

Features

- 8ns typ. propagation delay
- 5V to 12V input supply
- +2.7V to +5V output supply
- True-to-ground input
- Rail-to-rail outputs
- Active low latch
- Single available (EL5181C)
- Quad available (EL5481C & EL5482C)
- Pin-compatible 4ns family available (EL5x85C, EL5287C & EL5486C)

Applications

- Threshold detection
- High speed sampling circuits
- High speed triggers
- Line receivers
- PWM circuits
- High speed V/F converters

Ordering Information

Part No.	Package	Tape & Reel	Outline #
EL5283CY	10-Pin MSOP	-	MDP0043
EL5283CY-T13	10-Pin MSOP	13"	MDP0043

General Description

The EL5283C comparator is designed for operation in single supply and dual supply applications with 5V to 12V between V_{S+} and V_{S-} . For single supplies, the inputs can operate from 0.1V below ground for use in ground-sensing applications.

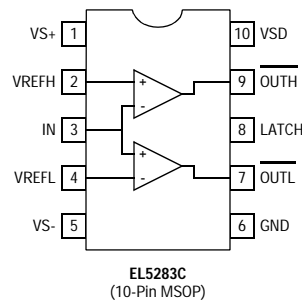
The output side of the comparators can be supplied from a single supply of 2.7V to 5V. The rail-to-rail output swing enables direct connection of the comparator to both CMOS and TTL logic circuits.

The latch input of the EL5283C can be used to hold the comparator output value by applying a low logic level to the pin.

The EL5283C contains two comparators set up as a window comparator. A single input is compared with a high and low reference. When the output goes beyond one of these reference signals, the relevant output goes high.

The EL5283C in the 10-pin MSOP package and is specified for operation over the full -40°C to $+85^{\circ}\text{C}$ temperature range. Also available are a single (EL5181C) and quad versions (EL5481C and EL5482C).

Pin Configurations



Note: All information contained in this data sheet has been carefully checked and is believed to be accurate as of the date of publication; however, this data sheet cannot be a "controlled document". Current revisions, if any, to these specifications are maintained at the factory and are available upon your request. We recommend checking the revision level before finalization of your design documentation.

EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

Absolute Maximum Ratings ($T_A = 25^\circ\text{C}$)

Absolute maximum ratings are those values beyond which the device could be permanently damaged. Absolute maximum ratings are stress ratings only and functional device operation is not implied.

Analog Supply Voltage (V_{S+} to V_{S-})	+13.2V
Digital Supply Voltage (V_{SD} to GND)	+7V
Differential Input Voltage	$[(V_{S-}) - 0.2V]$ to $[(V_{S+}) + 0.2V]$

Common-mode Input Voltage	$[(V_{S-}) - 0.2V]$ to $[(V_{S+}) + 0.2V]$
Latch Input Voltage	-0.2V to $[(V_{SD}) + 0.2V]$
Storage Temperature Range	-65°C to +150°C
Ambient Operating Temperature	-40°C to +85°C
Operating Junction Temperature	125°C
Power Dissipation	See Curves

Important Note:

All parameters having Min/Max specifications are guaranteed. Typ values are for information purposes only. Unless otherwise noted, all tests are at the specified temperature and are pulsed tests, therefore: $T_J = T_C = T_A$.

Electrical Characteristics

$V_S = \pm 5V$, $V_{SD} = 5V$, $R_L = 2.3k\Omega$, $C_L = 15pF$, $T_A = 25^\circ\text{C}$, unless otherwise specified.

