

# 24V/50mA General-Purpose Output-Driver



# **General Description**

The epc70x family is a general purpose low-side power switch for 24V interfaces. A high-side switch is also available, please refer to the separate data sheet of epc701/703. The device is very easy to use and capable to drive a load of 50mA at 30VDC. If a higher driving current is necessary or if the output voltage shall be higher than 30VDC, these chips can be used to drive an external power transistor. In this mode of operation, the external transistor is fully protected against over-current by the epc chip.

If the current through the external load exceeds a specified threshold during a longer time period than a predefined time, the output is turned off to protect the output switch. The switch is turned on again after a predefined off-time, thus enabling the load again in a self-healing mechanism. The over-current information is indicated on the STATUS pin.

epc700 and epc702 are easy to use and reduce the need of external components to the minimum, thus saving pcb space and money.

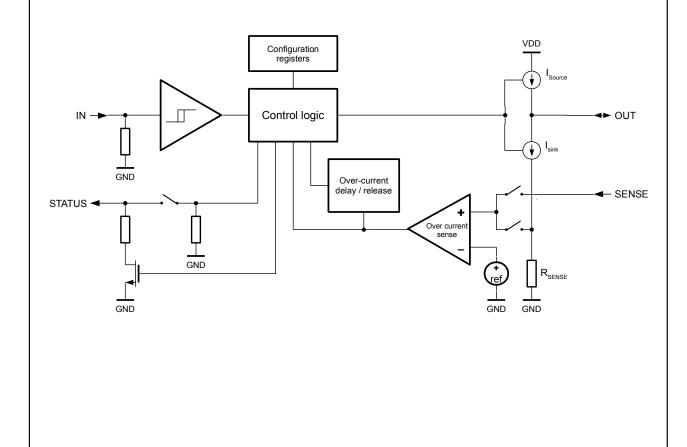
### **Features**

- Low-side power switch
- Driving capability without external transistor 50 mA/30VDC
- Programmable short-circuit detection delay-time and recovery time
   Static (epc700) or flashing (epc702) indication of the over-current
- status
  Self-healing output mode
- Available in CSP6 package with very small footprint and standard QFN16 package

# **Applications**

- PLC
- Sensors
- Controllers

# **Functional Block Diagram**





| Absolute Maximum Rat                              | Absolute Maximum Ratings (Note 1) |                            |      | Recommended Operating Conditions |       |  |  |  |  |  |  |
|---|-----------------------------------|----------------------------|------|----------------------------------|-------|--|--|--|--|--|--|
|   |                                   |                            | Min. | Max.                             | Units |  |  |  |  |  |  |
| Power Supply Voltage V <sub>DD</sub>              | -0.3 to +36.0 V (Note 2)          | Power Supply Voltage (VDD) | 9.6  | 30                               | V     |  |  |  |  |  |  |
| Maximum Power Dissipation                         | 100mW                             |                            |      |                                  |       |  |  |  |  |  |  |
| Storage Temperature Range (T <sub>s</sub> )       | -40°C to +85°C                    | Operating Temperature (To) | -40° | +85°                             | С     |  |  |  |  |  |  |
| Lead Temperature solder, 4 sec. (T <sub>L</sub> ) | +260°C                            | Humidity (non-condensing)  | +5   | +95                              | %     |  |  |  |  |  |  |

**Note 1:** Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Recommended operating conditions indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see Electrical Characteristics.

**Note 3:** This device is a highly sensitive CMOS amplifier with an ESD rating of JEDEC HBM class 1C (>1kV). Handling and assembly of this device should only be done at ESD protected workstations.

