TOSHIBA Bipolar Linear Integrated Circuit Silicon Monolithic

## TA31275FN, TA31275FNG

## AM/FM RF/IF Detector IC for Low Power Wireless System

The TA 31275 FN is an RF/IF detector IC for AM/F M radio.
The IC incorporates an RF amp, 2-level comparator, and local $\times 8$ circuit

## Features

- RF frequency: 240 to 450 MHz (multiplication is used) 100 to 450 MHz (multiplication is not used)
- IF frequency: 10.7 MHz
- Operating voltage range: 2.4 to 5.5 V
- Current dissipation: $5.8 \mathrm{~mA}(\mathrm{FM}) / 5.4 \mathrm{~mA}(\mathrm{AM})$
(except current at oscillator circuit)
- Current dissipation at BS: $0 \mu \mathrm{~A}$ (typ.)


Weight: 0.09 g (typ.)

- Small package: 24-pin SSOP ( 0.65 mm pitch)


## Block Diagram



## Pin Description

(the values of resistor and capacitor in the internal equivalent circuit are typical.)

| Pin No. | Pin Name | Function | Internal Equivalent Circuit |
| :---: | :---: | :---: | :---: |
| 1 | OSC IN | Local oscillator input pin. |  |
| 2 | V CC -Lo | Local' power supply pin | - |
| 3 | LOBS | Lo switch pin. <br> $\mathrm{H}: \times 8$ circuit pin. <br> L: Through pass |  |
| 4 | MIX OUT | Mixer output pin. <br> The output impedance of the pin is typically $330 \Omega$. | (4) |
| 5 | $\mathrm{V}_{\mathrm{CC} 1}$ | Power supply pin 1. | - |
| 6 <br>  <br> 7 | IF IN | IF amp input pin. <br> IF amp input pin. <br> Used as a bias coupling pin. |  |
| 8 | GND2 | GND pin 2. | - |
| 9 | BS | Battery saving pin. |  |


| Pin No. | Pin Name | Function | Internal Equivalent Circuit |
| :---: | :---: | :---: | :---: |
| 10 | QUAD | Phase-shift input terminal for the FSK Demodulator. <br> Connect to the discriminator or LC. |  |
| 11 | VCC2 | Power supply pin 2. | - |
| 12 | DATA | FM/AM waveform shaping output pin. Open collector output. Connect a pull-up resistor. | (12) |
| 13 | RF IN | RF signal input pin. | (16) |
| 14 | RF DEC | Emitter pin for internal transistor. | $3 \mathrm{k}$ |
| 16 | RF OUT | RF amp output pin. | (14) |
| 15 | CHARGE | Control terminal for quick charge circuit. To use the quick charge circuit, attach a capacitor. | (15) |
| 17 | GND1 | GND pin 1. | - |
| 18 | MIX IN | Mixer input pin. |  |
| 19 | AM/FM | Changeover switch for ASK/FSK. <br> Hi: AM <br> Lo: FM |  |


| Pin No. | Pin Name | Function | Internal Equivalent Circuit |
| :---: | :---: | :---: | :---: |
| 20 | REF | Threshold input terminal for 2-level FM/AM comparator. |  |
| 21 | RSSI | RSSI output pin. | (21) |
| 22 | AFOUT | Output terminal for FM demodulator. |  |
| 23 | LPF IN | FM/AM LPF input pin. |  |
| 24 | LPF OUT | FM/AM LPF output pin. |  |

Equivalent circuits are given to help understand design of the external circuits to be connected. They do not accurately represent the internal circuits.

## Functions

## 1. Waveform Shaper Circuit (comparator)

The output data (pin 12) are inverted.

## 2. RSSI Function

DC potential corresponding to the input level of IF IN (pin 6) is output to RSSI (pin 21). Output to RSSI (pin 21) is converted to a voltage by the internal resistance. Thus, connecting external resistance $R$ to pin 21 varies the gradient of the RSSI output as shown below. Note that due to the displacement of temperature coefficients between external resistor R and the internal IC resistor IC resistor, the temperature characteristic of the RSSI output may change. Also, the maximum RSSI value should be VCC -1 V or less, because AM doesn't correct movement Filter AMP when voltage of RSSI high.


Figure 1


Figure 2

## 3. $V_{c c}$ Pin and GND Pin

Use the same voltage supply for VCC - Lo (2 pin) and VCC1 (5 pin) and VCC2 (11 pin) (or connect them). Also, use the same voltage supply source for GND1 (17 pin) and GND2 (8 pin) (or connect them).