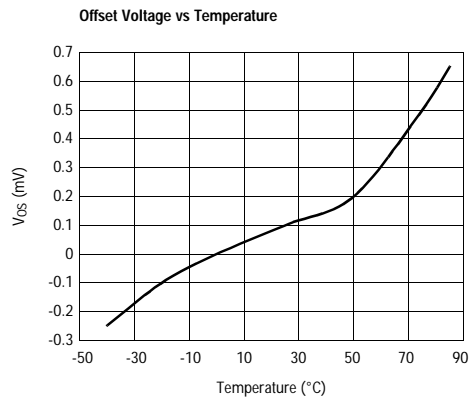
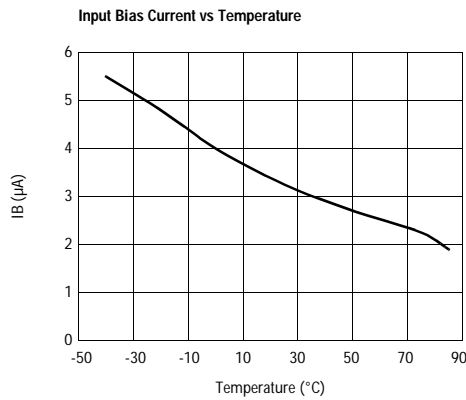
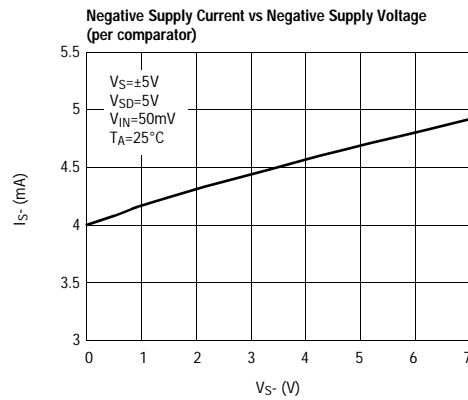
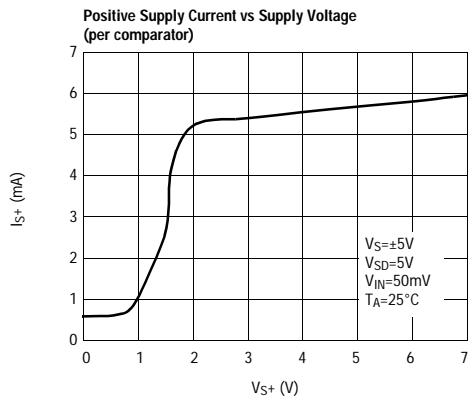
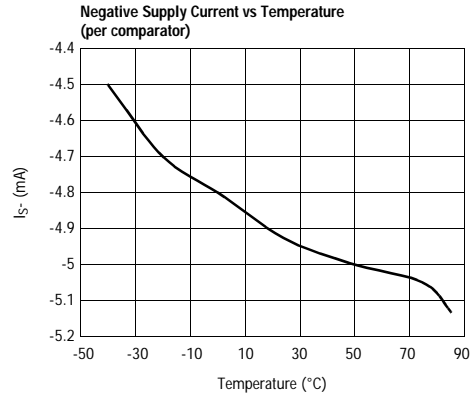
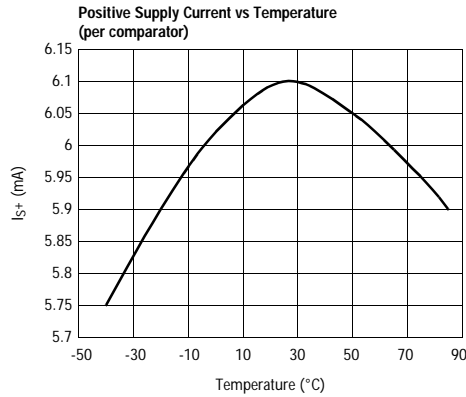
Parameter	Description	Condition	Min	Typ	Max	Unit
Input						
V_{OS}	Input Offset Voltage	$V_{CM} = 0V$, $V_O = 2.5V$		1	4	mV
I_B	Input Bias Current		-6	-3.5		μA
C_{IN}	Input Capacitance			5		pF
I_{OS}	Input Offset Current	$V_{CM} = 0V$, $V_O = 2.5V$	-2.5	0.5	2.5	μA
V_{CM}	Input Voltage Range		$(V_{S-}) - 0.1$		$(V_{S+}) - 2.25$	V
CMRR	Common-mode Rejection Ratio	$-5.1V < V_{CM} < +2.75V$	65	90		dB
Output						
V_{OH}	Output High Voltage	$V_{IN} > 250mV$	$V_{SD} - 0.6$	$V_{SD} - 0.4$		V
V_{OL}	Output Low Voltage	$V_{IN} > 250mV$		GND + 0.25	GND + 0.5	V
Dynamic Performance						
t_{pd+}	Positive Going Delay Time	$V_{IN} = 1V_{P-P}$, $V_{OD} = 50mV$		8	12	ns
t_{pd-}	Negative Going Delay Time	$V_{IN} = 1V_{P-P}$, $V_{OD} = 50mV$		8	12	ns
Supply						
I_{S+}	Positive Analog Supply Current	(per comparator)		7	8.2	mA
I_{S-}	Negative Analog Supply Current	(per comparator)		5	6.5	mA
I_{SD}	Digital Supply Current	(per comparator) All inputs high		4	5	mA
		(per comparator) All inputs low		0.75	1	mA
PSRR	Power Supply Rejection Ratio		60	80		dB

EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

EL5283C - Preliminary

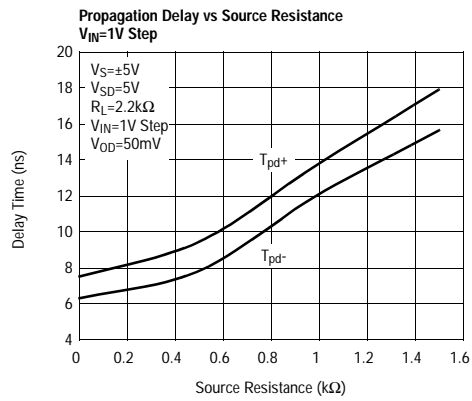
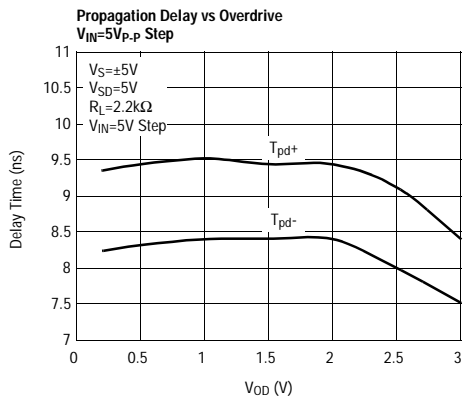
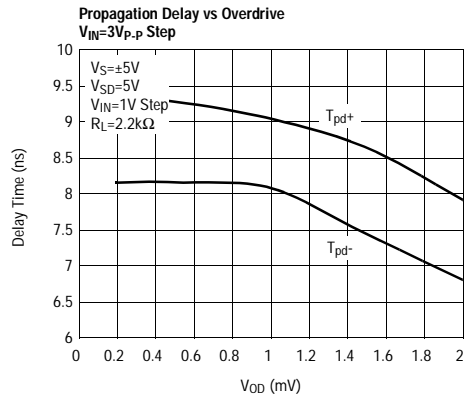
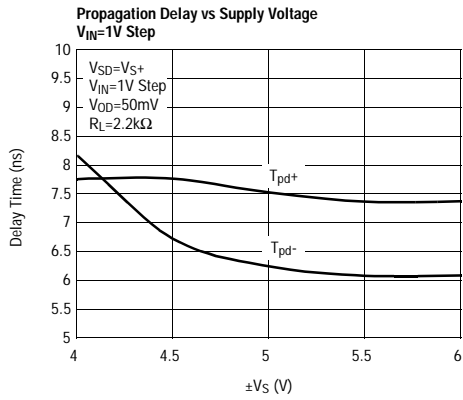
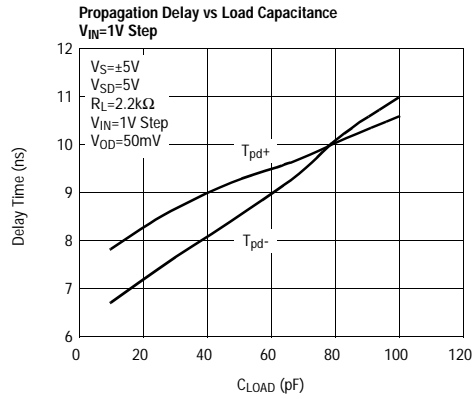
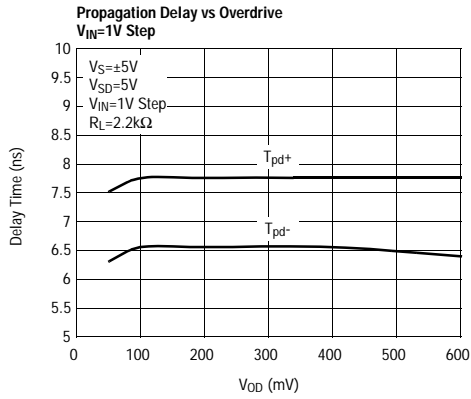
Typical Performance Curves



EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

Typical Performance Curves

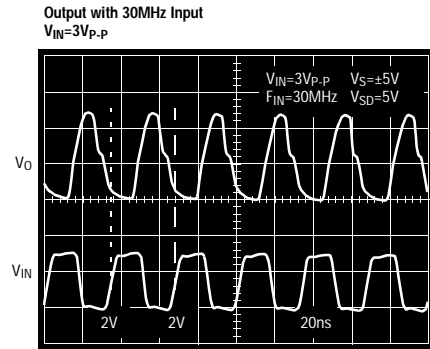
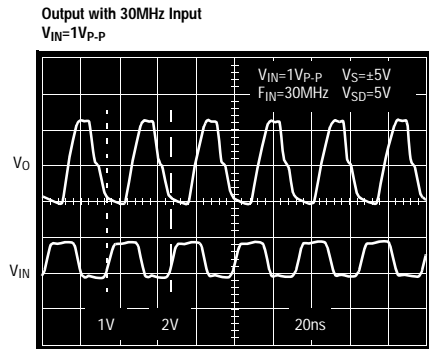
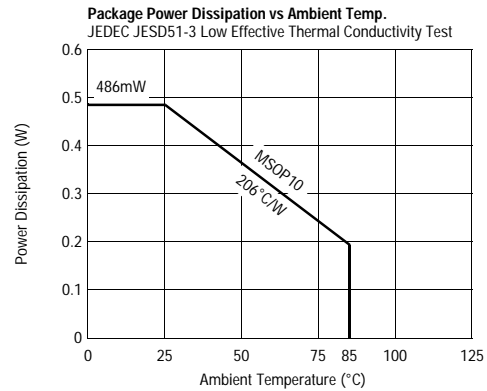
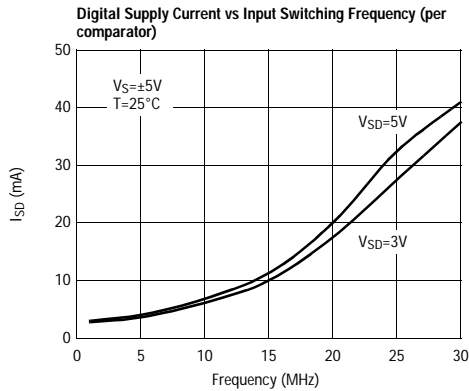
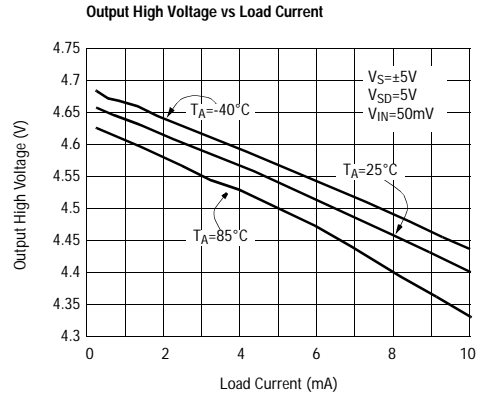
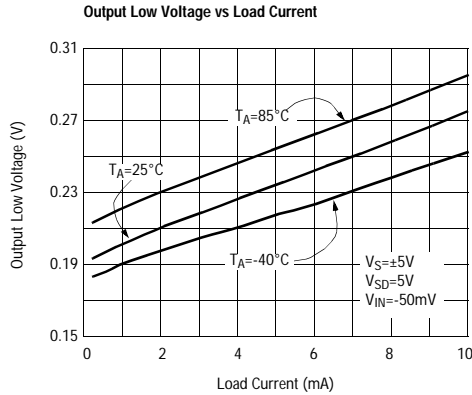


EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

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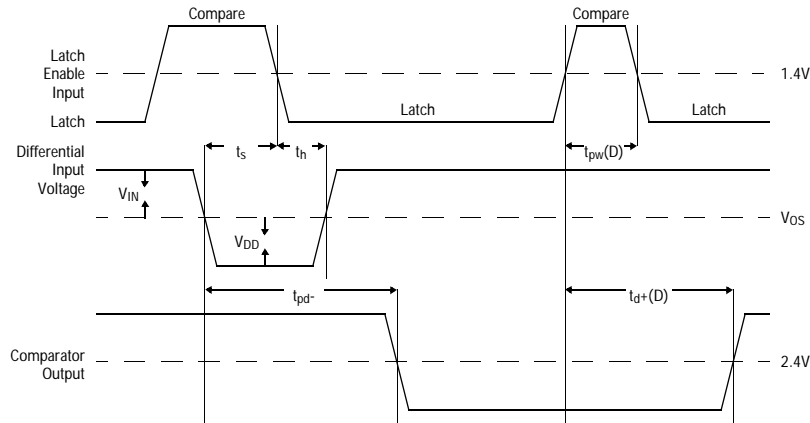
Typical Performance Curves



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Dual and Window 8ns High-Speed Comparators

Timing Diagram



Definition of Terms

Terms	Definition
V_{OS}	Input Offset Voltage - Voltage applied between the two input terminals to obtain CMOS logic threshold at the output
V_{IN}	Input Voltage Pulse Amplitude - Usually set to 100mV for comparator specifications
V_{OD}	Input Voltage Overdrive - Usually set to 5mV and in opposite polarity to V_{IN} for comparator specifications
t_{pd+}	Input to Output High Delay - The propagation delay measured from the time the input signal crosses the input offset voltage to the CMOS logic threshold of an output low to high transition
t_{pd-}	Input to Output Low Delay - The propagation delay measured from the time the input signal crosses the input offset voltage to the CMOS logic threshold of an output high to low transition
t_{d+}	Latch Disable to Output High Delay - The propagation delay measured from the latch signal crossing the CMOS threshold in a low to high transition to the point of the output crossing CMOS threshold in a low to high transition
t_{d-}	Latch Disable to Output Low Delay - The propagation delay measured from the latch signal crossing the CMOS threshold in a low to high transition to the point of the output crossing CMOS threshold in a high to low transition
t_s	Minimum Setup Time - The minimum time before the negative transition of the latch signal that an input signal change must be present in order to be acquired and held at the outputs
t_h	Minimum Hold Time - The minimum time after the negative transition of the latch signal that an input signal must remain unchanged in order to be acquired and held at the output
$t_{pw(D)}$	Minimum Latch Disable Pulse Width - The minimum time that the latch signal must remain high in order to acquire and hold an input signal change

EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

EL5283C - Preliminary

Pin Descriptions

Pin Number	Pin Name	Function	Equivalent Circuit
1	VS+	Positive supply voltage	<p style="text-align: center;">Circuit 4</p>
2	VREFH	Upper voltage reference	
3	IN	Input	(Reference Circuit 4)
4	VREFL	Lower voltage reference	(Reference Circuit 4)
5	VS-	Negative supply voltage	
6	GDN	Digital ground	
7	OUTL	Low output	(Reference Circuit 2)
8	LATCH	Latch	(Reference Circuit 3)
9	OUTH	High output	(Reference Circuit 2)
10	VSD	Digital supply voltage	

Applications Information

Power Supplies and Circuit Layout

The EL5283C comparator operates with single and dual supply with 5V to 12V between V_{S+} and V_{S-} . The output side of the comparators is supplied by a single supply from 2.7V to 5V. The rail to rail output swing enables direct connection of the comparator to both CMOS and TTL logic circuits. As with many high speed devices, the supplies must be well bypassed. Elantec recommends a 4.7 μ F tantalum in parallel with a 0.1 μ F ceramic. These should be placed as close as possible to the supply pins. Keep all leads short to reduce stray capacitance and lead inductance. This will also minimize unwanted parasitic feedback around the comparator. The device should be soldered directly to the PC board instead of using a socket. Use a PC board with a good, unbroken low inductance ground plane. Good ground plane construction techniques enhance stability of the comparators.

