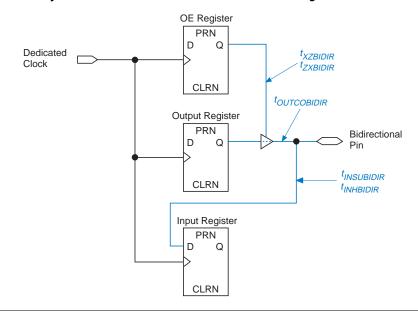


Figure 28. Synchronous Bidirectional Pin External Timing Model

Tables 32 through 36 describe the FLEX 10K device internal timing parameters. These internal timing parameters are expressed as worst-case values. Using hand calculations, these parameters can be used to estimate design performance. However, before committing designs to silicon, actual worst-case performance should be modeled using timing simulation and analysis. Tables 37 through 38 describe FLEX 10K external timing parameters.

Symbol	Parameter	Conditions
t _{LUT}	LUT delay for data-in	
t _{CLUT}	LUT delay for carry-in	
t _{RLUT}	LUT delay for LE register feedback	
t _{PACKED}	Data-in to packed register delay	
t _{EN}	LE register enable delay	
t _{CICO}	Carry-in to carry-out delay	
t _{CGEN}	Data-in to carry-out delay	
t _{CGENR}	LE register feedback to carry-out delay	
t _{CASC}	Cascade-in to cascade-out delay	
$t_{\mathbb{C}}$	LE register control signal delay	
$t_{\rm CO}$	LE register clock-to-output delay	
t _{COMB}	Combinatorial delay	

Table 32. LE	Table 32. LE Timing Microparameters (Part 2 of 2) Note (1)			
Symbol	Parameter	Conditions		
t _{SU}	LE register setup time for data and enable signals before clock; LE register recovery time after asynchronous clear, preset, or load			
t_H	LE register hold time for data and enable signals after clock			
t _{PRE}	LE register preset delay			
t _{CLR}	LE register clear delay			
t _{CH}	Minimum clock high time from clock pin			
t_{CL}	Minimum clock low time from clock pin			

Table 33. IOE	Table 33. IOE Timing Microparameters Note (1)		
Symbol	Parameter	Conditions	
t_{IOD}	IOE data delay		
t_{IOC}	IOE register control signal delay		
t _{IOCO}	IOE register clock-to-output delay		
t _{IOCOMB}	IOE combinatorial delay		
t _{IOSU}	IOE register setup time for data and enable signals before clock; IOE register recovery time after asynchronous clear		
t _{IOH}	IOE register hold time for data and enable signals after clock		
t _{IOCLR}	IOE register clear time		
t _{OD1}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = V _{CCINT}	C1 = 35 pF (2)	
t _{OD2}	Output buffer and pad delay, slow slew rate = off, V _{CCIO} = low voltage	C1 = 35 pF (3)	
t _{OD3}	Output buffer and pad delay, slow slew rate = on	C1 = 35 pF (4)	
t_{XZ}	IOE output buffer disable delay		
t_{ZX1}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = V _{CCINT}	C1 = 35 pF (2)	
t_{ZX2}	IOE output buffer enable delay, slow slew rate = off, V _{CCIO} = low voltage	C1 = 35 pF (3)	
t_{ZX3}	IOE output buffer enable delay, slow slew rate = on	C1 = 35 pF (4)	
t _{INREG}	IOE input pad and buffer to IOE register delay		
t _{IOFD}	IOE register feedback delay		
t _{INCOMB}	IOE input pad and buffer to FastTrack Interconnect delay		

Symbol	Parameter	Conditions
t _{EABDATA1}	Data or address delay to EAB for combinatorial input	
t _{EABDATA2}	Data or address delay to EAB for registered input	
t _{EABWE1}	Write enable delay to EAB for combinatorial input	
t _{EABWE2}	Write enable delay to EAB for registered input	
t _{EABCLK}	EAB register clock delay	
t _{EABCO}	EAB register clock-to-output delay	
t _{EABBYPASS}	Bypass register delay	
t _{EABSU}	EAB register setup time before clock	
t _{EABH}	EAB register hold time after clock	
t_{AA}	Address access delay	
t_{WP}	Write pulse width	
t _{WDSU}	Data setup time before falling edge of write pulse	(5)
t_{WDH}	Data hold time after falling edge of write pulse	(5)
t _{WASU}	Address setup time before rising edge of write pulse	(5)
t _{WAH}	Address hold time after falling edge of write pulse	(5)
t_{WO}	Write enable to data output valid delay	
t_{DD}	Data-in to data-out valid delay	
t _{EABOUT}	Data-out delay	
t _{EABCH}	Clock high time	
t _{EABCL}	Clock low time	

Table 35. EAL	B Timing Macroparameters Notes (1), (6)	
Symbol	Parameter	Conditions
t _{EABAA}	EAB address access delay	
t _{EABRCCOMB}	EAB asynchronous read cycle time	
t _{EABRCREG}	EAB synchronous read cycle time	
t_{EABWP}	EAB write pulse width	
t _{EABWCCOMB}	EAB asynchronous write cycle time	
t _{EABWCREG}	EAB synchronous write cycle time	
t_{EABDD}	EAB data-in to data-out valid delay	
$t_{\it EABDATACO}$	EAB clock-to-output delay when using output registers	
t _{EABDATASU}	EAB data/address setup time before clock when using input register	
$t_{EABDATAH}$	EAB data/address hold time after clock when using input register	
t _{EABWESU}	EAB WE setup time before clock when using input register	
t _{EABWEH}	EAB WE hold time after clock when using input register	
t _{EABWDSU}	EAB data setup time before falling edge of write pulse when not using input registers	
t _{EABWDH}	EAB data hold time after falling edge of write pulse when not using input registers	
t _{EABWASU}	EAB address setup time before rising edge of write pulse when not using input registers	
t _{EABWAH}	EAB address hold time after falling edge of write pulse when not using input registers	
t _{EABWO}	EAB write enable to data output valid delay	

Symbol	Parameter	Conditions		
t _{DIN2IOE}	Delay from dedicated input pin to IOE control input			
t _{DCLK2LE}	Delay from dedicated clock pin to LE or EAB clock	(7)		
t _{DIN2DATA}	Delay from dedicated input or clock to LE or EAB data	(7)		
t _{DCLK2IOE}	Delay from dedicated clock pin to IOE clock	(7)		
t _{DIN2LE}	Delay from dedicated input pin to LE or EAB control input	(7)		
t _{SAMELAB}	Routing delay for an LE driving another LE in the same LAB			
t _{SAMEROW}	Routing delay for a row IOE, LE, or EAB driving a row IOE, LE, or EAB in the same row	(7)		
t _{SAME} COLUMN	Routing delay for an LE driving an IOE in the same column	(7)		
t _{DIFFROW}	Routing delay for a column IOE, LE, or EAB driving an LE or EAB in a different row	(7)		
t _{TWOROWS}	Routing delay for a row IOE or EAB driving an LE or EAB in a different row	(7)		
t _{LEPERIPH}	Routing delay for an LE driving a control signal of an IOE via the peripheral control bus	(7)		
t _{LABCARRY}	Routing delay for the carry-out signal of an LE driving the carry-in signal of a different LE in a different LAB			
t _{LABCASC}	Routing delay for the cascade-out signal of an LE driving the cascade-in signal of a different LE in a different LAB			

Table 37. External Timing ParametersNotes (8), (10)					
Symbol	Parameter	Conditions			
t _{DRR}	Register-to-register delay via four LEs, three row interconnects, and four local interconnects	(9)			
t _{INSU}	Setup time with global clock at IOE register				
t _{INH}	Hold time with global clock at IOE register				
t _{OUTCO}	Clock-to-output delay with global clock at IOE register				

Table 38. External Bidirectional Timing Parameters Note (10)						
Symbol	Parameter	Condition				
t _{INSUBIDIR}	Setup time for bidirectional pins with global clock at adjacent LE register					
t _{INHBIDIR}	Hold time for bidirectional pins with global clock at adjacent LE register					
t _{OUTCOBIDIR}	Clock-to-output delay for bidirectional pins with global clock at IOE register					
t _{XZBIDIR}	Synchronous IOE output buffer disable delay					
t _{ZXBIDIR}	Synchronous IOE output buffer enable delay, slow slew rate = off					

Notes to tables:

- (1) Microparameters are timing delays contributed by individual architectural elements. These parameters cannot be measured explicitly.
- (2) Operating conditions: $V_{CCIO} = 5.0 \text{ V} \pm 5\%$ for commercial use in FLEX 10K devices.

 $V_{CCIO} = 5.0 \text{ V} \pm 10\%$ for industrial use in FLEX 10K devices.

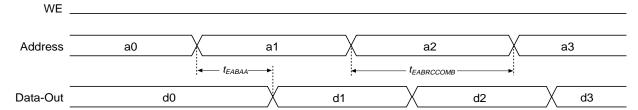
 V_{CCIO} = 3.3 V $\pm\,10\%$ for commercial or industrial use in FLEX 10KA devices.

- (3) Operating conditions: $V_{CCIO} = 3.3 \text{ V} \pm 10\%$ for commercial or industrial use in FLEX 10K devices.
 - V_{CCIO} = 2.5 V \pm 0.2 V for commercial or industrial use in FLEX 10KA devices.
- (4) Operating conditions: $V_{CCIO} = 2.5 \text{ V}$, 3.3 V, or 5.0 V.
- (5) Because the RAM in the EAB is self-timed, this parameter can be ignored when the WE signal is registered.
- (6) EAB macroparameters are internal parameters that can simplify predicting the behavior of an EAB at its boundary; these parameters are calculated by summing selected microparameters.
- (7) These parameters are worst-case values for typical applications. Post-compilation timing simulation and timing analysis are required to determine actual worst-case performance.
- (8) External reference timing parameters are factory-tested, worst-case values specified by Altera. A representative subset of signal paths is tested to approximate typical device applications.
- (9) Contact Altera Applications for test circuit specifications and test conditions.
- (10) These timing parameters are sample-tested only.

Figures 29 and 30 show the asynchronous and synchronous timing waveforms, respectively, for the EAB macroparameters in Table 34.

Figure 29. EAB Asynchronous Timing Waveforms

EAB Asynchronous Read



EAB Asynchronous Write

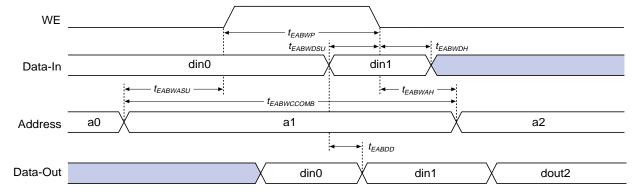
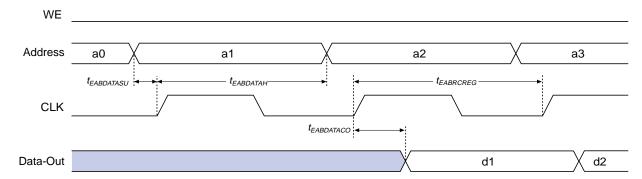
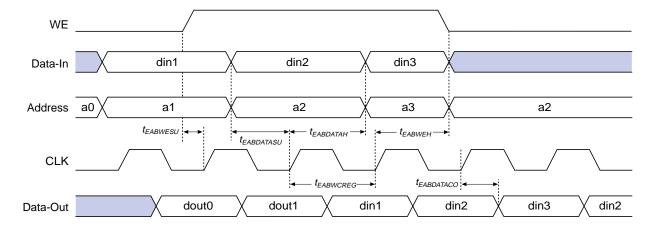


Figure 30. EAB Synchronous Timing Waveforms

EAB Synchronous Read



EAB Synchronous Write (EAB Output Registers Used)



Tables 39 through 47 show EPF10K10 and EPF10K20 device internal and external timing parameters.