## 4. Local Oscillator Circuit

The local oscillator circuit is external-input-only. The device incorporates no transistor for oscillation. Input to pin 1 at a level from 95 to $105 \mathrm{~dB} \mu \mathrm{~V}$.
Adjust the values of constants C107 and C108 shown in the application circuit diagram so that the input level will become approximately $100 \mathrm{~dB} \mu \mathrm{~V}$.

By switching the Lo switch (LOBS), the frequency set by the external circuit can be used as-is without using the $\times 8$ circuit.

| Lo Switch (LOBS) | H | L |
| :---: | :---: | :---: |
| Local oscillation <br> status | $\times 8$ circuit in operation | $\times 8$ circuit halted/through pass |

## 5. RF Amp Current Adjustment

The RF amp current dissipation can be regulated by varying resistor R as shown in the figure below. When $R=1 \mathrm{k} \Omega$, the current dissipation is approximately $600 \mu \mathrm{~A}$.


Figure 3
6. Battery-Saving (BS) Function and Lo Switch LOBS Function

The IC incorporates a battery-saving function and a Lo switch function. These function offer the following selection.

FM Mode (FM/AM pin: L)

| BS Pin/LOBS Pin | Circuit Status in the IC | IC Current Dissipation (at no signal) |
| :---: | :---: | :---: |
| H/H | Circuits in operation: <br> $\square \times 8$ circuit <br> $\square$ Mixer <br> RF amp <br> -Comparator <br> -IF amp <br> -Detector circuit <br> -RSSI <br> $\square$ Comparator capacitor charger circuit | 5.8 mA (typ.) |
| H/L | $\times 8$ circuit only halted, Frequency set by External circuit can be used as-is. | 3.5 mA (typ.) |
| L/H | $\times 8$ circuit only in operation | 2.6 mA (typ.) |
| L/L | All circuits | 0 mA (typ.) |

AM Mode (FM/AM pin: H)

| BS Pin/LOBS Pin | Circuit Status in the IC | IC Current Dissipation (at no signal) |
| :---: | :---: | :---: |
| H/H | Circuits in operation: <br> $0 \times 8$ circuit <br> 0 Mixer <br> -RF amp <br> -Comparator <br> -IF amp <br> -RSSI <br> -Comparator capacitor charger circuit | 5.4 mA (typ.) |
| H/L | $\times 8$ circuit only halted, Frequency set by External circuit can be used as-is. | 3.1 mA (typ.) |
| L/H | $\times 8$ circuit only in operation | 2.6 mA (typ.) |
| L/L | All circuits | 0 mA (typ.) |

## 7. RF Amp Gain 2

$R F$ amp gain $2\left(G_{v}(R F) 2\right)$ is a reference value calculated as follows. Measure $G_{R F}$ in the following figure. $\mathrm{G}_{\mathrm{v}}$ (RF) 2 is calculated as follows:

$$
\mathrm{G}_{\mathrm{v}}(\mathrm{RF}) 2=\mathrm{G}_{\mathrm{RF}}-\mathrm{G}_{\mathrm{v}}(\mathrm{MIX})
$$



Figure 4

## 8. IF Amp Gain

The intended value is 75 dB .

## 9. Waveform-Shaping Output Duty Cycle

The specified range of electrical characteristics is only available for single-tone.

## 10. Local Frequency Range (after multiplying frequency by 8)

When the multiplier circuit is used, the local frequency will be in the range 250.7 MHz to 439.3 MHz .

## 11. Treatment of FM Terminal when Using AM

When using AM, it is not necessary to treat the QUAD pin (pin 10). Leave it open or connected to an FM external circuit. To use the bit rate filter, connect the RSSI pin (pin 21) to the bit rate filter through a resistor. The AF-OUT pin (pin 22) should be left open.


Bit rate filter for FM

Figure 5


Bit rate filter for AM

Figure 6

Using AM causes current to flow through theAM/FM pin (pin 19). Ground theAM/FM pin (pin 19) or connect it to the BS pin (pin 9).