Input Voltage Considerations

The EL5283C input range is specified from 0.1V below V_{S-} to 2.25V below V_{S+} . The criterion for the input limit is that the output still responds correctly to a small differential input signal. The differential input stage is a pair of PNP transistors, therefore, the input bias current flows out of the device. When either input signal falls below the negative input voltage limit, the parasitic PN junction formed by the substrate and the base of the PNP will turn on, resulting in a significant increase of input bias current. If one of the inputs goes above the positive input voltage limit, the output will still maintain the correct logic level as long as the other input stays within the input range. However, the propagation delay will increase. When both inputs are outside the input voltage range, the output becomes unpredictable. Large differential voltages greater than the supply voltage should be avoided to prevent damages to the input stage. Inputs of unused channels should not be left floating. They should be driven to a known state. For example, one input can

EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

be tied to ground and the other input can be connected to some voltage reference (like $\pm 100\text{mV}$) to avoid oscillation in the output due to unwanted output to input feedback.

Input Slew Rate

Most high speed comparators oscillate when the voltage of one of the inputs is close to or equal to the voltage on the other input due to noise or undesirable feedback. For clean output waveform, the input must meet certain minimum slew rate requirements. In some applications, it may be helpful to apply some positive feedback (hysteresis) between the output and the positive input. The hysteresis effectively causes one comparator's input voltage to move quickly past the other, thus taking the input out of the region where oscillation occurs. For the EL5283C, the propagation delay increases when the input slew rate increases for low overdrive voltages. With high overdrive voltages, the propagation delay does not change much with the input slew rate.

Latch Pin Dynamics

The EL5283C contains a "transparent" latch for each channel. The latch pin is designed to be driven with either a TTL or CMOS output. When the latch is connected to a logic high level or left floating, the comparator is transparent and immediately responds to the changes at the input terminals. When the latch is switched to a logic low level, the comparator output latches remains latched to its value just before the latch high-to-low transition. To guarantee data retention, the input signal must remain the same state at least 1ns (hold time) after the latch goes low and at least 2ns (setup time) before the latch goes low. When the latch goes high, the new data will appear at the output in approximately 8ns (latch propagation delay).

Power Dissipation

When switching at high speeds, the comparator's drive capability is limited by the rise in junction temperature caused by the internal power dissipation. For reliable operation, the junction temperature must be kept below T_{JMAX} (125°C).

An approximate equation for the device power dissipation is as follows. Assume the power dissipation in the load is very small:

$$P_{DISS} = (V_S \times I_S + V_{SD} \times I_{SD}) \times N$$

where:

V_S is the analog supply voltage from V_{S+} to V_{S-}

I_S is the analog quiescent supply current per comparator

V_{SD} is the digital supply voltage from V_{SD} to ground

I_{SD} is the digital supply current per comparator

N is the number of comparators in the package

I_{SD} strongly depends on the input switching frequency. Please refer to the performance curve to choose the input driving frequency. Having obtained the power dissipation, the maximum junction temperature can be determined as follows:

$$T_{JMAX} = T_{MAX} + \theta_{JA} \times P_{DISS}$$

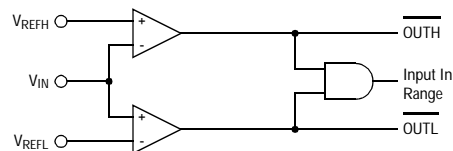
where:

T_{MAX} is the maximum ambient temperature

θ_{JA} is the thermal resistance of the package

Window Detector

If V_{IN} is in the range of $V_{REFL} < V_{IN} < V_{REFH}$, both outputs go high and the input in range is high. If V_{IN} is out of the range set by V_{REFH} and V_{REFL} , the input in range is low.



EL5283C - Preliminary

Dual and Window 8ns High-Speed Comparators

General Disclaimer

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HIGH PERFORMANCE ANALOG INTEGRATED CIRCUITS

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