

FLS0116 MOSFET Integrated Smart LED Lamp Driver IC with PFC Function

Features

Built-in MOSFET(1A/550V)

SEMICONDUCTOR

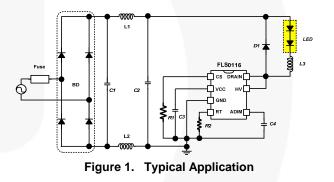
- Digitally Implemented Active-PFC Function
- No Additional Circuit for Achieving High PF
- Application Input Range: 80V_{AC} ~ 308V_{AC}
- Built-In HV Supplying Circuit: Self Biasing
- AOCP Function with Auto-Restart Mode
- Built-In Over-Temperature Protection (OTP)
- Cycle-by-Cycle Current Limit
- Current Sense Pin Open Protection
- Low Operating Current: 0.85mA (Typical)
- Under-Voltage Lockout with 5V Hysteresis
- Programmable Oscillation Frequency
- Programmable LED Current
- Analog Dimming Function
- Soft-Start Function
- Precise Internal Reference: ±3%

Applications

- LED Lamp for Decorative Lighting
- LED Lamp for Low-Power Lighting Fixture

Description

The FLS0116 LED lamp driver is a simple IC with integrated MOSFET and PFC function. The special "adopted digital" technique automatically detects input voltage condition and sends an internal reference signal to achieve high power factor. When AC input is applied to the IC, the PFC function is automatically enabled. When DC input is applied to the IC, the PFC function is automatically disabled. The FLS0116 does not need a bulk (electrolytic) capacitor for supply rail stability, which significantly improves LED lamp life.

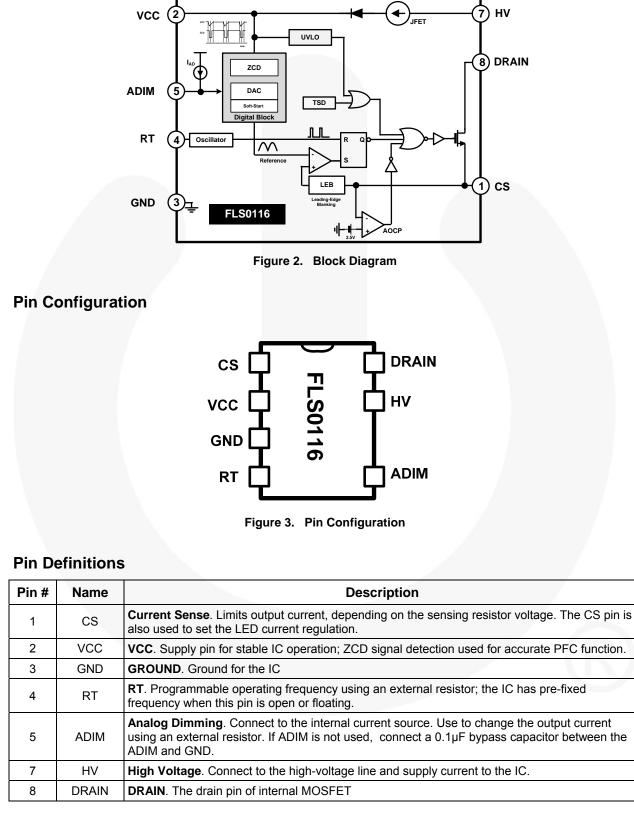


Ordering Information

Part Number	Operating Temperature Range	Package	Packing Method	
FLS0116MX	-40°C to +125°C	7-Lead, Small-Outline Integrated Circuit (SOIC), JEDEC MS-012, .150-inch, Narrow Body	Tape & Reel	

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2

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Block Diagram

Absolute Maximum Ratings

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Parameter		Min.	Max.	Unit
V _{CC}	IC Supply Voltage			20	V
HV	High Voltage Sensing			550	V
DRAIN	Internal Drain Voltage			550	V
V _{ADIM}	Analog Dimming			5	V
V _{RT}	RT Pin Voltage			5	V
V _{CS}	Allowable Current Sensing Detection Voltage			5	V
T _A	Operating Ambient Temperature Ra	nge	-40	+125	°C
TJ	Operating Junction Temperature		-40	+150	°C
T _{STG}	Storage Temperature Range		-65	+150	°C
θ_{JA}	Thermal Resistance Junction-Air ^(1,2)			135	°C/W
PD	Power Dissipation			660	mW
FOD	Electrostatic Discharge Capability	Human Body Model, JESD22-A114		2000	v
ESD		Charged Device Model, JESD22-C101		1000	v

Notes:

1. Thermal resistance test board. Size: 76.2mm x 114.3mm x 1.6mm (1S0P); JEDEC standard: JESD51-2, JESD51-3.

2. Assume no ambient airflow.

Electrical Characteristics

Typical values are at $T_A = +25^{\circ}$ C. Specifications to -40° C ~ 125° C are guaranteed by design based on final characterization results.