## 12. Control Terminal for Quick Charge Circuit (CHARGE)

CHARGE (15 pin) is control terminal for quick charge circuit. REF (20 pin) control terminal for quick charge a given period by time constant of internal resistance and outside capacitance. Enabling the CHARGE pin requires an external capacitor. In normal operation, connect a capacitor having the same capacitance as that of the capacitor connected to the REF pin (pin 20).
If the connected external capacitor (C11) is $0.1 \mu \mathrm{~F}$, the quick charge time is 7 ms (typically).

## 13. Bit Rate Filter for FM

The current FM bit rate filter is used as a tertiary filter.
If the filter is to be used at a rate other than 1200 bps , please change the filter constant.

Quadratic Filter (NRZ)

|  | R 10 | R 9 | R 8 | C 20 | C 19 | C 18 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1200 bps | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $0.01 \mu \mathrm{~F}$ | 560 pF | 3300 pF |
| 2400 bps | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 4700 pF | 270 pF | 1500 pF |
| 4800 bps | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 2200 pF | 150 pF | 680 pF |

## 14. Bit Rate Filter for AM

The current AM bit rate filter is used as a quadratic filter.
If the filter is to be used at a rate other than 1200 bps , please change the filter constant.

Quadratic Filter (NRZ)
(the bit rate filter time constant takes into account the internal resistance RSSI (30 k $\Omega$ ))

|  | R | R 10 | C 20 | C 19 |
| :---: | :---: | :---: | :---: | :---: |
| 1200 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 4700 pF | 1500 pF |
| 2400 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 2200 pF | 680 pF |
| 4800 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 1000 pF | 390 pF |

When the filter constants shown below are used, it is not necessary to set the $R$ constant value.

|  | R | R 10 | C 20 | C 19 |
| :---: | :---: | :---: | :---: | :---: |
| 1200 bps | - | $30 \mathrm{k} \Omega$ | 6800 pF | 2200 pF |
| 2400 bps | - | $30 \mathrm{k} \Omega$ | 3300 pF | 1500 pF |
| 4800 bps | - | $30 \mathrm{k} \Omega$ | 1800 pF | 820 pF |

In addition, the current AM bit rate filter can be used as a tertiary filter.
If the filter is to be used at a rate other than 1200 bps, please change the filter constant.

## Quadratic Filter (NRZ)

(the bit rate filter time constant takes into account the internal resistance RSSI (30 k $\Omega$ ))

|  | R | R 9 | R 10 | C 20 | C 19 | C 18 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1200 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $0.01 \mu \mathrm{~F}$ | 560 pF | 3300 pF |
| 2400 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 4700 pF | 270 pF | 1500 pF |
| 4800 bps | $36 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | $68 \mathrm{k} \Omega$ | 2200 pF | 150 pF | 680 pF |

When the filter constants shown below are used, it is not necessary to set the $R$ constant value.

|  | R | R 9 | R 10 | C 20 | C 19 | C 18 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1200 bps | - | $30 \mathrm{k} \Omega$ | $30 \mathrm{k} \Omega$ | $0.033 \mu \mathrm{~F}$ | 2200 pF | 8200 pF |
| 2400 bps | - | $30 \mathrm{k} \Omega$ | $30 \mathrm{k} \Omega$ | $0.015 \mu \mathrm{~F}$ | 1000 pF | 3900 pF |
| 4800 bps | - | $30 \mathrm{k} \Omega$ | $30 \mathrm{k} \Omega$ | 6800 pF | 470 pF | 1800 pF |

F or the cutoff frequency of the bit rate filter, specify a sufficiently high value for the bit rate to be used. Specifying a relatively high cutoff frequency for the bit rate filter enables a low capacitor to be used at the REF pin, therefore making the pulse rise quickly.

When AM is used, the internal resistance of RSSI is used. So, take the output resistance into account when specifying a cutoff frequency.

## Cautions for Designing Circuit Board Patterns

Observe the following cautions when designing circuit patterns for this product.

## Local Oscillator Circuit (pin 1)

Isolate the local oscillator circuit block sufficiently from the RF amp block.
Isolate the local oscillator circuit block securely so that its output will not get in the IF input, IF filter, or mixer input.
Do not place the local oscillator circuit block too close to the ceramic filter.
Subdivide the ground pattern for the local oscillator circuit block, and connect the subdivisions with thin lines.

## Mixer Output Block (pin 4) to IF Input Block (pin 6)

Isolate the input and output patterns of the IF filter securely from each other.

## Demodulator Circuit Block (pin 10)

Isolate the demodulator circuit block sufficiently from the IF input block (pin 6).
Do not place the LC too close to the IC device.