# **Electrical Characteristics**

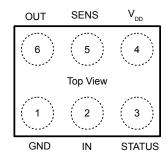
 $V_{DD} = 9.6V < V_{DD} < 30V, -40^{\circ}C < T_{A} < +85^{\circ}C$ 

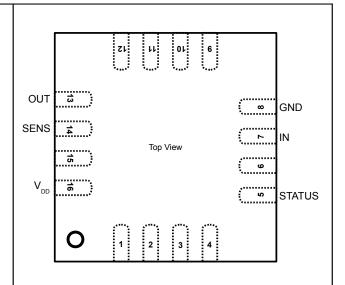
| Symbol               | Parameter                                | Conditions/Comments   |           | l,                   |            |                  |
|----------------------|--|---|-----------|----------------------|------------|------------------|
|                      |  |   | Min.      | Тур.                 | Max.       | Units            |
| $V_{DD}$             | Supply Voltage                           |   | 9.6       |                      | 30         | V                |
| $\Delta V_{DD}$      | Ripple on Supply Voltage                 | Peak-Peak   |           |                      | 10         | %V <sub>DD</sub> |
| I <sub>DD</sub>      | Supply Current                           | no load   |           | 300                  | 400        | μA               |
| $V_{\text{OUT}}$     | Output Voltage                           |   | 0         |                      | 30         | V                |
| $V_{Sat}$            | Output Saturation Voltage                | @50mA output current  |           | 1                    | 2          | V                |
| I <sub>SENS</sub>    | Sens Current                             | Current trigger threshold   | -50       | -60                  | -70        | mA               |
| $V_{\text{SENS}}$    | Current Sens Voltage                     | Over-current trigger threshold voltage (by using an external power switch)  | 0.18      | 0.2                  | 0.25       | V                |
| I <sub>Peak</sub>    | Short Circuit Peak Current               | Initial current during a short circuit (<1ms, 50Ω series resistor)  |           |                      | -0.5       | Α                |
| V <sub>Status</sub>  | Status Output                            | Logical high  | 2         |                      | 5.5        | .,               |
|                      |  | Logical low   | -0.3      |                      | 0.8        | V                |
|                      |  | Sink driving capability   | -8        | -10                  | -12        | mA               |
| f <sub>Status</sub>  | Status Output Frequency                  | epc702 only, duty cycle 50%   | 1.5       | 1.7                  | 1.9        | Hz               |
| V <sub>IN</sub>      | Input                                    | Logical high  | 2.0       |                      | 5.5        |                  |
|                      |  | Logical low   | -0.3      |                      | 0.8        | V                |
|                      |  | Hysteresis  | 0.25      |                      |            |                  |
|                      |  | Pull-down resistance  | 100       | 150                  | 200        | kΩ               |
| P <sub>DIS</sub>     | Power Dissipation                        | On-chip power dissipation   |           |                      | 100        | mW               |
| t <sub>on</sub>      | Response Time                            | On  |           | 1.0                  | 1.2        | μs               |
| t <sub>OFF</sub>     | Response Time                            | Off   |           | 0.7                  | 1.0        | μs               |
| $t_{\sf del}$        | Off-delay Time                           | Time between over-current detection and STATUS/OUT change (default value), refer to section Programming                                       | 40.0      | 50.0                 | 60.0       | μs               |
|                      |  | Programmable off-delay values   | 5/10/20/5 | 0/100/200            | /500/1,000 |                  |
| t <sub>minOFF</sub>  | Recovery Time                            | Minimum down time of the OUT pin to protect the external transistor (default value), refer to section Programming                             | 400       | 500                  | 600        | ms               |
|                      |  | Programmable values   | 10/20/50  | 0/100/200/5<br>1,500 | 500/1,000/ |                  |
| f <sub>rt</sub>      | Short circuit recovery delay time factor | $t_{\text{minorff}}/t_{\text{delay}} = f_{rt}$ (when used without external driver transistor, refer to section "Over-current Reset Sequence") | 1,000     |                      |            |                  |
| t <sub>STARTUP</sub> | Start-up time                            | VDD ramp > 100 V/ms   |           |                      | 200        | μs               |
| C <sub>L_max</sub>   | External Load Capacitance                | Load capacitance that can be driven through OUT without triggering over-current @2kOhm load and 5µs delay time                                |           | 30                   |            | nF               |

Note 2: Supply voltages up to 36 Volts may be present for 10 seconds only.