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{LUT}		1.4		1.7	ns
t _{CLUT}		0.6		0.7	ns
t _{RLUT}		1.5		1.9	ns
t _{PACKED}		0.6		0.9	ns
t _{EN}		1.0		1.2	ns
t _{CICO}		0.2		0.3	ns
t _{CGEN}		0.9		1.2	ns
t _{CGENR}		0.9		1.2	ns
t _{CASC}		0.8		0.9	ns
$t_{\mathbb{C}}$		1.3		1.5	ns
t_{CO}		0.9		1.1	ns
t _{COMB}		0.5		0.6	ns
t _{SU}	1.3		2.5		ns
t _H	1.4		1.6		ns
t _{PRE}		1.0		1.2	ns
^t CLR		1.0		1.2	ns
^t cн	4.0		4.0		ns
t_{CL}	4.0		4.0		ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{IOD}		1.3		1.6	ns
t _{IOC}		0.5		0.7	ns
t_{IOCO}		0.2		0.2	ns
t_{IOCOMB}		0.0		0.0	ns
t_{IOSU}	2.8		3.2		ns
t_{IOH}	1.0		1.2		ns
t _{IOCLR}		1.0		1.2	ns
t_{OD1}		2.6		3.5	ns
t_{OD2}		4.9		6.4	ns
t _{OD3}		6.3		8.2	ns
t_{XZ}		4.5		5.4	ns
t_{ZX1}		4.5		5.4	ns
t_{ZX2}		6.8		8.3	ns
t_{ZX3}		8.2		10.1	ns
t _{INREG}		6.0		7.5	ns
t _{IOFD}		3.1		3.5	ns
t _{INCOMB}		3.1		3.5	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t _{EABDATA1}		1.5		1.9	ns
t _{EABDATA2}		4.8		6.0	ns
t _{EABWE1}		1.0		1.2	ns
t _{EABWE2}		5.0		6.2	ns
t _{EABCLK}		1.0		2.2	ns
t _{EABCO}		0.5		0.6	ns
t _{EABBYPASS}		1.5		1.9	ns
t _{EABSU}	1.5		1.8		ns
t _{EABH}	2.0		2.5		ns
t_{AA}		8.7		10.7	ns
t_{WP}	5.8		7.2		ns
t _{WDSU}	1.6		2.0		ns
t _{WDH}	0.3		0.4		ns
t _{WASU}	0.5		0.6		ns
t _{WAH}	1.0		1.2		ns
t_{WO}		5.0		6.2	ns
t_{DD}		5.0		6.2	ns
t _{EABOUT}		0.5		0.6	ns
t _{EABCH}	4.0		4.0		ns
t _{EABCL}	5.8		7.2		ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t_{EABAA}		13.7		17.0	ns
t _{EABRCCOMB}	13.7		17.0		ns
t _{EABRCREG}	9.7		11.9		ns
t _{EABWP}	5.8		7.2		ns
$t_{EABWCCOMB}$	7.3		9.0		ns
t _{EABWCREG}	13.0		16.0		ns
t_{EABDD}		10.0		12.5	ns
t _{EABDATACO}		2.0		3.4	ns
t _{EABDATASU}	5.3		5.6		ns
t _{EABDATAH}	0.0		0.0		ns
t _{EABWESU}	5.5		5.8		ns
t _{EABWEH}	0.0		0.0		ns
t _{EABWDSU}	5.5		5.8		ns
t _{EABWDH}	0.0		0.0		ns
t _{EABWASU}	2.1		2.7		ns
t _{EABWAH}	0.0		0.0		ns
t_{EABWO}		9.5		11.8	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t _{DIN2IOE}		4.8		6.2	ns
t _{DIN2LE}		2.6		3.8	ns
t _{DIN2DATA}		4.3		5.2	ns
t _{DCLK2IOE}		3.4		4.0	ns
t _{DCLK2LE}		2.6		3.8	ns
t _{SAMELAB}		0.6		0.6	ns
t _{SAMEROW}		3.6		3.8	ns
t _{SAME} COLUMN		0.9		1.1	ns
t _{DIFFROW}		4.5		4.9	ns
t _{TWOROWS}		8.1		8.7	ns
t _{LEPERIPH}		3.3		3.9	ns
t _{LABCARRY}		0.5		0.8	ns
t _{LABCASC}		2.7		3.0	ns

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	
t _{DIN2IOE}		5.2		6.6	ns
t _{DIN2LE}		2.6		3.8	ns
t _{DIN2DATA}		4.3		5.2	ns
t _{DCLK2IOE}		4.3		4.0	ns
t _{DCLK2LE}		2.6		3.8	ns
t _{SAMELAB}		0.6		0.6	ns
t _{SAMEROW}		3.7		3.9	ns
t _{SAME} COLUMN		1.4		1.6	ns
t _{DIFFROW}		5.1		5.5	ns
t _{TWOROWS}		8.8		9.4	ns
t _{LEPERIPH}		4.7		5.6	ns
t _{LABCARRY}		0.5		0.8	ns
t _{LABCASC}		2.7		3.0	ns

Table 45. EPF10K10 & EPF10K20 Device External Timing Parameters Note (1)							
Symbol -3 Speed Grade -4 Speed Grade							
	Min	Max	Min	Max			
t _{DRR}		16.1		20.0	ns		
t _{INSU} (2), (3)	5.5		6.0		ns		
t _{INH} (3)	0.0		0.0		ns		
t _{outco} (3)	2.0	6.7	2.0	8.4	ns		

Table 46. EPF10K10 Device External Bidirectional Timing Parameters Note (1)							
Symbol	-3 Speed Grade		-4 Speed Grade		Unit		
	Min	Max	Min	Max	-		
t _{INSUBIDIR}	4.5		5.6		ns		
t _{INHBIDIR}	0.0		0.0		ns		
toutcobidir	2.0	6.7	2.0	8.4	ns		
t _{XZBIDIR}		10.5		13.4	ns		
t _{ZXBIDIR}		10.5		13.4	ns		

Table 47. EPF10K20 Device External Bidirectional Timing Parameters Note (1)							
Symbol	-3 Speed Grade		-4 Speed Grade		Unit		
	Min	Max	Min	Max			
t _{INSUBIDIR}	4.6		5.7		ns		
t _{INHBIDIR}	0.0		0.0		ns		
t _{OUTCOBIDIR}	2.0	6.7	2.0	8.4	ns		
txzbidir		10.5		13.4	ns		
tzxbidir		10.5		13.4	ns		

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- Using an LE to register the signal may provide a lower setup time.
 This parameter is specified by characterization.

Tables 48 through 56 show EPF10K30, EPF10K40, and EPF10K50 device internal and external timing parameters.

Symbol	-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	1
t_{LUT}		1.3		1.8	ns
t _{CLUT}		0.6		0.6	ns
t _{RLUT}		1.5		2.0	ns
t _{PACKED}		0.5		0.8	ns
t _{EN}		0.9		1.5	ns
t _{CICO}		0.2		0.4	ns
t _{CGEN}		0.9		1.4	ns
t _{CGENR}		0.9		1.4	ns
t _{CASC}		1.0		1.2	ns
$t_{\mathbb{C}}$		1.3		1.6	ns
t_{CO}		0.9		1.2	ns
t _{COMB}		0.6		0.6	ns
t _{SU}	1.4		1.4		ns
t_H	0.9		1.3		ns
t _{PRE}		0.9		1.2	ns
t _{CLR}		0.9		1.2	ns
t _{CH}	4.0		4.0		ns
t_{CL}	4.0		4.0		ns

Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	
t_{IOD}		0.4		0.6	ns
t _{IOC}		0.5		0.9	ns
t_{IOCO}		0.4		0.5	ns
t_{IOCOMB}		0.0		0.0	ns
t _{IOSU}	3.1		3.5		ns
t_{IOH}	1.0		1.9		ns
t _{IOCLR}		1.0		1.2	ns
t_{OD1}		3.3		3.6	ns
t_{OD2}		5.6		6.5	ns
t_{OD3}		7.0		8.3	ns
t_{XZ}		5.2		5.5	ns
t _{ZX1}		5.2		5.5	ns
t _{ZX2}		7.5		8.4	ns
t_{ZX3}		8.9		10.2	ns
t _{INREG}		7.7		10.0	ns
t _{IOFD}		3.3		4.0	ns
t _{INCOMB}		3.3		4.0	ns

Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	
t _{EABDATA1}		1.5		1.9	ns
t _{EABDATA2}		4.8		6.0	ns
t _{EABWE1}		1.0		1.2	ns
t _{EABWE2}		5.0		6.2	ns
t _{EABCLK}		1.0		2.2	ns
t _{EABCO}		0.5		0.6	ns
t _{EABBYPASS}		1.5		1.9	ns
t _{EABSU}	1.5		1.8		ns
t _{EABH}	2.0		2.5		ns
t_{AA}		8.7		10.7	ns
t_{WP}	5.8		7.2		ns
t _{WDSU}	1.6		2.0		ns
t _{WDH}	0.3		0.4		ns
t _{WASU}	0.5		0.6		ns
t _{WAH}	1.0		1.2		ns
t_{WO}		5.0		6.2	ns
t_{DD}		5.0		6.2	ns
t _{EABOUT}		0.5		0.6	ns
t _{EABCH}	4.0		4.0		ns
t _{EABCL}	5.8		7.2		ns

Symbol	-3 Spee	d Grade	-4 Spee	ed Grade	Unit
	Min	Max	Min	Max	
t_{EABAA}		13.7		17.0	ns
t _{EABRCCOMB}	13.7		17.0		ns
t _{EABRCREG}	9.7		11.9		ns
t _{EABWP}	5.8		7.2		ns
t _{EABWCCOMB}	7.3		9.0		ns
t _{EABWCREG}	13.0		16.0		ns
t _{EABDD}		10.0		12.5	ns
t _{EABDATA} CO		2.0		3.4	ns
t _{EABDATASU}	5.3		5.6		ns
t _{EABDATAH}	0.0		0.0		ns
t _{EABWESU}	5.5		5.8		ns
t _{EABWEH}	0.0		0.0		ns
t _{EABWDSU}	5.5		5.8		ns
t _{EABWDH}	0.0		0.0		ns
t _{EABWASU}	2.1		2.7		ns
t _{EABWAH}	0.0		0.0		ns
t_{EABWO}		9.5		11.8	ns

Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit		
	Min	Max	Min	Max			
t _{DIN2IOE}		6.9		8.7	ns		
t _{DIN2LE}		3.6		4.8	ns		
t _{DIN2DATA}		5.5		7.2	ns		
t _{DCLK2IOE}		4.6		6.2	ns		
t _{DCLK2LE}		3.6		4.8	ns		
t _{SAMELAB}		0.3		0.3	ns		
t _{SAMEROW}		3.3		3.7	ns		
t _{SAME} COLUMN		2.5		2.7	ns		
t _{DIFFROW}		5.8		6.4	ns		
t _{TWOROWS}		9.1		10.1	ns		
t _{LEPERIPH}		6.2		7.1	ns		
t _{LABCARRY}		0.4		0.6	ns		
t _{LABCASC}		2.4		3.0	ns		

Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit	
	Min	Max	Min	Max		
t _{DIN2IOE}		7.6		9.4	ns	
t _{DIN2LE}		3.6		4.8	ns	
t _{DIN2DATA}		5.5		7.2	ns	
t _{DCLK2IOE}		4.6		6.2	ns	
t _{DCLK2LE}		3.6		4.8	ns	
t _{SAMELAB}		0.3		0.3	ns	
t _{SAMEROW}		3.3		3.7	ns	
t _{SAME} COLUMN		3.1		3.2	ns	
t _{DIFFROW}		6.4		6.4	ns	
t _{TWOROWS}		9.7		10.6	ns	
t _{LEPERIPH}		6.4		7.1	ns	
t _{LABCARRY}		0.4		0.6	ns	
t _{LABCASC}		2.4		3.0	ns	

Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit		
	Min	Max	Min	Max			
t _{DIN2IOE}		8.4		10.2	ns		
t _{DIN2LE}		3.6		4.8	ns		
t _{DIN2DATA}		5.5		7.2	ns		
t _{DCLK2IOE}		4.6		6.2	ns		
t _{DCLK2LE}		3.6		4.8	ns		
t _{SAMELAB}		0.3		0.3	ns		
t _{SAMEROW}		3.3		3.7	ns		
t _{SAME} COLUMN		3.9		4.1	ns		
t _{DIFFROW}		7.2		7.8	ns		
t _{TWOROWS}		10.5		11.5	ns		
t _{LEPERIPH}		7.5		8.2	ns		
t _{LABCARRY}		0.4		0.6	ns		
t _{LABCASC}		2.4		3.0	ns		

Table 55. EPF10K30, EPF10K40 & EPF10K50 Device External Timing Parameters Note (1)								
Symbol	-3 Spee	d Grade	-4 Spee	d Grade	Unit			
	Min	Max	Min	Max				
t _{DRR}		17.2		21.1	ns			
t _{INSU} (2), (3)	5.7		6.4		ns			
t _{INH} (3)	0.0		0.0		ns			
t _{оитсо} (3)	2.0	8.8	2.0	11.2	ns			

Table 56. EPF10K30, EPF10K40 & EPF10K50 Device External Bidirectional Timing Parameters Note (1)								
Symbol	-3 Spe	ed Grade	-4 Spec	Unit				
	Min	Max	Min	Max				
t _{INSUBIDIR}	4.1		4.6		ns			
t _{INHBIDIR}	0.0		0.0		ns			
toutcobidir	2.0	8.8	2.0	11.2	ns			
t _{XZBIDIR}		12.3		15.0	ns			
t _{ZXBIDIR}		12.3		15.0	ns			

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 57 through 63 show EPF10K70 device internal and external timing parameters.

Table 57. EPF10K70	Device LE Tim	ing Micropa	rameters	Note (1)			
Symbol	-2 Spee	d Grade	-3 Spec	ed Grade	-4 Spec	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.3		1.5		2.0	ns
t _{CLUT}		0.4		0.4		0.5	ns
t_{RLUT}		1.5		1.6		2.0	ns
t _{PACKED}		0.8		0.9		1.3	ns
t_{EN}		0.8		0.9		1.2	ns
t_{CICO}		0.2		0.2		0.3	ns
t _{CGEN}		1.0		1.1		1.4	ns
t _{CGENR}		1.1		1.2		1.5	ns
t_{CASC}		1.0		1.1		1.3	ns
$t_{\rm C}$		0.7		0.8		1.0	ns
$t_{\rm CO}$		0.9		1.0		1.4	ns
t_{COMB}		0.4		0.5		0.7	ns
t_{SU}	1.9		2.1		2.6		ns
t_H	2.1		2.3		3.1		ns
t _{PRE}		0.9		1.0		1.4	ns
t_{CLR}		0.9		1.0		1.4	ns
t _{CH}	4.0		4.0		4.0		ns
$t_{\rm CL}$	4.0		4.0		4.0		ns

Symbol	-2 Spee	d Grade	-3 Spee	ed Grade	-4 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		0.0		0.0		0.0	ns
t _{IOC}		0.4		0.5		0.7	ns
t _{IOCO}		0.4		0.4		0.9	ns
t _{IOCOMB}		0.0		0.0		0.0	ns
t _{IOSU}	4.5		5.0		6.2		ns
t _{IOH}	0.4		0.5		0.7		ns
t _{IOCLR}		0.6		0.7		1.6	ns
t _{OD1}		3.6		4.0		5.0	ns
t _{OD2}		5.6		6.3		7.3	ns
t _{OD3}		6.9		7.7		8.7	ns
t_{XZ}		5.5		6.2		6.8	ns
t_{ZX1}		5.5		6.2		6.8	ns
t_{ZX2}		7.5		8.5		9.1	ns
t_{ZX3}		8.8		9.9		10.5	ns
t _{INREG}		8.0		9.0		10.2	ns
t_{IOFD}		7.2		8.1		10.3	ns
t _{INCOMB}		7.2		8.1		10.3	ns

Symbol	-2 Spee	d Grade	-3 Spee	ed Grade	-4 Spec	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.3		1.5		1.9	ns
t _{EABDATA2}		4.3		4.8		6.0	ns
t _{EABWE1}		0.9		1.0		1.2	ns
t _{EABWE2}		4.5		5.0		6.2	ns
t _{EABCLK}		0.9		1.0		2.2	ns
t _{EABCO}		0.4		0.5		0.6	ns
t _{EABBYPASS}		1.3		1.5		1.9	ns
t _{EABSU}	1.3		1.5		1.8		ns
t _{EABH}	1.8		2.0		2.5		ns
t_{AA}		7.8		8.7		10.7	ns
t_{WP}	5.2		5.8		7.2		ns
t _{WDSU}	1.4		1.6		2.0		ns
t_{WDH}	0.3		0.3		0.4		ns
t _{WASU}	0.4		0.5		0.6		ns
t _{WAH}	0.9		1.0		1.2		ns
t_{WO}		4.5		5.0		6.2	ns
t_{DD}		4.5		5.0		6.2	ns
t _{EABOUT}		0.4		0.5		0.6	ns
t _{EABCH}	4.0		4.0		4.0		ns
t _{EABCL}	5.2		5.8		7.2		ns

Symbol	-2 Speed	l Grade	-3 Spee	ed Grade	-4 Spec	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		12.1		13.7		17.0	ns
t _{EABRCCOMB}	12.1		13.7		17.0		ns
t _{EABRCREG}	8.6		9.7		11.9		ns
t _{EABWP}	5.2		5.8		7.2		ns
t _{EABWCCOMB}	6.5		7.3		9.0		ns
t _{EABWCREG}	11.6		13.0		16.0		ns
t _{EABDD}		8.8		10.0		12.5	ns
t _{EABDATA} CO		1.7		2.0		3.4	ns
t _{EABDATASU}	4.7		5.3		5.6		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	4.9		5.5		5.8		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	1.8		2.1		2.7		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	4.1		4.7		5.8		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		8.4		9.5		11.8	ns

Symbol	-2 Spee	d Grade	-3 Spee	ed Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		6.6		7.3		8.8	ns
t _{DIN2LE}		4.2		4.8		6.0	ns
t _{DIN2DATA}		6.5		7.1		10.8	ns
t _{DCLK2IOE}		5.5		6.2		7.7	ns
t _{DCLK2LE}		4.2		4.8		6.0	ns
t _{SAMELAB}		0.4		0.4		0.5	ns
t _{SAMEROW}		4.8		4.9		5.5	ns
t _{SAME} COLUMN		3.3		3.4		3.7	ns
t _{DIFFROW}		8.1		8.3		9.2	ns
t _{TWOROWS}		12.9		13.2		14.7	ns
t _{LEPERIPH}		5.5		5.7		6.5	ns
t _{LABCARRY}		0.8		0.9		1.1	ns
t _{LABCASC}		2.7		3.0		3.2	ns

Table 62. EPF10K70 Device External Timing Parameters Note (1)										
Symbol	-2 Spee	ed Grade	-3 Spee	ed Grade	-4 Spee	Unit				
	Min	Max	Min	Max	Min	Max				
t _{DRR}		17.2		19.1		24.2	ns			
t _{INSU} (2), (3)	6.6		7.3		8.0		ns			
t _{INH} (3)	0.0		0.0		0.0		ns			
t _{outco} (3)	2.0	9.9	2.0	11.1	2.0	14.3	ns			

Table 63. EPF10K70 Device External Bidirectional Timing Parameters Note (1)											
Symbol	-2 Spec	ed Grade	-3 Spe	ed Grade	-4 Spee	Unit					
	Min	Max	Min	Max	Min	Max					
t _{INSUBIDIR}	7.4		8.1		10.4		ns				
t _{INHBIDIR}	0.0		0.0		0.0		ns				
toutcobidir	2.0	9.9	2.0	11.1	2.0	14.3	ns				
t _{XZBIDIR}		13.7		15.4		18.5	ns				
t _{ZXBIDIR}		13.7		15.4		18.5	ns				

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables $64\,\mathrm{through}\,70\,\mathrm{show}\,\mathrm{EPF10K100}\,\mathrm{device}\,\mathrm{internal}\,\mathrm{and}\,\mathrm{external}\,\mathrm{timing}\,\mathrm{parameters}.$

Symbol	-3DX Spe	ed Grade	-3 Spee	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.5		1.5		2.0	ns
t _{CLUT}		0.4		0.4		0.5	ns
t _{RLUT}		1.6		1.6		2.0	ns
t _{PACKED}		0.9		0.9		1.3	ns
t_{EN}		0.9		0.9		1.2	ns
t _{CICO}		0.2		0.2		0.3	ns
t _{CGEN}		1.1		1.1		1.4	ns
t _{CGENR}		1.2		1.2		1.5	ns
t _{CASC}		1.1		1.1		1.3	ns
$t_{\mathbb{C}}$		0.8		0.8		1.0	ns
t_{CO}		1.0		1.0		1.4	ns
t _{COMB}		0.5		0.5		0.7	ns
t_{SU}	2.1		2.1		2.6		ns
t_H	2.3		2.3		3.1		ns
t _{PRE}		1.0		1.0		1.4	ns
t _{CLR}		1.0		1.0		1.4	ns
t _{CH}	4.0		4.0		4.0		ns
t _{CL}	4.0		4.0		4.0		ns

Symbol	-3DX Spe	eed Grade	-3 Spee	ed Grade	-4 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		0.0		0.0		0.0	ns
t _{IOC}		0.5		0.5		0.7	ns
t _{IOCO}		0.4		0.4		0.9	ns
t _{IOCOMB}		0.0		0.0		0.0	ns
t _{IOSU}	5.5		5.5		6.7		ns
t _{IOH}	0.5		0.5		0.7		ns
t _{IOCLR}		0.7		0.7		1.6	ns
t _{OD1}		4.0		4.0		5.0	ns
t _{OD2}		6.3		6.3		7.3	ns
t_{OD3}		7.7		7.7		8.7	ns
t_{XZ}		6.2		6.2		6.8	ns
t_{ZX1}		6.2		6.2		6.8	ns
t_{ZX2}		8.5		8.5		9.1	ns
t_{ZX3}		9.9		9.9		10.5	ns
t _{INREG} without ClockLock or ClockBoost circuitry		9.0		9.0		10.5	ns
t _{INREG} with ClockLock or ClockBoost circuitry		3.0		_		_	ns
t _{IOFD}		8.1		8.1		10.3	ns
t _{INCOMB}		8.1		8.1		10.3	ns

