ST62T15C/T25C/E25C

## 8-BIT OTP/EPROM MCUs WITH A/D CONVERTER, OSCILLATOR SAFEGUARD, SAFE RESET AND 28 PINS

- 3.0 to 6.0V Supply Operating Range
- 8 MHz Maximum Clock Frequency
- -40 to $+125^{\circ} \mathrm{C}$ Operating Temperature Range
- Run, Wait and Stop Modes
- 5 Interrupt Vectors
- Look-up Table capability in Program Memory
- Data Storage in Program Memory:

User selectable size

- Data RAM: 64bytes
- User Programmable Options
- 20 I/O pins, fully programmable as:
- Input with pull-up resistor
- Input without pull-up resistor
- Input with interrupt generation
- Open-drain or push-pull output
- Analog Input
- 4 I/O lines can sink up to 20 mA to drive LEDs or TRIACs directly
- 8-bit Timer/Counter with 7-bit programmable prescaler
- Digital Watchdog
- Oscillator Safe Guard
- Low Voltage Detector for Safe Reset
- 8-bit A/D Converter with 16 analog inputs
- On-chip Clock oscillator can be driven by Quartz Crystal Ceramic resonator or RC network
- Power-on Reset
- One external Non-Maskable Interrupt
- ST626x-EMU2 Emulation and Development System (connects to an MS-DOS PC via a parallel port).
DEVICE SUMMARY

| DEVICE | OTP <br> (Bytes) | EPROM <br> (Bytes) | I/O Pins |
| :---: | :---: | :---: | :---: |
| ST62T15C | 1836 | - | 20 |
| ST62T25C | 3884 | - | 20 |
| ST62E25C |  | 3884 | 20 |



Rev. 2.8
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## 1 GENERAL DESCRIPTION

### 1.1 INTRODUCTION

The ST62T15C,T25C and ST62E25C devices are low cost members of the ST62xx 8-bit HCMOS family of microcontrollers, which is targeted at low to medium complexity applications. All ST62xx devices are based on a building block approach: a common core is surrounded by a number of onchip peripherals.
The ST62E25C is the erasable EPROM version of the ST62T15C,T25C device, which may be used to emulate the ST62T15C,T25C device, as well as the respective ST6215C,25C ROM devices.

OTP and EPROM devices are functionally identical. The ROM based versions offer the same functionality selecting as ROM options the options de-
fined in the programmable option bytes of the OTP/EPROM versions.

OTP devices offer all the advantages of user programmability at low cost, which make them the ideal choice in a wide range of applications where frequent code changes, multiple code versions or last minute programmability are required.

These compact low-cost devices feature a Timer comprising an 8-bit counter and a 7-bit programmable prescaler, an 8-bit A/D Converter with 16 analog inputs and a Digital Watchdog timer, making them well suited for a wide range of automotive, appliance and industrial applications.

Figure 1. Block Diagram


### 1.2 PIN DESCRIPTIONS

$\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{SS}}$. Power is supplied to the MCU via these two pins. $V_{D D}$ is the power connection and $V_{S S}$ is the ground connection.

OSCin and OSCout. These pins are internally connected to the on-chip oscillator circuit. A quartz crystal, a ceramic resonator or an external clock signal can be connected between these two pins. The OSCin pin is the input pin, the OSCout pin is the output pin.

RESET. The active-low RESET pin is used to restart the microcontroller. Internal pull-up is provided at this pin.
TEST/ $\mathbf{V}_{\text {Pp. }}$. The TEST must be held at $\mathrm{V}_{\text {SS }}$ for normal operation. If TEST pin is connected to a +12.5 V level during the reset phase, the EPROM programming Mode is entered.
NMI. The NMI pin provides the capability for asynchronous interruption, by applying an external non maskable interrupt to the MCU. The NMI input is falling edge sensitive. The user can select as option the availability of an on-chip pull-up at this pin.
TIMER. This is the timer I/O pin. In input mode it is connected to the prescaler and acts as external timer clock input or as control gate input for the internal timer clock. In output mode the timer pin outputs the data bit when a time-out occurs. The user can select as option the availability of an on-chip pull-up at this pin.

PAO-PA7. These 8 lines are organized as one I/O port (A). Each line may be configured under software control as inputs with or without internal pullup resistors, interrupt generating inputs with pullup resistors, open-drain or push-pull outputs. PAOPA3 can also sink 20mA for direct LED driving while PA4-PA7 can be programmed as analog in-
puts for the $A / D$ converter.
PB0-PB7. These 8 lines are organized as one I/O port (B). Each line may be configured under software control as inputs with or without internal pullup resistors, interrupt generating inputs with pullup resistors, open-drain or push-pull outputs, or as analog inputs for the A/D converter.
PC4-PC7. These 4 lines are organized as one I/O port (C). Each line may be configured under software control as inputs with or without internal pullup resistors, interrupt generating inputs with pullup resistors, open-drain or push-pull outputs, or as analog inputs for the $A / D$ converter.

Figure 2. ST62T15C,T25C and E25C Pin Configuration


### 1.3 MEMORY MAP

### 1.3.1 Introduction

The MCU operates in three separate memory spaces: Program space, Data space, and Stack space. Operation in these three memory spaces is described in the following paragraphs.

Briefly, Program space contains user program code in OTP and user vectors; Data space contains user data in RAM and in OTP, and Stack space accommodates six levels of stack for subroutine and interrupt service routine nesting.

Figure 3. Memory Addressing Diagram


## MEMORY MAP (Cont'd)

### 1.3.2 Program Space

Program Space comprises the instructions to be executed, the data required for immediate addressing mode instructions, the reserved factory test area and the user vectors. Program Space is addressed via the 12-bit Program Counter register (PC register) Program Memory Protection.
The Program Memory in OTP or EPROM devices can be protected against external readout of memory by selecting the READOUT PROTECTION option in the option byte.

In the EPROM parts, READOUT PROTECTION option can be disactivated only by U.V. erasure that also results into the whole EPROM context erasure.

Note: Once the Readout Protection is activated, it is no longer possible, even for STMicroelectronics, to gain access to the OTP contents. Returned parts with a protection set can therefore not be accepted.

Figure 4. Program Memory Map


## MEMORY MAP (Cont'd)

### 1.3.3 Data Space

Data Space accommodates all the data necessary for processing the user program. This space comprises the RAM resource, the processor core and peripheral registers, as well as read-only data such as constants and look-up tables in OTP/ EPROM

### 1.3.3.1 Data ROM

All read-only data is physically stored in program memory, which also accommodates the Program Space. The program memory consequently contains the program code to be executed, as well as the constants and look-up tables required by the application.
The Data Space locations in which the different constants and look-up tables are addressed by the processor core may be thought of as a 64-byte window through which it is possible to access the read-only data stored in OTP/EPROM.

### 1.3.3.2 Data RAM

In ST6225 devices, the data space includes 60 bytes of RAM, the accumulator ( A ), the indirect registers $(\mathrm{X}),(\mathrm{Y})$, the short direct registers $(\mathrm{V})$, (W), the I/O port registers, the peripheral data and control registers, the interrupt option register and the Data ROM Window register (DRW register).

### 1.3.4 Stack Space

Stack space consists of six 12-bit registers which are used to stack subroutine and interrupt return addresses, as well as the current program counter contents.

Table 1. ST6225 Data Memory Space

| RESERVED | 000h |
| :---: | :---: |
| DATA ROM WINDOW AREA 64 BYTES |  |
| X REGISTER | 080h |
| Y REGISTER | 081h |
| V REGISTER | 082h |
| W REGISTER | 083h |
| DATA RAM 60 BYTES |  |
| PORT A DATA REGISTER | OCOh |
| PORT B DATA REGISTER | 0 C 1 h |
| PORT C DATA REGISTER | 0C2h |
| RESERVED | 0C3h |
| PORT A DIRECTION REGISTER | 0C4h |
| PORT B DIRECTION REGISTER | 0C5h |
| PORT C DIRECTION REGISTER | 0C6h |
| RESERVED | 0C7h |
| INTERRUPT OPTION REGISTER | 0C8h* |
| DATA ROM WINDOW REGISTER | 0C9h* |
| RESERVED | $\begin{aligned} & \text { OCAh } \\ & \text { OCBh } \end{aligned}$ |
| PORT A OPTION REGISTER | 0CCh |
| PORT B OPTION REGISTER | 0CDh |
| PORT C OPTION REGISTER | OCEh |
| RESERVED | OCFh |
| A/D DATA REGISTER | ODOh |
| A/D CONTROL REGISTER | 0D1h |
| TIMER PRESCALER REGISTER | 0D2h |
| TIMER COUNTER REGISTER | 0D3h |
| TIMER STATUS CONTROL REGISTER | 0D4h |
| RESERVED | 0D6h |
| WATCHDOG REGISTER | 0D8h |
| RESERVED | 0FEh |
| ACCUMULATOR | OFFh |

## MEMORY MAP (Cont'd)

### 1.3.5 Data Window Register (DWR)

The Data read-only memory window is located from address 0040 h to address 007Fh in Data space. It allows direct reading of 64 consecutive bytes located anywhere in program memory, between address 0000h and 0FFFh (top memory address depends on the specific device). All the program memory can therefore be used to store either instructions or read-only data. Indeed, the window can be moved in steps of 64 bytes along the program memory by writing the appropriate code in the Data Window Register (DWR).
The DWR can be addressed like any RAM location in the Data Space, it is however a write-only register and therefore cannot be accessed using singlebit operations. This register is used to position the 64-byte read-only data window (from address 40h to address 7Fh of the Data space) in program memory in 64-byte steps. The effective address of the byte to be read as data in program memory is obtained by concatenating the 6 least significant bits of the register address given in the instruction (as least significant bits) and the content of the DWR register (as most significant bits), as illustrated in Figure 5 below. For instance, when addressing location 0040h of the Data Space, with 0 loaded in the DWR register, the physical location addressed in program memory is 00h. The DWR register is not cleared on reset, therefore it must be written to prior to the first access to the Data readonly memory window area.

## Data Window Register (DWR)

Address: 0C9h - Write Only


Bits 6, $7=$ Not used.
Bit 5-0 = DWR5-DWR0: Data read-only memory Window Register Bits. These are the Data readonly memory Window bits that correspond to the upper bits of the data read-only memory space.
Caution: This register is undefined on reset. Neither read nor single bit instructions may be used to address this register.
Note: Care is required when handling the DWR register as it is write only. For this reason, the DWR contents should not be changed while executing an interrupt service routine, as the service routine cannot save and then restore the register's previous contents. If it is impossible to avoid writing to the DWR during the interrupt service routine, an image of the register must be saved in a RAM location, and each time the program writes to the DWR, it must also write to the image register. The image register must be written first so that, if an interrupt occurs between the two instructions, the DWR is not affected.

Figure 5. Data read-only memory Window Memory Addressing


### 1.4 PROGRAMMING MODES

### 1.4.1 Option Bytes

The two Option Bytes allow configuration capability to the MCUs. Option byte's content is automatically read, and the selected options enabled, when the chip reset is activated.
It can only be accessed during the programming mode. This access is made either automatically (copy from a master device) or by selecting the OPTION BYTE PROGRAMMING mode of the programmer.
The option bytes are located in a non-user map. No address has to be specified.

## EPROM Code Option Byte (LSB)



## EPROM Code Option Byte (MSB)



D15-D10. Reserved. Must be cleared
EXTCNTL. External STOP MODE control.. When EXTCNTL is high, STOP mode is available with watchdog active by setting NMI pin to one.. When EXTCNTL is low, STOP mode is not available with the watchdog active.

LVD. LVD RESET enable.When this bit is set, safe RESET is performed by MCU when the supply voltage is too low. When this bit is cleared, only power-on reset or external RESET are active.
PROTECT. Readout Protection. This bit allows the protection of the software contents against piracy. When the bit PROTECT is set high, readout of the OTP contents is prevented by hardware.. When this bit is low, the user program can be read.
OSCIL. Oscillator selection. When this bit is low, the oscillator must be controlled by a quartz crystal, a ceramic resonator or an external frequency. When it is high, the oscillator must be controlled by an RC network, with only the resistor having to be externally provided.
D5-D4. Reserved. Must be cleared to zero.
NMI PULL. NMI Pull-Up. This bit must be set high to configure the NMI pin with a pull-up resistor. When it is low, no pull-up is provided.
TIM PULL.TIM Pull-Up. This bit must be set high to configure the TIMER pin with a pull-up resistor. When it is low, no pull-up is provided.
WDACT. This bit controls the watchdog activation. When it is high, hardware activation is selected. The software activation is selected when WDACT is low.
OSGEN. Oscillator Safe Guard. This bit must be set high to enable the Oscillator Safe Guard. When this bit is low, the OSG is disabled.
The Option byte is written during programming either by using the PC menu (PC driven Mode) or automatically (stand-alone mode)

## PROGRAMMING MODES (Cont'd)

### 1.4.2 Program Memory

EPROM/OTP programming mode is set by a +12.5 V voltage applied to the TEST/VPP pin. The programming flow of the ST62T15C,T25C/E25C is described in the User Manual of the EPROM Programming Board.

Table 2. ST62T15C Program Memory Map

| Device Address | Description |
| :---: | :---: |
| 0000h-087Fh | Reserved |
| 0880h-0F9Fh | User ROM |
| 0FAOh-OFEFh | Reserved |
| 0FFOh-0FF7h | Interrupt Vectors |
| 0FF8h-0FFBh | Reserved |
| OFFCh-0FFDh | NMI Interrupt Vector |
| 0FFEh-0FFFh | Reset Vector |

Table 3. ST62T25C,E25C Program Memory Map

| Device Address | Description |
| :---: | :---: |
| 0000h-007Fh | Reserved |
| 0080h-0F9Fh | User ROM |
| OFAOh-0FEFh | Reserved |
| OFFOh-0FF7h | Interrupt Vectors |
| OFF8h-0FFBh | Reserved |
| OFFCh-0FFDh | NMI Interrupt Vector |
| OFFEh-0FFFh | Reset Vector |

Note: OTP/EPROM devices can be programmed with the development tools available from STMicroelectronics (ST62E2X-EPB or ST622X-KIT).

### 1.4.3 EPROM Erasing

The EPROM of the windowed package of the MCUs may be erased by exposure to Ultra Violet light. The erasure characteristic of the MCUs is such that erasure begins when the memory is exposed to light with a wave lengths shorter than approximately $4000 \AA$. It should be noted that sunlights and some types of fluorescent lamps have wavelengths in the range $3000-4000 \AA$.

It is thus recommended that the window of the MCUs packages be covered by an opaque label to prevent unintentional erasure problems when testing the application in such an environment.
The recommended erasure procedure of the MCUs EPROM is the exposure to short wave ultraviolet light which have a wave-length 2537A. The integrated dose (i.e. U.V. intensity $x$ exposure time) for erasure should be a minimum of 30W$\mathrm{sec} / \mathrm{cm}^{2}$. The erasure time with this dosage is approximately 30 to 40 minutes using an ultraviolet lamp with $12000 \mu \mathrm{~W} / \mathrm{cm}^{2}$ power rating. The ST62E25C should be placed within 2.5 cm (1Inch) of the lamp tubes during erasure.

## 2 CENTRAL PROCESSING UNIT

### 2.1 INTRODUCTION

The CPU Core of ST6 devices is independent of the I/O or Memory configuration. As such, it may be thought of as an independent central processor communicating with on-chip I/O, Memory and Peripherals via internal address, data, and control buses. In-core communication is arranged as shown in Figure 6; the controller being externally linked to both the Reset and Oscillator circuits, while the core is linked to the dedicated on-chip peripherals via the serial data bus and indirectly, for interrupt purposes, through the control registers.

### 2.2 CPU REGISTERS

The ST6 Family CPU core features six registers and three pairs of flags available to the programmer. These are described in the following paragraphs.
Accumulator (A). The accumulator is an 8 -bit general purpose register used in all arithmetic calculations, logical operations, and data manipulations. The accumulator can be addressed in Data space as a RAM location at address FFh. Thus the ST6 can manipulate the accumulator just like any other register in Data space.

Figure 6. ST6 Core Block Diagram

Indirect Registers (X, Y). These two indirect registers are used as pointers to memory locations in Data space. They are used in the register-indirect addressing mode. These registers can be addressed in the data space as RAM locations at addresses $80 \mathrm{~h}(\mathrm{X})$ and $81 \mathrm{~h}(\mathrm{Y})$. They can also be accessed with the direct, short direct, or bit direct addressing modes. Accordingly, the ST6 instruction set can use the indirect registers as any other register of the data space.
Short Direct Registers (V, W). These two registers are used to save a byte in short direct addressing mode. They can be addressed in Data space as RAM locations at addresses $82 \mathrm{~h}(\mathrm{~V})$ and 83 h (W). They can also be accessed using the direct and bit direct addressing modes. Thus, the ST6 instruction set can use the short direct registers as any other register of the data space.
Program Counter (PC). The program counter is a 12-bit register which contains the address of the next ROM location to be processed by the core. This ROM location may be an opcode, an operand, or the address of an operand. The 12 -bit length allows the direct addressing of 4096 bytes in Program space.