# **Connection Diagrams**





| 6-Pin | Chip | Scale | Pac | kage | (CSP) |  |
|-------|------|-------|-----|------|-------|--|
|-------|------|-------|-----|------|-------|--|

16-Pin QFN Package

| 6-Pin<br>CSP | 16-Pin<br>QFN       | Pin Name | Description   |
|--------------|---------------------|----------|---|
| 1            | 8                   | GND      | Negative power supply pin   |
| 2            | 7                   | IN       | Input from the controller   |
| 3            | 5                   | STATUS   | Output status signal to the controller  |
| 4            | 16                  | $V_{DD}$ | Positive power supply pin   |
| 5            | 14                  | SENS     | Input either to switch internal/external mode or to measure the voltage drop on an external shunt resistor to detect over-current |
| 6            | 13                  | OUT      | Output  |
| n/a          | 1-4, 6,<br>9-12, 15 | NC       | Not connected. Connect these pins to GND.   |

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# **Functional Description**

## **Normal Operation**

During typical operation the OUT follows the IN as shown in Figure 1. As long as the output current does not exceed the current threshold, OUT is stable ON and STATUS is stable HIGH-Z. The delay from IN to OUT is defined as  $t_{\text{ON}}$  and  $t_{\text{OFF}}$  for the rising and the falling edge respectively.

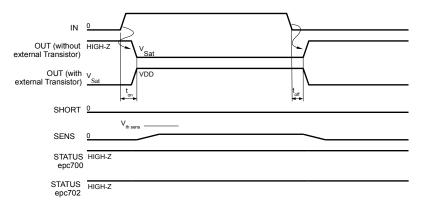


Figure 1: Normal operation

Please note that the status of the pin OUT is dependent whether the chip is operated with or without external driver transistor. In the subsequent diagrams, the version with external driver transistor is shown only.

### **Over-current Sequence**

A short on the load side will lead to an over-current through OUT. If an over-current stays longer than the time  $t_{\text{del}}$  and  $t_{\text{ext}}$ , OUT is turned off as shown in the figure below. At the same time STATUS changes its state to indicate an over-current situation to an external controller. STATUS can also be used to drive directly an indicator LED due to its 10mA driving capability. epc700 delivers a constant on-signal, whereas epc702 has a flashing output.

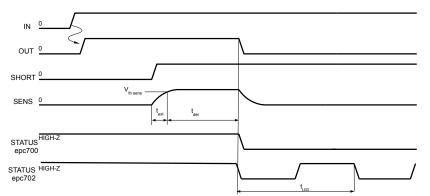


Figure 2: Short circuit detection

Note: The parameter  $t_{\text{ext}}$  used in this paper is described in Figure 7.



### **Current Peak at OUT**

A short current peak when OUT is turned on, typically generated by a capacitive load, could trigger the short-circuit protection logic. However, if the current peak is shorter than  $t_{\text{ext}}$  plus  $t_{\text{del}}$ , the over-current peak will be ignored.

Make sure that the energy drawn by such a current peak does not destroy the internal/external output transistor.

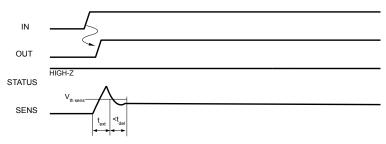


Figure 3: Short over-current pulse, i.e., by switching a capacitive load

### **Over-current Reset Sequence**

If there is a permanent short circuit at OUT, such short circuit will be detected and OUT is turned off. After a waiting time t<sub>minOFF</sub>, the device tries to turn on OUT again. If the short circuit is still present, OUT is immediately turned off again. This sequence continues until the short circuit is removed or IN goes to LOW or power is turned off. This mode is called self-healing since the device tries to self-heal the short circuit end to switch back into normal operation.

As a consequence in the case of a permanent over-current, short current peaks are issued into the load, respectively short.

The time  $t_{minOFF}$  has to be set to a value that the internal/external switch cannot be damaged by a too high power dissipation. Without an external switch, the time  $t_{minOFF}$  has to be 1,000 times longer than the time  $t_{del}$ .