Symbol	-3DX Spe	ed Grade	-3 Spee	d Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.5		1.5		1.9	ns
t _{EABDATA2}		4.8		4.8		6.0	ns
t _{EABWE1}		1.0		1.0		1.2	ns
t _{EABWE2}		5.0		5.0		6.2	ns
t _{EABCLK}		1.0		1.0		2.2	ns
t _{EABCO}		0.5		0.5		0.6	ns
t _{EABBYPASS}		1.5		1.5		1.9	ns
t _{EABSU}	1.5		1.5		1.8		ns
t _{EABH}	2.0		2.0		2.5		ns
t_{AA}		8.7		8.7		10.7	ns
t_{WP}	5.8		5.8		7.2		ns
t _{WDSU}	1.6		1.6		2.0		ns
t _{WDH}	0.3		0.3		0.4		ns
t _{WASU}	0.5		0.5		0.6		ns
t_{WAH}	1.0		1.0		1.2		ns
t_{WO}		5.0		5.0		6.2	ns
t _{DD}		5.0		5.0		6.2	ns
t _{EABOUT}		0.5		0.5		0.6	ns
t _{EABCH}	4.0		4.0		4.0		ns
t _{EABCL}	5.8		5.8		7.2		ns

Symbol	-3DX Spe	ed Grade	-3 Spee	d Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		13.7		13.7		17.0	ns
t _{EABRCCOMB}	13.7		13.7		17.0		ns
t _{EABRCREG}	9.7		9.7		11.9		ns
t _{EABWP}	5.8		5.8		7.2		ns
t _{EABWCCOMB}	7.3		7.3		9.0		ns
t _{EABWCREG}	13.0		13.0		16.0		ns
t _{EABDD}		10.0		10.0		12.5	ns
t _{EABDATA} CO		2.0		2.0		3.4	ns
t _{EABDATASU}	5.3		5.3		5.6		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	5.5		5.5		5.8		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	5.5		5.5		5.8		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	2.1		2.1		2.7		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		9.5		9.5		11.8	ns

Symbol	-3DX Sp	eed Grade	-3 Spee	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		10.3		10.3		12.2	ns
t _{DIN2LE}		4.8		4.8		6.0	ns
t _{DIN2DATA}		7.3		7.3		11.0	ns
$t_{DCLK2IOE}$ without ClockLock or ClockBoost circuitry		6.2		6.2		7.7	ns
$t_{DCLK2IOE}$ with ClockLock or ClockBoost circuitry		2.3		_		_	ns
t _{DCLK2LE} without ClockLock or ClockBoost circuitry		4.8		4.8		6.0	ns
$t_{DCLK2LE}$ with ClockLock or ClockBoost circuitry		2.3		_		_	ns
$t_{SAMELAB}$		0.4		0.4		0.5	ns
t _{SAMEROW}		4.9		4.9		5.5	ns
t _{SAME} COLUMN		5.1		5.1		5.4	ns
t _{DIFFROW}		10.0		10.0		10.9	ns
t _{TWOROWS}		14.9		14.9		16.4	ns
t _{LEPERIPH}		6.9		6.9		8.1	ns
t _{LABCARRY}		0.9		0.9		1.1	ns
t _{LABCASC}		3.0		3.0		3.2	ns

Symbol	-3DX Sp	-3DX Speed Grade		ed Grade	-4 Spee	Unit		
	Min Max		Min	Max	Min Max		7	
t _{DRR}		19.1		19.1		24.2	ns	
t _{INSU} (2), (3), (4)	7.8		7.8		8.5		ns	
t _{оитсо} (3), (4)	2.0	11.1	2.0	11.1	2.0	14.3	ns	
t _{INH} (3)	0.0		0.0		0.0		ns	
t _{INSU} (2), (3), (5)	6.2		_		-		ns	
t _{OUTCO} (3), (5)	2.0	6.7		_		_	ns	

Symbol	-3DX Sp	-3DX Speed Grade		d Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR} (4)	8.1		8.1		10.4		ns
t _{INHBIDIR} (4)	0.0		0.0		0.0		ns
toutcobidir (4)	2.0	11.1	2.0	11.1	2.0	14.3	ns
t _{XZBIDIR} (4)		15.3		15.3		18.4	ns
t _{ZXBIDIR} (4)		15.3		15.3		18.4	ns
t _{INSUBIDIR} (5)	9.1		_		-		ns
t _{INHBIDIR} (5)	0.0		-		-		ns
toutcobidir (5)	2.0	7.2	-	-	_	_	ns
t _{XZBIDIR} (5)		14.3		-		_	ns
t _{ZXBIDIR} (5)		14.3		_		_	ns

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.
- (4) This parameter is measured without the use of the ClockLock or ClockBoost circuits.
- (5) This parameter is measured with the use of the ClockLock or ClockBoost circuits.

Tables 71 through 77 show EPF10K50V device internal and external timing parameters.

Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spe	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.9		1.0		1.3		1.6	ns
t _{CLUT}		0.1		0.5		0.6		0.6	ns
t _{RLUT}		0.5		0.8		0.9		1.0	ns
t _{PACKED}		0.4		0.4		0.5		0.7	ns
t _{EN}		0.7		0.9		1.1		1.4	ns
t _{CICO}		0.2		0.2		0.2		0.3	ns
t _{CGEN}		0.8		0.7		8.0		1.2	ns
t _{CGENR}		0.4		0.3		0.3		0.4	ns
t _{CASC}		0.7		0.7		8.0		0.9	ns
t_C		0.3		1.0		1.3		1.5	ns
t_{CO}		0.5		0.7		0.9		1.0	ns
t _{COMB}		0.4		0.4		0.5		0.6	ns
t _{SU}	0.8		1.6		2.2		2.5		ns
t _H	0.5		0.8		1.0		1.4		ns
t _{PRE}		0.8		0.4		0.5		0.5	ns
t _{CLR}		0.8		0.4		0.5		0.5	ns
t _{CH}	2.0		4.0		4.0		4.0		ns
t_{CL}	2.0		4.0		4.0		4.0		ns

Table 72. EPI	F10K50V D	evice IOE T	iming Mid	croparamet	ers No	ote (1)			
Symbol	-1 Spec	ed Grade	-2 Spee	ed Grade	-3 Spee	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.2		1.6		1.9		2.1	ns
t _{IOC}		0.3		0.4		0.5		0.5	ns
t_{IOCO}		0.3		0.3		0.4		0.4	ns
t_{IOCOMB}		0.0		0.0		0.0		0.0	ns
t _{IOSU}	2.8		2.8		3.4		3.9		ns
t _{IOH}	0.7		0.8		1.0		1.4		ns
t _{IOCLR}		0.5		0.6		0.7		0.7	ns
t _{OD1}		2.8		3.2		3.9		4.7	ns
t _{OD2}		_		_		_		_	ns
t_{OD3}		6.5		6.9		7.6		8.4	ns
t_{XZ}		2.8		3.1		3.8		4.6	ns
t_{ZX1}		2.8		3.1		3.8		4.6	ns
t_{ZX2}		_		_		_		_	ns
t_{ZX3}		6.5		6.8		7.5		8.3	ns
t _{INREG}		5.0		5.7		7.0		9.0	ns
t _{IOFD}		1.5		1.9		2.3		2.7	ns
t _{INCOMB}		1.5		1.9		2.3		2.7	ns

Symbol	-1 Spee	ed Grade	-2 Spee	ed Grade	-3 Speed Grade		-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.7		2.8		3.4		4.6	ns
t _{EABDATA2}		4.9		3.9		4.8		5.9	ns
t _{EABWE1}		0.0		2.5		3.0		3.7	ns
t _{EABWE2}		4.0		4.1		5.0		6.2	ns
t _{EABCLK}		0.4		0.8		1.0		1.2	ns
t _{EABCO}		0.1		0.2		0.3		0.4	ns
t _{EABBYPASS}		0.9		1.1		1.3		1.6	ns
t _{EABSU}	0.8		1.5		1.8		2.2		ns
t _{EABH}	0.8		1.6		2.0		2.5		ns
t_{AA}		5.5		8.2		10.0		12.4	ns
t_{WP}	6.0		4.9		6.0		7.4		ns
t _{WDSU}	0.1		0.8		1.0		1.2		ns
t_{WDH}	0.1		0.2		0.3		0.4		ns
t _{WASU}	0.1		0.4		0.5		0.6		ns
t _{WAH}	0.1		0.8		1.0		1.2		ns
t_{WO}		2.8		4.3		5.3		6.5	ns
t_{DD}		2.8		4.3		5.3		6.5	ns
t _{EABOUT}		0.5		0.4		0.5		0.6	ns
t _{EABCH}	2.0		4.0		4.0		4.0		ns
t _{EABCL}	6.0		4.9		6.0		7.4		ns

Symbol	-1 Spee	ed Grade	-2 Speed Grade		-3 Speed Grade		-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t _{EABAA}		9.5		13.6		16.5		20.8	ns
t _{EABRCCOMB}	9.5		13.6		16.5		20.8		ns
t _{EABRCREG}	6.1		8.8		10.8		13.4		ns
t _{EABWP}	6.0		4.9		6.0		7.4		ns
t _{EABWCCOMB}	6.2		6.1		7.5		9.2		ns
t _{EABWCREG}	12.0		11.6		14.2		17.4		ns
t _{EABDD}		6.8		9.7		11.8		14.9	ns
t _{EABDATA} CO		1.0		1.4		1.8		2.2	ns
t _{EABDATASU}	5.3		4.6		5.6		6.9		ns
t _{EABDATAH}	0.0		0.0		0.0		0.0		ns
t _{EABWESU}	4.4		4.8		5.8		7.2		ns
t _{EABWEH}	0.0		0.0		0.0		0.0		ns
t _{EABWDSU}	1.8		1.1		1.4		2.1		ns
t _{EABWDH}	0.0		0.0		0.0		0.0		ns
t _{EABWASU}	4.5		4.6		5.6		7.4		ns
t _{EABWAH}	0.0		0.0		0.0		0.0		ns
t _{EABWO}		5.1		9.4		11.4		14.0	ns

Symbol	-1 Spee	ed Grade	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		4.7		6.0		7.1		8.2	ns
t _{DIN2LE}		2.5		2.6		3.1		3.9	ns
t _{DIN2DATA}		4.4		5.9		6.8		7.7	ns
t _{DCLK2IOE}		2.5		3.9		4.7		5.5	ns
t _{DCLK2LE}		2.5		2.6		3.1		3.9	ns
t _{SAMELAB}		0.2		0.2		0.3		0.3	ns
t _{SAMEROW}		2.8		3.0		3.2		3.4	ns
t _{SAME} COLUMN		3.0		3.2		3.4		3.6	ns
t _{DIFFROW}		5.8		6.2		6.6		7.0	ns
t _{TWOROWS}		8.6		9.2		9.8		10.4	ns
t _{LEPERIPH}		4.5		5.5		6.1		7.0	ns
t _{LABCARRY}		0.3		0.4		0.5		0.7	ns
t _{LABCASC}		0.0		1.3		1.6		2.0	ns

Table 76. EPF	10K50V De	evice Exter	nal Timing	g Paramete	ers Not	e (1)			
Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spee	d Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	Min	Max	
t _{DRR}		11.2		14.0		17.2		21.1	ns
t _{INSU} (2), (3)	5.5		4.2		5.2		6.9		ns
t _{INH} (3)	0.0		0.0		0.0		0.0		ns
t _{оитсо} (3)	2.0	5.9	2.0	7.8	2.0	9.5	2.0	11.1	ns

Table 77. EPF10K50V Device External Bidirectional Timing Parameters Note (1)											
Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Speed Grade		-4 Speed Grade		Unit		
	Min	Max	Min	Max	Min	Max	Min	Max			
t _{INSUBIDIR}	2.0		2.8		3.5		4.1		ns		
t _{INHBIDIR}	0.0		0.0		0.0		0.0		ns		
toutcobidir	2.0	5.9	2.0	7.8	2.0	9.5	2.0	11.1	ns		
t _{XZBIDIR}		8.0		9.8		11.8		14.3	ns		
t _{ZXBIDIR}		8.0		9.8		11.8		14.3	ns		

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 78 through 84 show EPF10K130V device internal and external timing parameters.