	Parameter	Condition	Min.	Тур.	Max.	Unit	
V _{cc} Bias Se	ection		1			•	
V _{CC}	V _{CC} Regulator Output Voltage	V _{HV} =100V _{DC}	14.0	15.5	17.0	V	
V _{CCST+}	UVLO Positive-Going Threshold	V _{cc} Increasing	12	13	14	V	
V _{CCST-}	UVLO Negative-Going Threshold	V _{cc} Decreasing	7	8	9	V	
V _{CCHYS}	UVLO Hysteresis		4	5	6	V	
I _{HV}	HV Pin Current	V _{HV} =100V _{DC} , RT=Open		0.85	1.2	mA	
I _{ST}	Startup Current			120	150	μA	
Switching S	Section						
		R_T =5.95k Ω	200	250	300	kHz	
f _{OSC} Operat	Operating Frequency	$R_T=87k\Omega$	16	20	24	kHz	
		R _T Open	40.5	45.0	49.5	kHz	
t _{MIN}	Minimum On Time ⁽³⁾			400		ns	
D _{MAX}	Maximum Duty Cycle			50		%	
t _{LEB}	Leading Edge Blanking Time ⁽³⁾			350		ns	
V _{RT}	Voltage Reference of RT Pin			1.5		V	
Soft-Start S	Section						
	Soft-Start Time ⁽³⁾	DC Mode	48	60	72	ms	
t _{ss}	Solt-Start Time	AC Mode		7		Periods	
Reference S	Section		1		•		
V _{CS1}		DC Mode	0.354	0.365	0.376		
V _{CS2}	Internal Reference Voltage of CS Pin	AC Mode ⁽³⁾	0.485	0.500	0.515	- V	
Protection	Section		1		•		
OVP _{VCC}	Over-Voltage Protection on VCC Pin		17.7	18.7	19.7	V	
V _{AOCP}	Abnormal OCP Level at CS Pin ⁽³⁾			2.5		V	
t _{AOCP}	Abnormal Detection Time ⁽³⁾			70		ns	
T _{TSDH}	Thermal Shutdown Threshold ⁽³⁾		140	150	X	°C	
T _{TSDHY}	Thermal Shutdown Threshold Hysteresis ⁽³⁾			50		°C	
Dimming So	ection						
V _{ADIM(ST+)}	Analog Dimming Positive Going Threshold ⁽³⁾		3.15	3.50	3.85	V	
V _{ADIM(ST-)}	Analog Dimming Negative Going Threshold ⁽³⁾			0.50	0.75	V	
I _{AD}	Internal Current Source for ADIM Pin		9	12	15	μA	

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Electrical Characteristics (Continued)

Typical values are at $T_A = +25^{\circ}$ C. Specifications to -40° C ~ 125° C are guaranteed by design based on final characterization results.

Symbol	Parameter	Condition	Min.	Тур.	Max.	Unit
MOSFET S	MOSFET Section					
BV _{DSS}	Breakdown Voltage	V _{CC} =0V, I _D =250µA	550			V
I _{LKMOS}	Internal MOSFET Leakage Current	V _{DS} =550V _{DC} , V _{GS} =0V			250	μA
R _{ON(ON)}	Drain-Source On Resistance ⁽³⁾	V_{GS} =10V, V_{DGS} =0V, T_{C} =25°C		7.3	10.0	Ω
C _{ISS}	Input Capacitance ⁽³⁾	V _{GS} =0V,V _{DS} =25V, f=1MHz		135		pF
C _{OSS}	Output Capacitance ⁽³⁾	V _{GS} =0V,V _{DS} =25V, f=1MHz		21		pF
C _{RSS}	Reverse Transfer Capacitance ⁽³⁾	V _{GS} =0V,V _{DS} =25V, f=1MHz		3.2		pF
t _{d(ON)}	Turn-On Delay ⁽³⁾	V _{DD} =350V, I _D =1A		10		ns
t _r	Rise Time ⁽³⁾	V _{DD} =350V, I _D =1A		13.4		ns
t _{d(OFF)}	Turn-Off Delay ⁽³⁾	V _{DD} =350V, I _D =1A		14.9		ns
t _f	Fall Time ⁽³⁾	V _{DD} =350V, I _D =1A		36.8		ns

