## Datasheet

The LXT974 and LXT975 are four-port PHY Fast Ethernet Transceivers which support IEEE 802.3 physical layer applications at both 10 Mbps and 100 Mbps . They provide all of the active circuitry to interface four 802.3 Media Independent Interface (MII) compliant controllers to 10BASE-T and/or 100BASE-TX media.

This data sheet applies to all versions of the LXT974 and LXT975 products including LXT974A, LXT974B, LXT975A, and LXT975B. As a result of product changes, Revision 4 parts are labeled LXT974B and LXT975B. Revision 3 parts are labeled LXT974A and LXT975A. The differences in these product revisions are described in the LXT974/975 Specification Update.

All four ports on the LXT974 provide a combination twisted-pair (TP) or pseudo-ECL (PECL) interface for a 10/100BASE-TX or 100BASE-FX connection.

The LXT975 is pin compatible with the LXT974 except for the network ports. The LXT975 is optimized for dual-high stacked RJ-45 modular applications and provides a twisted-pair interface on every port, but the PECL interface on only two.

The LXT974/975 provides three separate LED drivers for each of the four PHY ports and a serial LED interface. In addition to standard Ethernet, each chip supports full- duplex operation at 10 Mbps and 100 Mbps . The LXT974/975 requires only a single 5 V power supply. The MII may be operated independently with either a 3.3 V or 5 V supply.

## Applications

- 10BASE-T, 10/100-TX, or 100BASEFX Switches and multi-port NICs.
- LXT975 optimized for dual-high stacked modular RJ-45 applications.


## Product Features

- Four independent IEEE 802.3compliant 10BASE-T or 100BASE-
TX ports in a single chip.
- 100BASE-FX fiber-optic capable.
- Standard CSMA/CD or full-duplex operation.
- Supports auto-negotiation and legacy systems without auto-negotiation capability.
■ Baseline wander correction.
- 100BASE-TX line performance over 130 meters.
- Configurable LED drivers and serial LED output.
- Configurable through MII serial port or via external control pins.
- Available in 160-pin PQFP with heat spreader.
- Commercial temperature range $\left(0-70^{\circ} \mathrm{C}\right.$ ambient).
- Part numbers:
—LXT974AHC
—LXT974BHC
—LXT975AHC
—LXT975BHC

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The LXT974/LXT975 may contain design defects or errors known as errata which may cause the product to deviate from published specifications. Current characterized errata are available on request.

Contact your local Intel sales office or your distributor to obtain the latest specifications and before placing your product order.
Copies of documents which have an ordering number and are referenced in this document, or other Intel literature may be obtained by calling 1-800-548-4725 or by visiting Intel's website at http://www.intel.com.

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## Revision History

| Revision | Date | Description |
| :---: | :---: | :--- |
| 1.4 | $11 / 00$ | Replace all references to LXT974A and LXT975A with LXT974 and LXT975 <br> (applied to all versions, including A and B) |

Figure 1. LXT974/975 Block Diagram


### 1.0 Pin Assignments and Signal Descriptions

Figure 2. LXT974 Pin Assignments


Table 1. LXT974 Signal Detect/TP Select Signal Descriptions

| Pin\# ${ }^{2}$ | Symbol | Type ${ }^{1}$ | Signal Description |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 158 \\ & 149 \\ & 139 \\ & 130 \end{aligned}$ | SDO/TP0 <br> SD1/TP1 <br> SD2/TP2 <br> SD3/TP3 | 1 | Signal Detect - Ports 0-3. When SD/TPn pins are tied High or to a 5V PECL input, bit $19.2=1$ and the operating mode of each respective port is forced to FX mode. In this mode, full-duplex is set via pin 117 (FDE_FX). When not using FX mode, SD/TPn pins should be tied to GNDT. <br> TP Select-Ports 0-3. When SD/TPn pins are tied Low, bit 19.2 = 0. The operating mode of each port can be set to 10BASE-T, 100BASE-TX, or 100BASE-FX via the hardware control interface pins as shown in Table 8 on page 16. <br> Note: Hardware control interface pins (CFG_0, CFG_1, CFG_2, FDE, BYPSCR, and AUTOENA) are global and set all ports simultaneously. <br> In TP mode, network pins operate as described in Table 2. <br> In FX mode, network pins are re-mapped and operate as described in Table 3. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output. <br> 2. When not using fiber mode, SD/TPn pins should be tied to GNDT. |  |  |  |

Table 2. LXT974 Twisted-Pair Interface Signal Descriptions

| Pin\# | Symbol | Type ${ }^{1}$ | Signal Description |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 154,157 \\ & 145,148 \\ & 135,138 \\ & 126,129 \end{aligned}$ | TPOPO, TPONO <br> TPOP1, TPON1 <br> TPOP2, TPON2 <br> TPOP3, TPON3 | 0 | Twisted-Pair Outputs, Positive \& Negative - Ports 0-3. <br> During 100BASE-TX or 10BASE-T operation, TPO pins drive 802.3 compliant pulses onto the line. |
| $\begin{aligned} & 151,152 \\ & 142,143 \\ & 132,133 \\ & 123,124 \end{aligned}$ | TPIPO, TPINO <br> TPIP1, TPIN1 <br> TPIP2, TPIN2 <br> TPIP3, TPIN3 | 1 | Twisted-Pair Inputs, Positive \& Negative - Ports 0-3. <br> During 100BASE-TX or 10BASE-T operation, TPI pins receive differential 100BASE-TX or 10BASE-T signals from the line. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output. |  |  |  |

Table 3. LXT974 Fiber Interface Signal Descriptions

| Pin\# | Symbol | Type $^{1}$ | Signal Description |
| :---: | :--- | :---: | :--- |
| 154,157 | FIBINO, FIBIP0 |  | Fiber Inputs, Positive \& Negative - Ports 0-3. |
| 145,148 | FIBIN1, FIBIP1 | I | During 100BASE-FX operation, FIBI pins receive differential PECL inputs <br> from fiber transceivers. |
| 135,138 | FIBIN2, FIBIP2 |  |  |
| 126,129 | FIBIN3, FIBIP3 |  |  |
| 151,152 | FIBOP0, FIBON0 |  | Fiber Outputs, Positive \& Negative - Ports 0-3. |
| 142,143 | FIBOP1, FIBON1 | O | During 100BASE-FX operation, FIBO pins produce differential PECL <br> 132, 133 |
| FIBOP2, FIBON2 |  | outputs for fiber transceivers. |  |
| 123,124 | FIBOP3, FIBON3 |  |  |

Figure 3. LXT975 Pin Assignments


Table 4. LXT975 Signal Detect/TP Select Signal Descriptions

| Pin\# ${ }^{2}$ | Symbol | Type $^{1}$ | Signal Description |
| :---: | :---: | :---: | :--- |
| 149 |  | SD1/TP1 |  |
| 130 | SD3/TP3 | Signal Detect - Ports 1 \& 3. When SD/TPn pins are tied High or to a 5V PECL input, bit <br> 19.2 = 1 and the operating mode of each respective port is forced to FX mode. In this <br> mode, full-duplex is set via pin 117 (FDE_FX). When not using fiber mode, SD/TPn pins <br> should be tied to GNDT. <br> TP Select - Ports 1 \& 3. When SD/TPn pins are tied Low, bit 19.2 $=0$. The operating mode <br> of each port can be set to 10BASE-T, 100BASE-TX, or 100BASE-FX via the hardware <br> control interface pins as shown in Table 8 on page 16. <br> Note: Hardware control interface pins (CFG_0, CFG_1, CFG_2, FDE, BYPSCR, and <br> AUTOENA) are global and set all ports simultaneously. <br> In TP mode, network pins operate as described in Table 5. <br> In FX mode, network pins are re-mapped and operate as described in Table 6. |  |
| 1. Type Column Coding: I = Input, O = Output. <br> 2. When not using fiber mode, SD/TPn pins should be tied to GNDT. |  |  |  |

Table 5. LXT975 Twisted-Pair Interface Signal Descriptions

| Pin\# | Symbol | Type ${ }^{1}$ | Signal Description |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 155,152 \\ & 145,148 \\ & 136,133 \\ & 126,129 \end{aligned}$ | TPOPO, TPONO <br> TPOP1, TPON1 <br> TPOP2, TPON2 <br> TPOP3, TPON3 | 0 | Twisted-Pair Outputs, Positive \& Negative - Ports 0-3. <br> During 100BASE-TX or 10BASE-T operation, TPO pins drive 802.3 compliant pulses onto the line. |
| $\begin{aligned} & 158,157 \\ & 142,143 \\ & 139,138 \\ & 133,124 \end{aligned}$ | TPIPO, TPINO <br> TPIP1, TPIN1 <br> TPIP2, TPIN2 <br> TPIP3, TPIN3 | 1 | Twisted-Pair Inputs, Positive \& Negative - Ports 0-3. <br> During 100BASE-TX or 10BASE-T operation, TPI pins receive differential 100BASE-TX or 10BASE-T signals from the line. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output. |  |  |  |

Table 6. LXT975 Fiber Interface Signal Descriptions

| Pin\# | Symbol | Type $^{1}$ | Signal Description |
| :---: | :---: | :---: | :--- |
| $\begin{array}{l}145,148 \\ 126,129\end{array}$ | FIBIN1, FIBIP1 | FIBIN3, FIBIP3 | I | \(\left.\begin{array}{l}Fiber Network Interface - Ports 1 and 3 <br>

During 100BASE-FX operation, FIBI pins receive differential PECL inputs <br>
from fiber transceivers.\end{array}\right]\)

Table 7. LXT974 and LXT975 MII Signal Descriptions

| Pin\# ${ }^{3}$ | Symbol | Type ${ }^{1}$ | Signal Description ${ }^{2}$ |
| :---: | :---: | :---: | :---: |
| MII Data Interface Pins |  |  |  |
| $\begin{aligned} & \hline 33 \\ & 34 \\ & 35 \\ & 36 \end{aligned}$ |  | 1 | Transmit Data - Port 0. Inputs containing NRZ data to be transmitted from port 0. |
| $\begin{aligned} & 52 \\ & 53 \\ & 54 \\ & 55 \end{aligned}$ |  | 1 | Transmit Data-Port 1. Inputs containing NRZ data to be transmitted from port 1. |
| $\begin{aligned} & 71 \\ & 72 \\ & 73 \\ & 74 \end{aligned}$ |  | 1 | Transmit Data-Port 2. Inputs containing NRZ data to be transmitted from port 2. |
| $\begin{aligned} & 89 \\ & 90 \\ & 91 \\ & 92 \end{aligned}$ |  | 1 | Transmit Data - Port 3. Inputs containing NRZ data to be transmitted from port 3. |
| $\begin{aligned} & \hline 32 \\ & 51 \\ & 70 \\ & 88 \end{aligned}$ | $\begin{aligned} & \hline \text { TX_ENO } \\ & \text { TX_EN1 } \\ & \text { TX_EN2 } \\ & \text { TX_EN3 } \end{aligned}$ | 1 | Transmit Enable - Ports 0-3. Active High input enables respective port transmitter. This signal must be synchronous to the TX_CLK. |
| $\begin{aligned} & 31 \\ & 50 \\ & 69 \\ & 87 \end{aligned}$ | $\begin{aligned} & \text { TX_CLKO } \\ & \text { TX_CLK1 } \\ & \text { TX_CK2 } \\ & \text { TX_CLK3 } \end{aligned}$ | 0 | Transmit Clock - Ports 0-3.25 MHz for 100 Mbps operation, 2.5 MHz for 10 Mbps operation. The transmit data and control signals must always be synchronized to TX_CLK by the MAC. The LXT974/975 normally samples these signals on the rising edge of TX_CLK. However, Advanced TX_CLK Mode is available by setting MII register bit 19.5=1. In this mode, the LXT974/975 samples the transmit data and control signals on the falling edge of TX_CLK. |
| $\begin{aligned} & 30 \\ & 49 \\ & 68 \\ & 86 \end{aligned}$ | $\begin{aligned} & \text { TX_ERO } \\ & \text { TX_ER1 } \\ & \text { TX_ER2 } \\ & \text { TX_ER3 } \end{aligned}$ | 1 | Transmit Coding Error - Ports 0-3. This signal must be driven synchronously to TX_CLK. When High, forces the respective port to transmit Halt (H) code group. |
| $\begin{aligned} & 26 \\ & 25 \\ & 24 \\ & 23 \end{aligned}$ |  | 0 | Receive Data - Port 0. Receive data signals (4-bit parallel nibbles) are driven synchronously to RX_CLKO. |
| $\begin{aligned} & 45 \\ & 44 \\ & 43 \\ & 42 \end{aligned}$ |  | 0 | Receive Data - Port 1. Receive data signals (4-bit parallel nibbles) are driven synchronously to RX_CLK1. |
| $\begin{aligned} & 64 \\ & 63 \\ & 62 \\ & 61 \end{aligned}$ |  | 0 | Receive Data - Port 2. Receive data signals (4-bit parallel nibbles) are driven synchronously to RX_CLK2. |
| $\begin{aligned} & 82 \\ & 81 \\ & 80 \\ & 79 \end{aligned}$ |  | 0 | Receive Data - Port 3. Receive data signals (4-bit parallel nibbles) are driven synchronously to RX_CLK3. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output, $\mathrm{OD}=$ Open Drain <br> 2. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ). <br> 3. Unused pins should be tied Low. |  |  |  |