## CPU REGISTERS (Cont'd)

However, if the program space contains more than 4096 bytes, the additional memory in program space can be addressed by using the Program Bank Switch register.
The PC value is incremented after reading the address of the current instruction. To execute relative jumps, the PC and the offset are shifted through the ALU, where they are added; the result is then shifted back into the PC. The program counter can be changed in the following ways:

- JP (Jump) instructionPC=Jump address
- CALL instructionPC= Call address
- Relative Branch Instruction.PC= PC +/- offset
- Interrupt PC=Interrupt vector
- Reset PC= Reset vector
- RET \& RETI instructionsPC= Pop (stack)
- Normal instructionPC= PC + 1

Flags (C, Z). The ST6 CPU includes three pairs of flags (Carry and Zero), each pair being associated with one of the three normal modes of operation: Normal mode, Interrupt mode and Non Maskable Interrupt mode. Each pair consists of a CARRY flag and a ZERO flag. One pair (CN, ZN) is used during Normal operation, another pair is used during Interrupt mode ( $\mathrm{Cl}, \mathrm{ZI}$ ), and a third pair is used in the Non Maskable Interrupt mode (CNMI, ZNMI ).
The ST6 CPU uses the pair of flags associated with the current mode: as soon as an interrupt (or a Non Maskable Interrupt) is generated, the ST6 CPU uses the Interrupt flags (resp. the NMI flags) instead of the Normal flags. When the RETI instruction is executed, the previously used set of flags is restored. It should be noted that each flag set can only be addressed in its own context (Non Maskable Interrupt, Normal Interrupt or Main routine). The flags are not cleared during context switching and thus retain their status.
The Carry flag is set when a carry or a borrow occurs during arithmetic operations; otherwise it is cleared. The Carry flag is also set to the value of the bit tested in a bit test instruction; it also participates in the rotate left instruction.
The Zero flag is set if the result of the last arithmetic or logical operation was equal to zero; otherwise it is cleared.
Switching between the three sets of flags is performed automatically when an NMI, an interrupt or a RETI instructions occurs. As the NMI mode is
automatically selected after the reset of the MCU, the ST6 core uses at first the NMI flags.
Stack. The ST6 CPU includes a true LIFO hardware stack which eliminates the need for a stack pointer. The stack consists of six separate 12-bit RAM locations that do not belong to the data space RAM area. When a subroutine call (or interrupt request) occurs, the contents of each level are shifted into the next higher level, while the content of the PC is shifted into the first level (the original contents of the sixth stack level are lost). When a subroutine or interrupt return occurs (RET or RETI instructions), the first level register is shifted back into the PC and the value of each level is popped back into the previous level. Since the accumulator, in common with all other data space registers, is not stored in this stack, management of these registers should be performed within the subroutine. The stack will remain in its "deepest" position if more than 6 nested calls or interrupts are executed, and consequently the last return address will be lost. It will also remain in its highest position if the stack is empty and a RET or RETI is executed. In this case the next instruction will be executed.

Figure 7. ST6 CPU Programming Mode


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## 3 CLOCKS, RESET, INTERRUPTS AND POWER SAVING MODES

### 3.1 CLOCK SYSTEM

The MCU features a Main Oscillator which can be driven by an external clock, or used in conjunction with an AT-cut parallel resonant crystal or a suitable ceramic resonator, or with an external resistor $\left(R_{\text {NET }}\right)$. In addition, a Low Frequency Auxiliary Oscillator (LFAO) can be switched in for security reasons, to reduce power consumption, or to offer the benefits of a back-up clock system.
The Oscillator Safeguard (OSG) option filters spikes from the oscillator lines, provides access to the LFAO to provide a backup oscillator in the event of main oscillator failure and also automatically limits the internal clock frequency ( $\mathrm{f}_{\mathrm{INT}}$ ) as a function of $\mathrm{V}_{\mathrm{DD}}$, in order to guarantee correct operation. These functions are illustrated in Figure 9, Figure 10, Figure 11 and Figure 12.
Figure 8 illustrates various possible oscillator configurations using an external crystal or ceramic resonator, an external clock input, an external resistor ( $\mathrm{R}_{\text {NET }}$ ), or the lowest cost solution using only the LFAO. $\mathrm{C}_{\mathrm{L} 1}$ an $\mathrm{C}_{\mathrm{L} 2}$ should have a capacitance in the range 12 to 22 pF for an oscillator frequency in the 4-8 MHz range.
The internal MCU clock frequency ( $\mathrm{f}_{\mathrm{INT}}$ ) is divided by 12 to drive the Timer, the A/D converter and the Watchdog timer, and by 13 to drive the CPU core, as may be seen in Figure 11.
With an 8 MHz oscillator frequency, the fastest machine cycle is therefore $1.625 \mu \mathrm{~s}$.
A machine cycle is the smallest unit of time needed to execute any operation (for instance, to increment the Program Counter). An instruction may require two, four, or five machine cycles for execution.

### 3.1.1 Main Oscillator

The oscillator configuration may be specified by selecting the appropriate option. When the CRYSTAL/ RESONATOR option isselected, itmustbeusedwith a quartz crystal, a ceramic resonator or an external signal provided onthe OSCinpin. When theRCNETWORK option is selected, the system clock is generated by an external resistor.
The main oscillator can be turned off (when the OSG ENABLED option is selected) by setting the OSCOFF bit of the ADC Control Register. The Low Frequency Auxiliary Oscillator is automatically started.

Figure 8. Oscillator Configurations


## CLOCK SYSTEM (Cont'd)

Turning on the main oscillator is achieved by resetting the OSCOFF bit of the A/D Converter Control Register or by resetting the MCU. Restarting the main oscillator implies a delay comprising the oscillator start up delay period plus the duration of the software instruction at flFAO clock frequency.

### 3.1.2 Low Frequency Auxiliary Oscillator (LFAO)

The Low Frequency Auxiliary Oscillator has three main purposes. Firstly, it can be used to reduce power consumption in non timing critical routines. Secondly, it offers a fully integrated system clock, without any external components. Lastly, it acts as a safety oscillator in case of main oscillator failure.
This oscillator is available when the OSG ENABLED option is selected. In this case, it automatically starts one of its periods after the first missing edge from the main oscillator, whatever the reason (main oscillator defective, no clock circuitry provided, main oscillator switched off...).
User code, normal interrupts, WAIT and STOP instructions, are processed as normal, at the reduced flFAO frequency. The A/D converter accuracy is decreased, since the internal frequency is below 1 MHz .
At power on, the Low Frequency Auxiliary Oscillator starts faster than the Main Oscillator. It therefore feeds the on-chip counter generating the POR delay until the Main Oscillator runs.
The Low Frequency Auxiliary Oscillator is automatically switched off as soon as the main oscillator starts.

## ADCR

Address: OD1h - Read/Write
7

| ADCR | ADCR | ADCR | ADCR | ADCR | OSC | ADCR | ADCR |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 7 | 6 | 5 | 4 | 3 | OFF | 1 | 0 |

Bit 7-3, 1-0= ADCR7-ADCR3, ADCR1-ADCR0: ADC Control Register. These bits are not used.
Bit 2 = OSCOFF. When low, this bit enables main oscillator to run. The main oscillator is switched off when OSCOFF is high.

### 3.1.3 Oscillator Safe Guard

The Oscillator Safe Guard (OSG) affords drastically increased operational integrity in ST62xx devices. The OSG circuit provides three basic func-
tions: it filters spikes from the oscillator lines which would result in over frequency to the ST62 CPU; it gives access to the Low Frequency Auxiliary Oscillator (LFAO), used to ensure minimum processing in case of main oscillator failure, to offer reduced power consumption or to provide a fixed frequency low cost oscillator; finally, it automatically limits the internal clock frequency as a function of supply voltage, in order to ensure correct operation even if the power supply should drop.
The OSG is enabled or disabled by choosing the relevant OSG option. It may be viewed as a filter whose cross-over frequency is device dependent.
Spikes on the oscillator lines result in an effectively increased internal clock frequency. In the absence of an OSG circuit, this may lead to an over frequency for a given power supply voltage. The OSG filters out such spikes (as illustrated in Figure 9). In all cases, when the OSG is active, the maximum internal clock frequency, $\mathrm{f}_{\mathrm{INT}}$, is limited to $\mathrm{f}_{\mathrm{OSG}}$, which is supply voltage dependent. This relationship is illustrated in Figure 12.
When the OSG is enabled, the Low Frequency Auxiliary Oscillator may be accessed. This oscillator starts operating after the first missing edge of the main oscillator (see Figure 10).
Over-frequency, at a given power supply level, is seen by the OSG as spikes; it therefore filters out some cycles in order that the internal clock frequency of the device is kept within the range the particular device can stand (depending on $\mathrm{V}_{\mathrm{DD}}$ ), and below $\mathrm{f}_{\mathrm{OSG}}$ : the maximum authorised frequency with OSG enabled.
Note. The OSG should be used wherever possible as it provides maximum safety. Care must be taken, however, as it can increase power consumption and reduce the maximum operating frequency to fosg.
Warning: Care has to be taken when using the OSG, as the internal frequency is defined between a minimum and a maximum value and is not accurate.
For precise timing measurements, it is not recommended to use the OSG and it should not be enabled in applications that use the SPI or the UART.
It should also be noted that power consumption in Stop mode is higher when the OSG is enabled (around $50 \mu \mathrm{~A}$ at nominal conditions and room temperature).

CLOCK SYSTEM (Cont'd)
Figure 9. OSG Filtering Principle


Figure 10. OSG Emergency Oscillator Principle


## CLOCK SYSTEM (Cont'd)

Figure 11. Clock Circuit Block Diagram


Figure 12. Maximum Operating Frequency ( $\mathrm{f}_{\mathrm{MAX}}$ ) versus Supply Voltage ( $\mathrm{V}_{\mathrm{DD}}$ )


Notes:

1. In this area, operation is guaranteed at the quartz crystal frequency.
2. When the OSG is disabled, operation in this area is guaranteed at the crystal frequency. When the OSG is enabled, operation in this area is guaranteed at a frequency of at least $\mathrm{f}_{\mathrm{OSG}}$ Min.
3. When the OSG is disabled, operation in this
area is guaranteed at the quartz crystal frequency. When the OSG is enabled, access to this area is prevented. The internal frequency is kept a $\mathrm{f}_{\mathrm{OSG}}$.
4. When the OSG is disabled, operation in this area is not guaranteed
When the OSG is enabled, access to this area is prevented. The internal frequency is kept at $f_{\text {OSG. }}$

### 3.2 RESETS

The MCU can be reset in four ways:

- by the external Reset input being pulled low;
- by Power-on Reset;
- by the digital Watchdog peripheral timing out.
- by Low Voltage Detection (LVD)


### 3.2.1 RESET Input

The RESET pin may be connected to a device of the application board in order to reset the MCU if required. The RESET pin may be pulled low in RUN, WAIT or STOP mode. This input can be used to reset the MCU internal state and ensure a correct start-up procedure. The pin is active low and features a Schmitt trigger input. The internal Reset signal is generated by adding a delay to the external signal. Therefore even short pulses on the RESET pin are acceptable, provided $V_{D D}$ has completed its rising phase and that the oscillator is running correctly (normal RUN or WAIT modes). The MCU is kept in the Reset state as long as the RESET pin is held low.
If RESET activation occurs in the RUN or WAIT modes, processing of the user program is stopped (RUN mode only), the Inputs and Outputs are configured as inputs with pull-up resistors and the main Oscillator is restarted. When the level on the RESET pin then goes high, the initialization sequence is executed following expiry of the internal delay period.
If RESET pin activation occurs in the STOP mode, the oscillator starts up and all Inputs and Outputs are configured as inputs with pull-up resistors. When the level of the RESET pin then goes high, the initialization sequence is executed following expiry of the internal delay period.

### 3.2.2 Power-on Reset

The function of the POR circuit consists in waking up the MCU by detecting around 2 V a dynamic (rising edge) variation of the VDD Supply. At the beginning of this sequence, the MCU is configured in the Reset state: all I/O ports are configured as inputs with pull-up resistors and no instruction is executed. When the power supply voltage rises to a sufficient level, the oscillator starts to operate, whereupon an internal delay is initiated, in order to allow the oscillator to fully stabilize before executing the first instruction. The initialization sequence
is executed immediately following the internal delay.
To ensure correct start-up, the user should take care that the VDD Supply is stabilized at a sufficient level for the chosen frequency (see recommended operation) before the reset signal is released. In addition, supply rising must start from 0 V .
As a consequence, the POR does not allow to supervise static, slowly rising, or falling, or noisy (presenting oscillation) VDD supplies.
An external RC network connected to the RESET pin, or the LVD reset can be used instead to get the best performances.

Figure 13. Reset and Interrupt Processing


## RESETS (Cont'd)

### 3.2.3 Watchdog Reset

The MCU provides a Watchdog timer function in order to ensure graceful recovery from software upsets. If the Watchdog register is not refreshed before an end-of-count condition is reached, the internal reset will be activated. This, amongst other things, resets the watchdog counter.

The MCU restarts just as though the Reset had been generated by the RESET pin, including the built-in stabilisation delay period.

### 3.2.4 LVD Reset

The on-chip Low Voltage Detector, selectable as user option, features static Reset when supply voltage is below a reference value. Thanks to this feature, external reset circuit can be removed while keeping the application safety. This SAFE RESET is effective as well in Power-on phase as in power supply drop with different reference val-
ues, allowing hysteresis effect. Reference value in case of voltage drop has been set lower than the reference value for power-on in order to avoid any parasitic Reset when MCU start's running and sinking current on the supply.
As long as the supply voltage is below the reference value, there is a internal and static RESET command. The MCU can start only when the supply voltage rises over the reference value. Therefore, only two operating mode exist for the MCU: RESET active below the voltage reference, and running mode over the voltage reference as shown on the Figure 14, that represents a powerup, power-down sequence.
Note: When the RESET state is controlled by one of the internal RESET sources (Low Voltage Detector, Watchdog, Power on Reset), the RESET pin is tied to low logic level.

Figure 14. LVD Reset on Power-on and Power-down (Brown-out)


### 3.2.5 Application Notes

No external resistor is required between $\mathrm{V}_{\mathrm{DD}}$ and the Reset pin, thanks to the built-in pull-up device.

Direct external connection of the pin RESET to $V_{D D}$ must be avoided in order to ensure safe behaviour of the internal reset sources (AND.Wired structure).

## RESETS (Cont'd)

### 3.2.6 MCU Initialization Sequence

When a reset occurs the stack is reset, the PC is loaded with the address of the Reset Vector (located in program ROM starting at address OFFEh). A jump to the beginning of the user program must be coded at this address. Following a Reset, the Interrupt flag is automatically set, so that the CPU is in Non Maskable Interrupt mode; this prevents the initialisation routine from being interrupted. The initialisation routine should therefore be terminated by a RETI instruction, in order to revert to normal mode and enable interrupts. If no pending interrupt is present at the end of the initialisation routine, the MCU will continue by processing the instruction immediately following the RETI instruction. If, however, a pending interrupt is present, it will be serviced.

Figure 15. Reset and Interrupt Processing


Figure 16. Reset Block Diagram


## RESETS (Cont'd)

## Table 4. Register Reset Status

| Register | Address(es) | Status | Comment |
| :---: | :---: | :---: | :---: |
| Oscillator Control Register <br> Port Data Registers <br> Port Direction Register <br> Port Option Register Interrupt Option Register <br> TIMER Status/Control | oDCh <br> 0COh to 0C2h <br> 0C4h to 0C6h <br> OCCh to OCEh <br> 0C8h <br> 0D4h | 00h | $\mathrm{f}_{\text {INT }}=\mathrm{f}_{\text {Osc }} ;$ OSG disabled I/O are Input with pull-up I/O are Input with pull-up I/O are Input with pull-up Interrupt disabled TIMER disabled |
| X, Y, V, W, Register <br> Accumulator <br> Data RAM <br> Data ROM Window Register <br> A/D Result Register | 080H TO 083H <br> 0FFh <br> 084h to 0BFh <br> 0C9h <br> 0D0h | Undefined | As written if programmed |
| TIMER Counter Register TIMER Prescaler Register Watchdog Counter Register A/D Control Register | 0D3h <br> 0D2h <br> 0D8h <br> 0D1h | $\begin{aligned} & \text { FFh } \\ & 7 \text { Fh } \\ & \text { FEh } \\ & 40 \mathrm{~h} \end{aligned}$ | Max count loaded <br> A/D in Standby |

### 3.3 DIGITAL WATCHDOG

The digital Watchdog consists of a reloadable downcounter timer which can be used to provide controlled recovery from software upsets.
The Watchdog circuit generates a Reset when the downcounter reaches zero. User software can prevent this reset by reloading the counter, and should therefore be written so that the counter is regularly reloaded while the user program runs correctly. In the event of a software mishap (usually caused by externally generated interference), the user program will no longer behave in its usual fashion and the timer register will thus not be reloaded periodically. Consequently the timer will decrement down to 00h and reset the MCU. In order to maximise the effectiveness of the Watchdog function, user software must be written with this concept in mind.
Watchdog behaviour is governed by two options, known as "WATCHDOG ACTIVATION" (i.e. HARDWARE or SOFTWARE) and "EXTERNAL STOP MODE CONTROL" (see Table 5).
In the SOFTWARE option, the Watchdog is disabled until bit C of the DWDR register has been set.

When the Watchdog is disabled, low power Stop mode is available. Once activated, the Watchdog cannot be disabled, except by resetting the MCU.
In the HARDWARE option, the Watchdog is permanently enabled. Since the oscillator will run continuously, low power mode is not available. The STOP instruction is interpreted as a WAIT instruction, and the Watchdog continues to countdown.
However, when the EXTERNAL STOP MODE CONTROL option has been selected low power consumption may be achieved in Stop Mode.
Execution of the STOP instruction is then governed by a secondary function associated with the NMI pin. If a STOP instruction is encountered when the NMI pin is low, it is interpreted as WAIT, as described above. If, however, the STOP instruction is encountered when the NMI pin is high, the Watchdog counter is frozen and the CPU enters STOP mode.
When the MCU exits STOP mode (i.e. when an interrupt is generated), the Watchdog resumes its activity.