Note:

3. These parameters, although guaranteed, are not 100% tested in production.

Functional Description

The FLS0116 is a basic PWM controller for buck converter topology in Continuous Conduction Mode (CCM) with an intelligent PFC function that uses a digital control algorithm. An internal self-biasing circuit uses the high-voltage switching device. The IC does not need an auxiliary powering path to the VCC pin typical in flyback control ICs or PSR product family.

When the input voltage applied to the HV pin is within operating range (25V to 500V), the FLS0116 maintains a 15.5V DC voltage at the VCC pin for stable operation. The UVLO block functions such that when the V_{CC} voltage rises higher than V_{CCST+} , the internal UVLO block releases and starts operation. Otherwise, the V_{CC} goes down to the V_{CCST-} and IC operation stops. Normally, the hysteresis function provides stable operation even if the input voltage is operating under very noisy or unstable circumstances.

The FLS0116 has a "smart" internal digital block for determining input condition: AC or DC. When an AC source with 50Hz or 60Hz is applied to the IC, the IC automatically changes its internal reference signal, which is similar to input signal, for creating high power factor. When a DC source connects to the IC, the internal reference immediately changes to DC.

Soft-Start Function

The FLS0116 has an internal soft-start function to reduce inrush current at startup. When the IC starts operation following an internal sequence, the internal reference slowly increases for a pre-determined fixed time. After this transient period, the internal reference goes to a steady-state level. In this time, the IC continually tries to find phase information from the VCC pin. If the IC succeeds in getting phase information, it automatically follows a similar shape reference made during the transient times, 7 periods. If not, the IC has a DC reference level.

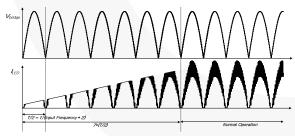


Figure 4. Soft-Start Function in AC Input Mode

Internal PFC Function: How to Achieve High Power Factor

The FLS0116 has a simple, "smart", internal PFC function that does not require additional pins for detecting input phase information or an electrolytic capacitor for supply voltage stabilization. For achieving high PF, the FLS0116 does not use the rectification capacitor after the bridge diode. This is important because the IC instead uses fluctuation in the signal on the VCC pin. Basically, the VCC pin, which is supplies

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power for the IC, has voltage ripple as well as the rectification voltage after bridge, changing voltage level according to the V_{CC} capacitor value. Using this kind of voltage fluctuation on the VCC pin, the IC can detect the time reference and create the internal ZCD signal.

For precise and reliable internal reference for input voltage signal, the FLS0116 uses a digital technique (sigma/delta modulation) and creates a new internal signal (DAC_OUT) that has the same phase as the input voltage, as shown in Figure 5. This signal enters the final comparator and is compared with current information from the sensing resistor.

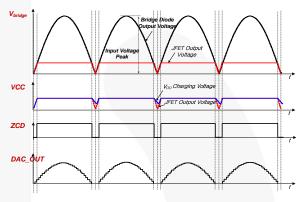


Figure 5. Internal PFC Function

Self-Biasing Function

The self-biasing function, using an HV device, can supply enough operating current to the IC and guarantee similar startup time across the whole input voltage range ($80V \sim 308V_{AC}$). However, self-biasing has a weakness in high-voltage condition. Normally, the HV device acts as constant current source, so the internal HV device has power loss when high input voltage connects to the HV pin. This power loss is proportional to input voltage. To reduce this power loss, one of the possible solutions is an additional resistor between the input voltage source and the HV pin, as shown in Figure 6.