Table 7. LXT974 and LXT975 MII Signal Descriptions (Continued)

| Pin\# ${ }^{3}$ | Symbol | Type ${ }^{1}$ | Signal Description ${ }^{2}$ |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & 27 \\ & 46 \\ & 65 \\ & 83 \end{aligned}$ | RX_DV0 RX_DV1 RX_DV2 RX_DV3 | 0 | Receive Data Valid - Ports 0-3. These signals are synchronous to the respective RX_CLKn. Active High indication that received code group maps to valid data. |
| $\begin{aligned} & 29 \\ & 48 \\ & 67 \\ & 85 \end{aligned}$ |  | 0 | Receive Error - Ports 0-3. These signals are synchronous to the respective RX_CLKn. Active High indicates that received code group is invalid, or that PLL is not locked. |
| $\begin{aligned} & 28 \\ & 47 \\ & 66 \\ & 84 \end{aligned}$ | RX_CLKO RX_CLK1 RX_CLK2 RX_CLK3 | 0 | Receive Clock - Ports 0-3.25 MHz for 100 Mbps and 2.5 MHz for 10 Mbps . |
| $\begin{aligned} & 37 \\ & 57 \\ & 75 \\ & 93 \end{aligned}$ | COLO COL1 COL2 COL3 | 0 | Collision Detected - Ports 0-3. Active High outputs asserted upon detection of a collision. Remain High for the duration of the collision. These signals are generated asynchronously. Inactive during full-duplex operation. |
| $\begin{aligned} & 38 \\ & 58 \\ & 76 \\ & 94 \end{aligned}$ | CRSO CRS1 CRS2 CRS3 | 0 | Carrier Sense-Ports 0-3. Active High signals. During half-duplex operation (bit $0.8=0$ ), CRS $n$ is asserted when either transmit or receive medium is non-idle. During full-duplex operation (bit $0.8=1$ ), CRSn is asserted only when the receive medium is nonidle. |
|  |  |  | MII Control Interface Pins |
| 97 | MDIO | I/O | Management Data Input/Output. Bidirectional serial data channel for PHY/STA communication. |
| 98 | MDINT | OD | Management Data Interrupt. An active Low output on this pin indicates status change. Interrupt is cleared by sequentially reading Register 1, then Register 18. |
| 99 | MDC | 1 | Management Data Clock. Clock for the MDIO serial data channel. Maximum frequency is 2.5 MHz . |
| 100 | MDDIS | 1 | Management Disable. <br> When MDDIS is High, the MDIO is restricted to Read Only and the Hardware Control Interface pins provide continual control of their respective bits. <br> When MDDIS is Low at power up or Reset, the Hardware Control Interface pins control only the initial or "default" values of their respective register bits. After the power-up/reset cycle is complete, bit control reverts to the MDIO serial channel. |
| $\begin{aligned} & 106 \\ & 105 \\ & 104 \\ & 103 \end{aligned}$ |  | 1 | Tristate - Ports 0-3. This bit controls bit 0.10 (Isolate bit). When TRSTE $n$ is High, the respective port isolates itself from the MII Data Interface. <br> When MDDIS is High, TRSTE provides continuous control over bit 0.10. When MDDIS is Low, TRSTE sets the initial (default) value of bit 0.10 at Reset and then bit control reverts back to the MDIO interface. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output, $\mathrm{OD}=$ Open Drain <br> 2. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number $(0-15)$. <br> 3. Unused pins should be tied Low. |  |  |  |

Table 8. LXT974 and LXT975 Hardware Control Interface Signal Descriptions

| Pin\# | Symbol | Type ${ }^{1}$ | Signal Description ${ }^{2}$ |
| :---: | :---: | :---: | :---: |
| 116 | CFG_0 <br> (Global) | 1 | Configuration Control 0. <br> When A/N is enabled, Low to High transition on CFG_0 causes auto-negotiate to restart on all ports and $0.9=1$. <br> When $\mathbf{A} / \mathbf{N}$ is disabled, this input selects operating speed and directly affects bit 0.13. <br> When CFG_0 is High, 100 Mbps is selected and bit $0.13=1$. <br> When CFG_0 is Low, 10 Mbps is selected and bit $0.13=0$. |
| 115 | CFG_1 <br> (Global) | 1 | Configuration Control 1. <br> When $\mathbf{A} / \mathbf{N}$ is enabled, CFG_1 determines operating speed advertisement capabilities in combination with CFG_2 and FDE on all ports. See Table 16 on page 26 for details. When $\mathbf{A} / \mathbf{N}$ is disabled, CFG_1 enables 10 Mbps link test and directly affects bit 19.8. When CFG_1 is High, 10 Mbps link test is disabled and bit $19.8=1$. When CFG_1 is Low, 10 Mbps link test is enabled and bit $19.8=0$. |
| 114 | CFG_2 <br> (Global) | 1 | Configuration Control 2. <br> When $\mathbf{A} / \mathbf{N}$ is enabled, CFG_2 determines operating speed advertisement capabilities in combination with CFG_1 on all ports. See Table 16 on page 26 for details. <br> When A/N is disabled, this input selects either TP or FX interface. When FX interface is selected, the LXT974/975 automatically disables the scrambler. For correct FX operation, 100 Mbps operation must also be selected. <br> Note: It is recommended to set the network interface for each port independently, via the SD/ TPn pins. See Table 1 and Table 4 for Signal Detect / TP Select signal descriptions and operation. <br> When CFG_2 is Low, TP is enabled and bit $19.2=0$. <br> When CFG_2 is High, FX is enabled and bit $19.2=1$. |
| 110 | FDE (Global) | 1 | Full-Duplex Enable - All Ports. <br> When High, enables full-duplex operation on all ports. |
| 117 | FDE_FX | 1 | Full-Duplex Enable - FX Ports only. <br> When High, enables full-duplex operation on all ports set for FX mode operation. This pin is ignored on ports set for TP mode. |
| 113 | BYPSCR <br> (Global) | 1 | Bypass Scrambler. <br> In TP mode, enables or bypasses Scrambler operation and directly affects MDIO register bit 19.3. <br> When High, Scrambler is bypassed and bit 19.3=1. <br> When Low, Scrambler is enabled and bit $19.3=0$. <br> In FX mode, the LXT974/975_ automatically bypasses the Scrambler. This pin has no effect selecting Scrambler bypass. |
| 111 | AUTOENA (Global) | 1 | Auto-Negotiation Enable. When High, enables auto-negotiation on all ports. |
| 1. Type Column Coding: $\mathrm{I}=$ Input, $\mathrm{O}=$ Output, $\mathrm{OD}=$ Open Drain. <br> 2. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ). |  |  |  |

Table 9. LXT974 and LXT975 Miscellaneous Signal Descriptions

| Pin\# | Symbol | Type ${ }^{1}$ | Signal Description ${ }^{2}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & 20 \\ & 19 \\ & 18 \end{aligned}$ | ADD4 ADD3 ADD2 | $\begin{aligned} & \text { I } \\ & \text { \| } \end{aligned}$ | Address <4:2>. Set upper three bits of PHY address. ADD<1:0> are set internally to match port number as shown at right. | ADD1 | ADDO | Port |
|  |  |  |  | 0 | 0 | 0 |
|  |  |  |  | 0 | 1 | 1 |
|  |  |  |  | 1 | 0 | 2 |
|  |  |  |  | 1 | 1 | 3 |
| 101, 112, 159 | TEST | 1 | Test. Must be tied Low. |  |  |  |
| 140 | RBIAS | 1 | Bias. This pin provides bias current for the internal circuitry. Must be tied to ground through a $22 \mathrm{k} \Omega$ resistor. |  |  |  |
| 118 | CLK25M | 1 | Clock Input. A 25 MHz clock input is required at this pin. Refer to Functional Description for detailed clock requirements. |  |  |  |
| 109 | RESET | 1 | Reset. This active Low input is OR'ed with the control register Reset bit (0.15). The LXT974/975 reset cycle is extended $205 \mu$ s (nominal) after Reset is deasserted. |  |  |  |
| 102 | PWRDN | 1 | Power Down. When High, forces LXT974/975 into power down mode. This pin is OR'ed with the Power Down bit (0.11). Refer to Table 44 on page 64 for more information. |  |  |  |
| 41, 119, 120 | N/C | - | No Connection. Leave open. |  |  |  |

1. Type Column Coding: $\mathrm{I}=\mathrm{Input}, \mathrm{O}=$ Output, $\mathrm{A}=$ Analog.
2. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ).

Table 10. LXT974 and LXT975 LED Indicator Signal Descriptions

| Pin\# ${ }^{2}$ | Symbol | Type ${ }^{1}$ | Signal Description ${ }^{3}$ |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & \hline 11 \\ & 8 \\ & 4 \\ & 1 \end{aligned}$ |  | OD | LED0 - Ports 0-3. In default mode, active Low output indicates transmitter active. However, LEDO is programmable and may also be set to indicate receiver active, link status or duplex status. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options. |
| $\begin{gathered} \hline 12 \\ 9 \\ 5 \\ 2 \end{gathered}$ | LEDO_1 LED1_1 LED2_1 LED3_1 | OD | LED1 - Ports 0-3. In default mode, active Low output indicates receiver active. However, LED1 is programmable and may also be set to indicate link status, duplex status, or operating speed. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options. |
| $\begin{aligned} & \hline 13 \\ & 10 \\ & 6 \\ & 3 \end{aligned}$ | $\begin{aligned} & \text { LEDO_2 } \\ & \text { LED1_2 } \\ & \text { LED2_2 } \\ & \text { LED3_2 } \end{aligned}$ | OD | LED2 - Ports 0-3. In default mode, active Low output indicates link up. However, LED2 is programmable and may also be set to indicate duplex status, operating speed or collision. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options. |
| 17 | LEDENA | 0 | LED Enable. Active High output signals external device that LEDDAT is active. |
| 15 | LEDCLK | 0 | LED Clock. 25 MHz clock for LED serial data output. |
| 16 | LEDDAT | 0 | LED Data. Serial data output for 24 LEDs ( $6 \times 4$ ports) data. |
| 1. Type Column Coding: I = Input, $\mathrm{O}=$ Output, $\mathrm{OD}=$ Open Drain. <br> 2. Unused pins should be tied Low. <br> 3. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ). |  |  |  |

Table 11. LXT974 Power Supply Signal Descriptions

| Pin\# | Symbol | Type | Signal Description |
| :--- | :--- | :---: | :--- |
| $22,60,96$ | VCC | - | Power Supply. +5V supply for all digital circuits. |
| 40,78 | VCCMII | - | MII Supply. +3.3 V or +5V supply for MII. A decoupling capacitor <br> to digital ground should be supplied for these pins. |
| $7,14,39,56,59,77,95,160$ | GND | - | Digital Ground. Ground return for digital supply. |
| 21 | GNDA | - | Analog Ground. Ground return for analog supply. |
| 108 | GNDH | - | Ground. Ground return for core analog circuitry. |
| 107 | VCCH | - | Supply. +5V supply for core analog circuitry. |
| $128,137,147,156$ | VCCT | - | Transmit Power Supply. +5V supply for transmit circuits. |
| $127,136,146,155$ | GNDT | - | Transmit Ground. Ground return for transmit supply. |
| $125,134,144,153$, | VCCR | - | Receive Power Supply. +5V supply for all receive circuits. |
| $121,122,131,141,150$ | GNDR | - | Receive Ground. Ground return for receive supply. |

Table 12. LXT975 Power Supply Signal Descriptions

| Pin\# | Symbol | Type | Signal Description |
| :--- | :--- | :---: | :--- |
| $22,60,96$ | VCC | - | Power Supply. +5V supply for all digital circuits. |
| 40,78 | VCCMII | - | MII Supply. +3.3 V or +5 V supply for MII. A decoupling capacitor <br> to digital ground should be supplied for these pins. |
| $7,14,39,56,59,77,95,160$ | GND | - | Digital Ground. Ground return for digital supply. |
| 21 | GNDA | - | Analog Ground. Ground return for analog supply. |
| 108 | GNDH | - | Ground. Ground return for core analog circuitry. |
| 107 | VCCH | - | Supply. +5V supply for core analog circuitry. |
| $128,135,147,154$ | VCCT | - | Transmit Power Supply. +5V supply for transmit circuits. |
| $127,134,146,153$ | GNDT | - | Transmit Ground. Ground return for transmit supply. |
| $125,132,144,151$, | VCCR | - | Receive Power Supply. +5V supply for all receive circuits. |
| $121,122,131,137,141,150,156$ | GNDR | - | Receive Ground. Ground return for receive supply. |

### 2.0 Functional Description

### 2.1 Introduction

The LXT974 and LXT975 are four-port Fast Ethernet 10/100 Transceivers that support 10 Mbps and 100 Mbps networks. They comply with all applicable requirements of IEEE 802.3. Each port can directly drive either a 100BASE-TX line ( $>130$ meters) or a 10BASE-T line ( $>185$ meters). Figure 4 shows the LXT974 in a typical switch application.

Figure 4. LXT974 Switch Application


On power-up, the LXT974/975 uses auto-negotiation/parallel detection on each port to automatically determine line operating conditions. If the PHY device on the other side of the link supports auto-negotiation, the LXT974/975 auto-negotiates with it using Fast Link Pulse (FLP) Bursts. If the PHY partner does not support auto-negotiation, the LXT974/975 automatically detects the presence of either link pulses ( 10 Mbps PHY ) or Idle symbols ( 100 Mbps PHY ) and set its operating conditions accordingly.

The LXT974/975 interfaces to four 10/100 Media Access Controllers (MAC)s through the MII interfaces. It performs all functions of the Physical Coding Sublayer (PCS) and Physical Media Attachment (PMA) sublayer as defined in the IEEE 802.3 100BASE-X specification. This device also performs all functions of the Physical Media Dependent (PMD) sublayer for 100BASE-TX connections. The MII speeds are automatically set once port operating conditions have been determined.

The LXT974/975 provides half-duplex and full-duplex operation at 100 Mbps and 10 Mbps . It also offers standard Loopback Mode for switch applications. The LXT974/975 supports the 802.3 MDIO register set. Specific bits in the registers are referenced using an "X.Y" notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ).

The LXT975 is pin compatible with the LXT974 except for the network ports. Each port on the LXT974 provides a combination twisted-pair or PECL interface for a 10/100BASE-TX or 100BASE-FX connection.

The LXT975 is optimized for stacked RJ-45 modular applications as shown in Figure 5. Ports 1 and 3 support the PECL interface for fiber connections and all four ports support the twisted-pair interface for 10/100BASE-TX connections.

Figure 5. LXT975 Switch Application


### 2.2 Network Media / Protocol Support

The LXT974/975 supports both 10BASE-T and 100BASE-TX Ethernet over twisted-pair, or 100 Mbps Ethernet over fiber media (100BASE-FX). A Media Independent Interface (MII) is used for communication with the Media Access Controller (MAC).

### 2.2.1 $\quad 10 / 100 \mathrm{Mbps}$ Network Interface

Each of the four network interface ports consists of four external pins (two differential signal pairs). The pins are shared between twisted-pair (TP) and fiber. Signal assignments (input or output, positive or negative) vary depending on whether the port is configured for TP or fiber media. Refer to Table 1 through Table 6 for specific pin assignments.

The LXT974/975 output drivers generate either 100BASE-TX, 10BASE-T, or 100BASE-FX output. When not transmitting data, the LXT974/975 generates 802.3-compliant link pulses or idle code. Input signals are decoded either as a 100BASE-TX, 100BASE-FX, or 10BASE-T input, depending on the mode selected. Auto-negotiation/parallel detection or manual control is used to determine the speed of this interface.

### 2.2.1.1 Twisted-Pair Interface

When operating at 100 Mbps , MLT3 symbols are continuously transmitted and received. When not transmitting data, the LXT974/975 generates "IDLE" symbols.

During 10 Mbps operation, Manchester-encoded data is exchanged. When no data is being exchanged, the line is left in an idle state.

In 100 Mbps mode, the LXT974/975 is capable of driving a 100BASE-TX connection over $100 \Omega$, Category 5, Unshielded Twisted Pair (UTP). A 10BASE-T connection can be supported using $100 \Omega$ Category 3, UTP.