Table 5. Recommended Option Choices

| Functions Required | Recommended Options |
| :---: | :---: |
| Stop Mode \& Watchdog | "EXTERNAL STOP MODE" \& "HARDWARE WATCHDOG" |
| Stop Mode | "SOFTWARE WATCHDOG" |
| Watchdog | "HARDWARE WATCHDOG" |

## DIGITAL WATCHDOG (Cont'd)

The Watchdog is associated with a Data space register (Digital WatchDog Register, DWDR, location 0D8h) which is described in greater detail in Section 3.3.1 Digital Watchdog Register (DWDR). This register is set to OFEh on Reset: bit C is cleared to " 0 ", which disables the Watchdog; the timer downcounter bits, TO to T5, and the SR bit are all set to " 1 ", thus selecting the longest Watchdog timer period. This time period can be set to the user's requirements by setting the appropriate value for bits T0 to T5 in the DWDR register. The SR bit must be set to " 1 ", since it is this bit which generates the Reset signal when it changes to " 0 "; clearing this bit would generate an immediate Reset.
It should be noted that the order of the bits in the DWDR register is inverted with respect to the associated bits in the down counter: bit 7 of the DWDR register corresponds, in fact, to T0 and bit 2 to T5. The user should bear in mind the fact that these bits are inverted and shifted with respect to the physical counter bits when writing to this register. The relationship between the DWDR register bits and the physical implementation of the Watchdog timer downcounter is illustrated in Figure 17.
Only the 6 most significant bits may be used to define the time period, since it is bit 6 which triggers the Reset when it changes to " 0 ". This offers the user a choice of 64 timed periods ranging from 3,072 to 196,608 clock cycles (with an oscillator frequency of 8 MHz , this is equivalent to timer periods ranging from $384 \mu$ s to 24.576 ms ).

Figure 17. Watchdog Counter Control


## DIGITAL WATCHDOG (Cont'd)

### 3.3.1 Digital Watchdog Register (DWDR)

Address: 0D8h — Read/Write
Reset status: 1111 1110b


Bit $0=\mathbf{C}$ : Watchdog Control bit
If the hardware option is selected, this bit is forced high and the user cannot change it (the Watchdog is always active). When the software option is selected, the Watchdog function is activated by setting bit C to 1 , and cannot then be disabled (save by resetting the MCU).
When C is kept low the counter can be used as a 7 -bit timer.
This bit is cleared to " 0 " on Reset.
Bit 1 = SR: Software Reset bit
This bit triggers a Reset when cleared.
When $\mathrm{C}=$ " 0 " (Watchdog disabled) it is the MSB of the 7-bit timer.
This bit is set to " 1 " on Reset.
Bits 2-7 = T5-T0: Downcounter bits
It should be noted that the register bits are reversed and shifted with respect to the physical counter: bit-7 (TO) is the LSB of the Watchdog downcounter and bit-2 (T5) is the MSB.
These bits are set to " 1 " on Reset.

### 3.3.2 Application Notes

The Watchdog plays an important supporting role in the high noise immunity of ST62xx devices, and should be used wherever possible. Watchdog related options should be selected on the basis of a trade-off between application security and STOP mode availability.
When STOP mode is not required, hardware activation without EXTERNAL STOP MODE CONTROL should be preferred, as it provides maximum security, especially during power-on.
When STOP mode is required, hardware activation and EXTERNAL STOP MODE CONTROL should be chosen. NMI should be high by default, to allow STOP mode to be entered when the MCU is idle.
The NMI pin can be connected to an I/O line (see Figure 18) to allow its state to be controlled by software. The I/O line can then be used to keep NMI low while Watchdog protection is required, or to avoid noise or key bounce. When no more processing is required, the I/O line is released and the device placed in STOP mode for lowest power consumption.
When software activation is selected and the Watchdog is not activated, the downcounter may be used as a simple 7-bit timer (remember that the bits are in reverse order).
The software activation option should be chosen only when the Watchdog counter is to be used as a timer. To ensure the Watchdog has not been unexpectedly activated, the following instructions should be executed within the first 27 instructions:

```
jrr 0, WD, #+3
ldi WD, OFDH
```


## DIGITAL WATCHDOG (Cont'd)

These instructions test the C bit and Reset the MCU (i.e. disable the Watchdog) if the bit is set (i.e. if the Watchdog is active), thus disabling the Watchdog.
In all modes, a minimum of 28 instructions are executed after activation, before the Watchdog can generate a Reset. Consequently, user software should load the watchdog counter within the first 27 instructions following Watchdog activation (software mode), or within the first 27 instructions executed following a Reset (hardware activation). It should be noted that when the GEN bit is low (interrupts disabled), the NMI interrupt is active but cannot cause a wake up from STOP/WAIT modes.

Figure 19. Digital Watchdog Block Diagram

Figure 18. A typical circuit making use of the EXERNAL STOP MODE CONTROL feature



### 3.4 INTERRUPTS

The CPU can manage four Maskable Interrupt sources, in addition to a Non Maskable Interrupt source (top priority interrupt). Each source is associated with a specific Interrupt Vector which contains a Jump instruction to the associated interrupt service routine. These vectors are located in Program space (see Table 6).
When an interrupt source generates an interrupt request, and interrupt processing is enabled, the PC register is loaded with the address of the interrupt vector (i.e. of the Jump instruction), which then causes a Jump to the relevant interrupt service routine, thus servicing the interrupt.
Interrupt sources are linked to events either on external pins, or on chip peripherals. Several events can be ORed on the same interrupt source, and relevant flags are available to determine which event triggered the interrupt.
The Non Maskable Interrupt request has the highest priority and can interrupt any interrupt routine at any time; the other four interrupts cannot interrupt each other. If more than one interrupt request is pending, these are processed by the processor core according to their priority level: source \#1 has the higher priority while source \#4 the lower. The priority of each interrupt source is fixed.

## Table 6. Interrupt Vector Map

| Interrupt Source | Priority | Vector Address |
| :--- | :---: | :---: |
| Interrupt source \#0 | 1 | (FFCh-FFDh) |
| Interrupt source \#1 | 2 | (FF6h-FF7h) |
| Interrupt source \#2 | 3 | (FF4h-FF5h) |
| Interrupt source \#3 | 4 | (FF2h-FF3h) |
| Interrupt source \#4 | 5 | (FF0h-FF1h) |

### 3.4.1 Interrupt request

All interrupt sources but the Non Maskable Interrupt source can be disabled by setting accordingly the GEN bit of the Interrupt Option Register (IOR). This GEN bit also defines if an interrupt source, including the Non Maskable Interrupt source, can restart the MCU from STOP/WAIT modes.
Interrupt request from the Non Maskable Interrupt source \#0 is latched by a flip flop which is automat-
ically reset by the core at the beginning of the nonmaskable interrupt service routine.
Interrupt request from source \#1 can be configured either as edge or level sensitive by setting accordingly the LES bit of the Interrupt Option Register (IOR).
Interrupt request from source \#2 are always edge sensitive. The edge polarity can be configured by setting accordingly the ESB bit of the Interrupt Option Register (IOR).
Interrupt request from sources \#3 \& \#4 are level sensitive.
In edge sensitive mode, a latch is set when a edge occurs on the interrupt source line and is cleared when the associated interrupt routine is started. So, the occurrence of an interrupt can be stored, until completion of the running interrupt routine before being processed. If several interrupt requests occurs before completion of the running interrupt routine, only the first request is stored.
Storage of interrupt requests is not available in level sensitive mode. To be taken into account, the low level must be present on the interrupt pin when the MCU samples the line after instruction execution.
At the end of every instruction, the MCU tests the interrupt lines: if there is an interrupt request the next instruction is not executed and the appropriate interrupt service routine is executed instead.

Table 7. Interrupt Option Register Description

| GEN | SET | Enable all interrupts |
| :--- | :--- | :--- |
|  | CLEARED | Disable all interrupts |
| ESB | SET | Rising edge mode on inter- <br> rupt source \#2 |
|  | CLEARED | Falling edge mode on inter- <br> rupt source \#2 |
|  | SET | Level-sensitive mode on in- <br> terrupt source \#1 |
|  | CLEARED | Falling edge mode on inter- <br> rupt source \#1 |
| OTHERS | NOT USED |  |

## INTERRUPTS (Cont'd)

### 3.4.2 Interrupt Procedure

The interrupt procedure is very similar to a call procedure, indeed the user can consider the interrupt as an asynchronous call procedure. As this is an asynchronous event, the user cannot know the context and the time at which it occurred. As a result, the user should save all Data space registers which may be used within the interrupt routines. There are separate sets of processor flags for normal, interrupt and non-maskable interrupt modes, which are automatically switched and so do not need to be saved.
The following list summarizes the interrupt procedure:

## MCU

- The interrupt is detected.
- The C and Z flags are replaced by the interrupt flags (or by the NMI flags).
- The PC contents are stored in the first level of the stack.
- The normal interrupt lines are inhibited (NMI still active).
- The first internal latch is cleared.
- Theassociated interrupt vector is loaded inthe PC.

WARNING: In some circumstances, when a maskable interrupt occurs while the ST6 core is in NORMAL mode and especially during the execution of an "Idi IOR, 00h" instruction (disabling all maskable interrupts): if the interrupt arrives during the first 3 cycles of the "Idi" instruction (which is a 4-cycle instruction) the core will switch to interrupt mode BUT the flags CN and ZN will NOT switch to the interrupt pair Cl and ZI .

## User

- User selected registers are saved within the interrupt service routine (normally on a software stack).
- The source of the interrupt is found by polling the interrupt flags (if more than one source is associated with the same vector).
- The interrupt is serviced.
- Return from interrupt (RETI)


## MCU

- Automatically the MCU switches back to the normal flag set (or the interrupt flag set) and pops the previous PC value from the stack.
The interrupt routine usually begins by the identifying the device which generated the interrupt request (by polling). The user should save the registers which are used within the interrupt routine in a software stack. After the RETI instruction is executed, the MCU returns to the main routine.

Figure 20. Interrupt Processing Flow Chart


## INTERRUPTS (Cont'd)

### 3.4.3 Interrupt Option Register (IOR)

The Interrupt Option Register (IOR) is used to enable/disable the individual interrupt sources and to select the operating mode of the external interrupt inputs. This register is write-only and cannot be accessed by single-bit operations.
Address: 0C8h — Write Only
Reset status: 00h


Bit 7, Bits 3-0 = Unused.
Bit 6 = LES: Level/Edge Selection bit.
When this bit is set to one, the interrupt source \#1 is level sensitive. When cleared to zero the edge sensitive mode for interrupt request is selected.

Bit 5 = ESB: Edge Selection bit.
The bit ESB selects the polarity of the interrupt source \#2.
Bit $4=$ GEN: Global Enable Interrupt. When this bit is set to one, all interrupts are enabled. When this bit is cleared to zero all the interrupts (excluding NMI ) are disabled.
When the GEN bit is low, the NMI interrupt is active but cannot cause a wake up from STOP/WAIT modes.
This register is cleared on reset.

### 3.4.4 Interrupt sources

Interrupt sources available on the ST62E25C/ T25C are summarized in the Table 8 with associated mask bit to enable/disable the interrupt request.

Table 8. Interrupt Requests and Mask Bits

| Peripheral | Register | Address <br> Register | Mask bit | Masked Interrupt Source | Interrupt <br> vector |
| :--- | :--- | :--- | :--- | :--- | :--- |
| GENERAL | IOR | C8h | GEN | All Interrupts, excluding NMI |  |
| TIMER | TSCR | D4h | ETI | TMZ: TIMER Overflow | Vector 3 |
| A/D CONVERTER | ADCR | D1h | EAI | EOC: End of Conversion | Vector 4 |
| Port PAn | ORPA-DRPA | C4h-CCh | ORPAn-DRPAn | PAn pin | Vector 1 |
| Port PBn | ORPB-DRPB | C5h-CDh | ORPBn-DRPBn | PBn pin | Vector 2 |
| Port PCn | ORPC-DRPC | C6h-CEh | ORPCn-DRPCn | PCn pin | Vector 2 |

## INTERRUPTS (Cont'd)

Figure 21. Interrupt Block Diagram


### 3.5 POWER SAVING MODES

The WAIT and STOP modes have been implemented in the ST62xx family of MCUs in order to reduce the product's electrical consumption during idle periods. These two power saving modes are described in the following paragraphs.
In addition, the Low Frequency Auxiliary Oscillator (LFAO) can be used instead of the main oscillator to reduce power consumption in RUN and WAIT modes.

### 3.5.1 WAIT Mode

The MCU goes into WAIT mode as soon as the WAIT instruction is executed. The microcontroller can be considered as being in a "software frozen" state where the core stops processing the program instructions, the RAM contents and peripheral registers are preserved as long as the power supply voltage is higher than the RAM retention voltage. In this mode the peripherals are still active.
WAIT mode can be used when the user wants to reduce the MCU power consumption during idle periods, while not losing track of time or the capability of monitoring external events. The active oscillator (main oscillator or LFAO) is not stopped in order to provide a clock signal to the peripherals. Timer counting may be enabled as well as the Timer interrupt, before entering the WAIT mode: this allows the WAIT mode to be exited when a Timer interrupt occurs. The same applies to other peripherals which use the clock signal.
If the power consumption has to be further reduced, the Low Frequency Auxiliary Oscillator (LFAO) can be used in place of the main oscillator, if its operating frequency is lower. If required, the LFAO must be switched on before entering the WAIT mode.

If the WAIT mode is exited due to a Reset (either by activating the external pin or generated by the Watchdog), the MCU enters a normal reset procedure. If an interrupt is generated during WAIT mode, the MCU's behaviour depends on the state of the processor core prior to the WAIT instruction, but also on the kind of interrupt request which is generated. This is described in the following paragraphs. The processor core does not generate a delay following the occurrence of the interrupt, because the oscillator clock is still available and no stabilisation period is necessary.

### 3.5.2 STOP Mode

If the Watchdog is disabled, STOP mode is available. When in STOP mode, the MCU is placed in the lowest power consumption mode. In this operating mode, the microcontroller can be considered as being "frozen", no instruction is executed, the oscillator is stopped, the RAM contents and peripheral registers are preserved as long as the power supply voltage is higher than the RAM retention voltage, and the ST62xx core waits for the occurrence of an external interrupt request or a Reset to exit the STOP state.
If the STOP state is exited due to a Reset (by activating the external pin) the MCU will enter a normal reset procedure. Behaviour in response to interrupts depends on the state of the processor core prior to issuing the STOP instruction, and also on the kind of interrupt request that is generated.
This case will be described in the following paragraphs. The processor core generates a delay after occurrence of the interrupt request, in order to wait for complete stabilisation of the oscillator, before executing the first instruction.

## POWER SAVING MODE (Cont'd)

### 3.5.3 Exit from WAIT and STOP Modes

The following paragraphs describe how the MCU exits from WAIT and STOP modes, when an interrupt occurs (not a Reset). It should be noted that the restart sequence depends on the original state of the MCU (normal, interrupt or non-maskable interrupt mode) prior to entering WAIT or STOP mode, as well as on the interrupt type.
Interrupts do not affect the oscillator selection, consequently, when the LFAO is used, the user program must manage oscillator selection as soon as normal RUN mode is resumed.

### 3.5.3.1 Normal Mode

If the MCU was in the main routine when the WAIT or STOP instruction was executed, exit from Stop or Wait mode will occur as soon as an interrupt occurs; the related interrupt routine is executed and, on completion, the instruction which follows the STOP or WAIT instruction is then executed, providing no other interrupts are pending.

### 3.5.3.2 Non Maskable Interrupt Mode

If the STOP or WAIT instruction has been executed during execution of the non-maskable interrupt routine, the MCU exits from the Stop or Wait mode as soon as an interrupt occurs: the instruction which follows the STOP or WAIT instruction is executed, and the MCU remains in non-maskable interrupt mode, even if another interrupt has been generated.

### 3.5.3.3 Normal Interrupt Mode

If the MCU was in interrupt mode before the STOP or WAIT instruction was executed, it exits from STOP or WAIT mode as soon as an interrupt occurs. Nevertheless, two cases must be considered:

- If the interrupt is a normal one, the interrupt routine in which the WAIT or STOP mode was en-
tered will be completed, starting with the execution of the instruction which follows the STOP or the WAIT instruction, and the MCU is still in the interrupt mode. At the end of this routine pending interrupts will be serviced in accordance with their priority.
- In the event of a non-maskable interrupt, the non-maskable interrupt service routine is processed first, then the routine in which the WAIT or STOP mode was entered will be completed by executing the instruction following the STOP or WAIT instruction. The MCU remains in normal interrupt mode.


## Notes:

To achieve the lowest power consumption during RUN or WAIT modes, the user program must take care of:

- configuring unused I/Os as inputs without pull-up (these should be externally tied to well defined logic levels);
- placing all peripherals in their power down modes before entering STOP mode;
- selecting the Low Frequency Auxiliary Oscillator (provided this runs at a lower frequency than the main oscillator).
When the hardware activated Watchdog is selected, or when the software Watchdog is enabled, the STOP instruction is disabled and a WAIT instruction will be executed in its place.
If all interrupt sources are disabled (GEN low), the MCU can only be restarted by a Reset. Although setting GEN low does not mask the NMI as an interrupt, it will stop it generating a wake-up signal.
The WAIT and STOP instructions are not executed if an enabled interrupt request is pending.


## 4 ON-CHIP PERIPHERALS

### 4.1 I/O PORTS

The MCU features Input/Output lines which may be individually programmed as any of the following input or output configurations:

- Input without pull-up or interrupt
- Input with pull-up and interrupt
- Input with pull-up, but without interrupt
- Analog input
- Push-pull output
- Open drain output

The lines are organised as bytewise Ports.
Each port is associated with 3 registers in Data space. Each bit of these registers is associated with a particular line (for instance, bits 0 of Port A Data, Direction and Option registers are associated with the PAO line of Port A).
The DATA registers (DRx), are used to read the voltage level values of the lines which have been configured as inputs, or to write the logic value of the signal to be output on the lines configured as outputs. The port data registers can be read to get the effective logic levels of the pins, but they can
be also written by user software, in conjunction with the related option registers, to select the different input mode options.
Single-bit operations on I/O registers are possible but care is necessary because reading in input mode is done from I/O pins while writing will directly affect the Port data register causing an undesired change of the input configuration.
The Data Direction registers (DDRx) allow the data direction (input or output) of each pin to be set.
The Option registers (ORx) are used to select the different port options available both in input and in output mode.
All I/O registers can be read or written to just as any other RAM location in Data space, so no extra RAM cells are needed for port data storage and manipulation. During MCU initialization, all I/O registers are cleared and the input mode with pull-ups and no interrupt generation is selected for all the pins, thus avoiding pin conflicts.

Figure 22. I/O Port Block Diagram


## I/O PORTS (Cont'd)

### 4.1.1 Operating Modes

Each pin may be individually programmed as input or output with various configurations.
This is achieved by writing the relevant bit in the Data (DR), Data Direction (DDR) and Option registers (OR). Table 9 illustrates the various port configurations which can be selected by user software.