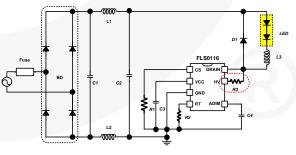
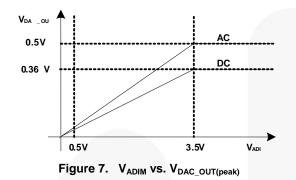


Figure 6. High-Voltage Application

Dimming Function

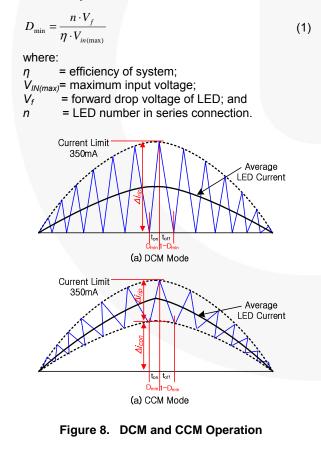
The FLS0116 uses the ADIM pin for analog or 0V to 10V dimming. The peak voltage of internal reference, which is DAC_OUT signal in Figure 5, is changed by the V_{ADIM} level, as shown in Figure 7, and has different peak level according to the operating mode.



Inductor Design

The fixed internal duty ratio range is below 50%, or around 400ns, from a timing point of view. The range is dependent on the input voltage and number of LEDs in its string.

Minimum duty is calculated as:



In DCM Mode, inductance is:

$$L_m = \frac{n \cdot V_f \cdot (1 - D_{\min})}{f_s \cdot \Delta i_{rip}} [H]$$
⁽²⁾

If the peak current is fixed at 350mApk, the formula for the peak current is:

$$I_{o(peak)} = \Delta i_{con} + \frac{\Delta i_{rip}}{2} = 350 \, [\text{mA}]$$
(3)

In CCM Mode, determine the necessary inductance using Equations (2) and (3). Assuming the high-frequency current ripple (Δi_{rip}) as in the DCM, the average LED current in CCM is defined as:

$$I_{LED(average)} = I_{LED(PK)} - \frac{\Delta i_{rip}}{2} \quad [mA]$$
(4)

LED RMS current determined as:

2 12

 $\sqrt{2}$

$$I_{LED(rms)} = \frac{1}{2} \left(\Delta i_{rip(rms)} + \Delta i_{con(rms)} \right) = \frac{1}{2} \left(\frac{I_{LED(PK)}}{\sqrt{2}} + \frac{I_{LED(PK)} - \Delta i_{rip}}{\sqrt{2}} \right)$$
(5)
= $\left(I_{LED(PK)} - \Delta i_{rip} \right)$ [mA] (6)

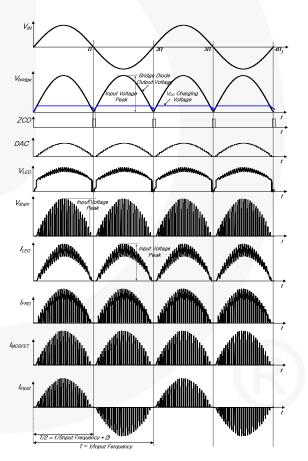
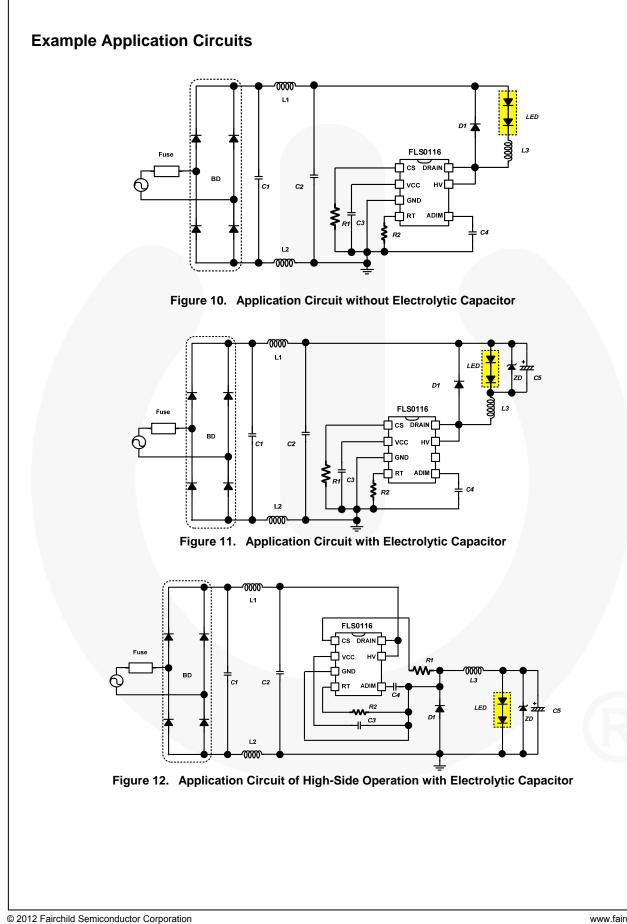