Only a transformer (1:1 on receive side, 2:1 on transmit side), load resistors, and bypass capacitors are needed to complete this interface. Using Intel's patented waveshaping technology, the transmitter pre-distorts the outgoing signal to reduce the need for external filters for EMI compliance.

A $4 \mathrm{k} \Omega$ passive load is always present across the twisted-pair inputs. When enabled, the twistedpair inputs are actively biased to approximately 2.8 V .

### 2.2.1.2 Fiber Interface

The LXT974/975 provides a PECL interface that complies with the ANSI X3. 166 specification. This interface is suitable for driving a fiber-optic coupler.

The twisted-pair pin assignments are remapped to support the PECL interface. The LXT974 supports both the twisted-pair and fiber interface on all four ports. The LXT975, optimized for TP operation with dual-high RJ-45 connectors, provides dual interfaces on ports 1 and 3 .

During 100BASE-FX operation, the FIBI pins receive differential PECL signals and the FIBO pins produce differential PECL output signals.

Fiber ports cannot be enabled via auto-negotiation; they must be enabled via the Hardware Control Interface or MDIO registers.

### 2.2.2 MII Interface

The LXT974/975 supports four standard MIIs (one per port). This interface consists of a data interface and a management interface. The MII Data Interface passes data between the LXT974/ 975 and one or more Media Access Controllers (MACs). Separate signals are provided for transmit and receive. This interface operates at either 10 Mbps or 100 Mbps . The speed is set automatically, once the operating conditions of the network link have been determined.

Nine signals are used to pass received data to the MAC: RXD $<3: 0>$, RX_CLK, RX_DV, RX_ER, COL and CRS. Seven signals are used to transmit data from the MAC: TXD<3:0>, TX_CLK, TX_EN, and TX_ER.

### 2.2.2.1 MII Data Interface

Figure 6 shows the data portion of the MII interface. Separate channels are provided for transmitting data from the MAC to the LXT974/975 (TXD), and for receiving data (RXD) from the line.

Each channel has its own clock, data bus, and control signals. The LXT974/975 supplies both clock signals as well as separate outputs for carrier sense and collision. Data transmission across the MII is implemented in 4-bit-wide nibbles.

## Tristating the MII

The LXT974/975 asserts RX_DV, RXD, RX_CLK and RX_ER as soon as it receives a packet from the network. When TRSTE $n$ is High, the associated port output signals are tristated.

Figure 6. MII Data Interface


## Transmit Clock

The LXT974/975 is the master clock source for data transmission. The LXT974/975 automatically sets the speed of TX_CLK to match port conditions. If the port is operating at 100 Mbps , TX_CLK is set to 25 MHz . If the port is operating at 10 Mbps , TX_CLK is set to 2.5 MHz . The transmit data and control signals must always be synchronized to TX_CLK by the MAC. The LXT974/975 normally samples these signals on the rising edge of TX_CLK.

However, Advanced TX_CLK Mode is available by setting MII register bit 19.5=1. In this mode, the LXT974/975 samples the transmit data and control signals on the falling edge of TX_CLK. When operating under MDIO Control, the user can advance the transmit clock relative to TXD $<3: 0>$ and TX_ER. When Advance TX_CLK Mode is selected, the LXT974/975 clocks TXD data in on the falling edge of TX_CLK, instead of the rising edge. This mode provides an increase in timing margins of TXD, relative to TX_CLK. Advance TX_CLK Mode is enabled when bit 19.5 $=1$.

## Transmit Enable

The MAC must assert TX_EN the same time as the first nibble of preamble, and de-assert TX_EN after the last bit of the packet.

## Receive Data Valid

The LXT974/975 asserts RX_DV when it receives a valid packet. Timing changes depend on line operating speed:

- For 100TX and 100FX links, RX_DV is asserted from the first nibble of preamble to the last nibble of the data packet.
- For 10BT links, the entire preamble is truncated. RX_DV is asserted with the first nibble of the Start of Frame Delimiter (SFD) " 5 D " and remains asserted until the end of the packet.


## Error Signals

Whenever the LXT974/975 receives an errored symbol from the network, it asserts RX_ER and drives " 1110 " on the RXD pins.

When the MAC asserts TX_ER, the LXT974/975 drives "H" symbols out on the line.

## Carrier Sense

Carrier sense (CRS) is an asynchronous output. It is always generated when a packet is received from the line and in some modes when a packet is transmitted.

On transmit, CRS is asserted on a 10 Mbps or 100 Mbps half-duplex link. Carrier sense is not generated on transmit when the link is operating in full-duplex mode.

Usage of CRS for Interframe Gap (IFG) timing is not recommended for the following reasons:

- De-assertion time for CRS is slightly longer than assertion time. This causes IFG intervals to appear somewhat shorter to the MAC than it actually is on the wire.
- CRS de-assertion is not aligned with TX_EN de-assertion on transmit loopbacks in halfduplex mode.


## Operational Loopback

Operational loopback is provided for 10 Mbps half-duplex links when bit $19.11=0$. Data transmitted by the MAC is looped back on the receive side of the MII. Operational loopback is not provided for 100 Mbps links, full-duplex links, or when $19.11=1$.

## Test Loopback

A test loopback function is provided for diagnostic testing of the LXT974/LXT975. During test loopback, twisted-pair and fiber interfaces are disabled. Data transmitted by the MAC is internally looped back by the LXT974/975 and returned to the MAC.

Test loopback is available for $100 \mathrm{TX}, 100 \mathrm{FX}$, and 10 T operation. Test loopback is enabled by setting bit $0.14=1$, bit $0.8=1$ (full-duplex), and bit $0.12=0$ (disable auto-negotiation). The desired mode of operation for test loopback is set using bits 0.13 and 19.2 as shown in Table 13. Loopback paths for the three modes of operation are shown in Figure 7.

Table 13. Test Loopback Operation

| Mode of Operation | Bit |  |
| :--- | :---: | :---: |
|  | 19.2 | 0.13 |
| 10T Test Loopback | 0 | 0 |
| 100TX Test Loopback | 0 | 1 |
| 100 FX Test Loopback | 1 | 1 |
| 1. Bit $0.14=1$, bit $0.8=1$, and $0.12=0$ must also be set to enable Test Loopback. |  |  |

Figure 7. Loopback Paths


## Collision

The LXT974/975 asserts its collision signal, asynchronously to any clock, whenever the line state is half-duplex and the transmitter and receiver are active at the same time. Table 14 summarizes the conditions for assertion of carrier sense, collision, and data loopback signals.

Table 14. Carrier Sense, Loopback, and Collision Conditions

| Speed \& Duplex Condition | Carrier Sense | Loopback | Collision |
| :--- | :---: | :---: | :---: |
| Full-Duplex at 10 Mbps or 100 Mbps | Receive Only | None | None |
| 100 Mbps, Half-Duplex | Transmit or Receive | None | Transmit and Receive |
| 10 Mbps , Half-Duplex, $19.11=0$ | Transmit or Receive | Yes | Transmit and Receive |
| 10 Mbps , Half-Duplex, $19.11=1$ | Transmit or Receive | None | Transmit and Receive |

### 2.2.2.2 MII Management Interface

The LXT974/975 supports the IEEE 802.3 MII Management Interface also known as the Management Data Input/Output (MDIO) Interface. This interface allows upper-layer devices to monitor and control the state of the LXT974/975. The MDIO interface consists of a physical connection, a specific protocol that runs across the connection, and an internal set of addressable registers. Some registers are required and their functions are defined by the IEEE 802.3 specification. Additional registers are allowed for expanded functionality. The LXT974/975 is configured with both sets of registers.

The physical interface consists of a data line (MDIO) and clock line (MDC). Operation of this interface is controlled by the MDDIS input pin. When MDDIS is High, the MDIO operates as a read-only interface. When MDDIS is Low, both read and write are enabled. The timing for the MDIO Interface is shown in Table 40 on page 61. The protocol is shown in Figure 8 and Figure 9 (read and write). The protocol allows one controller to communicate with up to eight LXT974/975 chips. Bits A4:2 of the 5-bit PHY address are assigned as the LXT974/975 address. Bits A1:0 are assigned as port addresses 0 through 3. The LXT974/975 supports 12 internal registers per port (48 total), each of which is 16 bits wide.

Figure 8. Management Interface - Read Frame Structure


Figure 9. Management Interface - Write Frame Structure


MII Interrupts
The LXT974/975 provides interrupt signals in two ways. The MDIO interrupt reflects the interrupt status of each port addressed by the read. Details are shown in Figure 10.

Setting bit $17.1=1$ on all four ports, enables global interrupts using the MDINT pin. An active Low on this pin indicates a status change on the LXT974/975. Interrupts may be caused by:

- Link status change
- Auto-negotiation complete
- Full-duplex status change
- Jabber detect

Figure 10. MDIO Interrupt Signaling


### 2.2.3 Hardware Control Interface

The Hardware Control Interface is used to configure operating characteristics of the LXT974/975. When MDDIS is Low, this interface provides initial values for the MDIO registers, and then passes control to the MDIO Interface. When MDDIS is High, this interface provides continuous control over the LXT974/975.

Individual chip addressing allows multiple LXT974/975 devices to share the MII in either mode. Table 15 through Table 17 show how to set up the desired operating configurations using the Hardware Control Interface.

Table 15. Configuring the LXT974/975 via Hardware Control

| Desired Configuration | Pin Name | Input Value | MDIO Registers |
| :--- | :---: | :---: | :---: |
| Auto-Negotiation Enabled on all ports ${ }^{1,2,3}$ | AUTOENA | High | $0.12=1$ |
|  | SD/TP $n$ | Low | $19.2=0$ |
| Auto-Negotiation Disabled on all ports ${ }^{4}$ | AUTOENA | Low | $0.12=0$ |
| Scrambler Bypassed on all ports | BYPSCR | High | $19.3=1$ |
| Scrambler Enabled on all ports | BYPSCR | Low | $19.3=0$ |

1. SD/TP $n$ must be set Low for Auto-Negotiation operation.
2. Refer to Table 16 for Hardware Control Interface functions advertised when auto-negotiation is enabled.
3. Fiber operation can be forced per port via SD/TPn pins when auto-negotiation is enabled. See Table 17 for details.
4. Refer to Table 17 for Hardware Control Interface functions available when auto-negotiation is disabled.

Table 16. Configuring LXT974/975 Auto-Negotiation Advertisements Via Hardware Control

| Desired Configuration ${ }^{1,2}$ | Pin Settings |  |  |  |  | MDIO Registers |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SD/TPn (per port) | FDE (global) | CFG_2 <br> (global) | CFG_1 <br> (global) | CFG_0 ${ }^{3}$ <br> (global) | 4.5 | 4.6 | 4.7 | 4.8 |
| Advertise All | Low | Ignore | Low | Low | Ignore | 1 | 1 | 1 | 1 |
| Advertise 100 HD | Low | Low | High | Low | Ignore | 0 | 0 | 1 | 0 |
| Advertise 100 HD/FD | Low | High | High | Low | Ignore | 0 | 0 | 1 | 1 |
| Advertise 10 HD | Low | Low | Low | High | Ignore | 1 | 0 | 0 | 0 |
| Advertise 10 HD/FD | Low | High | Low | High | Ignore | 1 | 1 | 0 | 0 |
| Advertise 10/100 HD | Low | Low | High | High | Ignore | 1 | 0 | 1 | 0 |

1. Refer to Table 15 for basic configurations.
2. Refer to Table 17 for Hardware Control Interface functions available when auto-negotiation is disabled.
3. Auto-Negotiation is not affected by CFG_0.

Table 17. Configuring the LXT974/975 with Auto-Negotiation Disabled

| Desired Configuration ${ }^{1,2}$ | Pin Settings |  |  |  |  | MDIO Registers |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | SD/TPn per port | CFG 2 global | CFG_0 global | FDE global | FDE_FX | 0.8 | 0.13 | 19.2 |
| Per Port (Fiber) Configuration |  |  |  |  |  |  |  |  |
| Fiber operation can be forced per port via SD/TPn pins when auto-negotiation is enabled. Per-port settings override the global pin settings. |  |  |  |  |  |  |  |  |
| 100FX Full-Duplex Operation. | High or PECL ${ }^{3}$ | Ignored | Ignored | Ignored | High | 1 | 1 | 1 |
| 100FX Half-Duplex Operation. | High or PECL ${ }^{3}$ | Ignored | Ignored | Ignored | Low | 0 | 1 | 1 |
| Global (Twisted-Pair) Configuration ${ }^{5}$ |  |  |  |  |  |  |  |  |
| Force 100TX Full-Duplex Operation on all ports. ${ }^{4}$ | Low | Low | High | High | Ignored | 1 | 1 | 0 |
| Force 100TX Half-Duplex Operation on all ports. ${ }^{4}$ | Low | Low | High | Low | Ignored | 0 | 1 | 0 |
| Force 10T Full-Duplex Operation on all ports. | Low | Low | Low | High | Ignored | 1 | 0 | 0 |
| Force 10T Half-Duplex Operation on all ports. | Low | Low | Low | Low | Ignored | 0 | 0 | 0 |
| 1. Refer to Table 15 for basic configurations. <br> 2. Refer to Table 16 for Hardware Control Interface functions advertised when auto-negotiation is enabled. <br> 3. When SD/TP $n$ is set High or to PECL levels, auto-negotiation is disabled and FDE_FX determines the duplex mode of the port. <br> 4. CFG_2, CFG_0, and SD/TPn must all be set for 100TX operation. <br> 5. Fiber configuration must be selected on a per-port basis. |  |  |  |  |  |  |  |  |

### 2.3 Initialization

At power-up or reset, the LXT974/975 performs the initialization as shown in Figure 11. Control mode selection is provided via the MDDIS pin as shown in Table 18. When MDDIS (pin 100) is High, the LXT974/975 operates in Manual Control Mode. When MDDIS is Low, the LXT974/975 operates in MDIO Control Mode.

### 2.3.1 MDIO Control Mode

In the MDIO Control Mode, the LXT974/975 uses the Hardware Control Interface to set up initial (default) values of the MDIO registers. The MDIO Register set for the LXT974/975 is described in Table 44 through Table 55. Specific bits in the registers are referenced using an "X. Y " notation, where X is the register number ( $0-6$ or $16-20$ ) and Y is the bit number ( $0-15$ ). Once initial values are set, bit control reverts to the MDIO interface.

### 2.3.2 Manual Control Mode

In the Manual Control Mode, LXT974/975 disables direct write operations to the MDIO registers via the MDIO interface. The Hardware Control Interface is continuously monitored and the MDIO registers are updated accordingly.

### 2.3.3 Link Configuration

When the LXT974/975 is first powered on, reset, or encounters a link failure state, it must determine the line speed and operating conditions to use for the network link.