### 4.1.1.1 Input Options

Pull-up, High Impedance Option. All input lines can be individually programmed with or without an internal pull-up by programming the OR and DR registers accordingly. If the pull-up option is not selected, the input pin will be in the high-impedance state.

### 4.1.1.2 Interrupt Options

All input lines can be individually connected by software to the interrupt system by programming the OR and DR registers accordingly. The interrupt trigger modes (falling edge, rising edge and low level) can be configured by software as described in the Interrupt Chapter for each port.

### 4.1.1.3 Analog Input Options

Some pins can be configured as analog inputs by programming the OR and DR registers accordingly. These analog inputs are connected to the onchip 8-bit Analog to Digital Converter. ONL Y ONE pin should be programmed as an analog input at any time, since by selecting more than one input simultaneously their pins will be effectively shorted.

Table 9. I/O Port Option Selection

| DDR | OR | DR | Mode |  |
| :---: | :---: | :---: | :---: | :--- |
| 0 | 0 | 0 | Input | With pull-up, no interrupt |
| 0 | 0 | 1 | Input | No pull-up, no interrupt |
| 0 | 1 | 0 | Input | With pull-up and with interrupt |
| 0 | 1 | 1 | Input | Analog input (when available) |
| 1 | 0 | $X$ | Output | Open-drain output (20mA sink when available) |
| 1 | 1 | $X$ | Output | Push-pull output (20mA sink when available) |

Note: X = Don't care

## I/O PORTS (Cont'd)

### 4.1.2 Safe I/O State Switching Sequence

Switching the I/O ports from one state to another should be done in a sequence which ensures that no unwanted side effects can occur. The recommended safe transitions are illustrated in Figure 23. All other transitions are potentially risky and should be avoided when changing the I/O operating mode, as it is most likely that undesirable sideeffects will be experienced, such as spurious interrupt generation or two pins shorted together by the analog multiplexer.
Single bit instructions (SET, RES, INC and DEC) should be used with great caution on Ports Data registers, since these instructions make an implicit read and write back of the entire register. In port input mode, however, the data register reads from the input pins directly, and not from the data register latches. Since data register information in input mode is used to set the characteristics of the input pin (interrupt, pull-up, analog input), these may be unintentionally reprogrammed depending on the state of the input pins. As a general rule, it is better to limit the use of single bit instructions on data registers to when the whole (8-bit) port is in output mode. In the case of inputs or of mixed inputs and
outputs, it is advisable to keep a copy of the data register in RAM. Single bit instructions may then be used on the RAM copy, after which the whole copy register can be written to the port data register:

```
SET bit, datacopy
LD a, datacopy
LD DRA, a
```

Warning: Care must also be taken to not use instructions that act on a whole port register (INC, DEC, or read operations) when all 8 bits are not available on the device. Unavailable bits must be masked by software (AND instruction).

The WAIT and STOP instructions allow the ST62xx to be used in situations where low power consumption is needed. The lowest power consumption is achieved by configuring I/Os in input mode with well-defined logic levels.
The user must take care not to switch outputs with heavy loads during the conversion of one of the analog inputs in order to avoid any disturbance to the conversion.

Figure 23. Diagram showing Safe I/O State Transitions


Note *. xxx = DDR, OR, DR Bits respectively

I/O PORTS (Cont'd)
4.1.3 I/O Port Option Registers

ORA/B/C (CCh PA, CDh PB, CEh PC)
Read/Write


Bit 7-0 = Px7-Px0: Port A, B and C Option Register bits.
4.1.4 I/O Port Data Direction Registers DDRA/B/C (C4h PA, C5h PB, C6h PC)
Read/Write


Bit 7-0 = Px7-Px0: Port A, B and C Data Direction Registers bits.

### 4.1.5 I/O Port Data Registers

 DRA/B/C (C0h PA, C1h PB, C2h PC) Read/Write7

| $P$ | 0 |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $P 7$ | $P x 6$ | $P x 5$ | $P x 4$ | $P x 3$ | $P x 2$ | $P x 1$ | $P x 0$ |

Bit 7-0 = Px7 - Px0: Port A, B and C Data Registers bits.
Note: X = Don't care

I/O PORTS (Cont'd)
Table 10. I/O Port Option Selections

| MODE | AVAILABLE ON ${ }^{(1)}$ | SCHEMATIC |
| :---: | :---: | :---: |
| Input | $\begin{aligned} & \text { PAO-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \end{aligned}$ |  |
| Input with pull up | $\begin{aligned} & \text { PAO-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \end{aligned}$ |  |
| Input with pull up with interrupt | $\begin{aligned} & \text { PAO-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \end{aligned}$ |  |
| Analog Input | $\begin{aligned} & \text { PA4-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \end{aligned}$ |  |
| Open drain output <br> 5 mA <br> Open drain output 20 mA | $\begin{aligned} & \text { PA4-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \\ & \text { PA0-PA3 } \end{aligned}$ |  |
| Push-pull output <br> 5 mA <br> Push-pull output 20 mA | $\begin{aligned} & \text { PA4-PA7 } \\ & \text { PB0-PB7 } \\ & \text { PC4-PC7 } \\ & \text { PA0-PA3 } \end{aligned}$ |  |

Note 1. Provided the correct configuration has been selected.

### 4.2 TIMER

The MCU features an on-chip Timer peripheral, consisting of an 8-bit counter with a 7-bit programmable prescaler, giving a maximum count of $2^{15}$. The peripheral may be configured in three different operating modes.
Figure 24 shows the Timer Block Diagram. The external TIMER pin is available to the user. The content of the 8-bit counter can be read/written in the Timer/Counter register, TCR, while the state of the 7-bit prescaler can be read in the PSC register. The control logic device is managed in the TSCR register as described in the following paragraphs.
The 8-bit counter is decremented by the output (rising edge) coming from the 7-bit prescaler and can be loaded and read under program control. When it decrements to zero then the TMZ (Timer Zero) bit in the TSCR is set to " 1 ". If the ETI (Enable Timer Interrupt) bit in the TSCR is also set to " 1 ", an interrupt request is generated as described in the Interrupt Chapter. The Timer interrupt can be used to exit the MCU from WAIT mode.

The prescaler input can be the internal frequency $f_{\text {INT }}$ divided by 12 or an external clock applied to the TIMER pin. The prescaler decrements on the rising edge. Depending on the division factor programmed by PS2, PS1 and PS0 bits in the TSCR. The clock input of the timer/counter register is multiplexed to different sources. For division factor 1, the clock input of the prescaler is also that of timer/ counter; for factor 2, bit 0 of the prescaler register is connected to the clock input of TCR. This bit changes its state at half the frequency of the prescaler input clock. For factor 4, bit 1 of the PSC is connected to the clock input of TCR, and so forth. The prescaler initialize bit, PSI, in the TSCR register must be set to " 1 " to allow the prescaler (and hence the counter) to start. If it is cleared to " 0 ", all the prescaler bits are set to " 1 " and the counter is inhibited from counting. The prescaler can be loaded with any value between 0 and 7Fh, if bit PSI is set to " 1 ". The prescaler tap is selected by means of the PS2/PS1/PS0 bits in the control register.

Figure 25 illustrates the Timer's working principle.

Figure 24. Timer Block Diagram


## TIMER (Cont'd)

### 4.2.1 Timer Operating Modes

There are three operating modes, which are selected by the TOUT and DOUT bits (see TSCR register). These three modes correspond to the two clocks which can be connected to the 7-bit prescaler ( $\mathrm{f}_{\mathrm{INT}} \div 12$ or TIMER pin signal), and to the output mode.

### 4.2.1.1 Gated Mode

(TOUT = " 0 ", DOUT = " 1 ")
In this mode the prescaler is decremented by the Timer clock input ( $\mathrm{f}_{\mathrm{INT}} \div 12$ ), but ONLY when the signal on the TIMER pin is held high (allowing pulse width measurement). This mode is selected by clearing the TOUT bit in the TSCR register to " 0 " (i.e. as input) and setting the DOUT bit to " 1 ".

### 4.2.1.2 Event Counter Mode

(TOUT = "0", DOUT = "0")
In this mode, the TIMER pin is the input clock of the prescaler which is decremented on the rising edge.

### 4.2.1.3 Output Mode

(TOUT = "1", DOUT = data out)
The TIMER pin is connected to the DOUT latch, hence the Timer prescaler is clocked by the prescaler clock input ( $\mathrm{f}_{\mathrm{INT}} \div 12$ ).

The user can select the desired prescaler division ratio through the PS2, PS1, PS0 bits. When the TCR count reaches 0 , it sets the TMZ bit in the TSCR. The TMZ bit can be tested under program control to perform a timer function whenever it goes high. The low-to-high TMZ bit transition is used to latch the DOUT bit of the TSCR and transfer it to the TIMER pin. This operating mode allows external signal generation on the TIMER pin.

Table 11. Timer Operating Modes

| TOUT | DOUT | Timer Pin | Timer Function |
| :---: | :---: | :--- | :--- |
| 0 | 0 | Input | Event Counter |
| 0 | 1 | Input | Gated Input |
| 1 | 0 | Output | Output "0" |
| 1 | 1 | Output | Output "1" |

### 4.2.2 Timer Interrupt

When the counter register decrements to zero with the ETI (Enable Timer Interrupt) bit set to one, an interrupt request is generated as described in the Interrupt Chapter. When the counter decrements to zero, the TMZ bit in the TSCR register is set to one.

Figure 25. Timer Working Principle


## TIMER (Cont'd)

### 4.2.3 Application Notes

The user can select the presence of an on-chip pull-up on the TIMER pin as option.
TMZ is set when the counter reaches zero; however, it may also be set by writing 00h in the TCR register or by setting bit 7 of the TSCR register. The TMZ bit must be cleared by user software when servicing the timer interrupt to avoid undesired interrupts when leaving the interrupt service routine. After reset, the 8 -bit counter register is loaded with OFFh, while the 7-bit prescaler is loaded with 07Fh, and the TSCR register is cleared. This means that the Timer is stopped ( $\mathrm{PSI}={ }^{\prime} 0$ ") and the timer interrupt is disabled.
If the Timer is programmed in output mode, the DOUT bit is transferred to the TIMER pin when TMZ is set to one (by software or due to counter decrement). When TMZ is high, the latch is transparent and DOUT is copied to the timer pin. When TMZ goes low, DOUT is latched.
A write to the TCR register will predominate over the 8 -bit counter decrement to 00 h function, i.e. if a write and a TCR register decrement to 00h occur simultaneously, the write will take precedence, and the TMZ bit is not set until the 8 -bit counter reaches 00 h again. The values of the TCR and the PSC registers can be read accurately at any time.

### 4.2.4 Timer Registers

Timer Status Control Register (TSCR)
Address: 0D4h — Read/Write


Bit 7 = TMZ: Timer Zero bit
A low-to-high transition indicates that the timer count register has decrement to zero. This bit must be cleared by user software before starting a new count.
Bit 6 = ETI: Enable Timer Interrupt
When set, enables the timer interrupt request (vector \#3). If $\mathrm{ETI}=0$ the timer interrupt is disabled. If $E T I=1$ and $T M Z=1$ an interrupt request is generated.
Bit 5 = TOUT: Timers Output Control
When low, this bit selects the input mode for the TIMER pin. When high the output mode is selected.

## Bit 4 = DOUT: Data Output

Data sent to the timer output when TMZ is set high (output mode only). Input mode selection (input mode only).

## Bit 3 = PSI: Prescaler Initialize Bit

Used to initialize the prescaler and inhibit its counting. When $\mathrm{PSI}=$ " 0 " the prescaler is set to 7Fh and the counter is inhibited. When PSI="1" the prescaler is enabled to count downwards. As long as PSI="0" both counter and prescaler are not running.
Bit 2, 1, 0 = PS2, PS1, PS0: Prescaler Mux. Select. These bits select the division ratio of the prescaler register.

Table 12. Prescaler Division Factors

| PS2 | PS1 | PS0 | Divided by |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | 2 |
| 0 | 1 | 0 | 4 |
| 0 | 1 | 1 | 8 |
| 1 | 0 | 0 | 16 |
| 1 | 0 | 1 | 32 |
| 1 | 1 | 0 | 64 |
| 1 | 1 | 1 | 128 |

## Timer Counter Register TCR

Address: 0D3h — Read/Write


Bit 7-0 = D7-D0: Counter Bits.

## Prescaler Register PSC

Address: OD2h — Read/Write


Bit $7=$ D7: Always read as " 0 ".
Bit 6-0 = D6-D0: Prescaler Bits.

### 4.3 A/D CONVERTER (ADC)

The A/D converter peripheral is an 8-bit analog to digital converter with analog inputs as alternate I/O functions (the number of which is device dependent), offering 8-bit resolution with a typical conversion time of 70us (at an oscillator clock frequency of 8 MHz ).
The ADC converts the input voltage by a process of successive approximations, using a clock frequency derived from the oscillator with a division factor of twelve. With an oscillator clock frequency less than 1.2 MHz , conversion accuracy is decreased.
Selection of the input pin is done by configuring the related I/O line as an analog input via the Option and Data registers (refer to I/O ports description for additional information). Only one I/O line must be configured as an analog input at any time. The user must avoid any situation in which more than one I/O pin is selected as an analog input simultaneously, to avoid device malfunction.
The ADC uses two registers in the data space: the ADC data conversion register, ADR, which stores the conversion result, and the ADC control register, ADCR, used to program the ADC functions.
A conversion is started by writing a " 1 " to the Start bit (STA) in the ADC control register. This automatically clears (resets to " 0 ") the End Of Conversion Bit (EOC). When a conversion is complete, the EOC bit is automatically set to " 1 ", in order to flag that conversion is complete and that the data in the ADC data conversion register is valid. Each conversion has to be separately initiated by writing to the STA bit.
The STA bit is continuously scanned so that, if the user sets it to " 1 " while a previous conversion is in progress, a new conversion is started before completing the previous one. The start bit (STA) is a write only bit, any attempt to read it will show a logical "0".
The A/D converter features a maskable interrupt associated with the end of conversion. This interrupt is associated with interrupt vector \#4 and occurs when the EOC bit is set (i.e. when a conversion is completed). The interrupt is masked using the EAI (interrupt mask) bit in the control register.
The power consumption of the device can be reduced by turning off the ADC peripheral. This is done by setting the PDS bit in the ADC control register to " 0 ". If PDS=" 1 ", the A/D is powered and enabled for conversion. This bit must be set at least one instruction before the beginning of the conver-
sion to allow stabilisation of the A/D converter. This action is also needed before entering WAIT mode, since the A/D comparator is not automatically disabled in WAIT mode.
During Reset, any conversion in progress is stopped, the control register is reset to 40h and the ADC interrupt is masked (EAI=0).

Figure 26. ADC Block Diagram


### 4.3.1 Application Notes

The A/D converter does not feature a sample and hold circuit. The analog voltage to be measured should therefore be stable during the entire conversion cycle. Voltage variation should not exceed $\pm 1 / 2$ LSB for the optimum conversion accuracy. A low pass filter may be used at the analog input pins to reduce input voltage variation during conversion.
When selected as an analog channel, the input pin is internally connected to a capacitor $\mathrm{C}_{\mathrm{ad}}$ of typically 12 pF . For maximum accuracy, this capacitor must be fully charged at the beginning of conversion. In the worst case, conversion starts one instruction ( $6.5 \mu \mathrm{~s}$ ) after the channel has been selected. In worst case conditions, the impedance, ASI, of the analog voltage source is calculated using the following formula:

$$
6.5 \mu \mathrm{~s}=9 \times \mathrm{C}_{\mathrm{ad}} \times \mathrm{ASI}
$$

(capacitor charged to over 99.9\%), i.e. $30 \mathrm{k} \Omega$ including a $50 \%$ guardband. ASI can be higher if $\mathrm{C}_{\mathrm{ad}}$ has been charged for a longer period by adding instructions before the start of conversion (adding more than 26 CPU cycles is pointless).

## A/D CONVERTER (Cont'd)

Since the ADC is on the same chip as the microprocessor, the user should not switch heavily loaded output signals during conversion, if high precision is required. Such switching will affect the supply voltages used as analog references.
The accuracy of the conversion depends on the quality of the power supplies ( $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\mathrm{SS}}$ ). The user must take special care to ensure a well regulated reference voltage is present on the $\mathrm{V}_{\mathrm{DD}}$ and $\mathrm{V}_{\text {SS }}$ pins (power supply voltage variations must be less than $5 \mathrm{~V} / \mathrm{ms}$ ). This implies, in particular, that a suitable decoupling capacitor is used at the $\mathrm{V}_{\mathrm{DD}}$ pin.
The converter resolution is given by::

$$
\frac{V_{D D}-V_{S S}}{256}
$$

The Input voltage (Ain) which is to be converted must be constant for $1 \mu$ s before conversion and remain constant during conversion.
Conversion resolution can be improved if the power supply voltage ( $\mathrm{V}_{\mathrm{DD}}$ ) to the microcontroller is lowered.

In order to optimise conversion resolution, the user can configure the microcontroller in WAIT mode, because this mode minimises noise disturbances and power supply variations due to output switching. Nevertheless, the WAIT instruction should be executed as soon as possible after the beginning of the conversion, because execution of the WAIT instruction may cause a small variation of the $\mathrm{V}_{\mathrm{DD}}$ voltage. The negative effect of this variation is minimized at the beginning of the conversion when the converter is less sensitive, rather than at the end of conversion, when the less significant bits are determined.
The best configuration, from an accuracy standpoint, is WAIT mode with the Timer stopped. Indeed, only the ADC peripheral and the oscillator are then still working. The MCU must be woken up from WAIT mode by the ADC interrupt at the end of the conversion. It should be noted that waking
up the microcontroller could also be done using the Timer interrupt, but in this case the Timer will be working and the resulting noise could affect conversion accuracy.