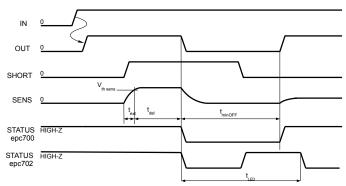


Figure 4: Over-current reset sequence

If one wants to reduce the waiting time in case of a short circuit situation, it can be done by simply taking IN to a low state and then to high again. Please refer to the timing diagram in Figure 5. However, make sure that the maximum power dissipation of the chip or the external switching transistor will never be exceeded above the allowed maximum.

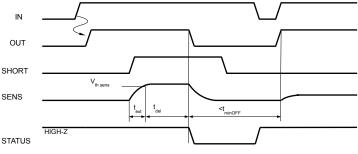


Figure 5: Over-current status reset



# **Application Information**

epc700 and epc702 have two modes of operation, where the SENS pin is used to define the mode. When SENS is tied to VDD, the chip operates as a sink driver capable to sink max. 50mA at 30VDC (refer to Figure 6). The load is connected directly between  $V_L$  and the OUT pin. If the current through the internal switch exceeds  $I_{SENS}$ , the switch is turned off.

If the SENS pin is at low level, the OUT pin is driven by a source driver also capable to drive 50mA into an external power transistor. This mode is used if the required output current has to be higher than 50mA, e.g. 1A and the output voltage exceeds 30VDC (refer to Figure 7).

The load current is measured by monitoring the voltage drop over a resistor. If the internal switch is used, also the current measurement resistor is located internally (Figure 6). In the case of using an external power transistor as shown in Figure 7, the current measurement resistor  $R_s$  has to be placed externally. If the voltage drop at  $R_s$  exceeds the threshold of 200 mV, the output stage is deactivated. The timing diagrams of the signals can be found in section "Functional Description".

The IN signal must be low during power-up  $(t_{STARTUP})$  for proper function of the chip. The epc70x has a built in pull down resistor, so not external active driving is needed during startup.

### epc700 or epc702 Using the Internal Switch

Figure 6 shows the epc700/702 in the mode using the internal switch. To enable this mode, the SENS pin has to be connected to VDD. Note that the VDD of the chip and  $V_L$  at the load can have a different value. The values for both VDD and  $V_L$  need to be between 9.6 and 30V.

The factor  $f_{rt}$  between minimum off-time and delay-time must be maintained in order not to damage the chip due to overheating. This factor has to be higher than 1,000. In the worst case scenario a peak current of approx. 0.5A is flowing from  $V_L$  at 30V into the chip with a  $t_{\text{Del}}$  set at 50 $\mu$ s if a short-circuit occurs. If the recovery time  $t_{\text{minoff}}$  in this case is smaller then 50ms, the average power dissipation would exceed the safe operation condition and the device will get damage d.

The diode  $D_1$  is to protect the internal switch against voltage surges when inductive loads are turned off.

R1 is to protect the internal switch in case of a short circuit on the load when a very low impedance power supply is used.

The voltage  $V_L$  can be higher than VDD in this configuration. However, it must not be above the maximum value of 'Supply Voltage' stated in the table Electrical Characteristics

# VDD V\_≤VDD max VDD epc700 epc700 epc702 IN STATUS SENSE GND GND

Figure 6: epc700 or epc702 using the internal switch to drive a load of up to 50mA/30VDC

### epc700 or epc702 Using an External Switching Transistor

Figure 7 shows the operation mode using an external switch  $T_1$  in order to extend the output current/voltage drive capability. In this example, a bipolar transistor is used, whereas the base current is limited by the resistor  $R_B$ . The maximum base current is 50mA. In order not to damage the chip, the user has to select the resistor  $R_B$  such that chip does not need to drive more than 50mA. Other possible switches are a NPN BJT or an n-channel MOSFET.