Table 78. EPF10K13	OV Device LE	Timing Micro	oparameters	Note (1)		
Symbol	-2 Spec	ed Grade	-3 Spe	ed Grade	-4 Spec	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.3		1.8		2.3	ns
t _{CLUT}		0.5		0.7		0.9	ns
t _{RLUT}		1.2		1.7		2.2	ns
t _{PACKED}		0.5		0.6		0.7	ns
t _{EN}		0.6		0.8		1.0	ns
t _{CICO}		0.2		0.3		0.4	ns
t _{CGEN}		0.3		0.4		0.5	ns
t _{CGENR}		0.7		1.0		1.3	ns
t _{CASC}		0.9		1.2		1.5	ns
$t_{\rm C}$		1.9		2.4		3.0	ns
$t_{\rm CO}$		0.6		0.9		1.1	ns
t _{COMB}		0.5		0.7		0.9	ns
t_{SU}	0.2		0.2		0.3		ns
t _H	0.0		0.0		0.0		ns
t _{PRE}		2.4		3.1		3.9	ns
t _{CLR}		2.4		3.1		3.9	ns
t _{CH}	4.0		4.0		4.0		ns
t_{CL}	4.0		4.0		4.0		ns

Symbol	-2 Spee	d Grade	-3 Spec	ed Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.3		1.6		2.0	ns
t _{IOC}		0.4		0.5		0.7	ns
t _{IOCO}		0.3		0.4		0.5	ns
t_{IOCOMB}		0.0		0.0		0.0	ns
t _{IOSU}	2.6		3.3		3.8		ns
t _{IOH}	0.0		0.0		0.0		ns
t _{IOCLR}		1.7		2.2		2.7	ns
t _{OD1}		3.5		4.4		5.0	ns
t_{OD2}		-		_		-	ns
t _{OD3}		8.2		8.1		9.7	ns
t_{XZ}		4.9		6.3		7.4	ns
t_{ZX1}		4.9		6.3		7.4	ns
t_{ZX2}		-					ns
t_{ZX3}		9.6		10.0		12.1	ns
t _{INREG}		7.9		10.0		12.6	ns
t _{IOFD}		6.2		7.9		9.9	ns
t _{INCOMB}		6.2		7.9		9.9	ns

Symbol	-2 Spee	d Grade	-3 Spee	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.9		2.4		2.4	ns
t _{EABDATA2}		3.7		4.7		4.7	ns
t _{EABWE1}		1.9		2.4		2.4	ns
t _{EABWE2}		3.7		4.7		4.7	ns
t _{EABCLK}		0.7		0.9		0.9	ns
t _{EABCO}		0.5		0.6		0.6	ns
t _{EABBYPASS}		0.6		0.8		0.8	ns
t _{EABSU}	1.4		1.8		1.8		ns
t _{EABH}	0.0		0.0		0.0		ns
t_{AA}		5.6		7.1		7.1	ns
t_{WP}	3.7		4.7		4.7		ns
t _{WDSU}	4.6		5.9		5.9		ns
t_{WDH}	0.0		0.0		0.0		ns
t _{WASU}	3.9		5.0		5.0		ns
t _{WAH}	0.0		0.0		0.0		ns
t_{WO}		5.6		7.1		7.1	ns
t_{DD}		5.6		7.1		7.1	ns
t _{EABOUT}		2.4		3.1		3.1	ns
t _{EABCH}	4.0		4.0		4.0		ns
t _{EABCL}	4.0		4.7		4.7		ns

Symbol	-2 Spee	d Grade	-3 Spee	ed Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		11.2		14.2		14.2	ns
t _{EABRCCOMB}	11.1		14.2		14.2		ns
t _{EABRCREG}	8.5		10.8		10.8		ns
t _{EABWP}	3.7		4.7		4.7		ns
t _{EABWCCOMB}	7.6		9.7		9.7		ns
t _{EABWCREG}	14.0		17.8		17.8		ns
t _{EABDD}		11.1		14.2		14.2	ns
t _{EABDATA} CO		3.6		4.6		4.6	ns
t _{EABDATASU}	4.4		5.6		5.6		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	4.4		5.6		5.6		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	4.6		5.9		5.9		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	3.9		5.0		5.0		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		11.1		14.2		14.2	ns

Symbol	-2 Spee	ed Grade	-3 Spee	d Grade	-4 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		8.0		9.0		9.5	ns
t _{DIN2LE}		2.4		3.0		3.1	ns
t _{DIN2DATA}		5.0		6.3		7.4	ns
t _{DCLK2IOE}		3.6		4.6		5.1	ns
t _{DCLK2LE}		2.4		3.0		3.1	ns
t _{SAMELAB}		0.4		0.6		0.8	ns
t _{SAMEROW}		4.5		5.3		6.5	ns
t _{SAME} COLUMN		9.0		9.5		9.7	ns
t _{DIFFROW}		13.5		14.8		16.2	ns
t _{TWOROWS}		18.0		20.1		22.7	ns
t _{LEPERIPH}		8.1		8.6		9.5	ns
t _{LABCARRY}		0.6		0.8		1.0	ns
t _{LABCASC}		0.8		1.0		1.2	ns

Table 83. EPF10K130V Device External Timing Parameters Note (1)										
Symbol	-2 Spec	ed Grade	-3 Speed Grade -4 Speed Grade			d Grade	Unit			
	Min	Max	Min	Max	Min	Max				
t _{DRR}		15.0		19.1		24.2	ns			
t _{INSU} (2), (3)	6.9		8.6		11.0		ns			
t _{INH} (3)	0.0		0.0		0.0		ns			
t _{outco} (3)	2.0	7.8	2.0	9.9	2.0	11.3	ns			

Table 84. EPF10K130V Device External Bidirectional Timing Parameters Note (1)										
Symbol	-2 Spee	ed Grade	-3 Speed Grade -4 Speed Gra			d Grade	Unit			
	Min	Max	Min	Max	Min	Max				
t _{INSUBIDIR}	6.7		8.5		10.8		ns			
t _{INHBIDIR}	0.0		0.0		0.0		ns			
toutcobidir	2.0	6.9	2.0	8.8	2.0	10.2	ns			
t _{XZBIDIR}		12.9		16.4		19.3	ns			
t _{ZXBIDIR}		12.9		16.4		19.3	ns			

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 85 through 91 show EPF10K10A device internal and external timing parameters.

Symbol	-1 Spee	d Grade	-2 Spec	ed Grade	-3 Spec	ed Grade	Unit
•	Min	Max	Min	Max	Min	Max	_
t_{LUT}		0.9		1.2		1.6	ns
t_{CLUT}		1.2		1.4		1.9	ns
t _{RLUT}		1.9		2.3		3.0	ns
t _{PACKED}		0.6		0.7		0.9	ns
t_{EN}		0.5		0.6		0.8	ns
t _{CICO}		02		0.3		0.4	ns
t _{CGEN}		0.7		0.9		1.1	ns
t _{CGENR}		0.7		0.9		1.1	ns
t _{CASC}		1.0		1.2		1.7	ns
t_C		1.2		1.4		1.9	ns
t_{CO}		0.5		0.6		0.8	ns
t _{COMB}		0.5		0.6		0.8	ns
t _{SU}	1.1		1.3		1.7		ns
t _H	0.6		0.7		0.9		ns
t _{PRE}		0.5		0.6		0.9	ns
t _{CLR}		0.5		0.6		0.9	ns
t _{CH}	3.0		3.5		4.0		ns
t_{CL}	3.0		3.5		4.0		ns

Table 86. EPF10	K10A Device	IOE Timing I	Microparamet	ers Note	(1) (Part 1 of	2)	
Symbol	-1 Spec	ed Grade	-2 Spee	d Grade	-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
		1.3		1.5		2.0	ns
t_{IOC}		0.2		0.3		0.3	ns
t_{IOCO}		0.2		0.3		0.4	ns
t _{IOCOMB}		0.6		0.7		0.9	ns
t_{IOSU}	0.8		1.0		1.3		ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t_{IOH}	0.8		1.0		1.3		ns
t _{IOCLR}		1.2		1.4		1.9	ns
t_{OD1}		1.2		1.4		1.9	ns
t _{OD2}		2.9		3.5		4.7	ns
t_{OD3}		6.6		7.8		10.5	ns
t_{XZ}		1.2		1.4		1.9	ns
t_{ZX1}		1.2		1.4		1.9	ns
t_{ZX2}		2.9		3.5		4.7	ns
t_{ZX3}		6.6		7.8		10.5	ns
t _{INREG}		5.2		6.3		8.4	ns
t _{IOFD}		3.1		3.8		5.0	ns
t _{INCOMB}		3.1		3.8		5.0	ns

Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		3.3		3.9		5.2	ns
t _{EABDATA2}		1.0		1.3		1.7	ns
t _{EABWE1}		2.6		3.1		4.1	ns
t _{EABWE2}		2.7		3.2		4.3	ns
t _{EABCLK}		0.0		0.0		0.0	ns
t _{EABCO}		1.2		1.4		1.8	ns
t _{EABBYPASS}		0.1		0.2		0.2	ns
t _{EABSU}	1.4		1.7		2.2		ns
t _{EABH}	0.1		0.1		0.1		ns
t_{AA}		4.5		5.4		7.3	ns
t_{WP}	2.0		2.4		3.2		ns
t _{WDSU}	0.7		0.8		1.1		ns
t _{WDH}	0.5		0.6		0.7		ns
t _{WASU}	0.6		0.7		0.9		ns
t _{WAH}	0.9		1.1		1.5		ns
t_{WO}		3.3		3.9		5.2	ns
t_{DD}		3.3		3.9		5.2	ns
t _{EABOUT}		0.1		0.1		0.2	ns
t _{EABCH}	3.0		3.5		4.0		ns
t _{EABCL}	3.03		3.5		4.0		ns

Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		8.1		9.8		13.1	ns
t _{EABRCCOMB}	8.1		9.8		13.1		ns
t _{EABRCREG}	5.8		6.9		9.3		ns
t _{EABWP}	2.0		2.4		3.2		ns
t _{EABWCCOMB}	3.5		4.2		5.6		ns
t _{EABWCREG}	9.4		11.2		14.8		ns
t _{EABDD}		6.9		8.3		11.0	ns
t _{EABDATA} CO		1.3		1.5		2.0	ns
t _{EABDATASU}	2.4		3.0		3.9		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	4.1		4.9		6.5		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	1.4		1.6		2.2		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	2.5		3.0		4.1		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		6.2		7.5		9.9	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		4.2		5.0		6.5	ns
t _{DIN2LE}		2.2		2.6		3.4	ns
t _{DIN2DATA}		4.3		5.2		7.1	ns
t _{DCLK2IOE}		4.2		4.9		6.6	ns
t _{DCLK2LE}		2.2		2.6		3.4	ns
t _{SAMELAB}		0.1		0.1		0.2	ns
t _{SAMEROW}		2.2		2.4		2.9	ns
t _{SAME} COLUMN		0.8		1.0		1.4	ns
t _{DIFFROW}		3.0		3.4		4.3	ns
t _{TWOROWS}		5.2		5.8		7.2	ns
t _{LEPERIPH}		1.8		2.2		2.8	ns
t _{LABCARRY}		0.5		0.5		0.7	ns
t _{LABCASC}		0.9		1.0		1.5	ns

Table 90. EPF10	K10A Externa	al Reference 1	Timing Paran	neters No	nte (1)		
Symbol	-1 Spec	ed Grade	-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	1
t _{DRR}		10.0		12.0		16.0	ns
t _{INSU} (2), (3)	1.6		2.1		2.8		ns
t _{INH} (3)	0.0		0.0		0.0		ns
t _{outco} (3)	2.0	5.8	2.0	6.9	2.0	9.2	ns

Table 91. EPF10K10/	4 Device Exteri	nal Bidirectio	nal Timing I	Parameters	Note (1)		
Symbol	-2 Spec	ed Grade	-3 Spe	ed Grade	-4 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR}	2.4		3.3		4.5		ns
t _{INHBIDIR}	0.0		0.0		0.0		ns
toutcobidir	2.0	5.8	2.0	6.9	2.0	9.2	ns
t _{XZBIDIR}		6.3		7.5		9.9	ns
t _{ZXBIDIR}		6.3		7.5		9.9	ns

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 92 through 98 show EPF10K30A device internal and external timing parameters.