## Data Output Block (pin 12)

Isolate the data output block sufficiently from the IF input block (pin 6).
I solate the output pattern of the data output block from other circuits as much as possible, so any noise from a stage subsequent to the output will not affect them.

## RF Amp Circuit Block

(1) Preventing RF amp oscillation

Do not place the patterns connected to pins 13 and 14 too close to each other.
I sol ate the patterns connected to the input block (pin 13) and output block (pin 16) from each other. Make the RF input signal line relatively thin.
Place a relatively wide ground pattern between the RF-IN pin (pin 13) and RF-DEC pin (pin 14). Connect the RF-OUT pin (pin 16) and MIX-IN pin (pin 18) with the shortest possible pattern.
(2) Attaining a sufficient gain

To attain a sufficient RF amp gain, select an optimum value for the input matching circuit block (pin 13) according to the board circuit pattern.

## IC Mounting Area

Provide a ground pattern under the IC device, and prepare relatively many through holes.

## Maximum Ratings

(unless otherwise specified, $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$. the voltage is with reference to the ground level.)

| Characteristics | Symbol | Rating | Unit |
| :--- | :---: | :---: | :---: |
| Supply voltage | $\mathrm{V}_{\mathrm{CC}}$ | 6 | V |
| Power dissipation | $\mathrm{P}_{\mathrm{D}}$ | 780 | mW |
| Operating temperature range | $\mathrm{T}_{\text {opr }}$ | -40 to 85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature range | $\mathrm{T}_{\text {stg }}$ | -55 to 150 | ${ }^{\circ} \mathrm{C}$ |

The maximum ratings must not be exceeded at any time. Do not operate the device under conditions outside the above ratings.

## Operable Range

(unless otherwise specified, $\mathbf{T a}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$. the voltage is with reference to the ground level.)

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Operating voltage range | $\mathrm{V}_{\mathrm{CC}}$ | - | - | 2.4 | 5.0 | 5.5 | V |
| RF operating frequency 1 | $\mathrm{fr}_{\mathrm{R}} 1$ | - | When frequency multiplication is used | 240 | - | 450 | MHz |
| RF operating frequency 2 | $\mathrm{f}_{\mathrm{RF}}$ | - | When frequency multiplication is not used | 100 | - | 450 | MHz |
| Local frequency | flo | - | When frequency multiplication is used ( $\times 8$ ) | 250.7 | - | 439.3 | MHz |

Operating ranges indicate the conditions for which the device is intended to be functional even with the electrical changes.

Electrical Characteristics (unless otherwise specified: $\mathrm{Ta}=\mathbf{2 5}{ }^{\circ} \mathrm{C}, \mathrm{V} \mathrm{CC}=5 \mathrm{~V}$, fin (RF) $=\operatorname{fin}(\mathrm{MIX})=\mathbf{3 1 4 . 9} \mathbf{~ M H z}$, fin (IF) $=\mathbf{1 0 . 7} \mathrm{MHz}$ )

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Current dissipation at battery saving | $\mathrm{I}_{\text {cco }}$ | 3 | $\mathrm{BS}=$ "L", LOBS = "L" | - | 0 | 5 | $\mu \mathrm{A}$ |
| RF amp gain 1 | $\mathrm{G}_{\mathrm{V} \text { (RF) } 1}$ | 1 (5) | The input and output impedances are $50 \Omega$. | -9.0 | -6.0 | -3.0 | dB |
| Mixer conversion gain | $\mathrm{G}_{\mathrm{v} \text { (MIX) }}$ | - | - | 17 | 21 | 25 | dB |
| RSSI output voltage 1 | $\mathrm{V}_{\text {RSSI1 }}$ | - | $\mathrm{V}_{\text {in }}(\mathrm{IF})=35 \mathrm{~dB} \mu \mathrm{VEMF}$ | 0.05 | 0.25 | 0.45 | V |
| RSSI output voltage 2 | VRSSI2 | - | $\mathrm{V}_{\text {in }}(\mathrm{IF})=65 \mathrm{~dB} \mu \mathrm{VEMF}$ | 0.8 | 1.05 | 1.3 | V |
| RSSI output voltage 3 | $\mathrm{V}_{\text {RSSI3 }}$ | - | $\mathrm{V}_{\text {in }}(\mathrm{IF})=100 \mathrm{~dB} \mu \mathrm{VEMF}$ | 1.6 | 1.95 | 2.3 | V |
| RSSI output resistance | RRSSI | - | - | 22 | 30 | 38 | $\mathrm{k} \Omega$ |
| Comparator input resistance | RCOMP | - | - | 75 | 100 | 125 | $\mathrm{k} \Omega$ |
| Data output voltage (L level) | V ${ }_{\text {datal }}$ | 1 (3) | IDATAL $=500 \mu \mathrm{~A}$ | - | - | 0.4 | V |
| Data output leakage current (H level) | IDATAH | 1 (4) | - | - | - | 2 | $\mu \mathrm{A}$ |
| BS pin H-level input voltage | $V_{\text {BSH }}$ | - | - | 2.2 | - | 5.5 | V |
| BS pin L-level input voltage | $\mathrm{V}_{\text {BSL }}$ | - | - | 0 | - | 0.2 | V |
| LOBS pin H-level input voltage | VLOBSH | - | - | 2.2 | - | 5.5 | V |
| LOBS pin L-level input voltage | VLOBSL | - | - | 0 | - | 0.2 | V |