The load is turned on and off by setting the pin IN to high respectively to low level. When the load is turned on, the load current flows from  $V_{\rm L}$  through the resistor  $R_{\rm S}$  and through the transistor  $T_1$  to GND. This current creates a voltage drop over  $R_{\rm S}.$  The resulting voltage is applied to pin SENS, which measures the voltage drop. If it exceeds the threshold of an internal comparator, set to 200mV, the output is turned off after the given delay time  $t_{\rm del}.$ 

If the delay time should be extended to a value above the possible settings of  $t_{\text{del}}$  (refer to Table 3), an RC network can added, designated as  $R_{\text{T}}$  and  $C_{\text{T}}$  in Figure 7. The additional time delay  $t_{\text{ext}}$  can be calculated approx. as  $R_{\text{T}}$  x  $C_{\text{T}}$ . However, the time varies according to the current through  $R_{\text{S}}$ . This design concept is especially useful, when a large capacitor in the load path needs to be charged. The additional delay in the over-current detection helps in such a situation.

Note that the VDD of the chip and  $V_L$  on the load are different in most of the applications. The value of VDD must be between 9.6 and 30V.  $V_L$  instead, can be on a level which is appropriate to the external switching transistor.

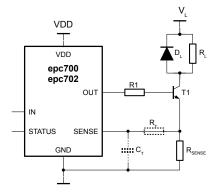


Figure 7: epc700 or epc702 operation mode using an external switching transistor.

In case of a short-circuit in the load the turn-off delay can be extended by an external RC network.

This network adds t<sub>ext</sub> to the internal delay t<sub>del</sub>.

The diode D<sub>1</sub> is to protect the transistor T<sub>1</sub> against voltage surges when inductive loads are turned off.



# **Programming**

The time delay ( $t_{del}$ ), until the output is turned off after the detection of an over-current condition can be programmed in order to adapt the timing to specific requirements, i.e., if a capacitive load has to be operated or an external transistor allows other values. The default value is 50µs which allows to charge a load capacitor of approx. 100 – 500nF without an external power transistor, dependent on the source impedance, the load impedance and the voltage  $V_{\perp}$ .

The time until the output is turned on again after a short circuit can be programmed as well ( $t_{minOFF}$ ). This "self-healing" mechanism is very useful because no operator interaction is necessary after a short circuit to enable normal function once the short circuit has been eliminated. The default value is 500ms which means that the device tries to turn the output on after 0.5s waiting time in the short circuit mode. This waiting time is recommended as long as  $t_{del}$  is not changed. If  $t_{del}$  has been changed, the parameter  $t_{minOFF}$  shall be changed accordingly in order to respect parameter  $t_{nl}$ . The user has to ensure that the maximum operation conditions never exceed in order to avoid damage of the device.

It is to note that the parameters programmed are stored in a non-volatile random access memory. Thus, the parameters can be lost after a power down for longer than 5ms (data retention time @ 25°C: min. 100ms). The corresponding requirements for safety applications have to take in consideration. Parameters can be changed as many times as necessary and even under operation to change the behavior of an output. During power-on, the default values are restored automatically.

### **Programming Interface**

The interface to store changed parameters are the pins IN and STATUS. IN is the chip select pin and STATUS, which is under normal operation an output, is used as an input pin. As long as the IN pin is at low state, parameters can be stored through the STATUS pin. Since IN is low during the programming of new parameters, OUT is low as well.

The digital input high threshold is typically at 2.2V, thus a 5V compatible communication. Please note that the voltage at STATUS should not exceed 5.5V.

### **Single Wire Communication Interface**

The epc70x is based on a single wire communication interface by using the STATUS pin. Programming is done by a 21-Bit Manchester code according to IEEE 802.4.

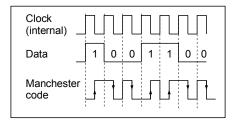


Figure 8: Manchester encoding sample

Figure 8 shows such a sample Manchester encoded data-stream. The clock and the corresponding data is used to generate the Manchester data-stream. Each positive clock-edge in the Manchester encoded data (indicated with the up-arrow) corresponds to a 1 and each negative clock-edge (indicated with the down-arrow) corresponds to a 0.