## A/D Converter Control Register (ADCR)

Address: 0D1h — Read/Write


Bit $7=$ EAI: Enable A/D Interrupt. If this bit is set to " 1 " the $A / D$ interrupt is enabled, when EAI=0 the interrupt is disabled.
Bit 6 = EOC: End of conversion. Read Only. This read only bit indicates when a conversion has been completed. This bit is automatically reset to " 0 " when the STA bit is written. If the user is using the interrupt option then this bit can be used as an interrupt pending bit. Data in the data conversion register are valid only when this bit is set to " 1 ".
Bit $5=$ STA: Start of Conversion. Write Only. Writing a " 1 " to this bit will start a conversion on the selected channel and automatically reset to " 0 " the EOC bit. If the bit is set again when a conversion is in progress, the present conversion is stopped and a new one will take place. This bit is write only, any attempt to read it will show a logical zero.
Bit 4 = PDS: Power Down Selection. This bit activates the A/D converter if set to " 1 ". Writing a "0" to this bit will put the ADC in power down mode (idle mode).
Bit 3-0 = D3-D0. Not used

## A/D Converter Data Register (ADR)

Address: ODOh — Read only


Bit 7-0 = D7-D0 : 8 Bit A/D Conversion Result.

## 5 SOFTWARE

### 5.1 ST6 ARCHITECTURE

The ST6 software has been designed to fully use the hardware in the most efficient way possible while keeping byte usage to a minimum; in short, to provide byte efficient programming capability. The ST6 core has the ability to set or clear any register or RAM location bit of the Data space with a single instruction. Furthermore, the program may branch to a selected address depending on the status of any bit of the Data space. The carry bit is stored with the value of the bit when the SET or RES instruction is processed.

### 5.2 ADDRESSING MODES

The ST6 core offers nine addressing modes, which are described in the following paragraphs. Three different address spaces are available: Program space, Data space, and Stack space. Program space contains the instructions which are to be executed, plus the data for immediate mode instructions. Data space contains the Accumulator, the $\mathrm{X}, \mathrm{Y}, \mathrm{V}$ and W registers, peripheral and Input/ Output registers, the RAM locations and Data ROM locations (for storage of tables and constants). Stack space contains six 12-bit RAM cells used to stack the return addresses for subroutines and interrupts.
Immediate. In the immediate addressing mode, the operand of the instruction follows the opcode location. As the operand is a ROM byte, the immediate addressing mode is used to access constants which do not change during program execution (e.g., a constant used to initialize a loop counter).

Direct. In the direct addressing mode, the address of the byte which is processed by the instruction is stored in the location which follows the opcode. Direct addressing allows the user to directly address the 256 bytes in Data Space memory with a single two-byte instruction.

Short Direct. The core can address the four RAM registers $X, Y, V, W$ (locations $80 \mathrm{~h}, 81 \mathrm{~h}, 82 \mathrm{~h}, 83 \mathrm{~h}$ ) in the short-direct addressing mode. In this case, the instruction is only one byte and the selection of the location to be processed is contained in the opcode. Short direct addressing is a subset of the direct addressing mode. (Note that 80h and 81h are also indirect registers).
Extended. In the extended addressing mode, the 12-bit address needed to define the instruction is obtained by concatenating the four less significant
bits of the opcode with the byte following the opcode. The instructions (JP, CALL) which use the extended addressing mode are able to branch to any address of the 4 K bytes Program space.

An extended addressing mode instruction is twobyte long.

Program Counter Relative. The relative addressing mode is only used in conditional branch instructions. The instruction is used to perform a test and, if the condition is true, a branch with a span of -15 to +16 locations around the address of the relative instruction. If the condition is not true, the instruction which follows the relative instruction is executed. The relative addressing mode instruction is one-byte long. The opcode is obtained in adding the three most significant bits which characterize the kind of the test, one bit which determines whether the branch is a forward (when it is 0 ) or backward (when it is 1) branch and the four less significant bits which give the span of the branch (Oh to Fh) which must be added or subtracted to the address of the relative instruction to obtain the address of the branch.

Bit Direct. In the bit direct addressing mode, the bit to be set or cleared is part of the opcode, and the byte following the opcode points to the address of the byte in which the specified bit must be set or cleared. Thus, any bit in the 256 locations of Data space memory can be set or cleared.

Bit Test \& Branch. The bit test and branch addressing mode is a combination of direct addressing and relative addressing. The bit test and branch instruction is three-byte long. The bit identification and the tested condition are included in the opcode byte. The address of the byte to be tested follows immediately the opcode in the Program space. The third byte is the jump displacement, which is in the range of -127 to +128 . This displacement can be determined using a label, which is converted by the assembler.

Indirect. In the indirect addressing mode, the byte processed by the register-indirect instruction is at the address pointed by the content of one of the indirect registers, X or $\mathrm{Y}(80 \mathrm{~h}, 81 \mathrm{~h})$. The indirect register is selected by the bit 4 of the opcode. A register indirect instruction is one byte long.

Inherent. In the inherent addressing mode, all the information necessary to execute the instruction is contained in the opcode. These instructions are one byte long.

### 5.3 INSTRUCTION SET

The ST6 core offers a set of 40 basic instructions which, when combined with nine addressing modes, yield 244 usable opcodes. They can be divided into six different types: load/store, arithmetic/logic, conditional branch, control instructions, jump/call, and bit manipulation. The following paragraphs describe the different types.
All the instructions belonging to a given type are presented in individual tables.

Load \& Store. These instructions use one, two or three bytes in relation with the addressing mode. One operand is the Accumulator for LOAD and the other operand is obtained from data memory using one of the addressing modes.
For Load Immediate one operand can be any of the 256 data space bytes while the other is always immediate data.

Table 13. Load \& Store Instructions

| Instruction | Addressing Mode | Bytes | Cycles | Flags |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |
| LD A, X | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD A, Y | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD A, V | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD A, W | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD X, A | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD Y, A | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD V, A | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD W, A | Short Direct | 1 | 4 | $\Delta$ | $*$ |
| LD A, rr | Direct | 2 | 4 | $\Delta$ | $*$ |
| LD rr, A | Direct | 2 | 4 | $\Delta$ | $*$ |
| LD A, (X) | Indirect | 1 | 4 | $\Delta$ | $*$ |
| LD A, (Y) | Indirect | 1 | 4 | $\Delta$ | $*$ |
| LD (X), A | Indirect | 1 | 4 | $\Delta$ | $*$ |
| LD (Y), A | Indirect | 1 | 4 | $\Delta$ | $*$ |
| LDI A, \#N | Immediate | 2 | 4 | $\Delta$ | $*$ |
| LDI rr, \#N | Immediate | 4 | $*$ | $*$ |  |

## Notes:

X,Y. Indirect Register Pointers, V \& W Short Direct Registers
\# . Immediate data (stored in ROM memory)
rr. Data space register
$\Delta$. Affected
*. Not Affected

## INSTRUCTION SET (Cont'd)

Arithmetic and Logic. These instructions are used to perform the arithmetic calculations and logic operations. In AND, ADD, CP, SUB instructions one operand is always the accumulator while the other can be either a data space memory con-
tent or an immediate value in relation with the addressing mode. In CLR, DEC, INC instructions the operand can be any of the 256 data space addresses. In COM, RLC, SLA the operand is always the accumulator.

Table 14. Arithmetic \& Logic Instructions

| Instruction | Addressing Mode | Bytes | Cycles | Flags |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  | Z | C |
| ADD A, (X) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| ADD A, (Y) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| ADD A, rr | Direct | 2 | 4 | $\Delta$ | $\Delta$ |
| ADDI A, \#N | Immediate | 2 | 4 | $\Delta$ | $\Delta$ |
| AND A, (X) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| AND A, (Y) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| AND A, rr | Direct | 2 | 4 | $\Delta$ | $\Delta$ |
| ANDI A, \#N | Immediate | 2 | 4 | $\Delta$ | $\Delta$ |
| CLR A | Short Direct | 2 | 4 | $\Delta$ | $\Delta$ |
| CLR r | Direct | 3 | 4 | * | * |
| COM A | Inherent | 1 | 4 | $\Delta$ | $\Delta$ |
| CP A, (X) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| CP A, (Y) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| CP A, rr | Direct | 2 | 4 | $\Delta$ | $\Delta$ |
| CPI A, \#N | Immediate | 2 | 4 | $\Delta$ | $\Delta$ |
| DEC X | Short Direct | 1 | 4 | $\Delta$ | * |
| DEC Y | Short Direct | 1 | 4 | $\Delta$ | * |
| DEC V | Short Direct | 1 | 4 | $\Delta$ | * |
| DEC W | Short Direct | 1 | 4 | $\Delta$ | * |
| DEC A | Direct | 2 | 4 | $\Delta$ | * |
| DEC rr | Direct | 2 | 4 | $\Delta$ | * |
| DEC (X) | Indirect | 1 | 4 | $\Delta$ | * |
| DEC (Y) | Indirect | 1 | 4 | $\Delta$ | * |
| INC X | Short Direct | 1 | 4 | $\Delta$ | * |
| INC Y | Short Direct | 1 | 4 | $\Delta$ | * |
| INC V | Short Direct | 1 | 4 | $\Delta$ | * |
| INC W | Short Direct | 1 | 4 | $\Delta$ | * |
| INC A | Direct | 2 | 4 | $\Delta$ | * |
| INC rr | Direct | 2 | 4 | $\Delta$ | * |
| INC (X) | Indirect | 1 | 4 | $\Delta$ | * |
| INC (Y) | Indirect | 1 | 4 | $\Delta$ | * |
| RLC A | Inherent | 1 | 4 | $\Delta$ | $\Delta$ |
| SLA A | Inherent | 2 | 4 | $\Delta$ | $\Delta$ |
| SUB A, (X) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| SUB A, (Y) | Indirect | 1 | 4 | $\Delta$ | $\Delta$ |
| SUB A, rr | Direct | 2 | 4 | $\Delta$ | $\Delta$ |
| SUBI A, \#N | Immediate | 2 | 4 | $\Delta$ | $\Delta$ |

## Notes:

X,Y.Indirect Register Pointers, V \& W Short Direct RegistersD. Affected
\# . Immediate data (stored in ROM memory)* . Not Affected
rr. Data space register

## INSTRUCTION SET (Cont'd)

Conditional Branch. The branch instructions achieve a branch in the program when the selected condition is met.

Bit Manipulation Instructions. These instructions can handle any bit in data space memory. One group either sets or clears. The other group (see Conditional Branch) performs the bit test branch operations.

Control Instructions. The control instructions control the MCU operations during program execution.

Jump and Call. These two instructions are used to perform long (12-bit) jumps or subroutines call inside the whole program space.

Table 15. Conditional Branch Instructions

| Instruction | Branch If | Bytes | Cycles | Flags |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  |  | Z | C |
| JRC e $=1$ | 1 | 2 | $*$ | $*$ |  |
| JRNC e | C $=0$ | 1 | 2 | $*$ | $*$ |
| JRZ e | Z $=1$ | 1 | 2 | $*$ | $*$ |
| JRNZ e | Z $=0$ | 3 | 2 | $*$ | $\Delta$ |
| JRR b, rr, ee | Bit $=0$ | 3 | 5 | $*$ | $\Delta$ |
| JRS b, rr, ee | Bit $=1$ |  |  |  |  |

Notes:
b. 3-bit address
e. 5 bit signed displacement in the range -15 to $+16<\mathrm{F} 128 \mathrm{M}>$
ee. 8 bit signed displacement in the range -126 to +129
rr. Data space register
$\Delta$. Affected. The tested bit is shifted into carry.
*. Not Affected

Table 16. Bit Manipulation Instructions

| Instruction | Addressing Mode | Bytes | Cycles | Flags |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  |  | Z | C |
| SET b,rr | Bit Direct | 2 | 4 | ${ }^{*}$ | ${ }^{*}$ |
| RES b,rr | Bit Direct | 2 | 4 | ${ }^{*}$ |  |

Notes:
b. 3-bit address;
rr. Data space register;
Table 17. Control Instructions

| Instruction | Addressing Mode | Bytes | Cycles | Flags |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  |  | C |  |
| NOP | Inherent | 1 | 2 | $*$ | $*$ |
| RET | Inherent | 1 | 2 | $*$ | $\Delta$ |
| RETI | Inherent | 1 | 2 | $*$ | $*$ |
| STOP (1) | Inherent | 1 | 2 | $*$ | $*$ |
| WAIT | Inherent | 1 | 2 | $*$ |  |

Notes:

1. This instruction is deactivated<N>and a WAIT is automatically executed instead of a STOP if the watchdog function is selected.
$\Delta$. Affected
*. Not Affected
Table 18. Jump \& Call Instructions

| Instruction | Addressing Mode | Bytes | Cycles | Flags |  |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  |  |  |  | Z | C |
| CALL abc | Extended | 2 | 4 | $*$ | $*$ |
| JP abc | Extended | 2 | 4 | $*$ | $*$ |

Notes:
abc. 12-bit address;
*. Not Affected

Opcode Map Summary. The following table contains an opcode map for the instructions used by the ST6