Symbol	-1 Spee	ed Grade	-2 Spee	d Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.8		1.1		1.5	ns
t _{CLUT}		0.6		0.7		1.0	ns
t_{RLUT}		1.2		1.5		2.0	ns
t _{PACKED}		0.6		0.6		1.0	ns
t_{EN}		1.3		1.5		2.0	ns
t_{CICO}		0.2		0.3		0.4	ns
t _{CGEN}		0.8		1.0		1.3	ns
t _{CGENR}		0.6		0.8		1.0	ns
t_{CASC}		0.9		1.1		1.4	ns
t_C		1.1		1.3		1.7	ns
t_{CO}		0.4		0.6		0.7	ns
t _{COMB}		0.6		0.7		0.9	ns
t _{SU}	0.9		0.9		1.4		ns
t _H	1.1		1.3		1.7		ns
t _{PRE}		0.5		0.6		0.8	ns
t _{CLR}		0.5		0.6		0.8	ns
t _{CH}	3.0		3.5		4.0		ns
t_{CL}	3.0		3.5		4.0		ns

Symbol	-1 Spee	-1 Speed Grade		-2 Speed Grade		d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.2		2.6		3.4	ns
t _{IOC}		0.3		0.3		0.5	ns
t _{IOCO}		0.2		0.2		0.3	ns
t _{IOCOMB}		0.5		0.6		0.8	ns
t _{IOSU}	1.4		1.7		2.2		ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t_{IOH}	0.9		1.1		1.4		ns
t _{IOCLR}		0.7		8.0		1.0	ns
t_{OD1}		1.9		2.2		2.9	ns
t_{OD2}		4.8		5.6		7.3	ns
t_{OD3}		7.0		8.2		10.8	ns
t_{XZ}		2.2		2.6		3.4	ns
t_{ZX1}		2.2		2.6		3.4	ns
t_{ZX2}		5.1		6.0		7.8	ns
t_{ZX3}		7.3		8.6		11.3	ns
t _{INREG}		4.4		5.2		6.8	ns
t_{IOFD}		3.8		4.5		5.9	ns
t _{INCOMB}		3.8		4.5		5.9	ns

Symbol	-1 Spee	d Grade	-2 Spee	d Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		5.5		6.5		8.5	ns
t _{EABDATA2}		1.1		1.3		1.8	ns
t _{EABWE1}		2.4		2.8		3.7	ns
t _{EABWE2}		2.1		2.5		3.2	ns
t _{EABCLK}		0.0		0.0		0.2	ns
t _{EABCO}		1.7		2.0		2.6	ns
t _{EABBYPASS}		0.0		0.0		0.3	ns
t _{EABSU}	1.2		1.4		1.9		ns
t _{EABH}	0.1		0.1		0.3		ns
t_{AA}		4.2		5.0		6.5	ns
t_{WP}	3.8		4.5		5.9		ns
t _{WDSU}	0.1		0.1		0.2		ns
t _{WDH}	0.1		0.1		0.2		ns
t _{WASU}	0.1		0.1		0.2		ns
t _{WAH}	0.1		0.1		0.2		ns
t_{WO}		3.7		4.4		6.4	ns
t_{DD}		3.7		4.4		6.4	ns
t _{EABOUT}		0.0		0.1		0.6	ns
t _{EABCH}	3.0		3.5		4.0		ns
t _{EABCL}	3.8		4.5		5.9		ns

Symbol	-1 Speed Grade		-2 Spee	d Grade	-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	•
t _{EABAA}		9.7		11.6		16.2	ns
t _{EABRCCOMB}	9.7		11.6		16.2		ns
t _{EABRCREG}	5.9		7.1		9.7		ns
t _{EABWP}	3.8		4.5		5.9		ns
t _{EABWCCOMB}	4.0		4.7		6.3		ns
t _{EABWCREG}	9.8		11.6		16.6		ns
t _{EABDD}		9.2		11.0		16.1	ns
t _{EABDATA} CO		1.7		2.1		3.4	ns
t _{EABDATASU}	2.3		2.7		3.5		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	3.3		3.9		4.9		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	3.2		3.8		5.0		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	3.7		4.4		5.1		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		6.1		7.3	_	11.3	ns

Symbol	-1 Speed Grade		-2 Spee	d Grade	-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		3.9		4.4		5.1	ns
t _{DIN2LE}		1.2		1.5		1.9	ns
t _{DIN2DATA}		3.2		3.6		4.5	ns
t _{DCLK2IOE}		3.0		3.5		4.6	ns
t _{DCLK2LE}		1.2		1.5		1.9	ns
t _{SAMELAB}		0.1		0.1		0.2	ns
t _{SAMEROW}		2.3		2.4		2.7	ns
t _{SAME} COLUMN		1.3		1.4		1.9	ns
t _{DIFFROW}		3.6		3.8		4.6	ns
t _{TWOROWS}		5.9		6.2		7.3	ns
t _{LEPERIPH}		3.5		3.8		4.1	ns
t _{LABCARRY}		0.3		0.4		0.5	ns
t _{LABCASC}		0.9		1.1		1.4	ns

Table 97. EPF10K30A External Reference Timing Parameters Note (1)									
Symbol	bol -1 Speed Grade		-2 Spec	ed Grade	-3 Spee	Unit			
	Min	Max	Min	Max	Min	Max			
t _{DRR}		11.0		13.0		17.0	ns		
t _{INSU} (2), (3)	2.5		3.1		3.9		ns		
t _{INH} (3)	0.0		0.0		0.0		ns		
t _{оитсо} (3)	2.0	5.4	2.0	6.2	2.0	8.3	ns		

Table 98. EPF10K30A Device External Bidirectional Timing Parameters Note (1)									
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit		
	Min	Max	Min	Max	Min	Max			
t _{INSUBIDIR}	4.2		4.9		6.8		ns		
t _{INHBIDIR}	0.0		0.0		0.0		ns		
toutcobidir	2.0	5.4	2.0	6.2	2.0	8.3	ns		
t _{XZBIDIR}		6.2		7.5		9.8	ns		
t _{ZXBIDIR}		6.2		7.5		9.8	ns		

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 99 through 105 show EPF10K100A device internal and external timing parameters.

Table 99. EPF10	K100A Device	e LE Timing N	<i>Nicroparame</i>	ters Note	e (1)		
Symbol	-1 Spee	d Grade	-2 Spe	ed Grade	-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		1.0		1.2		1.4	ns
t _{CLUT}		0.8		0.9		1.1	ns
t _{RLUT}		1.4		1.6		1.9	ns
t _{PACKED}		0.4		0.5		0.5	ns
t _{EN}		0.6		0.7		0.8	ns
t_{CICO}		0.2		0.2		0.3	ns
t _{CGEN}		0.4		0.4		0.6	ns
t _{CGENR}		0.6		0.7		0.8	ns
t _{CASC}		0.7		0.9		1.0	ns
t_C		0.9		1.0		1.2	ns
$t_{\rm CO}$		0.2		0.3		0.3	ns
t _{COMB}		0.6		0.7		0.8	ns
t_{SU}	0.8		1.0		1.2		ns
t _H	0.3		0.5		0.5		ns
t _{PRE}		0.3		0.3		0.4	ns
t _{CLR}		0.3		0.3		0.4	ns
t _{CH}	2.5		3.5		4.0		ns
t_{CL}	2.5		3.5		4.0		ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		2.5		2.9		3.4	ns
t_{IOC}		0.3		0.3		0.4	ns
t _{IOCO}		0.2		0.2		0.3	ns
t_{IOCOMB}		0.5		0.6		0.7	ns
t _{IOSU}	1.3		1.7		1.8		ns
t_{IOH}	0.2		0.2		0.3		ns
t _{IOCLR}		1.0		1.2		1.4	ns
t_{OD1}		2.2		2.6		3.0	ns
t_{OD2}		4.5		5.3		6.1	ns
t_{OD3}		6.8		7.9		9.3	ns
t_{XZ}		2.7		3.1		3.7	ns
t_{ZX1}		2.7		3.1		3.7	ns
t _{ZX2}		5.0		5.8		6.8	ns
t_{ZX3}		7.3		8.4		10.0	ns
t _{INREG}		5.3		6.1		7.2	ns
t _{IOFD}		4.7		5.5		6.4	ns
t _{INCOMB}		4.7		5.5		6.4	ns

Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.8		2.1		2.4	ns
t _{EABDATA2}		3.2		3.7		4.4	ns
t _{EABWE1}		0.8		0.9		1.1	ns
t _{EABWE2}		2.3		2.7		3.1	ns
t _{EABCLK}		0.8		0.9		1.1	ns
t _{EABCO}		1.0		1.1		1.4	ns
t _{EABBYPASS}		0.3		0.3		0.4	ns
t _{EABSU}	1.3		1.5		1.8		ns
t _{EABH}	0.4		0.5		0.5		ns
t_{AA}		4.1		4.8		5.6	ns
t_{WP}	3.2		3.7		4.4		ns
t _{WDSU}	2.4		2.8		3.3		ns
t_{WDH}	0.2		0.2		0.3		ns
t _{WASU}	0.2		0.2		0.3		ns
t _{WAH}	0.0		0.0		0.0		ns
t_{WO}		3.4		3.9		4.6	ns
t_{DD}		3.4		3.9		4.6	ns
t _{EABOUT}		0.3		0.3		0.4	ns
t _{EABCH}	2.5		3.5		4.0		ns
t _{EABCL}	3.2		3.7		4.4		ns

Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		6.8		7.8		9.2	ns
t _{EABRCCOMB}	6.8		7.8		9.2		ns
t _{EABRCREG}	5.4		6.2		7.4		ns
t _{EABWP}	3.2		3.7		4.4		ns
t _{EABWCCOMB}	3.4		3.9		4.7		ns
t _{EABWCREG}	9.4		10.8		12.8		ns
t _{EABDD}		6.1		6.9		8.2	ns
t _{EABDATACO}		2.1		2.3		2.9	ns
t _{EABDATASU}	3.7		4.3		5.1		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	2.8		3.3		3.8		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	3.4		4.0		4.6		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	1.9		2.3		2.6		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t_{EABWO}		5.1		5.7		6.9	ns