FM Mode $\left(\mathrm{Ta}=\mathbf{2 5}{ }^{\circ} \mathrm{C}, \mathrm{V} \mathrm{VC}=5.0 \mathrm{~V}\right.$, $\mathrm{fin}(\mathrm{RF})=\mathrm{fin}(\mathrm{MIX})=314.9 \mathrm{MHz}, \mathrm{fin}(\mathrm{IF})=10.7 \mathrm{MHz}$, dev $= \pm 20 \mathrm{kHz}$, fmod $=600 \mathrm{~Hz}$ (single wave))

| Characteristics | Symbol | Test Circuit | Test Condition | Min | Typ. | Max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Quiescent current consumption (for FM) | $\mathrm{I}_{\text {ccafm }}$ | 2 (1) | $\begin{aligned} & \text { BS/LOBS/FMAM = "H/H/L" } \\ & \text { Fin (Lo) }=40.7 \mathrm{MHz} \end{aligned}$ | 4.3 | 5.8 | 7.3 | mA |
| Demodulated output level | Vod | - | $\mathrm{V}_{\text {in }}(\mathrm{IF})=80 \mathrm{~dB} \mu \mathrm{VEMF}$ | 30 | 40 | 55 | mVrms |
| Waveform shaping duty ratio | DRfm | 1 (2) | $\mathrm{V}_{\mathrm{in}(\mathrm{IF})}=80 \mathrm{~dB} \mu \mathrm{VEMF}$ <br> For single tone | 45 | 50 | 55 | \% |

AM Mode $\left(\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{V} \mathrm{Cc}=5.0 \mathrm{~V}, \mathrm{fin}(\mathrm{RF})=\mathrm{fin}(\mathrm{MIX})=314.9 \mathrm{MHz}, \mathrm{fin}(\mathrm{IF})=10.7 \mathrm{MHz}\right.$, $\mathrm{AM}=\mathbf{9 0 \%}, \mathrm{fmod}=\mathbf{6 0 0 ~ H z}$ (square wave))

| Characteristics | Symbol | Test <br> Circuit | Test Condition | Min | Typ. | Max |
| :--- | :---: | :---: | :--- | :---: | :---: | :---: | Unit $\mid$

## Reference Characteristic Data*

| Characteristics | Symbol | Test Circuit | Test Condition | Typ. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: |
| IF amp input resistance | R (IF) IN | - | - | 330 | $\Omega$ |
| RF amp gain 2 | $\mathrm{G}_{\mathrm{v} \text { (RF) } 2}$ | - | - | 31 | dB |
| RF amp input resistance | $\mathrm{R}_{\text {(RF) }} \mathrm{IN}$ | - | - | 1.2 | $\mathrm{k} \Omega$ |
| RF amp input capacitance | $\mathrm{C}_{\text {(RF) }} \mathrm{IN}$ | - | - | 2.0 | pF |
| RF amp output capacitance | C (RF) OUT | - | - | 2.0 | pF |
| Mixer input resistance | $\mathrm{R}_{\text {(MIX) }} \mathrm{IN}$ | - | - | 1.5 | k $\Omega$ |
| Mixer input capacitance | $\mathrm{C}_{(\mathrm{MIX})} \mathrm{IN}$ | - | - | 1.5 | pF |
| Mixer output resistance | $\mathrm{R}_{\text {(MIX) }}$ OUT | - | - | 330 | $\Omega$ |
| Mixer intercept point | IP3 | - | - | 96 | $\mathrm{dB} \mu \mathrm{V}$ |