| LOW | $\begin{gathered} 0 \\ 0000 \end{gathered}$ | $\begin{gathered} 1 \\ 0001 \end{gathered}$ | $\stackrel{2}{0010}$ | $\begin{gathered} 3 \\ 0011 \end{gathered}$ | $\begin{gathered} 4 \\ 0100 \end{gathered}$ | $\begin{gathered} 5 \\ 0101 \end{gathered}$ | $\begin{gathered} 6 \\ 0110 \end{gathered}$ | $\begin{gathered} 7 \\ 0111 \end{gathered}$ | LOW |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} 0 \\ 0000 \end{gathered}$ | 2 JRNZ $^{2}$ <br> 1  <br> $\mathrm{e}^{2}$ pcr | $\begin{array}{lc}4 & \text { CALL } \\ 2 & \text { abc }\end{array}$ | $\begin{array}{ll}2 & \text { JRNC } \\ 1 & e \\ 1 & \\ \text { pcr }\end{array}$ | $\left\|\begin{array}{\|ccc} 5 & \text { JRR } \\ & \mathrm{b} 0, \mathrm{rr}, \mathrm{ee} & \\ 3 & \mathrm{bt} \end{array}\right\|$ | $\begin{array}{ccc} \hline 2 & & \text { JRZ } \\ & \text { e } & \\ 1 & & \text { pcr } \end{array}$ | \# | $\begin{array}{\|lll\|} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & \mathrm{a},(\mathrm{x}) & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} 0 \\ 0000 \end{gathered}$ |
| $\begin{gathered} 1 \\ 0001 \end{gathered}$ | $\begin{array}{\|lll\|} \hline 2 & \text { JRNZ } \\ & e^{\prime} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | 4 CALL <br> abc  <br> 2 ext | $\begin{array}{\|lll} \hline 2 & \text { JRNC }^{\text {JRNC }} \\ & e^{2} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | $$ | $\begin{array}{lll} \hline 2 & & \text { JRZ } \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | 4  INC <br>  $x$  <br> 1  sd | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \mathrm{e} & \\ 1 & & \mathrm{prc} \\ \hline \end{array}$ | $\begin{array}{\|ccc} \hline 4 & \text { LDI } \\ & \mathrm{a}, \mathrm{nn} & \\ 2 & \text { imm } \\ \hline \end{array}$ | $\begin{gathered} 1 \\ 0001 \end{gathered}$ |
| $\begin{gathered} 2 \\ 0010 \end{gathered}$ | 2 JRNZ <br> 1 $e$ <br>   | $\begin{array}{lc}4 & \text { CALL } \\ 2 & \text { exc }\end{array}$ | $\begin{array}{lll}2 & \text { JRNC } \\ 1 & e & \text { pcr }\end{array}$ | 5 JRR <br> 3 b4,rr,ee <br> $5 t$  | $\begin{array}{llll} \hline 2 & & \text { JRZ } \\ & \text { e } & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | $\begin{array}{\|ccc} \hline 4 & & C P \\ & \mathrm{a},(\mathrm{x}) & \\ 1 & & \text { ind } \end{array}$ | $\stackrel{2}{0} 010$ |
| $\begin{gathered} 3 \\ 0011 \end{gathered}$ | 2 JRNZ <br> 1 $e \quad$ pcr | $\begin{array}{lc}4 & \text { CALL } \\ 2 & \text { abc }\end{array}$ | $\begin{array}{ll}2 & \text { JRNC } \\ 1 & e^{\text {JRCr }}\end{array}$ | 5 JRS <br> 3 $\mathrm{~b} 4, \mathrm{rr}, \mathrm{ee}$ <br> $5 t$  | 2 JRZ <br> e  <br> 1 pcr | $\begin{array}{\|ccc} \hline 4 & & \text { LD } \\ & \mathrm{a}, \mathrm{x} & \\ 1 & & \mathrm{sd} \end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | $\begin{array}{lr}4 & \text { CPI } \\ 2 & \text { a,nn } \\ 2 m m\end{array}$ | $\begin{gathered} 3 \\ 0011 \end{gathered}$ |
| $\begin{gathered} 4 \\ 0100 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  <br> pcr  | 4 CALL <br>  abc | $\begin{array}{\|lll} \hline 2 & \text { JRNC } \\ & e^{\prime} & \\ 1 & & \mathrm{pcr} \end{array}$ | $\left.\begin{array}{\|cc\|} \hline 5 & \text { JRR } \\ & \text { b2,rr,ee } \\ 3 & b t \end{array} \right\rvert\,$ | $\begin{array}{lll} \hline 2 & & \text { JRZ } \\ & \text { e } & \\ 1 & & \text { pcr } \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { ADD } \\ & \mathrm{a},(\mathrm{x}) & \\ 1 & & \text { ind } \\ \hline 1 \end{array}$ | $\begin{gathered} 4 \\ 0100 \end{gathered}$ |
| $\begin{gathered} 5 \\ 0101 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  <br> $\mathrm{e}^{2}$ pcr | $\begin{array}{\|lll} \hline 4 & \text { CALL } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | $\begin{array}{\|lll} \hline 2 & \text { JRNC } \\ & e^{\text {J }} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | $\left.\begin{array}{\|cc\|} \hline 5 & \text { JRS } \\ & \mathrm{b} 2, \mathrm{rr}, \mathrm{ee} \\ 3 & \\ 3 & \mathrm{bt} \end{array} \right\rvert\,$ | $\begin{array}{lll} \hline 2 & & \text { JRZ } \\ & \text { e } & \\ 1 & & \text { pcr } \\ \hline \end{array}$ | $\begin{array}{lll}4 & & \text { INC } \\ & y & \\ 1 & & \text { sd }\end{array}$ | $\begin{array}{\|lll\|} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | $\begin{array}{\|cc\|} \hline 4 & \text { ADDI } \\ & \mathrm{a}, \mathrm{nn} \\ 2 & \text { imm } \end{array}$ | $\begin{gathered} 5 \\ 0101 \end{gathered}$ |
| $\begin{gathered} 6 \\ 0110 \end{gathered}$ | 2 $\mathrm{e}^{\text {JRNZ }}$ <br> 1  <br> $\mathrm{e}^{2}$ pcr | $\begin{array}{\|lll} \hline 4 & \text { CALL } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRNC } \\ & e^{\prime} & \\ 1 & & \mathrm{pcr} \end{array}$ | $\begin{array}{\|cc\|} \hline 5 & \text { JRR } \\ & \mathrm{b} 6, \mathrm{rr}, \mathrm{ee} \\ 3 & \mathrm{bt} \end{array}$ | $\begin{array}{lll} \hline 2 & & J R Z \\ & \text { e } & \\ 1 & & \mathrm{pcr} \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \mathrm{JRC} \\ & \mathrm{e} & \\ 1 & & \mathrm{prc} \end{array}$ | 4  INC <br>  (x)  <br> 1  ind | $\begin{gathered} 6 \\ 0110 \end{gathered}$ |
| $\begin{gathered} 7 \\ 0111 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | 4 CALL <br> abc  <br> 2 ext | $\begin{array}{\|lll} \hline 2 & \text { JRNC } \\ & e^{\text {JRN }} & \mathrm{pcr} \\ \hline \end{array}$ | $\begin{array}{\|cc\|} \hline 5 & \mathrm{JRS} \\ & \mathrm{~b} 6, \mathrm{rr}, \mathrm{ee} \\ 3 & \mathrm{bt} \end{array}$ | $\begin{array}{\|llll} \hline 2 & & \text { JRZ } \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | $\begin{array}{\|lll} \hline 4 & & L D \\ & \mathrm{a}, \mathrm{y} & \\ 1 & & \mathrm{sd} \\ \hline \end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \mathrm{e} & \\ 1 & & \mathrm{prc} \\ \hline \end{array}$ | \# | $\begin{gathered} 7 \\ 0111 \end{gathered}$ |
| $\begin{gathered} 8 \\ 1000 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | 4 CALL <br> 2 exc | 2 $e^{\text {JRNC }}$ <br> 1  <br> pcr  | 5 JRR <br> 3 b1,rr,ee <br> 3 bt | 2 JRZ  <br> 1   | \# | 2  JRC <br> 1   <br>   prc | $\begin{array}{\|ccc} \hline 4 & & \text { LD } \\ & (\mathrm{x}), \mathrm{a} & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} 8 \\ 1000 \end{gathered}$ |
| $\begin{gathered} 9 \\ 1001 \end{gathered}$ | 2 RNZ <br>  e <br> 1  | 4 CALL <br> 2 abc | 2 JRNC <br> 1 err | 5 JRS <br> 3 $\mathrm{~b} 1, \mathrm{rr}, \mathrm{ee}$ <br> 3 bt | 2 JRZ  <br> 1   <br>  pcr  | 4  INC <br>  v  <br> 1  sd | 2  JRC <br> 1  prc | \# | $\begin{gathered} 9 \\ 1001 \end{gathered}$ |
| $\underset{1010}{A}$ | 2 $e^{\text {JRNZ }}$ <br> 1  <br> pcr  | 4 CALL <br> abc  <br> 2 ext | $\begin{array}{\|lll} \hline 2 & { }^{\text {JRNC }} \\ & e^{\text {JRN }} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | $\left.\begin{array}{\|ccc} \hline 5 & \text { JRR } \\ & \mathrm{b} 5, \mathrm{rr}, \mathrm{ee} & \\ 3 & \mathrm{bt} \end{array} \right\rvert\,$ | $\begin{array}{\|lll} \hline 2 & & J R Z \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | $\begin{array}{\|lll} \hline 4 & \text { AND } \\ & \mathrm{a},(\mathrm{x}) & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} \text { A } \\ 1010 \end{gathered}$ |
| $\begin{gathered} B \\ 1011 \end{gathered}$ | 2 JRNZ <br> 1 $e$ <br>  pcr | 4 CALL <br> 2 abc | 2 JRNC <br> 1 $e^{\text {pcr }}$ | 5  <br> 3 JR, rr,ee <br> 3 bt | 2  JRZ <br>  e  <br> 1  pcr | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & a, v & \\ 1 & & s d \end{array}$ | 2  JRC <br> 1   | 4 ANDI <br> 2 ann <br> 2 imm | $\begin{gathered} B \\ 1011 \end{gathered}$ |
| $\begin{gathered} C \\ 1100 \end{gathered}$ | $\begin{array}{ll}2 & e^{\text {JRNZ }} \\ 1 & \\ & \mathrm{pcr}\end{array}$ | $\begin{array}{lc}4 & \text { CALL } \\ 2 & \text { abc }\end{array}$ | $\begin{array}{ll}2 & \text { JRNC } \\ 1 & e^{\text {jpr }}\end{array}$ | $\begin{array}{\|ccc} \hline 5 & \text { JRR } \\ & \text { b3, rr,ee } \\ 3 & b t \end{array}$ | $\begin{array}{\|llll} \hline 2 & & J R Z \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | \# | $\begin{array}{lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & \text { SUB } \\ & \mathrm{a},(\mathrm{x}) & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} C \\ 1100 \end{gathered}$ |
| $\begin{gathered} \text { D } \\ 1101 \end{gathered}$ | 2 JRNZ <br> 1 $e^{\text {JRN }}$ <br> 1  | 4 CALL <br> 2 abc | 2 JRNC <br> 1 $e^{\text {pcr }}$ | 5 JRS <br> $3^{b} 3, \mathrm{rr}, \mathrm{ee}$  <br> bt  | 2  JRZ <br> 1  e <br>   pcr | 4  INC <br>  w  <br> 1  sd | 2   <br>  JRC  <br> 1   | 4 SUBI <br>  a,nn <br> 2 imm | $\begin{gathered} \text { D } \\ 1101 \end{gathered}$ |
| $\begin{gathered} E \\ 1110 \end{gathered}$ | 2 JRNZ <br> 1 $\mathrm{e}^{\mathrm{JRN}}$ <br> 1 pcr | 4 CALL <br> 2 abc | $\begin{array}{\|lll} \hline 2 & & \text { JRNC } \\ & e^{\prime} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | 5 JRR <br> 3 b7,rr,ee <br> 3 bt | $\begin{array}{\|lll} \hline 2 & & \text { JRZ } \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | 4 DEC <br>  (x) <br> 1 ind | $\begin{gathered} E \\ 1110 \end{gathered}$ |
| $\stackrel{F}{F}$ | 2 $\mathrm{JRNZ}^{\text {JRN }}$ <br> 1  <br>  pcr | $\begin{array}{\|lcc} \hline 4 & \text { CALL } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | $\begin{array}{lll}2 & \text { JRNC } \\ 1 & e & \mathrm{pcr}\end{array}$ | $\left\|\begin{array}{\|cc\|} \hline 5 & \text { JRS } \\ & \text { b7, rr,ee } \\ 3 & b t \end{array}\right\|$ | $\begin{array}{\|lll\|} \hline 2 & & \text { JRZ } \\ & \text { e } & \\ 1 & & \text { pcr } \end{array}$ | $\begin{array}{lll}4 & & \text { LD } \\ & \text { a,w } & \\ 1 & & \text { sd }\end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \\ \hline \end{array}$ | \# | $\underset{1111}{F}$ |

## Abbreviations for Addressing Modes:

| dir | Direct |
| :--- | :--- |
| sd | Short Direct |
| imm | Immediate |
| inh | Inherent |
| ext | Extended |
| b.d | Bit Direct |
| bt | Bit Test |
| pcr | Program Counter Relative |
| ind | Indirect |

Legend:

| \# | Indicates III egal Instructions |
| :--- | :--- |
| e | 5 Bit Displacement |
| b | 3 Bit Address |
| rr | 1byte dataspace address |
| nn | 1 byte immediate data |
| abc | 12 bit address |
| ee | 8 bit Displacement |



Opcode Map Summary (Continued)

| LOW | $\begin{gathered} 8 \\ 1000 \end{gathered}$ | $\begin{gathered} 9 \\ 1001 \end{gathered}$ | $\begin{gathered} \text { A } \\ 1010 \end{gathered}$ | $\begin{gathered} B \\ 1011 \end{gathered}$ | $\begin{gathered} \text { C } \\ 1100 \end{gathered}$ | $\begin{gathered} \text { D } \\ 1101 \end{gathered}$ | $\underset{1110}{E}$ | $\underset{1111}{\mathrm{~F}}$ | $\begin{array}{\|ll\|}\text { LOW } & \\ & \\ & \mathrm{HI}\end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} 0 \\ 0000 \end{gathered}$ | $\begin{array}{ccc} \hline 2 & & \text { JRNZ } \\ & \text { e } & \\ 1 & & \mathrm{pcr} \end{array}$ | $\left\|\begin{array}{lll} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}\right\|$ | 2 JRNC <br> 1 $e^{\text {pcr }}$ | $\left.\begin{array}{\|cc\|} \hline 4 & \text { RES } \\ & \text { b0,rr } \\ 2 & \text { b.d. } \end{array} \right\rvert\,$ | 2   <br>  JRZ  <br> 1   | $\begin{array}{\|lcc\|} \hline 4 & \text { LDI } \\ & \mathrm{rr}, \mathrm{nn} \\ 3 & \mathrm{imm} \end{array}$ | $\begin{array}{\|llll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & \mathrm{a},(\mathrm{y}) & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} 0 \\ 0000 \end{gathered}$ |
| $\begin{gathered} 1 \\ 0001 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | $\left.\begin{array}{\|ccc} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array} \right\rvert\,$ | 2 JRNC  <br> 1  pcr | $\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b0,rr } & \\ 2 & \text { b.d.d } \end{array}$ | 2   <br>  JRZ  <br> 1  pcr | $\begin{array}{\|lll} \hline 4 & & \text { DEC } \\ & x & \\ 1 & & s d \end{array}$ | 2  JRC <br> 1   | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & \text { a,rr } & \\ 2 & & \operatorname{dir} \end{array}$ | $\begin{gathered} 1 \\ 0001 \end{gathered}$ |
| $\stackrel{2}{0010}$ | $\begin{array}{lll} \hline 2 & & \text { JRNZ } \\ & \text { e } & \\ 1 & & \mathrm{pcr} \end{array}$ | $\begin{array}{\|ccc} \hline 4 & & J P \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 JRNC <br> 1 $e^{\text {JR }}$ | $\begin{array}{\|cc\|} \hline 4 & \text { RES } \\ & \text { b4,rr } \\ 2 & \text { b.d } \end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRZ } \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | $\begin{array}{\|ll\|} \hline 4 & \text { COM } \\ \hline \end{array}$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { CP } \\ & \text { a,(y) } & \\ 1 & & \text { ind } \end{array}$ | $\stackrel{2}{0010}$ |
| $\begin{gathered} 3 \\ 0011 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | $\left.\begin{array}{\|lll} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array} \right\rvert\,$ | 2 JRNC  <br> 1 $e^{\text {JR }}$  | $\left.\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b4,rr } & \\ 2 & \text { b.d.d } \end{array} \right\rvert\,$ | 2 JRZ <br> e  <br> 1 pcr | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & \mathrm{x}, \mathrm{a} & \\ 1 & & \mathrm{sd} \end{array}$ | 2 JRC  <br> 1 e  | $\begin{array}{\|lll} \hline 4 & & C P \\ & \text { a,rr } & \\ 2 & & \operatorname{dir} \end{array}$ | $\begin{gathered} 3 \\ 0011 \end{gathered}$ |
| $\begin{gathered} 4 \\ 0100 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | 4  JP <br>  abc  <br> 2  ext | 2   <br> $\mathrm{JRNC}^{2}$   <br> 1  pcr | 4 RES <br>  b2,rr <br> 2 b.d | 2   <br>  JRZ  <br> 1   | 2 RETI <br> 1 inh | 2  JRC <br> 1  prc | 4 ADD <br>  $\mathrm{a},(\mathrm{y})$ <br> 1 ind | $\stackrel{4}{0100}$ |
| $\begin{gathered} 5 \\ 0101 \end{gathered}$ | $\begin{array}{lll} \hline 2 & & \text { JRNZ } \\ & \text { e } & \\ 1 & & \mathrm{pcr} \end{array}$ | $\left\|\begin{array}{lll} \hline 4 & & J P \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}\right\|$ | 2 JRNC <br> 1  <br> ecr  | $\left.\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b2,rr } & \\ 2 & \text { b.d.d } \end{array} \right\rvert\,$ | $\left.\begin{array}{\|lll} \hline 2 & & J R Z \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \end{array} \right\rvert\,$ | $\left\|\begin{array}{lll} \hline 4 & & D E C \\ & y & \\ 1 & & s d \end{array}\right\|$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \mathrm{e} & \\ 1 & & \mathrm{prc} \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { ADD } \\ & \text { a,rr } & \\ 2 & & \operatorname{dir} \end{array}$ | $\begin{gathered} 5 \\ 0101 \end{gathered}$ |
| $\begin{gathered} 6 \\ 0110 \end{gathered}$ | 2 JRNZ <br> 1 $e^{\text {J }}$ <br>   <br> pcr  | $\left\|\begin{array}{lll} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}\right\|$ | 2 JRNC  <br> 1 e  <br> 1  pcr | $\left.\begin{array}{\|cc\|} \hline 4 & \text { RES } \\ & \text { b6,rr } \\ 2 & \text { b.d } \end{array} \right\rvert\,$ | 2  JRZ <br>  e  <br> 1  pcr | 2 STOP <br> 1 inh | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \mathrm{e} & \\ 1 & & \mathrm{prc} \end{array}$ | 4  INC <br>  (y)  <br> 1  ind | $\begin{gathered} 6 \\ 0110 \end{gathered}$ |
| $\begin{gathered} 7 \\ 0111 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | $\left\|\begin{array}{lll} \hline 4 & & J P \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}\right\|$ | 2 JRNC  <br> 1 $e^{\text {JRN }}$  | $\left.\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b6,rr } & \\ 2 & \text { b.d.d } \end{array} \right\rvert\,$ | 2  JRZ <br>  e  <br> 1  pcr | $\left\|\begin{array}{lll} \hline 4 & & L D \\ & \mathrm{y}, \mathrm{a} & \\ 1 & & \mathrm{sd} \end{array}\right\|$ | 2 JRC  <br> 1 e  | 4  INC <br>  rr  <br> 2  $\operatorname{dir}$ | $\begin{gathered} 7 \\ 0111 \end{gathered}$ |
| $\begin{gathered} 8 \\ 1000 \end{gathered}$ | $\begin{array}{lll} \hline 2 & & \text { JRNZ } \\ & e^{\prime} & \\ 1 & & \mathrm{pcr} \end{array}$ | $\begin{array}{\|ccc\|} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 JRNC <br> 1 $e^{\text {por }}$ | $\left.\begin{array}{\|cc\|} \hline 4 & \text { RES } \\ & \text { b1,rr } \\ 2 & \text { b.d. } \end{array} \right\rvert\,$ | $\begin{array}{\|ccc\|} \hline 2 & & J R Z \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \\ \hline \end{array}$ | \# | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \mathrm{e} & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll\|} \hline 4 & & \text { LD } \\ & (y), a & \\ 1 & & \text { ind } \end{array}$ | $\begin{gathered} 8 \\ 1000 \end{gathered}$ |
| $\begin{gathered} 9 \\ 1001 \end{gathered}$ | 2 RNZ  <br>  e  <br> 1  pcr | $\begin{array}{\|ccc} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 JRNC  <br> 1   <br>   pcr | $\left.\begin{array}{\|cc\|} \hline 4 & \text { SET } \\ & \text { b1,rr } \\ 2 & \text { b.d. } \end{array} \right\rvert\,$ | $\left.\begin{array}{\|lll} \hline 2 & & \text { JRZ } \\ & \mathrm{e} & \\ 1 & & \mathrm{pcr} \end{array} \right\rvert\,$ | $\left\|\begin{array}{lll} \hline 4 & & \text { DEC } \\ & \mathrm{v} & \\ 1 & & \mathrm{sd} \end{array}\right\|$ | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | $\begin{array}{\|lll} \hline 4 & & \text { LD } \\ & \text { rr,a } & \\ 2 & & \operatorname{dir} \end{array}$ | $\begin{gathered} 9 \\ 1001 \end{gathered}$ |
| $\underset{1010}{A}$ | 2 $e^{\text {JRNZ }}$ <br> 1 $e^{2}$ | $\begin{array}{\|lll} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2  <br> $\mathrm{~J}^{\text {JRNC }}$  <br> 1  <br> e pcr | 4 RES <br>  b5,rr <br> 2 b.d | 2   <br>  JRZ  <br> 1  pcr | 4  $R C L$ <br>  a  <br> 1  inh | 2 JRC  <br> 1  prc | $\begin{array}{ll} \hline 4 & \text { AND } \\ & \mathrm{a},(\mathrm{y}) \\ 1 & \text { ind } \end{array}$ | $\begin{gathered} \text { A } \\ 1010 \end{gathered}$ |
| $\begin{gathered} B \\ 1011 \end{gathered}$ | 2 JRNZ <br> 1 $e^{\text {JRCr }}$ | $\begin{array}{\|ccc} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 JRNC  <br> 1 $e^{\text {JR }}$  | $\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b5,rr } & \\ 2 & \text { b.d.d } \end{array}$ | 2  JRZ <br>  e  <br> 1  pcr | $\left.\begin{array}{\|ccc\|} \hline 4 & & \text { LD } \\ & \mathrm{v}, \mathrm{a} & \\ 1 & & \mathrm{sd} \end{array} \right\rvert\,$ | 2  JRC <br> 1 e  <br> 1  prc | $\begin{array}{\|lll} \hline 4 & & \text { AND } \\ & \text { a,rr } & \\ 2 & & \operatorname{dir} \end{array}$ | $\begin{gathered} B \\ 1011 \end{gathered}$ |
| $\underset{1100}{C}$ | 2 $e^{\text {JRNZ }}$ <br> 1  | $\begin{array}{\|ccc} \hline 4 & & J P \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 $e^{\text {JRNC }}$ <br> 1  <br> pcr  | $\left.\begin{array}{\|cc\|} \hline 4 & \text { RES } \\ & \text { b3,rr } \\ 2 & \text { b.d } \end{array} \right\rvert\,$ | 2  JRZ <br>  e  <br> 1  pcr | 2 RET <br> 1 inh | 2 JRC  <br> 1   <br> prc   | 4 SUB <br>  $\mathrm{a},(\mathrm{y})$ <br> 1  <br> ind  | $\begin{gathered} \text { C } \\ 1100 \end{gathered}$ |
| $\begin{gathered} \text { D } \\ 1101 \end{gathered}$ | 2 $e^{\text {JRNZ }}$ <br> 1  <br>  pcr | $\left.\begin{array}{\|ccc} \hline 4 & & \mathrm{JP} \\ & \text { abc } & \\ 2 & & \text { ext } \end{array} \right\rvert\,$ | 2 JRNC <br> 1 $e^{\text {pcr }}$ | $\left.\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b3,rr } & \\ 2 & \text { b.d.d } \end{array} \right\rvert\,$ | 2  JRZ <br>  e  <br> 1  pcr | 4  DEC <br>  w  <br> 1  sd | $\begin{array}{\|lll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | 4  SUB <br>  a,rr  <br> 2  $\operatorname{dir}$ | $\begin{gathered} \text { D } \\ 1101 \end{gathered}$ |
| $\underset{1110}{E}$ | 2 $e^{\text {JRNZ }}$ <br> 1  <br> pcr  | $\begin{array}{llll}4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext }\end{array}$ | 2 JRNC <br> 1 $e^{\text {pcr }}$ | 4 RES <br>  b7,rr <br> 2 b.d | 2  JRZ <br> 1 $e$  <br> 1  pcr | $\begin{array}{\|cc} \hline 2 & \text { WAIT } \\ 1 & \text { inh } \end{array}$ | $\begin{array}{llll} \hline 2 & & \text { JRC } \\ & \text { e } & \\ 1 & & \text { prc } \end{array}$ | 4 DEC <br>  (y) <br> 1 ind | $\underset{1110}{E}$ |
| $\stackrel{F}{1111}$ | 2  JRNZ <br>  $e^{\text {JRN }}$  <br> 1  pcr | $\begin{array}{\|ccc\|} \hline 4 & & \text { JP } \\ & \text { abc } & \\ 2 & & \text { ext } \end{array}$ | 2 JRNC <br> 1 $e^{2}$ <br> 1  | $\begin{array}{\|ccc} \hline 4 & \text { SET } \\ & \text { b7,rr } & \\ 2 & \text { b.d } \end{array}$ | 2  JRZ <br>  e  <br> 1  pcr | $\begin{array}{\|ccc\|} \hline 4 & & \text { LD } \\ & \mathrm{w}, \mathrm{a} & \\ 1 & & \mathrm{sd} \end{array}$ | 2  JRC <br> 1   <br> prc   | $\begin{array}{\|lll} \hline 4 & & \text { DEC } \\ & \mathrm{rr} & \\ 2 & & \mathrm{dir} \end{array}$ | $\underset{1111}{F}$ |