Figure 9. Typical Performance Characteristics

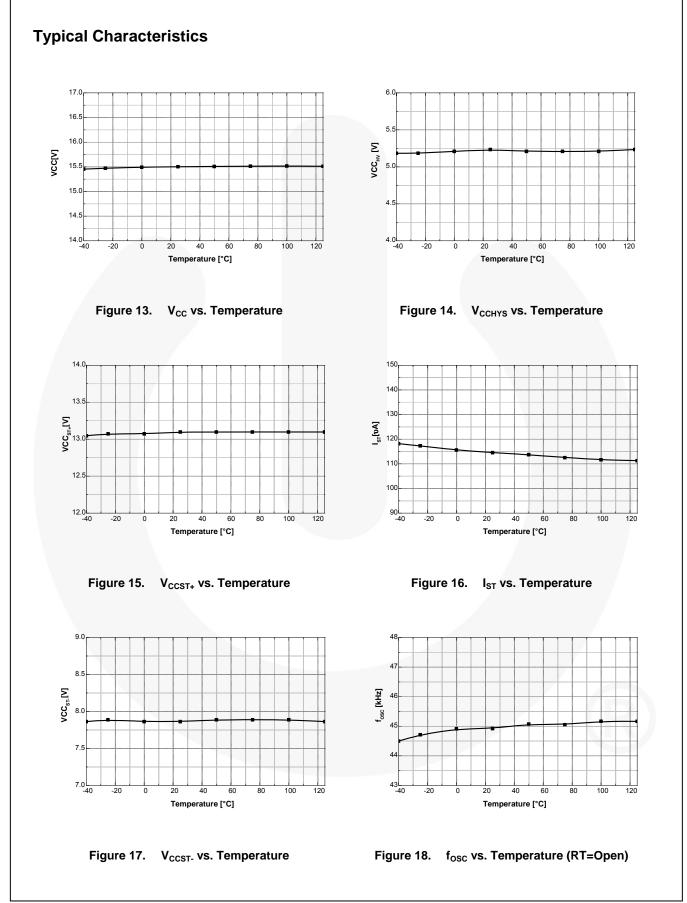
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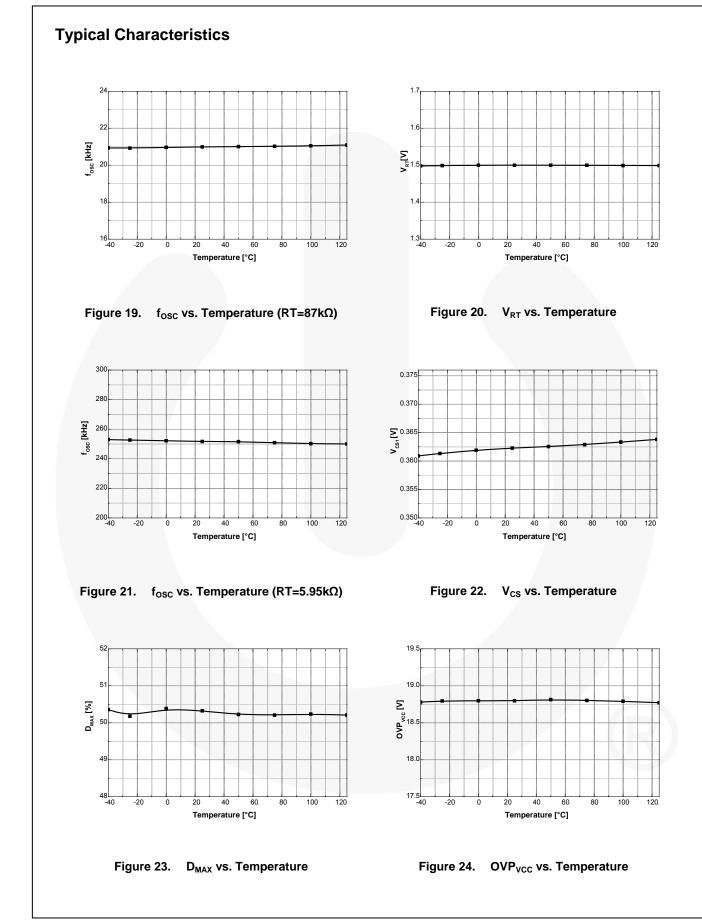