Symbol	-1 Speed Grade		-2 Spee	d Grade	-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		4.8		5.4		6.0	ns
t _{DIN2LE}		2.0		2.4		2.7	ns
t _{DIN2DATA}		2.4		2.7		2.9	ns
t _{DCLK2IOE}		2.6		3.0		3.5	ns
t _{DCLK2LE}		2.0		2.4		2.7	ns
t _{SAMELAB}		0.1		0.1		0.1	ns
t _{SAMEROW}		1.5		1.7		1.9	ns
t _{SAME} COLUMN		5.5		6.5		7.4	ns
t _{DIFFROW}		7.0		8.2		9.3	ns
t _{TWOROWS}		8.5		9.9		11.2	ns
t _{LEPERIPH}		3.9		4.2		4.5	ns
t _{LABCARRY}		0.2		0.2		0.3	ns
t _{LABCASC}		0.4		0.5		0.6	ns

Table 104. EPF1	IOK100A Devi	ce External Ti	iming Param	eters Not	te (1)		
Symbol	-1 Spec	ed Grade	-2 Speed Grade -3 Speed Grade			d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{DRR}		12.5		14.5		17.0	ns
t _{INSU} (2), (3)	3.7		4.5		5.1		ns
t _{INH} (3)	0.0		0.0		0.0		ns
t _{outco} (3)	2.0	5.3	2.0	6.1	2.0	7.2	ns

Table 105. EPF10K10	OOA Device Ext	ernal Bidirec	tional Timin	g Parameters	Note (1))	
Symbol	-1 Spee	d Grade	-2 Spec	ed Grade	-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR}	4.9		5.8		6.8		ns
t _{INHBIDIR}	0.0		0.0		0.0		ns
toutcobidir	2.0	5.3	2.0	6.1	2.0	7.2	ns
t _{XZBIDIR}		7.4		8.6		10.1	ns
t _{ZXBIDIR}		7.4		8.6		10.1	ns

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 38 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

Tables 106 through 112 show EPF10K250A device internal and external timing parameters.

Table 106. EPF1	OK250A Devi	ce LE Timing	Microparam	eters No	ete (1)		
Symbol	-1 Spee	ed Grade	-2 Spee	ed Grade	-3 Spee	ed Grade	Unit
	Min	Max	Min	Max	Min	Max	
t_{LUT}		0.9		1.0		1.4	ns
t_{CLUT}		1.2		1.3		1.6	ns
t_{RLUT}		2.0		2.3		2.7	ns
t _{PACKED}		0.4		0.4		0.5	ns
t _{EN}		1.4		1.6		1.9	ns
t_{CICO}		0.2		0.3		0.3	ns
t _{CGEN}		0.4		0.6		0.6	ns
t _{CGENR}		0.8		1.0		1.1	ns
t _{CASC}		0.7		0.8		1.0	ns
t_C		1.2		1.3		1.6	ns
$t_{\rm CO}$		0.6		0.7		0.9	ns
t _{COMB}		0.5		0.6		0.7	ns
t _{SU}	1.2		1.4		1.7		ns
t _H	1.2		1.3		1.6		ns
t _{PRE}		0.7		0.8		0.9	ns
t _{CLR}		0.7		0.8		0.9	ns
t _{CH}	2.5		3.0		3.5		ns
t_{CL}	2.5		3.0		3.5		ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t_{IOD}		1.2		1.3		1.6	ns
t_{IOC}		0.4		0.4		0.5	ns
t_{IOCO}		0.8		0.9		1.1	ns
t_{IOCOMB}		0.7		0.7		0.8	ns
t_{IOSU}	2.7		3.1		3.6		ns
t_{IOH}	0.2		0.3		0.3		ns
t _{IOCLR}		1.2		1.3		1.6	ns
t _{OD1}		3.2		3.6		4.2	ns
t_{OD2}		5.9		6.7		7.8	ns
t_{OD3}		8.7		9.8		11.5	ns
t_{XZ}		3.8		4.3		5.0	ns
t_{ZX1}		3.8		4.3		5.0	ns
t_{ZX2}		6.5		7.4		8.6	ns
t_{ZX3}		9.3		10.5		12.3	ns
t _{INREG}		8.2		9.3		10.9	ns
t _{IOFD}		9.0		10.2		12.0	ns
t _{INCOMB}		9.0	_	10.2		12.0	ns

Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Spee	d Grade	Unit
	Min	Max	Min	Max	Min	Max	
t _{EABDATA1}		1.3		1.5		1.7	ns
t _{EABDATA2}		1.3		1.5		1.7	ns
t _{EABWE1}		0.9		1.1		1.3	ns
t _{EABWE2}		5.0		5.7		6.7	ns
t _{EABCLK}		0.6		0.7		0.8	ns
t _{EABCO}		0.0		0.0		0.0	ns
t _{EABBYPASS}		0.1		0.1		0.2	ns
t _{EABSU}	3.8		4.3		5.0		ns
t _{EABH}	0.7		0.8		0.9		ns
t_{AA}		4.5		5.0		5.9	ns
t_{WP}	5.6		6.4		7.5		ns
t _{WDSU}	1.3		1.4		1.7		ns
t _{WDH}	0.1		0.1		0.2		ns
t _{WASU}	0.1		0.1		0.2		ns
t _{WAH}	0.1		0.1		0.2		ns
t_{WO}		4.1		4.6		5.5	ns
t_{DD}		4.1		4.6		5.5	ns
t _{EABOUT}		0.1		0.1		0.2	ns
t _{EABCH}	2.5		3.0		3.5		ns
t _{EABCL}	5.6		6.4		7.5		ns

Symbol	-1 Spee	d Grade	-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{EABAA}		6.1		6.8		8.2	ns
t _{EABRCCOMB}	6.1		6.8		8.2		ns
t _{EABRCREG}	4.6		5.1		6.1		ns
t _{EABWP}	5.6		6.4		7.5		ns
t _{EABWCCOMB}	5.8		6.6		7.9		ns
t _{EABWCREG}	15.8		17.8		21.0		ns
t _{EABDD}		5.7		6.4		7.8	ns
t _{EABDATA} CO		0.7		0.8		1.0	ns
t _{EABDATASU}	4.5		5.1		5.9		ns
t _{EABDATAH}	0.0		0.0		0.0		ns
t _{EABWESU}	8.2		9.3		10.9		ns
t _{EABWEH}	0.0		0.0		0.0		ns
t _{EABWDSU}	1.7		1.8		2.1		ns
t _{EABWDH}	0.0		0.0		0.0		ns
t _{EABWASU}	0.9		0.9		1.0		ns
t _{EABWAH}	0.0		0.0		0.0		ns
t _{EABWO}		5.3		6.0		7.4	ns

Symbol	-1 Speed Grade		-2 Speed Grade		-3 Spee	Unit	
	Min	Max	Min	Max	Min	Max	
t _{DIN2IOE}		7.8		8.5		9.4	ns
t _{DIN2LE}		2.7		3.1		3.5	ns
t _{DIN2DATA}		1.6		1.6		1.7	ns
t _{DCLK2IOE}		3.6		4.0		4.6	ns
t _{DCLK2LE}		2.7		3.1		3.5	ns
t _{SAMELAB}		0.2		0.3		0.3	ns
t _{SAMEROW}		6.7		7.3		8.2	ns
t _{SAME} COLUMN		2.5		2.7		3.0	ns
t _{DIFFROW}		9.2		10.0		11.2	ns
t _{TWOROWS}		15.9		17.3		19.4	ns
t _{LEPERIPH}		7.5		8.1		8.9	ns
t _{LABCARRY}		0.3		0.4		0.5	ns
t _{LABCASC}		0.4		0.4		0.5	ns

Table 111. EPF10K250A Device External Reference Timing Parameters Note (1)							
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{DRR}		15.0		17.0		20.0	ns
t _{INSU} (2), (3)	6.9		8.0		9.4		ns
t _{INH} (3)	0.0		0.0		0.0		ns
t _{оитсо} (3)	2.0	8.0	2.0	8.9	2.0	10.4	ns

Table 112. EPF10K250A Device External Bidirectional Timing Parameters Note (1)							
Symbol	-1 Speed Grade		-2 Speed Grade		-3 Speed Grade		Unit
	Min	Max	Min	Max	Min	Max	
t _{INSUBIDIR}	9.3		10.6		12.7		ns
t _{INHBIDIR}	0.0		0.0		0.0		ns
toutcobidir	2.0	8.0	2.0	8.9	2.0	10.4	ns
t _{XZBIDIR}		10.8		12.2		14.2	ns
t _{ZXBIDIR}		10.8		12.2		14.2	ns

Notes to tables:

- (1) All timing parameters are described in Tables 32 through 37 in this data sheet.
- (2) Using an LE to register the signal may provide a lower setup time.
- (3) This parameter is specified by characterization.

ClockLock & ClockBoost Timing Parameters

For the ClockLock and ClockBoost circuitry to function properly, the incoming clock must meet certain requirements. If these specifications are not met, the circuitry may not lock onto the incoming clock, which generates an erroneous clock within the device. The clock generated by the ClockLock and ClockBoost circuitry must also meet certain specifications. If the incoming clock meets these requirements during configuration, the ClockLock and ClockBoost circuitry will lock onto the clock during configuration. The circuit will be ready for use immediately after configuration. Figure 31 illustrates the incoming and generated clock specifications.

Figure 31. Specifications for the Incoming & Generated Clocks

The t_l parameter refers to the nominal input clock period; the t_0 parameter refers to the nominal output clock period.

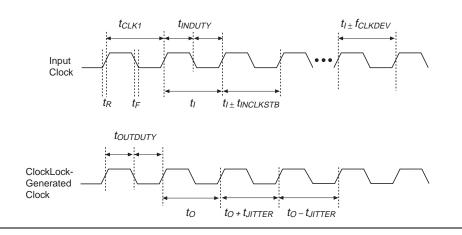


Table 113 summarizes the ClockLock and ClockBoost parameters.

Symbol	Parameter	Min	Тур	Max	Unit
t_R	Input rise time			2	ns
t _F	Input fall time			2	ns
t _{INDUTY}	Input duty cycle	45		55	%
f _{CLK1}	Input clock frequency (ClockBoost clock multiplication factor equals 1)	30		80	MHz
t _{CLK1}	Input clock period (ClockBoost clock multiplication factor equals 1)	12.5		33.3	ns
f _{CLK2}	Input clock frequency (ClockBoost clock multiplication factor equals 2)	16		50	MHz
t _{CLK2}	Input clock period (ClockBoost clock multiplication factor equals 2)	20		62.5	ns

Table 113. ClockLock & ClockBoost Parameters (Part 2 of 2)						
Symbol	Parameter	Min	Тур	Max	Unit	
f _{CLKDEV1}	Input deviation from user specification in MAX+PLUS II (ClockBoost clock multiplication factor equals 1) (1)			±1	MHz	
f _{CLKDEV2}	Input deviation from user specification in MAX+PLUS II (ClockBoost clock multiplication factor equals 2) (1)			±0.5	MHz	
t _{INCLKSTB}	Input clock stability (measured between adjacent clocks)			100	ps	
t _{LOCK}	Time required for ClockLock or ClockBoost to acquire lock (2)		μs			
t _{JITTER}	Jitter on ClockLock or ClockBoost-generated clock (3)			1	ns	
t _{OUTDUTY}	Duty cycle for ClockLock or ClockBoost-generated clock	40	50	60	%	

Notes:

- (1) To implement the ClockLock and ClockBoost circuitry with the MAX+PLUS II software, designers must specify the input frequency. The MAX+PLUS II software tunes the PLL in the ClockLock and ClockBoost circuitry to this frequency. The f_{CLKDEV} parameter specifies how much the incoming clock can differ from the specified frequency during device operation. Simulation does not reflect this parameter.
- (2) During device configuration, the ClockLock and ClockBoost circuitry is configured before the rest of the device. If the incoming clock is supplied during configuration, the ClockLock and ClockBoost circuitry locks during configuration, because the t_{LOCK} value is less than the time required for configuration.
- (3) The t_{IITTER} specification is measured under long-term observation.