Abbreviations for Addressing Modes:

| dir | Direct |
| :--- | :--- |
| sd | Short Direct |
| imm | Immediate |
| inh | Inherent |
| ext | Extended |
| b.d | Bit Direct |
| bt | Bit Test |
| pcr | Program Counter Relative |
| ind | Indirect |

Legend:

 Indirect

## 6 ELECTRICAL CHARACTERISTICS

### 6.1 ABSOLUTE MAXIMUM RATINGS

This product contains devices to protect the inputs against damage due to high static voltages, however it is advisable to take normal precaution to avoid application of any voltage higher than the specified maximum rated voltages.

For proper operation it is recommended that $\mathrm{V}_{\mathrm{I}}$ and $V_{0}$ be higher than $V_{S S}$ and lower than $V_{D D}$. Reliability is enhanced if unused inputs are connected to an appropriate logic voltage level ( $\mathrm{V}_{\mathrm{DD}}$ or $\mathrm{V}_{\mathrm{SS}}$ ).

Power Considerations.The average chip-junction temperature, Tj, in Celsius can be obtained from:

> Tj=TA + PD x RthJA

Where:TA = Ambient Temperature.
RthJA =Package thermal resistance (junc-tion-to ambient).
PD = Pint + Pport.
Pint =IDD $\times$ VDD (chip internal power).
Pport =Port power dissipation (determined by the user).

| Symbol | Parameter | Value | Unit |
| :---: | :--- | :---: | :---: |
| $\mathrm{V}_{\mathrm{DD}}$ | Supply Voltage | -0.3 to 7.0 | V |
| $\mathrm{~V}_{\mathrm{I}}$ | Input Voltage | $\mathrm{V}_{\mathrm{SS}}-0.3$ to $\mathrm{V}_{\mathrm{DD}}+0.3^{(1)}$ | V |
| $\mathrm{V}_{\mathrm{O}}$ | Output Voltage | $\mathrm{V}_{\mathrm{SS}}-0.3$ to $\mathrm{V}_{\mathrm{DD}}+0.3^{(1)}$ | V |
| $\mathrm{V}_{\mathrm{DD}}$ | Total Current into $\mathrm{V}_{\mathrm{DD}}$ (source) | 80 | mA |
| $\mathrm{IV}_{\mathrm{SS}}$ | Total Current out of $\mathrm{V}_{\mathrm{SS}}$ (sink) | 100 | mA |
| Tj | Junction Temperature | 150 | ${ }^{\circ} \mathrm{C}$ |
| $\mathrm{T}_{\mathrm{STG}}$ | Storage Temperature | -60 to 150 | ${ }^{\circ} \mathrm{C}$ |

Notes:

- Stresses above those listed as "absolute maximum ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.
- (1) Within these limits, clamping diodes are guarantee to be not conductive. Voltages outside these limits are authorised as long as injection current is kept within the specification.


### 6.2 RECOMMENDED OPERATING CONDITIONS

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $\mathrm{T}_{\mathrm{A}}$ | Operating Temperature | 6 Suffix Version 1 Suffix Version 3 Suffix Version | $\begin{gathered} \hline-40 \\ 0 \\ -40 \end{gathered}$ |  | $\begin{gathered} \hline 85 \\ 70 \\ 125 \end{gathered}$ | ${ }^{\circ} \mathrm{C}$ |
| $V_{\text {DD }}$ | Operating Supply Voltage | $\begin{aligned} & \text { fosc }=4 \mathrm{MHz}, 1 \& 6 \text { Suffix } \\ & \text { fosc }=4 \mathrm{MHz}, 3 \text { Suffix } \\ & \text { fosc }=8 \mathrm{MHz}, 1 \& 6 \text { Suffix } \\ & \text { fosc }=8 \mathrm{MHz}, 3 \text { Suffix } \end{aligned}$ | $\begin{aligned} & 3.0 \\ & 3.0 \\ & 3.6 \\ & 4.5 \end{aligned}$ |  | $\begin{aligned} & \hline 6.0 \\ & 6.0 \\ & 6.0 \\ & 6.0 \end{aligned}$ | V |
| ${ }_{\text {fosc }}$ | Oscillator Frequency ${ }^{2}$ | $\begin{aligned} & \hline V_{D D}=3.0 \mathrm{~V}, 1 \& 6 \text { Suffix } \\ & \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, 3 \text { Suffix } \\ & \mathrm{V}_{\mathrm{DD}}=3.6 \mathrm{~V}, 1 \& 6 \text { Suffix } \\ & \mathrm{V}_{\mathrm{DD}}=3.6 \mathrm{~V}, 3 \text { Suffix } \\ & \hline \end{aligned}$ | $\begin{aligned} & \hline 0 \\ & 0 \\ & 0 \\ & 0 \\ & \hline \end{aligned}$ |  | $\begin{aligned} & 4.0 \\ & 4.0 \\ & 8.0 \\ & 4.0 \\ & \hline \end{aligned}$ | MHz |
| $\mathrm{I}_{\text {INJ }+}$ | Pin Injection Current (positive) | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V |  |  | +5 | mA |
| $\mathrm{I}_{\text {INJ }-}$ | Pin Injection Current (negative) | $\mathrm{V}_{\mathrm{DD}}=4.5$ to 5.5 V |  |  | -5 | mA |

Notes:

1. Care must be taken in case of negative current injection, where adapted impedance must be respected on analog sources to not affect the A/D conversion. For a -1 mA injection, a maximum $10 \mathrm{~K} \Omega$ is recommended.
2.An oscillator frequency above 1 MHz is recommended for reliable $A / D$ results

Figure 27. Maximum Operating FREQUENCY (Fmax) Versus SUPPLY VOLTAGE (VDD


The shaded area is outside the recommended operating range; device functionality is not guaranteed under these conditions.

### 6.3 DC ELECTRICAL CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=-40$ to $+125^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $\mathrm{V}_{\text {IL }}$ | Input Low Level Voltage All Input pins |  |  |  | $V_{D D} \times 0.3$ | V |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Level Voltage All Input pins |  | $V_{D D} \times 0.7$ |  |  | V |
| $\mathrm{V}_{\text {Hys }}$ | Hysteresis Voltage ${ }^{(1)}$ All Input pins | $\begin{aligned} & V_{D D}=5 \mathrm{~V} \\ & V_{D D}=3 \mathrm{~V} \end{aligned}$ | $\begin{aligned} & 0.2 \\ & 0.2 \end{aligned}$ |  |  | V |
| $\mathrm{V}_{\text {up }}$ | LVD Threshold in power-on |  |  | 4.1 | 4.3 |  |
| $\mathrm{V}_{\mathrm{dn}}$ | LVD threshold in powerdown |  | 3.5 | 3.8 |  |  |
| $\mathrm{V}_{\text {OL }}$ | Low Level Output Voltage All Output pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+3 \mathrm{~mA} \end{aligned}$ |  |  | $\begin{aligned} & \hline 0.1 \\ & 0.8 \\ & \hline \end{aligned}$ | V |
|  | Low Level Output Voltage 20 mA Sink I/O pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+7 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+15 \mathrm{~mA} \\ & \hline \end{aligned}$ |  |  | $\begin{aligned} & 0.1 \\ & 0.8 \\ & 1.3 \end{aligned}$ |  |
| $\mathrm{V}_{\mathrm{OH}}$ | High Level Output Voltage All Output pins | $\begin{array}{\|l} \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OH}}=-10 \mu \mathrm{~A} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OH}}=-3.0 \mathrm{~mA} \\ \hline \end{array}$ | $\begin{aligned} & 4.9 \\ & 3.5 \end{aligned}$ |  |  | V |
| $\mathrm{R}_{\mathrm{PU}}$ | Pull-up Resistance | All Input pins | 40 | 100 | 350 | $\mathrm{K} \Omega$ |
|  |  | RESET pin | 150 | 350 | 900 |  |
| $\begin{aligned} & I_{\mathrm{IL}} \\ & I_{\mathrm{IH}} \end{aligned}$ | Input Leakage Current All Input pins but RESET | $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\text {SS }}$ (No Pull-Up configured) $V_{I N}=V_{D D}$ |  | 0.1 | 1.0 | $\mu \mathrm{A}$ |
|  | Input Leakage Current RESET pin | $\begin{aligned} & V_{I N}=V_{S S} \\ & V_{I N}=V_{D D} \end{aligned}$ | -8 | -16 | $\begin{gathered} -30 \\ 10 \end{gathered}$ |  |
| IDD | Supply Current in RESET Mode | $\begin{aligned} & \mathrm{V}_{\mathrm{RESET}}=\mathrm{V}_{\mathrm{SS}} \\ & \mathrm{f}_{\mathrm{OSC}}=8 \mathrm{MHz} \end{aligned}$ |  |  | 3.5 | mA |
|  | Supply Current in RUN Mode (2) | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \quad \mathrm{f}_{\text {INT }}=8 \mathrm{MHz}$ |  |  | 3.5 | mA |
|  | Supply Current in WAIT Mode ${ }^{(3)}$ | $\mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} \quad \mathrm{f}_{\text {INT }}=8 \mathrm{MHz}$ |  |  | 500 | $\mu \mathrm{A}$ |
|  | Supply Current in WAIT Mode ${ }^{(4)}$ | $\mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V} \quad \mathrm{f}_{\text {INT }}=32 \mathrm{~K}$ |  |  | 30 | $\mu \mathrm{A}$ |
|  | $\begin{aligned} & \hline \text { Supply Current in STOP } \\ & \text { Mode, with LVD disabled }{ }^{(3)} \end{aligned}$ | $\begin{aligned} & \hline \mathrm{LOAD}=0 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} \end{aligned}$ |  |  | 20 | $\mu \mathrm{A}$ |
|  | Supply Current in STOP <br> Mode, with LVD enabled ${ }^{(3)}$ | $\begin{aligned} & \mathrm{I}_{\mathrm{LOAD}}=0 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} \end{aligned}$ |  |  | 500 |  |
| Retention | EPROM Data Retention | $\mathrm{T}_{\mathrm{A}}=55^{\circ} \mathrm{C}$ | 10 |  |  | years |

## Notes:

(1) Hysteresis voltage between switching levels
(2) All peripherals running
(3) All peripherals in stand-by + option byte programmed (except LVD)
(4) Characterized but not tested; option byte programmed except LVD

## DC ELECTRICAL CHARACTERISTICS (Cont'd)

( $\mathrm{T}_{\mathrm{A}}=-40$ to $+85^{\circ} \mathrm{C}$ unless otherwise specified))

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $\mathrm{V}_{\text {up }}$ | LVD Threshold in power-on |  | $\mathrm{V}_{\mathrm{dn}}+50 \mathrm{mV}$ | 4.1 | 4.3 | V |
| $\mathrm{V}_{\mathrm{dn}}$ | LVD threshold in powerdown |  | 3.6 | 3.8 | $\mathrm{V}_{\text {up }}-50 \mathrm{mV}$ | V |
| VoL | Low Level Output Voltage All Output pins | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OL}}=+5 \mathrm{~mA} \\ & \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+10 \mathrm{mAv} \\ & \hline \end{aligned}$ |  |  | $\begin{aligned} & \hline 0.1 \\ & 0.8 \\ & 1.2 \\ & \hline \end{aligned}$ | V |
|  | Low Level Output Voltage 20 mA Sink I/O pins | $\begin{array}{\|l\|} \hline \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+10 \mu \mathrm{~A} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+10 \mathrm{~mA} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+20 \mathrm{~mA} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OL}}=+30 \mathrm{~mA} \\ \hline \end{array}$ |  |  | $\begin{aligned} & \hline 0.1 \\ & 0.8 \\ & 1.3 \\ & 2.0 \\ & \hline \end{aligned}$ |  |
| $\mathrm{V}_{\mathrm{OH}}$ | High Level Output Voltage <br> All Output pins | $\begin{array}{\|l} \hline \mathrm{V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{I}_{\mathrm{OH}}=-10 \mu \mathrm{~A} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} ; \mathrm{l}_{\mathrm{OH}}=-5.0 \mathrm{~mA} \\ \hline \end{array}$ | $\begin{aligned} & 4.9 \\ & 3.5 \end{aligned}$ |  |  | V |
| IDD | Supply Current in STOP Mode, with LVD disabled ${ }^{(*)}$ | $\begin{aligned} & \begin{array}{l} \mathrm{IOAD}=0 \mathrm{~mA} \\ \mathrm{~V}_{\mathrm{DD}}=5.0 \mathrm{~V} \end{array} \\ & \hline \end{aligned}$ |  |  | 10 | $\mu \mathrm{A}$ |

Note:
(*) All Peripherals in stand-by. $^{\text {a }}$

### 6.4 AC ELECTRICAL CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=-40$ to $+125^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $t_{\text {REC }}$ | Supply Recovery Time ${ }^{(1)}$ |  | 100 |  |  | ms |
| $\mathrm{f}_{\text {LFAO }}$ | Internal frequency with LFAO active |  | 200 | 400 | 800 | kHz |
| ${ }_{\text {fosG }}$ | Internal Frequency with OSG enabled ${ }^{2)}$ | $\begin{aligned} & \hline \mathrm{V}_{\mathrm{DD}}=3 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DD}}=3.6 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DD}}=4.5 \mathrm{~V} \end{aligned}$ | 2 2 4 |  | ${ }_{\text {fosc }}$ | MHz |
| $\mathrm{f}_{\mathrm{RC}}$ | Internal frequency with RC oscillator and OSG disabled ${ }^{2)}$ 3) | $\begin{aligned} & \mathrm{VDD}=5.0 \mathrm{~V} \\ & \mathrm{R}=47 \mathrm{k} \Omega \\ & \mathrm{R}=100 \mathrm{k} \Omega \\ & \mathrm{R}=470 \mathrm{k} \Omega \end{aligned}$ | $\begin{gathered} 4 \\ 2.7 \\ 800 \end{gathered}$ | $\begin{gathered} 5 \\ 3.2 \\ 850 \end{gathered}$ | $\begin{aligned} & 5.8 \\ & 3.5 \\ & 900 \end{aligned}$ | $\begin{aligned} & \mathrm{MHz} \\ & \mathrm{MHz} \\ & \mathrm{kHz} \end{aligned}$ |
| $\mathrm{C}_{\text {IN }}$ | Input Capacitance | All Inputs Pins |  |  | 10 | pF |
| $\mathrm{C}_{\text {OUT }}$ | Output Capacitance | All Outputs Pins |  |  | 10 | pF |

Notes:

1. Period for which $V_{D D}$ has to be connected at $O V$ to allow internal Reset function at next power-up

2 An oscillator frequency above 1 MHz is recommended for reliable A/D results.
3. Measure performed with OSCin pin soldered on PCB, with an around $2 p F$ equivalent capacitance.