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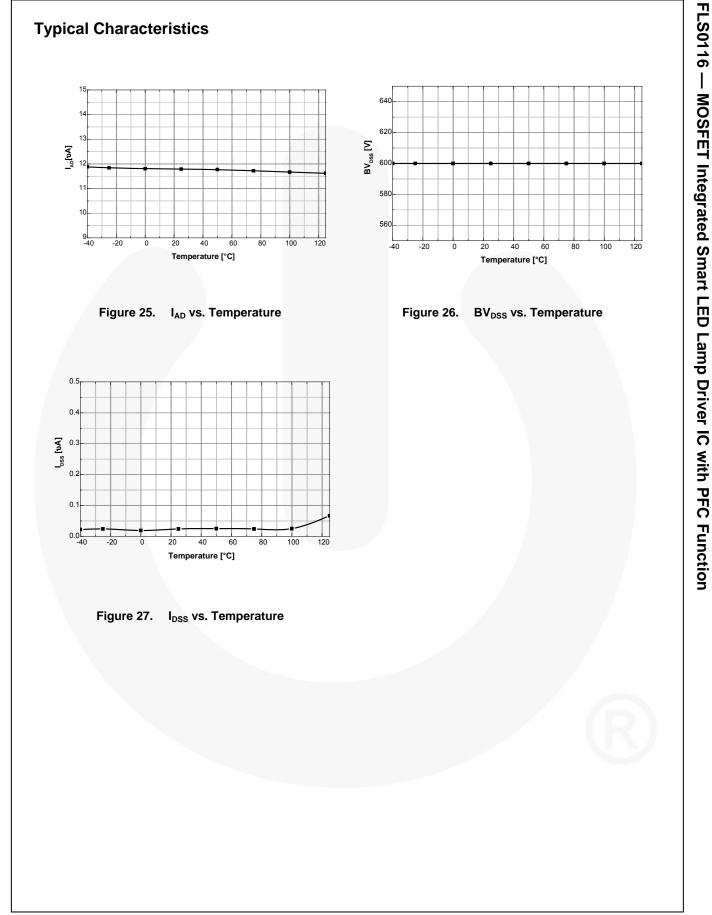
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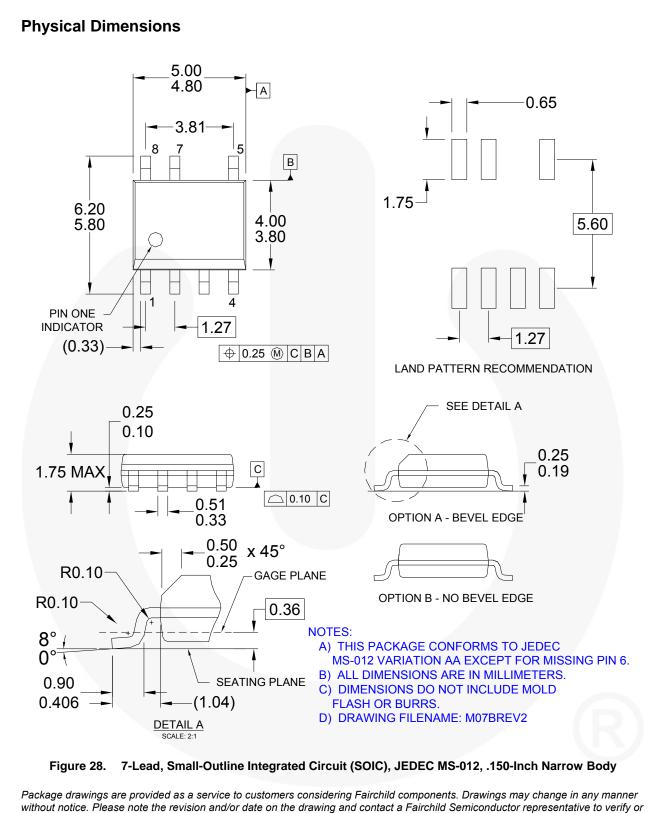


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