### **Data Clock Frequency Range**

The communication frequency range has to be according to Table 1.

|            | minimal | typical | maximum |  |  |
|------------|---------|---------|---------|--|--|
| Data clock | 396kHz  | 450kHz  | 540kHz  |  |  |

Table 1: Frequency range for programming

### Configuration Bit Stream for changing the delay time

In order to guarantee a reliable communication with the Manchester encoded bit stream on STATUS, some additional bits have been added to the configuration bits. Table 2 shows the digital pattern for the delay time configuration and the recovery time configuration.

| Bit #                     | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | 10 | 11 | 12 | 13 | 14 | 15 | 16 | 17 | 18 | 19 | 20 |
|---------------------------|---|---|---|---|---|---|---|---|---|---|----|----|----|----|----|----|----|----|----|----|----|
| t <sub>Del</sub> Value    | 1 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 1  | 1  | 1  | 0  | D1 | D2 | D3 | 1  | d1 | d2 | d3 |
| T <sub>minOFF</sub> Value | 1 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 1 | 0 | 0  | 0  | 1  | 0  | 01 | 02 | О3 | 1  | 01 | o2 | о3 |

Table 2: Configuration of the delay and the recovery time



The delay time is set with  $d_1$ ,  $d_2$ , and  $d_3$ . The bits  $D_1$ ,  $D_2$ , and  $D_3$  are the inverted values of  $d_1$ ,  $d_2$ , and  $d_3$ . Table 3 shows the value mapping table for the 8 different delay times.

| delay [µs] | D1 | D2 | D3 | d1 | d2 | d3 | Comments      |
|------------|----|----|----|----|----|----|---------------|
| 50         | 1  | 1  | 1  | 0  | 0  | 0  | Default value |
| 5          | 1  | 1  | 0  | 0  | 0  | 1  |               |
| 10         | 1  | 0  | 1  | 0  | 1  | 0  |               |
| 20         | 1  | 0  | 0  | 0  | 1  | 1  |               |
| 100        | 0  | 1  | 1  | 1  | 0  | 0  |               |
| 200        | 0  | 1  | 0  | 1  | 0  | 1  |               |
| 500        | 0  | 0  | 1  | 1  | 1  | 0  |               |
| 1000       | 0  | 0  | 0  | 1  | 1  | 1  |               |

Table 3: Delay-time programming table

An un-configured chip is applying a default delay time of 50µs corresponding to [d 1, d2, d3] = 000 and [D1, D2, D3] = 111.

The recovery time is set to  $o_1$ ,  $o_2$ , and  $o_3$ , resp.  $O_1$ ,  $O_2$ , and  $O_3$  which are the inverted values of  $o_1$ ,  $o_2$ , and  $o_3$ . Table 4 shows the value mapping table for the 8 different recovery time delay values.

| Recovery time [ms] | <b>O</b> <sub>1</sub> | O <sub>2</sub> | <b>O</b> <sub>3</sub> | <b>O</b> <sub>1</sub> | 02 | <b>O</b> <sub>3</sub> | Comments      |
|--------------------|-----------------------|----------------|-----------------------|-----------------------|----|-----------------------|---------------|
| 500                | 1                     | 1              | 1                     | 0                     | 0  | 0                     | Default value |
| 10                 | 1                     | 1              | 0                     | 0                     | 0  | 1                     |               |
| 20                 | 1                     | 0              | 1                     | 0                     | 1  | 0                     |               |
| 50                 | 1                     | 0              | 0                     | 0                     | 1  | 1                     |               |
| 100                | 0                     | 1              | 1                     | 1                     | 0  | 0                     |               |
| 200                | 0                     | 1              | 0                     | 1                     | 0  | 1                     |               |
| 1000               | 0                     | 0              | 1                     | 1                     | 1  | 0                     |               |
| 1500               | 0                     | 0              | 0                     | 1                     | 1  | 1                     |               |

Table 4: Recovery time programming table

An un-configured chip is applying the default recovery time of 500ms corresponding to [o 1, o2, o3] = 000 and [O1, O2, O3] = 111.