The LXT974/975 first checks the Hardware Control Interface pins and MDIO registers. Using these mechanisms, the user can command the LXT974/975 to do one of the following:

- Force network link to 100FX (Fiber).
- Force network link operation to:

100TX, Full-Duplex
100TX, Half-Duplex
10BASE-T, Full-Duplex
10BASE-T, Half-Duplex

- Allow auto-negotiation/parallel-detection. The Hardware Control Interface pins are used to set the state of the MDIO advertisement registers.

When forcing the network link, the LXT974/975 immediately begins operating the network interface as commanded. When auto-negotiation is enabled, the auto-negotiation / paralleldetection operation begins.

Table 18. Mode Control Settings

| Mode | MDDIS <br> Pin 100 | RESET <br> Pin 109 | PWR <br> DWN <br> Pin 102 |
| :---: | :---: | :---: | :---: |
| MDIO Control | Low | High | Low |
| Manual Control | High | High | Low |
| Reset | - | Low | Low |
| Power Down | - | - | High |

Figure 11. Hardware Interface Mode Selection


### 2.4 Auto-Negotiation

The LXT974/975 attempts to auto-negotiate with its counterpart across the link by sending Fast Link Pulse (FLP) bursts. Each burst consists of 33 link pulses spaced $62.5 \mu \mathrm{~s}$ apart. Odd link pulses (clock pulses) are always present. Even link pulses (data pulses) may be present or absent to indicate a " 1 " or a " 0 ". Each FLP burst exchanges 16 bits of data, which are referred to as a "page". All devices that support auto-negotiation must support a "Base Page" as defined in the IEEE 802.3 standard.

By exchanging Base Pages, the LXT974/975 and its link partner communicate their capabilities to each other. Both sides must receive at least three identical base pages for negotiation to proceed. Each side finds the highest common capabilities that both sides support. Both sides then exchange more pages, and finally agree on the operating state of the line.

### 2.4.1 Parallel Detection

In parallel with auto-negotiation, the LXT974/975 also monitors for 10 Mbps Normal Link Pulses (NLP) or 100 Mbps Idle symbols. If either is detected, the device automatically reverts to the corresponding operating mode. Parallel detection allows the LXT974/975 to communicate with devices that do not support auto-negotiation.

### 2.4.2 Controlling Auto-Negotiation

When auto-negotiation is controlled by software, the following steps are recommended:

- After power-up, power-down, or reset, the power- down recovery time, as specified in Table 41 on page 62 , must be exhausted before proceeding.
- Set the auto-negotiation advertisement register bits.
- Enable auto-negotiation by setting MDIO bit $0.12=1$.


### 2.4.3 Monitoring Auto-Negotiation

When auto-negotiation is being monitored, the following apply:

- Bit 20.13 is set to 1 once the link is established.
- Bits 20.12 and 20.11 can be used to determine the link operating conditions (speed and duplex).

Figure 12. LXT974/975 Auto-Negotiation Operation


## 2.5 $\quad 100 \mathrm{Mbps}$ Operation

### 2.5.1 100BASE-X MII Operations

The LXT974/975 encodes and scrambles the data sent by the MAC, and then transmits it using MLT3 signaling. The LXT974/975 descrambles and decodes MLT3 data received from the network.

When the MAC is not actively transmitting data, the LXT974/975 sends out Idle symbols.
The 100BASE-X protocol specifies the use of a 5-bit symbol code on the network media. However, data is normally transmitted across the MII interface in 4-bit nibbles. The LXT974/975 incorporates a 4B/5B encoder/decoder circuit that translates 4-bit nibbles from the MII into 5-bit symbols for the 100BASE-X connection, and translates 5-bit symbols from the 100BASE-X connection into 4 -bit nibbles for the MII. Table 12 shows the data conversion flow from nibbles to symbols. Table 19 on page 32 shows 4B/5B symbol coding (not all symbols are valid).

### 2.5.2 100BASE-X Network Operations

During 100BASE-X operation, the LXT974/975 transmits and receives 5-bit symbols across the network link. Figure 14 shows the structure of a standard frame packet. When the MAC is not actively transmitting data, the LXT974/975 sends out Idle symbols on the line.

In 100TX mode, the LXT974/975 scrambles the data and transmits it to the network using MLT-3 line code. The MLT-3 signals received from the network are descrambled and decoded and sent across the MII to the MAC.

In 100FX mode, the LXT974/975 transmits and receives NRZI signals across the PECL interface. An external 100FX transceiver module is required to complete the fiber connection.

As shown in Figure 14, the MAC starts each transmission with a preamble pattern. As soon as the LXT974/975 detects the start of preamble, it transmits a J/K symbol (Start of Stream Delimiter, SSD) to the network. It then encodes and transmits the rest of the packet, including the balance of the preamble, the Start of Frame Delimiter (SFD), packet data, and CRC. Once the packet ends, the LXT974/975 transmits the T/R symbol End-of-Stream Delimiter (ESD) and then returns to transmitting Idle symbols.

Figure 13. 100BASE-TX Data Flow


1. Four independent MII ports serve four independent Network ports. Network port configurations are independently selectable. MII port speed is set to match the associated Network port.
2. The Scrambler can be bypassed by setting 19.3=1.

Figure 14. 100BASE-TX Frame Structure


Table 19. 4B/5B Coding

| Code Type | $\begin{gathered} \text { 4B Code } \\ 3210 \end{gathered}$ | Name | 5B Code 43210 | Interpretation |
| :---: | :---: | :---: | :---: | :---: |
|  | 0000 | 0 | 11110 | Data 0 |
|  | 0001 | 1 | 01001 | Data 1 |
|  | 0010 | 2 | 10100 | Data 2 |
|  | 0011 | 3 | 10101 | Data 3 |
|  | 0100 | 4 | 01010 | Data 4 |
|  | 0101 | 5 | 01011 | Data 5 |
|  | 0110 | 6 | 01110 | Data 6 |
| DATA | 0111 | 7 | 01111 | Data 7 |
|  | 1000 | 8 | 10010 | Data 8 |
|  | 1001 | 9 | 10011 | Data 9 |
|  | 1010 | A | 10110 | Data A |
|  | 1011 | B | 10111 | Data B |
|  | 1100 | C | 11010 | Data C |
|  | 1101 | D | 11011 | Data D |
|  | 1110 | E | 11100 | Data E |
|  | 1111 | F | 11101 | Data F |
| IDLE | undefined | $\left.\right\|^{1}$ | 11111 | Idle. Used as inter-stream fill code |
|  | 0101 | $J^{2}$ | 11000 | Start-of-Stream Delimiter (SSD), part 1 of 2 |
| CONTROL | 0101 | K ${ }^{2}$ | 10001 | Start-of-Stream Delimiter (SSD), part 2 of 2 |
|  | undefined | $\mathrm{T}^{3}$ | 01101 | End-of-Stream Delimiter (ESD), part 1 of 2 |
|  | undefined | $\mathrm{R}^{3}$ | 00111 | End-of-Stream Delimiter (ESD), part 2 of 2 |
|  | undefined | $\mathrm{H}^{4}$ | 00100 | Transmit Error. Used to force signaling errors |
|  | undefined | Invalid | 00000 | Invalid |
|  | undefined | Invalid | 00001 | Invalid |
|  | undefined | Invalid | 00010 | Invalid |
| INVALID | undefined | Invalid | 00011 | Invalid |
|  | undefined | Invalid | 00101 | Invalid |
|  | undefined | Invalid | 00110 | Invalid |
|  | undefined | Invalid | 01000 | Invalid |
|  | undefined | Invalid | 01100 | Invalid |
|  | undefined | Invalid | 10000 | Invalid |
|  | undefined | Invalid | 11001 | Invalid |

1. The /I/ (Idle) code group is sent continuously between frames.
2. The $/ \mathrm{J} /$ and $/ \mathrm{K} /(\mathrm{SSD})$ code groups are always sent in pairs; /K/ follows $/ \mathrm{J} /$.
3. The /T/ and /R/ (ESD) code groups are always sent in pairs; /R/ follows /T/.
4. An /H/ (Error) code group is used to signal an error condition.

### 2.5.3 100BASE-X Protocol Sublayer Operations

With respect to the 7-layer communications model, the LXT974/975 is a Physical Layer 1 (PHY) device. The LXT974/975 implements the Physical Coding Sublayer (PCS), Physical Medium Attachment (PMA), and Physical Medium Dependent (PMD) sublayers of the reference model defined by the IEEE 802.3u specification. The following paragraphs discuss LXT974/975 operation from the reference model point of view.

### 2.5.4 PCS Sublayer

The Physical Coding Sublayer (PCS) provides the MII interface, as well as the 4B/5B encoding/ decoding function.

For 100TX and 100FX operation, the PCS layer provides IDLE symbols to the PMD-layer line driver as long as TX_EN is de-asserted.

For 10T operation, the PCS layer merely provides a bus interface and serialization/de-serialization function. 10T operation does not use the 4B/5B encoder.

### 2.5.4.1 Preamble Handling

When the MAC asserts TX_EN, the PCS substitutes a /J/K symbol pair, also known as the Start of Stream Delimiter (SSD), for the first two nibbles received across the MII. The PCS layer continues to encode the remaining MII data, following Table 19 on page 32, until TX_EN is de-asserted. It then returns to supplying IDLE symbols to the line driver.

In the receive direction, the PCS layer performs the opposite function, substituting two preamble nibbles for the SSD.

Figure 15. LXT974/975 Protocol Sublayers


### 2.5.4.2 Data Errors

Figure 16 shows normal reception. When the LXT974/975 receives invalid symbols from the line, it asserts RX_ER, as shown in Figure 17.

### 2.5.4.3 Collision Indication

Figure 18 shows normal transmission. The LXT974/975 detects a collision if transmit and receive are active at the same time. As shown in Figure 19, upon detection of a collision, the COL output is asserted and remains asserted for the duration of the collision.

Figure 16. 100BASE-TX Reception with No Errors


Figure 17. 100BASE-TX Reception with Invalid Symbol
$\square$
Figure 18. 100BASE-TX Transmission with No Errors


Figure 19. 100BASE-TX Transmission with Collision


### 2.5.5 PMA Sublayer

### 2.5.5.1 Link

The LXT974/975 supports a Standard link algorithm or Enhanced link algorithm, which can be set via bit 16.1. Link is established when the symbol error rate is less than 64 errors out of 1024 symbols received. Once the link is established:

When standard link algorithm is selected (default, bit $16.1=0$ ), the link goes down when the symbol error rate becomes greater than 64 out of 1024.

When enhanced link algorithm is selected (bit $16.1=1$ ), the link goes down if twelve idle symbols in a row are not received within 1 to 2 ms . This mode makes it more difficult to bring the link down.

In either mode, the LXT974/975 reports link failure via the MII status bits (1.2, 18.15, and 20.13) and interrupt functions. If auto-negotiate is enabled, link failure causes the LXT974/975 to renegotiate.

### 2.5.5.2 Link Failure Override

The LXT974/975 normally transmits 100 Mbps data packets or Idle symbols only if it detects the link is up, and transmits only FLP bursts if the link is not up. Setting bit $19.14=1$ overrides this function, allowing the LXT974/975 to transmit data packets even when the link is down. This feature is provided as a diagnostic tool. Note that auto-negotiation must be disabled to transmit data packets in the absence of link. If auto-negotiation is enabled, the LXT974/975 automatically begins transmitting FLP bursts if the link goes down.

### 2.5.5.3 Carrier Sense (CRS)

For 100TX and 100FX links, a start of stream delimiter or $/ \mathrm{J} / \mathrm{K}$ symbol pair causes assertion of carrier sense (CRS). An end-of-stream delimiter, or /T/R symbol pair causes de-assertion of CRS. The PMA layer also de-asserts CRS if IDLE symbols are received without/T/R; however, in this case RX_ER is asserted for one clock cycle when CRS is de-asserted.

For 10T links, CRS assertion is based on reception of valid preamble, and de-assertion on reception of an end-of-frame (EOF) marker.

### 2.5.6 Twisted-Pair PMD Sublayer

The twisted-pair Physical Medium Dependent (PMD) layer provides the signal scrambling and descrambling, line coding and decoding (MLT-3 for 100TX, Manchester for 10T), as well as receiving, polarity correction, and baseline wander correction functions.

### 2.5.6.1 Scrambler/Descrambler (100TX Only)

The purpose of the scrambler is to spread the signal power spectrum and further reduce EMI using an 11-bit, non-data-dependent polynomial. The receiver automatically decodes the polynomial whenever IDLE symbols are received.

The scrambler/descrambler can be bypassed by either setting bit $19.3=1$ or setting pin (BYPSCR) High. The scrambler is automatically bypassed when the fiber port is enabled. Scramber bypass is provided for diagnostic and test support.

### 2.5.6.2

Polarity Correction
The LXT974/975 automatically detects and corrects for the condition where the receive signal (TPIP/N) is inverted. Reversed polarity is detected if eight inverted link pulses, or four inverted end-of-frame (EOF) markers, are received consecutively. If link pulses or data are not received by the maximum receive time-out period, the polarity state is reset to a non-inverted state.

### 2.5.7 Fiber PMD Sublayer

The LXT974/975 provides a PECL interface for connection to an external fiber-optic transceiver. (The external transceiver provides the PMD function for fiber media.) The LXT974/975 uses an NRZI format for the fiber interface. The fiber interface operates at 100 Mbps and does not support 10FL applications.

### 2.6 10 Mbps Operation

The LXT974/975 operates as a standard 10BASE-T transceiver. Data transmitted by the MAC as 4-bit nibbles is serialized, Manchester-encoded, and transmitted on the TPOP/N outputs. Received data is decoded and de-serialized into 4-bit nibbles. The LXT974/975 supports all the standard 10 Mbps functions.

### 2.6.1 10BASE-T MII Operation

The MAC transmits data to the LXT974/975 via the MII interface. The LXT974/975 converts the digital data from the MAC into an analog waveform that is transmitted to the network via the copper interface. The LXT974/975 converts analog signals received from the network into a digital format suitable for the MAC. The LXT974/975 sends the received data to the MAC via the MII.

### 2.6.2 10BASE-T Network Operations

During 10BASE-T operation, the LXT974/975 transmits and receives Manchester-encoded data across the network link. When the MAC is not actively transmitting data, the LXT974/975 sends out link pulses on the line.

In 10BASE-T mode, the polynomial scrambler/descrambler is inactive. Manchester-encoded signals received from the network are decoded by the LXT974/975 and sent across the MII to the MAC.

The LXT974/975 does not support fiber connections at 10 Mbps .

### 2.6.2.1 Preamble Handling

In 10BASE-T Mode, the LXT974/975 strips the entire preamble off of received packets. CRS is asserted a few bit times after carrier is detected. RX_DV is held Low for the duration of the preamble.