### 6.5 A/D CONVERTER CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=-40$ to $+125^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| Res | Resolution |  |  | 8 |  | Bit |
| A $_{\text {TOT }}$ | Total Accuracy ${ }^{(1)(2)}$ | $\begin{aligned} & \mathrm{f}_{\mathrm{Osc}}>1.2 \mathrm{MHz} \\ & \mathrm{f}_{\mathrm{osc}}>32 \mathrm{kHz} \\ & \hline \end{aligned}$ |  |  | $\begin{array}{r}  \pm 2 \\ \pm 4 \\ \hline \end{array}$ | LSB |
| ${ }^{\text {t }}$ | Conversion Time | $\begin{aligned} & \mathrm{f}_{\mathrm{OSC}}=8 \mathrm{MHz}\left(\mathrm{~T}_{\mathrm{A}}<85^{\circ} \mathrm{C}\right) \\ & \mathrm{f}_{\mathrm{SSC}}=4 \mathrm{MHz} \end{aligned}$ |  | $\begin{gathered} \hline 70 \\ 140 \end{gathered}$ |  | $\mu \mathrm{S}$ |
| ZIR | Zero Input Reading | Conversion result when $\mathrm{V}_{\text {IN }}=\mathrm{V}_{\mathrm{SS}}$ | 00 |  |  | Hex |
| FSR | Full Scale Reading | Conversion result when $\mathrm{V}_{\mathrm{IN}}=\mathrm{V}_{\mathrm{DD}}$ |  |  | FF | Hex |
| $A D_{1}$ | Analog Input Current During Conversion | $V_{D D}=4.5 \mathrm{~V}$ |  |  | 1.0 | $\mu \mathrm{A}$ |
| $\mathrm{AC}_{\text {IN }}$ | Analog Input Capacitance |  |  | 2 | 5 | pF |

## Notes:

1. Noise at VDD, VSS $<10 \mathrm{mV}$
2. With oscillator frequencies less than 1 MHz , the $\mathrm{A} / \mathrm{D}$ Converter accuracy is decreased.

### 6.6 TIMER CHARACTERISTICS

( $\mathrm{T}_{\mathrm{A}}=-40$ to $+125^{\circ} \mathrm{C}$ unless otherwise specified)

| Symbol | Parameter | Test Conditions | Value |  |  | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | Min. | Typ. | Max. |  |
| $\mathrm{f}_{\mathrm{IN}}$ | Input Frequency on TIMER Pin |  |  |  | $\frac{\mathrm{IINT}}{4}$ | MHz |
| tw | Pulse Width at TIMER Pin | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V} \\ & \mathrm{~V}_{\mathrm{DD}}>4.5 \mathrm{~V} \end{aligned}$ | $\begin{gathered} \hline 1 \\ 125 \end{gathered}$ |  |  | $\begin{aligned} & \mu \mathrm{s} \\ & \mathrm{~ns} \end{aligned}$ |

Figure 28.. RC frequency versus Vcc


Figure 29. LVD thresholds versus temperature

Figure 30. Idd WAIT versus Vcc at $\mathbf{8} \mathbf{~ M h z ~ f o r ~ O T P ~ d e v i c e s ~}$


Figure 31. Idd STOP versus Vcc for OTP devices


Figure 32. Idd STOP versus Vcc for ROM devices


This curves represents typical variations and is given for guidance only

Figure 33. Idd WAIT versus Vcc at 8 Mhz for ROM devices


Figure 34. Idd WAIT ( $\mu \mathrm{A}$ ) Fosc= $=32 \mathrm{KHz}$ (option byte programmed to 00h)


Figure 35. Idd RUN versus Vcc at 8 Mhz for ROM and OTP devices


This curves represents typical variations and is given for guidance only

Figure 36. Vol versus Iol on all I/O port at Vdd=5V


This curves represents typical variations and is given for guidance only
Figure 37. Vol versus lol on all I/O port at $\mathrm{T}=\mathbf{2 5}^{\circ} \mathrm{C}$


Figure 38. Vol versus lol for High sink (20mA) I/Oports at $\mathrm{T}=\mathbf{2 5}^{\circ} \mathrm{C}$


Figure 39. Vol versus lol for High sink ( 20 mA ) I/O ports at Vdd=5V


Figure 40 . Voh versus loh on all I/O port at $25^{\circ} \mathrm{C}$


Figure 41. Voh versus loh on all I/O port at Vdd=5V


## 7 GENERAL INFORMATION

### 7.1 PACKAGE MECHANICAL DATA

Figure 42. 28-Pin Plastic Dual In-Line Package, 600-mil Width


Figure 43. 28-Pin Plastic Small Outline Package, 300-mil Width


PACKAGE MECHANICAL DATA (Cont'd)
Figure 44. 28-Pin Ceramic Side-Brazed Dual In-Line Package


Figure 45. 28-Pin Plastic Shrink Small Outline Package, 0.209" Width


## 7.2 . ORDERING INFORMATION

Table 19. OTP/EPROM VERSION ORDERING INFORMATION

| Sales Type | Program Memory (Bytes) | I/O | Temperature Range | Package |
| :---: | :---: | :---: | :---: | :---: |
| ST62E25CF1 | 3884 (EPROM) | 20 | 0 to $+70^{\circ} \mathrm{C}$ | CDIP28W |
| ST62T15CB6 |  |  | -40 to $+85^{\circ} \mathrm{C}$ | PDIP28 |
| ST62T15CM6 | 1836 (OTP) |  |  | PSO28 |
| ST62T25CB6 | 3884 (OTP) |  |  | PDIP28 |
| ST62T25CM6 |  |  |  | PSO28 |
| ST62T25CN6 |  |  |  | SSOP28 |
| ST62T25CB3 | 3884 (OTP) |  | -40 to $+125^{\circ} \mathrm{C}$ | PDIP28 |
| ST62T25CM3 |  |  |  | PSO28 |
| ST62T25CN3 |  |  |  | SSOP28 |

# 8-BIT FASTROM MCUs WITH A/D CONVERTER, OSCILLATOR SAFEGUARD, SAFE RESET AND 28 PINS 

- 3.0 to 6.0 V Supply Operating Range
- 8 MHz Maximum Clock Frequency
- -40 to $+125^{\circ} \mathrm{C}$ Operating Temperature Range
- Run, Wait and Stop Modes
- 5 Interrupt Vectors
- Look-up Table capability in Program Memory
- Data Storage in Program Memory:

User selectable size

- Data RAM: 64bytes
- 20 I/O pins, fully programmable as:
- Input with pull-up resistor
- Input without pull-up resistor
- Input with interrupt generation
- Open-drain or push-pull output
- Analog Input
- 4 I/O lines can sink up to 20 mA to drive LEDs or TRIACs directly
- 8-bit Timer/Counter with 7 -bit programmable prescaler
- Digital Watchdog
- Oscillator Safe Guard
- Low Voltage Detector for Safe Reset
- 8-bit A/D Converter with 16 analog inputs
- On-chip Clock oscillator can be driven by Quartz Crystal Ceramic resonator or RC network
- Power-on Reset
- One external Non-Maskable Interrupt
- ST626x-EMU2 Emulation and Development System (connects to an MS-DOS PC via a parallel port).


PDIP28


PS028


SSOP28
(See end of Datasheet for Ordering Information)

## DEVICE SUMMARY

| DEVICE | ROM <br> (Bytes) |
| :---: | :---: |
| ST62P15C | 1836 |
| ST62P25C | 3884 |

## 1 GENERAL DESCRIPTION

### 1.1 INTRODUCTION

The ST62P15C/P25C are the Factory Advanced Service Technique ROM (FASTROM) versions of ST62T15C,T25C OTP devices.

They offer the same functionality as OTP devices, selecting as FASTROM options the options defined in the programmable option byte of the OTP version.

### 1.2 ORDERING INFORMATION

The following section deals with the procedure for transfer of customer codes to STMicroelectronics.

### 1.2.1 Transfer of Customer Code

Customer code is made up of the ROM contents and the list of the selected FASTROM options. The ROM contents are to be sent on diskette, or by electronic means, with the hexadecimal file generated by the development tool. All unused bytes must be set to FFh.
The selected options are communicated to STMicroelectronics using the correctly filled OPTION LIST appended.

### 1.2.2 Listing Generation and Verification

When STMicroelectronics receives the user's ROM contents, a computer listing is generated from it. This listing refers exactly to the ROM contents and options which will be used to produce

Table 3. FASTROM version Ordering Information

| Sales Type | ROM | Temperature Range | Package |
| :---: | :---: | :---: | :---: |
| ST62P15CB1/XXX <br> ST62P15CB6/XXX <br> ST62P15CB3/XXX (*) | 1836 Bytes | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PDIP28 |
| ST62P15CM1/XXX ST62P15CM6/XXX ST62P15CM3/XXX (*) |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PSO28 |
| $\begin{aligned} & \text { ST62P25CB1/XXX } \\ & \text { ST62P25CB6/XXX } \\ & \text { ST62P25CB3/XXX (*) } \end{aligned}$ | 3884 Bytes | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PDIP28 |
| ST62P25CM1/XXX ST62P25CM6/XXX ST62P25CM3/XXX (*) |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PSO28 |
| ST62P25CN1/XXX ST62P25CN6/XXX ST62P25CN3/XXX (*) |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | SSOP28 |

(*) Advanced information $^{\text {( }}$

## ST62P15C/P25C FASTROM MICROCONTROLLER OPTION LIST

| Customer <br> Address |  |  |  |
| :---: | :---: | :---: | :---: |
|  |  |  |  |
| Contact |  |  |  |
| Phone No |  |  |  |
| Reference |  |  |  |
| STMicroelectronics references |  |  |  |
| Device: | [] ST62P15C | [] ST62P25C |  |
| Package: | [] Dual in Line Plastic |  |  |
|  | [] Small Outline Plastic with conditioning: |  |  |
|  | [ ] Standard (Stick) |  |  |
|  |  | [ ] Tape \& Reel |  |
|  | [] Shrink Small Outline Plastic |  |  |
| Temperature Range: | [] $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | [] - $40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | [] $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Oscillator Source Selection: [ ] Crystal Quartz/Ceramic resonator |  |  |  |
|  | [] RC Network |  |  |
| Watchdog Selection: | [] Software Activation[] Hardware Activation |  |  |
| Readout Protection: | [] Disabled |  |  |
|  | [] Enabled |  |  |
| External STOP Mode Control[ ] Enabled |  | []Disabled |  |
| LVD Reset | [] Enabled | []Disabled |  |
| TIMER pin pull-up | [] Enabled | []Disabled |  |
| NMI pin pull-up | [] Enabled | []Disabled |  |
| OSG | [] Enabled | [] Disabled |  |
| Comments: |  |  |  |
| Supply Operating Range in the application: |  |  |  |
| Oscillator Frequency in the application: |  |  |  |
| Notes |  |  |  |
| Signature |  |  |  |
| Date |  |  |  |

Notes:

## 8-BIT ROM MCUs WITH A/D CONVERTER, OSCILLATOR SAFEGUARD, SAFE RESET AND 28 PINS

- 3.0 to 6.0V Supply Operating Range
- 8 MHz Maximum Clock Frequency
- -40 to $+125^{\circ} \mathrm{C}$ Operating Temperature Range
- Run, Wait and Stop Modes
- 5 Interrupt Vectors
- Look-up Table capability in Program Memory
- Data Storage in Program Memory:

User selectable size

- Data RAM: 64bytes
- 20 I/O pins, fully programmable as:
- Input with pull-up resistor
- Input without pull-up resistor
- Input with interrupt generation
- Open-drain or push-pull output
- Analog Input
- 4 I/O lines can sink up to 20 mA to drive LEDs or TRIACs directly
- 8-bit Timer/Counter with 7 -bit programmable prescaler
- Digital Watchdog
- Oscillator Safe Guard
- Low Voltage Detector for Safe Reset
- 8-bit A/D Converter with 16 analog inputs
- On-chip Clock oscillator can be driven by Quartz Crystal Ceramic resonator or RC network
- Power-on Reset
- One external Non-Maskable Interrupt
- ST626x-EMU2 Emulation and Development System (connects to an MS-DOS PC via a parallel port).


PDIP28


PS028


SSOP28
(See end of Datasheet for Ordering Information)

## DEVICE SUMMARY

| DEVICE | ROM <br> (Bytes) |
| :---: | :---: |
| ST6215C | 1836 |
| ST6225C | 3884 |

Rev. 2.8

## 1 GENERAL DESCRIPTION

### 1.1 INTRODUCTION

The ST6215C/25C are mask programmed ROM version of ST62T15C,T25C OTP devices.
They offer the same functionality as OTP devices, selecting as ROM options the options defined in the programmable option byte of the OTP version.

Figure 1. Programming wave form


### 1.2 ROM READOUT PROTECTION

If the ROM READOUT PROTECTION option is selected, a protection fuse can be blown to prevent any access to the program memory content. In case the user wants to blow this fuse, high voltage must be applied on the TEST pin.

Figure 2. Programming Circuit


Note: ZPD15 is used for overvoltage protection

## ST6215C/25C MICROCONTROLLER OPTION LIST

Customer
Address
Contact
Phone No
Reference

STMicroelectronics references

| Device: | [] ST6215C | [] ST6225C |  |
| :---: | :---: | :---: | :---: |
| Package: | [] Dual in Line Plastic |  |  |
|  | [ ] Small Outline Plastic with conditioning: |  |  |
|  | [ ] Standard (Stick) |  |  |
|  | [ ] Tape \& Reel |  |  |
|  | [] Shrink Small Outline Plastic |  |  |
| Temperature Range: | [] $0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$ | [] $-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ | [] $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ |
| Special Marking: | []No | []Yes" |  |

Authorized characters are letters, digits, ' $\quad .$, ',',' '/' and spaces only.
Maximum character count: DIP28: 10
SO28: 8
Oscillator Source Selection: [ ] Crystal Quartz/Ceramic resonator
[] RC Network
Watchdog Selection: [] Software Activation
[] Hardware Activation
ROM Readout Protection: [ ] Disabled (Fuse cannot be blown)
[ ] Enabled (Fuse can be blown by the customer)
Note: $\quad$ No part is delivered with protected ROM.
The fuse must be blown for protection to be effective.

| External STOP Mode Controll ] Enabled | [] Disabled |  |
| :--- | ---: | ---: |
| LVD Reset | [] Enabled | [] Disabled |
| TIMER pin pull-up | [] Enabled | []Disabled |
| NMI pin pull-up | [] Enabled | [] Disabled |
| OSG | [] Enabled | []Disabled |

Comments:
Supply Operating Range in the application:
Oscillator Frequency in the application:
Notes
Signature
Date

### 1.3 ORDERING INFORMATION

The following section deals with the procedure for transfer of customer codes to STMicroelectronics.

### 1.3.1 Transfer of Customer Code

Customer code is made up of the ROM contents and the list of the selected mask options. The ROM contents are to be sent on diskette, or by electronic means, with the hexadecimal file generated by the development tool. All unused bytes must be set to FFh.
The selected mask options are communicated to STMicroelectronics using the correctly filled OPTION LIST appended.

### 1.3.2 Listing Generation and Verification

When STMicroelectronics receives the user's ROM contents, a computer listing is generated from it. This listing refers exactly to the mask which will be used to produce the specified MCU. The listing is then returned to the customer who must thoroughly check, complete, sign and return it to STMicroelectronics. The signed listing forms a part of the contractual agreement for the creation of the specific customer mask.

The STMicroelectronics Sales Organization will be pleased to provide detailed information on contractual points.

Table 1. ROM Memory Map for ST6215C

| Device Address | Description |
| :---: | :---: |
| 0000h-087Fh | Reserved |
| 0880h-0FFFh | User ROM |
| OFAOh-0FEFh | Reserved |
| OFFOh-0FF7h | Interrupt Vectors |
| 0FF8h-0FFBh | Reserved |
| OFFCh-0FFDh | NMI Interrupt Vector |
| 0FFEh-0FFFh | Reset Vector |

Table 2. ROM Memory Map for ST6225C

| Device Address | Description |
| :---: | :---: |
| 0000h-007Fh | Reserved |
| 0080h-0F9Fh | User ROM |
| OFAOh-0FEFh | Reserved |
| OFFOh-0FF7h | Interrupt Vectors |
| OFF8h-0FFBh | Reserved |
| OFFCh-0FFDh | NMI Interrupt Vector |
| OFFEh-0FFFh | Reset Vector |

## ORDERING INFORMATION (Cont'd)

Table 3. ROM version Ordering Information

| Sales Type | ROM | Temperature Range | Package |
| :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { ST6215CB1/XXX } \\ & \text { ST6215CB6/XXX } \\ & \text { ST6215CB3/XXX } \end{aligned}$ | 1836 Bytes | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PDIP28 |
| ST6215CM1/XXX ST6215CM6/XXX ST6215CM3/XXX |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PSO28 |
| $\begin{aligned} & \text { ST6225CB1/XXX } \\ & \text { ST6225CB6/XXX } \\ & \text { ST6225CB3/XXX } \end{aligned}$ | 3884 Bytes | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PDIP28 |
| $\begin{aligned} & \text { ST6225CM1/XXX } \\ & \text { ST6225CM6/XXX } \\ & \text { ST6225CB3/XXX } \end{aligned}$ |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | PSO28 |
| $\begin{aligned} & \hline \text { ST6225CN1/XXX } \\ & \text { ST6225CN6/XXX } \\ & \text { ST6225CN3/XXX } \end{aligned}$ |  | $\begin{gathered} 0 \text { to }+70^{\circ} \mathrm{C} \\ -40 \text { to }+85^{\circ} \mathrm{C} \\ -40 \text { to }+125^{\circ} \mathrm{C} \end{gathered}$ | SSOP28 |

## Notes:

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