Power Consumption

The supply power (P) for FLEX 10K devices can be calculated with the following equation:

$$P = P_{INT} + P_{IO} = (I_{CCSTANDBY} + I_{CCACTIVE}) \times V_{CC} + P_{IO}$$

Typical $I_{CCSTANDBY}$ values are shown as I_{CC0} in the FLEX 10K device DC operating conditions tables on pages 46, 49, and 52 of this data sheet. The $I_{CCACTIVE}$ value depends on the switching frequency and the application logic. This value is calculated based on the amount of current that each LE typically consumes. The P_{IO} value, which depends on the device output load characteristics and switching frequency, can be calculated using the guidelines given in *Application Note* 74 (Evaluating Power for Altera Devices).



Compared to the rest of the device, the embedded array consumes a negligible amount of power. Therefore, the embedded array can be ignored when calculating supply current.

The I_{CCACTIVE} value is calculated with the following equation:

$$I_{CCACTIVE} = K \times f_{MAX} \times N \times tog_{LC} \times \frac{\mu A}{MHz \times LE}$$

The parameters in this equation are shown below:

 f_{MAX} = Maximum operating frequency in MHz

N = Total number of logic cells used in the device

tog_{LC} = Average percent of logic cells toggling at each clock

(typically 12.5%)

K = Constant, shown in Tables 114 and 115

Table 114. FLEX 10K K Constant Values				
Device	K Value			
EPF10K10	82			
EPF10K20	89			
EPF10K30	88			
EPF10K40	92			
EPF10K50	95			
EPF10K70	85			
EPF10K100	88			

Table 115. FLEX 10KA K Constant Values				
Device	K Value			
EPF10K10A	17			
EPF10K30A	17			
EPF10K50V	19			
EPF10K100A	19			
EPF10K130V	22			
EPF10K250A	23			

This calculation provides an I_{CC} estimate based on typical conditions with no output load. The actual I_{CC} should be verified during operation because this measurement is sensitive to the actual pattern in the device and the environmental operating conditions.

To better reflect actual designs, the power model (and the constant *K* in the power calculation equations) for continuous interconnect FLEX devices assumes that logic cells drive FastTrack Interconnect channels. In contrast, the power model of segmented FPGAs assumes that all logic cells drive only one short interconnect segment. This assumption may lead to inaccurate results, compared to measured power consumption for an actual design in a segmented interconnect FPGA.

Figure 32 shows the relationship between the current and operating frequency of FLEX 10K devices.

Figure 32. I_{CCACTIVE} vs. Operating Frequency (Part 1 of 3)

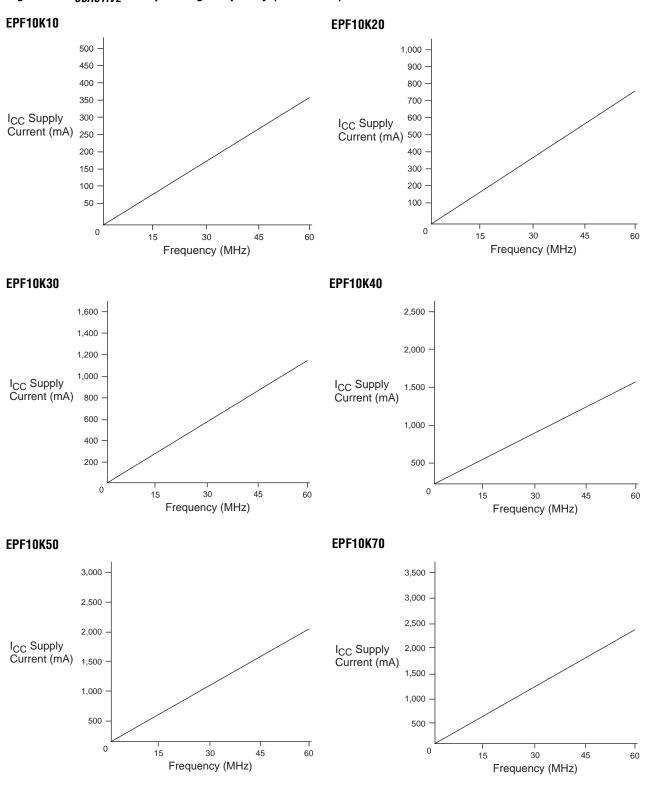
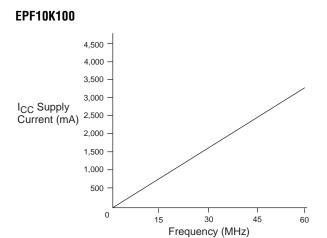
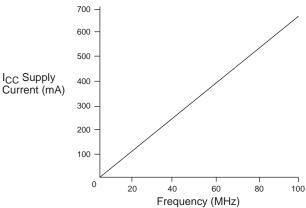


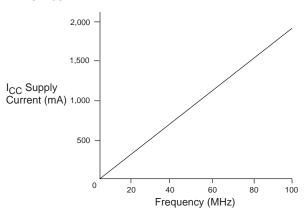
Figure 32. I_{CCACTIVE} vs. Operating Frequency (Part 2 of 3)



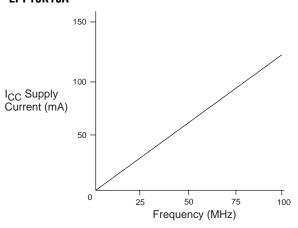




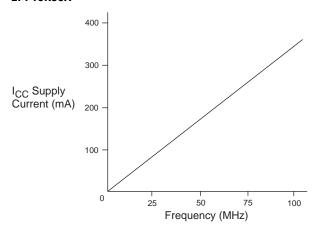
EPF10K130V



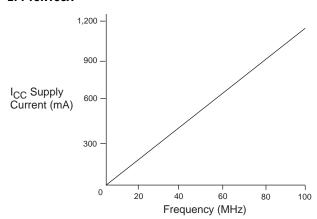
EPF10K10A



EPF10K30A



EPF10K100A



EPF10K250A

3,500 - 3,000 - 2,500 - 2,500 - 1,500 - 1,500 - 1,000 - 1,500 - 1,000 - 1,

Figure 32. I_{CCACTIVE} vs. Operating Frequency (Part 3 of 3)

Configuration & Operation

The FLEX 10K architecture supports several configuration schemes. This section summarizes the device operating modes and available device configuration schemes.



See *Application Note* 116 (*Configuring APEX 20K, FLEX 10K & FLEX 6000 Devices*) for detailed descriptions of device configuration options, device configuration pins, and for information on configuring FLEX 10K devices, including sample schematics, timing diagrams, and configuration parameters.

Operating Modes

The FLEX 10K architecture uses SRAM configuration elements that require configuration data to be loaded every time the circuit powers up. The process of physically loading the SRAM data into the device is called *configuration*. Before configuration, as VCC rises, the device initiates a Power-On Reset (POR). This POR event clears the device and prepares it for configuration. The FLEX 10K POR time does not exceed 50 μs .

During initialization, which occurs immediately after configuration, the device resets registers, enables I/O pins, and begins to operate as a logic device. The I/O pins are tri-stated during power-up, and before and during configuration. Together, the configuration and initialization processes are called *command mode*; normal device operation is called *user mode*.

SRAM configuration elements allow FLEX 10K devices to be reconfigured in-circuit by loading new configuration data into the device. Real-time reconfiguration is performed by forcing the device into command mode with a device pin, loading different configuration data, reinitializing the device, and resuming user-mode operation.

The entire reconfiguration process may be completed in less than 320 ms using an EPF10K250A device with a DCLK frequency of 10 MHz. This process can be used to reconfigure an entire system dynamically. In-field upgrades can be performed by distributing new configuration files.



Refer to the configuration device data sheet to obtain the POR delay when using a configuration device method.

Programming Files

Despite being function- and pin-compatible, FLEX 10KA and FLEX 10KE devices are not programming- or configuration-file compatible with FLEX 10K devices. A design should be recompiled before it is transferred from a FLEX 10K device to an equivalent FLEX 10KA or FLEX 10KE device. This recompilation should be performed to create a new programming or configuration file and to check design timing on the faster FLEX 10KA or FLEX 10KE device. The programming or configuration files for EPF10K50 devices can program or configure an EPF10K50V device. However, Altera recommends recompiling a design for the EPF10K50V device when transferring it from the EPF10K50 device.

Configuration Schemes

The configuration data for a FLEX 10K device can be loaded with one of five configuration schemes (see Table 116), chosen on the basis of the target application. An EPC1, EPC2, EPC16, or EPC1441 configuration device, intelligent controller, or the JTAG port can be used to control the configuration of a FLEX 10K device, allowing automatic configuration on system power-up.

Multiple FLEX 10K devices can be configured in any of the five configuration schemes by connecting the configuration enable (nCE) and configuration enable output (nCEO) pins on each device.

Table 116. Data Sources for Configuration				
Configuration Scheme	Data Source			
Configuration device	EPC1, EPC2, EPC16, or EPC1441 configuration device			
Passive serial (PS)	BitBlaster, MasterBlaster, or ByteBlasterMV download cable, or serial data source			
Passive parallel asynchronous (PPA)	Parallel data source			
Passive parallel synchronous (PPS)	Parallel data source			
JTAG	BitBlaster, MasterBlaster, or ByteBlasterMV download cable, or microprocessor with Jam STAPL file or Jam Byte-Code file			

Device Pin-Outs

See the Altera web site (http://www.altera.com) or the Altera Digital Library for pin-out information.

Revision History

The information contained in the *FLEX 10K Embedded Programmable Logic Device Family Data Sheet* version 4.2 supersedes information published in previous versions.

Version 4.2 Changes

The following change was made to version 4.2 of the *FLEX 10K Embedded Programmable Logic Device Family Data Sheet*: updated Figure 13.

Version 4.1 Changes

The following changes were made to version 4.1 of the FLEX 10K Embedded Programmable Logic Device Family Data Sheet.

- Updated General Description section
- Updated I/O Element section
- Updated SameFrame Pin-Outs section
- Updated Figure 16
- Updated Tables 13 and 116
- Added Note 9 to Table 19
- Added Note 10 to Table 24
- Added Note 10 to Table 28





101 Innovation Drive San Jose, CA 95134 (408) 544-7000 http://www.altera.com Applications Hotline: (800) 800-EPLD **Customer Marketing:** (408) 544-7104 Literature Services: lit_req@altera.com

Copyright © 2003 Altera Corporation. All rights reserved. Altera, The Programmable Solutions Company, the stylized Altera logo, specific device designations, and all other words and logos that are identified as trademarks and/or service marks are, unless noted otherwise, the trademarks and service marks of Altera Corporation in the U.S. and other countries. All other product or service names are the property of their respective holders. Altera products are protected under numerous U.S. and foreign patents and pending applications, maskwork rights, and copyrights. Altera warrants performance of its semiconductor products to

current specifications in accordance with Altera's standard warranty, but reserves the right to make changes to any products and services at any time without notice. Altera assumes no responsibility or liability arising out of the application or use of any information, product, or service described herein except as expressly agreed to in writing by Altera Corporation. Altera customers are advised to obtain the latest version of device specifications before relying on any published information and before placing orders for products or services.

I.S. EN ISO 9001