When RX_DV is asserted, the very first two nibbles driven by the LXT974/975 are the SFD "5D" hex followed by the body of the packet. In 10T loopback, the LXT974/975 loops back whatever the MAC transmits to it, including the preamble.

### 2.6.2.2 Link Test

In 10 Mbps mode, the LXT974/975 always transmit link pulses. If the link test function is enabled, it monitors the connection for link pulses. Once link pulses are detected, data transmission are enabled and remain enabled as long as either the link pulses or data transmission continues. If the link pulses stop, the data transmission is disabled.

If the link test function is disabled, the LXT974/975 transmits to the connection regardless of detected link pulses. The link test function can be disabled by setting bit $19.8=1$ or by setting AUTOENA to disable auto-negotiation and setting CFG_1 input High.

### 2.6.2.3 Link Failure

Link failure occurs if Link Test is enabled and link pulses or packets stop being received. If this condition occurs, the LXT974/975 returns to the auto-negotiation phase if auto-negotiation is enabled.

### 2.6.2.4 SQE (Heartbeat)

By default, the SQE (heartbeat) function is disabled on the LXT974/975. To enable this function, set bit $19.10=1$. When this function is enabled, the LXT974/975 asserts its COL output after each transmit packet. See Figure 32 on page 58 for SQE timing parameters.

### 2.6.2.5 Jabber

If MAC transmission exceeds the jabber timer, the LXT974/975 disables the transmit and loopback functions and enables the COL pin. See Figure 33 on page 59 for jabber timing parameters.

The LXT974/975 automatically exits jabber mode after the unjab time has expired. This function can be disabled by setting bit $19.9=1$.

### 2.7 LED Functions

The LXT974/975 provides three programmable LEDs per port. Refer to Table 51 on page 68 for LED programming details. The LXT974/975 also provides a serial LED output.

### 2.7.1 Serial LED Output

The LXT974/975 provides a serial LED interface which should be attached to an external shift register. This interface provides 24 status bits ( $6 \times 4$ ports). Each port reports the following conditions:

- Transmit (T)
$0=$ Transmit active $1=$ Transmit inactive
- Receive (R)
$0=$ Receive active $\quad 1=$ Receive inactive
- Link (L)
$0=$ Link active $\quad 1=$ Link inactive
- Duplex (D)
$0=$ Half-Duplex $\quad 1=$ Full-Duplex
- Speed (S)
$0=100 \mathrm{Mbps} \quad 1=10 \mathrm{Mbps}$
- Collision (C)
$0=$ Collision active $1=$ Collision inactive
LED Data is output on LEDDAT in sets of 24 bits. The serial burst is repeated every 1 ms . A status change in any bit also triggers an immediate serial burst (following the minimum inter-burst gap of $10 \mu \mathrm{~s}$ ). LEDENA is driven High for the duration of the LEDDAT output.


### 2.7.2 Per Port LEDs

The LXT974/975 provides three LED outputs for each port (LEDn_0, LEDn_1 and LED $n \_2$, where $n=$ port number). These outputs can directly drive LEDs to indicate activity and collision status. The active Low "on" times are normally extended for improved LED visibility. The ontime extension can be disabled by setting bit $16.0=1$.

### 2.7.2.1 LEDn 0

In default mode, LED_0 indicates transmitter active. However, LED $n \_0$ is programmable and may also be set to indicate receiver active, link, or full-duplex status. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options.

### 2.7.2.2 LEDn_1

In default mode, LED_1 indicates receiver active. However, LED $n \_1$ is programmable and may also be set to indicate link status, full-duplex status or operating speed. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options.

### 2.7.2.3 LEDn_2

In default mode, active Low output indicates link up. However, LED $n \_2$ is programmable and may also be set to indicate full-duplex status, operating speed or collision. Refer to LED Configuration Register, Table 51 on page 68, for details on programming options.

Table 20. LED-DAT Serial Port Bit Assignments

| Port 0 |  |  |  |  |  | Port 1 |  |  |  |  |  | $\frac{\text { Port } 2}{11: 6}$ | Port 3 |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $23^{1}$ | 22 | 21 | 20 | 19 | 18 | 17 | 16 | 15 | 14 | 13 | 12 |  | 5 | 4 | 3 | 2 | 1 | 0 |
| T | R | L | D | S | C | T | R | L | D | S | C | TRLDSC | T | R | L | D | S | C |

## $2.8 \quad$ Operating Requirements

### 2.8.1 Power Requirements

The LXT974/975 requires four +5 V supply inputs (VCC, VCCR, VCCT, and VCCH). These inputs may be supplied from a single source although decoupling is required to each respective ground. As a matter of good practice, these supplies should be as clean as possible. Typical filtering and decoupling are shown in Figure 22 on page 46.

### 2.8.1.1 MII Power Requirements

An additional supply may be used for the MII (VCCMII). The supply may be either +5 V or +3.3 V . When the MII supply is 3.3 V , MII inputs may not be driven with 5 V levels. VCCMII should be supplied from the same power source used to supply the controller on the other side of the MII interface. Refer to Table 25 on page 51 for MII I/O characteristics.

### 2.8.1.2 Low-Voltage Fault Detect

The LXT974/975 has a low-voltage fault detection function that prevents transmission of invalid symbols when VCC goes below normal operating levels. This function disables the transmit outputs when a low- voltage fault on VCC occurs. If this condition happens, bit 20.2 is set High. Operation is automatically restored when VCC returns to normal. Table 27 on page 51 indicates voltage levels used to detect and clear the low-voltage fault condition.

### 2.8.1.3 Power Down Mode

The LXT974/975 goes into Power Down Mode when PWRDWN is asserted. In this mode, all functions are disabled except the MDIO. The power supply current is significantly reduced. This mode can be used for energy-efficient applications or for redundant applications where there are two devices and one is left as a standby. When the LXT974/975 is returned to normal operation, configuration settings of the MDIO registers are maintained. Refer to Table 23 on page 50 for power down specifications.

### 2.8.2 Clock Requirements

The LXT974/975 requires a constant 25 MHz clock (CLK25M) that must be enabled at all times. Refer to Test Specifications, Table 26 on page 51, for clock timing requirements.

### 3.0 Application Information

### 3.1 Design Recommendations

The LXT974/975 is designed to comply with IEEE requirements and to provide outstanding receive Bit Error Rate (BER) and long-line-length performance. Lab testing has shown that the LXT974/975 can perform well beyond the required distance of 100 meters. To achieve maximum performance from the LXT974/975, attention to detail and good design practices are required. Refer to the LXT974/975 Design and Layout Guide for detailed design and layout information.

### 3.1.1 General Design Guidelines

Adherence to generally accepted design practices is essential to minimize noise levels on power and ground planes. Up to 50 mV of noise is considered acceptable. 50 to 80 mV of noise is considered marginal. High-frequency switching noise can be reduced, and its effects can be eliminated, by following these simple guidelines throughout the design:

- Fill in unused areas of the signal planes with solid copper and attach them with vias to a VCC or ground plane that is not located adjacent to the signal layer.
- Use ample bulk and decoupling capacitors throughout the design (a value of $.01 \mu \mathrm{~F}$ is recommended for decoupling caps).
- Provide ample power and ground planes.
- Provide termination on all high-speed switching signals and clock lines.
- Provide impedance matching on long traces to prevent reflections.
- Route high-speed signals next to a continuous, unbroken ground plane.
- Filter and shield DC-DC converters, oscillators, etc.
- Do not route any digital signals between the LXT974/975 and the RJ-45 connectors at the edge of the board.
- Do not extend any circuit power and ground plane past the center of the magnetics or to the edge of the board. Use this area for chassis ground, or leave it void.


### 3.1.2 Power Supply Filtering

Power supply ripple and digital switching noise on the VCC plane can cause EMI problems and degrade line performance. It is generally difficult to predict in advance the performance of any design, although certain factors greatly increase the risk of having these problems:

- Poorly-regulated or over-burdened power supplies.
- Wide data busses (>32-bits) running at a high clock rate.
- DC-to-DC converters.

Many of these issues can be improved just by following good general design guidelines. In addition, Intel also recommends filtering between the power supply and the analog VCC pins of the LXT974/975. Filtering has two benefits. First, it keeps digital switching noise out of the analog
circuitry inside the LXT974/975, which helps line performance. Second, if the VCC planes are laid out correctly, it keeps digital switching noise away from external connectors, reducing EMI problems.

The recommended implementation is to divide the VCC plane into two sections. The digital section supplies power to the digital VCC pin, MII VCC pin, and to the external components. The analog section supplies power to VCCH, VCCT, and VCCR pins of the LXT974/975. The break between the two planes should run under the device. In designs with more than one LXT974/975, a single continuous analog VCC plane can be used to supply them all.

The digital and analog VCC planes should be joined at one or more points by ferrite beads. The beads should produce at least a $100 \Omega$ impedance at 100 MHz . The beads should be placed so that current flow is evenly distributed. The maximum current rating of the beads should be at least $150 \%$ of the current that is actually expected to flow through them. Each LXT974/975 draws a maximum of 500 mA from the analog supply so beads rated at 750 mA should be used. A bulk cap (2.2-10 $\mu \mathrm{F}$ ) should be placed on each side of each ferrite bead to stop switching noise from traveling through the ferrite.

In addition, a high-frequency bypass cap (. $01 \mu \mathrm{f}$ ) should be placed near each analog VCC pin.

### 3.1.2.1 Ground Noise

The best approach to minimize ground noise is strict use of good general design guidelines and by filtering the VCC plane.

### 3.1.3 Power and Ground Plane Layout Considerations

Great care needs to be taken when laying out the power and ground planes. The following guidelines are recommended:

- Follow the guidelines in the LXT974/975 Layout Guide for locating the split between the digital and analog VCC planes.
- Keep the digital VCC plane away from the TPOP/N and TPIP/N signals, away from the magnetics, and away from the RJ-45 connectors.
- Place the layers so that the TPOP/N and TPIP/N signals can be routed near or next to the ground plane. For EMI reasons, it is more important to shield TPOP/N and TPIP/N.


### 3.1.3.1 Chassis Ground

For ESD reasons, it is a good design practice to create a separate chassis ground that encircles the board and is isolated via moats and keep-out areas from all circuit-ground planes and active signals. Chassis ground should extend from the RJ-45 connectors to the magnetics, and can be used to terminate unused signal pairs ('Bob Smith' termination). In single-point grounding applications, provide a single connection between chassis and circuit grounds with a 2 kV isolation capacitor. In multi-point grounding schemes (chassis and circuit grounds joined at multiple points), provide 2 kV isolation to the Bob Smith termination.

### 3.1.4 MII Terminations

Series termination resistors are not required on the MII signals driven by the LXT974/975.

### 3.1.5 The RBIAS Pin

The LXT974/975 requires a $22 \mathrm{k} \Omega, 1 \%$ resistor directly connected between the RBIAS pin and ground. Place the RBIAS resistor as close to the RBIAS pin as possible. Run an etch directly from the pin to the resistor, and sink the other side of the resistor to a filtered ground. Surround the RBIAS trace with a filtered ground; do not run high-speed signals next to RBIAS.

### 3.1.6 The Twisted-Pair Interface

Because the LXT974/975 transmitter uses 2:1 magnetics, system designers must take extra precautions to minimize parasitic shunt capacitance in order to meet return loss specifications. These steps include:

- Use compensating inductor in the output stage (see Figure 23 on page 47).
- Place the magnetics as close as possible to the LXT974/975.
- Keep transmit pair traces short.
- Route the transmit pair adjacent to a ground plane. The optimum arrangement is to place the transmit traces two to three layers from the ground plane, with no intervening signals.
- Some magnetic vendors are producing magnetics with improved return loss performance. Use of these improved magnetics increases the return loss budget available to the system designer.
- Improve EMI performance by filtering the output center tap. A single ferrite bead may be used to supply center tap current to all 4 ports. All four ports draw a combined total of $\geq 270 \mathrm{~mA}$ so the bead should be rated at $\geq 400 \mathrm{~mA}$.

In addition, follow all the standard guidelines for a twisted-pair interface:

- Route the signal pairs differentially, close together. Allow nothing to come between them.
- Keep distances as short as possible; both traces should have the same length.
- Avoid vias and layer changes as much as possible.
- Keep the transmit and receive pairs apart to avoid cross-talk.
- Put all the components for the transmit network on the front side of the board (same side as the LXT974/975).
- Put entire receive termination network on the back side of the board.
- Bypass common-mode noise to ground on the in-board side of the magnetics using $0.01 \mu \mathrm{~F}$ capacitors.
- Keep termination circuits close together and on the same side of the board.
- Always put termination circuits close to the source end of any circuit.


### 3.1.7 The Fiber Interface

The fiber interface consists of a PECL transmit and receive pair to an external fiber-optic transceiver. The transmit pair should be AC-coupled to the transceiver, and biased to 3.7 V with a $50 \Omega$ equivalent impedance. The receive pair can be DC-coupled, and should be biased to 3.0 V with a $50 \Omega$ equivalent impedance. Figure 24 on page 48 shows the correct bias networks to achieve these requirements.

### 3.2 Magnetics Information

The LXT974/975 requires a 1:1 ratio for the receive transformers and a $2: 1$ ratio for the transmit transformers as shown in Table 21. The transformer isolation voltage should be rated at 2 kV to protect the circuitry from static voltages across the connectors and cables.

Refer to the Magnetic Manufacturers Cross Reference Guide (Application Note 73) for a list of suitable magnetic manufacturers and part numbers. The latest version is located on the Intel web site (http://developer.intel.com/design/network/). Suitable Magnetic part numbers are provided as a reference only. Before committing to a specific component, designers should contact the manufacturer for current product specifications, and validate the magnetics for a specific application.

### 3.2.1 Magnetics With Improved Return Loss Performance

Intel is working with magnetic vendors to develop magnetic modules with improved return loss characteristics. These improved magnetics simplify the design requirements for meeting ANSI X3.263 return loss specifications.

Table 21. Magnetics Requirements

| Parameter | Min | Nom | Max | Units | Test Condition |
| :--- | :---: | :---: | :---: | :---: | :--- |
| Rx turns ratio | - | $1: 1$ | - | - |  |
| Tx turns ratio | - | $2: 1$ | - | - |  |
| Insertion loss | 0.0 | - | 1.1 | dB | 80 MHz |
| Primary inductance | 350 | - | - | $\mu \mathrm{H}$ |  |
| Transformer isolation | - | 2 | - | kV |  |
| Differential to common mode rejection | 40 | - | - | dB | .1 to 60 MHz |
|  | 35 | - | - | dB | 60 to 100 MHz |
| Return Loss - Standard | - | - | -16 | dB | 30 MHz |
|  | - | - | -10 | dB | 80 MHz |
|  |  | - | - | -20 | dB |
| 30 MHz |  |  |  |  |  |

### 3.3 Twisted-Pair/ RJ-45 Interface

Figure 20 shows layout of the LXT974 twisted-pair interface in a single-high RJ-45 modular application. Figure 21 shows layout of the LXT975 twisted-pair interface in a dual-high (stacked) RJ-45 application.