*: These characteristic data values are listed just for reference purposes. They are not guaranteed values.
Reference Characteristic Data (FM mode)*

| Characteristics | Symbol | Test <br> Circuit | Test Condition | Typ. |
| :--- | :---: | :---: | :--- | :---: |
| Uimiting sensitivity | Vi (LIM) | - | IF input | 35 |
| Signal-to-noise ratio 1 | S/N1 | $1(8)$ | $V_{\text {in }(I F)}=40 \mathrm{~dB} \mu \mathrm{VEMF}$ | $\mathrm{dB} \mu \mathrm{V}$ |
| EMF |  |  |  |  |

*: These characteristic data values are listed just for reference purposes. They are not guaranteed values.

## Typical Test Circuit (FSK)



## Test Circuit 1

(1) $V_{\mathrm{RSSI}}$

(3) VDATAL

(2) $\mathrm{D}_{\mathrm{R}}$

(4) I DATAH

(5) Gv (RF) 1

(7) Gv (MIX) vs VLO

(6) Gv (MIX)

(8) $\mathrm{S} / \mathrm{N} 1,2$


## Test Circuit 2



## Test Circuit 3

I cco


Reference Data (This is characteristics data when it used evaluation boards. This is not guarantee on condition that it is stating except electrical characteristics.)






Reference Data (This is characteristics data when it used evaluation boards. This is not guarantee on condition that it is stating except electrical characteristics.)




RSSI Output Voltage Characteristics (MIX inputs)




Reference Data (This is characteristics data when it used evaluation boards. This is not guarantee on condition that it is stating except electrical characteristics.)



Mixer Conversion Gain Frequency Characteristics




Reference Data (This is characteristics data when it used evaluation boards. This is not guarantee on condition that it is stating except electrical characteristics.)


Demodulation Output -

## Reference Data (with a broadband ceramic filter ( $\mathbf{2 8 0}$ k) used)








## Reference Data (with a broadband ceramic filter (280 k) used)



Reference Data (with a narrowband ceramic filter ( 150 k ) used)


## Reference Data (with a narrowband ceramic filter (150 k) used)





## TOSHIBA

## Application Circuit (FSK)



## TOSHIBA

## Application Circuit (ASK)



## Package Dimensions



Weight: 0.09 g (typ.)

## RESTRICTIONS ON PRODUCT USE

000707EBA

- TOSHIBA is continually working to improve the quality and reliability of its products. Nevertheless, semiconductor devices in general can malfunction or fail due to their inherent electrical sensitivity and vulnerability to physical stress. It is the responsibility of the buyer, when utilizing TOSHIBA products, to comply with the standards of safety in making a safe design for the entire system, and to avoid situations in which a malfunction or failure of such TOSHIBA products could cause loss of human life, bodily injury or damage to property.
In developing your designs, please ensure that TOSHIBA products are used within specified operating ranges as set forth in the most recent TOSHIBA products specifications. Also, please keep in mind the precautions and conditions set forth in the "Handling Guide for Semiconductor Devices," or "TOSHIBA Semiconductor Reliability Handbook" etc..
- The TOSHIBA products listed in this document are intended for usage in general electronics applications (computer, personal equipment, office equipment, measuring equipment, industrial robotics, domestic appliances, etc.). These TOSHIBA products are neither intended nor warranted for usage in equipment that requires extraordinarily high quality and/or reliability or a malfunction or failure of which may cause loss of human life or bodily injury ("Unintended Usage"). Unintended Usage include atomic energy control instruments, airplane or spaceship instruments, transportation instruments, traffic signal instruments, combustion control instruments, medical instruments, all types of safety devices, etc.. Unintended Usage of TOSHIBA products listed in this document shall be made at the customer's own risk.
- The products described in this document are subject to the foreign exchange and foreign trade laws.
- The information contained herein is presented only as a guide for the applications of our products. No responsibility is assumed by TOSHIBA CORPORATION for any infringements of intellectual property or other rights of the third parties which may result from its use. No license is granted by implication or otherwise under any intellectual property or other rights of TOSHIBA CORPORATION or others.
- The information contained herein is subject to change without notice.