### **Programming Example**

An example for changing the delay time to  $100\mu s$  is shown in the following diagram:

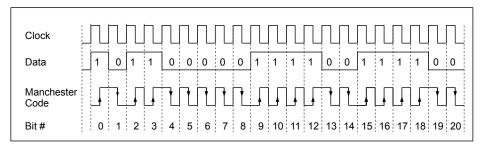
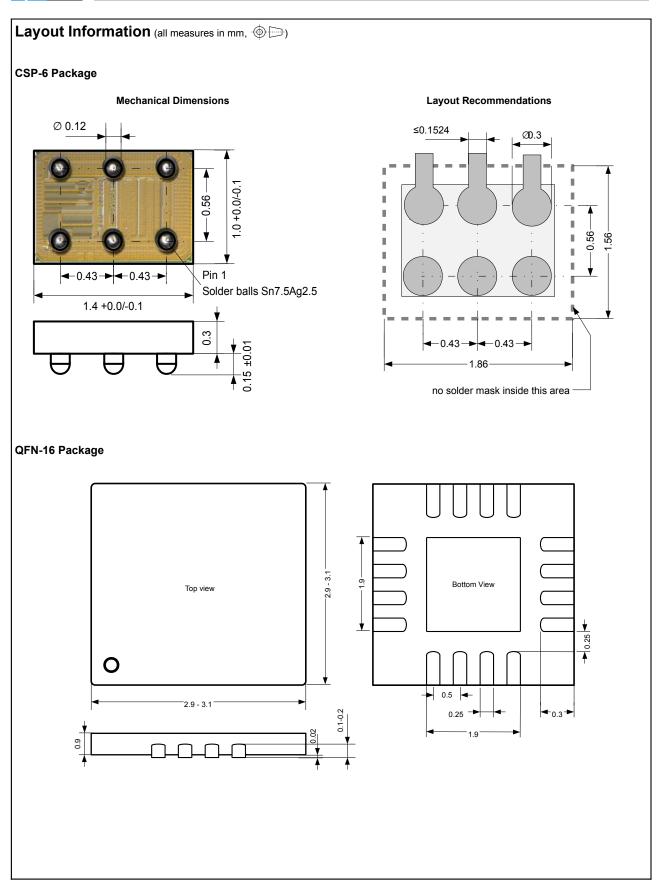


Figure 9: Programming example to set  $t_{del}$  to 100 $\mu$ s





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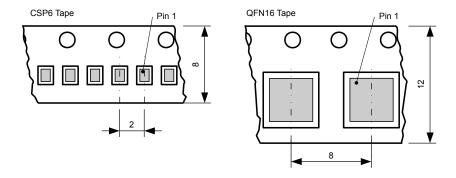


## **Reflow Solder Profile**

For infrared or conventional soldering the solder profile has to follow the recommendations of IPC/JEDEC J-STD-020C (min. revision C) for Pb-free assembly for both types of packages. The peak soldering temperature (T<sub>L</sub>) should not exceed +260°C for a maximum of 4 sec.

# Packaging Information (all measures in mm)

Tape & Reel Information
The devices are packaged into embossed tapes for automatic placement systems. The tape is wound on 178 mm (7 inch) or 330 mm (13 inch) reels and individually packaged for shipment. General tape-and-reel specification data are available in a separate data sheet and indicate the tape sizes for various package types. Further tape-and-reel specifications can be found in the Electronic Industries Association (EIA) standard 481-1, 481-2, 481-3.



epc does not guarantee that there are no empty cavities in the tape. Thus, the pick-and-place machine should check the presence of a chip during picking.

# **Ordering Information**

| Part Number  | Package | RoHS compliance | Packaging Method |
|--------------|---------|-----------------|------------------|
| epc700-CSP6  | CSP6    | Yes             | Reel             |
| epc700-QFN16 | QFN16   | Yes             | Reel             |
| epc701-CSP6  | CSP6    | Yes             | Reel             |
| epc701-QFN16 | QFN16   | Yes             | Reel             |
| epc702-CSP6  | CSP6    | Yes             | Reel             |
| epc702-QFN16 | QFN16   | Yes             | Reel             |
| epc703-CSP6  | CSP6    | Yes             | Reel             |
| epc703-QFN16 | QFN16   | Yes             | Reel             |



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