Figure 20. Typical LXT974 Twisted-Pair Single RJ-45 Modular Application


Figure 21. Typical LXT975 Twisted-Pair Stacked RJ-45 Modular Application


Figure 22. LXT974/975 Power and Ground Connections


Figure 23. Typical Twisted-Pair Interface and Supply Filtering


Figure 24. Typical Fiber Interface


Figure 25. Typical MII Interface


### 4.0 Test Specifications

Note: The minimum and maximum values in Table 22 through Table 42 and Figure 26 through Figure 39 represent the performance specifications of the LXT974/975 and are guaranteed by test, except where noted by design. Minimum and maximum values in Table 24 through Table 42 apply over the recommended operating conditions specific in Table 23.

Table 22. Absolute Maximum Ratings

| Parameter |  | Sym | Min | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Supply voltage |  | Vcc | -0.3 | 6 | V |
| Operating temperature | Ambient | TOPA | -15 | +85 | ${ }^{\circ} \mathrm{C}$ |
|  | Case | TOPC | - | +120 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature |  | TsT | -65 | +150 | ${ }^{\circ} \mathrm{C}$ |

Caution: Exceeding these values may cause permanent damage.
Functional operation under these conditions is not implied.
Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 23. Operating Conditions

| Parameter |  | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Recommended supply voltage ${ }^{2}$ | Except MII Supply | Vcc | 4.75 | 5.0 | 5.25 | V |
|  | MII Supply | VccMII | 3.125 | - | 5.25 | V |
| Recommended operating temperature | Ambient | TOPA | 0 | - | 70 | ${ }^{\circ} \mathrm{C}$ |
|  | Case | TOPC | 0 | - | 110 | ${ }^{\circ} \mathrm{C}$ |
| Vcc current | 100BASE-TX | ICC | - | - | 570 | mA |
|  | 100BASE-FX | ICC | - | - | 500 | mA |
|  | 10BASE-T | ICC | - | - | 570 | mA |
|  | Power Down Mode | ICC | - | 0.5 | 3.0 | mA |
|  | Auto-Negotiation | ICC | - | - | 570 | mA |

1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.
2. Voltages with respect to ground unless otherwise specified.

Table 24. Digital I/O Characteristics ${ }^{1}$

| Parameter | Symbol | Min | Typ $^{2}$ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Low voltage $^{3}$ | VIL | - | - | 0.8 | V | - |
| Input High voltage $^{3}$ | VIH | 2.0 | - | - | V | - |
| Input current | II | -100 | - | 100 | $\mu \mathrm{~A}$ | $0.0<\mathrm{VI}<\mathrm{VcC}$ |
| Output Low voltage | VOL | - | - | 0.4 | V | $\mathrm{IOL}=4 \mathrm{~mA}$ |
| Output High voltage | VOH | 2.4 | - | - | V | $\mathrm{IOH}=-4 \mathrm{~mA}$ |

1. Applies to all pins except MII pins. Refer to Table 25 for MII I/O Characteristics.
2. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.
3. Does not apply to CLK25M. Refer to Table 26 for clock input levels.

Table 25. Digital I/O Characteristics - MII Pins)

| Parameter | Symbol | Min | Typ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Low voltage | VIL | - | - | .8 | V | - |
| Input High voltage | VIH | 2.0 | - | - | V | - |
| Input current | II | -100 | - | 100 | $\mu \mathrm{~A}$ | $0.0<\mathrm{VI}<\mathrm{Vcc}$ |
| Output Low voltage | VoL | - | - | 0.4 | V | $\mathrm{IOL}=4 \mathrm{~mA}$ |
| Output High voltage | VoH | 2.2 | - | - | V | $\mathrm{IOH}=-4 \mathrm{~mA}, \mathrm{VcC}=3.3 \mathrm{~V}$ |
|  | VoH | 2.4 | - | - | V | $\mathrm{IOH}=-4 \mathrm{~mA}, \mathrm{VCC}=5.0 \mathrm{~V}$ |
| Driver output resistance <br> (Line driver output enabled) | $\mathrm{Ro}^{1}$ | 50 | 100 | 200 | $\Omega$ | $\mathrm{VCC}=3.3 \mathrm{~V}$ |
|  | $\mathrm{Ro}^{1}$ | 50 | 100 | 200 | $\Omega$ | $\mathrm{VCC}=5.0 \mathrm{~V}$ |
| 1. Parameter is guaranteed by design; not subject to production testing. |  |  |  |  |  |  |

Table 26. Required CLK25M Characteristics

| Parameter | Sym | Min | Typ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Input Low voltage | VIL | - | - | .8 | V | - |
| Input High voltage | VIH | 2.0 | - | - | V | - |
| Input clock frequency tolerance $^{1}$ | $\Delta \mathrm{f}$ | - | - | $\pm 100$ | ppm | Clock frequency is 25 MHz |
| Input clock duty cycle $^{1}$ | Tdc | 40 | - | 60 | $\%$ | - |
| 1. Parameter is guaranteed by design; not subject to production testing. |  |  |  |  |  |  |

Table 27. Low-Voltage Fault Detect Characteristics

| Parameter | Sym | Min | Typ $^{1}$ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Detect fault threshold | VLT | 3.4 | - | 4.0 | V | - |
| Clear fault threshold | VLH | 4.1 | - | 4.7 | V | - |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |  |

Table 28. 100BASE-TX Transceiver Characteristics

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Peak differential output voltage | Vp | 0.95 | - | 1.05 | V | Note 2 |
| Signal amplitude symmetry | Vss | 98 | - | 102 | $\%$ | Note 2 |
| Signal rise/fall time | TRF | 3.0 | - | 5.0 | ns | Note 2 |
| Rise/fall time symmetry | TrFs | - | - | 0.5 | ns | Note 2 |
| Jitter (measured differentially) | - | - | 0.7 | 1.4 | ns | - |
| Duty cycle distortion | - | - | - | $+/-0.5$ | ns | Offset from 16ns pulse width at <br> $50 \%$ of pulse peak |
| Overshoot |  |  |  |  |  |  |
| 1. Typical values are at $25{ }^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. <br> 2. Measured at the line side of the transformer, line replaced by $100 \Omega(+/-1 \%)$ resistor. |  |  |  |  |  |  |

Table 29. 100BASE-FX Transceiver Characteristics

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units | Test Conditions |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Transmitter |  |  |  |  |  |  |  |
| Peak differential output voltage <br> (single ended) | Vop | 0.6 | - | 1.5 | V |  |  |
| Signal rise/fall time | TRF | - | - | 1.6 | ns | $10<->90 \%$ | 2.0 pF load |
| Jitter (measured differentially) | - | - | - | 1.3 | ns | - |  |
| Receiver |  |  |  |  |  |  |  |
| Peak differential input voltage | VIP | 0.55 | - | 1.5 | V | - |  |
| Common mode input range | VCMIR | 2.25 | - | Vcc -0.5 | V | - |  |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |  |  |

Table 30. 10BASE-T Transceiver Characteristics

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Transmitter |  |  |  |  |  |  |
| Peak differential output voltage | Vop | 2.2 | 2.5 | 2.8 | V | With specified transformer, line replaced by $100 \Omega( \pm 1 \%)$ resistor |
| Link transmit period | - | 8 | - | 24 | ms | - |
| Transmit timing jitter added by the MAU and PLS sections ${ }^{2,3}$ | - | 0 | 2 | 11 | ns | After line model specified by IEEE 802.3 for 10BASE-T MAU |
| Receiver |  |  |  |  |  |  |
| Receive input impedance ${ }^{2}$ | ZIN | - | 3.6 | - | $k \Omega$ | Between TPIP and TPIN |
| Link min receive | - | 2 | - | 7 | ms | - |
| Link max receive | - | 50 | - | 150 | ms | - |
| Time link loss receive | - | 50 | - | 150 | ms | - |
| Differential squelch threshold | VDS | 300 | 390 | 585 | mV Peak | 5 MHz square wave input |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. <br> 2. Parameter is guaranteed by design; not subject to production testing. <br> 3. IEEE 802.3 specifies maximum jitter addition at 1.5 ns for the AUI cable, 0.5 ns from the encoder, and 3.5 ns from the MAU. |  |  |  |  |  |  |

Figure 26. MII - 100BASE-TX Receive Timing


Table 31. MII-100BASE-TX Receive Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| RXD, RX_DV, RX_ER setup to RX_CLK High | t1 | 10 | - | - | ns |
| RXD, RX_DV, RX_ER hold from RX_CLK High | t2 | 5 | - | - | ns |
| CRS asserted to RXD<3:0>, RX_DV | t4 | - | 8 | - | BT |
| Receive start of "J" to CRS asserted | t6 | 0 | $15-19$ | 20 | BT |
| Receive start of "T" to CRS de-asserted | t7 | 13 | $23-27$ | 28 | BT |
| Receive start of "J" to COL asserted | t8 | 0 | $15-19$ | 20 | BT |
| Receive start of " $T$ " to COL de-asserted | t9 | 13 | $23-27$ | 28 | BT |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 27. MII-100BASE-TX Transmit Timing
TXD<3:0>

Table 32. MII- 100BASE-TX Transmit Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| TXD<3:0>, TX_EN, TX_ER setup <br> to TX_CLK High | t 1 | 10 | - | - |  |
| TXD<3:0>, TX_EN, TX_ER hold <br> from TX_CLK High | t 2 | 0 | -1 | - |  |
| TX_EN sampled to CRS asserted | t 3 | - | 2 | 4 | BT |
| TX_EN sampled to CRS de-asserted | t 4 | - | 3 | 16 | BT |
| TX_EN sampled to TPOP out (Tx latency) | t 5 | 6 | 9 | 14 | BT |
| 1. Typical values are at $25{ }^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 28. MII-100BASE-FX Receive Timing


Table 33. MII-100BASE-FX Receive Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| RXD, RX_DV, RX_ER setup to RX_CLK High | t 1 | 10 | - | - | ns |
| RXD, RX_DV, RX_ER hold from RX_CLK High | t 2 | 5 | - | - | ns |
| CRS asserted to RXD<3:0>, RX_DV asserted | t 4 | - | 8 | - | BT |
| Receive start of "J" to CRS asserted | t 6 | 0 | $13-17$ | 20 | BT |
| Receive start of "T" to CRS de-asserted | t 7 | 13 | $21-25$ | 26 | BT |
| Receive start of "J" to COL asserted | t 8 | 0 | $13-17$ | 20 | BT |
| Receive start of "T" to COL de-asserted | ta | 13 | $21-25$ | 26 | BT |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 29. MII-100BASE-FX Transmit Timing


Table 34. MII-100BASE-FX Transmit Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| TXD<3:0>, TX_EN, TX_ER setup to TX_CLK High | t1 | 10 | - | - | ns |
| TXD<3:0>, TX_EN, TX_ER hold from TX_CLK High | t2 | 0 | -1 | - | ns |
| TX_EN sampled to CRS asserted | t3 | - | 2 | 4 | BT |
| TX_EN sampled to CRS de-asserted | t 4 | - | 3 | 16 | BT |
| TX_EN sampled to FIBOP out (Tx latency) | t 5 | 6 | 11 | 14 | BT |

1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.

Figure 30. MII - 10BASE-T Receive Timing


Table 35. MII-10BASE-T Receive Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units |
| :---: | :---: | :---: | :---: | :---: | :---: |
| RXD, RX_DV, RX_ER setup to RX_CLK High | t1 | 10 | - | - | ns |
| RXD, RX_DV, RX_ER hold from RX_CLK High | t2 | 10 | - | - | ns |
| TPI in to RXD out (Rx latency) | t3 | - | - | $73^{2}$ | BT |
| CRS asserted to RXD, RX_DV, RX_ER asserted | t4 | 0 | - | $69^{2}$ | BT |
| RXD, RX_DV, RX_ER de-asserted to CRS de-asserted | t5 | 0 | 2.5-5.5 | 6 | BT |
| TPI in to CRS asserted | t6 | 0 | 4 | 5 | BT |
| TPI quiet to CRS de-asserted | t7 | 0 | 18 | 19 | BT |
| TPI in to COL asserted | t8 | 0 | 4 | 5 | BT |
| TPI quiet to COL de-asserted | t9 | 0 | 18 | 19 | BT |

1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.
2. CRS is asserted. RXD/RX_DV are driven at the start of SFD (64 BT).

Figure 31. MII-10BASE-T Transmit Timing


Table 36. MII- 10BASE-T Transmit Timing Parameters

| Parameter | Sym | Min | Typ $^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| TXD, TX_EN, TX_ER setup to TX_CLK High | t 1 | 10 | - | - | ns |
| TXD, TX_EN, TX_ER hold from TX_CLK High | t 2 | 0 | -1 | - | ns |
| TX_EN sampled to CRS asserted | t 3 | - | 2 | 4 | BT |
| TX_EN sampled to CRS de-asserted | t 4 | - | $8-11$ | - | BT |
| TX_EN sampled to TPO out (Tx latency) | t 5 | - | $3-5$ | - | BT |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 32. 10BASE-T SQE (Heartbeat) Timing
TX_CLK $\square \square \square \square \square \square \square$


Table 37. 10BASE-T SQE (Heartbeat) Timing Parameters

| Parameter | Sym | Min | Typ $^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| COL (SQE) delay after TX_EN off | t 1 | 0.65 | 1.0 | 1.6 | $\mu \mathrm{~s}$ |
| COL (SQE) pulse duration | t 2 | .5 | 1.0 | 1.5 | $\mu \mathrm{~s}$ |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 33. 10BASE-T Jab and Unjab Timing


Table 38. 10BASE-T Jab and Unjab Timing Parameters

| Parameter | Sym | Min | Typ $^{1}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Maximum transmit time | t 1 | 20 | $96-128^{2}$ | 150 | ms |
| Unjab time | t 2 | 250 | 525 | 750 | ms |

1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.
2. Typical transmit time may be either of these values depending on internal 32 ms clock synchronization.

Figure 34. Auto Negotiation and Fast Link Pulse Timing


Figure 35. Fast Link Pulse Timing


Table 39. Auto Negotiation and Fast Link Pulse Timing Parameters

| Parameter | Sym | Min | Typ $^{\mathbf{1}}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Clock/Data pulse width | t 1 | - | 100 | - | ns |
| Clock pulse to Data pulse | t 2 | 55.5 | 62.5 | 69.5 | $\mu \mathrm{~s}$ |
| Clock pulse to Clock pulse | t 3 | 111 | 125 | 139 | $\mu \mathrm{~s}$ |
| FLP burst width | t 4 | - | 2 | - | ms |
| FLP burst to FLP burst | t 5 | 8 | 12 | 24 | ms |
| Clock/Data pulses per burst | - | 17 | - | 33 | ea |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |

Figure 36. MDIO Timing when Sourced by STA


Figure 37. MDIO Timing When Sourced by PHY


Table 40. MII Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{1}$ | Max | Units | Test Conditions |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| MDIO setup before MDC | - | 10 | - | - | ns | When sourced by STA |
| MDIO hold after MDC | - | 10 | - | - | ns | When sourced by STA |
| MDC to MDIO output delay | - | 0 | 27 | 300 | ns | When sourced by PHY |
| 1. Typical values are at $25^{\circ}$ C and are for design aid only; not guaranteed and not subject to production testing. |  |  |  |  |  |  |

Figure 38. Power Down Timing


Table 41. Power Down Timing Parameters

| Parameter | Sym | Min | Typ ${ }^{\mathbf{1}}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Power Down recovery time | tPDR | - | 50 | - | ms |

1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; not guaranteed and not subject to production testing.

Figure 39. Serial LED Timing


Table 42. Serial LED Timing Parameters

| Parameter | Symbol | Min | Typ $^{\mathbf{1}}$ | Max | Units |
| :--- | :---: | :---: | :---: | :---: | :---: |
| LEDENA setup to LEDCLK falling edge | tena1 | 5 | 12 | - | ns |
| LEDENA hold from LEDCLK falling edge | tena2 | 15 | 21 | - | ns |
| LEDDAT setup to LEDCLK falling edge | tdat1 | 5 | 12 | - | ns |
| LEDDAT hold from LEDCLK falling edge | tdat2 | 15 | 21 | - | ns |
| 1. Typical values are at $25^{\circ} \mathrm{C}$ and are for design aid only; they are not guaranteed and not subject to production testing. |  |  |  |  |  |

### 5.0 Register Definitions

The LXT974/975 register set includes a total of 48 16-bit registers, 12 registers per port. Refer to Table 43 for a complete register listing.

- Seven base registers (0 through 6) are defined in accordance with the "Reconciliation Sublayer and Media Independent Interface" and "Physical Layer Link Signaling for 10/100 Mbps AutoNegotiation" sections of the IEEE 802.3 specification (Register 7, Next Page, is not supported).
- Five additional registers (16 through 20) are defined in accordance with the IEEE 802.3 specification for adding unique chip functions.

Table 43. Register Set

| Address | Register Name | Bit Assignments |
| :---: | :--- | :--- |
| 0 | Control Register | Refer to Table 44 on page 64 |
| 1 | Status Register | Refer to Table 45 on page 65 |
| 2 | PHY Identification Register 1 | Refer to Table 46 on page 66 |
| 3 | PHY Identification Register 2 | Refer to Table 47 on page 66 |
| 4 | Auto-Negotiation Advertisement Register | Refer to Table 48 on page 67 |
| 5 | Auto-Negotiation Link Partner Ability Register | Refer to Table 49 on page 67 |
| 6 | Auto-Negotiation Expansion Register | Refer to Table 50 on page 68 |
| 16 | LED Configuration Register | Refer to Table 51 on page 68 |
| 17 | Interrupt Enable Register | Refer to Table 52 on page 69 |
| 18 | Interrupt Status Register | Refer to Table 53 on page 70 |
| 19 | Port Configuration Register | Refer to Table 55 on page 71 |
| 20 | Port Status Register |  |

Table 44. Control Register

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 0.15 | Reset | $\begin{aligned} & 1=\text { Reset port. } \\ & 0=\text { Enable normal operation. } \end{aligned}$ | $\begin{gathered} \mathrm{R} / \mathrm{W} \\ \mathrm{SC} \end{gathered}$ | 0 |
| 0.14 | Loopback | 1 = Enable loopback mode. $0=$ Disable loopback mode. | R/W | 0 |
| 0.13 | Speed Selection | $\begin{aligned} & 1=100 \mathrm{Mbps} . \\ & 0=10 \mathrm{Mbps} . \end{aligned}$ | R/W | Note 2 |
| 0.12 | Auto-Negotiation Enable | 1 = Enable auto-negotiate process (overrides speed select and duplex mode bits). <br> $0=$ Disable auto-negotiate process. | R/W | Note 3 |
| 0.11 | Power Down | 1 = Enable power down. <br> 0 = Enable normal operation. | R/W | Note 4 |
| 0.10 | Isolate | 1 = Electrically isolate port from MII. $0=$ Normal operation. | R/W | Note 5 |
| 0.9 | Restart AutoNegotiation | $1=$ Restart auto-negotiation process. <br> $0=$ Normal operation. | $\begin{gathered} \mathrm{R} / \mathrm{W} \\ \mathrm{SC} \end{gathered}$ | Note 6 |
| 0.8 | Duplex Mode | 1 = Enable full-duplex. 0 = Enable half-duplex. | R/W | Note 7 |
| 0.7 | Collision Test | 1 = Enable COL signal test. 0 = Disable COL signal test. | R/W | Note 8 |
| 0.6:4 | Transceiver Test Mode | Not supported. | RO | 0 |
| 0.3 | Master-Slave Enable | Not supported. | RO | 0 |
| 0.2 | Master-Slave Value | Not supported. | RO | 0 |
| 0.1:0 | Reserved | Write as 0 ; ignore on read. | R/W | N/A |
| 1. R/W = Read/Write <br> RO = Read Only <br> SC = Self Clearing <br> 2. If auto-negotiation is enabled, this bit is ignored. If auto-negotiation is disabled, the default value of bit 0.13 is determined by CFG_0. <br> 3. If $\mathrm{SD}_{-}^{-T X} n$ is tied High or to a 5 V PECL input (FX Mode), the default value of bit $0.12=0$. If $\mathrm{SD}_{-} \overline{T X} n$ is tied Low (TP Mode), the default value of bit 0.12 is determined by AUTOENA. <br> 4. The LXT974/975 internally holds all set values of the configuration registers upon exiting power down mode. A delay of 500 ns minimum is required from the time power down is cleared until any register can be written. <br> 5. The default value of bit 0.10 is determined by pin TRSTEn. <br> 6. If auto-negotiation is enabled, the default value of bit 0.9 is determined by CFG_0. If auto-negotiation is disabled, the bit is ignored. <br> 7. If auto-negotiation is enabled, this bit is ignored. If auto-negotiation is disabled and the port is operating in TX mode, the default value of bit 0.8 is determined by pin FDE. If auto-negotiation is disabled and the port is operating in FX mode, the default value of bit 0.8 is determined by pin FDE_FX. <br> 8. This bit is ignored unless loopback is enabled (bit $0.14=1$ ). |  |  |  |  |

Table 45. Status Register (Address 1)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 1.15 | 100BASE-T4 | Not supported. | RO | 0 |
| 1.14 | 100BASE-X <br> Full-Duplex | 1 = Port able to perform full-duplex 100BASE-X. | RO | 1 |
| 1.13 | 100BASE-X Half-Duplex | 1 = Port able to perform half-duplex 100BASE-X. | RO | 1 |
| 1.12 | 10 Mbps Full- Duplex | 1 = Port able to operate at 10 Mbps in full-duplex mode. | RO | 1 |
| 1.11 | 10 Mbps Half-Duplex | 1 = Port able to operate at 10 Mbps in half-duplex mode. | RO | 1 |
| 1.10 | 100BASE-T2 <br> Full-Duplex | Not supported. | RO | 0 |
| 1.9 | 100BASE-T2 <br> Half-Duplex | Not supported. | RO | 0 |
| 1.8 | Reserved | Ignore on read. | RO | 0 |
| 1.7 | Master-Slave Configuration Fault | Not supported. | RO | 0 |
| 1.6 | MF Preamble Suppression | $0=$ Port will not accept management frames with preamble suppressed. | RO | 0 |
| 1.5 | Auto-Negotiation Complete | 1 = Auto-negotiation process complete. <br> $0=$ Auto-negotiation process not complete. | RO | 0 |
| 1.4 | Remote Fault | 1 = Remote fault condition detected. $0=$ No remote fault condition detected. | RO/LH | 0 |
| 1.3 | Auto-Negotiation Ability | 1 = Port is able to perform auto-negotiation. | RO | 1 |
| 1.2 | Link Status | $\begin{aligned} & 1=\text { Link is up. } \\ & 0=\text { Link is down. } \end{aligned}$ | RO/LL | 0 |
| 1.1 | Jabber Detect (10BASE-T Only) | 1 = Jabber condition detected. <br> $0=$ No jabber condition detected. | RO/LH | 0 |
| 1.0 | Extended Capability | 1 = Extended register capabilities. | RO | 1 |
| 1. RO $=$ Read OnlyLL $=$ Latching LowLH $=$ Latching High |  |  |  |  |

Table 46. PHY Identification Register 1 (Address 2)

| Bit | Name | Description | Type $^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| $2.15: 0$ | PHY ID <br> Number | The PHY identifier composed of bits 3 through 18 of the OUI. | RO | 7810 hex |
| 1. RO $=$ Read Only |  |  |  |  |

Table 47. PHY Identification Register 2 (Address 3)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :--- | :--- | :--- | :---: | :---: |
| $3.15: 10$ | PHY ID number | The PHY identifier composed of bits 19 through 24 of the <br> OUI. | RO | 000000 |
| $3.9: 4$ | Manufacturer's <br> model number | 6 bits containing manufacturer's part number. | RO | $000100-$ LXT974 <br> $000101-$ LXT975 |
| $3.3: 0$ | Manufacturer's <br> revision number | 4 bits containing manufacturer's revision number. | RO | XXXX² |
| 1. RO = Read Only <br> 2. Default will be 0011 for LXT974A/975A and 0100 for LXT974B/975B. |  |  |  |  |

Figure 40. PHY Identifier Bit Mapping


The Intel OUI is 00207B hex.

Table 48. Auto Negotiation Advertisement Register (Address 4)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 4.15 | Next Page | Not supported. | RO | 0 |
| 4.14 | Reserved | Ignore. | RO | 0 |
| 4.13 | Remote Fault | 1 = Remote fault. <br> $0=$ No remote fault. | R/W | 0 |
| 4.12:11 | Reserved | Ignore. | R/W | 0 |
| 4.10 | Pause | 1 = Pause operation is enabled for full-duplex links. <br> $0=$ Pause operation is disabled. | R/W | 0 |
| 4.9 | 100BASE-T4 | $1=$ 100BASE-T4 capability is available. <br> $0=100 B A S E-T 4$ capability is not available. <br> (The LXT974/975 does not support 100BASE-T4 but allows this bit to be set to advertise in the Auto-Negotiation sequence for 100BASE-T4 operation. An external 100BASE-T4 transceiver could be switched in if this capability is desired.) | R/W | 0 |
| 4.8 | 100BASE-TX full-duplex | 1 = Port is 100BASE-TX full-duplex capable. <br> $0=$ Port is not 100BASE-TX full-duplex capable. | R/W | Note 2 |
| 4.7 | 100BASE-TX | 1 = Port is 100BASE-TX capable. <br> $0=$ Port is not 100BASE-TX capable. | R/W | Note 3 |
| 4.6 | 10BASE-T <br> full-duplex | 1 = Port is 10BASE-T full-duplex capable. <br> $0=$ Port is not 10BASE-T full-duplex capable. | R/W | Note 4 |
| 4.5 | 10BASE-T | 1 = Port is 10BASE-T capable. <br> $0=$ Port is not 10BASE-T capable. | R/W | Note 5 |
| 4.4:0 | Selector Field, S<4:0> | <00001> = IEEE 802.3. <br> <00010> = IEEE 802.9 ISLAN-16T. <br> <00000> = Reserved for future Auto-Negotiation development. <br> <11111> = Reserved for future Auto-Negotiation development. <br> Unspecified or reserved combinations should not be transmitted. | R/W | 00001 |
| 1. R/W = Read/Write <br> RO = Read Only <br> 2. The default value of bit 4.8 is determined by FDE ANDed with CFG_2. <br> 3. The default value of bit 4.7 is determined by CFG_2. <br> 4. The default value of bit 4.6 is determined by FDE AND'ed with CFG_1. <br> 5. The default value of bit 4.5 is determined by CFG_1. |  |  |  |  |

Table 49. Auto Negotiation Link Partner Ability Register (Address 5)

| Bit | Name | Description | Type $^{1}$ | Default |
| :--- | :--- | :--- | :---: | :---: |
| 5.15 | Next Page | 1 = Link Partner has ability to send multiple pages. <br> $0=$ Link Partner has no ability to send multiple pages. | RO | N/A |
| 5.14 | Acknowledge | 1 = Link Partner has received Link Code Word from LXT974/975. <br> O Link Partner has not received Link Code Word from the <br> LXT974/975. | RO | N/A |
| 5.13 | Remote Fault | $1=$ Remote fault. <br> $0=$ No remote fault. | RO | N/A |
| $5.12: 11$ | Reserved | Ignore. | RO | N/A |
| 1. RO $=$ Read Only |  |  |  |  |

Table 49. Auto Negotiation Link Partner Ability Register (Address 5) (Continued)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 5.10 | Pause | 1 = Pause operation is enabled for link partner. <br> $0=$ Pause operation is disabled. | RO | N/A |
| 5.9 | 100BASE-T4 | 1 = Link Partner is 100BASE-T4 capable. <br> $0=$ Link Partner is not 100BASE-T4 capable. | RO | N/A |
| 5.8 | 100BASE-TX full-duplex | 1 = Link Partner is 100BASE-TX full-duplex capable. <br> $0=$ Link Partner is not 100BASE-TX full-duplex capable. | RO | N/A |
| 5.7 | 100BASE-TX | 1 = Link Partner is $100 B A S E-T X$ capable. <br> $0=$ Link Partner is not 100BASE-TX capable. | RO | N/A |
| 5.6 | 10BASE-T <br> full-duplex | 1 = Link Partner is 10BASE-T full-duplex capable. <br> $0=$ Link Partner is not 10BASE-T full-duplex capable. | RO | N/A |
| 5.5 | 10BASE-T | $1=$ Link Partner is 10BASE-T capable. <br> $0=$ Link Partner is not 10BASE-T capable. | RO | N/A |
| 5.4:0 | Selector Field S<4:0> | <00001> = IEEE 802.3. <br> <00010> = IEEE 802.9 ISLAN-16T. <br> <00000> = Reserved for future Auto-Negotiation development. <br> <11111> = Reserved for future Auto-Negotiation development. <br> Unspecified or reserved combinations shall not be transmitted. | RO | N/A |
| 1. RO = Read Only |  |  |  |  |

Table 50. Auto Negotiation Expansion (Address 6)

| Bit | Name | Description | Type $^{1}$ | Default |
| :--- | :--- | :--- | :---: | :---: |
| $6.15: 5$ | Reserved | lgnore on read. | RO | 0 |
| 6.4 | Parallel <br> Detection Fault | $1=$ Parallel detection fault has occurred. <br> $0=$ Parallel detection fault has not occurred. | RO/ <br> LH | 0 |
| 6.3 | Link Partner <br> Next Page Able | $1=$ Link partner is next page able. <br> $0=$ Link partner is not next page able. | RO | 0 |
| 6.2 | Next Page Able | Not supported. | RO | 0 |
| 6.1 | Page Received | $1=$ Three identical and consecutive link code words have been received <br> from link partner. <br> $0=$ Three identical and consecutive link code words have not been <br> received from link partner. | RO <br> LH | 0 |
| 6.0 | Link Partner A/ <br> N Able | 1 = Link partner is auto-negotiation able. <br> $0=$ Link partner is not auto-negotiation able. | RO | 0 |
| 1. RO = Read Only LH = Latching High |  |  |  |  |

Table 51. LED Configuration Register (Address 16, Hex 10)

| Bit | Name | Description | Type $^{1}$ | Default |
| :--- | :--- | :--- | :---: | :---: |
| $16.15: 12$ | User Defined | No effect on chip operation. | R/W | N/A |
| $16.11: 9$ | Reserved | Ignore on read. | RO | N/A |
| 16.8 | TX Pulse Tuning | $1=$ Faster rise time - May be used to adjust output pulse to match <br> magnetic performance. <br> $0=$ Normal Operation - Provides best match for most magnetics. | R/W | 0 |
| 1. R/W $=$ Read /Write |  |  |  |  |

Table 51. LED Configuration Register (Address 16, Hex 10)

| Bit | Name | Description |  |  | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 16.7:6 | LED_2 Select | Determine condition indicated by LED_2 |  |  | R/W | 00 |
|  |  | bit 7 | bit 6 | Indication Setting |  |  |
|  |  | 0 | 0 | LEDn_2 indicates Link |  |  |
|  |  | 0 | 1 | LEDn_2 indicates Half-Duplex Status |  |  |
|  |  | 1 | 0 | LEDn_2 indicates 100 Mbps |  |  |
|  |  | 1 | 1 | LEDn_2 indicates Collision |  |  |
| 16.5:4 | LED_1 Select | Determine condition indicated by LED_1 |  |  | R/W | 00 |
|  |  | bit 5 | bit 4 | Indication Setting |  |  |
|  |  | 0 | 0 | LEDn_1 indicates Receive Activity |  |  |
|  |  | 0 | 1 | LEDn_1 indicates Link |  |  |
|  |  | 1 | 0 | LEDn_1 indicates Half-Duplex Status |  |  |
|  |  | 1 | 1 | LEDn_1 indicates 100 Mbps |  |  |
| 16.3:2 | LED_0 Select | Determine condition indicated by LED_0 |  |  | R/W | 00 |
|  |  | bit 3 | bit 2 | Indication Setting |  |  |
|  |  | 0 | 0 | LEDn_0 indicates Transmit Activity |  |  |
|  |  | 0 | 1 | LEDn_0 indicates Receive Activity |  |  |
|  |  | 1 | 0 | LEDn_0 indicates Link |  |  |
|  |  | 1 | 1 | LEDn_0 indicates Half-Duplex Status |  |  |
| 16.1 | Link Algorithm | 1 = Enhanced link algorithm - Link goes down when 12 idle symbols in a row are not received within 1 to 2 ms . <br> $0=$ Standard link algorithm - Link goes down when symbol error rate is greater than 64/1024. |  |  | R/W | 0 |
| 16.0 | LED Extension | 1 = Disable extension of LED active time for LEDn_<2:0>. $0=$ Enable extension of LED active time for LEDn_<2:0>. |  |  | R/W | 0 |
| 1. R/W = Read $/$ Write |  |  |  |  |  |  |

Table 52. Interrupt Enable Register (Address 17, Hex 11)

| Bit | Name | Description | Type $^{1}$ | Default |
| :--- | :--- | :--- | :---: | :---: |
| $17.15: 2$ | Reserved | Write as 0; ignore on read. | R/W | N/A |
| 17.1 | INTEN | $1=$ Enable interrupts. Must be enabled for bit 17.0 or 19.12 to be effective. <br> $0=$ Disable interrupts. | R/W | 0 |
| 17.0 | TINT | $1=$ Forces MDINT Low and sets bit $18.15=1$. Also forces interrupt pulse <br> on MDIO when bit $19.12=1$. <br> $0=$ Normal operation. <br> This bit is ignored unless the interrupt function is enabled $(17.1=1)$. | R/W | 0 |

Table 53. Interrupt Status Register (Address 18, Hex 12)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :--- | :--- | :---: | :---: |
| 18.15 | MINT | $1=$ Indicates MII interrupt pending. <br> $0=$ Indicates no MII interrupt pending. This bit is cleared by reading <br> Register 1 followed by reading Register 18. | RO | N/A |
| $18.14: 0$ | Reserved | Ignore | RO | 0 |
| 1. RO = Read Only |  |  |  |  |

Table 54. Port Configuration Register (Address 19, Hex 13)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 19.15 | Reserved | Write as 0; ignore on read. | R/W | N/A |
| 19.14 | Txmit Test Enable (100BASETX) | 1 = 100BASE-T transmit test enabled (Port transmits data regardless of receive status). <br> $0=$ Normal operation. | R/W | 0 |
| 19.13 | Reserved | Write as 0; ignore on read. | R/W | N/A |
| 19.12 | MDIO_INT | 1 = Enable interrupt signaling on MDIO (if 17.1 = 1). <br> $0=$ Normal operation (MDIO Interrupt disabled). <br> Bit is ignored unless the interrupt function is enabled $(17.1=1)$. | R/W | 0 |
| 19.11 | TP Loopback Enable (10BASE-T) | 1 = Disable 10BT Loopback - Data transmitted by the MAC will not loopback to the RXD and RX_DV pins. Only CRS is looped back. <br> 0 = Enable 10BT Loopback - Preamble, SFD, and data are directly looped back to the MII. | R/W | 0 |
| 19.10 | SQE Disable (10BASE-T) | $\begin{aligned} & 1=\text { Normal operation (SQE enabled). } \\ & 0=\text { Disable SQE. } \end{aligned}$ | R/W | 0 |
| 19.9 | Jabber Disable (10BASE-T) | $\begin{aligned} & 1=\text { Disable jabber. } \\ & 0=\text { Normal operation (jabber enabled). } \end{aligned}$ | R/W | 0 |
| 19.8 | Link Test Enable (10BASE-T) | 1 = Disable 10BASE-T link integrity test. <br> $0=$ Normal operation (10BASE-T link integrity test enabled). | R/W | Note 2 |
| 19.7:6 | Reserved | Write as 0; ignore on read. | R/W | N/A |
| 19.5 | Advance TX Clock | $1=$ TX clock is advanced relative to TXD<3:0> and TX_ER by $1 / 2$ TX_CLK cycle. <br> $0=$ Normal operation. | R/W | 0 |
| 19.4 | Reserved | Write as 0; ignore on read. | R/W | N/A |
| 1. $R / W=$ Read/Write <br> 2. If auto-negotiation is disabled, the default value of bit 19.8 is determined by pin 115 (CFG_1). If CFG_1 is High, the default value of bit $19.8=1$. <br> If CFG_1 is Low, the default value of bit $19.8=0$. If auto-negotiation is enabled, the default value of bit $19.8=0$. <br> 3. The default value of bit 19.3 is determined by BYPSCR. If BYPSCR is High, the default value of bit $19.3=1$. <br> If BYPSCR is Low, the default value of bit 19.3 $=0$. <br> 4. The default value of bit 19.2 is determined by the $\mathrm{SD} / \overline{\mathrm{TP}} n$ pin for the respective port. <br> If $\mathrm{SD} / \overline{\mathrm{TP}} n$ is tied Low, the default value of bit $19.2=0$. If $\mathrm{SD} / \overline{\mathrm{TP}} n$ is not tied Low, the default value of bit $19.2=1$. <br> On the LXT975, this bit is ignored on ports 0 and 2 that operate in twisted-pair mode only. |  |  |  |  |

Table 54. Port Configuration Register (Address 19, Hex 13) (Continued)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 19.3 | Scrambler <br> Bypass <br> (100BASE-T <br> only) | 1 = Bypass transmit scrambler and receive descrambler. <br> $0=$ Normal operation (scrambler and descrambler enabled). | R/W | Note 3 |
| 19.2 | 100BASE-FX | 1 = Enable 100BASE fiber interface. <br> 0 = Enable 100BASE twisted pair interface. | R/W | Note 4 |
| 19.1 | Reserved | Write as 0; ignore on read. | R/W | 0 |
| 19.0 | Transmit Disconnect | 1 = Disconnect TP transmitter from line. <br> $0=$ Normal operation. | R/W | 0 |
| 1. R/W = Read/Write <br> 2. If auto-negotiation is disabled, the default value of bit 19.8 is determined by pin 115 (CFG_1). If CFG_1 is High, the default value of bit $19.8=1$. <br> If CFG_1 is Low, the default value of bit $19.8=0$. If auto-negotiation is enabled, the default value of bit $19.8=0$. <br> 3. The default value of bit 19.3 is determined by BYPSCR. If BYPSCR is High, the default value of bit $19.3=1$. <br> If BYPSCR is Low, the default value of bit $19.3=0$. <br> 4. The default value of bit 19.2 is determined by the SD/TPn pin for the respective port. If $\mathrm{SD} / \mathrm{TP} n$ is tied Low, the default value of bit $19.2=0$. If $\mathrm{SD} / \overline{T P} n$ is not tied Low, the default value of bit $19.2=1$. On the LXT975, this bit is ignored on ports 0 and 2 that operate in twisted-pair mode only. |  |  |  |  |

Table 55. Port Status Register (Address 20, Hex 14)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 20.15:14 | Reserved | Write as 0, ignore on read. | R/W | N/A |
| 20.13 | Link | $1=\text { Link is up. }$ <br> $0=$ Link is down. <br> Link bit 20.13 is a duplicate of bit 1.2, except that it is a dynamic indication, whereas bit 1.2 latches Low. | RO | 0 |
| 20.12 | Duplex <br> Mode | $\begin{aligned} & 1=\text { Full-Duplex. } \\ & 0=\text { Half-Duplex. } \end{aligned}$ | RO | Note 2 |
| 20.11 | Speed | $1=100 \mathrm{Mbps}$ operation. <br> $0=10 \mathrm{Mbps}$ operation. | RO | Note 2 |
| 20.10 | Reserved | Ignore. | RO/LH | N/A |
| 20.9 | Auto-Negotiation Complete | 1 = Auto-negotiation process complete. <br> $0=$ Auto-negotiation process not complete. <br> Auto-Negotiation Complete bit 20.9 is a duplicate of bit 1.5. | RO/LH | 0 |
| 20.8 | Page Received | 1 = Three identical and consecutive link code words received. $0=$ Three identical and consecutive link code words not received. Page Received bit 20.8 is a duplicate of bit 6.1. | RO/LH | 0 |
| 20:7 | Reserved | Write as 0, ignore on read. | R/W | 0 |
| 20.6 | Stream cipher lock (100BASETX only) | 1 = Stream cipher locked. <br> 0 = Stream cipher not locked. | RO | 0 |
| 20.5 | Symbol Error | 1 = Symbol error detected. <br> $0=$ No symbol error detected. | $\begin{aligned} & \mathrm{RO} / \\ & \mathrm{LH} \end{aligned}$ | N/A |
| 1. $\mathrm{R} / \mathrm{W}=$ Read $/$ Write <br> RO = Read Only <br> LH = Latching High <br> 2. Bits 20.12 and 20.11 reflect the current operating mode of the LXT974/975. |  |  |  |  |

Table 55. Port Status Register (Address 20, Hex 14) (Continued)

| Bit | Name | Description | Type ${ }^{1}$ | Default |
| :---: | :---: | :---: | :---: | :---: |
| 20.4 | MLT3 <br> Encoding Error | 1 = MLT3 encoding error detected. <br> $0=$ No MLT3 encoding error detected. | $\begin{aligned} & \mathrm{RO} / \\ & \mathrm{LH} \end{aligned}$ | N/A |
| 20.3 | Reserved | Ignore. | RO | N/A |
| 20.2 | Low-Voltage Fault | 1 = Low-voltage fault on VCC has occurred. $0=$ No fault. | $\begin{gathered} \mathrm{RO} / \\ \mathrm{LH} \end{gathered}$ | N/A |
| 20.1 | Reserved | Write as 0 , ignore on read. | R/W | N/A |
| 20.0 | Reserved | Ignore. | $\begin{gathered} \mathrm{RO} / \\ \mathrm{LH} \end{gathered}$ | N/A |
| 1. R/W = Read $/$ Write <br> RO = Read Only <br> LH = Latching High <br> 2. Bits 20.12 and 20.11 reflect the current operating mode of the LXT974/975. |  |  |  |  |

### 6.0 Package Specification

Figure 41. LXT974/975 Package Specification

## 160-Pin PQFP with Heat Spreader

- (Commercial Temp 0-70 ${ }^{\circ} \mathrm{C}$ )
- Part Number LXT974AHC
- Part Number LXT975AHC
- Part Number LXT974BHC
- Part Number LXT975BHC


Table 56. QUAD FLAT PACKAGE

| Dim. | All Dimensions in millimeters |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. | Notes |
| A | --- | 3.70 | 4.07 |  |
| A $_{\mathbf{1}}$ | 0.25 | 0.33 | --- |  |
| $A_{\mathbf{2}}$ | 3.20 | 3.37 | 3.60 |  |
| D | 31.20 BSC |  |  | 5 |
| $\mathrm{D}_{\mathbf{1}}$ | 28.00 BSC |  |  |  |
| E | 31.20 BSC |  |  |  |
| $\mathrm{E}_{\mathbf{1}}$ | 28.00 BSC |  |  |  |
| L | 0.73 | 0.88 | 1.03 | 5 |
| M | --- | --- | --- | $6,7,8$ |
| b | 0.22 | --- | 0.38 | 8 |
| e | 0.65 BSC |  |  |  |

NOTE:

1. All dimensions are in millimeters.
2. This package conforms to JEDEC publication 95 registration MS-022, variation DD-1.
3. Datum plane -H - located at mold parting line and is coincident with leads where leads exit plastic body at bottom of parting line.
4. Measured at seating plane -C-
5. Measured at datum plane -H-.
6. Dimensions D1 and E1 do not include mold protrusion. Allowable mold protrusion is 0.254 mm .
7. Package top dimensions are smaller than bottom dimensions. Top of package will not overhang bottom of package.
8. Dimension b does not include dambar protrusion Allowable dambar protrusion is no more than 0.08 mm .

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